

N90-10108

**NUMERICALLY EFFICIENT ALGORITHM FOR MODEL DEVELOPMENT  
OF HIGH-ORDER SYSTEMS**

By

L. O. Parada  
Calspan Advanced Technology Center  
Buffalo, New York

**ABSTRACT**

Frequency domain parameter identification techniques provide a straightforward approach to transfer function estimation. However, for high-order systems, numerical difficulties may be encountered during the estimation process. Inaccuracies may result because of the large variation of the transfer function polynomial coefficients for high-order systems. The lack of numerical precision to represent this variation may cause the estimation process to break down.

This paper presents a technique for estimating transfer functions in partial fraction expansion form from frequency response data for a high-order system. The problem formulation avoids many of the numerical difficulties associated with high-order polynomials and has the advantage of having the option to fix the damping and frequency of a mode, if known, during the estimation process. The resulting transfer function(s) may be converted to Jordan-Form time domain equations directly.

During the implementation of this technique, a frequency and amplitude normalizing window was developed that maximized the efficiency of the optimization algorithm. The combination of estimating the transfer function in factored form, the ability to fix previously determined parameters and the effectiveness of the normalizing window led to a progressive approach to synthesizing transfer functions from frequency response data for high-order systems.

**PRECEDING PAGE BLANK NOT FILMED**

633

## Abstract

### Numerically Efficient Algorithm for Model Development of High Order Systems

L. O. Parada  
Calspan Advanced Technology Center  
P.O. Box 400  
Buffalo, NY 14225  
(716) 632-7500

for presentation at the

NASA Langley Research Center Workshop on  
Computational Aspects in the Control of Flexible Structures

Frequency domain parameter identification techniques provide a straightforward approach to transfer function estimation. However, for high order systems, numerical difficulties may be encountered during the estimation process. Inaccuracies may result because of the large variation of the transfer function polynomial coefficients for high order systems. The lack of numerical precision to represent this variation may cause the estimation process to break down.

This paper presents a technique for estimating transfer functions in partial fraction expansion form from frequency response data for a high order system. The problem formulation avoids many of the numerical difficulties associated with high order polynomials and has the advantage of having the option to fix the damping and frequency of a mode, if known, during the estimation process. The resulting transfer function(s) may be converted to Jordan-Form time domain equations directly.

During the implementation of this technique, a frequency and amplitude normalizing window was developed that maximized the efficiency of the optimization algorithm. The combination of estimating the transfer function in factored form, the ability to fix previously determined parameters and the effectiveness of the normalizing window led to a progressive approach to synthesizing transfer functions from frequency response data for high order systems.

635

PRECEDING PAGE BLANK NOT FILMED

# NUMERICALLY EFFICIENT ALGORITHM FOR MODEL DEVELOPMENT OF HIGH ORDER SYSTEMS

Statement of Problem

Development of Mathematical Models:

- Time Domain – Difficult to implement
  - Instrumentation Complement
  - Input Design
  - Noise
  - Computational Load
- Freq Domain – Simplified Implementation
  - Fewer parameters per computation cycle
  - Statistical methods applicable

## PREVIOUS WORK

Frequency domain parameter identification requires

Determination of characteristic equation  
(nonlinear or iterative techniques)

Estimation of numerator polynomials

Factor characteristic equation

Estimate zeros or residues

Inaccuracies (for high order systems) due to:

Variation of transfer function polynomial coefficients

Transformation errors

Sensitivity of polynomial roots to variations in polynomial coefficients

## SUMMARY OF CONTRIBUTIONS

- Development of technique to estimate transfer functions in partial fraction expansion form from frequency response (amplitude and phase) data
- Elimination of numerical difficulties associated with high order polynomials
- Incorporation of a priori knowledge of system modes (frequency and damping) directly into the estimation process
- Development of frequency and amplitude normalizing window that maximizes effectiveness of the optimization algorithm and eliminates the initial guess problem
- Stepwise approach for synthesizing transfer functions where order of system is high and unknown

## FACTORED FORM ESTIMATION

Classical Nonlinear Regression Problem

Estimate parameters from measured amplitude and phase data

Error Function:

Square of distance between measured and estimated frequency responses summed over all discrete frequency points

$$\epsilon = \sum_{i=1}^M [ F(j\omega_i) - G(j\omega_i) ]^2$$

where: M = # frequency points

F(j $\omega$ ) = measured frequency response

G(j $\omega$ ) = estimated frequency response

Estimated Transfer Function - G(j $\omega$ )

