# Harmonic Wavelet Transform and Image Approximation

Zhihua Zhang · Naoki Saito

Published online: 5 June 2010 © The Author(s) 2010. This article is published with open access at Springerlink.com

Abstract In 2006, Saito and Remy proposed a new transform called the Laplace Local Sine Transform (LLST) in image processing as follows. Let f be a twice continuously differentiable function on a domain  $\Omega$ . First we approximate f by a harmonic function u such that the residual component v = f - u vanishes on the boundary of  $\Omega$ . Next, we do the odd extension for v, and then do the periodic extension, i.e. we obtain a periodic odd function  $v^*$ . Finally, we expand  $v^*$  into Fourier sine series. In this paper, we propose to expand  $v^*$  into a periodic wavelet series with respect to biorthonormal periodic wavelet bases with the symmetric filter banks. We call this the Harmonic Wavelet Transform (HWT). HWT has an advantage over both the LLST and the conventional wavelet transforms. On the one hand, it removes the boundary mismatches as LLST does. On the other hand, the HWT coefficients reflect the local smoothness of fin the interior of  $\Omega$ . So the HWT algorithm approximates data more efficiently than LLST, periodic wavelet transform, folded wavelet transform, and wavelets on interval. We demonstrate the superiority of HWT over the other transforms using several standard images.

**Keywords** Harmonic wavelet transform · Laplace local sine transform · Biorthonormal wavelets · Periodic

Z. Zhang (🖂)

Z. Zhang · N. Saito Department of Mathematics, University of California, Davis, CA 95616, USA

N. Saito e-mail: saito@math.ucdavis.edu wavelets  $\cdot$  Folded wavelets  $\cdot$  Wavelets on interval  $\cdot$  Periodic wavelet coefficient  $\cdot$  Symmetry  $\cdot$  Odd extension

# **1** Introduction

Wavelet analysis is an important tool in image processing. In order to approximate or compress data, there are two common wavelet algorithms: the periodic wavelet algorithm, and the folded wavelet algorithm. More precisely, let an image f be supported on a square  $\Omega \in \mathbb{R}^2$ . In the periodic wavelet algorithm, one extends f to a periodic function  $f^*$ , and then expands  $f^*$  into a periodic wavelet series using Daubechies wavelets [1] or polyharmonic spline wavelets [9]. Since  $f^*$ is discontinuous at the boundary points of  $\Omega$  in general, the decay rate of the corresponding periodic wavelet coefficients is very slow. Hence, in order to obtain a good approximation of the image, we need many periodic wavelet coefficients. In the folded wavelet algorithm, in order to avoid the boundary mismatches caused by the brute-force periodization, one does an even extension of f, and then extends it to a periodic function  $f^*$ , and finally expands  $f^*$  into a periodic wavelet series with respect to a pair of biorthonormal periodic wavelet bases. If f is smooth, then  $f^* \in \text{lip1}$  on  $\mathbb{R}^2$ . This makes the decay rate of the corresponding wavelet coefficients faster than that in periodic wavelets algorithm. But, since  $\frac{\partial}{\partial x} f^*$  and  $\frac{\partial}{\partial y} f^*$  are discontinuous at the boundary points, the decay rate of the periodic wavelet coefficients is still relatively low.

It is natural to ask how to ensure the continuity of derivatives of the periodized function across the boundary. In 2006, Saito and Remy [8] presented a good method: the Laplace Local Sine Transform (LLST) that ensures the derivatives of the periodized functions are continuous across the boundary. It decomposes an image f supported on a

College of Global Change and Earth System Science, Beijing Normal University, Beijing 100875, China e-mail: zhangzh@math.ucdavis.edu

square  $\Omega$  into f = u + v, where u is a harmonic component satisfying Laplace's equation  $\Delta u = 0$  in  $\Omega$  and u = f on the boundary. After an odd extension, the residual component vis periodized to be  $v^*$ , i.e.,  $v^*$  is a periodic odd function. If f is smooth, then  $\frac{\partial}{\partial x}v^* \in \text{lip } 1$ ,  $\frac{\partial}{\partial y}v^* \in \text{lip } 1$  on  $\mathbb{R}^2$ . We call this method LLST decomposition. When one applies LLST to image approximation, one expands  $v^*$  into the Fourier sine series.

In this paper, combining the LLST decomposition and the wavelet algorithm, we expand  $v^*$  into the periodic wavelet series. More precisely, we propose the harmonic wavelet transform (HWT). In the HWT algorithm, the first step is the same as LLST, i.e., we decompose f = u + v and obtain a periodic odd function  $v^*$ . Next, we choose a pair of real-valued biorthonormal wavelets  $\psi$ ,  $\tilde{\psi}$  of  $L^2(\mathbb{R})$  generated by the real-valued even scaling functions  $\varphi$ ,  $\tilde{\varphi}$ . Using a method of the tensor product and periodization, one gets a pair of biorthonormal periodic wavelet bases. Finally, we expand  $v^*$  into the periodic wavelet series with respect to this pair of biorthonormal periodic wavelet bases. Since  $\frac{\partial}{\partial x}v^* \in \text{lip 1}$  and  $\frac{\partial}{\partial y}v^* \in \text{lip 1}$  on  $\mathbb{R}^2$ , the decay rate of the corresponding periodic wavelet algorithm or the folded wavelet algorithm.

In general, the decay rate of the Fourier sine coefficients depends on global smoothness, while that of the periodic wavelet coefficients depends on local smoothness. Since the global smoothness of a function is determined by its rough part, we need fewer periodic wavelet coefficients than the Fourier sine coefficients in order to approximate the image with the same quality. So, the HWT algorithm compresses data more efficiently than the LLST algorithm.

In the HWT algorithm, we carefully study the symmetry of the periodic wavelet coefficients and show where these coefficients vanish. From this, we see that in order to recover the image exactly, the number of the efficient coefficients is just the same as the size of the sample points of f. So the HWT is not a redundant transform.

Our HWT algorithm is quite different from polyharmonic wavelets proposed by Van De Ville et al. [9], which are a kind of wavelet bases constructed by polyharmonic functions. The HWT has a different strategy: we approximate a target function by a harmonic function such that the residual part vanishes on the boundary, and then expand the residual part into wavelet series.

This paper is organized as follows. In Sect. 2, we recall the notions of biorthonormal periodic wavelet bases and the corresponding Mallat algorithm as well as the LLST algorithm. In Sect. 3, we present the HWT algorithm and show that in the one-dimensional case, the periodic wavelet algorithm, the folded wavelet algorithm, and the HWT algorithm generate periodic wavelet coefficients at the level *m* with the decay rates  $O(2^{-\frac{m}{2}})$ ,  $O(2^{-\frac{3m}{2}})$ , and  $O(2^{-\frac{5m}{2}})$ , respectively. When we use first  $2^M$  periodic wavelet coefficients, the obtained approximation error are o(1),  $O(2^{-M})$ , and  $O(2^{-2M})$ , respectively. In the two-dimensional case, the decay rates of the corresponding periodic wavelet coefficients are  $O(2^{-m})$ ,  $O(2^{-2m})$ , and  $O(2^{-3m})$ , respectively. When we use first  $2^{2M}$  periodic wavelet coefficients, the obtained approximation error are o(1),  $O(2^{-M})$ , and  $O(2^{-2M})$ , respectively. In Sect. 4, first, we discuss one-dimensional discrete HWT and the symmetry of the sequences consisting of the corresponding periodic wavelet coefficients. From this, we precisely show the efficient number of the periodic wavelet coefficients in order to recover a signal perfectly. Second, we discuss two-dimensional discrete HWT and the symmetry of the matrices of the corresponding periodic wavelet coefficients. From this, we precisely show the efficient number of the periodic wavelet coefficients in order to recover an image perfectly. In Sects. 5-6, we apply the HWT algorithm to approximate images and show that the HWT algorithm approximates images better than the LLST algorithm, the periodic wavelet algorithm, the folded wavelet algorithm and the wavelets on the interval.

# 2 Preliminaries

We state the well-known notions [4, 7] of biorthonormal periodic wavelet bases. After that, we recall the corresponding Mallat algorithms.

#### 2.1 One-dimensional Biorthonormal Periodic Wavelets

It is well-known that using the method of periodization, one can construct biorthonormal periodic wavelet bases with the help of the biorthonormal wavelets [4, 7].

Let  $\psi$  and  $\tilde{\psi}$  be a pair of compactly supported smooth biorthonormal wavelets of  $L^2(\mathbb{R})$  generated by compactly supported smooth scaling functions  $\varphi$  and  $\tilde{\varphi}$ . For  $m \in \mathbb{Z}$ ,  $n \in \mathbb{Z}$ , we denote the function family  $g_{m,n} := 2^{\frac{m}{2}}g(2^m \cdot -n)$ . Its periodization (with period 1) is  $g_{m,n}^{\text{per}} := \sum_{l \in \mathbb{Z}} g_{m,n}(\cdot + l)$ . The families

$$\Psi_1^{\text{per}} := \{\varphi^{\text{per}}\} \cup \{\psi_{m,n}^{\text{per}}, m \in \mathbb{Z}_+, n = 0, \dots, 2^m - 1\},$$
  
$$\widetilde{\Psi}_1^{\text{per}} := \{\widetilde{\varphi}^{\text{per}}\} \cup \{\widetilde{\psi}_{m,n}^{\text{per}}, m \in \mathbb{Z}_+, n = 0, \dots, 2^m - 1\}$$

are called a pair of biorthonormal periodic wavelet bases for  $L^2([-\frac{1}{2},\frac{1}{2}]).$ 

Let  $f \in L^2([-\frac{1}{2}, \frac{1}{2}])$ . For  $m \in \mathbb{Z}_+$ ,  $n \in \mathbb{Z}$ , we denote the periodic wavelet coefficients

$$c_{m,n}^{(1)} := \int_{-\frac{1}{2}}^{\frac{1}{2}} f(t) \overline{\varphi_{m,n}^{\text{per}}(t)} \, \mathrm{d}t,$$
  
$$d_{m,n}^{(1)} := \int_{-\frac{1}{2}}^{\frac{1}{2}} f(t) \overline{\psi_{m,n}^{\text{per}}(t)} \, \mathrm{d}t.$$
 (2.1)

Deringer

Here the sequences  $\{c_{m,n}^{(1)}\}\$  and  $\{d_{m,n}^{(1)}\}\$  are periodic sequences (with period  $2^m$ ) with respect to the index *n*, i.e.,

$$c_{m,n+2^m l}^{(1)} = c_{m,n}^{(1)}, \qquad d_{m,n+2^m l}^{(1)} = d_{m,n}^{(1)} \quad (l \in \mathbb{Z}).$$

It is well known that

$$f = c_{0,0}^{(1)} + \sum_{m=0}^{\infty} \sum_{n=0}^{2^m - 1} d_{m,n}^{(1)} \widetilde{\psi}_{m,n}^{\text{per}}$$
(2.2)

in the space  $L^2([-\frac{1}{2}, \frac{1}{2}])$ . For  $M \in \mathbb{Z}_+$ , its  $2^M$ -partial sum is

$$S_{2^{M}}(f) = c_{0,0}^{(1)} + \sum_{m=0}^{M-1} \sum_{n=0}^{2^{m}-1} d_{m,n}^{(1)} \widetilde{\psi}_{m,n}^{\text{per}}.$$
(2.3)

Let the space  $\widetilde{V}_m$  be the span of  $\{\widetilde{\varphi}_{m,n}^{\text{per}}\}_{n=0,\dots,2^m-1}$ . One can decompose the projection of f in  $\widetilde{V}_m$  as follows [7].

$$P_{\widetilde{V}_{m}}(f) = \sum_{n=0}^{2^{m}-1} c_{m,n}^{(1)} \widetilde{\varphi}_{m,n}^{\text{per}} = \sum_{n=0}^{2^{m-1}-1} c_{m-1,n}^{(1)} \widetilde{\varphi}_{m-1,n}^{\text{per}} + \sum_{n=0}^{2^{m-1}-1} d_{m-1,n}^{(1)} \widetilde{\psi}_{m-1,n}^{\text{per}}.$$
 (2.4)

Denote the coefficients of the filter banks

$$a_{k} := \sqrt{2} \int_{\mathbb{R}} \varphi(t) \overline{\widetilde{\varphi}(2t-k)} \, \mathrm{d}t,$$
  

$$b_{k} := \sqrt{2} \int_{\mathbb{R}} \psi(t) \overline{\widetilde{\varphi}(2t-k)} \, \mathrm{d}t.$$
(2.5)

Without loss of generality, we assume that for some  $N \in \mathbb{Z}_+$ ,

$$a_n = b_{n+1} = 0 \quad (|n| \ge N).$$
 (2.6)

Mallat showed that the fast algorithm to compute periodic wavelet coefficients is similar to the fast algorithm of wavelet coefficients [7, Sect. 7.5.1]. Along Mallat's idea, one can obtain the following algorithm for the one-dimensional periodic wavelet coefficients.

