

AVOIDING INTERFERENCES IN PLANAR ARRAYS THROUGH THE USE OF ARTIFICIAL NEURAL NETWORKS

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Abstract

This article implements an Artificial Neural Network to find, through computer simulation, the excitations of a square planar array, composed by 52 uniformly spaced subarrays, with a quasi-null in its radiation diagram. This simulation model includes the reduction of any signal interference in the shaped radiating zone, after its position has been determined.

Introduction

In a previous article [1], the *Radiating System Group*¹ developed a synthesis method, which positions a quasi-null in the global beam radiation diagram of an array installed in a geostationary satellite, composed by 52 active and 12 inactive subarrays. That method has been accomplished through the use of the *simulated annealing* technique [2], an easy-implementing and very efficient tool widely utilized in computer-aided simulations. This method, although it has advantages, is too slow to be implemented in the array itself using real-time calculations. The algorithm of the cited article takes, in a desktop computer with an Athlon XP 1800 processor running at 1.53 GHz, about 40 minutes to find a solution.

Lately, the progress in the application of Artificial Neural Networks (ANN) in a variety of fields makes possible their use in signal processing, and synthesis or optimization for radiating problems. As interesting examples, Zooghy, Christodoulou et al. [3-4], implemented and then validated experimentally an ANN direction finder. A later publication of the same authors [5] presented a Radial Basis Function Neural Network (RBFNN) that can reject interferences through adaptive cancellation in circular arrays.

In this article a computer-aided simulating method that implements an ANN is presented. After its training, the output vector of this ANN specifies the excitation distribution of the array studied in article [1], taking as input vector a desired radiation mapping over a predetermined angular positions in space, with a quasi-null positioned in a pre-established direction. The fundamental advantage of this method, compared with that described in [5] for circular arrays, is that the radiation diagram of the antenna has low degradation before the quasi-null has been positioned, as seen below.

The method

The previous article [1] specifies the configuration and the relevant parameters of the antenna used in this work. Table 1 summarizes and lists its parameters. In this table, the reception and transmission modes are denoted as Rx and Tx , respectively. When the antenna is installed on a geostationary satellite, it can generate eight fundamental radiation patterns. These are described as: Rx mode, a global beam to cover a specified zone on the Earth's surface; one fixed pencil beam, that points to a certain pre-established angular direction; and finally two movable beams with the capability of scanning the complete hemisphere of the earth. The Tx mode uses the same number of beams and specifications as the Rx mode. The coverage zone, that is, the zone of illumination of the global beam of the antenna, is the cone enclosed by the variation of (θ, ϕ) coordinates spatial angles, with $0 \leq \theta \leq \theta_c$ and $0 \leq \phi \leq 2\pi$ (θ measured from the zenith of the plane of the array). Figure 1 represents antenna dimensions and its active and passive subarrays. It can be seen that 12 elements placed on the vertices of the square were inactivated.

For our requirements, a power pattern quasi-null is positioned at pre-established angles (θ_{qn}, ϕ_{qn}) inside the coverage zone of the global beam of the antenna, at the center frequency of Rx band. After this, a mapping of 121=11x11 points in the appropriate zone is taken, considering uniformly spaced (u, v) positions, where the transformations:

$$\begin{aligned} u &= \sin \theta \cos \phi \\ v &= \sin \theta \sin \phi \end{aligned} \tag{1}$$

replace the (θ, ϕ) values. The choice of $-0.2 \leq u \leq 0.2$ and $-0.2 \leq v \leq 0.2$, includes the (θ, ϕ) values within the coverage zone.

By definition, the synthesis problem in array design consists of finding the excitation distribution of the antenna elements, when a power -or field- pattern, with specified characteristics established by the designer. On the other hand, in the analysis problem, when an excitation distribution is given, the power -or field- pattern is obtained with an empirical or analytical formulation. Subsequently, the synthesis problem can be viewed as the inverse formulation for a certain (previously unknown) excitation distribution. Some authors, as Funahashi [6], have stated that neural networks with non-linear transfer functions can be considered as function universal approximators. Based on these justifications and on their own experience [7-9], a Multilayer Perceptron feed-forward architecture network (MLP) is proposed in this work by the group: *Artificial Neural Networks and Adaptive Systems Laboratory*² in order to obtain the excitation distribution of the array when a quasi-null position has been established in its power pattern.

