

Microfluidic EBG sensor based on phase shift method realized using 3D printing technology

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Abstract: In this article, we propose a novel microfluidic microstrip EBG sensor realized using cost-effective 3D printing technology. Microstrip sensor allows monitoring of the fluid properties flowing in the microchannel embedded between the microstrip line and ground plane. The sensor's operating principle is based on the phase shift method which allows the characterization at a single operating frequency of 6 GHz. The defected electromagnetic band gap (EBG) structure is realized as pattern in the microstrip ground plane to improve sensor sensitivity. The designed microfluidic channel is fabricated using fused deposition modelling (FDM) 3D printing process without additional supporting layers while the conductive layers are realized using sticky aluminium tape. The measurement results show that the change of permittivity of the fluid in the microfluidic channel from 1 to 80 results in the phase shift difference of almost 90°. The potential application is demonstrated through the implementation of proposed sensor for the detection of toluene concentration in toluene-methanol mixture where various concentrations of toluene were analysed.

Keywords: micorstrip sensor, electromagnetic band gap (EBG), microfluidics, 3D printing, fused deposition modelling (FDM), phase shift method.

1. Introduction

Microfluidics is a technology of manipulating of the small quantity of fluids in the range of microliters to picoliters in one or network of microchannels. Since microfluidic technology allows operation and control of fluids on sub-microscale, it found the applications in various scientific and engineering disciplines such as inkjet printing, chemistry, environment, biomedicine, etc. [1-5]. Nowadays, advanced microfluidic biochips integrate into a single chip a number of operations such as sample pre-treatment and preparation, cell separation, and transport, mixing and/or separation of fluids together with micromechanical, optical, and electronic components for sensing and detection.

In order to integrate advance functions into a singular chip, modern microfluidics combine multiple technologies including microelectromechanical systems technology, injection molding, photolithography and x-ray lithography, laser ablation and micromachining, etc. [1, 4-5]. However, all stated microfabrication techniques are relatively complex, time consuming processes that typically require additional manual manufacturing procedures.

Nowadays, 3D printing technology attracts significant attention due to their low-costs, simple fabrication process that can be realized in a single run, good system compatibility, and presence of number of different materials with good optical, biocompatible, chemical or mechanical properties, [6-8]. The 3D printing offers the opportunity to fabricate the whole microfluidic device in a single run without the need for additional assembly processes. A wide range of biomaterials, such as living cells and growth factors, could also be directly printed using 3D printing technology, [8].

The 3D printing technology is based on different manufacturing methods, such as stereolithography, multi-jet modelling, electron beam melting, bioprint, and fused deposition modelling (FDM). The most commercially available 3D printers operate according to FDM method that has relatively low accuracy and speed in comparison to other methods. FDM 3D printers build structure layer-by-layer from the bottom up by extruding thermoplastic filament through a heated

47 nozzle and deposit it in fine threads along the extrusion path. Recently, FDM technique has been
48 employed to produce LEDs, sensors, antennas, and electrodes within biological tissue [6-7].

49 Fabrication of microfluidic channels by FDM is still a challenge because of several limitations:
50 extruded filaments cannot be arbitrarily joined at intersections, the lack of structural integrity
51 between the layers results in weak seals, and the size of the extruded filament is larger than typical
52 channel used in microfluidics. So far, the utilisation of FDM 3D printing technology for applications
53 in different microfluidic devices has been reported in a number of publications in literature [6-15]. A
54 reactor with fluidic 3 mm tube was fabricated using ABS polymer in [9], but this device suffers from
55 leakage and low operation pressure. Microfluidic channel of 800 μm was used in the realization of
56 organic and inorganic synthesis reactionware for electrochemical and spectroscopic analysis, [10].
57 Furthermore, fluidic devices with the same size of the microchannel were used for nanoparticle
58 preparation and electrochemical sensing [11], and detection of cancer protein biomarkers based on
59 supercapacitor-powered immunoarray [12]. However, it is still challenging to fabricate leak-proof
60 microfluidic channels narrower than 800 μm with the help of commercially available filaments using
61 FDM 3D printing technology. In [13], a custom FDM printer was designed for enhanced resolution
62 of the microchannel, where 350 μm wide polycaprolactone microchannels have been applied in the
63 realization of 3D nervous-system-on-a-chip.

64 Recent studies on microfluidics have been expanded to fluid detection at RF and microwave
65 frequencies [14-20]. RF and microwave microfluidics uses fluids as a substitutive dielectric material
66 for microwave antennas [14-15], transmission lines [16], or resonators [17]. In that manner, the
67 characteristic parameters of the fluids were determined based on measured impedance [15-16]
68 resonant frequency [17-19], insertion loss, or the phase of the signal that propagates along the
69 transmission line [20]. For the sensing applications the most promising method is the method based
70 on the phase measurement of the propagating signal [20-21], since it has relatively fast response,
71 allows characterization at single frequency, and it is the least sensitive to insertion loss.