Sum of 1st and 2nd order terms

$$G(j\omega) = \frac{a_N}{b_N} + \sum_{k=1}^Q \frac{n_{1k}(j\omega) + N_{0k}}{(j\omega)^2 + d_{1k}(j\omega) + d_{0k}} + \sum_{\ell=1}^{N-2Q} \frac{a_{\ell}}{(j\omega) + b_{\ell}}$$

where: N = order of system

Q = # of second order terms

Express measured and estimated frequency responses in terms of real and imaginary components

$$F(j\omega) = R(\omega) + jI(\omega)$$

$$G(j\omega) = \frac{a_N}{b_N} + \sum_{k=1}^Q \frac{N_{0k}(d_{0k} - \omega^2) + N_{1k}d_{1k}^2}{(d_{0k} - \omega^2)^2 + d_{1k}^2\omega^2} + \sum_{\ell=1}^{N-2Q} \frac{a_\ell b_\ell}{b_\ell^2 + \omega^2}$$

$$\left[ +j \right] \sum_{k=1}^Q \frac{N_{1k}\omega(d_{0k} - \omega^2) - N_{0k}d_{1k}\omega}{(d_{0k} - \omega^2)^2 + d_{1k}^2\omega^2} - \sum_{\ell=1}^{N-2Q} \frac{a_\ell\omega}{b_\ell^2 + \omega^2}$$

640

Substitute for F(jω) & G(jω) into ε

$$\epsilon = \sum_{i=1}^M \left[ R(\omega_i) - \frac{a_N}{b_N} - \sum_{k=1}^Q \frac{N_{0k}(d_{0k} - \omega_i^2) + N_{1k}d_{1k}^2\omega_i^2}{(d_{0k} - \omega_i^2)^2 + d_{1k}^2\omega_i^2} + \sum_{\ell=1}^{N-2Q} \frac{a_\ell b_\ell}{b_\ell^2 + \omega_i^2} \right]$$

$$+ \left[ I(\omega_i) + \sum_{k=1}^Q \frac{N_{0k}d_{1k}\omega_i - N_{1k}\omega_i(d_{0k} - \omega_i^2)}{(d_{0k} - \omega_i^2)^2 + d_{1k}^2\omega_i^2} + \sum_{\ell=1}^{N-2Q} \frac{a_\ell\omega_i}{b_\ell^2 + \omega_i^2} \right]$$

Solve for unknown parameters

Set partial derivatives equal to zero

Solve using nonlinear optimization technique

## Fifth Order Single Precision Example

Simulate parameter identification of high order system

Modes distributed over wide frequency range

Single precision: Scale down problem  
Reduce number of variables

5th Order Transfer Function:

Cascade form

$$\frac{(s + 5 \times 10^{-3})(s + 5 \times 10^{-1})(s + 5 \times 10^{+3})}{(s^2 + 2 \times 10^{-3}s + 10^{-4})(s + 1)(s^2 + 1 \times 10^{+4}s + 10^{+8})}$$

Parallel form

$$\frac{1.253955 \times 10^{-3}s + 6.122844 \times 10^{-6}}{s^2 + 2 \times 10^{-3}s + 1 \times 10^{-4}} + \frac{1.221073 \times 10^{-3}}{s + 1} + \frac{9.975250 \times 10^{-1}s + 5.024754 \times 10^{+3}}{s^2 + 1 \times 10^{+4}s + 10^{+8}}$$

Frequency Range:  $1 \times 10^{-4}$  to  $1 \times 10^{+5}$  Hz.

# Points/Decade = 30



## DENOMINATOR COEFFICIENTS

<u>Term</u>	<u>Exact Coefficient</u>	<u>Additive Components</u>
$s^5$	1.000000	1.0
$s^4$	10001.002	$1.002 + 1.0 \times 10^4$
$s^3$	10001002.0021	$2.1 \times 10^{-3} + 1.002 \times 10^4 + 10^8$
$s^2$	10020021.0001	$1 \times 10^{-4} + 2.1 \times 10^{+1} + 1.002 \times 10^8$
$s^1$	210001.0	$1.0 + 2.1 \times 10^{+5}$
$s^0$	100004.0	$1.0 \times 10^{+4}$

