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On the Energy Efficiency of Hybrid Relaying Schemes in the Two-way Relay Channel

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Abstract: In this paper, hybrid relaying schemes are investigated in the two-way relay channel, where the relay node is able to adaptively switch between different forwarding schemes based on the current channel state and its decoding status and thus provides more flexibility as well as improved performance. The analysis is conducted from the energy efficiency perspective for two transmission protocols distinguished by whether exploiting the direct link between two main communicating nodes (the source and destination nodes, and vice versa since it is two way communication) or not. A realistic power model taking circuitry power consumption of all involved nodes into account is employed. The energy efficiency is optimized in terms of consumed energy per bit subject to the Quality of Service (QoS) constraint. Numerical results show that the hybrid schemes are able to achieve the highest energy efficiency due to its capability of adapting to the channel variations and the protocol where the direct link is exploited is more energy efficient.

Keywords: Energy efficiency, two-way relay channel, network coding, compress-and-forward, outage probability

I. Introduction

Global warming has become a more and more serious challenge across the whole world in the past century mostly caused by human activities that increase concentrations of greenhouse gases in the atmosphere. Some scientific findings reveal that the CO₂ emission of the ICT (Information & Communications Technology) industry has contributed to a considerable percentage to the world energy consumption budget and global warming [1, 2]. The arising challenge of the huge increase of energy consumption in the face of the exploding traffic demands of future communication systems motivates the need for greener and more energy efficient techniques. Among all the candidates, relaying technique has become increasingly important, motivated by its capability to provide uniform high coverage and potential energy savings.

The relay model is firstly introduced by van der Meulen [3] and substantially developed by Cover and El Gamal in their landmark work [4]. While the optimal relaying strategy in wireless networks has not yet been fully understood, several relaying schemes have been suggested in the literature including amplify-and-forward (AF), decode-andforward (DF) and compress-and-forward (CF) [5, 6]. More recently, advanced hybrid relaying schemes have been proposed and analyzed from both spectrum and energy efficiency perspectives [7]. However, these works are focused on one way relay where the communication is in one direction only, i.e. from the source to the destination. The communication systems, in nature, are two-way systems where two terminals transmit to each other simultaneously, i.e. a source can also be a destination. In this regard, a twoway relay channel (TWRC), where two nodes communicate with each other via a relay node, has drawn more and more attention recently. The well developed relaying strategies have been applied in the TWRC in [5, 6] for both full-duplex and half-duplex systems. A novel coding scheme called network coding (NC), where instead of simply relaying the received packets of information, the relay node take two packets from two main nodes and combine then together for transmission, has been proposed and investigated in [8, 9].

The existing analysis on the TWRC is mainly aiming at spectrum efficiency analysis, such as achievable rate region [10] and capacity bounds [8]. On the contrary, this work exploits the energy efficiency perspective with a realistic power consumption model to optimize the energy usage of the entire system. In addition to transmission power, circuitry power consumption of all involved nodes is taken into consideration to give insight of how to design a practical energy efficient system. Moreover, in order to further exploit the degrees of freedom of a time-variant wireless media, conventional forwarding mechanisms and network coding schemes are combined in a hybrid fashion to better adapt to the channel variations. Specifically, CF is used in conjunction with NC in this work. We focus on more feasible half-duplex mode where simultaneously transmission and reception are forbidden at each node. Two transmission protocols are studied. The direct link between two main nodes is exploited in the first protocol and the entire frame is divided into three phases. On the other hand, the second protocol does not use the direct link and therefore each frame only consists of two phases (the detailed frame structure will be described later in section II).

The rest of the paper is organized as follows: in the next section, the system model is introduced and problem formulation is presented. Section III presents the derivation of the outage probability for different relaying strategies. The energy consumption model and energy efficiency analysis is studied in section IV where an energy efficiency objective function is also established. Numerical results are shown in section V and the final section concludes.

