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5.9 GHz inter-vehicle communication at intersections: a validated non-line-of-sight path-loss and fading model

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Abstract

Inter-vehicle communication promises to prevent accidents by enabling applications such as cross-traffic assistance. This application requires information from vehicles in non-line-of-sight (NLOS) areas due to building at intersection corners. The periodic cooperative awareness messages are foreseen to be sent via 5.9 GHz IEEE 802.11p. While it is known that existing micro-cell models might not apply well, validated propagation models for vehicular 5.9 GHz NLOS conditions are still missing. In this article, we develop a 5.9 GHz NLOS path-loss and fading model based on real-world measurements at a representative selection of intersections in the city of Munich. We show that (a) the measurement data can very well be fitted to an analytical model, (b) the model incorporates specific geometric aspects in closed-form as well as normally distributed fading in NLOS conditions, and (c) the model is of low complexity, thus, could be used in large-scale packet-level simulations. A comparison to existing micro-cell models shows that our model significantly differs.

1 Introduction

Vehicular communication is envisioned to increase range and coverage of location and behavior awareness of vehicles, thus enabling highly developed pro-active safety systems.

The idea is that all vehicles communicate information like position, speed, and heading periodically to other vehicles in cooperative awareness messages to enable the derivation of an environment picture, used as basis for movement prediction. An *ad hoc* communication technology working on 10 MHz wide frequency bands centered around 5.9 GHz in the U.S./Europe is in development. Medium and physical access is standardized as IEEE 802.11p [1].

For this radio technology, cross-traffic assistance at inner city intersections is one of the most challenging use cases. It needs to monitor two spatial dimensions and is very sensitive to heading estimation inaccuracies [2], thus implying the need for high information update rates. At the same time, the corners at inner city intersections will often be occupied by buildings. They can

block the radio line-of-sight (LOS). If the LOS is blocked, diffraction, reflection, and refraction of radio waves can enable non-line-of-sight (NLOS) reception (Figure 1). But, the relatively high frequency of 5.9 GHz and a difficult radio fading environment might complicate the NLOS reception of packets.

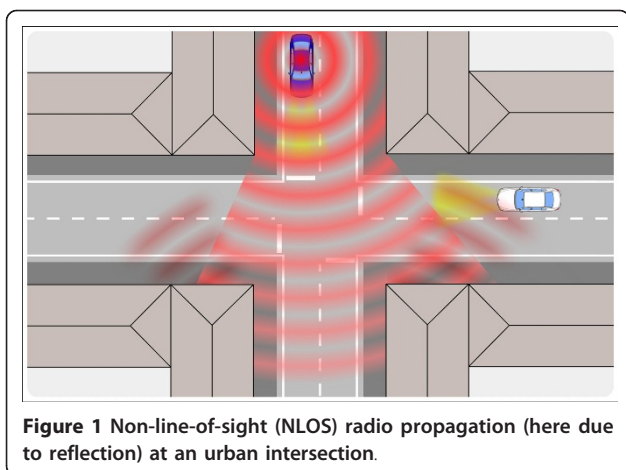
Our analysis of building locations at intersections in the city of Munich [3] showed a need for NLOS reception: If two cars are approaching an intersection with 50 km/h, only 30% of all intersection corners provide LOS at a desired warning point of 3 s [4,5] to a potential impact (\approx intersection center). While road side units (RSU) might re-broadcast messages and reduce the need for NLOS reception, urban street crossings with sparse traffic are likely not to be equipped with a dedicated RSU. Such scenarios require robust signal transmission between vehicles approaching a crossing. These situations predominantly exhibit NLOS radio link conditions between vehicles, motivating to investigate NLOS reception.

First measurements on 5.9 GHz NLOS reception were done with channel sounders in [6] and with off-the-shelf radios in [7-10]. While showing that NLOS reception is generally feasible, they did not investigate the influence of factors like building placement on reception quality.

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Also, all of them lack a systematic and representative test site selection, questioning the generalizability of results. NLOS path-loss models were—albeit simplistic—deduced in [9,10].

While NLOS models related to vehicle-to-vehicle (V2V) communication exist in cellular research—for below rooftop base stations [11-16], those were only validated at lower frequencies (0.9-2.1 GHz) and higher transmitter heights (3-4 m). Therefore, a validated 5.9 GHz NLOS model for V2V communication is still missing.

To gain more insight about 802.11p NLOS reception properties, we performed an extensive field test, specifically targeted to measure the quality of NLOS reception and to characterize propagation [17]. Special attention was paid to a well-founded selection of representative test cases and to find a test setup that allows for the derivation of predominant influence factors such as inter-building distance. A comparison of the data to the existing cellular models showed that they cannot be properly applied in the intended scenario.

In consequence, we developed—based on the measurement data—a specific 5.9 GHz NLOS propagation model for inter-vehicle communication. The proposed path-loss and fading model is intended to be used in packet level load simulations.

The article is organized as follows: our measurements will be presented in Section 2. Subsequently, we will deduce the NLOS path-loss model, called VirtualSource11p, and characterize NLOS small-scale fading in Section 3. Afterwards, we will compare our model to existing models in Section 4.

2 NLOS reception measurement campaign

Proper measurements are inevitable to judge existing models and deduce our new specific one. Therefore, a summarization of our measurements as published in [17] is provided here.

2.1 Test design

We used off-the-shelf radios to measure the channel in terms of per packet reception power values. While radios cannot provide detailed insight into the impulse response of the channel such as channel sounders do, the collected data provide the same per packet generalization as packet level simulators assume. With respect to limitations in reproducibility and traffic synchronization, we opted against a setup with two vehicles independently moving against each other. Our solution is a discretization of the transmitter (tx) position, with the receiver (rcv) driving on the crossing street. Consequently, we split the dimension $tx \leftrightarrow rcv$ into $tx \leftrightarrow center$ (of intersection) and $center \leftrightarrow rcv$. The measurement design is visualized in Figure 2, showing the fixed transmitter and moving receiver vehicle. By default, we tested two $tx \leftrightarrow center$ distances on each side street, 30 and 60 m to the center, corresponding to 2 and 4 s to center at 50 km/h. An additional position near the intersection center (0 m) was tested for LOS comparison.

