

A Statistical-Mechanical Analysis of Coded CDMA with Regular LDPC Codes

Toshiyuki Tanaka¹
Dept. of Electro. & Info. Eng.,
Tokyo Metropolitan Univ.
1-1 Minami-Osawa, Hachioji,
Tokyo, 192-0397 Japan
e-mail: tanaka@eei.metro-u.ac.jp

David Saad¹
Neural Computing Res. Grp.,
Aston Univ.
Aston Triangle, Birmingham,
B4 7ET United Kingdom
e-mail: d.saad@aston.ac.uk

I. INTRODUCTION

Performance of LDPC-coded CDMA system is analyzed, using the replica method of statistical mechanics (which has been separately applied to the analysis of CDMA [1] and of LDPC codes [2]), in the infinite code length and large system (i.e., infinite number of users) limits. Most existing studies of such systems consider the case of a finite number of users. Caire et al. [3] consider the same limits as the current study, but theirs is based on stripping and successive detection/decoding. Our analysis, on the other hand, gives the information-theoretic and decoding thresholds of the optimum joint detection/decoding scheme.

II. ANALYSIS

We consider the serially concatenated system of a bank of per-user LDPC encoders and the standard fully-synchronous, randomly-spread, BPSK CDMA channel. We assume that the users are of equal rate, that the LDPC codes used are drawn from the same random ensemble of a regular (C, L) -LDPC code (C and L are the column and row weights of the parity-check matrix), that the power control is perfect, and that the channel noise is additive white Gaussian with variance σ_0^2 . The code length M and the number of users K are sent to infinity, while the information rate $R \equiv N/M$ and the system load $\beta \equiv K/G$ are both kept to be $O(1)$, where N and G are the information-block length and the spreading factor, respectively.

Analytical evaluation of the performance of the optimum joint detection/decoding scheme requires solving the saddle-point equations for $\{m, q, E, F, \pi(x), \hat{\pi}(\hat{x})\}$:

$$m = \int \tanh\left(\sqrt{F}z + E + \sum_{c=1}^C \tanh^{-1} \hat{x}_c\right) Dz \mathcal{D}_{\hat{\pi}}^C \hat{x}$$

$$q = \int \tanh^2\left(\sqrt{F}z + E + \sum_{c=1}^C \tanh^{-1} \hat{x}_c\right) Dz \mathcal{D}_{\hat{\pi}}^C \hat{x}$$

$$E = \frac{1}{\sigma^2 + \beta(1-q)}, \quad F = \frac{\sigma_0^2 + \beta(1-2m+q)}{[\sigma^2 + \beta(1-q)]^2}$$

$$\pi(x) = \int \delta\left[x - \tanh\left(\sqrt{F}z + E + \sum_{c=1}^{C-1} \tanh^{-1} \hat{x}_c\right)\right] Dz \mathcal{D}_{\hat{\pi}}^{C-1} \hat{x}$$

$$\hat{\pi}(\hat{x}) = \int \delta\left(\hat{x} - \prod_{i=1}^{L-1} x_i\right) \mathcal{D}_{\pi}^{L-1} x$$

where $Dz \equiv (2\pi)^{-1/2} e^{-z^2/2} dz$ and $\mathcal{D}_{\hat{\pi}}^C \hat{x} \equiv \prod_{c=1}^C [\hat{\pi}(\hat{x}_c) d\hat{x}_c]$ etc. The bit error rate P_b (which is the same for all users since

¹Support from EPSRC research grant GR/N00562 (TT, DS), and from Grant-in-aid for Scientific Research on Priority Areas 14084209, MEXT, Japan, and from Bilateral Program between the UK and Japan, JSPS, Japan, (TT) is acknowledged.

they are statistically equivalent) is given by $P_b = \int_{-\infty}^0 P(u) du$ where

$$P(u) = \int \delta\left[u - \tanh\left(\sqrt{F}z + E + \sum_{c=1}^C \tanh^{-1} \hat{x}_c\right)\right] Dz \mathcal{D}^C \hat{x}.$$

When the saddle-point equations have multiple solutions, the one minimizing the associated free energy (not shown) is information-theoretically relevant. The other solutions are also significant because they may determine the decoding threshold. Analytically found are the “error-free” solution ($m = q = 1$, $\pi(x) = \delta(x-1)$, $\hat{\pi}(\hat{x}) = \delta(\hat{x}-1)$), and, in the limit $L \rightarrow \infty$, the one corresponding to the uncoded system [1]. Any other solutions are obtained numerically.

III. RESULTS

Figure 1 (a) shows the thresholds (decoding and information-theoretic), in terms of the critical load β , versus the noise level σ^2 for the joint detection/decoding scheme with (3,6)- and (4,8)-LDPC codes. The maximum achievable spectral efficiency $\rho = \beta f$ (f being the free energy or mutual information per symbol and user) versus signal-to-noise ratio E_b/N_0 is shown in Fig. 1 (b), which shows that, although the spectral efficiency of the joint detection/decoding scheme practically converges to a finite value, it diverges theoretically, approaching that of the AWGN single-user spectral efficiency (shown as dotted line), as $\beta \rightarrow \infty$, suggesting possible improvement in the decoding threshold by adopting irregular LDPC codes.

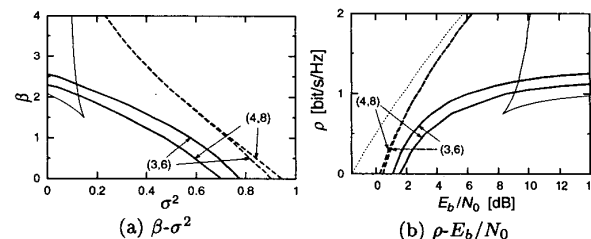


Fig. 1: Decoding (thick) and information-theoretic (dashed) thresholds for (3,6)- and (4,8)-LDPC coded CDMA systems.

REFERENCES

- [1] T. Tanaka, “A statistical-mechanics approach to large-system analysis of CDMA multiuser detectors,” *IEEE Trans. Inf. Theory*, vol. 48, pp. 2888–2910, 2002.
- [2] T. Murayama, Y. Kabashima, D. Saad, and R. Vicente, “Statistical physics of regular low-density parity-check error-correcting codes,” *Phys. Rev. E*, vol. 62, pp. 1577–1591, 2000.
- [3] G. Caire, S. Guemghar, A. Roumy, and S. Verdú, “Maximizing the spectral efficiency of LDPC-encoded CDMA,” submitted to *IEEE Trans. Inf. Theory*, 2002.