

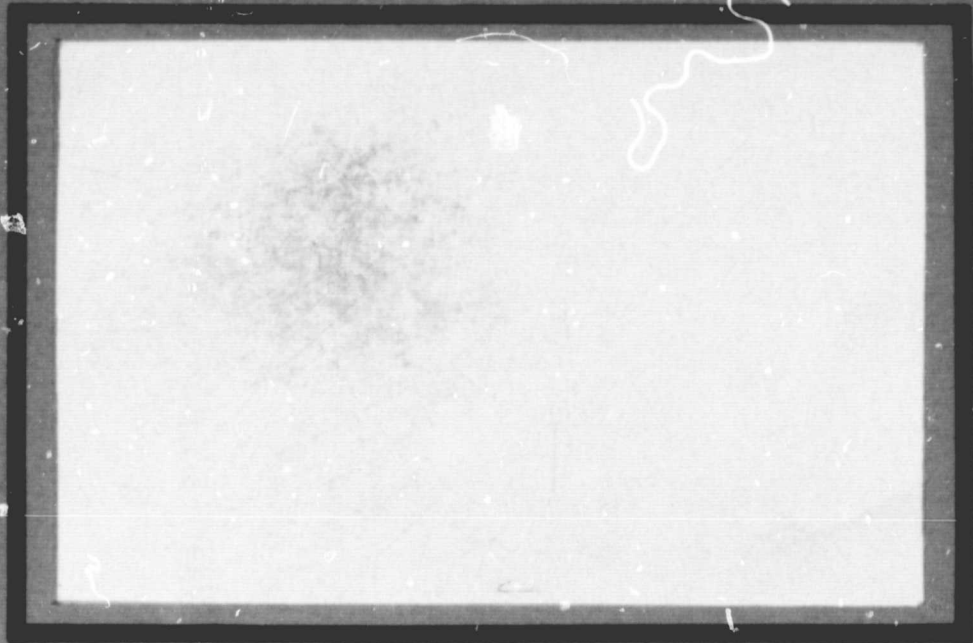
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SIMULATED LUMPED-PARAMETER SYSTEM
REDUCED-ORDER ADAPTIVE CONTROL
STUDIES

by

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1. INTRODUCTION

Very little conclusive insight is currently available for predicting the behavior of reduced-order adaptive controllers (ROAC), i.e. an adaptive controller based on a plant model of lower dimension than that of the actual plant. In [1] the ROAC problem is seen to be an unavoidable consequence of the application of any finite-dimensional, lumped-parameter system (LPS) adaptive control strategy to an infinite dimensional, distributed-parameter system (DPS). In [2] and [3] the relative orders of the plant, model, and controller are noted to be critical. In [4] the problem of adaptive parameterization of a C order controller for a M order model of a N order system is addressed when $N = M > C$. However, as noted in [2] and [3], the practical, poorly understood case is when $N > M \geq C$. This condition, i.e. the system order exceeds that of the plant model used for adaptive controller parameterization, will be considered to constitute the ROAC problem in this report.

Though numerous strategies have been espoused [5] for extracting a low order model from a too complex system description, none currently seem fully applicable to the real-time, recursive requirements of on-line adaptive control algorithms. These reduced-order modeling techniques seem to fall into two broad categories: (i) Extract those "modes" (or component-subsystems) from the full system description that are most influential in the performance of the model in its subsequent use. This strategy is followed, for example, in [6] and [7]. (ii) Parameterize the reduced-order structure to provide the best prediction of the desired output. This latter strategy will produce a model the "modes" of which need not correspond to any of those of the full system as noted in [3]. This latter strategy includes the model reference approach prominent in adaptive systems [8]. Two methods of interpreting the misbehavior of ROAC, one

based on system input-output description [3] [9] and one on state-variable description [1] [10], have emerged recently from attempts to develop adaptive controllers for flexible spacecraft.

Consider the implementation of the proper, single-input, single-output (SISO), autoregressive, moving-average (ARMA) system

$$y(k) = \sum_{i=1}^N [a_i y(k-i) + b_i u(k-i)] \quad (N \text{ even}), \quad (1-1)$$

where u is the system input and y the output, in parallel, i.e. partial fraction expanded, form as

$$y_\ell(k) = \sum_{i=1}^2 [\alpha_{i\ell} y_\ell(k-i) + \beta_{i\ell} u(k-i)] \quad (1-2)$$

$$y(k) = \sum_{\ell=1}^{N/2} y_\ell(k), \quad (1-3)$$

where y_ℓ is the output of the ℓ -th mode. As interpreted in [3] the first strategy in the preceding paragraph when used for identification would lead to approximation of (1-3) as

$$y_s(k) = \sum_{\ell}^{M/2 \text{ of } [1, N/2]} y_\ell(k) \quad (1-4)$$

with (1-2) describing the dynamics of each $y_\ell(k)$. The $M/2$ modes chosen in (1-4) may be selected as those $M/2$ from the $N/2$ in (1-3) that provide the "best" fit of y_s to y . Note that the order M of (1-4) refers to the order of the underlying reduced-order ARMA model. If the dimension of (1-4) (and (1-1)) were based on the number of quadratic "modes" then M (and N) would be replaced by $\bar{M} = \frac{M}{2}$ (and $\bar{N} = \frac{N}{2}$). This quadratic "modal" designation of order is common in the flexible spacecraft literature [3]. As noted in [1], [3], and [11] the extraction of the y_ℓ chosen in (1-4) from the y in

(1-3) remains an open question. However, these y_ℓ would be required for identification of the appropriate (1-2) parameters $\alpha_{i\ell}$ and $\beta_{i\ell}$. Alternatively, the second strategy would approximate (1-3) with

$$\hat{y}(k) = \sum_{\ell}^{M/2 \text{ of } [1, \frac{N}{2}]} \hat{y}_\ell(k) \quad (1-5)$$

in, e.g., a least-squares sense by selection of the $\hat{y}_\ell(k)$. These estimated "modal" outputs need not match the corresponding values in (1-2) and, therefore, need not lead to identification of the corresponding $\alpha_{i\ell}$ and $\beta_{i\ell}$ in (1-2) as noted in [3]. The successive \hat{y}_ℓ chosen to fit (1-5) to (1-3) may not even obey a time-invariant difference equation of the form of (1-2) with y_ℓ replaced by \hat{y}_ℓ . The problem with closing the adaptive control loop via a simultaneous identification and control strategy could be intensified by feeding back the \hat{y}_ℓ instead of the unavailable y_ℓ to meet a modal control objective.

The alternate interpretation of the ill-effects of ROAC, derived from the flexible spacecraft control problem [1] [10], begins with the separation of a SISO, state-space model for, e.g., (1-1) into the reduced-order and unmodeled (or residual) segments as

$$\begin{bmatrix} x_N(k+1) \\ x_R(k+1) \end{bmatrix} = \begin{bmatrix} A_N & A_{NR} \\ A_{RN} & A_R \end{bmatrix} \begin{bmatrix} x_N(k) \\ x_R(k) \end{bmatrix} + \begin{bmatrix} b_N \\ b_R \end{bmatrix} u(k) \quad (1-6)$$

$$y(k) = [c_N \quad c_R] \begin{bmatrix} x_N(k) \\ x_R(k) \end{bmatrix} \quad (1-7)$$

As outlined in [1] the derivation of (1-6) - (1-7) can be viewed as a

projection operation on the full system. From (1-7) the "spillover" of the residual modes into the observation of $y(k)$ via $c_R x_R(k)$ and from (1-6) the "spillover" of the control designed for the reduced model into the residual modes $x_R(k+1)$ via $b_R u(k)$ are clearly displayed. Also the possible coupling of reduced model modes and residual modes via A_{NR} and A_{RN} is immediately apparent. Shown in [1] and [10] is the predictable fact that if $A_{NR} = \underline{0}$ and $c_R = \underline{0}$ then, assuming the residual modes remain stable, any full-order (N), stable adaptive controller identifying A_N , b_N , and c_N explicitly or implicitly from y and u alone would retain its stability. If $A_{NR} = \underline{0}$ but c_R and b_R are nonzero then the degradation of ROAC has two sources. Unmodeled components in y via c_R will generate an error that is indistinguishable from parameter error thereby causing adaptation. The application of control u to the residual states x_R via b_R will contribute further to this unmodeled component of y . Not only will the parameter estimates be incorrect from use of y and u to identify only A_N , b_N , and c_N but also the state estimates provided by an adaptive observer will be incorrect, which if fed back could lead to an unpredictable "controlled" response.

These two problems of inappropriate parameter and state estimation are the same ones noted in the first interpretation with \hat{y} approximation of y in (1-5). The ability to extract the y_ℓ and obtain y_g corresponds to an effective zeroing of c_R . This report will, in part, attempt to implement, compare, and contrast these two strategies embodied in (1-4) and (1-5).

The next section details the specific objectives of this study. Section III presents the example autoregressive, moving-average plants that are to be used in the simulations. Section IV presents the adaptive

control algorithms to be used and their sources in the literature. Section V outlines the formats for the simulated tests including the description of numerical figures of merit to be tabulated in Section VI. Section VII offers interpretations of the test results and Section VIII draws conclusions relevant to the ROAC problem. The last sections of this report include the referenced literature and the appendices including computer program listings.

II. OBJECTIVES

The principal objective of this study is to test the usefulness of the folklore of reduced-order modelling with respect to adaptive control. In particular four "facts" will be tested:

- (i) Heavily damped modes may be neglected relative to more lightly damped modes in reduced-order-model derivation.
- (ii) Finite bandwidth actuators limit the number of modes necessary to be modeled.
- (iii) An "optimal" reduced-order controller neglects the modes contributing the least degradation in the control system performance measure.
- (iv) Indirect and direct adaptive control are essentially equivalent and interchangeable.

The implication to the ROAC problem of each of these statements will be developed in the following paragraphs. The simulations of the following sections will be chosen to test the veracity of these implications. The conclusion of this report will summarize the useful "facts" that either escape unscathed or emerge from these tests.

Since each of these facts has been accepted into the reduced-order and/or adaptive control folklore it is difficult to pinpoint particular references succinctly stating these points. However, several classical control texts contain the source of point (i) in the concept of dominant roots or poles. For example: "The complex conjugate roots near the origin of the s-plane relative to the other roots of the closed-loop system are labeled the dominant roots of the system since they represent or dominate the transient response. The relative dominance of the roots is determined by the ratio of the real parts of the complex roots and

will result in reasonable dominance for ratios exceeding five. ... Dominance .. also depends upon the relative magnitudes ... of the residues evaluated at the complex roots, [which] depend upon the location of the zeros in the s-plane" [12]. Or: "The relative dominance of closed-loop poles is determined by the ratio of the real parts of the closed-loop poles, as well as by the relative magnitudes of the residues evaluated at the closed-loop poles. The magnitudes of the residues depend upon both the closed-loop poles and zeros. If the ratios of the real parts exceed five, and there are no zeros nearby, then the closed-loop poles nearest the $j\omega$ axis will dominate in the transient-response behavior because these poles correspond to transient-response terms which decay slowly" [13, p. 251]. This separation concept has been formalized via singular perturbation theory [14]. Despite the concomitant warning in both [12] and [13] for caution in the use of this rule and the explication of its application to the closed-loop system, point (i) is commonly (though admittedly inappropriately) used for model reduction prior to control design. The (mis)implication for discrete systems is that if the open-loop singularities are separable into a group in the z-plane outside a radius of r (<1) and the other group inside the radius r^5 (since $|z|=e^{\text{Re}\{s\}T}$, where T is the sample period) then the first group alone provides a highly accurate input-output model of the entire system and therefore provides a useful dimension and parameterization for reduced-order controller design. One objective of this study is to test the usefulness of this guideline for ROAC. Predictably, such a rule will be valid only when the closed-loop system retains its open-loop singularity separability.

A frequently voiced (through unwritten) criticism of the pre-occupation with the effects of spillover is based on the assertion that the frequency content of the input to a plant has a strong effect on the most suitable reduced-order model. Since actuators do not have infinite bandwidth and are commonly modeled as essentially low pass filters [15] the higher frequency modes of the system will receive such insignificant excitation that they are ignorable as noted in point (ii). This again extrapolates a reasonable open-loop response mechanism to a closed-loop situation. Again, if the requirement that plant open-loop singularities were shifted only slightly by the feedback were included then this proposition would be strengthened. As such it narrows the more classical concept by utilizing the shape of the input spectrum in addition to the plant singularity constellation. One problem with the use of this idea in the ROAC problem is the nonlinear, time-varying character of the feedback, which does not fit the linear system character of this guideline. That is, during adaption a large, "high frequency" control effort could imbalance the roll-off provided by the low pass actuators. Another problem is the unmodeled phase shift induced by the actuators unless the actuator outputs are available as the identifier inputs. Due to the prevalence of this seemingly untested proposition, construction of a meaningful example for its examination represents another report objective. The influence of input frequency content on reduced-order controller selection will also be tested for various reference signal frequency content distributions.

Points (i) and (ii) rely on the near-equivalence of the two reduced-order modeling strategies interpreted in the introduction: modal selection and full-behavior approximation. The suggestion in point (iii) recognizes the distinction between these two strategies and represents the sensible result of pursuit of the first strategy embodied in (1-4). Also, clearly the modes of the plant may not be as separated as required in the discussion of point (i), either in the open-loop or closed-loop system, which requires a selection mechanism. As shown in [7] for lightly damped systems forced by infinite bandwidth inputs the modal costs are proportional to the product of the modal time constant, observability norm, and disturbability (or controllability) norm. Since for a partial fraction expansion the observability and disturbability norms are related to the modal residual, this suggestion can be viewed as a more sophisticated version of point (i). However, as demonstrated by example in [7] the low frequency modes need not always provide the best open-loop reduced-order model. Again this modal selection procedure is principally intended for an open-loop fit. For reduced-order control application the truncated modes can be viewed as the desired degrees of freedom omitted from the controller. However, a misinterpretation of this procedure would suggest that a reduced-order controller composed of a prespecified number of well-selected modal controllers would be optimal. This would suggest (falsely) that the second ROAC strategy of full approximation in the introduction, as represented by (1-5), could never prove better than the first of selective extraction, as represented by (1-4). The third objective of this study will be to test this implication by comparing single-mode controllers chosen to control single-mode reduced-models of a two mode system and a single-mode controller chosen to control the full system to "match" a single-mode objective. Such a

comparison will clearly rely on the example chosen and is expected, as with the first two objectives, to yield initially ambiguous results. Consider for example a second-order system controlled by a constant output feedback gain to meet a first order response matching the dominant pole of the root locus. The appropriate gain need not be provided by either separate modal controller but a close fit to the objective does exist for some reduced-order controller parameterization.

In the special case of full-order model following, indirect and direct adaptive control have recently been shown to be equivalent [16] [17]. Indirect adaptive control uses the current plant parameter (and state) estimates to solve for the controller parameters (and feedback signal). Direct adaptive control updates the controller parameters such that the control system response matches that of a prescribed model. Discovery of this equivalence has led to highly touted claims of the interchangeability of the algorithms resulting from these two approaches. The implication is that the two approaches are also equivalent for ROAC. Comparison of the indirect self-tuning scheme with the initial plant MA gain known, the direct input matching scheme, and the approximate direct output error identification interpretation with $q_1 = f_1$ in [17], all of which could be based on an equation error parameter estimator, does not dispute this claim for the specific model-following problem and equilibrating choice of adaptive gains. If the initial plant MA gain is also estimated then indirect and direct schemes can be expected to behave differently. Only recently has a simple provable direct adaptive control scheme, based on output error parameter estimation, been developed [18]. Like the approximate strategy in [19] a particular choice of designer selected constants reduces this strategy to the common

equation error based scheme in [17]. The complex generation of an additional reference model input required in [20] is avoided in [18].

As an aside, note that even though the reduction of a gradient based solution [21] and a stability theory based solution [22] to the equation error formulation of ARMA process identification prove identical this does not suggest the equivalence of gradient based [23] and stability theory based [24] solutions to the output error identification problem. At least currently, they are seen as special cases of a general, possible solution [25]. These two forms of adaptive parameter estimation underly the various possible adaptive control schemes suggesting differences for ROAC. Furthermore the distinction of equivalent convergence points versus convergence paths must be made. For full-order adaptive identification the asymptotic results may be similar though the transient behaviors are quite distinct. This transient disparity is amplified in reduced-order identification where the convergence points also become distinctly different [26]. Therefore, it would appear that the parameter estimation formulation and the explicit or implicit adaptive control strategy both lead to different ROAC behaviors. The fourth objective is to test the equivalence (or disparity) of indirect and direct adaptive controllers based on gradient and stability theory formulated equation and output error parameter estimators.

Achievement of the preceding objectives may provide suggestions for improving the presently available full-order adaptive controllers for ROAC use. Fulfillment of this hope is the ultimate objective of this report.

III. TEST EXAMPLES

Six test examples will be used to meet the objectives stated in the preceding section. They represent the following six categories:

- (i) Open-loop and closed-loop (in near satisfaction of desired model-following objective) system singularities are separable on a time basis due to significant damping differences.
- (ii) Open-loop system singularities are time separable but closed-loop singularities are only marginally separable.
- (iii) Neither open-loop nor closed-loop singularities are time separable but a near fit to desired behavior exists due to a lightly damped near-cancellation.
- (iv) Neither open-loop nor closed-loop singularities are time separable and no reduced-order controller gain closely satisfies the desired control objective.
- (v) Two nearly oscillatory open-loop modes of distinct frequency, which are only moderately shifted in the closed-loop, are preceded by an infinite bandwidth actuator.
- (vi) Two nearly oscillatory open-loop modes of the same distinct frequencies as in (v) are preceded by a low pass actuator providing at least -12 db (75%) attenuation at the resonant frequency of the second mode relative to the resonant frequency of the first.

Categories (i)-(iv) will test points (i) and (iii) of the preceding section. Categories (v) and (vi) will test point (ii). All of the categories will test point (iv).

Consider, for the first four categories, controlling a stable second

order plant with transfer function

$$\frac{Y(z)}{U(z)} = \frac{b}{z-a_1} + \frac{\epsilon}{z-a_2} = \frac{(b+\epsilon)z - (ba_2 + \epsilon a_1)}{(z-a_1)(z-a_2)} \quad (3-1)$$

and therefore difference equation description

$$\begin{aligned} y(k) &= (a_1 + a_2)y(k-1) - (a_1 a_2)y(k-2) \\ &+ (b+\epsilon)u(k-1) - (ba_2 + \epsilon a_1)u(k-2) \end{aligned} \quad (3-2)$$

and D.C. gain

$$\left. \frac{Y(z)}{U(z)} \right|_{z=1} = \frac{b}{1-a_1} + \frac{\epsilon}{1-a_2} \quad (3-3)$$

under the assumption of a first order model for (3-1) with the objective of following a first order model

$$s(k) = cr(k-1) + ds(k-1) . \quad (3-4)$$

Using the projection technique of (1-6)-(1-7), (3-1) can be rewritten as

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = \begin{bmatrix} a_1 & 0 \\ 0 & a_2 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} b \\ \epsilon \end{bmatrix} u(k) \quad (3-5)$$

$$y(k) = [1 \quad 1] \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} , \quad (3-6)$$

where, from (1-6)-(1-7), $A_N = a_1$, $A_R = a_2$, $A_{NR} = A_{RN} = 0$, $b_N c_N = b$, and $b_R c_R = \epsilon$. Note that any proportion of ϵ could be distributed to b_R and c_R . Therefore increased ϵ corresponds to increased spillover. Clearly when $\epsilon = b_R c_R = 0$, a first order model of (3-1) would be exact. In such a case, as noted earlier, any stable adaptive control scheme would be

successful. Also, equally apparent is the degradation of the reduced-order control system, adaptive or not, as ϵ becomes nonzero and the residual mode x_2 contributes significantly to y .

If ϵ were zero, then control of (3-1) via

$$u(k-1) = gr(k-1) + fy(k-1) \quad (3-7)$$

where

$$g = \frac{c}{b} \quad (3-8)$$

$$f = \frac{d-a_1}{b}, \quad (3-9)$$

would convert the plant output to

$$y(k) = a_1 y(k-1) + bu(k-1) = cr(k-1) + dy(k-1). \quad (3-10)$$

Therefore the model-following error

$$s(k) - y(k) = d[s(k-1) - y(k-1)] \quad (3-11)$$

would decay to zero if (3-4) were stable. If ϵ were not zero, use of (3-7) would convert (3-1) to

$$\begin{aligned} \frac{y(z)}{R(z)} &= g \left\{ \frac{(b+\epsilon)z - (ba_2 + \epsilon a_1)}{(z-a_1)(z-a_2) - f[(b+\epsilon)z - (ba_2 + \epsilon a_1)]} \right\} \\ &= \frac{g(b+\epsilon)(z-h)}{(z-p_1)(z-p_2)}, \end{aligned} \quad (3-12)$$

the roots of the characteristic equation of which, i.e. p_1 and p_2 , could be determined via a root locus. Note that f has a limited effect on the poles of (3-12) while g can be chosen to select the D.C. gain.

Alternatively the control of (3-1) (or (3-2)) could be chosen for (3-5)-(3-6) as

$$u(k) = gr(k) + fx_1(k), \quad (3-13)$$

if the reduced-order model state is assumed available. This assumption is equivalent to assuming the possibility of measuring $y_3(k)$ in (1-4) in the first reduced-order model strategy in the introduction. The control of (3-13) can be viewed as partial state feedback (impractically requiring state availability)

$$u(k) = gr(k) + [f \ 0] \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix}. \quad (3-14)$$

From (3-5), (3-6), and (3-14), (3-13) converts (3-1) to

$$\begin{aligned} \frac{Y(z)}{R(z)} &= g \begin{bmatrix} 1 & 1 \end{bmatrix} \left\{ zI - \begin{bmatrix} a_1 & 0 \\ 0 & a_2 \end{bmatrix} - \begin{bmatrix} b \\ c \end{bmatrix} [f \ 0] \right\}^{-1} \begin{bmatrix} b \\ c \end{bmatrix} \\ &= g \left\{ \frac{(b+\epsilon)z - (ba_2 + \epsilon a_1)}{(z-a_1 - bf)(z-a_2)} \right\} = \frac{g(b+\epsilon)(z-h)}{(z-q_1)(z-q_2)}. \end{aligned} \quad (3-15)$$

Note that relative to (3-12) only one pole is arbitrarily shifted, since $q_2 = a_2$, rather than both being shifted under the root locus constraint. Note that in either case, (3-12) or (3-15), ϵ can be chosen in constructing the test examples for a particular open-loop (and closed-loop, due to pole-shifting only by the controllers) zero location h

$$h = \frac{ba_2 + \epsilon a_1}{b + \epsilon} \Rightarrow \epsilon = \frac{b(a_2 - h)}{h - a_1}. \quad (3-16)$$

Example 1: $a_1 = 0.95$, $b = 0.065$, $a_2 = 0.2$, $\epsilon = 0.01$, $h = 0.3$, $c = 0.4$,
 $d = 0.8$

Recall that from (3-1) a_1 and a_2 are the open-loop plant poles, from (3-4) d is the desired closed-loop pole, from (3-12) and (3-15) p_1 , p_2 and q_1 ,

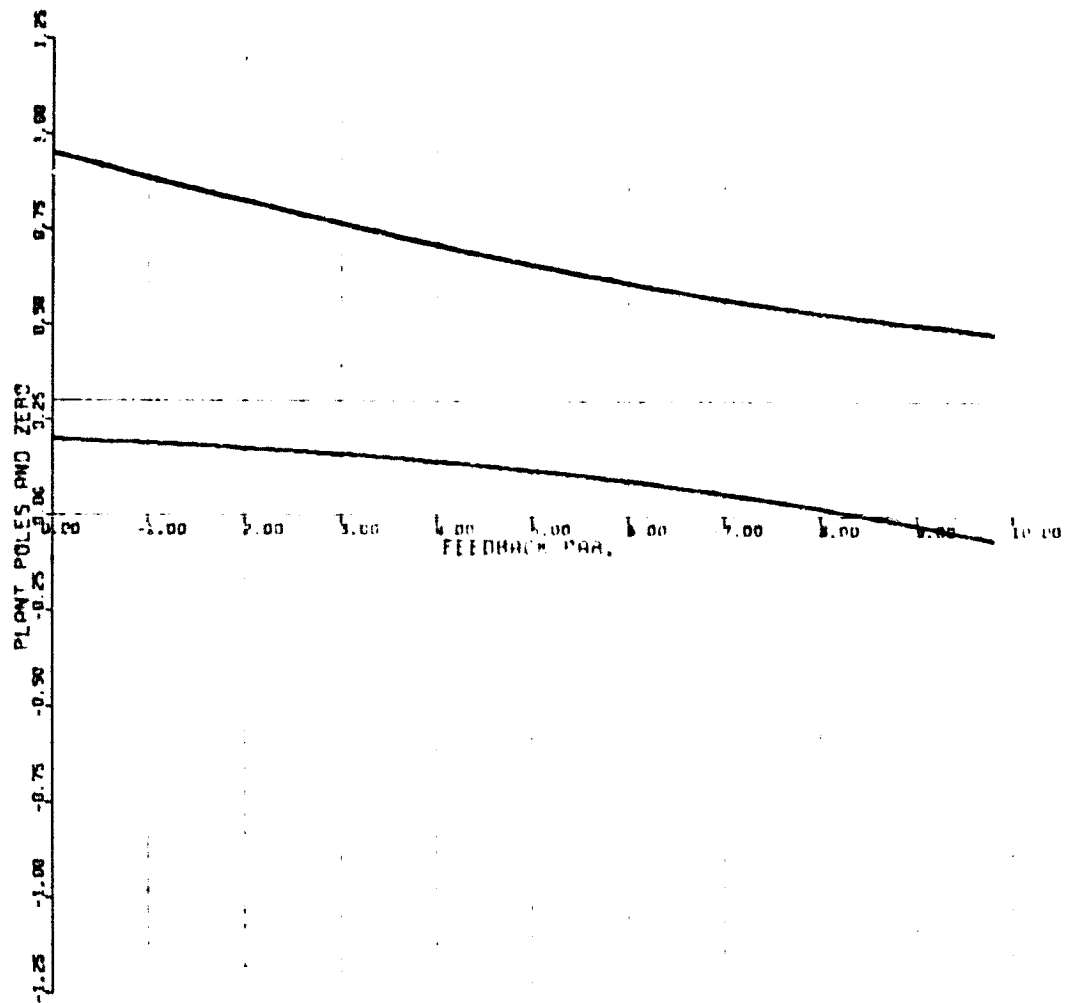
q_2 are the closed-loop system poles due to output and partial state feedback, respectively, and from (3-16) h is the open- and closed-loop system zero. The closed-loop migration due to various f in (3-7) and (3-13) appear in Fig. 3-1 and 3-2, respectively. Note that, since $a_1^5 = 0.77 > h > a_2$ and using (3-7) with $f = -2.4 \Rightarrow p_1^5 = (0.8)^5 = 0.33 > h > p_2 = 0.17$ and using (3-13) with $f = -2.3 \Rightarrow q_1^5 = (0.8)^5 = 0.33 > h > q_2 = 0.2$, this example falls in category (i). Note the near equivalence of the f 's from (3-7) or (3-13) for the same objective. (Refer to Appendix A for the program listings and tabular output of these and the following root locus plots.)

Example 2: $a_1 = 0.9, b = 0.1, a_2 = 0.1, \epsilon = -0.01, h = 0.011, c = 0.7,$
 $d = 0.65$

The closed-loop root migration due to various f in (3-7) and (3-13) appear in Fig. 3-3 and 3-4, respectively. Note that $a_1^5 = 0.59 > a_2 > h$ and $d^5 = 0.12 > a_2 > h$, i.e. the dominant open-loop pole and its desired location both dominate the other open-loop singularities, but for $f = -2.4$ in (3-7) the second closed-loop pole $p_2 = 0.13$ is crossing the dominance threshold with respect to $p_1 = 0.65$ as shown in Fig. 3-3, thereby putting this example in category (ii). To retain the open-loop dominance the partial state feedback of (3-14) in Fig. 3-4 must be used.

Example 3: $a_1 = 0.9, b = 0.08, a_2 = 0.8, \epsilon = 0.02, h = 0.82, c = 1,$
 $d = 0.5$

The closed-loop root migration due to various output feedback f in (3-7) is plotted in Fig. 3-5. Figure 3-6 shows the closed-loop roots for various partial state feedback gains in (3-13). Reversing the designation of the



— POLES

— ZERO

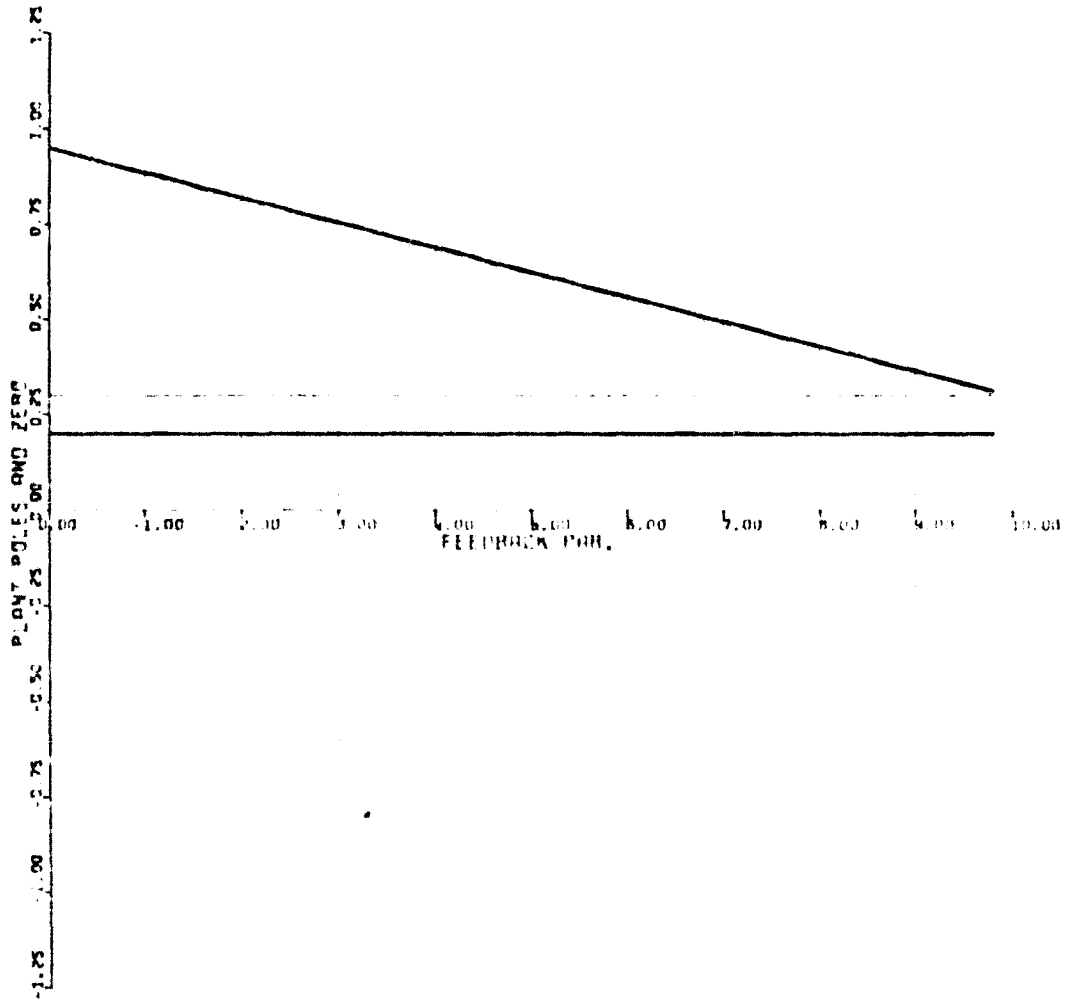
PLANT: $Y(K) = (A1+A2)Y(K-1) - (A1 \times A2)Y(K-2) + (B+E)U(K-1) - (A1 \times E + B \times A2)U(K-2)$

CONTROL: $U(K) = F1 \times R(K) + F2 \times Y(K)$

RESULTING CHAR. EQN.: $Z \times X^2 - Z \times (A1+A2+F \times (B+E)) + A1 \times A2 + F \times (B \times A2 + E \times A1)$

PARAMETERS: $A1=0.95$ $A2=0.2$ $B=0.065$ $E=0.01$

Fig. 3-1: Closed-loop Root Migration due to Output Feedback for Example 1.



— POLES

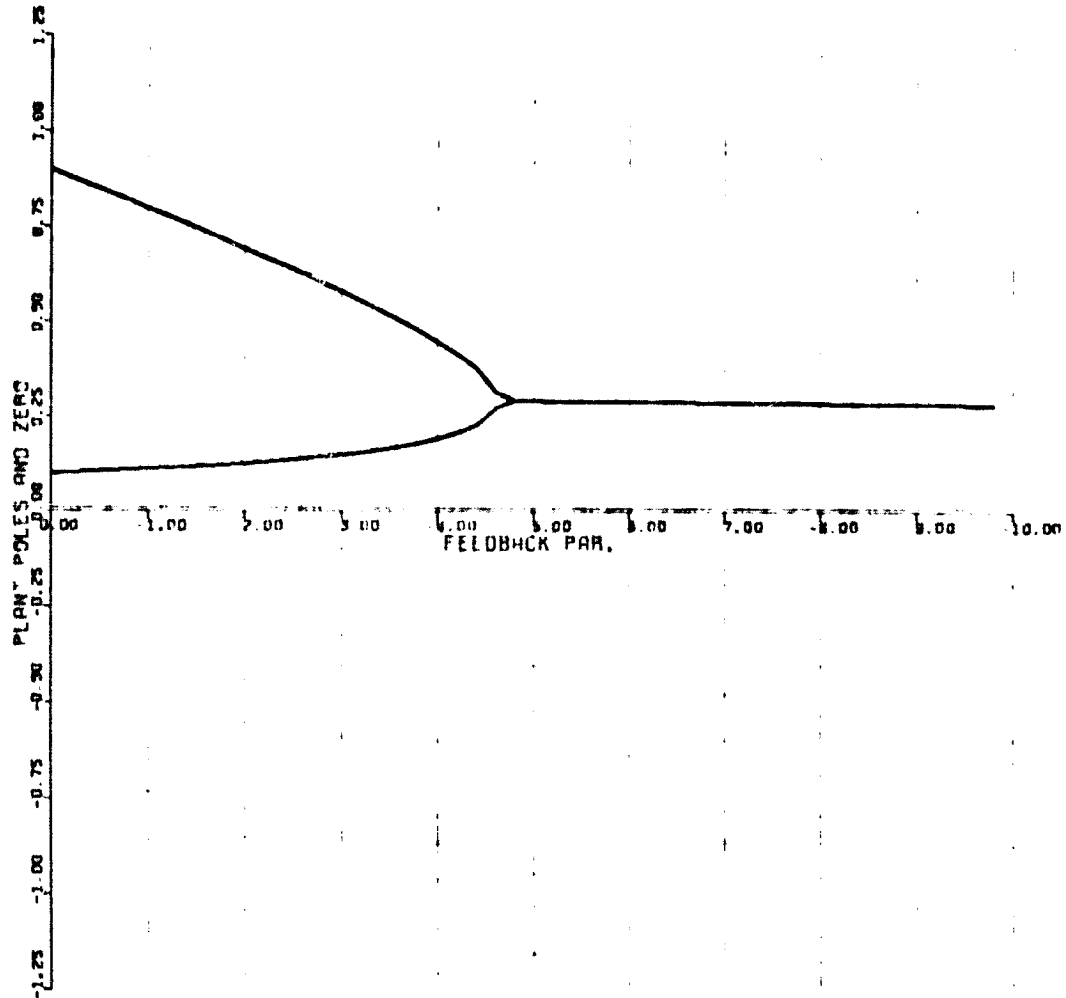
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PLANT: $Y(K) = (A_1 + B_2)Y(K-1) - (A_1 + A_2)Y(K-2) + (B_1 + E)U(K-1) - (A_1 + B_2)U(K-2)$

RESULTING CHAR. EQN.: $Z^2 - (A_1 + A_2)Z + (A_1 + B_2) = 0$

PARAMETERS: $A_1 = 0.95$ $A_2 = 0.2$ $B_1 = 0.05$ $E = 0.01$

Fig. 3-2: Closed-loop Root Migration due to Partial State Feedback for Example 1.



— POLES (Merged locus plots magnitude of complex conjugates)

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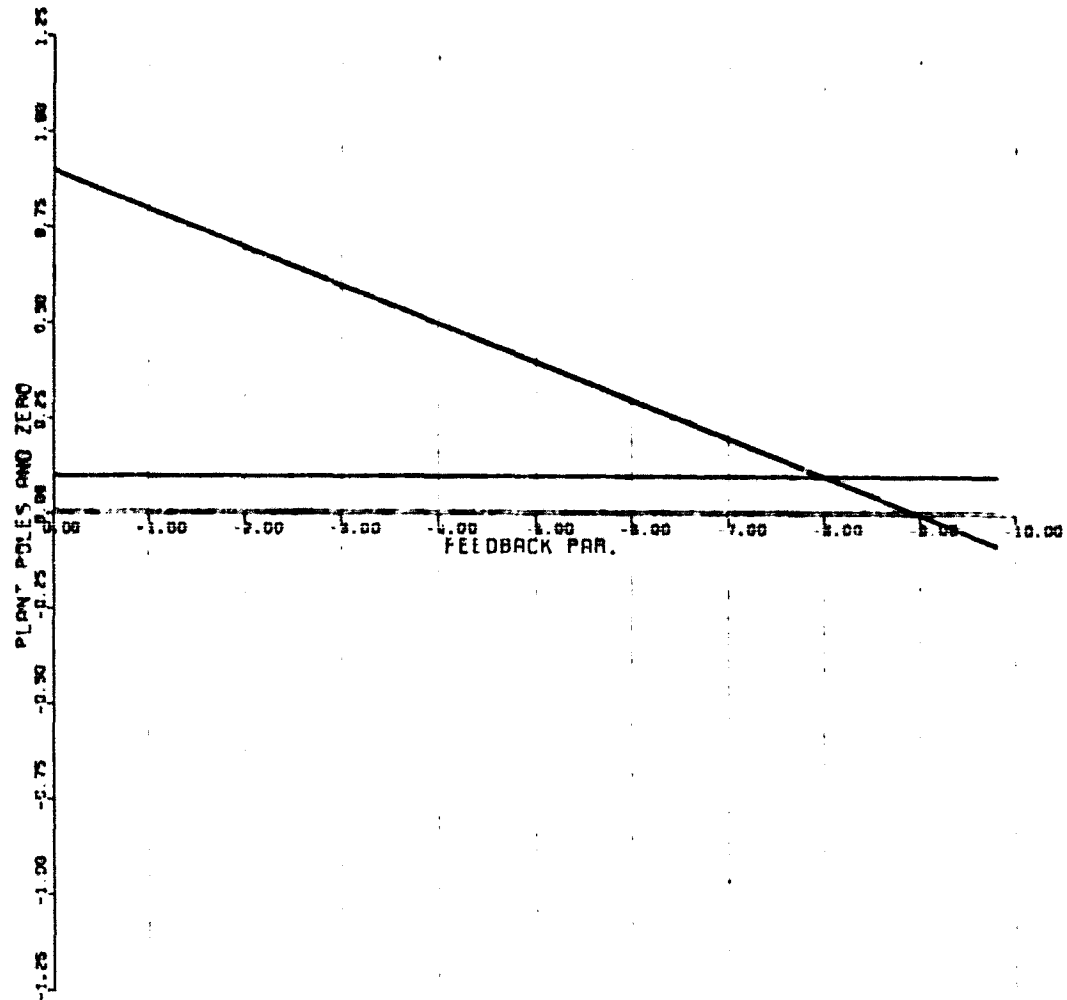
PLANT: $Y(K) = (A1 + A2)Y(K-1) - (A1 \times A2)Y(K-2) + (B+E)U(K-1) - (A1 \times E + B \times A2)U(K-2)$

CONTROL: $U(K) = F1 \times R(K) + F2 \times Y(K)$

RESULTING CHAR. EQN.: $Z^2 \times Z^2 - Z \times (A1 + A2 + F \times (B+E)) + A1 \times A2 + F \times (B \times A2 + E \times A1)$

PARAMETERS: $A1=0.9 \quad A2=0.1 \quad B=0.1 \quad E=-0.01$

Fig. 3-3: Closed-loop Root Migration due to Output Feedback for Example 2.



— POLES

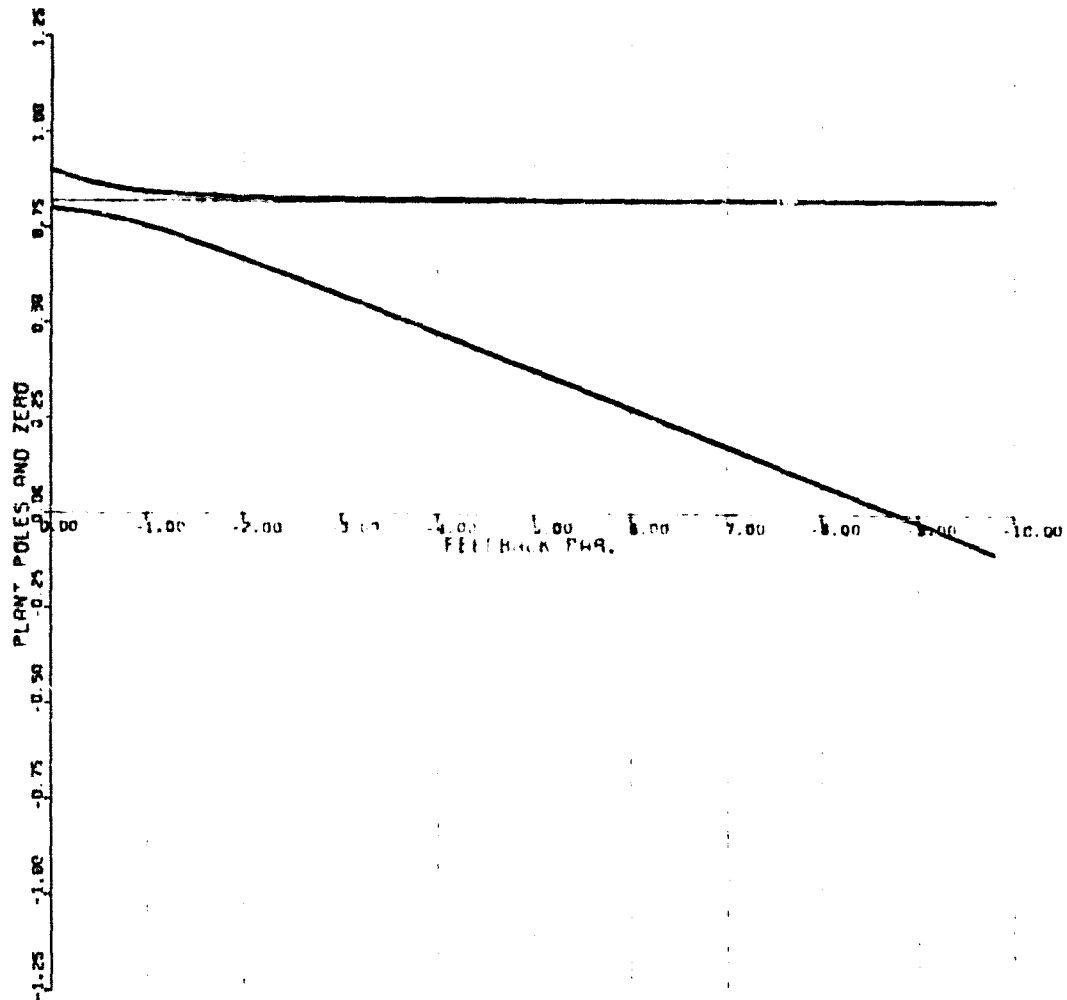
— ZERO

PLANT: $Y(K) = (A1+A2)Y(K-1) - (A1 \times A2)Y(K-2) + (B+F)U(K-1) - (A1 \times E + B \times A2)U(K-2)$

RESULTING CHAR. EQN.: $Z^2 - Z(A1+A2+F \times B) + A1 \times A2 + B \times A2 \times F$

PARAMETERS: $A1=0.9$ $A2=0.1$ $B=0.1$ $E=-0.01$

Fig. 3-4: Closed-loop Root Migration due to Partial State Feedback for Example 2.



