



Global Versus Local Lyapunov Approach Used in Disturbance Observer-Based Wind Turbine Control

Eckhard Gauterin*, Florian Pöschke and Horst Schulte

Control Engineering Group, Department of Engineering I, University of Applied Sciences HTW Berlin, Berlin, Germany

This contribution presents a Lyapunov-based controller and observer design method to achieve an effective design process for more dedicated closed-loop dynamics, i.e., a maximal flexibility in an observer-based controller design with a large consistency in desired and achieved closed-loop system dynamics is intended. The proposed, pragmatic approach enhances the scope for controller and observer design by using local instead of global Lyapunov functions, beneficial for systems with widely spaced pole locations. Within this contribution, the proposed design approach is applied to the complex control design task of wind turbine control. As the mechanical loads that affect the wind turbine components are very sensitive to the closed-loop system dynamic, a maximum flexibility in the control design is necessary for an appropriate wind turbine controller performance. Therefore, the implication of the local Lyapunov approach for an effective control design in the Takagi-Sugeno framework is discussed based on the sensitivity of the closed-loop pole locations and resulting mechanical loads to a variation of the design parameters.

Keywords: global and local Lyapunov approach, Takagi-Sugeno framework, model-based controller and observer design, feedforward-feedback control, linear-matrix-inequality and pole region-based controller design, wind turbine application, elaborated wind turbine simulation model, load analysis

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*Correspondence:

Eckhard Gauterin
gauterin@htw-berlin.de

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1 INTRODUCTION

The mechanical loads, affecting a wind turbine (WT), are very sensitive to the closed-loop system dynamics. Hence, for the design of an appropriate WT controller, a maximal flexibility is necessary to mitigate the resulting, mechanical loads. For this purpose, model-based and automated controller optimisation procedures are recommended. Until now, the authors achieved the intended flexibility with decomposed, structural dynamic models of the wind turbine (e.g., (Pöschke et al., 2020)) and with an observer-based controller structure (Gauterin et al. (2014), Pöschke et al. (2019)), whose separately designed observer and controller are based on a common and global Lyapunov approach, respectively. With the local Lyapunov approach, conceived in the outlook of (Pöschke et al., 2022) and presented in this work the first time, the evolution of the design process with an increased controller flexibility and improved consistency is proceeded by the implementation of a more effective controller design procedure.

In control theory, the Lyapunov approach (Lyapunov, 1992) is utilised for controller synthesis, i.e., to analyse the stability of closed-loop systems and simultaneously providing the related controller gains. Within a model-based control design, the Lyapunov approach enables an automated design process.

As most real world systems are characterised by complex, nonlinear dynamics, techniques to analyse stability and dynamical characteristics are needed. For this, the Takagi-Sugeno (TS)

