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VOLUME 1: USER'S MANUAL

A COMPUTER PROGRAM FOR POWER SPECTRAL ANALYSIS OF UNEQUALLY SPACED POINTS

By

Matthew Lybanon

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ABSTRACT

A power spectral density computer program with some unique capabilities is described. The special features include

- 1. The spectrum of unequally spaced data may be calculated.
- 2. The spectrum can be evaluated at any frequencies.
- Estimates of the parameters of dominant spectral terms can be improved.
- 4. Prewhitening can be done with no prior knowledge of the characteristics of the signal.

First, the mathematical basis of the conventional computer Fourier analysis technique is discussed, with special attention to the approximations involved. Next, the problem of spectral analysis is considered as a least-squares regression problem. Formulas are developed for continuous and discrete equally spaced data. Next, the interpretation of the power spectral estimates is explored, and the possibility of evaluating the spectrum at any set of frequencies is discussed. Then, the general formulas for unequally spaced points are derived.

After the mathematical formulation comes a description of the computer program implementing the theory. This includes a discussion of the program and its use, a description of the input cards and of the form of the output, and instructions for execution. Appendices include a flowchart and a sample run.

TABLE OF CONTENTS

Introduction	
Conventional Spectral Analysis - The Discrete Fourier Transform	1
Continuous and Equally Spaced Data Derivation of Formulas Effectiveness of Spectrum Estimates	8 8 11
Unequally Spaced Data	18
Computer Program General Information Functions of Programs in System Input Cards Output Executing the Program	24 24 28 29 32 35

APPENDIX 1 - Overall Logic Flowchart APPENDIX 2 - Sample Run

Page

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INTRODUCTION

This report describes a power spectral density computer program which has several features not commonly found in such programs. Perhaps the most unusual feature is its ability to analyze unequally spaced data. In addition, the spectrum can be evaluated at frequencies other than harmonics of the span of data, the maximum frequency at which the spectrum is evaluated can be chosen to be other than one-half the average data sampling rate, and the initial estimates of the frequencies and coefficients of dominant spectral terms can be improved. This last feature in effect permits a reconstruction of the original data as a sum of sinusoidal terms (not a classical Fourier series), which may accurately represent the signal without noise. In addition, it is used for "prewhitening" the spectrum, a procedure recommended by several authors.

CONVENTIONAL SPECTRAL ANALYSIS -THE DISCRETE FOURIER TRANSFORM

The scientist or engineer has an intuitive feeling about what he means by a "spectrum." However, there are certain differences between the Fourier transform and a computer-calculated spectrum. At present, the latter very often is obtained by the fast Fourier transform algorithm [References 1-5], which is an efficient way of calculating the finite or discrete Fourier transform. If X(j), $j = 0, 1, \ldots, N-1$, is a sequence of N complex numbers, the discrete Fourier transform of X(j) is defined

$$A(n) = \frac{1}{N} \sum_{j=0}^{N-1} X(j) \exp(-2\pi i n j/N)$$
(1)

X(j) can be expressed as the inverse finite Fourier transform of A(n)

$$X(j) = \sum_{n=0}^{N-1} A(n) \exp(2\pi i n j/N)$$
(2)

A(n) and X(j) are a transform pair, as can be seen by substituting (1) into (2) and using the orthogonality relationship

$$\sum_{j=0}^{N-1} \exp(2\pi i n j/N) \exp(-2\pi i n j/N) = \begin{cases} N \text{ if } n = m \mod N \\ 0 \text{ otherwise} \end{cases}$$
(3)

Let A(n) and X(j) be considered to be defined by (1) and (2) for all integers n and j. Then it follows from the definitions that

$$X(J) = X(kN+J) A(n) = A(kN+n)$$
 k = 0, ±1, ±2, . . . (4)

That is, X(j) and A(n) are both periodic, of period N.

Equations (1) and (2) are simply operations performed on sets of numbers, but form (1) is reminiscent of the Fourier integral transform, or of Fourier series coefficients. The discrete Fourier transform possesses an inverse, (2), it involves functions which satisfy an orthogonality relation, (3). The nature of the relationship between the Fourier transform and Fourier series on the one hand, and the discrete Fourier transform on the other, will now be examined further.

Two pertinent theorems involve the concept of a <u>periodic aliased function</u>. The periodic aliased version $a_{p}(f)$ of a function a(f) is defined to be

$$a_{p}(f) = \sum_{k=-\infty}^{\infty} a(f+kF)$$
(5)

where F is some interval. It is proved in References 2 and 3 that Theorem 1 If X(t), $-\infty < t < \infty$, and a(f), $-\infty < f < \infty$, are a Fourier integral transform pair, then TX_p (j Δ t) = T $\sum_{l=-\infty}^{\infty} X(j\Delta t+ 1T)$, j=0,1,...,N-1, and $a_p(n\Delta f) = \sum_{k=-\infty}^{\infty} a(n\Delta f+kF)$, n=0,1,...,N-1, are a finite Fourier transform pair. In these expressions F = 1/ Δ t = N Δ f, T = N Δ t = N/N Δ f = 1/ Δ t, FT=N. It is also shown in these references that

Theorem 2 If the periodic function X(t) with period T has the Fourier series expansion c(n) (i.e. c(n) are the coefficients in the complex form of the

Fourier series, $X(t) = \sum_{n=-\infty}^{\infty} c(n)e^{2\pi i(n+/T)}$, then the periodic sequence

 $X(j\Delta t)$ of period N, where $\Delta t = T/N$, has the finite Fourier transform

$$c_p(n) = \sum_{l=-\infty}^{\infty} c(n+lN).$$

These two theorems appear to say essentially the same thing. The resemblance is more than apparent, as the following heuristic argument shows. The usual "derivation" of the Fourier transform, by going to the limit in which the number of terms in the Fourier series increases while the frequency spacing decreases, and replacing the sum by an integral, shows that the Fourier series and integral are indeed very similar. The basic difference is that the integral transform maps one nondenumerably infinite number of points into another nondenumerably infinite set of numbers, whereas the Fourier series specifies the function by a denumerably infinite set of numbers. (Not all functions have Fourier series representations, only periodic ones.) In a numerical problem, one has a finite number of data points (containing a finite amount of information, hence an incomplete specification of the physical process), which can be completely specified by another equal-sized set of numbers. Whether these numbers are regarded as estimates of Fourier series coefficients, or of points on the Fourier transform curve, is largely a matter of taste. (This situation is discussed further in the next section.) At various places in the subsequent discussion, whichever supposition is more useful will be used.

The first problem is a consideration of how to apply Theorems 1 and 2 to the calculation of a spectrum when all that one has is a finite-length sample of experimental data, a situation which often arises in practice. If X(t)actually vanishes outside the interval $0 \le t < T$, then $X_p(j\Delta t) = X(j\Delta t)$, j=0, 1, ..., N-1. At this point it should be noted that it follows directly from (4) that

$$A(-n) = A(N-n) \tag{6}$$

Using this result it can be seen that if the Fourier transform a(f) is nonzero only for |f| < F/2 (recall that a(f) is defined for negative as well as positive f, and |a(-f)| = |a(f)| only if X(t) is real), then a $(n\Delta f) = a(n\Delta f)$, $n=0, 1, \ldots, N-1$. So in this case, Theorem 1 can be restated

If X(t), non-zero only in the range $0 \le t \le T$, and a(f), non-zero over $-F/2 \le f \le F/2$, are a Fourier integral transform pair, then TX(j Δ t), j=0,1, . . ., N-1, and a(n Δ f), n=0,1, . . ., N-1, are a finite Fourier transform pair.

For the simple case considered here, the relationship between the Fourier transform and the discrete Fourier transform can easily be found. The Fourier integral transform is

$$a(f) = \int_{-\infty}^{\infty} X(t) \exp(-2\pi i ft) dt$$
(7)

For X(t) non-zero only in the range $0 \le t < T$,

$$a(f) = \int_0^T X(t) \exp\left(-2\pi i f t\right) dt$$
(8)

The integral can be approximated by a sum

$$a(f) \approx \sum_{j=0}^{N-1} X(j\Delta t) \exp(-2\pi i f j\Delta t) \Delta t$$
$$= \frac{1}{N} \sum_{j=0}^{N-1} (N\Delta t) X(j\Delta t) \exp(-2\pi i f j\Delta t)$$
(9)

For $f=n\Delta f$, $n=0, 1, \ldots, N-1$,

$$a(n\Delta f) \approx \frac{1}{N} \sum_{j=0}^{N-1} (N\Delta t) \times (j\Delta t) \exp(-2\pi \ln\Delta f j \Delta t)$$

Using the relations $T=N\Delta t$, $\Delta f = F/N = 1/N\Delta t$,

$$a(n\Delta f) \approx \frac{1}{N} \sum_{j=0}^{N-1} T X(j\Delta t) \exp(-2\pi i n j/N)$$
(10)

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But by the modified Theorem 1 stated above, the equality in (10) is exact, that is, at the points $f=n\Delta f$, $n=0,1,\ldots$, N-1, no error is involved in replacing the integral of equation (8) by the sum of equation (9).

If X(t) does not vanish outside the interval $0 \le t \le T$, equation (8) is the Fourier transform of the product of X(t) and a square pulse of value unity on $0 \le t \le T$, and value zero elsewhere. This transform is the convolution of the transform of X(t) with the transform of the square pulse. The effect of this is to "smear out" each line in the spectrum of X(t).

This case can be viewed in another way. The equality in (10) is not necessarily exact, as was the case for the previous example, because $X(j\Delta t) \neq X_p(j\Delta t)$ (and $a(n\Delta f)$ will not equal $a_p(n\Delta f)$ in an arbitrary case). So there will be some error in representing the Fourier transform by the discrete Fourier transform, in general.

In fact there is always an error if a(f) does not vanish for all |f| > F/2. It is convenient for this demonstration to discuss Fourier series (as was stated before, as far as representing them by the discrete Fourier transform is concerned, Fourier series and the Fourier integral are essentially the same thing). The complex form of the Fourier series representation of a function X(t) is

$$X(t) = \frac{1}{2} \sum_{n=-\infty}^{\infty} C_{n} \exp (2\pi \ln t/T)$$

$$\therefore X(j\Delta t) = \frac{1}{2} \sum_{n=-\infty}^{\infty} C_{n} \exp (2\pi \ln j/N), \ j=0, 1, ..., N-1$$
(11)

Comparison with equation (2), using relation (6), shows that the finite Fourier transform gives a <u>finite</u>, i.e. truncated version of the Fourier series representation of a function, limited to frequencies under F/2. Reference 8 thoroughly discusses the error that this causes. Furthermore, the second of equations (4), A(n)=A(kN+n), $k=0,\pm 1,\pm 2,\ldots$, (and Theorems 1 and 2) implies that if there are frequencies above F/2, the discrete Fourier transform will show these frequencies as occurring below F/2. This is the well known phenomenon of <u>aliasing</u>, the possibility of which always exists when one is dealing with equally spaced data. This result can also be proved in another manner. [7]

At the conclusion of a discussion of aliasing in their important book [9], Blackman and Tukey make the comment "...It is not infrequently suggested that there should be a workable scheme of taking discrete data in some definite, but not uniformly spaced pattern, and estimating the power spectrum without aliasing. No such scheme seems so far to have been developed..." The subject of this report is a method of estimating a power spectrum from unequally spaced samples, in which case indeed there is no aliasing. However, before getting to a discussion of this technique, a few more preliminary remarks will be made.

Blackman and Tukey also discuss the obvious possibility that the length of the sampling interval might not be ideally chosen. For instance, a seemingly aperiodic function might be seen to repeat exactly if it were sampled long enough, or a seemingly periodic function might stop repeating eventually. Of course, it is impossible to know from examining a data sample what <u>might</u> happen if the function were sampled for a longer time, but another example can be given of a possible difficulty related to the size of the data interval. In Reference 7 the function

-6-

$$y(t) = B \cos \left[\frac{2\pi (k_0 + 1/2)}{T} t \right], \quad 0 \le t < T$$
 (12)

where k_0 is an integer, is considered. The key fact is that the interval does not contain a whole number of cycles of the cosine. The Fourier series coefficients of this function are

$$a_{k} = 0$$

$$b_{k} = \frac{B}{\pi} \left[\frac{1}{k + (k_{0} + 1/2)} + \frac{1}{k - (k_{0} + 1/2)} \right], \quad k = 1, 2, 3, \dots$$
(13)

Although the function in closed form is a cosine, the only non-zero Fourier series coefficients are the sine coefficients. But there is no mistake. The (infinite) Fourier series converges to y(t) at every interior point of the interval [10], and if y(t) is sampled at N equally spaced points spanning the interval, the (finite) discrete Fourier transform will specify a trigonometric series passing exactly through the N points.

The paradox is only apparent. Because the frequency of the cosine in (12) is not a harmonic of the basic period T, the Fourier series does not contain any terms of that frequency. The function y(t) may continue outside $0 \le t \le T$. However, the Fourier series is periodic with period T, so it does not represent y(t) in the interval $T \le t \le 2T$, for instance.

The (conventional, or restricted) sampling theorem summarizes some of the properties and limitations of conventional spectral analysis.[9] References 11 and 12 discuss the generalization of the sampling theorem to the case of unequally spaced samples, and applications to the reconstruction of signals from non-uniform samples. It is pointed out that a generalized sampling theorem was known to Cauchy over a hundred years ago.

CONTINUOUS AND EQUALLY SPACED DATA

Derivation of Formulas

The Fourier integral and Fourier series converge at every point to the functions they represent in many cases, and in every case they minimize the mean square error. Furthermore, the sum of the first M terms of a Fourier series, where M is any positive integer, is the best least-squares fit of a trigonometric sum of that order (involving those frequencies). [10]

The following discussion is presented in greater detail in References 6 and 7. Suppose the problem is approached from the opposite direction, as a regression problem in which it is desired to approximate a function y(t) in an interval $0 \le t \le T$ by a trigonometric sum, $y(t) = 0 \le t \le T$

$$y_{c}(t) = \frac{a_{0}}{2} + \sum_{k=1}^{L} \left(a_{k} \cos 2\pi k f t + b_{k} \sin 2\pi k f t \right), f=1/T$$
 (14)

Applying the requirement that the mean square error on the interval be a minimum gives the following formulas for the coefficients

$$a_{0} = \frac{2}{T} \int_{0}^{T} y(t) dt$$

$$a_{k} = \frac{2}{T} \int_{0}^{T} y(t) \cos 2\pi k ft dt$$

$$b_{k} = \frac{2}{T} \int_{0}^{T} y(t) \sin 2\pi k ft dt$$

$$k = 1, 2, \dots, L$$
(15)

which are the usual expressions for Fourier series coefficients. (The well - known orthogonality relationships make it unnecessary to invert a large matrix to obtain these formulas.)

