NASA Technical Memorandum 72866

## AIFTDS STAND-ALONE RMDU FLIGHT TEST REPORT

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Douglass O. Wilner

July 1979





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Dryden Flight Research Center Edwards, California



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#### 1.0 INTRODUCTION

The Airborne Integrated Flight Test Data System (AIFTDS) developed by NASA Dryden through contracts NAS4-1848 and NAS4-2161 has been subjected to two analytical studies in the past two years. The first study was designed to indicate the accuracy and operational characteristics in a hostile laboratory environment and was concluded with the publication of NASA TMX-56043. This report covers work, which commenced in mid-1976 and concluded with the last flight in July, 1977, and was designed to identify the AIFTDS performance characteristics in an actual flight environment.

### 1.1 Background on AIFTDS Stand-Alone Flight Test

Several documents and articles, published by both NASA and private industry; discuss the AIFTDS concept in its entirety; however, no document exists to date which analyzes the <u>AIFTDS</u> in a flight environment. NASA Dryden does not routinely evaluate electronic data acquisition systems in a flight environment; however, it was judged appropriate to do so for the AIFTDS. The reasoning for the decision was two-fold: first, the AIFTDS is an extremely complex, but flexible, pulse code modulation (PCM) system which requires considerable user familiarity for effective application. Second, it was deemed necessary to provide a baseline for future reference when the AIFTDS is in widespread use at Dryden Flight Research Center.

In conjunction with the two reasons stated above were the objectives of the flight test:

a. To familiarize the Dryden crew with the AIFTDS hardware and operation characteristics.

b. To study the accuracy and resolution of the system in a flight environment.

c. To determine shielding and instrumentation techniques for a system as sensitive as AIFTDS.

d. To obtain hard-copy PCM data from the flights through the Center's data reduction system.

Dryden's Measurement Engineering Branch proceeded with the flight test of the AIFTDS in 1976.

# 1.2 Discussion of Flight Test Plan

Because the AIFTDS is, as the name implies, a PCM data acquisition <u>system</u>, a flight test plan was formulated which divided the task into two phases. One phase was the flight test of the stand-alone remote multiplexer/digitizer unit (RMDU), while the second phase will be the flight test of the RMDU in conjunction with an airborne computer, an RMDU Controller Unit (RCU), and a Cockpit Control/Display Unit (CCDU), see Figure 1. This report covers only the first phase. The knowledge gained through the first phase of the flight test will be applied directly to the as yet uncompleted second phase; e.g., optimum RMDU wiring configurations as well as sensor instrumentation developed in phase one are extremely important in the configuration control of the second phase of the flight test.

#### 2.0 BACKGROUND ON AIFTDS

### 2.1 <u>Background on the Dryden Stand-Alone Remote Multipler/Digitizer</u> Unit (RMDU)

NASA Dryden's stand-alone configured RMDU used in the flight test program utilized the following plug in modules and/or cards: analog multiplexer (AMX) cards, presample filter/amplifier (PSF) card, excitation/bridge completion (EBC) card, analog data processor module (ADP-M), and the standalone timing module (SAT-M) see Figures 2a through 2e. A brief description of each is presented in Appendix 1.

2.2 Symbols and Abbreviations List

ADP-M	Analog Data Processor Module
AIFTDS	Airborne Integrated Flight Test Data System
ALT	Approach and Landing Test
AMX	Analog Multiplexer
AC.1 - AC.4	Accelerometer #1 - Accelerometer #4
BiØ~L	Bi-phase Level
BiØ-M	Bi-phase Mark
CPT 1 - CPT 4	Control Position Transducer #1 - Control Position Transducer #4
DC	Direct Current
DM-M	Delay Modulation Mark
EBC	Excitation/Bridge Completion
EPROM	Eraseable Programmable Read Only Memory
GPA-Ø	Gain Programmable Amplifier's Zero Input Response
HLC	High Level Calibration

ILT/Premod Filter	Isolation Line Terminator/Premodulation Filter Unit
ips	Inches per second
IRIG	Inter Range Instrumentation Group
LFC	Laminar Flow Control
LLC	Low Level Calibration
LSB	Least Significant Bit
MSB.	Most Significant Bit
MSBLS	Microwave Scanning Beam Landing System
NRZ-L	Non-Return to Zero Level
NRZ-M	Non-Return to Zero Mark
РСМ	Pulse Code Modulation
PSB	Power Supply Bite (Built In Test Equipment)
PSF	Presample,Amplifier Filter
PT1 - PT4	Pressure Transducer #1 - Pressure Transder #4
RCU-	RMDU Controller Unit
RMDU	Remote Multiplxer/Digitizer Unit
SAT-M	Stand-Alone Timing Module
тс	Thermocouple
ТМ	Telemetry
W.O.L.	Wallace O. Leonard

# 2.3 Flight Test Background

Normally, a flight test program of this scope would require a period of from 6 months to a year to accomplish. The flight testing of Dryden's

stand-alone system took approximately 19 months of which 9 months were actually devoted to flights. Several factors contributed to this time lag:

a. The Center viewed the AIFTDS as a low priority project.

b. The parameters originally requested had been wired in the aircraft (aircraft used was a modified Lockheed Jetstar - see Figure 3) circa 1970, and proved to be highly unreliable due to shielding and connector problems.

c. Project engineer and project instrumentation engineer were changed during the tests.

d. Higher priority programs (MSBLS, LFC, etc.) instituted after the start of the AIFTDS flight test program severely impacted the work schedule.

e. Time onboard the aircraft was restricted because of reason #4: "wringing out" or verification of the existing system was virtually impossible as was the opportunity to install new sensors in the aircraft.

f. Software was not available to analyze data on ground station.

The original parameter list agreed upon by the AIFTDS project engineer in 1975 included the following parameters, for which wiring was reported to be installed on the aircraft:

Early Jetstar Parameter List

Control Position Transducers: (Hi-Level Outputs)

- (1) Throttle Position
- (2) Rudder Pedal Position
- (3) Rudder Position
- (4) Aileron Position
- (5) Elevator Position
- (6&7) Stabilizer Position
- (8) Alpha Vane
  - Nose Boom
- (9) Beta Vane

Accelerometers:

- (10) Lateral CG Acceleration
- (11) Longitudinal CG Acceleration

(12)	Right Hand Horizontal Stabilizer Root Leading Edge Vertical Acceleration.
(13)	Right Hand Horizontal Stabilizer Root Trailing Edge Vertical Acceleration.
(14-17)	Tail Accelerometers #1, #3, #4, #5 - Hi-Level Output
(18&19)	Tail accelerometers #6, #7 - Low Level Output
Press	ure:
(20)	Air Speed Coarse
(21)	Air Speed Coarse Air Speed Fine
(22)	Altitude Coarse Altitude Fine $\left. \begin{array}{c} 0 - 2200 \text{ PSFA W.O.L.} \end{array} \right\}$
(23)	Altitude Fine
(24)	Rudder Pedal Force
(25)	Rudder Differential Pressure
(26)	Aileron Wheel Force
(27)	Total Reference Pressure
(28)	Elevator Differential Pressure
Gyros	:
(29)	Roll Rate
(30)	Yaw Rate
Presi	cison Voltages:

(31-42) Precision Voltages from Divider Box.

A pacer flight to calibrate the Jetstar's airspeed indicators with a calibrated craft was scheduled for June 3, 1976. This flight revealed that only 4 sensors of the 30 requested were operating. Six weeks later on July 14, another flight was flown. By this time 9 of the original 42 parameters had been installed and checked out. Two days later on July 16, another checkout flight was made; 8 sensors were operational, 14 precision voltages were employed, and 17 of the disabled sensor channels were shorted at the pallet. This flight was the first flight in which data was directed through the ground station. It then became painfully obvious that an excessive amount of time would be needed to insure the correct installation of the original sensors which were requested.

#### 3.0 STAND-ALONE FLIGHT TEST

#### 3.1 The AIFTDS Experimental Sensor Box Concept

It was also during this time frame (May-July 1976) when the Approach and Landing Testing (ALT - for the Shuttle program) was beginning to become active. The Jetstar was chosen as the test bed for the ALT Microwave Scanning Beam Landing System (MSBLS) as well as the facility for the Laminar Flow Control (LFC) project. The AIFTDS project had the lowest priority and hence, very little work was carried out on the Jetstar to satisfy the sensor configuration requirements.

It was at this time when the AIFTDS project leader decided to abandon the original idea of using the ship's sensors for the flight test phase. A plan was formulated which would benefit the project in two ways:

a. Trade time in the lab for actual time on the aircraft for wiring and testing of the system.

b. Would give the AIFTDS project leader complete control over sensor wiring configurations and shielding and excitation designs.

A sensor housing box was designed and constructed (See Figure 4) which was mounted inside the aircraft and fulfilled the original requirements of the AIFTDS flight test phase. In this box was a selection of commonly used sensors; i.e., strain gages, thermocouples, control position transducers (CPT's), accelerometers, and resistors. Four similar strain gages, four similar accelerometers, and four similar CPT's, as well as two similar TC's and two similar resistors were chosen. For the strain gages, accelerometers, and CPT's, there were four different methods of installation as follows:

> (1) Sensor excited by an integrated signal conditioning card, and having the shield referenced or driven at the sensor end.

(2) Sensor excited by integrated signal conditioning and grounding the shield at the RMDU.

(3) Sensor excited by DFRC "standard" external signal conditioning boxes, and having the shield referenced at the sensor end.

(4) Sensor excited externally with the shield grounded at the RMDU.

For the thermocouples and the resistors the configurations were limited to shield referencing either at the sensor end or at the RMDU end (see Figures 5 and 6). With the design and construction of this sensor box, the priority/time conflict on the Jetstar and configuration control problems were resolved. On the final two flights (June-July 1977) an active pre-sample filter was employed with four of the sixteen sensors. The results from the filtered data was compared with the unfiltered responses of previous flights.

#### 3.2 Jetstar Flight Schedules

As mentioned previously, several higher priority projects began using the aircraft originally chosen for the AIFTDS. Such additional projects, Microwave Scanning Beam Landing System (MSBLS) - part of the Shuttle Approach and Landing Tests, the Laminar Flow Control project and others, created an added burden on the flight scheduling. The MSBLS project needed a large number of flights to effectively checkout the landing system - sometimes multiple flights per day. This, of course, meant that any AIFTDS work on the aircraft had to be completed between pre-flight checks. This constraint also dictated the flight plan for the AIFTDS project when both AIFTDS and MSBLS were flown together. The Laminar Flow Control Experiment (LFC) required lengthy flights (> 90 minutes) to achieve its purpose.

The length of the LFC flights created two major problems: extending the use of the onboard tape recorder for the entire flight, and causing telemetry link dropouts by the extreme range involved. An additional problem of scheduling these flights was the time sharing of the instrumentation pallet for the multiple projects, the tape recorder and transmitter. Cooperation between the various projects was a must in order to satisfy both data requirements.

During the peak months of activity on the Jetstar as many as 4 or 5 flights were being scheduled weekly - any last minute cancellation or alteration of the flight schedule or plan would require additional manpower support from the telemetry ground station. Conflicts with the TM ground station coverage due to these cancellations or alterations often acted to eliminate the telemetry down link coverage of the AIFTDS PCM data. This, in effect, resulted in poor data visibility since no data could be verified until the ground station could play back the airborne record tape at a later date. Nonetheless, with all of these flight scheduling proglems, the AIFTDS project received the requested six data flights in a 9 month period.

#### 3.3 Discussion of AIFTDS Data Acquisition Procedure and Equipment

A block diagram of the AIFTDS data acquisition system utilizing the AIFTDS Experimental Sensor Box as sensor input, is shown in Figure 7. A photograph of the instrumentation pallet is shown in Figure 8.