= Single Precision Variable Representation

# LINEARIZED APPROACH

Initial Error Function:

$$E_k = F(j\omega_k) - \frac{P(j\omega_k)}{Q(j\omega_k)}$$

where:  $F(j\omega_k)$  = measured frequency response at  $\omega_k$

$P(j\omega_k)$  = estimated numerator polynomial at  $\omega_k$

$Q(j\omega_k)$  = estimated denominator polynomial at  $\omega_k$

Weighted Error Function:

$$E'_k = E_k Q(j\omega_k) = F(j\omega_k) Q(j\omega_k) - P(j\omega_k)$$

Iterative Error Function:

$$E''_k = \frac{E_k Q(j\omega_k)^L}{Q(j\omega_k)^{L-1}} = \frac{F(j\omega_k) Q(j\omega_k)^L}{Q(j\omega_k)^{L-1}} - \frac{P(j\omega_k)^L}{Q(j\omega_k)^{L-1}}$$

where:  $L$  = iteration #

Minimize  $E''_k$  by taking partial derivatives of  $E''_k$  with respect to each parameter  $x_i$

$$\frac{\partial E''_k}{\partial x_i} = 0$$

Rearrange equations to formulate problem as a set of linear simultaneous algebraic equations:

$$[A] [X] = [B]$$

Solve for parameter vector  $[x]$

Iterations converge to minimization of  $|E_k|^2$

# 5th Order example: Polynomial Results

Exact  
Transfer  
Function

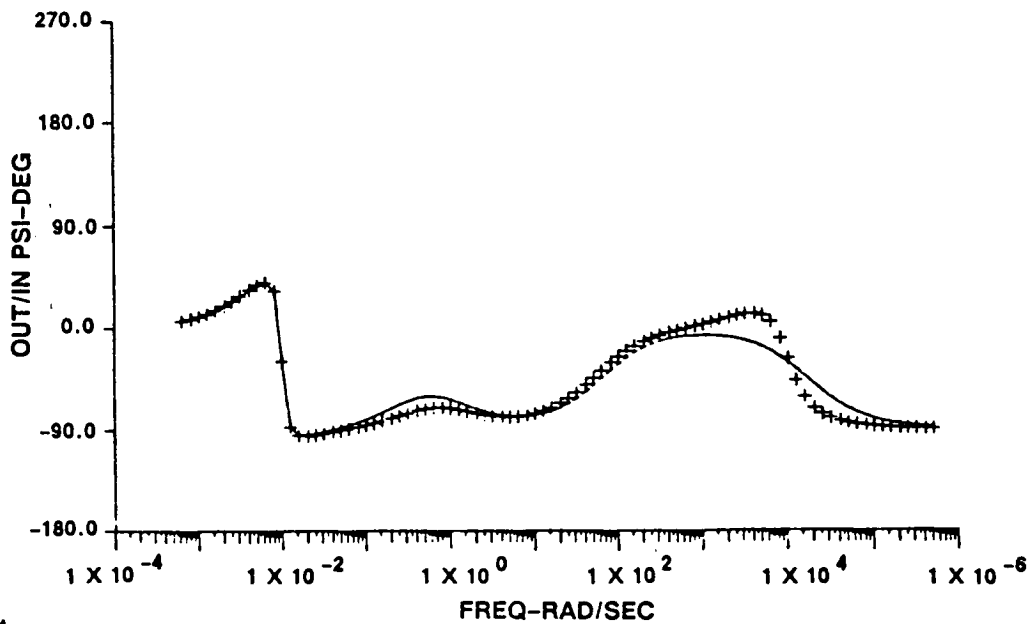
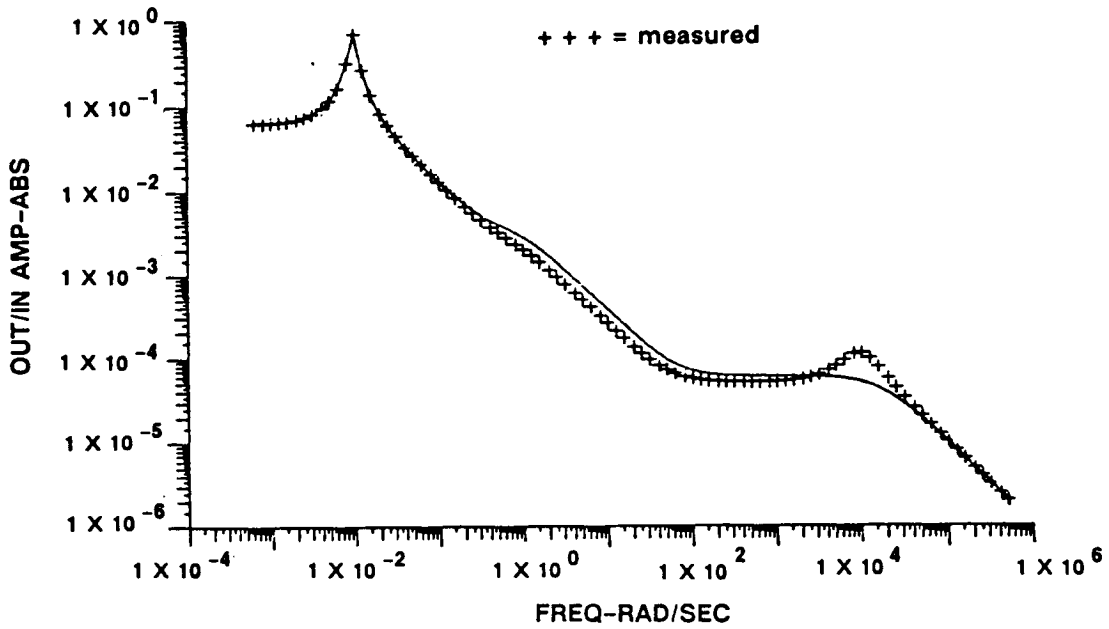
$$\frac{(s + 5 \times 10^{-3}) (s + 5 \times 10^{-1}) (s + 5 \times 10^{+1}) (s + 5 \times 10^{+3})}{(s^2 + 2 \times 10^{-3}s + 10^{-4})(s + 1) (s^2 + 1 \times 10^{+4} s + 10^{+8})}$$

