Hindawi Publishing Corporation EURASIP Journal on Wireless Communications and Networking Volume 2010, Article ID 618016, 14 pages doi:10.1155/2010/618016

Research Article

An Optimal Adaptive Network Coding Scheme for Minimizing Decoding Delay in Broadcast Erasure Channels

Parastoo Sadeghi, 1 Ramtin Shams, 1 and Danail Traskov 2

¹Research School of Information Sciences and Engineering, The Australian National University, Canberra ACT 0200, Australia

Correspondence should be addressed to Parastoo Sadeghi, parastoo.sadeghi@anu.edu.au

Received 31 August 2009; Accepted 3 March 2010

Academic Editor: Heung-No Lee

Copyright © 2010 Parastoo Sadeghi et al. This is an open access article distributed under the Creative Commons Attribution License, which permits unrestricted use, distribution, and reproduction in any medium, provided the original work is properly cited.

We are concerned with designing feedback-based adaptive network coding schemes with the aim of minimizing decoding delay in each transmission in packet-based erasure networks. We study systems where each packet brings new information to the destination regardless of its order and require the packets to be instantaneously decodable. We first formulate the decoding delay minimization problem as an integer linear program and then propose efficient algorithms for finding its optimal solution(s). We show that our problem formulation is applicable to memoryless erasures as well as Gilbert-Elliott erasures with memory. We then propose a number of heuristic algorithms with worst case linear execution complexity that can be used when an optimal solution cannot be found in a reasonable time. We verify the delay and speed performance of our techniques through numerical analysis. This analysis reveals that by taking channel memory into account in network coding decisions, one can considerably reduce decoding delays.

1. Introduction

In this paper, we are concerned with designing feedback-based adaptive network coding schemes that can deliver high throughputs and low decoding delays in packet erasure networks. We first present some background on existing work and emphasize that the notion of delay and the choice of a suitable network coding strategy are highly entangled with the underlying application.

1.1. Motivation and Background. Consider a broadcast packet-based transmission from one source to many destinations where erasures can occur in the links between the source and destinations. Two main throughput optimal schemes to deal with such erasures are fountain codes [1] and random linear network codes (RLNC) [2]. In the latter scheme, for example, the source transmits random linear mixtures of all the packets to be delivered. It is well-known that if the random coefficients are chosen from a finite field with a sufficiently large size, each coded packet will almost surely become linearly independent of all previously received

coded packets and hence, *innovative* for every destination [2]. The scheme is therefore almost surely throughput optimal. Another benefit of fountain codes and RLNC is that they do not require feedback about erasures in individual links in order to operate.

However in these schemes, throughput optimality comes at the cost of large decoding delays, as the receiver needs, in general, to collect all coded packets in a block before being able to decode. Despite this drawback, there are applications which are insensitive to such delays. Consider, for example, a simple software update (file download). The update only starts to work when the whole file is downloaded. In this case, the main desired properties are throughput optimality and the mean completion time and there is often little or no incentive to aim for partial "premature" decoding. The completion time performance of RLNC for rateless file download applications has been considered in [3]. In [3], the mean completion time of RLNC is shown to be much shorter than scheduling. Reference [4] considers time division duplex systems with large round-trip link latencies and proposes solutions for the number of coded packet

² Institute for Communications Engineering, Technische Universität München, D-80290 München, Germany

transmissions before waiting for acknowledgement on the received number of degrees of freedom.

There are applications where partial decoding can crucially influence the end user's experience. Consider, for example, broadcasting a continuous stream of video or audio in live or playback modes. Even though fountain codes and RLNC are throughput optimal, having to wait for the entire coded block to arrive can result in unacceptable delays in the application layer. But, we also note that partial decoding of packets out of their natural temporal order does not necessarily translate into low delivery delays desired by the application layer. The authors in [5, 6] have proposed feedback-based throughput-optimal schemes to deal with the transmitter queue size, as well as decoding and delivery delays at the destinations. When the traffic load approaches system capacity, their methods are shown to behave "gracefully" and meet the delay performance benchmark of singlereceiver automatic repeat request (ARQ) schemes.

There is yet another set of applications for which partial decoding is beneficial and can result in lower delays irrespective of the order in which packets are being decoded. Consider, for example, a wireless sensor network in which there is a fusion/command center together with numerous sensors/agents scattered in a region. Each sensor/agent has to execute or process one or more complex commands. Each command and its associated data is dispatched from the center in a packet. For coordination purposes, each agent needs to know its own and other agents' commands. Therefore, commands are broadcast to everyone in the network. In this application, in-order processing/execution of commands may not be a real issue. However, fast command execution may be crucial and therefore, it is imperative that innovative packets arrive and get decoded at the destinations as quickly as possible regardless of their order. As another example, consider emergency operations in a large geographical region where emergency-related updates of the map of the area need to be dispatched to all emergency crew members. In such situations too, updates of different parts of the map can be decoded in any order and still be useful for handling the emergency.

Finally, some applications may be designed in such a way that they are insensitive to in-order delivery. This can be particularly useful where the transport medium is unreliable. In such a case, it may be natural to use *multiple-description* source coding techniques [7], in which every decoded packet brings new information to the destination, irrespective of its order. In light of the emergency applications described above, one can perform multiple-description coding for map updates, so that updates of different subregions can be divided into multiple packets and each packet can provide an improved view of one region in a truly order-insensitive fashion.

1.2. Contributions. In this paper, we are inspired by the last set of order-insensitive packet delivery applications and hence, focus on designing network coding schemes that, with the help of feedback, can deliver innovative packets in any order to the destination and also guarantee fast

decoding of such packets. As a first step towards such goal, we limit ourselves to broadcast erasure channels, but emphasize that the ideas can be extended to other more complicated scenarios. We also consider the class of instantaneously decodable network coding schemes, in which each coded transmission contains at most one new source packet that a receiver has not decoded yet. The rationale is that in an orderinsensitive application, any innovative packet that cannot be decoded immediately incurs a unit of delay. Obviously, one other source of delay is when a coded packet does not contain any new information for a receiver and hence, is not innovative. A similar definition of the decoding delay was first considered in [8], where the authors presented a number of heuristic algorithms to reduce order-insensitive decoding delay. In this context, our main contributions are the following.

- (i) In Section 1.1, we have motivated the problem in light of possible applications in sensor and ad hoc networks. To the best of our knowledge, such application-dependent classification of network coding delays did not previously exist in the literature.
- (ii) In Section 3.1, we present a systematic framework for the minimization of decoding delay in each transmission subject to the instantaneous decodability constraint. We show that this problem can be cast into a special integer linear programming (ILP) framework, where instantaneously decodable packet transmission corresponds to a *set packing* problem [9] on an appropriately defined set structure.
- (iii) In Section 3.2, we provide a customized and efficient method for finding the optimal solution to the set packing problem (which is in general NP-hard). Our numerical results in Section 6 show that for reasonably sized number of receivers, the optimum solution(s) can be found in a time that is linearly proportional to the total number of packets.
- (iv) In Section 4, we discuss decoding delay minimization for an important class of erasure channels with memory, which can occur in wireless communication systems due to deep fades and shadowing [10]. We show that the general set packing framework in Section 3 can be easily modified to account for the erasure memory. Our results in Section 6 reveal that by adapting network coding decisions based on channel erasure conditions, significant improvements in delay are possible compared to when decisions are taken irrespective of channel states.
- (v) In Section 5, we provide a number of heuristic variations of the optimal search for finding (possibly suboptimal) solutions faster, if needed. Our results in Section 6 show that such heuristics work very well and often provide solutions that are very close to the search algorithm. Moreover, they improve on the proposed random opportunistic method in [8].

2. Network Model

Consider a single source that wants to broadcast some data to N receivers, denoted by R_i for i = 1, ..., N. The data to be broadcast is divided into K packets, denoted by m_j for j = 1, ..., K. Time is slotted and the source can transmit one (possibly coded) packet per slot.

A packet erasure link L_i connects the source to each individual receiver R_i . Erasures in different links can be independent or correlated with each other. Different erasures in a single link can be independent (memoryless) or correlated with each other (with memory) over time.