II. System Model

As shown in figure 1, two main nodes 1 and 2, denoted as S_1 and S_2 , respectively, exchange information through the relay node R. When the direct link S_1 - S_2 is exploited, the entire transmission can be separated into three phases as shown in figure 1(a): during the first phase, S_1 broadcasts and S_2 and R receive. Since it is half-duplex mode, S_2 cannot transmit simultaneously in phase 1. S_2 broadcasts and S_1 and S_2 receive in the second phase. In the final phase, S_2 broadcasts to S_1 and S_2 . Each phase lasts for S_2 0 records, where S_3 1 is the activation time and subject to the constraint

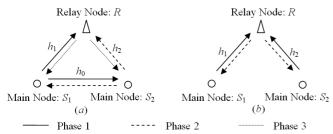


Fig.1 Two-way relay channel: (a) three-phase case (b) two-phase case. $T \le T_{max}$. The signals received at S_2 and R during phase 1 are given by

$$y_2^{(1)}(t) = h_0 x_1(t) + n_2^{(1)}(t), \quad y_r^{(1)}(t) = h_1 x_1(t) + n_r^{(1)}(t),$$

with
$$h_i = \frac{c_i}{\eta d_i^{\xi/2}}$$
, for $i = 0, 1, 2$, where the superscripts in

angle brackets indicate phases (only used when necessary), x_1 is the transmitted signal from S_1 , n_2 and n_r are the noise at S_1 and R, respectively, following circularly symmetric complex Gaussian (CSCG) distribution with zero mean, c_i represents fast fading of link i and follows CSCG distribution with unit variance and zero mean, d_i is the distance of link i, ξ is the path loss exponent, and η is a constant indicating the physical characteristics of the link and the power amplifier as in [7]. Similarly, the signal received during phase 2 are given by

$$y_1^{\langle 2 \rangle}(t) = h_0 x_2(t) + n_1^{\langle 2 \rangle}(t), \ y_r^{\langle 2 \rangle}(t) = h_2 x_2(t) + n_r^{\langle 2 \rangle}(t),$$
 at S_2 and R , respectively, where x_2 is the S_2 signal, and n_1 is the noise at S_1 , following the same distribution as n_r . The received two signals from two main nodes are processed at

the noise at S_1 , following the same distribution as n_r . The received two signals from two main nodes are processed at the relay node and a new signal is generated and broadcasted to S_1 and S_2 by R in the final phase. The received signals are

$$y_1^{\langle 3 \rangle}(t) = h_1 x_r(t) + n_1^{\langle 3 \rangle}(t), y_2^{\langle 3 \rangle}(t) = h_2 x_r(t) + n_2^{\langle 3 \rangle}(t),$$
 respectively.

On the contrary, when the direct link S_1 - S_2 is neglected, the entire transmission can be separated into two phases: R receiving phase and R broadcasting phase as shown in figure 1(b). Each phase lasts for T/2 seconds. The signal received at R during phase 1 is given by

$$y_r(t) = h_1 x_1(t) + h_2 x_2(t) + n_r(t),$$

During phase 2, R broadcasts and the received signal at S_1 and S_2 are

$$y_1(t) = h_1 x_r(t) + n_1(t), y_2(t) = h_2 x_r(t) + n_2(t),$$

respectively.

The objective of the optimization is to minimize the energy expenditure for each transmitted information bit. However, this optimization should be subject to some Quality of Service (QoS) constraints to guarantee that the user experience is maintained at a satisfactory level. As we know, outage probability indicates the possibility of unsuccessful reception and can be used as a QoS constraint in this work. We define the consumed transmission energy for each bit as E_b . Clearly, outage probability should a function of E_b and the activation time T, and thus can be expressed as $p(E_b,T)$. Considering there are three nodes in the TWRC, each nodes might have different level of energy expenditure and the outage probability should be rewritten as $p(E_{b,1}, E_{b,2}, P_r, T)$, where P_r is the transmission power of node R. Let's define the overall energy consumption of the entire system including circuitry power expenditure for each bit as E. It is straightforward to see that E should be a function of $E_{b,1}$, $E_{b,2}$ and P_r as well. It is also a function of activation time T because circuitry energy consumption is depending on it. Thus it takes the form as $E(E_{b,1}, E_{b,2}, P_r, T)$. Now the problem is to minimize the overall average energy

