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A novel control method to maximize the energy-harvesting capability of an adjustable slope angle wave energy converter

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ABSTRACT

This paper introduces a novel control approach to maximizing the output energy of an adjustable slope angle wave energy converter (ASAWEC) with oil-hydraulic power take-off. Different from typical floating-buoy WECs, the ASAWEC is capable of capturing wave energy from both heave and surge modes of wave motions. For different waves, online determination of the titling angle plays a significant role in optimizing the overall efficiency of the ASAWEC. To enhance this task, the proposed method was developed based on a learning vector quantitative neural network (LVQNN) algorithm. First, the LVQNN-based supervisor controller detects wave conditions and directly produces the optimal titling angles. Second, a so-called efficiency optimization mechanism (EOM) with a secondary controller was designed to regulate automatically the ASAWEC slope angle to the desired value sent from the supervisor controller. A prototype of the ASAWEC was fabricated and a series of simulations and experiments was performed to train the supervisor controller and validate the effectiveness of the proposed control approach with regular waves. The results indicated that the system could reach the optimal angle within 2s and subsequently, the output energy could be maximized. Compared to the performance of a system with a vertically fixed slope angle, an increase of 5% in the overall efficiency was achieved. In addition, simulations of the controlled system were performed with irregular waves to confirm the applicability of the proposed approach in practice.

Key Words: Wave energy converter, ASAWEC, oil-hydraulic power take-off, learning vector quantitative neural network, control systems.

1. Introduction

Among various sources of renewable energy, ocean wave energy is important, as it has significant potential in many locations, due to its relatively high power density and predictability

1 [1]. Therefore, wave energy conversion technologies have gained more attention to meet the
2 increasing demand for electrical power.

3 To harvest energy from waves, floating-buoy type wave energy converter (WEC) is the simplest
4 and most popular design, and has been a focus of research. To maximize energy extraction from
5 waves, the WEC must reach two optimal conditions, amplitude and phase, for each sinusoidal
6 incident wave [2]. To satisfy the amplitude condition, the amplitudes of the radiated waves must
7 be exactly half that of the incident waves [2,3]. To satisfy the phase condition, the oscillating
8 velocity of the body must be in phase with the excitation force on the body. This can be achieved
9 using phase control to obtain the resonance condition in which the wave frequency equals the
10 natural frequency of the body. Here, two main control strategies, reactive control and latching
11 control, are applied to WEC devices. Several interesting works have been reported [4-8]. Although
12 phase control is capable of substantially increasing the amount of absorbed energy,
13 implementation of this technique in real irregular waves has met both theoretical and practical
14 difficulties that have not been satisfactorily overcome [9]. In addition, a phase control strategy
15 requires governing equations for the body oscillations that are difficult to derive under real-world
16 conditions. Modeling the highly non-linear behavior of both friction and wave characteristics is
17 problematic.

18 To overcome the difficulty in regulating the power flow of WECs, neural networks (NNs) can
19 be used to adaptively control the power-take-off (PTO) mechanisms [10]. As reported in [11,12],
20 NNs were used to forecast the wave information in the near future to adjust in advance the PTO
21 force. In another study [13], NN was utilized to derive the heuristic relationship between the
22 system inputs and the control parameters. Although remarkable results were obtained using these
23 approaches, the development and use of the control logic is complex, which restricts their
24 applicability. Therefore, a simple and efficient way of maximizing WEC productivity without
25 information on system dynamics is required.

26 To meet that requirement, this study focuses on a particular type of floating-buoy WEC, which
27 was recently introduced: the adjustable slope angle wave energy converter (ASAWEC) [14]. The
28 ASAWEC comprises mainly a sliding-buoy structure and a hydrostatic transmission-based PTO
29 system. Different from typical floating-buoy designs, which are normally fixed vertically, an
30 efficiency optimization mechanism (EOM) was integrated in the ASAWEC to enable adjustment
31 of the system slope (or tilting angle) to increase the capture ratio from waves in both heave and
32 surge modes. An analytical study of the interaction between waves and a buoy in the horizontal
33 and vertical directions was carried out by Heikkinen et al. [15]. The results indicated that by
34 combining the two modes to create the cylinder movement, the amount of absorbed energy could
35 be increased. In the study by Think et al. [14], the effect of non-vertical linear motion of a

1 hemispherical-float wave energy converter was evaluated by both numerical simulations and
 2 experiments. The sliding angle exerted a significant impact on the energy capture ratio. Therefore,
 3 this paper aims to develop a simple control approach to maximize energy harvesting capability of
 4 the ASAWEC by regulating its slope angle. A control scheme that includes a supervisor controller
 5 and a secondary controller for the EOM mechanism is designed. The supervisor controller is
 6 constructed using a learning vector quantitative neural network (LVQNN) algorithm to classify
 7 wave conditions based on limited wave information, and so produces the optimal titling angles.
 8 The secondary controller with a simple control algorithm is used to drive the EOM mechanism to
 9 regulate automatically the ASAWEC slope angle to reach the value determined by the supervisor
 10 controller. In this way, the energy capture ratio is improved.

11 The remainder of this paper is organized as follows: in Section 2, the ASAWEC configuration
 12 is briefly introduced and, a mathematical model is developed for further investigation; in Section
 13 3, a prototype of the suggested ASAWEC is fabricated and the experimental apparatus is discussed;
 14 the LVQNN-based on EOM control scheme is constructed and optimized using training data in
 15 Section 4; the proposed approach is evaluated by both numerical simulations and real-time
 16 experiments in Section 5; and concluding remarks are provided in Section 6.

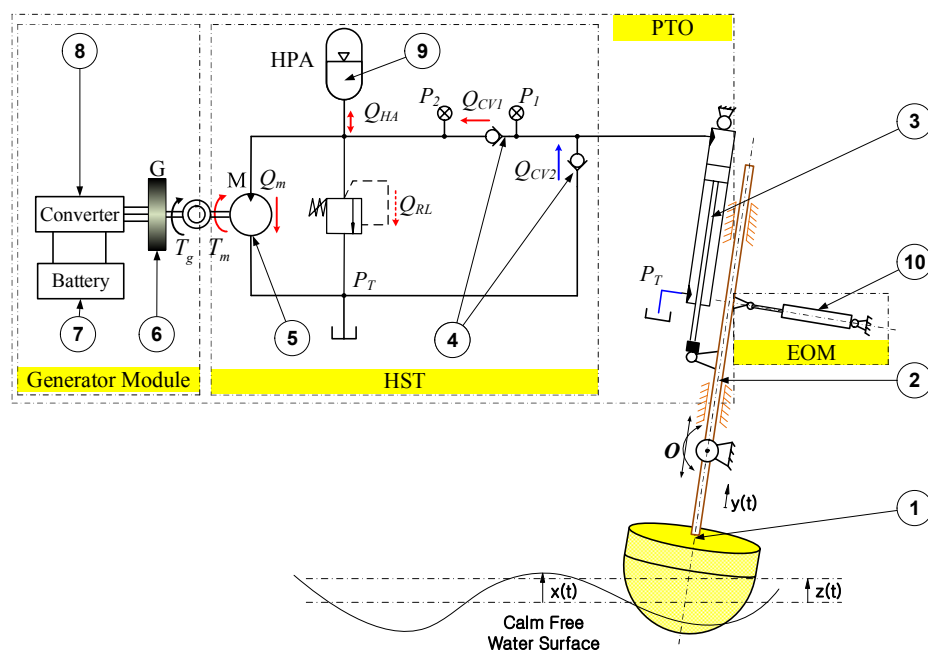
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18 2. EOM-Based ASAWEC design and modeling

19

20 2.1. EOM-based ASAWEC configuration

21 To maximize the ability to harvest energy from waves, an ASAWEC device design is suggested
 22 in Fig. 1.



23

(1) Sub-buoy

(2) Sliding shaft

- | | |
|-------------------------------------|------------------------|
| (3) Hydraulic cylinder | (4) Check valves |
| (5) One-directional hydraulic motor | (6) Electric generator |
| (7) Battery | (8) Converter |
| (9) High pressure accumulator | (10) EOM |

Fig. 1 Configuration of the proposed ASAWEC

The ASAWEC includes the following two modules:

- PTO module: converts wave energy into electric energy. This consists of a hydrostatic transmission (HST) and an electrical generator. The system interacts with waves through a sub-buoy jointed with a sliding shaft. This shaft can slide along sliding bearings or through rollers fixed on the device frame. The sliding shaft is then linked in parallel to a non-symmetric hydraulic cylinder. To convert mechanical energy into hydraulic energy, the HST is a simple hydraulic circuit with a one-directional hydraulic motor, a high-pressure accumulator (HPA), check valves and a small oil sump. The mechanical energy of the PTO is transmitted to the HST through the large chamber of the cylinder, while the small chamber is connected to the oil sump. For safety, a pressure relief valve is used to protect the system from damage due to extremely high power waves. Next, a generator block is employed to generate electric energy from hydraulic energy. This block comprises an electrical generator, a converter and an electricity storage device, such as battery. The output shaft of the hydraulic motor is coupled to the generator shaft to generate electricity.
- EOM module: designed based on the ASAWEC configuration. A linear actuator with an appropriate driver is selected as the power source for the EOM. A controller is designed for this mechanism in order to adjust automatically the tilting angle of the PTO according to wave conditions in such a way that the device can absorb most of the energy from waves.