The AIFTDS Experimental Sensor Box schematic and photograph is shown in Figures 4, 5, and 6. From the schematic it can be seen that sensors PT3, PT4, AC.3, AC.4, CPT3, and CPT4 have their power derived from the pallet external signal conditioning unit. It can also be seen that sensors PT1, PT2, AC.1, AC.2, CPT.1, and CPT.2 obtain power through either the RMDU internal signal conditioning card (EBC card) or through the RMDU power supply itself. All sensors in the AIFTDS Experimental Sensor Box, however, receive their stimuli from the box itself; each individual group of sensors is exposed to identical stimuli. This factor, in conjunction with the fact that the individual group of sensors are identical allows the investigator to accurately compare data responses for the four excitation and shielding combinations. The responses from the 16 sensors in the AIFTDS Experimental Sensor Box along with some internal RMDU "health check" signals (LLC for gains of 1000, 400, 100; HLC, GPA-Ø for gains of 400, 10; and PSB) and some shorted channels were processed and digitized by the RMDU. The RMDU has the capability of simultaneously generating two serial PCM bit streams in different codes for routing to different post-processing media. The output coding formats available are: NRZ-L, M; BiØ\_L,M; and DM-M. The output data stream being sent to the onboard tape recorder is in the format of DM-M (for improved low frequency response and minimum DC base line drift) while the data stream being sent to the transmitter is in NRZ-L format. The use of these two formats is standard at NASA Dryden.

The DM-M formatted PCM data stream is recorded on the Model AR-700 airborne tape recorder in the direct mode. The NRZ-L formatted PCM data stream sent to the transmitter is first directed to a premodulation filter having a cutoff frequency of 160 Kbits/sec. The filtered output is then used to frequency modulate the telemetry transmitter carrier of 1521.5 MHz. See Figures 9 and 10 for the circuit schematic and connector configuration, and the photograph in Figure 11 for a detailed view of the filter box.

A crucial part of the AIFTDS flight test was the use of the onboard flight test engineer. The flight test engineer recorded flight data by hand through the use of the flight line tester which was mounted on the instrumentation pallet. The flight line tester, being a portable PCM decommutator, allowed the flight test engineer to record data from any of the parameters during the flight to be compared with the data processed on the ground station.

#### 3.3.1 Real Time Processing of PCM Data

The 1521.5 MHz telemetry signal is received at the Dryden telemetry ground station. For the AIFTDS flights the PCM stream was received by the series 410 receivers and then processed by the decommutation equipment. The decommutation equipment (consisting of EMR Model 720 PCM bit synchronizer, Model 2731 PCM frame synchronizer, and Model 2736 PCM subframe synchronizer) is programmed and controlled by the SEL (Systems Engineering Laboratory) Model 86 and 810B computers.

Through this interaction between the SEL computer and the decommutation equipment, certain parameters were selected for real time display. The real time display was 32 tracks of stripchart recording in the high range ground station control room. In addition to the 32 displayed parameters, the IRIG timing track from the ground station was also displayed. IRIG-B time code generators are used to establish time correlation between real time data and that data which is recorded on the onboard tape recorder. Because of computer input/output limitations the stripchart outputs represent samples at a maximum rate of 5ms/sample (200 samples/Sec - which is the rate of the PCM system currently in use at Dryden).

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## 3.3.2 <u>Telemetry Dropouts</u>

Even with directional tracking antennas the maximum range attainable with approximately 5 watts of power at 1521 MHz into the blade transmitting antenna on the aircraft is 200-300 miles (320 Km-480 Km) taking into account propogation and the aircraft's altitude. Anything between the aircraft and the receiving antenna (over the horizon, mountains, buildings, etc) severely limits this maximum range. In fact, the only reliable reception is limited to line of sight ranges. Any distance greater than this or any obstacles between the aircraft and the receiving antenna can (and usually will) cause telemetry dropout. TM dropouts can cause the PCM decommutation equipment to lose synchronization and, hence, cause "noise" to appear at the output and on the stripcharts.

If the TM dropouts are of short duration the effect is transient irregularities on the stripchart outputs. These irregularities caused by TM dropouts interfere with the monitoring of the real time processed PCM data (in that the unadulterated data looks similar to that data which is contaminate by transient type noise). To examine the TM data during the times when dropouts occur is impossible and requires the onboard tape to be played back later to fill in data for dropout times.

The instrumentation pallet (which holds the entire data acquisition system for the project - see Figure 8) was shared with other ongoing experiments at this time (MSBLS, LFC, and others). During those portions of the flight when data was to be collected for the other project(s) the AIFTDS PCM bit stream was disconnected from the transmitter's modulation input. This effected a TM dropout, also which could extend for long periods of time. However, the onboard tape recorder was used for the collection of data during this time.

In the section of this report dealing with the analysis of the flight data, many dropouts are indicated by simultaneous loss of data on all channels (see appendix).

#### 3.3.3 PCM On Board Recording

Onboard recording of the PCM data was accomplished using an airborne AR-700 Model tape recorder. On all AIFTDS flights the tape speed was 15 ips (250 KHz bandwidth) thus giving approximately 90 minutes of recording time. The coding of the PCM data for the tape recorder was in DM-M (Delay Modulation-Mark), as well as NRZ-L (non-return to zero level), and the recorder was operated in the "direct" record mode. The bit packing density of the tape did not exceed 12,000 bits/inch throughout the flight test. The AR-700 is a 14 track recorder (when using 1 inch width tape); AIFTDS data (NRZ-L and DM-M formats) was recorded serially on two tracks. The IRIG-B time code generator signal was, recorded on another track, with the remaining eleven tracks being devoted to functions relating to the other projects which were sharing the pallet.

The tapes from the onboard recorder were held until time was available for post flight processing. The time delay was governed by the ongoing progress towards developing the software package necessary to process the flight data and the work load in the telemetry facility at Dryden. The plan of attack for processing the flight data was as follows: The data which was manually recorded (via the onboard AIFTDS flight test engineer) was correlated with the real time strip chart data. The IRIG-B time was noted for any data of interest and that time was used for selecting that portion of the flight data tape for processing through the ground station computer. See Appendix 3 for samples of data reduction in this manner.

#### 4.0 SUMMARY AND DISCUSSIONS

#### 4.1 Flight Objectives and Achievements

A measure of a flight test project success is to compare the results with the goals and determine if the achieved results satisfied these goals. Recalling the goals from Section 1, each will be compared with the associated results.

#### 4.1.1 <u>Goal #1: Familiarize the Dryden Crew with the</u> AIFTDS Hardware and Operational Characteristics

Throughout the flight test phase of the Stand-alone AIFTDS numberous people came into intimate contact with the AIFTDS.

a. The instrumentation crew involved with the wiring and checking out of the AIFTDS installation.

b. The personnel in the telemetry ground station responsible for the decommutation and processing of the real time data from the PCM telemetry link.

c. The project engineers and the project managers involved with the overall planning and direction of the project.

d. The persons involved with procurement and contract management were all involved with and familiarized with the AIFTDS hardware and/or the operational characteristics.

Although this goal was easily achieved, it would obviously be more effective if more people could have been trained and familiarized. However, not until the AIFTDS is in widespread use at the Center as its primary data acquisition system will all of the appropriate persons be familiarized with the AIFTDS system.

## 4.1.2 <u>Goal #2: To Determine Shielding and Instrumentation</u> Techniques

As was previously stated, the Dryden RMDU possesses a resolution of 12 bits; at a gain of 1000 each count represents  $5 \mu$  volts which is six times more sensitive than the existing systems presently in use at DFRC. In order to utilize this extreme sensitivity extra care must be exercised in the sensor wiring. It was shown early in the flight test program that the existing wiring and instrumentation (excitation and signal conditioning) of the Jetstar sensors was not adequate to make full use of this increased capability. Millivolt level noise (common mode and differential mode) destroys data validity when microvolt level measurements are desired.

The experimental sensor box data clearly demonstrates the trade-offs between excitation, signal conditioning, and shielding schemes (see Appendix 3). The best overall configuration of the sensors was to ground the shield at the sensor (to minimize common mode effects) and to utilize internal signal conditioning/excitation. This fact should not be too surprising to instrumentation engineers, however, the data from the flight test demonstrates this fact beyond any shadow of doubt.

In the process of determining the appropriate shielding and excitation techniques to use, care was exercised in not eliminating noise through the use of pre-sample filtering. This method of "cleaning up" the data can give false data in the area of shielding techniques. The AIFTDS was designed with increased flexibility so that pre-sample filtering is not needed in all cases. The need for filtering can sometimes be eliminated by increasing the sample rate. However, in cases where it is not possible or feasible to increase the sample rate (due to ground station limitations, tape recorder limitations, or any other reasons) the AIFTDS does have provisions for internal active-presample filter modules. As noted earlier in this report the presample filter used in the Dryden program was programmed for a gain of 400 on 3 of the 4 channels and a gain of 1 on one channel. The filters in all 4 cases were 4-pole Butterworth low pass filters having a cutoff frequency of 50 Hz. This presample filter module was used on a select few parameters during the last 2 flights; comparison between filtered data and non-filtered data is made in the Appendix section.

#### 4.1.3 <u>Goal #3: To Obtain Hard Copy PCM Flight Data</u> <u>Through the Ground Station Facility</u>

This goal was one of the most difficult to achieve: it certainly required the most time of all to achieve. The The Dryden PCM processing system is a complex system consisting of two SEL 32 bit computers controlling the PCM decommutators. with an SEL 86 (central processing system, a Control Data Corp. Cyber Model 7328 control computer facility, as well as many software operating systems. This processing system, being large and complex, is very difficult to change or modify. Since the AIFTDS PCM format was different for this test phase from the systems currently in use at Dryden, a modification to the presen Dryden PCM processing scheme was necessary in order to process AIFTDS data. Of course, the modifications and additions to the processing system required programming manpower; with the onslaught of the Shuttle ALT program, as well as higher priority projects at the Center very little manpower was available to work on AIFTDS requirements. Consequently, the hard copy data was received from the ground station 6 months after the last flight.

Normally, a flight project request for data processing requires pre and post-flight calibration data in order to perform the polynomial fit to the data as well as to convert the data to engineering units. The data processing also usually involves data compression by the computer controlled decommutators in the ground station. However, the requirements for the AIFTDS PCM data processing were extremely rudimentary, e.g., the only desired output from the processing was a straight decimal conversion of the raw offset binary data. In addition to the binary to decimal conversion, the display of stripchart data correlated with the IRIG-B time and the word frame and/or subframe number was requested. Data compression or engineering conversions were not performed on this data, however the gain tag bit conversion was accomplished (the gain tag bit has been discussed earlier) when converting the offset binary data to decimal value.

Because of the minimal requests for PCM data processing levied on the data reduction personnel, all the requested functional capabilities were provided in the telemetry ground station facility. This completely eliminated the need for Cyber system software and operational manpower. The PCM hard copy data was handled in a manner similar to that of the real-time strip chart outputs. The front end telemetry hardware (bit, frame, and subframe synchronizers) decommutated the flight data under control of the computer.

The flight data recording tape was played back at 1/4 speed so that the data sample rate would not exceed the output hardware  $I/\emptyset$  capabilities of the existing system software and hardware. Data then was provided at 5ms intervals (corresponding to Dryden's 200 samples per second existing data acquisition systems) on a hard copy line printer (see Appendix 3).

Flight data which was taken by hand in the aircraft was IRIG-B time correlated with that data which was displayed on the real time strip charts. Key time increments were then requested for the hard copy output. The hard copy PCM data was compared with that data obtained by hand and with that obtained from the real time strip charts in Appendix 3. From the examples illustrated in Appendix 3 it can be seen that the hand recorded data and the hard copy printout data correlate very closely with each other (±6 counts deviation). The strip chart data probably is also as accurate, however, the scale factor necessitates much coarser readings.