2.2 Used hardware

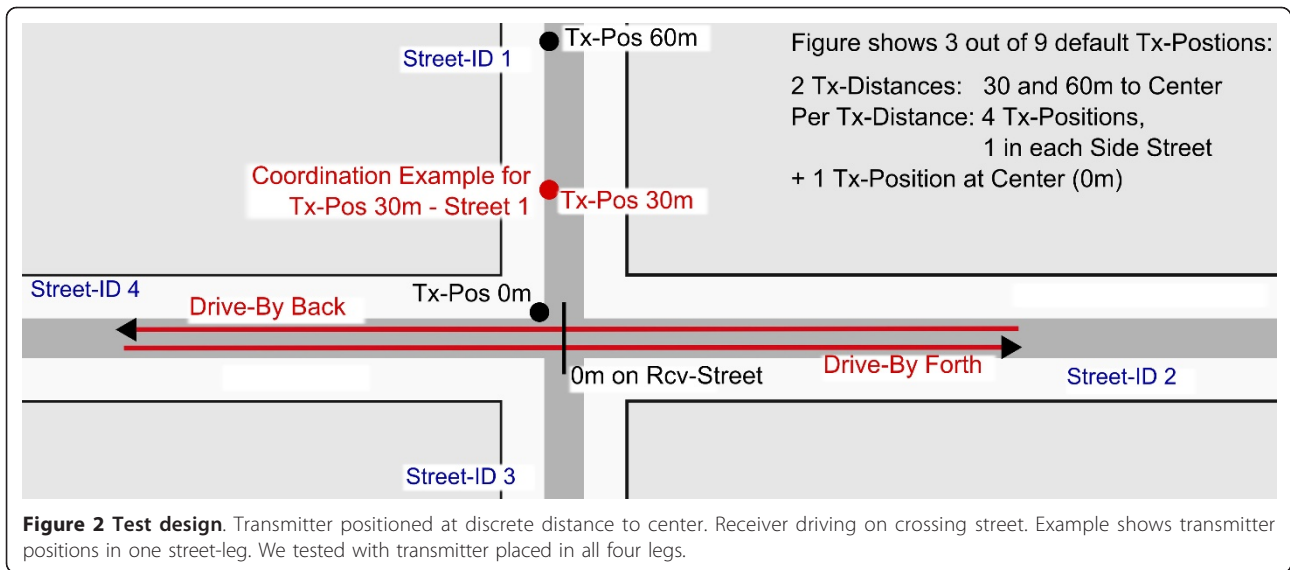
The receiver was integrated into a BMW 5 Series GT. GPS information was taken from the CAN bus, providing a high accuracy position (GPS data are enhanced by vehicle sensor data and map matching). A tripod with 35 × 35 cm metal plate at a height of 1.45 m was used as the transmitter mount as it is well placeable. Also, a vehicle would have blocked other traffic.

We used a LinkBird-MX V3 [18] 802.11p communication box by NEC containing two DCMA-82-N1 Mini-PCI cards with Atheros 802.11 radio chips. To resemble a close-to-production system, we used small low-profile puck antennas from Nippon Antenna. They provide a gain of 0 dB at 0° (= horizon) and +5 dB at 15°. Each antenna cable inherits a loss of 2.4 dB. These values are according to the corresponding data sheets.

2.3 Systematic intersection selection

The urban canyon micro-cellular NLOS models predict the influence of geometrical aspects, probably most important of the inter-building distance in a side street (\equiv side street width). The general assumption is: the smaller this distance, the less power in NLOS in the crossing street. An intuitive explanation of this observation—a better view into the side street—is visualized in Figure 3. We intended to verify this observation by selecting intersections with the same as well as with differing width.

To achieve generalizable results, a prototype intersection ("Main Case") was defined as found by clustering in [3], whose geometry is representative for $\approx 50\%$ of all traffic intersections in Munich. Those intersections exhibit an inter-building distance between 20 and 25 m in



each side street. Intersections with 90° angles were selected for comparability. Since the classification is based on nearest building vertex, and front gardens (with bushes and trees) can further limit the field of view, we suspected suburban intersections to provide worse reception conditions at similar distances into the crossing street compared to urban ones. To be able to check the influence, we selected three suburban and one urban ones with the “Main Case” dimensions, identified by ID 1, 2, 3 and 10 in Table 1.

To check the influence of street width, we selected one urban intersection with 30 m width (ID 20), and another in-between (ID 11). A really wide one (ID 21) and one with only two occupied corners (ID 9) complete the selection.

2.4 Parameters, evaluation and results

We tested, in general, with 3 Mbps and 20 dBm transmission power. The transmission frequency was 100 Hz and payload 200 Bytes. With headers, packets had a size of 258 Bytes.

The transmitter positioning and test of 4-leg intersections leads to 20 runs for a typical tested intersection (see Figure 2). In total, we tested 71 tx-positions and performed ≈170 runs. Table 2 shows an overview of the resulting setup. We generated reception power/rate versus $rcv \leftrightarrow center$ distance plots and a map-based result visualization for each run, available at [19]. One of those plots is shown in Figure 4. Averaged results over back/forth run showed that for one $tx \leftrightarrow center$ distance, the results from all four side streets are mostly very similar,

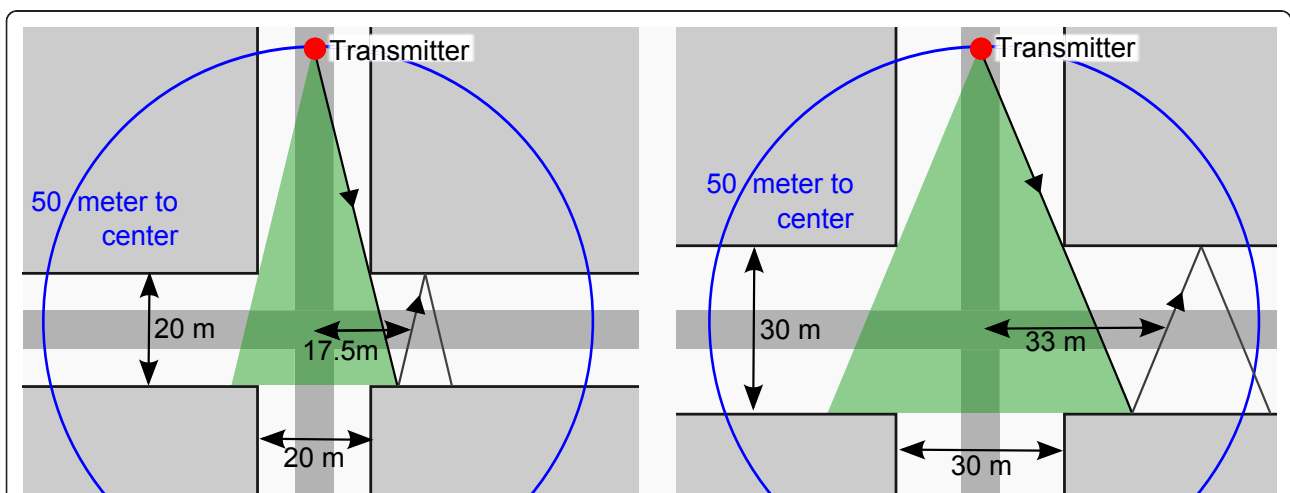


Figure 3 Exemplary visualization of change of LOS/NLOS area size and propagation by increase of inter-building distance from 20 to 30 m. The increased reflection angle in the right leads to a farther propagation of the reflected wave into the crossing street.