During operation, waves force the sub-buoy to move up and down based on the floating-buoy concept. Here, only upward motion of the sub-buoy is utilized for energy harvesting. During upward motion, the cylinder is retracted and a high-pressure flow is created in the large cylinder chamber. This pressurized flow enters the hydraulic circuit through the first check valve (CV1) and reaches the inlet of the hydraulic motor. Consequently, in this case the electric generator operates to generate electricity. In contrast, during downward motion of the cylinder, the low-pressure flow is supplied from the oil sump to fill the large cylinder chamber through the second check valve (CV2). The pressure at the hydraulic motor side is not affected in this case. To store redundant energy generated by large waves and to facilitate smooth performance of the electric generator between upward and downward motion of the sub-buoy, an HPA accumulator is employed. Using this HPA, the generator speed does not decrease to zero when the cylinder extends, and so, the system performance is improved.

1

2 *2.2. Modeling of the ASAWEC device*

3

4 *2.2.1. Equations of motion*

5 Generally, real ocean wave evaluation can be represented as [16]:

6
$$x = \sum_{n=-\infty}^{+\infty} Z_n \exp^{i2\pi nft} dt \quad (1)$$

7 where $Z_n = (1/T) \int_{-T/2}^{T/2} x(t) \exp^{-i2\pi nft} dt$, ($n = 0, 1, 2, \dots$), $f = 2\pi/T$ is the fundamental frequency.8 The mean wave power (P_{wave}) can be described as a function of the mean wave energy density
9 and the group velocity [2]:

10
$$P_{wave} = E c_g \quad (2)$$

11 where E is the mean wave energy density per horizontal unit area and is computed as:

12
$$E = E_k + E_p = \frac{1}{8} \rho g^2 H^2 \quad (3)$$

13 and c_g is the group velocity. For a constant water depth h at near shore locations (neither deep
14 nor shallow water), the group velocity is obtained as follows:

15
$$c_g = \frac{D(kh)}{2 \tanh(kh)} c_p = \frac{g}{2\omega} D(kh) \quad (4)$$

16 where $D(kh)$ is the depth function:

17
$$D(kh) = \left[1 + \frac{2kh}{\sinh(2kh)} \right] \tanh(kh) \quad (5)$$

18 here k is the wave number and is determined as follows: $k = 2\pi / \lambda$.19 An analysis of forces acting on the PTO mechanism is shown in Fig. 2. Assume that the PTO
20 tilting angle is adjusted to angle α_{WEC} by the EOM. The movement of the sub-buoy as well as the
21 cylinder rod is then on an incline axis tagged as $y(t)$. The buoy motion is then obtained following
22 Newton's second law:

23
$$M\ddot{y} = F_{WEC} + F_u = F_w \cos(\alpha_{WEC} - \alpha_w) + F_u \quad (6)$$

24 where M is the total mass of the moving parts, mainly the sub-buoy, sliding shaft and piston rod;
25 α_w is the direction of the sum of hydrodynamic forces; F_u is the force supported by a control
26 device assisting the oscillating body operation; and F_w is the sum of hydrodynamic forces acting
27 on the cylinder rod.

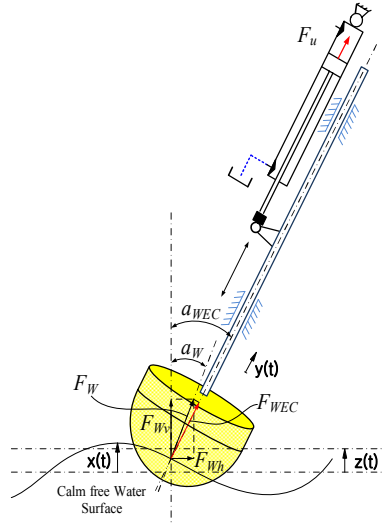


Fig. 2 Analysis of forces acting on the PTO mechanism

The sum of hydrodynamic forces acting on the cylinder rod can be obtained by the following relation:

$$\vec{F}_w = \vec{F}_{wv} + \vec{F}_{wh} \quad (7)$$

where \vec{F}_{wv} and \vec{F}_{wh} are in turn the hydrodynamic forces in the vertical and horizontal directions.

In the vertical direction with heave motion [2], the hydrodynamic forces can be derived as:

$$F_{wv} = F_{ev} + F_{rv} + F_b + F_v + F_f - Mg \quad (8)$$

where F_{ev} is excitation force, F_{rv} is the radiated force acting on the buoy in heave motion (vertical direction); F_b is hydrostatic buoyancy force; F_v is the viscosity effect and F_f is the friction effect.

In the horizontal direction with drag motion [2], the hydrodynamic force is computed as:

$$F_{wh} = F_{eh} + F_{rh} \quad (9)$$

where F_{eh} is excitation force; F_{rh} is the radiated force acting on the buoy in surge motion (horizontal direction).

The excitation force can be expressed as in equation (10):

$$F_w = \sqrt{F_{wh}^2 + F_{wv}^2} \quad (10)$$

2.2.2. HST model

From Fig. 2, the force F_u can be derived as:

$$F_u = F_{PTO} + F_{mf} \quad (11)$$

where F_{mf} is the total mechanical friction force existing in the system and F_{PTO} is the force generated by the fluid in the cylinder:

$$F_{PTO} = P_1 A_{LC} - P_T A_{SC} \quad (12)$$

1 where P_1/A_{LC} and P_T/A_{SC} are the pressure/section area of the cylinder large and small chambers,
 2 respectively. These areas are obtained from the piston bore, D_{cyl} , and piston rod, d_{cyl} , as:

$$3 \quad A_{LC} = \frac{\pi D_{cyl}^2}{4}; A_{SC} = \frac{\pi(D_{cyl}^2 - d_{cyl}^2)}{4} \quad (13)$$

4 In this system configuration, P_T is neglected because the small cylinder chamber is connected
 5 to the tank. Then, the reactive force F_{PTO} is generally determined from the characteristics of the
 6 HTS:

$$7 \quad F_{PTO} = P_1 A_{LC} \quad (14)$$

8 where P_1 is the pressure inside the bore chamber:

$$9 \quad \frac{dP_1}{dt} = \frac{\beta}{V_0 - A_{LC}z} (A_{LC}\dot{z} - Q_{cv1} + Q_{cv2}) \quad (15)$$

10 where β is the bulk modulus of fluid; L is the cylinder stroke, and Q_{cv1} and Q_{cv2} are the flow
 11 rates through check valve 1 and check valve 2, respectively. They are obtained using the following
 12 equations [17]:

$$13 \quad \begin{cases} Q_{cv1} = C_d A_{o1} \sqrt{\frac{2(P_1 - P_2)}{\rho_f}} & \text{if } P_1 > P_2 \\ Q_{cv1} = 0, & \text{if } P_1 \leq P_2 \end{cases} \quad (16)$$

$$14 \quad \begin{cases} Q_{cv2} = C_d A_{o2} \sqrt{\frac{2(P_t - P_1)}{\rho_f}} & \text{if } P_t > P_1 \\ Q_{cv2} = 0, & \text{if } P_t \leq P_1 \end{cases} \quad (17)$$

15 where C_d is the discharge coefficient, A_{o1} and A_{o2} are the working area of each check valve, ρ_f
 16 is the fluid density, P_t is the pressure of the fluid inside the tank, and P_2 is the pressure in the
 17 chamber connected to the high-pressure accumulator and can be obtained by the continuity:

$$18 \quad \frac{dP_2}{dt} = \frac{\beta}{V_p + V_{ha}} (Q_{cv1} - Q_{ha} - Q_m) \quad (18)$$

19 where Q_m and Q_{ha} are the flow rates through the hydraulic motor and into the HPA, respectively;
 20 V_p is the volume of fluid inside the segment pipe that connects check valve 1 to the input port of
 21 the motor, and V_{ha} is the fluid volume in the accumulator.

22 The flow rate Q_{ha} and fluid volume V_{ha} , which enter the accumulator, are calculated as:

$$V_{ha} = \begin{cases} 0, & \text{if } P_2 \leq P_0 \\ V_0 \left[1 - \left(\frac{P_0}{P_2} \right)^{\frac{1}{n}} \right], & \text{if } P_2 > P_0 \end{cases} \quad (19)$$

$$Q_{ha} = \dot{V}_{ha} = \begin{cases} 0, & \text{if } P_2 \leq P_0 \\ \frac{1}{n} V_0 \left(1 - \frac{P_0}{P_2} \right)^{\frac{1-n}{n}} \frac{P_0 \dot{P}_2}{P_2^2}, & \text{if } P_2 > P_0 \end{cases} \quad (20)$$

where V_0 is the accumulator capacity, P_0 is the pre-charge pressure, P_2 is the inlet gauge pressure of accumulator and n is the specific heat ratio.

And the flow rate via the hydraulic motor is defined as:

$$Q_m = D_m \omega_m / \eta_v \quad (21)$$

where D_m , ω_m and η_v are the displacement, angular speed and volumetric efficiency of the motor.

The actual output torque of the motor is given as:

$$\tau_m = \Delta p D_m \eta_m / 2\pi \quad (22)$$

where, Δp is the pressure difference between two ports of the motor, and η_m is the mechanical efficiency.

The electric power generated by the electric generator with overall efficiency η_g can be evaluated as:

$$P_{generator} = \eta_g 2\pi \omega_g \tau_g = \eta_g 2\pi \omega_m \tau_m \quad (23)$$

where τ_g and ω_g are in turn the torque and angular speed of the electric generator.