— POLES

— ZERO

PLANT: $Y(K) = (A1+A2)Y(K-1) - (A1 \times A2)Y(K-2) + (B+E)U(K-1) - (A1 \times E + B \times A2)U(K-2)$

CONTROL: $U(K) = F1 \times R(K) + F2 \times Y(K)$

RESULTING CHAR. EQN.: $Z^2 \times (1 - (A1+A2)Z + A1 \times A2) + F \times (B+E) - (A1 \times E + B \times A2)Z = 0$

PARAMETERS: $A1=0.9$ $A2=0.8$ $B=0.08$ $E=0.02$

Fig. 3-5: Closed-loop Root Migration due to Output Feedback for Example 3.

important mode for partial state feedback by either swapping the elements of the state feedback gain vector in (3-14) or swapping the designation of a_1 , b and a_2 , ϵ results in the closed-loop roots shown in Fig. 3-7. Note in Fig. 3-5 for $f = -3.8$ the near-cancellation of the slower mode while the faster mode matches that desired. The degree of closeness of this near cancellation is not matched in either Fig. 3-6 or 3-7. Note that in Fig. 3-6 if f is chosen as -5 to cause $q_1 = 0.5$, use of the same f in Fig. 3-5 would result in overcompensation. Neither in the open-loop plant nor for any f placing one pole near the desired location of $d = 0.5$ are the singularities separable as in category (iii).

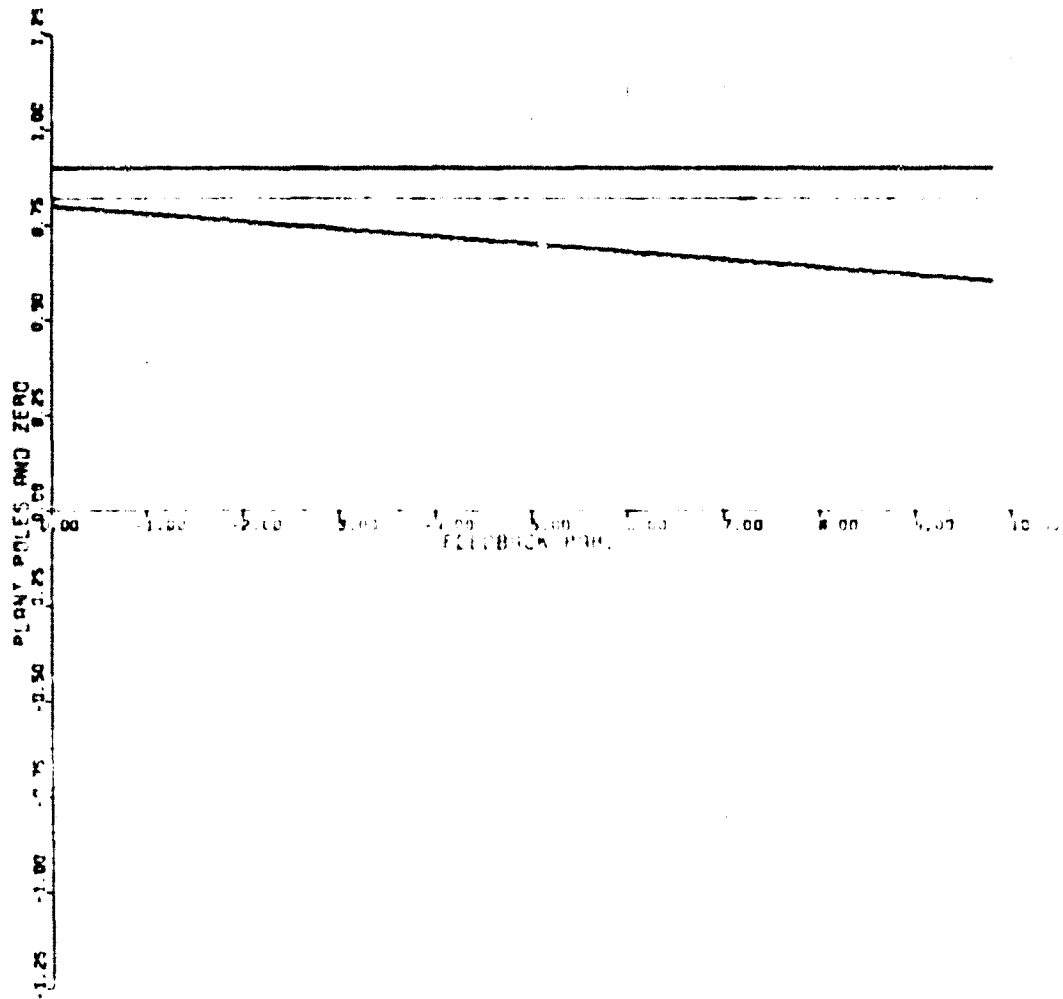
Example 4: $a_1 = 0.9$, $b = 0.5$, $a_2 = 0.7$, $\epsilon = -0.3$, $h = 0.4$, $c = 0.8$,
 $d = 0.6$

The pole migration due to the output feedback f from (3-7) displays in Fig. 3-8 the oscillatory character of the second-order poles in the region of desired pole radius. Even partial state feedback proves unsuccessful as shown in Fig. 3-9 due to the retention of one of the poles at its nearer unity value than the desired pole. Therefore, this example falls into category (iv).

Consider, for the last two example categories, digitally controlling a stable fourth-order discrete equivalent of a continuous plant composed of two quadratic, all-pole, modes

$$\frac{Y(z)}{U(z)} = \frac{B_{11}z + B_{12}}{z^2 - \alpha_{11}z - \alpha_{12}} + \frac{B_{21}z + B_{22}}{z^2 - \alpha_{21}z - \alpha_{22}}, \quad (3-17)$$

where the α and B are derived from the zero-order-hold equivalent



— POLES

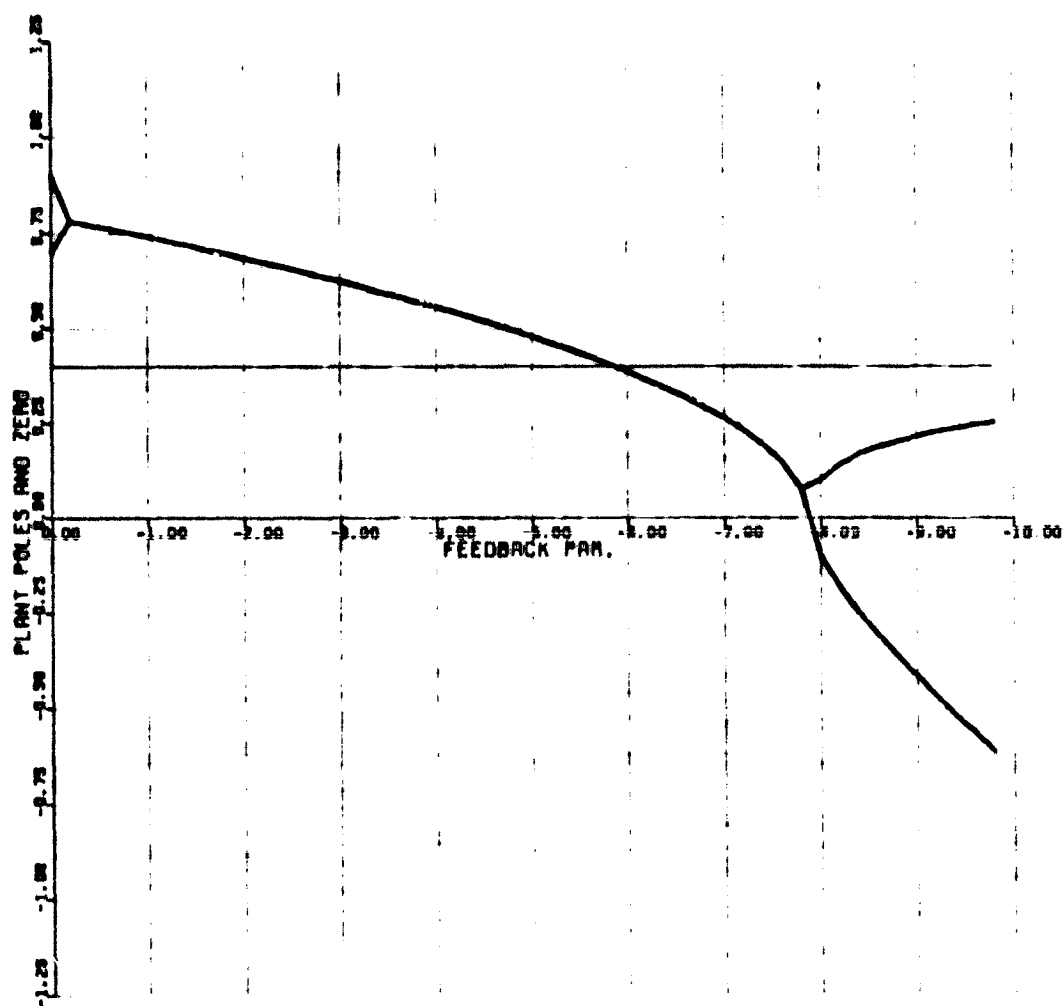
— ZERO

PLANT: $Y(s) = (A_1 + A_2)Y(s) + (B_1 + B_2)U(s) - (C_1 + C_2)Y(s) - (D_1 + D_2)U(s)$

RESULTING CHAR. EQN.: $Z(s) = (A_1 + A_2) + (B_1 + B_2)K - (C_1 + C_2) - (D_1 + D_2)K$

PARAMETERS: $A_1=0.9$ $A_2=0.8$ $B_1=0.08$ $B_2=0.02$

Fig. 3-7: Closed-loop Root Migration due to Partial State Feedback for Example 3.



— POLES (Merged locus plots magnitude of complex conjugates)

— ZERO

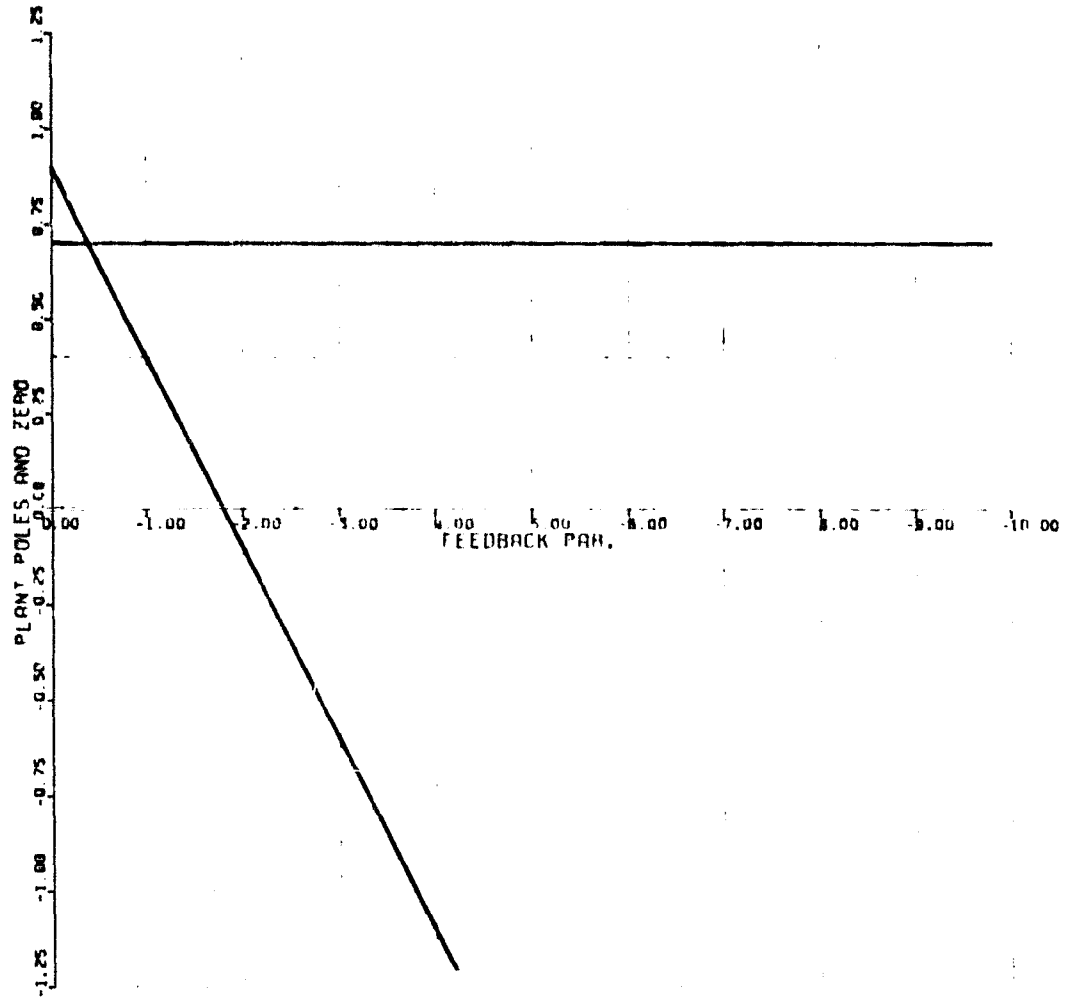
PLANT: $Y(K) = (A1+A2)Y(K-1) - (A1 \times A2)Y(K-2) + (B+E)U(K-1) - (A1 \times E + B \times A2)U(K-2)$

CONTROL: $U(K) = F_1 + F_2 Y(K)$

RESULTING CHAR. EQN.: $Z^2 - Z(A1+A2+F_1(B+E)) + A1 \times A2 + F_1(B \times A2 + E \times A1)$

PARAMETERS: $A1=0.9$ $A2=0.7$ $B=0.5$ $E=-0.3$

Fig. 3-8: Closed-loop Pole Migration due to Output Feedback for Example 4.



ORIGINAL PAGE IS
OF POOR QUALITY

— POLES

— ZERO

PLANT: $Y(K) = (A1+A2)Y(K-1) - (A1 \times A2)Y(K-2) + (B+E)U(K-1) - (A1 \times E + B \times A2) \times U(K-2)$

CONTROL: $U(K) = F1 \times R(K) + F2 \times Y(K)$

RESULTING CHAR. EQN.: $Z^2 - Z \times (A1+A2+F \times (B+E)) + A1 \times A2 + F \times (B \times A2 + E \times A1)$

PARAMETERS: $A1=0.9 \quad A2=0.7 \quad B=0.5 \quad E=-0.3$

Fig. 3-9: Closed-loop Pole Migration due to Partial State Feedback for Example 4.

$$\frac{\beta_{11}z + \beta_{12}}{z^2 - \alpha_{11}z - \alpha_{12}} = (1-z^{-1})z \left[\left(\frac{1}{s} \right) \left\{ \frac{\lambda_1}{s^2 + 2\zeta_1\omega_1 s + \omega_1^2} \right\} \right] \quad (3-18)$$

with [27]

$$\alpha_{11} = 2 e^{-\zeta_1\omega_1 T} \cos(\omega_1 T \sqrt{1-\zeta_1^2}) \quad (3-19)$$

$$\alpha_{12} = -e^{-2\zeta_1\omega_1 T} \quad (3-20)$$

$$\beta_{11} = \frac{\lambda_1}{\omega_1^2} \left\{ 1 - e^{-\zeta_1\omega_1 T} \left[\cos(\omega_1 T \sqrt{1-\zeta_1^2}) + (\zeta_1/\sqrt{1-\zeta_1^2}) \sin(\omega_1 T \sqrt{1-\zeta_1^2}) \right] \right\} \quad (3-21)$$

$$\beta_{12} = \frac{\lambda_1}{\omega_1^2} \left\{ e^{-\zeta_1\omega_1 T} \left[e^{-\zeta_1\omega_1 T} - \cos(\omega_1 T \sqrt{1-\zeta_1^2}) + (\zeta_1/\sqrt{1-\zeta_1^2}) \sin(\omega_1 T \sqrt{1-\zeta_1^2}) \right] \right\} \quad (3-22)$$

The reduced-order, model-following objective is to track the output of

$$\frac{S(z)}{R(z)} = \frac{\delta_1 z + \delta_2}{z^2 - \gamma_1 z - \gamma_2} \quad (3-23)$$

where γ_1 and γ_2 can represent the discrete equivalent of desired s-plane pole locations in terms of ζ and ω and translated via conversions similar to (3-19) and (3-20). For $\beta_{21} = \beta_{22} = 0$ this objective can only be achieved if the numerator of (3-17) is cancelled and replaced by that of (3-23). Therefore assume that the numerator zero of (3-23) matches that of (3-17) with $\beta_{21} = \beta_{22} = 0$ and only the poles require shifting. This converts the objective model in (3-23) to

$$\frac{S(z)}{R(z)} = \frac{\delta(\beta_{11}z + \beta_{12})}{z^2 - \gamma_1 z - \gamma_2} \quad (3-24)$$

The control effort

$$u(k) = \delta r(k) + \eta_1 u(k-1) + \eta_2 u(k-2) + v_1 y(k-1) + v_2 y(k-2), \quad (3-25)$$

where [28]

$$\begin{bmatrix} \eta_1 \\ \eta_2 \\ v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 0 & 0 \\ \alpha_{11} & -1 & -\beta_{11} & 0 \\ \alpha_{12} & \alpha_{11} & -\beta_{12} & -\beta_{11} \\ 0 & \alpha_{12} & 0 & -\beta_{12} \end{bmatrix}^{-1} \begin{bmatrix} \alpha_{11}^{-\gamma_1} \\ \alpha_{12}^{-\gamma_2} \\ 0 \\ 0 \end{bmatrix} \quad (3-26)$$

converts (3-17), with $\beta_{21} = \beta_{22} = 0$ to (3-24). An algebraic, sequentially calculated solution of (3-26) avoiding the matrix inversion is [3][9]

$$\eta_1 = \gamma_1 - \alpha_{11} \quad (3-27)$$

$$v_1 = \frac{[\eta_1 \alpha_{12} + (\eta_1 \alpha_{11} - \alpha_{12} + \gamma_2)(\alpha_{11} - \alpha_{12} \beta_{11} / \beta_{12})]}{[\beta_{11} \alpha_{11} + \beta_{12} - \alpha_{12} \beta_{11}^2 / \beta_{12}]} \quad (3-28)$$

$$\eta_2 = \alpha_{11} \eta_1 + \gamma_2 - \beta_{11} v_1 - \alpha_{12} \quad (3-29)$$

$$v_2 = \eta_2 \alpha_{12} / \beta_{12} \quad (3-30)$$

If β_{21} or $\beta_{22} \neq 0$ and (3-25) is used on (3-17) then

$$\begin{aligned} \frac{Y(z)}{R(z)} &= \delta \left\{ \frac{[1/(1 - \eta_1 z^{-1} - \eta_2 z^{-2})] [Y(z)/U(z)]}{1 - [(v_1 z^{-1} + v_2 z^{-2})/(1 - \eta_1 z^{-1} - \eta_2 z^{-2})] [Y(z)/U(z)]} \right\} \\ &= \frac{\delta z^2 [(\beta_{11} z + \beta_{12})(z^2 - \alpha_{21} z - \alpha_{22}) + \epsilon(\beta_{21} z + \beta_{22})(z^2 - \alpha_{11} z - \alpha_{12})]}{\{(z^2 - \eta_1 z - \eta_2)(z^2 - \alpha_{11} z - \alpha_{12})(z^2 - \alpha_{21} z - \alpha_{22})\}} \\ &\quad - (v_1 z + v_2) [(\beta_{11} z + \beta_{12})(z^2 - \alpha_{21} z - \alpha_{22}) + \epsilon(\beta_{21} z + \beta_{22})(z^2 - \alpha_{11} z - \alpha_{12})] \\ &= \frac{b_5 z^5 + b_4 z^4 + b_3 z^3 + b_2 z^2 + b_1 z + b_0}{a_6 z^6 + a_5 z^5 + a_4 z^4 + a_3 z^3 + a_2 z^2 + a_1 z + a_0} \quad (3-31) \end{aligned}$$

where

$$b_5 = \delta(\beta_{11} + \beta_{21}) \quad (3-32)$$

$$b_4 = \delta(-\beta_{11}\alpha_{21} + \beta_{12} - \beta_{21}\alpha_{11} + \beta_{22}) \quad (3-33)$$

$$b_3 = \delta(-\beta_{11}\alpha_{22} - \beta_{12}\alpha_{21} - \beta_{21}\alpha_{12} - \beta_{22}\alpha_{11}) \quad (3-34)$$

$$b_2 = \delta(-\beta_{12}\alpha_{22} - \beta_{22}\alpha_{12}) \quad (3-35)$$

$$b_1 = b_0 = 0 \quad (3-36)$$

$$a_6 = 1 \quad (3-37)$$

$$a_5 = -\alpha_{21} - \alpha_{11} - \eta_1 \quad (3-38)$$

$$a_4 = -\alpha_{22} + \alpha_{11}\alpha_{21} - \alpha_{12} + \eta_1\alpha_{21} + \eta_1\alpha_{11} - \eta_2 - \nu_1\beta_{11} - \nu_1\beta_{21} \quad (3-39)$$

$$a_3 = \alpha_{11}\alpha_{22} + \alpha_{12}\alpha_{21} + \eta_1\alpha_{22} - \eta_1\alpha_{11}\alpha_{21} + \eta_1\alpha_{12} + \eta_2\alpha_{21} + \eta_2\alpha_{11} \\ + \nu_1\beta_{11}\alpha_{21} - \nu_1\beta_{12} + \nu_1\beta_{21}\alpha_{11} - \nu_1\beta_{22} - \nu_2\beta_{11} - \nu_2\beta_{21} \quad (3-40)$$

$$a_2 = \alpha_{12}\alpha_{22} - \eta_1\alpha_{11}\alpha_{22} - \eta_1\alpha_{12}\alpha_{21} + \eta_2\alpha_{22} - \eta_2\alpha_{11}\alpha_{21} + \eta_2\alpha_{12} \\ + \nu_1\beta_{11}\alpha_{22} + \nu_1\beta_{12}\alpha_{21} + \nu_1\beta_{21}\alpha_{12} + \nu_1\beta_{22}\alpha_{11} \\ + \nu_2\beta_{11}\alpha_{21} - \nu_2\beta_{12} + \nu_2\beta_{21}\alpha_{11} - \nu_2\beta_{22} \quad (3-41)$$

$$a_1 = -\eta_1\alpha_{12}\alpha_{22} - \eta_2\alpha_{11}\alpha_{22} - \eta_2\alpha_{12}\alpha_{21} + \nu_1\beta_{12}\alpha_{22} + \nu_1\beta_{22}\alpha_{12} \\ + \nu_2\beta_{11}\alpha_{22} + \nu_2\beta_{12}\alpha_{21} + \nu_2\beta_{21}\alpha_{12} + \nu_2\beta_{22}\alpha_{11} \quad (3-42)$$

$$a_0 = -\eta_2\alpha_{12}\alpha_{22} + \nu_2\beta_{12}\alpha_{22} + \nu_2\beta_{22}\alpha_{12} \quad (3-43)$$

Note that $\frac{Y(z)}{R(z)}$ has the same numerator as $\frac{Y(z)}{U(z)}$. (See Appendix B for the supporting algebra and simulated check for (3-31)-(3-43) for the following example.)

Example 5: $T = 0.5$, $\lambda_1 = 1$, $\lambda_2 = 1$, $\zeta_1 = 0.2$, $\omega_1 = 0.5$, $\zeta_2 = 0.02$,
 $\omega_2 = 5$, $\delta = 0.5$, $\gamma_1 = 1.687$, $\gamma_2 = -0.741$.

The plant in (3-17) resulting from this parameterization has the pole-zero pattern shown in Fig. 3-10. Note $\zeta_1 \omega_1 T = 0.05$, i.e. both modes have the same settling time and are therefore not separable on a time basis. However, they are clearly separable on a frequency basis. The ζ_1 , ω_1 , and T were chosen to avoid aliasing, retain frequency separability, and model the lightly damped situation of flexible spacecraft [3][10].

Further, note that the low frequency mode has a DC gain which is 100 times that of the high frequency mode. Attempting to increase the damping ratio by a factor of three to 0.6 for the low frequency mode (and therefore third the settling time with ω_1 unchanged) leads to the stated objective, which from (3-26) or (3-27)-(3-30) yields the following controller parameterization for (3-25): $\delta = 0.5$, $\eta_1 = -0.159$, $\eta_2 = -0.0715$, $\nu_1 = -0.475$, and $\nu_2 = 0.556$.

Factorization of (3-31), where $b_5 = 0.952$, $b_4 = 1.188$, $b_3 = 1.128$, $b_2 = 0.83$, $b_1 = b_0 = 0$, $a_6 = 1$, $a_5 = -0.164$, $a_4 = -0.891$, $a_3 = -0.466$, $a_2 = 0.676$, $a_1 = 0.0625$, and $a_0 = -0.0338$ from (3-32)-(3-43), yields the pole-zero constellation in Fig. 3-11. (See Appendix C for the factorization and plotting routines used.) Note the closed-loop retention of frequency separability. Simply changing λ_2 to 10, thereby reducing the DC gain ratio from 100:1 to 10:1, leaves the plant poles of $0.923 \pm j0.231$

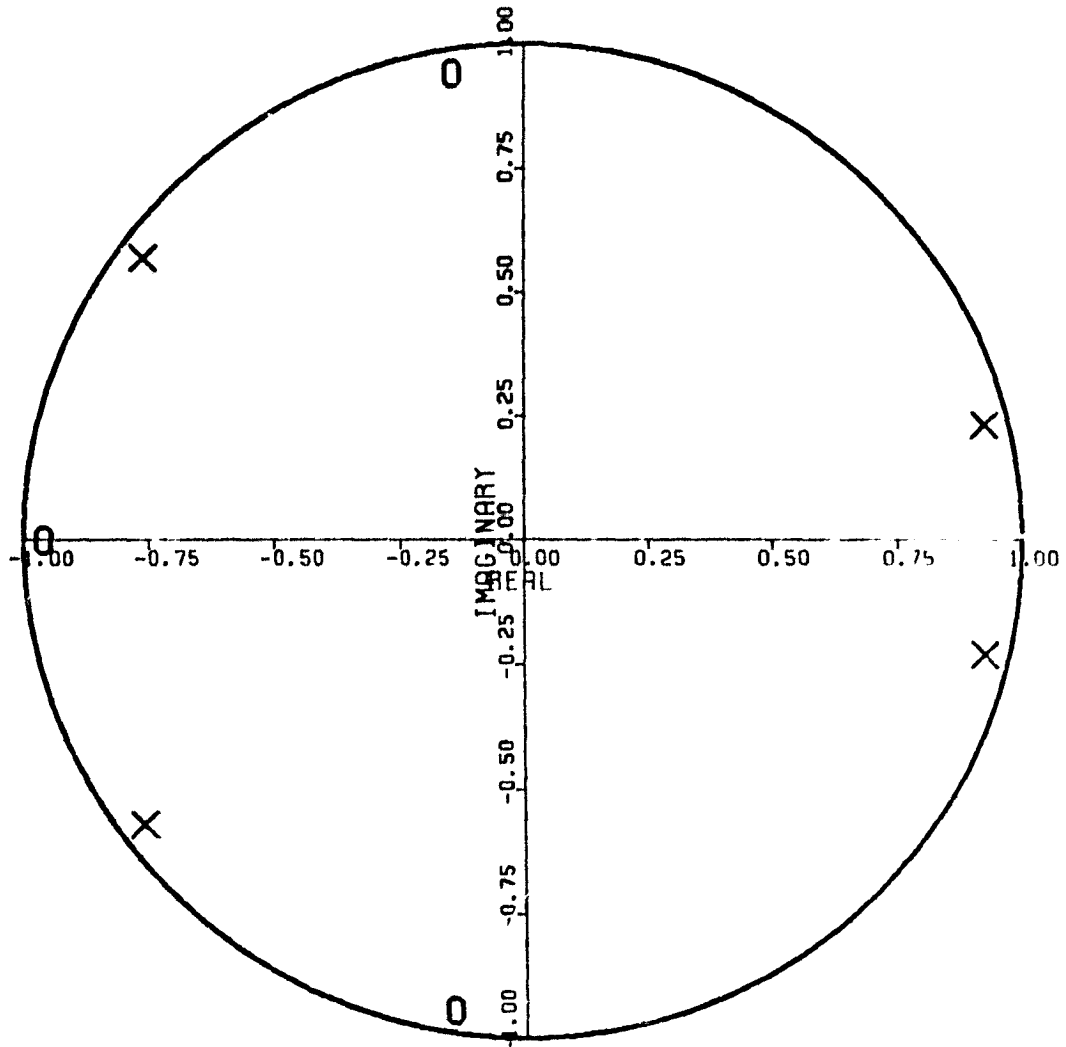


Figure 3-10: Plant Pole-zero Plot of Example 5

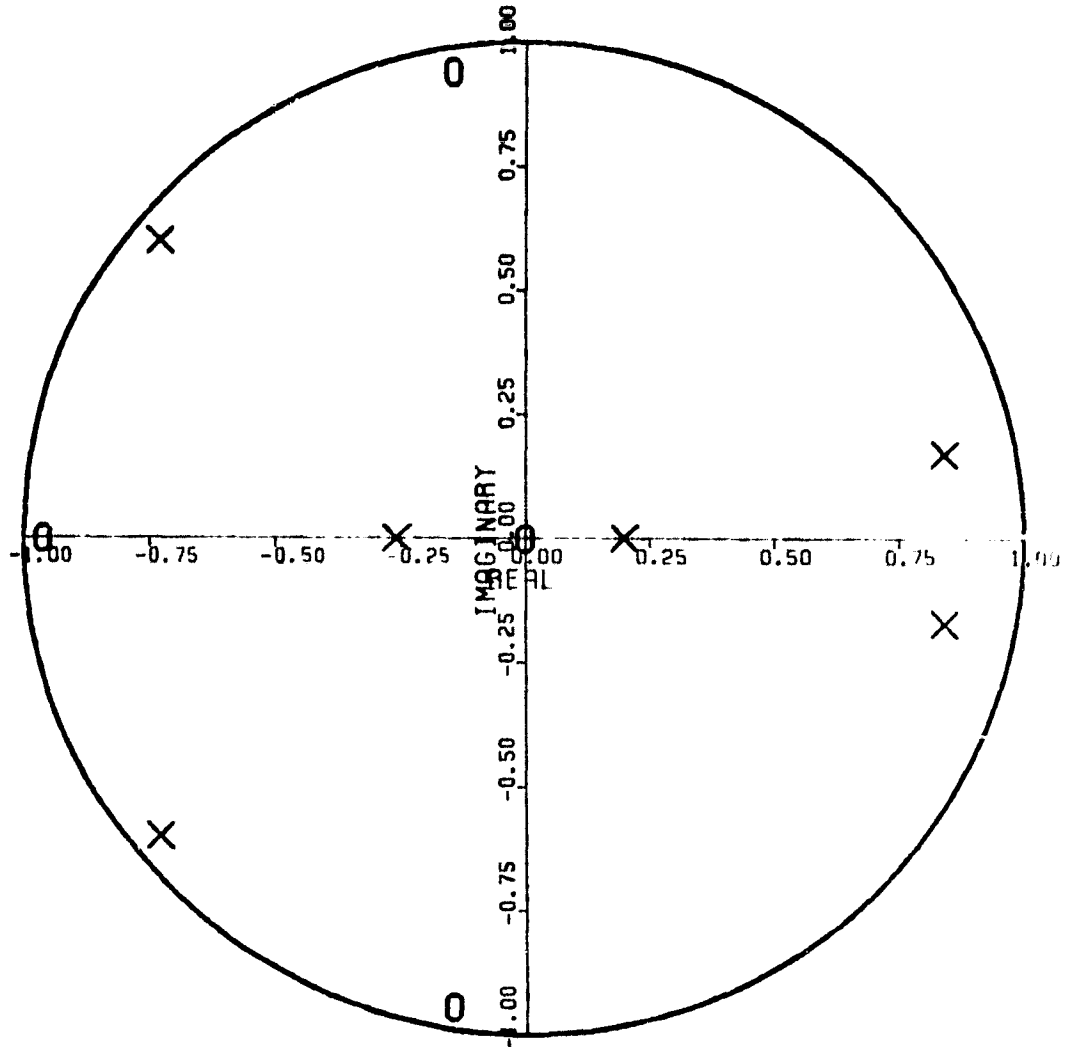


Figure 3-11: Reduced-Order Controlled Plant Pole-zero Plot for Example 5

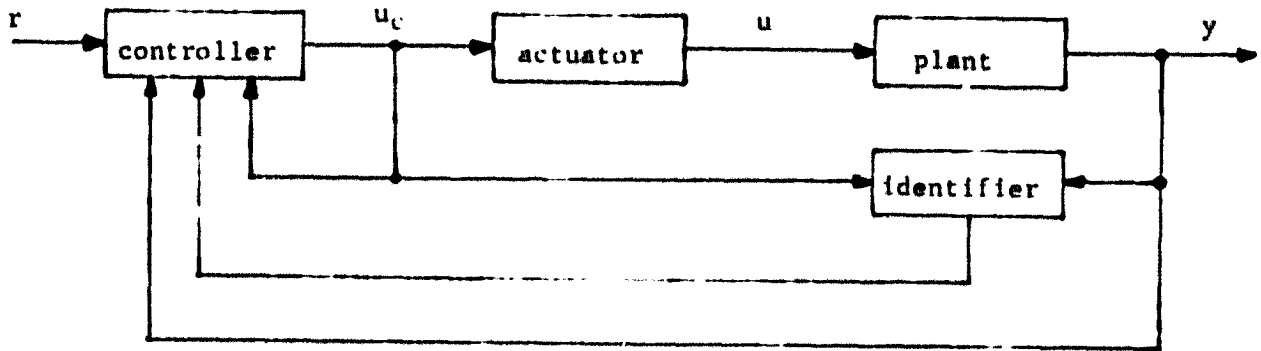
and $-0.762 \pm j0.570$ unaltered but changes the zeros from -0.960 and $-0.144 \pm j0.942$ to -0.959 and $0.675 \pm j0.671$. Since λ_2 does not affect the calculation of α_{1j} or β_{1j} in (3-19)-(3-22), the η_i and v_i selected in (3-27)-(3-30) will be the same. However, this control law results in an unstable system. (See Appendix C.) Therefore observation spillover can be seen to destabilize an otherwise stable system if neglected in reduced-order controller design.

The addition of a low-pass actuator in controlling (3-17) would occur in the feedback loop after the dynamic output feedback element of (3-23) (with $\delta = 1$), if the reference signal were considered an unmeasurable disturbance. However the scaling of r in (3-25) suggests a command signal designation of r . Therefore the actuator would precede the plant in the forward path. A remaining question centers on the availability of the actuator output for control and identification. The three meaningful possibilities are diagrammed in Fig. 3-12a-c, where the identifier feeds back the parameterization of the plant for adaptive controller parameterization. Retention of the controller dynamics from (3-25) suggests usage of a or c. The frequency limited identification concept of "fact" (ii) in the preceding section implicit in example category (vi) prompts usage of b or c. Therefore c, if physically feasible, appears to be the prime candidate. However a is the most reasonable if the actuator models a physical limitation rather than a control logic device.

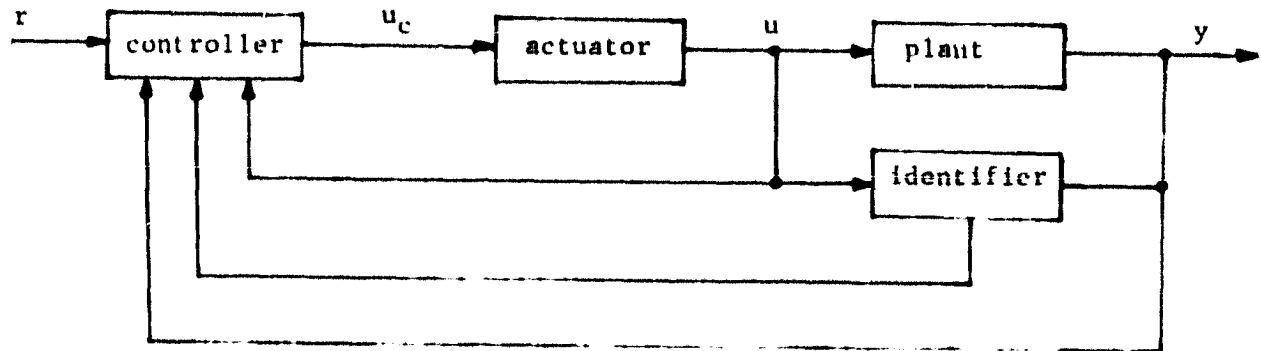
For the simple low-pass actuator model in [15] the zero-order-hold equivalent is [29]

$$H(z) = (1-z^{-1}) Z\left\{\left[\frac{1}{s}\right]\left[\frac{1}{s+\sigma}\right]\right\} = \frac{1 - e^{-\sigma T}}{z - e^{-\sigma T}} \quad (3-44)$$

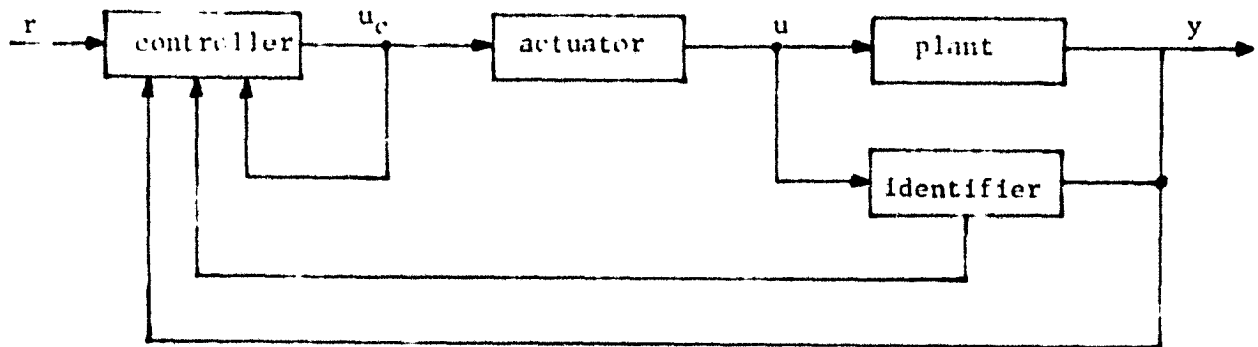
Therefore



(a) Actuator Incorporated in Plant



(b) Actuator Incorporated in Controller



(c) Actuator as Separable Element

Figure 3-12: Actuator Inclusion Possibilities

$$u(k+1) = e^{-\sigma T} u(k) + (1 - e^{-\sigma T}) u_c(k), \quad (3-45)$$

where from (3-25) for Fig. 3-12a and c

$$u_c(k) = \delta r(k) + \eta_1 u_c(k-1) + \eta_2 u_c(k-2) + v_1 y(k-1) + v_2 y(k-2) \quad (3-46)$$

and for Fig. 3-12b

$$u_c(k) = \delta r(k) + \eta_1 u(k-1) + \eta_2 u(k-2) + v_1 y(k-1) + v_2 y(k-2), \quad (3-47)$$

Applying (3-45)-(3-46) to (3-17) with the controller parameterization arising from e.g. (3-27)-(3-30), i.e. the reduced-order controller ignoring the actuator inclusion, yields the following overall transfer function (as shown in Appendix D)

$$\frac{Y(z)}{R(z)} = \frac{q_5 z^5 + q_4 z^4 + q_3 z^3 + q_2 z^2 + q_1 z + q_0}{z^7 + p_6 z^6 + p_5 z^5 + p_4 z^4 + p_3 z^3 + p_2 z^2 + p_1 z + p_0} \quad (3-48)$$

where

$$q_5 = \delta(1 - e^{-\sigma T})(\beta_{11} + \beta_{21}) \quad (3-49)$$

$$q_4 = \delta(1 - e^{-\sigma T})(-\beta_{11}\alpha_{21} + \beta_{12} - \beta_{21}\alpha_{11} + \beta_{22}) \quad (3-50)$$

$$q_3 = \delta(1 - e^{-\sigma T})(-\beta_{11}\alpha_{22} - \beta_{12}\alpha_{21} - \beta_{21}\alpha_{12} - \beta_{22}\alpha_{11}) \quad (3-51)$$

$$q_2 = \delta(1 - e^{-\sigma T})(-\beta_{12}\alpha_{22} - \beta_{22}\alpha_{12}) \quad (3-52)$$

$$q_1 = q_0 = 0 \quad (3-53)$$

$$p_6 = -e^{-\sigma T} - \alpha_{21} - \alpha_{11} - \eta_1 \quad (3-54)$$

$$p_5 = e^{-\sigma T} (\alpha_{21} + \alpha_{11} + \eta_1) - \alpha_{22} + \alpha_{11}\alpha_{21} - \alpha_{12} + \eta_1\alpha_{21} \\ + \eta_1\alpha_{11} - \eta_2 \quad (3-55)$$

$$p_4 = -e^{-\sigma T} (-\alpha_{22} + \alpha_{11}\alpha_{21} - \alpha_{12} + \eta_1\alpha_{21} + \eta_1\alpha_{11} - \eta_2) + \alpha_{11}\alpha_{22} \\ + \alpha_{12}\alpha_{21} + \eta_1\alpha_{22} - \eta_1\alpha_{11}\alpha_{21} + \eta_1\alpha_{12} + \eta_2\alpha_{21} + \eta_2\alpha_{11} \\ - v_1(\beta_{11} + \beta_{21})(1 - e^{-\sigma T}) \quad (3-56)$$

$$p_3 = -e^{-\sigma T} (\alpha_{11}\alpha_{22} + \alpha_{12}\alpha_{21} + \eta_1\alpha_{22} - \eta_1\alpha_{11}\alpha_{21} + \eta_1\alpha_{12} + \eta_2\alpha_{21} + \eta_2\alpha_{11}) \\ + \alpha_{12}\alpha_{22} - \eta_1\alpha_{11}\alpha_{22} - \eta_1\alpha_{12}\alpha_{21} + \eta_2\alpha_{22} - \eta_2\alpha_{11}\alpha_{21} + \eta_2\alpha_{12} \\ - v_1(-\beta_{11}\alpha_{21} + \beta_{22} + \beta_{12} - \beta_{21}\alpha_{11})(1 - e^{-\sigma T}) \\ - v_2(\beta_{11} + \beta_{21})(1 - e^{-\sigma T}) \quad (3-57)$$

$$p_2 = -e^{-\sigma T} (\alpha_{12}\alpha_{22} - \eta_1\alpha_{11}\alpha_{22} - \eta_1\alpha_{12}\alpha_{21} + \eta_2\alpha_{22} - \eta_2\alpha_{11}\alpha_{21} + \eta_2\alpha_{12}) \\ - \eta_1\alpha_{12}\alpha_{22} - \eta_2\alpha_{11}\alpha_{22} - \eta_2\alpha_{12}\alpha_{21} + v_1(\beta_{11}\alpha_{22} + \beta_{12}\alpha_{21} \\ + \beta_{21}\alpha_{12} + \beta_{22}\alpha_{11})(1 - e^{-\sigma T}) - v_2(-\beta_{11}\alpha_{21} + \beta_{22} + \beta_{12} \\ - \beta_{21}\alpha_{11})(1 - e^{-\sigma T}) \quad (3-58)$$

$$p_1 = e^{-\sigma T} (\eta_1\alpha_{12}\alpha_{22} + \eta_2\alpha_{11}\alpha_{22} + \eta_2\alpha_{12}\alpha_{21}) - \eta_2\alpha_{12}\alpha_{22} \\ + v_2(\beta_{11}\alpha_{22} + \beta_{12}\alpha_{21} + \beta_{21}\alpha_{12} + \beta_{22}\alpha_{11})(1 - e^{-\sigma T}) \\ + v_1(\beta_{12}\alpha_{22} + \beta_{22}\alpha_{12})(1 - e^{-\sigma T}) \quad (3-59)$$

$$p_0 = e^{-\sigma T} \eta_2\alpha_{12}\alpha_{22} + v_2(1 - e^{-\sigma T})(\beta_{12}\alpha_{22} + \beta_{22}\alpha_{12}) \quad (3-60)$$