Suppose we consider only one term of the series in (14),

$$y_{k}(t) = a_{k} \cos 2\pi kft + b_{k} \sin 2\pi kft$$
(16)

Let $\varepsilon(t)=y(t) - y_k(t)$. Then a straightforward calculation gives

$$\frac{1}{T} \left[\int_{0}^{T} y^{2}(t) dt - \int_{0}^{T} e^{2} dt \right] = \frac{1}{2} \left(a_{k}^{2} + b_{k}^{2} \right)$$
(17)

If y(t) represents voltage, for instance, then $y^2(t)$ is the power dissipated in a unit resistance, and $\varepsilon^2(t)$ is the power not contained in the term of frequency kf. Therefore, the left-hand side of (17) is the average power contained in the term (16). This analogy is the justification for plotting $a_k^2 + b_k^2$ against k or f and calling it the power spectrum of y(t).

The case of discrete equally spaced data is very similar. We want to fit a model

$$\int_{y}^{\Lambda} (n) = \frac{a_{0}}{2} + \sum_{k=1}^{L} \left(a_{k} \cos 2\pi k fn + b_{k} \sin 2\pi k fn \right), f=1/N$$
 (18)

to a function $y_n = y(n \Delta t)$ which takes on values at n=1,2,..., N (the y_n are though of as samples of a function y(t) of a continuous time variable). To do so we make use of identities (orthogonality relations) completely analogous to those for the continuous case,

$$\sum_{n=1}^{N} \cos(2\pi kn/N) = \sum_{n=1}^{N} \sin(2\pi kn/N) = 0$$

$$\sum_{n=1}^{N} \cos^{2}(2\pi kn/N) = \sum_{n=1}^{N} \sin^{2}(2\pi kn/N) = N/2$$

$$\sum_{n=1}^{N} \cos(2\pi kn/N) \cos(2\pi pn/N) = 0 \text{ if } p \neq k$$

$$\sum_{n=1}^{N} \sin(2\pi kn/N) \sin(2\pi pn/N) = 0 \text{ if } p \neq k$$

$$\sum_{n=1}^{N} \sin(2\pi kn/N) \cos(2\pi pn/N) = 0 \text{ if } p \neq k$$
(19)

where k and p are integers. Applying these to the problem gives

$$a_{0} = \frac{2}{N} \sum_{n=1}^{N} y_{n}$$

$$a_{k} = \frac{2}{N} \sum_{n=1}^{N} y_{n} \cos (2\pi k n/N)$$

$$b_{k} = \frac{2}{N} \sum_{n=1}^{N} y_{n} \sin (2\pi k n/N)$$
(20)
(20)

The similarity to (15) should be noted.

At this point it is appropriate to again mention aliasing, which is caused by the fact that $\cos \left[2\pi \left(\frac{N}{2} + \varepsilon\right) fn + \phi\right]$ and $\cos \left[2\pi \left(\frac{N}{2} - \varepsilon\right) fn - \phi\right]$ are equal for f=1/N and n=1,2,..., N. So there is a frequency $f_{max} = 1/2\Delta t$ such that frequencies above f_{max} appear to be below f_{max} , possibly with a phase shift. This is often discussed in connection with the sampling theorem, for equally spaced data, it is necessary to sample at a rate of at least twice the highest frequency in the signal to extract all the information. The frequency $f_{max} = 1/2\Delta t$ is known as the Nyquist or folding frequency.

From the point of view of fitting a function to N data points, aliasing is related to the fact that only N independent fit parameters can be found. If N is odd, (N-1)/2 is the greatest integer not exceeding N/2. If this value is chosen for L in (18), there is one a₀ coefficient, (N-1)/2 a_k coefficients, and (N-1)/2 b_k coefficients, making N in all If N is even, N/2 is an integer. It can be seen from (20) that b_k = 0 if k = N/2, so with L = N/2

-10-

there is one a $_0$, N/2 a $_k$, and $\left(\frac{N}{2}$ - 1 \right) b $_k$, making N all together.

Proceeding exactly as was done in obtaining (17), if $y_k(n) = a_k \cos 2\pi k fn + b_k \sin 2\pi k fn$ and $e_n = y_n - y_k(n)$, then it is easy to show that

$$\frac{1}{N} \left[\sum_{n=1}^{N} y_n^2 - \sum_{n=1}^{N} \varepsilon_n^2 \right] = \frac{1}{2} \left(a_k^2 + b_k^2 \right)$$
(21)

There are seen to be great similarities between the continuous case (Fourier series) and the discrete equally spaced case. Hamming [13], in discussing the use of orthogonal functions, explains that it is often difficult to decide between the two approaches of assuming that one has continuous functions and approximating the integrals giving the coefficients by numerical methods, or using the discrete orthogonal set of functions to fit the discrete samples. But he goes on to point out that in the Fourier series case these two approaches produce exactly the same computations.

Effectiveness of Spectrum Estimates

Before proceeding to the case of unequally spaced discrete data, some additional discussion of the interpretation of the power spectrum will be presented. Equations (17) and (21) are special ways of writing Parseval's theorem. This theorem, in the continuous frequency case (Fourier transform), can be written

$$\int_{-\infty}^{\infty} \left[X(t) \right]^{2} dt = \int_{-\infty}^{\infty} |a(f)|^{2} df$$
(22)

where the notation is that of Theorem 1 (X(t) is considered real). In the

discrete frequency case (Fourier series),

$$\frac{1}{T}\int_{0}^{T}\left[y(t)\right]^{2}dt = \frac{1}{2}\left[\frac{a_{0}^{2}}{2} + \sum_{k=1}^{\infty}\left(a_{k}^{2} + b_{k}^{2}\right)\right]$$
(23)

In case one is dealing with discrete equally spaced samples of data,

$$\frac{1}{N}\sum_{n=1}^{N}y_{n}^{2} = \frac{1}{2}\left[\frac{a_{0}^{2}}{2} + \sum_{k=1}^{N}\left(a_{k}^{2} + b_{k}^{2}\right)\right]$$
(24)

The discussion concerning equations (14) - (17) shows that obtaining the Fourier transform of a time series is equivalent to fitting a series of independent functions of the form

$$y_{k}(t) = a_{k} \cos 2\pi kft + b_{k} \sin 2\pi kft$$
(25)

where f = 1/T, T the length of the data interval. It was pointed out that the actual power in the signal might occur at any frequency. The digitized data to be analyzed in a practical case consists of discrete samples of some continuous process sampled over a time T. This time limit T may be imposed by practical considerations having nothing to do with true periodicities of the signal, the frequencies in the data may not be convenient integral multiples of 1/T. It would be wise, therefore, to explore the sensitivity of discrete spectrum estimates to the continuum of frequencies which can occur in the signal itself.

For simplicity (25) will be rewritten in the equivalent form

$$y_{l_{r}}(t) = A_{l_{r}} \sin(2\pi kft + \phi)$$
 (26)

where $A_k = (a_k^2 + b_k^2)^{\frac{1}{2}}$ and $\phi = \tan^{-1}(a_k/b_k)$. The absolute phase ϕ will not affect the results to be obtained and will be dropped.

From this point on, in anticipation of certain results, it will not be assumed that f = 1/T necessarily. The integral to be minimized in performing the least squares fit is

$$I = \int_{0}^{T} e^{2} dt = \int_{0}^{T} [y(t) - A_{k} \sin 2\pi kft]^{2} dt$$
 (27)

The value of A_k which minimizes I is

$$A_{k} = \frac{\int_{0}^{T} y(t) \sin 2\pi kft \, dt}{\int_{0}^{T} \sin^{2} 2\pi kft \, dt}$$
(28)

It can easily be shown that, if $F = (k + \delta)/T$, where k is an integer and $|\delta| < 1$,

$$J = \int_{0}^{T} \sin^{2} 2\pi Ft \, dt = \frac{T}{2} - \frac{T \sin 4\pi \delta}{8\pi (k+\delta)} = \frac{T}{2} \left[1 - \frac{\sin 4\pi \delta}{4\pi (k+\delta)} \right]$$
(29)

If $\delta = 0$ (or $\pm \frac{1}{4}, \pm \frac{1}{2}, \pm \frac{3}{4}$), J=T/2, a result which was used in obtaining equations (15). For arbitrary values of δ , even when k is a fairly small integer, the second term in (29) is much smaller than the first. So J is seldom very different from T/2. Assuming J=T/2, equation (28) becomes

$$A_{k} = \frac{2}{T} \int_{0}^{T} y(t) \sin 2\pi k ft dt$$
(30)

Substituting the expression (28) for A_{l_k} into (27), the result is

$$\int_{0}^{T} \varepsilon^{2} dt = \int_{0}^{T} \left[y(t) - A_{k} \sin 2\pi kft \right]^{2} dt$$
$$= \int_{0}^{T} y^{2}(t) dt - \left[\int_{0}^{T} y(t) \sin 2\pi kft dt \right]^{2} / J$$
(31)

Then as the discussion following equation (17) shows, the power associated with $y_k(t)$ is

$$P_{k} = \frac{1}{T} \left[\int_{0}^{T} y^{2}(t) dt - \int_{0}^{T} \varepsilon^{2} dt \right] = \frac{1}{T} \left[\int_{0}^{T} y(t) \sin 2\pi k ft dt \right]^{2} / J \quad (32)$$

Using the value J=T/2,

$$P_{k} = \frac{2}{T^{2}} \left[\int_{0}^{T} y(t) \sin 2\pi k ft dt \right]^{2}$$
(33)

Suppose y(t) contains a component of frequency θ , with relative phase α , so that the contribution of this term to the total signal is

$$y(\theta, \alpha) = C \sin(2\pi\theta t + \alpha)$$
 (34)

The contribution that this term makes to P_k is

$$P_{k}(\theta, \alpha) = \frac{C^{2}}{TJ} \left[\int_{0}^{T} \sin 2\pi k ft \sin(2\pi\theta t + \alpha) dt \right]^{2}$$
(35)

The expression in brackets gives the functional dependence on θ and α

$$\begin{bmatrix} \int_{0}^{T} \sin 2\pi \operatorname{kft} \sin(2\pi \theta t + \alpha) \, dt \end{bmatrix}^{2}$$

$$= \frac{1}{4} \left\{ \int_{0}^{T} \cos \left[2\pi \left(\operatorname{kf} - \theta \right) t - \alpha \right] \, dt - \int_{0}^{T} \cos \left[2\pi \left(\operatorname{kf} + \theta \right) t + \alpha \right] \, dt \right\}^{2}$$

$$= \frac{1}{4} \left\{ \frac{\sin \left[2\pi \left(\operatorname{kf} - \theta \right) T - \alpha \right] + \sin \alpha}{2\pi \left(\operatorname{kf} - \theta \right)} - \frac{\sin \left[2\pi \left(\operatorname{kf} + \theta \right) T + \alpha \right] - \sin \alpha}{2\pi \left(\operatorname{kf} + \theta \right)} \right\}^{2}$$
(36)

 P_k (θ, α) is a weighting function in a sum or integral over θ , to give the power estimate at frequency kf due to contributions from all possible frequencies θ . (A sum over α is also implied.) Unless kf is small, this weighting function will be relatively insignificant except for values of θ near kf. For such values of θ the second term is much smaller than the first. If the second term is dropped,

$$\begin{split} P_{k}(\theta, \alpha) &= \frac{C^{2}}{4T J} \left\{ \frac{\sin[2\pi(kf-\theta) T - \alpha] + \sin\alpha}{2\pi(kf-\theta)} \right\}^{2} \\ \left\{ \sin[2\pi(k-\theta) T - \alpha] + \sin\alpha \right\}^{2} \\ &\stackrel{\prime}{=} \left\{ \sin[2\pi(kf-\theta) T] \cos\alpha - \cos[2\pi(kf-\theta)T] \sin\alpha + \sin\alpha \right\}^{2} \\ &= \left\{ 2\sin[2\pi(kf-\theta) T/2] \cos[2\pi(kf-\theta)T/2] \cos\alpha + 2\sin^{2}[2\pi(kf-\theta)T/2] \sin\alpha \right\}^{2} \\ &= 4\sin^{2}[2\pi(kf-\theta)T/2] \left\{ \cos[2\pi(kf-\theta)T/2] \cos\alpha + \sin[2\pi(kf-\theta)T/2] \sin\alpha \right\}^{2} \\ &= 4\sin^{2}[2\pi(kf-\theta)T/2] \sin\alpha \right\}^{2} \\ &= 4\sin^{2}[2\pi(kf-\theta)T/2] \sin\alpha \right\}^{2} \\ &= 4\sin^{2}[2\pi(kf-\theta)T/2] \cos^{2}[2\pi(kf-\theta)T/2 - \alpha] \\ P_{k}(\theta, \alpha) &= \frac{C^{2}}{TJ} \sin^{2}[2\pi(kf-\theta)T/2] \cos^{2}[2\pi(kf-\theta)T/2 - \alpha] / [2\pi(kf-\theta)]^{2} \end{split}$$
(37)

Examination of the exact expression (36) shows that for kf a multiple of 1/T, $P_k(\theta, \alpha)$ vanishes when θ is also a multiple of 1/T. This is a manifestation of the orthogonality of the terms in a Fourier series. Equation (37) shows that this property is more general, $P_k(\theta, \alpha)$ vanishes whenever kf- θ is any integral multiple of 1/T, regardless of the values of kf and θ (unless kf is small and (37) doesn't hold).