Linear  
Results

$$\frac{(s + 5 \times 10^{-3}) (s + 5 \times 10^{-1}) (s + 4.24 \times 10^{+1}) (s - 7.17 \times 10^{+3})}{(s^2 + 2 \times 10^{-3} s + 10^{-4}) (s + 9.98 \times 10^{-1}) (s - 7.17 \times 10^{+3}) (s + 1.72 \times 10^{+4})}$$

Cost  
Function

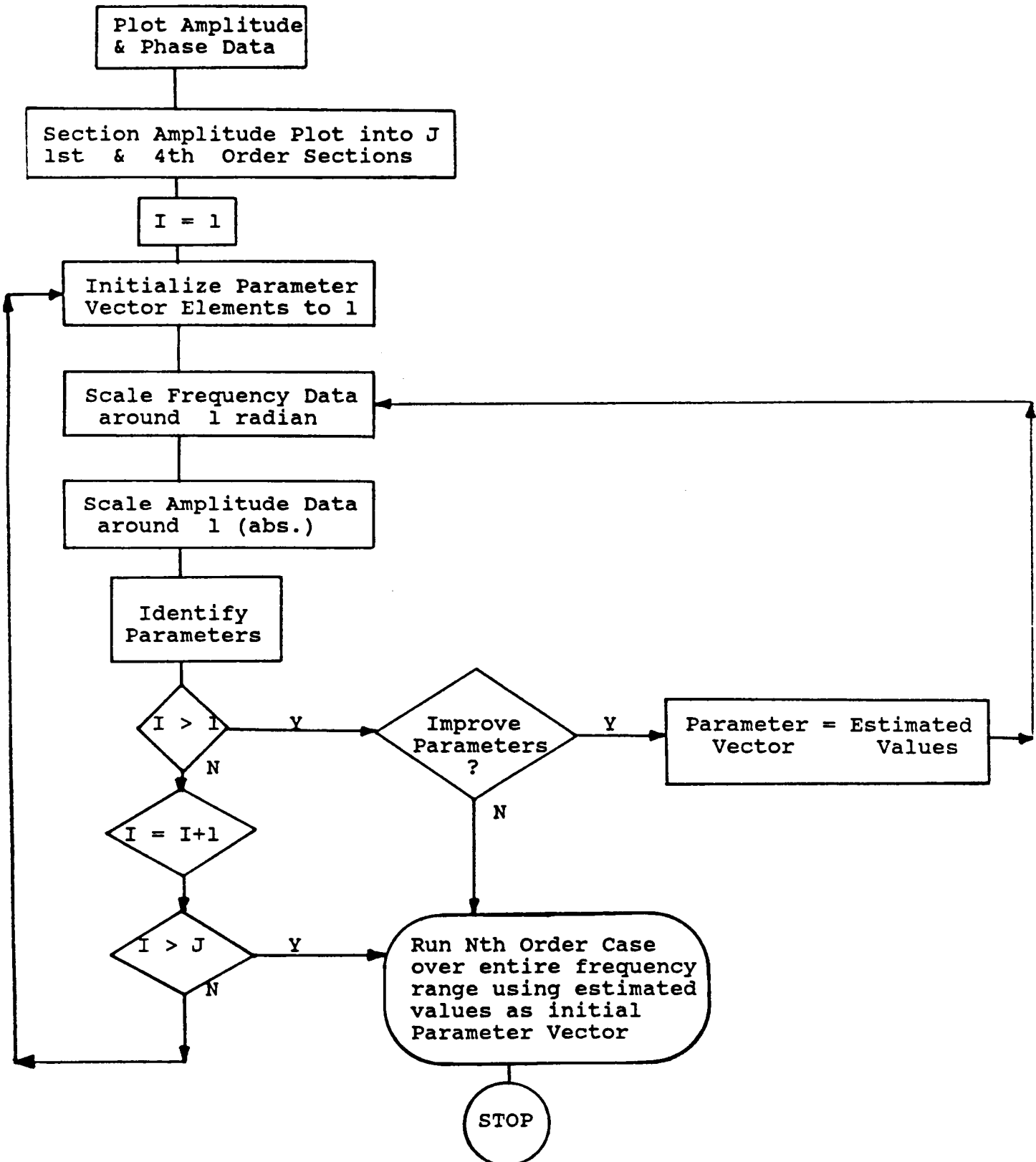
$$2.37 \times 10^{-10}$$



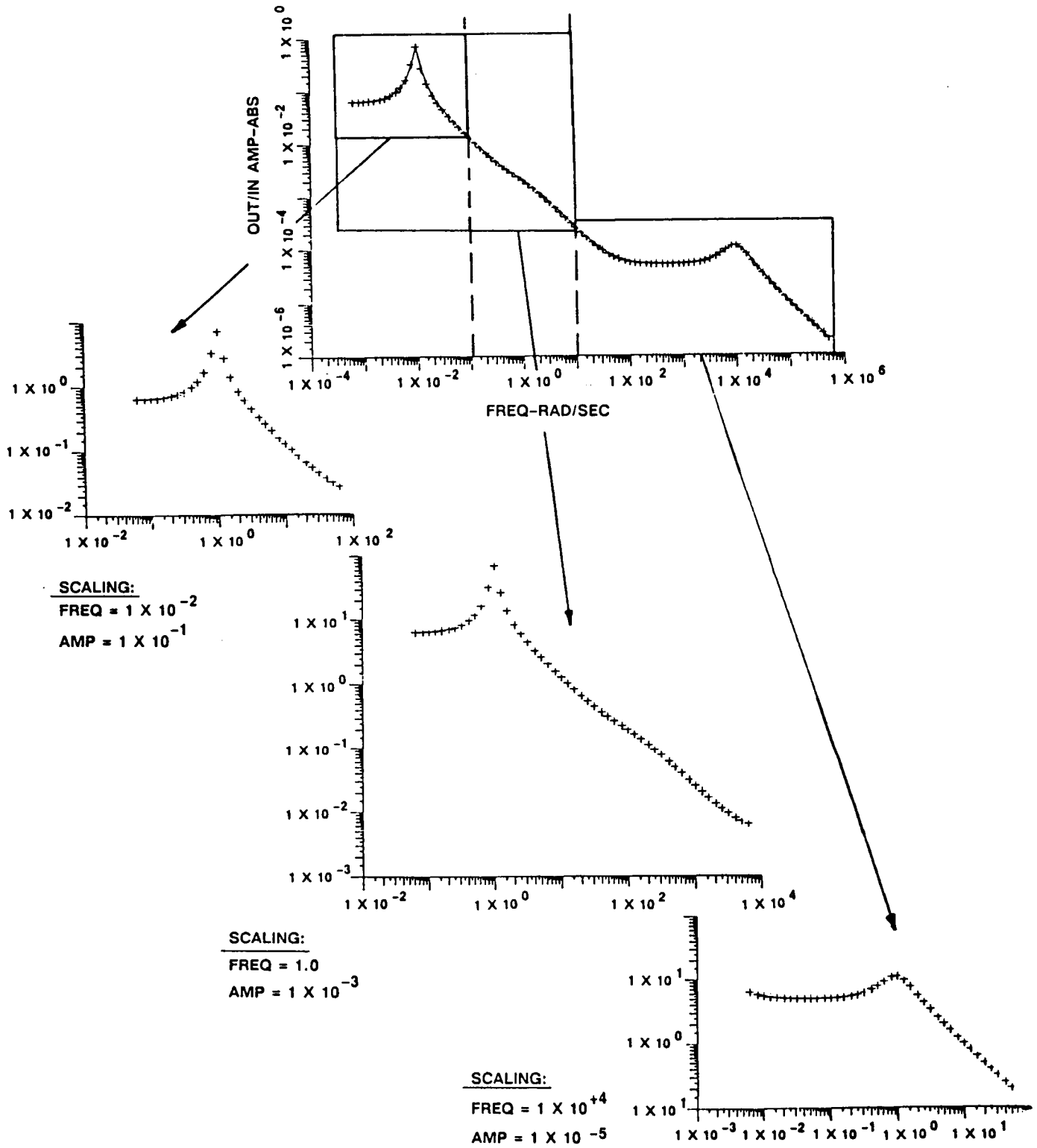
