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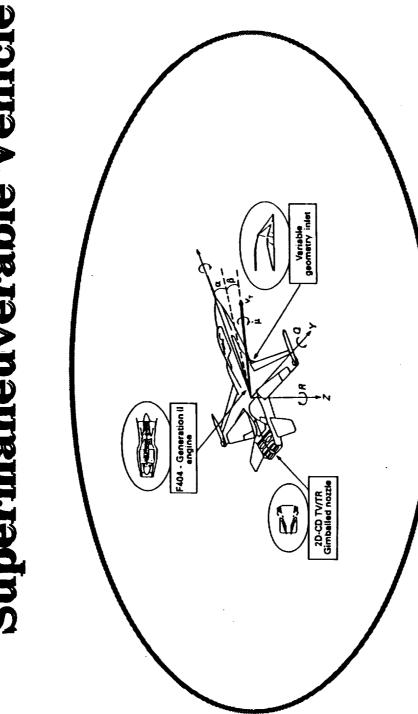
Introduction

Key Technical Challenges:

High Angle-of-Attack Flight Flying Qualities Robustness Multiple/Redundant Control Effectors

Solutions:

Control Selector/Command Limiting/Daisy Chaining Structured Singular Value Synthesis Dynamic Inversion



Supermaneuverable Vehicle

Equations of Motion

rotational dynamics of the aircraft in the body axis

$$\dot{\mathbf{q}} = \mathbf{I}/\mathbf{I}_{y} \left[m_{\text{aero}} + m_{\text{thrust}} + \mathbf{pr} \left(\mathbf{I}_{z} - \mathbf{I}_{x} \right) + \mathbf{I}_{xz} (\mathbf{r}^{2} - \mathbf{p}^{2}) \right]$$
$$\left[\dot{\mathbf{p}} \right]_{z} \left[\mathbf{I}_{x} - \mathbf{I}_{xz} \right]_{z} \left[l_{\text{aero}} + l_{\text{thrust}} + \mathbf{pqI}_{xz} + \mathbf{qr} \left(\mathbf{I}_{y} - \mathbf{I}_{z} \right) \right]$$
$$\left[\dot{\mathbf{p}} \right]_{z} \left[-\mathbf{I}_{xz} \mathbf{I}_{z} \right]_{z} \left[n_{\text{aero}} + n_{\text{thrust}} - \mathbf{qrI}_{xz} + \mathbf{pq} \left(\mathbf{I}_{x} - \mathbf{I}_{y} \right) \right]$$

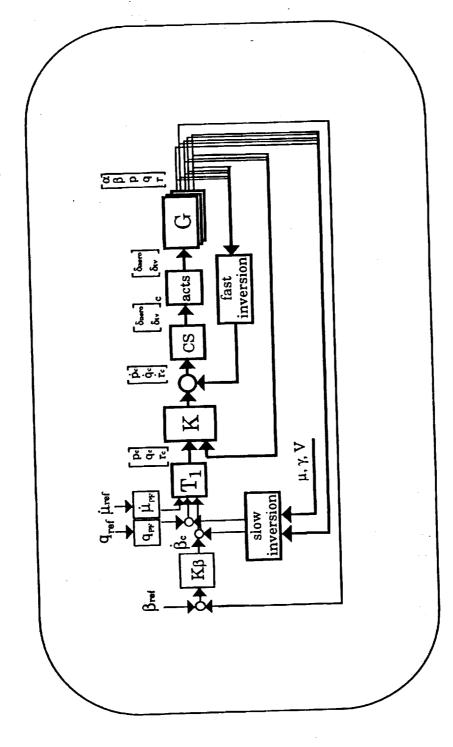
aircraft motion with respect to its velocity vector

 $\dot{\alpha} = q - (p \cos \alpha + r \sin \alpha) \tan \beta - (\cosh \beta)\dot{\gamma} - \sinh(\cos \beta)\dot{\chi}$

 $\dot{\beta} = p \sin \alpha - r \cos \alpha - \sinh \dot{\gamma} + \cosh \cos \dot{\chi}$

 $\dot{\mu} = (\cos\alpha/\cos\beta)p + (\sin\alpha/\cos\beta)r + \tan\beta\cos\mu \gamma + [\sin\gamma + \tan\beta \sinh \cos\gamma] \chi$

Control Law Structure



Fast Inversion

the rotational dynamics are broken into a linear parameter varying part and a nonlinear part $\begin{bmatrix} \dot{p} \\ \dot{q} \end{bmatrix} = \begin{bmatrix} 0 & L_{\mu} & L_{\nu} & 0 & L_{r} \\ M_{\alpha} & 0 & 0 & M_{q} & 0 \\ \dot{r} \end{bmatrix} \begin{bmatrix} \alpha & 1 & 0 & -I_{xz} \\ 0 & N_{\mu} & N_{\nu} & 0 & N_{r} \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \\ 0 \\ -I_{xz} & 0 \end{bmatrix} \begin{bmatrix} 1 & 0 & -I_{xz} \\ 0 \\ -I_{xz} & 0 \end{bmatrix} \begin{bmatrix} 0 & I_{x} + qr(I, -I_{z}) \\ -I_{xz} & 0 \end{bmatrix} \begin{bmatrix} 0 & I_{x} + qr(I, -I_{z}) \\ -I_{xz} & 0 \end{bmatrix} \begin{bmatrix} 0 & I_{x} + pq(I_{x} - I_{z}) \\ -qr & I_{xz} + pq(I_{x} - I_{z}) \end{bmatrix}$

$$\begin{bmatrix} \dot{\mathbf{p}} \\ \dot{\mathbf{q}} \end{bmatrix} = \cdots + \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \vec{\mathbf{r}}$$

Fast Inversion (cont)

 $\begin{bmatrix} \dot{p}_{des} \\ \dot{q}_{d} \\ \dot{r}_{c} \end{bmatrix} = \begin{bmatrix} \dot{p}_{des} \\ \dot{q}_{des} \\ \dot{r}_{des} \end{bmatrix} - \begin{bmatrix} 0 & L_{h} & L_{h} & 0 & L_{r} \\ M_{n} & 0 & 0 & M_{n} & 0 \\ 0 & N_{h} & N_{h} & 0 & N_{r} \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \\ p \end{bmatrix} - \begin{bmatrix} I_{x} & 0 & -I_{xz} \\ 0 & I_{y} & 0 \\ -I_{xz} & 0 & I_{z} \end{bmatrix} - \begin{bmatrix} pq & I_{x} + qr(I_{y} - I_{z}) \\ pr(I_{z} - I_{x}) + I_{xz}(r^{2} - p^{2}) \\ -qr & I_{xz} + pq(I_{x} - I_{y}) \end{bmatrix}$ inertial coupling cancel nonlinear cancel linear parameter varying dynamics

add desired dynamics

$$\begin{bmatrix} \dot{p}_{\text{des}} \\ \dot{q}_{\text{des}} \\ \dot{r}_{\text{des}} \end{bmatrix} = \begin{bmatrix} 0 & L'_{\text{B}} & L'_{\text{B}} & 0 & L'_{\text{T}} \\ M'_{\text{B}} & 0 & 0 & M'_{\text{B}} & 0 \\ 0 & N'_{\text{B}} & N'_{\text{B}} & 0 & N'_{\text{T}} \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \\ \beta \\ \beta \\ \beta \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \dot{p}_{\text{C}} \\ \dot{q}_{\text{C}} \\ \dot{r}_{\text{C}} \end{bmatrix}$$

ideal model $0 \quad 0 \quad \frac{3}{3+3}$ $\left[\frac{3}{s+3} \ 0 \ 0\right]$ 0 3+3 3+3 0 11 പിപ് ರ|ರೆ ម [អ] *r*,. ≩ີ Na robust performance/tracking **µ-Synthesis** W a H ٩٩٢ r,p,q, 80000 Ws ideal model ڻ ن fast inversion acts ڟ۫ڂؠ Zact Wact Wacth ă ở ử

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Slow Inversion

create new stability axis inputs

$$\begin{bmatrix} \mathbf{p}_{c} \\ \mathbf{q}_{c} \\ \mathbf{r}_{c} \end{bmatrix} = \begin{bmatrix} \cos\alpha & 0 & \sin\alpha \\ 0 & 1 & 0 \\ \sin\alpha & 0 & -\cos\alpha \end{bmatrix} \begin{bmatrix} \dot{\boldsymbol{\mu}}_{c} \\ \dot{\boldsymbol{\sigma}}_{c} \\ \dot{\boldsymbol{\beta}}_{c} \end{bmatrix}$$

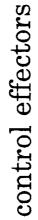
rewrite α and β equations with new inputs