Similarly the overall transfer function resulting from application of (3-45) and (3-47) to (3-17) is

$$\frac{Y(z)}{R(z)} = \frac{q_5 z^5 + q_4 z^4 + q_3 z^3 + q_2 z^2 + q_1 z + q_0}{z^7 + p_6 z^6 + p_5 z^5 + p_4 z^4 + p_3 z^3 + p_2 z^2 + p_1 z + p_0} \quad (3-61)$$

where the q_i are as in (3-49)-(3-53) and

$$p_6 = -\alpha_{21} - \alpha_{11} - e^{-\sigma T} \quad (3-62)$$

$$p_5 = -\alpha_{22} + \alpha_{11}\alpha_{21} - \alpha_{12} + e^{-\sigma T}(\alpha_{21} + \alpha_{11}) - \eta_1(1 - e^{-\sigma T}) \quad (3-63)$$

$$p_4 = \alpha_{11}\alpha_{22} + \alpha_{12}\alpha_{21} - e^{-\sigma T}(-\alpha_{22} + \alpha_{11}\alpha_{21} - \alpha_{12}) + \eta_1(1 - e^{-\sigma T})(\alpha_{21} + \alpha_{11}) - \eta_2(1 - e^{-\sigma T}) - \nu_1(1 - e^{-\sigma T})(\beta_{11} + \beta_{21}) \quad (3-64)$$

$$p_3 = \alpha_{12}\alpha_{22} - e^{-\sigma T}(\alpha_{11}\alpha_{22} + \alpha_{12}\alpha_{21}) - \eta_1(1 - e^{-\sigma T})(-\alpha_{22} + \alpha_{11}\alpha_{21} - \alpha_{12}) - \eta_2(1 - e^{-\sigma T})(-\alpha_{21} - \alpha_{11}) - \nu_1(1 - e^{-\sigma T})(-\beta_{11}\alpha_{21} + \beta_{22} + \beta_{12} - \beta_{21}\alpha_{11}) - \nu_2(1 - e^{-\sigma T})(\beta_{11} + \beta_{21}) \quad (3-65)$$

$$p_2 = -e^{-\sigma T}\alpha_{12}\alpha_{22} - \eta_1(1 - e^{-\sigma T})(\alpha_{11}\alpha_{22} + \alpha_{12}\alpha_{21}) - \eta_2(1 - e^{-\sigma T})(-\alpha_{22} + \alpha_{11}\alpha_{21} - \alpha_{12}) + \nu_1(1 - e^{-\sigma T})(\beta_{11}\alpha_{22} + \beta_{12}\alpha_{21} + \beta_{21}\alpha_{12} + \beta_{22}\alpha_{11}) - \nu_2(1 - e^{-\sigma T})(-\beta_{11}\alpha_{21} + \beta_{22} + \beta_{12} - \beta_{21}\alpha_{11}) \quad (3-66)$$

$$p_1 = -\eta_1(1 - e^{-\sigma T})\alpha_{12}\alpha_{22} - \eta_2(1 - e^{-\sigma T})(\alpha_{11}\alpha_{22} + \alpha_{12}\alpha_{21}) + \nu_2(1 - e^{-\sigma T})(\beta_{11}\alpha_{22} + \beta_{12}\alpha_{21} + \beta_{21}\alpha_{12} + \beta_{22}\alpha_{11}) + \nu_1(1 - e^{-\sigma T})(\beta_{12}\alpha_{22} + \beta_{22}\alpha_{12}) \quad (3-67)$$

$$p_0 = -\eta_2(1 - e^{-\sigma T})\alpha_{12}\alpha_{22} + \nu_2(1 - e^{-\sigma T})(\beta_{12}\alpha_{22} + \beta_{22}\alpha_{12}) \quad (3-68)$$

Example 6: $T = 0.5$, $\lambda_1 = i$, $\lambda_2 = 1$, $\zeta_1 = 0.2$, $\omega_1 = 0.5$, $\zeta_2 = 0.02$,
 $\omega_2 = 5$, $\sigma = 1$, $\delta = 0.5$, $\gamma_1 = 1.687$, $\gamma_2 = -0.741$

Note that

$$H(s) \Big|_{s=0.5j} = \frac{0}{0.5j + \sigma} = \frac{1}{1.12 \angle 26.57^\circ} = 0.89 \angle -26.57^\circ \quad (3-69)$$

and

$$H(s) \Big|_{s=5j} = \frac{1}{5.1 \angle 78.69^\circ} = 0.20 \angle -78.69^\circ \quad (3-70)$$

Therefore, since [30, p. 387] $\omega_r = \omega_n \sqrt{1 - 2\zeta^2} \approx \omega_n$ for $\zeta \leq 0.2$, the actuator provides $20 \log \left(\frac{0.20}{0.89} \right) = -13$ db magnitude attenuation of the frequency response at the resonance of the high frequency mode versus the resonance of the low frequency mode in Example 5. Therefore Example 6 falls into category (vi). Note that $z = e^{-\sigma T} = 0.61$, $e^{-5\zeta_1\omega_1 T} = e^{-5 \cdot 0.2 \cdot 0.5 \cdot 0.5} = 0.95$, $e^{-5\zeta_2\omega_2 T} = (0.95)^5 = 0.78$. However, the pole-shifting desired is to a radius of 0.861. Therefore the closed-loop transient effect of the actuator on the low frequency mode is expected to be appreciable but possibly tolerable. Despite the unity DC gain of $H(z)$, its effect on the pole-shifting will also alter the step response tracking accuracy. These effects are quantified in Table 3-1. Note that, for each actuator configuration, as λ_2 increases the controlled system exhibits instability as one or more of the closed-loop poles moves outside the unit circle; but the range of λ_2 retaining stability is increased by each actuator configuration relative to (3-61). (The factorization scheme used is the same as that in Appendix C. See Appendix E for details.)

Table 3-2 summarizes the six examples that will be used to meet the objectives of the preceding section. The next section presents the adaptive control algorithms to be applied.

actuator config.	λ_2	poles #1	#2,3	#4,5	#6,7	zeros: #1	#2,3	#4,5	DC gain
Fig. 3-12a	1	0.850 0°	0.890 ±18.4°	0.344 ±109.8°	0.950 ±143.9°	0.960 180°	0.953 ±98.7°	0 ±0°	2.2
	10	0.865 0°	0.896 ±18.2°	0.544 ±97.3°	0.964 ±149.3°	0.959 180°	0.952 ±44.8°	0 ±0°	2.5
	20	0.877 0°	0.902 ±17.9°	0.675 ±91.9°	0.997 ±153.2°	0.959 180°	0.952 ±34.0°	0 ±0°	2.8
	50	0.906 0°	0.915 ±17.4°	0.904 ±85.4°	1.080 ±159.7°	0.959 180°	0.951 ±24.3°	0 ±0°	-
Fig. 3-12b	1	0.820 0°	0.868 ±19.3°	0.154 ±89.3°	0.950 ±143.8°	0.960 180°	0.953 ±98.7°	0 ±0°	2.2
	10	0.841 0°	0.881 ±18.8°	0.469 ±86.1°	0.963 ±148.6°	0.959 180°	0.952 ±44.8°	0 ±0°	2.5
	20	0.862 0°	0.889 ±18.4°	0.629 ±83.8°	0.992 ±152.3°	0.959 180°	0.952 ±34.0°	0 ±0°	2.8
	50	0.897 0°	0.907 ±17.7°	0.886 ±80.3°	1.070 ±158.7°	0.959 180°	0.951 ±24.3°	0 ±0°	-

TABLE 3-1: Pole-Zero Locations for Reduced-Order Control of Example 6 with Different Actuator Connections.

Ex#	plant poles*	plant zero(s)*	DC gain	actuator pole*	actuator DC gain	desired pole(s)*	desired DC gain
1	0.95,0.2	0.3	1.31	-	-	0.8	2
2	0.9,0.1	0.011	0.99	-	-	0.65	2
3	0.9,0.8	0.82	0.9	-	-	0.5	2
4	0.9,0.7	0.4	4.0	-	-	0.6	2
5	0.923+j0.231 -0.762+j0.570	-0.960 -0.144+j0.942	4.04	-	-	0.844+j0.171	2.21
6	0.923+j0.231 -0.762+j0.570	-0.960 -0.144+j0.942	4.04	0.607	1	0.844+j0.171	2.21

* All singularities are z-plane equivalents

TABLE 3-2: Examples Summary.

IV. ADAPTIVE CONTROL ALGORITHMS

Presently, adaptive control schemes can be categorized by the following characteristics: (i) basis for parameter update scheme via gradient (G) search procedures or stability (S) theory analysis, (ii) formulation of parameter estimate error as equation (E) or output (O) error, and (iii) designation of parameters to be identified as plant parameters for indirect (I) adaptive control or as controller parameters for direct (D) adaptive control. Of the eight combinations only seven are presently available with GOD not yet developed. Of the remaining seven, two pairs are pairwise identical, GEI/SEI and GED/SED, due to the resulting equivalence of gradient and stability based equation error parameter estimation. The remaining five distinct classes providing adaptive solution to the model following problems of the preceding section are detailed below with appropriate source reference. The designer selected variables of each are listed following each algorithm statement.

A reduced-order adaptive model-following problem requires specification of a time-varying feedback control law causing the output of the plant

$$y(k) = \sum_{j=1}^{\bar{m}} \bar{b}_j u(k-j) + \sum_{i=1}^{\bar{n}} \bar{a}_i y(k-i) \quad (4-1)$$

to asymptotically "follow" that of the stable reference model

$$s(k) = cr(k-1) + \sum_{i=1}^n d_i s(k-i) \quad (4-2)$$

without a priori knowledge of the plant parameters \bar{a}_i and \bar{b}_j or plant

dimensions \bar{m} and \bar{n} under the assumption that the plant is described by

$$y(k) = \sum_{j=1}^{\bar{m}} b_j u(k-j) + \sum_{i=1}^{\bar{n}} a_i y(k-i), \quad (4-3)$$

where $\bar{m} > m$ and $\bar{n} > n$. Only if $\bar{m} = m$, $\bar{n} = n$, $\bar{b}_j = b_j$, and $\bar{a}_i = a_i$ are the following solutions expected to cause $y \rightarrow s$. If, as in examples 5 and 6 of the preceding section, the model of (4-2) includes zeros, which are restricted to match those of (4-3), this converts the problem to one of adaptive pole-shifting. Since these zeros are identified in indirect adaptive control the pole-shifting objective presents no apparent difficulty, though no globally convergent, even nonreduced-order, indirect adaptive pole-shifting scheme has been proven due to sufficient excitation requirements [30] or matrix polynomial solution singularity problems [31]. Direct adaptive controllers must assume the availability of the plant numerator coefficients, i.e. the plant zeros, [32] or the stability of the plant [33] to achieve globally stable pole-shifting (or replacement). Therefore such direct adaptive control schemes (or those requiring numerator cancellation [17]) are not presented for examples 5 and 6. These restrictions on direct adaptive control support the statement that presently indirect adaptive control appears more suited to the ROAC problem requiring pole-shifting.

The algorithms, specialized for the examples of the preceding section and written in implementation sequence, are:

For Examples 1-4:

GEI/SEI [17], [21], [22], [34] - [36]:

$$e(k-1) = y(k-1) - \hat{a}(k-1)y(k-2) - \hat{b}(k-1)u(k-2) \quad (4-4)$$

$$\hat{a}(k) = \hat{a}(k-1) + \left[\frac{\mu}{h + \mu y^2(k-2) + \gamma u^2(k-2)} \right] y(k-2)e(k-1) \quad (4-5)$$

$$\hat{b}(k) = \hat{b}(k-1) + \left[\frac{\rho}{h + \mu y^2(k-2) + \gamma u^2(k-2)} \right] u(k-2)e(k-1) \quad (4-6)$$

$$\hat{c}(k-1) = \frac{c}{\hat{b}(k)} \quad (4-7)$$

$$\hat{f}(k-1) = \frac{d - \hat{a}(k)}{\hat{b}(k)} \quad (4-8)$$

Designer-selected variable requirements:

$$\mu > 0 \quad (4-9)$$

$$\rho > 0 \quad (4-10)$$

$$0 < h < 2 \quad (4-11)$$

h is nominally chosen as one unless such a value would cause $\hat{b}(k) = 0$. Then h is chosen within the range in (4-11) such that $|\hat{b}(k)| > \bar{\epsilon} = 0^+$.

GOI [23], [25], [35]:

$$\hat{y}(k-1) = \hat{a}(k-1)\hat{y}(k-2) + \hat{b}(k-1)u(k-2) \quad (4-12)$$

$$\lambda(k-1) = \hat{y}(k-2) + \hat{a}(k-1)\lambda(k-2) \quad (4-13)$$

$$\gamma(k-1) = u(k-2) + \hat{a}(k-1)\gamma(k-2) \quad (4-14)$$

$$e(k-1) = y(k-1) - \hat{y}(k-1) \quad (4-15)$$

$$\hat{a}(k) = \hat{a}(k-1) + \left[\frac{\mu}{h + \mu \lambda^2(k-1) + \gamma \gamma^2(k-1)} \right] \lambda(k-1)e(k-1) \quad (4-16)$$

$$\hat{b}(k) = \hat{b}(k-1) + \left[\frac{\rho}{h + \mu \lambda^2(k-1) + \gamma \gamma^2(k-1)} \right] \gamma(k-1)e(k-1) \quad (4-17)$$

$$\hat{g}(k-1) = \frac{c}{\hat{b}(k)} \quad (4-18)$$

$$\hat{f}(k-1) = \frac{d-\hat{a}(k)}{\hat{b}(k)} \quad (4-19)$$

Designer-selected variable restrictions:

$$\mu > 0 \text{ and small} \quad (4-20)$$

$$\rho > 0 \text{ and small} \quad (4-21)$$

$$0 < h < 2 \quad (4-22)$$

h is nominally chosen as one unless such a value would cause $\hat{b}(k) = 0$.

Then h is chosen within the range in (4-22) such that $|\hat{b}(k)| > \bar{\epsilon} = 0^+$.

SOI: [8], [24], [35]:

$$\hat{y}(k-1) = \hat{a}(k-1)z(k-2) + \hat{b}(k-1)u(k-2) \quad (4-23)$$

$$v(k-1) = \frac{v(k-1) - \hat{y}(k-1) + q[y(k-2) - z(k-2)]}{h + \mu z^2(k-2) + \rho u^2(k-2)} \quad (4-24)$$

$$\hat{a}(k) = \hat{a}(k-1) + \mu z(k-2)v(k-1) \quad (4-25)$$

$$\hat{b}(k) = \hat{b}(k-1) + \rho u(k-2)v(k-1) \quad (4-26)$$

$$z(k-1) = \hat{a}(k)z(k-2) + \hat{b}(k)u(k-2) \quad (4-27)$$

$$\hat{g}(k-1) = \frac{c}{\hat{b}(k)} \quad (4-28)$$

$$\hat{f}(k-1) = \frac{d-\hat{a}(k)}{\hat{b}(k)} \quad (4-29)$$

Designer-selected variable restrictions:

$$\mu > 0 \quad (4-30)$$

$$\rho > 0 \quad (4-31)$$

$$\operatorname{Re} \left\{ \frac{1+qz^{-1}}{1-a_1 z^{-1}} \right\} > 0 \quad \forall |z|=1 \quad (4-32)$$

q can be chosen as zero for any stable a_1 and (4-32) will be satisfied.

$$1 \leq h < 2 \quad (4-33)$$

h is nominally chosen as one unless such a value would cause $\hat{b}(k) \approx 0$.

Then h is chosen within the range in (4-33) such that $|\hat{b}(k)| > \bar{\epsilon} \approx 0^+$.

GEL/SED [17][37]:

$$v(k-1) = cr(k-2) + d y(k-2) - y(k-1) \quad (4-34)$$

$$\hat{g}(k-1) = \hat{g}(k-2) + \frac{\rho r(k-2)v(k-1)}{h[1+\mu y^2(k-2)+\rho u^2(k-2)]} \quad (4-35)$$

$$\hat{f}(k-1) = \hat{f}(k-2) + \frac{\mu y(k-2)v(k-1)}{h[1+\mu y^2(k-2)+\rho u^2(k-2)]} \quad (4-36)$$

Designer-selected variable restrictions

$$\mu > 0 \quad (4-37)$$

$$\rho > 0 \quad (4-38)$$

$$|h| > \frac{|b|}{2} \text{ and } \operatorname{sgn}(h) = \operatorname{sgn}(b) \quad (4-39)$$

SOD [17]-[20]:

$$\beta(k-1) = \rho r^2(k-2) + \mu y^2(k-2) \quad (4-40)$$

$$\gamma(k-1) = (d+q)h\beta(k-2)v(k-2) + d\gamma(k-2) \quad (4-41)$$

$$v(k-1) = (1+h\beta(k-1))^{-1} \{s(k-1) - y(k-1) + q[s(k-2) - y(k-2)] - \gamma(k-1)\} \quad (4-42)$$

$$\hat{g}(k-1) = \hat{g}(k-2) + \rho r(k-2)v(k-1) \quad (4-43)$$

$$\hat{f}(k-1) = \hat{f}(k-2) + \mu y(k-2)v(k-1) \quad (4-44)$$

Designer-selected variable restrictions:

$$\mu > 0 \quad (4-45)$$

$$\rho > 0 \quad (4-46)$$

$$|h| > \frac{|b|}{2} \text{ and } \text{sgn}(h) = \text{sgn}(b) \quad (4-47)$$

$$\text{Re} \left\{ \frac{1+qz^{-1}}{1-fz^{-1}} \right\} > 0 \quad \forall |z| = 1 \quad (4-48)$$

This last condition is easily satisfied by $q = -d$, which should equate SOD to GED/SED [18].

Each of the preceding schemes (4-4) - (4-48) provides $\hat{g}(k-1)$ and $\hat{f}(k-1)$

for

$$u(k-1) = \hat{g}(k-1)r(k-1) + \hat{f}(k-1)y(k-1) \quad (4-49)$$

for application to (3-2) and generation of $y(k)$. Then these recursions are repeated.

For Examples 5 and 6:

GEI/SEI:

$$\begin{aligned} e(k-1) = & y(k-1) - \hat{\alpha}_1(k-1)v(k-2) - \hat{\alpha}_2(k-2)y(k-3) \\ & - \hat{\beta}_1(k-1)u(k-2) - \hat{\beta}_2(k-2)u(k-3) \end{aligned} \quad (4-50)$$

$$v(k-1) = e(k-1) / [h + \mu_1 y^2(k-2) + \mu_2 y^2(k-3) + \rho_1 u^2(k-2) + \rho_2 u^2(k-3)] \quad (4-51)$$

$$\hat{\alpha}_1(k) = \hat{\alpha}_1(k-1) + \mu_1 v(k-1)y(k-2) \quad (4-52)$$

$$\hat{\alpha}_2(k) = \hat{\alpha}_2(k-1) + \mu_2 v(k-1)y(k-3) \quad (4-53)$$

$$\hat{\beta}_1(k) = \hat{\beta}_1(k-1) + \rho_1 v(k-1)u(k-2) \quad (4-54)$$

$$\hat{\beta}_2(k) = \hat{\beta}_2(k-1) + \rho_2 v(k-1)u(k-3) \quad (4-55)$$

Designer-selected variable restrictions:

$$\mu_1 > 0 \quad (4-56)$$

$$\mu_2 > 0 \quad (4-57)$$

$$\rho_1 > 0 \quad (4-58)$$

$$\rho_2 > 0 \quad (4-59)$$

$$0 < h < 2 \quad (4-60)$$

h is nominally chosen as one unless such a value would cause $\hat{\beta}_2(k) \approx 0$ or $\hat{\beta}_1(k)\hat{\alpha}_1(k) + \hat{\beta}_2(k) - \hat{\alpha}_2(k)\hat{\beta}_1^2(k)/\hat{\beta}_2(k) \approx 0$. Then h is chosen within the range in (4-60) such that the absolute value of both terms $> \bar{\epsilon} \approx 0^+$.

GOI:

$$\hat{y}(k-1) = \hat{\alpha}_1(k-1)\hat{y}(k-1) + \hat{\alpha}_2(k-1)\hat{y}(k-2) + \hat{\beta}_1(k-1)u(k-2) + \hat{\beta}_2(k-1)u(k-3) \quad (4-61)$$

$$\lambda_1(k-1) = \hat{y}(k-2) + \hat{\alpha}_1(k-1)\lambda_1(k-2) + \hat{\alpha}_2(k-1)\lambda_1(k-3) \quad (4-62)$$

$$\lambda_2(k-1) = \hat{y}(k-3) + \hat{\alpha}_1(k-1)\lambda_2(k-2) + \hat{\alpha}_2(k-1)\lambda_2(k-3) \quad (4-63)$$

$$\gamma_1(k-1) = u(k-2) + \hat{\alpha}_1(k-1)\gamma_1(k-2) + \hat{\alpha}_2(k-1)\gamma_1(k-3) \quad (4-64)$$

$$\gamma_2(k-1) = u(k-3) + \hat{\alpha}_1(k-1)\gamma_2(k-2) + \hat{\alpha}_2(k-1)\gamma_2(k-3) \quad (4-65)$$

$$v(k-1) = [y(k-1) - \hat{y}(k-1)]/[h + \mu_1\lambda_1^2(k-1) + \mu_2\lambda_2^2(k-1) + \rho_1\gamma_1^2(k-1) + \rho_2\gamma_2^2(k-1)] \quad (4-66)$$

$$\hat{\alpha}_1(k) = \hat{\alpha}_1(k-1) + \mu_1v(k-1)\lambda_1(k-1) \quad (4-67)$$

$$\hat{\alpha}_2(k) = \hat{\alpha}_2(k-1) + \mu_2v(k-1)\lambda_2(k-1) \quad (4-68)$$

$$\hat{\beta}_1(k) = \hat{\beta}_1(k-1) + \rho_1v(k-1)\gamma_1(k-1) \quad (4-69)$$

$$\hat{\beta}_2(k) = \hat{\beta}_2(k-1) + \rho_2v(k-1)\gamma_2(k-1) \quad (4-70)$$

Designer-selected variable restrictions:

$$\mu_1 > 0 \text{ and small} \quad (4-71)$$

$$\mu_2 > 0 \text{ and small} \quad (4-72)$$

$$\rho_1 > 0 \text{ and small} \quad (4-73)$$

$$\rho_2 > 0 \text{ and small} \quad (4-74)$$

$$0 < h < 2 \quad (4-75)$$

h is nominally chosen as one unless such a value would cause $\hat{\beta}_2(k) = 0$ or $\hat{\beta}_1(k)\hat{\alpha}_1(k) + \hat{\beta}_2(k) - \hat{\alpha}_2(k)\hat{\beta}_1^2(k)/\hat{\beta}_2(k) = 0$. Then h is chosen within the range of (4-75) such that the absolute value of both terms $> \bar{\epsilon} = 0^+$.

SOI:

$$\hat{y}(k-1) = \hat{\alpha}_1(k-1)z(k-2) + \hat{\alpha}_2(k-1)z(k-3) + \hat{\beta}_1(k-1)u(k-2) + \hat{\beta}_2(k-1)u(k-3) \quad (4-76)$$

$$v(k-1) = \frac{\{y(k-1) - \hat{y}(k-1) + q_1[y(k-2) - z(k-2)] + q_2[y(k-3) - z(k-3)]\}}{\{h + \mu_1 z^2(k-2) + \mu_2 z^2(k-3) + \rho_1 u^2(k-2) + \rho_2 u^2(k-3)\}} \quad (4-77)$$

$$\hat{\alpha}_1(k) = \hat{\alpha}_1(k-1) + \mu_1 v(k-1)z(k-2) \quad (4-78)$$

$$\hat{\alpha}_2(k) = \hat{\alpha}_2(k-1) + \mu_2 v(k-1)z(k-3) \quad (4-79)$$

$$\hat{\beta}_1(k) = \hat{\beta}_1(k-1) + \rho_1 v(k-1)u(k-2) \quad (4-80)$$

$$\hat{\beta}_2(k) = \hat{\beta}_2(k-1) + \rho_2 v(k-1)u(k-3) \quad (4-81)$$

$$z(k-1) = \hat{\alpha}_1(k)z(k-2) + \hat{\alpha}_2(k)z(k-3) + \hat{\beta}_1(k)u(k-2) + \hat{\beta}_2(k)u(k-3) \quad (4-82)$$

Designer-selected variable restrictions:

$$\mu_1 > 0 \quad (4-83)$$

$$\mu_2 > 0 \quad (4-84)$$

$$\rho_1 > 0 \quad (4-85)$$

$$\rho_2 > 0 \quad (4-86)$$

$$\operatorname{Re} \left\{ \frac{1 + q_1 z^{-1} + q_2 z^{-2}}{1 - \alpha_{11} z^{-1} - \alpha_{12} z^{-2}} \right\} > 0 \quad \forall |z| = 1 \quad (4-87)$$

q_1 could be chosen as -0.98 and q_2 as zero for the rapidly sampled examples 5 and 6 and (4-87) would be satisfied.

$$1 \leq h \leq 2 \quad (4-88)$$

h is nominally chosen as one unless such a value would cause $\hat{\beta}_2(k) = 0$ or $\hat{\beta}_1(k)\hat{\alpha}_1(k) + \hat{\beta}_2(k) - \hat{\alpha}_2(k)\hat{\beta}_1^2(k)/\hat{\beta}_2(k) = 0$. Then h is chosen within the range of (4-88) such that the absolute value of both terms $\rightarrow 0^+$.

Each of the last three algorithms in (4-50)-(4-88) provides $\hat{\alpha}_1(k)$, $\hat{\alpha}_2(k)$, $\hat{\beta}_1(k)$, $\hat{\beta}_2(k)$ for use in (refer to (3-27)-(3-30)).

$$\hat{\eta}_1(k-1) = \gamma_1 - \hat{\alpha}_1(k) \quad (4-89)$$

$$\hat{v}_1(k-1) = \frac{[\hat{\eta}_1(k-1)\hat{\alpha}_2(k) + (\hat{\eta}_1(k-1)\hat{\alpha}_1(k) - \hat{\alpha}_2(k) + \gamma_2)(\alpha_1(k) - \hat{\alpha}_2(k)\hat{\beta}_1(k)/\hat{\beta}_2(k))]}{[\hat{\beta}_1(k)\hat{\alpha}_1(k) + \hat{\beta}_2(k) - \hat{\alpha}_2(k)\hat{\beta}_1^2(k)/\hat{\beta}_2(k)]} \quad (4-90)$$

$$\hat{\eta}_2(k-1) = \hat{\alpha}_1(k)\hat{\eta}_1(k-1) + \gamma_2 - \hat{\beta}_1(k)\hat{v}_1(k-1) - \hat{\alpha}_2(k) \quad (4-91)$$

$$\hat{v}_2(k-1) = \hat{\eta}_2(k-1)\hat{\alpha}_2(k)/\hat{\beta}_2(k) \quad (4-92)$$

for parameterization of

$$\begin{aligned} u(k-1) = & \delta r(k-1) + \hat{\eta}_1(k-1)u(k-2) + \hat{\eta}_2(k-1)u(k-3) \\ & + \hat{v}_1(k-1)y(k-2) + \hat{v}_2(k-1)y(k-3) \end{aligned} \quad (4-93)$$

for application to the difference equation representation of (3-17)

$$\begin{aligned} y(k) = & (\alpha_{21} + \alpha_{11})y(k-1) + (\alpha_{22} - \alpha_{11}\alpha_{21} + \alpha_{12})y(k-2) \\ & - (\alpha_{11}\alpha_{22} + \alpha_{12}\alpha_{21})y(k-3) - \alpha_{12}\alpha_{22}y(k-4) \\ & + (\beta_{11} + \beta_{21})u(k-1) - (\beta_{11}\alpha_{21} - \beta_{12} + \beta_{21}\alpha_{11} - \beta_{22})u(k-2) \\ & - (\beta_{11}\alpha_{22} + \beta_{12}\alpha_{21} + \beta_{21}\alpha_{12} + \beta_{22}\alpha_{11})u(k-3) \\ & - (\beta_{12}\alpha_{22} + \beta_{22}\alpha_{12})u(k-4) \end{aligned} \quad (4-94)$$

generating $y(k)$. Note that all of the preceding schemes in (4-4)-(4-93) use the current information as efficiently as possible and therefore result in causal rather than strictly causal adaptive controllers [17]. (See Appendix F for programs of each of these controllers.)

V. TEST FORMATS

Having specified the non-adaptive and adaptive control strategies and objectives the remaining items to be selected for the testing of the guidelines in the second section are the reference inputs, adaption step-size constants, parameter estimate initializations, and performance measures.

Selection of the reference signal is important for two reasons: richness and magnitude. The magnitude of the reference signal is important due to the nonlinearity of the adaptive control system and therefore the anticipated transient (at least) differences due to different signal levels. The richness is important because it is well-recognized [38] that a sufficient "input" richness is necessary to perturb all of the modes of the system for their "identification." This "input" differs for the indirect and direct approaches and the equation and output error formulations. For the equation error indirect approach the plant input and output must be sufficiently rich. Since the plant is linear this translates to a sufficient frequency content of the plant input (or control) signal, of which the reference is only one component. The difficulty of assuming adaptive control input richness based on reference signal richness is addressed in [30]. For the output error indirect approach the plant input and identifier output must be sufficiently rich. In an open-loop output error identification task the identifier output richness can be translated [39] solely to input richness necessity under the assumption that the identifier output asymptotically converges to the plant output, which requires sufficient identifier order absent from ROAC. For direct adaptive control the control parameter identification problem must be

recast as an open-loop identification task as in [19] and [37]. In the equation error case the richness requirements are converted principally to conditions on the model output to be tracked or, if generated by a linear, time-invariant model, to the forcing or reference signal [39]. Once again this assumes sufficient controller order. The requirements for output error based direct adaptive control, though as yet unspecified, are assumed to be similar.

All of the preceding richness requirements are for complete identification and assume non-reduced adaptive model order. As is well-recognized [22] [34] [37], for a control objective complete identification may not be necessary. Consider the control of any order system with a single constant output feedback gain and a reference signal gain. To achieve asymptotic convergence to a particular DC level set-point any feedback gain stabilizing the system in conjunction with the appropriate reference gain will be adequate. This applies readily to examples 1-4 in section three. As noted in [40] for open-loop output error (and by implication equation error) identification the richness of the input signal for stable reduced-order identification is dependent on the reduced-order model dimension, not that of the higher order plant. From a frequency response point of view, the viewpoint promoted in [41] for reduced-order modelling, the minimum number of sinusoids providing this richness matches the number of points at which the controller can specify the steady-state frequency response. Therefore, if this reduced-order controller stabilizes the system, asymptotically convergent steady-state control seems possible up to the richness level of the reduced-order model. When the signal to be tracked (or equivalently the reference signal) has a higher frequency content, reduced-order tracking will only be approximate at best. The reference

signal should be chosen to test this boundary condition, with the principal case of interest being reference signal over-richness.

Therefore the following reference signals (all zero prior to $k = 0$) will be considered:

INPUT 1 (unit step input):

$$r_1(k) = 1 \quad (5-1)$$

INPUT 2 (single low-frequency sinusoid):

$$r_2(k) = 2 \sin(k\omega_2 T) \quad (5-2)$$

Note that the sample period T is assumed to be one second for examples 1-4. The frequency ω_2 is chosen as 20% below the desired cutoff frequency for examples 1-4, i.e. 0.17851, 0.34463, 0.55452, and 0.40866 radians per second, respectively, and 20% below the dominant natural frequency for examples 5 and 6, i.e. 0.40 radians per second, with $T = 0.5$ seconds.

INPUT 3 (single high-frequency sinusoid):

$$r_3(k) = 2 \sin(k\omega_3 T) \quad (5-3)$$

Again the sample period is assumed to be one second for examples 1-4. The frequency ω_3 is double the desired cutoff for examples 1-4, i.e. respectively 0.44629, 0.86157, 1.38629, and 1.02170, and double the dominant natural frequency for examples 5 and 6, i.e. 1.00, with $T = 0.5$ seconds.

INPUT 4 (non-zero mean, uniform, white noise):

$$r_4(k) \in [-0.5, 1.5] \quad (5-4)$$

This is clearly the over-rich input of most interest.

Since the rate of convergence and parameter estimate variance near convergence are commonly acknowledged as being affected by the adaptive step-size (SS) coefficients μ_i and ρ_j , two cases will be considered.

$$SS_1 = 1:$$

$$\mu_i = \rho_j = 1 \quad (5-5)$$

$$SS_2 = 0.1:$$

$$\mu_i = \rho_j = 0.1. \quad (5-6)$$

Given the approximate nature of the gradient formulas for GOI embodied in the required μ and ρ smallness noted in (4-20)-(4-21) and (4-71)-(4-74) the μ and ρ of (5-5) may lead to unstable behavior for GOI while (5-6) should prove more satisfactory. (Note that h in the normalizing term of all of the algorithms is selected as unity).

For output error type algorithms such as SOI and SOD the error smoothing coefficients q in (4-24), (4-47), and (4-77) are to be selected in satisfaction of (4-32), (4-48), and (4-87), respectively. As noted in [42] selection of these q to make (4-32), (4-48), and (4-87) equal unity does not offer the most rapid convergence. Furthermore, in a reduced-order application these q can influence the mean convergence point. Such factors may be used to advantage in (4-48) where f is known a priori, but in (4-32) and (4-87) the strictly positive real (SPR) condition is dependent on unknown plant parameters. Therefore several values will be tested. They are summarized in Table 5-1.

Since only local convergence is currently anticipated three settings of the initial plant (and therefore controller) parameter estimates (EST) will be considered corresponding to the nominal values from control design based on neglecting the second mode (i.e. $\epsilon = 0$ for examples 1-4 and $\lambda_2 = 0$ for examples 5 and 6) and $\pm 20\%$ of these values. These values are summarized in Table 5-2. The controller parameter initializations for the direct schemes are based on the corresponding plant parameter initializations and

Example (n)	Algorithm	Error Smoothing Coefficient (a)	Label
1	SOI	-0.9	SC-
		-0.95	SCO
		-0.97	SC+
	SOD	-0.7	SC-
		-0.8	SCO
		-0.9	SC+
2	SOI	-0.9	SCO
	SOD	-0.65	SCO
3	SOI	-0.9	SCO
	SOD	-0.5	SCO
4	SOI	-0.8	SC-
		-0.9	SCO
		-0.95	SC+
	SOD	-0.4	SC-
		-0.6	SCO
		-0.8	SC+
5&6	SOI	$q_1 = -0.9, q_2 = 0$	SC-
		$q_1 = -1.846, q_2 = 0.9515$	SCO
		$q_1 = -0.96, q_2 = 0$	SC+

TABLE 5-1: Error Smoothing Coefficients

Example	Parameter	EST = +20%	EST = 0	EST = -20%
1	$\hat{b}(0)$	0.078	0.065	0.052
	$\hat{a}_1(0)$	1.14	0.95	0.76
2	$\hat{b}(0)$	0.12	0.1	0.08
	$\hat{a}_1(0)$	1.08	0.9	0.72
3	$\hat{b}(0)$	0.096	0.08	0.064
	$\hat{a}_1(0)$	1.08	0.9	0.72
4	$\hat{b}(0)$	0.6	0.5	0.4
	$\hat{a}_1(0)$	1.08	0.9	0.72
5&6	$\hat{p}_{11}(0)$	0.144	0.120	0.096
	$\hat{p}_{12}(0)$	0.139	0.116	0.093
	$\hat{a}_{11}(0)$	2.215	1.846	1.477
	$\hat{a}_{12}(0)$	-1.086	-0.905	-0.724

TABLE 5-2: Parameter Estimate Initializations.

(3-8) and (3-9) for examples 1-4 and (3-27)-(3-30) for examples 5 and 6. Note the instability of the EST = + 20% initializations.

Since the control objective of all of the examples in section 3 is model-following, the tracking error is the principal performance measure. For the expectation of asymptotic tracking, the figures of merit should be divisible into short and long term quantities. The mean and variance of the tracking errors should be normalized and tabulated for comparison among the examples and between the fixed and adaptive control schemes. Similarly the input cost and controller (and plant, for indirect schemes) parameter estimates should be observed. Therefore the following quantities will be tabulated at $k_i = 1, 2, 5, 10, 20, 50, 100, 200, 500,$ and 1000 iterations.

Instantaneous desired output:

$$IDO(k) = s(k) \quad (5-7)$$

Instantaneous controlled output:

$$IO(k) = y(k) \quad (5-8)$$

Normalized instantaneous tracking error:

$$NITE(k) = \frac{(s(k) - y(k))}{s(k)} \quad (5-9)$$

Normalized average segmented tracking error:

$$NASTE = \left(\frac{1}{k_i - k_{i-1}} \right) \sum_{j=k_{i-1}+1}^{k_i} NITE(j) \quad (5-10)$$

Normalized segmented tracking error variance

$$NSTEV = \left(\frac{1}{n(n-1)} \right) \left\{ n \sum_{j=k_{i-1}+1}^{k_i} NITE^2(j) - \left[\sum_{j=k_{i-1}+1}^{k_i} NITE(j) \right]^2 \right\} \quad (5-11)$$

(Note that n is $k_i - k_{i-1}$ minus the number of times $s(k) < 10^{-3}$, which are excluded due to (5-9). If NSTEV is negative due to round-off, it is printed as zero.)

Segmented average squared input

$$SASI = \left(\frac{1}{k_i - k_{i-1}} \right) \sum_{j=k_{i-1}+1}^{k_i} u^2(j) \quad (5-12)$$

Segmented average controller parameter estimate for f

$$SACPE F = \left(\frac{1}{k_i - k_{i-1}} \right) \sum_{j=k_{i-1}+1}^{k_i} \hat{f}(j) \quad (5-13)$$

(Similarly for g , a , b , α , β , v , and n .)

Segmented controller parameter estimate variance for g

$$SCPEV G = \left\{ \frac{1}{(k_i - k_{i-1})(k_i - k_{i-1} - 1)} \right\} \left\{ (k_i - k_{i-1}) \sum_{j=k_{i-1}+1}^{k_i} \hat{f}^2(j) - \left[\sum_{j=k_{i-1}+1}^{k_i} \hat{f}(j) \right]^2 \right\} \quad (5-14)$$

(Similarly for g , a , b , α , β , v , and n .)

These quantities are tabulated for each of the combinations of example, adaptive control algorithm, reference input, step-size weights, error smoothing coefficients (if necessary), and parameter estimate initialization. (See Appendix F for listings of the programs compiling these figures of merit.) Table 5-3 summarizes all the various combinations that were simulated.

Example	Control Alg.	Inputs	Par. Est. Init.	Step Size	Error Smooth Coeff.	Actuator Configuration	Total
1	nonadaptive	all 4	EST = 0	zero	-	-	4
	GEI/SEI (1)	all 4	all 3	both	-	-	24
	GOI (2)	all 4	all 3	both	-	-	24
	SOI (3)	all 4	all 3	both	all 3	-	72
	GED/SED (4)	all 4	all 3	both	-	-	24
	SOD (5)	all 4	all 3	both	all 3	-	72
2	nonadaptive	all 4	EST = 0	zero	-	-	4
	GEI/SEI (1)	all 4	EST = 0	SS ₁	-	-	4
	GOI (2)	all 4	EST = 0	SS ₁	-	-	4
	SOI (3)	all 4	EST = 0	SS ₁	SCO	-	4
	GED/SED (4)	all 4	EST = 0	SS ₁	-	-	4
	SOD (5)	all 4	EST = 0	SS ₁	SCO	-	4
3	nonadaptive	all 4	EST = 0	zero	-	-	4
	GEI/SEI (1)	all 4	EST = 0	SS ₁	-	-	4
	GOI (2)	all 4	EST = 0	SS ₁	-	-	4
	SOI (3)	all 4	EST = 0	SS ₁	SCO	-	4
	GED/SED (4)	all 4	EST = 0	SS ₁	-	-	4
	SOD (5)	all 4	EST = 0	SS ₁	SCO	-	4
4	nonadaptive	all 4	EST = 0	zero	-	-	4
	GEI/SEI (1)	all 4	all 3	both	-	-	24
	GOI (2)	all 4	all 3	both	-	-	24
	SOI (3)	all 4	all 3	both	all 3	-	72
	GED/SED (4)	all 4	all 3	both	-	-	24
	SOD (5)	all 4	all 3	both	all 3	-	72
5	nonadaptive	all 4	all 3	zero	-	-	12
	GEI/SEI (1)	all 4	all 3	both	-	-	24
	GOI (2)	all 4	all 3	both	-	-	24
	SOI (3)	all 4	all 3	both	all 3	-	72
6	nonadaptive	all 4	all 3	zero	-	Fig. 12a&b	24
	GEI/SEI (1)	all 4	all 3	both	-	all 3	72
	GOI (2)	all 4	all 3	both	-	all 3	72
	SOI (3)	all 4	all 3	both	all 3	all 3	216
							1008

TABLE 5-3: Tested Combinations.