consumption per bit E given the target outage probability p_t and maximum activation time T_{max} . It can be mathematically formulated as

minimize
$$E\left(E_{b,1}, E_{b,2}, P_r, T\right)$$

subject to $p\left(E_{b,1}, E_{b,2}, P_r, T\right) \le p_t$. (1)
 $T \le T_{\text{max}}$

The main contribution presented in this work is to solve this energy efficiency optimization problem in the TWRC for hybrid relaying schemes.

III. Outage Analysis

As we mentioned, outage probability $p(E_{b,1}, E_{b,2}, P_r, T)$ will be used as the QoS constraint when minimizing the energy consumption. In this section, we will investigate the outage performance of different relaying schemes.

We assume that the message from node S_1 to S_2 is m_1 belonging to a set $\Omega_1 = \{1, \dots, 2^{L_1}\}$, and message m_2 in the reverse direction belonging to a set $\Omega_2 = \{1, \dots, 2^{L_2}\}$. The transmission power of node S_1 and S_2 are P_1 and P_2 , respectively. Here for simplicity, we assume that $P_1 = P_2 = P_s$. In this regard, we have $L_1 E_{b,1} / T = L_2 E_{b,2} / T$. In such a case, $E_{b,1}$ can be eliminated from the overall energy consumption E and outage probability functions.

Network Coding (NC): In three-phase NC, the relay node firstly tries to decode two messages received in two separate phases before re-encoding them with network coding technique. During the first phase, the message broadcasted by S_1 can be decoded successfully at node R if

$$R_{NC,1\to R}^3 = \frac{1}{3} C \left(\frac{3\gamma_1 E_{b,1} L_1}{N_0 BT} \right) \ge \frac{L_1}{T},$$

where $C(x)=B\log_2(1+x)$ and the superscript of $R_{NC,1\to R}^3$ indicates that it is three-phase operation, $\gamma_i=|h_i|^2$, N_0 is the single-sided thermal noise spectral density and B (in Hertz) stands for the system bandwidth. Similarly, message m_2 can be decoded as long as

$$R_{NC,2\to R}^3 = \frac{1}{3}C\left(\frac{3\gamma_2 E_{b,2} L_2}{N_0 BT}\right) \ge \frac{L_2}{T}$$
.

Multiplexed coding [8] is then used for NC at node R. Without loss of generality, we assume that $L_1 \le L_2$. We first set up a mapping from Ω_2 to 2^{L_2} L_2 -dimentional vectors in the finite field of size $2 \, {\rm F_2}^{L_2}$. Then we associate each vector with an independent Gaussian codeword. We do the same mapping and codeword generation and association for Ω_1 . The only difference is that the mapping is to a sub-space of ${\rm F_2}^{L_2}$, denoted as ${\rm F}_{sub}$. The relay node transmits the codeword associated with m_1+m_2 , where the addition is performed using arithmetics in field ${\rm F_2}^{L_2}$. During phase 3, R broadcasts a network coded Gaussian codeword $x_r(m_1+m_2)$ to S_1 and S_2 . Upon reception of $x_r(m_1+m_2)$, S_1 tries to decode m_1+m_2 by jointly processing $y_1^{(3)}$ and its reception $y_1^{(2)}$ during the second phase. It can do so if

$$R_{NC,2\to 1}^{3} = \frac{1}{3}C\left(\frac{3\gamma_{0}E_{b,2}L_{2}}{N_{0}BT}\right) + \frac{1}{3}C\left(\frac{\gamma_{1}P_{r}}{N_{0}B}\right) \ge \frac{L_{2}}{T} .$$