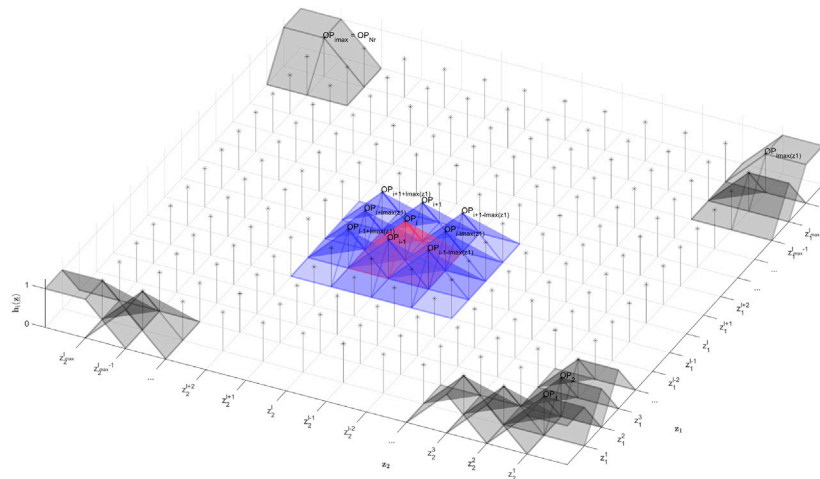


FIGURE 1 | Illustrative example for pyramid-shaped membership functions $h_i(\underline{z}) = h_i(z_1, z_2)$ with the two premise variables $\underline{z} = [z_1, z_2]^T$ (formed by the triangular-shaped weighting functions $w_{1,i}(z_1)$ and $w_{2,i}(z_2)$, which are discretised in $l(z_i) \in [1, l_{\max}(z_i)]$ linearisation points; see **Eq. 2** and exemplary visualisation of the eight direct adjacent steady Operating Points (and linearisation points, respectively) $OP_{i-1+l_{\max}(z_1)}$ to $OP_{i+1+l_{\max}(z_1)}$ adjoining OP_i with their membership functions (see the overlapping parts of the nine coloured pyramids) $h_{i-1+l_{\max}(z_1)}(z_1^{l-1}, z_2^{l-1})$ to $h_{i+1+l_{\max}(z_1)}(z_1^{l+1}, z_2^{l+1})$ (with $h_i(\underline{z}) \geq 0$ within $z_1^{l-2} \leq z_1 \leq z_1^{l+2}$ and $z_2^{l-2} \leq z_2 \leq z_2^{l+2}$ for each of the nine steady operating points, while $h_i(\underline{z}) = 0$ holds for all the other steady operation points and \underline{z} , respectively; with $i = (l(z_2) - 1)_{\max}(z_1) + l(z_1)$

framework (Takagi and Sugeno, 1985) may be used, that describes a nonlinear system as convexly blended linear submodels (A_i, B_i) (Tanaka and Sano, 1994a). A thorough discussion on the TS methodology is given in (Tanaka and Wang, 2001) and with a focus on observer-based methods in (Lendek et al., 2010). To form the TS model structure, the individual linear submodels (A_i, B_i) may be gained with the sector nonlinearity approach (Tanaka and Sano, 1994b), which yields an exact representation of the nonlinear system, or by linearisation (Johansen et al., 2000). As the identification and analysis of numerically derived linearised models is an established approach to investigate control properties in the wind turbine application (Bossanyi, 2000), the linearisation approach is used within this proceeding, too. To facilitate this, aero-elastic simulation programs like NREL FAST (Jonkman and Buhl, 2005; Jonkman, 2016) provide a linearisation feature to easily obtain the used matrices A_i, B_i as discussed in (Jonkman and Jonkman, 2016). The resulting TS model allows for the analysis of the dynamical properties and stability of both, the open- and closed-loop system.

Using the inequality of the Lyapunov approach on these convexly blended combinations of linear submodels in its matrix formulation, linear matrix inequalities (LMI, (Boyd et al., 1994)) are derived that describe the *stability condition* of the system dynamics (Tanaka and Wang (1997), Tanaka and Wang (2001), Lendek et al. (2010)). Additionally, *performance constraints*, in form of physically interpretable pole regions, can be specified and formulated in terms of LMIs (Chilali and Pascal, 1996). The combination of the stability condition with the performance constraints forms a collection of LMIs, which can be efficiently solved with numerical LMI solvers (VanAntwerp and Braatz, 2000). The LMIs' solution space and thereby the solution's conservativeness is restricted by the stability condition and the number of performance constraints taken into account. However, the conservativeness may be influenced, e.g.,

by the way the TS model is constructed and/or the use of relaxations in the resulting LMI (Tanaka et al. (1998), Tanaka and Wang (2001)).

For an observer-based controller, a separated controller and observer design can be used. The corresponding separation principle is also valid for observer-based TS controllers (Yoneyama et al. (1998), Ma et al. (1998)), but does not necessarily hold for parameter uncertainties or stochastic noise. Therefore, design procedures are investigated, which account for these restrictions (Zemouche et al., 2016), (Rauh et al., 2021). As an overall observer and controller design often results in conservative controller and control objective performances, respectively, pragmatic design syntheses are intended for real world application, rather than a guaranteed overall stability of the observer-based controller. From engineering point of view, a controller with guaranteed, overall system stability does not ensure stability for the closed-loop *real world* system dynamics, as the controller design model cannot cover *all* uncertainties (e.g., resulting from unpredicted or non-modelled environmental influences).