Application

For our purpose, a four-layered perceptron gives high-quality results, with the mapping of the power pattern (measured in dB) as input vector, and denoted as \mathbf{P}_p (with $1 \leq p \leq 121$). The 52-element excitation distribution is separated into real and imaginary parts, as output vector $\mathbf{Re}_q, \mathbf{Im}_q$ ($1 \leq q \leq 52$), and two hidden layers: \mathbf{h}_r^1 ($1 \leq r \leq 100$) and \mathbf{h}_s^2 ($1 \leq s \leq 80$). The training process of this particular ANN has been performed with 200 pairs of input-output vectors. In the validation process, 25 pairs were used. For better results, two ANN were constructed separately, with their processes of training-validation carried out as follows: on the one hand, an ANN with the \mathbf{P}_p and \mathbf{Re}_q as input and output vectors, respectively, and, on the other, with \mathbf{P}_p and \mathbf{Im}_q as before. Figure 2 gives the schematic of each ANN, with the generic output vector denoted as \mathbf{E}_q repre-

senting the \mathbf{Re}_q or \mathbf{Im}_q vector, as corresponds. Table 2 summarizes the fundamental characteristics of each ANN, and the errors obtained in their training and validation processes. Figure 3 represents an example of the desired power pattern, with a quasi-null positioned at $(u_{qn} = 0.08, v_{qn} = 0.08)$ and at -10.84 dB below the maximum power. The ripple in the coverage zone is ± 0.91 dB. Figure 4 represents the ANN response for the same case, after the training process has been completed, with a ripple slightly degraded to ± 1.15 dB, and the depth of the quasi-null raised at -10.17 dB.

Conclusions

The architecture of the multilayer perceptron presented here allows rejection of some interference in the coverage zone of a planar array radiation pattern, when its position has been established. Although the quasi-null positioning was moved within the $-0.1 \leq u \leq 0.1$ and $-0.1 \leq v \leq 0.1$ interval, before the training and validation processes, and in order to obtain the 225 pairs of input-output vectors, this must not be seen as a data base itself. The Δu and Δv of the equispaced angular points are large enough to admit quasi-null locations between them that change the excitation distribution of the array in a manner that can not be neglected.

The training of each network takes about 4 minutes, in a computer with an Athlon XP 1800 processor, running at 1.53 GHz. In the simulation process, after the neural network has been trained, the positioning of the quasi-null in a desired location can be done quickly (less than one second in the same PC), and with high-performance response of the radiation diagram. It can be seen that, as an immediate application, this scope could be expanded through the hardware implementation of the ANN in the array itself. As the ANN presented here represents, in a discretized way, the inverse equation of a radiation pattern of a planar antenna. This technique could be used, in a further in-

vestigation, to diagnosis array faults by using measured values of its own far-field amplitude [10], its input impedance [11] and / or its mutual coupling [12]. It would be very useful method for antennas permanently or semi-permanently deployed in space, and for any case that requires the use of costly near-field measurement facilities.

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TABLES AND FIGURES

Table 1. Antenna relevant parameters

Active subarrays	52
Passive subarrays	12
Number of elements of each subarray	8
Antenna dimensions	600 x 600 mm ²
Distance between subarrays	75 mm
Rx Band	7.90-8.15 (GHz)
Tx Band	7.25-7.50 (GHz)
Coverage zone	$\theta_c = 8.7^\circ$ (deg)

Table 2. Parameters of the two performed MLP and errors obtained in training and validation processes.

PARAMETER	VALUE
Network architecture	Multilayer Perceptron
Network training function	Resilient Backpropagation Algorithm
Number of input neurons	121
Number of neurons in first hidden layer	100
Number of neurons in second hidden layer	80
Number of neurons in output layer	52 (each MLP)
Transfer functions	Hyperbolic Tangent Sigmoid
Training error (real output vectors)	1.13%
Training error (imaginary output vectors)	1.37%
Validation error (real output vectors)	3.43%
Validation error (imaginary output vectors)	4.12%

Figure 1. Antenna spatial configuration. Filled squares represent active subarrays.