Table 1 Tested intersections

ID	Streets	Center lat/lon	Street width	Description
1	Pommer/Konstanzer	48.184, 11.560	23 m, 21 m	Suburban (main case)
2	Himmelschlüssel/Josef-Seifried	48.195, 11.532	23 m, 19 m	Suburban (main case)
3	Tizian/Kratzer	48.162, 11.525	≈21 m	Suburban (main case)
10	Agnes/Teng	48.158, 11.569	24 m, 21 m	Urban (main case)
11	Perlach/Untersberger	48.110, 11.586	27 m, 22 m	Urban increased width
20	Klug/Waisenh.	48.164, 11.529	≈30 m	Urban high street width
21	Hanauer/Gneisenau	48.179, 11.529	55 m, 30 m	Urban very wide, non-regular shape
9	Gotebold/Driesch	48.179, 11.442	18 m	Suburban, buildings at only two corners
100	Free Space, Country Road	48.247, 11.348		One street, no buildings, no trees, etc.

as can be seen in Figure 5. Therefore, we also produced intersection wide averages per $tx \leftrightarrow center$ distance. For 60 m $tx \leftrightarrow center$ distance, they are visualized in Figure 6.

Figure 6 shows that intersections with same street width and setting (suburban intersections 1, 2, 3) exhibit the same performance, and that path-loss decreases with increased street width (compare urban intersections 10, 11, 20, and 21 with increasing width). Suburban intersections have an estimated 3-4 dB less rcv-power compared to urbans with same width (urban intersection 10 against suburban 1, 2, 3). In general, NLOS reception is well feasible, with 50% or more reception rate at 50 m to center for transmitter and receiver.

3 NLOS propagation model development

Subsequent, we deduce a specific vehicular 5.9 GHz NLOS path-loss model—VirtualSource11p—from the measurements and characterize small-scale fading in NLOS areas.

Table 2 NLOS reception testing configuration

Parameter	Selected value
Tx-Power	20 dBm
Tx-DataRate	3.0 Mbps (BPSK modulation)
Channel	10 MHz, number 180 (5.9 GHz center freq.)
Packet Size	200 Byte payload + IP/MAC/PHY headers
Tx-TransmissionRate	100 Hz
Tx Dist. to ISect. Center	0, 30, 60 m (optional also 100 m)
Tx Street	1-4 (street legs) + 0 (center of intersection)
Drive Direction	Two directions per transmitter position
Communication Box	NEC LinkBird v3 (Atheros Chipset)
Antenna	Nippon DSRC Puck, 0°: 0 dB gain, 15°:+5 dB
Rx Antenna Position	Roof center (optimal position, evaluated in [17])
Cable (Box to Antenna)	SUCOFLEX_104, 4 m, data sheet: 2.4 dB loss
Transmitter	Tripod, 35 × 35 cm metal plate at 1.45 m height
Receiver	BMW 5 Series GT, no sunroof
Intersections	4 suburban + 4 urban + free space

3.1 Data quality and system loss

To deduce path-loss and fading from an of-the-shelf radio, it needs to provide reliable reception power values. We got per packet reception power values with 1 dBm resolution via the level value in the *iw_statistics*. *iw_quality* struct from a *SIOCGIWSTATS* socket call. Measured values equal to the reported values in iwconfig and the Linkbird wlan11p tool. Observed values range between -5 and -92 dBm. This corresponds well to the reception sensitivity of -92 as reported in data sheets such as [20]. Figure 7 shows the good quality of the reported power values over time for two exemplary 20 m street stretches. Differences in small-scale fading between NLOS and LOS are clearly visible. The resulting power histograms (compare Figure twelve in later Section 3.3) show very reasonable results too. More of these detailed plots are available at the website [19] covering this work.

There was only one issue: power histograms revealed that there are no packets reported with -69, -68 and -67 dBm. A figure illustrating this can be seen on the website [19]. The same gap can be seen in [21]. We believe that the chipset changes its sensitivity in this power range, and reports values above -69 dBm by 3 dB too strong. We corrected this by subtracting 3 dB from all reported values >-69 dBm. The reported reception sensitivity was not changed by the correction.

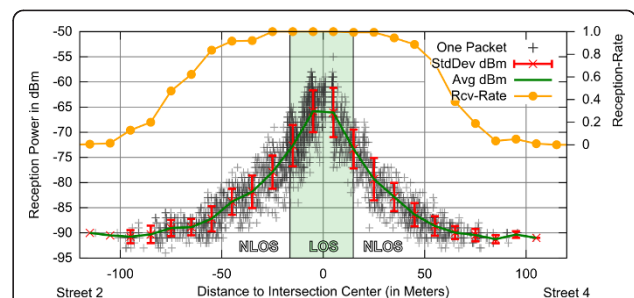
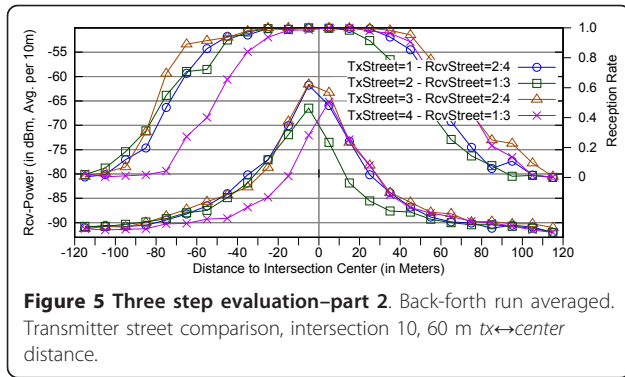


Figure 4 Three step evaluation—part 1. One of the 170 tests. Intersection 10, 60 m $tx \leftrightarrow center$ distance, Transmitter in street 1.



We measured received power, where in dBm space: $RxPower = Txpower - SystemLoss - PathLoss$. To determine path-loss, we need to know system loss. The cables lead to a combined loss of $\approx 4.8\text{dB}$ and the antennas to a gain of two times some value between $0 (0^\circ)$ and $5\text{ dB} (15^\circ)$.