VI. RESULTS

Due to the number (1008) of printouts for the examples shown in Table 5-3 the tabulated results are under separate cover. (See Appendix F for pertinent program listings.) The next section provides an evaluation of these numerical results.

VII. TEST RESULTS INTERPRETATIONS

Refer to section 2 for the statement of the objectives of this study. The four questions raised in that section will be addressed in this section with respect to the simulations of Section 6.

QUESTION (1):

Are heavily damped open-loop modes neglectable relative to more lightly damped modes? In particular:

- a) Are the accepted criteria for neglecting modes, based on relative damping, realistic for the examples considered here in a fixed (non-adaptive) reduced-order controller?
- b) For these examples, how do the various adaptive control schemes fare under the reduced order modeling "rules" concerning relative damping? More specific questions might be: Do the reduced-order adaptive control systems remain stable? Can the adaptive mechanism compensate for mis-modeling to yield improved performance over non-adaptive (modal) controllers of the same order?

In example 1, the open loop and closed loop modes are separable based on the accepted criteria for relative damping, so the good closed-loop (non-adaptive) reduced-order performance was to be expected (despite the fact that in this example and all the others the reduced-order modal model was based strictly on the partial fraction expansion term without DC gain correction). Example 2 represents a marginally separable system in closed loop, yet the (non-adaptive) reduced-order controller tracking errors are only slightly worse than in example 1. This indicates that, for these examples, this boundary between separable and inseparable

modes is reasonable but fuzzy (as expected).

Example 3 possesses neither closed loop nor open loop separability, but a pole-zero near-cancellation should leave one mode dominating the system behavior. As shown in the simulation, the reduced-order modeling causes significant errors. This is presumably due to the relative closeness of the "cancelled" pole to the unit circle compared to the "dominant" pole. Small errors in cancellation in this case may have large effects on system behavior and the performance of reduced-order control.

The approximation of a pair of complex poles by a simple real axis pole is studied in example 4. Here, no first order model can be found to closely represent this second order system, and considerable degradation in performance by reduced order control should be expected. The simulations show this to be true; the steady state tracking errors are significant.

Of the five adaptive control algorithms, all but the second one, the gradient-output-error-indirect (GOI) scheme, were stable. In example one, which has the most accurate reduced order model, a larger number of test combinations of inputs, step sizes, and initial parameter estimates produce stable responses than the other examples. Even so, the only consistent results were for zero initial parameter error (with respect to the extracted modal model) and the step and random inputs. In these cases, the step response is improved over the non-adaptive case to essentially zero steady state error, while the random input causes a stable yet more widely ranging response. The other four adaptive algorithms produced essentially equivalent responses for most inputs, step sizes and smoothing coefficients. Parameter convergence occurred between 50 and 500 iterations for

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most runs, the random input case being consistently faster. Also, the error between the desired and actual outputs always resulted in improvement over the non-adaptive output error. Some minor differences in responses were:

- Example 3 usually responded faster, due to the dominant pole being more highly damped.
- Algorithm 5 (stability-output-error-direct, SOD) produced consistently smaller steady state output tracking errors.
- Smaller step sizes improved the steady state tracking error variance, but lengthened convergence times.
- The smoothing coefficient SC in algorithm 3 produced more accurate steady state tracking than the other smoothing coefficients.
- The adaptive parameter estimates have higher variances for example 2 than example 1 indicating more active searching due to the marginal dominance of the desired reduced-order objective in the problem setup.

Except for the GOI algorithm (#2), all adaptive schemes provided improved output tracking of desired models over non-adaptive control in examples 1-4. This is the result of the adaptive algorithms inclination to select some controller parameterization to "better fit" some control objective. In the indirect algorithms, the identified plant often has parameters that do not match either those of the actual plant or those of the assumed reduced order (modal) model. These effects are explored further in the discussion of the third question, on choosing the "best" reduced order model.

QUESTION (11):

If actuator dynamics are included in the system such that the residual modes have frequencies within the stop band of the actuator frequency response, does improved reduced-order control result?

Examples 5 and 6 were used to test the effects of actuator inclusion. Example 5 included no actuator dynamics and was seen to be unstable for a large number of runs, particularly those with initial parameter estimate errors relative to the dominant mode. Example 6 included the three actuator configurations shown in Figure 3-12, where actuator configuration 2 in the simulations corresponded to actuator (a) in the figure, 3 to (b), and 4 to (c).

In the non-adaptive simulations, inclusion of an actuator provided improved tracking for those runs of example 5 that were stable based on the reduced-order modal model. For the fixed (initial) reduced order model parameter estimates that resulted in an unstable control system, actuator inclusion generally did not provide stability. Actuator configurations 2 and 4 provided up to an order of magnitude improvement in tracking compared to 3 and 1 (no actuator), which were similar in response.

Under adaption, the most improvement was obtained, conversely, by including an actuator as in configuration 3. This arrangement produced stable simulations for all but a few runs, where the output was apparently unstable but still bounded within the 1000 iterations tested. For the stable adaptive usage of actuator configuration 3, nearly all cases provided between 1 and 3 orders of magnitude improvement in tracking error over the other three actuator configurations.

Actuator arrangements 2 and 4 behaved almost identically during adaption under all conditions. They were seen to improve the response

over that without actuator dynamics for most runs, but occasionally their inclusion worsened the output tracking. The only consistent improvement over configuration 3 was for input 3, the high frequency sinusoid, where an order of magnitude difference was sometimes seen.

Among the three algorithms used on examples 5 and 6, the first (GED/SED) and third (GEI) behaved essentially the same, giving the responses discussed above. Algorithm 2 (GOI) caused unstable responses for all combinations of inputs, step sizes, initial parameter estimates, and actuator configurations.

For these examples, inclusion of a finite bandwidth actuator generally improved the performance of the reduced-order control. However, there were some combinations of inputs, initial parameter estimates, and adaptive step sizes that caused poorer behavior with an actuator than without. The actuator arrangement that allowed both the plant identifier and controller to sample the actuator output was clearly the best choice for the majority of cases here. This structure effectively includes the actuator as part of the controller. As noted in section 3, this is not the most physically feasible configuration since the actuator outputs can be inaccessible.

QUESTION (111):

Does an "optimal" reduced order controller result from retaining the dominant modes of the plant for the reduced-order plant model, or by choosing a model (or controller) parameterization that does not correspond to plant modes but is based on matching (in some sense) a reduced order control objective?

As phrased, the latter alternative intuitively seems "better". This question can be answered for these examples by comparing the "zero initial parameter estimate error" parameters with the steady state parameters

in the adaptive runs. If the difference in parameters is slight, then the selection of the dominant plant mode for the control model would seem to be a reasonable choice with respect to the adaptive parameter estimator update criterion. If the parameter difference is considerable and the adaptive responses are deemed better than the non-adaptive ones, then a reduced order model based on those parameters selected by the "behavior approximating" adaptive algorithms would seem the better choice.

As would be expected, the adaptive runs showing the greatest improvement over the fixed controller runs were those with the largest differences in initial and steady state parameters. The implication is that choosing dominant modes out of the set of plant modes for the reduced order model is often not as "optimal" as choosing model "modes" based on some performance criterion.

QUESTION (iv):

Are indirect and direct adaptive controllers equivalent and interchangeable in reduced-order application?

The absurdity of such an expectation is underscored by a comparison of the algorithms in section IV. Note that indirect schemes involve division by estimated quantities as in, e.g., (4-7) and (4-8), but that direct schemes do not. Clearly, if the estimated quantities ever approach zero the resulting quotient is quite large. A similar mechanism does not seem to appear in the direct algorithms. The claims of equivalence seem to rest on the (linear?) transformability between the indirect and direct approaches in simple cases, such as inverse or model-following control. If the algorithms are not identical then, as clearly demonstrated by this study, the transient differences in the full order case translate into steady-state differences in the reduced-order case. Unfortunately neither indirect or direct adaptive control seems consistently superior to the other in all situations.

VIII. CONCLUSION

The bulk of the data generated from the 1008 simulations attests to the difficulty of drawing definitive conclusions from such simulation studies. This is further complicated for studies of adaptive controllers by the number of variables, e.g. plant parameters, control objective, initial parameter estimates, step-size weights, error smoothing coefficients and input richness, effecting overall performance. Conclusively examining all possibilities is clearly unfeasible. Even a focused study, such as this one with respect to the four objectives commented on in the preceding section, can only be expected to provide heavily qualified remarks. At best, such a study can only be used to test and augment the "folklore". As particular effects are isolated without contradiction further simulation studies are in order to test seeming generalizations. The ones that survive closer scrutiny provide guideposts for restricted analytical studies, that could prove equally fruitless if not initially well-focused.

The data generated by this study may provide a database for judging eventual conclusions of future studies. A different format, e.g. graphical, for even the tests run here might facilitate evaluation of different conclusions. The one general (tautological) conclusion that seems well documented by this report is that all present adaptive control schemes will fail in some reduced-order application. Though most of the (stable) adaptive controllers tested seem to better the performance of fixed modal controllers this does not present a conclusive argument for unsupervised adaptive control. It is trite but true that further study is necessary.

VIII.2

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Appendix A: Root Locus Program Listings and Tabular Outputs

- Contents: p. A-1: Program determining p_1 and p_2 in (3-12) for Examples 1-4
- A-2: p_1 , p_2 , h , and f for Example 1 (Fig. 3-1)
- A-3: p_1 , p_2 , h , and f for Example 2 (Fig. 3-3)
- A-4: p_1 , p_2 , h , and f for Example 3 (Fig. 3-5)
- A-5: p_1 , p_2 , h , and f for Example 4 (Fig. 3-8)
- A-6: Program determining q_1 and q_2 in (3-15) for Examples 1-4
- A-7: q_1 , q_2 , h , and f for Example 1 (Fig. 3-2)
- A-8: q_1 , q_2 , h , and f for Example 2 (Fig. 3-4)
- A-9: q_1 , q_2 , h , and f for Example 3 (Fig. 3-6)
- A-10: q_1 , q_2 , h , and f for Example 4 (Fig. 3-9)
- A-11: Program determining $q_1 = a_1$, $q_2 = a_2 + \epsilon f$,
i.e. $[0 \ f]$ in (3-14), for Example 3
- A-12: q_1 , q_2 , h , and f for Example 3 (Fig. 3-7)
- A-13: Sample Plot Program (Fig. 3-1)

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1 JOB      WAIFIV,KP=29,PACE=100,TIME=100
C          RTLL:
C          ROOTLOCUS INTERPRETATION OF THE UNADAPTED PLANT.
C          T.F.: Y(Z)/F(Z) = ((B+E)*Z - (B*A2+E*A1)) / (Z**2 - (A1+A2+F*(B+E))*Z + A1*A
C          2+F*(B*A2+L*A1)
C          DIMENSION E(4),XX1(500),XX2(500),YY1(500),YY2(500),A2(500),F2(50)
C          DIMENSION A1(4),A2(4),B(4),YM1(500),YM2(500)
C          COMPLEX GC,FC,ZC(2)
C          DATA E/0.01,-0.01,0.02,-0.3/
C          DATA A1/0.95,0.9,0.4,0.9/
C          DATA A2/0.2,0.1,0.8,0.7/
C          DATA B/0.065,0.1,0.01,0.5/
C          DO 300 J=1,4
C          WRITE(6,200)A1(J),A2(J),B(J)
200        FORMAT(1X,'A1=',F13.6,5X,'A2=',F13.6,5X,'B=',F13.6)
C          WRITE(6,400)E(J)
400        FORMAT(//,'E=',F13.6)
C          DO 100 K=1,50
C          IF(J.EQ.4) GO TO 13
C          F2(K)=-0.2*(K-1)
C          GC=1
C          F2(K)=0.2*(K-1)
13         G=(A1(J)+A2(J)+F2(K)*(B(J)+E(J)))
1          F=(A1(J)*A2(J)+F2(K)*(B(J)*A2(J)+E(J)*A1(J)))
C          GC=CMPLX(G,0.0)
C          FC=CMPLX(F,0.0)
C          ZC(1)=(GC+(SQRT(GC**2-4.0*FC)))/2.0
C          ZC(2)=(GC-CSQRT(GC**2-4.0*FC))/2.0
C          ZR1=REAL(ZC(1))
C          ZR2=REAL(ZC(2))
C          ZI1=AIMAG(ZC(1))
C          ZI2=AIMAG(ZC(2))
C          XX1(K)=ZR1
C          XX2(K)=ZR2
C          YY2(K)=ZI2
C          YY1(K)=ZI1
C          YMAG1=SQRT(ZK1**2+ZI1**2)
C          YMAG2=SQRT(ZK2**2+ZI2**2)
C          YM1(K)=YMAG1
C          YM2(K)=YMAG2
C          V=R(J)+E(J)
C          W=(B(J)*A2(J)+E(J)*A1(J))
C          AZ(K)=W/V
100        CONTINUE
C          WRITE(6,60)
60         FORMAT(/,5X,'REAL',5X,'IMAGIN',8X,'MAG1',5X,'REAL',5X,'IMAGIN',3X
C          +,'MAG2',5X,'ZK1',8X,'FELDBACK')
C          DO 500 I=1,50
C          WRITE(6,29)XX1(I),YY1(I),YM1(I),XX2(I),YY2(I),YM2(I),AZ(I),F2(I)
29         FORMAT(8F13.6)
C          CONTINUE
500        CONTINUE
300        STOP
          END

```

A1= 0.950000 A2= 0.200000 E= 0.010000 E= 0.065000

REAL	IMAGIN	MAG1	E=	REAL	IMAGIN	MAG2	ZERO	FEEDBACK
0.949999	0.000000	0.949999	0.010000	0.200000	0.000000	0.200000	0.300000	0.000000
0.937034	0.000000	0.937034	0.010000	0.197965	0.000000	0.197965	0.300000	-0.200000
0.924141	0.000000	0.924141	0.010000	0.195857	0.000000	0.195857	0.300000	-0.400000
0.911325	0.000000	0.911325	0.010000	0.193674	0.000000	0.193674	0.300000	-0.600000
0.898588	0.000000	0.898588	0.010000	0.191411	0.000000	0.191411	0.300000	-0.800000
0.885932	0.000000	0.885932	0.010000	0.189066	0.000000	0.189066	0.300000	-1.000000
0.873364	0.000000	0.873364	0.010000	0.186635	0.000000	0.186635	0.300000	-1.200000
0.860887	0.000000	0.860887	0.010000	0.184112	0.000000	0.184112	0.300000	-1.400000
0.848504	0.000000	0.848504	0.010000	0.181496	0.000000	0.181496	0.300000	-1.599999
0.836216	0.000000	0.836216	0.010000	0.178781	0.000000	0.178781	0.300000	-1.799999
0.824037	0.000000	0.824037	0.010000	0.175963	0.000000	0.175963	0.300000	-1.999999
0.811962	0.000000	0.811962	0.010000	0.173038	0.000000	0.173038	0.300000	-2.200000
0.800000	0.000000	0.800000	0.010000	0.170000	0.000000	0.170000	0.300000	-2.400000
0.788154	0.000000	0.788154	0.010000	0.166845	0.000000	0.166845	0.300000	-2.599999
0.776431	0.000000	0.776431	0.010000	0.163569	0.000000	0.163569	0.300000	-2.799999
0.764834	0.000000	0.764834	0.010000	0.160165	0.000000	0.160165	0.300000	-2.999999
0.753370	0.000000	0.753370	0.010000	0.156629	0.000000	0.156629	0.300000	-3.200000
0.742043	0.000000	0.742043	0.010000	0.152956	0.000000	0.152956	0.300000	-3.400000
0.730860	0.000000	0.730860	0.010000	0.149139	0.000000	0.149139	0.300000	-3.599999
0.719825	0.000000	0.719825	0.010000	0.145174	0.000000	0.145174	0.300000	-3.799999
0.708945	0.000000	0.708945	0.010000	0.141055	0.000000	0.141055	0.300000	-3.999999
0.698224	0.000000	0.698224	0.010000	0.136776	0.000000	0.136776	0.300000	-4.199999
0.687668	0.000000	0.687668	0.010000	0.132331	0.000000	0.132331	0.300000	-4.400000
0.677283	0.000000	0.677283	0.010000	0.127716	0.000000	0.127716	0.300000	-4.599999
0.667075	0.000000	0.667075	0.010000	0.122925	0.000000	0.122925	0.300000	-4.799999
0.657048	0.000000	0.657048	0.010000	0.117952	0.000000	0.117952	0.300000	-4.999999
0.647207	0.000000	0.647207	0.010000	0.112792	0.000000	0.112792	0.300000	-5.199999
0.637556	0.000000	0.637556	0.010000	0.107441	0.000000	0.107441	0.300000	-5.400000
0.628106	0.000000	0.628106	0.010000	0.101894	0.000000	0.101894	0.300000	-5.599999
0.618854	0.000000	0.618854	0.010000	0.096145	0.000000	0.096145	0.300000	-5.799999
0.609807	0.000000	0.609807	0.010000	0.090192	0.000000	0.090192	0.300000	-5.999999
0.600968	0.000000	0.600968	0.010000	0.084031	0.000000	0.084031	0.300000	-6.199999
0.592342	0.000000	0.592342	0.010000	0.077658	0.000000	0.077658	0.300000	-6.400000
0.583930	0.000000	0.583930	0.010000	0.071070	0.000000	0.071070	0.300000	-6.599999
0.575734	0.000000	0.575734	0.010000	0.064266	0.000000	0.064266	0.300000	-6.799999
0.567757	0.000000	0.567757	0.010000	0.057243	0.000000	0.057243	0.300000	-6.999999
0.559999	0.000000	0.559999	0.010000	0.050000	0.000000	0.050000	0.300000	-7.199999
0.552463	0.000000	0.552463	0.010000	0.042537	0.000000	0.042537	0.300000	-7.399999
0.545146	0.000000	0.545146	0.010000	0.034853	0.000000	0.034853	0.300000	-7.599999
0.538050	0.000000	0.538050	0.010000	0.026949	0.000000	0.026949	0.300000	-7.799999
0.531173	0.000000	0.531173	0.010000	0.018826	0.000000	0.018826	0.300000	-7.999999
0.524514	0.000000	0.524514	0.010000	0.010486	0.000000	0.010486	0.300000	-8.199999
0.518069	0.000000	0.518069	0.010000	0.001930	0.000000	0.001930	0.300000	-8.399999
0.511837	0.000000	0.511837	0.010000	0.006838	0.000000	0.006838	0.300000	-8.599999
0.505816	0.000000	0.505816	0.010000	0.015616	0.000000	0.015616	0.300000	-8.799999
0.500000	0.000000	0.500000	0.010000	0.025000	0.000000	0.025000	0.300000	-8.999999
0.494386	0.000000	0.494386	0.010000	0.034386	0.000000	0.034386	0.300000	-9.199999
0.488970	0.000000	0.488970	0.010000	0.043370	0.000000	0.043370	0.300000	-9.399999
0.483747	0.000000	0.483747	0.010000	0.053747	0.000000	0.053747	0.300000	-9.599999
0.478712	0.000000	0.478712	0.010000	0.063712	0.000000	0.063712	0.300000	-9.799999

ORIGINAL PAGE IS OF POOR QUALITY

A1= C.500000 A2= C.500000 F= 0.0E0000

REAL	IMAGIN	MAG1	E=	PEAL	IMAGIN	MAG2	ZERO	FEEBACK
0.899999	0.000000	0.899999	0.800000	0.800000	0.000000	0.800000	0.819999	0.000000
0.884714	0.000000	0.884714	0.795265	0.795265	0.000000	0.795265	0.819999	-0.200000
0.871230	0.000000	0.871230	0.788770	0.788770	0.000000	0.788770	0.819999	-0.400000
0.859991	0.000000	0.859991	0.780008	0.780008	0.000000	0.780008	0.819999	-0.600000
0.851227	0.000000	0.851227	0.768773	0.768773	0.000000	0.768773	0.819999	-0.800000
0.844714	0.000000	0.844714	0.740004	0.740004	0.000000	0.740004	0.819999	-1.000000
0.839995	0.000000	0.839995	0.723437	0.723437	0.000000	0.723437	0.819999	-1.200000
0.836562	0.000000	0.836562	0.705972	0.705972	0.000000	0.705972	0.819999	-1.400000
0.834028	0.000000	0.834028	0.687892	0.687892	0.000000	0.687892	0.819999	-1.599999
0.832107	0.000000	0.832107	0.669379	0.669379	0.000000	0.669379	0.819999	-1.799999
0.830621	0.000000	0.830621	0.650560	0.650560	0.000000	0.650560	0.819999	-2.000000
0.829439	0.000000	0.829439	0.631516	0.631516	0.000000	0.631516	0.819999	-2.200000
0.828482	0.000000	0.828482	0.612299	0.612299	0.000000	0.612299	0.819999	-2.400000
0.827700	0.000000	0.827700	0.592957	0.592957	0.000000	0.592957	0.819999	-2.599999
0.827042	0.000000	0.827042	0.573511	0.573511	0.000000	0.573511	0.819999	-2.799999
0.826488	0.000000	0.826488	0.553989	0.553989	0.000000	0.553989	0.819999	-2.999999
0.826010	0.000000	0.826010	0.534399	0.534399	0.000000	0.534399	0.819999	-3.200000
0.825600	0.000000	0.825600	0.514761	0.514761	0.000000	0.514761	0.819999	-3.400000
0.825237	0.000000	0.825237	0.495078	0.495078	0.000000	0.495078	0.819999	-3.599999
0.824922	0.000000	0.824922	0.475360	0.475360	0.000000	0.475360	0.819999	-3.799999
0.824639	0.000000	0.824639	0.455611	0.455611	0.000000	0.455611	0.819999	-3.999999
0.824389	0.000000	0.824389	0.435837	0.435837	0.000000	0.435837	0.819999	-4.199999
0.824161	0.000000	0.824161	0.416041	0.416041	0.000000	0.416041	0.819999	-4.400000
0.823959	0.000000	0.823959	0.396227	0.396227	0.000000	0.396227	0.819999	-4.599999
0.823772	0.000000	0.823772	0.376395	0.376395	0.000000	0.376395	0.819999	-4.799999
0.823605	0.000000	0.823605	0.356550	0.356550	0.000000	0.356550	0.819999	-4.999999
0.823450	0.000000	0.823450	0.336691	0.336691	0.000000	0.336691	0.819999	-5.199999
0.823309	0.000000	0.823309	0.316822	0.316822	0.000000	0.316822	0.819999	-5.400000
0.823177	0.000000	0.823177	0.296943	0.296943	0.000000	0.296943	0.819999	-5.599999
0.823057	0.000000	0.823057	0.277055	0.277055	0.000000	0.277055	0.819999	-5.799999
0.822944	0.000000	0.822944	0.257159	0.257159	0.000000	0.257159	0.819999	-6.000000
0.822841	0.000000	0.822841	0.237256	0.237256	0.000000	0.237256	0.819999	-6.199999
0.822743	0.000000	0.822743	0.217347	0.217347	0.000000	0.217347	0.819999	-6.399999
0.822653	0.000000	0.822653	0.197432	0.197432	0.000000	0.197432	0.819999	-6.599999
0.822567	0.000000	0.822567	0.177511	0.177511	0.000000	0.177511	0.819999	-6.799999
0.822489	0.000000	0.822489	0.157586	0.157586	0.000000	0.157586	0.819999	-6.999999
0.822414	0.000000	0.822414	0.137656	0.137656	0.000000	0.137656	0.819999	-7.199999
0.822344	0.000000	0.822344	0.117723	0.117723	0.000000	0.117723	0.819999	-7.399999
0.822277	0.000000	0.822277	0.097786	0.097786	0.000000	0.097786	0.819999	-7.599999
0.822214	0.000000	0.822214	0.077845	0.077845	0.000000	0.077845	0.819999	-7.799999
0.822154	0.000000	0.822154	0.057955	0.057955	0.000000	0.057955	0.819999	-7.999999
0.822098	0.000000	0.822098	0.037955	0.037955	0.000000	0.037955	0.819999	-8.199999
0.822045	0.000000	0.822045	0.018006	0.018006	0.000000	0.018006	0.819999	-8.399999
0.821994	0.000000	0.821994	0.001945	0.001945	0.000000	0.001945	0.819999	-8.599999
0.821899	0.000000	0.821899	0.001699	0.001699	0.000000	0.001699	0.819999	-8.799999
0.821855	0.000000	0.821855	0.001455	0.001455	0.000000	0.001455	0.819999	-8.999999
0.821813	0.000000	0.821813	0.001181	0.001181	0.000000	0.001181	0.819999	-9.199999
0.821773	0.000000	0.821773	0.000817	0.000817	0.000000	0.000817	0.819999	-9.399999
0.821735	0.000000	0.821735	0.000410	0.000410	0.000000	0.000410	0.819999	-9.599999


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3JOB      WATF IV, KP=24, PAGE=100, TIME=100
C          RTL2:
C          ROOTLOCUS INTERPRETATION OF THE UNADAPTED PLANT.
C          T.F.:  $Y(Z)/R(Z) = ((5+Z)*Z - (3*A2+L*A1))/(Z**2 - (A1+A2+F*F) + A1*A2 + R*A$ 
C          F)
          DIMENSION F(4), XX1(500), XX2(500), YY1(500), YY2(500), AZ(500), F2(50)
          DIMENSION A1(4), A2(4), B(4)
          COMPLEX CC, FC, ZC(2)
          DATA E/0.01, -0.01, 0.02, -0.3/
          DATA A1/0.95, 0.9, 0.9, 0.9/
          DATA A2/0.2, 0.1, 0.8, 0.7/
          DATA B/0.005, 0.1, 0.08, 0.5/
          DO 300 J=1,4
200        WRITE(6,200) A1(J), A2(J), F(J)
          FC=FORMAT(1X, 'A1=', F13.6, 5X, 'A2=', F13.6, 5X, 'F=', F13.6)
400        WRITE(6,400) F(J)
          FC=FORMAT(//, 'F=', F13.6)
          DO 100 K=1,50
          IF (J.EQ.4) GO TO 17
          F2(K)=-0.2*(K-1)
          GO TO 1
17         F2(K)=0.2*(K-1)
1          G=(A1(J)+A2(J)+F2(F)*F(J))
          F=(A1(J)*A2(J)+F2(F)*A(J)*A2(J))
          GC=CMPLX(0,0.0)
          FC=CMPLX(F,0.0)
          ZC(1)=(GC+CMPLX(1, -4.0*FC))/2.0
          ZC(2)=(GC-CMPLX(1, -4.0*FC))/2.0
          ZR1=REAL(ZC(1))
          ZR2=REAL(ZC(2))
          ZI1=AIMAG(ZC(1))
          ZI2=AIMAG(ZC(2))
          XX1(K)=ZR1
          XX2(K)=ZR2
          YY1(K)=ZI1
          YY2(K)=ZI2
          V=R(J)+F(J)
          W=(B(J)*A2(J)+F(J)*A1(J))
          AZ(K)=W/V
100        CONTINUE
          WRITE(6,60)
60         FORMAT(/, 5X, 'REAL', 5X, 'IMAGIN', 5X, 'REAL', 5X, 'IMAGIN', 5X, 'ZERO', 5X
          , 'FEEDBACK')
          DO 500 I=1,50
20         WRITE(6,20) XX1(I), YY1(I), XX2(I), YY2(I), AZ(I), F2(I)
          FC=FORMAT(6F13.6)
29         CONTINUE
300        CONTINUE
          STOP
          END

```

AI= ... IF= ...

REAL	IMAGIN	REAL	IMAGIN	ZEROS	FEEDBACK
0.945594	0.000000	0.000000	0.000000	0.350000	0.000000
0.921900	0.000000	0.000000	0.000000	0.500000	0.000000
0.927960	0.000000	0.000000	0.000000	0.350000	0.000000
0.913060	0.000000	0.000000	0.000000	0.350000	0.000000
0.847997	0.000000	0.000000	0.000000	0.350000	1.000000
0.864663	0.000000	0.000000	0.000000	0.350000	1.200000
0.871999	0.000000	0.000000	0.000000	0.350000	1.200000
0.844943	0.000000	0.000000	0.000000	0.350000	1.400000
0.844903	0.000000	0.000000	0.000000	0.350000	1.400000
0.845569	0.000000	0.000000	0.000000	0.350000	1.700000
0.806663	0.000000	0.000000	0.000000	0.350000	1.700000
0.792669	0.000000	0.000000	0.000000	0.350000	1.700000
0.751000	0.000000	0.000000	0.000000	0.350000	2.400000
0.765000	0.000000	0.000000	0.000000	0.350000	2.400000
0.755000	0.000000	0.000000	0.000000	0.350000	2.700000
0.746000	0.000000	0.000000	0.000000	0.350000	2.700000
0.726000	0.000000	0.000000	0.000000	0.350000	3.000000
0.716000	0.000000	0.000000	0.000000	0.350000	3.000000
0.705000	0.000000	0.000000	0.000000	0.350000	3.400000
0.690000	0.000000	0.000000	0.000000	0.350000	3.400000
0.677000	0.000000	0.000000	0.000000	0.350000	3.700000
0.664000	0.000000	0.000000	0.000000	0.350000	3.700000
0.661000	0.000000	0.000000	0.000000	0.350000	4.000000
0.630000	0.000000	0.000000	0.000000	0.350000	4.000000
0.630000	0.000000	0.000000	0.000000	0.350000	4.700000
0.610000	0.000000	0.000000	0.000000	0.350000	4.700000
0.610000	0.000000	0.000000	0.000000	0.350000	5.000000
0.590000	0.000000	0.000000	0.000000	0.350000	5.000000
0.585000	0.000000	0.000000	0.000000	0.350000	5.500000
0.572000	0.000000	0.000000	0.000000	0.350000	5.500000
0.555000	0.000000	0.000000	0.000000	0.350000	5.700000
0.540000	0.000000	0.000000	0.000000	0.350000	5.700000
0.535000	0.000000	0.000000	0.000000	0.350000	6.000000
0.520000	0.000000	0.000000	0.000000	0.350000	6.000000
0.517000	0.000000	0.000000	0.000000	0.350000	6.700000
0.481500	0.000000	0.000000	0.000000	0.350000	6.700000
0.481500	0.000000	0.000000	0.000000	0.350000	7.000000
0.455000	0.000000	0.000000	0.000000	0.350000	7.000000
0.442000	0.000000	0.000000	0.000000	0.350000	7.500000
0.442000	0.000000	0.000000	0.000000	0.350000	7.500000
0.416000	0.000000	0.000000	0.000000	0.350000	7.900000
0.403500	0.000000	0.000000	0.000000	0.350000	7.900000
0.373500	0.000000	0.000000	0.000000	0.350000	8.300000
0.373500	0.000000	0.000000	0.000000	0.350000	8.300000
0.364500	0.000000	0.000000	0.000000	0.350000	8.700000
0.351000	0.000000	0.000000	0.000000	0.350000	8.700000
0.332500	0.000000	0.000000	0.000000	0.350000	9.000000
0.332500	0.000000	0.000000	0.000000	0.350000	9.000000
0.312500	0.000000	0.000000	0.000000	0.350000	9.700000
0.312500	0.000000	0.000000	0.000000	0.350000	9.700000


```

1000      WAIT IV, NF = 29, PAGE = 100, TIME = 100
C
C      KILL
C      ROOT LOCUS INTERPRETATION OF THE UNADAPTED PLANT.
C      T.F.:  $Y(Z)/K(Z) = (1+L)*Z - (A*Z + E*A1) / (Z**2 - (A1+A2 + F*Z) + A1*A2 + F1*Z$ 
C      *F)
C      DIMENSION XXI(500), XZ(500), YY1(500), YY2(500), AZ(500), F(500)
C      COMPLEX CC, FC, ZC(1)
C      A1 = 0.0
C      A2 = 0.0
C      E = 0.0
C      F = 0.0
C      WRITE(6,200) A1, A2, E
200      FORMAT(1X, 'A1 = ', F10.0, 'X, 'A2 = ', F10.0, 'X, 'E = ', F10.0)
C      WRITE(6,400) F
400      FORMAT(1Z, 'F = ', F10.0)
C      DO 100 K = 1, 500
C      F2(K) = -0.2*(K-1)
C      G = (A1+A2 + F2(K)*L)
C      F = (A1*A2 + A1*F1 + F2(K))
C      CC = CMPLX(G, 0.0)
C      FC = CMPLX(F, 0.0)
C      ZC(1) = (CC + CSQRT(4.0*FC - 4.0*G*FC)) / 2.0
C      ZC(2) = (CC - CSQRT(4.0*FC - 4.0*G*FC)) / 2.0
C      ZK1 = REAL(ZC(1))
C      ZK2 = REAL(ZC(2))
C      ZI1 = AIMAG(ZC(1))
C      ZI2 = AIMAG(ZC(2))
C      XXI(K) = ZK1
C      XX2(K) = ZK2
C      YY1(K) = ZI1
C      YY2(K) = ZI2
C      V = B*G
C      W = (1+A1 + F2(K)*A1)
C      AZ(K) = W/V
100      CONTINUE
C      WRITE(6,600)
600      FORMAT(2,2X, 'REAL', 2X, 'IMAG IN', 2X, 'REAL', 2X, 'IMAG IN', 2X, 'ZK1', 2X
C      , 'ZK2', 2X, 'ZI1', 2X, 'ZI2', 2X, 'AZ', 2X, 'F', 2X)
C      DO 500 I = 1, 500
C      WRITE(6,200) XZ(I), YY1(I), XX2(I), YY2(I), AZ(I), F(I)
200      FORMAT(13,*)
500      CONTINUE
C      STOP
C      END

```


C
C

```

PLOT1:
ROOTLOCUS INTERPRETATION OF THE UNADAPTED PLANT.
DIMENSION XX1(500),XX2(500),YY1(500),YY2(500),YY3(500),F2(51)
DIMENSION XXX(51),YYY(51),XX3(500)
COMPLEX GC,FC,ZC(2)
YMAX=1.25
YMIN=-1.25
A1=0.95
A2=0.2
B=0.065
E=0.01
N=50
DO 100 K=1,51
F2(K)=-0.2*(K-1)
G=(A1+A2+F2(K))*(B+E)
F=A1*A2+F2(K)*(B*A2+E*A1)
GC=CMPLX(G,0.0)
FC=CMPLX(F,0.0)
ZC(1)=(GC+CSQRT(GC**2-4.0*FC))/2.0
ZC(2)=(GC-CSQRT(GC**2-4.0*FC))/2.0
ZR1=REAL(ZC(1))
ZR2=REAL(ZC(2))
ZI2=AIMAG(ZC(2))
ZI1=AIMAG(ZC(1))
XX1(K)=ZI1
XX2(K)=ZI2
YY1(K)=ZR1
YY2(K)=ZR2
V=B+E
W=B*A2+E*A1
YY3(K)=W/V
XXX(K)=-0.2*(K-1)
CX=XXX(1)
CY=YY3(1)
100 CONTINUE
CALL PLOTS(0.0,0.0,51)
CALL PLOT(5.0,6.0,-3)
CALL FACTOR(0.5)
DT=XXX(51)/10.0
CALL AXIS (0.0,0.0,'FEEDBACK PAK.',-13,10.0,0.0,0.0,DT)
CALL GRID (0.0,-5.0,10,1.0,10,1.0,-30584)
YT=ABS(YMAX)
IF(ABS(YMIN).GT.YT) YT=ABS(YMIN)
YSTART=-YT
DY=YT/5.0
CALL PLOT(XXX(1)/DT,YY3(1)/DY,3)
10 DO 10 I=2,N
CALL PLOT(XXX(I)/DT,YY3(I)/DY,2)
CALL AXIS (0.0,-5.0,'PLANT POLES AND ZERU',20,10.0,90.0,YSTART,DY
*)
CALL NEWPEN(3)
CALL PLOT(XXX(1)/DT,YY1(1)/DY,3)
20 DO 20 I=2,N
CALL PLOT(XXX(I)/DT,YY1(I)/DY,2)
CALL PLOT(XXX(I)/DT,YY2(1)/DY,3)
30 DO 30 I=2,N
CALL PLOT(XXX(I)/DT,YY2(I)/DY,2)
CALL PLOT(-0.5,-6.5,3)
CALL PLOT(-0.2,-6.5,2)
CALL NEWPEN(1)
CALL PLOT(-0.5,-7.0,3)
CALL PLOT(-0.2,-7.0,2)
CALL PLOT(0.0,0.0,3)
CALL SYMBOL(-0.15,-6.5,0.15,'POLES',0.0,5)
CALL SYMBOL(-0.15,-7.0,0.15,'ZERU',0.0,4)
CALL SYMBOL(-0.5,-7.5,0.15,'PLANT: Y(K)=(A1+A2)Y(K-1)-(A1+A2)Y(K-
*2)+(B+E)*U(K-1)-(A1+E+B*A2)*U(K-2)',0.0,71)
CALL SYMBOL(-0.5,-8.0,0.15,'CONTROL: U(K)=F1*R(K)+F2*Y(K)',0.0,29
*)
CALL SYMBOL(-0.5,-8.5,0.15,'RESULTING CHAR. EQN.: Z**2-Z*(A1+A2+F
*2)+(B+E)+A1*A2+F*(B*A2+E*A1)',0.0,64)
CALL SYMBOL(-0.5,-9.0,0.15,'PARAMETERS: A1=0.95 A2=0.2 B=0.0
*5 E=0.01',0.0,44)
CALL PLOT(18.0,0.0,-3)
CALL PLOT(0.0,0.0,+999)
STOP
END

```


**Appendix B: Supporting Algebra for (3-31) - (3-43) and Simulated
Check**

Contents:

p.	B-1-3:	Supporting algebra
	B-4-5:	Step response test program
	B-6 :	Step response match

From (3-31)

$$\frac{Y(z)}{R(z)} = \frac{\delta z^2 [(\beta_{11}z + \beta_{12})(z^2 - \alpha_{21}z - \alpha_{22}) + (\beta_{21}z + \beta_{22})(z^2 - \alpha_{11}z - \alpha_{12})]}{\left[(z^2 - \eta_1z - \eta_2)(z^2 - \alpha_{11}z - \alpha_{12})(z^2 - \alpha_{21}z - \alpha_{22}) - (v_1z + v_2)[(\beta_{11}z + \beta_{12})(z^2 - \alpha_{21}z - \alpha_{22}) + (\beta_{21}z + \beta_{22})(z^2 - \alpha_{11}z - \alpha_{12})] \right]}$$

$$= \frac{b_5z^5 + b_4z^4 + b_3z^3 + b_2z^2 + b_1z + b_0}{a_6z^6 + a_5z^5 + a_4z^4 + a_3z^3 + a_2z^2 + a_1z + a_0} = \frac{N(z)}{D(z)} \quad (B-1)$$

Evaluating the numerator

$$\begin{aligned} N(z) &= \delta z^2 [\beta_{11}z^3 - \beta_{11}\alpha_{21}z^2 - \beta_{11}\alpha_{22}z + \beta_{12}z^2 - \beta_{12}\alpha_{21}z \\ &\quad - \beta_{12}\alpha_{22} + \beta_{21}z^3 - \beta_{21}\alpha_{11}z^2 - \beta_{21}\alpha_{12}z \\ &\quad + \beta_{22}z^2 - \beta_{22}\alpha_{11}z - \beta_{22}\alpha_{12}] \\ &= [\delta(\beta_{11} + \beta_{21})]z^5 + [\delta(-\beta_{11}\alpha_{21} + \beta_{12} - \beta_{21}\alpha_{11} + \beta_{22})]z^4 \\ &\quad + [\delta(-\beta_{11}\alpha_{22} - \beta_{12}\alpha_{21} - \beta_{21}\alpha_{12} - \beta_{22}\alpha_{11})]z^3 \\ &\quad + [\delta(-\beta_{12}\alpha_{22} - \beta_{22}\alpha_{12})]z^2 \end{aligned} \quad (B-2)$$

yields (3-32) - (3-36). Define the denominator as

$$D(z) \stackrel{\Delta}{=} D_1(z) - D_2(z) \quad (B-3)$$

where

$$D_1(z) = (z^2 - \eta_1z - \eta_2)(z^2 - \alpha_{11}z - \alpha_{12})(z^2 - \alpha_{21}z - \alpha_{22}) \quad (B-4)$$

and

$$D_2(z) = (v_1z + v_2) \frac{N(z)}{\delta z^2} \quad (B-5)$$

Expanding $D_1(z)$ yields

$$\begin{aligned}
 D_1(z) &= (z^2 - \eta_1 z - \eta_2)(z^4 - \alpha_{21} z^3 - \alpha_{22} z^2 - \alpha_{11} z + \alpha_{11} \alpha_{21} z^2 \\
 &\quad + \alpha_{11} \alpha_{22} z^2 - \alpha_{12} z^2 + \alpha_{12} \alpha_{21} z + \alpha_{12} \alpha_{22}) \\
 &= (z^2 - \eta_1 z - \eta_2)[z^4 - (\alpha_{21} + \alpha_{11})z^3 - (\alpha_{22} - \alpha_{11} \alpha_{21} + \alpha_{12})z^2 \\
 &\quad + (\alpha_{11} \alpha_{22} + \alpha_{12} \alpha_{21})z + \alpha_{12} \alpha_{22}] \\
 &= z^6 - (\alpha_{21} + \alpha_{11})z^5 - (\alpha_{22} - \alpha_{11} \alpha_{21} + \alpha_{12})z^4 + (\alpha_{11} \alpha_{22} + \alpha_{12} \alpha_{21})z^3 \\
 &\quad + \alpha_{12} \alpha_{22} z^2 - \eta_1 z^5 + \eta_1 (\alpha_{21} + \alpha_{11})z^4 + \eta_1 (\alpha_{22} - \alpha_{11} \alpha_{21} + \alpha_{12})z^3 \\
 &\quad - \eta_1 (\alpha_{11} \alpha_{22} + \alpha_{12} \alpha_{21})z^2 - \eta_1 \alpha_{12} \alpha_{22} z - \eta_2 z^4 + \eta_2 (\alpha_{21} + \alpha_{11})z^3 \\
 &\quad + \eta_2 (\alpha_{22} - \alpha_{11} \alpha_{21} + \alpha_{12})z^2 - \eta_2 (\alpha_{11} \alpha_{22} + \alpha_{12} \alpha_{21})z - \eta_2 \alpha_{12} \alpha_{22} \\
 &= z^6 + (-\alpha_{21} - \alpha_{11} - \eta_1)z^5 + (-\alpha_{22} + \alpha_{11} \alpha_{21} - \alpha_{12} + \eta_1 \alpha_{21} + \eta_1 \alpha_{11} - \eta_2)z^4 \\
 &\quad + (\alpha_{11} \alpha_{22} + \alpha_{12} \alpha_{21} + \eta_1 \alpha_{22} - \eta_1 \alpha_{11} \alpha_{21} + \eta_1 \alpha_{12} + \eta_2 \alpha_{21} + \eta_2 \alpha_{11})z^3 \\
 &\quad + (\alpha_{12} \alpha_{22} - \eta_1 \alpha_{11} \alpha_{22} - \eta_1 \alpha_{12} \alpha_{21} + \eta_2 \alpha_{22} - \eta_2 \alpha_{11} \alpha_{21} + \eta_2 \alpha_{12})z^2 \\
 &\quad + (-\eta_1 \alpha_{12} \alpha_{22} - \eta_2 \alpha_{11} \alpha_{22} - \eta_2 \alpha_{12} \alpha_{21})z - \eta_2 \alpha_{12} \alpha_{22} .
 \end{aligned} \tag{B-6}$$

Expanding $D_2(z)$ given (B-2) yields

$$\begin{aligned}
 D_2(z) &= (v_1 z + v_2)[(\beta_{11} + \beta_{21})z^3 + (-\beta_{11} \alpha_{21} + \beta_{12} - \beta_{21} \alpha_{11} + \beta_{22})z^2 \\
 &\quad + (-\beta_{11} \alpha_{22} - \beta_{12} \alpha_{21} - \beta_{21} \alpha_{12} - \beta_{22} \alpha_{11})z \\
 &\quad + (-\beta_{12} \alpha_{22} - \beta_{22} \alpha_{12})]
 \end{aligned}$$

$$\begin{aligned}
&= (v_1 \beta_{11} + v_1 \beta_{21})z^4 + (-v_1 \beta_{11} \alpha_{21} + v_1 \beta_{12} - v_1 \beta_{21} \alpha_{11} + v_1 \beta_{22} \\
&\quad + v_2 \beta_{11} + v_2 \beta_{21})z^3 + (-v_1 \beta_{11} \alpha_{22} - v_1 \beta_{12} \alpha_{21} - v_1 \beta_{21} \alpha_{12} \\
&\quad - v_1 \beta_{22} \alpha_{11} - v_2 \beta_{11} \alpha_{21} + v_2 \beta_{12} - v_2 \beta_{21} \alpha_{11} + v_2 \beta_{22})z^2 \\
&\quad + (-v_1 \beta_{12} \alpha_{22} - v_1 \beta_{22} \alpha_{12} - v_2 \beta_{11} \alpha_{22} - v_2 \beta_{12} \alpha_{21} \\
&\quad - v_2 \beta_{21} \alpha_{12} - v_2 \beta_{22} \alpha_{11})z + (-v_2 \beta_{12} \alpha_{22} - v_2 \beta_{22} \alpha_{12}). \tag{B-7}
\end{aligned}$$

subtracting (B-7) from (B-6) as in (B-3) yields (3-37) - (3-43).