**Proposition 2.1** Let the filters  $a_k$ ,  $b_k$  be defined in (2.5) and  $c_{m,n}^{(1)}$ ,  $d_{m,n}^{(1)}$  be defined in (2.1). Define

$$a_{n}^{*} := a_{n}, \qquad b_{n+1}^{*} := b_{n+1} \quad (|n| \le 2^{J-1}), a_{n+2^{J}}^{*} := a_{n}^{*}, \qquad b_{n+2^{J}}^{*} := b_{n}^{*} \quad (n \in \mathbb{Z}).$$

$$(2.7)$$

Then for  $2^{J-1} \ge N$ , we have

$$c_{J-1,k}^{(1)} = \sum_{n=0}^{2^{J}-1} \overline{a_{n-2k}^{*}} c_{J,n}^{(1)},$$

$$d_{J-1,k}^{(1)} = \sum_{n=0}^{2^{J}-1} \overline{b_{n-2k}^{*}} c_{J,n}^{(1)}, \quad k \in \mathbb{Z},$$
(2.8)

where N is stated in (2.6).

# 🖄 Springer

# 2.2 Two-dimensional Biorthonormal Periodic Wavelets

Let  $\psi$  and  $\tilde{\psi}$  be a pair of one-dimensional biorthonormal wavelets generated by the scaling functions  $\varphi$  and  $\tilde{\varphi}$ . Take the tensor products of  $\varphi$ ,  $\psi$ . Denote

$$\varphi_0(x, y) := \varphi(x)\varphi(y), \qquad \psi_1(x, y) := \varphi(x)\psi(y), 
\psi_2(x, y) := \psi(x)\varphi(y), \qquad \psi_3(x, y) := \psi(x)\psi(y).$$

Similarly, taking the tensor products of  $\tilde{\varphi}$  and  $\tilde{\psi}$ , we get  $\tilde{\varphi}_0(x, y)$ ,  $\tilde{\psi}_1(x, y)$ ,  $\tilde{\psi}_2(x, y)$ , and  $\tilde{\psi}_3(x, y)$ . Then  $\{\psi_{\mu}\}_1^3$  and  $\{\tilde{\psi}_{\mu}\}_1^3$  are a pair of two-dimensional biorthonormal wavelets generated by the scaling functions  $\varphi_0$  and  $\tilde{\varphi}_0$ , respectively.

The families

$$\Psi_2^{\text{per}} = \{\varphi_0^{\text{per}}\} \cup \{\psi_{\mu,m,n}^{\text{per}}, \ \mu = 1, 2, 3, \ m \in \mathbb{Z}_+, \\ n = (n_1, n_2), \ n_1, n_2 = 0, \dots, 2^m - 1\}, \\ \widetilde{\Psi}_2^{\text{per}} = \{\widetilde{\varphi}_0^{\text{per}}\} \cup \{\widetilde{\psi}_{\mu,m,n}^{\text{per}}, \ \mu = 1, 2, 3, \ m \in \mathbb{Z}_+, \\ n = (n_1, n_2), \ n_1, n_2 = 0, \dots, 2^m - 1\}$$

are called a pair of biorthonormal periodic wavelet bases for  $L^2([-\frac{1}{2},\frac{1}{2}]^2)$ .

Let  $f \in L^2([-\frac{1}{2}, \frac{1}{2}]^2)$ . For  $\mu = 1, 2, 3, m \in \mathbb{Z}_+, n \in \mathbb{Z}^2$ , we denote the periodic wavelet coefficients

$$c_{m,n}^{(2)} := \int_{\frac{1}{2}}^{\frac{1}{2}} \int_{-\frac{1}{2}}^{\frac{1}{2}} f(x, y) \overline{\varphi_{0,m,n}^{\text{per}}(x, y)} \, \mathrm{d}x \, \mathrm{d}y,$$
  
$$d_{\mu,m,n}^{(2)} := \int_{-\frac{1}{2}}^{\frac{1}{2}} \int_{-\frac{1}{2}}^{\frac{1}{2}} f(x, y) \overline{\psi_{\mu,m,n}^{\text{per}}(x, y)} \, \mathrm{d}x \, \mathrm{d}y.$$
 (2.9)

Here the sequences  $\{c_{m,n}^{(2)}\}$  and  $\{d_{\mu,m,n}^{(2)}\}$  are periodic sequences with respect to *n*, i.e.

$$c_{m,n+2^m l}^{(2)} = c_{m,n}^{(2)}, \qquad d_{\mu,m,n+2^m l}^{(2)} = d_{\mu,m,n}^{(2)} \quad (l \in \mathbb{Z}^2).$$

It is well known that

$$f = c_{0,0}^{(2)} + \sum_{\mu=1}^{3} \sum_{m=0}^{\infty} \sum_{n_1, n_2=0}^{2^m - 1} d_{\mu,m,n}^{(2)} \widetilde{\psi}_{\mu,m,n}^{\text{per}}$$
(2.10)

in the space  $L^2([-\frac{1}{2}, \frac{1}{2}]^2)$ , where  $n = (n_1, n_2)$ . For  $M \in \mathbb{Z}_+$ , its  $2^{2M}$ -partial sum is

$$S_{2^{2M}}(f) = c_{0,0}^{(2)} + \sum_{\mu=1}^{3} \sum_{m=0}^{M-1} \sum_{n_1,n_2=0}^{2^m-1} d_{\mu,m,n}^{(2)} \widetilde{\psi}_{\mu,m,n}^{\text{per}}$$

Let the space  $\widetilde{V}_m$  be the span of  $\{\widetilde{\varphi}_{0,m,n}^{\text{per}}\}_{n=(n_1,n_2), n_1,n_2=0,\dots,2^m-1}$ . One can decompose the project of f in  $\widetilde{V}_m$  as follows.

$$P_{\widetilde{V}_{m}}(f) = \sum_{n_{1},n_{2}=0}^{2^{m}-1} c_{m,n}^{(2)} \widetilde{\varphi}_{0,m,n}^{\text{per}}$$
  
$$= \sum_{n_{1},n_{2}=0}^{2^{m-1}-1} c_{m-1,n}^{(2)} \widetilde{\varphi}_{0,m-1,n}^{\text{per}}$$
  
$$+ \sum_{\mu=1}^{3} \sum_{n_{1},n_{2}=0}^{2^{m-1}-1} d_{\mu,m-1,n}^{(2)} \widetilde{\psi}_{\mu,m-1,n}^{\text{per}}.$$
 (2.11)

From the Mallat algorithm of the one-dimensional periodic wavelet, one can easily conclude the following Mallat algorithm for two-dimensional tensor product periodic wavelets.

**Proposition 2.2** Let the filters  $\{a_n^*\}, \{b_n^*\}$  be stated in (2.7) and the periodic wavelet coefficients  $c_{m,n}^{(2)}, d_{\mu,m,n}^{(2)}$  be stated in (2.9). Then for  $2^{J-1} \ge N$  and  $k = (k_1, k_2) \in \mathbb{Z}^2$ , we have

(i) 
$$c_{J-1,k}^{(2)} = \sum_{n_1,n_2=0}^{2^J-1} \overline{a_{n_1-2k_1}^*} \, \overline{a_{n_2-2k_2}^*} \, c_{J,n_1,n_2}^{(2)},$$

(ii) 
$$d_{1,J-1,k}^{(2)} = \sum_{n_1,n_2=0}^{2^J-1} \overline{a_{n_1-2k_1}^*} \, \overline{b_{n_2-2k_2}^*} \, c_{J,n_1,n_2}^{(2)},$$

(iii) 
$$d_{2,J-1,k}^{(2)} = \sum_{n_1,n_2=0}^{2^J-1} \overline{b_{n_1-2k_1}^*} \, \overline{a_{n_2-2k_2}^*} \, c_{J,n_1,n_2}^{(2)},$$

(iv) 
$$d_{3,J-1,k}^{(2)} = \sum_{n_1,n_2=0}^{2^J-1} \overline{b_{n_1-2k_1}^*} \, \overline{b_{n_2-2k_2}^*} \, c_{J,n_1,n_2}^{(2)}$$

where N is stated in (2.6).

#### 2.3 Laplace Local Sine Transforms

First we consider the one-dimensional case. Let  $\Omega_1 := [0, \frac{1}{2}]$ and let the function f be defined on  $\Omega_1$  and  $f \in C^2(\Omega_1)$ . We split the function f into two components

$$f(x) = u(x) + v(x)$$
 on  $\Omega_1$ 

The first function  $u(x) = 2(f(\frac{1}{2}) - f(0))x + f(0)$  and the second function satisfies

$$v(0) = v\left(\frac{1}{2}\right) = 0.$$

After an odd extension, the second function v is periodized to be  $v^*$ , i.e.,  $v^*$  is a periodic odd function. Finally we expand  $v^*$  into the Fourier sine series.

Next, we consider the two-dimensional case. Let  $\Omega_2 := [0, \frac{1}{2}]^2$  and the function f be defined on  $\Omega_2$  and  $f \in C^2(\Omega_2)$ . We split the function f into two components

$$f(x, y) = u(x, y) + v(x, y)$$
 on  $\Omega_2$ 

where u(x, y) is the harmonic function which satisfies Laplace's equation  $\Delta u(x, y) = 0$   $((x, y) \in \Omega_2)$  and u(x, y) = f(x, y) on the boundary of  $\Omega_2$ . So the residual satisfies

$$v(x, y) = 0$$
  $((x, y) \in \partial \Omega_2).$ 

After an odd extension, the residual component v is periodized to be  $v^*$ , i.e.,  $v^*$  is a periodic odd function. Finally we expand  $v^*$  into the Fourier sine series.

#### **3 Harmonic Wavelet Transform**

Now we present a notion of the *Harmonic Wavelet Transform* (HWT). We show that for the HWT algorithm, the decay rate of periodic wavelet coefficients is faster than those generated by the periodic wavelet transform algorithm and the folded wavelet algorithm.

#### 3.1 One-dimensional HWT Algorithm

We assume that f is defined on  $[0, \frac{1}{2}]$ . Let

$$f(t) = u(t) + v(t) \quad \left(t \in \left[0, \frac{1}{2}\right]\right),$$

where  $u(t) = 2(f(\frac{1}{2}) - f(0))t + f(0)$  and the residual component v(t) satisfies  $v(0) = v(\frac{1}{2}) = 0$ .