### 4.1.4 <u>Goal #4: Study of Accuracy and Resolution of</u> the AIFTDS System in a Flight Environment

Since hard copy PCM data has been made available to the AIFTDS team only recently, a complete analysis on the flight data may not be completed for some time. However, some accuracy and resolution analysis has been performed on some of the precision voltage inputs.

For these precision voltage input channels, several hundred samples were statistically analyzed. The occurrence of data values greater than 5 counts away from the norm were noted (see Appendix 3). Data point deviations on the order of 2 or 3 counts (for channels without presample filtering) were considered to be caused by instrumentation noise and/or thermal noise. From the analysis performed on these data it was determined that a deviation of 5 counts or more occurred less than once every 850 data values. For the most part, these large deviations were on the order of several hundred counts from the normal. The source of these large excursions has yet to be determined, although several possibilities have been brought to light. Instrumentation anomalies as well as anomalies in the ground station hardware are thought to be the most probable causes.

Appendix 3 illustrates several examples of precision voltage data. It is doubtful that the flight environment had any detrimental effects on the data validity save for an occasional noise anomaly. It is also obvious that the system can be made more immune to the aircraft noise environment with sound instrumentation engineering practices.

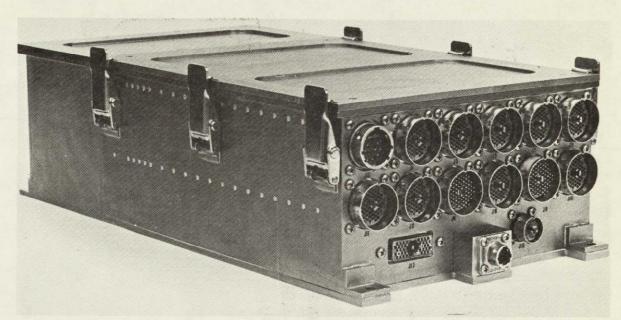
#### 4.2 Impact of 12 Bit PCM Systems in Today's Aerospace Environment

NASA Dryden Flight Research Center's experience in high resolution data acquisition systems took a major step forward in the mid 1960's with the advent of a 9 bit PCM system. This PCM system, called the CT-77, has served as the backbone of Dryden Flight Research Center's data acquisition systems. With the recent breakthroughs in hybrid analog circuitry, PCM systems with resolutions in excess of 10 bits have been finding their way into the instrumentation engineer's arsenal. However, it is too soon to tell what effect, if any, this increase in resolution will have on PCM data acquisition systems.

One major advantage of a higher resolution, higher bit PCM data acquisition system (say 12 bits) over a 9 bit system is the increase in usable dynamic range available in the 12 bit system. Thus, by allowing each count to represent the same level there will be nearly one decade of range increase in the 12 bit system over the 9 bit system. This obviously allows for greater simplicity in data reduction (engineering units conversion, etc.) as well as allows for more leeway in the original estimation of output signal levels.

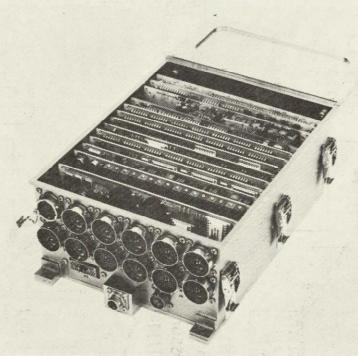
Another advantage is the obvious increase in sensitivity for the same dynamic full scale range. The same full scale range for a 12 bit system will allow an 8:1 increase in gain or resolution of each count over the 9 bit system. This advantage, however, also creates its own problems, e.g., the instrumentation engineer must properly shield and condition the transducers. Shielding and signal conditioning of transducers must be adequate to minimize common mode, differential mode, and ground loop coupling and other sources of data contamination. Since, in the case of aircraft instrumentation, there are usually many sensors located at remote locations referenced to the airframe potential in some manner, it is nearly impossible to eliminate one or all of these sources of error. This has a tendency to offset the advantages gained in the increased sensitivity (resolution per count) of the larger word size. There are, today, several groups of instrumentation engineers using the AIFTDS system with its increased word size and flexibility. NASA Ames, Douglas Aircraft Company, ERDA, and U.S. Air Force, to name only a few, have compiled many hours use with these systems - including many hours in a flight environment. Their experiences using the AIFTDS are well documented.

At NASA Dryden there are projects requiring accuracies on the order of 1/2% or better. These increases in accuracy requirements can be achieved easier with longer bit words; the quantizing error alone in an 8 bit system is on the order of .4% whereas it is only .025% in a 12 bit system. Hence, it is inevitable that the word length will have to increase if the dynamic range requirement or the accuracy requirement is to be satisfied. The penalties for this increase in word length are the increased care and engineering which must be exercised in instrumenting the newer system as well as the moderate increase in bandwidth required to keep the same word rate. The trade-offs are real and will have to be evaluated individually for each application.



Front View

Figure 1. AIFTDS standalone RMDU.



Top view (with cover removed) E 31053 Figure 1. Full AIFTDS system.



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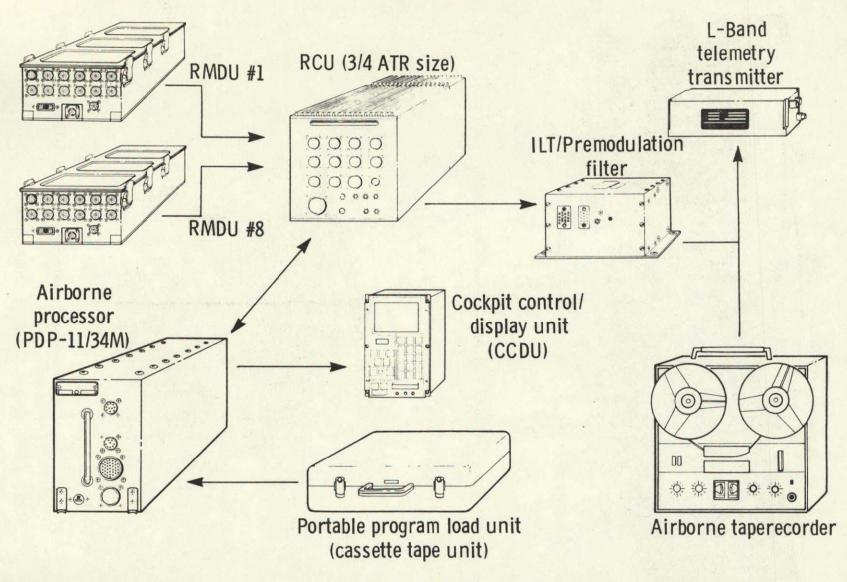
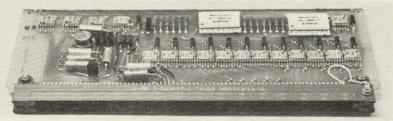


Figure 1. Full AIFTDS system.

.

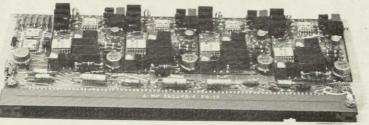
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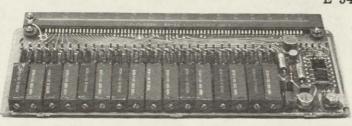


(a) Analog muliplexer(AMX) card.

(b) Presample filter/ amplifier (PSF) card.

E 34078



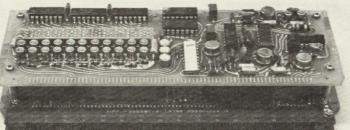


E 34077

(c) Excitation/bridge completion (EBC) card.

E 34075

E 34074

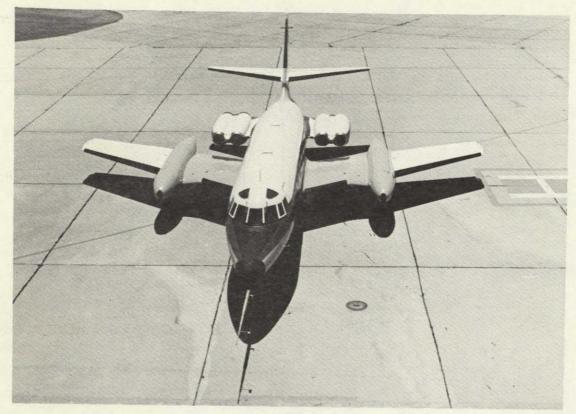


(d) Analog data processor module (ADP-M).

(e) Stand-alone timing module (SAT-M).

E 34073

Figure 2. RMDU plug-in modules.



Jetstar front view

ECN 4039

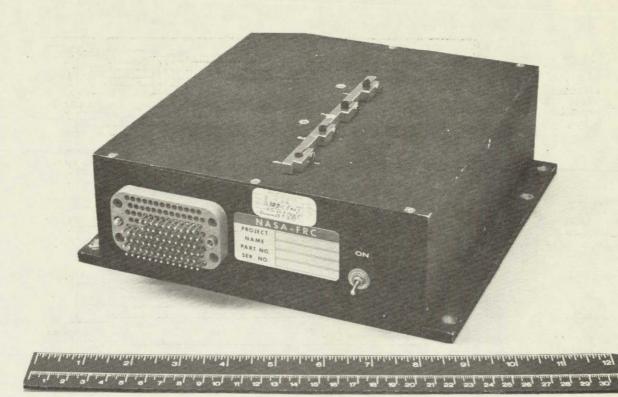


Jetstar side view

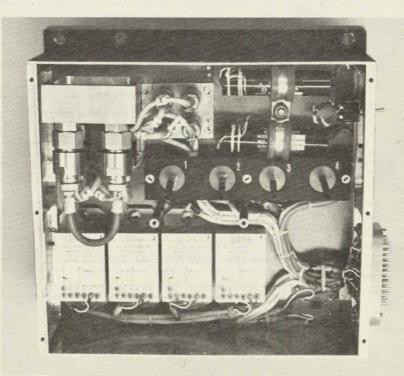
ECN 2401

Figure 3. Jetstar.

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Experimental sensor box with cover plate in place E 34082



Experimental sensor box with cover plate removed E 34081

Figure 4. Experimental sensor box.

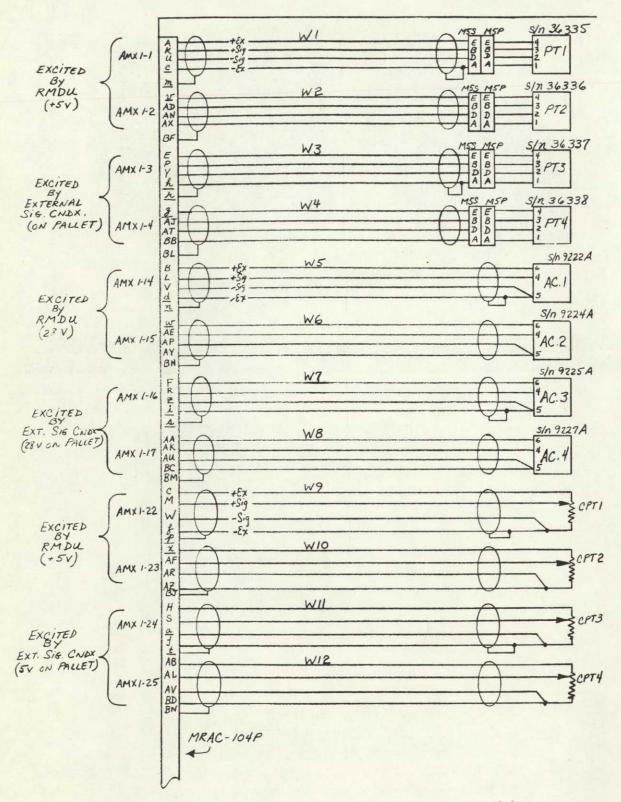
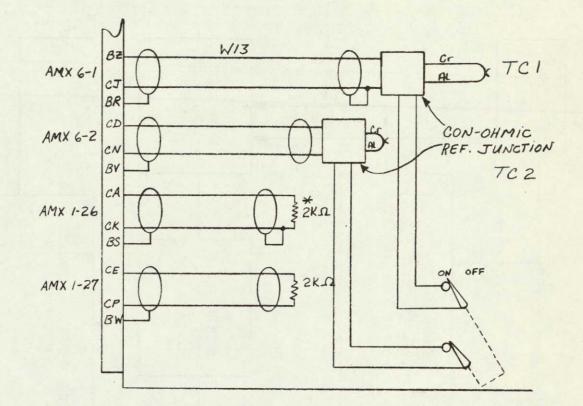


Figure 5. AIFTDS experimental box schematic (top section).