16

17 2.2.3. EOM force analysis

18 To drive the shaft-buoy to the optimal angle, an EOM using a linear actuator is integrated into
 19 the ASAWEC (Fig. 1). To select this actuator, which is capable of raising or lowering the shaft-
 20 buoy mechanism irrespective of wave lifting force, a moment analysis around the fulcrum ‘‘O’’ is
 21 carried out for the actuator force and shaft-buoy gravity force (Fig. 3). The moment balance
 22 equation, therefore, can be derived as follows:

$$Mg \sin \alpha_{WEC} a = F_{LA} \sin \alpha_{EOM} b \quad (24)$$

24 where F_{LA} is the desired force to rotate the shaft-buoy system, and a and b are instant distances
 25 from the buoy gravity mass center and linear actuator to the fulcrum ‘‘O’’ when the system is at
 26 angle α_{WEC} .

27

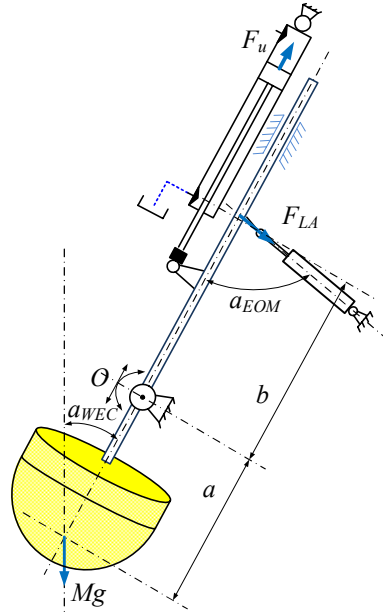


Fig. 3 Detail view and force analysis of system

3. ASAWEC prototype and experimental setup

To investigate the performance of the proposed ASAWEC and the effect of the EOM mechanism on energy harvesting efficiency, a mini-scale ASAWEC prototype was produced.

3.1. ASAWEC component selection

In the ASAWEC design, the buoy (Fig. 4) is of a composite material of which the outer and inner radiuses are R_b and r_b , respectively. The buoy mass is calculated as:

$$m_b = \rho(V_s + V_c + V_{c1}) = \rho \left(\frac{2\pi}{3}(R_b^3 - r_b^3) + \pi(R_b^2 - r_b^2)h_b + \pi R_b^2(H_b - R_b - h_b) \right) \quad (25)$$

where V_s, V_c and V_{c1} are the volumes of the spherical part, cylindrical part and top cover, ρ is the composite density, and h_b and H_b are in turn height of the cylinder part and total height of the buoy.

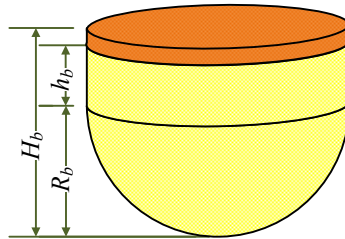


Fig. 4 Buoy shape

1 Using buoy parameters of $R_b=0.6m, r_b=0.55m, H_b=1m, h_b=0.37m$ and $\rho=800kg/m^3$, the mass of
 2 buoy is approximated as $m_b=200kg$

3 Here, a magneto-rheological brake (MR Brake) of which the load torque, τ_{MR} , is adjustable from
 4 0 to $5.6Nm$ is used to represent the electric generator. Based on the model developed in the previous
 5 section, the key components for the system are as follows:

6 From Eq. (22), the motor displacement is:

$$7 \quad D_m \geq \frac{2\pi \times \tau_m}{\Delta p \times \eta_m} = \frac{2\pi \times \tau_{MR \max}}{\Delta p \times \eta_m} \approx 11.16 cm^3 / rev \quad (26)$$

8 where, the motor mechanical efficiency and maximum working pressure are
 9 $\eta_m=0.9$ and $\Delta p=35bar$.

10 The maximum motor speed should be equal or greater to the maximum MR Brake speed
 11 ($n_{MR \max}=1000rpm$). Therefore, a fixed displacement gear motor with OML12.5 series
 12 manufactured by Danfoss with the following characteristics is selected:

$$13 \quad D_m=12.5cm^3 / rev, n_{m \max}=1280rpm, T_{\max}=11Nm, \eta_m=\eta_v=0.85.$$

14 The flow rate required to supply the motor for the maximum generator speed is then obtained:

$$15 \quad Q_{m \max} \geq \frac{D_m n_{MR \max}}{\eta_v} \approx 14.71 lpm \quad (27)$$

16 Here, the mini-scale ASAWEC is designed for evaluation with single waves of frequency of
 17 up to 1.5 Hz. Thus, based on the required flow rate, a cylinder with the following parameters is
 18 used: $D=0.025m, d=0.012m, L=0.5m$.

19 Next, the linear actuator is selected based on the power on demand. From (24), the force F_{LA}
 20 that must be supplied by this actuator can be calculated as:

$$21 \quad F_{LA} = \frac{F_G \sin \alpha_{WEC} a}{\sin \alpha_{POM} b} \quad (28)$$

22 Practical testing on the ASAWEC indicates that, the tilting angle should be limited to 0 to 20°.
 23 Initial values of the distances a and b corresponding to the initial tilting angle (0°) are 1500 and
 24 390 mm. The electric linear actuator made by Zhejiang Corp. with the following characteristics is
 25 used: 0.8kW power, 300mm stroke, and 7000N load capacity.

26 The specifications of the main components of the proposed system are summarized in Table
 27 A1 in the Appendix.

28

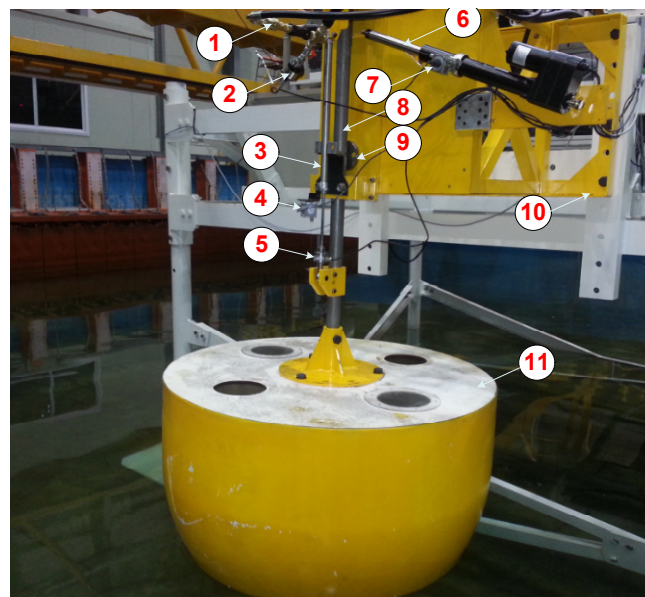
29 3.2. Experimental apparatus

30

1 Using the components outlined in the previous section, the scaled ASAWEC system with the
2 configuration shown in Fig. 2 was fabricated. A photograph of the system is shown in Fig. 5. To
3 acquire the necessary information for system management and evaluation, sensors were attached
4 to the system. The sub-buoy motion (y) was detected by a cable sensor, while the working
5 pressures (P_1 and P_2) and flow rate (Q_m) of the HST were measured by two pressure transducers
6 and a flow meter. To assess the output energy capable of generating electricity, a torque transducer
7 integrated with a speed sensor was used to obtain the generated torque and speed of the MR brake
8 (T_g and n_g , respectively). Furthermore, another cable sensor was installed along the linear actuator
9 to monitor indirectly the tilting angle of the sliding-buoy system. A personal computer (PC) was
10 used to receive signals fed-back from the sensors and derive outputs to control the ASAWEC via
11 an NI multi-function card. Software was developed within the Matlab/Simulink environment
12 combined with the Real-time Windows Target toolbox to facilitate this task.

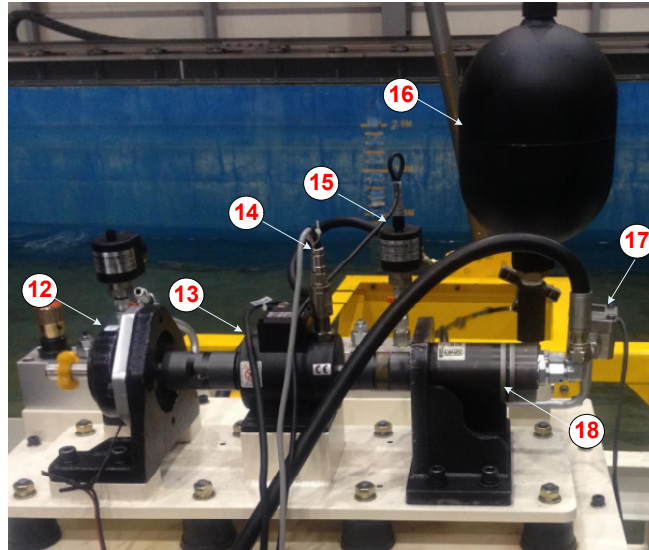
13 To investigate the performance of the EOM-based ASAWEC, a water tank integrated with a
14 wave maker was created (Fig. 6). The specifications of the water tank and wave maker are shown
15 in Table A2 in the Appendix.