The detailed behavior of $P_k(\theta, \alpha)$ depends on α . For instance, if $\alpha = 0$ equation (37) has zeros when kf- θ is any multiple of 1/2T, twice as often as in the general case. For certain values of α (depending on θ and T) $P_k(\theta, \alpha)$ is identically zero. But for an arbitrary signal the contribution at frequency θ may have any phase, so an expression independent of α would be useful. A reasonable estimate for the effect of the contribution at θ on the power estimate at kf is the expected value $P_k(\theta) = E_{\alpha} \left[P_k(\theta, \alpha) \right]$, where the

.

subscript α denotes that the average is over α . Assuming all values of α are equally likely, this involves

$$\frac{1}{2\pi} \int_{0}^{2\pi} \left\{ \cos\left[2\pi(\mathrm{kf}-\theta)\mathrm{T}/2\right]\cos\alpha + \sin\left[2\pi(\mathrm{kf}-\theta)\mathrm{T}/2\right]\sin\alpha \right\}^{2} \,\mathrm{d}\alpha = \frac{1}{2} \tag{38}$$

Therefore, the expected value of the contribution which a component of amplitude c, frequency θ , and arbitrary phase makes to the power estimate at frequency kf is

$$P_{k}(\theta) = \frac{T_{c}}{8J} \left[\frac{\sin 2\pi (kf - \theta) T/2}{2\pi (kf - \theta) T/2} \right]^{2}$$
(39)

The gain factor is actually independent of T, since J is proportional to T/2. Equation (39) can be written in a useful form by defining δ ,

$$kf - \theta = \delta/T$$
(40)

and using the (very good) approximation that J=T/2. Then

$$P_{k}(\theta) = \frac{c^{2}}{4} \left[\frac{\sin \pi \delta}{\pi \delta}\right]^{2}$$
(41)

The foregoing shows that the mathematical method of power spectrum analysis gives essentially the same result as filtering the signal through a bank of filters, each with frequency response given by (41), centered at the frequencies kf, $k=1,2,\ldots$. The power estimate at kf is the power at the output of the filter centered at kf. The shape (in particular, the width) of the filter is independent of the center frequency. This process of filtering is precisely the analog technique of spectral analysis. [9, 14]

An examination of the function $\left(\frac{\sin \pi \delta}{\pi \delta}\right)^2$ shows that the estimate of power at the frequency kf includes most of the power at frequencies close to kf. The power at frequencies $\pm 1/T$ away is completely filtered out, and frequencies

-16-

farther away than this make very little contribution. (An extremely large component occurring near the peak of one of the closest side lobes may cause an erroneously high power estimate at kf.) A "worst-case" example is f=1/T, and a component of the signal at (k+1/2)/T. For this case, about 41% of the component's power would be seen at k/T and the same amount at (k+1)/T.

The analysis from the paragraph following equation (25) to this point has shown that one need not be restricted to making power estimates only at the frequencies k/T, k=1,2,... If estimates are made at frequencies closer together than this the neighboring estimates will be correlated, but finer resolution of the spectrum will be obtained. If the spectrum varies slowly with frequency, computer time can be saved by making estimates further apart than 1/T, the correlation between estimates will be slight, if not zero. If there is some advance knowledge of the spectrum (perhaps based on the properties of the physical system producing the signal), f can be chosen equal to a fundamental frequency, possibly far from 1/T. In the case considered in equations (12) and (13) f (or kf, k=2, 3...) could be chosen equal to $(k_0 + 1/2)/T$, with results more appealing to intuition.

The approximations used in obtaining equations (37), (39), and (41) are good, except for values of k and f such that kf is of the order of or smaller than 1/T, particularly for $\theta < f$. Physically, this is connected with the fact that a very low frequency must be observed for a very long time to be accurately estimated. Therefore, a constant (bias, d.c. component) may "contaminate" low frequency power estimates and should be subtracted off before spectral analysis.

-17-

The mathematical problem for the unequally spaced case is almost the same as for points equally spaced in time. Formulas will be obtained proceeding by analogy with the equally spaced case, keeping in mind the results of the previous section.

A set of points is given at times t_1 with values $y_1 = y(t_1)$, i=1, 2, ..., N, sampled during an interval T. The model to be fit is

$$\frac{\Lambda}{y}(t_{1}) =, \frac{a_{0}}{2} + \sum_{k=1}^{L} \left(a_{k} \cos 2\pi k f t_{1} + b_{k} \sin 2\pi k f t_{1}\right)$$
(42)

where, keeping in mind the discussion of the previous section, no restrictions have been placed on the value of f_{c}

The least squares equations can be formed as was done in obtaining (20). However, in the general unequally spaced case there are no orthogonality relationships like (19), that is,

$$\sum_{i=1}^{N} \cos 2\pi kft_{i} \neq 0 \qquad \sum_{i=1}^{N} \sin 2\pi kft_{i} \neq 0$$

$$\sum_{i=1}^{N} \cos^{2} 2\pi kft_{i} \neq N/2 \qquad \sum_{i=1}^{N} \sin^{2} 2\pi kft_{i} \neq N/2$$

$$\sum_{i=1}^{N} \cos 2\pi kft_{i} \cos 2\pi pft_{i} \neq 0 \qquad \text{if } p \neq k \qquad (43)$$

$$\sum_{i=1}^{N} \sin 2\pi kft_{i} \sin 2\pi pft_{i} \neq 0 \qquad \text{if } p \neq k$$

$$\sum_{i=1}^{N} \sin 2\pi kft_{i} \cos 2\pi pft_{i} \neq 0$$

even for f=1/T. In other words, the estimates of a_k , b_k are not independent of a_1 , b_1 for $j \neq k$, and a_k is not independent of b_k .

At this point, it appears that it is necessary to invert a large matrix in order to fit (42) to the data. However, making use of the filter concept of the previous section, it will be assumed that frequencies 1/T apart do not not interfere with one another. (The same assumptions as before concerning changing the frequency spacing will also be made.) Therefore, the problem considered is that of fitting a series of (assumed) independent functions of the form

$$y_{k}(t_{1}) = a_{k} \cos 2\pi kft_{1} + b_{k} \sin 2\pi kft_{1}$$
 (44)

The least squares normal equations are (dropping subscripts on a and b)

1

$$\begin{bmatrix} \sum_{1} \cos^{2} 2\pi \operatorname{kft}_{1} & \sum_{1} \sin 2\pi \operatorname{kft}_{1} \cos 2\pi \operatorname{kft}_{1} \\ \sum_{1} \sin 2\pi \operatorname{kft}_{1} \cos 2\pi \operatorname{kft}_{1} & \sum_{1} \sin^{2} 2\pi \operatorname{kft}_{1} \\ 1 & 1 \end{bmatrix} \begin{bmatrix} a \\ a \\ b \\ \vdots \end{bmatrix}_{1} y_{1} \cos 2\pi \operatorname{kft}_{1} \\ y_{1} \sin 2\pi \operatorname{kft}_{1} \end{bmatrix}$$

$$(45)$$

with solutions

$$a = \frac{\sum_{l=1}^{1} \sin^{2} 2\pi k ft_{1} \sum_{l=1}^{l} y_{l} \cos 2\pi k ft_{1} - \sum_{l=1}^{1} \sin 2\pi k ft_{1} \cos 2\pi k ft_{1} \sum_{l=1}^{l} y_{1} \sin 2\pi k ft_{1}}{\sum_{l=1}^{l} \cos^{2} 2\pi k ft_{1} \sum_{l=1}^{l} \sin^{2} 2\pi k ft_{1} - (\sum_{l=1}^{l} \sin^{2}\pi k ft_{1} \cos 2\pi k ft_{1})^{2}}$$
(46)

$$b = \frac{\sum_{1}^{1} \cos^{2} 2\pi kft_{1} \sum_{1}^{1} y_{1} \sin 2\pi kft_{1} - \sum_{1}^{1} \sin 2\pi kft_{1} \cos 2\pi kft_{1} \sum_{1}^{1} y_{1} \cos 2\pi kft_{1}}{\sum_{1}^{1} \cos^{2} 2\pi kft_{1} \sum_{1}^{1} \sin^{2} 2\pi kft_{1} - {\sum_{1}^{1} \sin 2\pi kft_{1} \cos 2\pi kft_{1}}^{2}}$$

$$(47)$$

These are the estimates of the coefficients at a given frequency. To go with them we need an expression for the power at that frequency (in the sense discussed in the previous section). Since spectral analysis is being considered as a least-squares regression problem, the natural measure of the power at frequency kf is the variance $[E(y^2)-E(e^2)]$, where $e_1 = y(t_1)-y_k(t_1)$, and $y_k(t_1)$ is defined by equation (44). It is shown in Reference 6 that this definition makes possible meaningful estimates of the significance of contributions from specific frequencies. Furthermore, it will be shown that this definition leads to results consistent with equations (17) and (21), which express power as might be expected from an engineering point of view.

Before working out the expression for power, it is convenient to express a and b in slightly different forms --

$$a = \frac{\sum_{1}^{1} y_{1} \cos 2\pi kft_{1} - \delta \sum_{1}^{1} y_{1} \sin 2\pi kft_{1}}{\sum_{1}^{1} \cos^{2} 2\pi kft_{1} - \delta \sum_{1}^{1} \sin 2\pi kft_{1} \cos 2\pi kft_{1}}$$
(48a)

$$\delta = \frac{\sum \sin 2\pi k ft_1 \cos 2\pi k ft_1}{\sum \sin^2 2\pi k ft_1}$$
(48b)

$$\sum_{i=1}^{n} y_{i} \sin 2\pi k ft_{i} - \Delta \sum_{i=1}^{n} y_{i} \cos 2\pi k ft_{i}$$

$$b = \frac{1}{\sum_{i=1}^{n} y_{i} \cos 2\pi k ft_{i}}$$
(49a)

1

$$\sum_{i} \sin 2\pi k ft \cos 2\pi k ft$$

$$\Delta = \frac{1}{\sum_{1}^{1} \cos^2 2\pi k ft_1}$$
(49b)

It should be noted that, for equally spaced data and f=1/T, $\delta=\Delta=0$ and equations (48a) and (49a) reduce exactly to equations (20).

Let t_1 be considered a random variable satisfying a uniform distribution. Using some trigonometric identities

$$\delta = \frac{\sum_{i=1}^{n} \sin 2\pi k ft_{i} \cos 2\pi k ft_{i}}{\sum_{i=1}^{n} \sin^{2} 2\pi k ft_{i}} = \frac{\frac{1}{2} \sum_{i=1}^{n} \sin 2\pi (2kf) t_{i}}{\frac{1}{2} \sum_{i=1}^{n} \left[1 - \cos 2\pi (2kf) t_{i}\right]} = \frac{\sum_{i=1}^{n} \sin \theta_{i}}{\sum_{i=1}^{n} \left(1 - \cos \theta_{i}\right)}$$
(50)

where now $\theta_1 = 2\pi (2kf)t_1$ is regarded as a random variable. The sum $\sum_{l} \sin \theta_1$ is N times an estimate of the expected value of sin θ . Likewise, $\sum_{l} \cos \theta_l = 1$

NE($\cos\theta$). Therefore,

$$\delta = E \left(\sin \theta \right) / \left[1 - E(\cos \theta) \right]$$
(51)

If the interval contains any whole number of cycles, $E(\sin\theta) = E(\cos\theta) = 0$. If the interval is not an integral number of periods in length, $E(\sin\theta)$ and $E(\cos\theta)$ differ from zero by an amount of the order of 1/N or less. Therefore, (unless N is a small number) δ is a quantity of order 1/N. Likewise Δ is of the same magnitude.

Returning now to the expression for power, letting $\omega = 2\pi kf$ for conciseness.

$$\sum_{1}^{2} y_{1}^{2} - \sum_{1}^{2} \varepsilon_{1}^{2} = \sum_{1}^{2} y_{1}^{2} - \sum_{1}^{2} (y_{1} - a\cos \omega t_{1} - b\sin \omega t_{1})^{2}$$
$$= 2a \sum_{1}^{2} y_{1} \cos \omega t_{1} + 2b \sum_{1}^{2} y_{1} \sin \omega t_{1} - a^{2} \sum_{1}^{2} \cos^{2} \omega t_{1}$$
$$- 2ab \sum_{1}^{2} \sin \omega t_{1} \cos \omega t_{1} - b^{2} \sum_{1}^{2} \sin^{2} \omega t_{1}$$
(52)

Substituting (48a) - (50),

$$\sum_{1} y_{1}^{2} - \sum_{1} e_{1}^{2} = 2a \sum_{1} y_{1} \cos \omega t_{1} + 2b \sum_{1} y_{1} \sin \omega t_{1} - 2ab \sum_{1} \sin \omega t_{1} \cos \omega t_{1}$$

$$- a \left[\frac{\sum_{1} y_{1} \cos \omega t_{1} - \delta \sum_{1} y_{1} \sin \omega t_{1}}{\sum_{1} \cos^{2} \omega t_{1} - \delta \sum_{1} \sin \omega t_{1} \cos \omega t_{1}} \right] = \sum_{1} \cos^{2} \omega t_{1}$$

$$- b \left[\frac{\sum_{1} y_{1} \sin \omega t_{1} - \Delta \sum_{1} y_{1} \cos \omega t_{1}}{\sum_{1} \sin^{2} \omega t_{1} - \Delta \sum_{1} \sin \omega t_{1} \cos \omega t_{1}} \right] = \sum_{1} \sin^{2} \omega t_{1}$$

$$= 2a \sum_{1} y_{1} \cos \omega t_{1} + 2b \sum_{1} y_{1} \sin \omega t_{1} - 2ab \sum_{1} \sin \omega t_{1} \cos \omega t_{1}$$

$$- a^{*} \left[\frac{\sum_{1} y_{1} \cos \omega t_{1} - \delta \sum_{1} y_{1} \sin \omega t_{1} - 2ab \sum_{1} \sin \omega t_{1} \cos \omega t_{1}}{1 - \delta \Delta} \right]$$

$$- b \left[\frac{\sum_{1} y_{1} \cos \omega t_{1} - \delta \sum_{1} y_{1} \sin \omega t_{1}}{1 - \delta \Delta} \right]$$

Since both δ and Δ were found to be small quantities, terms of order $\dot{\delta}\Delta$ will be neglected.

$$\sum_{i} y_{i}^{2} - \sum_{i} y_{i}^{2} e_{i}^{2} = a \sum_{i} \cos\omega t_{i} + b \sum_{i} y_{i} \sin\omega t_{i} + a \delta \sum_{i} y_{i} \sin\omega t_{i} \\ + b\Delta \sum_{i} y_{i} \cos\omega t_{i} - 2ab \sum_{i} \sin\omega t_{i} \cos\omega t_{i} \\ = a \sum_{i} y_{i} \cos\omega t_{i} + b \sum_{i} y_{i} \sin\omega t_{i} - 2ab \sum_{i} \sin\omega t_{i} \cos\omega t_{i} \\ + a \left[\frac{\left(\sum_{i} \sin\omega t_{i} \cos\omega t_{i}\right) \left(\sum_{i} y_{i} \sin\omega t_{i}\right)}{\sum_{i} \sin^{2} \omega t_{i}} \right] \\ + b \left[\frac{\left(\sum_{i} \sin\omega t_{i} \cos\omega t_{i}\right) \left(\sum_{i} y_{i} \cos\omega t_{i}\right)}{\sum_{i} \cos^{2} \omega t_{i}} \right] \\ = a \sum_{i} y_{i} \cos\omega t_{i} + b \sum_{i} y_{i} \sin\omega t_{i}$$

$$+\sum_{1} \operatorname{sinwt}_{1} \operatorname{coswt}_{1} \left[\begin{array}{c} \sum_{1} y_{1} \operatorname{sinwt}_{1} & \sum_{1} y_{1} \operatorname{coswt}_{1} \\ \frac{1}{\sum_{1} \operatorname{sin}^{2} \operatorname{wt}_{1}} & + \operatorname{b} \frac{1}{\sum_{1} \operatorname{cos}^{2} \operatorname{wt}_{1}} & - 2\operatorname{ab} \end{array} \right]$$
(53)

After further manipulation, neglecting terms of order $\delta\,\Delta$, the entire last term vanishes, leaving

$$\sum_{1} y_{1}^{2} - \sum_{1} \varepsilon_{1}^{2} = a \sum_{1} y_{1} \cos \omega t_{1} + b \sum_{1} y_{1} \sin \omega t_{1}$$
(54)

For equally spaced data, with f=1/T, (54) becomes identical with (21).

