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## TORSIONAL OSCILLATIONS IN A GEARED

SYSTEM WITH CLEARANCES

A Dissertation

Submitted to the Faculty of Graduate Studies Through the Department of Mechanical Engineering in Partial Fulfillment of the Requirements for the Degree of Doctor of Philosophy at the University of Windsor

bу

Subhash C. Jasuja

Windsor, Ontario

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#### ABSTRACT

The torsional vibration characteristics of a geared system with clearance are investigated in this study. The clearance results in a bilinear restoring force characteristic which is non-linear in nature. This particular type of non-linearity causes the generation of ultraharmonic, harmonic and subharmonic resonances.

Analytical solutions are derived for ultraharmonic, harmonic and subharmonic resonances by the application of the Ritz averaging method with two or three term approximations for forcing functions of the type T cos  $\omega t$  and  $C \omega^2$  cos  $\omega t$ . T and C are constants and  $\omega$  is the forcing frequency. The accuracy of this approximate method is verified by means of analog computer simulation. lytical solutions agree quite closely with analog computer The analytical solutions are also compared with experimental results obtained from a mechanical model with a bilinear restoring force characteristic. The mechanical model exhibits ultraharmonic and harmonic resonances, but fails to develop distinct subharmonic resonances owing to inadequate power capacity of the vibrator system used. With the analog computer however, the subharmonic resonance is excited over a limited frequency range. Although the experimental results are lower in magnitude than predicted, they distinctly show the nature of non-linear response. The theory of limiting conditions for the generation of subharmonic resonances is developed in Appendix IV.

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#### NOMENCLATURE

C Rotating torque.

C' Damping coefficient.

C<sub>c</sub> = 2p.J Critical damping coefficient.

E(0) Error function or differential equation of motion.

f<sub>3</sub> Forcing function.

J Polar moment of inertia of oscillating mass.

K Linear torsional stiffness.

M Mean dynamic displacement from the static equilibrium position.

 $\overline{M} = \frac{M}{\theta}$  Non-dimensional mean displacement.

 $p = \sqrt{\frac{K}{J}}$  Equivalent linear natural frequency.

Q1 Amplitude of fundamental component of motion.

Q<sub>2</sub> Amplitude of the second harmonic component of motion.

Amplitude of the third harmonic or of the ultra-harmonic of 3rd order.

Q<sub>4</sub> Amplitude of the fourth harmonic component of motion.

Q<sub>1/3</sub> Amplitude of the 1/3 harmonic or of the subharmonic  $\overline{Q}_1 = \frac{Q_1}{A}$ 

 $\overline{Q}_3 = \frac{Q_3}{\theta_0}$  Non-dimensional amplitudes.

 $\overline{Q}_{1/3} = \frac{Q_{1/3}}{\theta}$ 

G<sub>1</sub>,G<sub>2</sub>,G<sub>3</sub>,G<sub>4</sub> etc. Amplitudes of Marmonic components of acceleration.



 $n_1, n_2, n_3, n_4$  etc. Magnitudes of harmonic components of acceleration in terms of the acceleration due to gravity.

Distance of the accelerometer axis from the center of the shaft.

Radial distance of the contact line on tooth from the center of the shaft.

 $\overline{S} = \frac{T}{K\theta_0}$  Non-dimensional forcing amplitude.

Period of oscillation of motion, period of disturbing force.

T Periodic torque.

T Static torque.

Wr,  $Wr_1$  Uncertainties involved in the measurements of r and  $r_1$  respectively.

 $\text{Wn}_1, \text{Wn}_2, \dots$  Uncertainties involved in measurement of acceleration harmonics.

Wt Uncertainty involved in the measurements of the period of the disturbing force.

 $W\theta$  Uncertainty involved in gap clearance measurements.

 $Z = \frac{C}{J}$  Disturbing torque constant.

 $Z' = \frac{Z}{\theta_0} = \frac{C}{J\theta_0}$  Non-dimensional forcing amplitude.

 $\eta = \frac{\omega}{p}$  Frequency ratio.

 $\mu$  Coefficient of non-linearity ( $L^{-2}$ ).

Non-linear frequency, frequency of disturbing force.

Angular displacements at the beginning and the end of angular clearance.

Angular displacement from the static equilibrium position.

 $2\theta_0$  Angular clearance (backlash).

Angular position of beginning of angular clearance measured from 0.

 $\overline{\theta} = \frac{\theta}{\theta_0}$  Non-dimensional displacement from the static equilibrium position.

 $\zeta = \frac{C'}{C_C}$  Damping ratio.

#### 1

#### 1. INTRODUCTION

## 1.1 CHARACTERISTICS OF NON-LINEAR SYSTEMS

Vibrating systems have generally non-linear restoring force characteristics. In most cases the magnitude of nonlinearity is sufficiently small to provide acceptable results with a linear approximation. Non-linearities are of two types; continuous and discontinuous. Continuous non-linear restoring forces are produced by imperfect elasticity of materials, such as rubber used for flexible couplings and vibration isolation suspensions, geometric configurations of linear springs and by large amplitudes" of angular motion. There are specially designed springs, such as Belleville, sine and wire mesh, which have continuous non-linear characteristics. Discontinuous nonlinearities are exhibited by systems with restoring force characteristics represented by simple combinations of straight lines. In practice, they often result from the dis-. continuous contact with elastic restraints due to clearances and backlash.

The vibration of systems with non-linear restoring forces is periodic, but contains harmonics of the fundamental oscillation. As a result three distinct types of forced vibration can occur, which are the harmonic, ultraharmonic and subharmonic resonances. The harmonic resonance is equivalent to a linear resonance. An ultraharmonic is the resonance of one of the higher harmonics of motion when

it reaches the region of the natural frequency of the system. Similarly a subharmonic is the resonance of one of the harmonics whose frequency is 1/n times the forcing frequency, where n is an integer. The type of resonance and its order are defined by the ratio of the frequency of the predominant component of the motion to the frequency of the disturbing force.

The ultraharmonic and subharmonic resonances are the general characteristics of a non-linear system. Their order and magnitude are typical for a specific system. For example a non-linear system with symmetrical restoring force characteristics may exhibit odd orders of resonances only. Asymmetrical restoring force characteristic, on the other hand, may produce both even and odd order resonances.

## 1.2 SUBJECT OF INVESTIGATION

In this study, the vibration response of a torsional system with bilinear type torsional stiffness has been investigated. This non-linear restoring force characteristic results from a clearance in a geared system vibrating torsionally. The restoring force is zero within the clearance and acts linearly when contact is made. This type of restoring force characteristic is found in actual power transmission systems with angular clearances. These clearances are caused by the backlash in gear teeth, wear of tooth flanks etc. In torsional vibrations, excessive

shocks and cyclic stresses may result from such nonlinearities causing premature failures. In a particular case a failure due to such a condition has been observed in the drive of a Krupp-Renn revolving kiln (12).\*

The aim of this investigation is to study this nonlinear phenomenon qualitatively and quantitatively.

## 1.3 ANALYTICAL METHODS FOR THE SOLUTION OF NONLINEAR VIBRATION EQUATIONS

In most cases it is not possible to obtain exact solutions for the differential equations representing nonlinear vibration. Approximate solutions of such equations to the first degree, are not difficult or laborious.

These are often preferred over exact solutions for the sake of simplicity. The latter, if at all possible, often require tedious calculations. On the other hand, one of the difficulties with approximate solutions is the selection of the most suitable method from the many which are in existence. The factors which have to be considered when the choice is made, are the magnitude of non-linearity and the accuracy of the solution, the laboriousness of the procedure and the type of vibration or resonance represented by the equation.

There are several methods which produce solutions in the closed form. These are:

<sup>\*</sup> Numbers in parentheses denote publications listed in the bibliography, pages 56 through 68.

- 1. Ritz Averaging Method
- 2. Krylov-Bogoliubov and Mitropolski Method
- .3. Perturbation Method
- 4. Substitution Method

The Ritz averaging method provides sufficiently accurate solutions. The effectiveness of the method improves with the magnitude of non-linearity. Non-autonomous systems and asymmetrical restoring force characteristics do not present any undue difficulties. Higher orders of approximations yield better accuracy, but the complexity of the problem increases. However, the difficulties are of a purely algebraic nature. There is no impediment to the application of the Ritz averaging method to systems of more than one degree of freedom, where the majority of other methods become extremely complicated or fail altogether.

The substitution method closely follows the Ritz averaging method in terms of accuracy which increases with the order of approximation. There is no difference in the laboriousness of the two methods, but the simplicity of the mathematical basis for the substitution method may be an advantage in some applications.

The perturbation method is very laboriuos and perhaps its only advantage is that the form of the solution need not be assumed before application. The Krylov-Boboliubov method is very useful in the analysis of self excited oscillations and non-linear damping. The perturbation and Krylov-Bogolibov methods are suitable only for quasilinear systems with symmetrical

restoring force characteristics. In case of forced vibration, the amplitude of the disturbing force is restricted to small values within the limits imposed by the magnitude of non-linearity.

There are also graphical procedures available, such as the Phase-Plane method and numerical integration techniques. Their range of application is generally restricted to point solutions. Furthermore, they are unsuitable for the analysis of the more complex types of non-linear effects such as subharmonic and ultraharmonic resonances.

## 1.3.1 RITZ AVERAGING METHOD

After careful evaluation, the Ritz averaging method was selected for the problem under study. Its application is described below:

Consider the non-linear differential equation

$$J^{\theta} + C'f_{1}(\theta) + K f_{2}(\theta) = f_{3}(t)$$
 (1.3.1.1)

in which the restoring force  $f_2(\theta)$  and the damping force  $f_1(\theta)$  are non-linear odd functions of the displacement and velocity respectively. In other words  $-f_2(\theta) = f_2(-\theta)$  and  $-f_1(\theta) = f_1(-\theta)$ . If we divide through by the constant moment of mass inertia J and place  $\frac{K}{J} = p^2$  and  $\frac{C}{J}' = 2$   $\zeta p$ , the differential equation  $E(\theta)$  becomes:

$$E(\theta) = \theta + 2 p \zeta f_1(\theta) + p^2 f_2(\theta) - \frac{1}{J} f_3(t) = 0$$
(1.3.1.1a)

We then assume our approximate solution of equation (1.3.1.1a) consisting of n terms and denote it by  $\overset{\sim}{\theta}$ , so that

$$\tilde{\theta} = a_1 \phi_1(t) + a_2 \phi_2(t) + \dots + a_n \phi_n(t)$$
 (1.3.1.2)

Upon substitution of the approximate solution  $\theta$ , the differential equation becomes  $E(\tilde{\theta})$  and  $E(\tilde{\theta}) \neq 0$ . This equation deficiency  $E(\tilde{\theta})$  will vary from instant to instant, but over an arbitrary duration of time T it will be possible to demand that each of the n weighted averages of the deficiency must vanish. In calculating the weighted averages of  $E(\tilde{\theta})$ , we postulate the existence of n weight functions and denote them by  $\omega_1(t)$ ,  $\omega_2(t)$ , ...  $\omega_n(t)$ . The Ritz averaging criterion then states that if  $\omega_1(t)$  is placed equal to  $\phi_1(t)$ ,  $\omega_2(t)$  placed equal to  $\phi_2(t)$ , etc. and finally  $\omega_n(t)$  placed equal to  $\phi_n(t)$ , the resulting n averaging integrals, each placed equal to zero will yield n algebraic equations from which the coefficients  $a_1, a_2, \ldots a_n$  can be found. Under these circumstances the approximate solution for  $\tilde{\theta}$  will be the best obtainable in the n terms chosen.

The Ritz averaging integrals are given by:

$$\int_{0}^{T} \{E(\tilde{\theta}) \phi_{1}(t)\} dt = 0$$

$$\int_{0}^{T} \{E(\tilde{\theta}) \phi_{2}(t)\} dt = 0$$

and similarly

$$\int_{0}^{T} \{E(\theta) \phi_{n}(t)\} dt = 0$$
 (1.3.1.3)

3

For a steady state vibration of a non-linear system acted upon by a sinusoidal excitation, the Ritz criterion (1.3.1.3) may be expressed for the duration  $T=\frac{2\pi}{\omega}$  or alternatively for the angle of  $2\pi$  radians as follows:

and

#### 1.3.2 OTHER METHODS

Application of remaining methods has been explained in various textbooks (83), (38), (15), (84), (68), (38) and various publications (76).

## 1.4 SIMULATION OF THE PHYSICAL SYSTEM

### 1.4.1 ANALOG COMPUTER

In practice, physical systems can in most cases be represented by mathematical equations or sets of equations. Their solutions are often difficult or practically impossible to obtain by the classic approach. The analog computer provides rapid solutions of linear or non-linear equations and makes possible qualitative surveys of the behavior of the simulated system.

The analog computer performs the required mathematical operations in a parallel manner on continuous variables. In electronic analog computers, the continuous variables are

d.c. voltages. The electronic analog computer makes it possible to build an electrical model of a physical system, where the voltages on the computer represent the dependent variables of the physical system. Except for a constant of proportionality or a scale factor, each voltage will behave with time in a manner similar to the physical system variable. The actual behavior is thus simulated because of the equivalence of operation of the electrical model. capability of the analog computer is of great value in performing scientific research or engineering design because it permits an insight into the relationship between the mathematical equations and the response of the physical system. Once the electrical model is completed, well controlled experiments can be performed quickly, inexpensively, and with great flexibility to predict the behavior of the physical system.

The analog computer is basically a set of mathematical building blocks, each able to perform specific mathematical operations on voltages. By appropriately interconnecting these building blocks, an electrical model is produced, in which the voltages at the outputs of the blocks obey the relation given in the mathematical description of a physical system. The standard components of an analog computer perform the following operations: inversion, algebraic summation, integraion with respect to time, multiplication and division, and function generation.

#### 1.4.2 DIGITAL COMPUTER

Non-linear differential equations are difficult to compute even with the help of a digital computer (53). Recently, the availability of the Continuous System Modelling Program (CSMP) has simplified the numerical solutions of non-linear ordinary differential equations (22).

The Continuous System Modelling Program is designed to facilitate the digital simulation of continuous processes. whose behavior follows a set of ordinary differential equa-This program provides an application oriented language that allows these problems to be prepared directly and simply from either a block diagram representation or a set of ordinary differential equations. This program includes a basic set of functional blocks, with which the components of a continuous system may be represented. It accepts application oriented statements for defining the connections between these functional blocks. CSMP also accepts FORTRAN statements to deal with non-linear and time variant problems of considerable complexity. The application of CSMP allows the user to concentrate upon the phenomenon being studied rather than upon the detailed technique for numerical computation.

### 2. LITERATURE SURVEY

## 2.1 LITERATURE RELATED WITH BASIC NON-LINEAR VIBRATION STUDIES

Several books are exclusively devoted to non-linear vibrations, such as Stoker (83), Minorsky (68) and Hayashi (32).

Some chapters deal with this subject in Timoshenko (84), Jacobson and Ayre (38) and DenHartog (15). In general these books describe non-linear systems and methods of obtaining analytical solutions. An interesting publication (14) presents a broad overview of non-linear systems and their characteristics as compared to linear systems.

Most of the past work is based on non-linear systems represented by the well known Duffing equation, which is

 $x + C' x + K(x \pm \mu x^n) = P \cos \omega t$ where n is an integer,

Klotter and Pinney (48) have studied forced vibrations in systems represented by a Duffing Equation with a hardening type restoring force characteristic. They also have established a stability criterion for vibrations of their system.

Similarly, Caughey (10) has studied conditions for the existence and the stability of ultraharmonics and subharmonics in forced oscillations of systems having a small cubic non-linearity represented by the Duffing equation.

Burgess (9) has published a report on the harmonic,

superharmonic and subharmonic response of a single degree of freedom system of the Duffing type. Atkinson (2) has obtained ultraharmonic oscillations as solutions to the Duffing equation using an electronic differential analyzer and verified results obtained earlier by Burgess.

Raganti (79) has obtained the subharmonic solutions of order 1/3 of the damped Duffing equation with large non-linearity in a suitable parametric form. The solutions are compared with the results obtained by direct numerical integration of the same equation, using the Runge-Kutta method.

Hayashi (35) has investigated the stability criteria of non-linear periodic oscillations and has studied subharmonic oscillations in non-linear systems (33). He has shown that the order of subharmonics has a close relationship with the form of the non-linear characteristics e.g. subharmonic oscillations of order 1/3 are related with the non-linear characteristic expressed by cubic and quantic functions.

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## 2.2 APPROXIMATE METHODS FOR THE SOLUTION OF NON-LINEAR VIBRATION PROBLEMS

The literature was reviewed for the most suitable approximate method available for obtaining analytical solutions for the bilinear type restoring force characteristics.

Special attention was given to the methods in which an approximate solution is assumed. In the search along these lines several possible approaches were observed. For instance,

Mahalingam (63) has proposed an approximate solution for forced vibration. In this method, the problem of free vibration is first solved and the response to forced vibration is then found using a frequency function in place of the actual restoring force characteristics. The author claims that this method is powerful when the restoring force characteristic is made up of a number of straight lines. Similarly Schweisinger and Manmouth (81) have proposed a one term approximation method based on the Ritz procedure. DenHartog (16) has introduced approximate graphical solutions for the problem of forced vibration of an undamped single degree of freedom vibrating system with a non-linear spring, whose characteristic is given in the form of a curve. His approach is based on the quarter cycle energy method. As a follow up of DenHartog's publication, Silverman (82) came up with an application of Hamilton's principle to obtain forced vibration displacement in the form of a Fourier Series.

Brock (6) has proposed an iterative procedure employing numerical integrations for the analysis of free and forced vibrations of undamped systems having non-linear elasticity. There are few publications suggesting graphical solution for solving transient vibration problems. Lemon (50) has given a procedure for adopting the phase plane method to the solution of a single degree of freedom system with a non-linear restoring force characteristic. Bruce (7) has shown application of the graphical method to a simple case of free vibrations of a non-linear system, with a restoring force characteristic

composed of linear segments. Bishop (5) has again applied the graphical method to the free and forced vibrations of a bilinear system. He has also suggested the approximation of any spring characteristic by a series of straight lines. Evaldson, Ayre and Jacobsen (25) have obtained the response of a non-linear system, with spring characteristics composed of straight lines, to transient disturbances, using the graphical solution technique.

There are several references (83), (45), (46) and (47) on the application of the Ritz averaging method for solving non-linear vibration problems. A technical report by Klotter (45) is of particular interest. In this publication, the author has introduced the Ritz averaging method and has applied it to cases for which exact solutions were possible. The first case studied was the undamped vibration of systems with restoring forces of the polynomial type. Exact solutions were compared with results obtained by means of the Ritz method using a single term approximation and good agreement was shown to exist.

The second case relates to free or forced vibrations of systems with several types of piecewise linear restoring force characteristics. The author has again compared the exact values with the averaging method using a single term solution and has shown that the single term approximate solutions are very close to the exact solutions when the driving frequency is not close to any type of non-linear resonance other than harmonic.

Reif (77), (78) has applied the Ritz averaging method to obtain

solutions of the ultraharmonic resonance of order 2 and the subharmonic resonance of order 1/2 for a non-linear system with an asymmetrical restoring force characteristic. He has also verified these results by means of an analog computer. Reif (76) has also solved the Duffing equation using the Ritz averaging, Krylov-Bogoliubov, perturbation and substitution methods and has compared the accuracy and relative advantages of these methods on the basis of an analog computer simulation.

Levenson (51) derived a numerical solution of subharmonic response for the Duffing equation.

Ashwell and Chauhan (1) have applied the method of harmonic balance to a study of subharmonic oscillations of order 1/2 with single degree of freedom systems having non-linear spring characteristics of skew symmetrical form.

### 2.3 STUDY OF PIECEWISE LINEAR SYSTEMS

DenHartog and Mikina (16) and DenHartog and Heiles (17) have studied forced vibration in non-linear systems with various combinations of linear springs. The authors have obtained the exact solutions for the steady state motion of these systems under the influence of a harmonic external force. The results have been plotted for stiffness ratios  $K_1/K_2 = 0$ , 0.5, 2 and  $\infty$  in the non-dimensional form. In these solutions the presence of higher order harmonics in the motion has been ignored. They however form a basis for approximate solutions that may be analytically simpler.

Klotter (45) has obtained approximate solutions for forced vibration in non-linear systems with various combinations of linear springs, using the Ritz averaging method. This method is much simpler than the exact solution (18) and provides an acceptable degree of accuracy for practical applications.

Atkinson and Heflinger (3) have studied a bilinear system by means of an electronic analog. They have shown the existence of ultraharmonic and subharmonic components in the system response. This publication provides a basic understanding of the bilinear system.

Chaloupka (12) has given a case study of the drive of a Krupp-Renn revolving kiln, in which excessive wear of the gears was observed after several years of operation. The study indicates that the failure was caused by a subharmonic resonance of order 1/2, which resulted from the non-linear character of the torsional stiffness of the drive due to excessive backlash in the gear teeth. The author has studied this problem experimentally and analytically.

Bruevich (8) has investigated the action of harmonic oscillations on a piecewise linear system. An equation is obtained for the periods of a piecewise linear system of any order in the case when the characteristic of the non-linear element is continuous and consists of segments of two intersecting lines, and the external perturbance is harmonic.

Maezawa (55) has introduced an analytical method to

obtain steady state solutions for forced vibration of asymmetric piecewise linear systems using a method utilizing appropriate perfect Fourier series expansion. method consists of 1. Linearizing the original non-linear equation of motion by expanding the non-linear part of the restoring and damping force into a Fourier series with the same period as the given exciting force; 2. Obtaining the formal solution of the linearized equation by taking the non-linear part to be an exciting force from without. solution contains certain unknown coefficients of the Fourier expansion; 3. Determining these unknown coefficients from the conditions that the formal solution satisfies the given piecewise linear characteristics of the system. quirements result in an infinite set of simultaneous linear equations for the non-dimensionalized coefficients of this Fourier expansion as an infinite number of unknowns. author has given a method utilizing the appropriate series transformation for the improvement of convergence. computations were performed and response curves for two typical cases were shown to illustrate this method. theory was also verified by means of an analog computer.

Maezawa (56),(57), (58), (61), has also shown the application of the Fourier series method for obtaining subharmonic and superharmonic resonance solutions in a piecewise linear system. He (62) has also discussed forced vibrations in an unsymmetrical piecewise linear system excited by general

periodic force functions. The superharmonic resonances up to the second order, are analyzed by means of a perfect Fourier series.

Tsuda (90) has obtained solutions for harmonic, subharmonic and ultraharmonic resonances of a one dimensional power transmission system having an angular clearance. The author has discussed stability discrimination and has solved approximately the maximum amplitude state of the system under velocity proportional damping and collision damping, which takes place at the clearance. The author has presented a series of numerical diagrams to determine maximum amplitude states of the system. He has also verified experimentally his theory on the approximation of collision damping.

Yeh and Yao (93) have obtained the response of bilinear structure systems to earthquake loads using the continuous system modelling program (CSMP) (95). The results of their study show that the bilinear system is more effective in resisting earthquake loads than the corresponding linear system with either spring component of the bilinear system.

Karasudhi, Tan and Lee (40) have analyzed the vibration of a single storey frame with bilinear hysteresis, supporting a rotating machine. The excitation force, caused by the rotating unbalanced mass of the machine, was sinusoidal with a frequency dependent amplitude. It was shown that the system exhibits unbounded resonance when the product of the machine unbalanced mass and its eccentricity exceeds a critical value.

Masri (65) has obtained an exact solution for the steady state motion of a sinusoidally excited single-degree-of-freedom system with bilinear hysteresis and viscous damping.

Dubowsky and Freudenstein (22) have presented the dynamic behavior of an elastic mechanical joint with clearances. The authors have formulated a mechanical model and have obtained the dynamic response analytically and numerically. The results include the frequency response, displacements, force amplification and vibrational characteristics of the system under various operating conditions. The authors have shown the application of the describing function technique (well known in control engineering) to obtain approximate solutions of the system.

# OVERALL REVIEW OF ANALYTICAL METHODS SUITABLE FOR THE SOLUTION OF THE PIECEWISE LINEAR SYSTEM

The exact analytical solutions for piecewise linear systems obtained by DenHartog and Mikina (16), and DenHartog and Heiles (17), provide a basis for the comparison of solutions obtained by other approximate methods. The exact solution, discussed in the above mentioned references, does not make allowance for higher order harmonics and thus is unsuitable for ultraharmonic and subharmonic resonance solutions. Moreover the exact solution is tedious to obtain.

Graphical methods appear to be predominantly suitable for free vibrations and cannot be applied to ultraharmonic

and subharmonic resonances.

There are several publications suggesting one term approximation methods (81), (63), (82) and (5) for obtaining solutions for forced vibrations of non-linear systems with piecewise linear restoring forces. However, these methods can be applied to the harmonic resonance, but cannot be extended to the ultraharmonic and subharmonic resonances.

Several authors have used the Ritz averaging method with various non-linear problems, for which either exact solutions or experimental results, such as analog computer, were known. They have shown that the method provided very satisfactory accuracy and has greater scope of applicability. Burgess (9) has obtained solutions for harmonic, ultraharmonic and subharmonic responses of a system of the Duffing type, using the Ritz averaging method. Klotter (45) has applied this method to free and forced vibrations of systems with several types of piecewise linear restoring force characteristics and has shown good agreement between exact solutions and the Ritz averaging method with a single term approximation.

Based on these considerations, the Ritz averaging method was selected for the theoretical analysis of the system under study.

## 3. ANALYTICAL SOLUTIONS

# 3.1 THE DEVELOPMENT OF THE GOVERNING DIFFERENTIAL EQUATIONS OF MOTION

## 3.1.a. FORCING INPUT OF THE TYPE T Cos wt

Figure 3.1.1.1b shows the restoring force characteristics of an asymmetrical bilinear system. The asymmetry is caused by the action of a constant torque  $T_0$ . The mean position of the vibration is shifted to 0, which is now used as the origin. The angular displacement  $\theta_1$  represents the shift due to the constant torque  $T_0$ . The range of displacement is divided into three regions. Within regions I and III, the stiffness is linear and the tooth contact is made with the restrainer walls. Region II represents motion within the clearance where the stiffness is zero. The equation of motion in region I is given by:

 $J\theta$  +  $K\theta$  =  $T_0$  + T Cos  $\omega t$  for  $\theta \ge -\theta_1$  (3.1.A.1) The equation of motion in region II is given by:

 $J\theta = T_0 + T \cos \omega t \text{ for } -(2\theta_0^{+}\theta_1^{-}) \leq \theta \leq -\theta_1$  (3.1.A.2) and the equation of motion in region III is given by:

$$J\theta + K[\theta - (-(2\theta_0 + \theta_1))] = T_0 + T \cos \omega t$$

for  $\theta < -(2\theta_0 + \theta_1)$ 

(3.1.A.3)

For the special case of a symmetrical bilinear restoring force characteristic, the mean torque  $T_0 = 0$  and thus the asymmetry factor,  $\theta_1 = 0$ .

# 3.1.b FORCING INPUT OF THE TYPE C $\omega^2$ Cos $\omega t$

The equations of motion are derived for a system with a symmetrical restoring force characteristic. Figure 3.1.1.1C shows the restoring force characteristic.

The equation of motion in region I is given by:

$$J\theta + K(\theta - \theta_0) = C \omega^2 \cos \omega t$$

or

$$\theta + p^{2}(\theta - \theta_{0}) = Z \omega^{2} \cos \omega t \quad \text{for } \theta \geq \theta_{0}$$
 (3.1.B.1)

where

$$Z = \frac{C}{J}$$
 and  $p^2 = \frac{K}{J}$ .

The equation of motion in region II is given by:

$$J\theta = C \omega^2 \cos \omega t$$

or

$$\theta = Z \omega^2 \cos \omega t$$
 for  $-\theta_0 \le \theta \le \theta_0$  (3.1.B.2)

The equation of motion in region III is given by:

$$J\theta + K[\theta - (-\theta_0)] = C \omega^2 \cos \omega t$$

or ' \*

$$\theta + p^{2} [\theta - (-\theta_{0})] = Z \omega^{2} \cos \omega t \operatorname{for} \theta \leq -\theta_{0}$$
 (3.1.B.3)

## 3.2 ANALYTICAL SOLUTIONS OF THE EQUATIONS OF MOTION

The equations of motion are solved by using the Ritz averaging method. To apply the Ritz averaging method a solution for the displacement must be assumed in the form of a series.

The series is truncated in accordance with the degree of accuracy required. The assumed approximation must satisfy the boundary conditions corresponding to a periodic

solution. It must also be consistent with the physical restraints of the system. The general form of the solution can be expressed as:

$$\tilde{\theta} = M + Q_1 \cos \omega t + Q_2 \cos 2\omega t + Q_3 \cos 3\omega t + Q_4 \cos 4\omega t + \dots$$
 (3.2.1)

The constant term M results from the asymmetry of the restoring force. The term  $Q_1$  Cos wt represents the fundamental harmonic component of the motion and the other terms represent higher harmonic components. The fundamental harmonic of the same frequency as the disturbing force must be included to allow for the transfer of energy to the vibrating system. The other harmonics become predominant components of motion at their corresponding orders of ultraharmonic resonances, e.g.  $Q_2$ ,  $Q_3$  and  $Q_4$  peak out at 2nd, 3rd and 4th order ultraharmonic resonances.

For the harmonic resonance, a two term solution of the following form is assumed:

$$\tilde{\theta} = M + Q_1 \cos \omega t \qquad (3.2.2)$$

The above is the simplest approximation which can be used with an asymmetrical restoring force. The term M represents the shift of the mean value, the term  $Q_1$  Cos  $\omega t$  is necessary to allow transfer of energy from the forcing function. The analytical solutions are fully developed in Appendix I.

For 3rd order ultraharmonic resonance, the simplest form of the solution must consist of the following three terms:

$$\theta = M + Q_1 \cos \omega t + Q_3 \cos 3\omega t$$
 (3.2.3)

In the above, the additional term  $Q_3$  Cos  $3\omega t$  is included to represent the predominant component of the ultraharmonic resonance. The analytical solutions are obtained in Appendix I.

For the 1/3rd order subharmonic resonance, a three term solution of the following form must be used:

$$\tilde{\theta} = M + Q_1 \cos \omega t + Q_{1/3} \cos \frac{\omega t}{3}$$
 (3.2.4)

The term  $Q_{1/3}$  Cos  $\frac{\omega t}{3}$  represents the predominant component of the 1/3rd order subharmonic resonance. The analytical solutions are presented in Appendix I.

## 4. EXPERIMENTAL STUDY

## 4.1 SIMULATION TECHNIQUES

## 4.1.1 ANALOG COMPUTER SIMULATION

An EAI model TR-20 analog computer was used for this study. This computer has twenty operational amplifiers, twenty coefficient potentiometers, six integrator modules, non-linear computing components and function generators.