With the proposed local Lyapunov approach, a pragmatic controller design procedure is introduced, that utilises a separated controller and observer design and reduces the stability conditions from *global stability conditions* (of the nonlinear system within the defined operational range of the TS description) to *local stability conditions* (at each considered operating point, i.e., small-signal stability). With the resulting reduction of the number of parallel to be solved LMIs, the constraints for the solver are reduced and thereby the flexibility in assigning desired pole locations of the closed-loop system is increased, i.e., the pole region bound modifiability and flexibility, respectively is significantly improved. The less conservative task for the LMI solver enables the specification of tighter performance constraints (e.g., smaller pole regions) and reaches an increased consistency in desired and achieved closed-loop dynamics. The

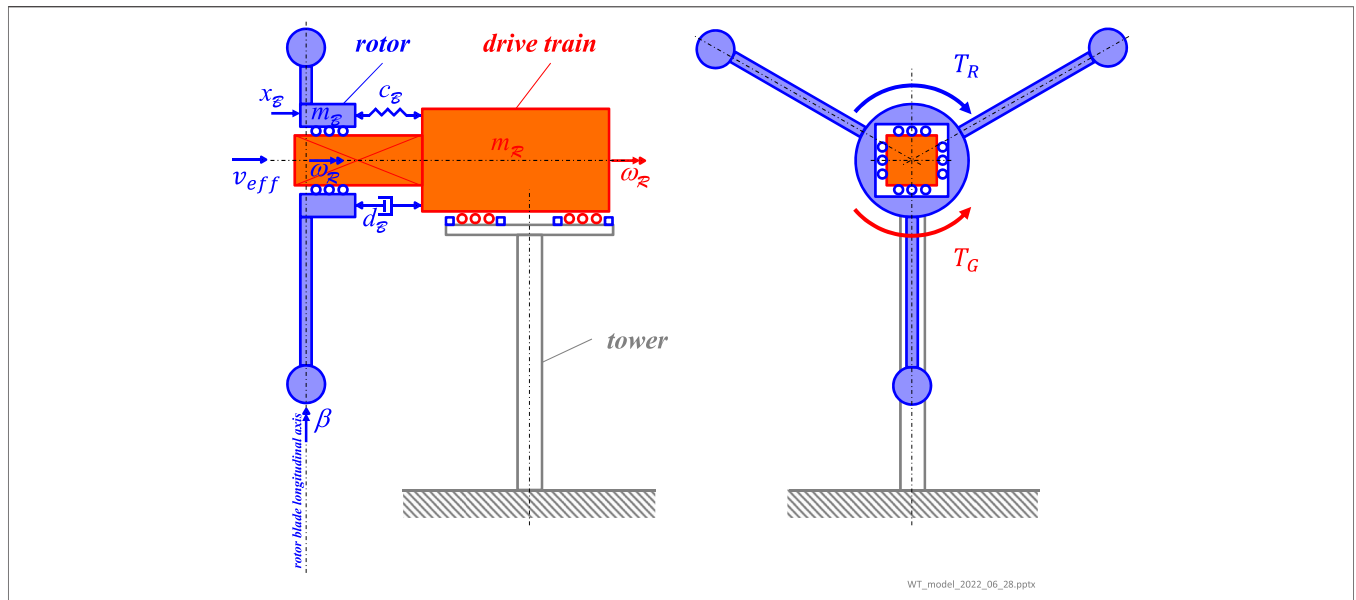


FIGURE 2 | Structural dynamical design model of a wind turbine (in side- and front-view) with the two degree of freedom:

- blade translation x_B (in up- and downwind direction) and
- drive train dynamic, i.e., rotor and generator rotation (speed) ω_R

for the rigid body model of the rotor and drive train.

The blue depicted ball bearings enable just the translation x_B of the discrete rotor model in relation to the discrete drive train model (i.e., in wind direction), while the red depicted bearings enable just the drive train rotation ω_R along its horizontal axis and suspend any other coupling of drive train rotation and tower translation.

Linear, single headed arrows represent translations, forces or translatory spring- and damper-elements, while linear, double headed arrows represent rotations.

The parameters m_K , c_K and d_K represent the discrete component masses, discrete spring coefficients and discrete damping coefficients of the discrete component model K (with $K = B$ (lade) or $K = R$ (otation) of the drive train.)