COMPUTER PROGRAM

General Information

The logical structure of the computer program implementing the mathematical method described in the previous sections has three main parts

- 1. The power spectrum of the data is calculated.
- Improved estimates of the coefficients and frequencies of dominant spectral terms are obtained.
- 3. The dominant terms are subtracted from the data (this is a form of prewhitening) and the spectrum of the residuals is calculated.

The user may control the calculations, and specify the form of the output, by means of parameters on input cards. Appendix 1 is a diagram showing the overall logic of the program.

The second step in the above list is done by searching the spectrum ("coarse spectrum") for relative maxima whose powers exceed a certain level and using the coefficients and frequencies of the maxima as initial estimates in a simultaneous least squares solution for the dominant terms. Specifically, the function is represented by

$$\overset{\Lambda}{y}(t) = \sum_{k=1}^{M} \left(a_k \sin 2\pi f_k t + b_k \cos 2\pi f_k t \right)$$
(55)

In a least squares fit to the data, in which a_k , b_k , and f_k , $k=1, \ldots, M$ are allowed to vary. Here $M \le 10$ is the number of relative maxima found in searching the coarse spectrum. Peaks are chosen in order of size, the largest first. It should be noted that in equation (55) a_k is the coefficient of the sine term and b_k that of the cosine term. This convention, which is opposite to that followed up to now in this report, is the one generally used in the program. Up to ten terms (the number is an input parameter) may be solved for simultaneously. The threshold which must be exceeded for a peak to be selected

-24-

("removal tolerance") may either be specified as input or computed by the program.

Parts 2 and 3 in the above list are done in a cyclic fashion until no more peaks can be found which exceed the removal tolerance. Then the removal tolerance is cut in half (once) and the process is continued until no more relative maxima can be found. The process is also terminated in the event that the number of terms removed exceeds 20% of the number of terms in the coarse spectrum.

This search-and-removal process is actually a method of prewhitening, a technique which is recommended for increasing the accuracy of power spectral density analysis. [9, 14] Prewhitening means filtering the data prior to analysis to remove spectral peaks and obtain as smooth a power spectrum as possible. One way to prewhiten data is to filter the signal as it is measured. To do this, of course, one must have some foreknowledge of the frequency content. In this program, the "preprocessing" is done in the course of the computations, and no advance knowledge of the properties of the signal is required. The terms removed, plus the spectrum of the prewhitened signal ("final spectrum"), give the most accurate representation of the frequency content of the signal

The subroutine which improves the estimates of dominant spectral terms uses an iterative linearized least squares formulation, employing Newton's method, to find the coefficients and frequencies of up to ten terms simultaneously There are several conditions besides convergence which will cause a return to the main program.

 If the frequency corrections to one or more terms are more than one-half of the frequency spacing of the coarse spectrum, those terms are tagged as false peaks. Upon return to the main program, the spectrum is searched for the ten largest relative maxima - not including those which are tagged - whose powers exceed the removal tolerance.

-25-

- 2. After the first two iterations the least squares process is checked for convergence (by observing the change in variance). If there is divergence on the nth iteration, the values for the (n-1) st iteration are chosen for the coefficients and frequencies.
- 3. If the convergence test is not satisfied after 50 iterations, an error message is printed and execution is terminated. The convergence criterion is that the relative change in Δy must be less than 10^{-8} (and Δy must decrease), where Δy is the sum of squares of residuals. The error message includes a list of the old and new values of the fit parameters and Δy .

Reference 6 shows how significance testing can be applied to the spectral powers computed by formula (54). On the program's graphical output, a dotted line appears approximately at a 95% significance level.

Let T(1), ..., T(N) be the times (for simplicity the independent variable will henceforth be called time) associated with the N values of the observed quantity. The program calculates T(N) - T(1) as the duration of the signal. However, in order that the statement following equation (49b) be true, that the general formulas for the coefficients reduce to equations (20) in the case of equally spaced data and f=1/T, [N/(N-1)]x duration must be used for T. This is equivalent to regarding the observations as forming a histogram.

$$\begin{vmatrix} -\Delta t \rightarrow | -\Delta t \rightarrow | -\Delta t \rightarrow | \cdots \\ | -\Delta t \rightarrow | -\Delta t \rightarrow | \cdots \\ \end{vmatrix}$$

For equally spaced times the time axis is divided into N intervals, each of width Δt (the sampling increment), with the sampling points at the midpoint of each interval. Thus, T must include portions before T(1) and after T(N). The basic frequency f is given in the program by (regardless of the time spacing)

$$f = \frac{1}{S\left(\frac{N}{N-1}\right) [T(N) - T(1)]}$$
(56)

where S is a factor which allows f to be different from 1/T. The spectrum is evaluated at kf, k=1, ..., k_{max} , where

$$k_{\max} = RSN/2$$
(57)

Here R is a factor which determines the maximum frequency at which the spectrum is evaluated, the <u>range factor</u>. When R=1 this maximum is the Nyquist frequency. Because of its role in determining frequency spacing, S is called the <u>spacing factor</u>. When S=1 the frequencies are harmonics of the fundamental period T.

A few comments on the use of unequally spaced data are appropriate. When the times are <u>randomly</u> spaced there is no aliasing. Therefore, components above the Nyquist frequency can be detected. When the times are in some regular pattern (but not equally spaced) there may be some misleading effects. For example, if the times are equally spaced except for regularly occurring gaps of a fixed size, there is an effect something like modulation by a square wave.

Often in using the program, changing one or more of the quantities S, R, the limit on the number of terms in the simultaneous solution (MTERMS), and the removal tolerance and rerunning will yield better results. Using a very large value of S may reveal several peaks so close together that they prevent proper operation of the removal feature with a spacing factor near unity. Setting MTERMS = 1 may help in this instance. A smaller removal tolerance may lead to the discovery of further underlying structure in the signal. It is recommended that the user make several runs with known data to gain familiarity with the operation of the program.

-27-

An option is available to smooth the spectrum estimates with low-pass convolution filters of up to 13 points. This feature is not currently activated, but the FORTRAN statements are in the program as comments. To activate this feature the "C" in column 1 should be removed.

The following sections contain a description of the system of programs and its use. Appendix 1 contains a flowchart showing the overall logic of the program. Appendix 2 is a sample run. A listing of the program which generated the (unequally spaced) data is given, along with the output. All of the output options are exercised. Values of S and R different from 1.0 are used. Removal is done, with the program calculating the removal tolerance.

For more information on the program the user is referred to Volume 2 of this report, the Maintenance Manual.

Functions of Programs in System

CUCPSA (main program)

Reads program control cards Reads data cards (or calls tape read subroutine) Calculates mean and a. c. power of total signal Calculates coefficients, power, and percent power of spectral terms. Prints power spectrum Calls two subroutines (GRAPH and RPLOT) to plot power spectrum Searches "coarse" spectrum for peaks.

Calls subroutine to solve for coefficients and frequencies of dominant terms. Removes dominant spectral terms.

GRAPH

Lists power and plots (on printer) percent power vs. frequency, 53-58⁻ points/page. A dotted line is drawn approximately at a 95% significance level.

RPLOT

Plots the spectrum on one page, with a significance indicator.

READ

Reads data from tape.

CAROL3

Performs simultaneous least squares fit of up to ten sinusoidal functions, solving for coefficients and frequencies.

BIORTH

Matrix inversion routine called by CAROL3.

Input Cards

There are two types of input cards program control cards and data cards. The latter may be replaced by a tape.

A description of the input cards follows. Everything pertaining to the spectrum smoothing feature is given in brackets. It should be noted that if the smoothing feature is not activated, card 4 is not used and there are four program control cards per case.

Card 1. Format Card



Observation card format. This is a FORTRAN FORMAT statement without the statement label and the word "FORMAT." It includes the beginning and ending parentheses. It must specify 3 double precision fields. This card should be left blank if the data are to be read from tape.

Card 2.	Observation	Definition	Card
---------	-------------	------------	------

1	56		10	11	15	16	25	26	35	
(15)		(15)	/	(15)	/	(F10.5)	<u>(</u> [F10.5)	
Field number for obser- vation (IY)	-	Field numbe for tir (IT)	er ne	Field numb for da (ID)	er Ly	Time un per com tational (TPC)	lits pu- unit	Cor unit (C	nputational ts per day CPD)	

NOTE "Field" means fields on observation cards as given on the Format Card.

t = T/TPC + (D-D₀) CPD
where t - time used in program
T- time on observation card
TPC-time scale factor
D- day on observation card
D₀- day on 1st observation card
CPD-day scale factor

For example, if the program is to compute in minutes and the times on the observation cards are in seconds and whole days, then TPC=60 and CPD=1440

If the data are on tape rather than cards there is no "day" variable. ID and CPD are ignored by the program. Subroutine READ assumes that the tape contains a file consisting of N (the number of points) records, each containing 7 (or more) single-precision words. The tape should be binary (unformatted). IT tells which of the 7 words is time, IY identifies the observed quantity. Time is calculated by t=T/TPC.

Card 3. J	Program	Option	Card
-----------	---------	--------	------

1 18	19	23 24	33 34	$43 \ 44$	45	46	55	56	
(3A6)	(I5)	(F1	0.5)(F1().5) (I1)	, (I1)	(F10	.5) (]	(1)	
Data set		Spacı	ng	Plot		Remo	oval		
name		factor	r	flag		tolera	ance		
(NAME)	i	(S)		(JPO)	(TO	L)		
	(S) Number of obser- vations		Ra fac	l nge tor	Remo flag	oval		 Print flag	
	vatı (N	ons)	(1	R)	(JF)			(JS)	

- NAME identifies data
 - N is number of observations (limited to 2000)
 - S is spacing factor in equation (56)
 - R is range factor in equation (57)
 - JPO = 0 program plots rough [and smooth] power spectrum
 - = 1 plots rough power spectrum only
 - = 2 [plots smooth power spectrum only]
 - = 3 no plots
 - JF = 0 program removes terms whose percent powers exceed TOL
 - = 1 no removal
 - TOL is calculated by program if not inputted
 - JS = 0 program prints power spectrum
 - = 1 program does not print spectrum
- NOTES The default value of TOL is the same as the 95% significance level (shown on the graphs by a dotted line). However, in this context it should not be regarded as a 95% significance level. Combinations of N, S, and R which would produce more than 1000 spectral terms are not allowed. In such a case, the program will compute the first 1000 terms.

TOL is a fraction which is used to compute a power level, that fraction of the total power of the signal. After removal of terms that level is not re-evaluated as a fraction of the remaining power, except for being halved once as described previously.

[Card 4. Spectrum Smoothing Filter Card



This card cannot be used unless the smoothing statements are activated and $S \neq 1$. If S=1 the values 1/4, 1/2, 1/4 ("Hanning") will be used for the filter.]

Card 5. Simultaneous Removal Card



These cards are followed by N observation cards, in the format specified on the Format Card, unless the data are on tape. Any number of sets of data may be analyzed sequentially by stacking the input cards to form one large data deck. After the last case, two more cards are needed to terminate execution The first may be blank. The second should have -999 punched in columns 2-5.

Output

The first page of output for each case gives a list of input quantities. The first group on this page lists, in order, NAME, N, S, R, JPO, JF, TOL, JS, and MTERMS, one per line. [If the spectrum smoothing feature is activated, the next group is a list of the filter factors.] The final group consists of FMT, IY, IT, ID, TPC, CPD, and DUR=T(N)-T(1), also one per line.

On the next page the title

INITIAL DATA SPECTRUM FOR SET {NAME} (TOTAL POWER = {YYBAR2})

is printed. Here $\{X\}$ means "the value of the quantity X." YYBAR2 is the total power, the sum of squares of the observations (after subtracting the mean) divided by N/2. This normalization for power is somewhat arbitrary, it amounts to a choice of the unit of power. The same normalization is used for the terms in the power spectrum, equation (54) is divided by N/2.

If the print option is used (JS=0), this is followed by a tabulation of the spectrum, 50-52 lines per page. The quantity C given at the top of each page is the mean of the observations. The first of the seven columns lists the frequency index k, which goes from 1 to k_{max} (equation (57)). The second column gives the period, the reciprocal of the frequency of the kth term. The third column is ω , $2\pi x$ frequency. The fourth and fifth columns list a_k , the coefficient of the sine term (called b in equation (47)), and b_k , the coefficient of the cosine term (equation (46)). Once again attention is called to the reversal of the usual naming convention for the coefficients. The next column lists the regression power of each term, equation (52) divided by N/2. The last column gives percent (really fractional) power, the preceding quantity divided by the total power YYBAR2.