## ANALOG COMPUTATION

The dynamic equation of motion of a bilinear system is given by:

Where  $f(\Theta)$  represents the bilinear restoring force  $f(\Theta)$  characteristics.

and  $T(t) = T \cos \omega t$  or  $C \omega^2 \cos \omega t$ 

p = circular linear natural frequency.

Generalized circuits for the solution of equation (4.1.1.1) are given in figures 4.1.1.1 to 4.1.1.4. In order to maintain reasonable machine accuracy and to reduce the effect of transient vibration, separate computing sequences of short duration were used for each frequency. Corresponding values of total amplitude and phase angle were set as initial conditions for the forced vibration. A small amount of damping was introduced at the start of computation and was then gradually reduced to zero before the output was recorded by means of a two channel pen recorder. This

technique was used to remove the transient free vibration which was excited by the start of computation. For each set of results, one cycle of vibration was divided into several parts and ordinates were scaled off, and a harmonic analysis was carried out by means of a digital computer.

## 4.1.2 DIGITAL COMPUTER SIMULATION

The Continuous System Modeling Program (CSMP) was used to simulate the bilinear vibrating system. The block diagram and the corresponding structural statements are shown in Figure 4.1.2.1. Two function generators were used to produce the bilinear effect and the periodic disturbing force input. The general forms of the generators are shown in Figure 4.1.2.2.

At each frequency, initial values were set for the amplitude and numerical integrations were performed from t=0 to  $t=150~{\rm secs.}$  Results were printed and plotted at an interval of 0.1 sec. A small amount of damping was introduced in the system to damp out the transient component. However, the system response seemed to be influenced by damping. This was evident by the presence of a small phase difference between the periodic disturbing input and the vibration output. The vibration output was steady after  $t=100~{\rm seconds.}$  A cycle of steady state vibration was divided into several parts and ordinates were recorded. These values of ordinates were fed to another digital computer program for harmonic analysis.

#### 4.2 MECHANICAL MODEL TECHNIQUE

#### 4.2.1 DESIGN OF THE MODEL OF THE BILINEAR NON-LINEAR SYSTEM

The design details of the experimental model are shown in Figure 4.2.1.1. It consists of a shaft with a circular cross section 'S', a rectangular vibrating arm 'A', four bush bearing 'B', a tooth block 'D' and a rigid restrainer 'E'.

The system has a natural frequency of 26 Hz. The shaft diameter is 5/8" and the length between the vibrating arm and the tooth block is 32". The dimensions of the vibrating arm are 8"  $\times 2\frac{1}{4}$ "  $\times 1\frac{1}{4}$ ". The fixed end consists of a case hardened restrainer with a cavity to accommodate the tooth block. The tooth block is clamped on the shaft. The tooth block is made in two halves. The upper half contains a rectangular tooth and is case hardened. The tooth block assembly slides into the restrainer cavity. A slot is cut in the restrainer to accommodate the tooth block and a fixed clearance (0.0015") is maintained between the tooth and outer restrainer slot walls. The tooth block assembly is shown in Figure 4.2.1.2.

The shaft is supported by four bush bearings. Each bearing consists of a phosphor bronze bushing contained in a mild steel housing. A slot is cut in the bush and a screw is provided to adjust pressure on the shaft. A hole is also provided for lubrication.

The bearing housings and restainer are mounted on the base plate 'C'. Three 3/4" holes are provided on the base

plate to clamp it on a vibration isolated table. The overall view of the set up is shown in Figure 4.2.1.3.

The vibration excitation system is also shown in Figure 4.2.1.3. The electro dynamic shaker is mounted upside down on a specially fabricated structure. The shaker pin is connected to the vibrating arm through a soft helical spring. A counter weight is attached on the vibrating arm to bring the shaker pin to its mechanical center as shown in Figure 4.2.1.3.

#### REMARKS ON THE DEVELOPMENT OF THE EXPERIMENTAL MODEL

The described apparatus was developed through several stages of improvement. Initially the vibrator was suspended by means of springs but it was not possible to maintain its stability under all operating conditions. The shaker pin was also connected to the vibrating arm, by a piano wire. It was, however, observed that the motion of the vibrating arm was interacting with the shaker pin movement, which could not be kept sinusoidal. The results were not repeatable. Hence this scheme was abondoned and a fixed suspension system for the shaker was designed with a spring coupling to the model. This arrangement proved satisfactory in operation.

Several tooth gap clearances  $(2\theta_0)$  of magnitudes 0.01", 0.005" and 0.003" were tried. The vibration excitation system could not maintain constant shaker pin displacement within this range at higher frequencies. Thus the subharmonic resonance could not be excited. Consequently, the experiments were restricted to the 0.003" clearance only.

#### 4.2.2 INSTRUMENTATION

A schematic diagram of the experimental set-up is shown in Figure 4.2.2.1. The arm 'A' is excited sinusoidally at point 'Q' and the system response is picked up by an accelerometer mounted at location 'P'. The accelerometer output is amplified and fed to a real time analyzer to obtain frequency analyses.

## 4.2.2.1 VIBRATION EXCITATION SYSTEM

The vibration excitation system consists of an automatic vibration exciter control (B&K 1025), excitation amplifier (GM, 5535), electrodynamic shaker (PR 7270) and accelerometer preamplifier (B&K 2622).

## 4.2.2.2 VIBRATION PICK-UP SYSTEM

The vibration pick-up system consists of an accelerometer (B&K 4343) and accelerometer preamplifier (B&K 2623).

# 4.2.2.3 VIBRATION READOUT/PROCESSING SYSTEM

The vibration readout system consists of a measuring amplifier (B&K 2607), oscilloscope (Tetronix type 564), X-Y display (SD 13116-2A) and X-Y recorder (Hewlett-Packard).

The vibration processing system consists of a frequency analyzer such as B&K 2107 or 2113 and a real time analyzer (SD 301).

## 4.2.3 EXPERIMENTAL PROCEDURE

Firstly, the vibration test system was set up to control the shaker motion. To achieve shaker control an accelerometer 'R' (B&K 4333) was installed on the shaker pin. The output of this accelerometer was fed back to the

automatic vibration exciter control via the accelerometer preamplifier. This forms a control feedback loop for the shaker and the exciter. Figure 4.2.3.1 shows a control accelerometer mounted on the shaker pin attachment. The automatic vibration exciter control was then set up for a constant displacement mode by selecting a compressor regulation speed and adjusting the output voltage for a required vibration level. Subsequently the frequency was swept through the desired range and the system response was monitored.

## 4.2.4 MEASUREMENT AND ANALYSIS OF EXPERIMENTAL DATA

An accelerometer 'P' was installed on the vibration arm 'A' at a distance 3.5" from the shaft axis to monitor the system vibration response. The accelerometer was connected to the measuring amplifier through an accelerometer preamplifier and an associated power supply unit (ZR 0024). The measuring amplifier was calibrated using an accelerometer calibrator (B&K 4291).

The measuring amplifier output was connected to a Spectral Dynamics real time analyzer unit, which consisted of a real time analyzer, ensemble average (SD 309), X-Y display and an X-Y recorder. The real time analyzer was also calibrated in conjunction with the measuring amplifier. The X-axis of the X-Y display represented frequency on a linear scale and its range was selected on the real time analyzer. The Y-axis represented component amplitudes and was set to a logarithmic (dB) scale.

The shaker pin displacement was preset to 0.1 in. peak-to-peak and the frequency was scanned to observe the harmonic resonance and thus obtain the natural frequency of the mechanical model. Using this technique, the natural frequency of the system was found to be 26 Hz. The same procedure was repeated with several other shaker pin displacements to verify the results.

The shaker pin displacement was preset to 0.056 in., 0.07 in., 0.105 in., 0.140 in., and 0.175 in. peak-to-peak to correspond with values of  $\overline{S}$  = 0.8, 1.0, 1.5, 2.0 and 2.5. The relationship between the shaker pin displacement and  $\overline{S}$  is shown in Appendix III. The frequency of the shaker pin was adjusted manually. At each frequency setting the output of the accelerometer mounted at 'P' (figure 4.2.2.1) was observed on the storage oscilloscope and a frequency analysis was carried out using the real time analyzer. The acceleration components were converted into non-dimensional amplitudes  $\overline{Q}_1$ ,  $\overline{Q}_2$  etc. (shown in Appendix III). Several frequency settings were taken to observe the ultraharmonic, harmonic and subharmonic resonance phenomena.

#### 5. DISCUSSION OF RESULTS

#### 5.1 GENERAL DISCUSSION

The analytical solutions are firstly developed for a system with an asymmetrical restoring force characteristic. The asymmetrical solutions can be simplified for symmetrical restoring force characteristics by assuming  $\overline{\theta}_1 = 0$ .

The main contribution of this study is the development of the analytical solutions by means of the Ritz averaging method for the ultraharmonic, harmonic and subharmonic resonances of a system with a bilinear asymmetrical restoring force characteristic. It is believed that the obtained results, in terms of the asymmetry factor  $\overline{\theta}_1$ , are in the simplest form for adequate definition of the behavior of the system.

It should be noted that in the graphs analytical results are represented by solid lines and experimental results by points.

## 5.1.a HARMONIC RESONANCE:

The harmonic resonance phenomenon in a non-linear system corresponds basically to the linear resonance and takes place at the natural frequency of the system. The expressions are developed in Appendix I. Section I.1.1 deals with the disturbing torque of type T Cos ωt. The general expressions for the analytical solutions with the asymmetrical restoring force characteristic are obtained by assuming a two term solution and applying the Ritz averaging method. Two simultaneous equations, (I.1.1.6)

and (I.1.1.7), are produced.

The equations (I.1.1.6) through (I.1.1.9) are solved for selected values of the parameters  $\overline{S}$  and  $\overline{\theta}_1$ . Corresponding values of  $\overline{M}$  and  $\overline{n}$  are calculated for set magnitudes of  $\overline{Q}_1$ . The analytical results are computed for the asymmetry factor,  $\overline{\theta}_1$  = 0 and 0.05. The value  $\overline{\theta}_1$  = 0 represents the special case of the symmetrical restoring force characteristic. The other value  $\overline{\theta}_1$  = 0.05, represents weak symmetry and thus corresponds best to the restoring force characteristic of the mechanical model. The analytical results for the symmetrical restoring force characteristic for different values of excitation amplitude  $\overline{S}$  are shown in figure 5.1.1. The backbone curve ( $\overline{S}$  = 0) represents the free vibration response of the system. The response curve has positive and negative branches. The motion is stable due to the absence of the jump phenomenon.

The analog computer results are superimposed on the analytical results, shown in figure 5.1.1. The analog computer results are in close agreement with the analytical solutions and thus verify the accuracy of the Ritz averaging method. It also indicates that a two term solution is a very good approximation.

It should be noted, however, that the two term solution is not valid in regions of low or high values of  $\eta$ , where ultraharmonic and subharmonic resonances occur and corresponding approximate solutions must have additional terms. The

results are shown by dotted lines for  $\eta < 0.4 \ \text{where}$  ultraharmonic resonances develop.

Solutions for the asymmetrical restoring force characteristic, with an asymmetry factor  $\theta_1 = 0.05$ , are shown in figures 5.1.2 through 5.1.4. The response curve of the amplitude  $\overline{\mathbb{Q}}_1$  is shown in figures 5.1.2 § 5.1.3 for  $\overline{S} = 2.0$  and 1.0 respectively. While the analog computer results agree very well with the analytical solution, indicating again that the Ritz method with a two term approximation yields very good results, the experimental values are generally lower in magnitude and there is a shift to the right. One possible explanation is that the experimental value for the natural frequency of the mechanical model was measured too low, which resulted in higher values of the frequency ratio n. The natural frequency of the mechanical model was determined by running a forced vibration test. cedure, is discussed in section 4.2.4. Thus some error in determining the natural frequency is possible due to the influence of damping and the difficulty in locating exactly the peak of resonance. As an estimation of the influence of this error in determining the natural freuqency, referring to figure 5.1.2, the point  $\overline{\mathbb{Q}}_1$  = 9.4049,  $\eta$  = 0.9111 will shift to  $\eta$  = 0.87 for an error of 1 Hz in the experimental value of the natural freugncy and to  $\eta = 0.85$  and  $\eta = 0.81$  for errors of. 2 Hz and 3 Hz respectively. Similarly a point  $\overline{Q}_1$  = 3.9143,  $\eta = 0.7023$  will shift to  $\eta = 0.68$  for an error of 1 Hz and  $\eta = 0.65$  and  $\eta = 0.63$  for errors of 2 Hz and 3 Hz respectively.

In the analytical and analog computer models damping was absent, but in the mechanical model a damping ratio  $\zeta=0.07$  was determined experimentally. The experimental points show a typical influence of damping, which reduces the amplitude of vibration. Some additional energy dissipation is also caused by the collision of the teeth with the clearance walls. Although the experimental results are consequently lower, the general trend is maintained.

In the mechanical model, the vibration output was monitored by an accelerometer mounted on the vibrating arm at point 'P' (shown in figure 4.2.2.1). The accelerometer outputs at different forcing frequencies are shown in figures 5.2.1 through 5.2.4 and their frequency spectra are shown in figures 5.2.5 through 5.2.8. The forcing frequencies are selected to show typical acceleration patterns in the low frequency region, where the ultraharmonic resonances develop, and in the harmonic resonance region. Figure 5.2.7 shows an acceleration spectrum at a forcing frequency, f = 19.9 Hz(n = 0.76), which is away from the ultraharmonic resonance region and near the harmonic resonance. The odd order harmonics are predominant. The 1st order harmonic, which is the main component of the motion, has the maximum amplitude. Figure 5.2.8 presents the acceleration spectrum at the forcing frequency f = 33.3 Hz ( $\eta = 1.28$ ), which is just beyond the harmonic reasonance peak. Odd order harmonics predominate. Figures 5.2.5 and 5.2.6 show the frequency contents of the acceleration at forcing frequencies f = 9.6 Hz

( $\eta$  = 0.37) and 13.8 Hz ( $\eta$  = 0.53) respectively. The development of the 3rd and 2nd ultraharmonic resonances can be clearly seen.

The vibration energy is spread over the harmonic components of the actual vibration motion. In the approximate analytical solution all the energy is represented by a single term, which consequently will be greater in magnitude than the corresponding single experimental term.

The existence of the constant term M in the analytical solutions of the symmetrical restoring force characteristics is not clearly evident in the first instance. In general, M should not be there with the symmetrical system but it must be used because the origin for the definition of stiffness has been chosen at the point 'O' (shown in figure 3.1.1.1b). When  $\overline{\theta}_1 = 0$ , the point 'O' moves to the end of the clearance. The natural mean position of the motion remains off-set by an amount  $-\theta_0$  from this new origin, hence the mean shift M is equal to -1. The plot of M versus n is shown in figure 5.1.4.

The analog and digital computer outputs are shown in figures 5.1.6 through 5.1.9. The free vibration response obtained on the analog computer for  $\overline{\theta}$  = 2.0, 3.0, 4.0 is shown in figure 5.1.6. Digital computer simulation results for  $\overline{\theta}$  = 2.0 are shown in figure 5.1.9. Several runs were made on the digital computer to obtain free vibration responses for  $\overline{\theta}$  = 2.0, 3.0 and 4.0. In both simulation

techniques identical results were obtained. Similarly, forced vibration responses were obtained on the analog computer for different forcing amplitudes  $\overline{S}=1.0$  and 2.0. Typical analog computer outputs are shown in figures 5.1.7 and 5.1.8. From these plots the phase shift of the amplitude  $\overline{\theta}$  before and after the harmonic resonance is evident.

The harmonic resonance of the system, with a symmetrical bilinear restoring force characteristic, to the centrifugal type input C  $\omega^2$  Cos  $\omega t$  is obtained by solving two simultaneous equations (I.2.1.5) and (I.2.1.6), which are developed in section I.2.1 (Appendix I). Both equations are solved for set values of parameters  $\overline{\mathbb{Q}}_1$  and  $\mathbb{Z}^1$  and corresponding values of n are calculated. Analytical results for a particular value of Z' = 2.0 are shown in figure 5.1.5. The backbone curve (Z' = 0) is also drawn in this figure. There are two distinct branches for a particular value of The negative branch becomes asymptotic to  $|\overline{Q}_1| = Z'$ at high values of  $\eta$ , but the single term solution is not valid in this region due to the presence of the subharmonic resonance. The single term solution is also not valid in the  $\eta$  = 1/3 region, where the 3rd order ultraharmonic resonance takes place. The analog computer results are superimposed on the analytical results. There is good agreement with analytical solutions, which supports the suitability of the single term approximation for this case and the

accuracy of the Ritz method.

#### 5.1.b 3RD ORDER ULTRAHARMONIC RESONANCE:

Ultraharmonic resonances take place when frequencies of the higher harmonics of motion coincide with the natural frequency of the system.

The general form of the displacement is expressed in series form as:

$$\theta = M + Q_1 \cos \omega t + Q_3 \cos 3\omega t + Q_5 \cos 5\omega t + \dots \text{ etc.}$$

In the above expression normally  $Q_3$  and  $Q_5$  etc. are negligibly small, unless  $3\omega$  = p or nearly so, then n = 1/3 and  $Q_3$  is magnified producing the ultraharmonic resonance of order 3. This condition is evident in the spectrum shown in figure 5. Similarly at or near  $5\omega$  = p,  $Q_5$  is magnified. The order of the ultraharmonic resonance is defined by the ratio  $\omega$ . The term  $Q_1$  must be included in the corresponding approximation solution for the transfer of energy to all vibration components, which can take place only at the frequency of the disturbing effect. Hence the simplest approximate solution for the 3rd order ultraharmonic resonance must have the first three terms. Thus,

$$\theta = M + Q_1 \cos \omega t + Q_3 \cos 3\omega t$$

The analytical solutions are developed in Appendix I. Section I.1.2 corresponds to the input of type T Cos  $\omega t$  and

section I.2.2 covers the input of the type C  $\omega^2$  Cos  $\omega t$ .

The application of the Ritz method with the disturbing torque T Cos wt results in five simultaneous equations (I.1.2.5) through (I.1.2.9). Similar expressions are developed for an input of the type C  $\omega^2$  Cos  $\omega t$ , by assuming a two term solution. The application of the Ritz averaging method results in three simultaneous equations (I.2.2.7) through (I.2.2.9). The equations (I.1.2.5) through (I.1.2.9)are solved for selected values of the parameters  $\overline{S}$  and  $\overline{\theta}_1$ . For set magnitudes of  $\overline{\mathbb{Q}}_3$  corresponding values of  $\overline{\mathbb{M}}$ ,  $\overline{\mathbb{Q}}_1$  and  $\mathbb{N}$ are calculated. The analytical results are obtained for  $\overline{\theta}_1 = 0$  and 0.05. The analytical results for the symmetrical restoring force characteristic are shown in figure 5.1.10 and 5.1.11. Figure 5.1.10 shows the response curve of the 3rd harmonic component  $\overline{\mathbb{Q}}_3$  and the harmonic component  $\overline{\mathbb{Q}}_1$  is shown in figure 5.1.11. The curve of  $\overline{\mathbb{Q}}_3$  has both positive and negative branches and the motion appears to be stable since conditions for a jump are not present. The component  $\overline{\mathbb{Q}}_3$  has a well defined resonance near n = 1/3. The component  $\overline{Q}_1$  has two positive branches, which correspond to the positive and negative branches of  $\overline{\mathbb{Q}}_3$ .  $\overline{\mathbb{Q}}_1$  does not depart significantly from the harmonic resonance curve developed in the previous  $\overline{\mathbb{Q}}_3$  becomes dominant only in the ultraharmonic resonance region: The relative magnitudes of the 3rd order ultraharmonic resonance are compared to the harmonic resonance in figure 5.1.34. Analytical and analog computer results show good agreement. Therefore, the validity of the three term approximation and the accuracy of the Ritz method are verified.

The analytical results for the asymmetrical restoring force characteristics with an asymmetry factor,  $\overline{\theta}_1 = 0.05$ and  $\overline{S}$  = 2.0 and 1.0 are shown in figures 5.1.12 through 5.1.17. The experimental results were obtained for the mechanical model for  $\overline{S} = 0.8$ , 1.0, 1.5, 2.0 and 2.5 and are shown in figures 5.2.9 through 5.2.20. The experimental data reduction technique is discussed in detail in section 5.2. Figures 5.1.12 and 5.1.13 show analytical and experimental results of  $\overline{\mathbb{Q}}_3$  and  $\overline{\mathbb{Q}}_1$  for a particular value of  $\overline{S}$  = 2.0. Referring to figure 5.1.12, experimental values of  $\overline{\mathbb{Q}}_3$  indicate a valley and a peak. The peak corresponds to the 3rd order ultraharmonic resonance and it is shifted towards the right. The explanations for this shift are the same as for the harmonic resonance. An additional factor, which is of greater influence here than in the harmonic resonance region, is the presence in the actual motion of significant higher hadmonics, as shown in figure 5.2.5. Since these are neglected in the theoretical solution, experimental values of  $Q_3$  must be lower due to the spread of energy over a greater number of components. The experimental results dip down between  $\eta = 0.2$  and 0.4 with a minimum at approximately n = 0.25. This is possibly due to the formation of the 4th order ultraharmonic. Under these conditions

the 4th order harmonic would absorb the major portion of the vibration energy at the expense of the 3rd order component. Similar conditions appear to prevail near  $\eta = 0.5$  where the 2nd order ultraharmonic resonance may be initiated. Referring to figure 5.2.11 and 5.2.12, it is observed that the component  $\overline{\mathbb{Q}}_4$  peaks out at n=0.25, then  $\overline{\mathbb{Q}}_3$  starts developing and reaches maximum value at about  $\eta$  = 0.36. Referring to figures 5.2.10 and 5.2.11, it is observed that the component  $\overline{\mathbb{Q}}_2$  peaks at about  $\eta$  = 0.53 and  $\overline{\mathbb{Q}}_3$  bottoms out. The experimental values of  $\overline{\mathbb{Q}}_1$  are consistently lower than the analytical results and they are also shifted towards the right (figure 5.1.13). As in the case of harmonic resonance the contributing factor for lowering of experimental results are damping, error in determination of the natural frequency of the mechanical model, collision and approximation in the solution. Although the experimental results are lower in magnitude, yet they show well defined 2nd, 3rd and 4th order ultraharmonic resonances. Experimental results for  $\overline{S}$  = 1.0 & 1.5 are shown in figure 5.1.15 through 5.1.17 and 5.2.13 through 5.2.16. The results indicate the same trends as for  $\overline{S}$  = 2.0, as discussed before.

The ultraharmonic resonance, produced by the forcing input of the type C  $\omega^2$  Cos  $\omega$ t, is obtained by solving the three simultaneous equations (I.2.2.7) through (I.2.2.9). For set magnitudes of  $\overline{\mathbb{Q}}_3$  corresponding values of  $\overline{\mathbb{Q}}_1$  and  $\mathfrak{q}$  are calculated for selected values of the parameter Z'.

Analytical results, for a particular value Z' = 10.0, are shown in figures 5.1.18 and 5.1.19. Figure 5.1.18 shows the curve of the 3rd harmonic component  $\overline{\mathbb{Q}}_3$  and that of the fundamental component  $\overline{\mathbb{Q}}_1$  is shown in figure 5.1.19. The curve of  $\overline{\mathbb{Q}}_3$  has positive and negative branches and the motion will be stable since the conditions for a jump are not present. The component  $\overline{\mathbb{Q}}_1$  has two positive branches corresponding to both branches of  $\overline{\mathbb{Q}}_3$ .

The analog and digital computer simulation results are superimposed on the analytical solutions, shown in figures 5.1.18 and 5.1.19. On the analog computer it was observed that the 3rd order ultraharmonic resonance could only be generated at comparatively high values of Z'.

An expression for the limiting value of Z', below which non-linear conditions do not exist, is developed in Appendix IV. With this criterion it can be shown that the system has a non-linear restoring force characteristic only when  $Z' \geq 1$ . Hence, the ultraharmonic resonance can be excited only with values of Z' greater than 1. However, the 3rd order ultraharmonic resonance could not be produced on the analog computer for magnitudes of Z' lower than 6.0. It appears that this is due to inherent damping present in the electronic components of the analog computer, the introduction of damping at the beginning of computation and the weak nature of the ultraharmonic resonance itself.

The analog and digital outputs are shown in figures

5.1.20 through 5.1.23. In figures 5.1.20 and 5.1.21 the existence of the 3rd order ultraharmonic resonance is clearly evident. Similarly the analog and digital computer outputs for the forcing function C  $\omega^2$  Cos  $\omega t$  and values of Z' = 10.0 and  $\eta$  = 0.22 are shown in figures 5.1.22 and 5.1.23.

## 5.1.c 1/3RD ORDER SUBHARMONIC RESONANCE:

When the frequency  $\omega$  of the disturbing torque rises well above p, the natural frequency of the system, new and additional components appear in the motion with frequencies of  $\frac{\omega}{3}$  and  $\frac{\omega}{5}$  etc. The original harmonics of frequencies  $2\omega$ ,  $3\omega$ , etc. become now negligibly small. The displacement function is:

$$\tilde{\theta} = M + Q_1 \cos \omega t + Q_{1/3} \cos \frac{\omega t}{3} + Q_{1/5} \cos \frac{\omega t}{5} + \dots$$

$$+ Q_3 \cos 3\omega t + Q_5 \cos 5\omega t + \dots \text{ etc.}$$

In the above expression  $Q_{1/3}$  and  $Q_{1/5}$  are small in comparison with  $Q_1$ , unless  $\frac{\omega}{3}$  = p or nearly so, when  $Q_{1/3}$  is magnified and the subharmonic resonance of order 1/3 takes place. Similarly at  $\frac{\omega}{5}$  = p,  $Q_{1/5}$  is magnified and subharmonic resonance of order 1/5 is generated. The term  $Q_1$  must be included in the assumed solution to allow for the transfer of vibratory energy at the frequency of the disturbing force. The simplest approximate solution for the 1/3rd order subharmonic resonance will thus have the following terms:

$$\tilde{\theta} = M + Q_1 \cos \omega t + Q_{1/3} \cdot \cos \frac{\omega t}{3}$$

The analytical solutions are derived in Appendix I. Sections I.1.3 and I.2.3 correspond to the inputs of the type T Cos  $\omega t$  and C  $\omega^2$  Cos  $\omega t$  respectively.

The application of the Ritz method to the asymmetrical case results in five simultaneous equations, (I.1.3.5) to (I.1.3.9). Similar expressions are developed for the symmetrical restoring force characteristic and an input of the type C  $\omega^2$  Cos  $\omega t$  by assuming a two term solution. simultaneous equations (I.2.3.7) to (I.2.3.9) are then pro-The solutions are shown in Appendix I. The equations (I.1.3.5) through (I.1.3.9) are solved for selected values of the parameter  $\overline{S}$  and  $\overline{\theta}_1$ . For set magnitudes of  $\overline{\mathbb{Q}}_{1/3}$  corresponding values of  $\overline{\mathtt{M}},~\overline{\mathbb{Q}}_1$  and  $\mathtt{n}$  are calculated. The analytical results are obtained for the symmetrical restoring force characteristic  $(\overline{\theta}_1 = 0)$  and are shown in figures 5.1.24 through 5.1.26. The figure 5.1.24 shows the response curve of the 1/3rd harmonic component  $\overline{\mathbb{Q}}_{1/3}.$  The harmonic component  $\overline{\mathbb{Q}}_1$  is shown in figure 5.1.25. Both are drawn for  $\overline{S}$  = 2.0. The curve of  $\overline{Q}_{1/3}$  has both positive and negative branches and the motion is stable since conditions for a jump are not present. In the analytical results the component  $\overline{\mathbb{Q}}_{1/3}$  has a well defined resonance at  $\eta$  = 3.0. The component  $\overline{Q}_1$  has two negative branches corresponding to both branches of  $\overline{\mathbb{Q}}_{1/3}$ . The comparison of relative magnitudes of  $\overline{\mathbb{Q}}_{1/3}$  and  $\mathbb{Q}_1$  indicates that in the subharmonic resonance region,  $\overline{\mathbb{Q}}_{1/3}$  develops very rapidly whereas  $\overline{\mathbb{Q}}_1$  drops down slowly. For instance, while  $\overline{\mathbb{Q}}_{1/3}$  increases from 2.0 to 14.0,  $\overline{\mathbb{Q}}_1$  drops down from 0.8 to 0.2, for a variation of  $\mathfrak{n}=1.8$  to 2.8. A plot of  $\overline{\mathbb{M}}$  versus  $\mathfrak{n}$  is shown in figure 5.1.26.

The physical system was simulated on the analog and digital computers for a symmetrical bilinear restoring force characteristic. Both types of inputs were used. T Cos.  $\omega$ t and  $\overline{S}$  = 2.0, the system produced 1/3rd order subharmonic resonance over a narrow frequency range,  $\eta = 1.6$  to 2.0, with the right initial conditions. As the frequency increased beyond  $\eta$  = 2.0, the overall amplitude of the vibration became less than the clearance, thus the tooth lost contact with the retainer walls and the system reverted to a linear restoring force characteristic. These results indicate that the excitation of this type of resonance depends upon the magnitude of the disturbing effect. It was expected that a disturbing force amplitude-frequency boundary was present, beyond which the subharmonic resonance cannot exist. mathematical development of the equation of this boundary is shown in Appendix IV. This criterion establishes a relationship between cut-off frequencies and forcing amplitudes. The cut-off frequency is the highest value, above which a forcing function cannot excite the subharmonic resonance.

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Although for  $\overline{S}=2.0$  the theoretical cut-off frequency is  $\eta=1.45$ , the analog computer results show the possible presence of the subharmonic resonance for values of  $\eta$  up to 2.0. The application of a larger amplitude of motion as the initial condition, will result in a wider frequency range of tooth contact, thus extending correspondingly the subharmonic resonance limit.

The sample wave forms for  $\overline{S}=2.0$  and  $\eta=1.6$  are shown in figure 5.1.30. The digital computer simulation response is shown in figure 5.1.31. This technique also verifies analytical and analog computer results as shown in figure 5.1.24.

To generate the 1/3rd order subharmonic resonance in the mechanical model, a stepped shaft of natural frequency 12 Hz was used in place of the 5/8" diameter shaft with 26 Hz natural frequency. The purpose of this change was to be able to vibrate the system at higher frequencies and still maintain constant displacement output within the power limitations of the shaker.

In the testing procedure the system was vibrated through a wide range of frequencies with a maximum value of  $\overline{S}=2.5$ . It was not possible to generate a subharmonic resonance, at higher frequencies the system just followed the harmonic resonance response. As indicated by the cut off frequency equation, at higher frequencies, the tooth lost contact with the walls of the clearance space and reverted to a linear restoring force. Owing to the inherent damping produced

primarily by the four bush bearings, the maximum output power of the shaker was insufficient to increase the amplitude for maintaining tooth contact and hence also the nonlinear restoring force in the subharmonic resonance region. Moreover, the dissipation of energy is proportional to the square of the frequency of vibration, and hence the subharmonic components are more sensitive to damping. It is expected that if a shaker with sufficient power had been available, the subharmonic resonance would have been generated.