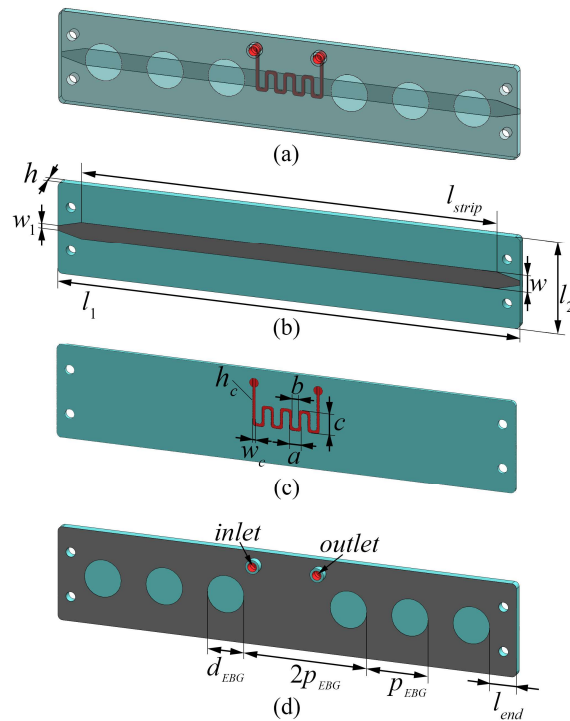
72 In this paper, we propose a novel microfluidic microstrip sensor realized using cost-effective 3D
73 printing technology. The 350 μm wide microfluidic channel embedded in the microstrip substrate
74 has been fabricated by conventional FDM 3D printing technology using polylactic acid (PLA)
75 filament without any supporting layer or soluble material. Leak-proof structure has been ensured
76 through careful optimization of the 3D model, infill factor and high infill/perimeter overlap settings.
77 The proposed sensor is designed to allow monitoring of the dielectric properties of the fluid that
78 flows in the microfluidic channel embedded between microstrip line and ground plane realized using
79 defected EBG structure. A concept to improve microstrip sensor sensitivity based on defected EBG
80 structure was proposed in [21], where it is demonstrated that the sensor sensitivity can be increased
81 by reducing the wave group velocity using periodic patterns in the ground plane, which exhibits EBG
82 effect [22]. The characteristics of different fluids in the microfluidic channel are analysed by the phase
83 shift measurements of the transmitted signal.

84 The potential application is demonstrated through the realization of the sensor for detection of
85 toluene concentration in toluene-methanol mixture. Toluene is an aromatic hydrocarbon solvent and
86 has numerous commercial and industrial applications, such as solvent in paints, thinners, and glues.
87 It is also used for printing and leather tanning processes, for disproportionation to a mixture of
88 benzene, methanol and xylene, and in production of number of synthetic drugs. Toluene, methanol
89 and their mixtures appear in a liquid state at room temperature. However, not only their liquids but
90 also their vapour concentrations in the air can become extremely high and can have negative effects
91 to work environment or can easily burn or explode, [23-24]. Hence, the existence of low-cost high-
92 sensitive sensor that is able to detect the exact concentrations of toluene or methanol in their mixtures
93 is essential in their operations, storage or transportation. The advantage of the proposed design has
94 been demonstrated through comparison between the fabricated sensor and other recently published
95 microfluidic sensors which operate according to the phase shift method.

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97
98

99 **2. Sensor design**

100 The 3D layout of the proposed microfluidic microstrip sensor with defected EBG etched in the
 101 ground plane is shown in the Figure 1a. Designed sensor consists of three main parts: microstrip line
 102 in the top layer, microfluidic channel in the middle, and ground plane patterned with defected EBG
 103 structure in the bottom layer. Top layer, shown in Figure 1b, consists of 50Ω microstrip line with
 104 length L_{strip} and width w , respectively, and two tapers. To avoid the short circuit between end-
 105 launched SMA (Southwest Microwave 292-04A-5) connectors' ground and the microstrip line two
 106 tapers were designed according to the recommendation from the manufacturer. The microfluidic
 107 channel bended in the shape of meander is embedded in the microstrip substrate above the defect in
 108 the EBG, i.e. the most sensitive location in the design, Figure 1c. Parameters a , b , and c denote the
 109 dimensions of meandered microchannel, while w_c and h_c are the width and height of the channel,
 110 respectively. The total thickness of the substrate with embedded microfluidic channel is denoted as
 111 h . The bottom layer of the sensor, shown in Figure 1d, represents the ground plane realized using
 112 defected EBG structure, periodical structure that consists of etched holes with diameter d_{EBG} placed
 113 at distance p_{EBG} . The defect in EBG is realized under the microchannel by removal of the one periodic
 114 element. Introduction of the defected EBG structure improves sensitivity of the sensor in comparison
 115 to the conventional microstrip line. In addition to EBG, two holes that serve as inlet and outlet of the
 116 channel were designed in the bottom layer for assembly microfluidic equipment to inject fluids into
 117 the channel, Figure 1d.

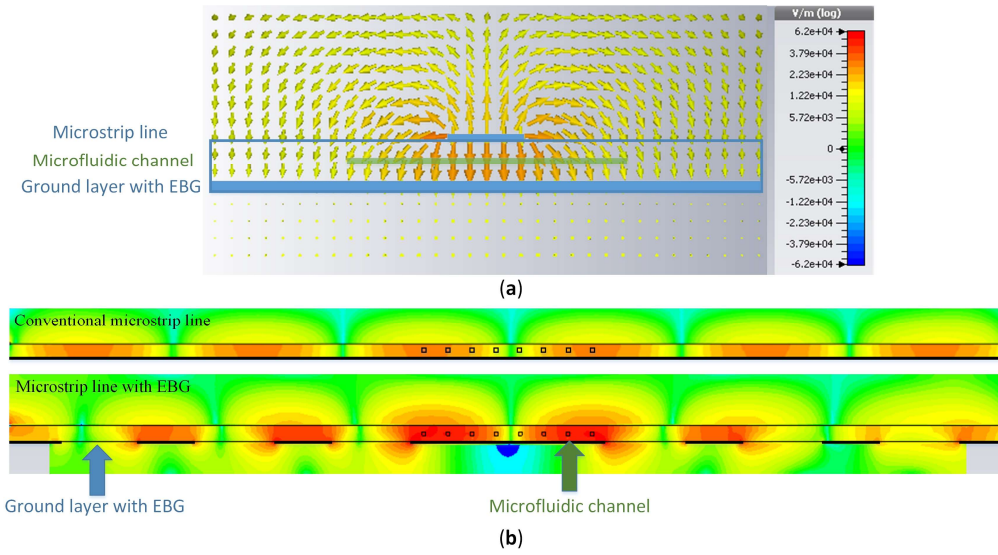


118

119 **Figure 1.** The layout of the proposed microfluidic EBG sensor: (a) 3D view, (b) top layer, (c) substrate
 120 with embedded microfluidic channel, (d) bottom layer with defected EBG structure.