For memoryless erasures, an erasure in link L_i can occur with a probability of $p_{e,i}$ in each packet transmission round independent of previous erasures.

For correlated erasures, we consider the well-known Gilbert-Elliott channel (GEC) [11], which is a Markov model with a good and a bad state. If the channel is in the good state, packets can be successfully received, while in the bad state packets are lost (e.g., due to deep fades or shadowing in the channel). The probability of moving from the good state G to the bad state B in link L_i is $b_i \triangleq \Pr(C_{i,\ell} = B \mid$ $C_{i,\ell-1} = G$) and the probability of moving from the bad state B to the good state G is $g_i \triangleq \Pr(C_{i,\ell} = G \mid C_{i,\ell-1} = B)$, where ℓ is the time slot index. Steady-state probabilities are given by $P_{G,i} \triangleq \Pr(C_i = G) = g_i/(b_i + g_i)$ and $P_{B,i} \triangleq$ $Pr(C_i = B) = b_i/(b_i + g_i)$. Following [12], we define the memory content of the GEC in link L_i as $0 \le \mu_i = 1$ $b_i - g_i < 1$, which signifies the persistence of the channel in remaining in the same state. A small μ means a channel with little memory and a large μ means a channel with large memory.

Before transmission of the next packet, the source collects error-free and delay-free 1-bit feedback from each destination indicating if the packet was successfully received or not. A successful reception generates an acknowledgement (ACK) and an erasure generates a negative acknowledgement (NAK). This feedback is used for optimizing network coding decisions at the source for the next packet transmission round, as described in future sections.

In this work, we consider linear network coding [2] in which coded packets are formed by taking linear combinations of the original source packets. Packets are vectors of fixed size over a finite field \mathbb{F}_q . The coefficient vector used for linear network coding is sent in the packet header so that each destination can at some point recover the original packets. Since in this paper we are only dealing with instantaneously decodable packet transmission, it suffices to consider linear network coding over \mathbb{F}_2 . That is, coded packets are formed using binary XOR of the original source packets. Thus, network coding is performed in a similar manner as in [13].

Definition 1. A transmitted packet is instantaneously decodable for receiver R_i if it is a linear combination of source packets containing at most one source packet that R_i has not decoded yet. A scheme is called instantaneously decodable if all transmissions have this property for all receivers.

Definition 2. At the end of transmission round ℓ in an instantaneously decodable scheme, the knowledge of receiver R_i is the set consisting of all packets that the receiver has decoded so far. The receiver can therefore, compute any linear combination of the packets that it has decoded for decoding future packets.

Definition 3. In an instantaneously decodable scheme, a coded packet is called non-innovative for receiver R_i if it only contains source packets that the receiver has decoded so far. Otherwise, the packet is innovative.

Definition 4. A scheme is called rate or throughput optimal if all transmissions are innovative for the entire set of receivers.

Definition 5. In time slot ℓ , receiver R_i experiences one unit of delay if it successfully receives a packet that is either non-innovative or not instantaneously decodable. If we impose instantaneous decodability on the scheme, a delay can only occur if the received packet is not innovative.

Note that in the last definition, we do not count channel inflicted delays due to erasures. The delay only counts "algorithmic" overhead delays when we are not able to provide innovative and instantaneously decodable packets to a receiver.

As an example, if the knowledge of R_1 is $\{m_1, m_2, m_3\}$, receiving $m_1 \oplus m_2$ will cause R_1 to experience one unit of delay, whereas $m_1 \oplus m_2 \oplus m_5$ is innovative and instantaneously decodable, hence does not incur any delay.

We note that a packet that is not transmitted yet or transmitted but not received by any receiver can be transmitted in an uncoded manner at any transmission slot without incurring any algorithmic delay. In fact, this is how the transmission starts: by sending m_1 uncoded, for example.

A zero-delay scheme would require all packets to be both innovative and instantaneously decodable to all receivers. Thus zero-delay implies rate optimality, but not vice versa. As the authors show in [8, Theorem 1] for the case of N=2 and N=3 receivers, there exists an *offline* algorithm that is both rate optimal and delay-free. For $N \ge 4$ the authors prove that a zero-delay algorithm does not exist. By offline we mean that the algorithm needs to know future realizations of erasures in broadcast links. In contrast, an *online* algorithm decides on what to send in the next time slot based on the information received in the past and in the current slot. In this paper, we focus on designing online algorithms.

3. Optimization Framework

3.1. Problem Formulation Based on Integer Linear Programming. Instantaneous decodability can be naturally cast into the framework of integer optimization. To this end, let us fix the packet transmission round to ℓ and consider the knowledge of all receivers, which is also available at the source because of the feedback. The state of the entire system at time index ℓ (in terms of packets that are still needed by

the receivers) can be described by an $N \times K$ binary *receiver*-packet incidence matrix A with elements

$$a_{ij} = \begin{cases} 1 & \text{if } R_i \text{ needs } m_j, \\ 0 & \text{otherwise.} \end{cases}$$
 (1)

Columns of matrix A are denoted by \mathbf{a}_1 to \mathbf{a}_K . We assume that packets received by all receivers are removed from the receiver-packet incidence matrix. Hence, A does not contain any all-zero columns.

Example 1. Consider N=2 receivers and K=3 packets. Before the transmission begins, the receiver-packet incidence matrix A is an all-one 2×3 matrix. If we send packet m_1 in the first transmission round $\ell=1$ and assuming that only receiver R_2 successfully receives it, A will become

$$A = \begin{bmatrix} 1 & 1 & 1 \\ 0 & 1 & 1 \end{bmatrix}. \tag{2}$$

If we send packet m_2 in the next transmission round $\ell = 2$ and assuming that only receiver R_1 successfully receives it, A will then be

$$A = \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 1 \end{bmatrix}. \tag{3}$$

The condition of instantaneous decodability means that at any transmission round we cannot choose more than one packet which is still unknown to a receiver R_i . In the example above, at $\ell = 3$, we cannot send $m_1 \oplus m_3$ because it contains more than one packet unknown to R_1 .

Let \mathbf{x} represent a binary decision vector of length Kthat determines which packets are being coded together. The transmitted packet consists of the binary XOR of the source packets for which $x_i = 1$. More formally, we can define the instantaneous decodability constraint for all receivers as $A\mathbf{x} \leq \mathbf{1}_N$, where $\mathbf{1}_N$ represents an all-one vector of length N and the inequality is examined on an element-by-element basis (Note that although x is a binary or Boolean vector, $A\mathbf{x}$ is calculated in real domain. Hence, $A\mathbf{x} \leq \mathbf{1}_N$ is in fact a pseudo-Boolean constraint.). This condition ensures that a transmitted coded packet contains at most one unknown source packet for each receiver. A vector **x** is called *infeasible* if it does not satisfy the instantaneous decodability condition. In other words, x is called *infeasible* if and only if there exists at least one p for which $b_p > 1$ in $A\mathbf{x} = \mathbf{b} =$ $[b_1, \ldots, b_p, \ldots, b_N]^T$. A vector **x** is called a *solution* if and only if it satisfies $A\mathbf{x} \leq \mathbf{1}_N$. In the rest of this paper, " $A\mathbf{x} \leq \mathbf{1}_N$ " and "x is a solution" are used interchangeably.

Now consider sets $M_1, \ldots, M_K \subset \{R_1, \ldots, R_N\}$, where M_j is the nonempty set of receivers that still *need* source packet m_j . Note that these sets can be easily determined by looking at the columns of matrix A. The "importance" of packet m_j can be, for example, taken to be the size of set M_j , which is the number of receivers that still need m_j .

We now formally describe the optimization procedure that should be performed at the transmitter. Maximizing the number of receivers for which a transmission is innovative, subject to the constraint of instantaneous decodability, can be posed as the following (binary-valued) integer linear program (ILP):

max
$$\mathbf{w}^T \mathbf{x}$$

subject to $A\mathbf{x} \le \mathbf{1}_N$, $\mathbf{x} \in \{0, 1\}^K$,

where $\mathbf{w}^T = (|M_1|, \dots, |M_K|)$. This is a standard problem in combinatorial optimization, usually called *set packing* [9]. Here the universe is the set of all receivers and we need to find disjoint (due to instantaneous decodability condition) subsets M_j with the largest total size. In the (most desirable) case when equality holds in $A\mathbf{x} \leq \mathbf{1}_N$ for every receiver, we also speak of a *set partition*. This is equivalent to a zero-delay transmission.