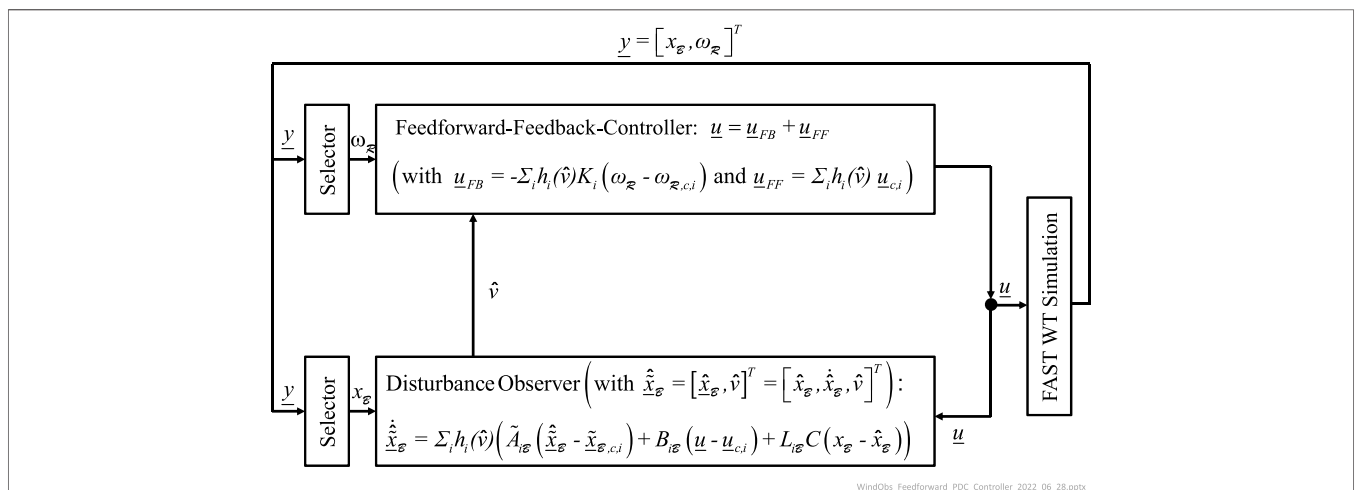
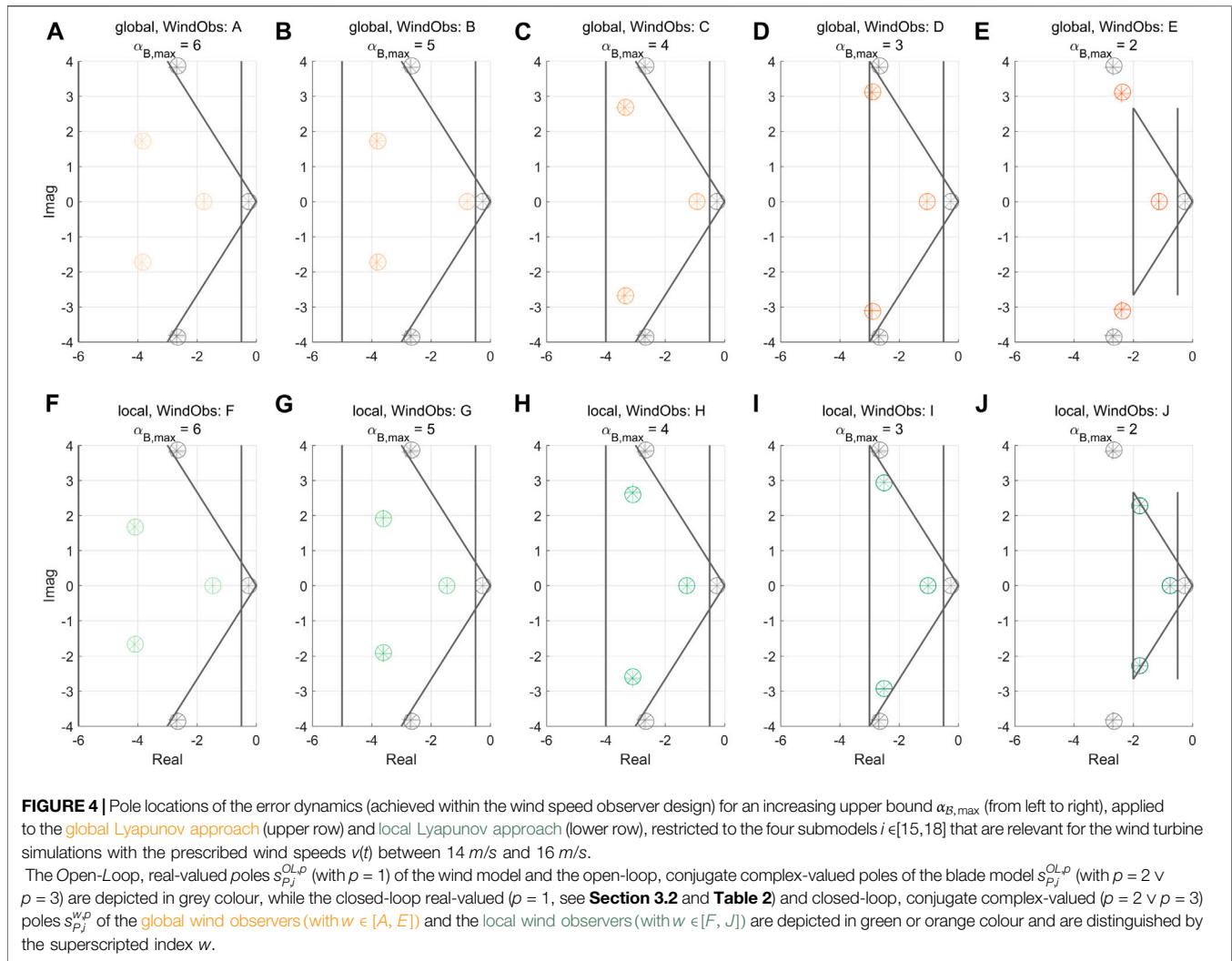