# FACTORED FORM APPROACH

- Nonlinear method of solution – strong initial guess required
- Initial Guess – Examine bode plots to approximate frequency and damping of modes
- Problem – Error function relatively insensitive to perturbations in parameters of high frequency modes
  - Gradient expressions small compared to those of the lower frequency parameters
- Solution – Normalizing window
- Scale Data Such That:
  - Frequency is centered around 1 rad.
  - Amplitude is centered around 1 (abs. units)
- Benefits – Need for strong initial guess eliminated
  - effectiveness of optimization algorithm maximized

STEPWISE FACTORED FORM TECHNIQUE



# 5th Order Example: Factored Form Approach



### 5th Order Example: Factored Form Results

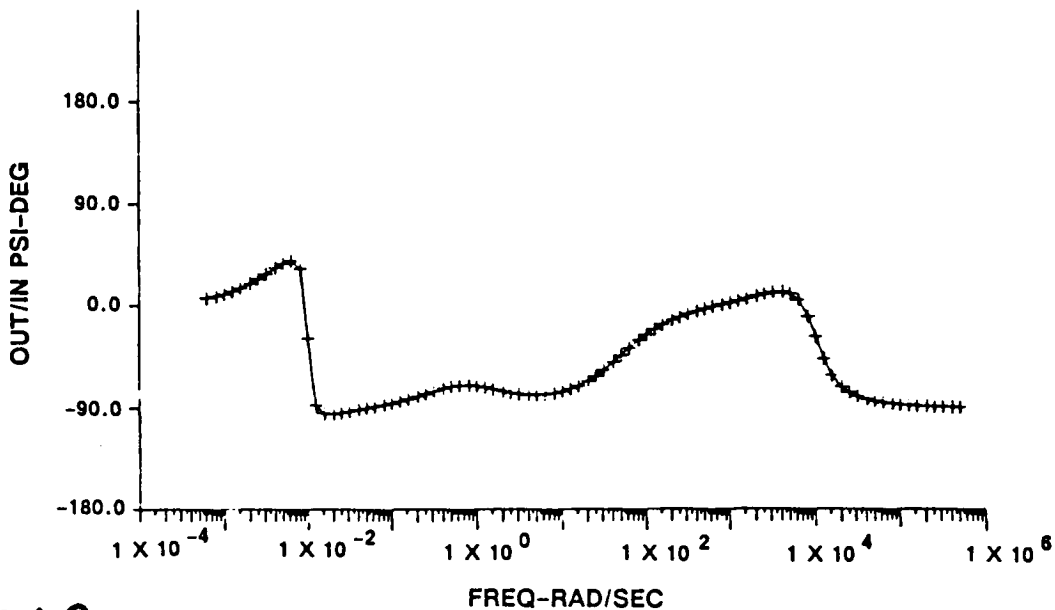
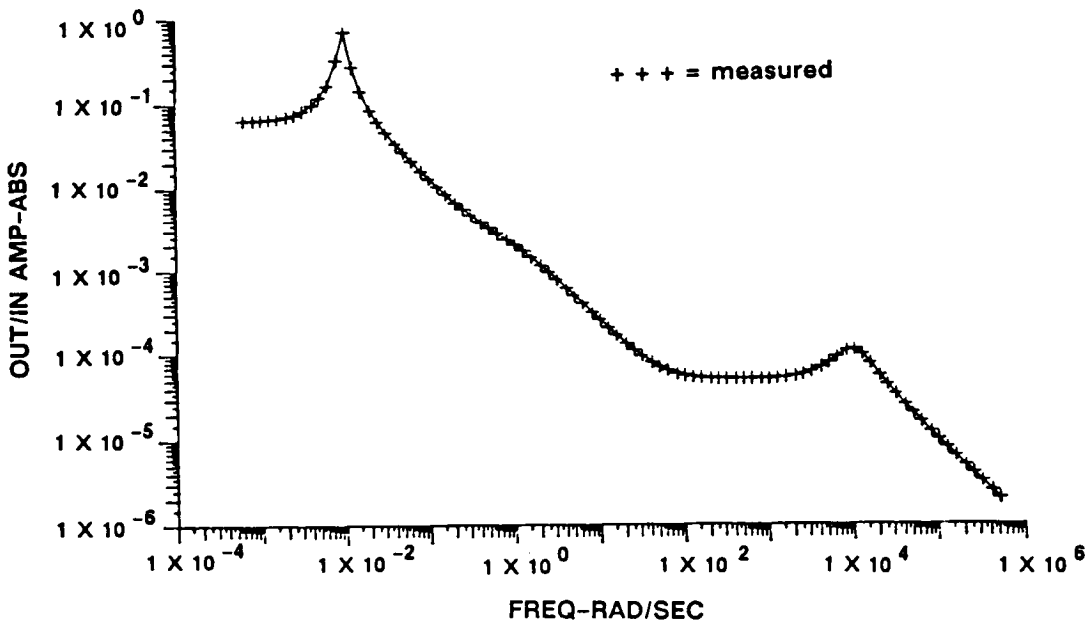
Exact Transfer Function