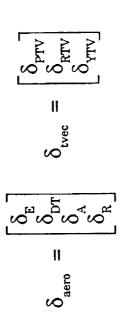
 $\dot{\alpha} = q - (p \cos \alpha + r \sin \alpha) \tan \beta - (\cosh / \cos \beta)\dot{\gamma} - \sinh (\cos / \cos \beta)\dot{\chi} + \dot{\alpha}_{e}$ $\dot{\beta} = p \sin \alpha - r \cos \alpha - \sinh \dot{\gamma} + \cosh \cos \dot{\chi} + \dot{\beta}_{c}$

cancel gravitational terms and add sideslip compensation

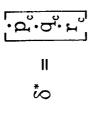
 $\dot{\alpha}_{c} = q_{c}^{c} - \cosh \cos{\gamma}/\cos{\beta} (g/V)$ $\dot{\beta}_{c} = K_{b}(\beta_{ref} - \beta) - \sinh{\alpha} \cos{\gamma} (g/V)$

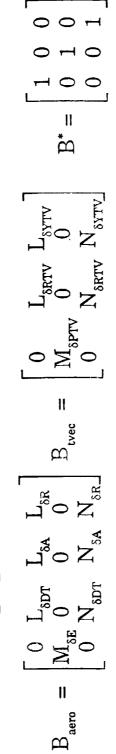
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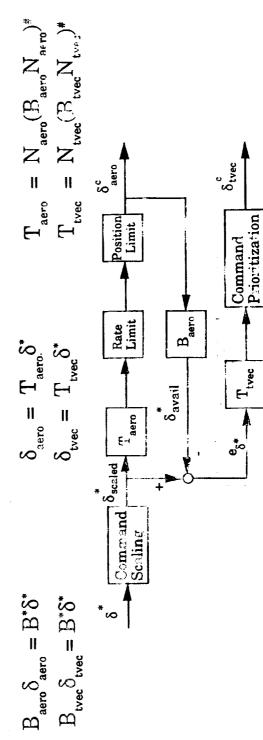




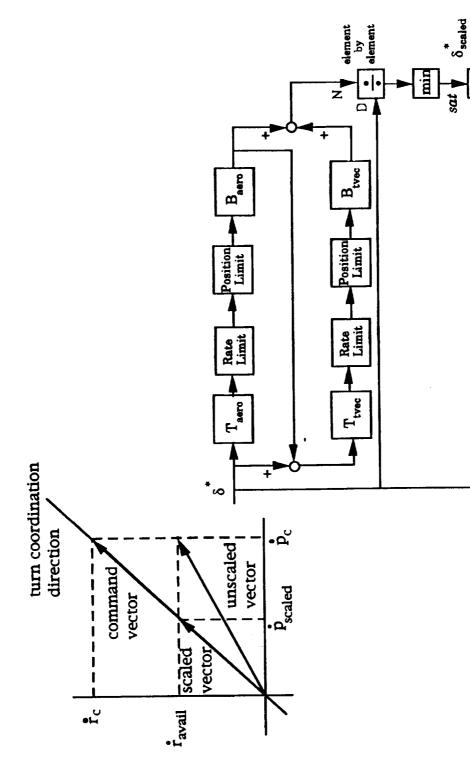




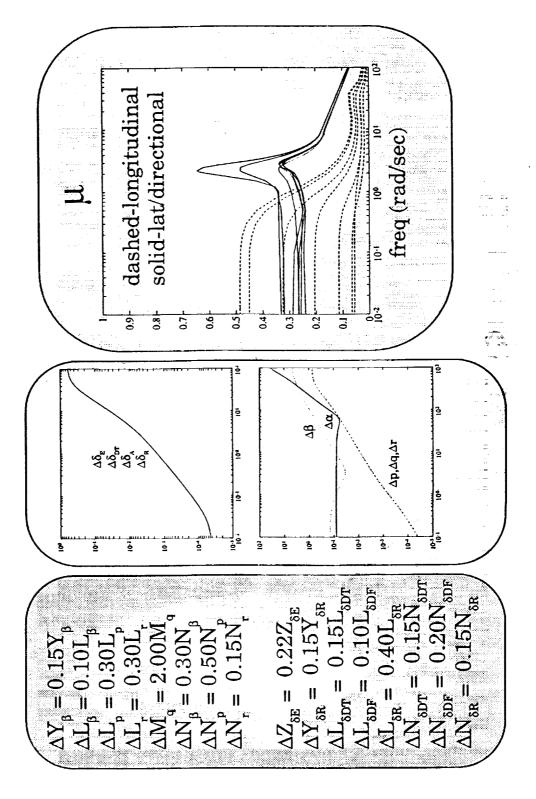


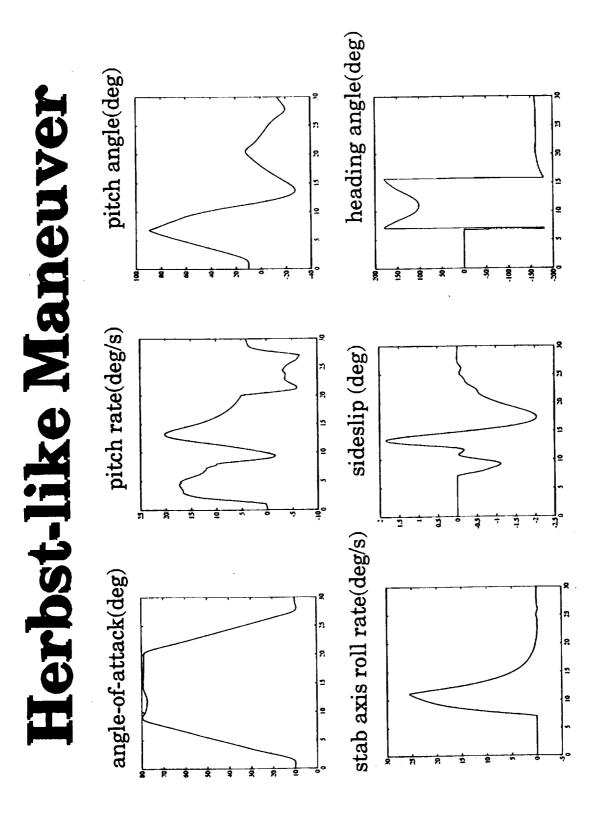




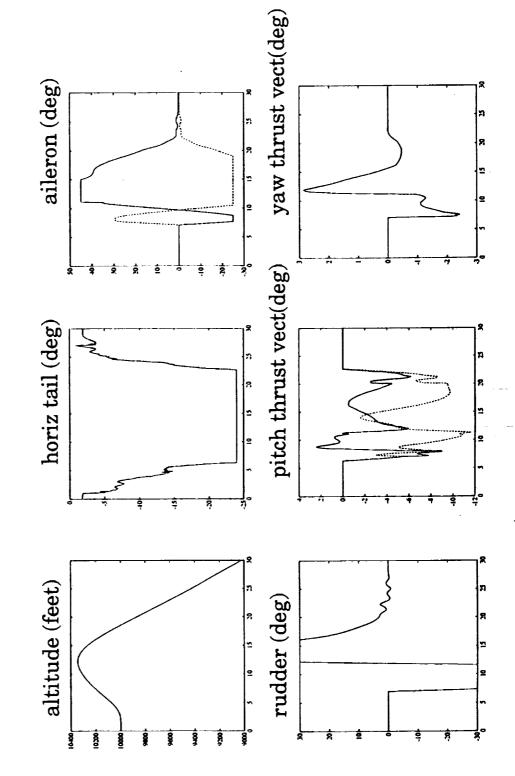


Structured Singular Value Analysis





Herbst-like Maneuver (cont)



ROBUST, NONLINEAR, HIGH ANGLE-OF-ATTACK CONTROL DESIGN FOR A SUPERMANEUVERABLE VEHICLE

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ABSTRACT

High angle-of-attack flight control laws are developed for a supermaneuverable fighter aircraft. The methods of dynamic inversion and structured singular value synthesis are combined into an approach which addresses both the nonlinearity and robustness problems of flight at extreme operating conditions. The primary purpose of the dynamic inversion control elements is to linearize the vehicle response across the flight envelope. Structured singular value synthesis is used to design a dynamic controller which provides robust tracking to pilot commands. The resulting control system achieves desired flying qualities and guarantees a large margin of robustness to uncertainties for high angle-of-attack flight conditions. The results of linear simulation and structured singular value stability analysis are presented to demonstrate satisfaction of the design criteria. High fidelity nonlinear simulation results show that the combined dynamic inversion/structured singular value synthesis control law achieves a high level of performance in a realistic environment.

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INTRODUCTION

Supermaneuverablity is defined as the ability to maneuver an aircraft up to and beyond the stall angle-of-attack. Some tactical payoffs of high angle-of-attack maneuvering include superior survivability, confusion of adversary pilots, and the ability to increase first-shot opportunities¹. Additional control power in the form of forebody vortex flow control and thrust vectoring can allow fighter aircraft to operate in the post-stall flight regime. Therefore, advanced control law design techniques must be found for robust high angle-of-attack stability augmentation and maneuvering. The resulting controllers must provide for the integration of both conventional and unconventional control effectors.