In Section 4, we will consider other measures of packet importance and discuss the role of **w** in tailoring the optimization problem according to the application requirements or channel conditions, such as memory in erasure links.

We assume that elements of \mathbf{w} , which signify packet importance, are all positive. If one has already found a solution such as $\mathbf{x}_1 = [x_1, \dots, x_{p-1}, 1, x_{p+1}, \dots, x_K]$ with $\mathbf{w}^T \mathbf{x}_1 = v_1$, then changing this solution into $\mathbf{x}_0 = [x_1, \dots, x_{p-1}, 0, x_{p+1}, \dots, x_K]$ by changing $x_p = 1$ into $x_p = 0$ can only result in a $\mathbf{w}^T \mathbf{x}_0 = v_0$ strictly smaller than v_1 . We say that *given* solution \mathbf{x}_1 , \mathbf{x}_0 is clearly *suboptimal* and hence, can be discarded in an algorithm that searches for the optimal solution(s).

3.2. Efficient Search Methods for Finding the Optimal Solution of (4). It is well known that the set packing problem is NP-hard [9]. Here, we present an efficient ILP solver designed to take advantage of the specific problem structure. Later, we will see that for many practical situations of interest, our method performs well empirically. Based on this framework, we will also present some heuristics in Section 5 to deal with more complicated and time-consuming problem instances.

We begin presenting our method by first defining constrained and unconstrained variables.

Definition 6. Two binary-valued variables are said to be constrained if they cannot be simultaneously 1 in a solution. Or formally, x_i and x_j are constrained if for any \mathbf{x} satisfying $A\mathbf{x} \leq \mathbf{1}_N$, $x_i + x_j \leq 1$ (Again, note that the addition of variables takes place in real domain.). We also say that x_j is constrained to x_i and vice versa. It can be proven that x_i and x_j are constrained if and only if there exits at least one row index p in A for which $a_{pi} = a_{pj} = 1$.

Definition 7. The set of all variables constrained to x_i is called the *constrained set* of x_i and is denoted by C_i . That is,

$$C_i = \left\{ x_j \mid j \neq i, A\mathbf{x} \le \mathbf{1}_N \Longrightarrow x_i + x_j \le 1 \right\}. \tag{5}$$

If x_i and x_j are not constrained to each other $(x_i \notin C_j$ and $x_j \notin C_i)$, then columns \mathbf{a}_i and \mathbf{a}_j in A cannot have nonzero elements in the same row position. That is, for each row index p, $a_{pi} = 1 \Rightarrow a_{pj} = 0$ and $a_{pj} = 1 \Rightarrow a_{pi} = 0$.

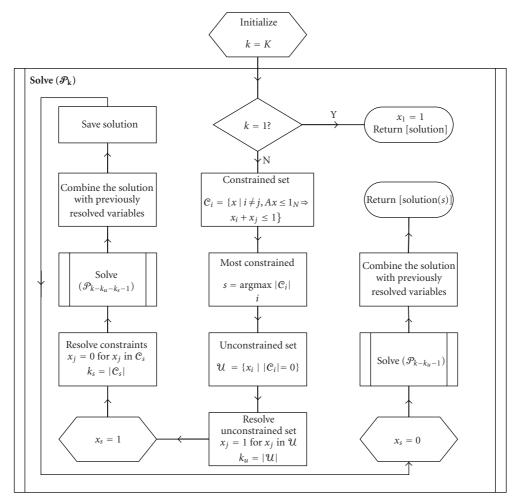


Figure 1: A schematic of Algorithm 1 with greedy pruning for finding the optimal network coding solution of (4). Note that the algorithm is recursive as it calls $\mathcal{P}_{k-k_u-k_s-1}$ and \mathcal{P}_{k-k_u-1} within itself.

Definition 8. A variable x_i is said to be unconstrained if $C_i = \emptyset$. The set of all unconstrained variables is denoted by \mathcal{U} and is referred to as the unconstrained set.

If x_i is an unconstrained variable, then for each row index p, $a_{pi} = 1 \Rightarrow a_{pj} = 0$ for all $j \neq i$ (otherwise, x_i and x_j would become constrained).

Example 2. Consider the following receiver-packet incidence matrix A

$$A = \begin{bmatrix} 1 & 0 & 1 & 0 & 0 & 0 \\ 1 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix}. \tag{6}$$

One can easily verify the relations defined above. For example, variables x_1 and x_3 are constrained because for p = 1, $a_{p1} = a_{p3} = 1$. Variables x_1 and x_4 are not constrained to each other because columns \mathbf{a}_1 and \mathbf{a}_4 do not

have a nonzero element in the same row position. Variable x_6 is unconstrained because no other column has a nonzero element in rows 6 or 7. In summary, $\mathcal{C}_1 = \{x_2, x_3\}$, $\mathcal{C}_2 = \{x_1\}$, $\mathcal{C}_3 = \{x_1, x_4\}$, $\mathcal{C}_4 = \{x_3\}$ and $\mathcal{C}_5 = \mathcal{C}_6 = \emptyset$.

To design an efficient search algorithm, one needs to efficiently prune the parameter space and reduce the problem size. We make the following observations for pruning of the parameter space.

- (1) Unconstrained variables must be set to 1. In other words, setting those variables to 0 does not contribute to the optimal solution (note that the elements in \mathbf{w} are positive). In the above example, x_5 and x_6 must be set to 1 because no other variable is constrained to them (we will make this statement formal in the optimality proof of the algorithm in the appendix).
- (2) If a constrained variable is set to 1 all members of its constrained set must be set to 0. In the above example, setting $x_1 = 1$ forces x_2 and x_3 to zero.

(3) At a given step, the parameter space can be pruned most by resolving the variable with the largest constrained set.

Application of the third observation, in a search algorithm results in greedy pruning of the parameter space. We note that greedy pruning is only optimal for a given step of the algorithm and is not guaranteed to result in the optimal reduction of the overall complexity of the search.

We now make a final remark before presenting the search algorithm. In particular, we have observed that finding constrained sets for each variable in each step of the algorithm can be somewhat time consuming. A very effective alternative is to first sort matrix A, columnwise, in descending order of the number of 1's in each column. Setting the "most important" head variable x_1 (with the highest $|M_1|$) to 1 is likely to result in the largest constrained set (because it potentially overlaps with many other variables) and hence, many variables will be resolved in the next recursion. We will refer to the approach based on finding the largest constrained set as the *greedy pruning* strategy and to the alterative approach as the *sorted pruning* search strategy.

The greedy pruning search strategy is shown in Figure 1, which with appropriate modifications can also represent the sorted pruning variation. Let \mathcal{P}_k denote the problem of size k whose input is an $N \times k$ receiver-packet incidence matrix A_k and whose output is a set of solutions of the form \mathbf{x} of length k which satisfy the instantaneous decodability condition $A_k\mathbf{x} \leq \mathbf{1}_N$. The algorithms can be described as shown in Algorithm 1.

In the appendix, we prove by structural induction that Algorithm 1 is guaranteed to return all optimal solutions of (4). However, we note that not every solution returned by Algorithm 1 is optimal. The nonoptimal solutions can be easily discarded by testing against the objective function (4) at the end of the algorithm. We also note that in Algorithm 1, we can simply remove those packets received by every receiver from the problem. If there are K_0 such variables, we can start step (1) above from $k = K - K_0$ instead of K. The Matlab code for both the greedy and sorted pruning algorithms can be found at http://users.rsise.anu.edu.au/~parastoo/netcod/.

We conclude this section by a brief note on the computational complexity of Algorithm 1. Let us denote the number of recursions required to solve the problem of size k by \mathfrak{C}_k . According to Algorithm 1, this problem is always broken into two smaller problems of size $k-k_u-k_s-1$ and $k-k_u-1$. Therefore, one can find the number of recursions required to solve \mathcal{P}_k by recursively computing $\mathfrak{C}_k = \mathfrak{C}_{k-k_u-k_s-1} + \mathfrak{C}_{k-k_u-1}$. The recursion stops when one reaches a problem of size 1 (only one packet to transmit) where $\mathfrak{C}_1 = 1$.