$$\frac{1.25 \times 10^{-3} s + 6.12 \times 10^{-6}}{s^2 + 2 \times 10^{-3} s + 10^{-4}} + \frac{1.22 \times 10^{-3}}{s + 1} + \frac{9.98 \times 10^{-1} s + 5.02 \times 10^3}{s^2 + 1 \times 10^4 s + 10^8}$$

Factored Form Results

$$\frac{1.25 \times 10^{-3} s + 6.12 \times 10^{-6}}{s^2 + 2.00 \times 10^{-3} s + 10^{-4}} + \frac{1.23 \times 10^{-3}}{s + 1.01} + \frac{9.79 \times 10^{-1} s + 5.21 \times 10^3}{s^2 + 9.95 \times 10^3 s + 1.03 \times 10^8}$$

Cost Function  $1.90 \times 10^{-12}$



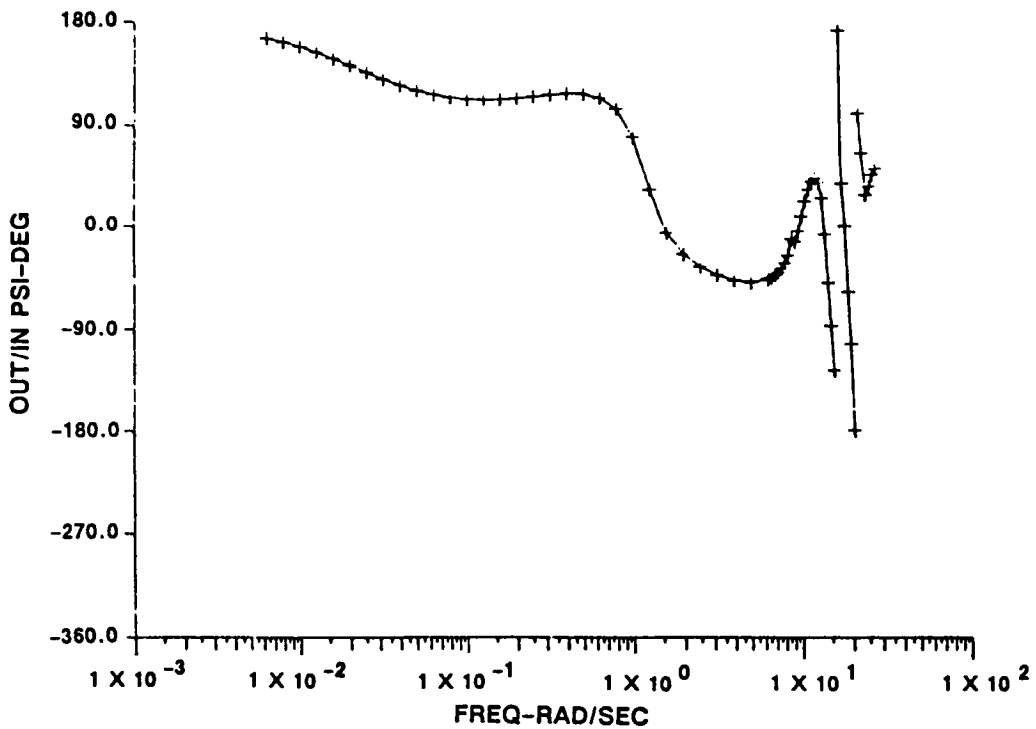
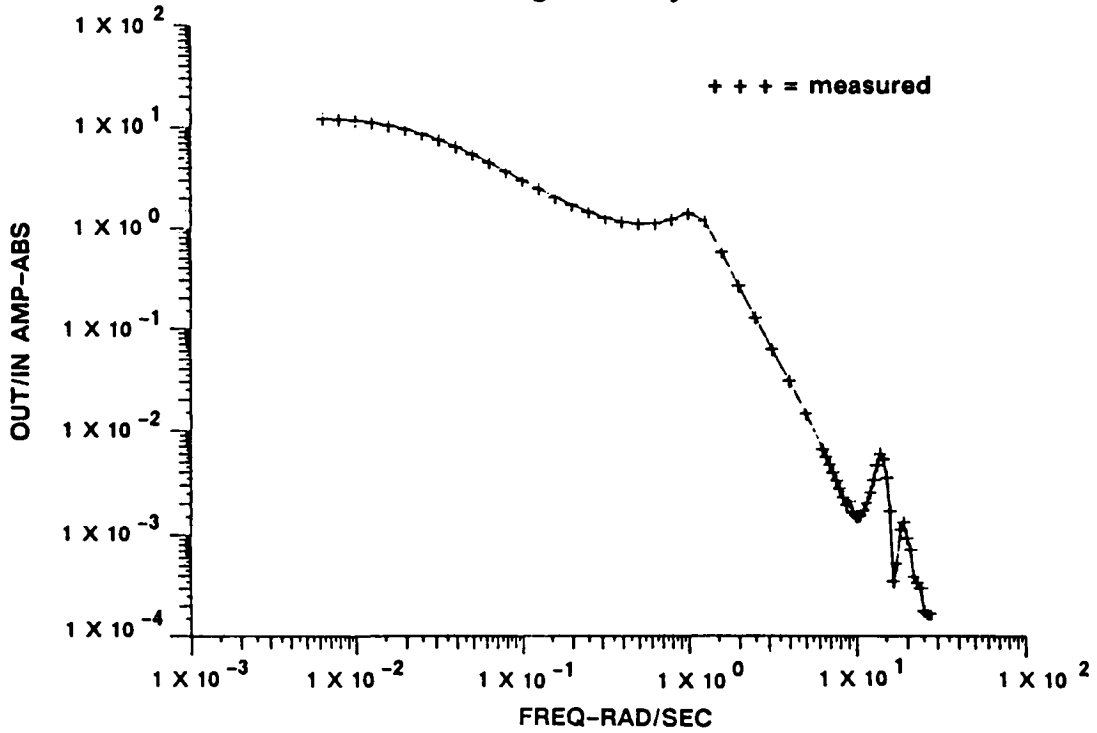
648

# 16th Order Transfer Function Estimation

Pcg/Vg: Roll Rate Measured at C.G. of Aircraft

vs.

Unit Gust Along Y-Body Axis



Cost Function:  $4.5 \times 10^{-12}$

649



## CONCLUSIONS

Development of technique to estimate transfer functions directly in factored form

### Advantages:

Ability to fix damping and frequency of a mode, if known, during the estimation process

Avoidance of numerical difficulties associated with high order polynomials

Ability to obtain Jordan-form time domain equations directly

Progressive approach to transfer function estimation through use of a frequency and amplitude normalizing window

Development of frequency and amplitude normalizing window that eliminates the initial guess problem and maximizes the effectiveness of the optimization algorithm