16
17



18
19
20
21

(a) PTO & EOM (1-Check Valve; 2-Pressure sensor 1; 3-Hydraulic cylinder; 4-Cable sensor 1;
5-Load cell; 6-Linear actuator; 7-Cable sensor 2; 8-Stell shaft; 9-Fulcrum “O”; 10-Base frame;
11-Sub-buoy)



(b) HTS & MR brake (12-MR brake; 13-Torque transducer; 14-Speed sensor; 15-Pressure sensor; 16-HP accumulator; 17-Flow meter; 18-Pump)

Fig. 5 Graphical view of the experimental EOM-based ASAWEC

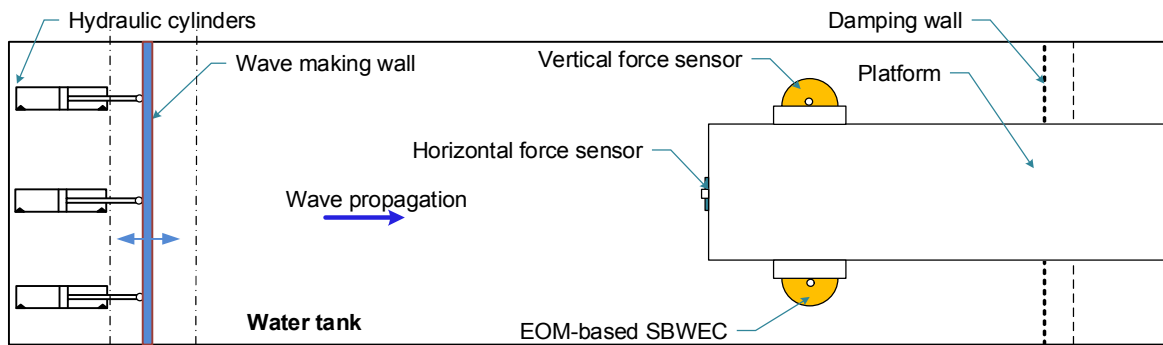


Fig. 6 System installation in water tank integrated wave maker

4. EOM controller design

4.1. EOM overall control concept and secondary controller

In this section, a controller is designed to drive the EOM to optimize the buoy-shaft tilting angle according to wave conditions. The control system configuration is depicted in Fig. 7. The controller consists of two sub-controllers: a supervisor controller and a secondary controller. The supervisor controller is designed using the LVQNN technique to detect each wave condition based on the wave information, tagged as η_{wave} , and then, to derive the optimal tilting angle (or desired angle) for the ASAWEC, tagged as α_r . The secondary controller uses a simple proportional-integral (PI) algorithm to drive the EOM to change the sliding-buoy mechanism to follow the desired angle. As a result, the maximum amount of wave energy can be absorbed.

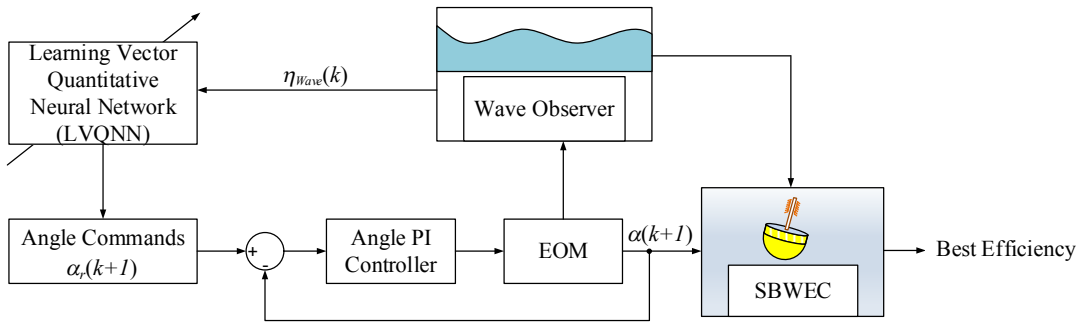


Fig. 7 EOM control architecture for maximizing energy performance of ASAWEC

The PI controller was designed based on the model-based PID tuning function of MATLAB. Using the system identification toolbox of MATLAB, the EOM model was derived based on the input and output data of the linear actuator (driving voltage command and cylinder displacement, respectively) installed in the ASAWEC. Next, the resultant EOM model was combined with the ASAWEC model to tune the PI controller. As a result, the PI proportional and derivative gains, were achieved in turn as: $k_p = 0.8$ and $k_i = 0.01$.

4.2. LVQNN-based supervisory control

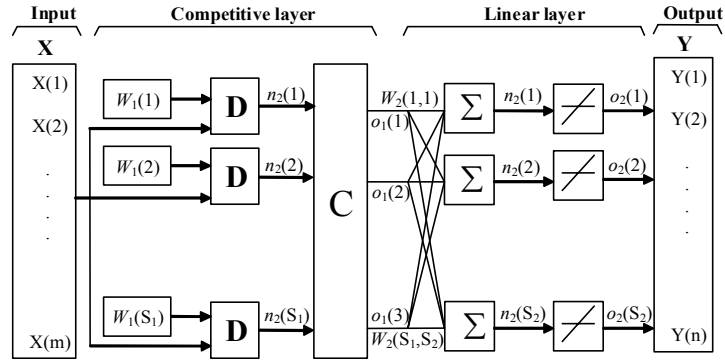
4.2.1. Learning vector quantitative neural network

Neural networks (NNs) emulate the activity of biological neural networks. They can learn from a set of data and construct weight matrices to represent the learning patterns. This technique modifies its behavior in response to the environment and is ideal for unknown expected mapping algorithms and when tolerance to faulty input information is required. NNs have been used successfully for applications such as pattern classification, decision making, forecasting, and adaptive control [18].

Generally, NNs can be classified according to the learning process: supervised and unsupervised learning. Supervised learning is training using the desired responses to given stimuli, while unsupervised learning is classification by “clustering” of stimuli, without specified responses. LVQNN is a hybrid network that uses advanced behaviors of both competitive learning and applies a Kohonen self-organizing map (SOM) or Kohonen feature map for classification.

Fig. 8 shows the structure of a LVQNN. This LVQNN contains four layers: one input layer with m nodes, two hidden layers, and one output layer with n nodes. The first hidden layer is termed the competitive layer with S_1 nodes, while the second output layer is termed the linear layer with S_2 nodes, (in this case, $S_2 \equiv n$).

1 During operation, the competitive layer first maps input vectors into the clusters found by the
 2 network through a training process. Secondly, the linear layer merges groups of these clusters into
 3 classes defined by the target data. The total number of clusters in the competitive layer is
 4 determined by the number of hidden neurons. The larger the hidden layer, the more clusters the
 5 competitive layer can learn, and the more complex mapping of input to target classes can be
 6 performed [19,20]. Therefore, with appropriate selection of the structure and training of the
 7 weighting factors, the LVQNN can classify the information of any system.



8
9 **Fig. 8** Structure of the LVQNN

10 The LVQNN is based on the nearest-neighbor method by calculating the Euclidean distance
 11 weight function, D , for each node, n_j , in the competitive layer as follows:

$$12 \quad n_j = D(X, W_1(j)) = \sqrt{\sum_{i=1}^m (X(i) - W_1(j,i))^2}, \quad j = 1, \dots, S_1 \quad (29)$$

13 where: X is the input vector and $W_1(j,i)$ is the weight of node j^{th} in the competitive layer
 14 corresponding to element i^{th} of the input vector.

15 Next, the Euclidean distances are fed into function C which is a competitive transfer function.
 16 This function returns an output vector o_1 , with 1 if each net input vector has its maximum value,
 17 and 0 otherwise. This vector is then input to the linear layer and derives the output vector o_2 of
 18 each element of which corresponds to a node of the output layer and is computed as:

$$19 \quad Y(k) = o_2(k) = k_w(k) n_2(k) = k_w(k) \sum_{j=1}^{S_1} W_2(k,j) o_1(j), \quad k = 1, \dots, n, (n \equiv S_2) \quad (30)$$

20 where: $W_2(k,j)$ is the weight of node k in the linear layer corresponding to element j of the
 21 competitive output vector; and $k_w(k)$ is the linearized gain of node k in the linear layer.

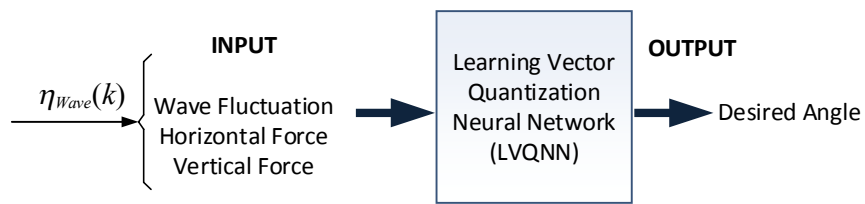
22 In the learning process, the weights of LVQNN are updated by the well-known Kohonen rule,
 23 which is shown as the following equation:

$$24 \quad \begin{cases} W_1^{t+1}(j) = W_1^t(j) + \mu(X - W_1^t(j)) & \text{IF: } X \text{ is classified correctly} \\ W_1^{t+1}(j) = W_1^t(j) - \mu(X - W_1^t(j)) & \text{IF: } X \text{ is classified incorrectly} \end{cases}, \quad j = 1, \dots, S_1 \quad (31)$$

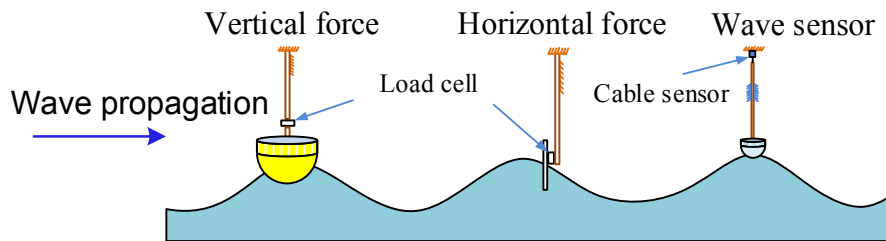
1 where μ is the learning ratio with positive and decreased with respect to the number of training
 2 iterations ($n_{iteration}$), $\mu = \frac{1}{n_{iteration}}$