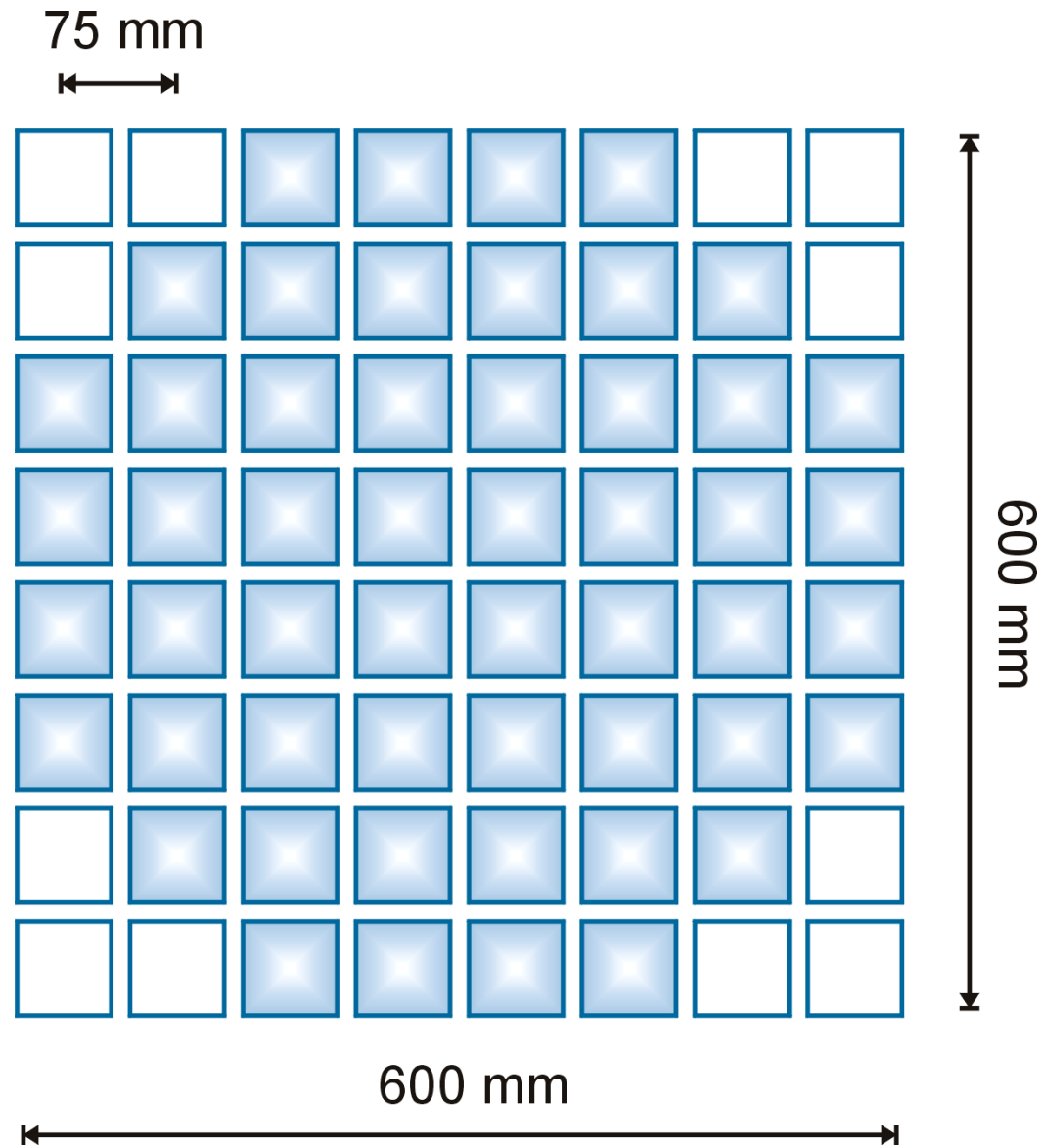


Figure 2. Architecture of the MLP feed-forward network used in this work.