121 Figure 2 shows the electrical field distribution in the orthogonal cross-section of the proposed
 122 sensor and comparison of the intensity of the electrical field along the microstrip line. The strongest
 123 electric field exists in the substrate between the microstrip line and the ground plane, i.e. the zone
 124 where the microfluidic channel is located. Moreover, in comparison with conventional microstrip
 125 line, the intensity of the electric field is stronger in the vicinity of a defect in the EBG. Therefore, the
 126 changes of the dielectric constant of the liquid that flows in the channel will have the highest impact
 127 to the sensor response. In order to further increase the sensitivity of the sensor, microfluidic channel
 128 is bended in the shape of the meander without changing the cross section dimensions of the channel.

129 In that manner, the effective area of the fluid exposed to the strongest electrical field increases
 130 compared to the topologies that use straight microfluidic channel.



131
 132 **Figure 2.** Electric field distribution of the proposed microfluidic EBG sensor: (a) orthogonal cross-
 133 section view, (b) comparison of the intensity of the electrical field along the microstrip line for
 134 conventional microstrip line and microstrip line with EBG.

135 3. Sensor operating principle: phase shift method

136 Phase shift method is based on a measurement of phase delay of the sinusoidal signal that
 137 propagates along transmission line. The phase shift of the signal is determined by velocity and
 138 frequency of the signal as well as physical properties of the transmission line:

$$139 \quad \Delta\varphi = \frac{\omega L_{TL}}{v_p}, \quad (1)$$

140 where ω denotes angular velocity, v_p is phase velocity of the signal, and L_{TL} is transmission line
 141 effective length. On the other hand, the phase velocity of the signal is dependent on the properties of
 142 the surrounding medium of the transmission line. In general, phase velocity of the signal can be
 143 determined as:

$$144 \quad v_p = \frac{\sqrt{2}}{\sqrt{\mu\epsilon}} \frac{1}{\sqrt{1 + \sqrt{1 + \frac{\sigma^2}{\omega^2 \epsilon^2}}}}, \quad (2)$$

145 where μ is magnetic permeability, ϵ is dielectric permittivity and σ is electric conductivity of the
 146 medium that surrounds the transmission line, [25]. If the operating frequency is high enough the
 147 influence of conductivity can be neglected and expression for phase velocity can be reduced to:

$$148 \quad v_p = \frac{1}{\sqrt{\mu\epsilon}}, \quad (3)$$

149 which exclusively depends on permittivity and permeability of surrounding medium. In this way,
 150 cross-sensitivity with the respect to conductivity can be avoided. This is particularly interesting if
 151 only the change in permittivity needs to be monitored, as in the case of detecting pollutants in
 152 water, [26].

153 The proposed sensor is based on a transmission line in the form of microstrip architecture with
 154 degeneration in the ground plane and inhomogeneous surrounding medium, Figure 3. The
 155 surrounding medium of the microstrip line consists of the air from top and the microfluidic channel
 156 embedded in 3D printed substrate.

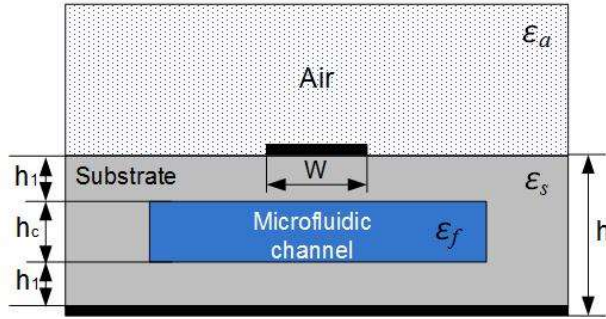


Figure 3. Cross section of the proposed sensor.

In order to apply above stated equations, the concept of effective permittivity can be introduced. In the simple case of the microstrip line, the effective permittivity can be approximated as:

$$\epsilon_{eff} = \frac{\epsilon_a + \epsilon_{sf}}{2} + \frac{\epsilon_a - \epsilon_{sf}}{2} \frac{1}{\sqrt{1 + 12 \frac{h}{W}}}, \quad (4)$$

where ϵ_a and ϵ_{sf} are permittivities of air and combination of 3D printed substrate and fluid in the microfluidic channel, respectively, [27].