For example 5 the unit step response from zero initial conditions of $\frac{Y(z)}{R(z)}$ in (B-1) with the parameter definitions of (3-32) - (3-43) was compared with that of (3-17) using the control law of (3-25). The program listing and outputs for this comparison complete this appendix. Note that the outputs are identical, which verifies (3-32) - (3-43).

```

1 JOB          WATFIV,KP=29,PAGE=100,TIME=100
DIMENSION G1(2),F1(2),ZLT(2),W(2),AL(2),G2(2),F2(2),AU(2),YC(200)
*,YC(200),DY(200)
DATA T/0.5/,AL/1.0,1.0/,E/1.0/,ZET/0.2,0.02/,W/0.5,5.0/ /0.5/,GM1
*/1.687/,GM2/-0.741/,R/1.0/,Y1/0.0/,Y2/0.0/,Y3/0.0/,Y4/0.0/,U1/0.0
/,U2/0.0/,U3/0.0/,U4/0.0/,R1/0.0/,R2/0.0/,R3/0.0/,R4/0.0/,R5/0.0/,
*,R6/0.0/
DO 10 I=1,2
G1(I)=2*EXP(-ZET(I)*W(I)*T)*COS(W(I)*T*SQRT(1-ZLT(I)**2))
IF(I.EQ.1)A11=G1(1)
IF(I.EQ.2)A12=G1(2)
G2(I)=-EXP(-2*ZET(I)*W(I)*T)
IF(I.EQ.1)A12=G2(1)
IF(I.EQ.2)A22=G2(2)
F1(I)=(AL(I)/W(I)**2)*(1-EXP(-ZET(I)*W(I)*T))*COS(W(I)*T*SQRT(1-ZE
T(I)**2))+(ZLT(I)/SQRT(1-ZET(I)**2))*SIN(W(I)*T*SQRT(1-ZET(I)**2))
IF(I.EQ.1)B11=F1(1)
IF(I.EQ.2)B12=F1(2)
F2(I)=(AL(I)/W(I)**2)*(EXP(-ZET(I)*W(I)*T)*(EXP(-ZET(I)*W(I)*T)-CO
S(W(I)*T*SQRT(1-ZET(I)**2)))+(ZLT(I)/SQRT(1-ZET(I)**2))*SIN(W(I)*T*
SQRT(1-ZET(I)**2))
IF(I.EQ.1)B12=F2(1)
IF(I.EQ.2)B22=F2(2)
10 CONTINUE
WRITE(6,15)
15 FORMAT(1X,'          A11',10X,'A12',10X,'B11',10X,'B12',10X,'A21',10
*, 'A22',10X,'B21',10X,'B22',10X)
WRITE(6,20)A11,A12,B11,B12,A21,A22,B21,B22
20 FORMAT(//,'F13.6)
E11=GM1-A11
ANU1=(E11*A12+(B11*A11-A12+GM2)*(A11-A12*B11/B12))/(B11*A11+B12-A1
2*B11**2/B12)
E12=A11*E11+GM1-B11*ANU1-A12
ANU2=E12+A12/B12
WRITE(6,25)
25 FORMAT(1X,'          B11',10X,'B12',10X,'NU1',10X,'NU2')
WRITE(6,30)E11,E12,ANU1,ANU2
30 FORMAT(//,'F13.6)
DO 35 J=1,200
Y=(A21+A11)*Y1+(A22-A11*A21+A12)*Y2-(A11*A22+A12*A21)*Y3-(A12*A22)
*,Y4+(B11+E*B21)*U1-(B11*A21-B12+E*B21*A11-E*B22)*U2-(B11*A22+B12*A
21+E*B21*A12+E*B22*A11)*U3-(B12*A22+E*B22*A12)*U4
U=U*R+E11*U1+E12*U2+ANU1*Y1+ANU2*Y2
YC(J)=Y
AU(J)=U
Y4=Y3
Y3=Y2
Y2=Y1
Y1=Y
U4=U3
U3=U2
U2=U1
U1=U
35 CONTINUE
B3=C*(B11+E*B21)
B4=D*(-B11*A21+B12-E*B21*A11+E*B22)
R3=D*(-B11*A22-B12*A21-E*B21*A12-E*B22*A11)
B1=L*(-B12*A22-E*B22*A12)
B1=0.0

```

```

B0=B1
A6=1.0
A5=-A21-A11-ET1
A4=-A22+A11*A21-A12+ET1*A21+ET1*A11-ET2-ANU1*B11-ANU1*E*B21
A3=A11*A22+A12*A21+ET1*A22-LT1*A11*A21+L11*A12+ET2*A21+ET2*A11+ANU
+1*B11*A21-ANU1*B12+ANU1*E*B21*A11-ANU1*E*B22-ANU2*B11-ANU2*E*B21
A2=A17*A22-ET1*A11*A22-ET1*A17*A21+ET1*A22-ET2*A11*A21+ET2*A12+ANU
+1*B11*A22+ANU1*B12*A21+ANU1*E*B21*A12+ANU1*E*B22*A11+ANU2*E*B21*A21-
+ANU2*B12+ANU2*E*B21*A11-ANU2*E*B22
A1=-ET1*A12*A22-ET2*A11*A22-ET2*A12*A21+ANU1*B12*A22+ANU1*E*B22*A1
+2*ANU2*B11*A22+ANU2*B12*A21+ANU2*E*B21*A12+ANU2*E*L22*A11
A0=-ET2*A12*A22+ANU2*E*B21*A22+ANU2*E*B22*A12
Y1=0.0
Y2=0.0
Y3=0.0
Y4=0.0
Y5=0.0
Y6=0.0
DO 40 K=1,200
Y=-A5*Y1-A4*Y2-A3*Y3-A2*Y4-A1*Y5-A0*Y6+B5*R1+B4*R2+B3*R3+B2*R4+B1*
+R5+B0*R6
Y0(K)=Y
Y6=Y5
Y5=Y4
Y4=Y3
Y3=Y2
Y2=Y1
Y1=Y
R6=R5
R5=R4
R4=R3
R3=R2
R2=R1
R1=R
40 CONTINUE
WRITE(6,60)
60 FORMAT(1X,'YC: CONTROLLED SYS',5X,'YU: OVERALL T.F.',12X,'(YC-YU)**
+2')
DO 50 L=1,200
DY(L)=(YC(L)-YU(L))**2
WRITE(6,55)YC(L),YU(L),DY(L)
55 FORMAT(1X,F13.6,10X,F13.6,10X,F13.6)
50 CONTINUE
STOP
END

```

	A11	A12	B11	B12	A21
	1.845669	-0.904637	0.120314	0.116359	-1.523571
	A22	B21	B22		
	-0.904637	0.070016	0.067121		
	DEL	ET1	E12	NU1	NU2
YC:	0.500000	-0.158009	-0.071517	-0.477284	0.556613
	CONTROLLED	SYS	OVERALL T.F	(YC-YU)**2	
	0.000000		0.000000	0.000000	
	0.095165		0.095165	0.000000	
	0.229498		0.229498	0.000000	
	0.449052		0.449052	0.000000	
	0.731993		0.731993	0.000000	
	0.472126		0.472126	0.000000	
	1.269010		1.269011	0.000000	
	1.509712		1.509711	0.000000	
	1.724968		1.724961	0.000000	
	1.940312		1.940311	0.000000	
	2.073265		2.073263	0.000000	
	2.213954		2.213952	0.000000	
	2.305182		2.305181	0.000000	
	2.356454		2.356450	0.000000	
	2.415576		2.415572	0.000000	
	2.417251		2.417256	0.000000	
	2.421103		2.421869	0.000000	
	2.423304		2.423298	0.000000	
	2.392859		2.372852	0.000000	
	2.385548		2.385542	0.000000	
	2.349551		2.349545	0.000000	
	2.325536		2.325580	0.000000	
	2.307442		2.307438	0.000000	
	2.273036		2.273030	0.000000	
	2.263845		2.263839	0.000000	
	2.241444		2.241438	0.000000	
	2.226110		2.226102	0.000000	
	2.222950		2.222944	0.000000	
	2.206086		2.206080	0.000000	
	2.207929		2.207924	0.000000	
	2.203193		2.203189	0.000000	
	2.198542		2.198537	0.000000	
	2.205502		2.205499	0.000000	
	2.199992		2.199988	0.000000	
	2.205641		2.205638	0.000000	
	2.208404		2.208402	0.000000	
	2.206746		2.206744	0.000000	
	2.214703		2.214701	0.000000	
	2.212575		2.212573	0.000000	
	2.216134		2.216133	0.000000	
	2.219666		2.219666	0.000000	
	2.217241		2.217240	0.000000	
	2.222535		2.222535	0.000000	
	2.220882		2.220881	0.000000	
	2.221516		2.221515	0.000000	
	2.224154		2.224151	0.000000	
	2.221069		2.221067	0.000000	
	2.223914		2.223912	0.000000	
	2.222715		2.222713	0.000000	
	2.221739		2.221736	0.000000	
	2.223776		2.223774	0.000000	
	2.221040		2.221037	0.000000	
	2.222480		2.222478	0.000000	
	2.222060		2.222057	0.000000	
	2.220642		2.220639	0.000000	
	2.222396		2.222394	0.000000	
	2.220490		2.220488	0.000000	
	⋮		⋮	⋮	
	2.220440		2.220482	0.000000	
	2.220992		2.220979	0.000000	
	2.220990		2.220953	0.000000	
	2.220991		2.220979	0.000000	
	2.220990		2.220981	0.000000	
	2.220990		2.220982	0.000000	
	2.220989		2.220979	0.000000	

Appendix C: Factorization of $\frac{Y}{R}$ for Example 5

- Contents: p. C-1-2: Factorization program
C-3-7: IMSL routine - ZRPOLY
C-8: Destabilizing λ_2 effect
C-9: Plot routine for Fig. 3-10
C-10: Plot routine for Fig. 3-11


```

C THIS PROGRAM DETERMINES PLANT PAR. AND ZING.
DIMENSION G(12),F1(2),Z(12),W(2),A1(2),G2(2),F2(2)
COMPLEX Z(4),Z1(2)
DATA 1/(.5),A1(2),G(12),Z(12),G2(2),W(2),Z1(2),A1(2)
DO 10 I=1,2
G1(I)=2*EXP(-Z1(I)*W(I)*I)*COS(W(I)*I)*SIN(I*(1-Z1(I)*W(I)))
IF(1.EC.1)A11=G1(I)
IF(1.EC.2)A12=G1(I)
G2(I)=-EXP(-Z1(I)*W(I)*I)
IF(1.EC.1)A12=G2(I)
IF(1.EC.2)A12=G2(I)
F1(I)=(A1(I)/W(I)**2)*(1-EXP(-Z1(I)*W(I)*I))*COS(W(I)*I)*SIN(I*(1-Z1(I)*W(I)))
+I(I)*Z1(I)/SIN(I*(1-Z1(I)*W(I))*I)*SIN(W(I)*I)*SIN(I*(1-Z1(I)*W(I)))
+I)
IF(1.EC.1)G11=F1(I)
IF(1.EC.2)G11=F1(I)
F2(I)=(A1(I)/W(I)**2)*(EXP(-Z1(I)*W(I)*I)*EXP(-Z1(I)*W(I)*I)-1)
+G2(I)*SIN(I*(1-Z1(I)*W(I))*I)+Z1(I)/SIN(I*(1-Z1(I)*W(I))*I)*SIN(W(I)*I)
+SIN(I*(1-Z1(I)*W(I))*I)
IF(1.EC.1)G12=F2(I)
IF(1.EC.2)G12=F2(I)
CONTINUE
10 G11=2*EXP(-Z1(1)*W(1)*I)*COS(W(1)*I)*SIN(I*(1-Z1(1)*W(1)))
G12=-EXP(-Z1(1)*W(1)*I)
G21=G11+G12
G22=-G11+G12
F11=-G11+G21-F12+G21-G11+G21
F12=-G12+G22-F11+G22-F11+G22
A4=1.0
A2=-A21-A11
A22=-A22+A11+A21-A12
A1=A11+A22+I1+I2
A0=A12+A22
WRITE(C,30)G11,G12,G21,G22
60 FORMAT(1X,'G11=',F7.0,'G12=',F7.0,'G21=',F7.0,'G22=',F7.0)
70 WRITE(C,30)A4,A2,A22,A1,A0
70 FORMAT(1X,'A4=',F7.0,'A2=',F7.0,'A22=',F7.0,'A1=',F7.0,'A0=',F7.0)
+I1=GM1-A11
ANU1=(I11*A12+I12*A21-A12+GM2)*(A11-A12+I11/I12)/(I11+I12+I1)
+2*I11*I12/I12
I12=A11*I11+GM1-F11*ANU1-A12
ANU2=I12*A12+I11
WRITE(C,30)I11,I12,ANU1,ANU2,GM1,GM2
80 FORMAT(1X,'I11=',F7.0,'I12=',F7.0,'ANU1=',F7.0,'ANU2=',F7.0)
+I1X,'GM1=',F7.0,'GM2=',F7.0)
90 WRITE(C,30)
90 PRINT(1X,'PLANT PARAMETERS AND ZINGS')
WRITE(C,30)
95 FORMAT(//,'          FIELDS          ZINGS')
WRITE(C,30)
95 FORMAT(//,'          FIELDS          REAL',F7.0,'IMAG',F7.0)
CALL ZCHI(Z,A0,A2,A22,A1,A0)
CALL ZCZ(Z1,W(1),W(2))
DO 10 I=1,2
Z1=REAL(Z(1))
Z1=AIMAG(Z(1))
Z1=REAL(Z1(I))
Z1=AIMAG(Z1(I))
IF(1.EC.4)GO TO 15
WRITE(C,30)Z1,Z1,Z1,Z1
95 FORMAT(//,'Z1',F7.0)
GO TO 10
15 WRITE(C,30)Z1,Z1
95 FORMAT(1X,Z1,F7.0)
95 CONTINUE
STOP
END

```

ORIGINAL PAGE IS OF POOR QUALITY

```

SUBROUTINE MCH1 (Z, A1, A3, A2, A1, A0)
INTEGER NDIG, IEN
REAL P(4)
COMPLEX Z(4)
NREG=4
P(1)=A4
P(2)=A3
P(3)=A2
P(4)=A1
P(5)=A0
CALL ZFFLY (P, NDIG, Z, IEN)
RETURN
END

```

```

SUBROUTINE MCH2 (Z, S1, S2, S3, S4, S5)
INTEGER NDIG, IEN
REAL S(4)
COMPLEX Z(4)
NREG=5
S(1)=S5
S(2)=S4
S(3)=S3
S(4)=S2
S(5)=S1
CALL ZFFLY (S, NDIG, Z, IEN)
RETURN
END

```

IMSL ROUTINE NAME - ZRPOLY

- COMPUTER - IBM/DOUBLE
- LATEST REVISION - JANUARY 1, 1978
- PURPOSE - ZEROS OF A POLYNOMIAL WITH REAL COEFFICIENTS (JENKINS-TRAUB)
- USAGE - CALL ZRPOLY (A,NDEG,Z,IER)
- ARGUMENTS
- A - INPUT REAL VECTOR OF LENGTH NDEG+1 CONTAINING THE COEFFICIENTS IN ORDER OF DECREASING POWERS OF THE VARIABLE.
 - NDEG - INPUT INTEGER DEGREE OF POLYNOMIAL. NDEG MUST BE GREATER THAN 0 AND LESS THAN 101.
 - Z - OUTPUT COMPLEX VECTOR OF LENGTH NDEG CONTAINING THE COMPUTED ROOTS OF THE POLYNOMIAL.
- NOTE - THE ROUTINE TREATS Z AS A REAL VECTOR OF LENGTH 2*NDEG. AN APPROPRIATE EQUIVALENCE STATEMENT MAY BE REQUIRED. SEE DOCUMENT EXAMPLE.
- IER - ERROR PARAMETER. (OUTPUT) TERMINAL ERROR
 - IER=129, INDICATES THAT THE DEGREE OF THE POLYNOMIAL IS GREATER THAN 100 OR LESS THAN 1.
 - IER=130, INDICATES THAT THE LEADING COEFFICIENT IS ZERO.
 - IER=131, INDICATES THAT ZRPOLY FOUND FEWER THAN NDEG ZEROS. IF ONLY M ZEROS ARE FOUND, Z(J), J=M+1, ..., NDEG ARE SET TO POSITIVE MACHINE INFINITY.
- PRECISION/HARDWARE - SINGLE AND DOUBLE/H32
- SINGLE/H36,H48,H60
- REQD. IMSL ROUTINES - UERTST,UGETIO,ZRPQLB,ZRPQLC,ZRPQLD,ZRPQLE,
ZRPQLF,ZRPQLG,ZRPQLH,ZRPQLI
- NOTATION - INFORMATION ON SPECIAL NOTATION AND CONVENTIONS IS AVAILABLE IN THE MANUAL INTRODUCTION OR THROUGH IMSL ROUTINE UHELP
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SUBROUTINE ZRPOLY (A,NDEG,Z,IER)

INTEGER NDEG,IER
DOUBLE PRECISION A(1),Z(1)

SPECIFICATIONS FOR ARGUMENTS

INTEGER N,NN,J,JJ,I,NM1,ICNT,N2,L,NZ,NPI
REAL ETA,RMRE,RINF,REPS,RADIX,RLO,XX,YY,SINR,
1 COSR,RMAX,RMIN,X,SC,XM,FF,DX,DF,BND,XXX,ARE
REAL PT(101)
1 DOUBLE PRECISION TEMP(101),P(101),QP(101),RK(101),QK(101),
SVK(101)
1 DOUBLE PRECISION SR,SI,U,V,RA,RB,C,D,A1,A2,A3,
1 A6,A7,E,F,G,H,SZR,SZI,RLZR,RLZI,
2 T,AA,BB,CC,FACTOR,REPSR1,ZERO,ONE, FN
LOGICAL ZEROK
COMMON /ZRPQLJ/
1

SPECIFICATIONS FOR LOCAL VARIABLES

P,QP,RK,QK,SVK,SR,SI,U,V,RA,RB,C,D,A1,A2,A3,A6,
A7,E,F,G,H,SZR,SZI,RLZR,RLZI,ETA,ARE,RMRE,N,NN

THE FOLLOWING STATEMENTS SET MACHINE
CONSTANTS USED IN VARIOUS PARTS OF
THE PROGRAM. THE MEANING OF THE
FOUR CONSTANTS ARE - REPSR1 THE
MAXIMUM RELATIVE REPRESENTATION
ERROR WHICH CAN BE DESCRIBED AS
THE SMALLEST POSITIVE FLOATING
POINT NUMBER SUCH THAT $1. + \text{REPSR1}$ IS
GREATER THAN 1
RINFP THE LARGEST FLOATING-POINT
NUMBER
REPSP THE SMALLEST POSITIVE
FLOATING-POINT NUMBER IF THE
EXPONENT RANGE DIFFERS IN SINGLE
AND DOUBLE PRECISION THEN REPSP
AND RINFP SHOULD INDICATE THE
SMALLER RANGE
RADIX THE BASE OF THE FLOATING-POINT
NUMBER SYSTEM USED

```
DATA RINFP/271FFFFF/
DATA REPSP/200100000/
DATA RADIX/16.0/
DATA REPSR1/234100000000000/
DATA ZERO/0.000/ONE/1.000/
```

ZRPOLY USES SINGLE PRECISION
CALCULATIONS FOR SCALING, BOUNDS
AND ERROR CALCULATIONS.
FIRST EXECUTABLE STATEMENT

```
IER = 0
IF (NDEG .GT. 100 .OR. NDEG .LT. 1) GO TO 165
ETA = REPSR1
ARE = ETA
RMRE = ETA
RLD = REPSP/ETA
```

INITIALIZATION OF CONSTANTS FOR
SHIFT ROTATION

```
XX = .7071068
YY = -XX
SINR = .9975641
COSR = -.06975647
N = NDEG
NN = N+1
```

ALGORITHM FAILS IF THE LEADING
COEFFICIENT IS ZERO.

```
IF (A(1).NE.ZERO) GO TO 5
IER = 130
GO TO 9000
```

REMOVE THE ZEROS AT THE ORIGIN IF
ANY

```
5 IF (A(NN).NE.ZERO) GO TO 10
J = NDEG-N+1
JJ = J+NDEG
Z(JJ) = ZERO
NN = NN-1
N = N-1
IF (NN.EQ.1) GO TO 9005
GO TO 5
```

MAKE A COPY OF THE COEFFICIENTS

```
10 DO 15 I=1,NN
P(I) = A(I)
15 CONTINUE
```

START THE ALGORITHM FOR ONE ZERO

```
20 IF (N.GT.2) GO TO 30
IF (N.LT.1) GO TO 9005
```

CALCULATE THE FINAL ZERO OR PAIR OF
ZEROS

```
IF (N.EQ.2) GO TO 25
Z(NDEG) = -P(2)/P(1)
Z(NDEG+NDEG) = ZERO
GO TO 145
25 CALL ZRPQL1 (P(1),P(2),P(3),Z(NDEG-1),Z(NDEG+NDEG-1),Z(NDEG),
1 Z(NDEG+NDEG))
GO TO 145
```

FIND LARGEST AND SMALLEST MODULI OF
COEFFICIENTS.

```
30 RMAX = 0.
RMIN = RINFP
DO 35 I=1,NN
X = ABS(SNGL(P(I)))
IF (X.GT.RMAX) RMAX = X
IF (X.NE.0. .AND. X.LT.RMIN) RMIN = X
35 CONTINUE
```

SCALE IF THERE ARE LARGE OR VERY SMALL COEFFICIENTS COMPUTES A SCALE FACTOR TO MULTIPLY THE COEFFICIENTS OF THE POLYNOMIAL. THE SCALING IS DONE TO AVOID OVERFLOW AND TO AVOID UNDETECTED UNDERFLOW INTERFERING WITH THE CONVERGENCE CRITERION. THE FACTOR IS A POWER OF THE BASE

```

SC = RLO/RMIN
IF (SC.GT.1.0) GO TO 40
IF (RMAX.LT.10.) GO TO 55
IF (SC.EQ.0.) SC = REPS*RADIX*RADIX
GO TO 45
40 IF (RINFP/SC.LT.RMAX) GO TO 55
45 L = ALOG(SC)/ALOG(RADIX)+.5
IF (L.EQ.0) GO TO 55
FACTOR = DBLE(RADIX)**L
DO 50 I=1,NN
50 P(I) = FACTOR*P(I)

```

COMPUTE LOWER BOUND ON MODULI OF ZEROS.

```

55 DO 60 I=1,NN
60 PT(I) = ABS(SNGL(P(I)))
PT(NN) = -PT(NN)

```

COMPUTE UPPER ESTIMATE OF BOUND

```

X = EXP((ALOG(-PT(NN))-ALOG(PT(1)))/N)
IF (PT(N).EQ.0.) GO TO 65

```

IF NEWTON STEP AT THE ORIGIN IS BETTER, USE IT.

```

XM = -PT(NN)/PT(N)
IF (XM.LT.X) X = XM

```

CHOP THE INTERVAL (0,X) UNTIL FF.LE.0

```

65 XM = X*.1
FF = PT(1)
DO 70 I=2,NN
70 FF = FF*XM+PT(I)
IF (FF.LE.0.) GO TO 75
X = XM
GO TO 65
75 DX = X

```

DO NEWTON ITERATION UNTIL X CONVERGES TO TWO DECIMAL PLACES

```

80 IF (ABS(DX/X).LE..005) GO TO 90
FF = PT(1)
DF = FF
DO 85 I=2,N
FF = FF*X+PT(I)
DF = DF*X+FF
85 CONTINUE
FF = FF*X+PT(NN)
DX = FF/DF
X = X-DX
GO TO 80
90 BND = X

```

COMPUTE THE DERIVATIVE AS THE INITIAL K POLYNOMIAL AND DO 5 STEPS WITH NO SHIFT

```

NM1 = N-1
FN = ONE/N
DO 95 I=2,N
95 RK(I) = (NN-1)*P(I)*FN
RK(1) = P(1)
AA = P(NN)
BB = P(N)
ZEROK = RK(N).EQ.ZERO
DO 115 JJ=1,5
CC = RK(N)
IF (ZEROK) GO TO 105

```

USE SCALED FORM OF RECURRENCE IF VALUE OF K AT 0 IS NONZERO

```

T = -AA/CC
DO 100 I=1,NM1
J = NN-I
RK(J) = T*RK(J-1)+P(J)
100 CONTINUE
RK(1) = P(1)
ZEROK = DABS(RK(N)).LE.DABS(BB)*ETA*10.
GO TO 115

```

USE UNSCALED FORM OF RECURRENCE

```

C      105  DO 110 I=1,NN1
           J = NN-1
           RK(J) = RK(J-1)
C      110  CONTINUE
           RK(1) = ZERO
           ZEROK = RK(N).EQ.ZERO
C      115  CONTINUE
C      120  DO 120 I=1,N
           TEMP(I) = RK(I)
C      C
C      C
C      C
C      C
C      C
           XXX = COSR*XX-SINR*YY
           YY = SINR*XX+COSR*YY
           XX = XXX
           SR = BND*XX
           SI = BND*YY
           U = -SR-SR
           V = BND*BND
C      C
           CALL ZRPQL6 (20*ICNT,NZ)
           IF (NZ.EQ.0) GO TO 130
C      C
C      C
C      C
           J = NDEG-N+1
           JJ = J+NDEG
           Z(J) = SZK
           Z(JJ) = SZI
           NN = NN-NZ
           N = NN-1
           DO 125 I=1,NN
C      125  P(I) = QP(I)
           IF (NZ.EQ.1) GO TO 20
           Z(J+1) = RLZR
           Z(JJ+1) = RLZI
           GO TO 20
C      C
C      C
           DO 130 I=1,N
C      130  RK(I) = TEMP(I)
C      140  CONTINUE
C      C
           IER = 131
C      C
C      145  DO 150 I=1,NDEG
           NP1 = NDEG+1
           P(1) = Z(NP1)
C      150  CONTINUE
           N2 = NDEG+NDEG
           J = NDEG
           DO 155 I=1,NDEG
           Z(N2-1) = Z(J)
           Z(N2) = P(J)
           N2 = N2-2
           J = J-1
C      155  CONTINUE
           IF (IER.EQ.0) GO TO 9005
C      C
           N2 = 2*(NDEG-NN)+3
           DO 160 I=1,N
           Z(N2) = KINFP
           Z(N2+1) = RINFP
           N2 = N2+2
C      160  CONTINUE
           GO TO 9000
C      165  IER = 129
C      9000  CONTINUE
           CALL UERTST (IER,6HZRPULY)
C      9005  RETURN
           END

```

SAVE K FOR RESTARTS WITH NEW SHIFTS

LOOP TO SELECT THE QUADRATIC
CORRESPONDING TO EACH NEW SHIFTQUADRATIC CORRESPONDS TO A DOUBLE
SHIFT TO A NON-REAL POINT AND ITS
COMPLEX CONJUGATE. THE POINT HAS
MODULUS BND AND AMPLITUDE ROTATED
BY 94 DEGREES FROM THE PREVIOUS
SHIFTSECOND STAGE CALCULATION, FIXED
QUADRATICTHE SECOND STAGE JUMPS DIRECTLY TO
ONE OF THE THIRD STAGE ITERATIONS
AND RETURNS HERE IF SUCCESSFUL.
DEFLATE THE POLYNOMIAL, STORE THE
ZERO OR ZEROS AND RETURN TO THE
MAIN ALGORITHM.IF THE ITERATION IS UNSUCCESSFUL
ANOTHER QUADRATIC IS CHOSEN AFTER
RESTORING KRETURN WITH FAILURE IF NO
CONVERGENCE WITH 20 SHIFTS

CONVERT ZEROS (Z) IN COMPLEX FORM

SET UNFOUND ROOTS TO MACHINE INFINITY

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Name Mohammed M. Alshook Department Electrical Engineering Phone 552-5544 Hou

Computer Center account 1 to be billed for cost of run 10667. 481-5156 affi

The charge for this service is the cost of the run plus \$2.00.

$\lambda_2 = 1: B3 = 0.19025$ $B4 = 1.167805$ $B5 = 1.128077$ $B6 = 0.830095$ $B7 = 0.0$ $B8 = 0.0$
 $A4 = 1.000000$ $A5 = -0.103530$ $A6 = -0.591490$ $A7 = -0.406287$ $A8 = 0.670072$ $A9 = 0.002909$ $A0 = 0.0$
 $E11 = -0.158567$ $E12 = -0.071468$ $E13 = -0.475215$ $E14 = -0.475215$ $E15 = 0.555756$ $E16 = 0.727825$ $E17 = 1.687101$ $E18 = -0.740818$
 PLANT POLES AND ZEROS NU1 = 0.71468 NU2 = 0.475215 NU3 = 0.555756 GM1 = 1.687101 GM2 = -0.740818

POLES

REAL IMAG
 0.922634
 0.922634
 -0.761785
 -0.761785

ZEROS

REAL IMAG
 -0.144123
 -0.144123
 -0.942301
 0.0

$B5 = 0.951647$ $B4 = 1.167805$ $B3 = 1.128077$ $B2 = 0.830095$ $B1 = 0.0$ $B0 = 0.0$
 $A6 = 1.000000$ $A5 = -0.103530$ $A4 = -0.591490$ $A3 = -0.406287$ $A2 = 0.670072$ $A1 = 0.002909$ $A0 = 0.0$
 $E11 = -0.158567$ $E12 = -0.071468$ $E13 = -0.475215$ $E14 = -0.475215$ $E15 = 0.555756$ $E16 = 0.727825$ $E17 = 1.687101$ $E18 = -0.740818$
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

POLES

REAL IMAG
 0.197403
 -0.259023
 -0.729755
 -0.729755
 0.842330
 0.642330

ZEROS

REAL IMAG
 0.0
 0.0
 -0.144123
 -0.144123
 -0.942301
 0.0

$\lambda_2 = 10: B3 = 0.820470$ $B4 = -0.321583$ $B5 = -0.319128$ $B6 = 0.712619$ $B7 = 0.0$ $B8 = 0.0$
 $A4 = 1.000000$ $A5 = -0.322058$ $A6 = -1.002222$ $A7 = -0.251446$ $A8 = 0.810751$ $A9 = 0.0$
 $E11 = -0.158567$ $E12 = -0.071468$ $E13 = -0.475215$ $E14 = -0.475215$ $E15 = 0.555756$ $E16 = 0.727825$ $E17 = 1.687101$ $E18 = -0.740818$
 PLANT POLES AND ZEROS NU1 = 0.71468 NU2 = 0.475215 NU3 = 0.555756 GM1 = 1.687101 GM2 = -0.740818

POLES

REAL IMAG
 0.922634
 0.922634
 -0.761785
 -0.761785

ZEROS

REAL IMAG
 0.670751
 0.670751
 -0.670751
 0.0

$B5 = 4.102349$ $B4 = -1.606913$ $B3 = -1.595765$ $B2 = 3.563096$ $B1 = 0.0$ $B0 = 0.0$
 $A6 = 1.000000$ $A5 = -1.635330$ $A4 = -1.592039$ $A3 = -1.082109$ $A2 = 0.727825$ $A1 = 0.625014$ $A0 = -0.337529$
 $E11 = -0.158567$ $E12 = -0.071468$ $E13 = -0.475215$ $E14 = -0.475215$ $E15 = 0.555756$ $E16 = 0.727825$ $E17 = 1.687101$ $E18 = -0.740818$
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

POLES

REAL IMAG
 0.513095
 -0.717872
 -0.645517
 -0.645517
 0.829670
 0.829670

ZEROS

REAL IMAG
 0.0
 0.0
 0.675252
 0.675252
 -0.670752
 -0.670752

C
C
C

```

UNIT CIRCLE:
THIS PROGRAM  DRAWS A UNIT CIRCLE AND LOCATES POLES & ZEROS OF A
SYSTEM IN IT.
DIMENSION X1(8001),X2(8001),X3(8001),X4(8001),Y1(8001),Y2(8001),Y3
(8001),Y4(8001),XP(4),YP(4),XZ(3),YZ(3),AXZ(3),AYZ(3)
DATA X1(1),X2(1),X3(1),X4(1)/0.0,4.0,0.0,-4.0/
DATA Y1(1),Y2(1),Y3(1),Y4(1)/4.0,0.0,-4.0,0.0/
DATA XP/C.922834,0.522834,-0.761785,-0.761785/
DATA YP/C.230681,-0.230682,0.569665,-0.569665/
DATA XZ/-0.144123,-0.144123,-0.959908/
DATA YZ/C.942301,-0.942301,0.0/
N=8000
YMAX=1.0
YMIN=-1.0
DO 25 I=2,8001
XZ(I)=XZ(I-1)-0.0005
YZ(I)=-ABS(SQRT(16.0-XZ(I)**2))
J=8003-I
X1(J)=X2(I-1)
Y1(J)=-Y2(I-1)
X3(J)=-X2(I-1)
Y3(J)=Y2(I-1)
X4(I)=-X2(I)
Y4(I)=-Y2(I)
25 CONTINUE
CALL PLOTS(0.0,0.0,8001)
CALL PLOT(0.0,0.0,-3)
CALL FACTOR(0.65)
XSTART=-1.0
DT=1/4.0
CALL AXIS (-4.0,0.0,'REAL',-9,8.0,0.0,XSTART,DT)
YT=ABS(YMAX)
IF(ABS(YMIN).GT.YT) YT=ABS(YMIN)
YSTART=-YT
DY=YI/4.0
CALL AXIS (0.0,-4.0,'IMAGINARY',9,8.0,90.0,YSTART,DY)
CALL NEWPEN(3)
CALL PLOT(X1(1),Y1(1),3)
10 DO 20 I=2,N
CALL PLOT(X1(I),Y1(I),2)
CALL PLOT(X2(1),Y2(1),3)
30 DO 30 I=2,N
CALL PLOT(X2(I),Y2(I),2)
CALL PLOT(X3(1),Y3(1),3)
40 DO 40 I=2,N
CALL PLOT(X3(I),Y3(I),2)
CALL PLOT(X4(1),Y4(1),3)
50 DO 50 I=2,N
CALL PLOT(X4(I),Y4(I),2)
DX=0.25
DY=0.25
DO 100 I=1,4
CALL SYMBOL(XP(I)/DX,YP(I)/DY,0.2,4,0.0,-1)
100 CONTINUE
DO 200 J=1,3
AXZ(J)=XZ(J)/DX-0.01515*4.0
AYZ(J)=YZ(J)/DY-0.02727*4.0
CALL SYMBOL(AXZ(J),AYZ(J),0.2,112,0.0,-1)
200 CONTINUE
CALL PLOT(18.0,0.0,-3)
CALL PLOT(0.0,0.0,4999)
STOP
END

```

```

C   UNIT CIRCLE:
C   THIS PROGRAM  DRAWS A UNIT CIRCLE AND LOCATES POLES & ZEROS OF A
C   SYSTEM IN IT.
DIMENSION X1(8001),X2(8001),X3(8001),X4(8001),Y1(8001),Y2(8001),Y3
+ (8001),Y4(8001),XP(5),YP(5),XZ(5),YZ(5),AXZ(5),AYZ(5)
DATA X1(1),X2(1),X3(1),X4(1)/0.0,4.0,0.0,-4.0/
DATA Y1(1),Y2(1),Y3(1),Y4(1)/4.0,0.0,-4.0,0.0/
DATA XP(1),YP(1),XZ(1),YZ(1),AXZ(1),AYZ(1)/0.197403,-0.259023,-0.729755,-0.729755,0.42350,0.842350/
DATA XZ(2),YZ(2),AXZ(2),AYZ(2)/0.600772,-0.600772,0.171181,-0.171181/
DATA XZ(3),YZ(3),AXZ(3),AYZ(3)/0.942301,-0.942301,0.0,0.0/
N=8000
YMAX=1.0
YMIN=-1.0
DO 25 I=2,8001
XZ(I)=XZ(I-1)-0.0005
Y2(I)=-ABS(SQRT(16.0-XZ(I)**2))
J=8003-I
X1(J)=XZ(I-1)
Y1(J)=-Y2(I-1)
X3(J)=-XZ(I-1)
Y3(J)=Y2(I-1)
X4(I)=-XZ(I)
Y4(I)=-Y2(I)
25 CONTINUE
CALL PLOTS(0.0,0.0,5001)
CALL PLOT(5.0,0.0,-3)
CALL FACTOR(0.65)
XSTART=-1.0
DT=1/4.0
CALL AXIS (-4.0,0.0,'REAL',-4.0,0.0,0.0,XSTART,DT)
YT=ABS(YMAX)
IF(ABS(YMIN).GT.YT) YT=ABS(YMIN)
YSTART=-YT
DY=Y1/4.0
CALL AXIS (0.0,-4.0,'IMAGINARY',9.8,0,90.0,YSTART,DY)
CALL NEWPEN(5)
CALL PLOT(X1(1),Y1(1),3)
DO 20 I=2,N
CALL PLOT(X1(I),Y1(I),2)
CALL PLOT(X2(I),Y2(I),3)
DO 30 I=2,N
CALL PLOT(X2(I),Y2(I),2)
CALL PLOT(X3(I),Y3(I),3)
DO 40 I=2,N
CALL PLOT(X3(I),Y3(I),2)
CALL PLOT(X4(I),Y4(I),3)
DO 50 I=2,N
CALL PLOT(X4(I),Y4(I),2)
DX=0.25
DY=0.25
DO 100 I=1,5
CALL SYMBOL(XP(I)/DX,YP(I)/DY,0.2,4,0.0,-1)
100 CONTINUE
DO 200 J=1,5
AXZ(J)=XZ(J)/DX-0.01515*4.0
AYZ(J)=YZ(J)/DY-0.02727*4.0
CALL SYMBOL(AXZ(J),AYZ(J),0.2,112,0.0,-1)
200 CONTINUE
CALL PLOT(16.0,0.0,-3)
CALL PLOT(0.0,0.0,+999)
STOP
END

```

Appendix D: Overall Transfer Functions for Example 6

Contents: p. D-1-2: Supporting algebra

D-3-4: Simulated step response check for (3-48)

D-5-6: Simulated step response check for (3-61)

**Note: a's and b's in printout correspond to p's and q's in (3-48)
and (3-61)**

Consider the control configuration in Fig. 3-12 a (or c) with a convergent identifier, i.e. the time-invariant, asymptotic case. For the plant in (3-17) with the controller in (3-46) and the actuator of (3-45) with

$$g \triangleq e^{-\sigma T} \quad (D-1)$$

then, since

$$U_c(z) = \frac{\delta R(z) + (v_1 z^{-1} + v_2 z^{-2})Y(z)}{(1 - \eta_1 z^{-1} - \eta_2 z^{-2})} \quad (D-2)$$

and

$$\frac{Y(z)}{U_c(z)} = \frac{U(z)}{U_c(z)} \quad \frac{Y(z)}{U(z)}, \quad (D-3)$$

$$\begin{aligned} \frac{Y(z)}{U_c(z)} &= \frac{(1 - \eta_1 z^{-1} - \eta_2 z^{-2}) Y(z)}{\delta R(z) + (v_1 z^{-1} + v_2 z^{-2}) Y(z)} \\ &= \left[\frac{\beta_{11} z + \beta_{12}}{z^2 - \alpha_{11} z - \alpha_{12}} + \frac{\beta_{21} z + \beta_{22}}{z^2 - \alpha_{21} z - \alpha_{22}} \right] \left(\frac{1 - g}{z - g} \right). \end{aligned} \quad (D-4)$$

Cross-multiplying in (D-4) and rearranging yields

$$\begin{aligned} \frac{Y(z)}{R(z)} &= \frac{z^2 \left[(\beta_{11} z + \beta_{12})(z^2 - \alpha_{21} z - \alpha_{22}) + (\beta_{11} z + \beta_{12})(z^2 - \alpha_{11} z - \alpha_{12}) \right] \delta (1 - g)}{\left\{ (z - g)(z^2 - \alpha_{11} z - \alpha_{12})(z^2 - \alpha_{21} z - \alpha_{22})(z^2 - \eta_1 z - \eta_2) \right.} \\ &\quad \left. - (v_1 + v_2)(1 - g) [(\beta_{11} z + \beta_{12})(z^2 - \alpha_{21} z - \alpha_{22}) + (\beta_{11} z + \beta_{12})(z^2 - \alpha_{11} z - \alpha_{12})] \right\}} \\ &= \frac{(1 - g) N(z)}{(z - g) D_1(z) - (1 - g) D_2(z)} = \frac{q_5 z^5 + q_4 z^4 + q_3 z^3 + q_2 z^2 + q_1 z + q_0}{[p_7 z^7 + p_6 z^6 + p_5 z^5 + p_4 z^4 + p_3 z^3 + p_2 z^2} \\ &\quad + p_1 z + p_0], \end{aligned} \quad (D-5)$$

where $N(z)$, $D_1(z)$, and $D_2(z)$ are from (B-2), (B-4), and (B-5), respectively.