Then it subtracts the vector corresponding to m_1 from the decoded message to get m_2 . In node S_2 , m_2 is known as well as the associated affine space m_2 + F_{sub} . It only needs to find the codeword in this affine space. By taking $y_2^{(1)}$ into account, it can successfully decode when

$$R_{NC,1\to 2}^3 = \frac{1}{3}C\left(\frac{3\gamma_0 E_{b,1} L_1}{N_0 BT}\right) + \frac{1}{3}C\left(\frac{\gamma_2 P_r}{N_0 B}\right) \ge \frac{L_1}{T}.$$

Then m_2 is subtracted to get m_1 . Since there are two messages exchanged between S_1 and S_2 , unsuccessful receptions can happen at either node. In this work, if either S_1 or S_2 detects errors, outage is claimed. In order to derive the outage probability, we can first look at the successful decoding probability which is complementary to outage probability. As we can see, correct reception requires both the relay node and two main nodes to successfully decode the message and outage probability can be expressed as

$$p(E_{b,1}, P_r, T) = 1 - \Pr\{\text{successful decoding}\}$$

$$= 1 - \Pr \left\{ R_{NC,1 \to R}^{3} \ge \frac{L_{1}}{T}, R_{NC,2 \to R}^{3} \ge \frac{L_{2}}{T}, \right\} . (2)$$

$$R_{NC,2 \to 1}^{3} \ge \frac{L_{2}}{T}, R_{NC,1 \to 2}^{3} \ge \frac{L_{1}}{T}$$

The outage probability of two-phase NC can be derived following the same line of argument and the derivation is given in Appendix A.

Hybrid NC/CF: In NC, the relay node needs to be able to decode two received messages before employing multiplexed coding. However, due to the time variation characteristics of the wireless media successful decoding might not be achieved when deep fading occurs. Here we introduce more flexibility to the relay's forwarding scheme by allowing CF operation when decoding fails at node R. For the three-phase scenario, we need to consider four kinds of decoding status as follows:

a) Message m_1 is decoded but m_2 is not. The relay node re-encodes m_1 and quantizes its phase 2 reception $y_r^{\langle 2 \rangle}$ into intermediate bin index u, which is then compressed to index v [11]. The minimum number of compressed information bits for v, such that the compressed signal can be recovered with the smallest distortion, is denoted as s and is given by

$$s = \frac{1}{3}TB\left(H\left(W \mid y_1^{(2)}\right) - H\left(W \mid y_r^{(2)}\right)\right)$$

$$= \frac{1}{3}TB\log_2\left(\frac{\gamma_2 P_s + N_0 B + \sigma_W^2 - \frac{\gamma_0 \gamma_2 P_s^2}{\gamma_0 P_s + N_0 B}}{\sigma_W^2}\right), (3)$$

where W is an auxiliary random variable with variance σ_W^2 . Finally, this index v is encoded as $x_c(v)$ and added together with the re-encoded message $x_1(m_1)$ to form the relay node's transmitted signal as

$$x_r = \beta_a x_c + \beta_b x_1.$$

with power constraint $\beta_a^2 P_c + \beta_b^2 P_s \le P_r$. At node S_1 , the known part $\beta_b h_1 x_1$ is cancelled from its reception $y_1^{\langle 3 \rangle}$ and v can be decoded if

$$R_{Hi,R\to 1}^3 = \frac{1}{3} C \left(\frac{\beta_a^2 \gamma_1 P_c}{N_0 B} \right) \ge \frac{s}{T}$$
 (4)