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```
PT1 : STATHAM PMISITC-202 - 350 ; 20 psiD; 5V; S/n 36335
PT2 :
         "
                                             S/n 36336
PT3 :
                                             S/n 36337
       11
PT4 :
                                             S/n 36338
AC.1: Donner; 28V; 105 max; .4v/G ; S/N 9222A
                                     ; S/N 9224A
AC.2
AC. 3
                                     ; S/N 9225A
AC.4
                                       S/N 9227A
CPTI-CPT4: 2KA HELIPOT
TCI: Chromel - Allumel 21/2 uv/oF
TC2:
2Ka*: 2Ka; 18W, . 1% Wire wound
2KA :
```

Figure 6. AIFTDS experimental box schematic (bottom section).

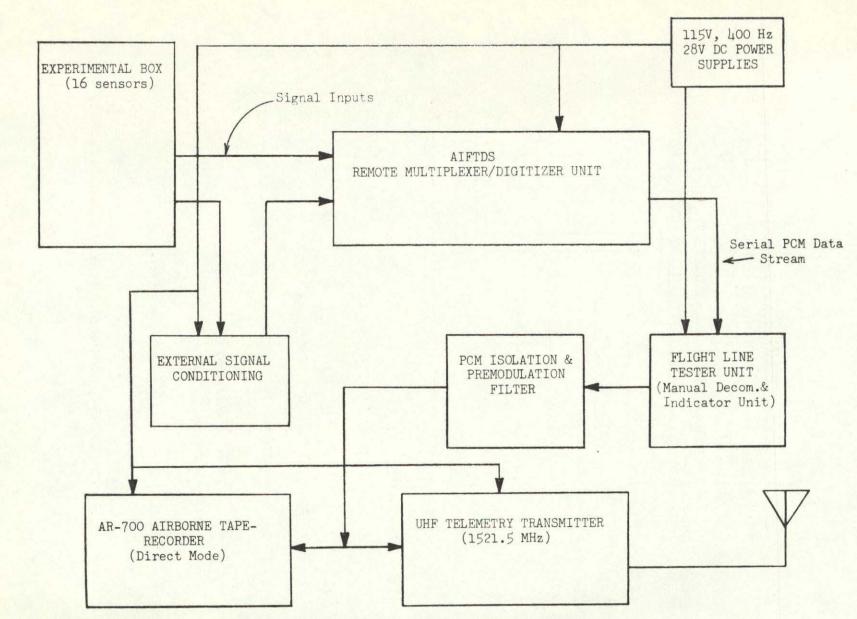


Figure 7. AIFTDS data acquisition system block diagram.

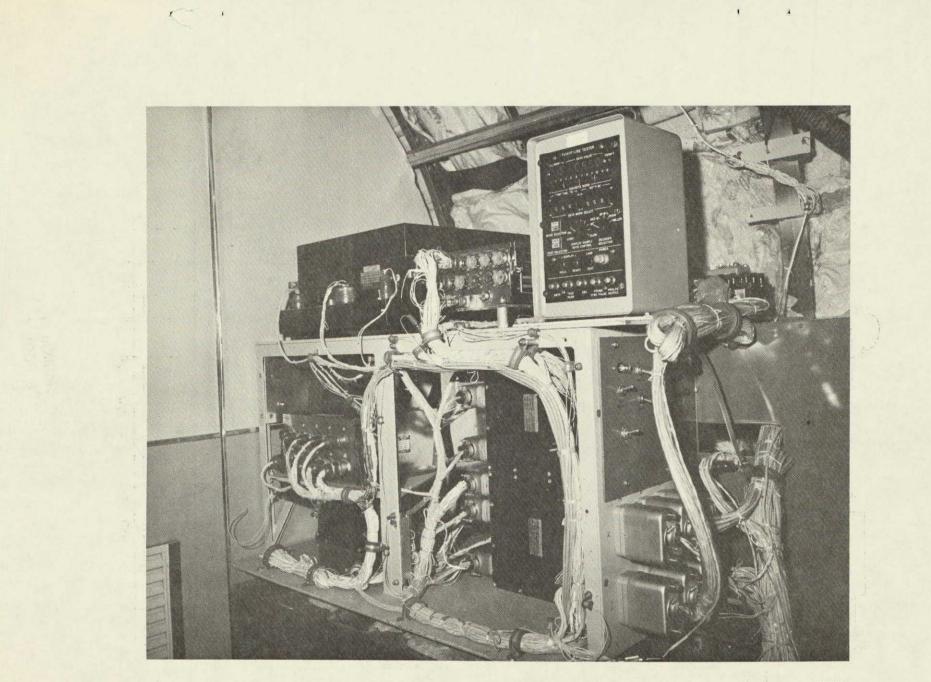


Figure 8. Instrumentation pallet.

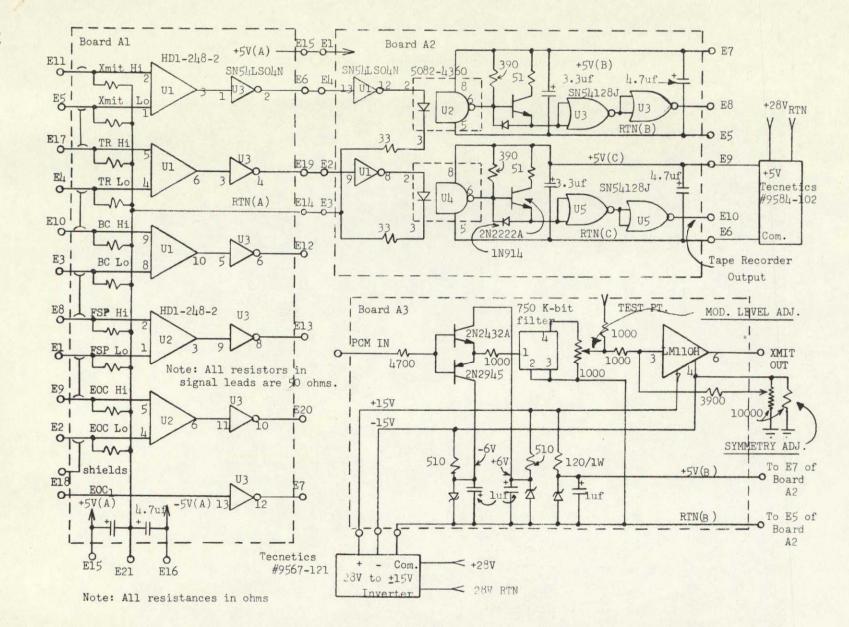


Figure 9. Schematic diagram of ILT/premodulation filter circuit.

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# INPUT PLUG (J20): MRAC-18P

Pin #

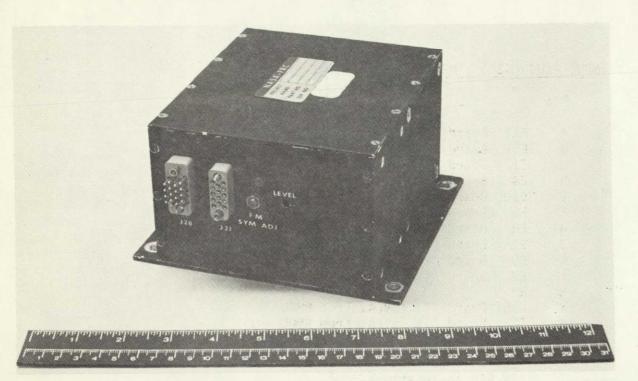
Destination

Α	E11	Board Al; XMIT Hi
В	E5	Board A1; XMIT Lo
С	E17	Board A1; TAPE RECORDER Hi
D	E4	Board A1; TAPE RECORDER Lo
Ε	E10	Board A1; BINARY CLOCK Hi
F	E3	Board A1; BINARY CLOCK Lo
Н	E8	Board A1; FRAME SYNC PULSE Hi
J	E1	Board A1; FRAME SYNC PULSE Lo
K		Board A1; END OF CYCLE Hi
L		Board A1; END OF CYCLE LO
М	N2	
N	E15	Board A1; +5V(A) from RMDU
Ρ		Board Al; -5V(A) from RMDU
R	E18	Board Al; EOC <sub>1</sub> TTL level signal from RMDU
S		to inverters
Т	28V	RTN to inverters
U	Shie	ds
۷	E14	Board A1; RTN(A) from RMDU

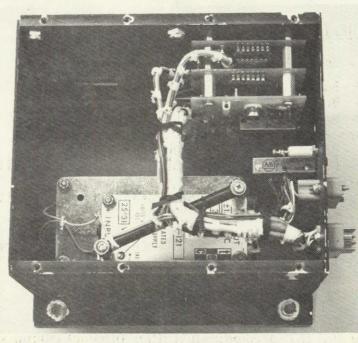
# OUTPUT PLUG (J21): MRAC-145

Α	E7 Board A1;	ECC <sub>1</sub> TTL level signal
В	E12 Board Al;	BINARY CLOCK TTL level
С	E13 Board Al;	FRAME SYNC PULSE TTL level
D	E20 Board Al;	END OF CYCLE TTL level
Ε	Board A3; XMIT	OUT RTN
F	E6 Board A2;	TAPE RECORD OUT RTN
Н	E7 Board A2;	+5V(B) to be used only as a monitor point
J	Board A3; XMIT	OUT Hi
K	E9 Board A2;	+5V(C) to be used only as a monitor point
L	E10 Board A2;	TAPE RECORD OUT Hi
М	NC	
Ν	E7 Board Al;	RTN(A) from RMDU to be used as reference with pins
	A,B,C,D	
Ρ	NC	
R	NC	

Figure 10. ILT/premodulation filter internal wiring schematic.



ILT/premodulation filter box with cover plate in place E 34083



E 34080

ILT/premodulation filter box with cover plate removed

Figure 11. ILT/premodulation filter box.

Format:

it:

MSB

LSB

Note: The MSB (Most Significant Bit) is the Gain Tag bit; when MSB = 0 the gain that was programmed is used, when MSB = 1 the programmed gain is reduced by ½ through the auto-ranging amplifier circuit. The MSB is transmitted first. The remaining 11 bits are in offset binary format.

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Examples:

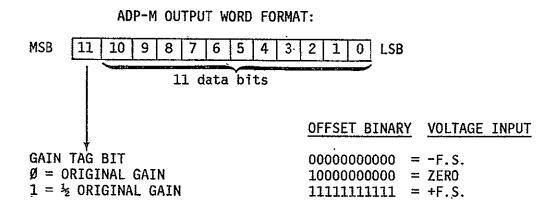
											Decimal Equivalent Translation		
	MSB LSB												LSB
~1024*	1	0	0	0	0	0	0	0	0	0	- 0	0	(0-1024) x 2 = -2048
<del>~</del> 768*	1	0	0	1	0	0	0	0	0	0	0	0	(256-1024) x 2 = -1536
-633*	1	0	0	1	1	0	0	0	0	1	1	1	(391-1024) x 2 = -1266
~512 <sup>`</sup>	0	0	1	0	0	0	·0	0	0	0	0	0	(512-1024) x 1 = -512
-256	0	0	1	1	0	0	0	0	0	0	0	0	(768-1024) x 1 = -256
-64	0	0	1	1	1	1	0	0	0	0	0	0	$(960-1024) \times 1 = -64$
-1	0	0	1	1	1	1	1	1	1	1	1	1	$(1023-1024) \times 1 = -1$
0	0	1	0	0	0	0	0	0	0	0	0	0	$(1024-1024) \times 1 = 0$
+120	0	1	0	0	0	1	1	1	1	0	0	0	$(1144-1024) \times 1 = 120$
+768	0	1	1	1	0	0	0	0	0	0	0	0	(1792-1024) x 1 = 768
+480*	1	1	0	1	1	1	1	0	0	0	0	0	$(1504-1024) \times 2 = 960$
+512*	1	1	1	0	0	0	0	0	0	0	0	0	$(1536-1024) \times 2 = 1024$
+896*	1	1	l	1	1	0	`0	0	0	0	0	0	(1920-1024) x 2 = 1792
+1023	1	1	1	1	1	1	1	1	1	1	1	1	$(2047-1024) \times 2 = 2045$

\* indicates that bit #11 (MSB:Gain Tab Bit) has been set

#### APPENDIX 1

## DESCRIPTION OF RMDU MODULES USED IN FLIGHT TEST

- AMX card: provides 33 addressable low level analog multiplexed channels, one channel of which is devoted to a precision voltage divider providing a 7.3237 mv signal useful for calibration purposes.
- PSF card: provides four independent 4-pole active filter, preamplifiers per card. Dryden's were designed as 50 Hz cutoff, low pass Butterworth filters with gains of 1, 2, and 400.
- ADP-M module: the analog data processing module, the heart of the analog processing portion of the RMDU, consists of a gain programmable amplifier, the autoranging circuitry, and the analog to digital (A/D) converter. The amplifier for the Dryden ADP-M has eight programmable gains (1,2,3,10,50,100,400, and 1000). The autoranging section samples the output of the gain programmable amplifier and automatically reduces the gain to  $\frac{1}{2}$  of its original value when the output exceeds approximately ±90% of full scale. Upon autoranging, a bit is appended to the most significant portion of the data word (see diagram below). The A/D converter portion of the ADP-M is an eleven-bit bipolar successive approximation converter with a resolution of 5 mv/count (thus giving an overall voltage range to the A/D converter of  $\pm 5.115$  volts). Dryden's ADP-M is configured for offset binary output coding to remain compatible with the Center's existing ground station capabilities.



The eight programmable gains give an overall resolution and full scale values as follows:

GAIN	RESOLUTION	FULL SCALE VOLTAGE INPUT						
1	5mv/count * 10 mv/count	±5.1150 Volts ±10.2300 Volts						
2	2½mv/count * 5 mv/count	±2.5575 Volts ±5.1150 Volts						
3	1-2/3mv/count * 3-1/3mv/count	±1.7050 Volts ±3.4100 Volts						
10	500µv/count * 1 mv/count	±0.5115 Volts ±1.023 Volts						
50	100µv/count * 200µv/count	±0.1023 Volts ±.2046 Volts						
100	50µv/count * 100µv/count	±51.150 mv ±.10230 Volts						
400	12 <sup>1</sup> <sub>2</sub> µv/count * 25µv/count	±12,789 mv ±25.5750 mv						
1000	$5\mu v/count$ * $10\mu v/count$	±5.115 mv ±10.2300 mv						

\* indicates autoranged values

SAT-M module: generates all internal and external time base signals, contains the sampling format memory for the data cycle, and formats the 12 bit digital data for transmission recording on an onboard tape recorder, or both. The available output formats are nonreturn to zero level (NRZ-L), nonreturn to zero Mark (NRZ-M) biphase level (Biø-L), biphase mark (Biø-M), and delay modulation mark (DM-M or Miller).