The effective permittivity of the combination of the inhomogeneous dielectric substrate can be calculated using equation for effective dielectric permittivity of the multilayered substrate, [28]:

$$\epsilon_{sf} = \frac{|d_1| + |d_2| + |d_3|}{\frac{|d_1|}{\epsilon_s} + \frac{|d_2|}{\epsilon_{s+f}} + \frac{|d_3|}{\epsilon_s}}, \quad (5)$$

where the coefficients d_n are:

$$d_1 = \frac{1}{\pi} \ln \left(2 \frac{1 + \sqrt{\frac{1}{\cosh\left(\frac{\pi W}{4h_1}\right)}}}{1 - \sqrt{\frac{1}{\cosh\left(\frac{\pi W}{4h_1}\right)}}} \right),$$

$$d_2 = \frac{1}{\pi} \ln \left(2 \frac{1 + \sqrt{\frac{1}{\cosh\left(\frac{\pi W}{4(h_1+h_c)}\right)}}}{1 - \sqrt{\frac{1}{\cosh\left(\frac{\pi W}{4(h_1+h_c)}\right)}}} \right),$$

$$d_3 = \frac{1}{\pi} \ln \left(2 \frac{1 + \sqrt{\frac{1}{\cosh\left(\frac{\pi W}{4(2h_1+h_c)}\right)}}}{1 - \sqrt{\frac{1}{\cosh\left(\frac{\pi W}{4(2h_1+h_c)}\right)}}} \right). \quad (6)$$

The ϵ_{s+f} is the dielectric constant of the middle layer with microchannel that can be calculated using Bruggeman formalism, [29]:

$$\epsilon_{s+f} = V\epsilon_s + (1 - V)\epsilon_f, \quad (7)$$

where V is the volumetric fraction of the microfluidic channel in the surrounding PLA substrate.

Based on the above equations, the operating principle of the sensor can be described. The change of the fluid's properties in the microfluidic channel causes the change of ϵ_{s+f} which results in the change of effective permittivity of the microstrip. Consequently, the phase velocity changes which alters the phase shift. It can be concluded that different values of fluid's permittivity is related to different phase shifts which is a necessary condition for constructing a calibration curve.

The defected EBG pattern in the sensor ground plane is used to improve the sensitivity of the microstrip sensor. The introduction of the uniform EBG structure in the ground plane forms a different frequency region where propagation is forbidden, i.e. bandgap in the transmission characteristic [22].

183 The group velocity that can be determined by the slopes of the bands of the propagation modes goes
184 to zero at the bandgap edges in the case of the microstrip EBG sensors [21]. In that manner, for a
185 constant operating frequency, a significant change in the group velocity can be observed in the case
186 of the microstrip with EBG in comparison to the conventional microstrip line. A large decrease in the
187 group velocity corresponds to the slow wave effect.

188 The defect in the EBG results in a resonance in the bandgap, which frequency is determined by
189 the size of the defect. By introducing defected EBG structure in the microstrip ground plane, the
190 phase change significantly increases especially at the frequencies that are close to the bandgap edges
191 and at the resonance in the bandgap. As it is shown in [21], at a constant frequency, the change in
192 wave vector, k is larger for the band with lower group velocity (Δk). The phase change ($\Delta\varphi$) for a
193 given change in permittivity is proportional to $\Delta k L_{TL}$, where L_{TL} is the effective length of the
194 microstrip sensor. This is the main cause of increased phase change and therefore sensitivity of the
195 proposed sensor.

196 It can be mentioned that from the phase difference measurement, the real part of dielectric
197 constant can be directly determined. The complex permittivity of the fluid can be obtained by
198 measuring both the amplitude and the phase of the transmitted signal of the sensor and incident
199 signal or it can be reconstructed from Kramers-Kronig dispersion relations.

200 The phase shift method allows characterization of sample on single frequency, unlike resonant
201 methods that require characterization over a range of frequencies. In addition, phase shift
202 measurement is less prone to the noise and less sensitive to the insertion loss. In that manner, this
203 method is suitable for sensing of the high loss materials.

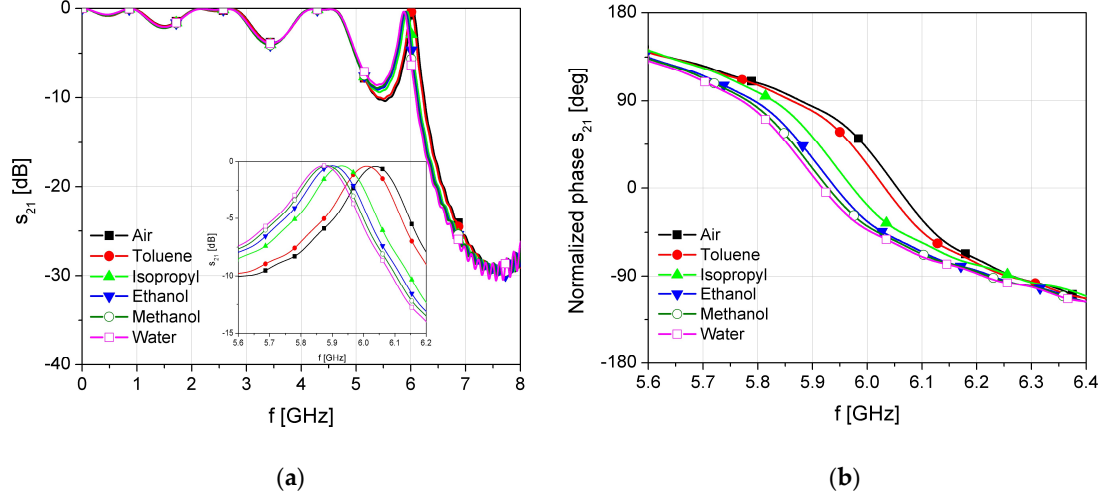
204 4. Simulation results

205 The characteristics of the proposed sensor, the influence of different geometrical parameters,
206 optimization and the influence of the different fluids in the microchannel have been analysed using
207 CST Microwave studio. PLA is chosen as substrate material since it is one of the most used
208 thermoplastics in the 3D printing. Initially, dielectric constant of the used 3D printed PLA material
209 printed with 100% infill was determined to be 2.7 with $\tan\delta$ equal to 0.01 at the frequency of 6 GHz.
210 The optimized dimensions of the microstrip line, microfluidic channel and EBG structure have been
211 determined to be: $h=1.5$ mm, $l_{strip}=89.84$ mm, $w_1=1$ mm, $w=3.6$ mm, $l_2=20$ mm, $l_1=100$ mm, $h_c=0.4$ mm,
212 $w_c=350$ μ m, $a=2.35$ mm, $b=1.65$ mm, $c=4.1$ mm, $d_{EBG}=8$ mm, $p_{EBG}=13.4$ mm, and $l_{end}=5.8$ mm.