Once v is obtained, with the help of side information $y_1^{\langle 2 \rangle}$, the estimation of the relay node's reception can be reconstructed and denoted as $\hat{y}_r^{\langle 2 \rangle}$. With joint processing of $y_1^{\langle 2 \rangle}$ and $\hat{y}_r^{\langle 2 \rangle}$, node S_1 can decode message m_2 as long as [6]

$$R_{Hi,2\to 1}^3 = \frac{1}{3} C \left(\frac{3\gamma_0 E_{b,2} L_2}{N_0 B T} + \frac{3\gamma_2 E_{b,2} L_2}{\left(N_0 B + \sigma_W^2 \right) T} \right) \ge \frac{L_2}{T} .$$

where σ_W^2 can be obtained from (3) and (4). At node S_2 , $\beta_a h_2 x_c$ is treated as noise. Based on joint processing of $y_2^{\langle 1 \rangle}$ and $y_2^{\langle 3 \rangle}$, m_1 can be successfully decoded if

$$R_{Hi,1\to 2}^{3} = \frac{1}{3} C \left(\frac{3\gamma_{0} E_{b,1} L_{1}}{N_{0} B T} \right) + \frac{1}{3} C \left(\frac{3\gamma_{2} \beta_{b}^{2} E_{b,1} L_{1}}{\left(\beta_{a}^{2} \gamma_{2} P_{c} + N_{0} B \right) T} \right) \ge \frac{L_{1}}{T}.$$

The successful decoding probability of both messages is given as

$$p_{a}(E_{b,1}, P_{r}, T) = \Pr \begin{cases} R_{NC, 1 \to R}^{3} \ge \frac{L_{1}}{T}, R_{NC, 2 \to R}^{3} < \frac{L_{2}}{T}, \\ R_{Hi, 2 \to 1}^{3} \ge \frac{L_{2}}{T}, R_{Hi, 1 \to 2}^{3} \ge \frac{L_{1}}{T} \end{cases} . (5)$$

- b) Message m_2 is decoded but m_1 is not. The decoding probability p_b can be derived following the same line of argument.
- c) Neither message m_1 nor m_2 is decoded. The relay node simply compresses the summation of $y_r^{\langle 1 \rangle}$ and $y_r^{\langle 2 \rangle}$. Similar compression/decompression procedure as described in case (a) is employed at both main nodes. Message m_1 and m_2 can be decoded if

$$\begin{split} R_{CF,2\to 1}^3 &= \frac{1}{3} \, C \left(\frac{3 \gamma_0 E_{b,2} L_2}{N_0 B T} + \frac{3 \gamma_2 E_{b,2} L_2}{\left(2 N_0 B + \sigma_W^2 \right) T} \right) \geq \frac{L_2}{T} \,, \\ R_{CF,1\to 2}^3 &= \frac{1}{3} \, C \left(\frac{3 \gamma_0 E_{b,1} L_1}{N_0 B T} + \frac{3 \gamma_1 E_{b,1} L_1}{\left(2 N_0 B + \sigma_W^2 \right) T} \right) \geq \frac{L_1}{T} \,. \end{split}$$

where the compression noise can be obtained by the constraint [10]

$$\max \left\{ H\left(W \mid y_{2}^{\langle 1 \rangle}\right) - H\left(W \mid y_{r}^{\langle 1 \rangle}\right), H\left(W \mid y_{1}^{\langle 2 \rangle}\right) - H\left(W \mid y_{r}^{\langle 2 \rangle}\right) \right\}$$

$$\leq \frac{1}{TB} \min \left\{ C\left(\frac{\gamma_{1} P_{r}}{N_{0} B}\right), C\left(\frac{\gamma_{2} P_{r}}{N_{0} B}\right) \right\}$$

It follows that the decoding probability is given as

$$p_{c}\left(E_{b,1}, P_{r}, T\right) = \Pr \left\{ \begin{cases} R_{NC, 1 \to R}^{3} < \frac{L_{1}}{T}, R_{NC, 2 \to R}^{3} < \frac{L_{2}}{T}, \\ R_{CF, 2 \to 1}^{3} \ge \frac{L_{2}}{T}, R_{CF, 1 \to 2}^{3} \ge \frac{L_{1}}{T} \end{cases} \right\}. (6)$$

d) Both messages are decoded. The decoding probability p_d is given as

$$p_{d}\left(E_{b,1}, P_{r}, T\right) = \Pr \begin{cases} R_{NC, 1 \to R}^{3} \ge \frac{L_{1}}{T}, R_{NC, 2 \to R}^{3} \ge \frac{L_{2}}{T}, \\ R_{NC, 2 \to 1}^{3} \ge \frac{L_{2}}{T}, R_{NC, 1 \to 2}^{3} \ge \frac{L_{1}}{T} \end{cases} . (7)$$