We do the odd extension  $v_{odd}$  of the residual component v to  $\left[-\frac{1}{2}, \frac{1}{2}\right]$ . We again do the 1-periodic extension of  $v_{odd}$ , denoted by  $v^*$ . Finally, we expand  $v^*$  into the biorthonormal periodic wavelet series. We call this process the one-dimensional HWT.

Now we examine the decay rates of the coefficients of various wavelet algorithms in the one-dimensional case.

(i) The 1D periodic wavelet algorithm. Let  $f \in C^2([-\frac{1}{2}, \frac{1}{2}])$ . In the periodic wavelet algorithm, we directly expand f into a biorthonormal periodic wavelet series. Clearly, the coefficients

$$d_{m,n}^{(1)} = \int_{-\frac{1}{2}}^{\frac{1}{2}} f(t) \overline{\psi_{m,n}^{\text{per}}(t)} \, \mathrm{d}t = \int_{\mathbb{R}} f^*(t) \overline{\psi_{m,n}(t)} \, \mathrm{d}t$$
$$= 2^{\frac{m}{2}} \int_{\mathbb{R}} f^*(t) \overline{\psi(2^m t - n)} \, \mathrm{d}t, \qquad (3.1)$$

Deringer

where  $f^*$  is the 1-periodic extension of f. Since  $f^* \in L^{\infty}(\mathbb{R}), \psi \in L^1(\mathbb{R})$ , we have

$$d_{m,n}^{(1)} = O(2^{\frac{m}{2}}) \int_{\mathbb{R}} |\psi(2^{m}t - n)| dt$$
  
=  $O(2^{-\frac{m}{2}}) \int_{\mathbb{R}} |\psi(t - n)| dt$   
=  $O(2^{-\frac{m}{2}}) \int_{\mathbb{R}} |\psi(t)| dt = O(2^{-\frac{m}{2}}).$  (3.2)

(ii) *The 1D folded wavelet algorithm*. Let  $f \in C^2([0, \frac{1}{2}])$ . In the folded wavelet algorithm, we do an even extension of f

$$f_{\text{even}}(t) = \begin{cases} f(t), & t \in [0, \frac{1}{2}], \\ f(-t), & t \in [-\frac{1}{2}, 0). \end{cases}$$

Then

$$f_{\text{even}}(t) \in C\left[-\frac{1}{2}, \frac{1}{2}\right]$$
 and  $f_{\text{even}}\left(-\frac{1}{2}\right) = f_{\text{even}}\left(\frac{1}{2}\right)$ .

Again let  $f^*(t)$  be a 1-periodic extension of  $f_{even}(t)$ , we have

**Proposition 3.1** *The periodic wavelet coefficients of*  $f^*$  *satisfy* 

$$d_{m,n}^{(1)} = O(2^{-\frac{3m}{2}}).$$
(3.3)

The partial sum  $S_{2^M}(f^*)$  of the periodic wavelet expansion of  $f^*$  satisfies

$$\|f^* - S_{2^M}(f^*)\|_{L^2([-\frac{1}{2},\frac{1}{2}])} = O(2^{-M}).$$
(3.4)

*Proof* Since  $f \in C^2[0, \frac{1}{2}]$ , we have  $f^* \in \text{lip } 1$  on  $\mathbb{R}$ , i.e., there is a constant K such that

$$\left|f^*(t) - f^*\left(\frac{n}{2^m}\right)\right| \le K \left|t - \frac{n}{2^m}\right| \quad (t \in \mathbb{R}).$$

Since  $f^* \in \lim 1$ ,  $\int_{\mathbb{R}} \psi(t) dt = 0$ , and (3.1), we have

$$\begin{aligned} \left| d_{m,n}^{(1)} \right| &= \left| 2^{\frac{m}{2}} \int_{\mathbb{R}} f^*(t) \overline{\psi(2^m t - n)} \, \mathrm{d}t \right| \\ &= \left| 2^{\frac{m}{2}} \int_{\mathbb{R}} \left( f^*(t) - f^*\left(\frac{n}{2^m}\right) \right) \overline{\psi(2^m t - n)} \, \mathrm{d}t \\ &\leq K \, 2^{\frac{m}{2}} \int_{\mathbb{R}} \left| t - \frac{n}{2^m} \right| |\psi(2^m t - n)| \, \mathrm{d}t \\ &= K \, 2^{-\frac{3m}{2}} \int_{\mathbb{R}} |t - n| |\psi(t - n)| \, \mathrm{d}t \\ &= K \, 2^{-\frac{3m}{2}} \int_{\mathbb{R}} |t\psi(t)| \, \mathrm{d}t = O(2^{-\frac{3m}{2}}). \end{aligned}$$

So (3.3) holds.

By (2.2) and (2.3), we know that the difference between  $f^*$  and the partial sum of its periodic wavelet series

$$f^* - S_{2^{M+1}}(f^*) = \sum_{m=M}^{\infty} \sum_{n=0}^{2^m - 1} d_{m,n}^{(1)} \,\overline{\widetilde{\psi}_{m,n}^{\text{per}}}$$

is in the space  $L^2([-\frac{1}{2}, \frac{1}{2}])$ . By (3.3), we get

$$\begin{split} \|f^* - S_{2^M}(f^*)\|_{L^2([-\frac{1}{2},\frac{1}{2}])}^2 &= O(1) \sum_{m=M}^{\infty} \sum_{n=0}^{2^m - 1} \left| d_{m,n}^{(1)} \right|^2 \\ &= O(1) \sum_{m=M}^{\infty} \sum_{n=0}^{2^m - 1} (2^{-\frac{3m}{2}})^2 \\ &= O(1) \sum_{m=M}^{\infty} 2^{-2m} = O(2^{-2M}). \end{split}$$

So (3.4) holds. Proposition 3.1 is proved.

(iii) The 1D HWT algorithm. Let  $f \in C^2([0, \frac{1}{2}])$ . Now we introduce the HWT algorithm.

(a) We first decompose f on  $[0, \frac{1}{2}]$  as follows

$$f(t) = u(t) + v(t) \quad \left(t \in \left[0, \frac{1}{2}\right]\right),$$

where  $u(t) = 2(f(\frac{1}{2}) - f(0))t + f(0)$ . So  $v(0) = v(\frac{1}{2}) = 0$ .

(b) We do the odd extension of v, i.e., let

$$v_{\text{odd}}(t) = v(t) \quad \left(t \in \left[0, \frac{1}{2}\right]\right) \text{ and}$$
  
 $v_{\text{odd}}(-t) = -v(t) \quad \left(t \in \left[0, \frac{1}{2}\right]\right).$ 

So  $v_{odd}(\frac{1}{2}) = v_{odd}(-\frac{1}{2}) = v_{odd}(0) = 0$ . Since  $\frac{v_{odd}(-t)}{-t} = \frac{v_{odd}(t)}{t}$ , letting  $t \to 0^+$ , we get  $(v_{odd})'_{-}(0) = (v_{odd})'_{+}(0)$ . Hence  $v_{odd}(t)$  is differentiable at t = 0.

(c) We do the 1-periodic extension  $v^*$  of  $v_{odd}$ , i.e.,  $v^*$  is a 1-periodic function and  $v^*(t) = v_{odd}(t)$   $(|t| \le \frac{1}{2})$ . Furthermore, we easily prove  $\frac{d}{dt}v^* \in \text{lip } 1$  on  $\mathbb{R}$ .

**Proposition 3.2** Let  $v^*$  be stated as above, i.e.,  $v^*$  is a 1periodic function and  $\frac{dv^*}{dt} \in lip 1$  on  $\mathbb{R}$ . Then the periodic wavelet coefficients of  $v^*$  satisfy

$$d_{m,n}^{(1)} = O(2^{-\frac{5m}{2}}).$$
(3.5)

The partial sum  $S_{2^M}(v^*)$  of the periodic wavelet expansion of  $v^*$  satisfies

$$\|v^* - S_{2^M}(v^*)\|_{L^2([-\frac{1}{2},\frac{1}{2}])} = O(2^{-2M}).$$
(3.6)

*Proof* Since  $\int_{\mathbb{R}} \psi(t) dt = 0$  and  $\int_{\mathbb{R}} t \psi(t) dt = 0$ , we have

$$d_{m,n}^{(1)} = 2^{\frac{m}{2}} \int_{\mathbb{R}} v^{*}(t) \overline{\psi(2^{m}t - n)} dt$$
  
=  $2^{\frac{m}{2}} \int_{\mathbb{R}} \left( v^{*}(t) - v^{*} \left( \frac{n}{2^{m}} \right) - v^{*\prime} \left( \frac{n}{2^{m}} \right) \left( t - \frac{n}{2^{m}} \right) \right) \overline{\psi(2^{m}t - n)} dt.$  (3.7)

Again, since  $\frac{dv^*}{dt} \in \text{lip 1}$  on  $\mathbb{R}$ , we know that there is a constant *K* such that for any  $t_0 \in \mathbb{R}$ ,

$$|v^{*'}(t) - v^{*'}(t_0)| \le K |t - t_0| \quad (t \in \mathbb{R}).$$

This implies that

$$|v^{*}(t) - v^{*}(t_{0}) - v^{*'}(t_{0})(t - t_{0})|$$
  
=  $\left| \int_{t_{0}}^{t} \left( v^{*'}(t') - v^{*'}(t_{0}) \right) dt' \right| \le K (t - t_{0})^{2} \quad (t \in \mathbb{R}).$ 

Taking  $t_0 = \frac{n}{2^m}$  in this formula, by (3.7) we have

$$\begin{aligned} \left| d_{m,n}^{(1)} \right| &\leq \frac{K}{2} \, 2^{\frac{m}{2}} \int_{\mathbb{R}} \left( t - \frac{n}{2^m} \right)^2 \left| \psi(2^m t - n) \right| \, \mathrm{d}t \\ &= \frac{K}{2} \, 2^{-\frac{5m}{2}} \int_{\mathbb{R}} t^2 |\psi(t)| \, \mathrm{d}t = O(2^{-\frac{5m}{2}}). \end{aligned} \tag{3.8}$$

By the similar argument to (3.4), from (3.8) we now have

$$\|v^* - S_{2^M}(v^*)\|_{L^2([-\frac{1}{2},\frac{1}{2}])} = O(2^{-2M}).$$
  
So (3.6) holds. Proposition 3.2 is proved.

### 3.2 Two-dimensional HWT Algorithm

We assume that f is defined on  $[0, \frac{1}{2}]^2$ . Let

$$f(x, y) = u(x, y) + v(x, y) \quad \left( (x, y) \in \left[ 0, \frac{1}{2} \right]^2 \right),$$

where u(x, y) is a harmonic function and v(x, y) satisfies v(x, y) = 0 for  $(x, y) \in \partial([0, \frac{1}{2}]^2)$ . We do the odd extension  $v_{\text{odd}}$  of the residual component v to  $[-\frac{1}{2}, \frac{1}{2}]^2$ , that is,

$$v_{\text{odd}}(x, y) = v(x, y) \quad \left( (x, y) \in \left[ 0, \frac{1}{2} \right]^2 \right)$$

and

$$v_{\text{odd}}(-x, y) = v_{\text{odd}}(x, -y)$$
$$= -v_{\text{odd}}(x, y) \quad \left((x, y) \in \left[-\frac{1}{2}, \frac{1}{2}\right]^2\right).$$

Again we do the 1-periodic extension of  $v_{odd}$ , denoted by  $v^*$ . Finally, we expand  $v^*$  into a biorthonormal periodic wavelet series. We call this process the two-dimensional HWT.

Now we examine the decay rates of the coefficients of various wavelet algorithms in the two-dimensional case.