To determine the average loss, we took the LOS measurements from most intersections (excluding free space, special case 9, and 1) and determined the deviation between average power curve and the theoretical limit. The fit equation used is

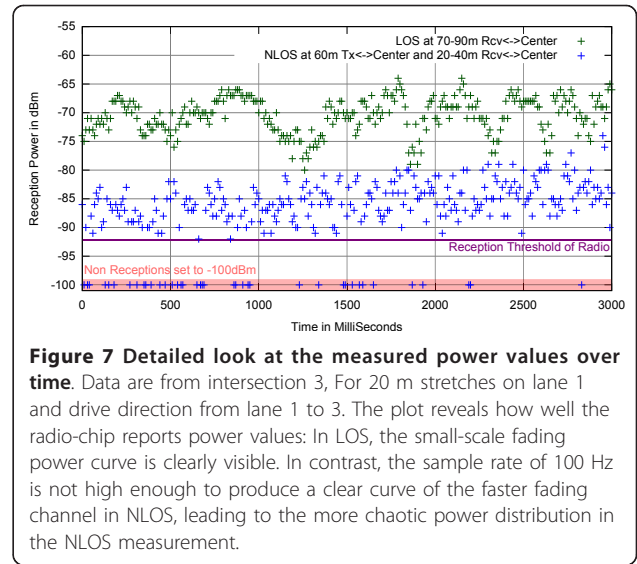
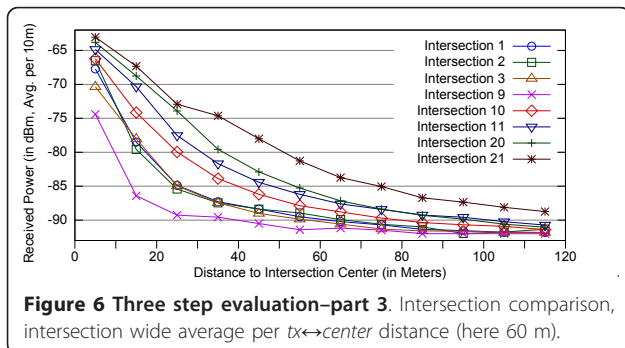
$$\text{LogDist}(x) = P_{tx} - L_S - PL_{ref} - 10 * E_L * \log_{10} \left(\frac{x}{T} \right)$$

$$PL_{ref} := \text{FSPL}(1) = 10 \log_{10} \left(\left(\frac{4\pi \cdot 1}{\lambda} \right)^2 \right) = 47.86 \text{ dB}$$

FitDimensions : $L_S = \text{SystemLoss}, E_L = \text{LossExponent}$.

It comprises the common Log Distance model and free space path loss (FSPL) for determination of reference loss. Unfortunately, curve attenuation interferes with slope variation. Therefore, three fits were performed (Variable, $SL = 0, LE = 2$). Figure 8 visualizes fit input and results. We limited the fit input to $20 < x < 150\text{ m}$, as packet loss occurred at $x < 150\text{ m}$ and small distances are inaccurate as the transmitter was not exactly positioned in intersection centers.

The fit reveals a loss of 2.75 dB with $LE = 2$. With $SL = 0$, it shows a loss exponent of 2.1 , being higher as in FSPL. Subsequently, we will assume 1.75 dB system loss,



as revealed by fitting both variables. The resulting average gain of 1.5 dB per antenna seems realistic, given its characteristic. Note that such loss determination absorbs the problem that real transmitted power might slightly differ from the configured value.

3.2 NLOS path-loss model development

To determine path-loss with respect to variable street-width and suburban/urban differences, we fit the intersection wide average power distance curves of multiple intersections (such as in Figure 6) to a unified path-loss equation. The basis for our fit equation is the cellular model proposed in [14]. The original VirtualSource equation (as given in [14], but indices modified to Figure 9 and as positive path-loss in dB) is

$$\text{PathLoss} = \begin{cases} 10 \log_{10} \left(\frac{1}{\alpha} \left(\sqrt{\frac{2\pi}{x_t w_r} \frac{4\pi d_t d_r}{\lambda}} \right)^2 \right), & d_r \leq d_b \\ 10 \log_{10} \left(\frac{1}{\alpha} \left(\sqrt{\frac{2\pi}{x_t w_r} \frac{4\pi d_t d_r^2}{\lambda d_b}} \right)^2 \right), & d_r > d_b \end{cases}$$

$$d_b = \frac{4h_t h_r}{\lambda} \text{ (BreakEven Dist.)} \quad \begin{matrix} h_t, h_r = \text{Tx, Rcv Height} \\ \alpha = \text{StreetParameter} \end{matrix}$$

The model takes the distance of transmitter and receiver to intersection center (d_t and d_r), receiver street width (w_r), and distance of transmitter to wall (x_t) as input. Last two values reflect building position influence. Adaption to differing streets is enabled by a street parameter (α). A higher loss is present at high rcv-distances (due to a diffraction, rather than reflection predominance), determined by a break even distance (d_b).

The geometric input parameters d_t, d_r, w_r, h_r, h_t , and x_t are given by the measured data. Therefore, we first fitted the path-loss exponent and α . As both are fitted globally, α represents a relative shift of the fitting curve.