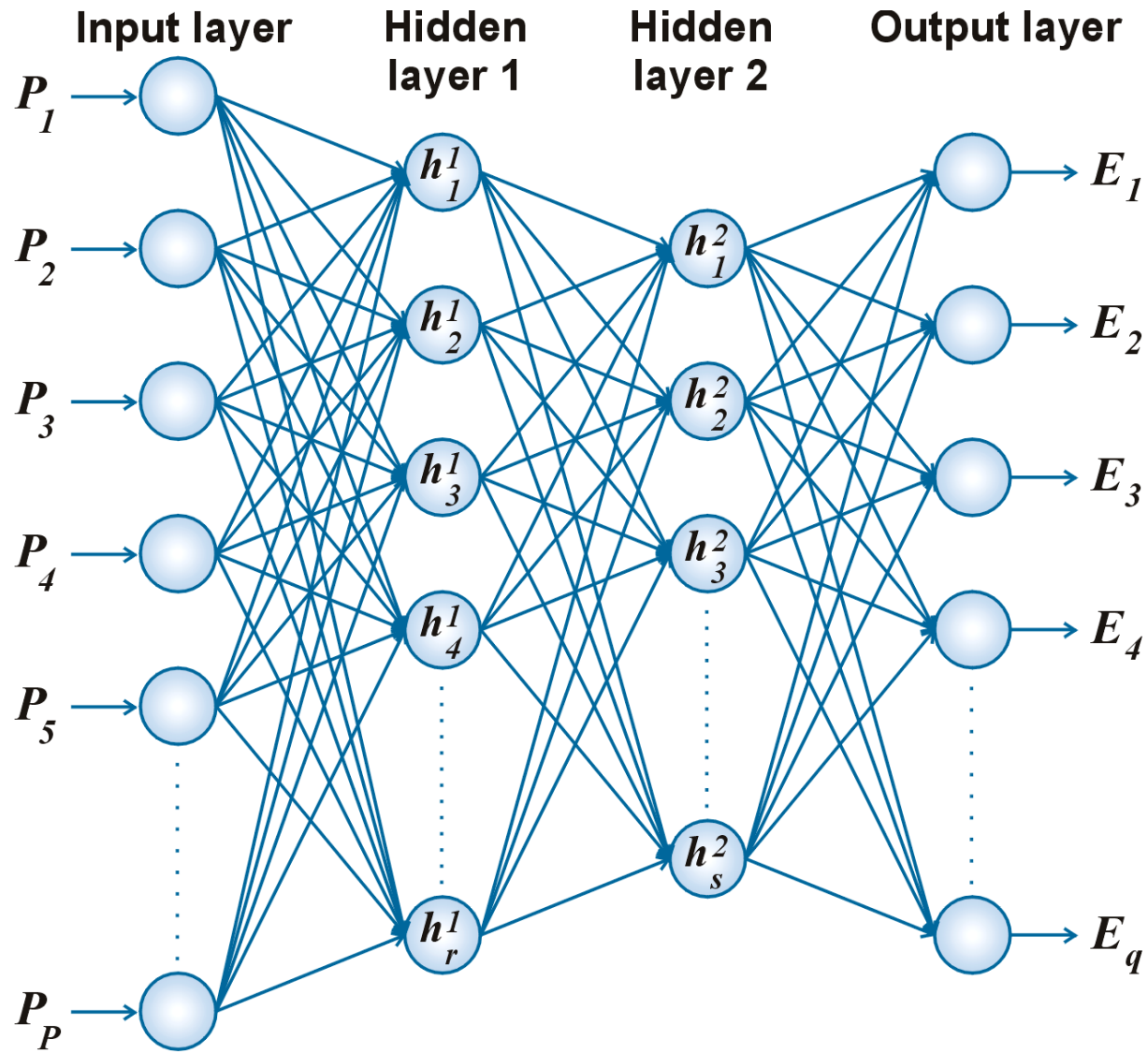


Figure 3. Desired power pattern taken with the quasi-null position at $(u_{qn}=0.08, v_{qn}=0.08)$ and depressed -10.84 dB below the maximum of the coverage zone. Ripple obtained: ± 0.91 dB.