213 The simulation results of the proposed sensor with different fluids placed in the microchannel
214 are shown in Figure 4. Each fluid in the simulation is modelled with its material parameters, i.e. its
215 permittivity and dissipation factor, as shown in Table 1, [30].

216 The EBG structure is designed to provide bandgap between 5 and 9 GHz, while the defect in the
217 EBG causes the resonant effect at 6 GHz. From the transmission characteristic it can be seen that the
218 resonant frequency of the defect in the band gap slightly shifts due to the change of the dielectric
219 constant of the material in microchannel, Figure 4a. On the other hand the phase change is increasing
220 due to decrease of the wave phase velocity, Figure 4b. The effect of the EBG structure is predominant
221 at the frequency of 6 GHz where the wave phase velocity is minimal. This frequency is recognized as
222 a frequency of interest since the change of the phase difference is the highest in that case. It should
223 be noted that changes of the resonance and loss in the transmitted signal depend on the material in
224 the microfluidic channel. However, the insertion losses do not fall below - 10 dB in the worst case at
225 operating frequency. Therefore, the phase of the transmitted signal can be measured using standard
226 phase comparators or detectors that determine phase difference as a subtraction of the phase shift of
227 the sensor transmitted signal and the phase of the excitation signal.

228 The simulation results show that the change of the fluid permittivity from 1 (air) to 80.1 (water)
229 causes the phase shift difference of 84° . Compared to the phase shift of the conventional microstrip
230 line without defected EBG which is only 10.2° at 6 GHz, the proposed design shows eight times higher
231 phase shift.



232 **Figure 4.** Simulation results of the proposed sensor with different fluids placed in the microchannel:
 233 (a) transmission characteristic, (b) normalized phase.

234 **Table 1.** Dielectric properties of analysed fluids at room temperature.

Fluid	ϵ_r	$\tan\delta$
Toluene	2.3	0.04
Isopropyl	19.7	0.799
Ethanol	24.5	0.941
Methanol	32	0.659
Water	80.1	0.123

235 5. Fabrication and measurement

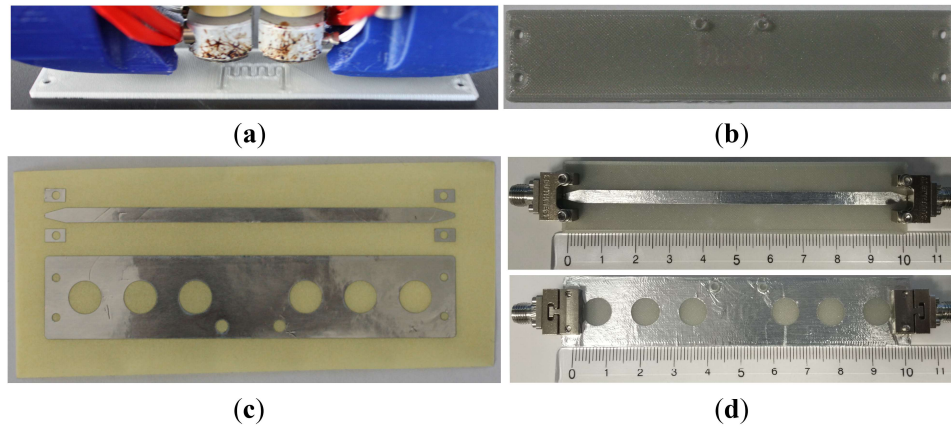
236 Microfluidic channel, embedded into the substrate, was designed using the 3D modelling CAD
 237 software. The effects that occur during 3D printing, such as shrinkage and variations of the final
 238 dimensions, were taken into account in the final model to obtain designed dimensions of the channel
 239 and thickness of the substrate. 3D model has been imported into KISSlicer software to create the
 240 G-code with the following slicing parameters: layer thickness of 0.1 mm, extrusion width of 0.25 mm,
 241 number of perimeter equal to 4, infill factor of 100%, and the print speed of 20 mm/s. In order to make
 242 leak-proof structure, the 3D printed model is designed using full infill factor and high infill/perimeter
 243 overlapping settings.

244 Microfluidic sensor's substrate and microfluidic channel are printed simultaneously using Felix
 245 3.1 3D printer based on FDM technology. Biodegradable PLA thermoplastic filament with diameter
 246 of 1.75 mm is used since it results in well-defined and quality printed structures. The extruder
 247 temperature was set to 190 °C for the first layer, while the 185 °C was used for other layers. It should
 248 be noted that designed microchannel is printed without any supporting material. Figure 5a shows
 249 the printing process of the 3D substrate with embedded microchannel.