Depending on whether the spectrum smoothing feature is activated, and on what plot option is specified, plots (on the printer) may be made of the raw spectrum, the smoothed spectrum, both, or neither. If plots of both the raw and smoothed spectrum are requested, the former appear first. Plots are made in pairs The first is on a single page, power vs. k. The second may occupy several pages, with the first 53 points on the first page and 58 on each succeeding page. The expanded plot shows percent power vs. k, with the k axis running down the page. To the left of the plot are two columns of figures giving the regression power and frequency corresponding to each value of k. Automatic scale selection is done for each plot, and on both plots a dotted line shows an approximate 95% significance level.

After plotting the spectrum, the program may be finished with that case. However, if the search-and-removal option is specified the title

REMOVAL TABLE FOR SET $\{NAME\}$,

column headings, and the first line of the removal table are printed This first line consists only of YYBAR2, the total power, under the heading "REMAINING TOTAL POWER."

The remainder of the removal table lists the results, if any, of the simultaneous least squares solution for dominant terms. The first five columns, in order, list the period, frequency, angular frequency (2 x frequency), and the sine and cosine coefficients of the terms in the solution. The column headed "SPECTRUM POWER" gives the regression power of each term. On the last line of a group of terms solved for simultaneously, the column headed "REMAINING TOTAL POWER" gives the power of the residuals after subtracting the solution terms from the original data. On <u>every</u> line of this group, the column headed "C" lists the mean of these residuals. There may be several such groups of terms. After each subtraction the residuals are treated as data, the spectrum is calculated and searched for peaks as before. When no more valid peaks exceeding the removal tolerance can be found, TOL is cut in half and the message "SEARCH WILL NOW BE MADE WITH SMALLER TOL" is printed Then the search-and-removal procedure continues until no more peaks are found, and the values found are printed in the removal table.

If at least one term is found and removed from the data, the spectrum of the residuals is printed and/or plotted according to the output options specified The form of this output is just the same as for the spectrum of the original 'data, except that the heading of the tabulation specifies that it is the "FINAL DATA SPECTRUM."

-34-

Executing the Program

The program is on the FASTRAND file LSQFIT*MLTSAN. The name of the absolute element is LSQFIT*MLTSAN. APSD. If input is from tape rather than cards, the tape is considered to be on logical unit 3. If it is necessary to reMAP the program for some reason, the MAP source element LSQFIT*MLTSAN. PSAMAP may be used.