Modern robust multivariable design methods provide an efficient means of developing linear controllers for aircraft. Since the flight control problem is inherently multivariable, and the linear aircraft model has associated uncertainties, robust multivariable methods are a good choice for flight control design when nonlinearities are not too severe. In a Flight Dynamics Directorate contracted effort^{2,3}, a robust H_{∞} controller within an inner/outer loop framework was designed for a supermaneuverable aircraft at a single flight condition, and robust performance was demonstrated for a Herbst-like maneuver. A robust controller for this same vehicle was designed by Sparks⁴ for a single flight condition using μ synthesis in a model-following framework to simultaneously incorporate flying qualities specifications and account for structured uncertainty. A recent Wright Laboratory technical report⁵ describes the design of a μ -synthesis controller for a supermaneuverable vehicle that is integrated into an inner/outer loop control structure to provide full-envelope robust stability and performance for angles-ofattack up to 25 degrees.

Traditionally, flight control law development for low to moderate angle-ofattack flight regimes has been accomplished using linear design methods on linearized models of the aircraft. However, the advantages of supermaneuverability dictate that future air combat will venture into high angle-of-attack, nonlinear flight regions. Purely linear controllers are not able to effectively control supermaneuverable aircraft for more than very limited flight envelopes. This limitation has motivated a number of researchers to explore nonlinear techniques such as dynamic inversion. Bugajski and Enns⁶ have used nonlinear dynamic inversion to control the HARV aircraft across a wide, high angle-of-attack flight envelope. Huang⁷ has used a dynamic inversion approach to develop high angle-ofattack control laws for the X-29.

High angle-of-attack maneuvering is still a relatively new area in flight controls. Venturing into the regions of post-stall flight can and should elicit serious questions about safety issues such as control effector saturation and departure susceptibility. Different methods have been successfully demonstrated that assist in preventing the destabilizing effects of control saturations. In an approach used by Bugajski and Enns⁶, loop bandwidths are reduced so that a scaled projection of the desired control vector is achieved and the control surfaces lie on the boundary of an achievable subspace. A method of allocating control effectors such that the maximum possible moment is generated within a constrained set of achievable values has been suggested by Durham⁸. Another approach introduces thrust vectoring controls when saturations occur in aerodynamic surfaces⁵.

The main contribution of the work presented in this paper is the integration of some the most promising approaches described above into a detailed design approach for achieving robust high angle-of-attack flight control designs. The most notable advancement is the integration of dynamic inversion and structured singular value synthesis. Linearization of the vehicle dynamics is accomplished through a nonlinear dynamic inversion scheme. A robust compensator is designed around the linearized plant using μ -synthesis in a model-following framework. The μ -synthesis design satisfies flying qualities requirements and robustness goals throughout the design envelope. A control allocation scheme is used which uses the pseudo-inverse of the control distribution matrix to allocate controls based on body axis rotational acceleration commands. A method known as daisy-chaining is used to generate thrust vectoring commands when aerodynamic control effector saturation occurs. Adverse control power saturation effects are minimized by scaling lateral commands based on an achievable control vector. Control effector prioritization is implemented through a daisy-chain technique that limits lateral control power demands that compete with longitudinal power requirements.

In the following sections, a description of a modified F-18 aircraft model is given followed by the definition of design requirements. A brief theoretical background is presented on nonlinear dynamic inversion and μ -synthesis. The

controller architecture and design is described followed by the control allocation scheme and departure resistance logic. Finally, linear robustness analysis results and the results of a high fidelity nonlinear simulation of a supermaneuver are presented.

AIRCRAFT MODEL

The aircraft model described in this paper is based upon a modified version of the F-18 aircraft. The vehicle is a twin engine fighter aircraft with a moderately swept wing, twin canted vertical tails, and a large leading edge root extension. The aircraft model is augmented with two dimensional thrust vectoring nozzles that provide pitch and yaw moments when deflected symmetrically and a roll moment when deflected asymmetrically. The aerodynamic control inputs to the aircraft dynamics are the elevators, the ailerons, the rudders, and the leading and trailing edge flaps. The aerodynamic surfaces are useful at normal flight conditions, where there is adequate aerodynamic control surface effectiveness. The thrust vectoring inputs are useful at high angle-of-attack, low dynamic pressure operating conditions, where the traditional aerodynamic control effectiveness is inadequate. The pilot inputs include a control stick and rudder pedals.

A nonlinear simulation model of this aircraft exists as modular FORTRAN code. The model consists of separate modules describing the atmosphere, nonlinear equations of motion, aerodynamics, engines, thrust vectoring nozzles, variable geometry inlets, sensors, and actuators which include rate and position limits. The high-fidelity model was developed as part of a previous effort which gives more detail than that presented here⁹. There are five pairs of aerodynamic surfaces: three pairs for active control and two pairs scheduled for optimum performance. The ailerons, rudders, and elevators are used for stability augmentation and flight path manipulation. The leading and trailing edge flaps are scheduled to maximize airframe performance across the flight envelope. The aerodynamic data are contained in tabular format and linear interpolation is used for traditional force and moment aerodynamic coefficient build-up. Thrust vectoring-induced aerodynamic effects are added to static and dynamic baseline aerodynamic coefficients to obtain total aerodynamic coefficients. The dynamics of this vehicle can be described by the following set of first order nonlinear differential equations 6,10 . The first three equations describe the rotational dynamics of the aircraft in the body axis.

$$\dot{q} = \frac{1}{Iy} [m_{aero} + m_{thrust} + pr(I_z - I_x) + I_{xz}(r^2 - p^2)]$$
(1)
$$\begin{bmatrix} \dot{p} \\ \dot{r} \end{bmatrix} = \begin{bmatrix} I_x & -I_{xz} \\ -I_{xz} & I_z \end{bmatrix}^{-1} \begin{bmatrix} l_{aero} + l_{thrust} + pqI_{xz} + qr(I_y - I_z) \\ n_{aero} + n_{thrust} - qrI_{xz} + pq(I_x - I_y) \end{bmatrix}$$
(2)

q, p, and r are the body axis pitch, roll, and yaw rates respectively. I_x , I_y , I_z , and I_{xz} are the moments of inertia. The m, l, and n terms are the aerodynamic and thrust moment contributions to the rotational equations of motion.

The next three equations describe the evolution of aircraft motion with respect to its velocity vector.

$$\dot{\alpha} = q - (p \cos \alpha + r \sin \alpha) \tan \beta - \frac{\cos \mu}{\cos \beta} \dot{\gamma} - \sin \mu \frac{\cos \gamma}{\cos \beta} \dot{\chi}$$
 (3)

$$\dot{B} = p \sin \alpha - r \cos \alpha - \sin \mu \dot{\gamma} + \cos \mu \cos \gamma \dot{\chi}$$
 (4)

$$\dot{\mu} = \frac{\cos\alpha}{\cos\beta}p + \frac{\sin\alpha}{\cos\beta}r + \tan\beta\cos\gamma + [\sin\gamma + \tan\beta\sin\mu\cos\gamma]\dot{\chi}$$
(5)

 α , β , and μ are the angle-of-attack, sideslip angle, and roll angle about the velocity vector.

The last three equations describe the orientation of the velocity vector with respect to inertial space.

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$$\dot{\gamma} = \frac{\cos\mu}{mV}L - \frac{\cos\beta\sin\mu}{mV}Y - \frac{\cos\gamma}{V}g + \frac{1}{mV}F_{\text{thrust }\gamma}$$
(6)

$$\dot{\chi} = \frac{1}{mV \cos\gamma} [\sin\mu L + \cos\beta \cos\mu Y] + \frac{1}{mV} F_{\text{thrust}\chi}$$
(7)

$$\dot{V} = -\frac{D}{m} - g \sin\gamma + \sin\beta Y + \frac{1}{m} F_{\text{thrust } V}$$
(8)

 γ is the flight path angle, χ is the ground track angle, and V is the velocity. The F_{thrust} terms are the linear contributions of vehicle thrust to the aircraft equations of motion, resolved into the respective vectors. L, D, and Y are lift, drag, and side force, respectively. The parameter m is vehicle mass, and g is gravitational acceleration.

DESIGN REQUIREMENTS

Flying qualities are the primary measures of performance for a manual flight control system. For conventional flight, specifications for flying qualities can be found in MIL-STD-1797A¹¹. While these requirements are not valid for high angleof-attack flight, enough guidance is given to provide a basis for extrapolation. A rigorous study of new flying qualities measures is beyond the scope of this paper. The following high- α requirements are defined here only as baselines for this design study.