The analytical results which are independent of damping, indicate with sufficiently large disturbing force amplitudes the presence of a strong 1/3rd order subharmonic resonance, as shown in figure 5.1.24. It could only be verified with a limited success by means of simulation techniques utilizing analog and digital computers.

The subharmonic resonance response to the forcing input of the type C  $\omega^2$  Cos  $\omega$ t is shown in figures 5.1.28 and 5.1.29. Figure 5.1.28 presents the curve of the 1/3rd harmonic component  $\overline{\mathbb{Q}}_{1/3}$ . Similarly, the harmonic component  $\overline{\mathbb{Q}}_1$  is shown in figure 5.1.29. The curve of  $\overline{\mathbb{Q}}_{1/3}$  has positive and negative branches and the motion will be stable due to the absence of the jump phenomenon. The component  $\overline{\mathbb{Q}}_1$  has two negative branches corresponding to both branches of  $\overline{\mathbb{Q}}_{1/3}$ . There is close agreement between simulation and analytical results for  $\overline{\mathbb{Q}}_{1/3}$ . But the analog computer values for  $\overline{\mathbb{Q}}_1$  are more scattered and generally are lower in magnitude. The dif-

ference is caused by damping introduced in the simulation technique. Typical analog and digital computer wave forms are given in figures 5.1.32 and 5.1.33.

With this type of disturbing effect it is also possible for the system to revert to linear restoring force conditions. The appropriate analysis is shown in Appendix IV. It establishes the limiting value of Z' below which non-linear conditions do not exist and hence the subharmonic resonance cannot be excited. Using this criterion the limiting value of Z' is 1. Therefore with  $Z' \geq 1$ , the subharmonic resonance can take place.

## 5.1.d OVERALL VIEW OF THE RESONANCE PHENOMENA

The analytical solutions for the 3rd order ultraharmonic, harmonic and 1/3rd order subharmonic resonances with the excitation torque of the type T Cos wt are grouped together in figure 5.1.34 for comparison. In the absence of damping, the amplitudes become asymptotic or indefinite near the resonance. Hence their peak values are indeterminate. The development of typical non-linear resonance is clearly observed. In comparison with the harmonic resonance, it is evident that the intensity of the 1/3rd order subharmonic resonance is at least as great. Although because of equipment limitations, a quantitative experimental verification was not possible, there is sufficient evidence to indicate that in Tightly damped systems subharmonic resonances can be destructive. The amplitudes of the 3rd order ultraharmonic resonance are relatively smaller and should not,

in general, cause serious problems. Vibration with a symmetrical bilinear restoring force characteristic contains predominantly odd order ultraharmonic and subharmonic resonances, as shown for example by Atkinson (3).

# 5.2 REVIEW OF RESULTS OBTAINED BY THE MECHANICAL MODEL TECHNIQUE

Acceleration wave forms and their frequency composition at several frequency settings for  $\overline{S}$  = 2.0 are shown in figures 5.2.1 through 5.2.8. The frequency spectra at these forcing frequencies indicate the presence of several higher harmonic components. For comparison of magnitudes of individual harmonic components, one should bear in mind that the amplitude of the first acceleration harmonic component is equal to  $\omega^2$   $\textbf{Q}_1$  and similarly the second, third and fourth components are  $4\omega^2$   $Q_2$ ,  $9\omega^2$   $Q_3$  and  $16\omega^2$   $Q_4$  respectively, where  $\mathbf{Q}_1$  ,  $\mathbf{Q}_2$  ,  $\mathbf{Q}_3$  amd  $\mathbf{Q}_4$  are the displacement amplitudes. each frequency setting, the first four harmonics were read from the real time analyzer output and were converted into non-dimensional amplitudes  $\overline{\mathbb{Q}}_1$  ,  $\overline{\mathbb{Q}}_2$  ,  $\overline{\mathbb{Q}}_3$  and  $\overline{\mathbb{Q}}_4$  (as shown in Appendix III.2). The plots of  $\overline{\mathbb{Q}}_1$ ,  $\overline{\mathbb{Q}}_2$ ,  $\overline{\mathbb{Q}}_3$  and  $\overline{\mathbb{Q}}_4$  versus  $\eta$ are presented in figures 5.2.9 through 5.2.12. A careful study of these results indicates the existence of 2nd, 3rd and 4th order ultraharmonic resonances. The even order ultraharmonic resonances are present because of effective asymmetry in the restoring force characteristic, which reresults from the inability to obtain perfect balance of the driving arm. The asymmetry is specified by the parameter  $\overline{\theta}_1$ , which has been defined before. Figure 5.2.9 shows, for  $\overline{\theta}_1$  = 0.05, the variation of  $\overline{Q}_1$ . In figure 5.2.10 the magnification of  $\overline{Q}_2$  in the neighbourhood of  $\eta = 0.5$  indicates the 2nd order ultraharmonic resonance. A subsidiary energy transfer from the 4th order resonance is also evident in the vicinity of  $\eta = 0.25$ . Similar conditions are evident in Figures 5.2.11 and 5.2.12 and are repeated for other values of  $\overline{S}$  in Figures 5.2.13 to 5.2.20.

In summary the mechanical model showed a well defined harmonic resonance and 2nd, 3rd and 4th order ultraharmonic resonances. The comparison of amplitudes indicates that the strength of resonance decreases inversely with the order magnitude.

Subharmonic resonances, as discussed previously, could not be generated on the mechanical model because of the inadequate power of the vibrator.

For quantitative evaluation, spectra of experimental values of  $\overline{\mathbb{Q}}_1$ ,  $\overline{\mathbb{Q}}_2$ ,  $\overline{\mathbb{Q}}_3$  and  $\overline{\mathbb{Q}}_4$  at different values of n, for  $\overline{\mathbb{S}}=1.0$ , 1.5, 2.0 and 2.5 are shown in figures 5.2.21 through 5.2.24. Referring to figure 5.2.21, before the harmonic resonance, the  $\overline{\mathbb{Q}}_1$  component developes in proportion with the forcing amplitude  $\overline{\mathbb{S}}$ . But beyond it decreases rapidly, hence the accuracy of the measured values in that region is in question. A greater dependence of the higher order components upon  $\overline{\mathbb{S}}$  in the range of their respective resonances is evident

from Figures 5.2.22 to 5.2.24.

## 5.3 ESTIMATE OF EXPERIMENTAL ERRORS

The errors are considered in two parts:

- 1. Transducer and readout equipment.
- 2. Data reduction.

## Transducer and Readout Errors:

Accelerometers, including cables, are individually calibrated by the manufacturer to a suggested accuracy of ± 2%, with flat frequency response within 2% from 1 Hz to 12000 Hz and with stability better than 2% per year. The mass of the accelerometer is quite small in comparison with the vibrating arm. Hence the mass loading effect of the accelerometer is negligible.

The accuracy of the accelerometer calibrator is stated to be better than ± 2% by the manufacturer.

The digital counter, which was used to determine frequency of the shaker pin, measured the period of the oscillation in milli seconds. The instrument error on frequency values was within  $\pm$  0.1%.

## Data Reduction:

The real time analyzer was calibrated for 1 'g' peak which corresponded with 50 dB on the Y-axis. During frequency analyses of the accelerometer output, the acceleration harmonics were read within ± 1 dB or 12.23% accuracy. The acceleration harmonics were converted into non-dimensional amplitudes as shown in Appendix III.2. Evaluation of equation

III.2.2 indicates that the non-dimensional amplitude of motion  $(\overline{\mathbb{Q}}_1)$  depends primarily upon the magnitude of the acceleration harmonic  $(\mathbf{n}_1)$  and the angular clearance  $(\theta_0)$ . The accuracy in the calculation of  $\overline{\mathbb{Q}}_1$  is a function of accuracy involved in acceleration measurements, which are within  $\pm 1$  dB and clearance measurements, which can be within  $\pm 16.66\%$ . The period 't,' can be measured within 0.1% accuracy. Hence it is not considered a source of errors.

The error involved in natural frequency measurements would influence the values of the frequency ratio  $\eta$ . As a rough estimate, an overall reading error of 2 Hz would result in an error of 7% in the value of  $\eta$ . An analytical approach of the experimental error analysis is presented in Appendix V.

## 5.4 REMARKS ON THE APPLICATION OF THE RESULTS

The piecewise linear restoring force characteristic is caused by discontinuous contact with elastic restraints due to clearances. The example of such phenomenon is a geared system in which torsional oscillations are generally present when transmitting power. Such a system will exhibit the harmonic, ultraharmonic and subharmonic resonances. Chaloupka (12) has shown the existence of a 1/2 order subharmonic resonance in a torsional system with gears. In this particular publication, the author has studied, the behavior of a system having piecewise linear restoring force characteristics caused by backlash in gear teeth.

The results obtained suggest that ultraharmonic resonances

can be easily excited, but because of the relatively small amplitudes they are not likely to cause serious problems in real systems. Subharmonic resonances, on the other hand, will only be excited with large disturbing forces in strongly non-linear systems with relatively low damping. Although this may not occur often in practice, when it does it is likely to be very destructive.

The relatively simple method developed in this project, of calculating the limiting frequency above which the subharmonic resonances cannot be excited, should be very useful in practical applications.

Although the study dealt with torsional vibration, the results obtained can be easily applied to the vibration any system with clearances, by replacing appropriately angular parameters and variables.

## 6. CONCLUSIONS AND SCOPE OF FUTURE WORK

#### 6.1 CONCLUSIONS

The following conclusions are drawn from the investigation undertaken:

- The Ritz averaging method with two or three term approximation provides satisfactory results.
- The mechanical model with slight asymmetrical restoring force characteristics develops even and odd order ultraharmonic resonances.
- 3. A system with symmetrical restoring force characteristics develops only odd order ultraharmonic and subharmonic resonances.
- 4. The ultraharmonic resonances can be easily generated, but their intensities are much smaller in comparison with the harmonic resonance.
- 5. Analytical solutions without damping indicate the possibility of large amplitude subharmonic resonances in the case of a forcing input of the type T Cos ωt. Experiments with the mechanical model and the analog computer show that it is difficult to generate this resonance in actual systems at low excitation amplitudes. The subharmonic resonance appears also to be proportionately more susceptible to damping. It must be therefore concluded that such a resonance is likely to occur only in systems with large non-linearity, very low damping and high excitation forces. If it is excited, however,

it is at least as destructive to the vibrating system as the harmonic resonance.

6. The analytical solutions indicate also the possibility of large amplitude ultraharmonic resonance with excitation of the type C  $\omega^2$  Cos  $\omega t$ . The system failed to show this resonance at low magnitudes of Z'. Experimental results however, suggest again that such a resonance can only be excited in system with very low damping and high excitation forces. On the other hand, as indicated by analog computer testing, this type of excitation appears to have a higher potential for generating subharmonic resonances. The magnitude of this resonance is significant and it is comparable with the harmonic resonance.

## 6.2 SCOPE OF FUTURE WORK

The most important consequence of non-linear vibration, caused by clearances, is the possibility of exciting subharmonic resonances. It is recommended that an improved mechanical model with a sufficiently powerful vibrator, be used to verify the theoretical results obtained in that frequency domain.

In geared systems there are additional sources of nonlinearity, which were not considered in this study. Since their combined effect may be greater than that due to clearances, it is also recommended to investigate the following:

1. Stiffness variations due to the number of teeth simul-

taneously in contact.

2. Stiffness variations due to the sliding of the point of contact along the tooth profile.

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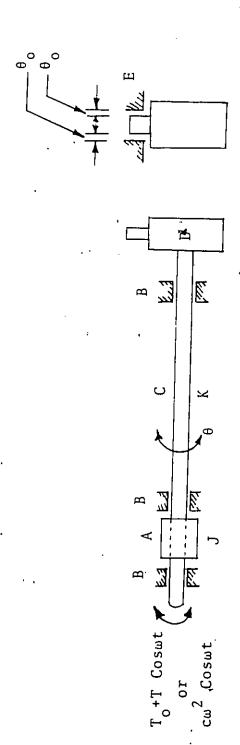
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A - Vibrating Arm

- Bush Bearings

- Shaft

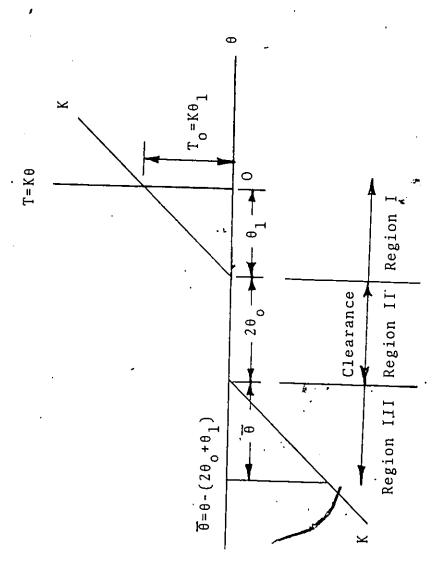
- Tooth Block

3 - Restrainer

J - Polar Moment of Inertia of the Vibrating Arm

K - Torsional Stiffness of the Shaft

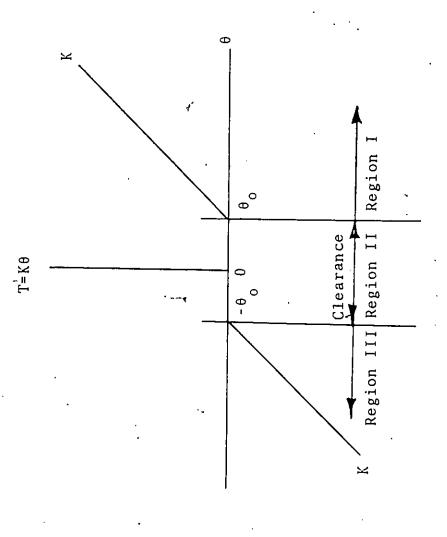
Representation of a Bilinear System Figure 3.1.1.1a.



Restoring Force Characteristics of a Bilinear System (Asymmetrical) Figure 3.1.1.1b

. i.

B



Restoring Force Characteristics of a Symmetrical Bilinear System Figure 3.1.1.1c

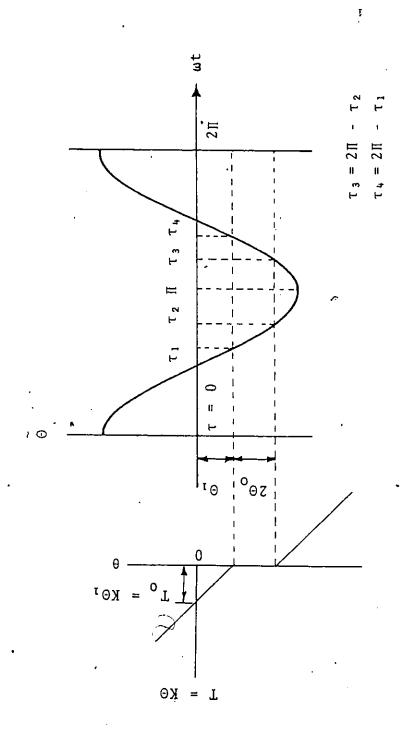


Figure 3.1.1.2 Amplitude Versus Angular Displacement Plot (Asymmetrical System). Harmonic Resonance.

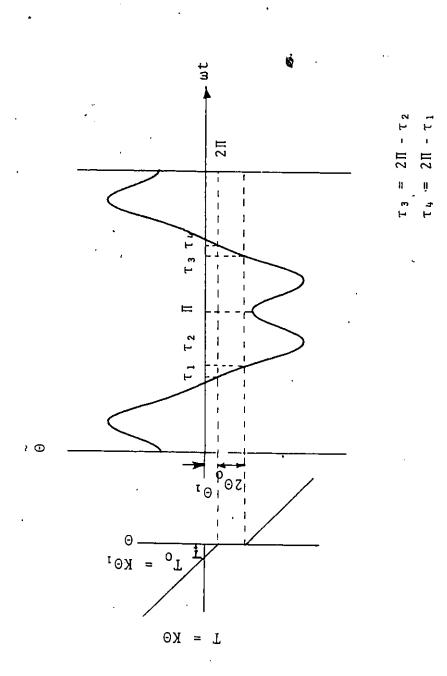
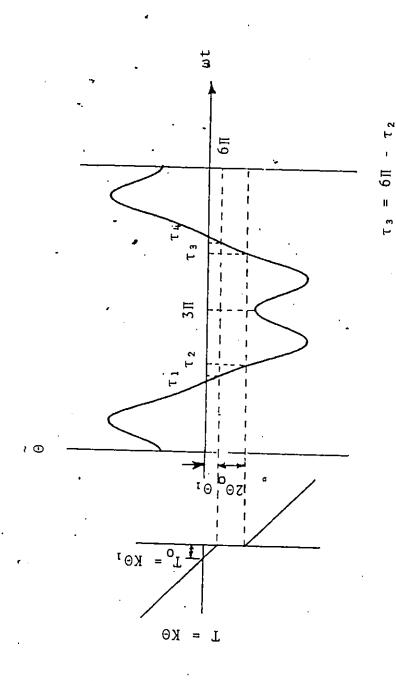
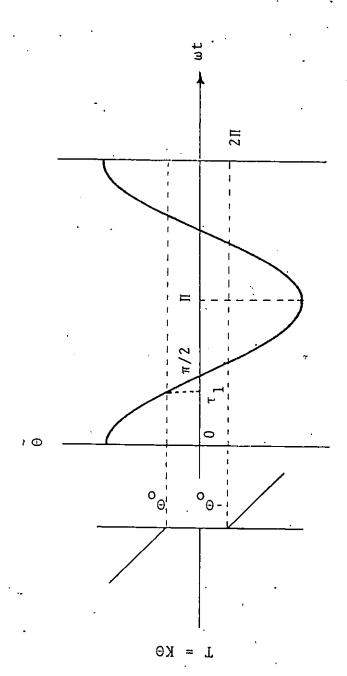


Figure 3.1.2.1

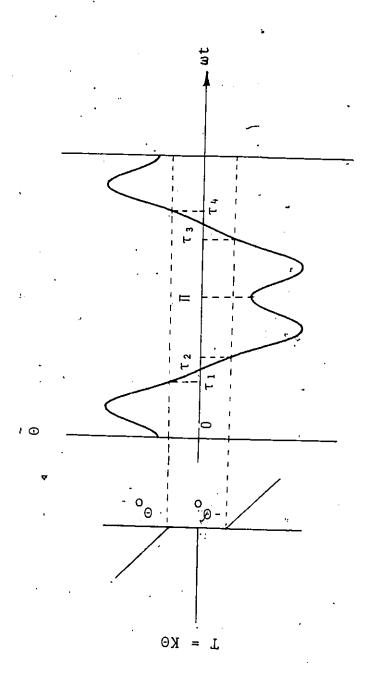
Amplitude Versus Angular Displacement Plot (Asymmetrical System), 3rd Order Ultraharmonic Resonance.



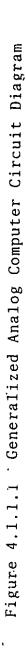
Amplitude Versus Angular Displacement Plot (Asymmetrical System). 1/3rd Order Subharmonic Resonance. Figure 3.1.3.1.

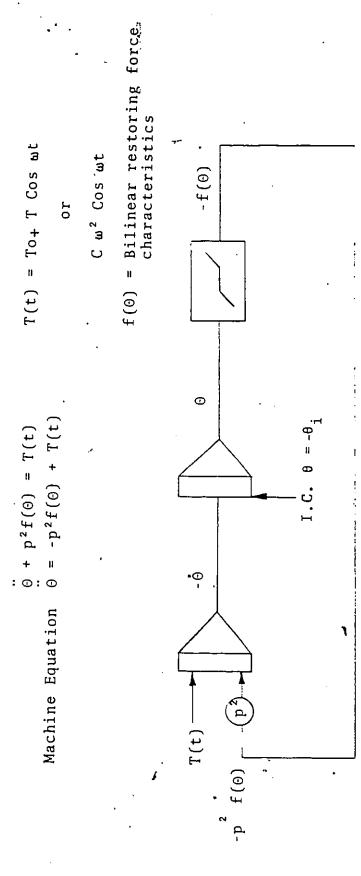


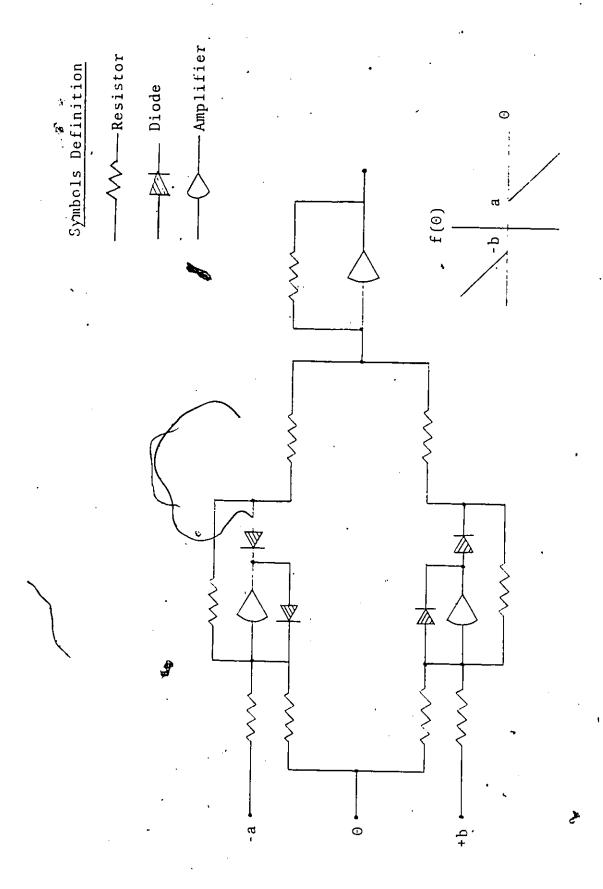
Amplitude Versus Angular Displacement Plot (Symmetrical System). Harmonic Resonance. Figure 3.2.1.2



Amplitude Versus Angular Displacement Plot (Symmetrical System). 3rd Order Ultraharmonic Résonance. Figure 3.2.2







Circuit Diagram for the Bilinear Restoring Force Characteristics Figure 4.1.1.2

Equation of undamped free vibrations is given by --

$$\ddot{2} + \omega^2 \ Z = 0$$

Solution of above equation is given by -

$$Z = A Cos \omega t + B Sin \omega t$$

$$=$$
  $Z_0$  Cos  $\omega$ t

73

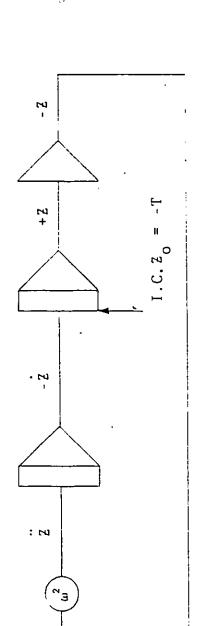


Figure 4.1.1.3 Generation of T Cos wt

$$\ddot{z} + \omega^2 z = 0$$

= 
$$-\omega^2 Z$$
 Machine Equation

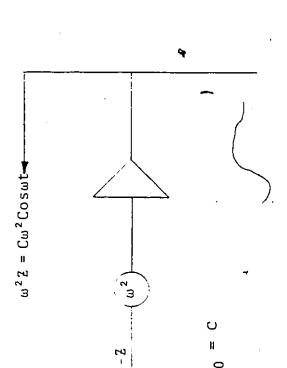
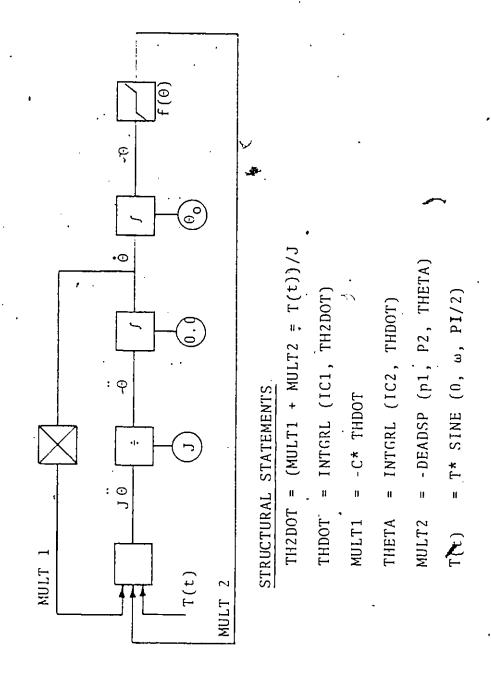
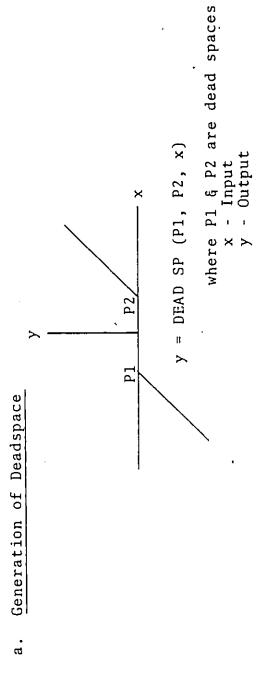


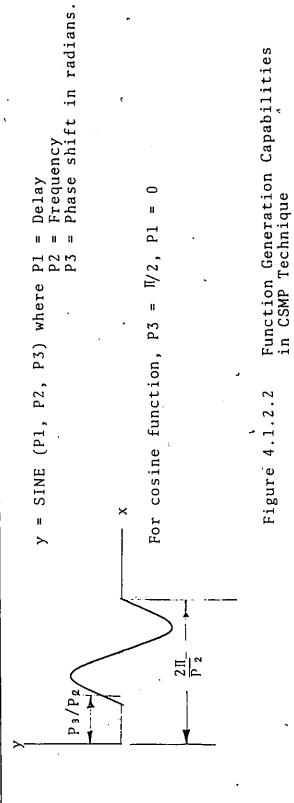
Figure 4.1.1.4 Generation of Cw² Coswt



Block Diagram and Structural Statements for a Bilinear System Figure 4.1.2.1



b. Generation of Sinusoidal Function



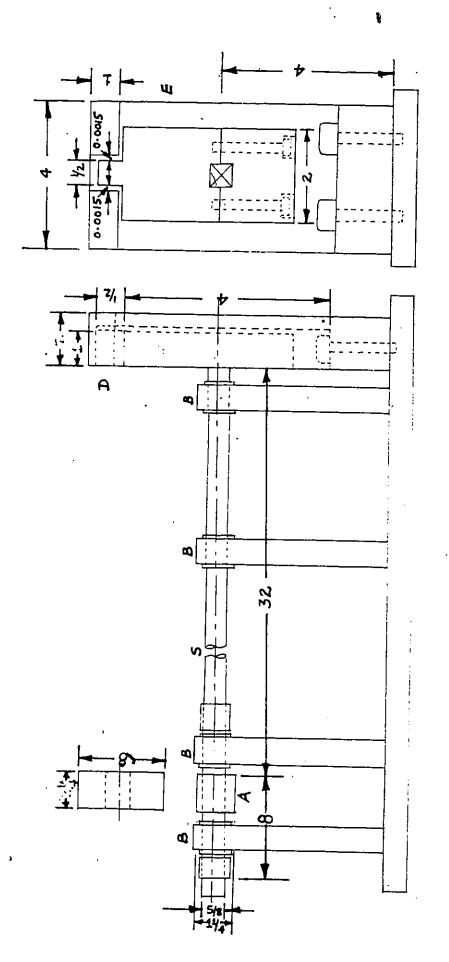


Figure 4.2.1.1 Design Details of the Mechanical Model

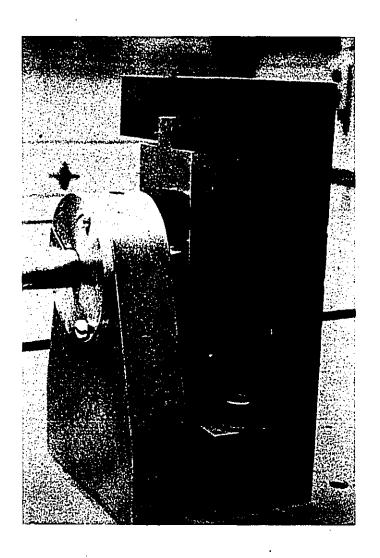


Figure 4.2.1.2 Tooth Block Assembly

0.