(i) The 2D periodic wavelet algorithm. Let  $f \in C^2([-\frac{1}{2}, \frac{1}{2}]^2)$ . Denote the 1-periodic extension of f on  $\mathbb{R}^2$  by  $f^*$ . By (2.9), the periodic wavelet coefficients are

$$d_{\mu,m,n}^{(2)} = \int_{\mathbb{R}} \int_{\mathbb{R}} f^*(x, y) \overline{\psi_{\mu,m,n}(x, y)} \, \mathrm{d}x \, \mathrm{d}y \quad (\mu = 1, 2, 3).$$

So, for each  $\mu$ , we have

$$d_{\mu,m,n}^{(2)} = \left| 2^m \int_{\mathbb{R}} \int_{\mathbb{R}} f^*(x, y) \times \overline{\psi_{\mu}(2^m x - n_1, 2^m y - n_2)} \, dx \, dy \right|$$
  
=  $O(2^{-m}) \int_{\mathbb{R}} \int_{\mathbb{R}} |\psi_{\mu}(x - n_1, y - n_2)| \, dx \, dy$   
=  $O(2^{-m}) \int_{\mathbb{R}} \int_{\mathbb{R}} |\psi_{\mu}(x, y)| \, dx \, dy = O(2^{-m}).$ 

(ii) The 2D folded wavelet algorithm. Let  $f \in C^2([0, \frac{1}{2}]^2)$ , we do an even extension to  $[-\frac{1}{2}, \frac{1}{2}]^2$ , and then we do a 1-periodic extension to  $\mathbb{R}^2$ , denoted by  $f_2$ . Then  $f_2 \in \text{lip 1 on } \mathbb{R}^2$ .

**Proposition 3.3** Let  $f_2$  be stated as above, i.e.,  $f_2$  is a 1periodic function and  $f_2 \in \lim 1$  on  $\mathbb{R}^2$ . Then the periodic wavelet coefficients of  $f_2$  satisfy

$$d_{\mu,m,n}^{(2)} = O(2^{-2m}).$$
(3.9)

The partial sum  $S_{2^{2M}}$  of the periodic wavelet expansion of  $f_2$  satisfies

$$\|f_2 - S_{2^{2M}}(f_2)\|_{L^2([-\frac{1}{2},\frac{1}{2}]^2)} = O(2^{-M}).$$
(3.10)

*Proof* For convenience, let z = (x, y) and dz = dx dy. Since  $f_2 \in \text{lip } 1(\mathbb{R}^2)$  and  $\int_{\mathbb{R}^2} \psi(z) dz = 0$ , we have

$$d_{\mu,m,n}^{(2)} = \int_{\mathbb{R}^2} f_2(z) \overline{\psi_{\mu,m,n}(z)} \, \mathrm{d}z$$
$$= \int_{\mathbb{R}^2} \left( f_2(z) - f_2\left(\frac{n}{2^m}\right) \right) \overline{\psi_{\mu,m,n}(z)} \, \mathrm{d}z,$$

so

$$\begin{aligned} \left| d_{\mu,m,n}^{(2)} \right| &= O(2^m) \int_{\mathbb{R}^2} \left| z - \frac{n}{2^m} \right| \left| \psi_{\mu}(2^m z - n) \right| \mathrm{d}z \\ &= O(2^{-2m}) \int_{\mathbb{R}^2} \left| \psi_{\mu}(z) \right| \mathrm{d}z = O(2^{-2m}). \end{aligned}$$

So (3.9) holds.

By (2.10), we know that the difference between  $f_2$  and the partial sum of its periodic wavelet series

$$f_2 - S_{2^{2M}}(f_2) = \sum_{\mu=1}^3 \sum_{m=M}^\infty \sum_{n=0}^{2^m-1} d_{\mu,m,n}^{(2)} \overline{\widetilde{\psi}_{\mu,m,n}^{\text{per}}}$$

is in the space  $L^2([-\frac{1}{2},\frac{1}{2}]^2)$ . So, we have

$$|f_2 - S_{2^{2M}}(f_2)||_{L^2([-\frac{1}{2},\frac{1}{2}]^2)}^2$$
  
=  $O(1) \sum_{\mu=1}^3 \sum_{m=M}^\infty \sum_{n_1=0}^{2^m-1} \sum_{n_2=0}^{2^m-1} |d_{\mu,m,n}^{(2)}|^2,$ 

where  $n = (n_1, n_2) \in \mathbb{Z}_+^2$ . Again, by  $d_{\mu,m,n}^{(2)} = O(2^{-2m})$ , we have

$$\|f_2 - S_{2^{2M}}(f_2)\|_{L^2([-\frac{1}{2},\frac{1}{2}]^2)}^2 = O(1) \sum_{m=M}^{\infty} 2^{-2m} = O(2^{-2M}).$$

So we get (3.10). Proposition 3.3 is proved.  $\hfill \Box$ 

(iii) *The 2D HWT algorithm.* We can easily prove that 1-periodic function  $v^*$  satisfies

$$\frac{\partial v^*}{\partial x}(x, y) \in \text{lip 1} \text{ and } \frac{\partial v^*}{\partial y}(x, y) \in \text{lip 1} \quad ((x, y) \in \mathbb{R}^2).$$
(3.11)

**Proposition 3.4** Let  $v^*$  be stated as above, i.e.,  $v^*$  is a 1-periodic function and (3.11) holds. Then the periodic wavelet coefficients of  $v^*$  satisfy

$$d_{\mu,m,n}^{(2)} = O(2^{-3m}).$$
(3.12)

The partial sum  $S_{2^{2M}}(v^*)$  of the periodic wavelet expansion of  $v^*$  satisfies

$$\|v^* - S_{2^{2M}}(v^*)\|_{L^2([-\frac{1}{2},\frac{1}{2}]^2)} = O(2^{-2M}).$$
(3.13)

*Proof* For z = (x, y) and  $z_0 = (x_0, y_0)$ , we have

$$v^{*}(z) - v^{*}(z_{0}) - \frac{\partial v^{*}}{\partial x}(z_{0})(x - x_{0}) - \frac{\partial v^{*}}{\partial y}(z_{0})(y - y_{0})$$

$$= \int_{x_{0}}^{x} \left(\frac{\partial v^{*}}{\partial x'}(x', y_{0}) - \frac{\partial v^{*}}{\partial x'}(x_{0}, y_{0})\right) dx'$$

$$+ \int_{y_{0}}^{y} \left(\frac{\partial v^{*}}{\partial y}(x, y) - \frac{\partial v^{*}}{\partial y}(x, y_{0})\right) dy$$

$$+ \int_{y_{0}}^{y} \left(\frac{\partial v^{*}}{\partial y}(x, y_{0}) - \frac{\partial v^{*}}{\partial y}(x_{0}, y_{0})\right) dy$$

$$=: J_{1} + J_{2} + J_{3}. \qquad (3.14)$$

By (3.11), we get

$$|J_1| \le K(x - x_0)^2 \le K|z - z_0|^2$$
,

$$|J_2| \le K (y - y_0)^2 \le K |z - z_0|^2,$$
  

$$|J_3| \le K |x - x_0| |y - y_0|$$
  

$$\le \frac{K}{2} ((x - x_0)^2 + (y - y_0)^2) = \frac{K}{2} |z - z_0|^2.$$

Again, by (3.14),

$$v^{*}(z) - v^{*}(z_{0}) - \frac{\partial v^{*}}{\partial x}(z_{0})(x - x_{0}) - \frac{\partial v^{*}}{\partial y}(z_{0})(y - y_{0}) \bigg|$$
  

$$\leq 3K|z - z_{0}|^{2}. \qquad (3.15)$$

By the definition of  $\psi_{\mu}(x, y)$  ( $\mu = 1, 2, 3$ ), we obtain that for  $\mu = 1, 2, 3$ ,

$$\int_{\mathbb{R}^2} x \psi_{\mu}(z) \, \mathrm{d}z = \int_{\mathbb{R}^2} y \psi_{\mu}(z) \, \mathrm{d}z = \int_{\mathbb{R}^2} \psi_{\mu}(z) \, \mathrm{d}z = 0.$$
(3.16)

By (2.9) and (3.16), taking  $z_0 = (x_0, y_0) = \frac{n}{2^m}$  (*n* =  $(n_1, n_2)$ ), we deduce that

$$d^{(2)}_{\mu,m,n} = 2^m \int_{\mathbb{R}^2} v^*(z) \,\overline{\psi_\mu(2^m z - n)} \,\mathrm{d}z$$
  
=  $2^m \int_{\mathbb{R}^2} \left( v^*(z) - v^*\left(\frac{n}{2^m}\right) - \frac{\partial v^*}{\partial x}\left(\frac{n}{2^m}\right) \left(x - \frac{n_1}{2^m}\right)$   
 $- \frac{\partial v^*}{\partial y} \left(\frac{n}{2^m}\right) \left(y - \frac{n_2}{2^m}\right) \right) \overline{\psi_\mu(2^m z - n)} \,\mathrm{d}z.$ 

From this and (3.15), noticing that  $x_0 = \frac{n_1}{2^m}$  and  $y_0 = \frac{n_2}{2^m}$ , we have

$$\begin{aligned} \left| d_{\mu,m,n}^{(2)} \right| &\leq 3K \, 2^m \int_{\mathbb{R}^2} \left| z - \frac{n}{2^m} \right|^2 \left| \psi_{\mu}(2^m z - n) \right| \mathrm{d}z \\ &\leq 3K \, 2^{-3m} \int_{\mathbb{R}^2} |z|^2 \left| \psi_{\mu}(z) \right| \mathrm{d}z = O(2^{-3m}). \end{aligned}$$
(3.17)

So (3.12) holds.

By the similar argument to (3.10) above, from (3.7) we now have

$$\|v^* - S_{2^{2M}}(v^*)\|_{L^2([-\frac{1}{2},\frac{1}{2}]^2)} = O(2^{-2M}).$$
  
Proposition 3.4 is proved.

# 4 Discrete HWT

In this section, we will discuss the discrete HWT and study the symmetry property of the coefficients.

We always assume that  $\psi$  and  $\tilde{\psi}$  be a pair of biorthonormal wavelets generated by the scaling functions  $\varphi$ ,  $\tilde{\varphi}$  that are compactly supported real-valued even functions. We also assume that  $\psi$  and  $\tilde{\psi}$  are real-valued functions. Let the filters  $\{a_k\}$  and  $\{b_k\}$  be defined in (2.5) and let us assume the formula (2.6) holds. Since  $\varphi$  and  $\tilde{\varphi}$  are symmetric at t = 0, by the known result [4], we know that  $\psi$  and  $\tilde{\psi}$  are symmetric at  $t = \frac{1}{2}$  and  $t = -\frac{1}{2}$ , respectively.

# 4.1 Definition of One-dimensional Discrete HWT

Let  $f \in C^2([0, \frac{1}{2}])$  be a signal of the interval  $[0, \frac{1}{2}]$ . We are given the discretized version of f sampled at  $\{\frac{n}{2^J}\}_{n=0}^{2^{J-1}}$ :

$$x_n = f\left(\frac{n}{2^J}\right) \quad (n = 0, \dots, 2^{J-1}).$$

Then  $x_0 = f(0)$  and  $x_{2^{J-1}} = f(\frac{1}{2})$ . Using the HWT decomposition, we get

$$f(t) = 2\left(f\left(\frac{1}{2}\right) - f(0)\right)t + f(0) + v(t)$$
  
=  $2(x_{2^{J-1}} - x_0)t + x_0 + v(t) \quad \left(0 \le t \le \frac{1}{2}\right).$ 

So, for  $n = 0, ..., 2^{J-1}$ ,

$$y_n := v\left(\frac{n}{2^J}\right) = x_n - (x_{2^{J-1}} - x_0)\frac{n}{2^{J-1}} - x_0$$

In particular,  $y_0 = y_{2^{J-1}} = 0$ .