Requirements for the short period mode include constraints on the frequency and damping of a low order fit of the transfer function between pilot inputs and aircraft pitch response. Appropriate forms for this low order transfer function and methods for deriving the low order fit are described in MIL-STD-1797A. Short period frequency, ω_{sp} , should be a function of equivalent airspeed, V_{eq} . An appropriate guideline for short period frequency is:

$$\omega_{\rm sp} \ ({\rm rad/s}) = 1.0 \times \frac{V_{\rm eq} \ ({\rm ft/s})}{100}$$
 (9)

Therefore, the desired pitch response speed to pilot commands should increase with equivalent airspeed. At flight conditions above 30 degrees angle-of-attack, short period damping should be at the high end of the military standard's level 1 and 2 requirements. For these conditions, the acceptable range for ζ_{sp} is between 0.7 and 2.0.

The primary roll subsidence mode flying quality parameter is roll mode time constant, T_R . The roll mode time constant is found from a first order fit of the transfer function between pilot input and roll rate response. Past experience with fighter aircraft has shown that desired values for T_R are a function of angle-of-

attack. MIL-STD-1797A level 2 requirements are used as a baseline range of acceptable roll mode time constants. The target values used in this study are: $T_R = 0.30$ at $\alpha = 0$ deg, $T_R = 0.75$ at $\alpha = 30$ deg, and $T_R = 1.40$ at $\alpha = 60$ deg.

The desired directional response to pilot inputs can be derived from requirements on the Dutch roll mode. Dutch roll frequency, ω_D , and damping, ζ_D , can be derived from a second order fit of the transfer function between pilot input and sideslip response. Because of the danger of departure susceptibility at high angles-of-attack, the Dutch roll damping is required to be greater than 0.8. The Dutch roll frequency requirement is taken directly form MIL-STD-1797A, $\omega_D \ge 1.0$ rad/s.

DYNAMIC INVERSION

The purpose of dynamic inversion is to develop a feedback control law that linearizes the plant response to commands. In general the nonlinear aircraft dynamics can take the form

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}), \qquad \mathbf{y} = \mathbf{C}\mathbf{x} \tag{10}$$

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where x is an *n*-dimensional state vector, u is an *m*-dimensional input vector, C is a $p \times n$ matrix, and y is a *p*-dimensional vector of output variables. A transformation is necessary to put the equations in a form from which the inverse dynamics can be constructed. Each controlled output, y_i, is differentiated until an input term from u appears¹². Only *m* outputs can be controlled independently by the *m* available inputs, therefore *p* must equal *m*. As shown by Lane and Stengel¹³, the output equations may now be written in the form,

$$y^{[d]} = \begin{bmatrix} y_1^{[d_1]} \\ y_2^{[d_2]} \\ \vdots \\ y_p^{[d_p]} \end{bmatrix} = h(x) + G(x)u$$
(11)

where $y_i^{[d_i]}$ represents the d_ith derivative of the output y_i . The inverse dynamics control law can be written as

$$u = G(x)^{-1}(v - h(x))$$
(12)

h(x) represents the nonlinear output dynamics and G(x) represents the nonlinear control distribution. The parameter \vee represents the desired linear dynamics of the closed loop system. With the inverse dynamics control law implemented, the closed loop system now has the form,

$$\mathbf{v}^{(d)} = \mathbf{v} \tag{13}$$

If the system is observable and $\sum_{i=1}^{p} d_i = n$, then all of the closed loop poles may be placed. If $\sum_{i=1}^{p} d_i < n$, then closed loop stability cannot be proven. In this case the unobserved dynamics or the *internal dynamics* of dynamic inversion must be checked at local operating points to insure stability¹³.

STRUCTURED SINGULAR VALUE SYNTHESIS

The structured singular value (μ) framework provides a unifying measure which can be used to simultaneously address stability and performance robustness specifications^{14,15}. If μ is less than unity for a properly scaled system, then the specifications are met. It is desirable to be able to address these multiple objectives directly within a design method. μ -synthesis provides for the direct incorporation of robust stability and performance goals into a design by combining H_{∞} design with structured singular value analysis^{16,17}. The μ -synthesis problem is described by the attempt to find a controller that minimizes an upper bound on the structured singular value,

min inf sup
$$\overline{\sigma}(DM(K)D^{-1})$$
. (14)
K $D \in D \omega$

M(K) is the weighted closed loop transfer function shown in Fig. 1.

One approach to this problem is the DK-iteration; it calls for alternately minimizing sup $\overline{\sigma}(DM(K)D^{-1})$ for either K or D while holding the other constant. First the controller synthesis problem is solved using H_{∞} design on the nominal design model, G_w . μ -analysis is then performed on the closed loop transfer function M(K), producing values of the D scaling matrices at each frequency. The resulting

frequency response data are fit with an invertible, stable, minimum phase transfer function which becomes part of the nominal synthesis structure. With D fixed, the controller synthesis problem is again solved by performing an H_{∞} design on the augmented system. The DK-iterations are continued until a satisfactory controller is found or a minimum is reached. Fig. 2 shows a flow diagram for the DK-iteration. The resulting controller order is the order of the design plant and weighting matrices, in addition to the order of the D-scale transfer function fits. With each iteration, the D-scale frequency response data from the previous iteration is combined with the current values, and then the transfer function fit is performed on the combined data. This approach avoids a built-in increase in controller order that would result if at each iteration new D-scale fit transfer functions were augmented into the synthesis model from the previous step. It is important to note that the DK-iteration is not guaranteed to converge to a global minimum, but practical experience has shown that the method works well for a broad class of problems¹⁷.

CONTROL LAW DEVELOPMENT

Dynamic inversion and structured singular value synthesis are combined to achieve robust manual control for high angle-of-attack flight. The controller structure is shown in Fig. 3. Pilot pitch stick inputs command a pitch rate, q_{ref} , roll stick inputs command a stability axis roll rate, μ_{ref} , and pedal inputs command a sideslip, β_{ref} . The following sections describe each element of the control design.

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Fast Inversion

The aerodynamic and thrust induced moments in eqs. (1) and (2) determine the classical linear stability and control characteristics of the aircraft. These terms may be expanded into derivative form.

$$m_{\text{aero}} = \left[C_{m\alpha} V \alpha + \frac{\tilde{c}}{2} C_{mq} q + C_{m\delta_{\text{E}}} \delta_{\text{E}} \right] \frac{1}{2} \rho V S \tilde{c}$$
(15)

$$l_{aero} = \left[C_{l\beta} V \beta + \frac{b}{2} C_{lp} p + \frac{b}{2} C_{lr} r + C_{l\delta_{DT}} \delta_{DT} + C_{l\delta_{A}} \delta_{A} + C_{l\delta_{R}} \delta_{R} \right] \frac{1}{2} \rho V S \bar{b}$$
(16)

$$n_{\text{aero}} = \left[C_{n\beta} V \beta + \frac{\tilde{b}}{2} C_{np} p + \frac{\tilde{b}}{2} C_{nr} r + C_{n\delta_{\text{DT}}} \delta_{\text{DT}} + C_{n\delta_{\text{R}}} \delta_{\text{A}} + C_{n\delta_{\text{R}}} \delta_{\text{R}} \right] \frac{1}{2} \rho V S \tilde{b} \quad (17)$$

$$m_{\rm thrust} = [C_{m\delta \rm PTV} \delta \rm PTV]$$
(18)

$$l_{\text{thrust}} = \left[C_{l_{\delta_{\text{RTV}}}} \delta_{\text{RTV}} + C_{l_{\delta_{\text{YTV}}}} \delta_{\text{YTV}} \right]$$
(19)

$$n_{\rm thrust} = \left[C_{n\delta_{\rm RTV}} \,\delta_{\rm RTV} + C_{n\delta_{\rm YTV}} \,\delta_{\rm YTV} \right] \tag{20}$$

where S is the wing area, \bar{c} is the mean aerodynamic chord, and \bar{b} is the wing span. δ_E is the symmetric elevator position, δ_{DT} is the asymmetric elevator position, δ_A is the aileron position, δ_R is the rudder position, δ_{PTV} is the symmetric pitch thrust vectoring nozzle position, δ_{RTV} is the asymmetric pitch (roll) thrust vectoring nozzle position, and δ_{YTV} is the yaw thrust vectoring nozzle position. The derivatives in eqs. (15-20) can be represented in dimensional form where:

$$M_{q} = \frac{\rho V S \tilde{c}^{2}}{4 I_{y}} C_{m_{q}}, \qquad M_{\alpha,\delta} = \frac{\rho V^{2} S \tilde{c}}{2 I_{y}} C_{m_{\alpha,\delta}}$$
(21)

$$L_{p,r}^{*} = \frac{\rho V S \tilde{b}^{2}}{4 I_{x}} C_{l_{p,r}}, \qquad L_{\beta,\delta}^{*} = \frac{\rho V^{2} S \tilde{b}}{2 I_{x}} C_{l_{\beta,\delta}}$$
(22)

$$N_{p,r}^{*} = \frac{\rho V S \tilde{b}^{2}}{4 I_{z}} C_{n_{p,r}}, \qquad N_{\beta,\delta}^{*} = \frac{\rho V^{2} S \tilde{b}}{2 I_{z}} C_{n_{\beta,\delta}}$$
(23)

$$L_{\beta,p,r,\delta} = (L_{\beta,p,r,\delta}^* + \frac{I_{xz}}{I_x} N_{\beta,p,r,\delta}^*) \frac{I_z I_x}{(I_z I_x - I_{xz}^2)}$$
(24)

$$N_{\beta,p,r,\delta} = (N_{\beta,p,r,\delta}^* + \frac{I_{xz}}{I_z} L_{\beta,p,r,\delta}^*) \frac{I_z I_x}{(I_z I_x - I_{xz}^2)}$$
(25)

$$Z_{\alpha,\delta} = \frac{\rho V^2 S}{2m} C_{z\alpha,\delta}$$
(26)

$$Y_{\beta,\delta} = \frac{\rho V^2 S}{2m} C_{\gamma\beta,\delta}$$
(27)