3
 4 **4.2.2. Application of LVQNN to a supervisor controller design**

5
 6 In this study, the LVQNN was employed to design the supervisor controller to distinguish
 7 different wave conditions in an online manner. In LVQNN, it is important to determine the input
 8 vector size and the number of data sequences. In general, a better result can be achieved with more
 9 inputs. However, this makes the network complex and difficult to train. For application to a
 10 specific water region, the ASAWEC was installed in such a way that its tilting plane was mostly
 11 parallel to the propagation direction of the wave components that contained the highest power
 12 density. Therefore, without loss of generality, to enhance the given task, three signals extracted
 13 from the feedback wave information, η_{wave} —wave fluctuation, hydrodynamic forces in the vertical
 14 direction F_{Wv} and horizontal direction F_{Wh} — were used to create the input vector for the LVQNN,
 15 while the output was a specific wave case used to determine the desired angle (Fig. 9a). To acquire
 16 this information, a cable sensor and two load cells were installed, in which the axis of load cell
 17 measurement of the horizontal force was parallel to the ASAWEC tilting plan (Fig. 9b). It should
 18 be noted that these sensors were not attached to the ASAWEC, being located near the device
 19 because the sensed wave information should not be influenced by the change in system slope angle.
 20 Furthermore, the supervisor controller required only the relative wave behavior to classify wave
 21 conditions, information on transient wave impacts on the system was not required.



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 23 (a)



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 25 (b)

26 **Fig. 9** LVQNN-based supervisor controller:

27 (a) Control configuration; (b) Hardware setup to observe wave information

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To classify ocean waves that contain many different frequencies, the output from the supervisor controller should be a wave condition (called a class), which is mixed from different waves by different mixing ratios. To facilitate this, a so-called smooth switching algorithm was proposed. The wave class was determined by the smooth combination of the current class detected by the LVQNN (Y) and the previous class as:

$$class(t) = \lambda \times class(t-1) + (1-\lambda) \times Y(t) \quad (32)$$

where λ is the forgetting factor.

Similarly, the desired tilting angle is produced by:

$$\alpha_r(t) = \lambda \times \alpha_r|_{class(t-1)} + (1-\lambda) \times \alpha_r|_{Y(t)} \quad (33)$$

Additionally, to prevent the influence of noise on classification performance, the forgetting factor was online tuned with respect to the change in the LVQNN output speed. This factor was tuned as follows:

Step 1: set the initial value of the forgetting factor, $\lambda = 0.5$; define a small positive threshold, $0 < \gamma_1 < \gamma_2$, for the change in LVQNN output speed, v_Y , which is defined by the number of sampling periods during which Y changes continuously.

Step 2: for each working step, check v_Y and update λ by comparing v_Y with γ using the following rule:

+ If: $v_Y = 0$, Then: $\lambda(t+1) = \lambda(t)$;

+ Else If: $(v_Y > \gamma_2)$, Then: $\lambda(t+1) = \lambda(t+1)/2$ and reset $v_Y = 0$;

+ Else If: $(v_Y \geq \gamma_1) \& (v_Y(t) \leq \gamma_2)$, Then: $\lambda(t+1) = \lambda(t+1) \times 2$ and reset $v_Y = 0$;

+ Otherwise, $\lambda(t+1) = \lambda(t)$.

In summary, the procedure to design the supervisor controller for application to real waves is as follows:

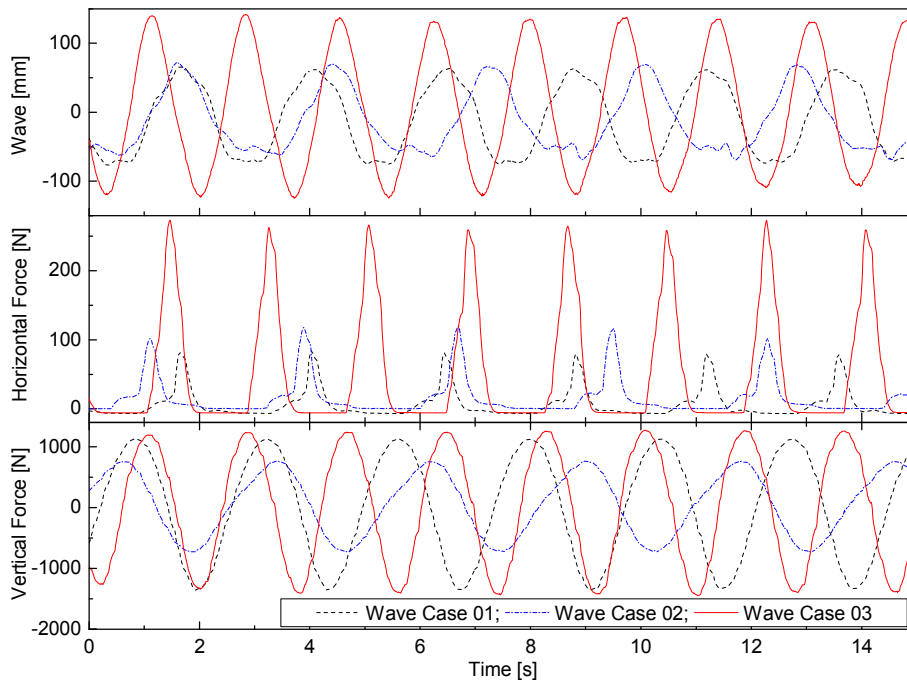
- Step 1: Observe of ocean waves at the location at which the ASAWEC system will be installed.
- Step 2: Analyze the observed waves to identify the major frequencies (for example, a set of seasonal main frequencies)
- Step 3: Train the LVQNN using the selected main frequencies (a set of LVQNN classifiers can be derived for a set of seasonal main frequencies and a real-time clock can be used to switch between these classifiers according to season)
- Step 4: Input the optimized LQVNN into the supervisor controller to regulate the ASAWEC slope.

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4.3. Training of the supervisor controller

4.3.1. Acquisition of data for system performance investigation and training

In a network-training problem, the prior task is to collect the system behavior data to improve the performance of the training process. To perform the investigation, single frequency waves whose specifications were shown in Table 1 were selected. Experiments on the ASAWEC were then carried out for each wave condition in which the tilting angle was altered smoothly within its range, 0 to 20°, using the designed secondary controller (PI). For these experiments, the ASAWEC setting parameters were identical and the overall efficiency of the full system was evaluated. As described in the previous section, the wave factors necessary for performing the LVQNN input vector were recorded for each test case and plotted in Fig. 10. Both the wave elevation and forces were differed and therefore, could be used for wave classification.



15
16

Fig. 10 Data acquisition of three wave factors according to three test cases

17

Table 1

18

Wave specifications and optimal tilting angles

Wave case	Wave specifications		Optimal tilting angle [degree]
	Amplitude [m]	Period [s]	
1	0.12	2.4	6.0
2	0.15	2.8	10.0
3	0.1	1.6	12.0

19

1 According to [21], the energy flux over duration T for the shallow water in the water tank can
 2 be calculated as:

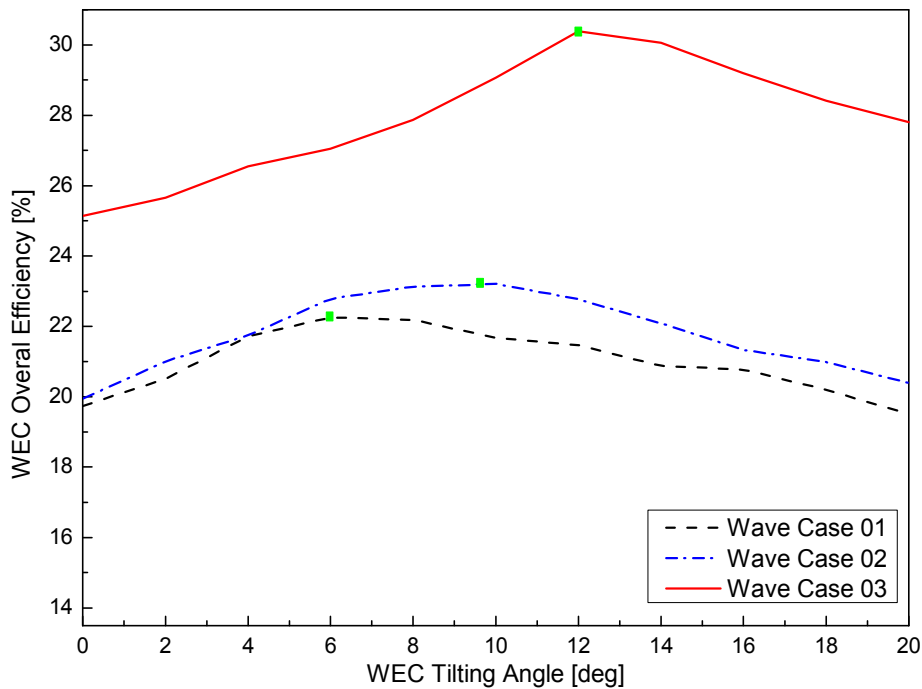
$$3 \quad E_{inTank} = \frac{\rho g^{3/2} H^2 \sqrt{h} T b}{8} \quad (34)$$

4 Since, the average energy conversion efficiency of the system is derived as:

$$5 \quad \eta_{ASAWEC} = \frac{\int_0^T P_{generator}}{E_{inTank}} \quad (35)$$

6
 7 The average efficiency curve of the system versus the tilting angle is shown in Fig. 11. It is
 8 clear that the efficiency differed according to the test case and tilting angle. The angles at which
 9 system efficiency was maximized for the three wave forms were determined (Table 1). In wave
 10 case 3, the greatest increase in overall efficiency of 5% was achieved compared to the case in
 11 which the sliding shaft was fixed in the vertical direction (zero degree, no change in the tilting
 12 angle).