Since the odd extension  $v_{\text{odd}}$  of v to  $\left[-\frac{1}{2}, \frac{1}{2}\right]$  satisfies

$$v_{\text{odd}}(t) = -v(-t) \quad \left(t \in \left[-\frac{1}{2}, 0\right]\right),$$

the sequence  $\{y_n^{\text{odd}}\}_{n=-2^{J-1}}^{2^{J-1}}$ , where  $y_n^{\text{odd}} := v_{\text{odd}}(\frac{n}{2^J})$ , satisfies the conditions

$$y_n^{\text{odd}} = -y_{-n}^{\text{odd}} \quad (-2^{J-1} \le n \le 2^{J-1}) \text{ and}$$
  
 $y_0^{\text{odd}} = y_{2^{J-1}}^{\text{odd}} = y_{-2^{J-1}}^{\text{odd}} = 0.$ 

Let  $v^*$  be a 1-periodic extension of  $v_{\text{odd}}$ . We define the sequence  $\{z_n\}_{n\in\mathbb{Z}}$ :

$$z_n := v^* \left(\frac{n}{2^J}\right) \quad (n \in \mathbb{Z}).$$
(4.1)

So we obtain that for  $n \in \mathbb{Z}$  and  $l \in \mathbb{Z}$ ,

$$z_{-n} = -z_n, \qquad z_{n+2^J} = z_n \quad \text{and} \quad z_{2^{J-1}l} = 0.$$
 (4.2)

Hence  $\{z_n\}_{n \in \mathbb{Z}}$  can be determined by  $2^{J-1} - 1$  different values  $z_1, \ldots, z_{2^{J-1}-1}$ .

# 4.2 The Relationships Between the Coefficients in the HWT Representation

Let  $\{c_{m,n}^{(1)}\}$  and  $\{d_{m,n}^{(1)}\}$  be periodic wavelet coefficients of  $v^* \in L^2([-\frac{1}{2}, \frac{1}{2}])$ . Taking J sufficiently large, the following

$$v^*(t) \simeq \sum_{n=0}^{2^J - 1} c_{J,n}^{(1)} \widetilde{\varphi}_{J,n}^{\text{per}}(t)$$

and

$$c_{J,n}^{(1)} \simeq v^* \left(\frac{n}{2^J}\right) = z_n.$$
 (4.3)

By (2.4), we get

$$\sum_{k=0}^{2^{J}-1} c_{J,k}^{(1)} \widetilde{\varphi}_{J,k}^{\text{per}} = \sum_{k=0}^{2^{J-1}-1} c_{J-1,k}^{(1)} \widetilde{\varphi}_{J-1,k}^{\text{per}} + \sum_{k=0}^{2^{J-1}-1} d_{J-1,k}^{(1)} \widetilde{\psi}_{J-1,k}^{\text{per}}.$$

By Proposition 2.1, we know that

$$c_{J-1,k}^{(1)} = \sum_{n=0}^{2^{J}-1} a_{n-2k}^{*} z_{n} \quad \text{and} \quad d_{J-1,k}^{(1)} = \sum_{n=0}^{2^{J}-1} b_{n-2k}^{*} z_{n},$$
(4.4)

where  $\{a_n^*\}$  and  $\{b_n^*\}$  are stated in (2.7).

Proposition 4.1 Let 
$$c_{J-1,k}^{(1)} = \alpha_k \ (k = 1, ..., 2^{J-2} - 1)$$
 and  
 $d_{J-1,k}^{(1)} = \beta_k \ (k = 0, ..., 2^{J-2} - 1)$ . Then  
(i)  $\{c_{J-1,k}^{(1)}\}_{k=0}^{2^{J-1}-1} = \{0, \alpha_1, ..., \alpha_{2^{J-2}-1}, ..., -\alpha_1\};$   
(ii)  $\{d_{J-1,k}^{(1)}\}_{k=0}^{2^{J-1}-1} = \{\beta_0, ..., \beta_{2^{J-2}-1}, ..., -\beta_0\}.$ 

We need a couple of lemmas.

**Lemma 4.2** Let  $\{a_n^*\}$  and  $\{b_n^*\}$  be stated in (2.7). Then, for any  $n \in \mathbb{Z}$ ,

$$a_{-n}^* = a_n^*$$
 and  $b_n^* = b_{2-n}^*$ 

*Proof* Since  $\varphi$ ,  $\tilde{\varphi}$  are both real-valued even functions and  $a_n = \sqrt{2} \int_{\mathbb{R}} \varphi(t) \overline{\tilde{\varphi}(2t-n)} dt$  (by (2.5)), we have

$$a_{-n} = \sqrt{2} \int_{\mathbb{R}} \varphi(t) \widetilde{\varphi}(2t+n) dt$$
$$= \sqrt{2} \int_{\mathbb{R}} \varphi(-t) \widetilde{\varphi}(-2t+n) dt$$
$$= \sqrt{2} \int_{\mathbb{R}} \varphi(t) \widetilde{\varphi}(2t-n) dt = a_n.$$

From this and (2.7), we get

$$a_{-n}^* = a_n^*, \qquad a_{2^J+n}^* = a_n^* \quad (n \in \mathbb{Z}).$$
 (4.5)

By (2.5),  $b_n = \sqrt{2} \int_R \psi(t) \widetilde{\varphi}(2t-n) dt$ . Since  $\psi(\frac{1}{2}-t) = \psi(\frac{1}{2}+t)$ , we have

$$b_n = \sqrt{2} \int_{\mathbb{R}} \psi\left(\frac{1}{2} + t\right) \widetilde{\varphi}(2t + 1 - n) dt$$
  
$$= \sqrt{2} \int_{\mathbb{R}} \psi\left(\frac{1}{2} - t\right) \widetilde{\varphi}(2t + 1 - n) dt$$
  
$$= \sqrt{2} \int_{\mathbb{R}} \psi(t) \widetilde{\varphi}(-2t + 2 - n) dt$$
  
$$= \sqrt{2} \int_{\mathbb{R}} \psi(t) \widetilde{\varphi}(2t - 2 + n) dt$$
  
$$= b_{2-n}.$$

So we get  $b_{1+n} = b_{1-n}$ . By (2.7), we have

$$b_{1+n}^* = b_{1-n}^* \quad (|n| \le 2^{J-1}).$$

Again since  $b_{n+2^J}^* = b_n^*$ , we have  $b_{1+n}^* = b_{1-n}^*$  for all  $n \in \mathbb{Z}$ . So we get

$$b_n^* = b_{2-n}^* \quad (n \in \mathbb{Z}).$$
 (4.6)

**Lemma 4.3** The periodic wavelet coefficients  $\{c_{m,n}^{(1)}\}$  and  $\{d_{m,n}^{(1)}\}$  satisfy the following relationships

$$c_{J-1,k}^{(1)} = -c_{J-1,2^{J-1}-k}^{(1)},$$
  

$$d_{J-1,k}^{(1)} = -d_{J-1,2^{J-1}-1-k}^{(1)} \quad (k \in \mathbb{Z}).$$
(4.7)

*Proof* From (4.2) and (4.5), it follows by (4.4) that

$$c_{J-1,2^{J-1}-k}^{(1)} = \sum_{n=0}^{2^{J}-1} a_{n+2k-2^{J}}^{*} z_{n}$$
$$= -\sum_{n=0}^{2^{J}-1} a_{n+2k}^{*} z_{-n} = -\sum_{n=-2^{J}+1}^{0} a_{-n+2k}^{*} z_{n}.$$

Again thanks to the periodicity of the  $\{a_n\}$  and  $\{z_n\}$ , we have

$$\sum_{n=-2^{J}+1}^{0} a_{-n+2k}^{*} z_{n} = \sum_{n=1}^{2^{J}} a_{-n+2k}^{*} z_{n}$$
$$= \sum_{n=1}^{2^{J}-1} a_{-n+2k}^{*} z_{n} + a_{2k-2^{J}}^{*} z_{2^{J}}$$
$$= \sum_{n=1}^{2^{J}-1} a_{-n+2k}^{*} z_{n} + a_{2k} z_{0}$$
$$= \sum_{n=0}^{2^{J}-1} a_{-n+2k}^{*} z_{n}.$$
(4.8)

So we have

$$c_{J-1,2^{J-1}-k}^{(1)} = -\sum_{n=0}^{2^{J}-1} a_{-n+2k}^* z_n$$
$$= -\sum_{n=0}^{2^{J}-1} a_{n-2k}^* z_n = -c_{J-1,k}^{(1)}.$$
(4.9)

On the other hand, by (4.4) and (4.6),

$$d_{J-1,k}^{(1)} = \sum_{n=0}^{2^{J}-1} b_{n-2k}^{*} z_{n} = \sum_{n=0}^{2^{J}-1} b_{2-n+2k}^{*} z_{n}$$
$$= -\sum_{n=0}^{2^{J}-1} b_{2-n+2k}^{*} z_{-n}$$
$$= -\sum_{n=-2^{J}+1}^{0} b_{2+n+2k}^{*} z_{n}.$$
(4.10)

By the similar argument to (4.8), we have

$$\sum_{n=-2^{J}+1}^{0} b_{2+n+2k}^{*} z_{n} = \sum_{n=0}^{2^{J}-1} b_{2+n+2k}^{*} z_{n}.$$

So we get

$$d_{J-1,k}^{(1)} = -\sum_{n=0}^{2^{J}-1} b_{n-(2^{J}-2k-2)}^{*} z_{n} = -d_{J-1,2^{J-1}-1-k}^{(1)}.$$

*Proof of Proposition 4.1* By the first formula of (4.7), we get  $c_{J-1,2^{J-2}}^{(1)} = -c_{J-1,2^{J-2}}^{(1)}$ , i.e.,

$$c_{J-1,2^{J-2}}^{(1)} = 0. (4.11)$$

Since  $a_n^* = a_{-n}^* = a_{2^J-n}^*$  and  $z_n = -z_{-n} = -z_{2^J-n}$ , by (4.4), we have

$$c_{J-1,0}^{(1)} = \sum_{n=0}^{2^{J}-1} a_{n}^{*} z_{n} = \sum_{n=1}^{2^{J}} a_{n}^{*} z_{n}$$
$$= -\sum_{n=1}^{2^{J}} a_{2^{J}-n}^{*} z_{2^{J}-n} = -\sum_{n=0}^{2^{J}-1} a_{n}^{*} z_{n} = -c_{J-1,0}^{(1)}.$$

So  $c_{J-1,0}^{(1)} = 0$ . Let  $c_{J-1,k}^{(1)} = \alpha_k \ (k = 0, \dots, 2^{J-1} - 1)$ . Then, by  $c_{J-1,0}^{(1)} = 0$ , and (4.9) and (4.11), we get

$$\alpha_0 = 0, \qquad \alpha_{2^{J-2}} = 0, \text{ and}$$
  
 $\alpha_k = -\alpha_{2^{J-1}-k} \quad (k = 1, \dots, 2^{J-2} - 1)$ 

i.e., (i) holds. Let  $d_{J-1,k}^{(1)} = \beta_k$   $(k = 0, ..., 2^{J-1} - 1)$ . By the second formula of (4.7), we have

$$\beta_k = -\beta_{2^{J-1}-1-k}$$
 (k = 0, ...,  $2^{J-2} - 1$ ),  
i.e., (ii) holds. Proposition 4.1 is proved.