250 The layout of the fabricated substrate and microfluidic channel filled with coloured fluid are
 251 shown in Figure 5b and Figure 6, respectively. The final dimensions of the fabricated microfluidic
 252 channel are determined by measuring the horizontal and vertical cross sections of the fabricated
 253 prototypes using the Huvitz HRM 300 profilometer, as shown in Figure 7. The measured dimensions
 254 of the fabricated microfluidic channel and substrate are: $h=1.52$ mm, $l_1=100.1$ mm, $h_c=367 - 383$ μ m,
 255 $w_c=343-384$ μ m, $a=2.15-2.71$ mm, $b=1.45 - 1.72$ mm, $c=4.02 - 4.23$ mm. Imperfections of the 3D printing
 256 process mostly affect the dimensions of the channel causing the variation up to 10% from the
 257 designed values.

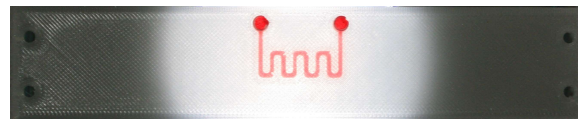
258 The conductive parts in the top and the bottom layers were realized using 40 μ m thick
 259 conductive aluminium sticky tape precisely cut with Rofin-Sinar PowerLine D laser, Figure 5c. The

260 sticky conductive tape is accurately positioned and affixed from both sides of the substrate. The final
261 layout of the proposed sensors with mounted end-lunch SMA connectors is shown in Figure 5d.



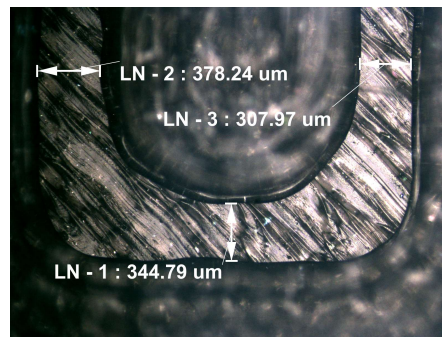
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263 **Figure 5.** Sensor fabrication process: (a) Printing process of the 3D substrate with embedded
264 microchannel, (b) Layout of the 3D printed substrate with embedded microfluidic channel,
265 (c) Conductive top and bottom layers precisely cut with laser, and (d) Layout of the proposed sensors,
266 top layer and bottom layer with mounted SMA connectors.



267

268 **Figure 6.** Layout of the 3D printed substrate with embedded microfluidic channel filled with coloured
269 fluid.



270

271 **Figure 7.** Cross section of the 3D printed microfluidic channel with measured widths.

272 The measurement setup is shown in the Figure 8, where the fluids are injected true additional
273 tubules into the microfluidic channel using syringe. The characteristic parameters of the fabricated
274 sensor were measured in the frequency range between 100 kHz and 8 GHz using two ports Agilent
275 8501C Vector Network Analyser.

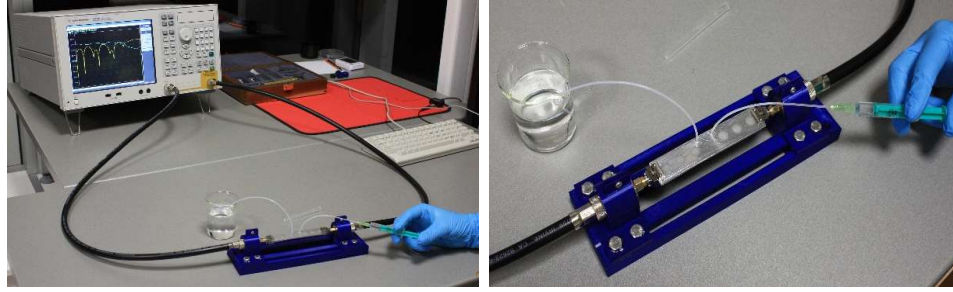


Figure 8. Measurement setup of the proposed sensor.

The simulation and the measurement results are compared in terms of transmitted amplitude and phase, Figure 9. For better visibility, the measurements and simulations were compared only for two fluids, i.e. fluids which have the lowest and the highest dielectric constant, i.e. air and water. It can be seen that the simulated and measured results are in excellent agreement. The frequency where the phase shift shows the highest changes is slightly shifted to 6.15 GHz due to imperfection of the fabrication process, while the phase difference is slightly increased. On the other hand, the insertion losses do not fall below - 10 dB in the worst case at the operating frequency.

Figure 10 shows the measured transmission and phase characteristics of the proposed sensor with different fluids in the microfluidic channel in the frequency range of interest. The empty air channel is used as a reference value. For better visibility, the phase characteristic is normalized in the range from -180° to 180° .

For more comprehensive analysis, Table 2 summarizes the phase shifts for different fluids in the microfluidic channel obtained by simulations and measurements, as well as the phase shifts calculated based on the above stated equations, where ϵ_{sf} denotes effective permittivity of combination of 3D printed substrate and fluid in the microfluidic channel, ϵ_{eff} is the total effective permittivity, while $\Delta\varphi_{cal}$, $\Delta\varphi_{sim}$, and $\Delta\varphi_{meas}$ are calculated, simulated and measured phase shifts, respectively. The calculated and simulated results agree well with measured results.