Note that the q_i of (D-5) are equal to the $(1-g) b_i$ of (3-32) - (3-36) as shown in (3-49) - (3-53). From (B-6) and (B-7) the p_i of (3-54) - (3-60) are readily formed.

Similarly for the control configuration of Fig. 3-12b after identifier convergence, from moving the pickoff point across the actuator

$$U_c(z) = \delta R(z) + [\eta_1 \left(\frac{1-g}{z-g} \right) z^{-1} + \eta_2 \left(\frac{1-g}{z-g} \right) z^{-2}] U_c(z) + [v_1 z^{-1} + v_2 z^{-2}] Y(z) \quad (D-6)$$

and

$$\frac{Y(z)}{U_c(z)} = \left(\frac{1-g}{z-g} \right) \frac{Y(z)}{U(z)} .$$

Therefore

$$\frac{[z^3 - gz^2 - \eta_1 (1-g) z - \eta_2 (1-g)] Y(z)}{z^2 \delta R(z) + [v_1 z + v_2] Y(z)} = (1-g) \left[\frac{\beta_{11} z + \beta_{12}}{z^2 - \alpha_{11} z - \alpha_{12}} + \frac{\beta_{21} z + \beta_{22}}{z^2 - \alpha_{21} z - \alpha_{22}} \right] . \quad (D-7)$$

Again cross-multiplying and rearranging yields

$$\begin{aligned} \frac{Y(z)}{R(z)} &= \frac{\left\{ z^2 \delta (1-g) [(\beta_{11} z + \beta_{12})(z^2 - \alpha_{21} z - \alpha_{22}) + (\beta_{21} z + \beta_{22})(z^2 - \alpha_{11} z - \alpha_{12})] \right\}}{\left\{ [z^3 - gz^2 - \eta_1 (1-g) z - \eta_2 (1-g)] (z^2 - \alpha_{11} z - \alpha_{12}) (z^2 - \alpha_{21} z - \alpha_{22}) \right.} \\ &\quad \left. - [v_1 z + v_2] (1-g) [(\beta_{11} z + \beta_{12})(z^2 - \alpha_{21} z - \alpha_{22}) + (\beta_{21} z + \beta_{22})(z^2 - \alpha_{11} z - \alpha_{12})] \right\}} \\ &= \frac{(1-g) N(z)}{\left\{ [z^3 - gz^2 - \eta_1 (1-g) z - \eta_2 (1-g)] (z^2 - \alpha_{11} z - \alpha_{12}) (z^2 - \alpha_{21} z - \alpha_{22}) - (1-g) D_2(z) \right\}} \end{aligned} \quad (D-8)$$

where $N(z)$ and $D_2(z)$ are from (B-2) and (B-7), respectively. Expansion yields (3-62) - (3-68).

```

DIMENSION G(2),F(2),ZET(2),W(2),AL(2),G2(2),F2(2)
DATA T/0.5/,AL/1.0,1.0/,ZET/0.2,0.2/,W/0.5,5.0/,U/0.5/,ZETA/0.6/
DATA Y1,Y2,Y3,Y4,Y5,U1,U2,U3,U4,U5/0.0,0.0,0.0,0.0,0.0,0.0,0.0,0.0,0.0/
*,0.0,0.0/,SIG/1.0/
DATA YC1,YC2,YC3,YC4,YC5,YC6,YC7/0.0,0.0,0.0,0.0,0.0,0.0,0.0/,R1,R
*,K3,K4,K5,K6,K7,R/0.0,0.0,0.0,0.0,0.0,0.0,1.0/
DO 10 I=1,2
G1(1)=2*EXP(-ZET(1)*W(1)*I)*COS(W(1)*I*SQRT(1-ZET(1)**2))
IF(1.EQ.1)A11=G1(1)
IF(1.EQ.2)A21=G1(2)
G2(1)=-EXP(-2*ZET(1)*W(1)*I)
IF(1.EQ.1)A12=G2(1)
IF(1.EQ.2)A22=G2(2)
F1(1)=(AL(1)/W(1)**2)*(1-EXP(-ZET(1)*W(1)*I)*(COS(W(1)*I*SQRT(1-ZE
*T(1)**2))+ZET(1)/SQRT(1-ZET(1)**2))*SIN(W(1)*I*SQRT(1-ZET(1)**2)
*))
IF(1.EQ.1)B11=F1(1)
IF(1.EQ.2)B21=F1(2)
F2(1)=(AL(1)/W(1)**2)*(EXP(-ZET(1)*W(1)*I)*(EXP(-ZET(1)*W(1)*I)-CO
*S(W(1)*I*SQRT(1-ZET(1)**2))+ZET(1)/SQRT(1-ZET(1)**2))*SIN(W(1)*I*
*SQRT(1-ZET(1)**2)))
IF(1.EQ.1)B12=F2(1)
IF(1.EQ.2)B22=F2(2)
10 CONTINUE
GM1=2*EXP(-ZETA*W(1)*I)*COS(W(1)*I*SQRT(1.0-ZETA**2))
GM2=-EXP(-2*ZETA*W(1)*I)
ET1=GM1-A11
ANU1=(ET1*A12+(ET1*A11-A12+GM2)*(A11-A12*B11/B12))/(B11*A11+B12-A1
*2*B11**2/B12)
ET2=A11*ET1+GM2-B11*ANU1-A12
ANU2=ET2*A12/B12
G=EXP(-SIG*I)
B5=D*(1.0-G)*(B11+B21)
B4=D*(1.0-G)*(-B11*A21+B12-B21*A11+B22)
B3=D*(1.0-G)*(-B11*A22-B12*A21-B21*A12-B22*A11)
B2=L*(1.0-G)*(-B12*A21-B22*A12)
E1=0.0
S0=E1
A7=1.0
A6=-G*A21-A11-ET1
A5=(G*(A21+A11*ET1)-A22+A11*A21-A12+ET1*A21+ET1*A11-ET2)
A4=(-G*(-A22+A11*A21-A12+ET1*A21+ET1*A11-ET2)+A11*A22+A12*A21+ET1
+A22-LT1*A11*A21+ET1*A12+ET2*A21+ET2*A11-ANU1*(B11+B21)*(1.0-G))
A3=(G*(A11*A22+A12*A21+ET1*A22-ET1*A11*A21+ET1*A12+ET2*A21+ET2*A1
+1)*A12+A22-LT1*A11*A22-ET1*A12*A21+ET2*A22-ET2*A11*A21+ET2*A12-ANU
+1*(B11*A21+B21*(1.0-G)-ANU2*(B11+B21)*(1.0-G))
A2=(G*(A12*A22-ET1*A11*A22-ET1*A12*A21+ET2*A22-LT2*A11*A21+LT2*A1
+2)-ET1*A12*A22-ET2*A11*A22-ET2*A12*A21+ANU1*(B11*A22+B12*A21+ET1*A
+12+B22*A11)*(1.0-G)-ANU2*(B11*A21+B22*B12-B21*A11)*(1.0-G))
A1=(G*(ET1*A12+A22+LT2*A11*A22+LT2*A12*A21)-LT2*A12*A22+ANU2*(B11*
+A22+B12*A21+B21*A12+B22*A11)*(1.0-G)+ANU1*(B12*A22+B22*A12)*(1.0-
+G))
A0=G*LT2*A12+A22+ANU2*(B12*A22+B22*A12)*(1.0-G)
60 WRITE(6,0)B5,B4,B3,B2,B1,B0
FORMAT(1X,'B5=',F9.6,2X,'B4=',F9.6,2X,'B3=',F9.6,2X,'B2=',F9.6,2X,
,'B1=',F9.6,2X,'B0=',F9.6,2X)
70 WRITE(6,70)A7,A6,A5,A4,A3,A2,A1,A0
FORMAT(1X,'A7=',F9.6,2X,'A6=',F9.6,2X,'A5=',F9.6,2X,'A4=',F9.6,2X,
,'A3=',F9.6,2X,'A2=',F9.6,2X,'A1=',F9.6,2X,'A0=',F9.6,2X)
30 WRITE(6,30)ET1,ET2,ANU1,ANU2,GM1,GM2
FORMAT(1X,'ET1=',F9.6,2X,'ET2=',F9.6,2X,'NU1=',F9.6,2X,'NU2=',F9.6
,2X,'GM1=',F9.6,2X,'GM2=',F9.6,2X)
100 WRITE(6,100)
FORMAT(1X,'Y1:CONTROLLED SYS.',3X,'Y2:OVERALL T.F.',2X,'(Y1-Y2)**2
*)
DO 35 J=1,100
Y=(G+A21+A11)*Y1-(G*(A21+A11)-A22+A11*A21-A12)*Y2-(A11*A22+A12*A21
-G*(-A22+A11*A21-A12))*Y3-(A12*A22-G*(A11*A22+A12*A21))*Y4+G*A12*
+A22*Y5+(1.0-G)*(B11+B21)*U2+(1.0-G)*(-B11*A21+B22+B12-B21*A11)*U3+
+1.0-G*(B11*A22-B12*A21-B21*A12-B22*A11)*U4+(1.0-G)*(-B12*A22-B2
+2*A12)*U5
U=D*R+ET1*U1+ET2*U2+ANU1*Y1+ANU2*Y2
Y5=Y4
Y4=Y3
Y3=Y2
Y2=Y1
Y1=Y
U5=U4
U4=U3
U3=U2
U2=U1
U1=U

```

```

YC=-A6*YCL-75*YCL-A4*YCL-A3*YCL-A2*YCL-A1*YCL-A0*YCL7+R5+R2+R4+R3+R
*5+R4+R2+R5+R1+R0+R0+R7
YC7=YC6
YC6=YC5
YC5=YC4
YC4=YC3
YC3=YC2
YC2=YC1
YC1=YC
K7=R6
K6=R5
K5=R4
R4=R3
R3=R2
K2=R1
R1=R
DY=(Y-YC)*+Z
WRITE(6,65)Y,YC,DY
FORMAT(1X,F13.6,9X,F13.3,5X,F13.8,5X)
CONTINUE
STOP
END

```

05
35

B5= 0.037444 B4= 0.040730 B3= 0.044380 B2= 0.032002
E1= 0.000000 E0= 0.000000
A7= 1.000000 A6=-0.770001 A5=-0.862752 A4= 0.157701
A3= 0.970814 A2=-0.525374 A1=-0.025889 A0=-0.011794
GM1= 1.007101 GM2=-0.740318
ET1=-0.156567 ET2=-0.071400 NU1=-0.472215 NU2= 0.555756

Y1: CONTROLLED SYS.	Y2: OVERALL T.F.	(Y1-Y2)**2
0.00000000	0.00000000	0.00000000
0.00000000	0.00000000	0.00000000
0.03744444	0.03744444	0.00000000
0.11501550	0.11501550	0.00000000
0.24665000	0.24665000	0.00000000
0.44656180	0.44656180	0.00000000
0.66967510	0.66967480	0.00000000
0.93212080	0.93212010	0.00000000
1.19115400	1.19115300	0.00000000
1.43985000	1.43984900	0.00000000
1.67098900	1.67098600	0.00000000
1.85116400	1.85118200	0.00000000
2.01068400	2.01068200	0.00000000
2.11737600	2.11737500	0.00000000
2.18552000	2.18552100	0.00000000
2.22514800	2.22515000	0.00000000
2.22875500	2.22875900	0.00000000
2.22224700	2.22230100	0.00000000
2.19892500	2.19892700	0.00000000
2.17125900	2.17126400	0.00000000
2.14916500	2.14916300	0.00000000
2.12490200	2.12490600	0.00000000
2.11508500	2.11508700	0.00000000
2.10947700	2.10947900	0.00000000
2.11212800	2.11213200	0.00000000
2.12521500	2.12521700	0.00000000
2.13777700	2.13778300	0.00000000
2.15871900	2.15873500	0.00000000
2.17743700	2.17744600	0.00000000
2.19540500	2.19541600	0.00000000
2.21593500	2.21394600	0.00000000
2.22497800	2.22499100	0.00000000
2.23651800	2.23657200	0.00000000
2.24204000	2.24205400	0.00000000
2.24422500	2.24424000	0.00000000
2.24582600	2.24584000	0.00000000
2.24201600	2.24203100	0.00000000
2.24011100	2.24012200	0.00000000
⋮	⋮	⋮
2.23283500	2.23284400	0.00000000
2.23286700	2.23289600	0.00000000
2.23286700	2.23287500	0.00000000
2.23283200	2.23284200	0.00000000

```

DIMENSION J1(2),F1(2),Z1(2),W1(2),A1(2),G2(2),F2(2)
DATA T/0.5/,AL/1.0/,L/0.7/,Z1/0.2/,W/0.5/,S/0.5/,ZETA/0.6/
DATA Y1,Y2,Y3,Y4,Y5,U1,U2,U3,U4,U5/0.0,0.0,0.0,0.0,0.0,0.0,0.0,0.0,0.0,0.0,0.0/
DATA YC1,YC2,YC3,YC4,YC5,YC6,YC7/0.0,0.0,0.0,0.0,0.0,0.0,0.0/R1,R
Z,K3,K4,K5,R0,K7,K/0.0,0.0,0.0,0.0,0.0,0.0,1.0/
DO 10 I=1,2
G1(I)=2*EXP(-Z1(I)*W1(I)*T)*COS(W1(I)*T*SQRT(1-Z1(I)**2))
IF(I.EQ.1)A11=G1(1)
IF(I.EQ.2)A21=G1(2)
G2(I)=-EXP(-2*Z1(I)*W1(I)*T)
IF(I.EQ.1)A12=G2(1)
IF(I.EQ.2)A22=G2(2)
F1(I)=(A1(I)/W1(I)**2)*(1-EXP(-Z1(I)*W1(I)*T))*(COS(W1(I)*T*SQRT(1-Z1(I)**2))+(Z1(I)/SQRT(1-Z1(I)**2))*SIN(W1(I)*T*SQRT(1-Z1(I)**2)))
F2(I)=(A1(I)/W1(I)**2)*(EXP(-Z1(I)*W1(I)*T)*(EXP(-Z1(I)*W1(I)*T)-COS(W1(I)*T*SQRT(1-Z1(I)**2))+(Z1(I)/SQRT(1-Z1(I)**2))*SIN(W1(I)*T*SQRT(1-Z1(I)**2)))
DO 20 I=1,2
CONTINUE
GM1=EXP(-ZETA*W1(I)*T)*COS(W1(I)*T*SQRT(1.0-ZETA**2))
GM2=-EXP(-2*Z1(I)*W1(I)*T)
ET1=GM1-A11
ANU1=(L11*A12+(L11*A11-A12+GM2)*(A11-A12*B11/B12))/(B11*A11+B12-A12*B11/B12)
ET2=A11*ET1+GM2-B11*ANU1-A12
ANU2=ET2*A12/B12
G=EXP(-SIG*T)
B5=D*(1.0-G)*(B11*B21)
B4=D*(1.0-G)*(-B11*A21+L12-B21*A11+B22)
B3=D*(1.0-G)*(-B11*A22-B12*A21-B21*A12-B22*A11)
B2=D*(1.0-G)*(-B12*A22-B22*A12)
B1=0.0
B0=B1
A7=1.0
A6=-A21-A11+G
A5=-A22+A11+A21-A12+G*(A21+A11)-ET1*(1.0-G)
A4=A11*A22+A12*A21-G*(-A22+A11*A21-A12)+ET1*(1.0-G)*(A21+A11)-ET2*(1.0-G)-ANU1*(1.0-G)*(B11+B21)
A3=A12*A22-G*(A11*A22+A12*A21)-L11*(1.0-G)*(-A22+A11*A21-A12)-ET2*(1.0-G)*(-A21-A11)-ANU1*(1.0-G)*(-B11*A21+B22+B12-B21*A11)-ANU2*(1.0-G)*(B11+B21)
A2=-G*A12+A22-L11*(1.0-G)*(A11*A22+A12*A21)-ET2*(1.0-G)*(-A22+A11*A21-A12)+ANU1*(1.0-G)*(B11*A22+B12*A21+B21*A12+B22*A11)-ANU2*(1.0-G)*(-B11*A21+B22+B12-B21*A11)
A1=-L11*(1.0-G)*A12+A22-ET2*(1.0-G)*(A11*A22+A12*A21)+ANU2*(1.0-G)*(B11*A22+B12*A21+B21*A11)+ANU1*(1.0-G)*(B12*A22+B22*A12)
A0=-L12*(1.0-G)*A12+A22+ANU2*(1.0-G)*(B12*A22+B22*A12)
WRITE(6,60)B5,B4,B3,B2,B1,B0
60  FORMAT(1X,'B5=',F9.6,2X,'B4=',F9.6,2X,'B3=',F9.6,2X,'B2=',F9.6,2X,'B1=',F9.6,2X,'B0=',F9.6,2X)
WRITE(6,70)A7,A6,A5,A4,A3,A2,A1,A0
70  FORMAT(1X,'A7=',F9.6,2X,'A6=',F9.6,2X,'A5=',F9.6,2X,'A4=',F9.6,2X,'A3=',F9.6,2X,'A2=',F9.6,2X,'A1=',F9.6,2X,'A0=',F9.6,2X)
WRITE(6,80)ET1,ET2,ANU1,ANU2,GM1,GM2
80  FORMAT(1X,'ET1=',F9.6,2X,'ET2=',F9.6,2X,'NU1=',F9.6,2X,'NU2=',F9.6,2X,'GM1=',F9.6,2X,'GM2=',F9.6,2X)
WRITE(6,100)
100  FORMAT(1X,'Y1:CONTROLLED SYS.',3X,'Y2:OVERALL T.F.',3X,'(Y1-Y2)**2')
DO 35 J=1,100
Y=(G+A21+A11)*Y1-(G*(A21+A11)-A22+A11*A21-A12)*Y2-(A11*A22+A12*A21-G*(A11*A22+A12*A21))*Y3-(A12*A22-G*(A11*A22+A12*A21))*Y4+G*A12*A22*Y5+(1.0-G)*(B11+B21)*U2+(1.0-G)*(-B11*A21+B22+B12-B21*A11)*U3+(1.0-G)*(-B11*A22-B12*A21-B21*A12-B22*A11)*U4+(1.0-G)*(-B12*A22-B22*A12)*U5
U=U1+G*(U2+ANU1*Y1-ANU1*G*Y2+ANU2*Y2-ANU2*G*Y3+G*U1+ET1*(1.0-G)*U2+ET2*(1.0-G)*U3)
Y2=Y4
Y4=Y3
Y3=Y2
Y2=Y1
Y1=Y
U5=U4
U4=U3
U3=U2
U2=U1
U1=U

```


YC=A6*YC1-A5*YC2-A4*YC3-A3*YC4-A2*YC5-A1*YC6-A0*YC7+B5*R2+B4*R3+B

3*R4+B2*R5+B1*K6+BU*R7
YC7=YL6
YC6=YL5
YC5=YL4
YC4=YL3
YC3=YL2
YC2=YL1
YC1=YL

R7=R6
R6=R5
R5=R4
R4=R3
R3=R2
R2=R1
R1=R

UY=(Y-YC)**2
WRITE(6,65)Y, YC, UY
FORMAT(1X, F13.0, 9X, F13.0, 5X, F13.0, 5X)
CONTINUE
STOP
END

65
55

B5= 0.037444 B4= 0.046736 B3= 0.044366 B2= 0.032662
B1= 0.000000 B0= 0.000000
A7= 1.000000 A6=-0.928628 A5=-0.744578 A4= 0.360112
A3= 0.926707 A2=-0.552717 A1= 0.024593 A0=-0.013281
GM1= 1.687101 GM2=-0.740818
E11=-0.158567 E12=-0.071466 NU1=-0.475215 NU2= 0.555756

Y1:CONTROLLED SYS.	Y2:OVERALL T.F.	(Y1-Y2)**2
0.00000000	0.00000000	0.00000000
0.00000000	0.00000000	0.00000000
0.03744444	0.03744444	0.00000000
0.11895280	0.11895280	0.00000000
0.26691040	0.26691040	0.00000000
0.48417500	0.48417480	0.00000000
0.73204700	0.73204680	0.00000000
1.01587800	1.01587700	0.00000000
1.29278600	1.29278500	0.00000000
1.55093600	1.55093200	0.00000000
1.78245800	1.78245000	0.00000000
1.96054800	1.96053200	0.00000000
2.10239800	2.10238100	0.00000000
2.19349000	2.19346800	0.00000000
2.24466100	2.24463400	0.00000000
2.26917900	2.26915200	0.00000000
2.26195600	2.26192700	0.00000000
2.24776500	2.24773500	0.00000000
2.22218400	2.22215100	0.00000000
2.19566700	2.19563200	0.00000000
2.17730100	2.17726600	0.00000000
2.15893200	2.15891300	0.00000000
2.15415200	2.15411300	0.00000000
2.15333200	2.15329200	0.00000000
2.15841000	2.15836800	0.00000000
2.17140600	2.17136700	0.00000000
:	:	:
2.25267900	2.25284000	0.00000000
2.23287500	2.23283600	0.00000000
2.23281900	2.23277600	0.00000000
2.23291500	2.23287200	0.00000000
2.23282200	2.23277700	0.00000000
2.23287100	2.23283200	0.00000000
2.23267500	2.23282900	0.00000000
2.23282500	2.23278200	0.00000000
2.23289800	2.23285600	0.00000000
2.23282400	2.23278600	0.00000000
2.23286700	2.23282900	0.00000000
2.23286500	2.23283000	0.00000000
2.23282800	2.23279300	0.00000000

**Appendix E: Factorization of (3-48) and (3-61) for various λ_2 as in
Table 3-1**

- Contents:**
- p.E-1-2: Sample program for (3-48) for $\lambda_2 = 1$**
 - E-3-4: Coefficients of (3-49)-(3-60) and singularities
of (3-48) for $\lambda_2 = 1, 10, 20,$ and 50**
 - E-5-6: Sample program for (3-61) for $\lambda_2 = 1$**
 - E-7-8: Coefficients of (3-62)-(3-68) and singularities
of (3-61) for $\lambda_2 = 1, 10, 20,$ and 50**

Note: a's and b's correspond to p's and q's in (3-48) and (3-61).

```

C THIS PROGRAM DETERMINES CONTROLLED SYSTEM PARS. AND SINGS.
  DIMENSION G1(2),F1(2),ZET(2),W(2),AL(2),G2(2),F2(2)
  COMPLEX Z(7),Z1(5)
  DATA T/0.5/,AL/1.0,1.0/,ZL/0.2,0.02/,W/0.5,*.0/,D/0.5/,ZETA/0.6/
  *,SIG/1.0/,PI/3.1415926/
  DO 10 I=1,2
    G1(I)=Z*EXP(-ZET(I)*W(I)*T)*COS(W(I)*T*SQRT(1-ZET(I)**2))
    IF(1.EQ.1)A11=G1(1)
    IF(1.EQ.2)A21=G1(2)
    G2(I)=-EXP(-2*ZL(I)*W(I)*T)
    IF(1.EQ.1)A12=G2(1)
    IF(1.EQ.2)A22=G2(2)
    F1(I)=(AL(I)/W(I)**2)*(1-EXP(-ZET(I)*W(I)*T)*(COS(W(I)*T*SQRT(1-ZL
  *(I)**2))+ZET(I)/SQRT(1-ZL(I)**2))*SIN(W(I)*T*SQRT(1-ZET(I)**2))
    IF(1.EQ.1)B11=F1(1)
    IF(1.EQ.2)B21=F1(2)
    F2(I)=(AL(I)/W(I)**2)*(EXP(-2*ZL(I)*W(I)*T)*(EXP(-ZET(I)*W(I)*T)-CU
  *SIN(W(I)*T*SQRT(1-ZET(I)**2))+ZL(I)/SQRT(1-ZET(I)**2))*SIN(W(I)*T*
  *SQRT(1-ZET(I)**2)))
    IF(1.EQ.1)B12=F2(1)
    IF(1.EQ.2)B22=F2(2)
10 CONTINUE
  GM1=Z*EXP(-ZETA*W(1)*T)*COS(W(1)*T*SQRT(1.0-ZETA**2))
  GM2=-EXP(-2*ZETA*W(1)*T)
  E1=GM1-A11
  ANU1=(E1*A12+(E1*A11-A12+GM2)*(A11-A12*B11/B12))/(B11*A11+B12-A1
  *2*B11**2/B12)
  E12=A11*E1+GM2-B11*ANU1-A12
  ANU2=E12*A12/B12
  G=EXP(-SIG*T)
  B5=D*(1.0-G)*(B11+B21)
  B4=D*(1.0-G)*(-B11*A21+B12-B21*A11+B22)
  B3=D*(1.0-G)*(-B11*A22-B12*A21+B21*A12-B22*A11)
  B2=D*(1.0-G)*(-B12*A22-B22*A12)
  B1=0.0
  B0=B1
  A7=1.0
  A6=-G*A21-A11-E11
  A5=(G*(A21+A11+E11)-A22+A11*A21-A12+E11*A21+E11*A11-E12)
  A4=(-G*(A22+A11*A21-A12+E11*A21+E11*A11-E12)+A11*A22+A12*A21+E11
  *A22-E11*A11*A21+E11*A12+E12*A21+E12*A11-ANU1*(B11+B21)*(1.0-G))
  A3=(-G*(A11*A22+A12*A21+E11*A22-E11*A11*A21+E11*A12+E12*A21+E12*A1
  *1)+A12*A22-L11*A11*A22-L11*A12*A21+E12*A22-E12*A11*A21+E12*A12-ANU
  *1*(B11*A21+B22+B12-B21*A11)*(1.0-G)-ANU2*(B11+B21)*(1.0-G))
  A2=-G*(A12*A22-E11*A11*A22-E11*A12*A21+E12*A22-E12*A11*A21+E12*A1
  *2)-E11*A12*A22-E12*A11*A22-E12*A12*A21+ANU1*(B11*A22+B12*A21+B21*A
  *12+B22*A11)*(1.0-G)-ANU2*(-B11*A21+B22+B12-B21*A11)*(1.0-G))
  A1=(G*(E11*A12+A22+E12*A11*A22+E12*A12*A21)-E12*A12*A21+ANU2*(B11*
  *A22+B12*A21+B21*A12+B22*A11)*(1.0-G)+ANU1*(B12*A22+B22*A12)*(1.0-
  *G))
  A0=G*E12*A12*A22+ANU2*(B12*A22+B22*A12)*(1.0-G)
60 WRITE(6,60)B5,B4,B3,B2,B1,B0
  FURMAT(1X,'B5=',F9.6,2X,'B4=',F9.6,2X,'B3=',F9.6,2X,'B2=',F9.6,2X,
  *,'B1=',F9.6,2X,'B0=',F9.6,2X)
70 WRITE(6,70)A7,A6,A5,A4,A3,A2,A1,A0
  FURMAT(1X,'A7=',F9.6,2X,'A6=',F9.6,2X,'A5=',F9.6,2X,'A4=',F9.6,2X,
  *,'A3=',F9.6,2X,'A2=',F9.6,2X,'A1=',F9.6,2X,'A0=',F9.6,2X)
  WRITE(6,30)E11,E12,ANU1,ANU2,GM1,GM2
30 FURMAT(1X,'E11=',F9.6,2X,'E12=',F9.6,2X,'NU1=',F9.6,2X,'NU2=',F9.6
  *,2X,'GM1=',F9.6,2X,'GM2=',F9.6,2X)
  WRITE(6,80)
80 FURMAT(1X,'OVERALL TRANSFER FUNCTION PULES AND ZERUS')
  WRITE(6,85)
85 FURMAT(7,20X,'***PULES***',40X,'***ZERUS***')
  WRITE(6,90)
90 FURMAT(7,9X,'REAL',9X,'IMAG',9X,'MAG.',9X,'ANG.',9X,'REAL',9X
  *,'IMAG',9X,'MAG.',9X,'ANG.')
  CALL MUH1(Z,A7,A6,A5,A4,A3,A2,A1,A0)
  CALL MUH2(Z1,B5,B4,B3,B2,B1,B0)
  DO 60 I=1,7
    ZK=REAL(Z(I))
    ZI=AIMAG(Z(I))
    ZMAG=SQRT(ZK**2+ZI**2)
    IF(ZI.NE.0.0.AND.ZK.NE.0.0)GO TO 12
    IF(ZI.EQ.0.0.AND.ZK.EQ.0.0)ZANG=0.0
    IF(ZI.EQ.0.0.AND.ZK.GT.0.0)ZANG=0.0
    IF(ZI.EQ.0.0.AND.ZK.LT.0.0)ZANG=180.0

```

```

IF(ZR.EQ.0.0.AND.ZI.GT.0.0)ZANG=90.0
IF(ZR.EQ.0.0.AND.ZI.LT.0.0)ZANG=270.0
GO TO 13
12 IF(ZR.LT.0.0)ZANG=(ATAN(ZI/ZR))*(180.0/PI)+180.0
13 IF(ZR.GT.0.0)ZANG=(ATAN(ZI/ZR))*(180.0/PI)
IF(1.GE.0)GO TO 15
Z11=AIMAG(Z1(1))
ZR1=REAL(Z1(1))
ZMAG1=SQRT(ZR1**2+Z11**2)
IF(Z11.NE.0.0.AND.ZR1.NE.0.0)GO TO 14
IF(Z11.LE.0.0.AND.ZR1.EQ.0.0)ZANG1=0.0
IF(Z11.LE.0.0.AND.ZR1.GT.0.0)ZANG1=0.0
IF(Z11.EQ.0.0.AND.ZR1.LT.0.0)ZANG1=180.0
IF(ZR1.EQ.0.0.AND.Z11.GT.0.0)ZANG1=90.0
IF(ZR1.EQ.0.0.AND.Z11.LT.0.0)ZANG1=270.0
GO TO 17
14 IF(ZR1.LT.0.0)ZANG1=(ATAN(Z11/ZR1))*(180.0/PI)+180.0
17 IF(ZR1.GT.0.0)ZANG1=(ATAN(Z11/ZR1))*(180.0/PI)
17 WRITE(6,65)ZR,ZI,ZMAG,ZANG,ZR1,Z11,ZMAG1,ZANG1
65 FORMAT(1X,6F13.6)
GO TO 66
15 WRITE(6,50)ZR,ZI,ZMAG,ZANG
50 FORMAT(1X,4F13.6)
66 CONTINUE
STOP
END

```

```

SUBROUTINE MUH1 (Z,A7,A6,A5,A4,A3,A2,A1,A0)
INTEGER NDEG,IER
REAL P(8)
COMPLEX Z(7)
NDEG=7
P(1)=A7
P(2)=A6
P(3)=A5
P(4)=A4
P(5)=A3
P(6)=A2
P(7)=A1
P(8)=A0
CALL ZRPOLY (P,NDEG,Z,IER)
RETURN
END

```

```

SUBROUTINE MOH2 (Z1,B5,B4,B3,B2,B1,B0)
INTEGER NDEG,IER
REAL Q(6)
COMPLEX Z1(5)
NDEG=5
Q(1)=B5
Q(2)=B4
Q(3)=B3
Q(4)=B2
Q(5)=B1
Q(6)=B0
CALL ZRPOLY (Q,NDEG,Z1,IER)
RETURN
END

```

B5= 0.037444 B4= 0.046736 B3= 0.044386 B2= 0.032662 B1= 0.0 BC= 0.0
 A7= 1.000000 A6=-0.770001 A5=-0.882752 A4= 0.157701 A3= 0.990814 A2=-0.325873 A1=-0.025869
 E11=-0.158507 E12=-0.071468 NU1=-0.475215 NU2= 0.555750 GM1= 1.687101 GM2=-0.740818 AO=-0.071794
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 1$

ZEROS

REAL	IMAG	MAG.	ANG.	REAL	IMAG	MAG.	ANG.
-0.110420	0.523645	0.543947	109.784531	0.0	0.0	0.0	0.0
-0.116420	0.523645	0.543947	250.215469	0.0	0.0	0.0	0.0
-0.767525	0.559482	0.949797	143.910065	-0.144124	0.942301	0.953259	98.695969
-0.767525	0.559482	0.949797	216.089955	-0.144124	-0.942301	0.953259	261.303955
0.844069	0.281373	0.889732	-16.435928	-0.955908	0.0	0.955908	180.000000
0.849812	0.0	0.849812	-0.0	0.0	0.0	0.0	0.0

POLES

B5= 0.161415 B4=-0.653227 B3=-0.662766 B2= 0.140197 B1= 0.0 B0= 0.0
 A7= 1.000000 A6=-0.770061 A5=-0.882752 A4= 0.275586 A3= 0.748507 A2=-0.305510 A1= 0.195442
 E11=-0.128507 E12=-0.071468 NU1=-0.475215 NU2= 0.555750 GM1= 1.687101 GM2=-0.740818 AC=-0.191321
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 10$

ZEROS

REAL	IMAG	MAG.	ANG.	REAL	IMAG	MAG.	ANG.
-0.089880	0.539954	0.544457	97.374191	0.0	0.0	0.0	0.0
-0.089880	0.539954	0.544457	262.625732	0.0	0.0	0.0	0.0
-0.128908	0.492095	0.564025	149.502654	0.670752	0.0	0.951774	44.808441
-0.128908	0.492095	0.564025	210.694366	0.670752	-0.0	0.951774	315.191559
0.851258	0.279411	0.895922	-16.171951	-0.958796	0.0	0.958796	180.000000
0.851258	0.0	0.851258	-0.0	0.0	0.0	0.0	0.0

POLES

B5= 0.299160 B4=-0.185409 B3=-0.181872 B2= 0.259680 B1= 0.0 B0= 0.0
 A7= 1.000000 A6=-0.770061 A5=-0.682752 A4= 0.406503 A3= 0.479276 A2=-0.282884 A1= 0.441365
 ET1=-0.158567 ET2=-0.071466 NU1=-0.475215 NU2= 0.555756 GM1= 1.687101 GM2=-0.740818 A0=-0.324128
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 20$

ZEROS

REAL	IMAG	MAG.	ANG.	REAL	IMAG	MAG.	ANG.
-0.022221	0.675100	0.675471	91.885239	0.0	0.0	0.0	0.0
-0.022221	-0.675100	0.675471	268.114746	0.0	0.0	0.0	0.0
-0.685574	0.449986	0.597267	153.178058	0.789246	0.531500	0.951526	33.957397
-0.689974	-0.449986	0.597267	206.821945	0.789246	-0.531500	0.951526	-33.957397
0.858554	0.277574	0.902310	-17.916214	-0.958726	0.0	0.958726	160.000000
0.858554	-0.277574	0.902310	0.0				
0.877340	0.0	0.877340					

POLES

B5= 0.712395 B4=-0.551954 B3=-0.539122 B2= 0.616131 B1= 0.0 B0= 0.0
 A7= 1.000000 A6=-0.770061 A5=-0.682752 A4= 0.799254 A3=-0.526415 A2=-0.215006 A1= 1.179135
 ET1=-0.158567 ET2=-0.071466 NU1=-0.475215 NU2= 0.555756 GM1= 1.687101 GM2=-0.740818 A0=-0.722550
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 50$

ZEROS

REAL	IMAG	MAG.	ANG.	REAL	IMAG	MAG.	ANG.
0.906295	0.0	0.906295	0.0	0.0	0.0	0.0	0.0
-1.013580	0.573931	1.080350	159.749924	0.0	0.0	0.0	0.0
-1.013580	-0.573931	1.080350	200.250076	0.866744	0.392230	0.951362	24.348297
0.072635	0.903709	0.903709	65.589847	0.866744	-0.392230	0.951362	-24.348297
0.072635	-0.903709	0.903709	-65.589847	-0.958700	0.0	0.958700	160.000000
0.872828	0.273215	0.914590	-17.381302				
0.872828	-0.273215	0.914590	0.0				

POLES

```

C THIS PROGRAM DETERMINES CONTROLLED SYSTEM PARAMS. AND SINGS.
  DIMENSION G1(2),F1(2),ZET(2),W(2),AL(2),G2(2),F2(2)
  COMPLEX Z(7),Z1(5)
  DATA T/0.5/,AL/1.0,1.0/,ZET/0.2,0.02/,W/0.5,5.0/,U/0.5/,ZETA/0.6/
  +,SIG/1.0/,PI/3.1415926/
  GO TO 1,2
  G1(1)=2*EXP(-ZET(1)*W(1)*T)*COS(W(1)*T)*SURT(1-ZET(1)**2)
  IF(1.EQ.1)A11=G1(1)
  IF(1.EQ.2)A21=G1(2)
  G2(1)=-EXP(-2*ZET(1)*W(1)*T)
  IF(1.EQ.1)A12=G2(1)
  IF(1.EQ.2)A22=G2(2)
  F1(1)=(AL(1)/W(1)**2)*(1-EXP(-ZET(1)*W(1)*T)*(COS(W(1)*T)*SURT(1-ZE
  +T(1)**2))+ZET(1)/SURT(1-ZET(1)**2))*SIN(W(1)*T)*SURT(1-ZET(1)**2)
  +))
  IF(1.EQ.1)B11=F1(1)
  IF(1.EQ.2)B21=F1(2)
  F2(1)=(AL(1)/W(1)**2)*(EXP(-ZET(1)*W(1)*T)*(LXP(-ZET(1)*W(1)*T)-CO
  +S(W(1)*T)*SURT(1-ZET(1)**2))+ZET(1)/SURT(1-ZET(1)**2))*SIN(W(1)*T*
  +SURT(1-ZET(1)**2)))
  IF(1.EQ.1)B12=F2(1)
  IF(1.EQ.2)B22=F2(2)
10 CONTINUE
  GM1=2*EXP(-ZETA*W(1)*T)*COS(W(1)*T)*SURT(1.0-ZETA**2)
  GM2=-EXP(-2*ZETA*W(1)*T)
  ET1=GM1-A11
  ANU1=(ET1*A12+LT1*A11-A12+GM2)*(A11-A12*B11/L12)/(B11*A11+B12-A1
  +2*B11**2/B12)
  ET2=A11*ET1+GM2-B11*ANU1-A12
  ANU2=ET2*A12/B12
  G=EXP(-SIG*T)
  B5=U*(1.0-G)*(B11+B21)
  B4=U*(1.0-G)*(-B11*A21+B12-B21*A11+B22)
  B3=U*(1.0-G)*(-B11*A22-B12*A21-B21*A12-B22*A11)
  B2=U*(1.0-G)*(-B12*A22-B22*A12)
  E1=C.U
  E0=B1
  A7=1.0
  A6=-A21-A11-G
  A5=-A22+A11*A21-A12+G*(A21+A11)-ET1*(1.0-G)
  A4=A11*A22+A12*A21-G*(-A22+A11*A21-A12)+ET1*(1.0-G)*(A21+A11)-ET2*
  +*(1.0-G)-ANU1*(1.0-G)*(B11+B21)
  A3=A12*A22-G*(A11*A22+A12*A21)-LT1*(1.0-G)*(-A22+A11*A21-A12)-ET2*
  +*(1.0-G)*(-A21-A11)-ANU1*(1.0-G)*(-B11*A21+B22+B12-B21*A11)-ANU2*(1
  +.0-G)*(B11+B21)
  A2=-G*A12*A22-LT1*(1.0-G)*(A11*A22+A12*A21)-ET2*(1.0-G)*(-A22+A11*
  +A21-A12)+ANU1*(1.0-G)*(B11*A22+B12*A21+B21*A12+B22*A11)-ANU2*(1.0-
  +G)*(-B11*A21+B22+B12-B21*A11)
  A1=-ET1*(1.0-G)*A12*A22-ET2*(1.0-G)*(A11*A22+A12*A21)+ANU2*(1.0-G)
  +*(B11*A22+B12*A21+B21*A12+B22*A11)+ANU1*(1.0-G)*(B12*A22+B22*A12)
  A0=-ET2*(1.0-G)*A12*A22+ANU2*(1.0-G)*(B12*A22+B22*A12)
60 WRITE(6,60)B5,B4,B3,B2,B1,E0
  FORMAT(1X,'B5=',F9.6,2X,'B4=',F9.6,2X,'B3=',F9.6,2X,'B2=',F9.6,2X,
  +,'B1=',F9.6,2X,'E0=',F9.6,2X)
70 WRITE(6,70)A7,A6,A5,A4,A3,A2,A1,A0
  FORMAT(1X,'A7=',F9.6,2X,'A6=',F9.6,2X,'A5=',F9.6,2X,'A4=',F9.6,2X,
  +,'A3=',F9.6,2X,'A2=',F9.6,2X,'A1=',F9.6,2X,'A0=',F9.6,2X)
80 WRITE(6,80)ET1,ET2,ANU1,ANU2,GM1,GM2
  FORMAT(1X,'ET1=',F9.6,2X,'ET2=',F9.6,2X,'NU1=',F9.6,2X,'NU2=',F9.6
  +,2X,'GM1=',F9.6,2X,'GM2=',F9.6,2X)
90 WRITE(6,90)
  FORMAT(1X,'OVERALL TRANSFER FUNCTION POLES AND ZERUS')
  WRITE(6,95)
65 FORMAT(17,20X,'***POLES***',40X,'***ZERUS***')
  WRITE(6,45)
45 FORMAT(17,20X,'REAL',9X,'IMAG',9X,'MAG.',9X,'ANG.',9X,'REAL',9X
  +,'IMAG',9X,'MAG.',9X,'ANG.')
  CALL MUH1(Z,A7,A6,A5,A4,A3,A2,A1,A0)
  CALL MUH2(Z1,B5,B4,B3,B2,B1,B0)
  UJ=0,1=1,7
  ZR=REAL(Z(1))
  ZI=AIMAG(Z(1))
  ZMAG=SURT(ZR**2+ZI**2)
  IF(ZI.NE.0.0.AND.ZR.NE.0.0)GO TO 12
  IF(ZI.EQ.0.0.AND.ZR.EQ.0.0)ZANG=0.0
  IF(ZI.EQ.0.0.AND.ZR.GT.0.0)ZANG=0.0
  IF(ZI.EQ.0.0.AND.ZR.LT.0.0)ZANG=180.0

```

```

IF(ZR.EQ.0.0.AND.ZI.GT.0.0)ZANG=90.0
IF(ZR.EQ.0.0.AND.ZI.LT.0.0)ZANG=270.0
GO TO 13
12 IF(ZR.LT.0.0)ZANG=(ATAN(ZI/ZR))*(180.0/PI)+180.0
13 IF(ZR.GT.0.0)ZANG=(ATAN(ZI/ZR))*(180.0/PI)
IF(1.GE.6)GO TO 15
Z11=AMAG(Z1(1))
ZR1=REAL(Z1(1))
ZMAG1=SQRT(ZR1**2+Z11**2)
IF(Z11.NE.0.0.AND.ZR1.NE.0.0)GO TO 14
IF(Z11.EQ.0.0.AND.ZR1.EQ.0.0)ZANG1=C.0
IF(Z11.EQ.0.0.AND.ZR1.GT.0.0)ZANG1=0.0
IF(Z11.EQ.0.0.AND.ZR1.LT.0.0)ZANG1=180.0
IF(ZR1.EQ.0.0.AND.Z11.GT.0.0)ZANG1=90.0
IF(ZR1.EQ.0.0.AND.Z11.LT.0.0)ZANG1=270.0
GO TO 17
14 IF(ZR1.LT.0.0)ZANG1=(ATAN(Z11/ZR1))*(180.0/PI)+180.0
IF(ZR1.GT.0.0)ZANG1=(ATAN(Z11/ZR1))*(180.0/PI)
17 WRITE(6,65)ZR,Z1,ZMAG,ZANG,ZR1,Z11,ZMAG1,ZANG1
65 FORMAT(1X,8F13.6)
GO TO 66
15 WRITE(6,50)ZR,Z1,ZMAG,ZANG
50 FORMAT(1X,4F13.6)
66 CONTINUE
STOP
END