Dryden's SAT-M was configured for NRZ-L format to the RF transmitter, and DM-M format to the onboard tape recorder. The maximum word rate obtained with the SAT-M is 125,000 WPS (1.5 Mbits/sec).

Dryden's was configured for a word rate of 13,888 WPS (167 Kbits/sec) - again keeping compatible with the existing station firmware setup. The sampling format for the data cycle map is stored in erasable programmable read-only memories (EPROM's) located in the SAT-M. The data cycle sampling format memory contains the information that directs the RMDU channel sampling sequence, gain, and sampling rate. The user has a great deal of flexibility in designing the sampling format memory, however, certain limits and constraints should be observed. Any number of mainframe channels can be specified up to and including 128 (which includes the synchronization words and the subframe identification word). Similarily, any number of subframe channels can be specified (not to exceed 128 unique parameters) as long as the product of the number of subframe columns and the depth of the deepest subframe column does not exceed 256. These constraints are entirely based upon the size of the memory circuits; newer SAT-M modules are available now from the manufacturer which have considerably larger memory circuits. The NASA Dryden sampling format was comprised of 16 words per frame with the deepest subframe channels (see figures in Appendix 2).

# APPENDIX 2

# DRYDEN'S SAMPLING FORMAT FOR USE IN FLIGHT TEST

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WORD NO. -----

$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	0	l	2	3	Ц	5	6	7	- 8	9	10	11	12	13	14	15		
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	1 1	r 1		HLC		1-5		1-1	1-10	1-1			1-14	1-14	AMX* 1-22 X2	AMX 1-23 X2	` 0	s V
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	L				<b>I</b>	<b>1</b>	G₽А-Ø	1-2	1-11	1-2		, i	1-15	1-15	АМХ 1-2Ц Х2	AMX 1-25 X2	1	B F R M
7ØØ38       1-4       1-27       1-4       1-13       1-17       1-17         X400       X1000       X1000       X1000       X100       X10       X3         SF-ID       7ØØ48       SF-ID       SF-ID       1000       X1000       X1000       X1000       X1000							(	1-3	1-26	1-3		1-12	1-16	1-16	AMX * 1-22 X1	AMX 1-23 X1	2	M E N O
70048 SF-ID								1-4	127	1-4		1-13	1-17	1-17	АМХ 1-24 Х1	АНХ 1-25 Х1	3	
															AMX 6-1 X400	AMX 6-2 X400	<u> </u>	
7005 8 Word Rate = 13,888 Words/sec No. Bits/Word = 12			sf-Id 7øø5 <sub>8</sub>											AMX 6-1 X1000	AMX 6-2 X1000	5	•	
SF-ID 7ØØ6 8 SF-ID 5 5 5 5 5 5 5 5 5 5 5 5 5						Sample Rate: Main Frame Words = 868 Samples/sec 2-Deep Subframe Words = 434 Samples/sec 4-Deep Subframe Words = 217 Samples/sec									АМХ 1-26 ХЦОО	амх 1-27 хцоо	6	
SF-ID 7ØØ7 <sub>8</sub> Note: astericks denote minor changes for Flights 179 and 181 in that those channels so marked are preceded by a 50 Hz					8-Deep Subframe Words = 108½ Samples/sec Note: astericks denote minor changes for Flights 479 and 481										PSB	LLC-1 X100	7	

\* overall gain of these channels are same; the gain programmable amplifier is reduced to unity and the remaining gain occurring in the active filter

\*\* overall gain of these channels are reduced from 1000 to 800; the gain programmable amplifier is induced to a gain of ? with the active filter providing a gain of h00.

LINE #	PARAMETER	DESCRIPTION	GAIN	<u>SAMPLE RATE</u> (Samples/sec)
1	SYNC-1	Barker Code: 0112) <sub>8</sub>	-	868
2	SYNC-2	Barker Code: 0270) <sub>8</sub>		868
3	SF-ID ∅	7000) <sub>8</sub>		108 <sup>1</sup> 2
4	HLC	High Level Calibration	-	868
5	LLC-1 (AMX 1-33)	Low Level Calibration	400	868
6	AMX 1-5	Shorted at RMDU plug	<b>40</b> 0	868
7	GPA-Ø	GPA zero calibration	400	434
8	AMX 1-1	PT1 Driven shield, RMDU excitation	400	217
9	AMX 1-10	Shorted at RMDU plug	1000	217
10	AMX 1-1	PT1 Driven Shield, RMDU excitation	1000	217
11	LLC-1(AMX 1-33)	Low level calibration	1000	217
12	GPA-Ø	GPA zero calibration	10	217 .
13	AMX 1-14	AC.1 Driven shield, RMDU excitation	10	217
14	AMX 1-14	AC.1 Driven shield, RMDU	3	217
15	AMX 1-22	CPT1 Driven shield, RMDU excitation		
36			2	108 <sup>1</sup> 2
16	SF - ID 1	7001) <sub>8</sub>	-	108 <sup>1</sup> 2
17	AMX 1-2	PT2 Non-driven shield, RMDU excitation	400	217
18	AMX 1-11	Shorted at RMDU plug	1000	217
19	AMX 1-2	PT2 Non-driven shield, RMDU bias	1000	217
20	AMX 1-15	AC-2 Non-driven shield, RMDU excitation	10	217

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LINE #	PARAMETER	DESCRIPTION	GAIN	SAMPLE RATE
21	AMX 1-15	AC-2 Non-driven shield, RMDU Excitation	3	217
22	AMX 1-24	CPT3 Driven shield External Excitation	2	108½
23	AMX 1-25	CPT4 Non-driven shield, External Excitation	2	108½
24	SF-ID 2	7002)8	-	$108\frac{1}{2}$
25	AMX 1-3	PT3 Driven shield <sup>C</sup> External Excitation	400	217
26 27	AMX 1-26 AMX 1-3	2KΩ Driven shield PT3 Driven shield,	1000	217
		External Excitation	1000	217
28	AMX 1-12	Shorted at RMDU plug	10	217
29 30	AMX 1-16 AMX 1-16	AC.3 Driven shield, External Excitation AC.3 Driven shield,	10	217
50	HHA I IO	External Excitation	3	217
31	AMX 1-22	CPT1 Driven shield,		
32	AMX 1-23	RMDU excitation CPT2 Non-driven shield	1	108 <sup>1</sup> 2
		RMDU excitation	1	108 <sup>1</sup> 2
33	SF-ID 3	7003) <sub>8</sub>	-	108½
34	AMX 1-4	PT4 Non-driven shield, External excitation	400	217
35	AMX 1-27	Tape recorder tachomete		
26	AMV 7 - 4	DTA New defines shield	1000	217
36	AMX 1-4	PT4 Non-driven shield, External Bias	1000	217
37	AMX 1-13	Shorted at RMDU plug	10	217
38	AMX 1-17	Ac4 Non-driven shield,	<b>+</b> •	<u> </u>
		External Excitation	10	217
39	AMX 1-17	Ac4 Non-driven shield		
		External Excitation	3	217
40	AMX 1-24	CPT3 Driven shield,		
41	AMX 1-25	External Excitation CPT4 Non-driven shield,	1	1081/2
		External Excitation	1	108 <sup>1</sup> 2
42	SF-ID 4	7004) <sub>8</sub>	-	108 <sup>1</sup> 2
43	AMX 6-1	TC-1 Driven shield	400	$108\frac{1}{2}$
44	AMX 6-2	TC-2 Non-driven shield	400	108 <sup>1</sup> 2
45	SF-ID 5	7005) <sub>8</sub>	-	108 <sup>1</sup> 2