4. Adaptive Network Coding in the Presence of Erasure Memory

Here, we present a generalization of the set packing approach for coded transmission in erasure channels with memory. The idea is that the importance of a packet m_j is no

longer determined by how many receivers need m_j , but by the probability that m_j will be successfully decoded by the receivers that need it. In computing this probability, one can use the fact that successive channel erasures in a link are usually correlated with each other and hence, their history can be used to make predictions about whether a receiver is going to experience erasure or not in the next time slot. To present the idea, we focus on the GEC model for representing channel erasures. More general memory models for erasure can also be incorporated into our framework.

We define the reward p_i of sending a packet to receiver R_i as the probability of successful reception by R_i in the next time slot: $p_i = \Pr(C_{i,\ell} = G \mid C_{i,\ell-1})$, where $C_{i,\ell-1}$ is the state of R_i in the previous transmission round (Statements like "state of R_i " should be interpreted as the state of the physical link L_i connecting the source to R_i .). The total reward or importance of sending packet m_i is then

$$w_j = \sum_{i \in M_j} p_i. \tag{7}$$

The above weight vector gives higher priority to a packet m_j for which there is a higher chance of successful reception, because the receivers that need m_j are more likely to be in good state in the next time slot. With this newly defined weight vector, one can try to solve the optimization problem given in (4) under the same instantaneous decodability condition.

Remark 1. We conclude this section by emphasizing that the optimization framework in (4) is very flexible in accommodating other possibilities for the weight vector w, which can be appropriately determined based on the application. For example, instead of allocating the same weight to a packet needed by a subset of receivers, one can allocate different weights to the same packet (looking column-wise at A) depending on the priorities or demands of each user. In the map update example described in the Introduction, different emergency units can adaptively flag to the base station different parts of the map as more or less important depending on their distance from a certain disaster zone. The task of the base station is then to send a packet combination that satisfies the largest total priority. One can also combine user-dependent packet weights with the channel state prediction outcomes in a GEC. One possibility is to multiply the probabilities p_i by the receiver priority. It could then turn out that although a receiver is more likely to be in erasure in the next transmission round, it may be served because of a high priority request.

5. Heuristic Search Algorithms

In Section 3.2, we proposed efficient search algorithms for finding the optimal solution(s) of (4). However, there may be situations where one would like to obtain a (possibly suboptimal) solution much more quickly. This may be the case, for example, when the total number of packets to be transmitted is very large. Therefore, designing efficient heuristic algorithms to complement the optimal search is

```
(1) Start with the original problem of size k = K.
(2) if sorted pruning strategy is desired then
(3) Rearrange the variables in A_k in descending order of packet importance (number of 1's in each column).
(4) end if
(5) Solve (\mathcal{P}_k):
(6) if k = 1 then
     Return x_1 = 1 (since the variable is not constrained).
(7)
(8) else
(9)
       if greedy pruning strategy is desired then
(10)
           Determine the constrained set for all variables x_1 to x_k.
(11)
           Denote the index of the variable with the largest constrained set by s and the cardinality of its constrained
(12)
        else
          Determine the constrained set for the head variable x_1 with cardinality k_1 and also the set of unconstrained
(13)
          variables (Note that we have overused index 1 to refer to the head variable in the reordered matrix at each
(14)
        Denote the cardinality of the unconstrained set \mathcal{U} by k_u.
(15)
        Set all the unconstrained variables to 1.
(16)
        Set x_s = 1 and the variables in its corresponding constrained set C_s to 0.
(17)
(18)
        Reduce the problem by removing resolved variables. Reduce A_k accordingly.
       Solve (\mathcal{P}_{k-k_u-k_s-1}) (Note that k_u unconstrained variables are set to one, x_s = 1 and k_s variables constrained by x_s
(19)
         are set to zero, hence a total of k_s + k_u + 1 variables are resolved.).
(20)
       Combine the solution with previously resolved variables. Save solution.
(21)
(22)
       Reduce the problem by removing resolved variables. Reduce A_k accordingly.
       Solve (\mathcal{P}_{k-k_u-1}) (Note that k_u unconstrained variables are set to one and x_s = 0, hence a total of k_u + 1 variables
(23)
        are resolved.).
(24)
       Combine the solution with previously resolved variables. Return solution(s).
(25) end if
```

Algorithm 1: Recursive search for the optimal solution(s) of (4).

important. In this section, we propose a number of such heuristics.

5.1. Heuristic 1—Weight Sorted Heuristic Algorithm. The idea behind this recursive algorithm is very simple. As in Algorithm 1, we start with the original problem of size k = K. We then rearrange the columns of the matrix A in descending order of $|w_i|$ (starting from the packet with the highest weight). Note that this is different from the sorted pruning version of the Algorithm 1, in which the columns of A were sorted in descending order of $|M_i|$ to potentially result in large constrained sets. We then set the head variable $x_1 = 1$ and find its corresponding constrained set \mathcal{C}_1 to resolve $k_1 = |\mathcal{C}_1|$ variables that are to be set to zero. We then solve the smaller problem of size \mathcal{P}_{k-k_1} and continue until the problem cannot be further reduced. One main difference between Heuristic 1 and Algorithm 1 is that at each recursion, the head variable is only set to one; the other possibility of $x_1 = 0$ is not pursued at all. In a sense, this heuristic algorithm finds greedy solutions to the problem at each recursion by serving the highest priority packet. In this heuristic algorithm, all k_u unconstrained variables are naturally set to 1 in the course of the algorithm. The computational complexity of this method is at worst proportional to K, which can happen when there is no constraint between packets.

5.2. Heuristic 2—Search Algorithm 1 with Maximum Recursions/Elapsed Time. It is possible to terminate the recursive search Algorithm 1 prematurely once it reaches a maximum number of allowed recursions/elapsed time. If the algorithm reaches this value and the search is not complete, it performs a termination procedure whereby it heuristically resolves the remaining unresolved packets in the current incomplete solution. That is, it performs Heuristic 1 on a smaller problem, which is yet to be solved. It then returns the best solution that has been found so far. We note that due the extra termination procedure, the actual number of recursions/elapsed time can be (slightly) higher than the preset value.

Two comments are in order here. Firstly, Algorithm 1 is designed to sort the matrix *A* based on the number of receivers that need a packet. It only reverts to sorting the unresolved variables based on the vector **w** in the termination process. Secondly, if the maximum number of recursions is set to one, Algorithm 1 just performs the termination process and becomes identical to Heuristic 1.

5.3. Heuristic 3—Dynamic Number of Recursions. This heuristic is based on Heuristic 2, where we dynamically increase the number of allowed recursions as needed. At each transmission round, we start with only one allowed

recursion (effectively run Heuristic 1). If the throughput (Let $\mathcal{Q} \subset \{1,\dots,N\}$ denote the index of receivers that still need at least one packet and $\mathcal{R}_{\mathcal{Q}}$ denote such receivers. The achieved throughput at time slot ℓ is defined as $\mathbf{w}^T\mathbf{x}/f(\mathcal{R}_{\mathcal{Q}})$, where \mathbf{x} is the found solution and $f(\mathcal{R}_{\mathcal{Q}})$ is an appropriate function of receivers' needs. For memoryless erasures $f(\mathcal{R}_{\mathcal{Q}}) = |\mathcal{R}_{\mathcal{Q}}|$ and for GEC's $f(\mathcal{R}_{\mathcal{Q}}) = \sum_{q \in \mathcal{Q}} |p_q|$ (refer to Section 4 and (7)).) is higher than a desired value, there is no need to proceed any further. Otherwise, we can gradually increase the number of recursions by an appropriate step size. This heuristic stops when it either reaches the maximum allowed recursions or when increasing the number of recursions does not result in a noticeable improvement in the throughput.

6. Numerical Results and Secondary Coding Considerations

We start this section by presenting *end-to-end* decoding delay results for memoryless erasure channels. We then specialize to erasure channels with memory. The end-to-end problem is the complete transmission of *K* packets. End-to-end decoding delay of a receiver is the sum of decoding delays for the receiver in each transmission step. In the following, when we say "the delay performance of method X", we are referring to the delay performance of the end-to-end transmission, where method X is applied at each step.