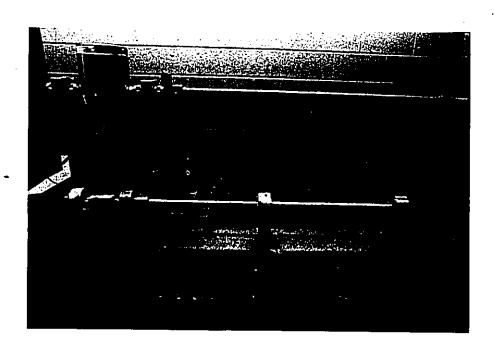


Figure 4.2.1.3 Overall View of the Experimental Set Up

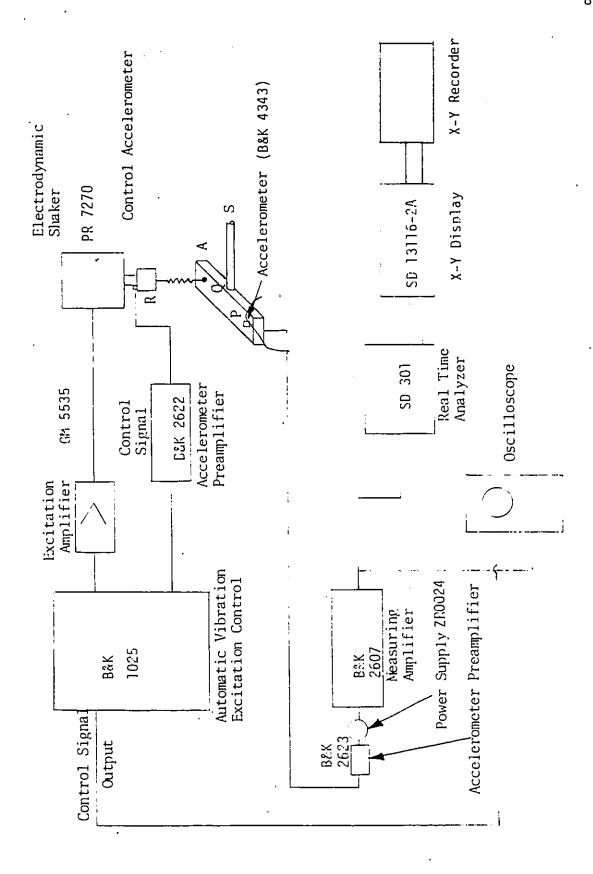


Figure 4.2.2.1 Schematic Diagram of Experimental Sct Up

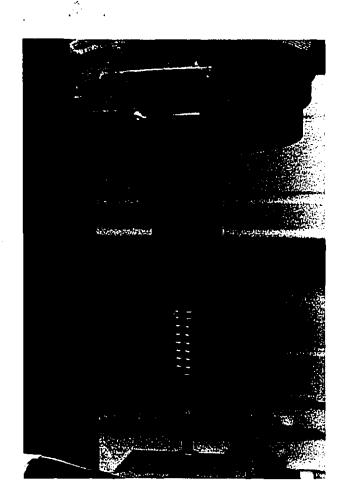
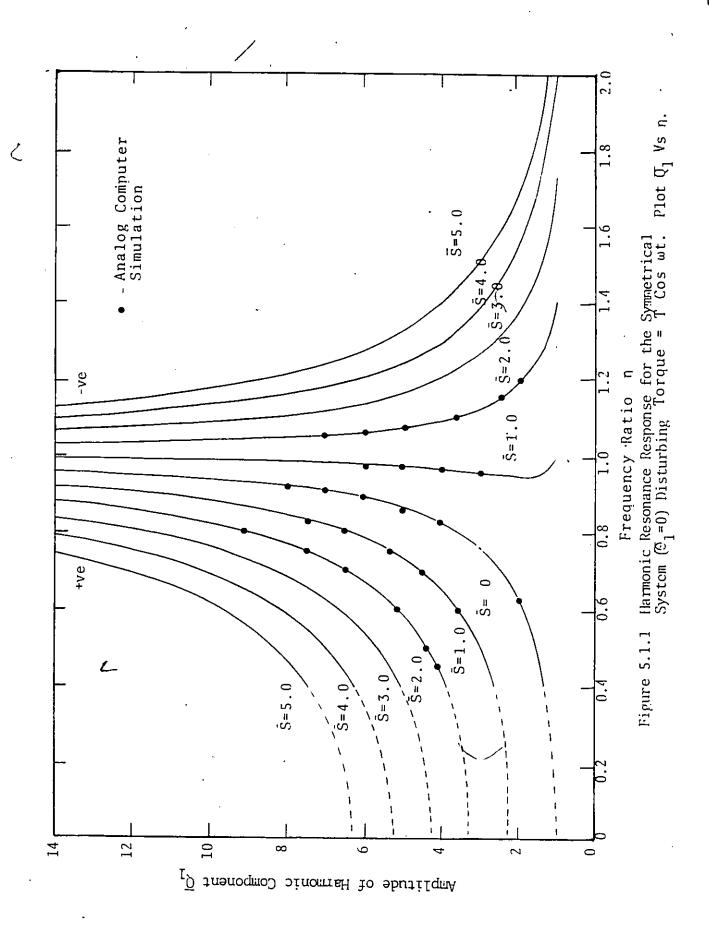
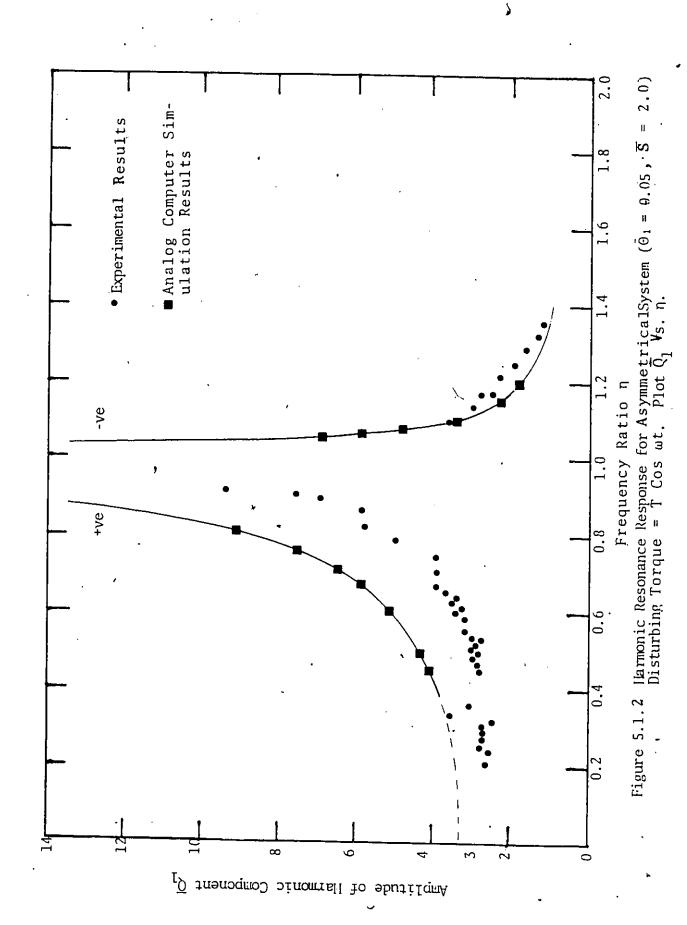
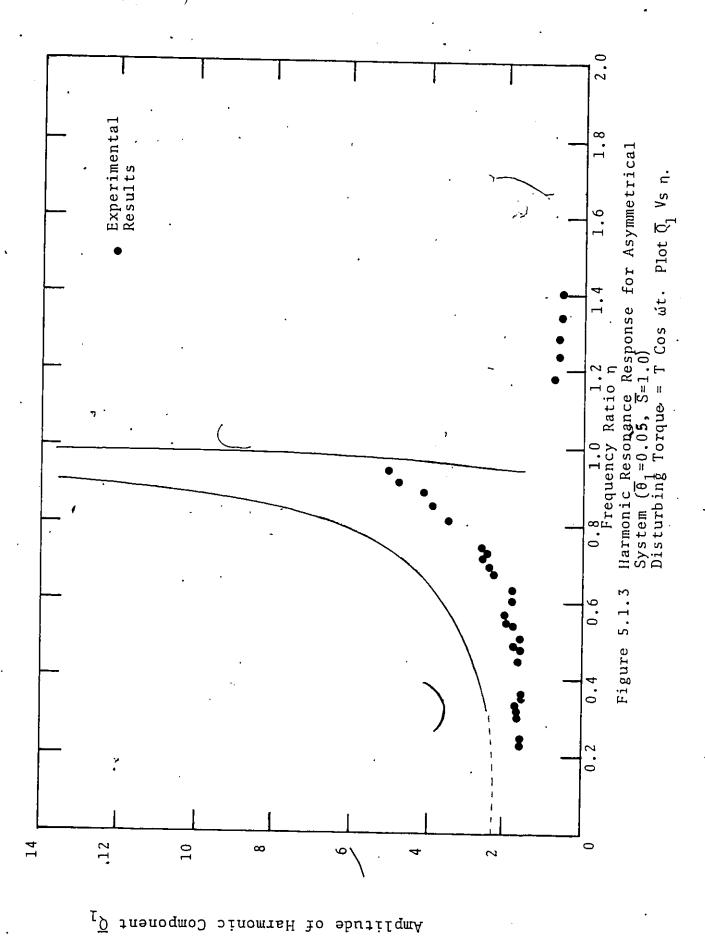


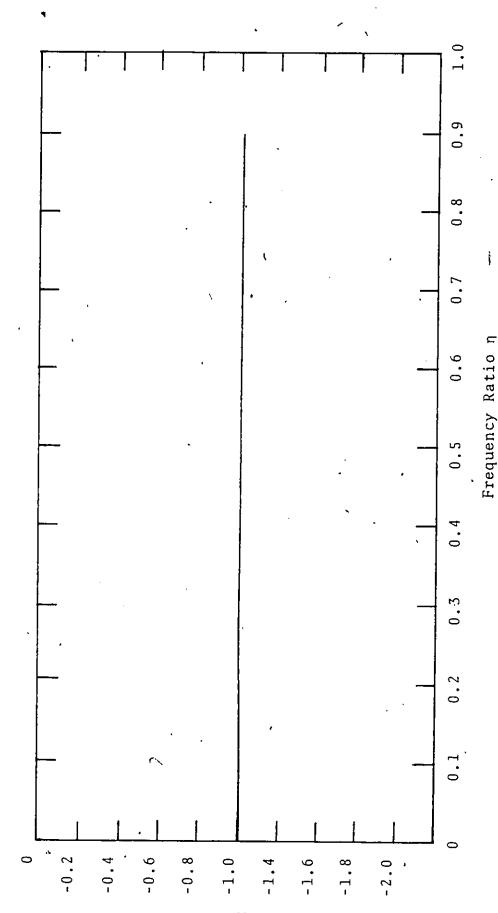
Figure 4.2.3.1 View Showing Shaker Pin
Attachment to the Vibrating
Arm 'A' and Control Accelerometer



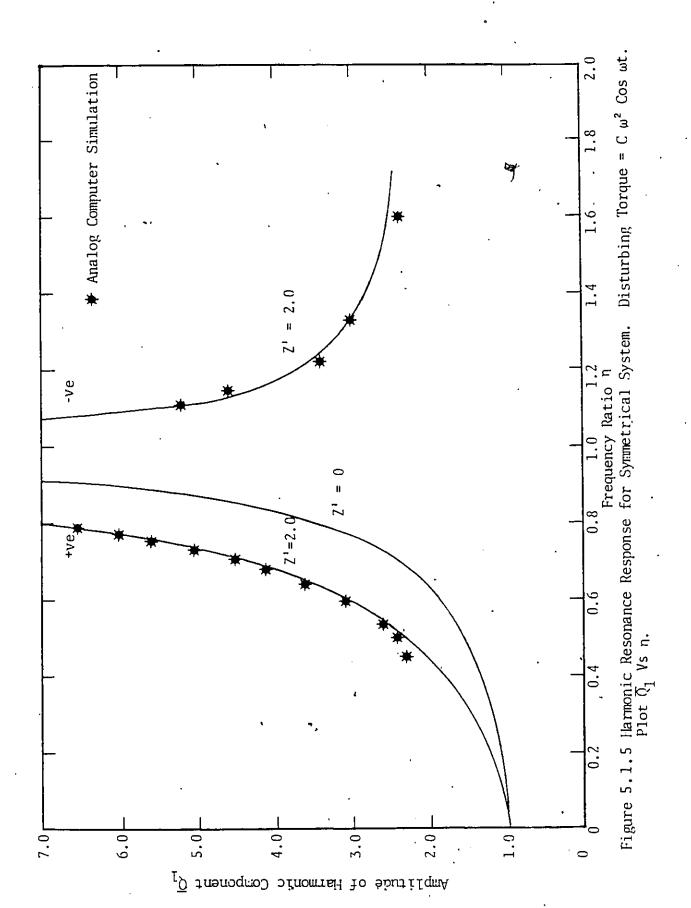
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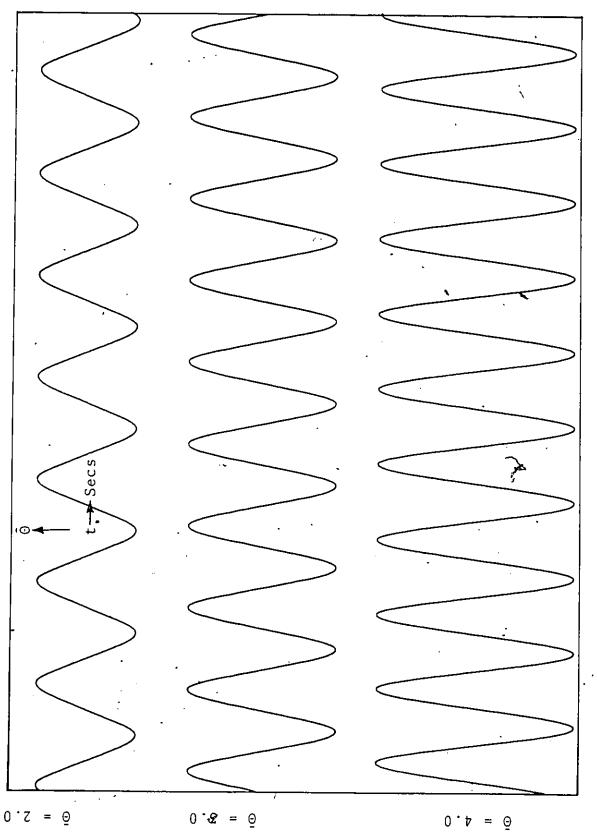




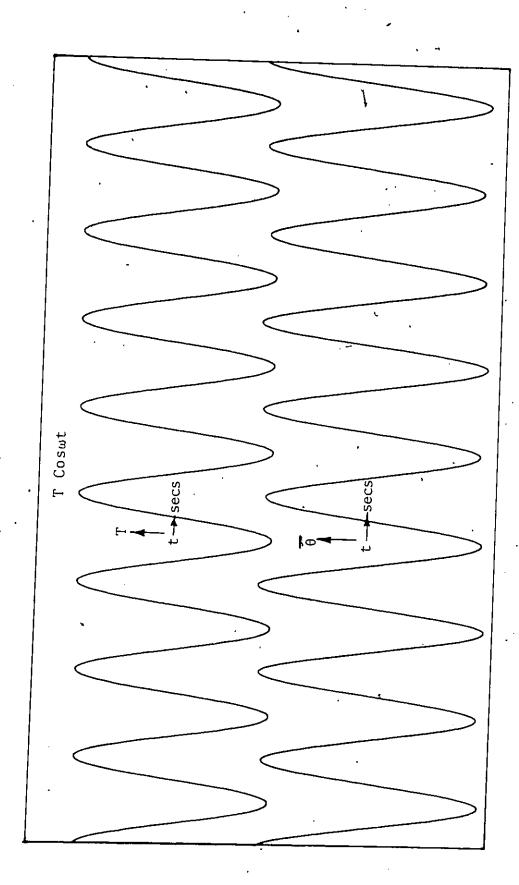
Harmonic Resonance Response for Asymmetrical System ( $\overline{\theta}_1$ =0.05,  $\overline{S}$ =1.0) Disturbing Torque = T Cos  $\omega t$ , Plot  $\overline{M}$  Vs n. Figure 5.1.4



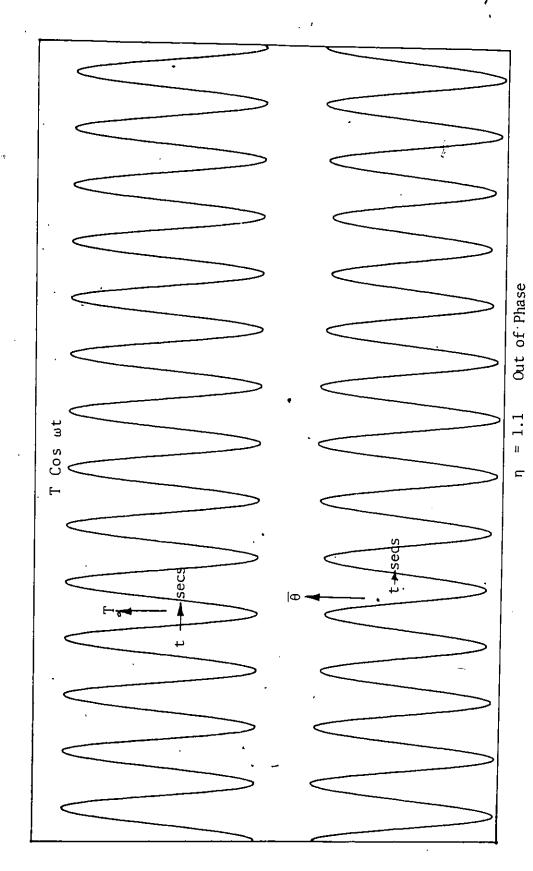
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Analog Computer Output, Free Vibration Response  $(\overline{S}{=}0)$  Disturbing Torque = T Cos $\omega t$ Figure 5.1.6



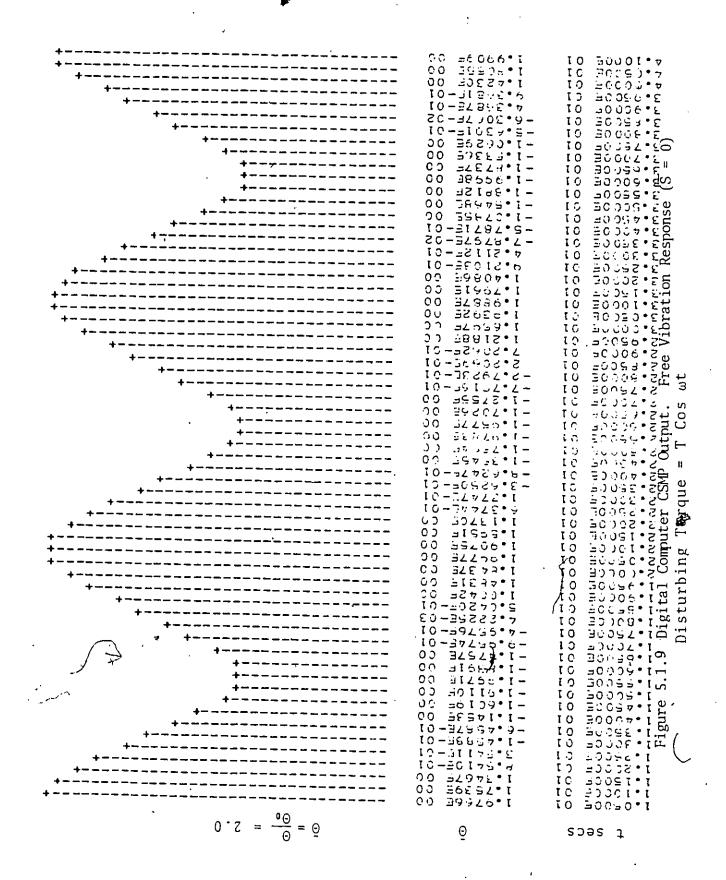
Analog Computer Output, Forced Vibration Response (n=0.7) In-Phase Motion— Disturbing Torque = T  $Cos\omega t$ ,  $\bar{S}$  = 2.0 Figure 5.1.7

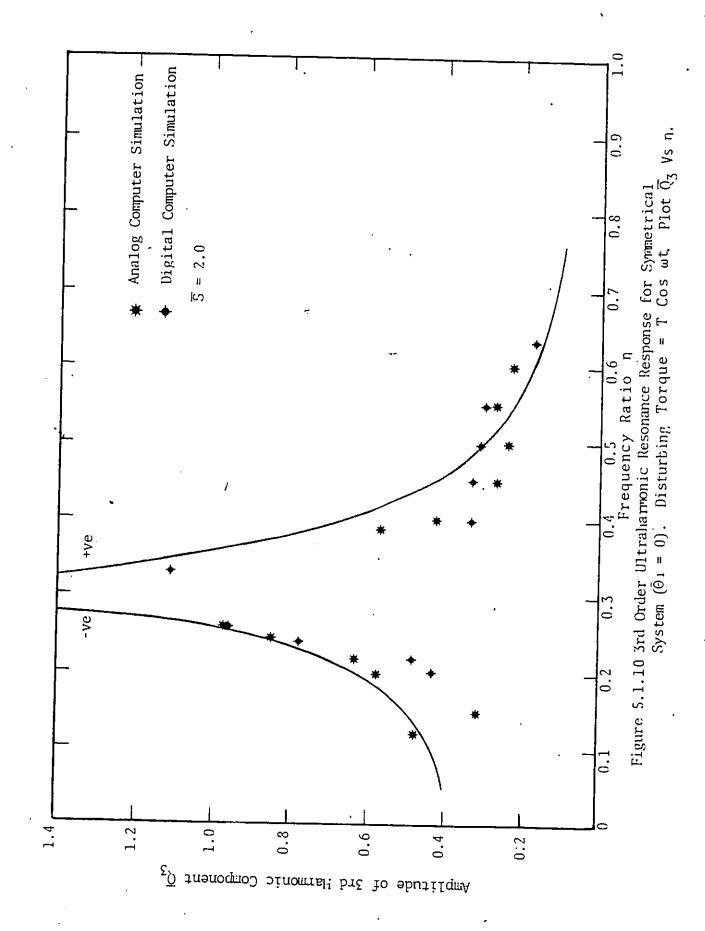


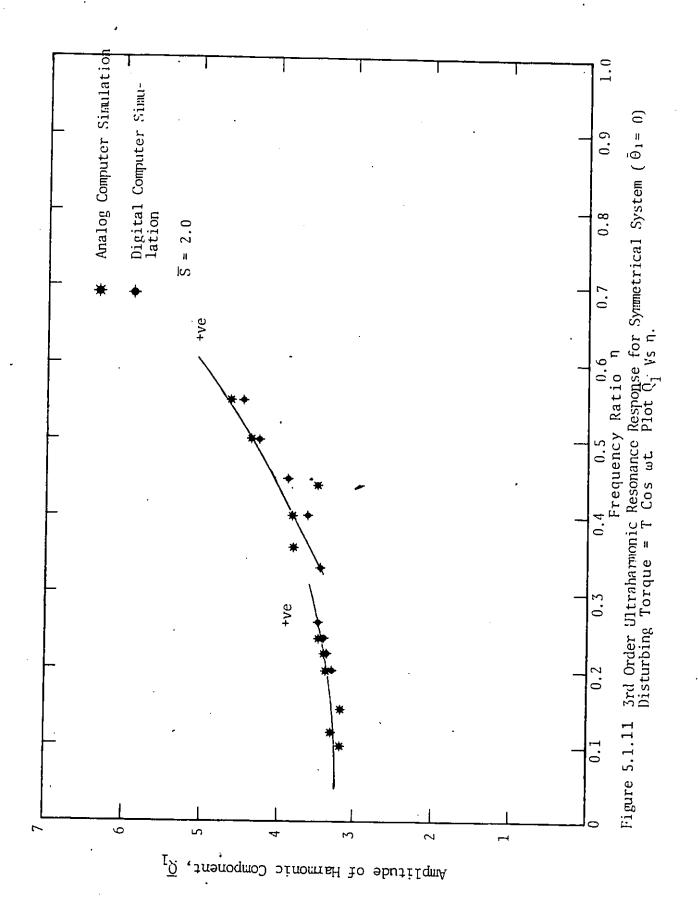
Analog Computer Output, Forced Vibration Response ( n= 1.1) Out of Phase Motion Disturbing Torque = T Cos  $\omega t$ ,  $\hat{S}$  = 2.0 Figure 5.1.8

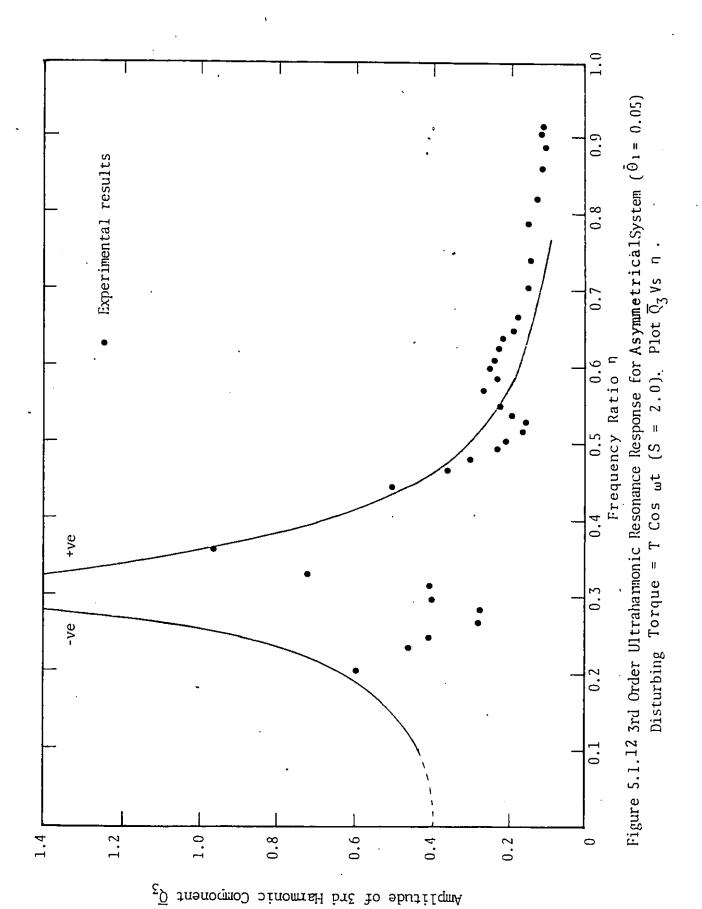
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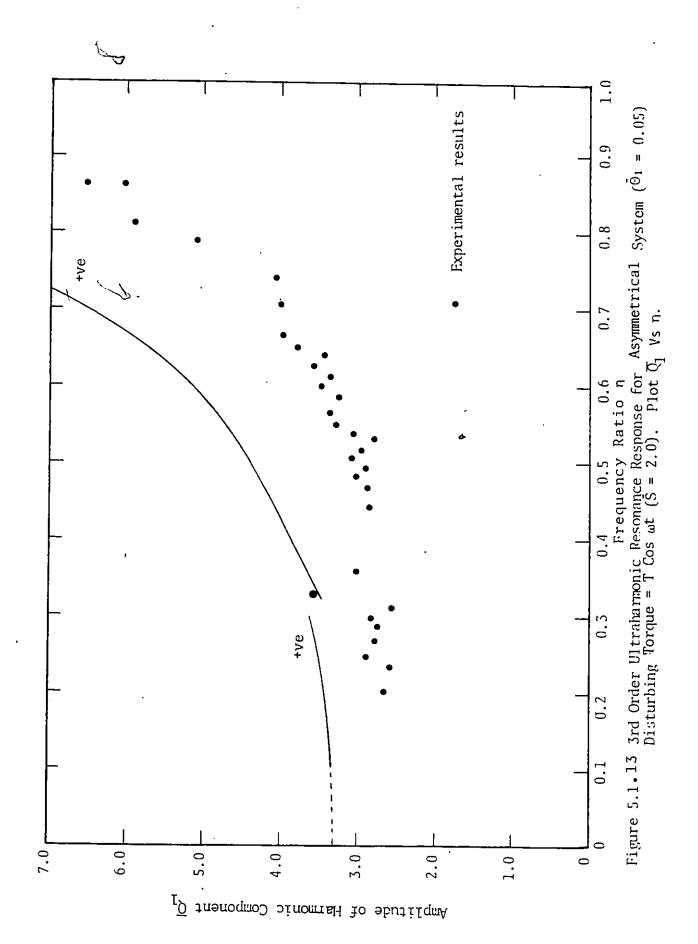
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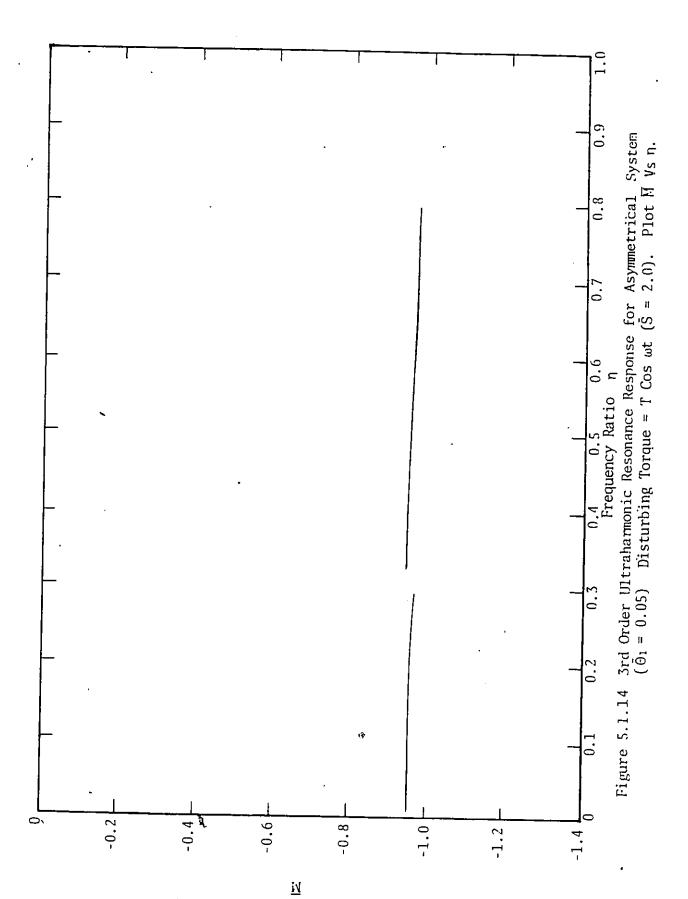