From Proposition 4.1, we see that in order to recover the signal, we only need  $2^{J-1} - 1$  periodic wavelet coefficients and two boundary values of the signal f. On the other hand, the number of the sampling points  $\{x_n\}_0^{2^{J-1}}$  is just  $2^{J-1} + 1$ . So the 1D HWT is not a redundant transform.

#### 4.3 Two-dimensional Discrete HWT

We will now discuss the two-dimensional discrete HWT and study the symmetry property of the coefficients.

Let an image  $f \in C^2([0, \frac{1}{2}]^2)$ . For some large J, take the  $(2^{J-1} + 1)^2$  sample points of f

$$x_{n_1,n_2} = f\left(\frac{n_1}{2^J}, \frac{n_2}{2^J}\right) \quad (n_1, n_2 = 0, \dots, 2^{J-1}).$$

Using the HWT decomposition, we get

$$f(x, y) = u(x, y) + v(x, y) \quad \left( (x, y) \in \left[ 0, \frac{1}{2} \right]^2 \right),$$

where u(x, y) is a harmonic function satisfying Laplace's equation  $\Delta u = 0$  and u = f on the boundary of  $[0, \frac{1}{2}]^2$ . We can efficiently and accurately compute *u* by using the Averbuch-Israeli-Vozovoi (AIV) method [2]. Let us now discuss the residual component *v*.

Let  $y_{n_1,n_2} = v(\frac{n_1}{2J}, \frac{n_2}{2J})$ . Then

$$y_{0,n_2} = y_{n_1,0} = y_{2^{J-1},n_2} = y_{n_1,2^{J-1}} = 0$$
  
 $(n_1, n_2 = 0, \dots, 2^{J-1}).$ 

Let  $v_{odd}$  be an odd extension of v to  $\left[-\frac{1}{2}, \frac{1}{2}\right]^2$ , i.e.,

$$v_{\text{odd}}(x, y) = v(x, y) \quad \left( (x, y) \in \left[ 0, \frac{1}{2} \right]^2 \right)$$

and

$$v_{\text{odd}}(x, y) = -v_{\text{odd}}(-x, y) = -v_{\text{odd}}(x, -y)$$
$$= v_{\text{odd}}(-x, -y) \quad \left( (x, y) \in \left[ -\frac{1}{2}, \frac{1}{2} \right]^2 \right).$$

Denote  $z_{n_1,n_2} = v_{\text{odd}}(\frac{n_1}{2^J}, \frac{n_2}{2^J})$ . Then

$$z_{n_1,n_2} = -z_{-n_1,n_2} = -z_{n_1,-n_2} = z_{-n_1,-n_2}$$
  
(n\_1, n\_2 = 0, ±1, ..., ±2<sup>J-1</sup>) (4.12)

and

$$z_{0,n_2} = z_{n_1,0} = z_{2^{J-1},n_2} = z_{n_1,2^{J-1}} = z_{-2^{J-1},n_2}$$
$$= z_{n_1,-2^{J-1}} = 0.$$

Let  $v^*$  be a 1-periodic extension of  $v_{odd}$  to  $\mathbb{R}^2$ . Denote  $z^*_{n_1,n_2} = v^*(\frac{n_1}{2J}, \frac{n_2}{2J})$ . Then

$$z_{n_1,n_2}^* = z_{n_1,n_2} \quad (n_1, n_2 = 0, \pm 1, \dots, \pm 2^{J-1}),$$
  

$$z_{n_1+2^J,n_2}^* = z_{n_1,n_2}^*, \qquad z_{n_1,n_2+2^J}^* = z_{n_1,n_2}^* \qquad (4.13)$$
  

$$(n_1 \in \mathbb{Z}, n_2 \in \mathbb{Z}).$$

For  $\mu = 1, 2, 3, m \in \mathbb{Z}, n \in \mathbb{Z}^2$ , let  $c_{m,n_1,n_2}^{(2)}$  and  $d_{\mu,m,n_1,n_2}^{(2)}$ be the periodic wavelet coefficients of  $v^* \in L^2([-\frac{1}{2}, \frac{1}{2}]^2)$ (see (2.9)).

Take J sufficiently large such that

$$v^*(x, y) \simeq \sum_{n_1, n_2=0}^{2^J - 1} c_{J,n}^{(2)} \widetilde{\varphi}_{0,J,n}^{\text{per}}$$
 and  
 $c_{J,n}^{(2)} \simeq v^* \left(\frac{n_1}{2^J}, \frac{n_2}{2^J}\right) = z_{n_1, n_2}^*, \quad n = (n_1, n_2).$ 

Again, by (2.11), we have

$$v^{*}(x, y) \simeq \sum_{n_{1}, n_{2}=0}^{2^{J-1}-1} c_{J-1, n_{1}, n_{2}}^{(2)} \widetilde{\varphi}_{0, J-1, n_{1}, n_{2}}^{\text{per}} + \sum_{\mu=1}^{3} \sum_{n_{1}, n_{2}=0}^{2^{J-1}-1} d_{\mu, J-1, n_{1}, n_{2}}^{(2)} \widetilde{\psi}_{\mu, J-1, n_{1}, n_{2}}^{\text{per}}.$$

$$(4.14)$$

Now we discuss the symmetry property of the coefficients  $c_{J-1,k_1,k_2}^{(2)}$  and  $d_{\mu,J-1,k_1,k_2}^{(2)}$ . From Proposition 2.2, Lemma 4.2, (4.12), and (4.13), we can get

**Proposition 4.4** For  $k_1, k_2 \in \mathbb{Z}$ , we have

(i) 
$$c_{J-1,2^{J-1}-k_{1},k_{2}}^{(2)} = -c_{J-1,k_{1},k_{2}}^{(2)},$$
  
 $c_{J-1,k_{1},2^{J-1}-k_{2}}^{(2)} = -c_{J-1,k_{1},k_{2}}^{(2)},$   
 $c_{J-1,2^{J-2},k_{2}}^{(2)} = c_{J-1,0,k_{2}}^{(2)} = c_{J-1,k_{1},2^{J-2}}^{(2)},$   
 $= c_{J-1,k_{1},0}^{(2)} = 0;$   
(ii)  $d_{1,J-1,2^{J-1}-k_{2},k_{2}}^{(2)} = -d_{1,J-1,k_{2},k_{2}}^{(2)},$ 

$$\begin{aligned} d_{1,J-1,2^{J-1}-k_{1},k_{2}}^{(2)} &= -d_{1,J-1,k_{1},k_{2}}^{(2)}, \\ d_{1,J-1,k_{1},2^{J-1}-k_{2}-1}^{(2)} &= -d_{1,J-1,k_{1},k_{2}}^{(2)}, \\ d_{1,J-1,2^{J-2},k_{2}}^{(2)} &= d_{1,J-1,0,k_{2}}^{(2)} = 0; \end{aligned}$$

(iii) 
$$d_{2, J-1, 2^{J-1}-k_1-1, k_2}^{(2)} = -d_{2, J-1, k_1, k_2}^{(2)},$$
  
 $d_{2, J-1, k_1, 2^{J-1}-k_2}^{(2)} = -d_{2, J-1, k_1, k_2}^{(2)},$   
 $d_{2, J-1, k_1, 2^{J-2}}^{(2)} = d_{2, J-1, k_1, 0}^{(2)} = 0;$  and  
(iv)  $d_{3, J-1, 2^{J-1}-k_1-1, k_2}^{(2)} = -d_{3, J-1, k_1, k_2}^{(2)},$ 

$$d_{3, J-1, k_1, 2^{J-1}-k_2-1}^{(2)} = -d_{3, J-1, k_1, k_2}^{(2)}.$$

From Proposition 4.4, the symmetry of the matrix  $(c_{J-1,k_1,k_2}^{(2)}), k_1, k_2 = 0, 1, \dots, 2^{J-1} - 1$  is described as follows:

( 0	0		0	0	0		0)
0	$\alpha_{1,1}$	•••	$\alpha_{1,2^{J-2}-1}$	0	$-\alpha_{1,2^{J-2}-1}$	•••	$-\alpha_{1,1}$
1 :	÷	÷		÷		÷	:
0	$\alpha_{2^{J-2}-1,1}$		$\alpha_{2^{J-2}-1,2^{J-2}-1}$	0	$-\alpha_{2^{J-2}-1,2^{J-2}-1}$		$-\alpha_{2^{J-2}-1,1}$
0	0		0	0	0		0
0	$-\alpha_{2^{J-2}-1,1}$	•••	$-\alpha_{2^{J-2}-1,2^{J-2}-1}$	0	$\alpha_{2^{J-2}-1,2^{J-2}-1}$		$\alpha_{2^{J-2}-1,1}$
1 :	:	÷		÷	:	÷	:
0	$-\alpha_{1,1}$		$-\alpha_{1,2^{J-2}-1}$	0	$\alpha_{2^{J-2}-1,2^{J-2}-1}$		$\alpha_{2^{J-2}-1,1}$

where

$$\alpha_{k_1,k_2} = c_{J-1,k_1,k_2}^{(2)}$$
  $(k_1,k_2=1,\ldots,2^{J-2}-1).$ 

Therefore, the matrix  $(c_{J-1,k_1,k_2}^{(2)}), k_1, k_2 = 0, 1, ..., 2^{J-1} - 1$ , is determined by  $(2^{J-2} - 1)^2$  values. Similarly, we have

- (i) the matrix (d<sup>(2)</sup><sub>1,J-1,k1,k2</sub>), k<sub>1</sub>, k<sub>2</sub> = 0, ..., 2<sup>J-1</sup> − 1, is determined by 2<sup>J-2</sup>(2<sup>J-2</sup> − 1) values;
  (ii) the matrix (d<sup>(2)</sup><sub>2,J-1,k1,k2</sub>), k<sub>1</sub>, k<sub>2</sub> = 0, ..., 2<sup>J-1</sup> − 1, is determined by 2<sup>J-2</sup>(2<sup>J-2</sup> − 1) values; and
- (iii) the matrix  $(d_{3,J-1,k_1,k_2}^{(2)})$ ,  $k_1, k_2 = 0, \dots, 2^{J-1} 1$ , is determined by  $2^{2J-4}$  values.

Noticing that

$$(2^{J-2} - 1)^2 + 2^{J-2}(2^{J-2} - 1) + 2^{J-2}(2^{J-2} - 1) + 2^{2J-4}$$
  
=  $(2^{J-1} - 1)^2$ 

we know that in order to recover  $v^*$ , we only need  $(2^{J-1}-1)^2$  periodic wavelet coefficients. To obtain the harmonic function *u*, we need  $4(2^{J-1}-1)+4$  boundary sample points of the image f on  $\partial([0, \frac{1}{2}]^2)$ . Since the number

$$(2^{J-1}-1)^2 + 4(2^{J-1}-1) + 4 = (2^{J-1}+1)^2$$

is exactly equal to the number of sampling points of f, the 2D HWT is not redundant.

# 5 Image Approximation Experiments via HWT

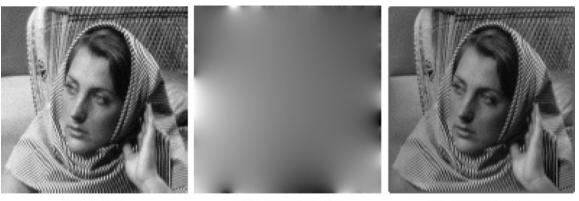
We will examine the approximation performance of the 2D HWT algorithm. The quality of approximation in this paper is measured by PSNR (or peak signal-to-noise ratio) defined as

$$PSNR := 20 \times \log_{10} \left( \max_{x \in \overline{\Omega}} |f(x)| / RMSE \right)$$

where RMSE is the absolute  $\ell^2$  error between the original and the approximation divided by the square root of the total number of pixels in the original image. The unit of PSNR is decibel (dB).