In the course of presenting the results and based on the observed trends, we will discuss some secondary coding techniques and post processing considerations that can improve the decoding delay. Throughout the analysis of this section, we assume independent erasures in different links with identical probabilities. Hence, we can drop subscript i when referring to link erasure probabilities.

Figure 2 shows the median of decoding delay for the transmission of K = 100 packets to N = 3 to N =100 receivers. Channel erasures are memoryless and occur with a high probability of p = 0.5 independently in every link. The median of delay is computed across all receivers and is, in fact, also the median across many stochastic runs of the algorithms. The first curve from below shows the delay obtained from Algorithm 1 (Throughout the numerical evaluations, we used the sorted pruning version of Algorithm 1.). The middle curve is the delay obtained by performing Heuristic 1. The top curve shows a reproduction of delay results reported in [8] which are based on a random opportunistic instantaneous network coding strategy. In this case, the transmitter first selects a packet needed by at least one receiver at random. Then, it goes over other packets in some order and adds a packet to the current choice only if their addition still results in instantaneous decodability. In comparison, Heuristic 1 performs noticeably better than that in [8] and more importantly, is not much far away from the results of Algorithm 1. This is specially important since for some number of receivers, Heuristic 1 can run considerably faster than Algorithm 1, which will be shown in the coming figures shortly.

Figure 3 compares the mean delay performance of different heuristics presented in Section 5 with that of

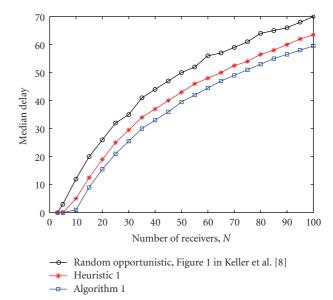


FIGURE 2: Median of decoding delay for the transmission of K=100 packets to N=3 to N=100 receivers. Channel erasures are memoryless and occur with a high probability of p=0.5 independently in every link. Algorithm 1, Heuristic 1 and random Heuristic [8] are compared with each other.

Algorithm 1. Similar to the previous figure, mean delay is computed across all receivers. The delay performance of Heuristic 2, Heuristic 3, and Algorithm 1 are close, whereas Heuristic 1 results in the largest delay. A careful reader may notice that the end-to-end performance of Heuristic 2 is at times better than Algorithm 1. While the difference is practically insignificant, this deserves some explanation. The end-to-end transmission problem involves making packet transmission decisions at each step. While all algorithms start with the same packet incidence matrix (all-ones), due to packet erasures and as they make decisions about transmission of packets at each step, they take diverging paths in the solution space. As a result, they end up with different packet incidence matrices to solve over time. Hence, it is conceivable for an algorithm to make suboptimal decisions at one or more steps and yet end up with a better end-to-end delay than Algorithm 1 that strictly makes optimal decisions at every step. Intuition suggests that an algorithm such as Heuristic 1 that consistently makes suboptimal decisions is unlikely to outperform Algorithm 1 end-to-end, which is confirmed by the numerical results. However, an algorithm such as Heuristic 2 which almost always makes optimal decisions with only infrequent exceptions, may outperform Algorithm 1. According to Figure 3, these perturbations in end-to-end performance are practically insignificant and the intuitive choice of the optimal or a largely optimal algorithm at each step will result in the best end-to-end performance.

We note that the delays presented here (and also in the following figures) are, in fact, excess median or mean delays beyond the minimum required number of transmissions, which is K. For example, a mean delay of 10 slots for K=100 packets signifies on average 10% overhead, which is the

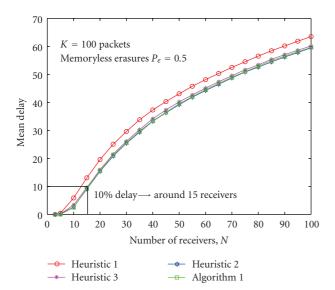


FIGURE 3: Mean decoding delay for the transmission of K=100 packets to N=3 to N=100 receivers. Algorithm 1 is compared with Heuristics 1–3. Both Heuristics 2 and 3 perform very closely to Algorithm 1. The maximum number of recursions for both Heuristic 2 and 3 is set to 100.

price for guaranteeing instantaneous decodability. In other words, one measure of throughput is $th_1 = K/(K + \overline{d})$, where \overline{d} is the mean delay across all receivers. An example is shown in Figure 3. For up to around 15 receivers in the system, Algorithm 1, Heuristics 2, and 3 ensure an average throughput loss of 10%.

It is quite possible that Algorithm 1 returns multiple network coding solutions all of which have the same objective value $\mathbf{w}^T \mathbf{x}$. A natural question that arises is whether systematic selection of a solution with a particular property is better than others in the presence of erasures in the channel. Our experiments verify that indeed some secondary post processing on the solutions can improve the end-toend delay. In particular, we compare two post processing techniques: (1) selecting a solution which involves minimum amount of coding (lowest number of 1's in the solution vector \mathbf{x}) and (2) selecting a solution with maximum amount of coding (highest number of 1's in the solution vector \mathbf{x}). Figure 4 shows the effects of such processing on the overall decoding delays. It is clear that maximum coding is not a reasonable choice and results in worse delays compared with minimum coding. We attempt to explain this behavior by means of an example and intuitive reasoning. Let us assume that there are K = 3 packets to be transmitted to N =3 receivers and at the beginning of the third transmission round, matrix A is given as follows

$$A = \begin{bmatrix} 0 & 1 & 1 \\ 1 & 0 & 1 \\ 0 & 1 & 1 \end{bmatrix}. \tag{8}$$

It is clear that there are two optimal solutions: we can either send packets $m_1 \oplus m_2$ or packet m_3 by itself, where the former involves coding and latter is uncoded. Now let us assume that

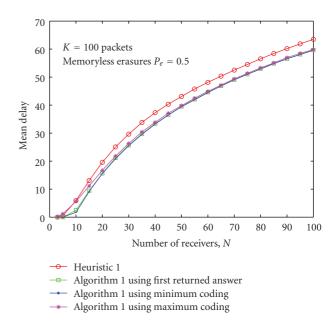


FIGURE 4: The effect of post processing on mean delay. Whenever Algorithm 1 returns multiple solutions, minimum amount of coding should be chosen. Heuristic 1 is shown for reference.

we select the maximum coding strategy and send $m_1 \oplus m_2$. If in the third transmission round only R_2 successfully receives, A will become

$$A = \begin{bmatrix} 0 & 1 & 1 \\ 0 & 0 & 1 \\ 0 & 1 & 1 \end{bmatrix}, \tag{9}$$

and clearly the optimal solution is sending packet m_3 . If in the fourth transmission round only R_1 successfully receives, A will become

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 1 & 1 \end{bmatrix}, \tag{10}$$

where it is evident that in the fifth transmission round, we cannot find a packet which is innovative and instantaneously decodable for all the three receivers. On the other hand, one can verify that if we adopt a minimum coding strategy and send packet m_3 in the third transmission round, we can always find innovative and instantaneously decodable packets for all three receivers in the future regardless of erasures in the channel. In summary, solutions with less coding tend to cause less constrains on the problem in the future.

It is noted in Figure 4 that the first solution returned by Algorithm 1 performs almost the same as the minimum coding solution. The reason for this is that Algorithm 1 first ranks the packets based on the number of receivers that need them. Therefore, the first solution picked by the algorithm is likely to contain packets with largest constrained sets and hence, many resolved packets are set to zero, which often translates into small amount of coding. Throughout this

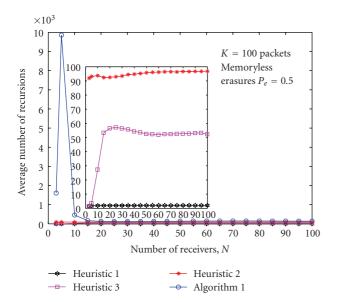


FIGURE 5: Average number of recursions in Algorithm 1 and Heuristics 1–3. The maximum number of recursions for both Heuristic 2 and 3 is set to 100. By referring to Figure 3, we observe that for small number of receivers, Heuristics 2-3 can provide same decoding delays at a fraction of computational complexity.

section, unless otherwise stated, we have shown the delay results based on the first returned solution of Algorithm 1.