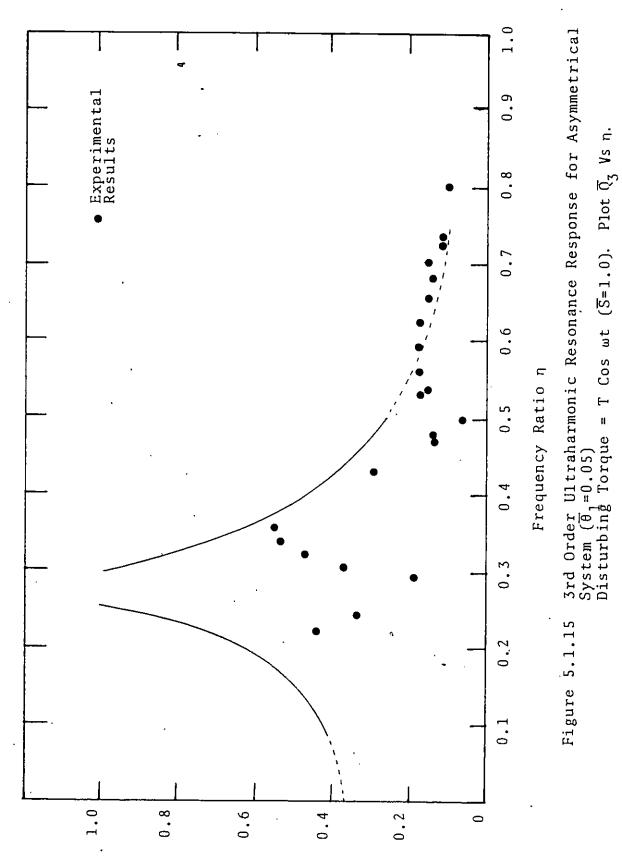




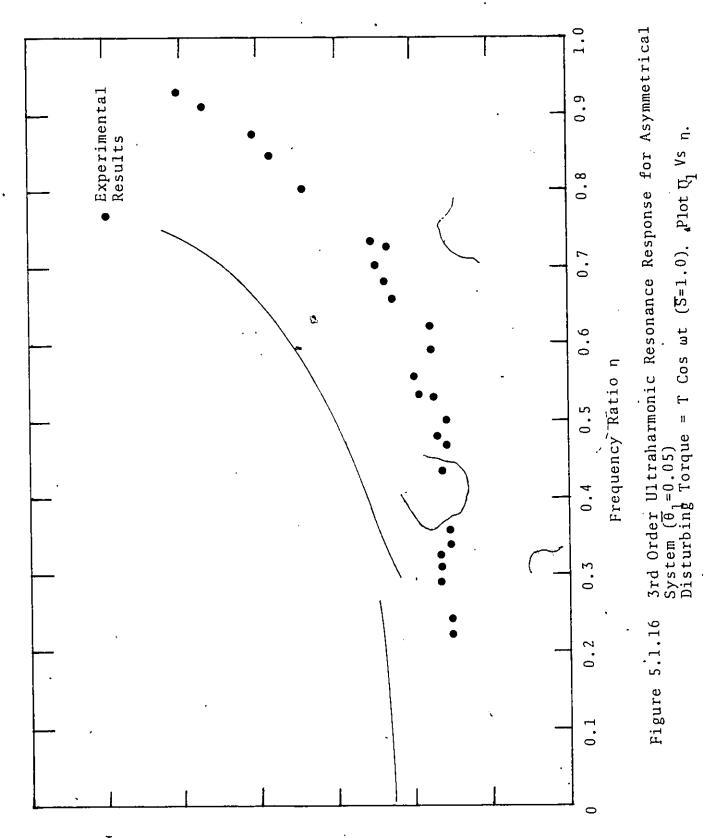




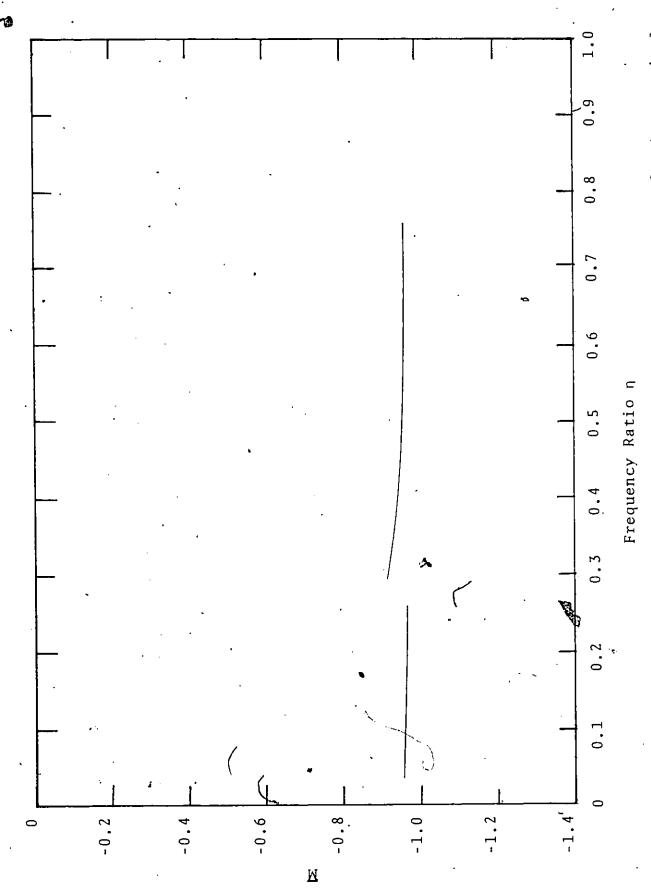




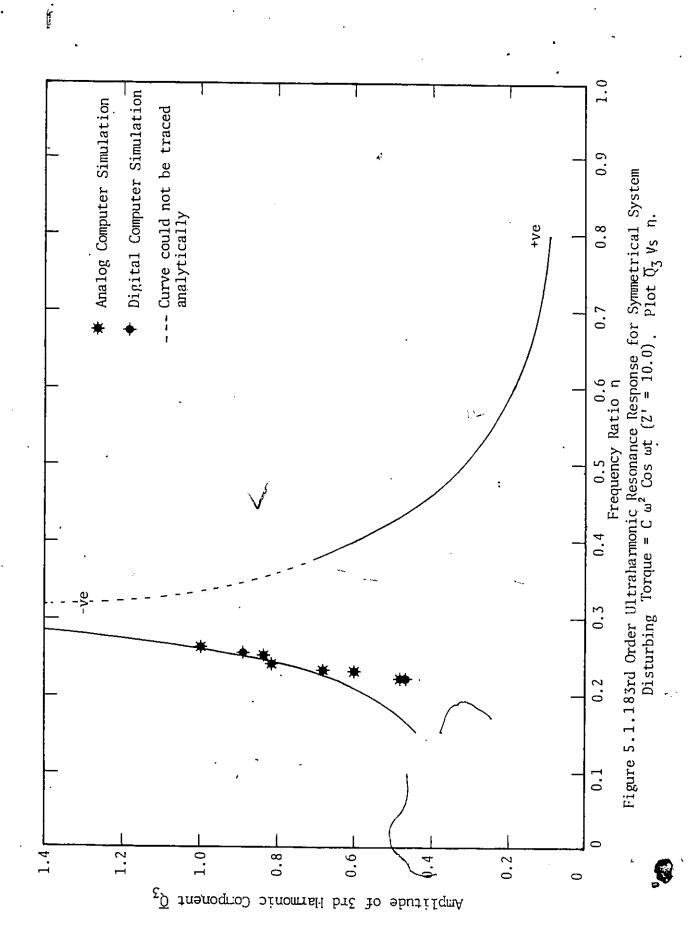
Amplitude of 3rd Harmonic Component  $\overline{Q}_3$ 

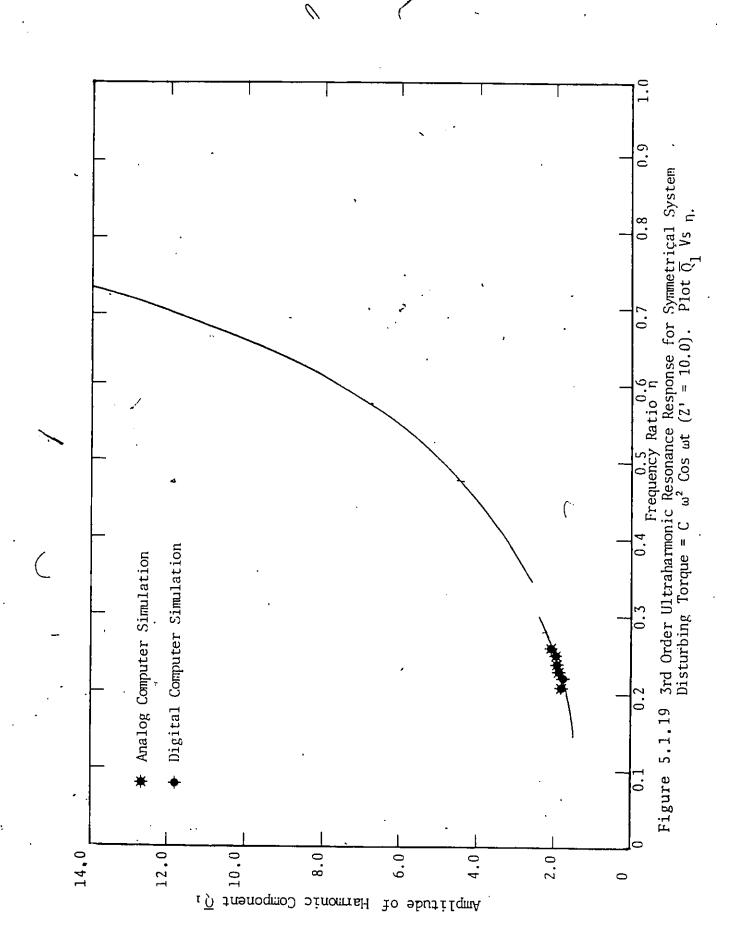


Amplitude of Harmonic Component  $\overline{Q}_{1}$ 



3rd Order Ultraharmonic Resonance Response for Asymmetrical System ( $\overline{\theta}_1$ =0.05) Disturbing Torque = T Cos  $\omega t$  ( $\overline{S}$ =1.0). Plot  $\overline{M}$  Vs  $\eta$ . Figure 5.1.17





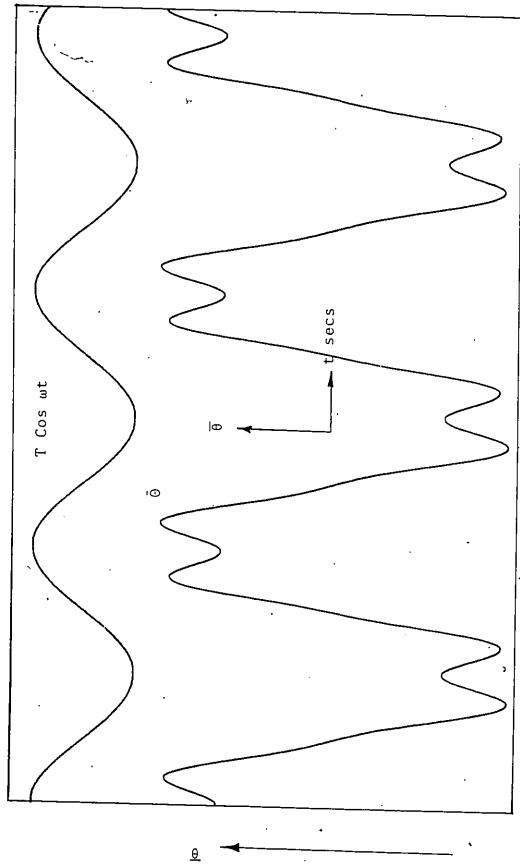
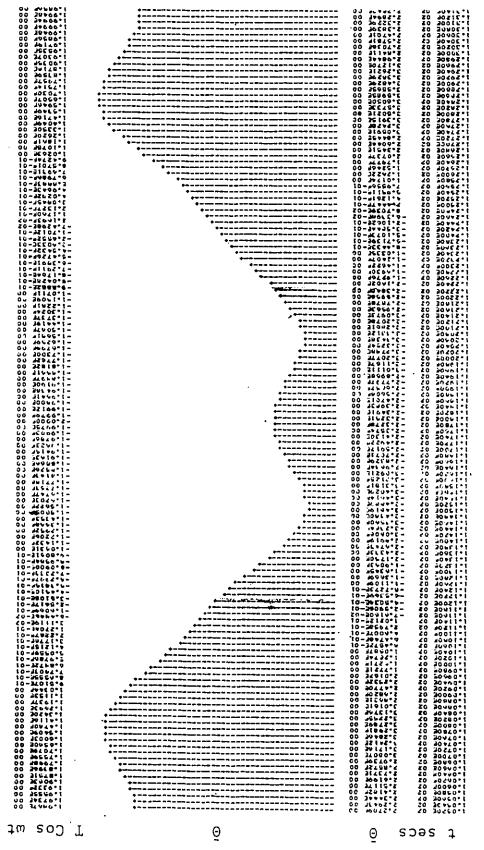


Figure .5.1.20 Analog Computer Output. 3rd Order Ultraharmoffic Resonance Response ( n=0.24) Disturbing Torque = T Cos  $\omega t$ ,  $\tilde{S}$  = 2.0



3rd Order Ultraharmonic ŝ) ωt Output Digital Computer CSMP Response ( $\eta = 0.24$ ) H Torque Disturbing

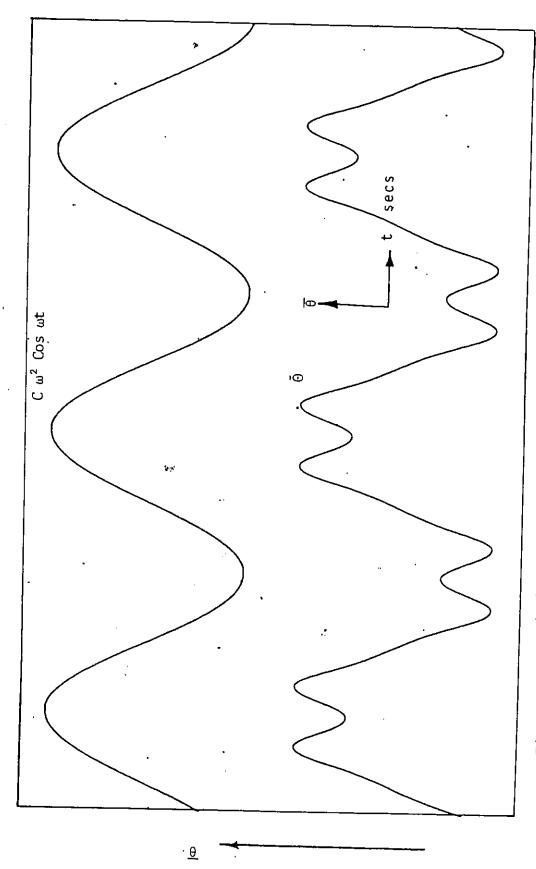
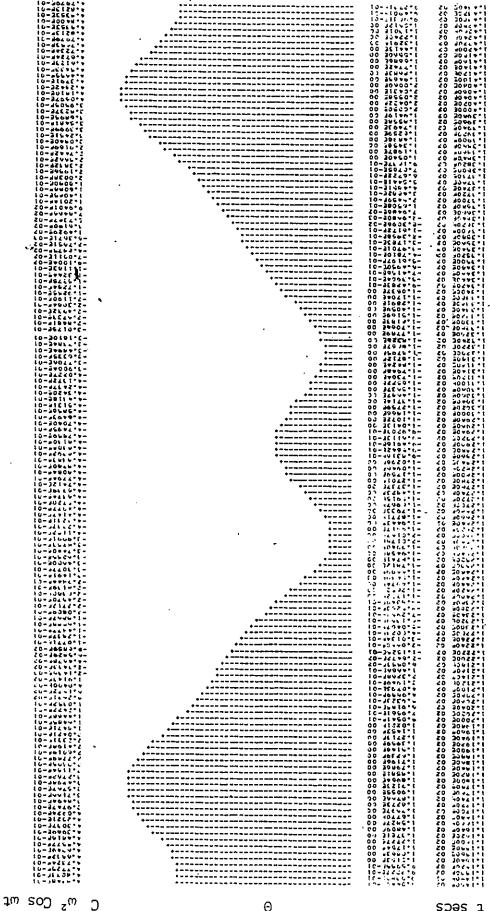


Figure 5.1.22 Analog Computer Output. 3rd Order Ultraharmonic Resonance Response for Symmetrical System ( $\Theta_1$  = 0,  $\eta$  = 0.22)  $\langle f_{\tau} \rangle$  Disturbing Torque = C  $\omega^2$  Cos  $\omega t$  ( $Z^1$  = 10.0)

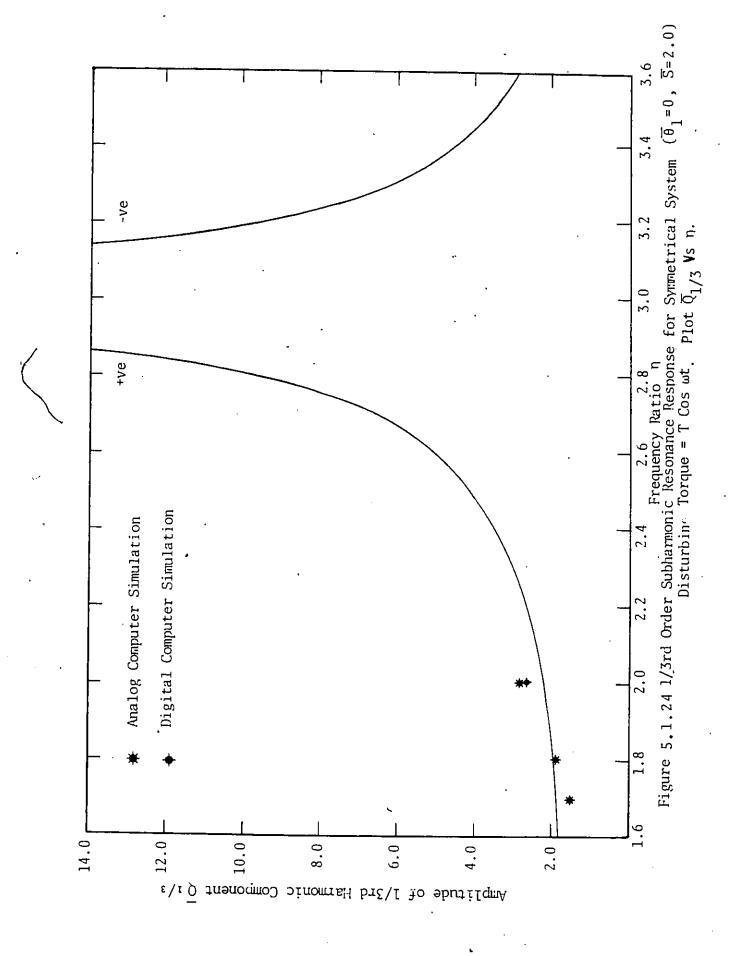
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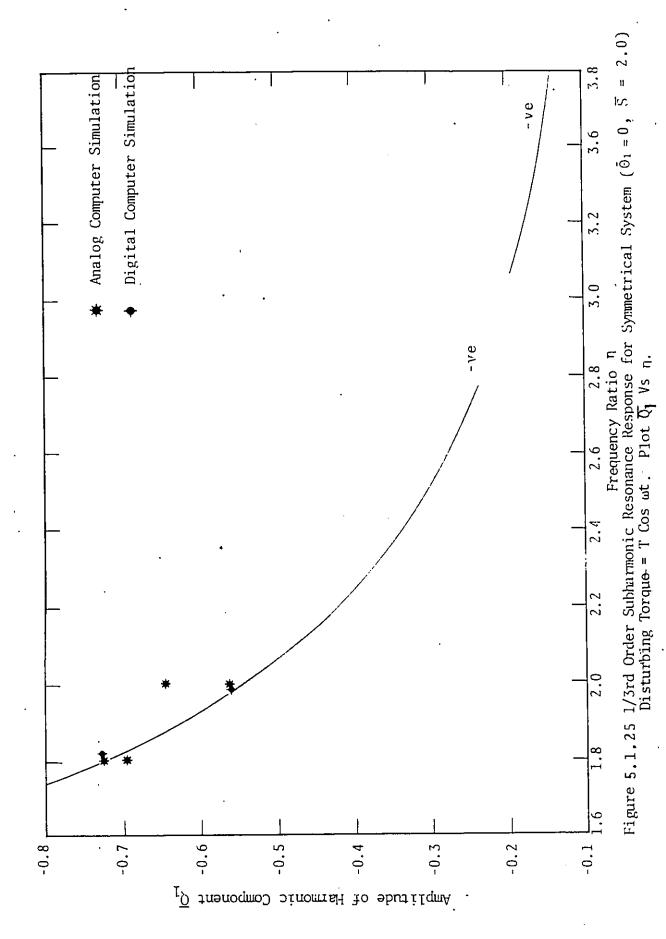
Resonance 3rd Order Ultraharmonic 0.22 = 10.0.2) = L t 0 Cos System ( $ue = C \omega^2$ Digital Computer CSMP for Symmetrical System Torque sturbing <u>.</u> Φ

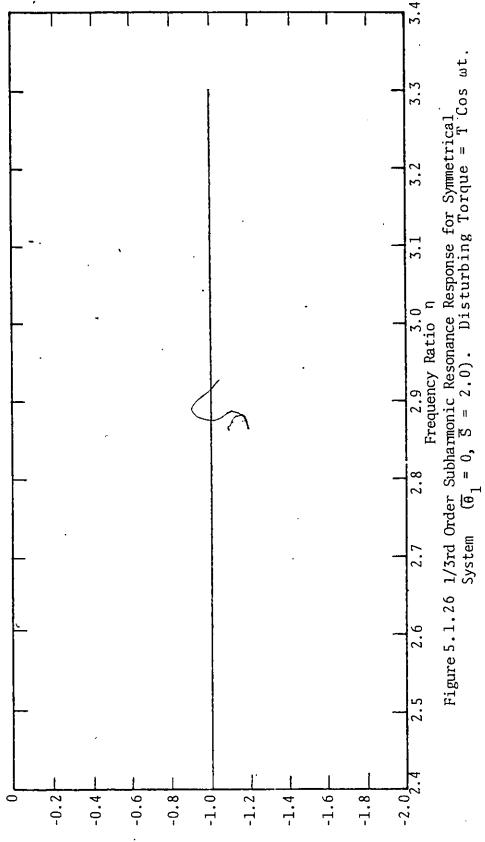
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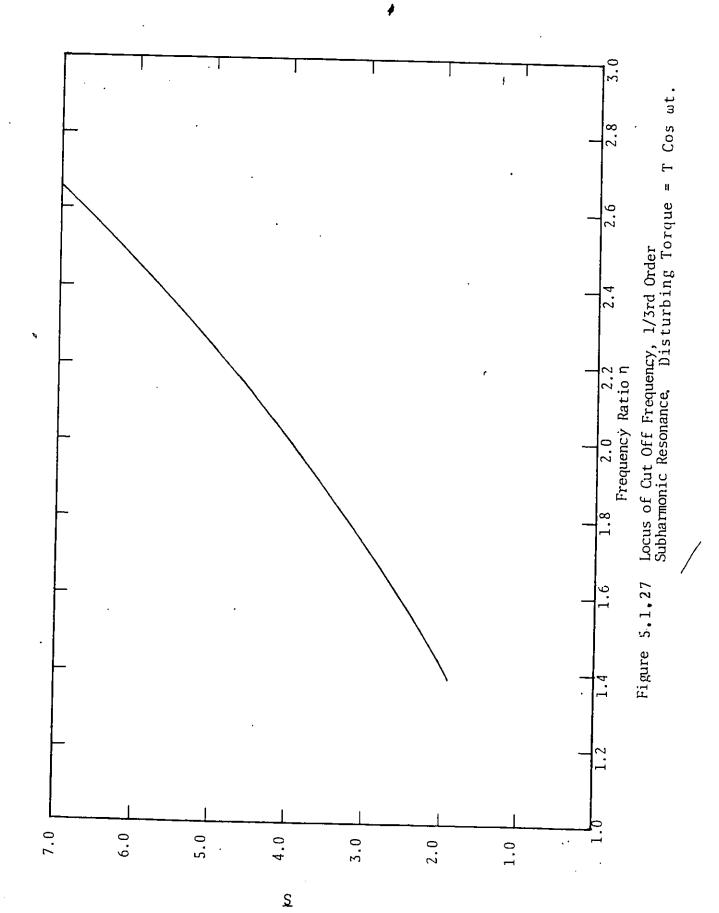


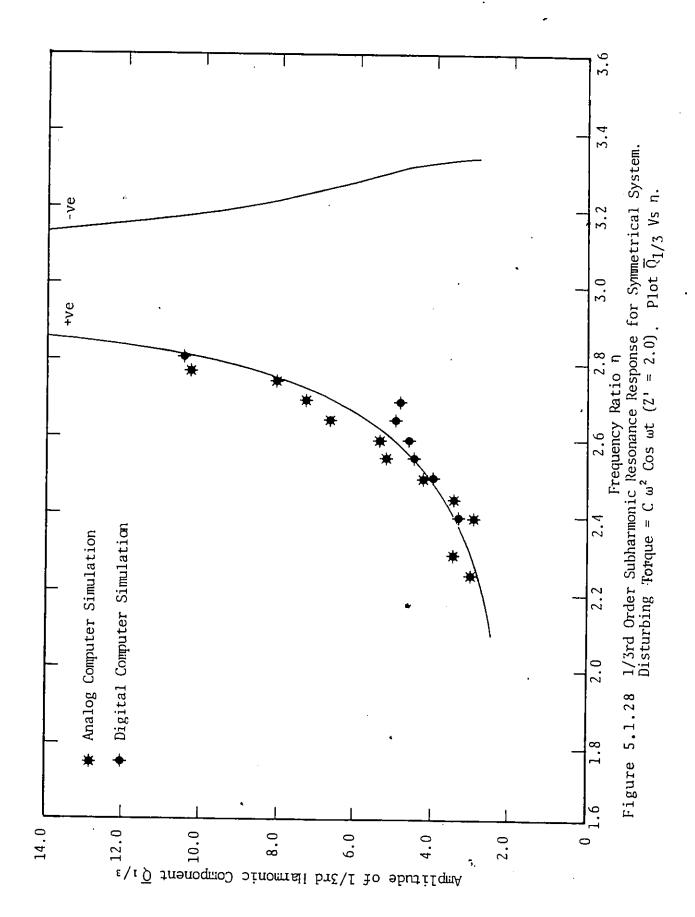
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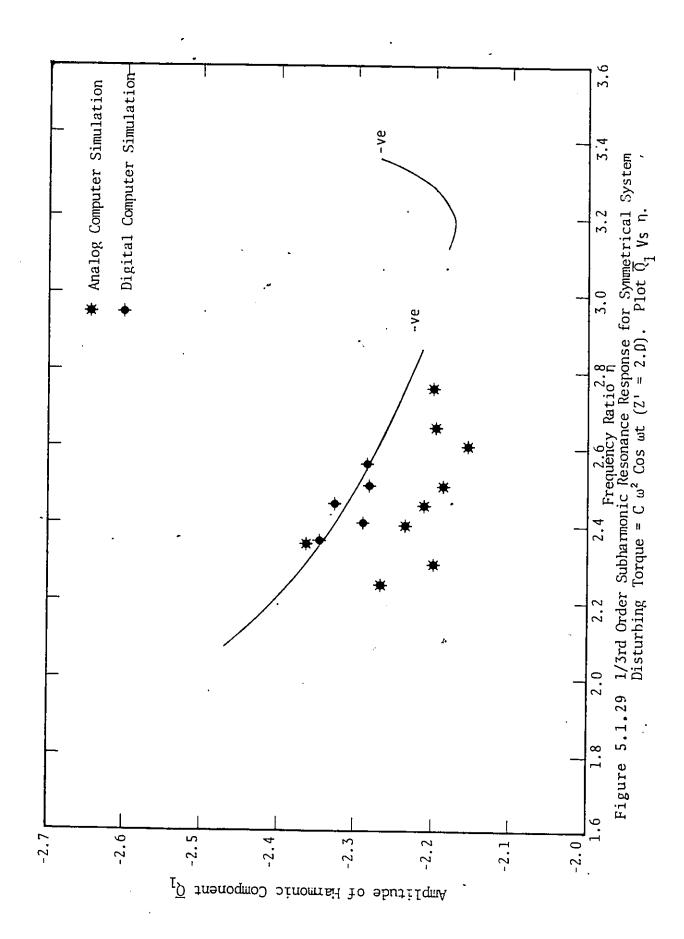


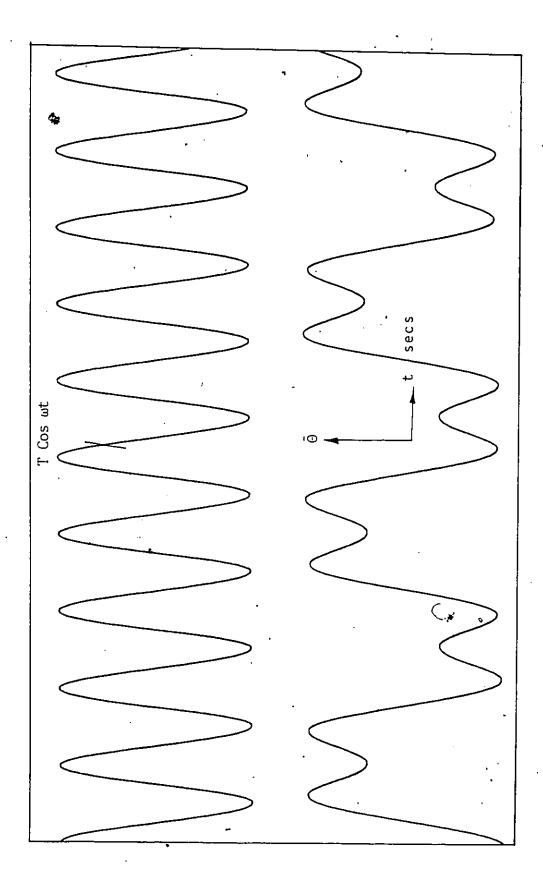


Plot M Vs n.

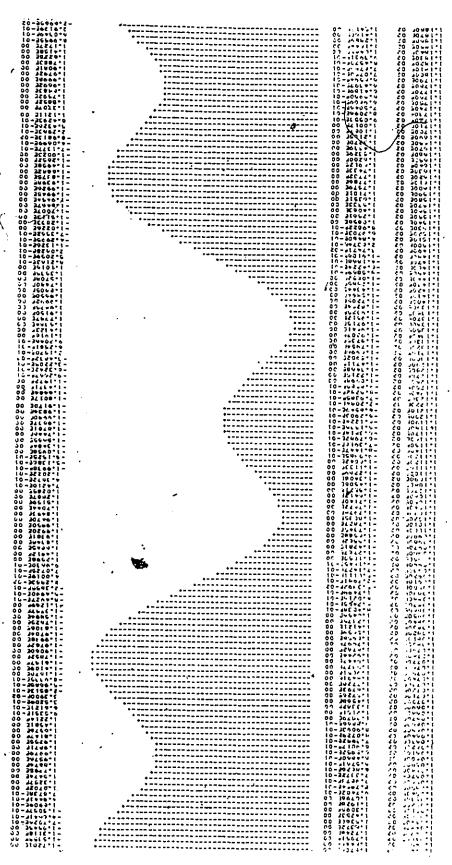








Analog Computer Output. 1/3rd Order Subharmonic Resonance Response for Symmetrical System ( $\bar{\Theta}_1$  = 0,  $\eta$  = 1.6) Disturbing Torque = T Cos  $\omega t$  ( $\bar{S}$  = 2.0) Figure 5.1.30

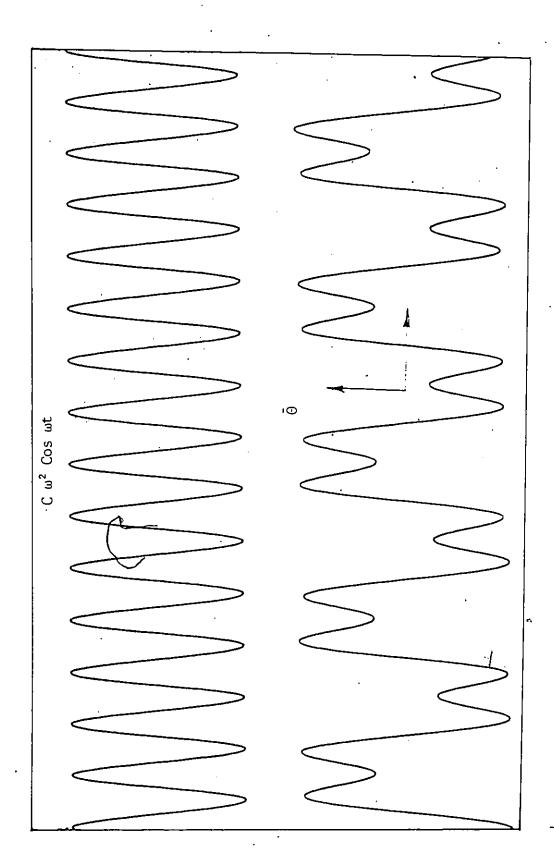


Response for Symmetrical System Disturbing Torque = T Cos wt (§. Digital Computer CSMP Output.