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APPENDIX 1

Overall Logic Flowchart



APPENDIX 2

Sample Run

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        RUN
                   MXL, D01, SPANAL, 10, 50
ø
                   GENERATE RANDOMLY SPACED DATA FOR SPECTRUM PROGPAM
        HDG
ធ
<u>a</u>
2
        FOR, IS RANDSD , RANDSD
       PROGRAM TO GENERATE KNOWN DATA FOP SPECTRAL ANALYSIS PROGRAM
 **
** ی
С
 **
                                   M
       Y = YBAR + SLOPE*T + \underline{SUM}(A(I)*SIN(2*PI*F(I)*T)
Ĉ
  **
C **
                                  J≃1
<u>C</u> **
C **
                                       + B(J)*COS(2*PI*F(I)*T))
C **
       DIMENSION A(19), B(10), F(10), X(10), RANDNO(1000)
C_**
       TIMES ARE RANDOMLY SPACED OVER (TO, TO + 1.0).
C_**
C ** THE PSEUDO - NYQUIST FREQUENCY IS (N - 1) / 2 CPS.
C **
 1000 FORMAT (1H1+21X+12/72H Y = YBAR + SLOPF*T + SUM(A(I)*STN(2*PI*F(I)
*T) + B(I)*COS(2*PI*F(I)*T))/22X3HI=1//10X7HYBAR = +F6.3,10X8HSLOF
      .E = ,F6.3//8X1HF,10X1HA,10X1HB/ (3(5XF6.3)/))
 1001 FORMAT (5XF5,2,5XE13.8)
 1002 FORMAT (3E20.8)
1004 FORMAT (F10.0,15)
 1005 FORMAT (15)
1006 FORMAT (8X1HT,13X1HY)
       READ (5,1002) YBAR, SLOPE
READ (5,1005) M
                                                  NUMBER OF FREQUENCIES
                                            R
       READ (5.1002) (F(I),A(I),B(I),I=1,M)
        WRITE (6,1000) M, YBAR, SLOPE, (F(1), A(1), B(1), I=1, M)
        WRITE (6,1006)
C **
       TO IS THE INITIAL TIME.
С
  **
       N IS THE NUMBER OF POINTS.
C
  **
       READ (5+1004) TO.N
       RANDNO(1) = 6.2831853076
       \frac{NM2}{CALL} = N - 2
CALL RANDU (RANDNO, NM2)
                                                  GENERATE ARRAY OF RANDOM NUMBERS.
                                            Ģ
        T =_T0
       NM = 0
        GO TO 30
    10 CONTINUE
       NM = NM + 1
T = TO + RANDNO(NM)
        GO TO 30
    20 CONTINUE
        \frac{NM = NM + 1}{I = T0 + 1.0}
    30 CONTINUE
```

		BIAS=	YBAR+SLOP	₽Ë * Ĩ		
		Y=RIA	5			
		DO 50	J≡1+M			
		YL.1)-	6.283185	3076*F(.1)*T		
	= ^		7 11 - CTN1/1	7070#1 (07#1 V(.1) \±0(.1)*	005(27.0)	
	20	DOTNE		X (0 /) + B (0) +	031/10//	
		PRINI	LUULPIP	¥		
		PUNCH	1002,T,	Ŷ		
		IF (N	M - NM2)	10,20,40		
	40	CONTI	NUÉ			
		STOP				
		FND				
Ð		YOT				
_10I		7(4)	5	_	0.	
	E		-	¢	- •	
	5		76		3	Ē.,
			30	•	.	2
			158	● 3		0 •
			220	•	<u>э</u> ,	U.
			250	•	υ.	_D+
			281	ę	7-,	-S •
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SET NAME	li H	RANDOM TIME	310568
SPACING	-	,500000000	
RANGE	÷	1.30000000	
PLOT FLAG	Ξ	0	
REMOVAL	4	₽	
TOLERANCE	Ξ	,02000	
PRINT FLAG	-=	ፁ	
MTERMS	Ξ	10	
FORMAT	=	(3020.0)	
Y FIELD	÷	-2	
T FIELD	Ξ	1	
DAY FIELD	1	3	
TIME SCALE	-	1,0000000) TIME UNITS PER COMPUTATIONAL UNIT
- DAY SCALE	=	+000000000	O COMPUTATIONAL UNITS PER DAY
DURATION	4	1.0000	COMPUTATIONAL UNITS

INITIAL DATA SPECTRUM FOR SET RANDON TIME 310568 (TOTAL POWER = +24051+03)

				REGRESSION	PERCENT		
PERIOD	OMEGA	A	च	PÓWER	POWER-	с =	•4444 9+61 -
.5010	12,541238	.38427+00	. 19315-01	149625	.000622		
.2505	25,082476	. 13933+00	78377+00	.635199	.002641		
.1670	37.623714	38234+00	19485+00	.181005	.000753		
.1253	50,164951	22804+00	10337+01	1,15615	.004807		
-1002	62,706189-	14735+00	•54906∓00 ⁻	323728	.001346		
.0835	75.247427	. 71437+00	49764-01	.531742	.002211		
.0710	87.788665	86443+00	. 65185+00-	1.16105	.004827		
.0626	100.329903	. 51587+00	•o7216+00	.702510	.002921		
.0557	112.871141	.26782+00	34871+00	185700	.000772		
.050L	125.412379	•54781+00	26621+00	•364523	.001516		
-0455	137-953617	• 60625+00 ·	+28622+00	457869	-001904		
•0413	150.494854	24629+00	•57047+00	•380804	.001584		
.0385	163.036092	. 12736+00	•69258+00	. 504103	.002096		
.0358	175.577330	.29731+00	.14428+00	109238	.000454		
.0334	188,118568	-,49089+00	.13375+00	.248718	.001034		
.0313	200.659806	.35153+00	25307+00	185071	.000769		
0295	213-201044	*80349+00	.57537+00	984454	.004093	-	
.0278	225,742282	. 12964+01	•56823+01	33,9198	.141031		
.0264	238,283520	-,49421+00	86820+00	.960870	.003995		
.0251	250,824757	14717+00	-,39713+00	.175213	.000728		
.0239	263,365995	. 79963+00	~,10045+01	1.69353	.007041		
.0228	275.907233	10662+01	17452+00	1,20290	.005001		
.0218	-288,448471	~ 31171+00	31962+00	-196311	⊾000816		
.0209	300,989709	. 21072+00	-,26428+00	•116106	.000483		
.0200	313.530947	14969+01	- •17975+00	2.31140	.009610		
.0195	326.072185	54436+00	10201+01	1,34717	.005601		
0180	338,613423	+ ,41553+00	-,78994-03	171249	.000712		
.0179	351.154600	. 68447+00	. 15912+01	2,95500	.012286		
+0173	363,695898-	. 37035+00	-+25410+00	202355	.000841		
.0167	376,237136	.38030+00	.617 40 +00	.529262	.002201		
.0162	388,778374	11352+01	. 15919+00	1,23824	.005148		
•0157	401.319612	25589+00	95412-01	.758469-01	.000315		
.0152	413.860850	91126+00	. 75881+00	1.38307	.005751		
.0147	426.402088	•73870+00	.77695+00	1.16997	.004864		
.0143	438.943326	-,47313+00	10268+01	1.27766	.005312		
.0139	451.484563	.10143+01	82577-01	1.01926	.004238		
.0135	464.025801	·16530+01	o4477+00	3.04091	.012643		
.0132	470.567039	.52508-01	.50513+00	.257990	.001073		
.0123	489.108277	29884+00	47652+00	308960	.001285		
.0125	501.649515	83427+00	10728+01	1.85894	.007729		
.0122	514,190733	.76261+00	.20013-01	.600436	.002496		
.0119	526.731991	12241+00	•65674+00	432352	.001798		
.0117	539.273229	. 35417+∪0	-,28731+00	.205242	.000353		
.0114	551.814466	12447+01	·22179+00	1.72498	.007172		
.0111	564 . 3557v4	-,40730+00	.38661+00	.299123	.001244		
.0109	576.896942	-, 38045+00	65127+00	•759659	.003158		
.0107	589,438180	.37710+00	56095+00	1.00642	.004184		
0104	601 979418	85103+00	.877 04+00	1.49809	.006229		
.0102	614.520656	06880+00	. 56267+00	∙5715 ა 2	.002376		
.0100	627,061894	.22249+00	78249+00	•645434	.002684		
	PER100 .5010 .2505 .1670 .1253 .1002 .0835 .0710 .0626 .0557 .0501 .0455 .0416 .0385 .0313 .0295 .0278 .0264 .0251 .0239 .0228 .0228 .0228 .0239 .0228 .0218 .0209 .0229 .0200 .0195 .0180 .0179 .0162 .0157 .0152 .0147 .0162 .0152 .0128 .0109	PER100OMEGA.501012.541238.250525.082476.167037.623714.125350.164951.100262.706189083575.247427.071087.788665.0626100.329903.0557112.871141.0501125.412379.0455137.953617.0413150.494854.0385163.036092.0358175.577330.0334188.118568.0313200.659806.0295213.201044.0278225.742282.0264238.283520.0251250.624757.0239263.365995.0228275.907233.0218-288.448471.0209300.989709.0200313.530947.0195326.072185.0186338.613423.0179351.154600.0173.363.6958980167376.237136.0162388.778374.0157401.319612.0152413.860850.0147426.402088.0143438.943326.0154438.943326.0155.0122.0120489.108277.0125501.649515.0122514.190733.0119526.731991.0117539.27329.011451.814466.0111564.3557.04.0109576.896942.0107589.438180.0104601.979418.0102614.520656 <td>PERIODOMEGAA.501012.541238.38427+00.250525.082476.13933+00.167037.623714.38234+00.125350.164951.22804+00.40262.70618914735+00.083575.247427.71437+00.071087.788665.8644,5400.0626100.329903.51587+00.0557112.871141.26782+00.0557112.871141.26782+00.0554137.953617.60622+00.0413150.494854.24629+00.0385163.036092.12736+00.0358175.577330.29731+00.0358175.577330.29731+00.0358163.036092.12736+00.0355175.577330.29731+00.0358163.03659806.35153+00.0295213.201044.80349+00.0276225.742282.12964+01.0264238.283520.49421+00.0251250.624757.14717+00.0229263.365995.79963+00.0228275.907233-10662+01.0193326.072185.54436+00.0193326.072185.54436+00.0193326.072185.54436+00.0193326.072185.37035+00.0162388.778374.11352+01.0157401.319612.25589+00.0157401.319612.25589+00.0152413.860850.91126+00.0143438.943326.47313+00.0154<</td> <td>PER100 OMEGA A -t .5010 12.541238 .38427+00 .19315-01 .78377+00 .2505 25.082476 .13933400 .78377+00 .78377+00 .1253 50.164951 .22804+00 .1948.500 .10337+01 .1253 50.164951 .22804+00 .1948.500 .99764-01 .0626 100.329903 .5187+00 7216+00 .5185+00 .0557 112.671141 .26762+00 3487100 .26621+00 .0551 125.412379 .54781+00 26621+00 .26621+00 .0355 175.577330 .29731+00 .14428400 .5704740 .0334 188.118568 .49089+00 .13375+00 .25307+00 .0264 23.283520 .49421+00 .56823+01 .56823+01 .0264 23.263595 .7963+00 .10428+00 .5737+00 .0264 23.263595 .7963+00 .10462+01 .17452+00 .0264 23.265954 .79765+00 .1042+10 .3682+010</td> <td>PERIOD OMEGA A T POWER .5010 12,541238 .3842740 .19315-01 .149625 .2505 25,082476 .13933400 .7837740 .635199 .1670 37,623714 .38234400 19485400 .181005 .1253 50,164951 .22604400 1037401 1.15615 .0710 87,788665 .86443400 .5515400 1.16105 .0266 100.329903 .51587400 .7216400 .702510 .0557 112.671141 .26782400 .5747400 .380844 .0455 137,953647 .66622400 .57047400 .38084 .0358 175,577330 .29731400 .1428400 .10375406 .0334 188,118568 .49089400 .3375400 .984454 .0278 225,742282 .12964401 .66823401 .635131 .0239 263,35595 .79963404 .163533 .998451 .0273 226,742282 .12964401 .56823401 .63511</td> <td>PERIOD OMEGA A T POWER POWER -5010 12.541238 .5842760 .19315-01 .149625 .000622 .2505 25.082476 .13933400 .78377+00 .635199 .002641 .1670 37.623714 .38234400 1948540 .181035 .004807 .1022 627706189- 14735400 .54906400- .323728 .004807 .00251 12.671427 .71437400 .49764-01 .531742 .002211 .00557 112.671414 .26782400 .5284200 .36871400 .105700 .004921 .0557 112.67141 .26782400 .57047400 .30864 .001584 .0455 133.7953617 .66622400 .504103 .00296 .0358 .001584 .0358 153.7063 .27731400 .14428400 .109235 .001344 .0313 200.659806 .35153400 .25307400 .948718 .00134 .0313 200.659806 .35153400 .2504103</td> <td>PERIOD OMEGA A T POWER POWER C = .5010 12.541238 .5842740 .19315-01 .149625 .000662 .2503 25.082476 .13933400 .7837740 .635199 .002641 .1253 50.164951 .22804400 .10337401 1.15615 .004807 .0835 75.247427 .71437400 .44764-01 .531742 .002211 .0716 87.78865647 .66444400 .65185400 .161015 .004807 .08557 122.671141 .2672200 .24621400 .726501 .002291 .0557 122.671141 .2672400 .34871400 .85185700 .002291 .0557 123.67144 .857047400 .85623 .001516 .0445 153.0492 .12734100 .42622400 .57047400 .80523 .001034 .0315 153.05092 .12735400 .6925840 .001034 .001334 .0334 186.118566 .49089400 .15375400 .260870</td>	PERIODOMEGAA.501012.541238.38427+00.250525.082476.13933+00.167037.623714.38234+00.125350.164951.22804+00.40262.70618914735+00.083575.247427.71437+00.071087.788665.8644,5400.0626100.329903.51587+00.0557112.871141.26782+00.0557112.871141.26782+00.0554137.953617.60622+00.0413150.494854.24629+00.0385163.036092.12736+00.0358175.577330.29731+00.0358175.577330.29731+00.0358163.036092.12736+00.0355175.577330.29731+00.0358163.03659806.35153+00.0295213.201044.80349+00.0276225.742282.12964+01.0264238.283520.49421+00.0251250.624757.14717+00.0229263.365995.79963+00.0228275.907233-10662+01.0193326.072185.54436+00.0193326.072185.54436+00.0193326.072185.54436+00.0193326.072185.37035+00.0162388.778374.11352+01.0157401.319612.25589+00.0157401.319612.25589+00.0152413.860850.91126+00.0143438.943326.47313+00.0154<	PER100 OMEGA A -t .5010 12.541238 .38427+00 .19315-01 .78377+00 .2505 25.082476 .13933400 .78377+00 .78377+00 .1253 50.164951 .22804+00 .1948.500 .10337+01 .1253 50.164951 .22804+00 .1948.500 .99764-01 .0626 100.329903 .5187+00 7216+00 .5185+00 .0557 112.671141 .26762+00 3487100 .26621+00 .0551 125.412379 .54781+00 26621+00 .26621+00 .0355 175.577330 .29731+00 .14428400 .5704740 .0334 188.118568 .49089+00 .13375+00 .25307+00 .0264 23.283520 .49421+00 .56823+01 .56823+01 .0264 23.263595 .7963+00 .10428+00 .5737+00 .0264 23.263595 .7963+00 .10462+01 .17452+00 .0264 23.265954 .79765+00 .1042+10 .3682+010	PERIOD OMEGA A T POWER .5010 12,541238 .3842740 .19315-01 .149625 .2505 25,082476 .13933400 .7837740 .635199 .1670 37,623714 .38234400 19485400 .181005 .1253 50,164951 .22604400 1037401 1.15615 .0710 87,788665 .86443400 .5515400 1.16105 .0266 100.329903 .51587400 .7216400 .702510 .0557 112.671141 .26782400 .5747400 .380844 .0455 137,953647 .66622400 .57047400 .38084 .0358 175,577330 .29731400 .1428400 .10375406 .0334 188,118568 .49089400 .3375400 .984454 .0278 225,742282 .12964401 .66823401 .635131 .0239 263,35595 .79963404 .163533 .998451 .0273 226,742282 .12964401 .56823401 .63511	PERIOD OMEGA A T POWER POWER -5010 12.541238 .5842760 .19315-01 .149625 .000622 .2505 25.082476 .13933400 .78377+00 .635199 .002641 .1670 37.623714 .38234400 1948540 .181035 .004807 .1022 627706189- 14735400 .54906400- .323728 .004807 .00251 12.671427 .71437400 .49764-01 .531742 .002211 .00557 112.671414 .26782400 .5284200 .36871400 .105700 .004921 .0557 112.67141 .26782400 .57047400 .30864 .001584 .0455 133.7953617 .66622400 .504103 .00296 .0358 .001584 .0358 153.7063 .27731400 .14428400 .109235 .001344 .0313 200.659806 .35153400 .25307400 .948718 .00134 .0313 200.659806 .35153400 .2504103	PERIOD OMEGA A T POWER POWER C = .5010 12.541238 .5842740 .19315-01 .149625 .000662 .2503 25.082476 .13933400 .7837740 .635199 .002641 .1253 50.164951 .22804400 .10337401 1.15615 .004807 .0835 75.247427 .71437400 .44764-01 .531742 .002211 .0716 87.78865647 .66444400 .65185400 .161015 .004807 .08557 122.671141 .2672200 .24621400 .726501 .002291 .0557 122.671141 .2672400 .34871400 .85185700 .002291 .0557 123.67144 .857047400 .85623 .001516 .0445 153.0492 .12734100 .42622400 .57047400 .80523 .001034 .0315 153.05092 .12735400 .6925840 .001034 .001334 .0334 186.118566 .49089400 .15375400 .260870

					REGRESSION	PERCENT		
к	PERIOU	OMEGA	А	В	POWER	POWER	с =	.44449+01
51	.0093	639.603132	- ,94956 - 01	10066+01	1.02168	.004248		