It is interesting to analyze the actual number of recursions that the search in Algorithm 1 takes to find the optimum solution. This is shown in Figure 5 for K = 100packets along with the number of recursions required in Heuristics 1, 2, and 3. Algorithm 1 shows three modes of behavior: low, medium, and high number of recursions. When the number of receivers is larger than N = 20, Algorithm 1 finds the optimal solution very quickly and the number of recursions is very close to the number of packets K. However, when the number of receivers is lower, the constraints that each receiver imposes on the network coding decisions cannot limit the search space enough and hence, a large number of combinations have to be tested. Obviously, Heuristic 1 has the lowest number of recursions. Compared to Heuristic 2 with 100 fixed recursions, dynamic Heuristic 3 can almost halve the number of recursions with negligible effect on delay performance (see Figure 3). By referring to Figure 3, we conclude that for the system under consideration, the excessive number of recursions in Algorithm 1 is not warranted as it does not result in any noticeable delay improvement compared to Heuristics 2 or

Figure 6 shows the effect of increasing the number of packets on the computational complexity of Algorithm 1 in terms of number of recursions to complete the search. Three different numbers of receivers $N=20,\,N=30,\,{\rm and}\,N=40$ are considered. The complexity remains linear with the number of packets for well-sized receiver populations (30 and 40 receivers). This is in agreement with observations in Figure 5. When the number of receivers is not so large

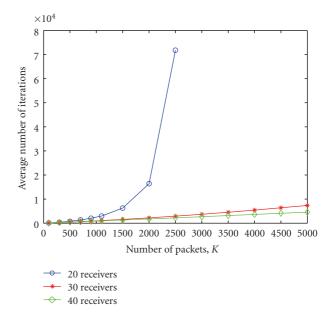


FIGURE 6: The effect of increasing the number of packets on the computational complexity of Algorithm 1 in terms of number of recursions. The complexity remains linear with the number of packets for well-sized receiver populations (30 and 40 receivers).

(see the blue curve in Figure 6 for N=20), we see a sudden growth in complexity, in terms of number of recursions, when $K \gtrsim 700$ packets. In such situations, truncating the number of recursion to be linear with the number of packets (Heuristic 2) is a good alternative.

Figure 7 shows the impact of the number of packets and also erasure probability on the decoding delay. The normalized mean delay versus number of packets K is plotted for three different erasure probabilities $P_e = 0.5$, $P_e = 0.4$, and $P_e = 0.2$, which are still high erasure probabilities. The number of receivers is fixed to N=20. The delay performance of Heuristics 1 and 3 are shown. A few observations are made. Firstly, as expected, the delay (both absolute and normalized measures) decreases as the erasure probability decreases. Secondly, the difference in the delay performance between Heuristics 1 and 3 decreases as the erasure probability decreases. This trend has also been observed for other number of receivers. Moreover, the difference between heuristics and Algorithm 1 decreases with erasure probability, which is not shown here for clarity of figure. Finally, the normalized delay decreases as the number of packets increases. We noted, however, that the absolute delay may increase or decrease depending on the number of receivers in the system. We attribute possible decrease in the normalized delay to the fact that when there are more packets to transmit, the transmitter has more options to choose from and hence, encounters delays less often in a normalized sense.

An important question that may arise in practical situations is how to choose the "block size" or the number of packets that are taken into account for making network coding decisions. If one has a total of *K* packets to transmit, does it make sense to divide them into subblocks of smaller

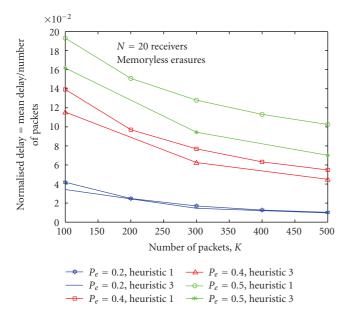


FIGURE 7: The effect of number of packets and erasure probabilities on the normalized delay. The maximum number of recursions for Heuristic 3 is set to 100. As the erasure probability decreases, the delay decreases as expected. The normalized delay decreases with *K* for this particular *N* (this is not always the case).

sizes or does it make sense to treat them as one single block of packets? The short answer is to include all "order-insensitive" packets in making transmission decisions and only break the packets into subblocks when the assumption of order insensitivity between subblocks breaks down. In the extreme case, an infinite number of order-insensitive packets provides an infinite pool of packets to choose from that can satisfy the demands of all receivers and are instantaneously decodable. Figure 8 shows the end-to-end delay when the number of packets in a block is finite and K = 100 packets is chosen as the reference for comparison. We can see that although the delay of transmitting λK packets, $d_{\lambda K}$, can be larger than that of transmitting K packets d_K , the delay does not increase by a factor of λ . That is $d_{\lambda K} < \lambda d_K$ and one does not benefit from breaking λK packets into λ subblocks of size K packets each. By treating λ subblocks of size K as one block of size λK , we add more degrees of freedom in making decisions.

Now we turn our attention to the delay performance of our algorithms in channels with memory. Figure 9 shows the mean delay of different algorithms for K=100 packets and N=3 receivers. The GEC parameters for all links are identical with b=g. The horizontal axis shows the memory content $\mu=1-2b$. The first curve from above shows the performance of Algorithm 1 when the transmitter does not take channel conditions into account in making coding decisions. In other words, $w_j=|M_j|$ is used in Algorithm 1 as if the channel states were memoryless. For relatively large memory contents, this method results in the largest mean delay. The next curve shows the delay performance of Heuristic 1. The next two curves, which are almost

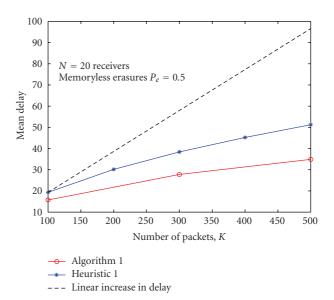


FIGURE 8: The effect of block size on the mean delay. If the delay of transmitting K=100 packets in Heuristic 1, d_{100} , is taken as the reference, we can see that the delay of including $\lambda \times 100$ packets in transmission is less than λd_{100} . The same observation applies to the delay of Algorithm 1. In general, it is recommended to include all "order-insensitive" packets in making transmission decisions and only break the packets into subblocks when the assumption of order insensitivity between subblocks breaks down.

indistinguishable, show the performance of Algorithm 1 which takes channel states into account (using (7)) and Heuristic 2 with 100 recursions. The last curve shows the best delay that can be achieved by occasionally violating the instantaneous decodability rule for one receiver in favor of the other two receivers that are predicted to be in good state in the next transmission round. More details can be found in [14].

Figure 10 shows the delay performance of Algorithm 1 using packet weights according to (7) for N=3 to N=15 receivers. Both the mean delay and mean delay plus one standard deviation of delay (across 1000 stochastic runs of the transmission) are shown. As expected, the delay increases as the number of receivers increases. Comparing the delay's standard deviation with its mean, we observe that when the number of receivers is 3–5, the delay is relatively more variant than when the number of receivers is 10–15. For example, for N=3 and $\mu=0.984$, the ratio of standard deviation to mean delay is around $3.225/0.8183 \approx 4$, whereas for N=15 and $\mu=0.94$ this ratio reduce to only $7.35/22.49 \approx 0.33$. One should keep these variations in mind when designing the transmission system.

We conclude this section with a brief look at the effect of post processing on the delay performance in channels with memory. Figure 11 shows different delays for N=15 receivers and K=100 packets. The figure confirms our earlier finding that selecting the maximum amount of coding among the optimal solutions provided by Algorithm 1 can result in larger end-to-end delays. We also note that serving the maximum number of receivers can have an adverse effect

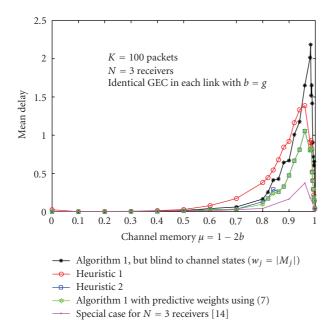


FIGURE 9: Delay performance of different algorithms in Gilbert-Elliott channels. The maximum number of recursions for Heuristic 2 is set to 100. By predicting next channel states and defining packet weights accordingly (see (7)), one can achieve considerably lower delays.

