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# High-Frequency InGaAs Tri-gate MOSFETs with fmax of 400 GHz

### C. B. Zota, F. Lindelöw, L.-E. Wernersson and E. Lind

We report on extremely scaled down tri-gate RF MOSFETs utilizing lateral nanowires as the channel, with gate length and nanowire width both of 20 nm. These devices exhibit simultaneous extrapolated  $f_t$  and  $f_{max}$  of 275 and 400 GHz at  $V_{DS} = 0.5$  V, which is the largest combined  $f_t$  and  $f_{max}$ , as well as the largest  $f_{max}$  reported for all III-V MOSFET.

Introduction: Tri-gate (or non-planar) MOSFETs for RF-applications are motivated by that the use of a high-k oxide, rather than a semiconductor barrier (as in HEMTs) allows for higher gate capacitance in the MOSFET [1-2]. Furthermore, the tri-gate architecture improves short-channel effects, allowing for shorter gate length,  $L_G$ , without degradation of performance due to short-channel effects. Both these points enable higher ideal transconductance,  $g_m$ , in MOSFETs compared to HEMTs, assuming similar electron mobility. In fact, stateof-the-art III-V MOSFET devices exhibit  $g_m$  larger than that of record HEMTs, although they presently do not allow RF-compatible device designs [3-5].

In this work, we present RF-compatible tri-gate  $In_{0.85}Ga_{0.15}As$  MOSFETs utilizing lateral nanowires (NWs) as the channel. Compared to our previous work, we have here further scaled down device dimensions,  $L_G$  and nanowire width,  $W_{NW}$  [6]. This enables higher  $g_m$  at  $V_{DS} = 0.5$  V, which significantly improves  $f_t/f_{max}$  from 220/305 GHz to 275/400 GHz. The combined  $f_t$  and  $f_{max}$ , as well as the  $f_{max}$  of these devices represent the highest reported values for all III-V MOSFETs.



Fig. 1 Device fabrication and device materials and design

*a* SEM image of the device after contact regrowth,  $L_G$  is defined as the distance between n<sup>+</sup> contacts. (111)B denotes the crystal facet of the contact layer.

b Schematic figure of the fabricated device.

*Fabrication:* The device fabrication process is similar to what has been described elsewhere [7]. The device channel consists of 200 lateral In<sub>0.85</sub>Ga<sub>0.15</sub>As nanowires, formed by selective area MOCVD growth on (100) InP:Fe (S.I.) substrate, split over two gate fingers. The nanowire width is 20 nm, and the height is 11 nm. The S/D highly doped regions are formed by a second MOCVD growth step of 40 nm n<sup>+</sup> In<sub>0.63</sub>Ga<sub>0.37</sub>As/100 nm InP with in-situ Sn-doping ( $N_D = 5 \times 10^{19}$  cm<sup>-3</sup>) in the doped layer (Fig. 1a). Subsequently, 1 nm/5 nm Al<sub>2</sub>O<sub>3</sub>/HfO<sub>2</sub> is deposited by ALD and Ti/Pd/Au by thermal evaporation, forming the gate stack. The regrown 100 nm InP is selectively etched by an HCl solution leaving a T-gate. S/D and pad metallization of Ti/Pd/Au completes the process (Fig. 1b).

*Results:* Fig. 2a shows transfer characteristics of a device with  $L_G = 20$  nm measured at DC with a Keithley 4200 semiconductor characterization system. All data is normalized to the total gated periphery of the NWs (7 µm). At  $V_{DS} = 0.5$  V, peak  $g_m$  is 2.1 mS/µm. Fig. 2b shows the scaling behaviour of peak  $g_m$  and on-resistance  $R_{on}$  versus  $L_G$ .  $R_{on}$  reaches 220  $\Omega$ µm at  $L_G = 20$  nm. The total access resistance is estimated to 130  $\Omega$ µm from transmission line measurements.

RF-measurements where performed at 40 MHz to 67 GHz with an Agilent E8361A vector network analyser. On-chip pad de-embedding as well as off-chip two-port load-reflect-reflect-match calibration was performed. The total pad capacitances were approximately 20 fF.