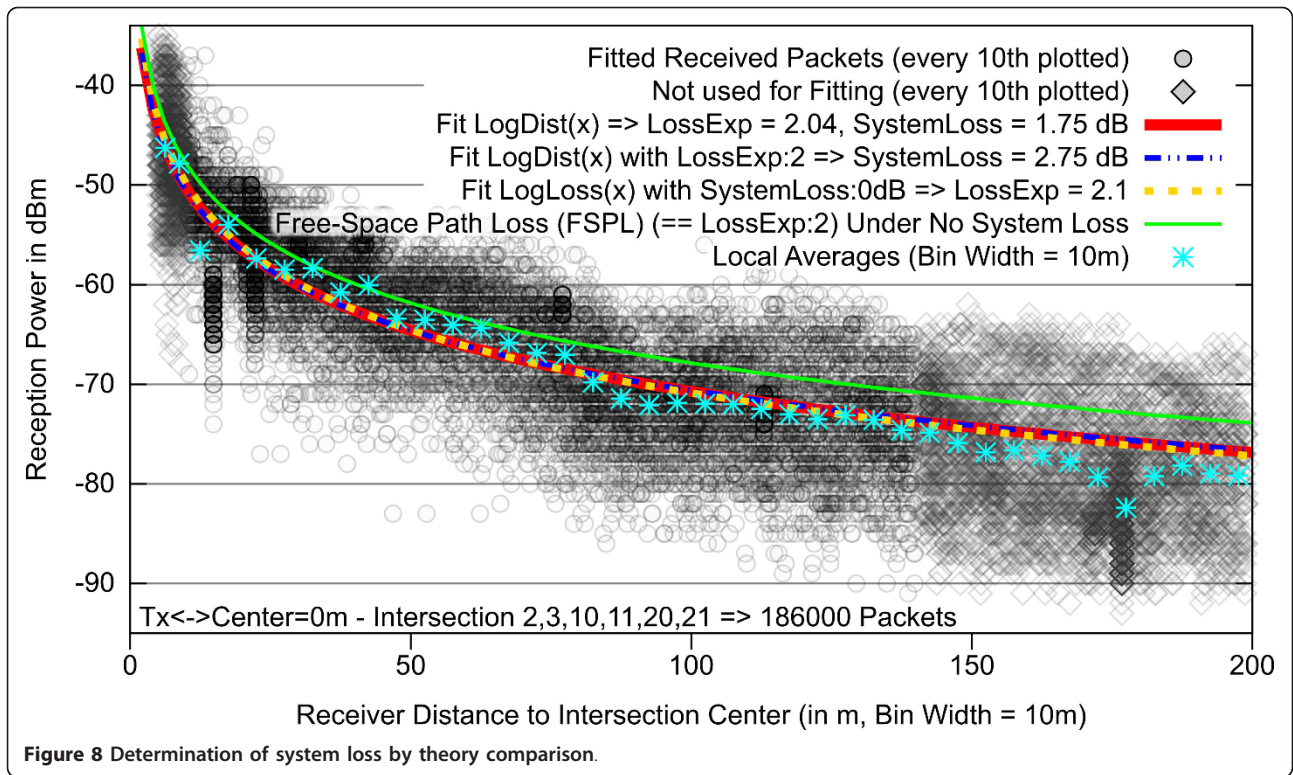


Figure 8 Determination of system loss by theory comparison.

These first fits showed that especially the influence of street width w_r is not properly reflected by the existing equation. Therefore, we replaced the fixed factor $\sqrt{2\pi}$ for $x_t w_r$ by a fittable exponent. As d_t also influences the “view” and therefore energy into a crossing street, it was also made fittable. A suburban loss factor was added to incorporate the observation of less power in suburban scenarios from Section 2.4. The following fit equation (in dBm) was found:

$$RxPower = P_{tx} - SystemLoss - PathLoss$$

$$PathLoss := VirtualSource11p(d_r, d_t, w_r, x_t, i_s) =$$

$$C + i_s L_{SU} + 10 \log_{10} \left(\left(\frac{d_r^{E_L}}{(x_t w_r)^{E_S}} \frac{4\pi d_r}{\lambda} \right)^{E_L} \right)$$

Fit Dimensions : $C = CurveShift$, $L_{SU} = SubUrbanLoss$,
 $E_L = LossExponent$, $E_S = StreetExp.$, $E_T = TxDistExp.$

Value i_s is specifying suburban ($i_s = 1$) or urban ($i_s = 0$). As $d_b \approx 180$ m for our setup, which exceeds the

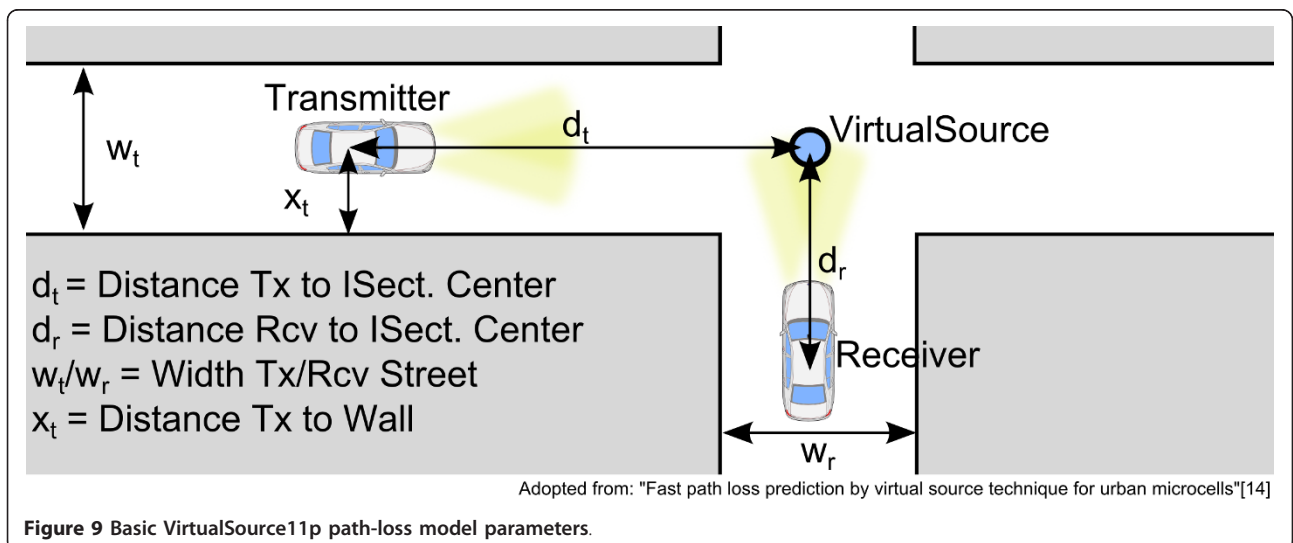
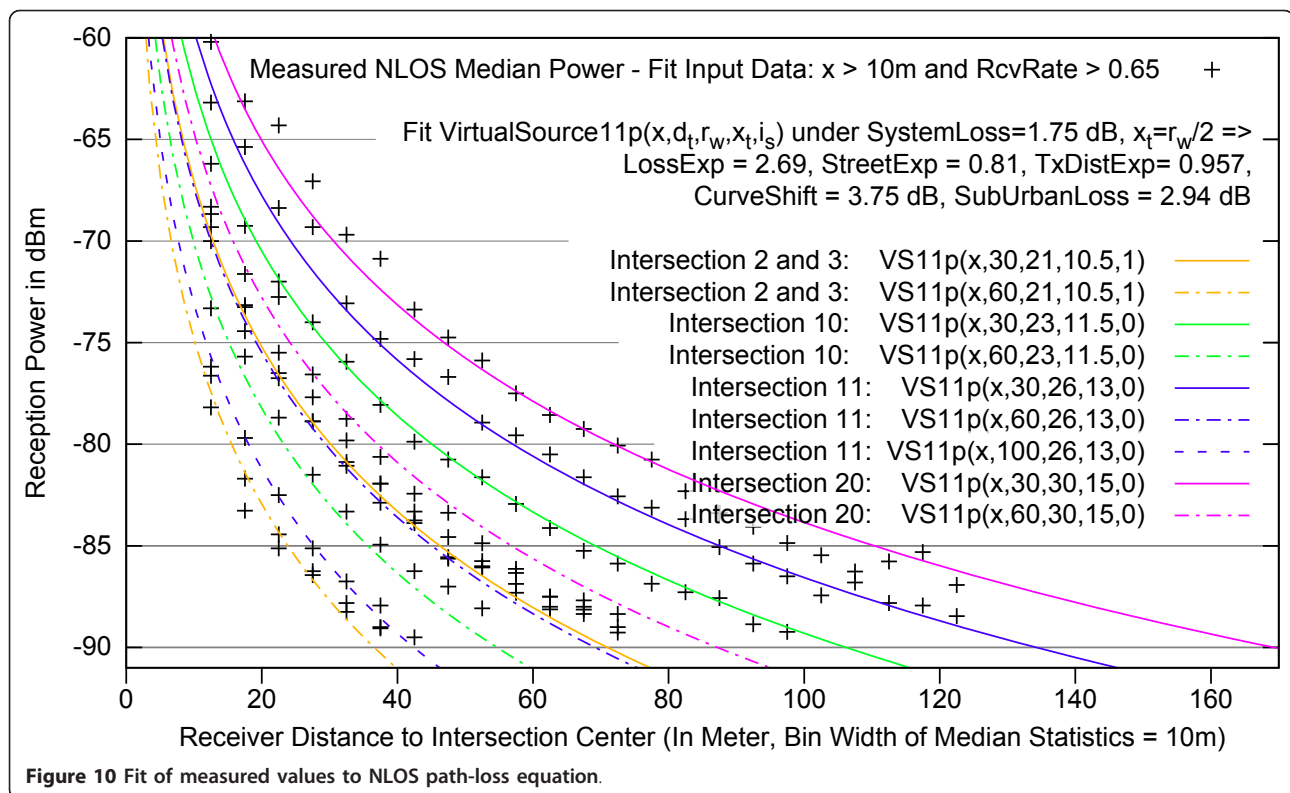