13 Therefore, the supervisor controller was designed to drive the system to the optimal working
 14 point.



15
 16 **Fig. 11** ASAWEC efficiency vs. tilting angle

17
 18 4.3.2. *LVQNN training*

19

In this section, the training of the LVQNN is discussed. The data acquired in Section 4.3.1 were used as the input-output vectors. The wave information and optimal tilting angles data of the three test cases (Table 1) were selected for training the LVQNN.

As mentioned before, determination of a suitable input vector size and number of hidden neurons in the competitive layer is critical for practical implementation of the neural classifier. Here, the input vector size was defined based on the numbers of sequences of the three wave signals (the hydrodynamic forces in the vertical and horizontal directions, the wave frequencies). To investigate the performance of the LVQNN with respect to different structures, training was performed using the selected data set by varying the number of inputs from 6 to 18, and the number of hidden neurons from 20 to 50. After each training process, the correlation between the simulation output and target output was taken as indicative of the success of training. The training results (goodness of fit [%]) of the LVQNN are shown in Fig. 12 and Table 2. These results indicate that the most suitable LVQNN structure comprised 12 nodes in the input vector and 40 nodes in the competitive layer. The highest learning success rate was ~86.84% in that case, which was sufficient for recognition of wave conditions.

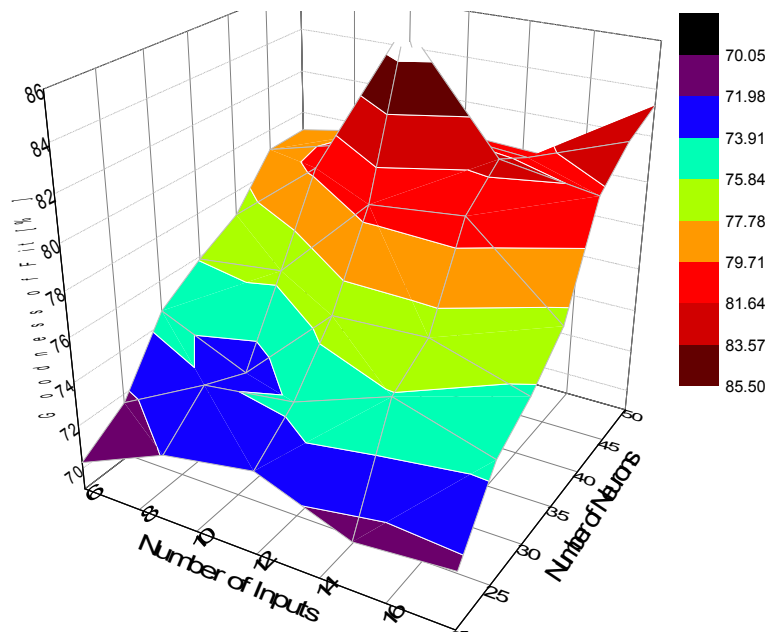


Fig. 12 Goodness of fit – 3D map

Table 2

Learning success rate of LVQNN [%]

Input Number	Number of nodes in hidden layer						
	20	25	30	35	40	45	50
6	70.27	71.59	74.62	75.88	76.91	79.08	79.15

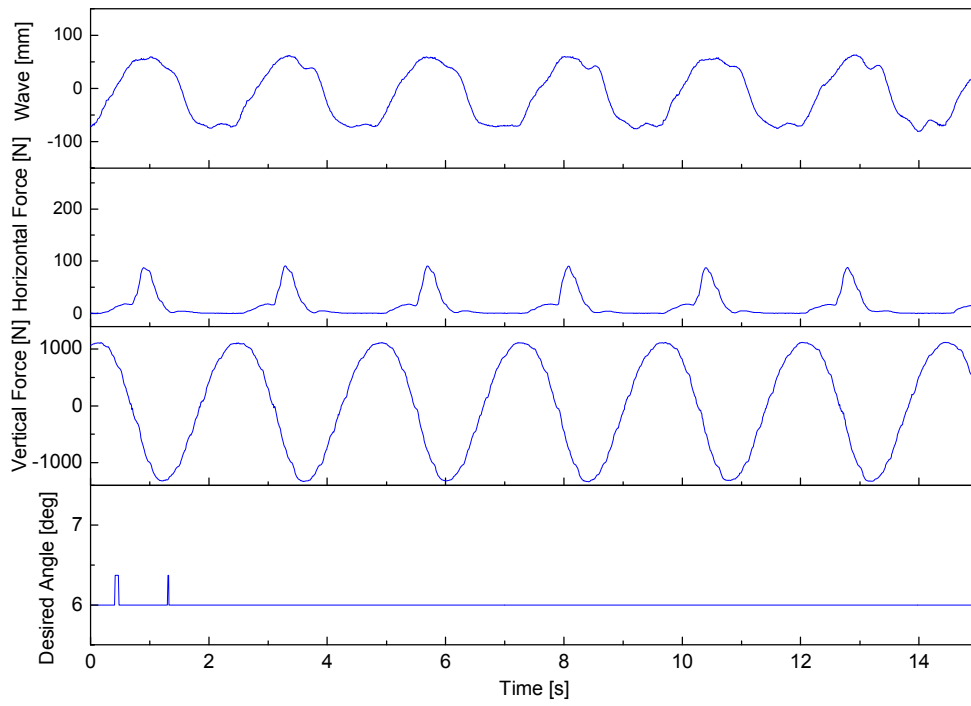
9	72.04	73.71	72.84	76.35	79.98	80.13	79.85
12	72.76	74.23	76.64	80.28	86.84	81.53	80.47
15	71.19	75.77	76.83	80.73	82.06	81.19	80.26
18	71.56	74.49	75.58	77.17	81.43	82.69	83.31
	Goodness of fit [%]						

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5. Supervisor control verification

In this section, the capability of the supervisor controller was evaluated using both single frequency waves and irregular waves. From the results obtained in the previous section, the controller was constructed using the LVQNN with the optimal structure.

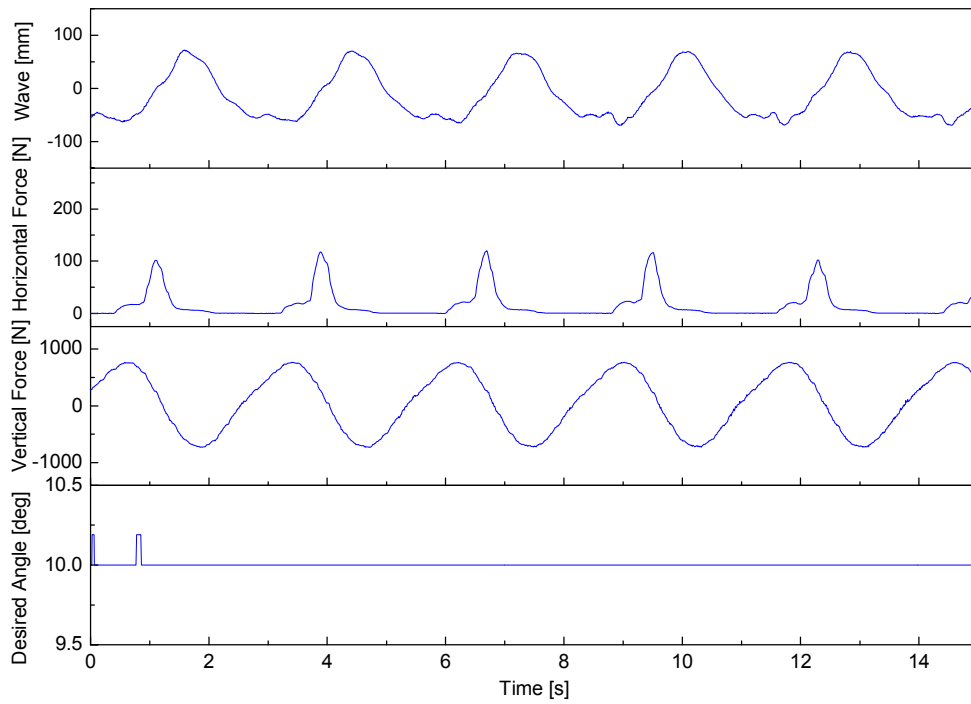
First, verification using single frequency waves by means of numerical simulations and real-time experiments is discussed. For the simulations, the three data sets (corresponding to the three cases in Table 3) were employed in turn. Then the classifier was operated to detect the wave conditions and the optimal tilting angles. Classification performance is shown in Figures 13 to 15 and compared with the target angles shown in Table 3. These results indicate that the classifier could detect the wave conditions and the output reach stably to the optimal angle in a short time; e.g., 6° output after ~1.3 s for wave case 1. In test case 2, the optimized controller detected the wave condition and the desired tilting angle within 1 s. In test case 3, the time required by the controller to make a decision on the wave condition was longer (~2.7 s), because the number of wave factors used as controller inputs was limited to 3, and these had similar forms among the various wave cases.