The rotational equations of motion can now be written as a combination of linear and nonlinear contributions.

$$\begin{bmatrix} \dot{\mathbf{p}} \\ \dot{\mathbf{q}} \\ \dot{\mathbf{r}} \end{bmatrix} = \begin{bmatrix} 0 & L_{\beta} & L_{p} & 0 & L_{r} \\ M_{\alpha} & 0 & 0 & M_{q} & 0 \\ 0 & N_{\beta} & N_{p} & 0 & N_{r} \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \\ p \\ q \\ r \end{bmatrix} + \begin{bmatrix} 1_{\mathbf{x}} & 0 & -I_{\mathbf{xz}} \\ 0 & I_{\mathbf{y}} & 0 \\ -I_{\mathbf{xz}} & 0 & I_{\mathbf{z}} \end{bmatrix}^{-1} \begin{bmatrix} pqI_{\mathbf{xz}} + qr(I_{\mathbf{y}} - I_{\mathbf{z}}) \\ pr(I_{\mathbf{z}} - I_{\mathbf{x}}) + I_{\mathbf{xz}}(r^{2} - p^{2}) \\ - qrI_{\mathbf{xz}} + pq(I_{\mathbf{x}} - I_{\mathbf{y}}) \end{bmatrix} + \begin{bmatrix} 0 & L_{\delta DT} & L_{\delta A} & L_{\delta R} \\ M_{\delta E} & 0 & 0 & 0 \\ 0 & N_{\delta DT} & N_{\delta A} & N_{\delta R} \end{bmatrix} \begin{bmatrix} \delta E \\ \delta DT \\ \delta A \\ \delta R \end{bmatrix} + \begin{bmatrix} 0 & L_{\delta RTV} & L_{\delta YTV} \\ M_{\delta PTV} & 0 & 0 \\ 0 & N_{\delta RTV} & N_{\delta YTV} \end{bmatrix} \begin{bmatrix} \delta PTV \\ \delta RTV \\ \delta YTV \end{bmatrix} (28)$$

At this point, it is assumed that a generalized control scheme has been implemented in the form of a control selector, described later in this paper. Eq. (28) can be rewritten in terms of the generalized controls: roll, pitch, and yaw acceleration commands.

$$\begin{bmatrix} \dot{\mathbf{p}} \\ \dot{\mathbf{q}} \\ \dot{\mathbf{r}} \end{bmatrix} = \begin{bmatrix} 0 & \mathbf{L}_{\beta} & \mathbf{L}_{p} & 0 & \mathbf{L}_{r} \\ M_{\alpha} & 0 & 0 & M_{q} & 0 \\ 0 & N_{\beta} & N_{p} & 0 & N_{r} \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \\ p \\ q \\ r \end{bmatrix} + \begin{bmatrix} \alpha \\ \beta \\ p \\ q \\ r \end{bmatrix} + \begin{bmatrix} \mathbf{I}_{\mathbf{X}} & 0 & -\mathbf{I}_{\mathbf{X}\mathbf{Z}} \\ 0 & \mathbf{I}_{\mathbf{Y}} & 0 \\ -\mathbf{I}_{\mathbf{X}\mathbf{Z}} & 0 & \mathbf{I}_{\mathbf{Z}} \end{bmatrix}^{-1} \begin{bmatrix} pq\mathbf{I}_{\mathbf{X}\mathbf{Z}} + qr(\mathbf{I}_{\mathbf{Y}} - \mathbf{I}_{\mathbf{Z}}) \\ pr(\mathbf{I}_{\mathbf{Z}} - \mathbf{I}_{\mathbf{X}}) + \mathbf{I}_{\mathbf{X}\mathbf{Z}}(r^{2} - p^{2}) \\ - qr\mathbf{I}_{\mathbf{X}\mathbf{Z}} + pq(\mathbf{I}_{\mathbf{X}} - \mathbf{I}_{\mathbf{Y}}) \end{bmatrix} + \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \dot{\mathbf{p}}_{\mathbf{c}} \\ \dot{\mathbf{q}}_{\mathbf{c}} \\ \dot{\mathbf{r}}_{\mathbf{c}} \end{bmatrix}$$
(29)

The first step in control law development is the implementation of a dynamic inversion loop that replaces the existing rotational aircraft dynamics with some set of desired dynamics. This step is called the fast inversion.

$$\begin{bmatrix} \dot{\mathbf{p}}_{c} \\ \dot{\mathbf{q}}_{c} \\ \dot{\mathbf{r}}_{c} \end{bmatrix} = \begin{bmatrix} \dot{\mathbf{p}}_{des} \\ \dot{\mathbf{q}}_{des} \\ \dot{\mathbf{r}}_{des} \end{bmatrix} - \begin{bmatrix} 0 & L_{\beta} & L_{p} & 0 & L_{r} \\ M_{\alpha} & 0 & 0 & M_{q} & 0 \\ 0 & N_{\beta} & N_{p} & 0 & N_{r} \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \\ p \\ q \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ \beta \\ p \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ \beta \\ r \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ \beta \\ r \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ \beta \\ r \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ \beta \\ r \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ \beta \\ r \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ p \\ r \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ p \\ r \\ r \end{bmatrix} - \begin{bmatrix} \alpha \\ p \\ r$$

Eq. (30) is the application of the dynamic inversion step in eq. (12) to the outputs p, q, and r. The stability derivatives and inertial properties in eq. (30) are found through linear interpolation of values stored in a tabular database. The desired dynamics contain a set of linear stability derivatives that provide satisfactory modal frequency and damping characteristics.

$$\begin{bmatrix} \dot{\mathbf{p}}_{\text{des}} \\ \dot{\mathbf{q}}_{\text{des}} \\ \dot{\mathbf{r}}_{\text{des}} \end{bmatrix} = \begin{bmatrix} 0 & \mathbf{L'}_{\beta} & \mathbf{L'}_{p} & 0 & \mathbf{L'}_{r} \\ \mathbf{M'}_{\alpha} & 0 & 0 & \mathbf{M'}_{q} & 0 \\ 0 & \mathbf{N'}_{\beta} & \mathbf{N'}_{p} & 0 & \mathbf{N'}_{r} \end{bmatrix} \begin{bmatrix} \alpha \\ \beta \\ p \\ q \\ r \end{bmatrix} + \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \dot{\mathbf{p}}_{c} \\ \dot{\mathbf{q}}_{c} \\ \dot{\mathbf{r}}_{c} \end{bmatrix}$$
(31)

Structured Singular Value Synthesis

The fast inversion control law provides equalization of the dominant dynamics across the flight envelope. This equalization effectively eliminates the requirement for gain scheduling. A robust controller may now be designed around the linearizing fast inversion loop to provide command tracking performance. Robust tracking of body axis rotational rate commands is achieved with a μ -synthesis controller. The structured singular value of two uncertainty and one performance block is minimized using a DK-iteration. As shown in Fig. 4, three frequency dependent weights are chosen to balance performance and robustness considerations.

The design plant is defined by $G^* = C^*(sI - A^*)^{-1}B^*$. A* is the system dynamics matrix at some flight condition which is considered to be central to the flight envelope in terms of modal frequencies and damping. B* is the normalized

control effectiveness matrix in eq. (29). C* produces the outputs α , β , p, q, and r. The fast inversion block represents the body axis rate inversions shown in eq. (30).

