LA-UR97-851

CONF-970669--2

Title:

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Submitted to:

International Conference on Engineering Applications of Neural Materia Stockholm, Sweden
June 16-18, 1997

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Optimisation of Circuits Using a Constructive Learning Algorithm

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Abstract—The paper presents an application of a constructive learning algorithm to optimisation of circuits. For a given Boolean function f, a fresh constructive learning algorithm builds circuits belonging to the smallest $F_{n,m}$ class of functions (n inputs and having m groups of ones in their truth table). The constructive proofs, which show how arbitrary Boolean functions can be implemented by this algorithm, are shortly enumerated. An interesting aspect is that the algorithm can be used for generating both classical Boolean circuits and threshold gate circuits (i.e. analogue inputs and digital outputs), or a mixture of them, thus taking advantage of mixed analogue/digital technologies. One illustrative example is detailed. The size and the area of the different circuits are compared (special cost functions can be used to closer estimate the area and the delay of VLSI implementations). Conclusions and further directions of research are ending the paper.

Keywords—constructive learning, circuit optimisation, VLSI complexity.

Introduction

In this paper we shall consider a neural network (NN) constructive algorithm (i.e. which determines both the number of layers and number of neurons, and their synaptic weights), which is guaranteed to converge in finite time. In particular, it can be used for synthesising Boolean circuits. We shall use the notations for feedforward NNs which classify m examples of n bits each $X_k = (x_0, ..., x_{n-1}) \in \mathbb{R}^n$, $k=1,\ldots,m$ where each neuron computes out $(In)=\sigma(\sum_{i=0}^{n-1}w_i\times In_i+\theta)$, with $w_i\in\mathbb{R}$ called the synaptic weights, $\theta \in \mathbb{R}$ known as the threshold, and σ a non-linear activation function. The cost functions commonly in use are depth and size. These can be linked to $T (\approx depth)$ and $A (\approx size)$ of a VLSI chip, but NNs do not closely follow such proportionalities (as the area of the connections counts and the area of one neuron is related to its associated weights, not to mention the complexity of the function realised by each neuron). That is why better alternative measures are: (i) the fan-in as the "area required for inter-node connectivity grows like the cube of a node's fan-in" [13]; (ii) the total number of connections [17]; (iii) the total number of bits needed to represent the weights [11, 20]; or even more precise approximations like (iv) the sum of all the weights and thresholds $\sum \left(\sum_{i=0}^{n-1} |w_i| + |\theta|\right)$ [3, 5], or $\sum \sum_{i=0}^{n-1} w_i^2$ [21]. Such cost measures can easily be understood as the *area* is related to the *weights* and the thresholds, as they have to be physically realised (transistors, conductances, resistances, etc.), and have been used as an optimum criteria for linear programming synthesis [16], or for defining the minimal integer realisation of one threshold gate (TG) [14].

A Direct Synthesis Algorithm

A novel synthesis algorithm has evolved from the decomposition of COMPARISON.

Proposition 1 [1] The computation of COMPARISON of two n-bit numbers can be realised by a Δ -ary tree of size $O(n/\Delta)$ and depth $O(\log n / \log \Delta)$ for any integer fan-in $2 \le \Delta \le 2n$.

Proposition 2 [1] The COMPARISON of two n-bit numbers can be computed by a Δ -ary tree neural network with polynomially bounded integer weights and thresholds, having size $O(n/\Delta)$ and depth $O(\log n/\log \Delta)$ for any integer fan-in (Δ) in the range 3 to $O(\log n)$.

Proposition 3 [3] The neural network with polynomially bounded integer weights (and thresholds) computing the COMPARISON of two n-bit numbers occupies an area of $O(n \cdot 2^{\Delta/2}/\Delta)$ for all the values of the fan-in (Δ) in the range 3 to $O(\log n)$.