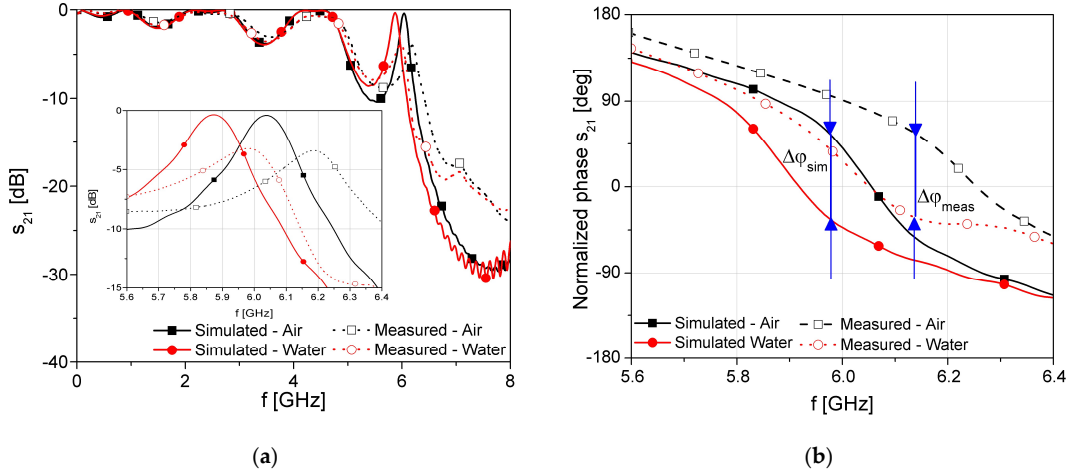
The sensitivity of the proposed sensor can be defined as a ratio of the phase difference of sensor response with fluid and air in the microchannel, divided by permittivity of used fluid. Figure 11 shows the phase shift change of the proposed microfluidic EBG sensor with the respect to change of permittivity of the fluid. For the better comparison the results of conventional microstrip line sensor as well as results obtained from simulations are added in Figure 11. The exponential fitting curves and corresponding equations that provide excellent curve-fitting are also presented in the Figure 11. The Curve Fitting Tool in the Matlab was used. It can be seen that proposed sensor shows relatively high and almost linear dependence for the fluid materials with permittivity lower than 30. For the higher values of the permittivity, the change in phase is relatively small and sensor goes to saturation. It can be seen that in the case of conventional microstrip line eight times lower sensitivity is achieved while the saturation occurs for permittivity of 20. Furthermore, the measurement results show the better linearity in comparison with the simulations, especially for the lower values of dielectric constant.

Table 2. Calculated, simulated and measured phase shifts for different fluids in the microchannel

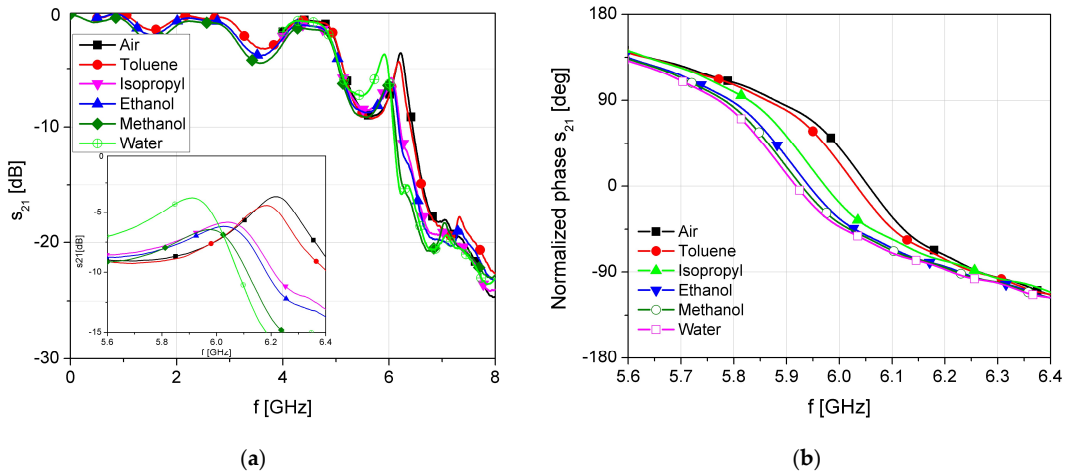
Fluid	ϵ_{sf}	ϵ_{eff}	$\Delta\varphi_{cal}$	$\Delta\varphi_{sim}$	$\Delta\varphi_{meas}$
Air	2.6946	1.5014	0	0	0
Toluene	2.6987	1.5026	17.02	16.39	17.7
Isopropyl	2.7518	1.5183	48.45	58.05	47.5
Ethanol	2.7657	1.5224	62.13	74.47	55.9
Methanol	2.7868	1.5287	76.12	79.12	71.7
Water	2.9066	1.5641	85.75	84	86.7

The potential application is demonstrated through the implementation of proposed sensors for the detection of toluene concentration in toluene-methanol mixture where various concentrations of

311 toluene were analysed. Figure 12 illustrates the measured phase difference when the toluene
 312 concentration varies from 0% to 100%. The linear and polynomial fitting curves and corresponding
 313 equations that better describe the measured phase dependence on the concentration of toluene in the
 314 toluene-methanol mixture are also shown in Figure 12. The experimental results show the phase
 315 difference changes for 54 degrees when the concentration of toluene is changed from 0% to 100%. In
 316 addition, the proposed sensor shows almost linear shift with sensitivity of 0.540° per percentage of
 317 toluene. These results successfully demonstrate and confirm the application of the proposed
 318 microfluidic EBG sensor as chemical sensor.



319 **Figure 9.** Comparison of the measured and simulated responses: (a) Transmission characteristic, and
 320 (b) Normalized phase.

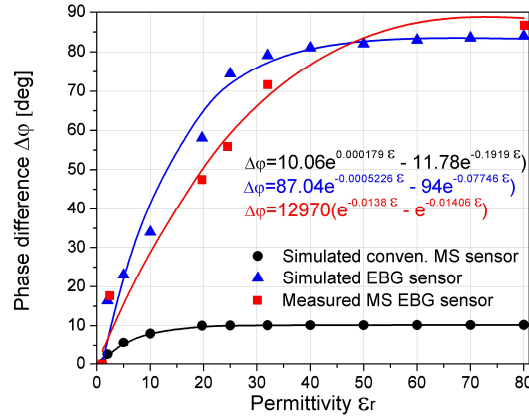


321 **Figure 10.** Measured results of the proposed sensor with different fluids inside microfluidic channel:
 322 (a) Transmission characteristic, and (b) Normalized phase.