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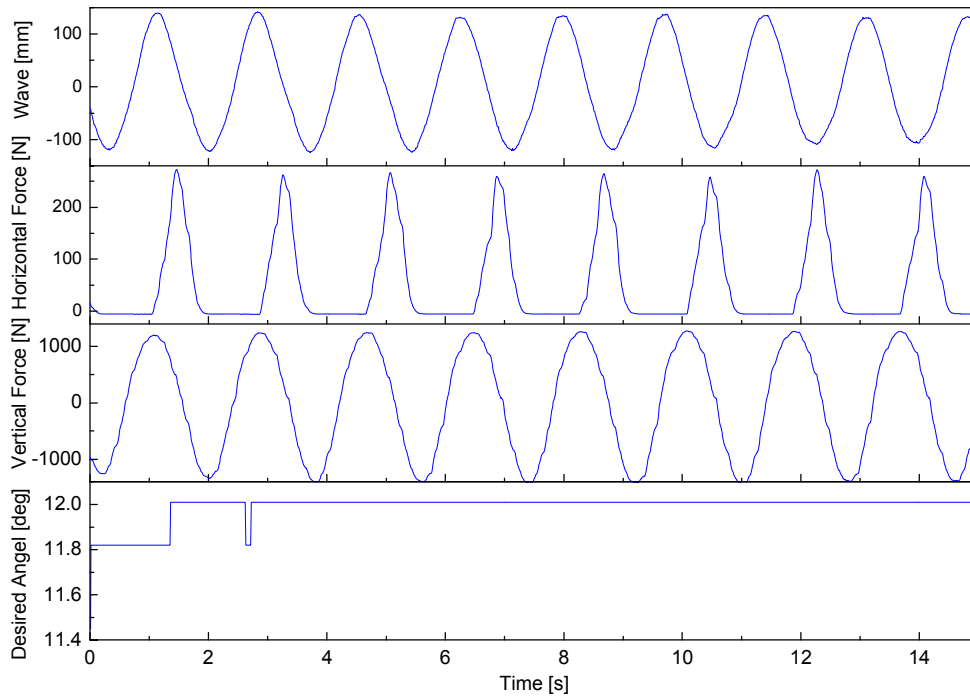
Fig. 13 Simulation verification of the optimized LVQNN with respect to test case 1



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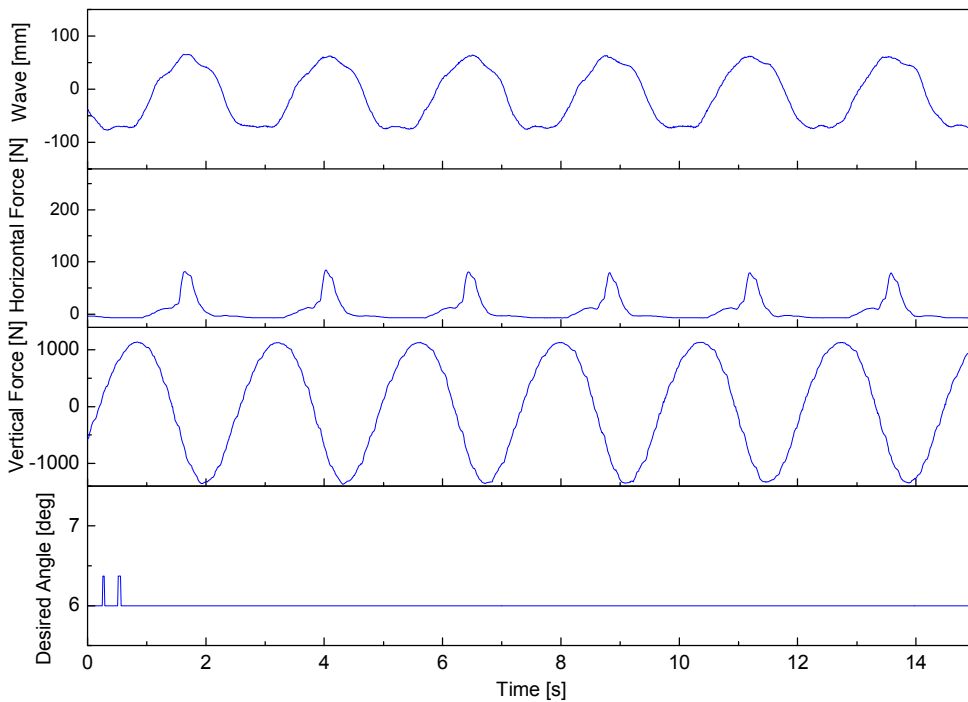
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Fig. 14 Simulation verification of the optimized LVQNN with respect to test case 2

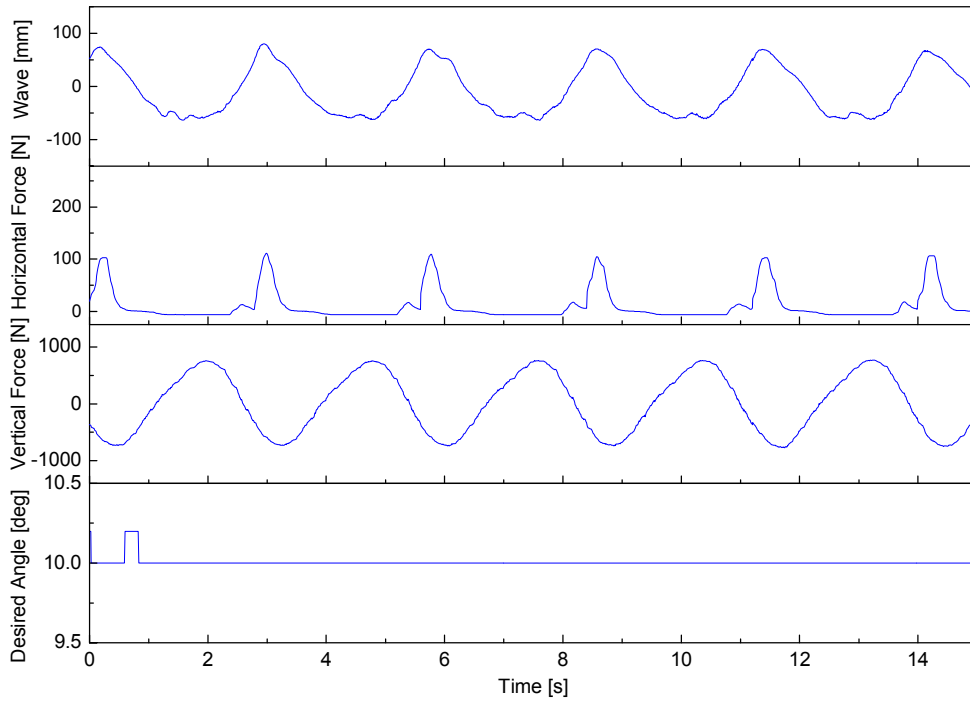


1
2 **Fig. 15** Simulation verification of the optimized LVQNN with respect to test case 3

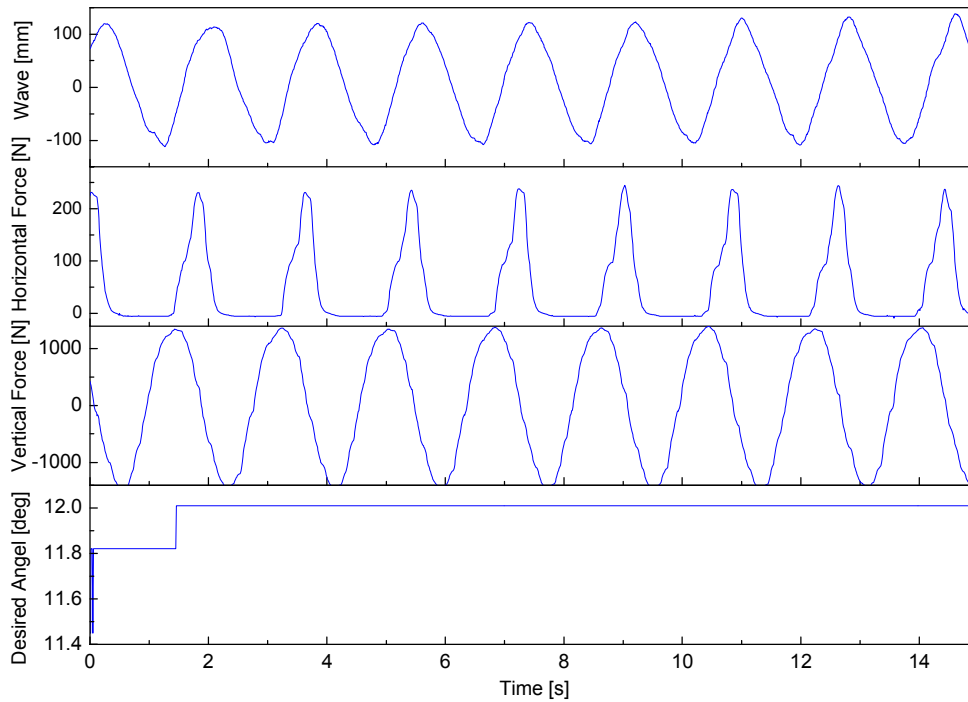
3
4 Next, the real-time ability of the supervisor controller was investigated. The system was
5 subjected to each of the three test cases in real-time. The wave classification and angle detection
6 results are shown in Figures 16 to 18. The real-time wave detection performance was similar to
7 the simulation performance. These results imply that the proposed classifier could detect the wave
8 conditions. Therefore, it made precise decisions of tilting angles to maximize the energy harvesting
9 capability of the device.