Figure 9 Basic VirtualSource11p path-loss model parameters.



highest distance d_r with reception, the $d_r > d_b$ equation is of no use for the fitting.

The final fit to determine the five variable parameters is visualized in Figure 10. We fitted the intersection wide average median reception power per tx-distance curve. Each of these curves abstracts (averages) eight measurements, thus providing a stable input to the fit and keeping the complexity on a moderate level. We showed in [17] that this averaging is viable, as the performance is very similar despite the transmitter being in the different side streets.

The fit input values are from the regular shaped ($\approx 90^\circ$ and $w_t \approx w_r$) intersections with buildings at each corner: intersection 2, 3, 10, 11, and 20, with w_r being 21, 21, 23, 26 and 30 m, respectively, and intersection 2 and 3 having $i_s = 1$. The fitted measurements have d_t values of 30 and 60 m (plus 100 m for intersection 11).

Each input value (visualized as a cross in Figure 10) is complemented by the reception rate in the bin and the intersection wide w_r and i_s values as input to the equation and for pre-selection. w_r is set to $\frac{w_t+w_r}{2}$ as w_t and w_r were selected similar per intersection and we fit intersection wide average values. System loss is set to 1.75 dB and $x_t = \frac{w_r}{2}$ (as this dimension was not tested).

We did not fit intersection 1, 9, and 21 due to differing reasons: intersection 1 was the very first tested intersection. Here, we measured with alternating

transmission power (20 dB, 10 dB) and rate (3 Mbps, 6 Mbps) in each second. In consequence, there are spatial gaps in the data for each of the four configurations, leading to empty bins at the 5 m bin width in the fit. Anyhow, the performance is very close to intersection 2 and 3, as shown in [17]. Intersection 9 has missing reflection facades. This dimension was not incorporated in the fit, as it would have complicated the fit by another dimension. Furthermore, we only tested one intersection of such type (as it is rare), leading to insufficient data to provide a reliable fit in this additional dimension. Intersection 21 was excluded due to two reasons: First, one of the street legs has a non 90° angle. Second, the inter-building distance in the two streets differs a lot (55 m against 30 m). It is questionable whether the averaging over the four side street simplification is applicable for this particular intersection. Despite the exclusion of these three intersections, the fit covers 11 data rows from five intersections, stemming from 88 test-runs.

We fitted the median reception power curve, as it is more stable at lower reception rates. The average reception power curve suffers (bin values are too high) from incomplete data as soon as the reception rate sinks below 1.0. The median is technically accurate as long as reception rate is greater than 0.5. However, due to small-scale fading leading to variations and potential

measurement inaccuracies around the reception threshold of the radio, median values also turn out to be slightly too high at reception rates close to 0.5. This is visible in plots. To prevent a negative influence on the fit, an exclusion criterion of reception-rate >0.65 was selected.

We also excluded small distances to center, as they are in LOS. The root mean square (RMS) error of the fit showed that $x > 10$ m is a good exclusion criterion: RMS error decreases from 2.4 with $x > 0$ to 0.8 with $x > 10$, but not much further with higher x . The very low RMS error of 0.8 corresponds to the good fit quality (compare input values to resulting model curves in Figure 10). The resulting VirtualSource11p path-loss equation, as determined by the fit, is

$$\text{VirtualSource11p}(d_r, d_t, w_r, x_t, i_s) = 3.75 + i_s 2.94 + \begin{cases} 10 \log_{10} \left(\left(\frac{d_t^{0.957}}{(x_t w_r)^{0.81}} \frac{4\pi d_t}{\lambda} \right)^{2.69} \right), & d_r \leq d_b \\ 10 \log_{10} \left(\left(\frac{d_t^{0.957}}{(x_t w_r)^{0.81}} \frac{4\pi d_r^2}{\lambda d_b} \right)^{2.69} \right), & d_r > d_b \end{cases}$$

Despite the no available measurement data for high d_r distances, an increased loss at high distances ($d_r > d_b$) due to diffraction rather than reflection being dominant is incorporated (as in [11,12,14,15]).

Note that close to intersection center, loss is really low (similar to FSPL having a heavy slope close to 0). Figure 11 depicts a representative example for intersection 10. In consequence, the VirtualSource11p path-loss equation only applies to NLOS conditions and not to the complete crossing street. At LOS on the crossing street, either the normal LOS path-loss should be used with distance as $d_t + d_r$ or a percental value between LOS at intersection center and NLOS value at the first point of NLOS. The latter is potentially more accurate.