FIGURE 3 | Block digram of the observer-based Takagi-Sugeno wind turbine controller.

increased consistency and increased flexibility, are hereinafter denoted as the *general objective* of the local Lyapunov approach (see **Section 2.3**). Hence, the pragmatic local Lyapunov approach provides an enhanced design scope for real world applications and results in a control design method, which is highly effective in posing more dedicated closed-loop dynamics, especially beneficial for systems with widely spaced pole locations.

Within this contribution the concept of a local Lyapunov approach is described in detail the first time and applied for wind turbine (WT) systems, characterised by widely spaced pole locations of the open-loop system, due to the divergent stiffness, damping and inertia of the main components, like rotor blades and tower. Also for these challenging system dynamics, WT controllers in a TS framework have



demonstrated to be capable for energy yield optimisation, load mitigation and active power reduction (Pöschke et al., 2020) or fault tolerant control (Georg (2015), Schulte and Gauterin (2015)). The utilised TS WT system model is achieved by linearising an elaborated WT simulation model (the NREL FAST 5 MW reference WT model (Jonkman et al., 2009)). Thereby, all mechanical couplings between the WT main components' degree of freedom are neglected to increase the flexibility for the controller design. That is, the overall, decomposed TS model comprises decoupled TS WT main component models (with each TS WT main component model consisting of convexly blended linear submodels) that are achieved from linearisation and afterwards combined to the overall, decomposed TS WT model.—As a precise wind speed measurement is hardly feasible with conventional, cost-effective anemometers on WT (e.g., due to high turbulences induced by the rotor (Ostergaard et al., 2007) and the significant variation of the local wind speeds within the enormous size of the rotor swept area), advanced, expensive sensors (like LiDaR systems (Schlipf et al., 2010)) and observer techniques are investigated

for the WT application ((Ma et al. (1995), Ostergaard et al. (2007), Jena and Rajendran (2015), Gauterin et al. (2015)). Within this contribution, a TS disturbance observer is used for an estimation of the unknown wind speed as premise variable, enabling a premise variable scheduled feedforward actuation for disturbance attenuation (Gauterin et al., 2014), while the—also premise variable scheduled—feedback-controller just compensates the control signal deviations resulting from model-uncertainties. To evaluate the observer-based TS feedforward-feedback controller achieved with the local Lyapunov observer design approach, the NREL FAST 5 MW reference WT model is used for WT operation simulation.