Since four events are mutually exclusive, it follows that the overall outage probability is $p(E_{b,1},P_r,T)=1-(p_a+p_b+p_c+p_d)$. The outage probability of two-phase case is derived in Appendix B.

IV. Energy Efficiency Analysis

The objective of the energy efficiency optimization is to minimize the overall energy expenditure including circuitry power consumption per transmitted information bit. In the three-phase case, during the first phase node S_1 broadcasts and nodes S_2 and R receive. The consumed energy for all three nodes is given as

$$E^{\langle 1 \rangle} = L_1 E_{b,1} + (P_{ct} + 2P_{cr}) T / 3,$$

where P_{ct} and P_{cr} are the transmission and reception circuitry power, respectively. The first term represents

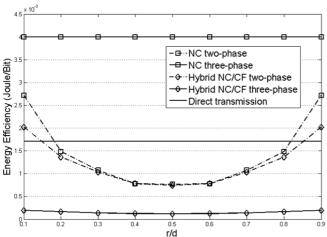


Fig. 2 Energy Efficiency, d=100m.

transmission power of S_1 and the second term stands for the circuitry power consumption of one transmitting node (S_1) and two receiving node (S_2) and (S_2) broadcasts to (S_1) and (S_2) and the associated energy consumption is expressed as

$$E^{\langle 2 \rangle} = L_2 E_{b,2} + (P_{ct} + 2P_{cr})T/3,$$

where the first term is the S_2 transmission power and the last term is circuitry power consumption. In the final phase, node R broadcasts and power consumption is given by

$$E^{\langle 3 \rangle} = P_r T / 3 + (P_{ct} + 2P_{cr}) T / 3,$$

where the first term is the relay node transmission power consumption. The overall energy consumed for transmitting message m_1 and m_2 is

$$E(E_{b,1}, E_{b,2}, T, P_r) = E^{\langle 1 \rangle} + E^{\langle 2 \rangle} + E^{\langle 3 \rangle}.$$

In order to optimize the energy efficiency, we set the objective function as

minimize
$$E\left(E_{b,1},T,P_r\right) = \frac{E\left(E_{b,1},T,P_r\right)}{L_1 + L_2}$$
 subject to:
$$p\left(E_{b,1},P_r,T\right) \le p_t \qquad . \tag{8}$$

$$T \le T_{\text{max}}$$

From the outage probability expressions, it is easy to see that with fixed L_1 and L_2 , the outage probability is a function of $E_{b,1}$, relay transmission power P_r , and activation time T. Once we fix the outage probability at target one, $E_{b,1}$ can be expressed as a function of P_r and T, which makes $E(E_{b,1}(P_r, T), P_r, T)$ a function of P_r and T as well. It means that the minimum energy per bit can be optimized by manipulating P_r and T. This problem can be solved numerically.

In the two-phase case, the energy consumption for each phase can be easily derived and given as

$$\begin{split} E^{\langle 1 \rangle} &= L_1 E_{b,1} + L_2 E_{b,2} + \left(2 P_{ct} + P_{cr} \right) T / 2, \\ E^{\langle 2 \rangle} &= P_r T / 2 + \left(2 P_{cr} + P_{ct} \right) T / 2, \end{split}$$

respectively. The optimization problem can be formulated in the same manner as in (8).