 W_{act} defines the uncertainty at the actuator input, W_s defines the uncertainty at the sensor output, and W_p weights the error between the complementary sensitivity function of the closed loop system and an ideal model of the system response. The actuator and sensor uncertainty models are taken from Haiges, et al.¹⁸. Wp is chosen such that the closed loop system follows the ideal model closely at frequencies below 10 rad/s. The ideal model represents the desired transfer function between body axis rate commands and roll, pitch, and yaw rate responses. For this problem it is defined as

$$Ideal Model = \begin{bmatrix} p \\ P_c \\ q \\ q_c \\ \frac{r}{r_c} \end{bmatrix} = \begin{bmatrix} \frac{3}{s+3} & 0 & 0 \\ 0 & \frac{3}{s+3} & 0 \\ 0 & 0 & \frac{3}{s+3} \end{bmatrix}$$
(32)

A successful μ -synthesis design will achieve this first order tracking response to body axis rotational rate commands. The diagonal structure of the ideal model will also force the response to be decoupled in roll, pitch, and yaw.

Slow Inversion

Because only three generalized controls are available, the first step in dynamic inversion ignores the dynamics associated with angle-of-attack and sideslip. These *internal dynamics* can be accounted for in a second application of dynamic inversion to these slower state dynamics. A simple unitary transformation can be made to translate stability axis rate commands into the body axis rate commands that are available to the μ -synthesis controller.

$$\begin{bmatrix} \mathbf{p}_{c} \\ \mathbf{q}_{c} \\ \mathbf{r}_{c} \end{bmatrix} = T_{1} \begin{bmatrix} \dot{\mu}_{c} \\ \dot{\alpha}_{c} \\ \dot{\beta}_{c} \end{bmatrix}, \qquad T_{1} = \begin{bmatrix} \cos\alpha & 0 & \sin\alpha \\ 0 & 1 & 0 \\ \sin\alpha & 0 & -\cos\alpha \end{bmatrix}$$
(33)

Notice that q_c is equal to α_c . Eqs. (3) and (4) can be rewritten with these stability axis rate commands.

$$\dot{\alpha} = q \cdot (p \cos \alpha + r \sin \alpha) \tan \beta - \frac{\cos \mu}{\cos \beta} \dot{\gamma} - \sin \mu \frac{\cos \gamma}{\cos \beta} \dot{\chi} + \dot{\alpha}_{c} \quad (34)$$
$$\dot{\beta} = p \sin \alpha - r \cos \alpha - \sin \mu \dot{\gamma} + \cos \mu \cos \gamma \dot{\chi} + \dot{\beta}_{c} \quad (35)$$

If sufficient frequency separation exists between the body axis rate command responses and a contribution to the $\dot{\alpha}$ and $\dot{\beta}$ equations, then that contribution can be canceled by a slow inversion loop. The inversion control law for $\dot{\alpha}_c$ includes only the nonlinear effect of gravitational acceleration due to vehicle orientation. The other terms in the $\dot{\alpha}$ equation are either considered negligible or too fast to be controlled.

$$\dot{\alpha}_{c} = q_{c}' - \frac{\cos\mu\cos\gamma}{V\cos\beta}g$$
 (36)

The inversion control law for the sideslip equation includes the nonlinear term representing gravity induced sideslip due to non-zero roll angle. Again all other terms are considered negligible or too fast to control.

$$\dot{\beta}_{c} = K_{\beta} \left(\beta_{ref} - \beta\right) - \sin\mu \cos\gamma \frac{g}{V}$$
(37)

The addition of a sideslip feedback term to this equation provides sideslip command tracking and increased turn coordination. The gain K_{β} is selected to provide a second order response that satisfies the frequency and damping requirements for the Dutch roll mode. Assume that the closed loop system exactly matches the desired first order response in eq. (32). Then it can be assumed that

$$\frac{\dot{\beta}}{\dot{\beta}_c} = \frac{3}{(s+3)} \text{ and } \frac{\beta}{\dot{\beta}_c} = \frac{3}{s(s+3)}$$
 (38)

With the addition of the sideslip feedback gain K_{β} , the simplified transfer function from β_c to β can be written as

$$\frac{\beta}{\beta_{\text{ref}}} = \frac{\frac{K_{\beta}}{s} \frac{3}{(s+3)}}{(1 + \frac{K_{\beta}}{s} \frac{3}{(s+3)})} \approx \frac{3 K_{\beta}}{s^2 + 3s + 3K_{\beta}}$$
(39)

A value of 0.5 is selected for K_{β} , so the sideslip response to commands should be second order with a frequency of 1.2 rad/s and a damping of 1.2.

Command Shaping

The desired flying qualities for the pitch and roll axis are achieved through the use of prefilters. By scheduling these prefilters, the response to pilot inputs can be shaped appropriately with flight condition. As described earlier, the desired roll response is first order with a time constant that is a function of angle-of-attack. With the μ -synthesis compensator implemented, we can assume that the stability axis roll rate transfer function is:

$$\frac{\dot{\mu}}{\mu_c} = \frac{3}{(s+3)}$$
(40)

so the response with a first order prefilter with gain K_{μ} is:

$$\frac{\dot{\mu}}{\dot{\mu}_{ref}} = [\dot{\mu}_{PF}] \frac{\dot{\mu}}{\dot{\mu}_{c}} = \frac{3K_{\mu}}{(s+3)(s+K_{\mu})} \approx \frac{3K_{\mu}}{(s^{2}+(3+K_{\mu})s+3K_{\mu})}$$
(41)

A schedule for the roll prefilter gain that achieves the desired equivalent system response with angle-of-attack is:

$$K_{\mu} = 3.65 - 0.0433 \alpha$$
 and $\min(K_{\mu}) = 0.5.$ (42)

The desired pitch response to pilot inputs is second order with a short period frequency that is a function of equivalent airspeed. The transfer function representing the pitch response to pilot commands is:

$$\frac{\mathbf{q}}{\mathbf{q}_{c}} = \frac{3}{(s+3)} \tag{43}$$

The pitch response with a first order prefilter with gain K_q is:

$$\frac{\mathbf{q}}{\mathbf{q}_{ref}} = [\mathbf{q}_{PF}] \frac{\mathbf{q}}{\mathbf{q}_{c}} = \frac{3K_{\mathbf{q}}}{(s+3)(s+K_{\mathbf{q}})} = \frac{3K_{\mathbf{q}}}{(s^{2}+(3+K_{\mathbf{q}})s+3K_{\mathbf{q}})}$$
(44)

The prefilter gain, K_q , is scheduled to provide the desired level of damping and increase in short period frequency with equivalent airspeed.

$$K_{q} = \frac{1}{3} \left(\frac{V_{eq}}{100} \frac{(ft/s)}{100} \right)^{2}$$
(45)

CONTROL LIMITING AND PRIORITIZATION

The control selector, sometimes referred to as pseudo-controls, has two functions. The first is to normalize control effectiveness by transforming generalized rotational rate commands into actuator position commands. The second is to take advantage of available control redundancy by allowing for control redistribution without changing the linear closed loop performance. The basic idea of the control selector is in redefining the control contribution to the state equation (28,29),

$$B\delta = B^* \delta^* \tag{46}$$

B and δ are the actual control effectiveness matrix and control vector. B^{*} and δ ^{*} are the generalized control effectiveness matrix and control vector. The actual control can now be defined in terms of the generalized control,

$$\delta = T\delta^* \tag{47}$$

The transformation, T, is the control selector. It is defined simply by

$$T = N(BN)^{\#}B^{*}$$
 (48)

The operation $()^{\#}$ is a pseudo-inverse and N is a matrix that may be used to combine controls or emphasize/de-emphasize a control channel in the case of redundant effectors. Because the B matrix in eq. (48) is a function of flight condition and aircraft state, the control selector is a function of parameters such as Mach number, altitude, angle-of-attack, and engine power level angle.