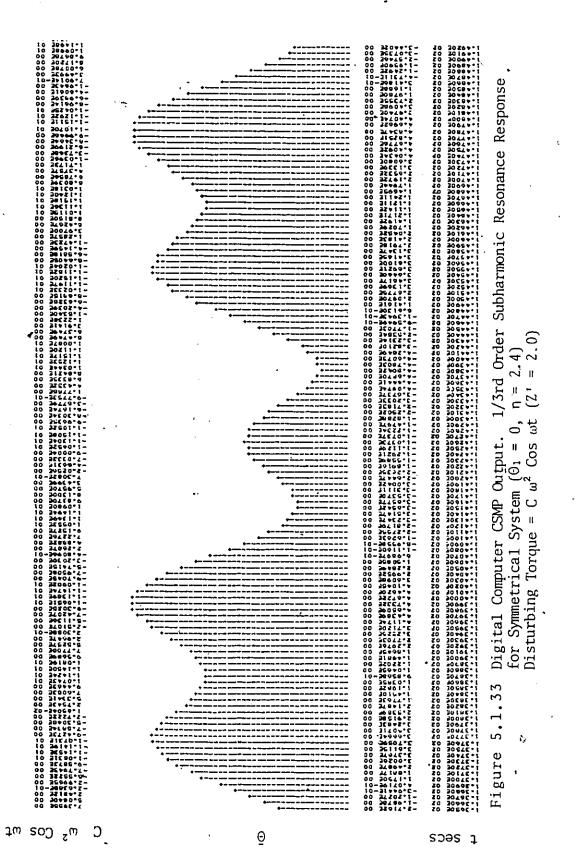
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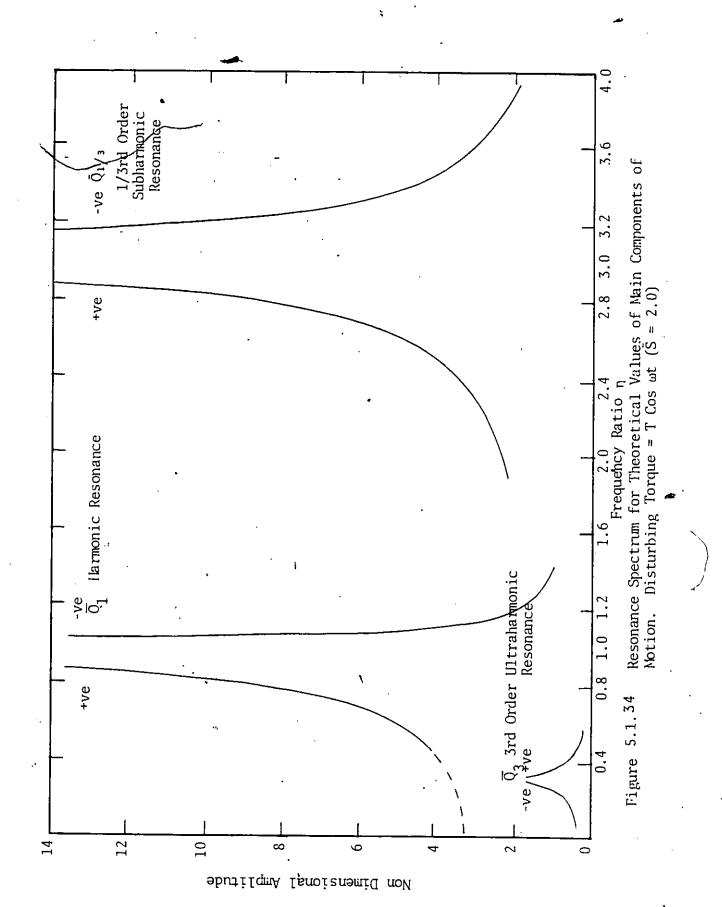
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Analog Computer Output 1/3rd Order Subharmonic Resonancé Response for Symmetrical System ( $\hat{\Theta}_1$  = 0, n= 2.4) Disturbing Torque = C  $\omega^2$  Cos  $\omega t$  (Z' = 2.0) Figure 5.1.32





Shaker Pin
Displacement

Acceleration
Waveform at
Point 'P'

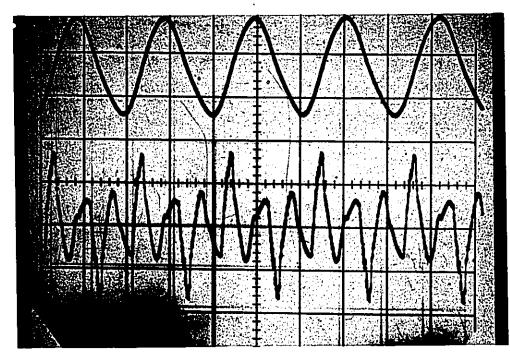


Figure 5.2.1 -Acceleration Waveform Display

$$f = 9.6 \text{ Hz}, p = 26.0 \text{ Hz},$$
  
 $\eta = 0.37, \overline{S} = 2.0$ 

Shaker Pin
Displacement

Acceleration
Waveform at
Point 'P'

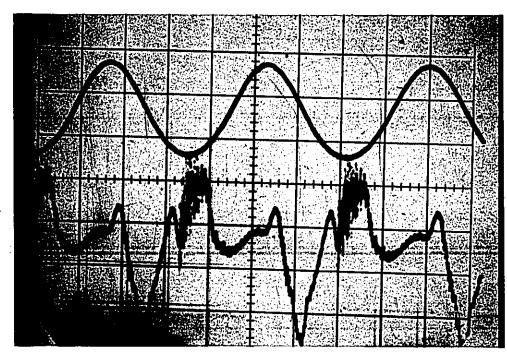


Figure 5.2.2 Acceleration Waveform Display f = 13.8 Hz, p = 26.0 Hz,

$$\eta = 0.53, 5 = 2.0$$

Shaker pin Displacement

Acceleration Waveform at Point 'P'

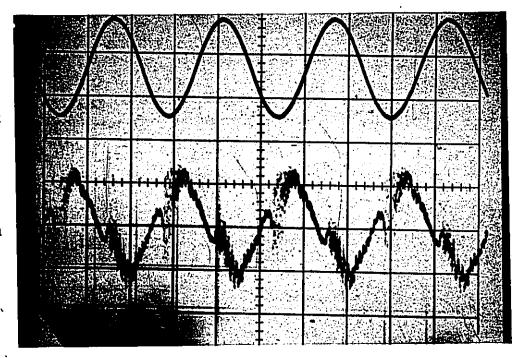


Figure 5.2.3 Acceleration Waveform Display

$$f = 19.9 Hz, p = 26.0 Hz,$$

$$\eta = 0.76$$
,  $\overline{S} = 2.0$ 

Shaker Pin Displacement

Acceleration Waveform at Point 'P'

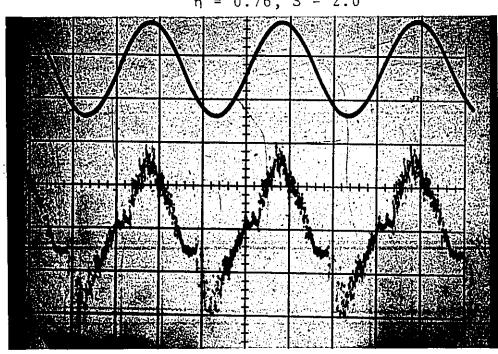
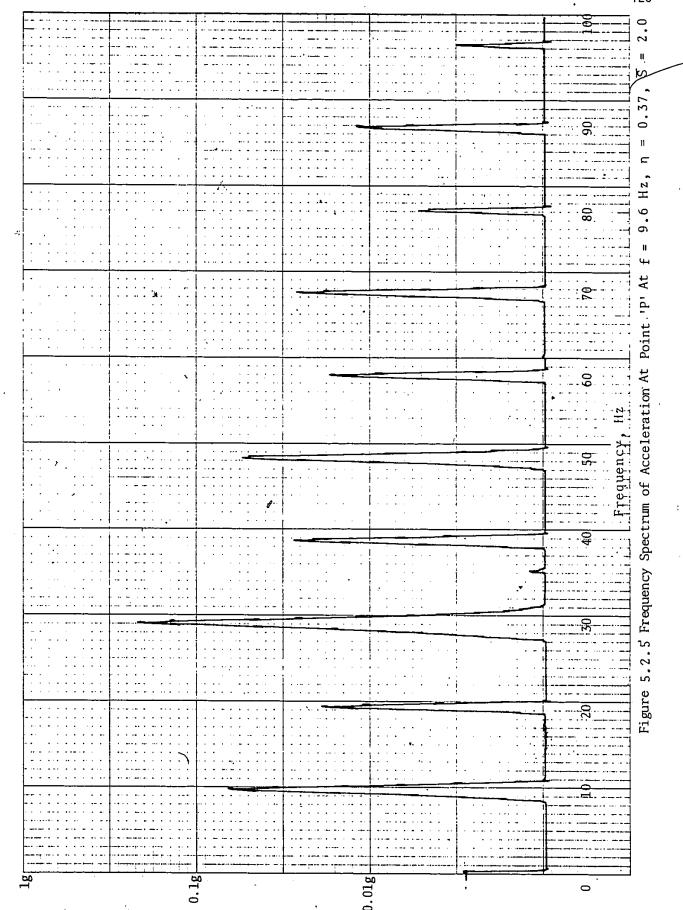
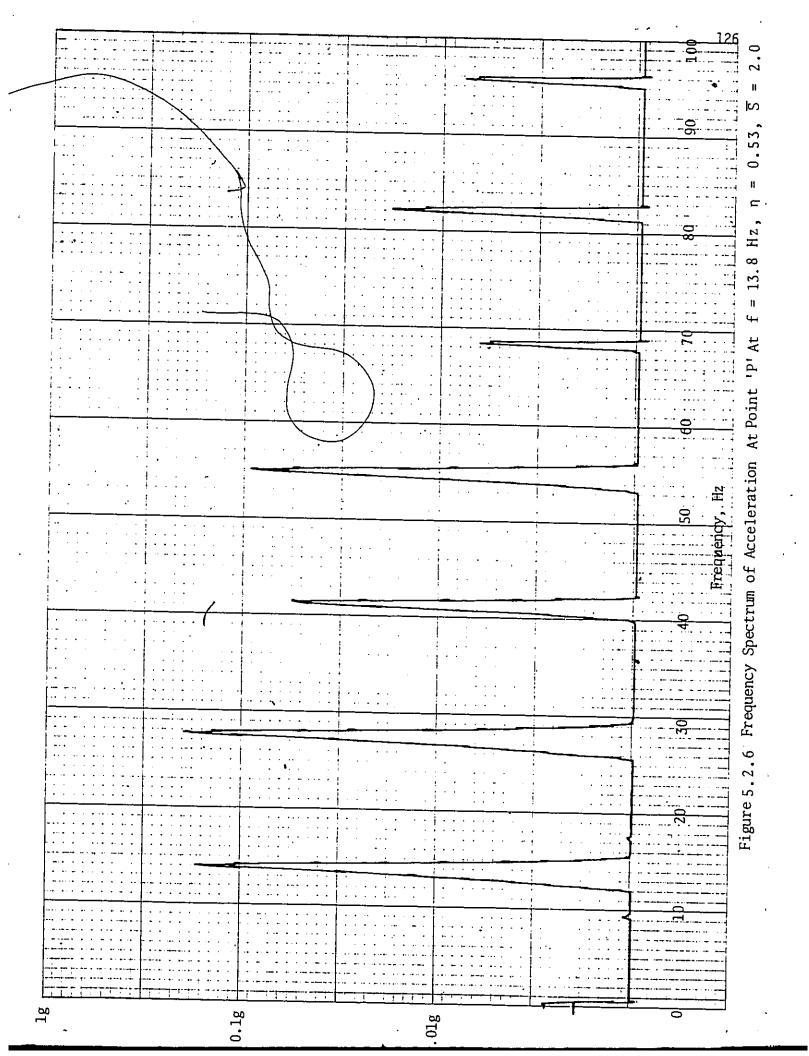


Figure 5.2.4 Acceleration Waveform Display

$$f = 33.3 Hz, p = 26.0 Hz,$$

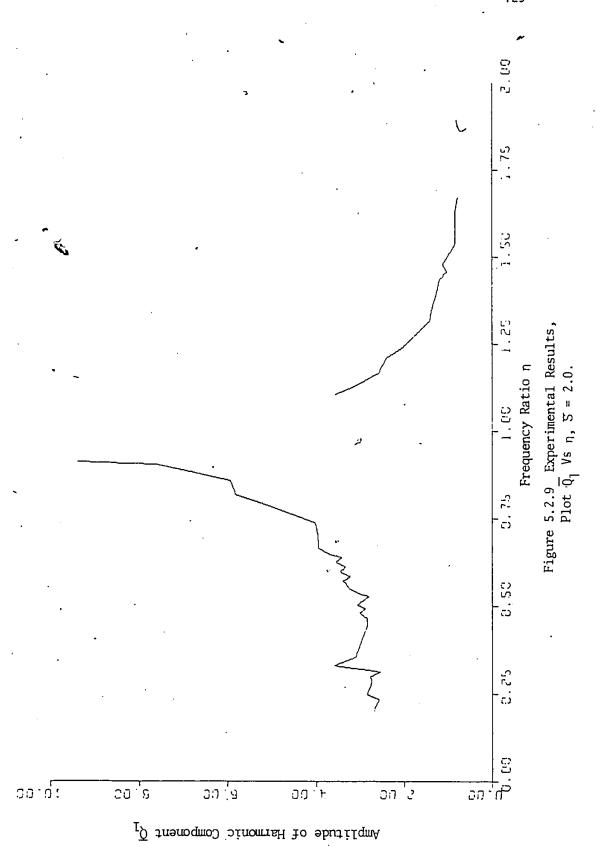
$$\eta = 1.28, \overline{S} = 2.0$$

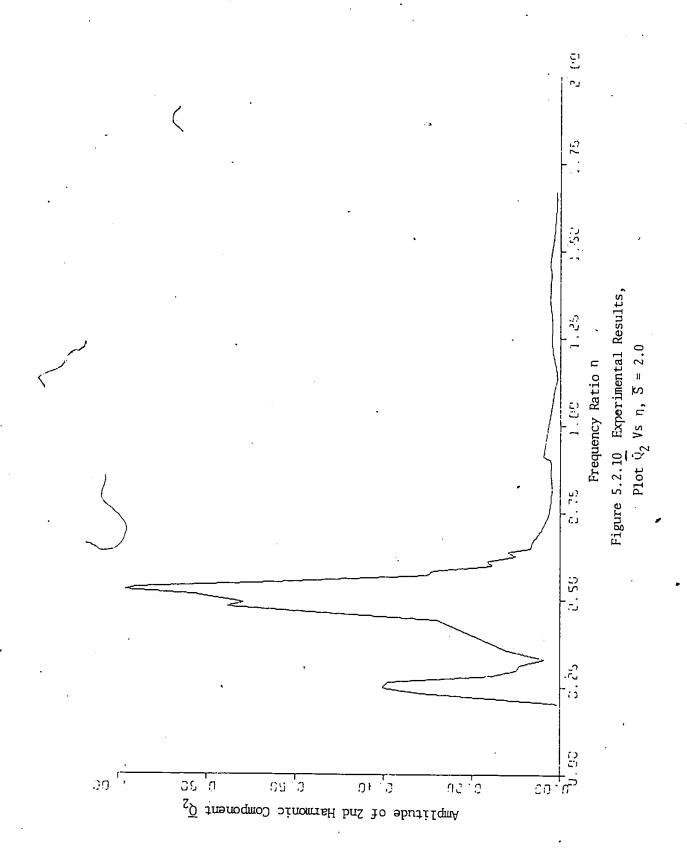


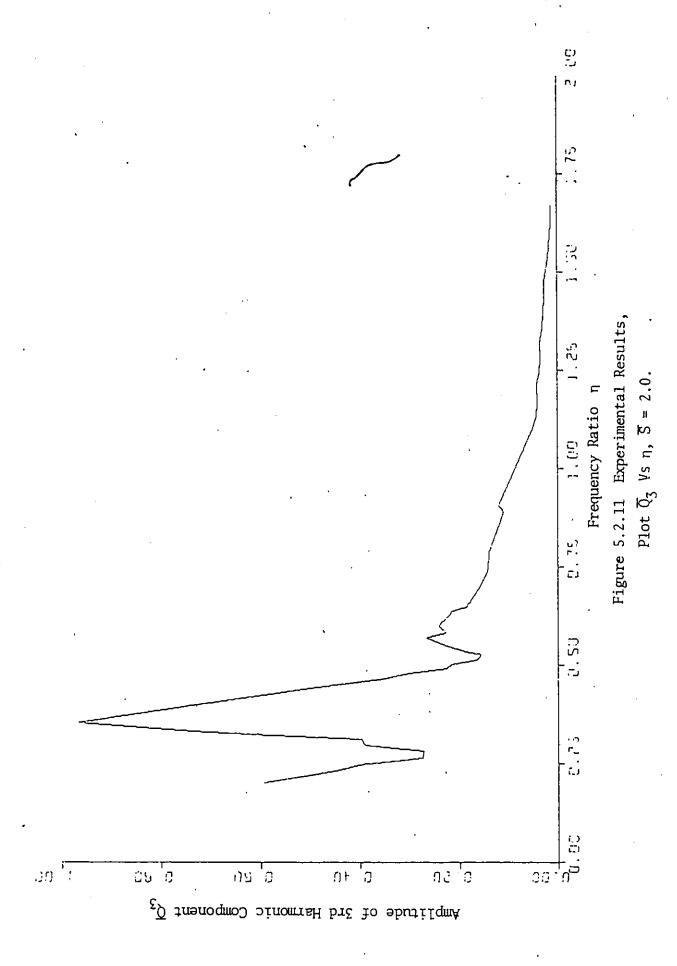


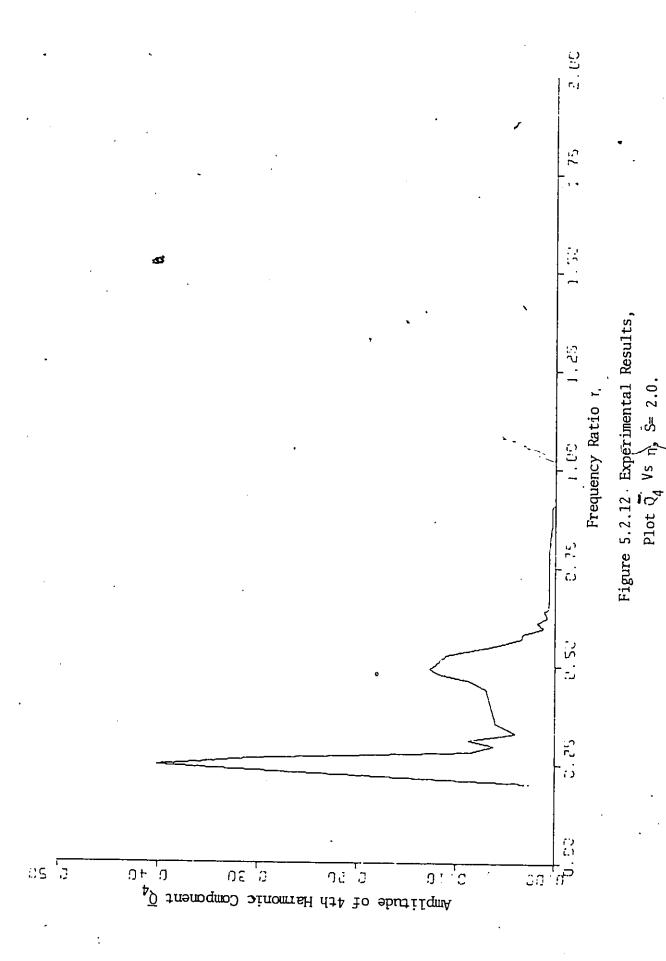
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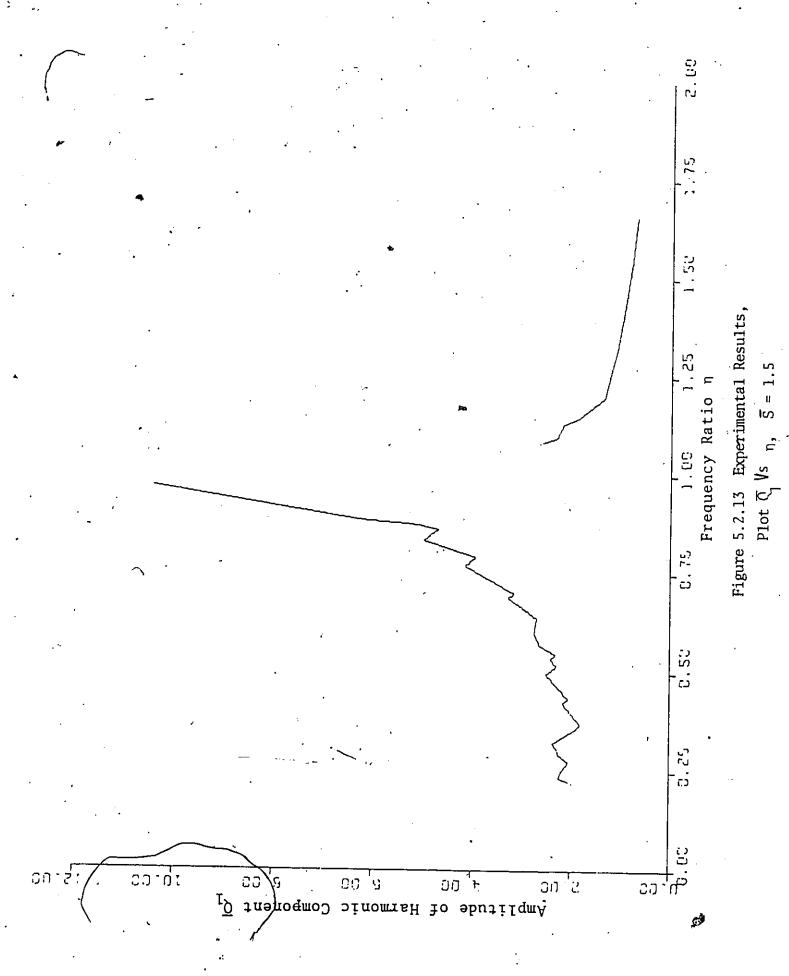


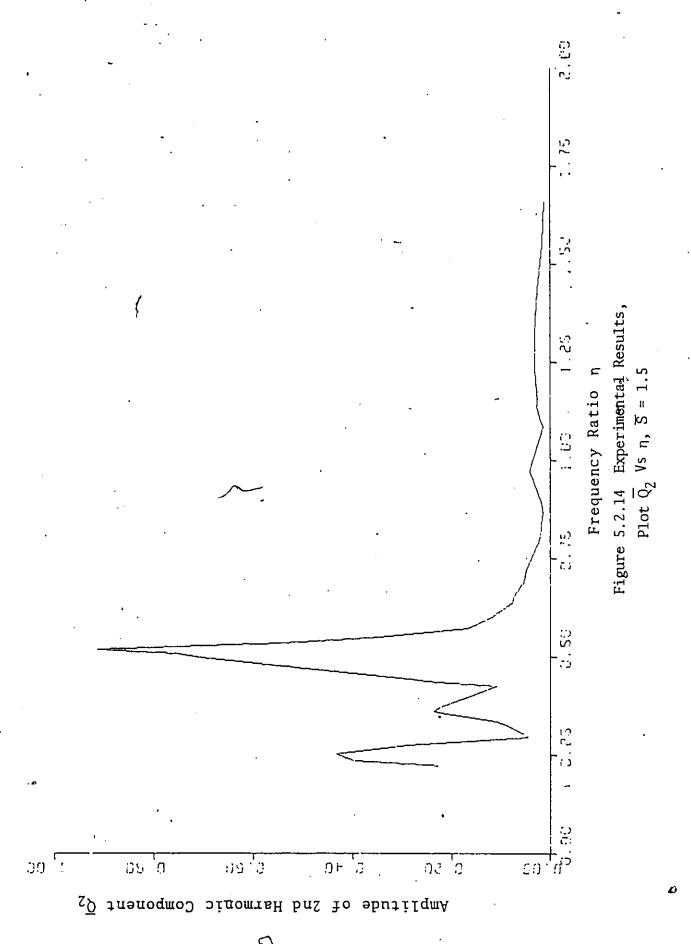


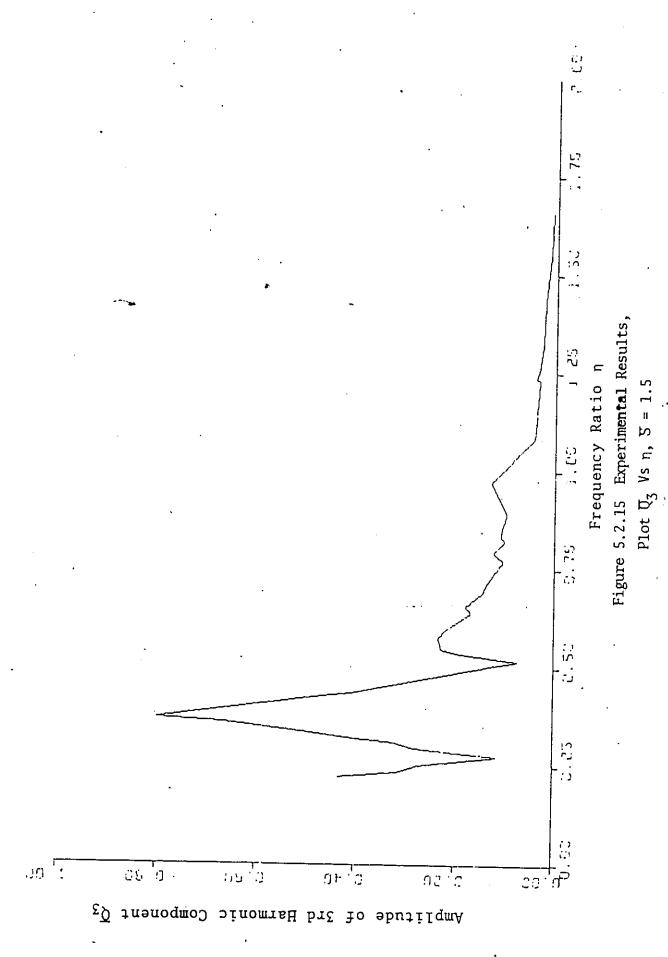


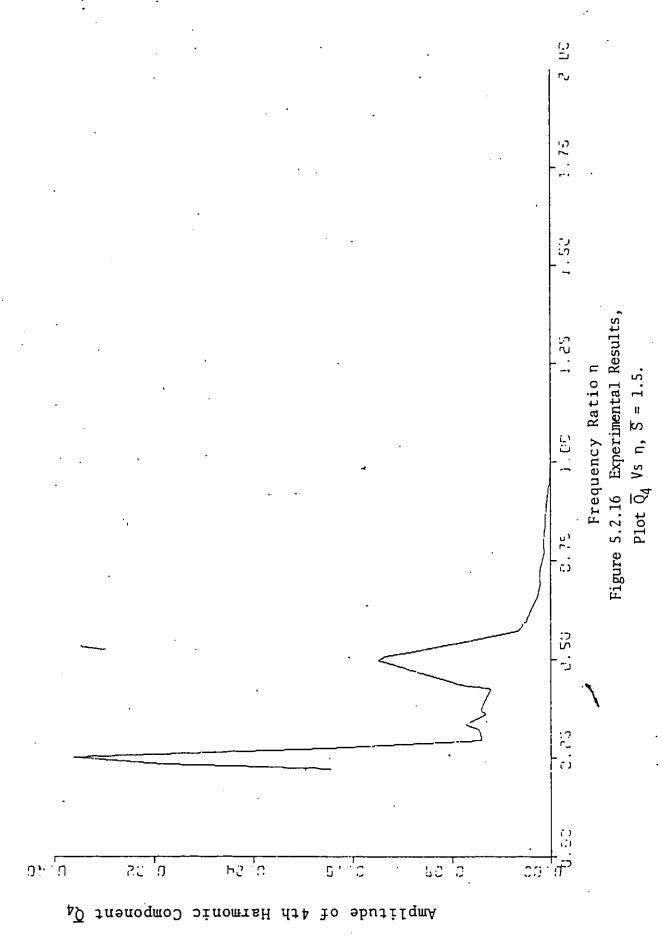


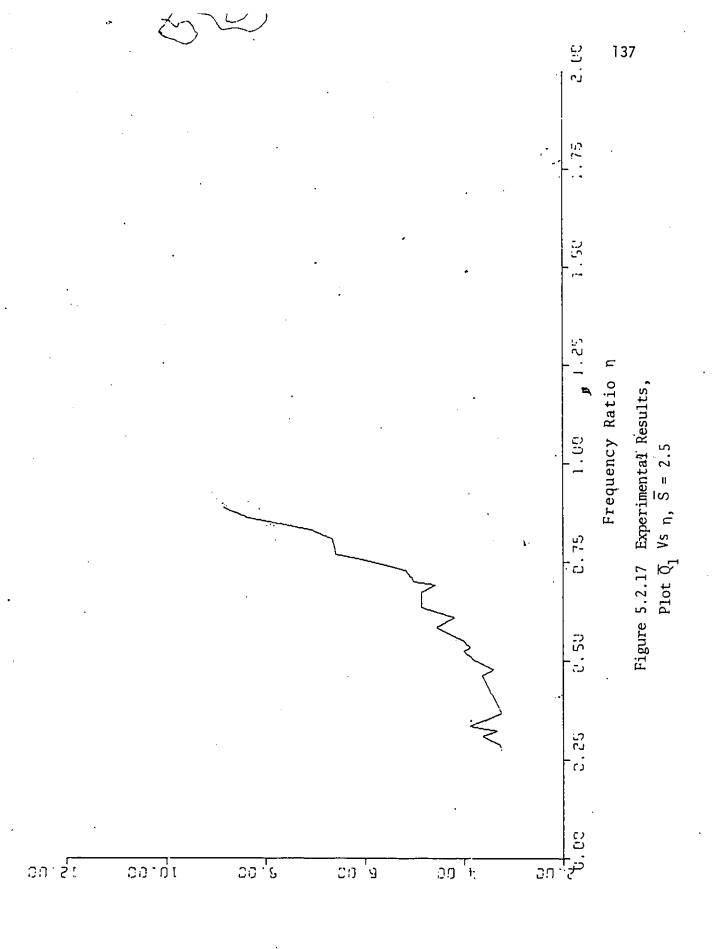
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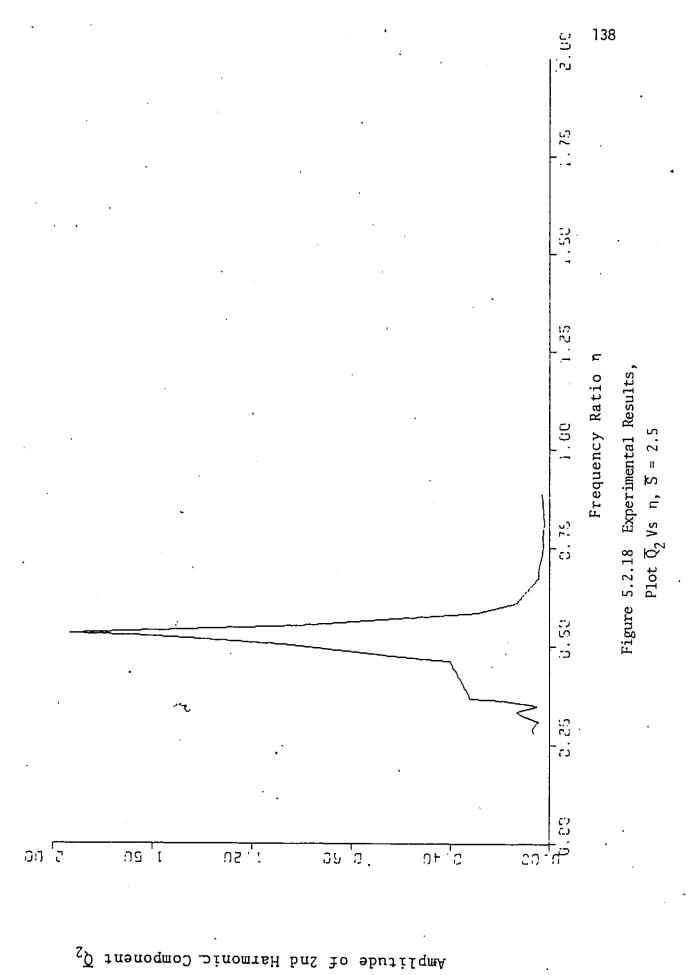