The paper is organised as follows: In **Section 2 Method**, the TS framework (**Section 2.1**), the global and local Lyapunov approach (**Section 2.2** and **Section 2.3**) and its application to WT control (**Section 2.4**) with additional performance constraints (**Section 2.5**) is introduced. **Section 3 Simulations and Results** describes the simulation design (**Section 3.1**) and the achieved results (**Section 3.2**). Finally, in **Section 4 Discussion** the achieved results are assessed and a **Conclusion** is drawn in **Section 5**.

2 METHOD: GLOBAL AND LOCAL LYAPUNOV APPROACH

In this Section the Takagi-Sugeno (TS) framework (Section 2.1) and the global Lyapunov approach are presented (Section 2.2). In Section 2.3 the local Lyapunov approach is introduced to system models described in the TS framework. Its application for wind turbine control is presented in Section 2.4 and additional controller design constraints to define pole regions with regard to wind turbine application are described in Section 2.5.

2.1 System Model, Controller and Observer in Takagi-Sugeno Framework

2.1.1 System Model

In this proceeding, a nonlinear system $\dot{\underline{x}} = f(\underline{x}, \underline{u})$ is described in the Takagi-Sugeno framework by convexly blended linear submodels, based on Taylor linearisation of the nonlinear system.

For the wind turbine (WT) application, the plant model is generated by linearising the nonlinear system model at Nr steady state operating points (piecewise equidistant regarding the disturbing wind speed v), resulting in a set of Nr linear submodels (A_i, B_i) (with $i \in [1, i_{\max}]$ and $i_{\max} \equiv Nr$). These submodels comprise the state matrices A_i and input matrices B_i for the state vector \underline{x} and input vector \underline{u} and their steady state values $\underline{x}_{c,i}$ and $\underline{u}_{c,i}$. In wind turbine application the nonlinearity mainly results from the aerodynamics of the rotor blades.

The linear submodels (A_i, B_i) are blended in a convex sum

$$\begin{aligned} \dot{\underline{x}} &= \sum_i^{Nr} h_i(\underline{z})(A_i(\underline{x} - \underline{x}_{c,i}) + B_i(\underline{u} - \underline{u}_{c,i})) \quad \text{with} \\ \underline{y} &= C\underline{x} \quad \text{and} \quad \sum_i^{Nr} h_i(\underline{z}) = 1 \quad (\text{with } 0 \leq h_i(\underline{z}) \leq 1) \end{aligned} \quad (1)$$

with the help of membership functions $h_i(\underline{z})$, which depend on the premise variable \underline{z} . Within this proceeding, the membership functions $h_i(\underline{z})$ are defined by triangular-shaped weighting functions $w_{k,l}(z_k)$ (with the actual premise variable $z_k(t)$ and z_k , respectively between the discrete linearisation points OP_i for z_k^{l-1}, z_k^l and z_k^{l+1} , see Figure 1) for each of the $k \in [1, k_{\max}]$ premise variables z_k (discretised in $l(z_k) \in [1, l_{\max}(z_k)]$ linearisation points with

$$w_{k,l}(z_k) = \begin{cases} \frac{z_k - z_k^{l-1}}{z_k^l - z_k^{l-1}} & \text{if } z_k^{l-1} < z_k \leq z_k^l \\ 1 - \frac{z_k - z_k^l}{z_k^{l+1} - z_k^l} & \text{if } z_k^l < z_k \leq z_k^{l+1} \\ 0 & \text{else.} \end{cases} \quad (2)$$

The weighting functions $w_{k,l}(z_k)$ are combined to form the $(i_{\max} \equiv) Nr = \prod_{k=1}^{k_{\max}} l_{\max}(z_k)$ membership functions $h_i(\underline{z})$ (with $i \in [1, Nr]$):

$$\sum_{i=1}^{Nr} h_i(\underline{z}(t)) = \prod_{k=1}^{k_{\max}} \sum_{l=1}^{l_{\max}(z_k)} w_{k,l}(z_k). \quad (3)$$

In this contribution, the membership functions $h_i(\underline{z})$ just blend the i submodels, which are *direct adjacent* to the actual operation point (with $h_i(\underline{z}) \neq 0$), while the membership functions of all other models are set to zero (i.e., $h_i(\underline{z}) = 0$), as described in e.g., (Pöschke et al., 2020) and illustrated in Figure 1.