# 5.1 Comparison with the Periodic and Folded Wavelet Algorithms

To approximate an image sampled on a square by the 2D HWT algorithm, we first decompose the image into the harmonic component u and the residual v (see Fig. 1). The harmonic component u is determined by the data at boundary of the square. Hence in order to approximate u, it suffices to approximate the data on the boundary of the square. Since the boundary consists of four segments, we simply apply one-dimensional LLST on each segment. For the residual v, we do an odd extension and a periodic extension (see Fig. 2) and expand it into a periodic wavelet series with respect to a biorthonormal periodic wavelet basis with the symmetric filter bank (see Fig. 3). Our image approximation and reconstruction strategy consists of the following steps: (1) Retain



(a) Original

(b) The u component

(c) The v component

Fig. 1 HWT decomposition of the Barbara face image. Each image is displayed using its full dynamic range



Fig. 2 The odd extension of the residual component v

all the "DC" components of the 1D LLST of each boundary segment of the square; (2) Select a certain number of the largest coefficients in terms of energy from the rest of the coefficients (both 1D and 2D); (3) Reconstruct u and vfrom these retained coefficients using the AIV algorithm [2] and the Mallat algorithm, respectively; and (4) Compute u + v.

For approximation and reconstruction of the image by the periodic wavelet algorithm and the folded wavelet algorithm, we simply retain a certain number of the largest coefficients in terms of energy from all the coefficients and reconstruct the image from them.

First, we compare the performance of HWT with that of the periodic wavelet transform (PWT) and the folded wavelet transform (FWT). We use the 9/7 biorthogonal filter bank (see [7, Sect. 7.4] for the actual filter coefficients). The depths of decomposition J we test here are  $J = 2, 4, \log_2(N)$  for an image of size  $N \times N$ . The original image sizes we use are all dyadic, i.e.,  $N = 2^n$ , which are

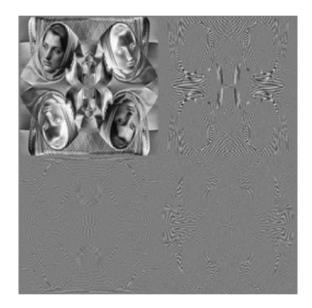
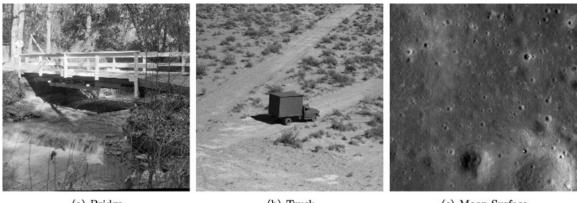


Fig. 3 The periodic biorthonormal wavelet coefficients of the odd extension of the residual component v when the depth of decomposition is one

suitable for PWT and FWT. Hence, for HWT, we duplicate the last column and row of each image to make it suitable for HWT.

As for the images, we use the face part of Barbara image (with  $128 \times 128$  pixels) as well as the standard images "Bridge", "Truck", and "Moon surface", each of which consists of  $256 \times 256$  pixels. Figure 4 shows the latter three images.

Figures 5, 6, and 7 show the quality of approximations of these four images when the depth of decomposition J is set to two, four, and maximum (seven for the Barbara face image and eight for the other three images), respectively. From these figures, the performance of HWT is consistently superior to that of PWT and FWT. For J = 2, we observe that the big performance difference between HWT and that of



(a) Bridge

(b) Truck

(c) Moon Surface

Fig. 4 The three more standard images used in our experiments

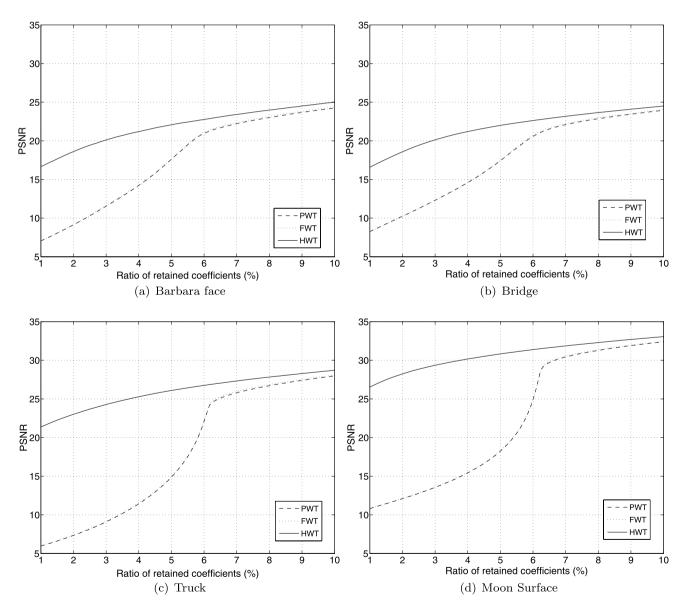


Fig. 5 Quality of approximation of the four standard images measured by PSNR values as a function of the ratio of the number of the retained coefficients to the total number of the coefficients. The depth of decomposition is set to two in each case

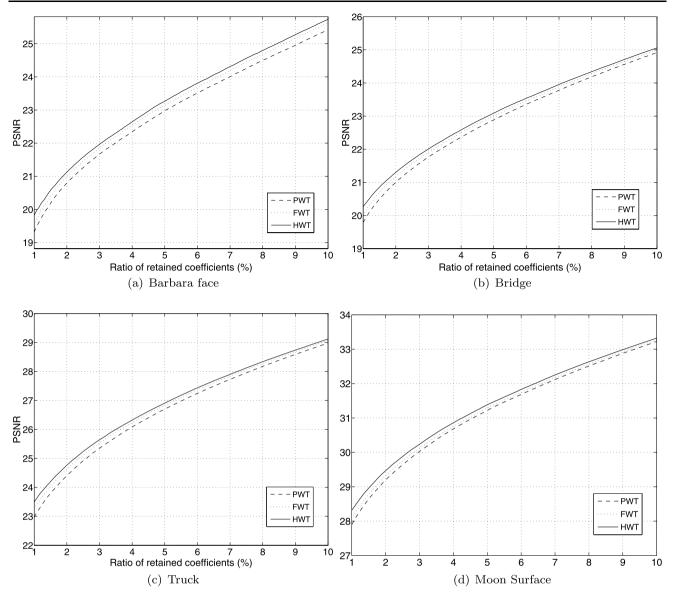


Fig. 6 Quality of approximation of the four standard images measured by PSNR values as a function of the ratio of the number of the retained coefficients to the total number of the coefficients. The depth of decomposition is set to four in each case

PWT and FWT particularly when the ratio f the retained coefficients is less than about 6%. This can be explained as follows. For any image of size  $N \times N$ , if the depth of decomposition J is set to 2, then the number of low-pass wavelet coefficients in PWT and FWT is  $N^2/16$ , i.e., very large. Since the low-pass wavelet coefficients are not sparse, we need many coefficients to approximate such an image well. In other words, we cannot efficiently approximate such an image using a small number of coefficients if J is set to a small number such as 2. In fact,  $N^2/16$  is 6.25% of the original image size  $N^2$ , which agrees with our observation. On the other hand, HWT decomposes an image into two components: the harmonic component and the residual. We can approximate the harmonic component very well us-

ing a few 1D LLST coefficients. For the residual, we still have the same problem as PWT and FWT if J = 2. However, since the norm of the residual is much smaller than that of the original image thanks to the removal of the harmonic component, the reconstruction error is kept small.

When J = 4 or J is set to its maximum, the performance of HWT is about  $0.2 \sim 0.5$  dB better than that of PWT, and the  $0.1 \sim 0.2$  dB better than that of FWT. In general, the smaller the number of the retained coefficients, the clearer the performance difference becomes. This is also due to the harmonic component in HWT.

We now would like to show the reconstructed images because the PSNR plots do not tell the whole story. Figures 8, 9, and 10 show the reconstructed Barbara face images using

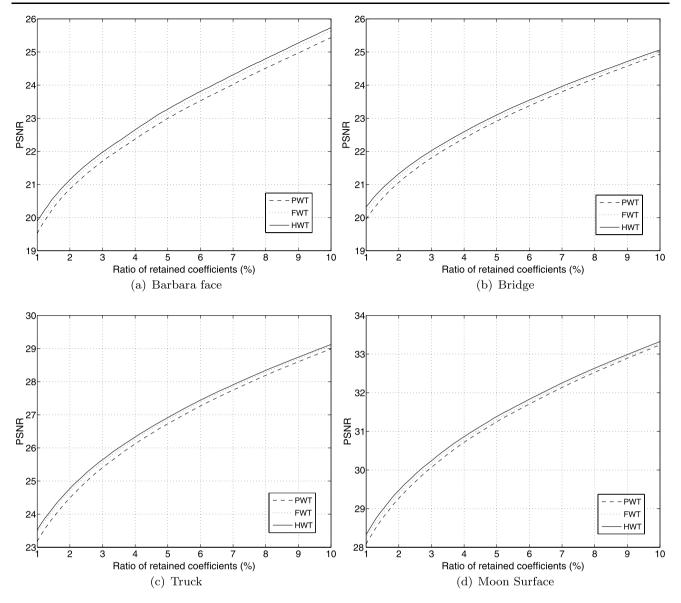


Fig. 7 Quality of approximation of the four standard images measured by PSNR values as a function of the ratio of the number of the retained coefficients to the total number of the coefficients. The depth of decomposition is set to the maximum in each case, i.e., seven for the Barbara face image and eight for the others

top 5% coefficients of PWT, FWT, and HWT for J = 2, 4, 7, respectively. From these figures, we can see that the HWT algorithm is also *perceptually* better than the PWT algorithm and the FWT algorithm, particularly for J = 2. We can also notice some difference around the frame boundary of the images. This is due to the use of LLST on the boundary pixels in HWT. Overall, for each algorithm, the difference between J = 4 and J = 7 is not noticeable.

# 5.2 Comparison with LLST

We now compare the HWT algorithm with the LLST algorithm. In general, the decay rate of LLST coefficients depends on global smoothness of the input data while that of HWT coefficients depends on local smoothness. Since the global smoothness of a function is determined by its rough part (even if that part is localized), we need fewer HWT coefficients than LLST coefficients in order to reconstruct the data to the same quality. For LLST, we divide the image into several blocks and do LLST on each block. The size of block we use in our experiments is  $9 \times 9$ ,  $17 \times 17$ ,  $33 \times 33$ ,  $65 \times 65$ ,  $129 \times 129$  for all the four images, and in addition, we use  $257 \times 257$  block for the "Bridge", "Truck", and "Moon Surface" images. Note that the largest block size for each image implies that LLST does not divide it into a set of smaller segments. We first retain all the corner pixel values of each block, select the certain number of coefficients



(a) Original Image





(c) FWT

(d) HWT

Fig. 8 Approximations of the Barbara face image using the top 5% coefficients when the depth of decomposition is two. The PSNR values (in dB) of PWT, FWT, and HWT are 17.6182, 17.5961, and 22.0776, respectively

with the largest energy among all the coefficients (both 1D and 2D) not yet used, reconstruct an approximation from these coefficients, and finally evaluate its quality of approximation. For HWT, the depth of decomposition is set to the maximum, and we also use the 9/7 biorthonormal filter bank. We apply HWT on the whole image and retain the coefficients with the largest energy. Figure 11 shows the quality of approximation by PSNR values when we retain 2%–20% of the original coefficients. From this figure, we again observe that HWT consistently outperforms LLST regardless of the block sizes. In particular, HWT's performance is

significantly better (about 1dB) than that of LLST for the "Bridge", "Truck", and "Moon Surface" images while its performance on "Barbara face" image is closely followed by that of LLST with  $17 \times 17$  blocks, especially around the ratio of the number of the retained coefficients to that of the original coefficients ranges around 8% to 10%. We believe that this is due to the existence of textures in the Barbara face image. We also observe that the performance of LLST at a particular ratio of the number of the retained coefficients to that of the that of the whole coefficient strongly depends on the block size. For example, for heavy compression (i.e., the ratio is



(c) FWT

(d) HWT

Fig. 9 Approximations of the Barbara face image using the top 5% coefficients when the depth of decomposition is four. The PSNR values (in dB) of PWT, FWT, and HWT are 22.9791, 23.1484, and 23.2651, respectively

about 2% to 8%), the LLST with the larger block sizes perform better than that with the smaller block sizes while the situation is opposite if one can afford to keep a larger number of coefficients (e.g., the ratio is larger than 10%). HWT does not require the user to choose any such block sizes, which is one of the major advantage of HWT over LLST.