3.3 NLOS small-scale fading classification

Small-scale fading leads to a distribution of power values around an average value. Figure 12 shows the different power probability distributions of received

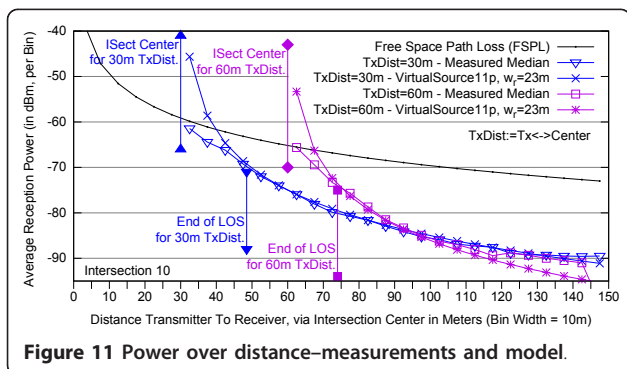


Figure 11 Power over distance—measurements and model.

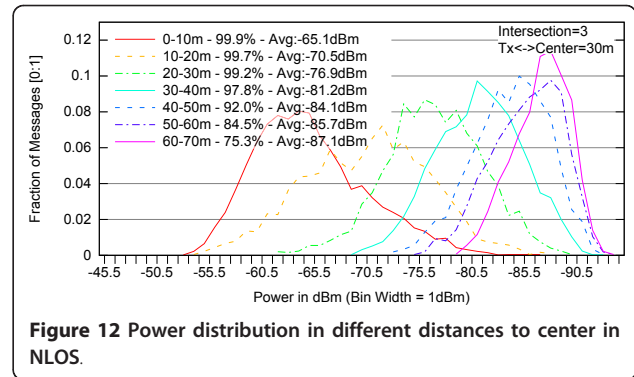


Figure 12 Power distribution in different distances to center in NLOS.

packets for different distances to the center in suburban intersection 3. It reveals e.g. a high variation in the 10-20 m bin as it includes LOS and NLOS conditions, or a variance limitation due to failed receptions (measurement limitation) for larger distances.

To determine fading in NLOS conditions, we centered the power probability distribution curves to their average and compared curves from different intersections for a certain street stretch in NLOS. Figure 13 shows the curves for $d_t = 30$ m and bin 40-50 m. This stretch exhibits NLOS conditions for all intersections and is in most cases (except probably intersection 2) not influenced by the radio reception limit. The curves from different intersections show a very similar shape. While they fit well to both, Nakagami-m and the Normal Distribution, the RMS error is slightly smaller for the Normal Distribution. Using visual judgment, they clearly match the Normal Distribution better. Therefore, we propose to model fading in NLOS as a normal distribution with $\sigma = 4.1$ dB.

For LOS conditions, we were able to verify that a 5.9 GHz vehicular channel is properly modeled by the often assumed Nakagami-m = 1 distribution. The corresponding fit is given in Figure 14. It reveals a good visible match to the Nakagami curve with fitted $m = 1.05$.

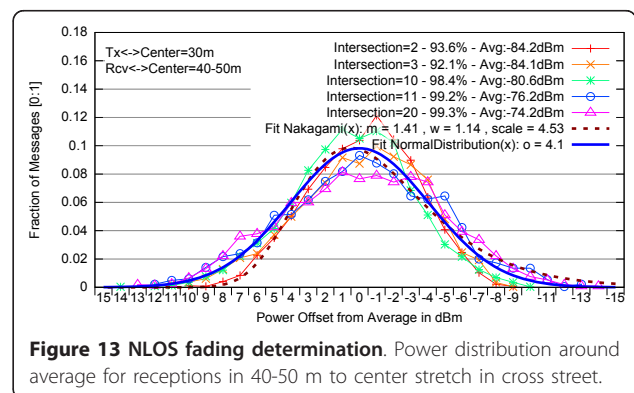
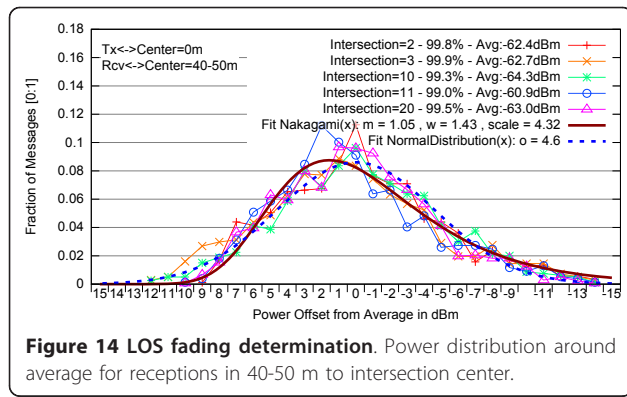


Figure 13 NLOS fading determination. Power distribution around average for receptions in 40-50 m to center stretch in cross street.



4 Path-loss model comparison

In this section, we compare our model to previously proposed NLOS models. One was measured under Car2X conditions at 790 MHz, the rest are micro-cellular models for urban street canyons where base stations are typically located inside street canyons, for example at signal lights. Note that we do not compare with the models in [9,10], as they do not take important influence factors such as street-width into account.

The models and their claimed validity (or verification setup) are given in Table 3. A model based on car to infrastructure measurements at 790 MHz was proposed in [22] by Toyota staff. Two different micro-cellular models are provided in the ITU-R P1411-5 recommendation for planning of short range communication systems [12]; one in section 4.2.3 for 0.8-2 GHz, the other in 4.2.4 for claimed 2-16 GHz and low height terminals, but receive street width w_r limited to <10 m. The second model seems to be based on [13]. Another urban micro-cell model stems from the WINNER II propagation measurement project [11]. It has been selected in the iTETRIS EU-project [23] to model urban 5.9 GHz V2V communication by adapting the frequency and transmitter height. At last, [15] is an analytical model based on propagation theory as used in ray-tracing. It

was used in a re-parameterized way in the CORNER 5.9 GHz V2V simulation implementation proposed in [8].

All models except the second ITU-R model are only specified for up to 2 GHz and $h_t > 4$ m. Also, those models are not based on measurements with vehicular ground plate antennas as in our measurements, except the receiver side in [22]. In iTETRIS, the model was not verified with the new environment parameters; in CORNER only very briefly. Our model was derived from measurements in 5 intersections and 11 loss-curves, based upon 88 test runs.