52	•009v	652.144309	•26399 + 00	. 93334∔00	. 941328	.003914		
53	.0095	664 . 6856u7	56910+00	. 06005+00	•796572	.003312		
54	•0093	677,226845	11276+01	11309+01	2.53085	.010523		
55	.0391	689,708033	10431+01	•54313+00	1.47515	.006133		
56	.0089	702,309321	95657+00	27778+00	•9 7 7725	.004065		
57	•0088	714.650559	. 42312+00	-*62821+00	•575362	.002392		
58	•0080	727.391797	.45594+00	. 54623+00	•504476	.002097		
59	.0085	739.933035	85177+00	•25604+00	•784240	.003261		
ъU	•0084	752.474272	-,98026+00	•45258+00	1.11401	.004632		
61	.0082	765.015510	. 47677+00	. 18912+00	267772	.001113		
62	.0081	777.556748	.26826+00	42422+00	•251261	. 001045		
63	•0060	790.097986	.10300+01	•b6934+00	1,48658	.006181		
64	.0078	802.639224	.51498+00	. 89129 - 01	266735	.001109		
65	.0077	815.180462	-,22314+00	21640+00	•952897 - 01	.000396		
66	• 00 70	827-721700	57385+00	•92 601+ 00	1.21919	+005069		
67	•Û075	840,262938	. 56949 + 00	•45505+00	. 542610	.002256		
68	•0074	852.804175	. 43582+00	74372+00	•744540	•003096		
69	.0073	865.3454 <u>1</u> 3	. 15415+00	. 40925-01	.252330-01	.000105		
70	.0072	877.886651	83827+00	21704+00	•767279	.003190		
71	، 0071	890.427889	33299+00	19359+00	148060	.000616		
72	. 0070	902.969127	.11274+01	28067+00	1-41627	005889		
73	.0069	915.510365	. 48744+00	•38295+00	.387168	.001610		
74	.0068	928.051603	15094+01	.55817-01	2,31363	.009620		
75	.0067	940.592841	-,46897+00	.31220+00	.309903	.001289		
76	•0060	953.134078	. 97677+00	.14951+01	3,08130	.012811		
77	.0065	965.675316	-,66324+00	•o6178+00	.883476	.003673		
78	.0064	978,216554	- 43452 - 01-	.10478+01	1.11218	.004624		
79	_ . 0063	990.757792	-,44950+00	•92348+01	80.8693	.336237		
80	.0063	1003.299030	-,52987+00	16084+01	3.00653	.012500		
81	.0062	1015,840268	14327+00	62794+00	.422924	.001758		
82	.0061	1028,381506	. 85875+00	10720+01	1.95762	.008139		
83	•0060	1040.922744	-,77599+00	• 79756+00	1.20470	.005009		
84	•0060	1053,463981	-,37793+00	.11232+00	.150910	,000627		
85	.0059	1066.005219	-,70092+00	13981+00	•518351	.002155		
86	.0058	1078,546457	.50951+00	- •43658+00	.439145	.001826		
87	∙0029	1091.087695	.57415-02	.72225+00	•555543	.002310		
88	.0057	1103.628933	39169+00	11261+00	.173927	.000723		
89	.0050	1116.170171	.89217+00	14781+00	.840841	.003496		
.90	-•0056	-1.128.711409	+63873+00-		401949	0016-7-1		
91	.0055	1141.252647	53876+00	35865+00	.402116	.001672		
92	-•0054	1153.793884	•43258+00		4.23148	+01/594		
93	.0054	1166.335122	-,35796+00	97009+00	1.09472	.004552		
94	-•0053	1178.876360	•71831-01	21185-02	•5240 <u>18</u> -02-	.000022		
95	·0053	1191.417598	14378-01	.35002+00	•119684	.000498		
96	.0052	1203,958836		=+98856+00	-11-3680-	004810		
97	.0052	1216,500074	.82687+00	45741+00	•865888 •865888	+UU35008		
98	.0051-	1229,041312		-+17566+00	+809/04	+003367		
99	.0051	1241.582550	→, 34946=01	···2/344+00	•//b204=01 • #7E%0	1000323		
100	+0050-	1254,123/87		-+10364+01	1.37.539-	.003707		
101	.0050	1266,665025	88731+00	26013+00	•910934	+003787		
102	-+0049	1279,206263	-+34879+00	+/4295+00	+663644-	+002-784		

					REGRESSION	PERCENT.		
к	PERIOU	OMEGA	Α	Ŗ	POWER	POWER	C ≠	.44449+01
107	0000	1001 707501		08041.00	767720	003192		
102	•0049	1291+14/301		+04041400 130cc+01	1 99534	007830		
104	.0048	304.200739	24012+00	+10400+01	x.00000	007039		
105	.0048	1316.829977	=.29756+00	92211+00	+946666	+003937 -004/135		
106	.0047	1329-571215	-59488100-		1040100	1001400		
107	.0047	1341.912453	.73795-01	.75390+00	•563381	.002342		
108	•0040	1354,453690	.33153+00	74528+00		.002000		
109	.0046	1366.994928	14302+00	13249+01	1.77845	.007394		
110	.0040	1379.536106	.51714+00	.31820+01	9.88808	.041112		
111	.0045	1392.077404	40759+00	19174+01	3.97349	.016521		
112	0045	1404-618642	41-319+00	-28124-01	•16/110	•000090°		
113	.0044	1417.159880	.59903+00	./3/06+00	•924658	.003043		
114	•0044	1429.701118	.27838-01	12339+01	1.57934	.005557		
115	•0044	1442,242356	.31064+00	.40138+00	.261149	.001086		
116	•0043	1454.783593	83642+00	55767+00	•982655	.004086		
117	.0043	1467,324831	17758+00	61605+00	.426968	.001/75		
<u>1-1-8</u>	-0042	1479-866069	-85 080+0 0		1-37112	-005/01		
119	+0042	1492,407307	71873+00	10616+01	1.75754	.007307		
120-	.0042	1504,948545	-,46421+00	*18446+00	.240911	.001002		
121	.0041	1517,489783	13792+00	12885+01	1.72020	.007152		
122	.004⊥	1530.031021	79100+00	-,27901+00	. 686745	.002855		
123	.0041	1542.572258	•50355+00	- ,474 <u>1</u> 7+00	. 469223	.001951		
124	- ,00 40-	1555-113496		.30 043+80	<u>∍66273</u> 1	.002755		
125	•004Ŭ	1567.654734	-,27506+01	72037+00	8,01840	.033339		
126	.0040	1580.195972	. 15665+01	65317+00	3,01800	.012548		
127	.0039	1592.737210	22458+00	13651+01	1,92079	.007986		
128	.0039	1605,278448-	.31494+00	₀ ₀5564+00	542729	.002257		
129	.0039	1617.819686	65080+00	53412-01	. 415851	.001729		
1-30		1630,360924	.38081+00	29434+00	240958	.001002		
131	.0038	1642.902101	67699+00	80907+00	1.09074	.004535		
132	•0038	1655.443399	. 51012+00	•27073+00	. 353060	.001468		
133	.0038	1667.984637	.15230+01	•69721+00	3,06986	•012764		
134	.0037	1680,525875	15742+00	.30219+00	.114383	.000476		
135	.0037	1693.067113	22318+00	. 89068+00	. 832057	.003460		
136	.0037	1705,608351	. 14442-01	= ,29624+00	. 875075-01	. 000364		
137	.0037	1718.149589	. 67503+00	16723+00	.492132	.002046		
138	.0036	1730.690827	10075+01	19629+01	5.05962	.021037		
139	.0036	1743.232064	-,76118+00	. 69200+00	1.07396	.004465		
140	.0036	1755.773302	.38359-01	15876+01	2.52726	.010508		
141	.0030	1768.314540	.19750+01	48866+01	26.8603	.111679		
142	.0035	1780-8557-78-	-81933+00	91042+00	1.52042	-006322		
143	.0035	1793.397016	74828+00	.85166-01	.593278	.002467		
144	.0035	1805,938254	- 26801+00	-,45227+00	.279174	.001161		
145	.00.05	1818.479492	83221+00	.17848+00	.719649	.002992		
146	.00.54	1831.020730	.66772+00	-,28876+00	.526707	.002190		
147	. 00.54	1843.561967	- 95728+00	-,56962+00	1,18280	.004918		
148	.0034	1856, 103205	53216+00	23699+00	.326881	.001359		
149	. กกรษ	1868 6444443	13840+01	74113-01	1.8596	.007758		
150	1034	1881.185681	- 85560+00	.51725+00	.971919	.004041		
151	1033	1893,726919	- 43287+00	·89613+00	981018	.004079		
152	.1033	1906.268167	49291+00	=.11639±01	1,60112	006657		
153	. 0033	1918,409395	=.11164+00	11084±00	243027-01	000101		
154	0000	1031_350633	=.30±5~±00	=.230K1±00	.147113	000612		
1.74	+0000	T-01**0000000	00+00+00	*C0331400	• • · · · · · · · · · · · · · · · · · ·	******		

к	PERIOD	OMEGA	A	Ŗ	-REGRESSION Power	PERCENT POWER	c =	.44449+01
155	.0u32	1943.891870	. 86103-01	-,43136+00	.188547	.000784		~
156	+0032	1950.433108	-,87079-01	. 39746+00	.166493	.000692		
157	.0032	1968.974346	.69862+0 0	12843+01	2,24692	.009342		
~ 1 58	,0032 -	1981.515584	•580 80+00 -	21431+00	• 38 2052	-00158 8		
159	.0032	1994.056822	- .17789+00	17584+00	•634460-0⊥	.000264		
160	.0031	2006.598060	-,39421+00	14294+00	175356	-000729		
161	.0031	2019.139298	.10091+00	22574+00	•583952 - 01	.000243		
162-	. −0+0-3-1	- 2031-6805 36	.45036-01	-46554+00	214520	.000892		

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PERCENT PU JER

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.2052419	45.8280	43	I *																				
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SAMPLE - SPECTRUM-PROGRAM-RUN---UNEQUALLY-SPAGEn-DATA-

2.530853	107.784	54 I	* -	
1.475148	109.780	55 I	* .	
9777251	1-11,776	56 I	* *	
5753624	113,772	57 I	* •	
5044759	115,768	-58 I	* -	
-7842400	117.764	59 I	* .	
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-7////5309	135 720	1 10	* •	
2623301-01	137 720	60 1	т т ш	
*2222201-01	1-20-200	70 7	* •	
-1072794	1097720	70 1	_* T	
.1480595	141.710	<u> </u>	* •	
207420207	143,712	72 1	·** •	
.38/16/8	145.708	<u>[</u>]	* •	
2-010027	147-704	<u>74 1</u>	* *	
*2032053	149,700	75 1	* •	
3-081-305	151-696	76 1	* ĕ	
.8834761	153,692	77 1	* •	
1-1121-79	155.680	78 1	-**- •	
80.86932	157.684	79 I	•	
3.006326	159,680	80 I	-* .	
,4229238	161,676	81 I	* .	
1-957619-	163-672	-82 I	-* *-	
1.204703	165.668	83 I	* •	
.1509102	167-664-	84 I	* *	
. 5183510	169,660	1 C8	* .	
a4391449	17-1-,656	86 I	-*-	
. 5555428	173.652	87 I	* .	
,1739272	1-7-5-648	88 I	-*	
.8408406	177,644	89 I	* •	
-4019491	179+640	60 1	* -	
4021158	181.636	91 I	*	
4.231475	183.632	92 1	*	
1.094715	185.628	93 I	* .	
5240179-02	187-624-	94 I		
.1196840	189,620	95 T	* .	
1.156798	191-616	96 1	*	
.8658885	193.612	97 1	* .	
.8097044	195.608	98 1	* •	
7762041-01	197 604	00 I		
1.375303	100 600	100 1	т. •	
9109436	201.596	101 1		
6606701	201,590	102 1	* •	
7677104	202 699	107 1		
1 985356	203,300	100 1	* •	
1+0000000	207-1004	105 7	-∓ ~∎" ″	
• 7400000 '7051760	207,300	100 1	₩	
+ JHJ±J04	2-1-1-0(0	107 1	-¥¥ ∎-∸	
*000001Z	213,572	10/ 1	* •	
	219,908	1 08 I	* -	
1.778452	217,564	109 [*•	
9-888076	219,560	-1-10 I	-	*-
3.973488	221,556	111 I	*.	

.1671130	223.552	112 I * •
9246581	225-548	113 🛙 * 🔹
1 570335	227 544	114 T *
2611486	229.540	115 T *
+2011400	227.576	116 T *
• 9020040	201.000	
4209070	200.002	
1.3/111/	239,020	110 1 T
1./5/539	237.524	
.2409107	239.520	
1.720197	241.510	
+6867451	243.012	^↓<< こ オ ・ 1 つ ス T ψ
.4692231	245.508	
.6627310	247.504	
8.018402	249.500	125 1 • *
3.017999	251,496	126 1 * •
1,920786	253,492	127 I * •
. 5427287	255,488	128 I * •
,4158510	257.484	129 I * •
. 2409580	259,480	1 3 0 I * •
1.090745	261,476	131 I * •
,3530597	263.472	132 I * •
3.069857	265,468	133 I *•
. 1143833	207.464	134 I * •
.8320565	269,460	135 I * •
8750748-01	271.456	136 I * •
4921323	273,452	137 I * •
5.059622	275.448	138 I 🛛 *
1.073950	277.444	139 I * •
2.527255	279,440	140 I * •
26.86028	281.436	141 I •
1,520425	283+432	142 I * +
.5932779	285,428	143 I * •
2791740	287 424	144 I * •
7196486	289.420	145 I * •
5267068	291.416	146 I * •
1.182803	293.412	147 I * •
-3268812	295.408	148 I *
1.865957	297,404	149 I *
0710101	299.400	150 T *
0010183	301.396	151 1 *
1.601124	303.392	152 T *
24302-7-01	305.388	153 1 *
1071134	307 384	154 J * -
1095470	303 380	155 T *
.1003470	309.300	
• 1004933	JII.J/D	157 1 -
2.246916	313.31Z	1507 ···
.3820518	313.368	
.6344597-01	317.364	
.1753564	319.360	TOA 7 🗶 🔹
,5839524-01	321.350	101 1 * •
,2145198	323.352	162 I *•

-	-	REMOVAL T	ABLE FOR SET	RANDOM TIME	- 310568		
-PERIOD-	- FREQUENCY	- OMEGA -	-A	- - B -	c	- SPECTRUM POWER -	REMAINING - TOTAL POWER
، بين يكري الأرب بير .			يو ڪي جديد ندي ڪ		_ ~~~~~~ ***		
.0063	158.000090	992.743847	.89995+01	. 59415+01	.53828+00	114821+03	
.0278	35,999333	226.190484	-30407+01	.50085+01	.53828+00	-340945+02	
,0036	280,997330	******	.69338+01	.20961+01	.53828+00	.539825+02	
0045	220-012693	********	-29707+01	- 71298-01	.53828+00	-119490+02	-
-0040	250.001495	*****	.57892-01	.50101+01	.53828+00	·250682+02	.180425-01
·		-					

SEARCH WILL NOW BE MADE WITH SMALLER TOL.

FINAL DATA SHECTRUM FOR SET RANDOM TIME 310464 (TOTAL POWER = .18043-01)

N.	DEDIGO	(MEC)	۵		REGRESSION		c -	53828+00
ĸ	PERIOD	CMEGA	R	4			C =	
1	.5010	12,541238	.36604-02	. 55947 - 05	.135935-04	·000753		
2	.250s	25,082476	31807-02	.33627-02	.220645-04	.001223		
3	.1070	37,623714	.43095-03	·o9240-02	. 482660-04	.002675		
4	.1253	50.164951	.11154-01	39911-03	124620-03	.006907		
5	.1002	62,706189	12462-02	. 90598 - 03	.234729-05	.000130		
6	.0835	75,247427	44771-03	•ó17 <u>1</u> 9-02	.368800-04	.002044		
7	.0710	87.788665	44042-03	16946-02	•293580−05	.000163		
8	.0o2o	100.329903	14194-02	17993-02	. 540332−05	.000299		
9	.0557	112,871141	43593-03	15018-01	.22b702−03	.012565		
10	.0501	125,412379	-,35427-02	49189-02	. 368973−04	.002045		
11	0455	137.953617	. 16881-02	.81117-03	.3571.2-05	.000198		
12	0413	150.494854	17947-02	.19678-02	,706216-05	.000391		
13	.0385	163.036092	23749-02	76750-02	•660643 - 04	.003662		
14	.0358	175.577330	11793-01	18792-02	. 140186-03	.007770		
15	0334	188,118568	-,38260-02	75570-02	. 743457 − 04	.004121		
16	0313	200.659806	10633-02	-,55307-02	.307016-04	.001702		
17	0295	213,201044	.22748-02	80740-02	+720281-04	.003992		
18	.0273	225.742282	44742-01	80156-02	204654-02	.113429		
19	.0264	238.283520	91116-03	.39261-03	166606-05	.000092		
20	.0251	250.824757	67349-02	.68709-02	.911221-04	.005050		
21	.0239	263.365995	.21716-02	30738-02	145042-04	.000804		
22	.022d	275.907233	58120-02	.26491-02	•414267 - 04	.002296		
23	.0218	288.448471	.49607-02	.50234-02	. 505900-04	.002804		
24	-0209	300.989709	34973-02	37464-02	.264944-04	.001468		
25	.0200	313.530947	14611-02	94732-03	.297267-05	.000165		
26	.0193	326.072185	58990-04	49394-02	.240527-04	.001333		
27	-0180	338.613423	08323-02	.79889-02	.105308-03	.005837		
28	.0179	351,154660	•98155 - 02	43504-02	.110993-03	.006152		
29	.0174	363.695898	11121-02	42481-02	.189510-04	.001050		
30	.0167	376.237136	.22107-02	.55350-02	.354877-04	.001967		
31	-0162	388.778374	48751-02	10269-01	.134366-03	.007447		
32	.0157	401.319612	.25084-02	41371-02	229554-04	.001272		
33	-0152	413.860850	62061-02	11953-02	J82463-04	.002120		
34	.0147	426.402088	.53587-02	- 96269-02	122139-03	.006770		
35	.014.5	438 943326	14145-01	46597-92	21-3411=03	-011828		
36	.01.9	451,484563	11613-02	.31665-02	112551-04	.000624		
37	.0135	464 (258)1	10160-01	-,77149-02	.158153-03	.008766		
38	.0132	476.507039	10008-01	21881-02	.105096-03 -	.005825		
39	.0128	489.108277	13542-02	-,46820-02	224732-04	.001246		
μn	.0125	501.649515	- 38146-02	35364-02	273796-04	.001518		
цці	.0122	514,190753	.8321p=02	10634-01	.181522-03	.010061		
42	.0119	526.731991	50989-02	79273-02	.852024-04	.004722		