In WT application, it is advantageous to define the reconstructed wind speed \hat{v} as the premise variable $\underline{z} \equiv \hat{v}$.

Within this contribution *individual* input-matrices B_i , e.g., depending on the wind speeds v and rotor rotation speed ω_R , and a *common* output-matrix $C_i \equiv C$, just describing the measurable states \underline{y} , are supposed.

2.1.2 Parallel-Distributed-Compensation-Controller With Feedforward Actuation

In the TS framework, a state-space controller $\underline{u} = -K\underline{x}$ with convexly blended state-feedback gains K_j —so-called *Parallel Distributed Compensation (PDC) Controller* (Wang et al. (1995), Tanaka and Wang (1997))—is usually used:

$$\underline{u} = - \sum_j^{Nr} h_j(\underline{z}) K_j (\underline{x} - \underline{x}_{c,j}). \quad (4)$$

In Eq. 4 and the following, it is assumed, that $\hat{\underline{z}} \rightarrow \underline{z}$ holds (as explained in Section 2.4.4).

If the disturbance attenuation is realised with a feedforward actuation, the PDC controller (Eq. 4) is extended to

$$\begin{aligned} \underline{u} &= - \underbrace{\sum_j^{Nr} h_j(\underline{z}) K_j (\underline{x} - \underline{x}_{c,j})}_{\underline{u}_{FB}} + \underbrace{\sum_j^{Nr} h_j(\underline{z}) \underline{u}_{c,j}}_{\underline{u}_{FF}} \\ &= - \sum_j^{Nr} h_j(\underline{z}) (K_j (\underline{x} - \underline{x}_{c,j}) - \underline{u}_{c,j}). \end{aligned} \quad (5)$$

With the feedforward signal \underline{u}_{FF} the disturbance is attenuated, while the feedback signal \underline{u}_{FB} compensates control errors resulting from design model uncertainties. With Eq. 5 and $\sum_i^{Nr} h_i(\underline{z}) = 1$ in Eq. 1 the closed-loop dynamics (for i individual input matrices B_i , see Section 2.1.1)¹ is described by

$$\begin{aligned} \dot{\underline{x}} &= \sum_i^{Nr} h_i(\underline{z}) \left(A_i (\underline{x} - \underline{x}_{c,i}) + B_i \left(- \sum_j^{Nr} h_j(\underline{z}) (K_j (\underline{x} - \underline{x}_{c,j}) - \underline{u}_{c,j}) \right) - \underline{u}_{c,i} \right) \\ &= \sum_i^{Nr} \sum_j^{Nr} h_i(\underline{z}) h_j(\underline{z}) (A_i - B_i K_j) (\underline{x} - \underline{x}_{c,i}). \end{aligned} \quad (6)$$

2.1.3 Takagi-Sugeno Observer

For nonlinear systems $\dot{\underline{x}} = f(\underline{x}, \underline{u})$ a weighted combination of linear Luenberger observers (Luenberger, (1971), Luenberger (1964)) can be used, which is denoted as Takagi-Sugeno Observer (TSO). The TS observer is obtained from the TS system Eq. 1 by introducing the

¹With *individual* input matrices B_i a weighted combination of the i th submodel with *all* and the *direct adjacent* submodels (see Section 2.1.1), respectively, is derived for the closed-loop dynamics, due to the individual weighting $\sum_i h_i(\underline{z}) B_i$ of the individual input matrices B_i for each submodel. Therefore, the double summation $\sum_i \sum_j$ is necessary in Eq. 6.

output error-feedback term $L_i(\underline{y} - \hat{\underline{y}})$ (Tanaka and Sano, 1994a). For a *common* output matrix $C_i \equiv C$ (see **Section 2.1.1**) it holds²:

$$\begin{aligned} \dot{\hat{\underline{x}}} &= \sum_i^{N_r} h_i(\underline{z})(A_i(\hat{\underline{x}} - \underline{x}_{ci}) + B_i(\underline{u} - \underline{u}_{ci})) + \sum_i^{N_r} h_i(\underline{z})(L_i(\underline{y} - \hat{\underline{y}})) \quad \left| \hat{\underline{y}} = C\hat{\underline{x}} \right. \\ &= \sum_i^{N_r} h_i(\underline{z})(A_i(\hat{\underline{x}} - \underline{x}_{ci}) + B_i(\underline{u} - \underline{u}_{ci}) + L_i C(\underline{x} - \hat{\underline{x}})) \end{aligned} \quad (7)$$

with the reconstructed states $\hat{\underline{x}}$, the reconstructed outputs $\hat{\underline{y}}$ and the error-feedback gains L_i .