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We have used these results for synthesising functions of n input variables having m groups of ones. **Proposition 4 [2, 3]** Any function $f \in F_{n,m}$ can be computed by a neural network with polynomially bounded integer weights (and thresholds) of depth $O[\log(mn)/\log\Delta]$ and size $O(mn/\Delta)$, and occupying $O(mn \cdot 2^{\Delta}/\Delta)$ area if $2m \le 2^{\Delta}$, for all the values of the fan-in (Δ) in the range 3 to $O(\log n)$. The well-known AT^2 becomes $O[(mn \cdot \log^2(mn) \cdot 2^{\Delta})/(\Delta \cdot \log^2\Delta)]$, which is minimised for small constant fan-in values 6...9 [7]. This method can be directly used to learn k functions from m examples. **Proposition 5 [3, 5]** Any finite set of k functions $f \in F_{n,i}$ defined by m examples ($i \le m \le ik$), can be computed by a neural network with polynomially bounded integer weights and thresholds having size $O[m(2n+k)/\Delta]$ and depth $O[\log(mn)/\log\Delta]$, and occupying $O(mn \cdot 2^{\Delta}/\Delta + mk)$ area, for all the values of the fan-in (Δ) in the range 3 to $O(\log n)$.

If $2m > 2^{\Delta}$ the tree structure has more COMPARATORS than all the different COMPARISONS (of limited fan-in), and by deleting the redundant ones the size and the area can be reduced by huge factors. **Proposition 6 [6]** The dichotomy of m examples from \mathbb{R}^n can always be solved with:

#bits
$$< m \cdot \lceil n \log (D/d) + 2.0471 n - (\log n)/2 - 0.8257 \rceil / 2$$
.

Proposition 7 [6] The entropy of a dichotomy of m examples from \mathbb{R}^n is bounded by $2^{m\lceil \log m \rceil}$. **Proposition 8 [9]** The dichotomy of $m = m_+ + m_-$ examples from \mathbb{R}^n can always be solved with:

#bits
$$< \max(m_+, m_-) \cdot [n \log(D/d) + 1.8396 n - 1.0800].$$

Proposition 9 [12] The number-of-bits necessary for the classification of m patterns in general position in \mathbb{R}^n using weights in the set $\{-p, -p+1, \ldots, 0, \ldots, p\}$ is:

$$\#bits > m n \cdot \lceil \log (2pD) \rceil = m n \cdot \lceil \log (D/d) \rceil$$
.

Proposition 10 [10] The dichotomy of $m = m_+ + m_-$ examples in general position in \mathbb{R}^n using weights in the set $\{-p, -p+1, \ldots, 0, \ldots, p\}$ requires:

$$\#bits > m \cdot \lceil n \log(D/d) + n + 0.6515 \rceil / 2$$
.

Proposition 11 [10] The dichotomy of $m = m_+ + m_-$ examples in general position in \mathbb{R}^n using weights in the set $\{-p, -p+1, \ldots, 0, \ldots, p\}$ requires:

#bits >
$$\max(m_+, m_-) \cdot \lceil n \log(D/d) - 0.4667n + \log n + 0.0665 \rceil$$
.

A first algorithm has been described [2, 4, 5, 8] as:

- 1. Quantize the space as described in *Propositions* 6 and 8; for the moment a logarithmic factor is lost as the intersection of two hypercubes is used instead of the intersection of two balls (see [6, 9, 19]).
- 2. From the set of examples, determine the 'constants' to be used by the COMPARATORS (the first layer).
- 3. Reduce the number of COMPARATORS by grouping together more hypercubes, or by sometimes using AND gates instead of COMPARATORS.

This is equivalent to finding hyperplanes parallel to the axes, which are COMPARISONS (first layer) with values deduced from the data-set. The desired function can be synthesised either by one more layer of TGs (in fact MAJORITY gates, see Fig. 1.a), or by an additional classical 2-layer AND-OR structure: a second hidden layer of AND gates (each gate corresponds to one hypercube, and at most 2 COMPARATORS are needed for each dimension), and a third layer of k OR gates represents the outputs (Fig. 1.b). For minimising the area some COMPARATORS can be replaced by AND gates [3] (see first layer in Fig. 1.b).

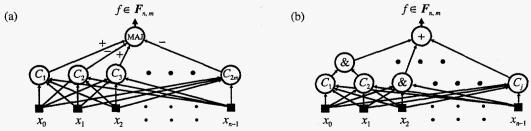


Figure 1. Two solutions for implementing $F_{n,m}$ functions: (a) COMPARISONS and one MAJORITY gate; (b) COMPARISONS or AND gates (for minimising size and area), followed by a classical AND-OR tree structure.