46	AMX 6-1	TC-1 Driven shield	1000	108 <sup>1</sup> 2
47	AMX 6-2	TC-2 Non-driven shield	1000	10812
48	SF-ID 6	7006)8	-	108 <sup>1</sup> 2
49	AMX 1-26	$2K\Omega^*$ Driven shield <sup>18</sup>	400	108 <sup>1</sup> 2

LINE #	PARAMETER	DESCRIPTION	GAIN	SAMPLE RATE
50	AMX 1-27	Tape Recorder		t
51	SF-ID 7	tachometer 7007) <sub>8</sub>	400	$108\frac{1}{2}$ $108\frac{1}{2}$
52 53	PSB LLC-1(AMX 1-33)	Power supply BITE Low level calibration	-	108½ 108½

$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	MEMODY			A C Code Co	G Y ode Code	X Code	C Code	
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$		PARAMETER	<u>GAIN</u>	<u>4 3 2 1</u> <u>4 3</u> D 7		<u>1 3 2 1</u>		<u>Octal</u>
129AMX 1-14000 1 0 0 0 0 1 0 0 0 0 1 0 0 0 0 0 0 0 1nsf102 001130AMX 1-1010000 1 0 0 0 0 0 0 1 0 0 0 0 0 0 0 0 0 0 1nsf101 055131AMX 1-110000 1 0 0 0 0 0 0 1 0 0 0 0 0 0 0 0 0 1nsf101 001132LLC-110000 1 0 0 0 0 0 0 1 1 0 1 0 0 0 0 0 0 0 1nsf101 251133GPA-Ø101 1 1 0 0 1 0 1 0 1 0 0 0 0 0 0 0 1 0 1	1 2 3 4	SYNC 2 SF-ID HLC LLC-1		0.0 0 0 1 0 1 1 1 0 0 0 1 1 1 1 0 0 0 1 0 0 0 0	1     1     1     0       0     0     0     0       0     0     0     0       1     0     1     0       1     0     1     0	0 0 0 0 0 0 0 0 0 0 0 0 1 0 1 0	00 nm 00 nm 00 nm 00 nm	013 200 340 000 360 000 102 250
153       AMX 1-16       10       0 1 0 0 0 1 0 1 0 1 0 1 0 1 0 1       nsf       105 115         154       AMX 1-16       3       0 1 0 0 1 0 0 0 0 1 0 0 1 1 0 1       nsf       110 115         155       AMX 1-22       1       0 1 0 0 1 0 1 0 0 1 1 0 1 1 0 1       nsf       112 155         156       AMX 1-23       1       0 1 0 0 1 0 1 0 0 1 1 0 0 1 1 0 0 1       10 0 1 2 162	129 130 131 132 133 134 135 136 137 138 139 140 141 142 143 144 145 146 147 148 149 150 151 152 153 154 155	AMX 1-1 AMX 1-10 AMX 1-1 LLC-1 GPA-Ø AMX 1-14 AMX 1-14 AMX 1-22 AMX 1-23 GPA-Ø AMX 1-23 GPA-Ø AMX 1-2 LLC-1 GPA-Ø AMX 1-15 AMX 1-15 AMX 1-15 AMX 1-25 AMX 1-25 AMX 1-25 AMX 1-3 LLC -1 AMX 1-16 AMX 1-16 AMX 1-22	$\begin{array}{c} 400\\ 1000\\ 1000\\ 1000\\ 10\\ 10\\ 2\\ 2\\ 400\\ 400\\ 1000\\ 1000\\ 1000\\ 1000\\ 10\\ 10\\ 3\\ 2\\ 2\\ 400\\ 1000\\ 1000\\ 1000\\ 1000\\ 1000\\ 100\\ 10\\ 3\\ 1\\ \end{array}$	$\begin{array}{c} 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 \\ \end{array}$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	$\begin{array}{c} 0 & 0 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \\ 1 & 0 & 0 \\ 1 & 0 & 0 \\ 1 & 0 & 0 \\ 1 & 0 & 0 \\ 1 & 0 & 0 \\ 1 & 0 & 0 \\ 1 & 0 & 0 \\ 1 & 0 \\ 1 & 0 \\ 1 & 0 \\ 1 & 0 \\ 1 & 0 \\ 1 & 0 \\ 1 & 0 \\ 1 & 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1 \\ 1$	0 1 nsf 0 1 nsf	$\begin{array}{cccccccccccccccccccccccccccccccccccc$

# EPROMS 1 and 2 MAIN FORMAT MEMORY PROGRAM

MEMORY			A Codé	G Code	Y Code	X C Code Code	
ADDRES		GAIN	<u>4321</u> D 7	4'32	$ \begin{array}{c} 1 3 2 1 \\ \overline{D D} \\ 0 7 \end{array} $	$\frac{321}{21}$	<u>Octal</u>
158 159 160 161 162 163 164 165 166 167 168 169 170 171 172 173	AMX 1-27 AMX 1-4 LLC-1 AMX 1-13 AMX 1-17 AMX 1-17 AMX 1-24 AMX 1-25 AMX 6-1 AMX 6-2 AMX 6-1 AMX 6-2 AMX 6-1 AMX 6-2 AMX 1-26 AMX 1-27 PSB LLC-1	$     \begin{array}{r}       1000\\       1000\\       10\\       10\\       10\\       3\\       1\\       1\\       400\\       400\\       1000\\       1000\\       400\\       -100     \end{array} $	1001 0100 0100 0000	0 0 0 0 0 0 0 0 0 1 0 0 1 0 0 0 1 0 1 0 1 0 1 0 1 0 1 0 0 0 0 0 0 0 1 0 0 0 1 0 0 0 1 0 1 0 0 0 1 0 0 0 0 0 0 0	$ \begin{array}{c} 1 & 1 & 0 & 0 \\ 1 & 0 & 0 & 0 \\ 1 & 1 & 0 & 1 & 0 \\ 1 & 0 & 1 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 \\ 1 & 1 & 0 & 1 & 0 & 1 \\ \end{array} $	$\begin{array}{c} 0 1 1 0 1 \\ 0 1 0 0 1 \\ 0 0 0 1 \\ 1 0 0 0 1 \\ 1 0 0 0 1 \\ 1 0 0 0 1 \\ 1 0 1 0$	nsf 101 211 nsf 101 015 nsf 101 251 nsf 105 101 nsf 105 121 nsf 110 121 nsf 110 121 nsf 112 165 eof 112 202 nsf 222 001 eof 222 006 nsf 221 001 eof 221 006 nsf 102 205 eof 102 212 nsf 000 001 eoc 103 253
N	nm ≕ nsf	Code: next mai next su end of	bframe				

eoc = end of cycle
A Code = Card Select
G Code = Gain Select
YX Code = Channel Select

# EPROMS 1 and 2 MAIN FORMAT MEMORY PROGRAM (CONCLUDED)

DATA CONTENTS

DATA CONTENTS

$\begin{array}{c c c c c c c c c c c c c c c c c c c $		A .F LLOWD . Y LYES			and the Alexander Server	
ADDRESSVALUE $D_7D$ DD1131100000111111111111111111111111111111111111 <t< td=""><td>MEMORY</td><td>DECIMAL</td><td>BINARY</td><td>MEMORY</td><td>DECIMAL</td><td>BINARY</td></t<>	MEMORY	DECIMAL	BINARY	MEMORY	DECIMAL	BINARY
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$						D_D D D D D D D
$\begin{array}{cccccccccccccccccccccccccccccccccccc$			-/			- / 0
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	0	128	10000000	40	128	10000000
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	1	129	10000001	41	129	10000001
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	2	130	10000010	42	130	1000010
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	3		10000011			
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$		132	10000100			10000100
$\begin{array}{cccccccccccccccccccccccccccccccccccc$		133	10000101	45		10000101
$\begin{array}{cccccccccccccccccccccccccccccccccccc$			10000110			10000110
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		135	10000111			10000111
$\begin{array}{cccccccccccccccccccccccccccccccccccc$		136	10001000	48		10100110
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$		137				10100111
11139100001011511391000010111214010001100521401000110013141100011005314110000110014142100011100541421000011100151431000100005614410000000016144100100000561441000000000171451000100000561441000000000000000000000000000000000000	10	138	10001010	50		10001010
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	11	139΄		51		10001011
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	12	140	10001100			
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	13	141	10001101			
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	14	142	10001110			10001110
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	15	143	10001111	55		10001111
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	16	144	10010000			10010000