```

```

SUBROUTINE MUH1 (Z,A7,A6,A5,A4,A3,A2,A1,A0)
INTEGER NDEG,IER
REAL P(8)
COMPLEX Z(1)
NDEG=7
P(1)=A7
P(2)=A6
P(3)=A5
P(4)=A4
P(5)=A3
P(6)=A2
P(7)=A1
P(8)=A0
CALL ZKPOLY (P,NDEG,Z,IER)
RETURN
END

```

```

SUBROUTINE MUH2 (Z1,B5,B4,B3,B2,B1,B0)
INTEGER NDEG,IER
REAL C(6)
COMPLEX Z1(5)
NDEG=5
C(1)=B5
C(2)=B4
C(3)=B3
C(4)=B2
C(5)=B1
C(6)=B0
CALL ZKPOLY (C,NDEG,Z1,IER)
RETURN
END

```


b5= 0.0037444 b7= 0.046730 b3= 0.044386 b2= 0.032002 b1= 0.0
 A7= 1.000000 A6=-0.928028 A5=-0.744578 A4= 0.360112 A3= 0.926707 A2=-0.522717 A1= 0.024593
 ET1=-0.156567 ET2=-0.071468 NU1=-0.475215 NU2= 0.555756 GM1= 1.667101 GM2=-0.740818 A0=-0.013261
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 1$

PULES

REAL 0.001859
 0.001859
 -0.767098
 -0.767098
 0.819335
 0.819335
 0.820430

IMAG 0.154233
 -0.154233
 0.561032
 -0.561032
 0.280425
 -0.280425
 0.00

MAG. 0.154244
 0.154244
 0.950366
 0.950366
 0.867937
 0.867937
 0.820430

ANG. 69.309494
 -89.309494
 143.819427
 216.180573
 -19.268692
 -19.268692
 0.00

REAL 0.00
 0.00
 -0.144124
 -0.144124
 -0.959908
 -0.959908

IMAG 0.00
 0.00
 0.942301
 -0.942301
 0.00

MAG. 0.00
 0.00
 0.953239
 0.953239
 0.959908
 0.959908

ANG. 0.00
 0.00
 98.655969
 261.303955
 161.000000
 161.000000

ZEROS

b5= 0.101415 b7=-0.003227 b3=-0.002766 b2= 0.140197 b1= 0.0
 A7= 1.000000 A6=-0.928028 A5=-0.744578 A4= 0.360112 A3= 0.926707 A2=-0.522717 A1= 0.024593
 ET1=-0.156567 ET2=-0.071468 NU1=-0.475215 NU2= 0.555756 GM1= 1.667101 GM2=-0.740818 A0=-0.013261
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 10$

PULES

REAL 0.031599
 0.031599
 -0.821860
 -0.821860
 0.834234
 0.834234
 0.040722

IMAG 0.467440
 -0.467440
 0.501240
 -0.501240
 0.283348
 -0.283348
 0.00

MAG. 0.468513
 0.468513
 0.962667
 0.962667
 0.861105
 0.861105
 0.040722

ANG. 66.132690
 -86.132690
 140.644236
 211.377762
 -18.772369
 -18.772369
 0.00

REAL 0.00
 0.00
 0.675232
 0.675232
 -0.952752
 -0.952752

IMAG 0.00
 0.00
 0.670752
 -0.670752
 0.00

MAG. 0.00
 0.00
 0.951774
 0.951774
 0.952752
 0.952752

ANG. 0.00
 0.00
 94.809441
 -94.809441
 161.000000
 161.000000

ZEROS

ORIGINAL PAGE IS OF POOR QUALITY

B2= 0.255160 B4= 0.162409 B3= -0.161872 B2= 0.025080 B1= 0.0 B0= 0.0
 A7= 1.000000 A6= -0.928628 A5= -0.744578 A4= 0.608824 A3= 0.412169 A2= -0.209727 A1= 0.091847
 E11= -0.158267 E12= -0.071468 NU1= -0.475213 NU2= 0.322726 GM1= 1.657101 GM2= -0.740818 AU= -0.265014
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 20$

POLES

ZEROS

REAL	IMAG	MAG.	ANG.	REAL	IMAG	MAG.	ANG.
0.007900	0.025055	0.629329	83.805298	0.00	0.00	0.00	0.00
0.007900	-0.025055	0.629329	-83.805298	0.00	0.00	0.00	0.00
-0.078130	0.401252	0.991940	124.291107	0.551500	0.951526	33.957397	0.00
-0.078130	-0.401252	0.991940	-124.291107	-0.551500	0.951526	-33.957397	0.00
0.042529	0.281092	0.685131	10.429811	0.00	0.958770	180.050000	0.00
0.042529	-0.281092	0.685131	-10.429811	0.00	0.00	0.00	0.00

B2= 0.712550 B4= -0.021924 B3= -0.239122 B2= 0.018131 B1= 0.0 B0= 0.0
 A7= 1.000000 A6= -0.928628 A5= -0.744578 A4= 1.000000 A3= -0.352522 A2= -0.441850 A1= 1.229617
 E11= -0.158267 E12= -0.071468 NU1= -0.475213 NU2= 0.322726 GM1= 1.687101 GM2= -0.740818 AU= -0.664037
 OVERALL TRANSFER FUNCTION POLES AND ZEROS

$\lambda_2 = 50$

POLES

ZEROS

REAL	IMAG	MAG.	ANG.	REAL	IMAG	MAG.	ANG.
0.004092	0.276116	0.907136	17.721008	0.00	0.00	0.00	0.00
0.004092	-0.276116	0.907136	-17.721008	0.00	0.00	0.00	0.00
-0.090985	0.589109	1.0070249	128.077517	0.592230	0.951362	34.346297	0.00
-0.090985	-0.589109	1.0070249	-128.077517	-0.592230	0.951362	-34.346297	0.00
0.148511	0.873011	0.885553	61.0545551	0.00	0.558700	180.050000	0.00
0.148511	-0.873011	0.885553	-61.0545551	0.00	0.00	0.00	0.00

**Appendix F: Sample Programs for the Five Adaptive Controllers
in Section 4 and the Run Statistics in Section 5.**

BLOCK DATA

IMPLICIT REAL*8 (A-H,O-Z)

REAL*8 IDO (10), IO (10), NITE (10), SASI (10), SAPPEA (10),

- * SAPPEB (10), SPPEVA (10), SPPEVB (10), SACPEF (10), SACPEG (10),
- * SCPEVF (10), SCPEVG (10), NASTE (10), NSTEV (10)

INTEGER KD (12)

COMMON /STATS/IDO, IO, NITE, SASI, SAPPEA, SAPPEB,

- * SPPEVA, SPPEVB, SACPEF, SACPEG, SCPEVF, SCPEVG, NASTE, NSTEV,
- * STE, SSPPA, SSPPB, SSCPF, SSCPG, M, KD

DATA KD/0, 1, 2, 5, 10, 20, 50, 100, 200, 500, 1000, 2000/

END

IMPLICIT REAL*8 (A-H,O-Z)

EXTERNAL OVFL

REAL*8 A1(5), A2(5), B(5), E(5), AMP(4), W(4, 5)

REAL*8 C(5), D(5), AHAT(4), BHAT(4), FHAT(4), GHAT(4)

REAL YMIN(4), YMAX(4)

REAL*8 IDO (10), IO (10), NITE (10), SASI (10), SAPPEA (10),

- * SAPPEB (10), SPPEVA (10), SPPEVB (10), SACPEF (10), SACPEG (10),
- * SCPEVF (10), SCPEVG (10), NASTE (10), NSTEV (10)

REAL*8 ESTMUL(3)

INTEGER PCTL, PTEMI, PXCHG, ESTSTR(3)

INTEGER KD (12)

COMMON /STATS/IDO, IO, NITE, SASI, SAPPEA, SAPPEB,

- * SPPEVA, SPPEVB, SACPEF, SACPEG, SCPEVF, SCPEVG, NASTE, NSTEV,
- * STE, SSPPA, SSPPB, SSCPF, SSCPG, M, KD

COMMON /LUN/LP

COMMON /CTRL/F, G, AHAT, BHAT, FHAT, GHAT

DATA K/4/

DATA A1/0.95D0, 0.9D0, 0.9D0, 0.9D0, 0.83D0/

DATA A2/0.2D0, 0.1D0, 0.8D0, 0.7D0, 0.3D0/

DATA B/0.065D0, 0.1D0, 0.08D0, 0.5D0, -0.193D0/

DATA E/0.01D0, -0.01D0, 0.02D0, -0.3D0, 0.208D0/

DATA C/0.4D0, 0.7D0, 1.0D0, 0.8D0, 0.6D0/

DATA D/0.8D0, 0.65D0, 0.5D0, 0.6D0, 0.7D0/

DATA YMIN/-1.0, -4.0, -4.0, -1.0/

DATA YMAX/2.0, 4.0, 4.0, 2.0/

DATA MAX, NSEED/1000, 13579/

DATA AMP/1.0D0, 2.0D0, 2.0D0, 2.0D0/

DATA W(1, 1), W(1, 2), W(1, 3), W(1, 4), W(1, 5)/

- * 0.17851D0, 0.34463D0, 0.55452D0, 0.40866D0, 0.28534D0/

DATA W(2, 1), W(2, 2), W(2, 3), W(2, 4), W(2, 5)/

- * 0.44629D0, 0.86157D0, 1.38629D0, 1.0217D0, 0.71335D0/

DATA W(3, 1), W(3, 2), W(3, 3), W(3, 4), W(3, 5)/

- * 0.17851D0, 0.34463D0, 0.55452D0, 0.40866D0, 0.28534D0/

DATA W(4, 1), W(4, 2), W(4, 3), W(4, 4), W(4, 5)/

- * 0.44629D0, 0.86157D0, 1.38629D0, 1.0217D0, 0.71335D0/

DATA ESTMUL/0.8D0, 1.0D0, 1.2D0/

DATA PCTL, PTEMP/'1', '0'/

DATA ESTSTR/'-20%', '0', '+20%'/

LP = 6

CALL ERNSET(207, 1000, -1, 1, OVFL, 0)

WRITE(6, 10)

10 FORMAT(' PRINTOUT INTERVAL = ')

```

READ (5,*)INTV

WRITE (6,20)
20 FORMAT (' STARTING AND ENDING EXAMPLES = ')
READ (5,*)NEXST,NEXEND
WRITE (6,30)
30 FORMAT (' STARTING AND ENDING ALGORITHMS = ')
READ (5,*)NALGST,NALGED
NSC = 2
DO 1000 NEX=NEXST,NEXEND
DO 100 NALG=NALGST,NALGED
DO 100 NEST=1,3
DO 100 INP=1,4
AHAT (K) = A1 (NEX) * ESTMUL (NEST)
AHAT (K-1) = AHAT (K)
BHAT (K) = B (NEX) * ESTMUL (NEST)
BHAT (K-1) = BHAT (K)
GHAT (K-1) = C (NEX) / B (NEX)
GHAT (K-1) = GHAT (K-1)
GHAT (K-2) = GHAT (K-1)
G = GHAT (K-1)
FHAT (K-1) = (D (NEX) - A1 (NEX)) / B (NEX)
FHAT (K-1) = FHAT (K-1)
FHAT (K-2) = FHAT (K-1)
F = FHAT (K-1)
WRITE (10,50) PCTL,NEX,INP,ESTSTR (NEST)
50 FORMAT (A1,15X,'EXAMPLE',12,' INPUT',12,' NON-ADAPTIVE',
* ' EST = ',A4/)
WRITE (6,50) PCTL,NEX,INP,ESTSTR (NEST)
PXCHG = PCTL
PCTL = PTEMP
PTEMP = PXCHG
CALL SIMUL (A1 (NEX),A2 (NEX),B (NEX),E (NEX),AMP (INP),W (INP,NEX),
* NSEED,INP,NALG,NEX,NSC,C (NEX),D (NEX),YMIN (INP),YMAX (INP),
* INTV,MAX)
100 CONTINUE
000 CONTINUE
ENDFILE 10
STOP
END
SUBROUTINE OVFL
LOGICAL OVFLOW
COMMON /BOMB/OVFLOW
WRITE (6,10)
10 FORMAT (' *** OVERFLOW *** ')
OVFLOW = .TRUE.
RETURN
END

```

BLOCK DATA

IMPLICIT REAL*8 (A-H,O-Z)

REAL*8 IDO(10), IO(10), NITE(10), SASI(10), SAPPEA(10),

* SAPPEB(10), SPPEVA(10), SPPEVB(10), SACPEF(10), SACPEG(10),

* SCPEVF(10), SCPEVG(10), NASTE(10), NSTEV(10)

INTEGER KD(12)

COMMON /STATS/IDO, IO, NITE, SASI, SAPPEA, SAPPEB,

* SPPEVA, SPPEVB, SACPEF, SACPEG, SCPEVF, SCPEVG, NASTE, NSTEV,

* STE, SSPPA, SSPPB, SSCPF, SSCPG, M, KD

DATA KD/0, 1, 2, 5, 10, 20, 50, 100, 200, 500, 1000, 2000/

END

IMPLICIT REAL*8 (A-H,O-Z)

EXTERNAL OVFL

REAL*8 A1(5), A2(5), B(5), E(5), AMP(4), W(4, 5)

REAL*8 C(5), D(5), AHAT(4), BHAT(4), PHAT(4), GHAT(4)

REAL YMIN(4), YMAX(4)

REAL*8 IDO(10), IO(10), NITE(10), SASI(10), SAPPEA(10),

* SAPPEB(10), SPPEVA(10), SPPEVB(10), SACPEF(10), SACPEG(10),

* SCPEVF(10), SCPEVG(10), NASTE(10), NSTEV(10)

REAL*8 ESTMUL(3)

INTEGER PCTL, PTEMP, PXCHG, ESTSTR(3), SCSTR(3)

INTEGER KD(12)

COMMON /STATS/IDO, IO, NITE, SASI, SAPPEA, SAPPEB,

* SPPEVA, SPPEVB, SACPEF, SACPEG, SCPEVF, SCPEVG, NASTE, NSTEV,

* STE, SSPPA, SSPPB, SSCPF, SSCPG, M, KD

COMMON /LUN/LP

COMMON /CONTRL/F, G, AHAT, BHAT, PHAT, GHAT

DATA K/4/

DATA A1/0.95D0, 0.9D0, 0.9D0, 0.9D0, 0.83D0/

DATA A2/0.2D0, 0.1D0, 0.8D0, 0.7D0, 0.3D0/

DATA B/0.065D0, 0.1D0, 0.08D0, 0.5D0, -0.193D0/

DATA E/0.01D0, -0.01D0, 0.02D0, -0.3D0, 0.208D0/

DATA C/0.4D0, 0.7D0, 1.0D0, 0.8D0, 0.6D0/

DATA D/0.8D0, 0.65D0, 0.5D0, 0.6D0, 0.7D0/

DATA YMIN/-1.0, -4.0, -4.0, -1.0/

DATA YMAX/2.0, 4.0, 4.0, 2.0/

DATA MAX, NSEED/1000, 13579/

DATA AMP/1.0D0, 2.0D0, 2.0D0, 2.0D0/

DATA W(1, 1), W(1, 2), W(1, 3), W(1, 4), W(1, 5)/

* 0.17851D0, 0.34463D0, 0.55452D0, 0.40866D0, 0.28534D0/

DATA W(2, 1), W(2, 2), W(2, 3), W(2, 4), W(2, 5)/

* 0.44629D0, 0.86157D0, 1.38629D0, 1.0217D0, 0.71335D0/

DATA W(3, 1), W(3, 2), W(3, 3), W(3, 4), W(3, 5)/

* 0.17851D0, 0.34463D0, 0.55452D0, 0.40866D0, 0.28534D0/

DATA W(4, 1), W(4, 2), W(4, 3), W(4, 4), W(4, 5)/

* 0.44629D0, 0.86157D0, 1.38629D0, 1.0217D0, 0.71335D0/

DATA ESTMUL/0.8D0, 1.0D0, 1.2D0/

DATA PCTL, PTEMP/'1', '0'/

DATA ESTSTR/'-20%', '0', '+20%'/

DATA SCSTR/'SC-', 'SC0', 'SC+'/

LP = 6

CALL ERRSET(207, 1000, -1, 1, OVFL, 0)

WRITE(6, 10)

```

10 FORMAT (' PRINTOUT INTERVAL = ')
   READ (5,*) INTV

   WRITE (6,20)
20 FORMAT (' STARTING AND ENDING EXAMPLES = ')
   READ (5,*) NEXST,NEXEND
   WRITE (6,30)
30 FORMAT (' STARTING AND ENDING ALGORITHMS = ')
   READ (5,*) NALGST,NALGED
   WRITE (6,40)
40 FORMAT (' SMOOTHING COEFFICIENT = ')
   READ (5,*) NSC
   DO 1000 NEX=NEXST,NEXEND
   DO 100 NALG=NALGST,NALGED
   DO 100 NEST=1,3
   DO 100 INP=1,4
   AHAT(K) = A1(NEX)*ESTMUL(NEST)
   AHAT(K-1) = AHAT(K)
   BHAT(K) = B(NEX)*ESTMUL(NEST)
   BHAT(K-1) = BHAT(K)
   GHAT(K-1) = C(NEX)/B(NEX)
   GHAT(K-1) = GHAT(K-1)
   GHAT(K-2) = GHAT(K-1)
   G = GHAT(K-1)
   PHAT(K-1) = (D(NEX)-A1(NEX))/B(NEX)
   PHAT(K-1) = PHAT(K-1)
   PHAT(K-2) = PHAT(K-1)
   F = PHAT(K-1)
   WRITE (10,50) PCTL,NEX,INP,NALG,ESTSTR(NEST),SCSTR(NSC)
50 FORMAT (A1,15X,'EXAMPLE',12,' INPUT',12,' ALGORITHM',12,
* ' EST = ',A4,' SS2 ',A4/)
   WRITE (6,50) PCTL,NEX,INP,NALG,ESTSTR(NEST),SCSTR(NSC)
   PXCHG = PCTL
   PCTL = PTEMP
   PTEMP = PXCHG
   CALL SIMUL (A1(NEX),A2(NEX),B(NEX),E(NEX),AMP(INP),W(INP,NEX),
* NSEED,INP,NALG,NEX,NSC,C(NEX),D(NEX),YMIN(INP),YMAX(INP),
* INTV,MAX)
100 CONTINUE
1000 CONTINUE
   ENDFILE 10
   STOP
   END
   SUBROUTINE OVPL
   LOGICAL OVFLOW
   COMMON /BOMB/OVFLOW
   WRITE (6,10)
10 FORMAT (/ ' *** OVERFLOW *** ' /)
   OVFLOW = .TRUE.
   RETURN
   END

```

```

SUBROUTINE SIMUL(A1,A2,B,E,AMP,W,NSEED,
*  MODE,NALG,NEX,NSC,C,D,YMIN,YMAX,INTV,MAX)
IMPLICIT REAL*8 (A-H,O-Z)
REAL*8 MU,U(4),Y(4),R(4),S(4)
REAL*8 AHAT(4),BHAT(4),FHAT(4),GHAT(4)
REAL SIG(4),TIME
INTEGER SYMBOL(4)
REAL*8 IDO(10),IO(10),NITE(10),SASI(10),SAPPEA(10),
*  SAPPEB(10),SPPEVA(10),SPPEVB(10),SACPEF(10),SACPEG(10),
*  SCPEVF(10),SCPEVG(10),NASTE(10),NSTEV(10)
REAL*8 T1(4),T2(4),T3(4),T4(4),T5(4),T6(4),T7(4),
*  T8(4),T9(4)
REAL*8 SC(4,3,2)
INTEGER KD(12)
LOGICAL OVFLOW
COMMON /STATS/IDO,IO,NITE,SASI,SAPPEA,SAPPEB,
*  SPPEVA,SPPEVB,SACPEF,SACPEG,SCPEVF,SCPEVG,NASTE,NSTEV,
*  STE,SSPPA,SSPPB,SSCPF,SSCPG,M,KD
COMMON /LUN/LP
COMMON /BOMB/OVFLOW
COMMON /SIGNAL/R,U,Y,S
COMMON /CONTRL/F,G,AHAT,BHAT,FHAT,GHAT
COMMON /MISC1/T1,T2,T3
COMMON /MISC2/T4,T5,T6
COMMON /MISC3/T7,T8,T9
DATA K/4/
DATA SC(1,1,1),SC(1,2,1),SC(1,3,1)/-0.9,-0.95,-0.97/
DATA SC(1,1,2),SC(1,2,2),SC(1,3,2)/-0.7,-0.8,-0.9/
DATA SC(2,1,1),SC(2,2,1),SC(2,3,1)/-0.9,-0.9,-0.9/
DATA SC(2,1,2),SC(2,2,2),SC(2,3,2)/-0.65,-0.65,-0.65/
DATA SC(3,1,1),SC(3,2,1),SC(3,3,1)/-0.9,-0.9,-0.9/
DATA SC(3,1,2),SC(3,2,2),SC(3,3,2)/-0.5,-0.5,-0.5/
DATA SC(4,1,1),SC(4,2,1),SC(4,3,1)/-0.8,-0.9,-0.95/
DATA SC(4,1,2),SC(4,2,2),SC(4,3,2)/-0.4,-0.6,-0.8/
DATA SYMBOL/'U','R','S','Y'/

```

```
OVFLOW = .FALSE.
```

```

DO 100 I=1,4
U(I) = 0.0D0
Y(I) = 0.0D0
R(I) = 0.0D0
S(I) = 0.0D0
T1(I) = 0.0D0
T2(I) = 0.0D0
T3(I) = 0.0D0
T4(I) = 0.0D0
T5(I) = 0.0D0
T6(I) = 0.0D0
T7(I) = 0.0D0
T8(I) = 0.0D0
T9(I) = 0.0D0

```

```
100 CONTINUE
```

```

NSD = NSEED
M = 1

```



```

DO 200 I=1,10
IDO(I) = 0.0D0
IO(I) = 0.0D0
SASI(I) = 0.0D0
NASTE(I) = 0.0D0
NSTEV(I) = 0.0D0
NITE(I) = 0.0D0
SAPPEA(I) = 0.0D0
SAPPEB(I) = 0.0D0
SACPEF(I) = 0.0D0
SACPEG(I) = 0.0D0
SPPEVA(I) = 0.0D0
SPPEVB(I) = 0.0D0
SCPEVF(I) = 0.0D0
SCPEVG(I) = 0.0D0

```

```
200 CONTINUE
```

```

KOUNT = 0
STE = 0.0D0
SSPPA = 0.0D0
SSPPB = 0.0D0
SSCPF = 0.0D0
SSCPG = 0.0D0
CY1 = A1+A2
CY2 = -A1*A2
CU1 = B+E
CU2 = -B*A2-E*A1
IPRT = 0
INIT = 1

```

```

MU = 1.0D0
RHO = 1.0D0

```

```

H = 1.0D0
IF(NALG.EQ.3) Q = SC(NEX,NSC,1)
IF(NALG.EQ.5) Q = SC(NEX,NSC,2)

```

```
DO 5000 N=1,MAX
```

```
IF(OVFLOW) GOTO 6000
```

```

Y(K) = CY1*Y(K-1)+CY2*Y(K-2)+CU1*U(K-1)+CU2*U(K-2)
S(K) = C*R(K-1)+D*S(K-1)

```

```

COMMENT OUT THIS COMPUTED GOTO AND ALSO THE
FOLLOWING CALL'S FOR THE NON-ADAPTIVE CASE

```

```

GOTO (210,220,230,240,250),NALG
: 210 CALL ADAPT1(C,D,MU,RHO,H)
: GOTO 300
: 220 CALL ADAPT2(C,D,MU,RHO,H)
: GOTO 300
: 230 CALL ADAPT3(C,D,MU,RHO,H,Q)
: GOTO 300
: 240 CALL ADAPT4(C,D,MU,RHO,H)
: GOTO 300

```

```
250 CALL ADAPT5(C,D,MU,RHO,H,Q)
300 CONTINUE
```

```
CALL INP(MODE,N,AMP,W,NSD,R(K))
U(K) = G*R(K)+F*Y(K)
TIME = FLOAT(N)
```

```
CALL STAT(N,KOUNT)
```

```
IF((N/INTV)*INTV.NE.N) GOTO 1000
SIG(1) = SNGL(U(K))
SIG(2) = SNGL(R(K))
SIG(3) = SNGL(S(K))
SIG(4) = SNGL(Y(K))
CALL TTPLOT(TIME,SIG,4,SYMBOL,YMIN,YMAX,IPRT,INIT)
```

```
000 CONTINUE
```

```
R(K-2) = R(K-1)
R(K-1) = R(K)
U(K-2) = U(K-1)
U(K-1) = U(K)
Y(K-2) = Y(K-1)
Y(K-1) = Y(K)
S(K-2) = S(K-1)
S(K-1) = S(K)
```

```
5000 CONTINUE
```

```
GOTO 6500
```

```
6000 CONTINUE
```

```
IDO(M) = 0.000
IO(M) = 0.000
SASI(M) = 0.000
NASTE(M) = 0.000
NSTEV(M) = 0.000
NITE(M) = 0.000
SAPPEA(M) = 0.000
SAPPEB(M) = 0.000
SACPEF(M) = 0.000
SACPEG(M) = 0.000
SPPEVA(M) = 0.000
SPPEVB(M) = 0.000
SCPEVF(M) = 0.000
SCPEVG(M) = 0.000
```

```
6500 CONTINUE
```

```
WRITE(10,7000) (KD(I),I=2,6)
```

```
7000 FORMAT(11X,' K',6X,5(2X,I5,4X)/)
```

```
WRITE(10,7010) (IDO(I),I=1,5)
```

```
7010 FORMAT(11X,' IDO',4X,5(1PD11.3))
```

```
WRITE(10,7020) (IO(I),I=1,5)
```

```
7020 FORMAT(11X,' IO',5X,5(1PD11.3))
```

```
WRITE(10,7030) (NITE(I),I=1,5)
```

```
7030 FORMAT(11X,' NITE',3X,5(1PD11.3))
```

```
WRITE(10,7040) (NASTE(I),I=1,5)
```

```
7040 FORMAT(11X,' NASTE',2X,5(1PD11.3))
```

```
WRITE(10,7050) (NSTEV(I),I=1,5)
```

```
7050 FORMAT(11X,' NSTEV',2X,5(1PD11.3))
```

```

WRITE(10,7060) (SASI(I),I=1,5)
060 FORMAT(11X,' SASI',3X,5(1PD11.3))

```

COMMENT OUT THESE WRITE'S FOR THE NON-ADAPTIVE
CASE BECAUSE THE PARAMETERS DO NOT CHANGE

```

WRITE(10,7070) (SACPEF(I),I=1,5)
070 FORMAT(11X,' SACPE F',5(1PD11.3))
WRITE(10,7080) (SACPEG(I),I=1,5)
080 FORMAT(11X,' SACPE G',5(1PD11.3))
WRITE(10,7090) (SCPEVF(I),I=1,5)
090 FORMAT(11X,' SCPEV F',5(1PD11.3))
WRITE(10,7100) (SCPEVG(I),I=1,5)
100 FORMAT(11X,' SCPEV G',5(1PD11.3))
IF(NALG.GT.3) GOTO 7500
WRITE(10,7110) (SAPPEA(I),I=1,5)
110 FORMAT(11X,' SAPPE A',5(1PD11.3))
WRITE(10,7120) (SAPPEB(I),I=1,5)
120 FORMAT(11X,' SAPPE B',5(1PD11.3))
WRITE(10,7130) (SPPEVA(I),I=1,5)
130 FORMAT(11X,' SPPEV A',5(1PD11.3))
WRITE(10,7140) (SPPEVB(I),I=1,5)
140 FORMAT(11X,' SPPEV B',5(1PD11.3)/)
GOTO 7800

```

```

000 CONTINUE
WRITE(10,7750)
050 FORMAT(////)

```

```

000 CONTINUE
WRITE(10,7000) (KD(I),I=7,11)
WRITE(10,7010) (IDO(I),I=6,10)
WRITE(10,7020) (IO(I),I=6,10)
WRITE(10,7030) (NITE(I),I=6,10)
WRITE(10,7040) (NASTE(I),I=6,10)
WRITE(10,7050) (NSTEV(I),I=6,10)
WRITE(10,7060) (SASI(I),I=6,10)

```

COMMENT OUT THESE WRITE STATEMENTS FOR THE
NON-ADAPTIVE CASE, SINCE F,G,A,B DO NOT VARY

```

WRITE(10,7070) (SACPEF(I),I=6,10)
WRITE(10,7080) (SACPEG(I),I=6,10)
WRITE(10,7090) (SCPEVF(I),I=6,10)
WRITE(10,7100) (SCPEVG(I),I=6,10)
IF(NALG.GT.3) GOTO 8000
WRITE(10,7110) (SAPPEA(I),I=6,10)
WRITE(10,7120) (SAPPEB(I),I=6,10)
WRITE(10,7130) (SPPEVA(I),I=6,10)
WRITE(10,7140) (SPPEVB(I),I=6,10)
RETURN
000 CONTINUE
WRITE(10,7750)
RETURN
END

```

```
SUBROUTINE INP(MODE,K,AMP,W,NSEED,U)
  IMPLICIT REAL*8 (A-H,O-Z)
  REAL RANNO
  IF(K.LT.0) GOTO 1000
  GOTO(100,200,300,400),MODE
100 U = AMP
   RETURN
200 CONTINUE
300 U = AMP*DSIN(W*FLOAT(K))
   RETURN
400 CALL RANDU(NSEED,M,RANNO)
     NSEED = M
     U = AMP*(RANNO-0.25)
     RETURN
000 U = 0.000
     RETURN
     END
```

```

SUBROUTINE STAT(N,KOUNT)
  IMPLICIT REAL*8 (A-H,O-Z)
  REAL*8 R(4),U(4),Y(4),S(4)
  REAL*8 IDO(10),IO(10),NITE(10),SASI(10),SAPPEA(10),
  * SAPPEB(10),SPPEVA(10),SPPEVB(10),SACPEF(10),SACPEG(10),
  * SCPEVF(10),SCPEVG(10),NASTE(10),NSTEV(10)
  REAL*8 AHAT(4),BHAT(4),PHAT(4),GHAT(4)
  INTEGER KD(12)
  COMMON /STATS/IDO,IO,NITE,SASI,SAPPEA,SAPPEB,
  * SPPEVA,SPPEVB,SACPEF,SACPEG,SCPEVF,SCPEVG,NASTE,NSTEV,
  * STE,SSPPA,SSPPB,SSCPF,SSCPG,M,KD
  COMMON /CONTRL/F,G,AHAT,BHAT,PHAT,GHAT
  COMMON /SIGNAL/R,U,Y,S
  DATA K/4/
  IF(DABS(S(K)).LT.1.0D-03) GOTO 100
  TE = (S(K)-Y(K))/S(K)
  GOTO 200
100 TE = 0.0
  KOUNT = KOUNT+1
200 NASTE(M) = NASTE(M)+TE**2
  STE = STE+TE
  SASI(M) = SASI(M)+U(K)**2
  SAPPEA(M) = SAPPEA(M)+AHAT(K-1)
  SAPPEB(M) = SAPPEB(M)+BHAT(K-1)
  SACPEF(M) = SACPEF(M)+F
  SACPEG(M) = SACPEG(M)+G
  SSPPA = SSPPA+AHAT(K-1)**2
  SSPPB = SSPPB+BHAT(K-1)**2
  SSCPF = SSCPF+F**2
  SSCPG = SSCPG+G**2
300 TIME = FLOAT(N)
  IF(N.LT.KD(M+1)) RETURN
  IDO(M) = S(K)
  IO(M) = Y(K)
  RANGE = FLOAT(KD(M+1)-KD(M))
  DIV = (RANGE-FLOAT(KOUNT))*(RANGE-FLOAT(KOUNT)-1)
  RNG = RANGE-FLOAT(KOUNT)
  NSTEV(M) = 0.0
  IF(DIV.GT.0.0) NSTEV(M) = (RNG*NASTE(M)-STE**2)/DIV
  IF(NSTEV(M).LT.0.0D0) NSTEV(M) = 0.0D0
  NITE(M) = TE
  DIVSOR = RANGE*(RANGE-1)
  IF(DIVSOR.LE.0.0) GOTO 1000
  SPPEVA(M) = (RANGE*SSPPA-SAPPEA(M)**2)/DIVSOR
  IF(SPPEVA(M).LT.0.0D0) SPPEVA(M) = 0.0D0
  SPPEVB(M) = (RANGE*SSPPB-SAPPEB(M)**2)/DIVSOR
  IF(SPPEVB(M).LT.0.0D0) SPPEVB(M) = 0.0D0
  SCPEVF(M) = (RANGE*SSCPF-SACPEF(M)**2)/DIVSOR
  IF(SCPEVF(M).LT.0.0D0) SCPEVF(M) = 0.0D0
  SCPEVG(M) = (RANGE*SSCPG-SACPEG(M)**2)/DIVSOR
  IF(SCPEVG(M).LT.0.0D0) SCPEVG(M) = 0.0D0
  GOTO 2000
1000 SPPEVA(M) = 0.0D0
  SPPEVB(M) = 0.0D0

```

```
SCFEVF (M) = 0.000
SCPEVG (M) = 0.000
000 CONTINUE
SASI (M) = SASI (M) / RANGE
SAPPEA (M) = SAPPEA (M) / RANGE
SAPPEB (M) = SAPPEB (M) / RANGE
SACPEF (M) = SACPEF (M) / RANGE
SACPEG (M) = SACPEG (M) / RANGE
NASTE (M) = NASTE (M) / RANGE
KOUNT = 0
STE = 0.000
SSPPA = 0.000
SSPPB = 0.000
SSCPF = 0.000
SSCPG = 0.000
M = M + 1
RETURN
END
```

CC

GEI/SEI ALGORITHM

EQUATIONS (4-4) - (4-8)

SUBROUTINE ADAPT1(C,D,MU,RHO,H)
 IMPLICIT REAL*8 (A-H,O-Z)
 REAL*8 MU,R(4),U(4),Y(4),S(4),AHAT(4),BHAT(4),E(4)
 REAL*8 PHAT(4),GHAT(4)
 COMMON /SIGNAL/R,U,Y,S
 COMMON /CONTRL/P,G,AHAT,BHAT,PHAT,GHAT
 DATA K/4/

AHAT(K-2) = AHAT(K-1)

AHAT(K-1) = AHAT(K)

BHAT(K-2) = BHAT(K-1)

BHAT(K-1) = BHAT(K)

E(K-2) = E(K-1)

E(K-1) = E(K)

PHAT(K-2) = PHAT(K-1)

PHAT(K-1) = PHAT(K)

GHAT(K-2) = GHAT(K-1)

GHAT(K-1) = GHAT(K)

$E(K-1) = Y(K-1) - AHAT(K-1) * Y(K-2) - BHAT(K-1) * U(K-2)$

$DENOM = H + MU * Y(K-2) ** 2 + RHO * U(K-2) ** 2$

$AHAT(K) = AHAT(K-1) + MU * Y(K-2) * E(K-1) / DENOM$

$BHAT(K) = BHAT(K-1) + RHO * U(K-2) * E(K-1) / DENOM$

$GHAT(K-1) = C / BHAT(K)$

$PHAT(K-1) = (D - AHAT(K)) / BHAT(K)$

$F = PHAT(K-1)$

$G = GHAT(K-1)$

RETURN

END

CC

GOI ALGORITHM

EQUATIONS (4-12) - (4-19)

SUBROUTINE ADAPT2 (C, D, MU, RHO, H)
 IMPLICIT REAL*8 (A-H, O-Z)
 REAL*8 MU, R(4), U(4), Y(4), S(4), YHAT(4), AHAT(4), BHAT(4)
 REAL*8 E(4), FHAT(4), GHAT(4)
 REAL*8 LAMBDA(4), GAMMA(4)
 COMMON /SIGNAL/R, U, Y, S
 COMMON /CONTRL/F, G, AHAT, BHAT, FHAT, GHAT
 COMMON /MISC1/GAMMA, LAMBDA, YHAT
 DATA K/4/

LAMBDA (K-2) = LAMBDA (K-1)
 LAMBDA (K-1) = LAMBDA (K)
 GAMMA (K-2) = GAMMA (K-1)
 GAMMA (K-1) = GAMMA (K)
 AHAT (K-2) = AHAT (K-1)
 AHAT (K-1) = AHAT (K)
 BHAT (K-2) = BHAT (K-1)
 BHAT (K-1) = BHAT (K)
 YHAT (K-2) = YHAT (K-1)
 YHAT (K-1) = YHAT (K)
 E (K-2) = E (K-1)
 E (K-1) = E (K)
 FHAT (K-2) = FHAT (K-1)
 FHAT (K-1) = FHAT (K)
 GHAT (K-2) = GHAT (K-1)
 GHAT (K-1) = GHAT (K)

 YHAT (K-1) = AHAT (K-1) * YHAT (K-2) + BHAT (K-1) * U (K-2)
 LAMBDA (K-1) = YHAT (K-2) + AHAT (K-1) * LAMBDA (K-2)
 GAMMA (K-1) = U (K-2) + AHAT (K-1) * GAMMA (K-2)
 E (K-1) = Y (K-1) - YHAT (K-1)
 DENOM = H + MU * LAMBDA (K-1) ** 2 + RHO * GAMMA (K-1) ** 2
 AHAT (K) = AHAT (K-1) + MU * LAMBDA (K-1) * E (K-1) / DENOM
 BHAT (K) = BHAT (K-1) + RHO * GAMMA (K-1) * E (K-1) / DENOM
 GHAT (K-1) = C / BHAT (K)
 FHAT (K-1) = (D - AHAT (K)) / BHAT (K)
 F = FHAT (K-1)
 G = GHAT (K-1)

RETURN
 END

CC

SOI ALGORITHM

EQUATIONS (4-23) - (4-29)

```

SUBROUTINE ADAPT3(C,D,MU,RHO,H,Q)
IMPLICIT REAL*8 (A-H,O-Z)
REAL*8 MU,R(4),U(4),Y(4),S(4),YHAT(4),AHAT(4),BHAT(4)
REAL*8 V(4),Z(4),PHAT(4),GHAT(4)
COMMON /SIGNAL/R,U,Y,S
COMMON /CONTRL/P,G,AHAT,BHAT,PHAT,GHAT
COMMON /MISC2/YHAT,Z,V
DATA K/4/

```

$$Z(K-2) = Z(K-1)$$

$$Z(K-1) = Z(K)$$

$$AHAT(K-2) = AHAT(K-1)$$

$$AHAT(K-1) = AHAT(K)$$

$$BHAT(K-2) = BHAT(K-1)$$

$$BHAT(K-1) = BHAT(K)$$

$$YHAT(K-2) = YHAT(K-1)$$

$$YHAT(K-1) = YHAT(K)$$

$$V(K-2) = V(K-1)$$

$$V(K-1) = V(K)$$

$$PHAT(K-2) = PHAT(K-1)$$

$$PHAT(K-1) = PHAT(K)$$

$$GHAT(K-2) = GHAT(K-1)$$

$$GHAT(K-1) = GHAT(K)$$

$$YHAT(K-1) = AHAT(K-1) * Z(K-2) + BHAT(K-1) * U(K-2)$$

$$DENOM = H + MU * Z(K-2) ** 2 + RHO * U(K-2) ** 2$$

$$V(K-1) = (Y(K-1) - YHAT(K-1) + Q * (Y(K-2) - Z(K-2))) / DENOM$$

$$AHAT(K) = AHAT(K-1) + MU * Z(K-2) * V(K-1)$$

$$BHAT(K) = BHAT(K-1) + RHO * U(K-2) * V(K-1)$$

$$Z(K-1) = AHAT(K) * Z(K-2) + BHAT(K) * U(K-2)$$

$$GHAT(K-1) = C / BHAT(K)$$

$$PHAT(K-1) = (D - AHAT(K)) / BHAT(K)$$

$$F = PHAT(K-1)$$

$$G = GHAT(K-1)$$

RETURN

END

CC

GED/SED ALGORITHM

EQUATIONS (4-34) - (4-36)

SUBROUTINE ADAIT4 (C,D,MU,RHO,H)
 IMPLICIT REAL*8 (A-H,O-Z)
 REAL*8 MU,R(4),U(4),Y(4),S(4),V(4),PHAT(4),GHAT(4)
 REAL*8 AHAT(4),BHAT(4)
 COMMON /SIGNAL/R,U,Y,S
 COMMON /CONTRL/F,G,AHAT,BHAT,PHAT,GHAT
 DATA K/4/

GHAT(K-2) = GHAT(K-1)
 GHAT(K-1) = GHAT(K)
 PHAT(K-2) = PHAT(K-1)
 PHAT(K-1) = PHAT(K)
 V(K-2) = V(K-1)
 V(K-1) = V(K)

V(K-1) = C*R(K-2) + D*Y(K-2) - Y(K-1)
 DENOM = H*(1.0 + MU*Y(K-2)**2 + RHO*U(K-2)**2)
 GHAT(K-1) = GHAT(K-2) + RHO*R(K-2)*V(K-1)/DENOM
 PHAT(K-1) = PHAT(K-2) + MU*Y(K-2)*V(K-1)/DENOM
 G = GHAT(K-1)
 F = PHAT(K-1)

RETURN
 END

CC

SOD ALGORITHM

EQUATIONS (4-40) - (4-44)

SUBROUTINE ADAFT5 (C,D,MU,RHO,H,Q)
 IMPLICIT REAL*8 (A-H,O-Z)
 REAL*8 MU,R(4),U(4),Y(4),S(4),V(4),PHAT(4),GHAT(4)
 REAL*8 GAMMA(4),BETA(4),AHAT(4),BHAT(4)
 COMMON /SIGNAL/R,U,Y,S
 COMMON /CONTRL/P,G,AHAT,BHAT,PHAT,GHAT
 COMMON /MISC3/BETA,GAMMA,V
 DATA K/4/

GHAT(K-2) = GHAT(K-1)
 GHAT(K-1) = GHAT(K)
 PHAT(K-2) = PHAT(K-1)
 PHAT(K-1) = PHAT(K)
 BETA(K-2) = BETA(K-1)
 BETA(K-1) = BETA(K)
 GAMMA(K-2) = GAMMA(K-1)
 GAMMA(K-1) = GAMMA(K)
 V(K-2) = V(K-1)
 V(K-1) = V(K)

BETA(K-1) = RHO*R(K-2)**2+MU*Y(K-2)**2
 GAMMA(K-1) = (D+Q)*H*BETA(K-2)*V(K-2)+D*GAMMA(K-2)
 DENOM = 1.0+H*BETA(K-1)
 V(K-1) = (S(K-1)-Y(K-1)+Q*(S(K-2)-Y(K-2))-GAMMA(K-1))/DENOM
 GHAT(K-1) = GHAT(K-2)+RHO*R(K-2)*V(K-1)
 PHAT(K-1) = PHAT(K-2)+MU*Y(K-2)*V(K-1)
 G = GHAT(K-1)
 F = PHAT(K-1)