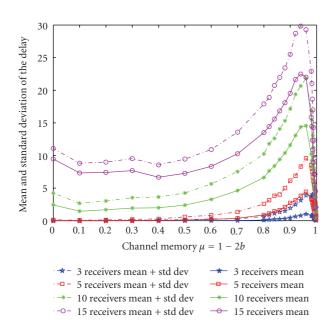


FIGURE 10: Delay performance of Algorithm 1 with weights defined using (7) for different number of receivers. As expected, the delay increases with the number of receivers. Both the mean delay (solid curves) and mean delay plus one standard deviation of delay (dashed-dotted curves) across 1000 stochastic runs of the transmission are shown.

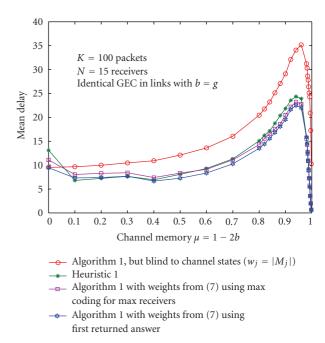


FIGURE 11: The effect of post processing on mean delay. As explained in the main text, whenever Algorithm 1 returns multiple solutions, choosing the maximum amount of coding and serving maximum number of receivers can often have adverse effects on the delay.

on the delay in GEC's. To explain this, consider an example where there are K=2 left packets to be transmitted to N=100 receivers. Packet 1 is needed by R_1 to R_{99} and packet 2 is needed by R_{99} and R_{100} . Since both packets are needed by R_{99} , we can either send packet 1 or 2, but not both. Now assume that R_1 to R_{99} are all predicted to be in good state with probability 0.01 and R_{100} is predicted to be in good state with probability 0.98, so that $w_1=w_2=0.99$ according to (7). Therefore, transmission of either packet seems to be equally optimal. However, one can easily verify that the probability of at least one receiver among R_1 to R_{99} receiving packet 1 is only $1-0.99^{99}=0.63$, whereas the probability of either R_{99} or R_{100} receiving packet 2 is 1-0.99*0.02=0.9802. Therefore, it makes sense to satisfy only two receivers, one of which has a high priority due its good channel conditions.

7. Conclusions

In this paper, we provided an online optimal network coding scheme with feedback to minimize decoding delay in each transmission round in erasure broadcast channels. Efficient search algorithms for the optimal network coding solution, as well as heuristic methods were presented and their delay and computational performance were tested in several system scenarios. We found that adopting an optimized approach using as much information about the channel as possible, such as memory, leads to a significantly better decoding delay. An interesting problem for future research is

to relax the instantaneous decodability condition to *L*-step decodability and investigate the delay-throughput tradeoff.

Appendix

Here we prove by structural induction that (a) every result returned by Algorithm 1 is a solution of (4) and (b) the set of solutions returned by the algorithm contains all the optimal solutions. We note that the algorithm is designed to discard infeasible vectors and those solutions that are clearly suboptimal at each recursion to improve performance. The latter is based on positiveness of the elements of ${\bf w}$ as explained below.

The algorithm generates a binary tree. Each node represents a problem of size k and \mathcal{P}_k , and branches into two subproblems of size $\mathcal{P}_{k-k_u-k_s-1}$ and \mathcal{P}_{k-k_u-1} . The former subproblem is a result of setting $x_s = 1$ and the latter a result of setting $x_s = 0$. A leaf is reached when we need to solve \mathcal{P}_1 . Without loss of generality let us assume that the variable to be examined is the first variable (s = 1) which is followed by k_s variables (x_2 to x_{k_s+1}) that are constrained to x_1 , $k-k_u-k_s-1$ variables (x_{k_s+2} to x_{k-k_u}) that are constrained but not to x_1 , and finally k_u unconstrained variables x_{k-k_u+1} to x_k . This can be easily accomplished by rearranging the columns of A.

For k=1, it is clear that the only optimal solution to \mathcal{P}_1 is $x_1=1$ which is returned by the algorithm. Hence, the *minimal structure* of the algorithm returns the optimal solution and our claim is true for k=1. The induction hypothesis is that the two subproblems $\mathcal{P}_{k-k_u-k_s-1}$ and \mathcal{P}_{k-k_u-1} have only discarded infeasible vectors and some suboptimal solutions. We need to prove that the same statement applies to the parent problem \mathcal{P}_k .

We first look at the left branch where $x_1 = 1$. According to the construction of the algorithm, any solution such as \mathbf{x}_1 of length $k - k_s - k_u - 1$ provided by the left branch $\mathcal{P}_{k-k_s-k_u-1}$ is appended by the parent problem \mathcal{P}_k to form

$$\mathbf{x} = \left[1, \underbrace{0, 0, \dots, 0}_{k_s}, \mathbf{x}_1, \underbrace{1, 1, \dots, 1}_{k_u}\right],\tag{A.1}$$

where the head variable x_1 is set to one, variables constrained to x_1 are set to zero and all unconstrained variables are set to one. We first prove that \mathbf{x} is indeed a solution and then show that changing any element of \mathbf{x} results in either an infeasible or a clearly suboptimal \mathbf{x} . We use Definitions 6–8.

(i) For

$$\mathbf{x} = \left[1, \underbrace{0, 0, \dots, 0}_{k_r}, \mathbf{x}_1, \underbrace{1, 1, \dots, 1}_{k_r}\right],\tag{A.2}$$

we write the condition $A\mathbf{x}$ as a weighted sum of columns of A. That is, $A\mathbf{x} = 1\mathbf{a}_1 + A'\mathbf{x}_1 + 1\mathbf{a}_{k-k_u+1} + \cdots + 1\mathbf{a}_k$, where A' is a submatrix of A of size $N \times (k-k_s-k_u-1)$, which is input to $\mathcal{P}_{k-k_s-k_u-1}$, and according to the induction hypothesis $A'\mathbf{x}_1 \le \mathbf{1}_N$. But since no variable in $\mathcal{P}_{k-k_u-k_s-1}$ is constrained to x_1 , no column in A' and \mathbf{a}_1 can have ones in the same row position. Therefore, $1\mathbf{a}_1 + A'\mathbf{x}_1 \le \mathbf{1}_N$.

- (ii) Since x_{k-k_u+1} to x_k are unconstrained, no column \mathbf{a}_{k-k_u+1} to \mathbf{a}_k can have ones in the same row position. Hence, $1\mathbf{a}_{k-k_u+1}+\cdots+1\mathbf{a}_k \leq 1_N$.
- (iii) Using similar arguments, we can assert that no column in A' or \mathbf{a}_1 can have ones in the same row positions as \mathbf{a}_{k-k_u+1} to \mathbf{a}_k do. Therefore, $1\mathbf{a}_1 + A'\mathbf{x}_1 + 1\mathbf{a}_{k-k_u+1} + \cdots + 1\mathbf{a}_k \le \mathbf{1}_N$ and \mathbf{x} is a solution.
- (iv) We now argue that variables x_2 to x_{k_s+1} cannot be anything other than zero. This directly follows from the fact that x_1 is constrained with x_i for $2 \le i \le k_s + 1$ and hence, in any given solution they cannot be simultaneously one.
- (v) Since we have already found a solution \mathbf{x} where the first and last k_u variables are one, we know that any other solution such as \mathbf{x}' with one or more zeros in these positions becomes suboptimal and can be discarded. That is, $\mathbf{w}^T\mathbf{x}' < \mathbf{w}^T\mathbf{x}$ due to positiveness of elements of \mathbf{w} .
- (vi) Finally, according to induction hypothesis, we know that \mathbf{x}_1 cannot be changed into anything other than what $\mathcal{P}_{k-k_s-k_u-1}$ provides without making it either infeasible or suboptimal.