18       146       1 0 0 1 0 0 1 0       58       168       1 0 1 0 1 0 0 0         19       147       1 0 0 1 0 0 1 1       59       169       1 0 1 0 1 0 0 1         20       128       1 0 0 0 0 0 0 0       60       128       1 0 0 0 0 0 0       0         21       148       1 0 0 1 0 1 0 1       61       148       1 0 0 1 0 1 0 1       0         22       149       1 0 0 1 0 1 0 1       62       149       1 0 0 1 0 1 0 1       1         23       150       1 0 0 1 0 1 1 0       63       150       1 0 0 1 0 1 1 0       1         24       151       1 0 0 1 0 1 1 1       64       151       1 0 0 1 0 1 1 1	17	145	10010001			10010001
19       147       1 0 0 1 0 0 1 1       59       169       1 0 1 0 1 0 0 1         20       128       1 0 0 0 0 0 0 0       60       128       1 0 0 0 0 0 0         21       148       1 0 0 1 0 1 0 1 0       61       148       1 0 0 1 0 1 0 1         22       149       1 0 0 1 0 1 0 1       62       149       1 0 0 1 0 1 0 1         23       150       1 0 0 1 0 1 1 0       63       150       1 0 0 1 0 1 1 0         24       151       1 0 0 1 0 1 1 1       64       151       1 0 0 1 0 1 1 1	18	146	10010010	58		10101000
20       128       10000000       60       128       10000000       0000         21       148       10010100       61       148       10010100       10100         22       149       10010101       62       149       10010101       10101         23       150       10010110       63       150       10010110       10010110         24       151       10010111       64       151       10010111	19	147	10010011			10101001
22       149       10010101       62       149       10010101       101         23       150       10010110       63       150       10010110       10010110         24       151       10010111       64       151       10010111	20	128	10000000		128	10000000
22       149       10010101       62       149       10010101         23       150       10010110       63       150       10010110         24       151       10010111       64       151       10010111	21	148	10010100			10010100
24 151 10010111 64 151 10010111	22	149	10010101	62	149	10010101
	23	150		63	150	10010110
	24	151	10010111	64	151	10010111
	25	152				10011000
26 153 10011001 66 153 10011001		153	10011001			10011001
27 154 10011010 67 154 10011010	27	154	10011010			10011010
28 155 10011011 68 170 10101010	28	155	10011011	68	170	10101010
29 156 10011100 69 171 10101011	29	156	10011100		171	10101011
30 138 10001010 70 138 10001010	30	138	10001010			10001010
31 157 10011101 71 157 10011101	31	157	10011101			
32 158 10011110 72 158 10011110	32	158	10011110			
33 159 10011111 73 159 10011111	33	159	10011111			
34 160 1010000 74 160 1010000						
35 161 10100001 75 161 10100001						
36 162 10100010 76 162 10100010			10100010			
37 163 10100011 77 163 10100011						
38 164 10100100 78 172 10101100						
39 165 10100101 79 173 10101101	39	165	10100101			

SUBFRAME STEERING (EPROM 3) PROGRAM

			CONTENTS					
MEMORY ADDRESS	MSI D.	B 7 <sup>D</sup>	D	D	D	D	D	LSB D <sub>0</sub>
0	1	1	1	1	0	0	1	0
1	1	1	1	1	1	0	1	0
2	1	1	1	1	1	0	0	1
3	1	1	1	1	1	1	1	1
•	٠	•	•	٠	•	•	•	•
•	•	•	•	•	•	•	٠	•
•	•	٠	•	٠	٠	•	٠	•
•	•	•	٠	•	•	•	٠	•
255	1	1	1	1	1	1	1	1

•

.

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Note:	0 in D <sub>0</sub> SYNC word(s)
	0 in D <sub>1</sub> SF-ID word
	0 in D <sub>2</sub> SYNC or ID word(s)
	0 in D <sub>3</sub> 1 <sup>st</sup> word in frame

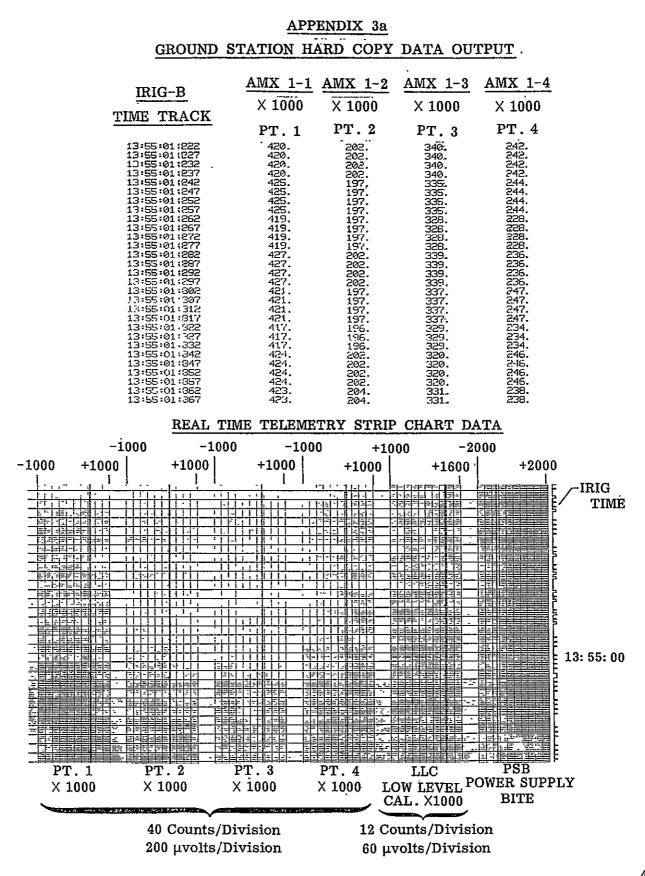
-

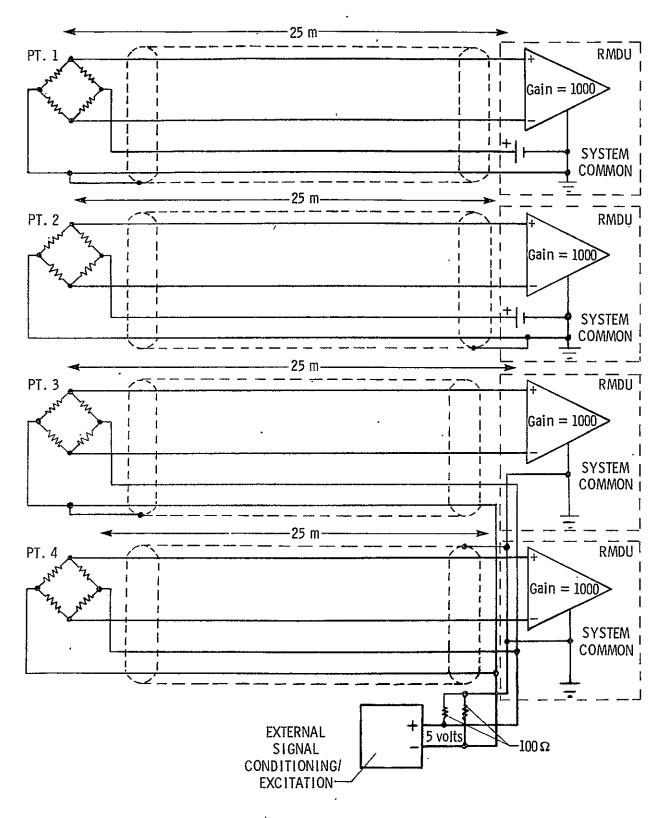
OVERHEAD WORD LOCATOR (EPROM 4) PROGRAM

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# APPENDIX 3a

3a. Real Time TM Data and ground station hard copy data for PT.  $1 \rightarrow$  PT.4





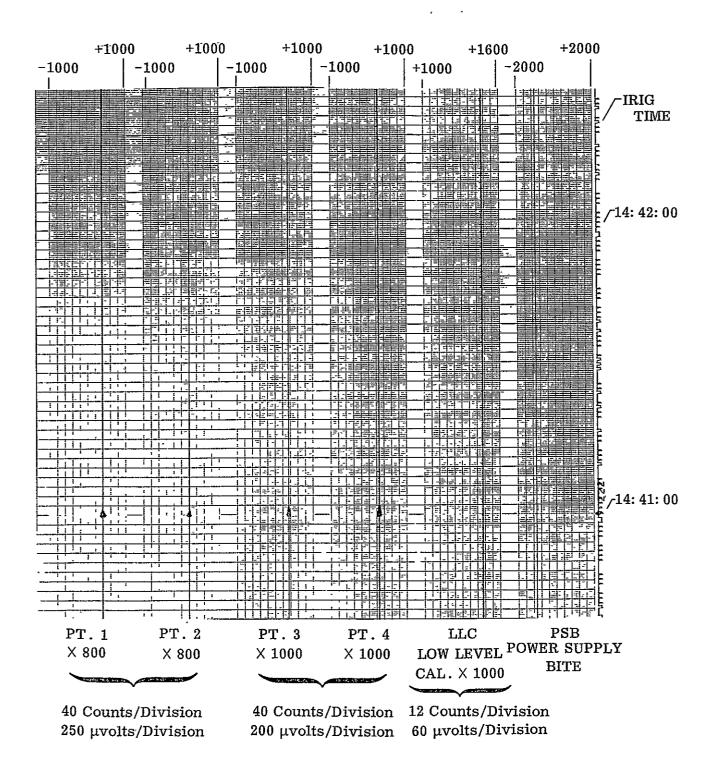
SHIELDING AND EXCITATION CONFIGURATION FOR PT. 1—PT. 4

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# APPENDIX 3b

3b. Real Time TM Data and ground station hard copy data for PT.  $1 \rightarrow$  PT. 4

## REAL TIME TELEMETRY STRIP CHART DATA



ORIGINAL PAGE IS OF POOR QUALITY

#### 12/29/77 14:48:28 FOHC

SYSTEMS REAL-TIME MONITOR-5.2

LISTING END TIME: 23:23:23:000

LISTING START TIME: 00:00:00:000

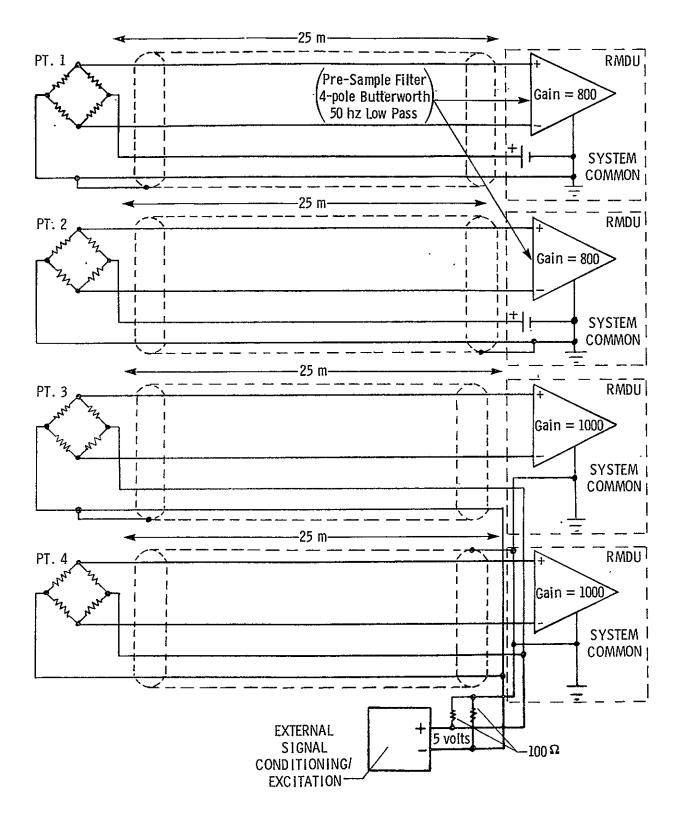
### JETSTAR HLT 479 AIFTDS SYSTEM COUNTS

IRIG-B	
-	

TIME	TRACK	

	-
1 4 . 41 .00 .007	
14:41:00:067	
14:41:00:082	
14:41:00:087	
14:41:00:092	
14:41:00:097	
14:41:00:102	
14:41:00:107	
14:41:00:112	
14:41:00:117	
14:41:00:122	
14:41:00:127	
14:41:00:132	
14:41:00:137	
14:41:00:142	
14:41:00:147	
14:41:00:152	
14:41:00:157	
14:41:00:162	
14:41:00:167	
14:41:00:172	
14:41:00:177	
14:41:00:182	
14:41-100:187	
14:41:00:192	
14:41:00:197	
14:41:00:202	
14:41:00:207	
14:41:00(212	
14:41:00:217	
14:41:00:222	
14:41:00:227	
14:41:00:232	
14:41:00:237	
14:41:00:242	
14:41:00(247	
14:41:00:252	
14:41:00(257	
14:41:00:262	
14:41:00(267	
14:41:00:277	
14:41:00:282	
14:41:00(287	
14:41:00(292	
14:41:00 (297	
14:41:00:002	
14:41:00:307	
14:41:00:812	
14:41:00:317	
14:41:00:822	
4-1-11 100 10CC	

PT. 1	PT. 2	PT.3 CH 83	PT.4 CH 84
CH 81	CH 82		
A1X800	A2X800	A3X1000	A4X1000
195. 195.		345. 339.	250. 252. 252.
195.	376.	339. 339.	252.
195. 195. 195.	376. 376.	339.	252.
195. 195.	376. 376.	341. 341.	253.
195.	376.	341.	252. 253. 253. 253. 253. 253.
195. 195.	376. 377.	341. 344.	250.
195. 195.	377.	344. 344.	250. 250.
195.	377. 377.	344.	250.
100	376. 376.	314. 344.	252.
195. 195. 195. 195. 195. 195.	376.	344.	252. 252. 252.
195.	376. 376.	344. 343	252.
195.	376.	343 343. 343.	244.
195.	376. 376.	343.	244. 244.
195. 195. 195.	376. 376.	345.	247. 247.
195.	376.	345.	247.
195. 196.	376. 376.	345. 339.	247. 252.
195 196.	276.	<b>33</b> 9.	252.
196.	376. 376.	339. 339.	252.
195.	376. 376.	342. 342.	248. 248.
195. 195. 195. 195.	376	342	248.
195.	376. 376.	342. 344.	248. 256.
195.	376. 376.	344. 344.	256
195. 195.	375.	344.	2556. 2557. 257.
195.	376. 376.	344. 344.	257.
195.	376.	344.	257
L95. 195.	377. 377.	345. 345.	249.
195.	377.	345.	249.
(95. 195.	377. 377.	345. 344.	249. 256.
195. 195.	377. 377.	344. 344.	256.
195.	377	344.	256.
195. 195.	376. 376.	340. 340.	236.



SHIELDING AND EXCITATION CONFIGURATION FOR PT.  $1 \rightarrow$  PT. 4

PT.1 AND PT.2 (From Flight 439 - Unfiltered Gain = 1000)