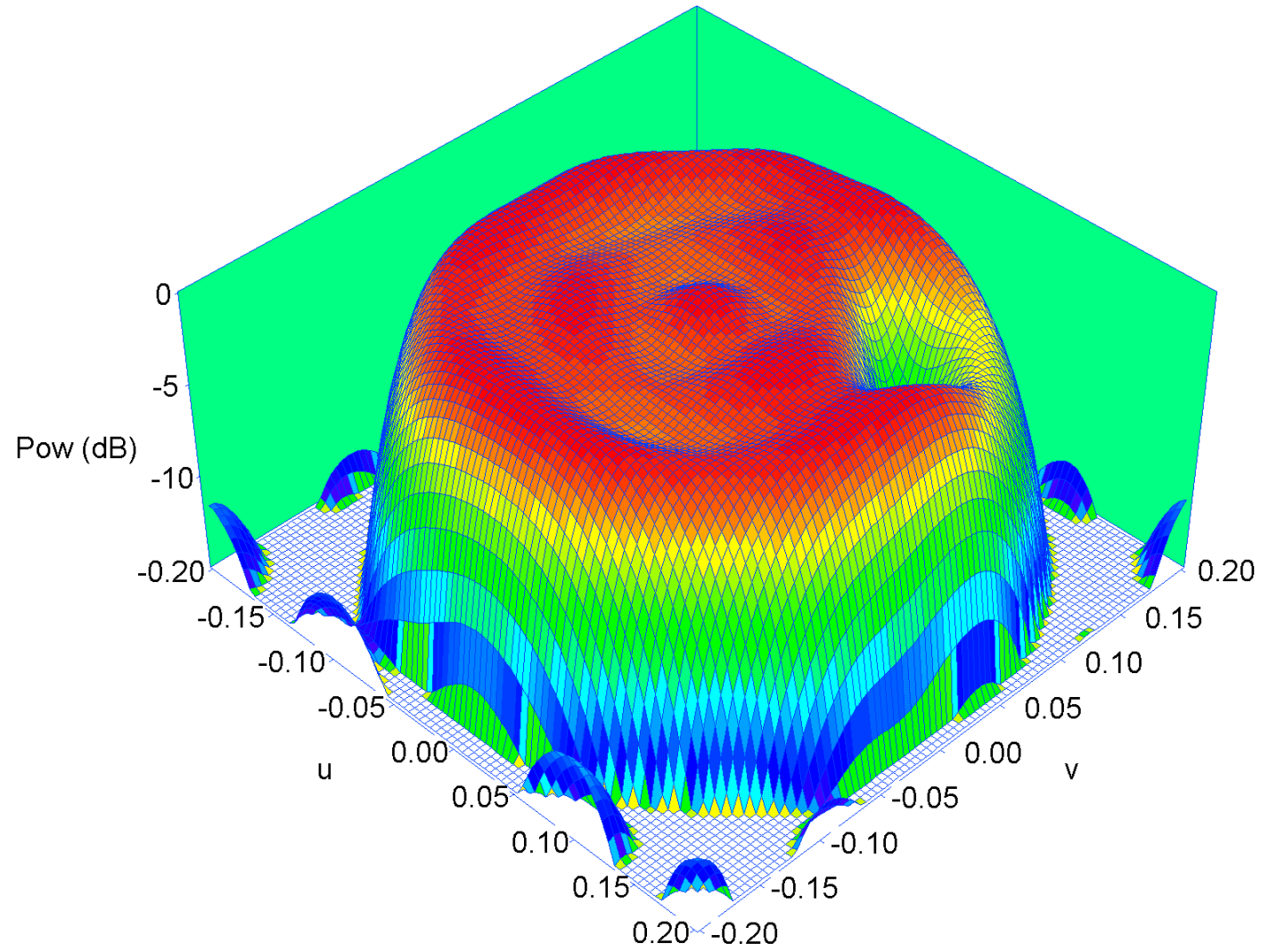


Figure 4. Power pattern obtained after the neural network has been trained, corresponding this example to that given in figure 3. Quasi-null depressed -10.17 dB below the maximum. Ripple obtained: ± 1.15 dB.