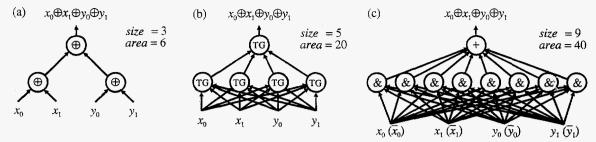


Figure 2. The PARITY problem: (a) solution with XORS; (b) TG solution; (c) classical AND-OR solution (brute force).

An immediate but significant extension for analog inputs has been to use analogue COMPARATORS in the first layer [5]. Such an extension—beside allowing analog inputs—has also several other advantages: (i) the overall area is reduced (a 2-input analog COMPARATOR is smaller than a digital COMPARATOR requiring several digital inputs); (ii) it is faster (smaller depth); (iii) dissipates less power (due to reduced area and fewer transistors needed to implement analog COMPARATORS). The only disadvantage is that the design and the technology is more complex.

Illustrative Example

As an example, we consider the PARITY function of four bits. Here we estimate the *area* as the sum of the *fan-ins* [17] (see [5, 8] for more precise cost functions). It is known that PARITY can be implemented with just three 2-input XOR gates (Fig. 2.a), or with five 4-input TGs (Fig. 2.b). It is also known that, in general, a 4-input Boolean function requires $2\sqrt{16} + 3 = 11$ TGs [15, 18]. A classical Boolean solution requires eight (one for each minterm) AND gates (Fig. 2.c). Using the constructive algorithm we obtain a brute force solution (see Fig. 3.a and 3.b); a hand crafted solution can be seen in Fig. 3.c.

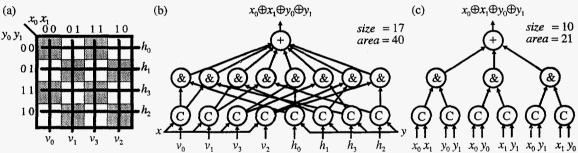


Figure 3. The PARITY problem: (a) Karnaugh map (for COMPARATORS); (b) brute force solution; (c) hand crafted.

If the inputs are considered 'analogues', the solution given by our algorithm is improved, and looks like the one in Fig. 4.b, or Fig. 4.c (using MAJORITY gates, the size could still be reduced).

Finally, going closer towards the TG solution from Fig. 2.b, we can allow COMPARATORS between

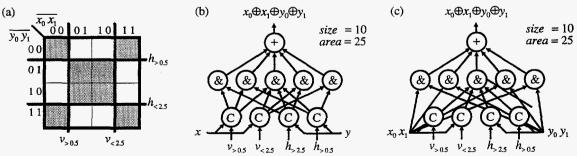


Figure 4. The PARITY problem: (a) Karnaugh map (for analogue COMPARATORS); (b) solution using COMPARATORS (with constants); (c) mixed solution using COMPARATORS with constants and classical digital circuitry.

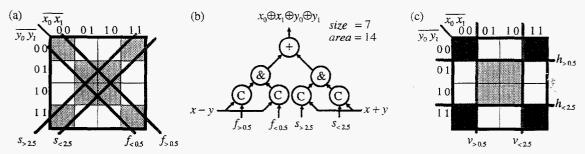


Figure 5. The PARITY problem: (a) Karnaugh map for analogue COMPARATORS between inputs 1; (b) the resulting circuit; (c) Karnaugh map for a possible mixed solution (hand crafted).

several inputs and a constant (e.g. the inputs are combined by a pre-processing layer of op-amps); this is equivalent to allowing hyperplanes which are equally spaces from the axes (Fig. 5.a). Now, the algorithm generates an even smaller circuit (Fig. 5.b); a mixed solution can be hand crafted (Fig. 5.c).

Conclusions

This paper has shown that constructive learning could be used for optimising circuits. Such an approach can take advantage of mixed analogue/digital implementations as it builds either TG circuits, Boolean circuits, or a mixture of them (for a solution to the two-spirals problem see [5, 8]). For the particular example considered, the solutions given by our algorithm can still be enhanced (see hand crafted versions), but the starting points are always in the solutions given by the algorithm. We are currently working on designing another constructive algorithm based on *Propositions 9*, 10 and 11.

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¹ Many thanks to Kalpak Dighe who suggested the particular approach from Fig. 5.a for the 'analogue' 4-input PARITY problem.