We implemented all path-loss equations in gnuplot to compare them against our inter-vehicle measurements and the proposed VirtualSource11p model. Figure 15 shows the resulting received power when configured same as our measurements in intersection 10 with $tx \leftrightarrow center$ distance $d_t = 30$ m. All models are configured with their proposed default configuration. It can be seen that the models selected in CORNER and iTETRIS differ to our measurements by 10 dB and more in critical NLOS areas. Only the P1411-5 2-16 GHz and the Toyota model come close to the measurements in this scenario. Note that the very low path-loss of the Toyota model in LOS is a strange behavior of their formula when parameterized to 5.9 GHz; at specified 0.79 GHz, the curve is just below FSPL.

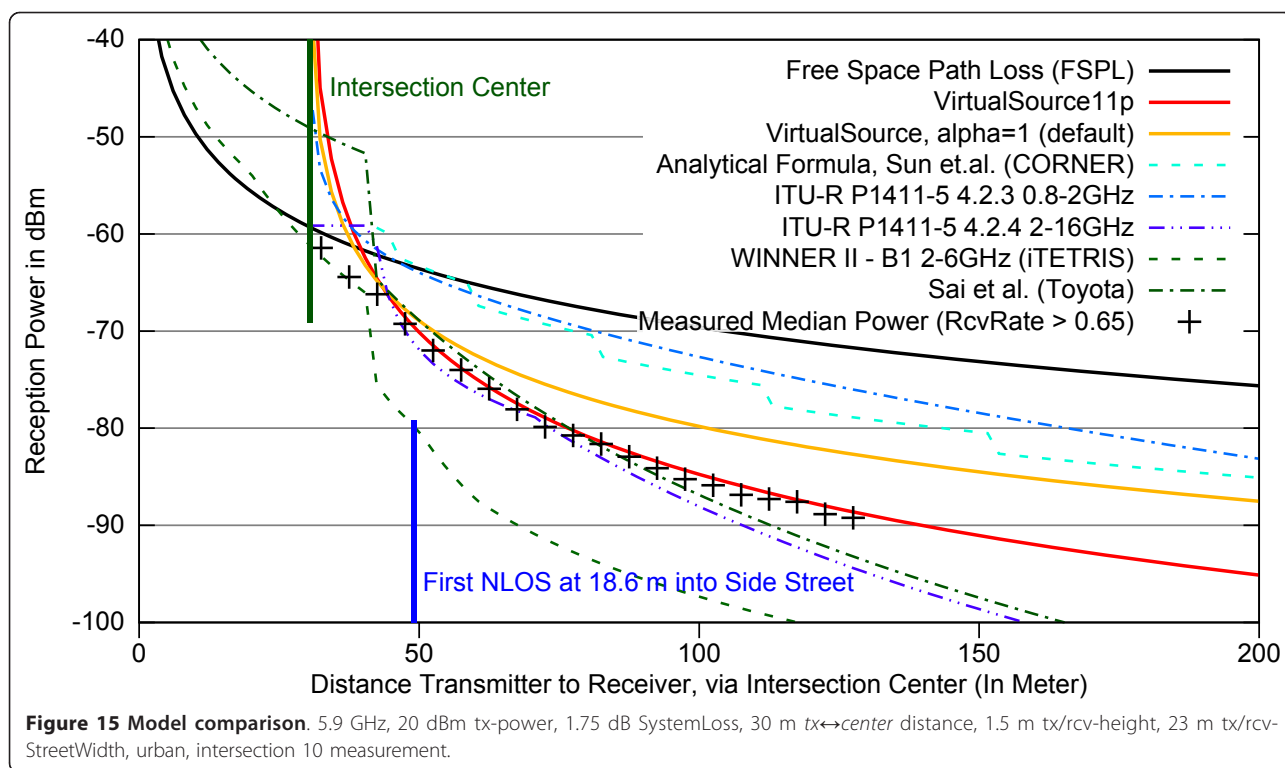
Obviously, the VirtualSource11p model shows a very good accordance to measurements due to its fitting based development. Comparisons to the other tested scenarios reveal similar results and show that our model especially follows changes in the street width better than the P1411-5 2-16GHz and the Toyota model.

5 Conclusion

With VirtualSource11p, we present a well validated, low complexity NLOS path-loss model for 5.9 GHz V2V communications at intersections. It was deduced from data collected in an extensive measurement campaign, specifically targeted to measure NLOS reception quality.

Table 3 Street canyon NLOS model comparison

	Claimed validity or verification (V)					
	Freq. GHz	Tx,Rcv height	Street width	Corner angle	# Verific. scenarios/measurem.	Vehicle G-plate antenna
VirtualSource11p	5.9 (V)	1.5, 1.5 m (V)	15-40 (V)	90° (V)	5/88	Yes
Virtual Source [14]	0.9/1.5 (V)	5-10, 2-3 m (V)	15-40 (V)	90° (V)	3/3	No
Sai et al. [22] (Toyota)	0.79 (V)	1.8-5 m, 1.9 m (V)	n.a.	90° (V)	1/>6	Only at Rcv.
ITU-R P1411-5 [12] (4.2,3,0.8-2 GHz)	0.8-2	4-50, 1-3 m	n.a.	n.a.	n.a.	n.a. (No)
ITU-R P1411-5 [12] (4.2,4,2-16 GHz)	2-16	4-50, 1-3 m	$w_r < 10$ m	90°	n.a. (3/3)	n.a. (No)
Winner II - B1 [11] (u. in iTETRIS [23])	2-6	10, 1.5 m (V)	n.a.	90°	n.a.	No
Sun et al. Paper [15] (u. in CORNER [8])	0.9/1.5/2.1 (V)	n.a.	10-40 m (V)	90° (V)	5/5	No



A founded selection of test intersections enabled to quantify the influence of inter-building distance and suburban/urban differences in a single path-loss equation with only a few simple geometric input-values.

Due to its fitting-based generation, the equation corresponds well to measurement data in different scenarios, especially with changing inter-building distances. In contrast, existing street canyon NLOS path-loss models for micro-cellular environments at lower frequencies differ mostly substantially from the measurements when parameterized to 5.9 GHz V2V communication. Of course, our model is also limited in its validity to the measurement environment it is based on. While we certainly did not cover every special case, we selected the test intersections as representative as possible, building upon an own building positioning investigation [3]. To the best of our knowledge, this is a novel approach and has not been done before.

In addition to path-loss, we investigated NLOS fading, finding a normal distributed power variation around average. In conclusion, this article provides a well-founded general framework to include NLOS propagation conditions into packet level simulations of 5.9 GHz inter-vehicle communication, thus enabling large-scale load simulations in intersection environments.

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Competing interests

The authors declare that they have no competing interests.

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