323 The proposed sensor is compared with recently published microfluidic sensors that operate
 324 according to the phase shift method [31-34]. The parameters for these sensors are summarized in the
 325 Table 3, where f_{opr} denotes sensor operating frequency and $\Delta\phi_{max}$ is the maximal phase shift. Table 3
 326 also shows specific application of the compared sensors and used fabrication technologies.

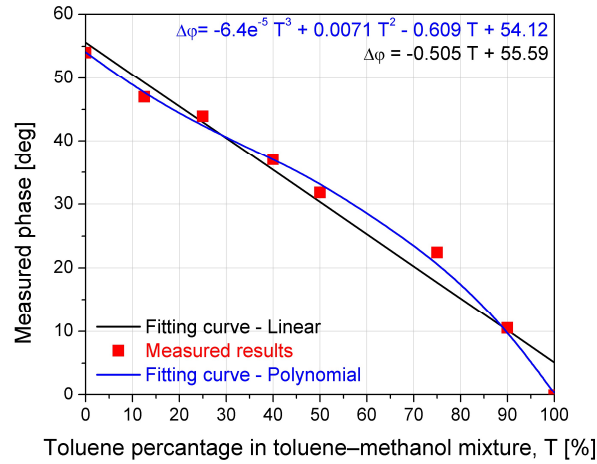
327 The proposed sensor has comparable characteristics and its fabrication process is the simplest
 328 and the cheapest comparing to other published sensors. Although one may argue that the sensor
 329 published in [34] has the largest phase shift, it should be noted that this sensor requires the most
 330 demanding fabrication process. Comparing with sensor proposed in [31] which fabrication
 331 complexity is fairly similar, the proposed EBG sensor shows 65% better sensitivity. Therefore, the

332 proposed design represents a good candidate for the design of high-performance sensor since it
 333 reconciles the requirements for good sensitivity, compactness, and simple fabrication.



334

335 **Figure 11.** Sensitivity of the proposed EBG sensor compared to conventional microstrip line sensor
 336 (dots) and the corresponding fitting curves (lines) with equations.



337

338 **Figure 12.** Measurement of the toluene concentration in toluene-methanol mixture: sensor measured
 339 response (dots) and the corresponding fitting curves (lines) with equations.

340 **Table 3.** Comparison of the characteristics of the proposed sensor and other recently published
 341 sensors that operate according to the phase shift method.

Ref.	f_{opr}	$\Delta\phi_{max}$ [deg]	Fabrication technologies	Applications
[31]	900 MHz	52 ($\epsilon_r=1-80$)	PMMA + micromachining CPW (CRLH TL with shorted stub)	General microfluidic
[32]	1.84 GHz	94 ($\epsilon_r=1-80$)	CPW with PDMS microfluidic channel on top of glass wafer Borofloat33	General microfluidic
[33]	1 Hz-10 MHz	90 ($\epsilon_r=1-80$)	SOI wafer with Cr/Au electrodes, PDMS fluidic	Red blood cell characterization
[34]	2-10 GHz	136 (at 6 GHz) ($\epsilon_r=1:36$)	RF MEMF CVD Parylene, Parylene surface- micromachining, Silicon, Cr/Au electrodes	CWP phase shifter with integrated micropumps
This work	6 GHz	86 ($\epsilon_r=1-80$)	3D printing Microstrip	General microfluidic/ Toluene sensor

342 **6. Conclusions**

343 A novel microfluidic microstrip EBG sensor realized using simple 3D printing process has been
344 proposed in this article. The proposed sensor is composed of the microstrip line, defected EBG
345 structure etched in the ground plane and microfluidic channel embedded in the microstrip substrate.
346 The EBG structure with single defect placed beneath the channel is used as pattern in the ground
347 plane to improve sensor sensitivity. The microfluidic channel is fabricated using conventional 3D
348 printing technique without any supporting material simultaneously with the microstrip substrate.

349 The operating principle of the sensor is based on phase shift measurement of the propagation
350 signal at single operating frequency. When a fluid flows in the microfluidic channel, the phase of the
351 propagating signal changes due to the different permittivity of the fluid. The sensor dimensions were
352 optimized using electromagnetic simulations while performances of the proposed sensor were
353 validated by measuring the phase response for different fluids in the microchannel.

354 The measurement results of the fabricated sensor show that the change of the permittivity of
355 fluids in the microchannel from 1 to 80 results in the phase shift of 86° . Moreover, the proposed sensor
356 shows relatively high and almost linear sensitivity for fluids which dielectric constant is lower than
357 30. The potential application is demonstrated by the implementation of proposed sensor for the
358 detection of toluene concentration in toluene-methanol mixture. The experimental results show that
359 the phase difference linearly changes when the concentration of toluene changes from 0% to 100%
360 with sensitivity of 0.54° per percentage of toluene.

361 In this paper, we propose a novel low-cost, reusable, and easily fabricated design that uses small
362 volumes of fluid. The proposed sensor is characterized with relatively high sensitivity and linearity
363 which makes it suitable candidate for monitoring small concentration of specific fluid in different
364 mixtures. For future consideration, we are planning to design supporting electronic, to minimize the
365 sensor dimensions, and improve its sensitivity. The sensor will be further tested on various fluids
366 and mixtures used in biomedical applications and industry.

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