The generalized and actual controls for the supermaneuverable vehicle are given by

$$\delta^{*} = \begin{bmatrix} \dot{p}_{c} \\ \dot{q}_{c} \\ \dot{r}_{c} \end{bmatrix}, \quad \delta = \begin{bmatrix} \delta_{E} \\ \delta_{DT} \\ \delta_{A} \\ \delta_{R} \\ \delta_{PTV} \\ \delta_{RTV} \\ \delta_{RTV} \end{bmatrix}$$
(49)

Consider the following partitioning of the control effector vector as shown in eq. (28)

$$\delta = \begin{bmatrix} \delta_{aero} \\ \delta_{tvec} \end{bmatrix}, \quad \text{where} \qquad \delta_{aero} = \begin{bmatrix} \delta_E \\ \delta_{DT} \\ \delta_A \\ \delta_R \end{bmatrix}, \qquad \delta_{tvec} = \begin{bmatrix} \delta_{PTV} \\ \delta_{RTV} \\ \delta_{YTV} \end{bmatrix}$$
(50)

resulting in

$$B_{aero} = \begin{bmatrix} 0 & L_{\delta DT} & L_{\delta A} & L_{\delta R} \\ M_{\delta E} & 0 & 0 & 0 \\ 0 & N_{\delta DT} & N_{\delta A} & N_{\delta R} \end{bmatrix}$$

$$B_{\text{tvec}} = \begin{bmatrix} 0 & L_{\delta RTV} & L_{\delta YTV} \\ M_{\delta PTV} & 0 & 0 \\ 0 & N_{\delta RTV} & N_{\delta YTV} \end{bmatrix}$$
(51)

With the above partitions, eq. (46) is written as:

$$[B_{aero} \quad B_{tvec}] \begin{bmatrix} \delta_{aero} \\ \delta_{tvec} \end{bmatrix} = B^* \delta^* , \text{ where } \quad B^* = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$
(52)

A daisy-chain method is used to generate thrust vector commands. Thrust vectoring is used only when the aerodynamic surfaces are not able to generate the necessary forces and moments required for commanded maneuvers. Therefore, the computation of aerodynamic control commands is independent of thrust vectoring control commands. The control selector is defined by

$$δaero = Taero δ*$$
 $δ*tvec = Ttvec δ*$
(53)

and

$$T_{aero} = N_{aero} (B_{aero} N_{aero})^{\#} \qquad T_{tvec} = N_{tvec} (B_{tvec} N_{tvec})^{\#}$$
(54)

where N_{aero} and N_{tvec} are used to weight the redundant control effectors. Since the ailerons contribute more to the roll acceleration and the first priority of the horizontal tail should be pitch control, the differential horizontal tail command is reduced by weighting the command to be a quarter of the other aerodynamic commands. There is no redundancy for the thrust vectoring control effectors, and thus, the weighting matrices become

$$N_{aero} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & .25 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix} \qquad N_{tvec} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$
(55)

Computation of the control selector eq. (53) depends on flight condition. Therefore, the elements of B_{aero} and B_{tvec} are found using linear interpolation of stored table values.

Nonlinear elements, such as position and rate limits, are required to implement the daisy-chain. Fig. 5 shows the structure of the nonlinear control selector. A limited aerodynamic surface command (δ_{aero}^{c}) is generated from a

rotational acceleration command (δ^*) via the aerodynamic control selector (T_{aero}) , the aerodynamic surface limits, and command scaling logic. An achievable aerodynamic rotational acceleration vector (δ^*_{avail}) is computed from the limited aerodynamic surface command using the control distribution (B_{aero}) . The difference of the commanded and achievable rotational acceleration vectors (e_{δ^*}) is transformed to a thrust vector command (δ^c_{tvec}) using the thrust vector control selector (T_{tvec}) and command prioritization logic.

The command scaling logic limits the acceleration command in the event of control effector saturation. The lateral/directional generalized control command that is generated by the control system can be thought of as a vector. This concept is illustrated in Fig. 6. When saturation occurs in one axis, the resulting control vector loses both the magnitude and the direction of the desired control. By scaling the command vector in both axis, an achievable control vector can be realized that preserves the direction of the desired command and holds the limiting controls on their limits. A block diagram of the command scaling logic is shown in Fig. 7. The figure shows that the scaled vector is the product of the commanded vector and the minimum ratio of available and commanded acceleration, $\delta^*_{scaled} = \delta^* \times sat$. The scaling parameter, sat, is always less than or equal to unity. It can be argued that when saturations occur, control bandwidth is too high. An interpretation can be made that this scaling logic acts to reduce the control bandwidth in the event of control power saturation⁶.

The command prioritization logic limits the amount of commanded differential pitch (roll) thrust vectoring. By using models of rate and position limiters within a daisy-chain, roll thrust vectoring is commanded only when the thrust vectoring nozzles are not saturated due to symmetric pitch thrust vectoring commands. For commanded rolls at high angles-of-attack, it can be interpreted that rolls commands correspond to performance and pitch commands correspond to stability. Therefore, the pitch thrust vectoring command, and thus stability, has top priority.

ROBUSTNESS ANALYSIS

A range of flight conditions must be selected for the purpose of linear analysis. Table 1 describes the seven conditions that span a broad range of Mach numbers, altitudes, and angles-of-attack. Linearized models of the vehicle dynamics at these points are used for robustness analysis.

| Flight Condition | Mach | Altitude (ft) | Angle-of-Attack (deg) | Equivalent Airspeed (ft/s) |
|---------------------|------|------------------|--------------------------|-------------------------------|
| 1 | 0.2 | 10,000 | 30 | 109.7 |
| 2 | 0.2 | 10,000 | 45 | 109.7 |
| 3 | 0.2 | 10,000 | 60 | 109.7 |
| 4 | 0.2 | 30,000 | 75 | 72.09 |
| 5 | 0.4 | 30,000 | 50 | 144.2 |
| 6 | 0.6 | 30,000 | 20 | 216.3 |
| 7 | 0.6 | 30,000 | 30 | 216.3 |

Table 1 Flight Conditions for Linear Analysis

The linear analysis models at each of these test conditions include high order actuator models, vehicle dynamics, and control elements shown in Fig. 3.

The robustness of the closed loop system is tested to simultaneous structured and unstructured uncertainties. The structured uncertainties consist of perturbations in aerodynamic stability and control derivatives. They are shown in Table 2. The structured uncertainties are presented in additive form because uncertainty percentages can vary greatly with flight condition, especially when the nominal value of a parameter approaches zero. The values shown in parentheses in Table 2 are the uncertainty percentages at flight condition 1. These are presented to show the relative degree of uncertainty in the different coefficients. Uncertainties in thrust vectoring are not included because the control distribution logic dictates that those effectors are only used at conditions where linear analysis is no longer appropriate. The dimensional form of these uncertainties can be found using eqs. (21-27).

| stability derivatives | control derivatives | |
|--|--|--|
| $\Delta C_{Z\alpha} = 0.150 \ (20\%)$ | $\Delta C_{z\delta E} = 0.0100 \ (21\%)$ | |
| $\Delta C_{y\beta} = 0.0150$ (12%) | $\Delta C_{y\delta R} = 0.0016 \ (10\%)$ | |
| $\Delta C_{l\beta} = 0.0030 \ (10\%)$ | $\Delta C_{l\delta DT} = 0.0015$ (2%) | |
| $\Delta C_{lp} = 0.3000 \ (66\%)$ | $\Delta C_{l\delta DA} = 0.0010$ (2%) | |
| $\Delta C_{lr} = 0.1000 \ (31\%)$ | $\Delta C_{l\delta R} = 0.0010$ (40%) | |
| $\Delta C_{m\alpha} = 0.150 \ (40\%)$ | $\Delta C_{m\delta E} = 0.0200 (5\%)$ | |
| $\Delta C_{mg} = 6.0000 \ (86\%)$ | $\Delta C_{n \delta DT} = 0.0015$ (6%) | |
| $\Delta C_{n\beta} = 0.0030 \ (10\%)$ | $\Delta C_{n\delta DA} = 0.0015 (6\%)$ | |
| $\Delta C_{np} = 0.1250 \ (140\%)$ | $\Delta C_{n\delta R} = 0.0020 (5\%)$ | |
| $\Delta C_{n_{\rm f}} = 0.0750 \ (24\%)$ | | |

Table 2 Structured Uncertainty Levels

Unstructured uncertainties include uncertain actuator and sensor dynamics. Fig. 8 shows the levels of multiplicative uncertainty that must be tolerated for each actuator and sensor channel. The quantities were derived as part of the work presented by Haiges, et al.¹⁸. The same level of unstructured uncertainty is assumed for all of the aerodynamic control effectors.

The results of structured singular value analysis indicate that the closed loop system is robust to the levels of uncertainty considered. Fig. 9 shows the upper bounds for the structured singular values at each of the linear test points. The fact that these bounds are less than unity at all frequencies provides a sufficient condition for robust stability. The peak in the lateral/directional bounds at 2-3 rad/sec indicates that Dutch roll mode is the most sensitive to plant uncertainties.

NONLINEAR RESULTS

In order to test the nonlinear performance of the flight control system, batch simulations are run on a high fidelity six degree-of-freedom simulation of the supermaneuverable vehicle. A challenging supermaneuver that tests the performance of the control laws and the control distribution logic is a very high angle-of-attack velocity vector roll. Fig. 10 shows such a maneuver where the aircraft is pitched up to 80 degrees angle-of-attack and then rolled 180 degrees about the velocity vector. This supermaneuver creates a rapid 180 degrees change in heading angle. The solid arrows represent the aircraft's velocity vector. Fig. 11 shows the time histories for this maneuver. Actuator positions are given in terms of left and right tail (δ_{TL} , δ_{TR}), left and right aileron (δ_{AL} , δ_{AR}), rudder (δ_R), left and right pitch thrust vectoring (δ_{PTVL} , δ_{PTVR}), and yaw thrust vectoring (δ_{YTV}). The left and right convention is used in place of symmetric and asymmetric so that control effector saturations are properly represented.