# 5.3 Comparison with Wavelets on the Interval

It is appropriate to compare HWT with wavelets on the interval (WOI) introduced by Cohen, Daubechies and Vial [5] because WOI also tries to overcome boundary effects. Compared to HWT, the periodic wavelets, and folded wavelets, however, the construction of WOI is quite complicated. Their starting point is Daubechies's compactly supported scaling functions and wavelets. They first construct scaling functions on the interval consisting of three parts: the left edge scaling function, the interior scaling function, and the right edge scaling function. Then, they use these scaling functions on the interval to construct the corresponding wavelets also consisting of three parts: the left edge wavelet, the interior wavelet, and the right edge wavelet. Although



(c) FWT

(d) HWT

Fig. 10 Approximations of the Barbara face image using the top 5% coefficients when the depth of decomposition is seven. The PSNR values (in dB) of PWT, FWT, and HWT are 22.9962, 23.1476, and 23.2688, respectively

these wavelets on the interval have high vanishing moments, we cannot apply its discrete version to image approximation immediately. This is because their corresponding high pass filter cannot map a simple polynomial sequence to zero. So when one uses WOI to approximate images, one has to perform a prefiltering (or preconditioning) on the data first; see [5] for the detail.

We now compare the performance of HWT with that of WOI. Since the harmonic component of HWT takes care of linear parts on the boundary, in order to be a fair comparison with HWT, we use WOI with two vanishing moments. For HWT, we use Villasenor 5/3 filter bank [10], which also has two vanishing moments. Figure 12 shows the quality of approximation measured by PSNR values when we retain 1%–10% of the original coefficients. The depth of decomposition for both HWT and WOI is set to five for the Barbara face image and six for the other three images since these are the maximal depth of decomposition that the WOI can take. From this figure, we observe that HWT again consistently outperforms WOI. Considering the implementation complexity of WOI, HWT should be used if one wants to reduce the boundary effects.

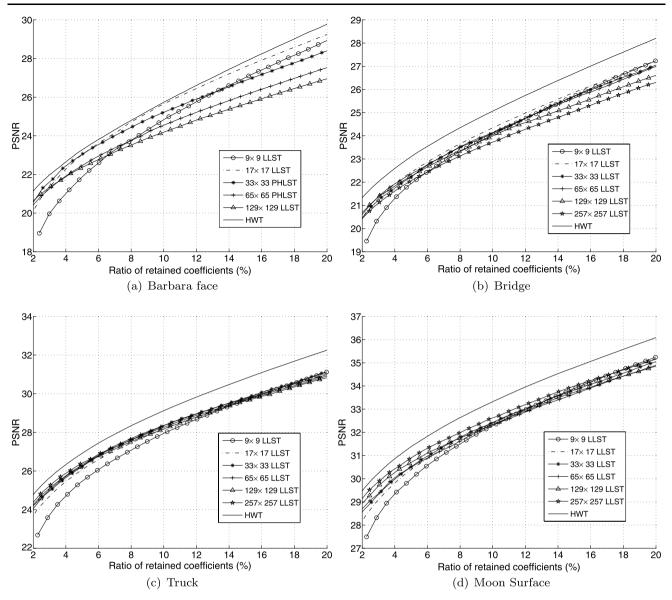


Fig. 11 Quality of approximation of the four standard images measured by PSNR values as a function of the ratio of the number of the retained coefficients to the total number of the coefficients. These figures compare the performance of HWT with that of LLST with different block sizes. For HWT, the depth of decomposition is set to its maximum level

# 6 Conclusion

In this paper, we improved PWT and FWT and proposed the Harmonic Wavelet Transform (HWT) that is not affected by the boundary of input data. The idea of removing the boundary mismatches of input data originally proposed in LLST [8] is quite important since the residual component after odd reflection and periodization, say, the  $v^*$  component, does not contain any artificial discontinuities caused by the boundary mismatches. Hence, the expansion coefficients of  $v^*$  with respect to the periodic wavelet basis truly reflect the local smoothness of an input image. Moreover, HWT captures the intrinsic singularities in the interior of the domain more efficiently than LLST that uses the Fourier sine series expansion

for  $v^*$ . Also, the implementation of HWT is simpler than WOI because HWT does not need any special boundarydependent filter banks. Finally, our image approximation experiments using four standard images demonstrated the superiority of HWT over LLST, PWT, FWT, and WOI. We also note that the extension of HWT to a higher dimension is straightforward thanks to the efficient Laplace solver for higher dimension [3].

Acknowledgements This research was partially supported by the NSF grant DMS-0410406, the ONR grants N00014-07-1-0166, N00014-09-1-0041, and N00014-09-1-0318. For computing the periodic and folded wavelet transforms as well as the wavelets on interval, we used the WaveLab software developed by Prof. David Donoho of Stanford Univ. and his team [6].

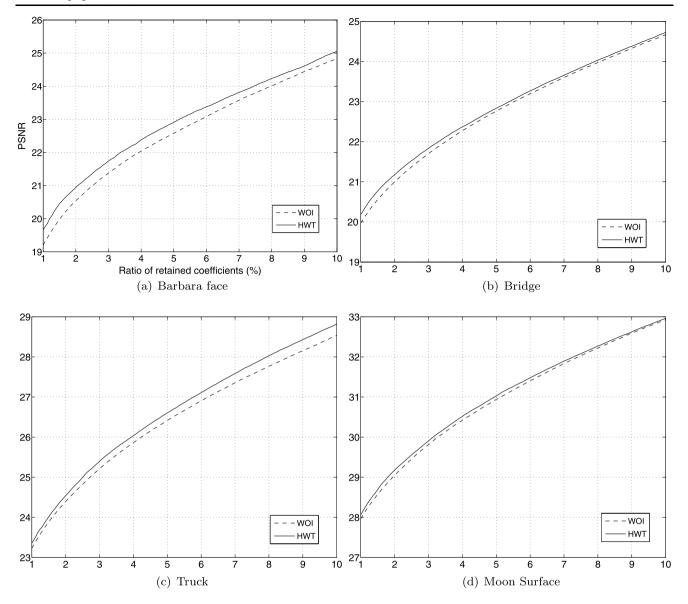


Fig. 12 Quality of approximation of the four standard images measured by PSNR values as a function of the ratio of the number of the retained coefficients to the total number of the coefficients. These figures compare the performance of HWT with that of WOI. For both HWT and WOI, the depth of decomposition is set to its maximum possible level

**Open Access** This article is distributed under the terms of the Creative Commons Attribution Noncommercial License which permits any noncommercial use, distribution, and reproduction in any medium, provided the original author(s) and source are credited.

# References

- Antonini, M., Barlaud, M., Mathieu, P., Daubechies, I.: Image coding using wavelet transform. IEEE Trans. Image Process. 1(2), 205–220 (1992)
- Averbuch, A., Israeli, M., Vozovoi, L.: A fast Poisson solver of arbitrary order accuracy in rectangular regions. SIAM J. Sci. Comput. 19, 933–952 (1998)

- Braverman, E., Israeli, M., Averbuch, A., Vozovoi, L.: A fast 3D Poisson solver of arbitrary order accuracy. J. Comput. Phys. 144, 109–136 (1998)
- 4. Chui, C.K.: An Introduction to Wavelets. Academic Press, San Diego (1992)
- Cohen, A., Daubechies, I., Vial, P.: Wavelet bases on the interval and fast algorithms. Appl. Comput. Harmon. Anal. 1, 54–81 (1993)
- Donoho, D., et al.: http://www-stat.stanford.edu/~wavelab. Accessed April 2010
- Mallat, S.: A Wavelet Tour of Signal Processing, 3rd edn. Academic Press, San Diego (2009)
- Saito, N., Remy, J.: The polyharmonic local sine transform: A new tool for local image analysis and synthesis without edge effect. Appl. Comput. Harmon. Anal **31**, 41–73 (2006)

- Van De Ville, D., Blu, T., Unser, M.: Isotropic polyharmonic B-splines: Scaling functions and wavelets. IEEE Trans. Image Process. 14(11), 1798–1813 (2005)
- Villasenor, J.D., Belzer, B., Liao, J.: Wavelet filter evaluation for image compression. IEEE Trans. Image Process. 4(8), 1053–1060 (1995)

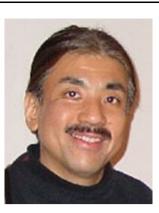


Zhihua Zhang received the B.S. degree from Shandong University (China) in 2001, the M.S. degree from Chinese Academy of Sciences in 2004, and the Ph.D. degree from the University of California (USA) in 2007. After that, he was a post-doc/lecturer at the University of California in 2007–2009. Now he is a senior scientist in the College of Global Change and Earth System Science, Beijing Normal University, China.

Dr. Zhang received the Chinese Government Award for Outstanding

Students Abroad in 2007. This award is the highest honor bestowed by the Chinese Government for outstanding Chinese students studying abroad. Dr. Zhang also received Zolk Fellowship, Block Grant Fellowship, Wakeham Fellowship and Alice Leung scholarship from the University of California in 2004–2007, and Presidential Excellence Award of Chinese Academy of Sciences in 2003.

Dr. Zhang has published 30 papers. His research interests include: Global climate change, wavelet analysis, image processing, astronomical image processing, and data classification.



Naoki Saito received the B.Eng. and the M.Eng. degrees in mathematical engineering from the University of Tokyo, Japan, in 1982 and 1984, respectively. In 1984, he joined Nippon Schlumberger K.K., Fuchinobe, Japan, and in 1986, he transferred to Schlumberger-Doll Research (SDR), Ridgefield, CT where he worked as a research scientist until 1997. While working at SDR, he also pursued his Ph.D. degree in applied mathematics and received it from Yale University in 1994. In 1997, he joined the Depart-

ment of Mathematics at the University of California, Davis, where he is currently a professor and chair of the Graduate Group in Applied Mathematics.

Dr. Saito received the Best Paper Award from SPIE for the wavelet applications in signal and image processing conference in 1994 and the Henri Doll Award (the highest honor for technical papers presented at the annual symposium within the Schlumberger organization) in 1997. He also received the Young Investigator Award from the Office of Naval Research, and the Presidential Early Career Award for Scientists and Engineers (PECASE) from the White House, both in 2000. Since 2008, he has been listed in Marquis Who's Who in America.

His research interests include: applied and computational harmonic analysis; feature extraction; pattern recognition; data analysis; Laplacian eigenvalue problems; elliptic boundary value problems; data compression; statistical signal processing and analysis; human and machine perception; and geophysical inverse problems. He has been also serving on an editorial board of the two journals: Applied and Computational Harmonic Analysis; Inverse Problems and Imaging.

He is a senior member of IEEE as well as a member of IMS, SIAM, and JSIAM.