V. Numerical Results and Discussions

In this section, we compare the energy efficiency performance of different schemes. We use the following settings: L_1 = L_2 =10, P_{ct} =98mW, P_{cr} =112.4mW, η =7.78×10⁴, ζ =3, N_0 =-171dBm/Hz, B=1, ρ =0.5 and T_{max} =10. The distance between S_1 and S_2 is d and the distance between S_1 and R is expressed by r. Three nodes are assumed to be in a straight line with R in the middle and moving towards S_2 from S_1 .

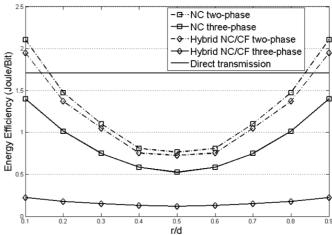
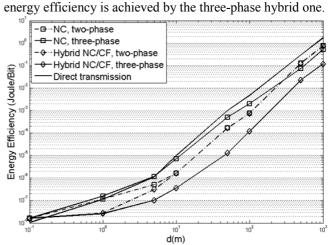


Figure 2 shows the energy efficiency performance in

terms of consumed energy per bit in short transmission range (d=100m). We see that hybrid schemes demonstrate better energy efficiency performance compared to NC and direct transmission due to its flexibility (except the extremely asymmetric case in the two-phase scenario, where R is very close to either S_1 or S_2). For three-phase schemes, their consumed energy per bit is almost irrelevant with the position of node R. This is caused by the fact that two nodes mainly communicate through the direct link since their distance is small and the link quality is naturally high. Using a relaying node usually helps to improve the outage behaviour. With a fixed outage target, it means lower transmission power consumption. However, when circuitry power consumption is taken into account, especially in short distance, where it accounts for a major part of overall energy expenditure, the additional energy consumption of the relay node might overweight the benefit of improved outage behaviour. This is why direct transmission outperforms twophase NC in this relatively short distance. Another interesting observation is that the best energy efficiency is achieved by the three-phase hybrid scheme. When the transmission range is increased to 1000m, as shown in figure 3, most of the relaying strategies totally or partly outperform direct transmission because the transmission energy is the main part of the total energy consumption and adding a relay node can significantly reduce the transmission power. Hybrid schemes still out perform NC and achieve much high



energy efficiency compared to direct transmission. The best

Fig. 4 Energy efficiency when d is increasing.

Figure 4 shows the energy efficiency when the distance between two main nodes increases from 0.1 to 1000m. Node *R* is in the middle of two nodes. Direct transmission achieves the best performance in very short distance; while

three-phase hybrid NC/CF scheme has the lowest energy expenditure when the distance is increased.

VI. Conclusions and Future Work

In this paper, we investigate, with a practical power model, the energy efficiency performance of hybrid relaying strategies switching between NC and CF in the two-way relay channel. Two system settings are assumed: with or without exploiting the direct link and the frame structure is divided into two and three phases, respectively. The energy efficiency in terms of energy per bit is minimized by manipulating the transmission power and the activation time of all involved nodes. The results show that the hybrid schemes which enjoy more flexibility and are able to adapt to the channel variation outperform non-hybrid ones. However, this improvement is only achieved when the distance between two main nodes is not very close and the circuitry power consumption accounts for only a small part of the overall energy expenditure. An interesting observation is that, compared to the two-phase hybrid relaying scheme, the three-phase hybrid one achieves a better energy efficiency performance by exploiting the direct link between two main nodes.

In the current settings, we only consider balanced scenario where the transmission powers of S_1 and S_2 are the same, non-balanced cases will be addressed in the future. In addition, the application of multiple relay nodes is also an interesting topic to study in our future work.