The 180 degree change in roll and heading angle is achieved by holding a 30 deg/sec stability axis roll rate command for six seconds. The performance of the dynamic inversion/ μ -synthesis control system is demonstrated by the smooth, well damped stability axis roll rate response and the excellent turn coordination at 80 degrees angle-of-attack. Less than 2 degrees of sideslip is generated during the supermaneuver. Notice that all of the aerodynamic surfaces saturate during this maneuver, forcing the control distribution, scaling, and prioritization logic to be activated. Command scaling comes into effect due to rate saturations in yaw thrust vectoring at the application and removal of the stability axis roll rate command. The pitch thrust vectoring prioritization logic is activated when symmetrical horizontal tail saturates, causing a requirement for symmetrical pitch thrust vectoring.

CONCLUSIONS

High angle-of-attack control laws have been developed for a supermaneuverable vehicle with thrust vectoring capability. The methods of dynamic inversion and structured singular value synthesis are successfully integrated into a design approach which achieves desired performance and robustness levels. An advanced generalized controls approach is demonstrated for the allocation of redundant aerodynamic and thrust vectoring effectors. Command scaling and prioritization are implemented to minimize the destabilizing effects of saturations during demanding supermaneuvers. The design goals are achieved across a broad range of airspeeds, altitudes, and angles-of-attack. High fidelity simulations show that the nonlinear aspects of the control laws perform well in a highly dynamic, nonlinear environment.

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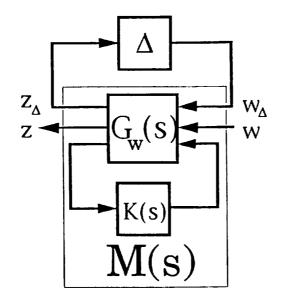


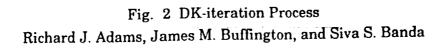
Fig. 1 M Δ Form for μ -synthesis Richard J. Adams, James M. Buffington, and Siva S. Banda

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$$Z_{m} \longrightarrow U_{m} \longrightarrow U_{m$$

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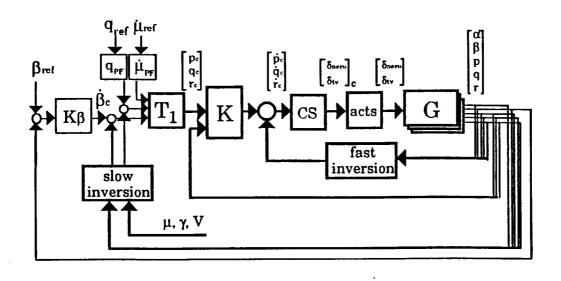


Fig. 3 Closed Loop System Richard J. Adams, James M. Buffington, and Siva S. Banda

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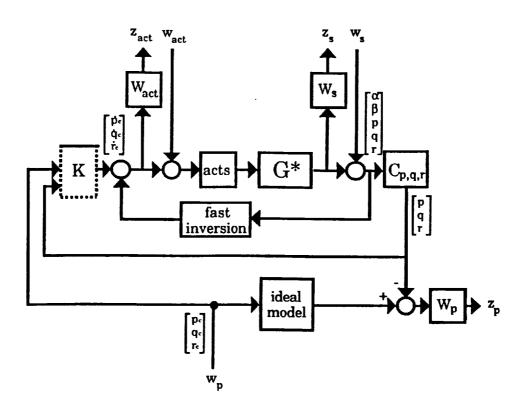


Fig. 4 Design Model for µ-synthesis Richard J. Adams, James M. Buffington, and Siva S. Banda

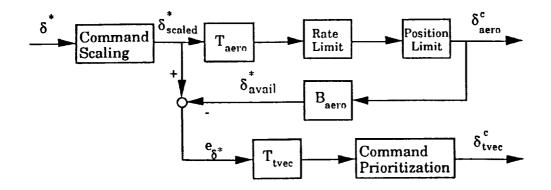


Fig. 5 Nonlinear Control Selector Richard J. Adams, James M. Buffington, and Siva S. Banda i

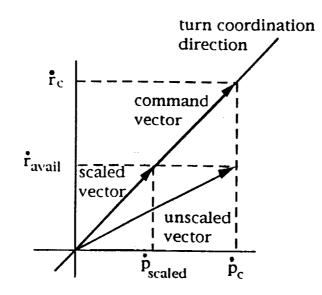


Fig. 6 Lateral/Directional Command Scaling Richard J. Adams, James M. Buffington, and Siva S. Banda

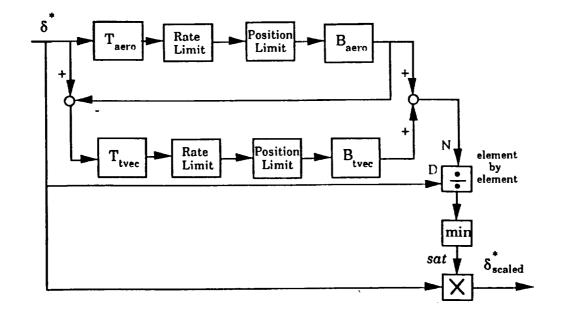
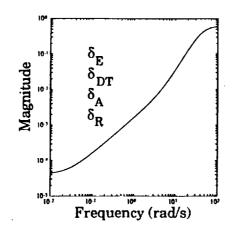


Fig. 7 Command Scaling Logic Richard J. Adams, James M. Buffington, and Siva S. Banda

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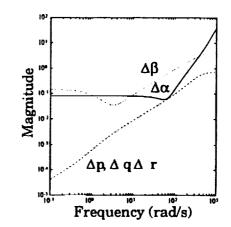


Fig. 8 Magnitudes of Unstructured Uncertainties Richard J. Adams, James M. Buffington, and Siva S. Banda

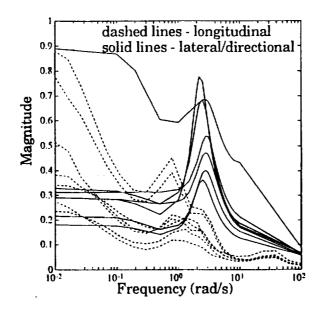


Fig. 9 Upper Bounds on Structured Singular Value Richard J. Adams, James M. Buffington, and Siva S. Banda ÷

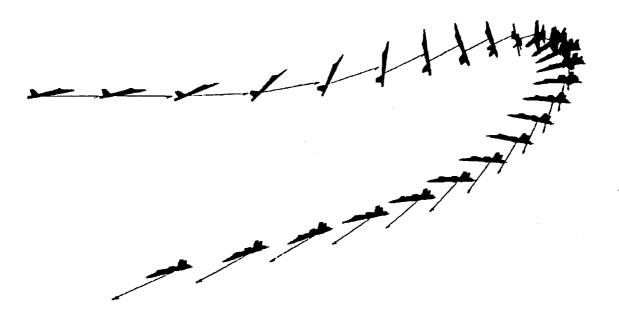


Fig. 10 High Angle-of-Attack Supermaneuver Richard J. Adams, James M. Buffington, and Siva S. Banda

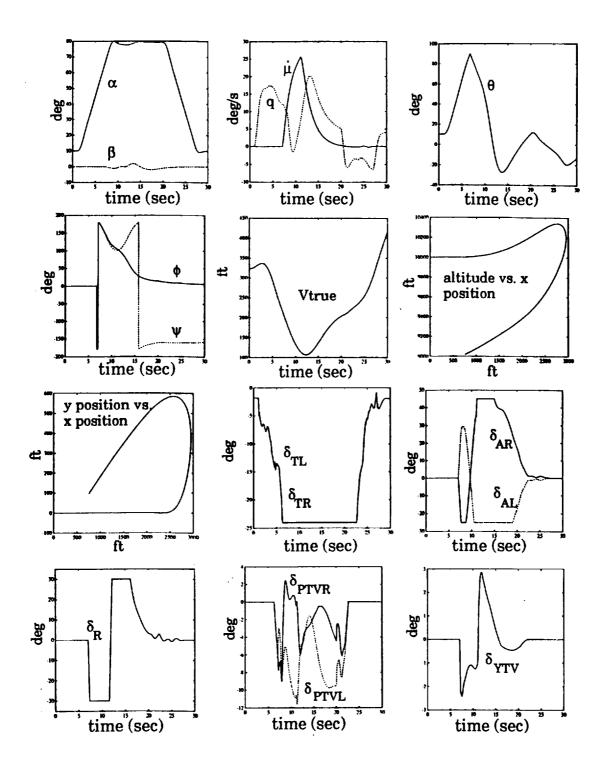


Fig. 11 Nonlinear Time Responses Richard J. Adams, James M. Buffington, and Siva S. Banda

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