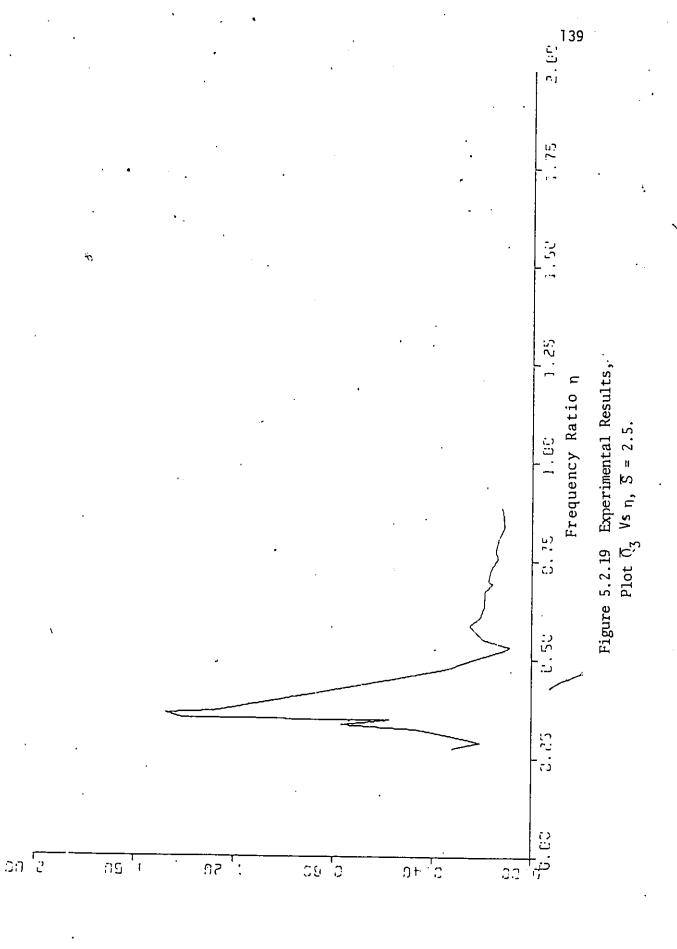




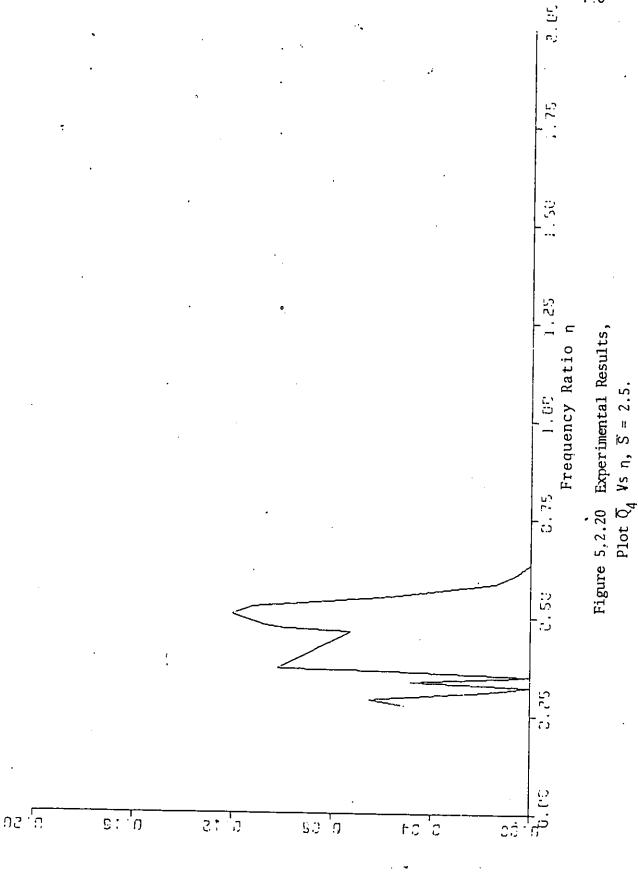


. Amplitude of Harmonic Component  $\overline{\mathbb{Q}}_{1}$ 

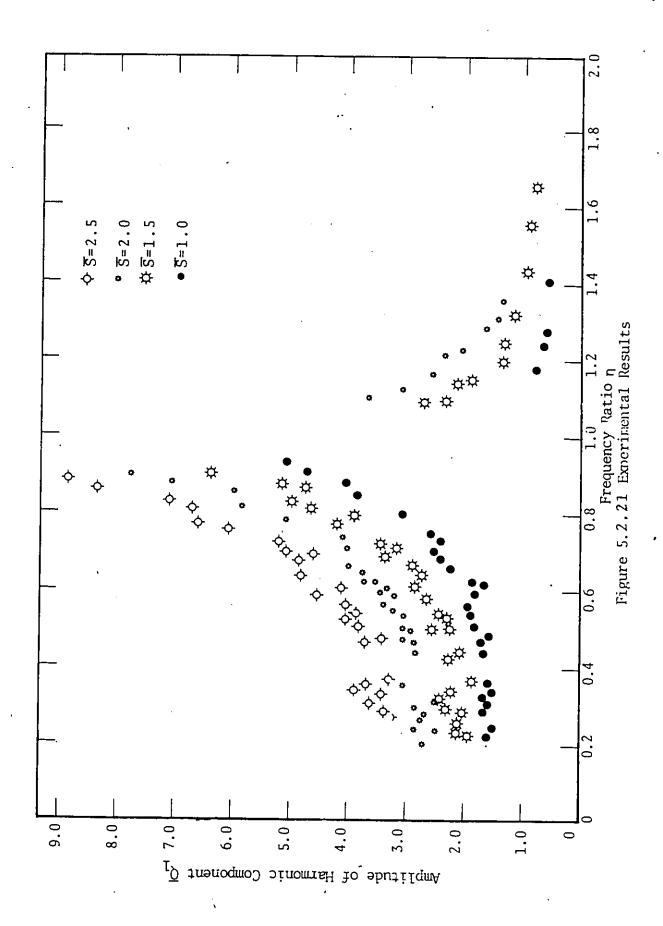


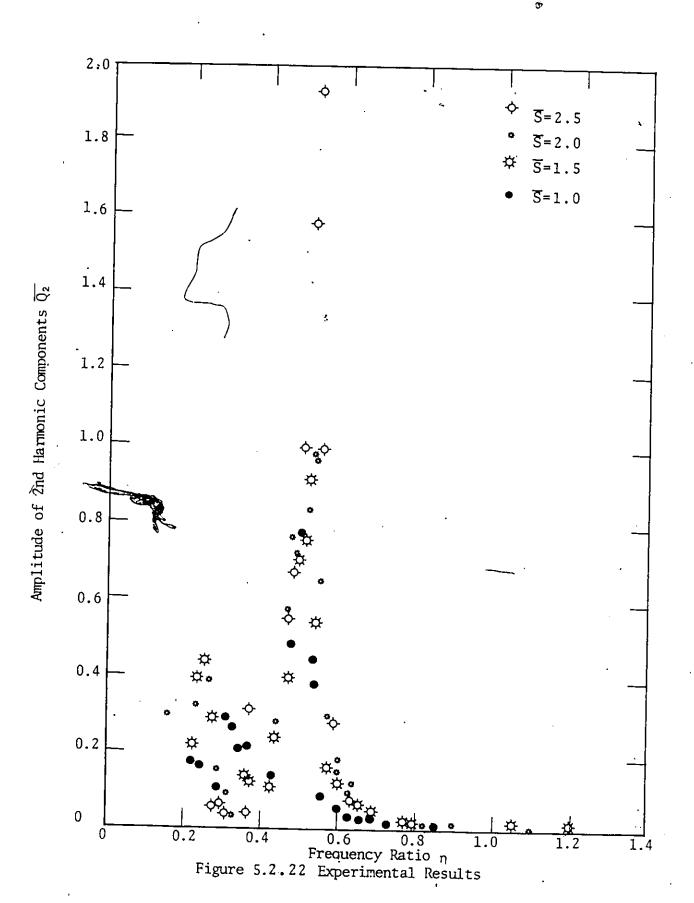


Amplitude of 3rd Component Harmonic  $\overline{Q}_3$ 



Amplitude of 4th Harmonic Component  $\overline{\mathbb{Q}}_{\mathfrak{q}}$ 





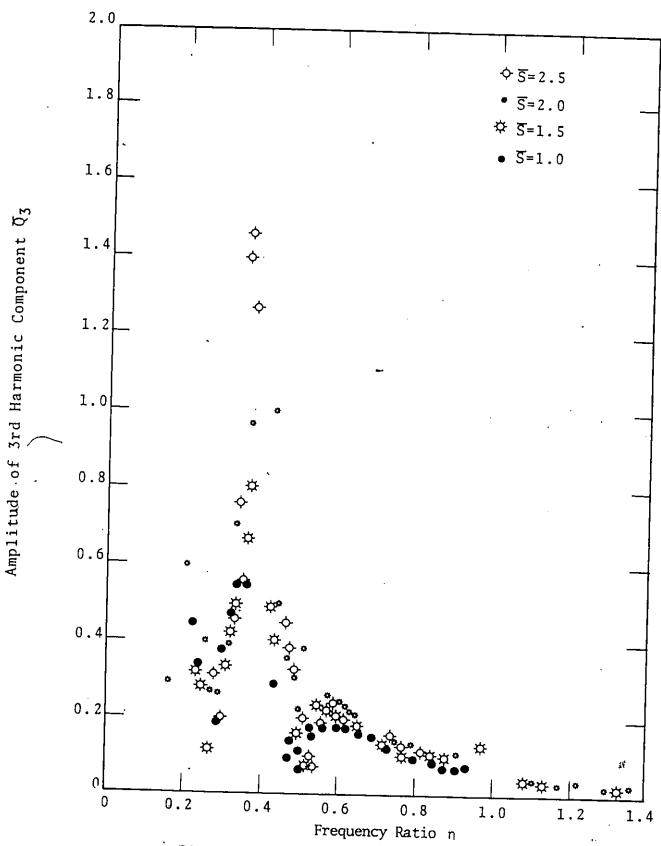
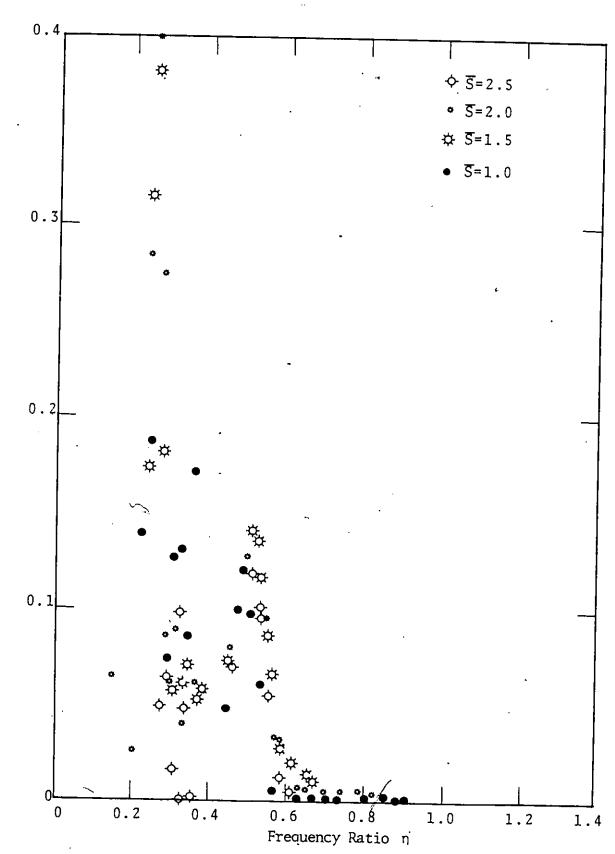


Figure 5.2,23 Experimental Results

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Amplitude of 4th Harmonic Component  $\overline{\mathbb{Q}}_4$ 

Figure 5.2.24 Experimental Results

#### APPENDIX I

#### ANALYTICAL SOLUTION DEVELOPMENT

## I.1 SYSTEM RESPONSE TO A DISTURBING TORQUE OF TYPE

#### T Cos wt

General solutions for the asymmetrical case are derived here. These can be modified for the special case of a symmetrical system by omitting  $\overline{\theta}_1$  from the solution.

## I.1.1 HARMONIC RESONANCE:

Rewriting equations (3.1.A.1) through (3.1.A.3) in the following form:

$$E_{1}(\theta) = J\theta + K\theta - T_{0} - T \cos \omega t = 0$$
 $E_{2}(\theta) = J\theta - T_{0} - T \cos \omega t = 0$ 
 $E_{3}(\theta) = J\theta + K[\theta - \{-(2\theta_{0} + \theta_{1})\}] - T_{0} = T \cos \omega t = 0$ 

(I.1.1.1)

The simplest form of the solution, which can be assumed in this case is:

$$\theta = M + Q_1 \cos \omega t$$
 (1.1.1.2)

Substituting (I.1.1.2) for the approximate value of the displacement  $(\theta)$  in (I.1.1.1). The resulting equation is defined as  $E(\theta)$ , in order to simplify the notation. Applying the Ritz averaging method, the following equations are obtained:

$$\int_{0}^{2\pi} E(\theta) d(\omega t) = 0$$

$$\int_{0}^{2\pi} E(\theta) \cos \omega t d(\omega t) = 0$$
(I.1.1.3)

In equation (I.1.1.3)  $E(\theta)$  is represented by  $E_1(\theta)$  between limits 0 to  $\tau_1$ ,  $E_2(\theta)$  between  $\tau_1$  to  $\tau_2$  and  $E_3(\theta)$  between  $\tau_2$  and  $\tau_3$ . The  $E_1(\theta)$ ,  $E_2(\theta)$  and  $E_3(\theta)$  are expressed as:

$$E_{1}(\tilde{\theta}) = -J\omega^{2} Q_{1} \cos \omega t + K(M+Q_{1} \cos \omega t) - T_{0} - \tilde{T} \cos \omega t = 0$$

$$E_{2}(\tilde{\theta}) = -J\omega^{2} Q_{1} \cos \omega t - T_{0} - T \cos \omega t = 0$$

$$E_{3}(\tilde{\theta}) = -J\omega^{2} Q_{1} \cos \omega t + K(M+Q_{1} \cos \omega t) + K(2\theta_{0}+\theta_{1})$$

$$- T_{0} - T \cos \omega t = 0$$

Referring to figure 3.1.1.2, equation (I.1.1.3) can be written in complete form as:

Integration of expressions (I.1.1.4) and (I.1.1.5) yields two non-linear algebraic equations. From equation (I.1.1.4):

1.

$$\overline{M} = \frac{\overline{Q}_{1}(\sin \tau_{2} - \sin \tau_{1}) - (2 + \overline{\theta}_{1})(\pi - \tau_{2}) + \overline{\theta}_{1} \pi}{(\pi + \tau_{1} - \tau_{2})}$$
 (I.1.1.6)

From equation (I.1.1.5)

$$\eta^{2} = -\frac{S}{\overline{Q}_{1}} + (\frac{2}{\pi}) \left(\frac{2}{\pi} + \frac{\tau_{1}}{2} - \frac{\tau_{2}}{2} + \frac{\sin 2\tau_{1}}{4} - \frac{\sin 2\tau_{2}}{4}\right) + (\frac{2}{\pi}) \left(\frac{M}{\overline{Q}_{1}}\right)$$

$$(\sin \tau_{1} - \sin \tau_{2}) - (\frac{2}{\pi \overline{Q}_{1}}) (2 + \overline{\theta}_{1}) \sin \tau_{2} \qquad (I.1.1.7)$$

From figure 3.1.1.2

$$\theta = -\theta_1$$
 at  $\omega t = \tau_1$ 

and .  $\theta = -(2\theta_0 + \theta_1)$  at  $\omega t = \tau_2$ 

Applying these conditions to expression (I.1.4.2)

$$-\theta_1 = M + Q_1 \cos \tau_1$$

. Dividing throughout by  $\theta_{o}$  and rewriting

$$-\overline{\theta}_1 = \overline{M} + \overline{Q}_1 \cos \tau_1$$

Therefore

$$\tau_1 = \cos^{-1}\left[\frac{-\overline{\theta}_1 - \overline{M}}{\overline{Q}_1}\right] \tag{I.1.1.8}$$

Similarly,

$$-(2\theta_0 + \theta_1) = M + Q_1 \cos \tau_2$$

Dividing by  $\theta_0$  throughout and rewriting:

$$-(2+\overline{\theta}_1) = \overline{M} + \overline{Q}_1 \cos \tau_2$$

Therefore

$$\tau_2 = \cos^{-1}\left[\frac{-2-\overline{\theta}_1-\overline{M}}{\overline{Q}_1}\right] \tag{I.1.1.9}$$

### I.1.2 3RD ORDER ULTRAHARMONIC RESONANCE:

The simplest form of the solution, which can be used in this case, consists of three terms, thus

$$\theta = \overline{M} + Q_1 \cos \omega t + Q_3 \cos 3\omega t$$
 (I.1.2.1)

Substituting (I.1.2.1) into (I.1.1.1) and redefining  $E_1(\theta)$ ,  $E_2(\theta)$  and  $E_3(\theta)$  as:

$$E_{1}(\tilde{\theta}) = -J\omega^{2} Q_{1} \cos \omega t - 9J\omega^{2} Q_{3} \cos 3\omega t + K(M+Q_{1} \cos \omega t) + Q_{3} \cos 3\omega t - T_{0} - T \cos \omega t = 0$$

$$E_{2}(\tilde{\theta}) = -J\omega^{2} Q_{1} \cos \omega t - 9J\omega^{2} Q_{3} \cos 3\omega t - T_{0} - T \cos \omega t = 0$$

$$E_{3}(\tilde{\theta}) = -J\omega^{2} Q_{1} \cos \omega t - 9J\omega^{2} Q_{3} \cos 3\omega t + K(M+Q_{1} \cos \omega t) + Q_{3} \cos 3\omega t + K(M+Q_{1} \cos \omega t) + C(2\theta_{0}+\theta_{1}) - T_{0} - T \cos \omega t = 0$$

Applying the Ritz averaging method and referring to figure 3.1.2.1, the following equations are obtained:

$$2 \int_{0}^{\tau_{1}} E_{1}(\tilde{\theta}) d(\omega t) + 2 \int_{0}^{\tau_{2}} E_{2}(\tilde{\theta}) d(\omega t) + 2 \int_{0}^{\pi} E_{3}(\tilde{\theta}) d(\omega t) = 0$$

$$0 \qquad \qquad \tau_{1} \qquad \qquad (I.1.2.2)$$

$$2 \int_{0}^{\tau_{1}} E_{1}(\tilde{\theta}) Cos \omega t d(\omega t) + 2 \int_{0}^{\tau_{2}} E_{2}(\tilde{\theta}) Cos \omega t d(\omega t)$$

$$\qquad \qquad + 2 \int_{0}^{\tau_{2}} E_{3}(\tilde{\theta}) Cos \omega t d(\tilde{\omega} t) = 0$$

$$\qquad \qquad \tau_{2} \qquad \qquad (I.1.2.3)$$
and 
$$2 \int_{0}^{\tau_{1}} E_{1}(\tilde{\theta}) Cos 3\omega t d(\omega t) + 2 \int_{0}^{\tau_{2}} E_{2}(\tilde{\theta}) Cos 3\omega t d(\omega t)$$

and 
$$2\int_{0}^{\tau_{1}} E_{1}(\tilde{\theta}) \cos 3\omega t \ d(\omega t) + 2\int_{0}^{\tau_{2}} E_{2}(\tilde{\theta}) \cos 3\omega t \ d(\omega t) + 2\int_{0}^{\tau_{2}} E_{3}(\tilde{\theta}) \cos 3\omega t \ d(\omega t) = 0$$

$$(1.1.2.4)$$

Integration of expressions (I.1.2.2) to (I.1.2.4) yields three non-linear algebraic equations.

From equation (I.1.2.2):

$$M = \frac{\overline{\mathbb{Q}}_{1}(\sin \tau_{2} - \sin \tau_{1}) + \overline{\mathbb{Q}}_{3} \frac{1}{3}(\sin 3\tau_{2} - \sin 3\tau_{1}) - (2 + \overline{\theta}_{1})(\pi - \tau_{2}) + \overline{\theta}_{1}\pi}{(\pi + \tau_{1} - \tau_{2})}$$

$$(I.1.2.5)$$

From equation (I.1.2.3):

$$\overline{Q}_{1} = \frac{[(\overline{S})\pi - \overline{Q}_{3}\{\frac{1}{4}(\sin 4\tau_{1} - \sin 4\tau_{2}) + \frac{1}{2}(\sin 2\tau_{1} - \sin 2\tau_{2})\}}{\frac{-2\overline{M}(\sin \tau_{1} - \sin \tau_{2}) + 2(2 + \overline{\theta}_{1})\sin \tau_{2}]}{[(\pi + \tau_{1} - \tau_{2}) + \frac{1}{2}(\sin 2\tau_{1} - \sin 2\tau_{2}) - \pi\eta^{2}]}}$$
(I.1.2.6)

From equation (I.1.2.4):

$$\eta^{2} = \left(\frac{1}{9} \pi\right) \left(\frac{\overline{Q}_{1}}{\overline{Q}_{3}}\right) \left[\frac{1}{4} \left(\sin 4\tau_{1} - \sin 4\tau_{2}\right) + \frac{1}{2} \left(\sin 2\tau_{1} - \sin 2\tau_{2}\right)\right] + \left(\frac{1}{9} \pi\right) \left[\pi + (\tau_{1} - \tau_{2}) + \frac{1}{6} \left(\sin 6\tau_{1} - \sin 6\tau_{2}\right)\right] + \left(\frac{2}{27\pi}\right) \left(\frac{\overline{M}}{\overline{Q}_{3}}\right)$$

$$\left(\sin 3\tau_{1} - \sin 3\tau_{2}\right) - \left(\frac{2}{27\pi}\right) \left(\frac{1}{\overline{Q}_{3}'}\right) \left(2 + \overline{\theta}_{1}\right) \sin 3\tau_{2}$$

$$\left(1.1.2.7\right)$$

From figure 3.1.2.1

$$\begin{array}{lll}
\tilde{\theta} &= -\theta_1 & \text{at } \omega t = \tau_1 \\
\tilde{\theta} &= -(2\theta_0 + \theta_1) & \text{at } \omega t = \tau_2
\end{array}$$

Applying these conditions to expression (I.1.2.1)

$$-\theta_{1} = M + Q_{1} \cos \tau_{1} + Q_{3} \cos 3\tau_{1}$$

Dividing throughout by  $\theta_{o}$  and rewriting

$$-\overline{\theta}_1 = \overline{M} + \overline{Q}_1 \cos \tau_1 + \overline{Q}_3 \cos 3\tau_1$$

Therefore,

$$\tau_1 = \cos^{-1}\left[\frac{-\overline{\theta}_1 - \overline{M} - \overline{Q}_3 \cos 3\tau_1}{\overline{Q}_1}\right]$$
 (I.1.2.8)

Similarly,

$$-(2\theta_0 + \theta_1) = M + Q_1 \cos \tau_2 + Q_3 \cos 3\tau_2$$

Dividing throughout by  $\boldsymbol{\theta}_{0}$  and rewriting

$$-(2+\overline{\theta}_1) = \overline{M} + \overline{Q}_1 \cos \tau_2 + \overline{Q}_3 \cos 3\tau_2$$

hence

$$\tau_2 = \cos^{-1}\left[\frac{-2 - \overline{\theta}_1 - \sqrt{M} - \overline{Q}_3 \cos 3\tau_2}{\overline{Q}_1}\right]$$
 (I.1.2.9)

## I.1.3 1/3RD ORDER SUBHARMONIC RESONANCE:

The simplest form of the solution, which can be used in this case is:

$$\tilde{\theta} = M + Q_1 \cos \omega t + Q_{1/3} \cos \frac{\omega t}{3}$$
 (I.1.3.1)

Substituting (I.1.3.1) into (I.1.1.1) and redefining  $E_1(\theta)$ ,  $E_2(\theta)$  and  $E_3(\theta)$  as:

$$E_{1}(\theta) = -J\omega^{2} Q_{1} \cos \omega t - \frac{J\omega^{2}}{9} Q_{1/3} \cos \frac{\omega t}{3} + K(M + Q_{1} \cos \omega t) + Q_{1/3} \cos \frac{\omega t}{3} - T_{0} - T \cos \omega t = 0$$

$$E_2(\theta) = -J\omega^2 Q_1 \cos \omega t - \frac{J}{9}\omega^2 Q_{1/3} \cos \frac{\omega t}{3} - T_0 - T \cos \omega t = 0$$

$$E_{3}(\theta) = -J\omega^{2} Q_{1} \cos \omega t - \frac{J}{9} \omega^{2} Q_{1/3} \cos \frac{\omega t}{3\epsilon} + K(M + Q_{1} \cos \omega t) + Q_{1/3} \cos \frac{\omega t}{3} + K(2\theta_{0} + \theta_{1}) - T_{0} - T \cos \omega t = 0$$

Applying the Ritz averaging method, and referring to figure 3.1.3.1, the following equations are obtained:

(I.1.3.2)

and

Integration of expressions (I.1.3.2) to (I.1.3.4) yields three non-linear algebraic equation.