RETURN
 END

CC
C THIS PROGRAM PROVIDES THE INITIAL VALUES NECESSARY
C TO SIMULATE EXAMPLES 5 AND 6. THE SIMULATION IS
C PROVIDED BY SUBROUTINE SIMULZ.
C

BLOCK DATA

IMPLICIT REAL*8 (A-H,O-Z)
REAL*8 I00(10), I01(10), N11(10), SAS1(10),
* SAPEA1(10), SAPEA2(10), SAPEB1(10), SAPEB2(10),
* SPLVA1(10), SPLVA2(10), SPLVB1(10), SPLVB2(10),
* SACEE1(10), SACEE2(10), SACEN1(10), SACEN2(10),
* SCEV1(10), SCEV2(10), SCEVN1(10), SCEVN2(10),
* NASTE(10), NSTEV(10)
INTEGER K01(2)
COMMON /STATS/ I00, I01, N11, SAS1, SAPEA1, SAPEA2, SAPEB1, SAPEB2,
* SPLVA1, SPLVA2, SPLVB1, SPLVB2, SACEE1, SACEE2, SACEN1, SACEN2,
* SCEV1, SCEV2, SCEVN1, SCEVN2,
* NASTE, NSTEV, STE,
* SSPA1, SSPA2, SSPD1, SSPD2, SSCB1, SSCB2, SSCN1, SSCN2, M, KD
DATA KD/0, 1, 2, 5, 10, 20, 50, 100, 200, 500, 1000, 2000/
END

IMPLICIT REAL*8 (A-H,O-Z)
EXTERNAL OVFL
REAL*8 A1(5), A2(5), B1(5), B2(5), AMP(4)
REAL*8 ALPHA(2,2), BETA(2,2)
REAL*8 A1(5), B1(5)
REAL*8 OMEGA(2), ZETA(2), LAMBDA(2), GAMMA(2)
REAL*8 EXPORT, COSWT, I
REAL*8 ESTIMUL(5)
REAL YMIN(4), YMAX(4)
INTEGER FCIL, FLEMP, PROTO, ESTISIR(5), SCSTR(5)
REAL*8 I00(10), I01(10), N11(10), SAS1(10),
* SAPEA1(10), SAPEA2(10), SAPEB1(10), SAPEB2(10),
* SPLVA1(10), SPLVA2(10), SPLVB1(10), SPLVB2(10),
* SACEE1(10), SACEE2(10), SACEN1(10), SACEN2(10),
* SCEV1(10), SCEV2(10), SCEVN1(10), SCEVN2(10),
* NASTE(10), NSTEV(10)
REAL*8 I1(10), I2A(5), NO1(5), NO2(5)
INTEGER K01(2)

COMMON /STATS/ I00, I01, N11, SAS1, SAPEA1, SAPEA2, SAPEB1, SAPEB2,
* SPLVA1, SPLVA2, SPLVB1, SPLVB2, SACEE1, SACEE2, SACEN1, SACEN2,
* SCEV1, SCEV2, SCEVN1, SCEVN2,
* NASTE, NSTEV, STE,
* SSPA1, SSPA2, SSPD1, SSPD2, SSCB1, SSCB2, SSCN1, SSCN2, M, KD
COMMON /C/ OMEGA1, OMEGA2, I01, I02, I03, I04, I05, I06, I07, I08, I09, I10, I11, I12
COMMON /LEN/

DATA YMIN/-1.0, 0.0, 10.0, -10.0, -10.0/
DATA YMAX/10.0, 0.0, 0.0, 10.0, 10.0/
DATA MPA, SCSTR/1000, 1575.00/
SET PARAMETERS FOR EXAMPLES 5 AND 6
DATA SICR/1.0/
DATA LMEGA/0.5, 5.0/
DATA ZETA/0.2, 0.02/
DATA GAMMA/1.0, -0.141/
DATA LAMBDA/1.0, 1.0/
DATA DELTA/0.5/
DATA T/0.1/
DATA AMP/1.000, 2.000, 2.0, 0.2, 0.000/
DATA ESTIMUL/0.000, 1.000, 1.200/
DATA FCIL/'1'
DATA FLEMP/'0'
DATA ESTISIR/'-20%', '0', '0', '+20%'
DATA SCSTR/'SC-', 'SCU', 'SC+'

ORIGINAL PAGE IS
OF POOR QUALITY

LP = 0
CALL ERKSET(207, 1000, -1, 1, OVFL, 0)
CALL CRFSET(208, 1000, -1, 1, 1, 1)

CALCULATE COEFFICIENTS FOR THE Z-PLANE EQUIVALENT OF
THE n TH ORDER CONTINUOUS PLANT (EQNS. (3-18)-(3-22))
DO 5 I=1,2
EXPNT = DEXP(-ZETA(I)*OMEGA(I)*I)
COSWT = COS(OMEGA(I)*I*OSCRT(1.0-ZETA(I)**2))
ALPHA(I,1) = 2.0*EXPNT*COSWT
ALPHA(I,2) = -(EXPORT**2)
BETA(I,1) = (LAMBDA(I)/OMEGA(I)**2)*I*(1.0-EXPNT*
* (COSWT+(ZETA(I)/OSCRT(1.0-ZETA(I)**2)))*
* DSIN(OMEGA(I)*I*OSCRT(1.0-ZETA(I)**2)))

```

BETA(1,2) = (LAMBDA(1)/SIGMA(1)**2)*(LXPCW1*
* (LXPCW1-USHW1+(ZETA(1)/DSQR(1.0-ZETA(1)**2))*
* DSIN(UMD(1))*1+DSQR(1.0-ZETA(1)**2)))
5 CONTINUE

C
C
C CALCULATE ACTUATOR COEFFICIENTS
ESIG1 = DEXP(-SIGMA*1)
ESIG2 = 1.-DEXP(-SIGMA*1)

C
C
C DEFINE NEW DIFFERENCE EQUATION PARAMETERS FOR THE 4TH ORDER PLANT
FROM EQS. (4-74).
A(1) = ALPHA(1,1)+ALPHA(2,1)
A(2) = ALPHA(1,2)-ALPHA(1,1)*ALPHA(2,1)+ALPHA(2,2)
A(3) = -ALPHA(1,2)*ALPHA(2,1)-ALPHA(1,1)*ALPHA(2,2)
A(4) = -ALPHA(1,2)*ALPHA(2,2)
B(1) = BETA(1,1)+BETA(2,1)
B(2) = BETA(1,2)+BETA(2,2)-ALPHA(2,1)*BETA(1,1)
* -ALPHA(1,1)*BETA(2,1)
B(3) = -ALPHA(2,2)*BETA(1,1)-ALPHA(2,1)*BETA(1,2)
* -ALPHA(1,2)*BETA(2,1)-ALPHA(1,1)*BETA(2,2)
B(4) = -ALPHA(2,2)*BETA(1,2)-ALPHA(1,2)*BETA(2,2)

C
C
C WRITE(6,10)
10 FORMAT(' PRINTOUT INTERVAL = ')
READ(5,*)INTV

C
C
C WRITE(6,30)
30 FORMAT(' STARTING AND ENDING ALGORITHMS = ')
READ(5,*)NALG1,NALG2

C
C
C WRITE(6,40)
40 FORMAT(' SMOOTHING COEFFICIENT FOR SUI ALGORITHM (NS) = ')
READ(5,*)NSC
NEX = 5

C
C
C WRITE(6,55)
55 FORMAT(' ADAPTIVE SUI SIZE = ')
READ(5,57) STPSIZ
57 FORMAT('4.2)

C
C
C ACTUATOR CONFIGURATION LOOP (1=NO ACTUATOR,
2=ACTUATOR CONFIGURATION (A), 3=ACTUATOR
CONFIG. (B), 4=ACTUATOR CONFIG. (C) ).
DO 100 NACT=1,4

C
C
C ADAPTIVE ALGORITHMS LOOP
42 DO 100 NALG=NALG1,NALG2

C
C
C INITIAL PLANT PARAMETER ESTIMATES LOOP
DO 100 NEST=1,3

C
C
C INPUTS LOOP
DO 100 INP=1,4

C
C
C IF(NACT.GT.1) NEX = 0

C
C
C SET INITIAL PARAMETER ESTIMATES (ALPHA(1,1) BECOMES
A(1)=ALPHA(1)-HAT ETC., IN EQS. (4-69)-(4-92)).
DO 45 I=1,3
A(1) = ALPHA(1,1)*ESTIMUL(NEST)
A(2) = ALPHA(1,2)*ESTIMUL(NEST)
B(1) = BETA(1,1)*ESTIMUL(NEST)
B(2) = BETA(1,2)*ESTIMUL(NEST)
45 CONTINUE

C
C
C CALCULATE INITIAL CONTROLLER PARAMETERS
CALL CNISSET(GAMMA)

C
C
C WRITE(6,50)PCIL,NEX,INP,NALG,ESTSTR(NEST),SCSTR(NSC),
* NACT,STPSIZ
50 FORMAT('A1,ZX,EXAMPLE',12,' INPUT',12,' ALGORITHM',12,
* EST = ',A4,' SS2 = ',A4,' NACT = ',12,' ST = ',F4.2)
WRITE(6,50)PCIL,NEX,INP,NALG,ESTSTR(NEST),SCSTR(NSC),NACT
* ,STPSIZ
PXCHG = PCTL
PCTL = PTEMP
ENTER SIMULATION SUBROUTINE
PTEMP = PXCHG
CALL SIMUL2(ALPHA,BETA,A,B,GAMMA,DELTA,AMP(INP),
* USEED,INP,NALG,NEX,NSC,YMIN(INP),YMAX(INP),
* INTV,MAX,ESIG1,ESIG2,STPSIZ)
100 CONTINUE
ENDFILE 10

```

```

      STOP
      END
C
      SUBROUTINE OVFL
      LOGICAL OVFLOW
      COMMON /BOMB/OVFLOW
      WRITE(C,10)
10  FORMAT(/' *** OVERFLOW *** /')
      OVFLOW = .TRUE.
      RETURN
      END

      SUBROUTINE SIMULZ(ALPHA,BETA,A,B,GAMMA,DELTA,AMP,USEED,
      * MODE,NALC,NEX,NSC,YMIN,YMAX,INIV,MAX,ESIG1,LSIG2,STPS1Z)
      IMPLICIT REAL*8 (A-H,O-Z)
      REAL*8 ALPHA(2,2),BETA(2,2)
      REAL*8 A(5),B(5)
      REAL*8 U(5),Y(5),R(5),S(5),CC(5)
      REAL*8 ML(2),RHU(2),GAMMA(2),L(2)
      REAL SIG(5),TIM
      INTEGER SYMUL(4)
      REAL*8 T1(5),T2(5),T3(5),T4(5),T5(5),T6(5),T7(5),
      * T8(5)
      REAL*8 SC(3,2)
      REAL*8 IDU(10),IDU(10),NITE(10),SASI(10),
      * SAPEA1(10),SAPEA2(10),SAPEL1(10),SAPEL2(10),
      * SPEVA1(10),SPEVA2(10),SPEVL1(10),SPEVL2(10),
      * SACE1(10),SACE2(10),SACV1(10),SACV2(10),
      * SCEVE1(10),SCEVE2(10),SCEVN1(10),SCEVN2(10),
      * NASTE(10),NSTEV(10)
      REAL*8 A1(5),A2(5),B1(5),B2(5),ETA1(5),ETA2(5),NU1(5),NU2(5)
      INTEGER KU(12)
      LOGICAL OVFLOW
C
      COMMON /STATS/IDU,IO,NITE,SASI,SAPEA1,SAPEA2,SAPEL1,SAPEL2,
      * SPEVA1,SPEVA2,SPEVL1,SPEVL2,SACE1,SACE2,SACV1,SACV2,
      * SCEVE1,SCEVE2,SCEVN1,SCEVN2,
      * NASTE,NSTEV,STE,
      * SSPA1,SSPA2,SSPB1,SSPB2,SSCE1,SSCE2,SSCN1,SSCN2,M,KD
C
      COMMON /LUN/LP
      COMMON /BOMB/OVFLOW
      COMMON /SIGNAL/R,U,Y,S,CC
      COMMON /CONTRL/A1,A2,B1,B2,ETA1,ETA2,NU1,NU2,NACT
      COMMON /MISC1/T1,T2,T3,T4,T5
      COMMON /MISC2/T6,T7,T8
C
      DATA K/5/
      DATA SYMUL/'U','R','S','Y'/
      DATA SC(1,1),SC(1,2)/-0.9,0.0/
      DATA SC(2,1),SC(2,2)/-1.848,0.9515/
      DATA SC(3,1),SC(3,2)/-0.90,0.0/
C
      OVFLOW = .FALSE.
C
      DO 100 I=1,5
      U(I) = 0.000
      UC(I) = 0.000
      Y(I) = 0.000
      R(I) = 0.000
      S(I) = 0.000
      T1(I) = 0.000
      T2(I) = 0.000
      T3(I) = 0.000
      T4(I) = 0.000
      T5(I) = 0.000
      T6(I) = 0.000
      T7(I) = 0.000
      T8(I) = 0.000
100  CONTINUE
      DSD = DSEED
      M = 1
      DO 110 I=1,10
      IDU(I) = 0.000
      IO(I) = 0.000
      SASI(I) = 0.000
      NASTE(I) = 0.000
      NSTEV(I) = 0.000
      NITE(I) = 0.000

```

```

SAPEA1(1) = 0.000
SAPEA2(1) = 0.000
SAPEB1(1) = 0.000
SAPEB2(1) = 0.000
SACLE1(1) = 0.000
SACLE2(1) = 0.000
SACEN1(1) = 0.000
SACEN2(1) = 0.000
SPLVA1(1) = 0.000
SPEVA2(1) = 0.000
SPEVB1(1) = 0.000
SPEVB2(1) = 0.000
SCEVE1(1) = 0.000
SCEVE2(1) = 0.000
SCEVN1(1) = 0.000
SCEVNE(1) = 0.000

```

110

```

CONTINUE
KOUNT = 0
STE = 0.000
SSPA1 = 0.000
SSPA2 = 0.000
SSPB1 = 0.000
SSPB2 = 0.000
SSCL1 = 0.000
SSCL2 = 0.000
SSCN1 = 0.000
SSCN2 = 0.000
IFRT = 0
INIT = 1

```

```

MU(1) = STPS12
MU(2) = STPS12
RHU(1) = STPS12
RHU(2) = STPS12

```

```

H = 1.000

```

```

Q(1) = SC(NSC,1)
Q(2) = SC(NSC,2)

```

```

ENTER MAIN SIMULATION LOOP

```

```

DO 5000 N=1,MAX

```

```

IF(DVFLOW) GOTO 2000

```

```

Y(K) = 0.000

```

```

CALCULATE NEXT PLANT OUTPUT

```

```

DO 150 I=1,4

```

```

150 Y(K) = Y(K)+A(I)*Y(K-1)+E(I)*U(K-1)

```

```

CALCULATE DESIRED OUTPUT

```

```

* S(K) = DELTA*(BETA(1,1)*K(K-1)+BETA(1,2)*K(K-2))
+ GAMMA(1)*S(K-1)+GAMMA(2)*S(K-2)

```

```

COMPUTE STATISTICS AND PRINT VALUES

```

```

TIME = FLOAT(N)

```

```

CALL STAT2(N,KOUNT)

```

```

IF((N/INTV)*INTV.NE.N) GOTO 1000

```

```

SIG(1) = SNGL(UC(K-1))

```

```

SIG(2) = SNGL(R(K-1))

```

```

SIG(3) = SNGL(S(K))

```

```

SIG(4) = SNGL(Y(K))

```

```

CALL TPLCOT(TIME,SIG,4,SYMBOL,YMIN,YMAX,IFRT,INIT)

```

1000

```

SHIFT VARIABLES TO ALLOW CALCULATION OF NEW VALUES

```

```

R(K-4) = R(K-3)

```

```

R(K-3) = R(K-2)

```

```

R(K-2) = R(K-1)

```

```

U(K-4) = U(K-3)

```

```

U(K-3) = U(K-2)

```

```

U(K-2) = U(K-1)

```

```

UC(K-4) = UC(K-3)

```

```

UC(K-3) = UC(K-2)

```

```

UC(K-2) = UC(K-1)

```

```

Y(K-4) = Y(K-3)
Y(K-3) = Y(K-2)
Y(K-2) = Y(K-1)
Y(K-1) = Y(K)
S(K-4) = S(K-3)
S(K-3) = S(K-2)
S(K-2) = S(K-1)
S(K-1) = S(K)
C   UPDATE PLANT PARAMETER ESTIMATES
C
210  GOTO (210,220,230),NALL
    CALL GE1(GAMMA,ML,KNTOT,I)
    GOTO 300
220  CALL GE1(GAMMA,ML,KML,I)
    GOTO 300
230  CALL SU1(GAMMA,ML,KSU,I,C)
    GOTO 300
300  CONTINUE

C   C
C   C   FIND NEW INPUT R(K-1)
    CALL INF(MODE,N,AMP,DSB,R(K-1))
C   C   C   FIND NEW CONTROLLER OUTPUT UC(K-1) AND ACTUATOR OUTPUT U(K-1)
    IF(INACT.EQ.3) GOTO 410
C   C   C   CALCULATIONS FOR ACTUATOR CONFIG. 1, 2 AND 4
    UC(K-1) = DELTA*R(K-1)+ETA1(K-1)*UC(K-2)+ETA2(K-1)*UC(K-3)
    *   +NU1(K-1)*Y(K-2)+NU2(K-1)*Y(K-3)
C   C   C   IF(INACT.EQ.1) GOTO 420
    GOTO 415
C   C   C   CALCULATION FOR ACTUATOR CONFIG. 3:
410  UC(K-1) = DELTA*R(K-1)+ETA1(K-1)*U(K-2)+ETA2(K-1)*U(K-3)
    *   +NU1(K-1)*Y(K-2)+NU2(K-1)*Y(K-3)
415  U(K-1) = ESIG1*U(K-2)+ESIG2*UC(K-2)
    GOTO 5000
420  U(K-1) = UC(K-1)
5000 CONTINUE
    GOTO 6500
6000 CONTINUE
    IDG(M) = 0.000
    IO(M) = 0.000
    SASI(M) = 0.000
    NASTE(M) = 0.000
    NSTEV(M) = 0.000
    NITE(M) = 0.000
    SAPLAI(M) = 0.000
    SAPEAZ(M) = 0.000
    SAPEB1(M) = 0.000
    SAPEB2(M) = 0.000
    SACEE1(M) = 0.000
    SACEE2(M) = 0.000
    SACLN1(M) = 0.000
    SACEN2(M) = 0.000
    SPEVA1(M) = 0.000
    SPEVA2(M) = 0.000
    SPEVB1(M) = 0.000
    SPEVB2(M) = 0.000
    SCEVE1(M) = 0.000
    SCLEVE2(M) = 0.000
    SCEVNI(M) = 0.000
    SCLEVN2(M) = 0.000
6500 CONTINUE
    WRITE(10,7000)(KD(I),I=2,6)
7000  FORMAT(11X,' K',7X,5(2X,15,+X)/)
    WRITE(10,7010)(IDG(I),I=1,5)
7010  FORMAT(11X,' IDG',5X,5(1PD11.3))
    WRITE(10,7020)(IO(I),I=1,5)
7020  FORMAT(11X,' IO',6X,5(1PD11.3))
    WRITE(10,7030)(NITE(I),I=1,5)
7030  FORMAT(11X,' NITE',4X,5(1PD11.3))
    WRITE(10,7040)(NASTE(I),I=1,5)
7040  FORMAT(11X,' NASTE',5X,5(1PD11.3))
    WRITE(10,7050)(NSTEV(I),I=1,5)
7050  FORMAT(11X,' NSTEV',5X,5(1PD11.3))
    WRITE(10,7060)(SASI(I),I=1,5)
7060  FORMAT(11X,' SASI',4X,5(1PD11.3))
    WRITE(10,7070)(SACEE1(I),I=1,5)

```



```

7070 FORMAT(11X,' SACPE B1',5(1PD11.3))
WRITE(10,7075)(SACPEZ(1),I=1,5)
7075 FORMAT(11X,' SACPE B2',5(1PD11.3))
WRITE(10,7080)(SACPEV1(1),I=1,5)
7080 FORMAT(11X,' SACPE V1',5(1PD11.3))
WRITE(10,7085)(SACPEV2(1),I=1,5)
7085 FORMAT(11X,' SACPE V2',5(1PD11.3))
WRITE(10,7090)(SCEV1(1),I=1,5)
7090 FORMAT(11X,' SCEV B1',5(1PD11.3))
WRITE(10,7095)(SCEV2(1),I=1,5)
7095 FORMAT(11X,' SCEV B2',5(1PD11.3))
WRITE(10,7100)(SCEVN1(1),I=1,5)
7100 FORMAT(11X,' SCEV N1',5(1PD11.3))
WRITE(10,7105)(SCEVN2(1),I=1,5)
7105 FORMAT(11X,' SCEV N2',5(1PD11.3))
WRITE(10,7110)(SAPL A1(1),I=1,5)
7110 FORMAT(11X,' SAPPE A1',5(1PD11.3))
WRITE(10,7115)(SAPL A2(1),I=1,5)
7115 FORMAT(11X,' SAPPE A2',5(1PD11.3))
WRITE(10,7120)(SAPL B1(1),I=1,5)
7120 FORMAT(11X,' SAPPE B1',5(1PD11.3))
WRITE(10,7125)(SAPL B2(1),I=1,5)
7125 FORMAT(11X,' SAPPE B2',5(1PD11.3))
WRITE(10,7130)(SPEVA1(1),I=1,5)
7130 FORMAT(11X,' SPPLEV A1',5(1PD11.3))
WRITE(10,7135)(SPEVA2(1),I=1,5)
7135 FORMAT(11X,' SPPLEV A2',5(1PD11.3))
WRITE(10,7140)(SPEVB1(1),I=1,5)
7140 FORMAT(11X,' SPPLEV B1',5(1PD11.3))
WRITE(10,7145)(SPEVB2(1),I=1,5)
7145 FORMAT(11X,' SPPLEV B2',5(1PD11.3))
GO TO 7600
7500 CONTINUE
WRITE(10,7750)
7750 FORMAT(7777)
7800 CONTINUE
WRITE(10,7000)(K0(1),I=7,11)
WRITE(10,7010)(I00(1),I=6,10)
WRITE(10,7020)(I1(1),I=6,10)
WRITE(10,7030)(V1(1),I=6,10)
WRITE(10,7040)(NS1(1),I=6,10)
WRITE(10,7050)(NS1V1(1),I=6,10)
WRITE(10,7060)(SAS1(1),I=6,10)
WRITE(10,7070)(SAC1(1),I=6,10)
WRITE(10,7075)(SAC1Z(1),I=6,10)
WRITE(10,7080)(SAC1V1(1),I=6,10)
WRITE(10,7085)(SAC1V2(1),I=6,10)
WRITE(10,7090)(SCEV1(1),I=6,10)
WRITE(10,7095)(SCEV2(1),I=6,10)
WRITE(10,7100)(SCEVN1(1),I=6,10)
WRITE(10,7105)(SCEVN2(1),I=6,10)
WRITE(10,7110)(SAPL A1(1),I=6,10)
WRITE(10,7115)(SAPL A2(1),I=6,10)
WRITE(10,7120)(SAPL B1(1),I=6,10)
WRITE(10,7125)(SAPL B2(1),I=6,10)
WRITE(10,7130)(SPEVA1(1),I=6,10)
WRITE(10,7135)(SPEVA2(1),I=6,10)
WRITE(10,7140)(SPEVB1(1),I=6,10)
WRITE(10,7145)(SPEVB2(1),I=6,10)
RETURN
6000 CONTINUE
WRITE(10,7750)
RETURN
END
C
SUBROUTINE INP(MODE,K,AMP,USEL,U)
IMPLICIT REAL*8 (A-H,J-Z)
REAL KANNE(1)
M = 1
IF(K.LT.0) GO TO 1000
CALL(100,200,300,400),MODE
100 U = AMP
RETURN
200 U = AMP*DSIN(PI*U*(U.20*FLOAT(K)))
RETURN
300 U = AMP*DSIN(PI*U*(U.5*FLOAT(K)))
RETURN
400 CALL GODES(USEL,M,KANNE)
U = AMP*(KANNE(1)-0.25)
RETURN
1000 U = 0.000
RETURN
END

```

CC

GE1/SE1 ALGORITHM

EQUATIONS (4-50) - (4-55)

SUBROUTINE GE1(GAMMA,MU,RHO,H)
 IMPLICIT REAL*8 (A-H,O-Z)
 REAL*8 R(5),U(5),Y(5),S(5),A1(5),A2(5),B1(5),B2(5),UC(5)
 REAL*8 MU(2),RHO(2),GAMMA(2)
 REAL*8 E(5),V(5)
 REAL*8 ETA1(5),ETA2(5),NU1(5),NU2(5)
 COMMON /SIGNAL/R,U,Y,S,UC
 COMMON /CNTRL/A1,A2,B1,B2,ETA1,ETA2,NU1,NU2,NACT
 DATA K/5/

SHIFT VARIABLES TO ALLOW CALCULATION OF NEW VALUES

A1(K-2) = A1(K-1)
 A1(K-1) = A1(K)
 A2(K-2) = A2(K-1)
 A2(K-1) = A2(K)
 B1(K-2) = B1(K-1)
 B1(K-1) = B1(K)
 B2(K-2) = B2(K-1)
 B2(K-1) = B2(K)
 E(K-2) = E(K-1)
 E(K-1) = L(K)
 V(K-2) = V(K-1)
 V(K-1) = V(K)

IF(NACT.GE.2) GOTO 560

CALCULATIONS FOR ACTUATOR CONFIG. 1 AND 2:

E(K-1) = Y(K-1)-A1(K-1)*Y(K-2)-A2(K-2)*Y(K-3)
 * -B1(K-1)*UC(K-2)-B2(K-2)*UC(K-3)
 DENOM = H+MU(1)*Y(K-2)**2+MU(2)*Y(K-3)**2
 * +RHO(1)*UC(K-2)**2+RHO(2)*UC(K-3)**2
 GOTO 570

CALCULATIONS FOR ACTUATOR CONFIG. 3 AND 4

560 E(K-1) = Y(K-1)-A1(K-1)*Y(K-2)-A2(K-2)*Y(K-3)
 * -B1(K-1)*U(K-2)-B2(K-2)*U(K-3)
 DENOM = H+MU(1)*Y(K-2)**2+MU(2)*Y(K-3)**2
 * +RHO(1)*U(K-2)**2+RHO(2)*U(K-3)**2
 570 V(K-1) = L(K-1)/DENOM

CALCULATE NEW PLANT PARAMETER ESTIMATES

A1(K) = A1(K-1)+MU(1)*V(K-1)*Y(K-2)
 A2(K) = A2(K-1)+MU(2)*V(K-1)*Y(K-3)
 B1(K) = B1(K-1)+RHO(1)*V(K-1)*U(K-2)
 B2(K) = B2(K-1)+RHO(2)*V(K-1)*U(K-3)

UPDATE CONTROLLER PARAMETER ESTIMATES

CALL CNTSET(GAMMA)
 RETURN
 END

CC

GO1 ALGORITHM

EQUATIONS (4-61) - (4-70)

SUBROUTINE GO1(GAMMA,MU,RHL,F)
 IMPLICIT REAL*8 (A-H,O-Z)
 REAL*8 R(5),U(5),Y(5),UC(5),S(5),YHA1(5)
 REAL*8 A1(5),A2(5),B1(5),B2(5)
 REAL*8 MU(2),RHO(2),GAMMA(2)
 REAL*8 E(5),V(5),ETA1(5),ETA2(5),NU1(5),NU2(5)
 REAL*8 LAM1(5),LAM2(5),GAM1(5),GAM2(5)
 COMMON /SIGNAL/R,U,Y,S,UC
 COMMON /CNTRL/A1,A2,B1,B2,ETA1,ETA2,NU1,NU2,NACT
 COMMON /MISC/GAM1,GAM2,LAM1,LAM2,YHA1
 DATA K/5/

SHIFT VARIABLES TO ALLOW NEW VALUES TO BE CALCULATED

LAM1(K-3) = LAM1(K-2)
 LAM1(K-2) = LAM1(K-1)

```

LAM1(K-1) = LAM1(K)
LAM2(K-3) = LAM2(K-2)
LAM2(K-2) = LAM2(K-1)
LAM2(K-1) = LAM2(K)
GAM1(K-3) = GAM1(K-2)
GAM1(K-2) = GAM1(K-1)
GAM1(K-1) = GAM1(K)
GAM2(K-3) = GAM2(K-2)
GAM2(K-2) = GAM2(K-1)
GAM2(K-1) = GAM2(K)
A1(K-2) = A1(K-1)
A1(K-1) = A1(K)
A2(K-2) = A2(K-1)
A2(K-1) = A2(K)
B1(K-2) = B1(K-1)
B1(K-1) = B1(K)
B2(K-2) = B2(K-1)
B2(K-1) = B2(K)
YHAT(K-3) = YHAT(K-2)
YHAT(K-2) = YHAT(K-1)
YHAT(K-1) = YHAT(K)
E(K-2) = E(K-1)
E(K-1) = E(K)

```

```

IF (INACT.GT.2) GO TO 760

```

```

CALCULATIONS FOR ACTUATOR CONFIG. 1 AND 2:

```

```

YHAT(K-1) = A1(K-1)*YHAT(K-1)+A2(K-1)*YHAT(K-2)
* +B1(K-1)*UC(K-2)+B2(K-1)*UC(K-3)
GAM1(K-1) = UC(K-2)+A1(K-1)*GAM1(K-2)+A2(K-1)*GAM1(K-3)
GAM2(K-1) = UC(K-3)+A1(K-1)*GAM2(K-2)+A2(K-1)*GAM2(K-3)
GO TO 770

```

```

CALCULATIONS FOR ACTUATOR CONFIG. 3 AND 4:

```

```

760 YHAT(K-1) = A1(K-1)*YHAT(K-1)+A2(K-1)*YHAT(K-2)
* +B1(K-1)*U(K-2)+B2(K-1)*U(K-3)
GAM1(K-1) = U(K-2)+A1(K-1)*GAM1(K-2)+A2(K-1)*GAM1(K-3)
GAM2(K-1) = U(K-3)+A1(K-1)*GAM2(K-2)+A2(K-1)*GAM2(K-3)
770 LAM1(K-1) = YHAT(K-2)+A1(K-1)*LAM1(K-2)+A2(K-1)*LAM1(K-3)
LAM2(K-1) = YHAT(K-3)+A1(K-1)*LAM2(K-2)+A2(K-1)*LAM2(K-3)
E(K-1) = Y(K-1)-YHAT(K-1)
DENOM = H*MU(1)*LAM1(K-1)**2+MU(2)*LAM2(K-1)**2
* +RHU(1)*GAM1(K-1)**2+RHU(2)*GAM2(K-1)**2
IF (E(K-1).LT.1.0D-10) E(K-1) = 0.0D0
IF (DENOM.GT.1.0D05) DENOM = 1.0D05
V(K-1) = E(K-1)/DENOM

```

```

CALCULATE NEW PLANT PARAMETER ESTIMATES

```

```

A1(K) = A1(K-1)+MU(1)*V(K-1)*LAM1(K-1)
A2(K) = A2(K-1)+MU(2)*V(K-1)*LAM2(K-1)
B1(K) = B1(K-1)+RHU(1)*V(K-1)*GAM1(K-1)
B2(K) = B2(K-1)+RHU(2)*V(K-1)*GAM2(K-1)

```

```

CALCULATE NEW CONTROLLER PARAMETERS

```

```

CALL CNTSET(GAMMA)
RETURN
END

```

```

CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC

```

```

SOI ALGORITHM

```

```

EQUATIONS (4-76) - (4-81)

```

```

SUBROUTINE SOI(GAMMA,MU,RHU,H,Q)

```

```

IMPLICIT REAL*8 (A-H,O-Z)

```

```

REAL*8 R(5),U(5),Y(5),S(5),YHAT(5),UC(5)

```

```

REAL*8 GAMMA(2),MU(5),RHU(5),Q(2)

```

```

REAL*8 A1(5),A2(5),B1(5),B2(5)

```

```

REAL*8 V(5),Z(5),ETA1(5),ETA2(5),NU1(5),NU2(5)

```

```

COMMON /SIGNAL/R,U,Y,S,UC

```

```

COMMON /CTRLR/A1,A2,B1,B2,ETA1,ETA2,NU1,NU2,NACT

```

```

COMMON /MISC/Z,YHAT,Z,V

```

```

DATA K/5/

```

```

SHIFT VARIABLES TO ALLOW CALCULATION OF NEW VALUES

```

```

Z(K-3) = Z(K-2)

```

```

Z(K-2) = Z(K-1)

```

```

Z(K-1) = Z(K)

```

```

A1(K-2) = A1(K-1)

```

```

A1(K-1) = A1(K)

```

```

A2(K-2) = A2(K-1)
A2(K-1) = A2(K)
U1(K-2) = U1(K-1)
U1(K-1) = U1(K)
U2(K-2) = U2(K-1)
U2(K-1) = U2(K)
YHAT(K-2) = YHAT(K-1)
YHAT(K-1) = YHAT(K)
V(K-2) = V(K-1)
V(K-1) = V(K)

```

```
IF (NACT.GT.2) GOTO 660
```

```

CALCULATIONS FOR ACTUATOR CONFIG. 1 AND 2:
YHAT(K-1) = A1(K-1)*Z(K-2)+A2(K-1)*Z(K-3)
* B1(K-1)*UC(K-2)+B2(K-1)*UC(K-3)
DENOM = 1+MU(1)*Z(K-2)**2+MU(2)*Z(K-3)**2
* RHO(1)*UC(K-2)**2+RHO(2)*UC(K-3)**2
GOTO 670

```

```

CALCULATIONS FOR ACTUATOR CONFIG. 3 AND 4
660 YHAT(K-1) = A1(K-1)*Z(K-2)+A2(K-1)*Z(K-3)
* B1(K-1)*U(K-2)+B2(K-1)*U(K-3)
DENOM = 1+MU(1)*Z(K-2)**2+MU(2)*Z(K-3)**2
* RHO(1)*U(K-2)**2+RHO(2)*U(K-3)**2
670 V(K-1) = (Y(K-1)-YHAT(K-1)+U(1)*(Y(K-2)-Z(K-2))
* U(2)*(Y(K-3)-Z(K-3)))/DENOM

```

```

CALCULATE NEW PLANT PARAMETER ESTIMATES
A1(K) = A1(K-1)+MU(1)*V(K-1)*Z(K-2)
A2(K) = A2(K-1)+MU(2)*V(K-1)*Z(K-3)
U1(K) = U1(K-1)+RHO(1)*V(K-1)*U(K-2)
U2(K) = U2(K-1)+RHO(2)*V(K-1)*U(K-3)

```

```
IF (NACT.GT.2) GOTO 680
```

```
Z(K-1) = A1(K)*Z(K-2)+A2(K)*Z(K-3)+B1(K)*UC(K-2)+B2(K)*UC(K-3)
GOTO 690
```

```
680 Z(K-1) = A1(K)*Z(K-2)+A2(K)*Z(K-3)
* B1(K)*U(K-2)+B2(K)*U(K-3)
```

```
690 CONTINUE
```

```

CALCULATE NEW CONTROLLER PARAMETERS
CALL CNIS1(GAMMA)
RETURN
END

```

```

SUBROUTINE STAT2(N,KOUNT)
IMPLICIT REAL*8 (A-H,O-Z)
REAL*8 R(5),U(5),Y(5),S(5),UC(5)
REAL*8 IDO(10),IO(10),NITE(10),SAS1(10),
* SAPEA1(10),SAPEA2(10),SAPEB1(10),SAPEB2(10),
* SPEVA1(10),SPEVA2(10),SPEVB1(10),SPEVB2(10),
* SACEE1(10),SACEE2(10),SACEN1(10),SACEN2(10),
* SCEVE1(10),SCEVE2(10),SCEVN1(10),SCEVN2(10),
* NASTE(10),NSTEV(10)
REAL*8 A1(5),A2(5),B1(5),B2(5),ETA1(5),ETA2(5),NU1(5),NU2(5)
INTEGER KD(12)
COMMON /STATS/IDO,IO,NITE,SAS1,SAPEA1,SAPEA2,SAPLB1,SAPEB2,
* SPEVA1,SPEVA2,SPEVB1,SPEVB2,SACEE1,SACEE2,SACEN1,SACEN2,
* SCEVE1,SCEVE2,SCEVN1,SCEVN2,
* NASTE,NSTEV,STE,
* SSPA1,SSPA2,SSPB1,SSPB2,SSCE1,SSCE2,SSCN1,SSCN2,M,KD
COMMON /CONTRL/A1,A2,B1,B2,ETA1,ETA2,NU1,NU2,NACT
COMMON /SIGNAL/R,U,Y,S,UC
DATA K/5/

```

```
IF (DABS(S(K)).LT.1.0D-03) GOTO 100
```

```
TE = (S(K)-Y(K))/S(K)
```

```
GOTO 200
```

```
100 TE = 0.0
```

```
KOUNT = KOUNT+1
```

```
200 NASTE(M) = NASTE(M)+TE**2
```

```
STE = STE+TE
```

```
SAS1(M) = SAS1(M)+UC(K-1)**2
```

```
SAPEA1(M) = SAPEA1(M)+A1(K-1)
```

```
SAPEA2(M) = SAPEA2(M)+A2(K-1)
```

```
SAPEB1(M) = SAPEB1(M)+B1(K-1)
```

```
SAPEB2(M) = SAPEB2(M)+B2(K-1)
```

```

SACEE1(M) = SACEE1(M)+ETA1(K-1)
SACEE2(M) = SACEE2(M)+ETA2(K-1)
SACEN1(M) = SACEN1(M)+NU1(K-1)
SACEN2(M) = SACEN2(M)+NU2(K-1)
SSPA1 = SSPA1+A1(K-1)**2
SSPA2 = SSPA2+A2(K-1)**2
SSPB1 = SSPB1+B1(K-1)**2
SSPB2 = SSPB2+B2(K-1)**2
SSCE1 = SSCE1+ETA1(K-1)**2
SSCE2 = SSCE2+ETA2(K-1)**2
SSCN1 = SSCN1+NU1(K-1)**2
SSCN2 = SSCN2+NU2(K-1)**2

```

C

```

300 TIME = FLOATIN)
IF (N.LT.KD(M+1)) RETURN
IDD(M) = S(K)
IO(M) = Y(K)
RANGE = FLOAT(KD(M+1)-KD(M))
DIV = (RANGE-FLOAT(KOUNT))*(RANGE-FLOAT(KOUNT)-1)
RNG = RANGE+FLOAT(KOUNT)
NSTEV(M) = 0.0
IF (DIV.GT.0.0) NSTEV(M) = (RNG*NASTE(M)-STE**2)/DIV
IF (NSTEV(M).LT.0.000) NSTEV(M) = 0.000
NITE(M) = TE
DIVSOR = RANGE*(RANGE-1)
IF (DIVSOR.LE.0.0) GOTU 1000
SPEVA1(M) = (RANGE*SSPA1-SAPEA1(M)**2)/DIVSOR
IF (SPEVA1(M).LT.0.000) SPEVA1(M) = 0.000
SPEVA2(M) = (RANGE*SSPA2-SAPEA2(M)**2)/DIVSOR
IF (SPEVA2(M).LT.0.000) SPEVA2(M) = 0.000
SPEVB1(M) = (RANGE*SSPB1-SAPEB1(M)**2)/DIVSOR
IF (SPEVB1(M).LT.0.000) SPEVB1(M) = 0.000
SPEVB2(M) = (RANGE*SSPB2-SAPEB2(M)**2)/DIVSOR
IF (SPEVB2(M).LT.0.000) SPEVB2(M) = 0.000

```

C

```

SCEVE1(M) = (RANGE*SSCE1-SACEE1(M)**2)/DIVSOR
IF (SCEVE1(M).LT.0.000) SCEVE1(M) = 0.000
SCEVE2(M) = (RANGE*SSCE2-SACEE2(M)**2)/DIVSOR
IF (SCEVE2(M).LT.0.000) SCEVE2(M) = 0.000
SCEVN1(M) = (RANGE*SSCN1-SACEN1(M)**2)/DIVSOR
IF (SCEVN1(M).LT.0.000) SCEVN1(M) = 0.000
SCEVN2(M) = (RANGE*SSCN2-SACEN2(M)**2)/DIVSOR
IF (SCEVN2(M).LT.0.000) SCEVN2(M) = 0.000
GOTO 2000
1000 SPEVA1(M) = 0.000
SPEVA2(M) = 0.000
SPEVB1(M) = 0.000
SPEVB2(M) = 0.000
SCEVE1(M) = 0.000
SCEVE2(M) = 0.000
SCEVN1(M) = 0.000
SCEVN2(M) = 0.000
2000 CONTINUE
SAS1(M) = SAS1(M)/RANGE
SAPEA1(M) = SAPEA1(M)/RANGE
SAPEA2(M) = SAPEA2(M)/RANGE
SAPEB1(M) = SAPEB1(M)/RANGE
SAPEB2(M) = SAPEB2(M)/RANGE
SACEE1(M) = SACEE1(M)/RANGE
SACEE2(M) = SACEE2(M)/RANGE
SACEN1(M) = SACEN1(M)/RANGE
SACEN2(M) = SACEN2(M)/RANGE
NASTE(M) = NASTE(M)/RANGE
KOUNT = 0
STE = 0.000
SSPA1 = 0.000
SSPA2 = 0.000
SSPB1 = 0.000
SSPB2 = 0.000
SSCE1 = 0.000
SSCE2 = 0.000
SSCN1 = 0.000
SSCN2 = 0.000
M = M+1
RETURN
END

```

CC

SUBROUTINE - CONTROLLER SETTINGS

EQUATIONS (4-89) TO (4-92)

SUBROUTINE CNTSET(GAMMA)
 IMPLICIT REAL*8 (A-H,I-Z)
 REAL*8 A1(5),A2(5),B1(5),B2(5)
 REAL*8 ETA1(5),ETA2(5),NU1(5),NU2(5)
 REAL*8 GAMMA(2)

COMMON /CONTRL/A1,A2,B1,B2,ETA1,ETA2,NU1,NU2,NACT
 DATA K/5/

CALCULATE CONTROLLER COEFFICIENTS

ETA1(K-1) = GAMMA(1)-A1(K)

NU1(K-1) = (ETA1(K-1)*A2(K)+(ETA1(K-1)*A1(K)-A2(K)+

* GAMMA(2))*(A1(K)-A2(K)+B1(K)/B2(K))/
 * (B1(K)+A1(K)+B2(K)-A2(K)+B1(K)*2/B2(K))

ETA2(K-1) = A1(K)+ETA1(K-1)+GAMMA(2)-B1(K)+NU1(K-1)-A2(K)

NU2(K-1) = ETA2(K-1)+A2(K)/B2(K)

RETURN
 END