43	.0117	539,273229	.99858-03	-,46805-02	.213173-04	.001182		
44	.0114	551.8144-6	31725-02	58790-02	.420738-04	002332		
45	.0111	564.355704	82422-03	.65512-02	.449862-04	.002493		
46	.0109	576.896942	32568-02	29220-02	.191217-04	.001060		
47	.0107	589.438180	.39541-03	55276-02	-313596-04	.001738		
48	.0104	p01.979418	38372-02	,45657-02	.356193-04	.001974		
49	.0102	614.520656	20494-02	13936-02	.611097-05	.000339		
50	.0104	627.061894	18480-03	39966-02	.154631-04	.000857		
00	.0100	021002094	170,00,00	101700 OE	· · · · · · · · ·			

					REGRESSION	PERCENT		
к	PERIOD	OMEGA	Α	ß	POWER	POWER	c =	. 53828+00
				•				
51	.0098	639,603132	.48871-03	64025-02	.411603-04	.002281		
52	+0090	652.144369	-,48842-02	19121-02	.281386-04	.001560		
53	.0095	664.685607	•5024o-02	. 60103 - 02	•591330-04	.003277		
54	•0 0 90	677.26845	. 8700 0 −02	. 22884-02	∗841405-0 4	.00 46 63		
55	.0091	689,768033	88014-03	39134-02	150495-04	.000834		
56	•0089	702.309321	. 55371-02	42472-02	.479847- 04	.002660		
57	•0088	714.650559	. 10833−01	21317-02	•126403 - 03	.007006		
58	.0060	727.391797	.32955 - 02	•74291-02	. 648 3 17 - 04	.003593		
59	•0085	739,933035	84419-04	36654-02	.133760-04	.000741		
-60-	~ 0084	752,474272	5757d-02	82593-03	.331952-04	.001840		
61	•0082	765.015510	•79837-03	19190-02	424344-05	.000235		
62	•9081	777.556748	.11519-01	•99014-02	.232015-03	.012859		
63	•0080	790,097986	.32564-02	•08476-02	.550246-04	.003050		
64	.007a	802,639224	4183d-02	.11970-02	.178006-04	.000987		
65	.0077	815.180462	.18188-02	.57891-02	.373420-04	.002070		
66	.0070	827,721700	67738-02	57007-02	.776867-04	.004306		
67	.0075	840,262938	+40491-02	73343-02	+688730-04	.003817		
68	.0074	852,804175	66944-02	.15856-03	.451363-04	.002502		
69	.0073	865,345413	-,51923-03	.36359-03	.349017-06	.000022		
70	.0072	877.886651	=.11343=01	•73963 - 02	.185850 - 03	.010301		
71	.0071	890.427839	.11900-02	54398-03	.169553-05	.000094		
72	•0070	902,969127	.31620-02	40154-02	.261932-04	.001452		
73	•0069	915.510365	.56108-02	70879-02	.811013-04	+004495		
74	.0063	928,051603	106/4-01	.10740-02	+116529 - 03	.006439		
75	.0067	940.592841	10961-01	•14366-01	+32125/=03	.01/000		
/6	+0066	953,134078	.10699-02	29179-02	+998185-05	.0000000		
77	.0065	965.675316	+97568=U3	.13988-01	•196965-00 F70070-00	•U1U917		
78	.0064	978.210004		40180-02	+370979-04	4 00 0100		
79	.0063	990+757792	-+37001-01	+ 30340-01 - 803-0	•231003-02 \\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	+140T//		
80	+0063	1015 6002.9	36703-02	- 10901-01	134300-03	007666		
01	+000Z	1010.040200	+J070J-02 - #8270+02		.197587-0U	.007000		
02	-0001 -0060	1020-001000	- 89354-02	- 34567-02	.866859=04	.004805		
80 81	•0000 (ມ)6.0	1053.463981		.18568-02	.343264-B4	.001903		
04 25	0000	1066 005219		-10080-02	.444317-04	.002463		
86	.0059	1078.546457	.49218+02	-70035-02	.733271-04	.004064		
87	.0058	1091.087695	.37565=02	25572-02	1998p6=04	.001108		
88	- 0055	1103-628933	-28263-02	70363-03	.873306-05	000484		
29	0057	1116.170171	.50517-02	-17119-01	-313193-00	-017359		
07	•0000 ເປີໂດ	1128.711409	. 34614-02	-10629-82	.127837=04	.000709		
90	- 0050	1141.252647	.13235+01	27572-02	.172708-03	.009572		
92	- 1055	1153.793884	72534-03	10669-02	1700p3-05	.000094		
93	. 0054	1166.335122	25732-03	10827-01	122059-03	.006765		
94	- 105-1	1174.876300	.20585-04	10950-02	.118266-05	.000066		
95	.0055	1191,417598	2882.0-02	.93771-02	.927000-04	.005138		
96	-11052	1203.958836	25479-02	57725-02	.410322-04	.002274		
97	.0052	1216,500074	.60530-02	.92479-02	.126447 − 0J	.007008		
98	.0051	1229.041312	53317-02	-,46559-02	.507892-04	.002815		
99	.005r	1241 582550	- 18470-02	31031-02	.130678-04	.000724		
100	. ບປວກ	1254 123737	-,58454-02	10648-02	.326338-04	.001809		
101	0050	1266 665025	08320-02	-,76763-04	.491394-04	.002724		
102	.0049	1279,206263	.53029-02	. 30732-02	.515329-04	.002856		
				-				

					REGRESSION	PERCENT		
к	ΡΕΚΙΟΟ	OMEGA	A	2	POWER	POWER	с =	•53 ⁸ 28+00
1.113	11010.4	1001 787501	5850	.916:1=02	-8318d3=04	.004610		
103	.0049	1271+141301	-31(53.03	10003 01	1620-8-03	008984		
104	10040	1304.200739	- (1009-02 10007-02	+10223-01	4427-0-0A	003673		
102	• 7 0 4 3	1316.829977	.10627-02	+0091)7-02	+002/00-04 0661 //-0+	01/098		
106	.0047	1329.371215	•16429-01	+66941-0J	•200104-00 J6 \670-00	000227		
107	.0347	1341.912455	.85170-02	16582-02	•/020/4T04	+004444		
108	.0046	1354.453690	50107-02	-,28433-02	• 346298-04	•UU1717 002001		
109	+ÛU40	1366,994928	61829-02	22530-02	•433206-04	00000101		
110	.0040	1379.536166	.44983-01	,50461-01	•44/884 ⁻⁰ 2	000676		
111	ຸງບໍ4ວ	1392.677404	10325-02	02450-02	•121941-04	.000070		
112	. ປີປ4ວ	1404.618642	21453-01	83114-02	.507071-05	•U∠0104		
113	.0044	1417.159830	.77047-03	-,98852-03	.154437-05	.0000000		
114	.0044	1429,701118	.48615-02	24293-02	.298/42-04	.001055		
115	.0044	1442.242356	.42651-02	53134-02	.469464-04	.002602		
116	.0043	1454.783593	.26351-02	14178-01	.207840-03	*011519		
117	.0043	1467.324831	.50405-02	-,45674-02	.441526-04	.002447		
118	.0042	1479.866069	.14937-02	.10385-01	.110834-03	.006143		
119	.0042	1492.407307	.40518-02	47298-02	.414913-04	.002300		
120	,0042	1504.948545	29658-02	11802-02	.100096-04	.000555		
121	.0041	1517.489783	88015-02	81503-02	<u>148297-03</u>	.008219		
122	.0041	1530.031021	59101-02	90406-03	.355013-04	.001968		
123	.0041	1542.572258	.27946-03	97102-02	891132-04	.004939		
124	.0040	1555.113496	35095-02	25970-02	• <u>184626-</u> 04	.001023		
125	.0040	1567,654734	-,95547-02	11981-01	•2418u6−03	.013402		
126	•0040	1580.195972	. 45377-03	. 10188-01	. 102334−03	.005672		
127	.0039	1592.737210	. 41888-02	•14526-01	•229546 − 03	.012722		
128	.0039	1605.278448	.24240-02	-,50797-02	•301005 - 04	.001668		
129	.0039	1617.819686	•51747 − 03	48330-02	<u>.</u> 239555−04	.001328		
130	.0039	1630,360924	-,15773-03	. 70792-02	.489358-0 4	.002712		
131	•ປປວີອີ	1642.902101	32202-02	.57156-02	•417246 ~ 04	.002313		
132	.0038	1655,443399	. 57049 - 02	90465-02	.107981 - 03	.005985		
153	.0038	1667.984637	. 4857υ−02	. 53843-02	. 558756-04	.003097		
134	.0037	1680,525875	-,28200-03	.38799-02	. 142918 − 04	.000792		
135	.0037	1693.067113	.78113-03	.15002-04	.617366-06	.000034		
136	.0037	1705-608351	. 59086-02	-,59709-02	.665268-04	.003687		
137	.0037	1718.149589	.14447-01	43797-02	•232524-03	.012888		
1.38	•003n	1730.690827	13349-02	79498-02	.663906-04	.003680		
139	.00.56	1743.2320.4	.21768-02	.67415-02	•512524-04	.002841		
140	-0036	1755.773302	.75124-02	.14790-01	.279768-03	.015506		
141	.0036	1768.314540	14647-01	59307-01	.354561-0∠	.196525		
142		1780-855778			·299257-0s	.016586		
143	.0035	1793.397016	.89901-02	61263-02	•117195 - 0s	.006495		
146	0035	1805.938254	17016-02	85949-03	.360934-05	.000200		
145	.0035	1818-479492	90803-03	41100-02	.178370-04	.000989		
146	0034	1831.020730	18815-02	35125-02	.154321-04	.000855		
140	+0004 003/L	1843.561967	- 48900-02	-58768-02	.604962-04	.003353		
147	0004	1056 103205	– 07//6//→02	- u9us9-02	-115755-03	006416		
140	*0030 *0030	1868 CUPTAN	-2554Q=02	- 685//4-02	568220-04	003149		
150	.UUJ4 njæ	1881 185601	- 73747-02		.136162-03	007547		
161	+0033	1001+10001		17528-02	390510-04	002164		
150	.0033	100- 040157	25820-02	=,13871_01	195199-03	010819		
152	.0033	1010 000302 1010 500101	+20020-02	85616-03	-767083-05	000043		
122	•0033	1910-909342	+01040-04	+00010-00	10207000-00	. 005990		
154	. 0∪≾3	1931,350633	.lu149−U1		∎-I-U-OU-F H U-O	+043220		

к	PERIOD	OMEGA	A	З	REGRESSION POWER	PERCENT POWER	c =	.53828+00
155	.0032	1943.891870	.39724-02	24817-02	.225792-04	.001251		
156	.0032	1956,433108	53312-02	32057-02	.379312-04	.002102		
157	.00.32	1968,974346	. 8614o-02	18288-02	802928-04	.004450		
158	.0032	1981 .5 15584	-10552-01		.140244-03	.007773		
159	0052	1994,056822	99427-02	• 52826-02	•131549−0ა	.007291		
100	.0031	2006.598060	14332-02	13484-02	•395745 − 05	.000219		
161	.0031	2019.139298	16569-02	. 68240 - 02	. 500789−04	.002776		
162	.0031	2031.680536	.89621-02	35528-03	. 813186-04	.004507		

SAMPLE SPECTRAN PROGRAM KUN - ULEQUALLY SPACE) MATA

					0.0.00	• • • •				
POWER										
,448 - ∪2 I										
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1359351-04	1.99600	1	I *	•																		
.2206446-04	3,99200	2	1*	٠																		
.4825002-04	5,98800	<u>ວ</u>	1 *	•																		
.1246200-03	7,90400	4	1 *	•																		
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.0606427-04	25.9480	13	1 *	•																		
,1401360-03	27.9440	14	I a	¥ .																		
, 74345od-04	29.9400	15	I *	•																		
.3070155-04	31.9360	16	I *	•																		
,7202810-04	33.9320	17	I *	٠																		
.2046543-02	35.9280	18	Į	•																		
+1006008-05 011000-00	37.9240	19	1 * 1	•																		
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.2972609-05	49,9000	25	- I *																			
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.1109926-03	ა5.5680	28	I *	•																		
1895104-04	57.8840	29	I *	•																		
.3548767-04	59.8600	30	I *	•																		
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.1050956-03	75+8480	38	1 ×																			
.2247323-04	77+8440	39	I *	•																		
.2737959-04	79.8400	40	I *	•																		
.1815218-03	ძ⊥,მა60	41	I X	ĸ .																		
.8520237-04	03.8320	42	I *	•																		
,2131729-04	45.8280	43	I *	•																		
+207302-04	87.8240	44	1 ¥	•																		
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.3135955-04	91.0100	47	⊥ ~ ⊺ *																			
.3561930-04	95.8080	48	тж Тж																			
.6110971-05	97.6040	49	I *	:																		
1546309-04	99.8000	50	I *	-																		
.4116030-04	101.796	51	I *																			
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. 591329o-04	103.788	53	I *	•																		
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.8414059-04	107.784	54 Ť *
150005-00	100 700	
.1304933-04	103.190	55 1 * .
, 4798471 * 04	111-776	56 I * 🔸
. 1264u30-03	113.772	57I * .
6483166-04	115.768	58 T * .
1337595-04	117 764	50 T +
.1007090-04	117.704	391*
.3319924-04	119-760	60 F * •
, 4243437 − 05	121.756	61 I * 🔸
.2320152-03	123-752	62 I * .
5502456-04	125.748	63 ī *
1780065-04	127.744	64 T *
3730 300-00	100 740	
.5734200-04	129.740	651* .
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. 6887302 - 04	133.732	67 I * .
4513628-04	135.728	68 I * .
3990174-06	137.724	69 7 *
1050114 00	130 720	
.1000499-00	1390120	10 1 *
*T9A2222-02	141.716	/1 1 * •
<u>,2619319-04</u>	143.712	72 F * 🕌
.8110125-04	145.708	73 I * .
1165286-03	147-704	74 1 *
3212673-03	100 700	75 1 4
.0212070-00	149.700	
*AA91925-02	107-000	<u>76 1 * -</u>
, 1969⊳52−03	153.692	77 I * •
-570 979 0-04	155.688	7 8 I * -
.2318050-02	157.684	79 I
1405202-03	169-680	80 7
1393 200-03	161 676	
,1303224-03	101.0/0	<u> 01 1 * •</u>
-1975875=04	163.672	821*
.8668885-04	165.668	83 I * .
-3432642=04	167-664	84] * .
4443169-04	169.660	85 1 *
7332715-04	171 656	86 T *
1002110-04	471.000	
.1998004-04	173.652	8/1* .
. 8733064-05	175,648	. * I 88
.3131926-03	177.644	89 I *.
-1278373=04	179.640	90.1*
1727077-03	191.636	01 T +
1700431-05	102 (30	
F1700632=03	103.032	96 5 * •
,1220587-03	185.628	93 I * •
. 1182656-05	187.624	94 I * 🕝
.9270005-04	189,620	95 I * .
4103220=04	101_616	96 I V
10600200-01	104 610	07 1 +
.1204471-00	190.012	
.5078920-04	195,608	981*
.1306782-04	197.604	99 I * •
.3263881-04	199.600	100-1 *
4913939-04	201.596	101 T *
5153200-00	203 502	
0710074-04		
.8318033-04	205.588	1031 * •
.1620879-03	207,584	-104 I 🛶
. 6627601 - 04	209.580	105 I * .
-2651841-03	211-576	106 t 🛶 🔒
7626743-04	213.572	107 1 *
3462086-04	216 640	109 7
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.4332060-04	217.564	1091* .
,4478839-02	219.560	110 I .
. 1219407-04	221.556	111 I * .
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-5070707-03	223 552	112 I *
.1544367-05	225,548	113 I * .
-2987420-04	227-544	114 I-+ ·
,4694641-04	229.540	115 I * •
-2078403-03	231-536	116 I -* •
4415257-04	233.532	117 I * .
-1108337-03	235.528	118 i * •
,4149130-04	237.524	119 I * •
-1000959-04	239-520	120 f * -
. 1482972-03	241.516	121 I * .
-35501-33-04	243-512	1-22 f * ;
. 89 11 321-04	245,508	123 T * 🔸
.1846263-04	247.50 4	124
.2418063-03	249,500	125 I 🛛 \star 🔒
.1023336-03	251-496-	126 I * .
<u>2295457-03</u>	253.492	127 I *•
-3010049-0 4	255,488	128 I * .
.2395549-04	257.484	129 I * .
~4893581-04	-259-480	130 I * •
4172458-04	261.476	131 I * •
-1079808-03	-26-3-472	1-32 I * .
•5587557 − 04	265 468	133 I * •
-1429183-04	267,464	134 [* •
.6173660-06	269.460	135 [* .
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.2325240-03	273,452	137 I * .
	275,448-	138 1 * -
•5125242 ⁻⁰⁴	277,444	1391*
7545010-00	279-440-	-14-11 × -
• 3345810-02 3083570-03	281.436	
1171046-07	203,432	
-11/1940-US	285,428	
1783407-00	2017424	
1543208-04	207,420	
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-1361618=03	200_400	150 I w -
3905103-04	401-396	151 T *
1951993-03	303.392	152 I
7670826-06	305.388	153 T *
1080738-03	<u>307 384</u>	<u>154 I × _</u>
2257922-04	309.380	155 T *
3793119-04	311 376	156 I *
8029276-04	313.372	157 T *
1402444-03	315 368	158 I *
.1315485-03	317.364	159 I *
-3957450=05	319.360	160 I *
.5007891-04	321.356	161 I *
-8131857-04	-323-352-	162 I * -

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