In summary, for each solution \mathbf{x}_1 provided by the left branch $\mathcal{P}_{k-k_s-k_u-1}$, the constructed vector

$$\mathbf{x} = \left[1, \underbrace{0, 0, \dots, 0}_{k_{s}}, \mathbf{x}_{1}, \underbrace{1, 1, \dots, 1}_{k_{s}}\right], \tag{A.3}$$

is the only solution that is not trivially suboptimal.

Now we look at the right branch where $x_1 = 0$. According to the construction of the algorithm, a given solution such as \mathbf{x}_0 of length $k - k_u - 1$ provided by the right branch \mathcal{P}_{k-k_u-1} is appended by the parent problem \mathcal{P}_k to form

$$\mathbf{x} = \left[0, \mathbf{x}_0, \underbrace{1, 1, \dots, 1}_{k_n}\right],\tag{A.4}$$

where the head variable is set to zero and all unconstrained variables are set to one. We need to show that for a given \mathbf{x}_0 this is indeed a solution. We then show that changing any element of \mathbf{x} can only result in an infeasible vector, a clearly suboptimal solution, or a duplicate solution already provided by the left branch and hence, can be discarded. We use Definitions 6–8.

(i) We write $A\mathbf{x}$ as $A\mathbf{x} = 0\mathbf{a}_1 + A''\mathbf{x}_0 + 1\mathbf{a}_{k_s+2} + \cdots + 1\mathbf{a}_k$, where A'' is a submatrix of A of size $N \times (k - k_u - 1)$, which is input to \mathcal{P}_{k-k_u-1} , and according to the induction hypothesis $A''\mathbf{x}_0 \leq \mathbf{1}_N$. Similar to the arguments for the left branch, we can assert that no column \mathbf{a}_{k-k_u+1} to \mathbf{a}_k corresponding to unconstrained variables can have ones in the same row position. Hence, $1\mathbf{a}_{k-k_u+1} + \cdots + 1\mathbf{a}_k \leq \mathbf{1}_N$. Furthermore, that no column in A'' can have ones in the same row positions as \mathbf{a}_{k-k_u+1} to \mathbf{a}_k . Therefore, $A''\mathbf{x}_0 + 1\mathbf{a}_{k-k_u+1} + \cdots + 1\mathbf{a}_k \leq \mathbf{1}_N$ and \mathbf{x} is a solution.

- (ii) Since we have already found a solution \mathbf{x} where the last k_u variables are one, we know that any other solution such as \mathbf{x}' with one or more zeros in these positions becomes suboptimal and can be discarded.
- (iii) Finally, we show that any vector of the form

$$\mathbf{x}' = \left[1, \mathbf{x}_0, \underbrace{1, 1, \dots, 1}_{k_n}\right] \tag{A.5}$$

with a one in the first variable is either infeasible or is already constructed based on solutions from the left branch and hence, need not be considered twice. We consider two possibilities for $\mathbf{x}_0 = [x_2, \dots, x_{k_s+1}, x_{k_s+2}, \dots, x_{k-k_u}]$. If $x_i = 1$ for any $2 \le i \le k_s + 1$, then we have already shown in the analysis of the left branch that

$$\mathbf{x}' = \left[1, \mathbf{x}_0, \underbrace{1, 1, \dots, 1}_{k_u}\right] \tag{A.6}$$

is infeasible because x_1 and x_i are constrained to each other. If none of x_2 to x_{k_s+1} are one, then \mathbf{x}' will be of the form

$$\mathbf{x}' = \left[1, \underbrace{0, 0, \dots, 0, \mathbf{x}_1}_{\mathbf{x}_0}, \underbrace{1, 1, \dots, 1}_{k_n}\right] \tag{A.7}$$

for some \mathbf{x}_1 . But, \mathbf{x}_1 has to be a solution of $\mathcal{P}_{k-k_s-k_u-1}$. Hence, considering vectors of the form

$$\mathbf{x}' = \left[1, \underbrace{0, 0, \dots, 0, \mathbf{x}_1}_{\mathbf{x}_0}, \underbrace{1, 1, \dots, 1}_{k_u}\right] \tag{A.8}$$

does not lead to any new solution.

In summary, for each solution \mathbf{x}_0 provided by the right branch \mathcal{P}_{k-k_u-1} , the constructed vector

$$\mathbf{x} = \left[0, \mathbf{x}_0, \underbrace{1, 1, \dots, 1}_{h}\right] \tag{A.9}$$

is the only novel solution that is not trivially suboptimal. By combining the arguments of left and right branch, the induction claim is proven.

Acknowledgments

The authors wish to thank anonymous reviewers for their valuable comments which helped to improve the presentation of this paper. In the early stages of this work, the authors benefited from fruitful discussions with Ralf Koetter. This paper is dedicated to his memory. Preliminary results of this paper were presented in the 2009 Workshop on Network Coding, Theory and Applications (NetCod 2009), Lausanne, Switzerland. The work of P. Sadeghi was supported under ARC Discovery Projects funding scheme (Project no. DP0984950). The work of D. Traskov was supported by the European Commission in the framework of the FP7 Network of Excellence in Wireless COMmunications NEWCOM++ (Contract no. 216715).

References

- [1] A. Shokrollahi, "Raptor codes," *IEEE Transactions on Information Theory*, vol. 52, no. 6, pp. 2551–2567, 2006.
- [2] T. Ho, M. Medard, R. Koetter, et al., "A random linear network coding approach to multicast," *IEEE Transactions on Information Theory*, vol. 52, no. 10, pp. 4413–4430, 2006.
- [3] A. Eryilmaz, A. Ozdaglar, and M. Medard, "On delay performance gains from network coding," in *Proceedings of the IEEE Annual Conference on Information Sciences and Systems (CISS '06)*, pp. 864–870, Princeton, NJ, USA, March 2006.
- [4] D. E. Lucani, M. Stojanovic, and M. Medard, "Random linear network coding for time division duplexing: when to stop talking and start listening," in *Proceedings of the IEEE Conference on Computer Communications (INFOCOM '09)*, pp. 1800–1808, April 2009.
- [5] J.-K. Sundararajan, D. Shah, and M. Medard, "Feedback-based online network coding," Submitted to *IEEE Transactions on Information Theory*, http://arxiv.org/pdf/0904.1730v1.
- [6] J.-K. Sundararajan, P. Sadeghi, and M. Medard, "A feedback-based adaptive broadcast coding scheme for reducing in-order delivery delay," in *Proceedings of the Workshop on Network Coding, Theory, and Applications (NetCod '09)*, Lausanne, Switzerland, June 2009.
- [7] V. K. Goyal, "Multiple description coding: compression meets the network," *IEEE Signal Processing Magazine*, vol. 18, no. 5, pp. 74–93, 2001.
- [8] L. Keller, E. Drinea, and C. Fragouli, "Online broadcasting with network coding," in *Proceedings of the 4th Workshop on Network Coding, Theory, and Applications (NetCod '08)*, Hong kong, January 2008.
- [9] D. Bertsimas and R. Weissmantel, *Optimization Over Integers*, Dynamic Ideas, Belmont, Mass, USA, 2005.
- [10] T. S. Rappaport, Wireless Communications, Principles and Practice, Prentice Hall, Upper Saddle River, NJ, USA, 2nd edition, 2002.
- [11] P. Sadeghi, R. A. Kennedy, P. B. Rapajic, and R. Shams, "Finite-state Markov modeling of fading channels: a survey of principles and applications," *IEEE Signal Processing Magazine*, vol. 25, no. 5, pp. 57–80, 2008.
- [12] M. Mushkin and I. Bar-David, "Capacity and coding for the Gilbert-Elliot channels," *IEEE Transactions on Information Theory*, vol. 35, no. 6, pp. 1277–1290, 1989.
- [13] S. Katti, H. Rahul, W. Hu, D. Katabi, M. Medard, and J. Crowcroft, "XORs in the air: practical wireless network coding," in *Proceedings of the ACM Computer Communication Review (SIGCOMM '06)*, vol. 36, pp. 243–254, ACM Press, October 2006.
- [14] P. Sadeghi, D. Traskov, and R. Koetter, "Adaptive network coding for broadcast channels," in *Proceedings of the Workshop* on *Network Coding, Theory, and Applications (NetCod '09)*, pp. 80–85, Lausanne, Switzerland, June 2009.