٠

PT.1 Gain = 1000 Unfiltered

Median value	=	374.00	Counts
Mean value X	=	374.44	Counts
Maximum value	=	383.00	Counts
Minimum value	=	367.00	Counts
RMS deviation $\sigma$	=	2.89	Counts

## PT.2 Gain = 1000 Unfiltered

Median value		=	158.00	Counts
Mean value $\overline{X}$		=	163.74	Counts
Maximum value		=	256.00	Counts
Minimum value		=	85.00	Counts
RMS deviation	σ	=	54.32	Counts

# PT.1 AND PT.2 GAIN = 800 FILTERED (From Flight #479)

## PT.1 Gain = 800 Filtered

.

Median value	-	376.00	Counts
Mean value X	=	376.33	Counts
Maximum value	=	377.00	Counts
Minimum value	=	376.00	Counts
RMS deviation	=	0.47	Counts

## PT.2 Gain = 800 Filtered

Median value		=	195.00	Counts
Mean value X		=	195.04	Counts
Maximum value		=	196.00	Counts
Minimum value		=	195.00	Counts
RMS deviation	σ	=	0.20	Counts

.

# APPENDIX 3c

3c. Internal precision voltage response for gains of 400 and 1000

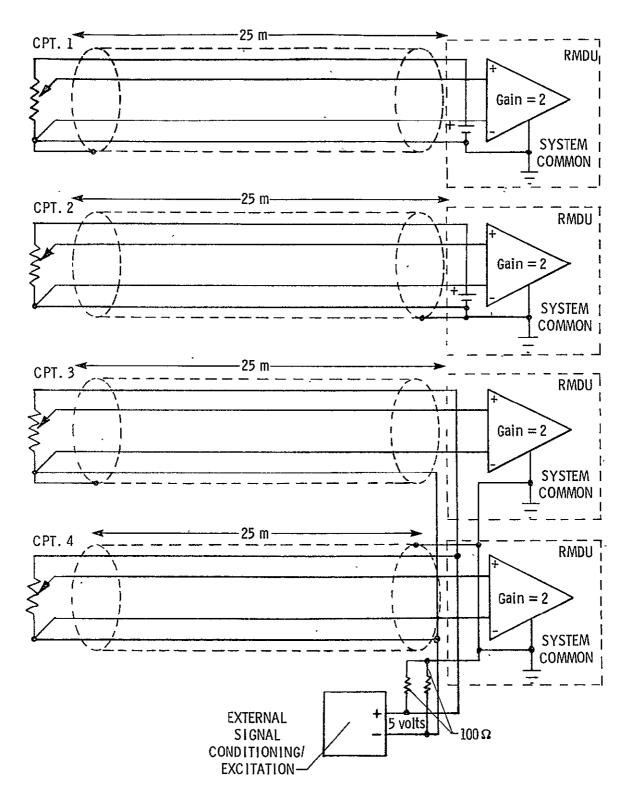
PRECISION VOLTAGE CHANNELS: (From Flight #481)

Low Level Calibration (7.3237 mv) at Gain = 1000: Median value Mean value X = 1458.00 Counts 1457.89 Counts = Maximum value = 1460.00 Counts = 1454.00 Counts Minimum value RMS deviation  $\sigma$  = 1.15 Counts Shorted Channel Response at Gain = 400: Median value = 2.00 Counts Mean value  $\overline{X}$ = 1.99 Counts Maximum value = 8.00 Counts Minimum value = 0.00 Counts RMS diviation  $\sigma$  = 1.08 Counts Gain Programmable Amplifier's Zero Response ( $GPA-\emptyset$ ) at Gain = 400: Median value = 4.00 Counts = 3.90 Counts Mean value Maximum value = 6.00 Counts Minimum value = 1.00 Counts RMS deviation  $\sigma = 0.62$  Counts

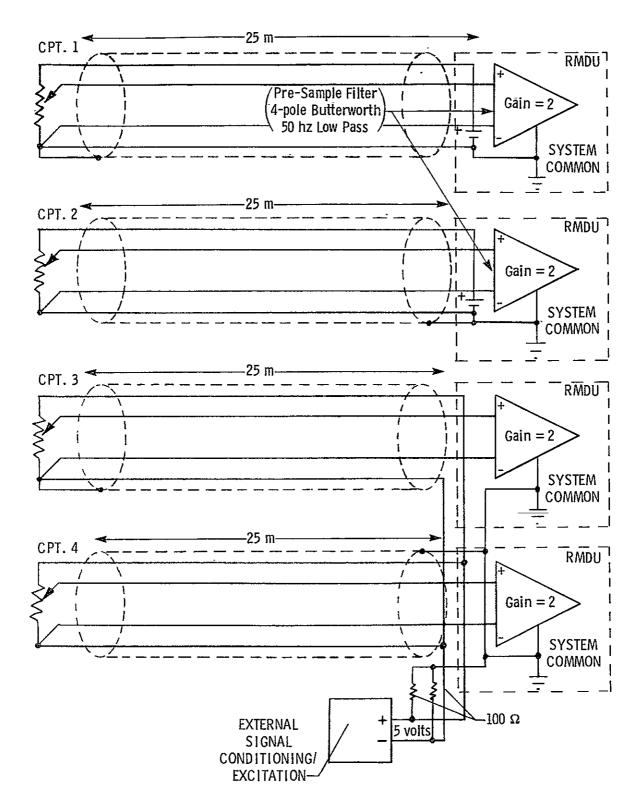
NOTE: For Gains of 1000 each Count =  $5\mu$  volts For Gains of 400 each Count =  $12\mu$  volts

## APPENDIX 3d

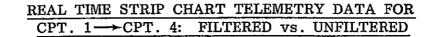
3d. Strip chart and ground station hard copy data for  $CPT1 \rightarrow CPT4$  showing shielding, excitation, and pre-sample filter response.

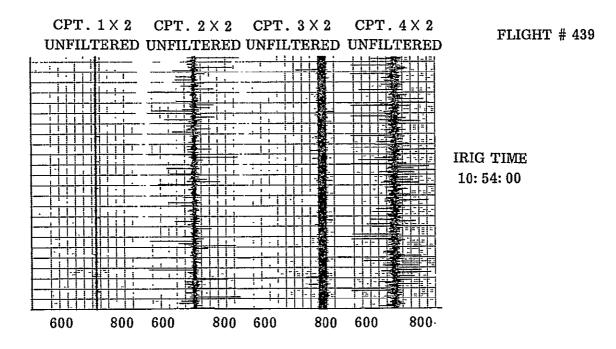


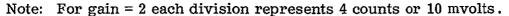
SHIELDING AND EXCITATION FOR CPT. 1→CPT. 4; FLIGHT #439

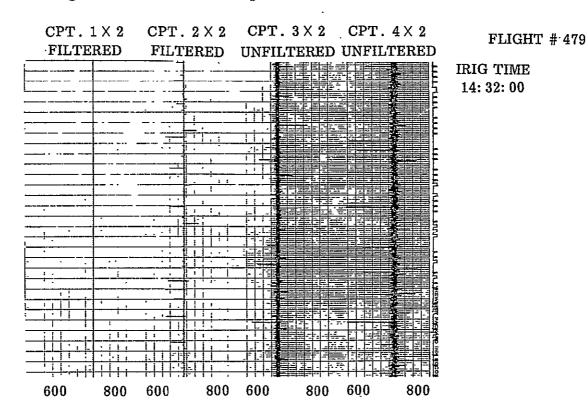


SHIELDING AND EXCITATION - FILTERING FOR CPT. 1-+ CPT. 4; FLIGHT #479









00:09.26 POHC

SYSTEMS REAL-TIME MONITOR-5.2

Page 2

LISTING END TIME: 23:23:23:000

LISTING START TIME: 00:00:00:000

## JETSTAR FLT 4 9 AIFTDS SYSTEM COUNTS

TIME	ICH131	CH141 A23X2	CH132 A24X2	CH142 A25X2	(CH133 (A22X1	CH143 A23X1	CH134 A24X1	CH144 A25X1		CH104 13X10
$10:54:00:413\\10:54:00:428\\10:54:00:428\\10:54:00:428\\10:54:00:433\\10:54:00:433\\10:54:00:433\\10:54:00:433\\10:54:00:453\\10:54:00:453\\10:54:00:458\\10:54:00:458\\10:54:00:458\\10:54:00:458\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:463\\10:54:00:453\\10:54:00:453\\10:54:00:453\\10:54:00:453\\10:54:00:453\\10:54:00:453\\10:54:00:453\\10:54:00:453\\10:54:00:4558\\10:54:00:453\\10:558\\10:558\\10:558\\10:558\\10:558\\10:558$	710. 710. 710. 710. 710. 710. 710. 710.	711. 711. 711. 711. 711. 711. 711. 711.	780. 780. 7772. 7775. 77	696. 696. 896. 896. 896. 710. 710. 710. 710. 710. 710. 710. 710	356. 356. 356. 3556. 3554. 44. 3554. 3555.	2 2 CPT.	3900. 39905. 399		1. 1. 1. 1. 1. 1. 1. 1. 1. 1.	1. 1. 1. 1. 0. 0. 0. 0. 0. 0. 0. 0. 0. 0
	X 2	X 2	X 2	X 2	X 1	X 1	X 1	X.1	CHANNEL	CHANNEL
									X 1	X 1

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00:09:26 POHC

### SYSTEMS REAL-TIME MONITOR-5.2

### PAGE

LISTING END TIME: 23:23:23:000

LISTING START TIME: 00:00:00:000

### JETSTAR RLT 439 AIFTDS SYSTEM COUNTS

TIME	ICH131 IA22X2	CH141 A23X2	CH132 A24X2	CH1 42 A25X2	ICH123 IA22X1	CH143 A23X1	CH134 A24X1	Ch144 A25X1	1CH103 A12X10 A	CH104 13X10
$\begin{array}{l} 10:54:00:163\\ 10:54:00:173\\ 10:54:00:173\\ 10:54:00:173\\ 10:54:00:133\\ 10:54:00:133\\ 10:54:00:133\\ 10:54:00:133\\ 10:54:00:203\\ 10:54:00:203\\ 10:54:00:203\\ 10:54:00:203\\ 10:54:00:223\\ 10:54:00:223\\ 10:54:00:223\\ 10:54:00:223\\ 10:54:00:223\\ 10:54:00:223\\ 10:54:00:223\\ 10:54:00:233\\ 10:54:00:$	71177111777777777777777777777777	700. 702. 702. 702. 702. 702. 702. 702.	774. 774. 774. 774. 781. 781. 781. 781. 781. 781. 782. 762. 762. 762. 762. 762. 762. 762. 76	687. 700. 700. 700. 700. 700. 700. 700. 7	35555554444444444444444444444444444444			349. 3555. 3555. 3555. 3555. 3555. 3555. 3555. 3555. 3553. 3553. 3553. 3553. 3553. 3553. 3553. 3553. 3553. 3553. 3553. 3553. 3555. 3	***************************************	1
	CPT.1 X1	CPT.2 ×1	CPT.3 ×1	CPT.4 ×1	CPT.1 × 2	CPT.2 ×2	CPT.3 × 2	CPT.4 X2	SHORTED CHANNEL × 1	

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LISTING START TIME: 00:00:00:000

LISTING END TIME: 23:23:23:000

## JETSTAR FLT 479 AIFTDS SYSTEM COUNTS

TIME	ICH131 IA22X2	CH141 A23X2	CH132 A24X2	CH142 A25X2	ICH133 IA22X1	CH143 A23X1	CH134 A24X1	CH144 A25X1	ICH103 A12X10	CH104 A13X10	
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#### SYSTEMS REAL-TIME MONITOR-5.2

#### PAGE 5

LISTING START TIME: 00:00:00:000

### LISTING END TIME: 23:23:23:000

JETSTAR HLT 479 AIFTDS SYSTEM COUNTS

TIME	ICH131 1A22X2	CH141 A23X2	CH132 A24X2	CH142 A25X2	ICH133 IA22X1	CH143 A23X1	CH134 A24X1	CH144 A25X1	TICH103 CH10 A12X10 A13X1	
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UNFILTERED DATA FROM CPT.1 AND CPT.2 (From Flight #439)

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# FILTERED DATA FROM CPT.1 AND CPT.2 (From Flight #479)

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Maximum value			716.00	
Minimum value		Ħ	716.00	
RMS deviation	σ	Ħ	0.00	Counts

# CPT.2 Gain = 2 Filtered

Median value		=	692.00	Counts
Mean value		=	692.32	Counts
Maximum value		=	694.00	Counts
Minimum value		=	692.00	Counts
RMS deviation	σ	=	0.61	Counts

1. Report No. NASA TM-72	867	2 Government Access	ion No	3. Recipient's Catalog No								
4 Title and Subtitl	e I			5. Report Date July 1979								
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7. Author(s)		<u> </u>		8. Performing Organiza	ation Report No.							
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Figure C-6. Fan Calibration, Fan No. 5, Ser. No. 421 (Concluded)

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