From equation (I.1.3.2):

$$\overline{M} = \frac{\overline{Q}_{1}(\sin \tau_{2} - \sin \tau_{1}) - 3\overline{Q}_{1/3}(\sin \frac{\tau_{1}}{3} - \sin \frac{\tau_{2}}{3}) - (2 + \overline{\theta}_{1})(3\pi - \tau_{2}) + 3\pi\overline{\theta}_{1}}{(3\pi + \tau_{1} - \tau_{2})}$$

$$(I.1.3.5)$$
From equation (I.1.3.3):
$$[3\pi \overline{S} - \overline{Q}_{1/3}\{\frac{3}{4}(\sin \frac{4\tau_{1}}{3} - \sin \frac{4\tau_{2}}{3}) + \frac{3}{2}(\sin \frac{2\tau_{1}}{3} - \sin \frac{2\tau_{2}}{3})\}$$

$$= \frac{2\overline{M}(\sin \tau_{1} - \sin \tau_{2}) - 2(2 + \overline{\theta}_{1}) \sin \tau_{2}]}{[3\pi + \tau_{1} - \tau_{2} + \frac{1}{2}(\sin 2\tau_{1} - \sin 2\tau_{2}) - 3\pi \eta^{2}]}$$

$$(I.1.3.6)$$

ß

From equation (I.1.3.4)

$$\eta^{2} = \left(\frac{6}{\pi}\right) \left(\frac{\overline{Q}_{1}}{\overline{Q}_{1/3}}\right) \left[\frac{3}{8} \left(\sin \frac{4\tau_{1}}{3} - \sin \frac{4\tau_{2}}{3}\right) + \frac{3}{4} \left(\sin \frac{2\tau_{1}}{3} - \sin \frac{2\tau_{2}}{3}\right)\right] + \left(\frac{6}{\pi}\right) \left[\frac{3\pi}{2} + \frac{1}{2} \left(\tau_{1} - \tau_{2}\right) + \frac{3}{4} \left(\sin \frac{2\tau_{1}}{3} - \sin \frac{2\tau_{2}}{3}\right)\right] + \left(\frac{18}{\pi}\right) \left(\frac{\overline{M}}{\overline{Q}_{1/3}}\right) \left(\sin \frac{\tau_{1}}{3} - \sin \frac{\tau_{2}}{3}\right) - \left(\frac{18}{\pi}\overline{Q}_{1/3}\right) \left(2 + \overline{\theta}_{1}\right) \sin \frac{\tau_{2}}{3}$$
(I.1.3.7)

From figure 3.1.3.1

$$\theta = -\theta_1 \qquad \text{at} \qquad \omega t = \tau_1$$

$$\theta = -(2\theta_0 + \theta_1) \qquad \text{at} \qquad \omega t = \tau_2$$

Applying these conditions to expression (I.1.3.1)

$$-\theta_1 = M + Q_1 \cos \tau_1 + Q_{1/3} \cos \frac{\tau_1}{3}$$

Dividing throughout by  $\theta_0$  and rewriting

$$. -\overline{\theta}_1 = \overline{M} + \overline{Q}_1 \cos \tau_1 + \overline{Q}_{1/3} \cos \frac{\tau_1}{3}$$

Therefore,

$$\tau_{\downarrow} = 3\cos^{-1} \left[ \frac{-\overline{\theta}_{1} - \overline{M} - \overline{Q}_{1} \cos \tau_{1}}{\overline{Q}_{1/3}} \right]$$
 (I.1.3.8)

Similarly,

$$-(2\theta_0 + \theta_1) = \overline{M} + \overline{Q}_1 \cos \tau_2 + \overline{Q}_{1/3} \cos \frac{\tau_2}{3}$$

Dividing throughout by  $\theta_0$  and rewriting

$$-(2+\overline{\theta}_1) = \overline{M} + \overline{Q}_1 \cos \tau_2 + \overline{Q}_{1/3} \cos \frac{\tau_2}{3}$$

hence

$$\tau_2 = 3\cos^{-1}\left[\frac{-2 - \overline{\theta}_1 - \overline{M} - \overline{Q}_1 \cos \tau_2}{\overline{Q}_{1/3}}\right]$$
 (I.1.3.9)

## I.2 SYSTEM RESPONSE TO A DISTURBING TORQUE OF TYPE C $\omega^2$ Cos $\omega t$ I.2.1 HARMONIC RESONANCE:

Rewriting equations (3.1.B.1) through (3.1.B.3) in the following form:

$$E_{1}(\theta) = \overset{\cdot \cdot \cdot}{\theta} + p^{2} (\theta - \theta_{0}) - Z \omega^{2} \cos \omega t = 0$$

$$E_{2}(\theta) = \overset{\cdot \cdot \cdot}{\theta} - Z \omega^{2} \cos \omega t = 0$$

$$E_{3}(\theta) = \overset{\cdot \cdot \cdot}{\theta} + p^{2} [\theta - (-\theta_{0})] - Z \omega^{2} \cos \omega t = 0 \quad (I.2.1.1)$$

The simplest form of the solution, which can be assumed in this case is:

$$\tilde{\theta} = Q_1 \cos \omega t$$
 (I.2.1.2)

Expression (I.2.1.2) for the approximate value of the displacement  $\theta$  is substituted for  $\theta$  in expression (I.2.1.1). The resulting equation is defined hereby as  $E(\theta)$ . Applying the Ritz averaging method, the following equations are obtained:

$$E(\theta) \quad \cos \omega t \cdot d(\omega t) = 0 \qquad (I.2.1.3)$$

Because the characteristic is symmetrical with respect to the origin, the upper limit in equation (I.2.1.3) may be changed to  $\pi/2$ . In equation (I.2.1.3),  $E(\theta)$  is represented by  $E_1(\theta)$  between limits 0 to  $\tau_1$ ,  $E_2(\theta)$  between  $\tau_1$  and  $\pi/2$  with  $\theta_0 = Q_1 \cos \tau_1$ . The  $E_1(\theta)$  and  $E_2(\theta)$  are represented as:

$$E_{1}(\theta) = -\omega^{2} Q_{1} \cos \omega t + p^{2}(Q_{1} \cos \omega t - \theta_{0}) - Z \omega^{2} \cos \omega t = 0$$

$$E_{2}(\theta) = -\omega^{2} Q_{1} \cos \omega t - Z \omega^{2} \cos \omega t = 0$$

Thus, referring to figure 3.2.1.2, equation (I.2.1.3) may be written in the more complete form:

Integration of expression (I.2.1.4) yields:

$$(-\eta^2 \overline{Q}_1 - \frac{z}{\theta_0} \eta^2) \pi + \overline{Q}_1 (2\tau_1 + \sin 2\tau_1) - 4\sin \tau_1 = 0$$

or

$$\eta^{2} = \frac{4\sin \tau_{1} - \overline{Q}_{1}(2\tau_{1} + \sin 2\tau_{1})}{(-\overline{Q}_{1} - Z')\pi}$$
 (I.2.1.5)

From figure 3.2.1.2

$$\tilde{\theta} = \theta_0$$
 at  $\omega t = \tau_1$ 

Applying this condition to expression (I.2.1.2)

$$\theta_0 = Q_1 \cos \tau_1$$

Dividing throughout by  $\theta_0$  and rewriting  $1 = \overline{Q}_1 \cos \tau_1$  hence,

$$\tau_1 = \cos^{-1} \left[ \frac{1}{\overline{Q}_1} \right]$$
 (1.2.1.6)

3

## I.2.2 3RD ORDER ULTRAHARMONIC RESONANCE:

The simplest form of the solution, which can be used in this case, consists of two terms, thus

$$\theta = Q_1 \cos \omega t + Q_3 \cos 3\omega t$$
 (I.2.2.1)

Applying the Ritz averaging method, the following equations are obtained:

$$\int_{0}^{2\pi} E(\theta) \cos \omega t \, d(\omega t) = 0$$

$$2\pi$$

$$\int_{0}^{2\pi} E(\theta) \cos 3\omega t \, d(\omega t) = 0$$

$$(I.2.2.2)$$

In equation (I.2.2.2),  $E(\theta)$  is represented by  $E_1(\theta)$  between limits 0 to  $\tau_1$ ,  $E_2(\theta)$  between  $\tau_1$  to  $\tau_2$  and  $E_3(\theta)$  between  $\tau_2$  to  $\pi$ .

The  $E_1(\theta)$ ,  $E_2(\theta)$  and  $E_3(\theta)$  are expressed as:  $E_1(\theta) = -\omega^2 Q_1 \cos \omega t - 9\omega^2 Q_3 \cos 3\omega t + p^2(Q_1 \cos \omega t + Q_3 \cos 3\omega t - \theta_0) - Z \omega^2 \cos \omega t = 0$   $E_2(\theta) = -\omega^2 Q_1 \cos \omega t - 9\omega^2 Q_3 \cos 3\omega t - Z \omega^2 \cos \omega t = 0$  $E_3(\theta) = -\omega^2 Q_1 \cos \omega t - 9\omega^2 Q_3 \cos 3\omega t + p^2(Q_1 \cos \omega t + Q_3 \cos 3\omega t + \theta_0) - Z \omega^2 \cos \omega t = 0$ 

Referring to figure 3.2.2.1, equations (I.2.2.2) can be written in the complete form as:

Integration of expressions (I.2.2.3) and (I.2.2.4) yields:

$$(-\eta^2 \ \overline{Q}_1 - \frac{z}{\theta_0} \ \eta^2) \pi + \overline{Q}_1 (2\tau_1 + \sin 2\tau_1) + \overline{Q}_3 (\sin 2\tau_1 + 1/2)$$

$$\sin 4\tau_1) - 4\sin \tau_1 = 0$$
 (I.2.2.5)

and,

$$-9\pi \eta^{2} \overline{Q}_{3} + \overline{Q}_{1}(\sin 2\tau_{1} + \frac{1}{2} \sin 4\tau_{1}) + \overline{Q}_{3}(2\tau_{1} + \frac{1}{3} \sin 6\tau_{1})$$

$$- \frac{4}{3} \sin 3\tau_{1} = 0 \qquad (I.2.2.6)$$

From expression (I.2.2.5):

$$\overline{Q}_{1} = \frac{[4\sin \tau_{1} - \overline{Q}_{3}(\sin 2\tau_{1} + \frac{1}{2}\sin 4\tau_{1}) + z' \pi \eta^{2}]}{(2\tau_{1} + \sin 2\tau_{1} - \pi \eta^{2})}.$$
(I.2.2.7)

From expression (I.2.2.6):

$$\eta^{2} = \left(\frac{\overline{Q}_{1}}{9\pi \overline{Q}_{3}}\right) \left(\sin 2\tau_{1} + \frac{1}{2} \sin 4\tau_{1}\right) + \left(\frac{1}{9\pi}\right) \left(2\tau_{1} + \frac{1}{3} \sin 6\tau_{1}\right)$$

$$- \frac{4}{3} \left(\frac{1}{9\pi \overline{Q}_{3}}\right) \sin 3\tau_{1} \qquad (I.2.2.8)$$

From figure 3.2.2.1

$$\tilde{\theta} = \theta_0$$
 at  $\omega t = \dot{\tau}_1$ 

Applying this condition to expression (I.2.2.1)

$$\theta_{o} = Q_{1} \cos \tau_{1} + Q_{3} \cos 3\tau_{1}$$
(I.2.2.9)

Dividing throughout by  $\theta_0$  and rewriting

$$1 = \overline{Q}_1 \cos \tau_1 + \overline{Q}_3 \cos 3\tau_1$$

hence,

$$\overline{Q}_1 = \cos^{-1}\left[\frac{1-\overline{Q}_3 \cos 3\tau_1}{\overline{Q}_1}\right]$$

## I.2.3 1/3RD ORDER SUBHARMONIC RESONANCE:

The simplest form of the solution, which can be used in this case, consists of two terms, thus

$$\tilde{\theta} = Q_1 \cos \omega t + Q_{1/3} \cos \frac{\omega t}{3}$$
 (I.2.3.1)

Applying the Ritz averaging method, the following equations are obtained:

$$\int_{0}^{6\pi} \tilde{E(\theta)} \quad \cos \omega t \, d(\omega t) = 0$$

and,

$$\int_{0}^{6\pi} E(\theta) \cos \frac{\omega t}{3} d(\omega t) = 0 \qquad (I.2.3.2)$$

In equation (I.2.3.2),  $E_1(\theta)$  is represented by  $E_1(\theta)$  between limits 0 to  $\tau_1$ ,  $E_2(\theta)$  between  $\tau_1$  to  $\tau_2$  and  $E_3(\theta)$  between  $\tau_2$  to  $3\pi$ .

The  $E_1(\theta)$ ,  $E_2(\theta)$  and  $E_3(\theta)$  are expressed as:

$$E_{1}(\tilde{\theta}) = -\omega^{2} Q_{1} \cos \omega t - \frac{\omega^{2}}{9} Q_{1/3} \cos \frac{\omega t}{3} + p^{2}(Q_{1} \cos \omega t) + Q_{1/3} \cos \frac{\omega t}{3} - \theta_{0}) - Z \omega^{2} \cos \omega t = 0$$

$$E_{2}(\tilde{\theta}) = -\omega^{2} Q_{1} \cos \omega t - \frac{\omega^{2}}{9} Q_{1/3} \cos \frac{\omega t}{3} - Z \omega^{2} \cos \omega t = 0$$

$$E_{3}(\tilde{\theta}) = -\omega^{2} Q_{1} \cos \omega t - \frac{\omega^{2}}{9} Q_{1/3} \cos \frac{\omega t}{3} + p^{2}(Q_{1} \cos \omega t) + Q_{1/3} \cos \frac{\omega t}{3} + \theta_{0}) - Z \omega^{2} \cos \omega t = 0$$

Referring to figure 3.2.3.1, equations (I.2.3.2) can now be written as:

Integration of expressions (1.2.3.3) and (1.2.3.4) Fields:

$$(-\eta^2 \ \overline{Q}_1 - \frac{Z}{\theta_0} \ \eta^2) \ \frac{3\pi}{2} + \overline{Q}_1(\tau_1 + \frac{1}{2} \cdot \sin 2\tau_1) + \overline{Q}_{1/3}(\frac{3}{4} \cdot \sin \frac{4\tau_1}{3} + \frac{3}{2} \cdot \sin \frac{2\tau_1}{3}) - 2\sin \tau_1 = 0$$
 (I.2.3.5)

and,

$$\overline{Q}_{1}\left[\begin{array}{cccc} \frac{3}{4} \sin \frac{4\tau_{1}}{3} + \frac{3}{2} \sin \frac{2\tau_{1}}{3}\right] - \frac{3\pi}{2} \left(\frac{\eta^{2}}{9}\right) \overline{Q}_{1/3} + \overline{Q}_{1/3}(\tau_{1} + \frac{3}{2}) \\ \sin \frac{2\tau_{1}}{3} - 6 \sin \frac{\tau_{1}}{3} = 0 \end{array}$$
(I.2.3.6)

From expression (I.2.3.5)

$$\overline{Q}_{1} = \frac{4\sin \tau_{1} + Z' 3\pi \eta^{2} - \overline{Q}_{1/3}(\frac{3}{2}\sin \frac{4\tau_{1}}{3} + 3\sin \frac{2\tau_{1}}{3})}{(2\tau_{1} + \sin 2\tau_{1} - 3\pi \eta^{2})}$$
(I.2.3.7)

and from expression (1.2.3.6)

$$\eta^{2} = \left(\frac{\overline{Q}_{1}}{\overline{Q}_{1/3}}\right) \left(\frac{6}{\pi}\right) \left(\frac{3}{4} \sin \frac{4\tau_{1}}{3} + \frac{3}{2} \sin \frac{2\tau_{1}}{3}\right) + \left(\frac{6}{\pi}\right) \left(\tau_{1} + \frac{3}{2} \sin \frac{2\tau_{1}}{3}\right) - \left(\frac{36}{\pi} \cdot \frac{1}{\overline{Q}_{1/3}}\right) \sin \frac{\tau_{1}}{3} \qquad (I.2.3.8)$$

From figure (3.2.3.1)

$$\theta = \theta_0$$
 at  $\omega t = \tau_1$ 

Applying this condition to expression (I.2.3.1):

$$\theta_0 = Q_1 \cos \tau_1 + Q_{1/3} \cos \frac{\tau_1}{3}$$

Dividing throughout by  $\theta_0$  and rewriting

$$1 = \overline{Q}_1 \cos \tau_1 + \overline{Q}_{1/3} \cos \frac{\tau_1}{3}$$

hence,

$$\tau_1 = 3\cos^{-1} \left[ \frac{1 - \overline{Q}_1 \cos \tau_1}{\overline{Q}_{1/3}} \right]$$
 (1.2.3.9)



#### APPENDIX II

#### ITERATIVE PROCEDURE FOR THE SOLUTION OF SIMULTANEOUS EQUATIONS:

To illustrate the iterative procedure, simultaneous equations(I.1.1.6), (I.1.1.7), (I.1.1.8) and (I.1.1.9) will be solved using this technique.

$$\tau_1 = \cos^{-1}\left[\frac{-\overline{\theta}_1 - \overline{M}}{\overline{Q}_1}\right] \tag{I.1.1.8}$$

$$\tau_2 = \cos^{-1} \left[ -\frac{2 - \overline{\theta}_1 - \overline{M}}{\overline{Q}_1} \right]$$
 (I.1.1.9)

$$M = \frac{\overline{Q}_{1}(\sin \tau_{2} - \sin \tau_{1}) - (2 + \overline{\theta}_{1})(\pi - \tau_{2}) + \overline{\theta}_{1} \pi}{(\pi + \tau_{1} - \tau_{2})}$$
 (I.1.1.6)

$$\eta^{2} = -(\frac{\overline{S}}{\overline{Q}_{1}}) + (\frac{2}{\pi}) \left[ \frac{\pi}{2} + 1/2(\tau_{1} - \tau_{2}) + 1/4(\sin 2\tau_{1} - \sin 2\tau_{2}) \right]$$

$$+ (\frac{2}{\pi}) \left( \frac{M}{\overline{Q}_{1}} \right) \left( \sin \tau_{1} - \sin \tau_{2} \right) - \left( \frac{2}{\pi \overline{Q}_{1}} \right) \left( 2 + \overline{\theta}_{1} \right) \sin \tau_{2}$$

$$(1.1.1.7)$$

To initiate iteration, starting values are chosen for M,  $\tau_1$  and  $\tau_2$ . New values of  $\tau_1$  and  $\tau_2$  are calculated by substituting the previously assumed starting values. Again, these new calculated values are taken as starting values and next new values are calculated. This procedure is continued until there is a sufficient convergence. The conditions for convergence are taken as --

$$\tau_1(I) - \tau_1(I-1) \le \varepsilon$$
  $\varepsilon = 0.001$  (error function)

and

$$\tau_2(I) - \tau_2(I-1) \leq \varepsilon$$

The iterated values of  $\tau_1$  and  $\tau_2$  are substituted into the expression for M(I.1.1.6). The calculated value of M is substituted into (I.1:1.7) to calculate  $\eta^2$ .

The iterated values of  $\tau_1$  and  $\tau_2$  and calculated value of  $\overline{M}$  become starting values for the iterations of  $\tau_1$  and  $\tau_2$ . The new iterated values of  $\tau_1$  and  $\tau_2$  are used to calculate new values of  $\overline{M}$  and  $\eta^2$ .

This procedure is continued until there is a sufficient convergence for  $\overline{M}$  and  $\eta^2$ . The test of convergence is defined as --

: 
$$\overline{M}(I) - \overline{M}(I-1) \le \varepsilon$$
  $\varepsilon = 0.001$  (error function) or  $\eta^2(I) - \eta^2(I-1) \le \varepsilon$ 

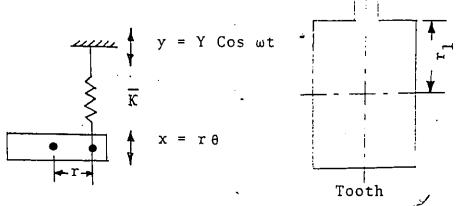
The computer program for the solution of the simultaneous equations using the iterative procedure is given on page 162.

```
('5%,'101111111') X''O BAR''BX,'HDA'', DX''GTASQ''11X''TAU!''11X
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                 10.04(J),441AR(W),ETASD(W),TADI(J),TADR(L),SMAR,ETA(W)
|U.a.dx,Fl0,b.5x,Fl0,6,5x,Fl0,6,5x,Fl0,6,5,Fl0,6,5x
            45744ET21CAL SYSTEM)
•0 11AP(1000)•MHAK(1000)•ŤAU2(1000)•ETA(1000)
                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                            (1-(1)*5*0+1.1=(1)
EC 1.3
```

#### APPENDIX III

#### III.1 RELATIONSHIP BETWEEN SHAKER PIN MOTION AND INPUT TORQUE

TO THE SHAFT:



Equation of motion is given by:

$$\Sigma M = J\Theta$$

$$J\Theta + K\Theta + \overline{K}(x^{\alpha}y) r = 0$$

$$J\Theta + (K+\bar{K}r^2)\Theta = \bar{K} y r$$

$$J\Theta + K'\Theta = \bar{K} Y r \cos \omega t$$

$$\Theta + p^2 \Theta = \frac{\bar{K}}{J} Y r \cos \omega t$$

$$\frac{\pi}{\Theta} + p^2 \Theta = \frac{T}{J} \cos \omega t$$

Where K = shaft torsional stiffness

$$\bar{K}$$
 = spring constant

$$K' = K + \overline{K} r^2$$

$$p^2 = \frac{K'}{J}$$

$$T = \bar{K} Y r$$

Since

$$p^{2} = \frac{K'}{J} = \frac{(K + \bar{K} r^{2})}{J}$$
or  $K + \bar{K} r^{2} = Jp^{2}$  (1)

For the vibrating arm and the counter weight

$$\dot{J} = 0.179 \text{ lb in } \sec^2$$

For the shaft .

$$K = \frac{\pi d^4 G}{32k}$$

$$K = 5617.60$$
 lb in/rad



From the experiment p = 29.3 c/s

From equation (1) --

$$K + \bar{K}r^2 = Jp^2$$

$$5617.60 + \bar{K}(3.5)^2 = 0.179 (2\pi \times 29.3)^2$$

$$\bar{K} = 34.5 \text{ lb/in}$$

Since

$$T = \bar{K} Yr$$

And

Therefore

$$\bar{S} = \frac{\bar{K}}{K} \frac{Yr}{\theta_0}$$

$$Y = \frac{\tilde{S} K \theta}{\tilde{K} r}$$
 inches

Y = 0.175'' peak-peak

Y = 0.140" peak-peak

Y = 0.105'' / peak-peak

Y = 0.070''' peak-peak

Y = 0.056" peak-peak

 $Y = \frac{T}{\bar{K}r}$ 

$$K = 5617.60, \ \theta_0 = \frac{0.0015}{r_1}$$
  
 $\bar{K} = 34.5 \quad r = 3.5, \ r_1^1 = 2.0$ 

for  $\bar{S} = 2.5$ 

Š = 2.0

 $\bar{S} = 1.5$ 

 $\ddot{S} = 1.0$ 

 $\bar{S} = 0.8$ 

## CALCULATION OF THE NON-DIMENSIONAL AMPLITUDES OF MOTION.

Let  $G_1$ ,  $G_2$ ,  $G_3$  &  $G_4$  be the amplitudes of the 1st, 2nd, 3rd and 4th harmonic components of acceleration at a certain forcing frequency.

Since  $a = \theta$ , and  $\theta = Q_1 \cos \omega t$ 

$$G_1 = -\omega^2 Q_1 \cos \omega t$$

$$G_1 = -\omega^2 Q_1 \qquad \qquad \text{III.2.1}$$

$$G_2 = -4\omega^2 Q_2 \cos 2\omega t, \text{ since } \theta = Q_2 \cos 2\omega t$$

$$|G_2| = 4\omega^2 Q_2$$

or 
$$|G_2'| = 4\omega^2 Q_2$$

Similarly  $|G_3| = 9\omega^2 Q_3$ 

and 
$$|G_4| = 16\omega^2 Q_4$$

From III.2.1. 
$$Q_1 = \frac{G_1}{\omega^2}$$

 $= \frac{n_1 \times 386.4}{(2\pi/t)^2}$  inches where  $n_1 = \text{magnitude}$ 

of acceleration in 'g' and t = periodic time.

$$Q_1(Angular) = (\frac{1}{r}). \frac{n_1 \times 386.4}{(2\pi/t)^2}$$
 rad.

 $\theta_{o}$  (clearance) = 0.0015 inches

$$\frac{\theta}{\text{clearance}} = \frac{\frac{\theta}{0}}{r_1} \text{ rad. } \theta$$

$$\overline{Q}_{1} = -\frac{Q_{1}}{\theta_{0}} = (\frac{1}{r}) \frac{n_{1} \times 386.4}{(2\pi/t)^{2}} / (\frac{\theta_{0}}{r_{1}})$$

$$= (\frac{r_{1}}{r}) \frac{n_{1} \times 386.4}{(2\pi/t)^{2}} \times \frac{1}{\theta_{0}} \text{ III.2}$$

 $\delta v$ 

Similarly 
$$\overline{Q}_2 = (\frac{r_1}{r}) \frac{n_2 \times 386.4}{4(2\pi/t)^2} \times \frac{1}{\theta_0}$$
 III.2.3  $\overline{Q}_3 = (\frac{r_1}{r}) \frac{n_3 \times 386.4}{9(2\pi/t)^2} \times \frac{1}{\theta_0}$  III.2.4 and  $\overline{Q}_4 = (\frac{r_1}{r}) \frac{n_4 \times 386.4}{16(2\pi/t)^2} \times \frac{1}{\theta_0}$  III.2.5

 $n_2$ ,  $n_3$ ,  $n_4$  are the magnitudes of acceleration in 'g's.

 $\mathcal{N}$ 

#### APPENDIX IV

# LIMITING CONDITIONS FOR THE GENERATION OF THE HIGHER ORDER RESONANCES IN SYMMETRICAL SYSTEM

For a certain value of a forcing amplitude 5, after passing through the harmonic resonance, the amplitude of motion decreases with the increase of n. If the first subharmonic resonance is not initiated before the tooth loses contact with the restrainer walls, the restoring force and the system become linear and subharmonic resonances cannot be generated. The value of n at which the tooth loses contact is called the cut-off frequency, which can be used as a critical value for defining the limit of the subharmonic generation capacity. Hence, the term "cut-off" frequency means the frequency above which a forcing function of a given amplitude cannot excite the subharmonic resonance.

The relationships between the forcing amplitude  $\overline{S}$  and the corresponding cut off frequency  $\eta$  for the two types of disturbing torques are derived below.

### a. Disturbing torque = T Cos ωt

The equation of motion within the clearance is given

by

or

$$J\theta = T \cos \omega t$$

$$\theta = \frac{T}{I} \cos \omega t \qquad (IV.1)$$

Substituting the approximation for displacement, which is valid within the range of the harmonic resonance,

$$\tilde{\theta} = Q_1 \cos \omega t$$

into (IV.1) we have

$$-\omega^{2} Q_{1} \cos \omega t = \frac{T}{J} \cos \omega t$$
or
$$|Q_{1}| = \frac{T}{J\omega^{2}}$$
Since  $\overline{Q}_{1} = \frac{Q_{1}}{Q_{1}}$  and  $\overline{Q}_{2} = \frac{Q_{1}}{Q_{2}}$ 

Since 
$$\overline{Q}_1 = \frac{Q_1}{\theta_0}$$
 and  $\eta = \frac{\omega}{p}$ 

hence 
$$|\overline{Q}_1| = \frac{T}{J_{p^2\eta^2\theta_0}}$$

But

$$\overline{S} = \frac{T}{K\theta_0}$$
 and  $p^2 = \frac{K}{J}$ 

therefore 
$$|\overline{Q}_1| = \frac{TJ}{JK\theta_0^2\eta^2} = \frac{\overline{S}}{\eta^2}$$

The condition for loss of tooth contact and hence for the limiting angular amplitude below which non-linear conditions do not exist is given by  $|\overline{\mathbb{Q}}_1|=1$ . Therefore,

or

i.e. "cut-off" frequency.

Thus if the conditions for the initiation of the subharmonic resonance at a specific value of  $\overline{S}$  occur at a frequency above  $\eta = \sqrt{S}$ , then this resonance will not take place. The graph of  $\overline{S}$  vs "cut off"  $\eta$  is shown in figure 5.1.27.

## b. Disturbing torque = $C \omega^2 \cos \omega t$ .

The equation of motion within the clearance is given by  ${}^{\mathfrak{g}}$ 

$$J\theta = C \omega^{2} \cos \omega t$$
 or 
$$\theta = Z \omega^{2} \cos \omega t$$
 (IV.2) where 
$$Z = \frac{C}{T} \quad .$$

Assuming the simpest solution which is valid near harmonic resonance  $\theta = Q_1$  Cos  $\omega t$  and substituting into (IV.2) yields:

$$-\omega^2 Q_1 \cos \omega t = Z \omega^2 \cos \omega t$$

or

$$|Q_1| = Z$$

or

$$|\overline{Q}_1| = Z'$$

The limiting value of Z' for the "cut-off" frequency below which non-linear conditions do not exist, is given by:

$$Z' = |\overline{Q}_1| = 1$$

Therefore it is only possible to excite the subharmonic resonance at values Z' > 1.

#### APPENDIX V

#### EXPERIMENTAL ERROR ANALYSIS

The textbook (36) and reference (43) § (86) describe a method to estimate the uncertainty in the calculated result on the basis of the uncertainties in the primary measurements. The result R is given as a function of the independent variables  $x_1, x_2, \ldots, x_n$ . Thus

$$R = R(x_1, x_2, x_3, ..., x_n)$$

Let  $W_r$  be the uncertainty in the result and  $W_1,\ W_2,\ \ldots,\ W_n$  be the uncertainties in the independent variables. If the uncertainties in the measurement of independent variables are all given with the same odds, then the uncertainty in the result having these odds is given as

$$W_{\mathbf{r}} = \left[ \left( \frac{\partial R}{\partial \mathbf{x}_1} \cdot W_1 \right)^2 + \left( \frac{\partial R}{\partial \mathbf{x}_2} \cdot W_2 \right)^2 + \dots + \left( \frac{\partial R}{\partial \mathbf{x}_n} \mathbf{x} \cdot W_n \right)^2 \right]^{1/2}$$

...V.1

#### Estimation of Errors in Measurements of the Amplitude of Motion:

From equation III.2.2

$$\overline{Q}_{1} = \left(\frac{r_{1}}{r}\right) \cdot \frac{n_{1} \times 386.4}{(2\pi/t)^{2}} \cdot \frac{1}{\theta_{0}}$$

$$= f(r_{1}, r, n_{1}, t, \theta_{0})$$

Using equation V.1

$$W_{R} = \left[ \left( \frac{\partial \overline{Q}_{1}}{\partial r_{1}} \cdot Wr_{1} \right)^{2} + \left( \frac{\partial \overline{Q}_{1}}{\partial r} \cdot Wr \right)^{2} + \left( \frac{\partial \overline{Q}_{1}}{\partial n_{1}} \cdot Wn_{1} \right)^{2} + \left( \frac{\partial \overline{Q}_{1}}{\partial r} \cdot Wr \right)^{2} + \left( \frac{\partial \overline{Q}_{1}}{\partial r} \cdot Wr_{0} \right)^{2} \right]^{1/2}$$

where Wr<sub>1</sub>, Wr, Wn<sub>1</sub>, Wt and Wθ<sub>o</sub> are uncertainties in measurements of radial distances  $r_1$  & r, acceleration components, time period t and the gap clearance  $\theta_o$ . The uncertainties Wr<sub>1</sub> and Wr are of the order of ± 0.1%. The uncertainty W<sub>T</sub> is of the order ± 0.1%, since the period was measured in milli seconds employing a digital counter. The uncertainty Wθ<sub>o</sub> is of the order of ± 0.1%.

Hence neglecting Wr  $_{1},$  Wr  $_{0}$  Wt and W  $_{0}$  in the above relationship yields

$$W_{R} = (\frac{\partial Q_{1}}{\partial n_{1}} . Wn_{1})$$
or
$$W_{R} = (\frac{r_{1}}{r}) \frac{386.4}{(2\pi/t)^{2}} . (\frac{1}{\theta_{0}} . Wn_{1}) ... V.2$$

The above expression V.2 shows that uncertainties of measurements of the amplitudes of motion is proportional to the uncertainties involved in acceleration measurements. When using the real time spectrum analyser, the amplitude of the acceleration component can be read within ±1 dB or 12.23% accuracy. The error of the accelerometer and the associated instrumentation is well within 3%. Hence the encertainty of the calculated values of the amplitude of motion are within 15.23%.

Similarly for higher harmonics  $Q_3$ ,  $Q_3$ ,  $Q_4$ , etc. From equation III.2.3

$$\overline{Q}_2 = (\frac{r_1}{r}) \cdot \frac{386.4}{4(2\pi/t)^2} \times \frac{1}{\theta_0} \dots V.3$$

From equation V.2

$$W_{R} = (\frac{r_{1}}{r}) = \frac{386.4}{4(2\pi/t)^{2}} \cdot \frac{1}{\theta_{0}} \cdot Wn_{2}$$

Similarly uncertainty involved in  $Q_3$  is given by

$$W_{R} = (\frac{r_{1}}{r}) = \frac{386.4}{9(2\pi/t)^{2}} \cdot \frac{1}{\theta_{0}} \cdot Wn_{3} + \dots V.4$$

From the above expressions it is evident that uncertainties involved in higher order amplitudes of motion are 1/4, 1/9 & 1/16 of the uncertainties of measurements of fundamental component.

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#### VITA

- 1942 Born in Multan, W. Pakistan
- 1965 Received the degree of Bachelor of Science in Mechanical Engineering, Ranchi University, Bihar, \_ India.
- 1965-1969 Worked as a Mechanical Engineer for several companies in India.
- 1970 Received M.Sc. Degree in Applied Dynamics and Mechanical Vibrations, University of Surrey, Surrey, England.
- 1974-1975 Worked for Zinder Engineering Inc., Ann Arbor, Michigan as an Acoustic Engineer.
- 1976-1977 Worked for Schwitzer Engineered Components, Indianapolis, Indiana as a Project Engineer.
- 1977- Presently employed by Ford Motor Co., Dearborn, Michigan as a Research Engineer.

A candidate for the degree of Doctor of Philosophy in Mechanical Engineering, University of Windsor, Windsor, Ontario, Canada.