Fig. 2 Device characteristics at DC.

*a* Transfer characteristics of a  $L_{\rm G} = 20$  nm device.

b Scaling behaviour of peak  $g_m$  and  $R_{on}$  versus  $L_G$ .

A small-signal model was determined from the measured Sparameters, with a good fit to the measurement data [8]. Fig. 3 shows measured and modelled (dashed traces) unilateral power gain |U|, current gain  $|h_{21}|^2$  and maximum available/stable gain (|MAG| and |MSG|) for a device with  $L_G = 20$  nm. Extrapolated cut-off frequency  $f_t$ is 275 GHz and maximum oscillation frequency  $f_{max}$  is 400 GHz.



**Fig. 3** Measured and modelled (dashed traces) gain of an  $L_G = 20$  nm device at  $V_{DS} = 0.5$  V.

The small-signal model, which is similar to that in [6], includes both the effect of border traps in the oxide, and impact ionization. Border traps are modelled using the distributed border trap model in [9]. Border traps introduce a frequency-dependency to  $g_{\rm m}$  and  $g_{\rm d}$ , as well as a frequency-dependency to  $g_{\rm m}$  and  $g_{\rm d}$ , as well as a frequency-dependent oxide loss, and explain the -10 dB slope of |U| versus f [10]. Fig. 4a shows  $g_{\rm m,peak}$  for an  $L_{\rm G} = 20$  nm device extracted from the small-signal model at DC and 67 GHz (RF).  $g_{\rm m,peak}$  increases by approximately 13% in the latter case, to a maximum of 2.9 mS/µm at  $V_{\rm DS} = 1.25$  V, which is attributable to that trap responses are partially disabled at high frequency.

The effective gate resistance is ~5  $\Omega$ , and the source and drain resistances are ~2  $\Omega$ . The gate-to-source and gate-to-drain capacitances,  $C_{\rm GS}$  and  $C_{\rm GD}$ , are shown in Fig. 4b. At  $V_{\rm DS} = 0.5$  V, the total gate capacitance  $C_{\rm GS} + C_{\rm GD}$  is 15 fF at peak  $g_{\rm m}$ . This includes both the parasitic capacitance from the source and drain gate overlaps, and the intrinsic gate capacitance. The latter is estimated as  $C_{\rm gg,int} = (2/3)WLC_{\rm ox}/(C_{\rm q} + C_{\rm ox})$ , with the quantum capacitance  $C_{\rm q} = q^2 m^*/\pi \hbar^2$ , which is ~2 fF with m\* = 0.04 $m_0$ . Thus, RF-performance is primarily limited by the parasitic overlap capacitance, which can be lowered by implementation of source and drain spacers.



**Fig. 4** *Peak* g<sub>m</sub> and capacitances

*a* Peak  $g_m$  measured at both 40 MHz (DC) and 67 GHz (RF), for an  $L_G$  = 20 nm device.

*b* Gate-to-source,  $C_{GS}$ , and gate-to-drain,  $C_{GD}$ , capacitances measured at different  $V_{DS}$ .

Fig. 5 shows a benchmark of  $f_i$ ,  $f_{max}$  and the geometric mean  $\sqrt{f_t} \times f_{max}$  (dashed traces) for state-of-the-art III-V MOSFETs [11-18]. The geometric mean is 330 GHz for these devices, which is the highest reported value for a III-V MOSFET. Squares show planar devices, and triangles show non-planar devices.



Fig. 5 Benchmark of RF-performance for III-V MOSFETs

Squares show planar devices, triangles show non-planar devices.  $V_{DS}$  and  $L_G$  varies between devices, but is 0.5 V and 20 nm, respectively, for this work. Dashed traces show the geometric mean.

*Conclusion:* We have demonstrated  $L_{\rm G} = 20$  nm In<sub>0.85</sub>Ga<sub>0.15</sub>As tri-gate MOSFETs with record high-frequency performance,  $f_{\rm t} = 275$  GHz and  $f_{\rm max} = 400$  GHz at  $V_{\rm DS} = 0.5$  V.

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