10
11 **Fig. 16** Experimental verification of the optimized control LVQNN with respect to test case 1



1
2 **Fig. 17** Experimental verification of the optimized control LVQNN with respect to test case 2

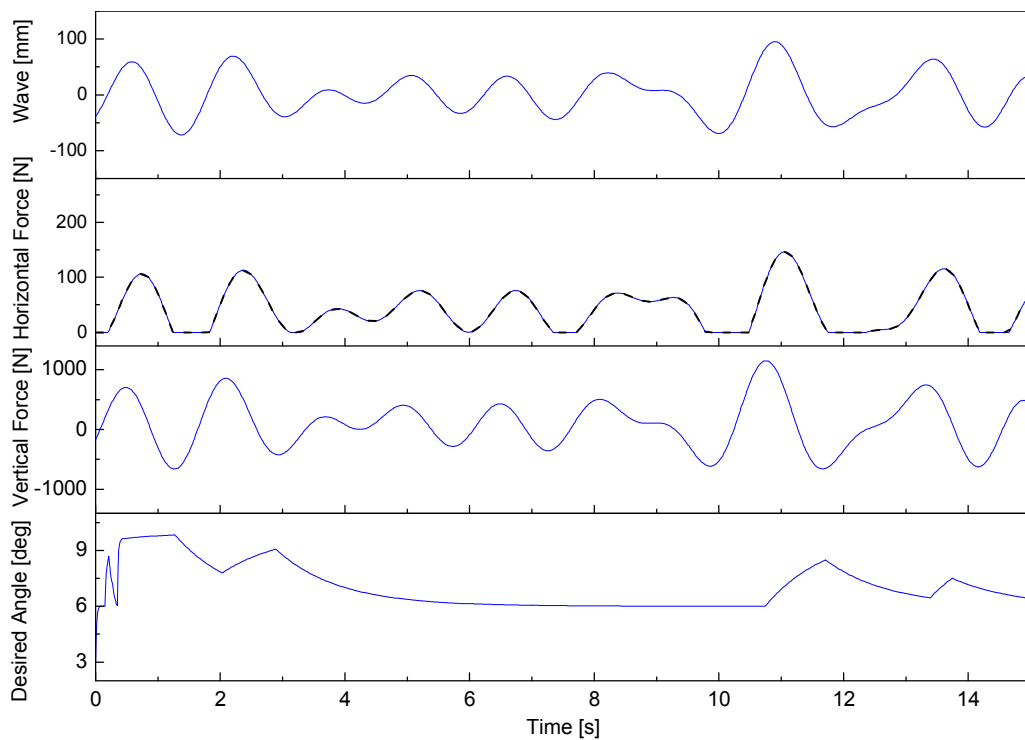


3
4 **Fig. 18** Experimental verification of the optimized LVQNN with respect to test case 3

5
6 Second, simulations were carried out for irregular waves to demonstrate the applicability of the
7 proposed controller for real waves. Herein, irregular waves were simulated using the Pierson-
8 Moskowitz spectrum formulation [2, 16]. To utilize the results in Section 4 in which the LVQNN
9 was optimized for the wave cases in Table 1, a peak period of 2 s ($T_p = 2s$), a significant wave
10 height of 270 mm ($H = 270$ mm), and a frequency range of 0.25-0.75 Hz were selected to generate
11 the Pierson-Moskowitz wave spectra. Thus, the irregular waves covered the three wave cases in

1 Table 1; therefore, the procedure to derive the supervisor controller (as mentioned in Section 4.2.2)
2 was employed. Simulations using the ASAWEC model developed in Section 2 were then
3 performed with irregular waves for two cases: using the supervisor controller to regulate the slope
4 angle, and with the slope fixed at an appropriate angle. The simulation with the supervisor
5 controller was performed first (Fig. 19). The system slope was smoothly varied according to the
6 wave variation. Subsequently, the simulated system efficiency was obtained (solid blue line in the
7 top sub-plot of Fig. 20). Using the mean value of the desired angle regulation in Fig. 19, the second
8 ASAWEC simulation was performed in which the slope angle was fixed at 7° . The simulated
9 system efficiency was achieved (dash-dot black line in the top sub-plot of Fig. 20). The difference
10 in efficiency due to use of the supervisor controller is depicted in the bottom sub-plot of Fig. 20.
11 A maximum improvement in system efficiency of 2.8% was recorded in the irregular wave
12 condition. These results demonstrate that system performance could be improved using the
13 proposed control methodology.

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Fig. 19 Simulation LVQNN classification result in irregular wave condition

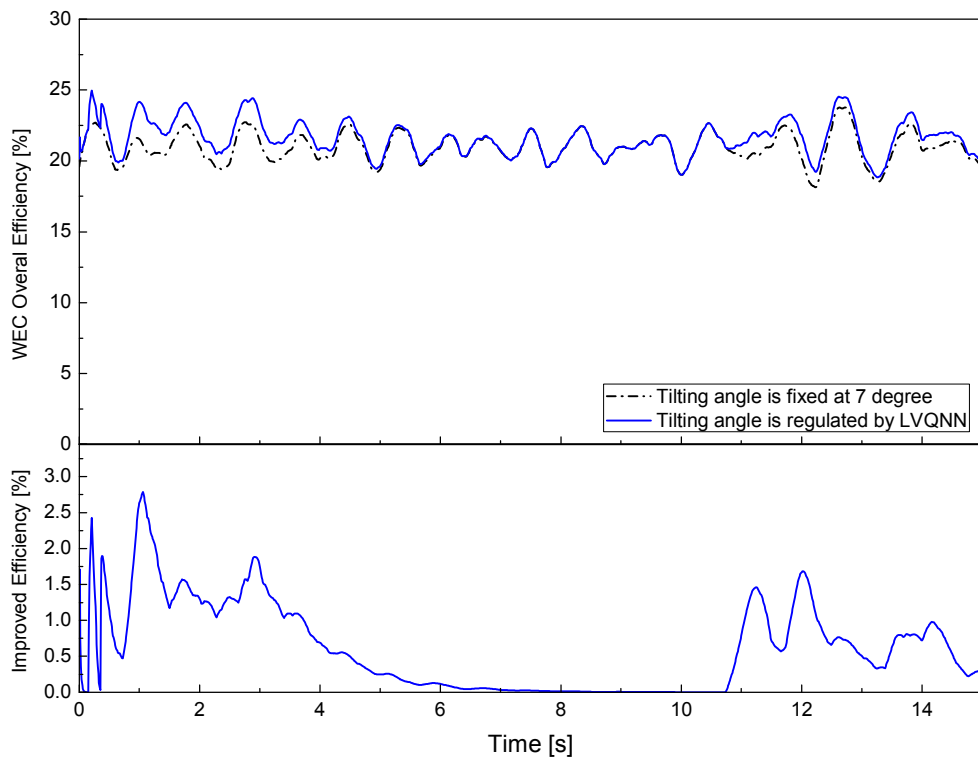


Fig. 20 Simulation system efficiency comparison in irregular wave condition

6. Conclusions

This paper presented a simple approach to maximizing the energy harvesting capability of the LVQNN algorithm based ASAWEC. The supervisor controller, optimized by the training data set, classified wave conditions and generated the desired tilting angle for the secondary controller. At the same time, the PI controller drove the EOM in order to set the ASAWEC at an optimal slope angle. The experimental results with regular waves indicated that by using the proposed control scheme, a decision on the optimal tilting angle could be made within 2s. At these positions, the absorbed energy and energy capture ratio were maximized. In wave case 3, the highest increase in overall efficiency of 5% was achieved compared to the case in which the sliding shaft was fixed in the vertical direction (without changing the tilting angle). In addition, the applicability of the proposed approach in practice was demonstrated by means of simulations of irregular waves.

Use of the LVQNN-based EOM control system does not require system governing equations or information on the transient influence of the waves on the system performance. Therefore, it can be applied to any floating-buoy WEC integrated with an EOM mechanism. Future research works including real-time tests of the controlled system in irregular waves are now being carried out in order to validate the applicability of the proposed method. Then, the suitability of this method for other WEC technologies will be evaluated.

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17 Press, UK, 2010.

1 **APPENDIX**

2 **Table A1**

3 Component selection for the experimented ASAWEC

Component	Characteristic	Value
Floating buoy	Diameter	$D_b = 1.2\text{m}$
	Height	$H_b = 1\text{m}$
	Mass	$m_b = 200\text{kg}$
Cylinder	Bore diameter	$D = 0.025\text{m}$
	Rod diameter	$d = 0.012\text{m}$
	Length of stroke	$L = 0.5\text{m}$
Hydraulic motor	Displacement	$D_m = 12.5\text{cc/rev}$
Accumulator	Volume	$V_{gas0} = 2.8\text{L}$
	Pre-charge pressure	$P_{gas0} = 5\text{bar}$
Relief valve	Cracking pressure	$P_{crack} = 35\text{bar}$
Hydraulic oil	Effective bulk modulus	$1.5 \times 10^9 \text{ [Pa]}$
Flow coefficient	Discharge coefficient	0.7
MR Rotary brake	Maximum current	1A
	Maximum torque	5.6Nm
	Maximum speed	1000rpm
Linear actuator	Input voltage	12/24 VDC
	Max load capacity	1600 lbs/ 7000N
	Stroke length	25mm - 600mm
	Speed	6mm/s - 42mm/s

4

5 **Table A2**

6 Water tank and wave maker specifications

Equipment	Specification	Value
Water tank	Length	50m
	Width	20m
	Depth	2m
Wave maker	Maximum wave amplitude	0.5m
	Shortest wave period	0.5s
	Cylinder number	10

7