Appendix A

For two-phase NC, the first phase is a MAC channel and two messages can be decode successfully at node R if [12]

$$\begin{split} R_{NC,1\to R}^2 &= \frac{1}{2} C \Bigg(\frac{2\gamma_1 E_{b,1} L_1}{N_0 B T} \Bigg) \geq \frac{L_1}{T}, R_{NC,2\to R}^2 = \frac{1}{2} C \Bigg(\frac{2\gamma_2 E_{b,2} L_2}{N_0 B T} \Bigg) \geq \frac{L_2}{T}, \\ R_{NC,12\to R}^2 &= \frac{1}{2} C \Bigg(\frac{2\gamma_1 E_{b,1} L_1 + 2\gamma_2 E_{b,2} L_2}{N_0 B T} \Bigg) \geq \frac{L_1 + L_1}{T}. \end{split}$$

Multiplexed coding can also be used here and the same decoding procedure is employed in both main nodes. The

$$R_{NC,R\to 1}^2 = \frac{1}{2}C\left(\frac{2\gamma_1 P_r}{N_0 B}\right) \ge \frac{L_2}{T}, R_{NC,R\to 2}^2 = \frac{1}{2}C\left(\frac{2\gamma_2 P_r}{N_0 B}\right) \ge \frac{L_1}{T}.$$

The outage probability can be given in a similar formation as in (2).

Appendix B

In two-phase hybrid scheme, the fact that only message m_1 is decoded during the first phase means [12]

$$R_{NC,2\to R}^2 < \frac{L_2}{T}, \frac{1}{2}C\left(\frac{2\gamma_1 E_{b,1} L_1}{\left(N_0 B + P_s\right)T}\right) \ge \frac{L_1}{T}.$$

The relay subtracts h_1x_1 from the received signal, reencodes m_1 and compresses the residual signal $\overline{y}_r(t) = h_2 x_2(t) + n_r(t)$. Then it broadcasts the weighted summation as $x_r = \beta_a x_c + \beta_b x_1$. At node S_1 , the known part $\beta_b h_1 x_1$ is cancelled and tries to reconstruct estimation of \overline{y}_r . If can do so if [13]

$$I\left(\overline{y}_r; \hat{\overline{y}}_r\right) \le \frac{1}{2BT} C\left(\frac{\beta_a^2 \gamma_1 P_c}{N_0 B}\right).$$

It follows that m_2 can be decoded if

$$R_{Hi,2\to 1}^2 = \frac{1}{2}C\left(\frac{2\gamma_2 E_{b,2} L_2}{\left(N_0 B + \sigma_W^2\right)T}\right) \ge \frac{L_2}{T}$$

At node S_2 , $\beta_a h_2 x_c$ is treated as noise. The message m_1 can be decoded if

$$R_{Hi,1\to 2}^2 = \frac{1}{2} C \left(\frac{2 \gamma_2 \beta_b^2 E_{b,1} L_1}{\left(N_0 B + \beta_a^2 \gamma_2 P_c \right) T} \right) \ge \frac{L_1}{T} .$$

The successful decoding probability is given as

$$p_{a}\left(E_{b,1}, P_{r}, T\right) = \Pr \left\{ \begin{cases} R_{NC, 2 \to R}^{2} < \frac{L_{2}}{T}, \frac{1}{2}C\left(\frac{2\gamma_{1}E_{b,1}L_{1}}{\left(N_{0}B + P_{s}\right)T}\right) \geq \frac{L_{1}}{T}, \\ R_{HI, 2 \to 1}^{2} \geq \frac{L_{2}}{T}, R_{HI, 1 \to 2}^{2} \geq \frac{L_{1}}{T} \end{cases} \right\}$$

Probability p_b and p_c can be derived in a similar way and p_d is given by

$$p_{d}\left(E_{b,1},P_{r},T\right) = \Pr \left\{ \begin{aligned} R_{NC,1 \to R}^{2} &\geq \frac{L_{1}}{T}, R_{NC,2 \to R}^{2} \geq \frac{L_{2}}{T}, R_{NC,12 \to R}^{2} \geq \frac{L_{1} + L_{1}}{T}, \\ R_{NC,2 \to 1}^{2} &\geq \frac{L_{2}}{T}, R_{NC,1 \to 2}^{2} \geq \frac{L_{1}}{T} \end{aligned} \right\}.$$

 $(p_a+p_b+p_c+p_d)$.

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