For the error dynamics \underline{e}

$$\underline{e} := \underline{x} - \hat{\underline{x}} \Rightarrow \dot{\underline{e}} = \dot{\underline{x}} - \dot{\hat{\underline{x}}} \quad (8)$$

it follows with **Eqs 7, 1** in **Eq. 8** (with the *common* output matrix C , see **Section 2.1.1** and **Section 2.2.2**):

$$\begin{aligned} \dot{\underline{e}} &= \sum_i^{N_r} h_i(\underline{z})(A_i(\underline{x} - \hat{\underline{x}}) + L_i C(\underline{x} - \hat{\underline{x}})) \\ &= \sum_i^{N_r} h_i(\underline{z})(A_i + L_i C) \underbrace{(\underline{x} - \hat{\underline{x}})}_{\underline{e}}. \end{aligned} \quad (9)$$

Within this contribution (and the previously published works in (Gauterin et al., 2014) and (Pöschke et al., 2020)) the TS-observer is implemented to reconstruct the disturbing wind speed $\hat{\underline{v}}$, that is used as the premise variable \underline{z} , scheduling the feedforward and feedback signal (see **Eq. 5** and **Figure 3** with $\underline{z} = \hat{\underline{v}}$). Therefore, the TS-observer does not represent a typical *disturbance* observer for explicit disturbance rejection, rather than a *premise* observer to reconstruct the disturbance signal $\hat{\underline{v}}$ as premise variable \underline{z} , that is used to influence the controllable system inputs with the premise variable governed control signal scheduling (see explanation for **Eq. 5** and **Section 2.4.3**). For the wind turbine application, the premise variable \underline{z} often comprises the reconstructed, *disturbing* wind speed $\hat{\underline{v}}$ (within this contribution, the premise variable consists just of the reconstructed wind speed $\hat{\underline{v}}$, i.e. $\underline{z} = \hat{\underline{v}}$ holds), therefore the premise observer is also denoted as *disturbance* observer in the following (see also the observer classification in **Section 2.4.3**).

2.2 Global Lyapunov Approach Based Takagi-Sugeno Controller and Observer Design

The state-feedback gains K_j and error-feedback gains L_i are achieved from the stability condition, based on a Lyapunov approach: Assigning a simple, quadratic Lyapunov function, the following *global* stability condition yields

²With a *common* output matrix $C_i \equiv C$ the weighted combination of the i th submodel with all and the direct adjacent submodels, respectively, is superfluous for the reconstructed closed-loop dynamics, as $\sum_i h_i(\underline{z})C = C \sum_i h_i(\underline{z}) = C \cdot 1 = C$ holds for all submodel. Therefore, the single summation \sum_i is sufficient in **Eq. 7** and **Eq. 9**.

$$\begin{aligned} V &:= \underline{x}^T P \underline{x} > 0 \\ \Leftrightarrow \dot{V} &= \dot{\underline{x}}^T P \underline{x} + \underline{x}^T P \dot{\underline{x}} < 0 \quad \text{with} \quad P^T = P > 0 \end{aligned} \quad (10)$$

for the *common* and *global*, respectively, symmetric, positive definite matrix $P > 0$ and the system-states $(\underline{x}, \dot{\underline{x}}) \equiv (\underline{x}, \dot{\underline{x}})$ or error-states $(\underline{e}, \dot{\underline{e}}) \equiv (\underline{e}, \dot{\underline{e}})$. That is, if the common, symmetric and positive definite matrix P exists, which holds the stability condition for the common and global, respectively, quadratic Lyapunov function V **Eq. 10**, the system is globally asymptotically stable and the state- and error-feedback-gain K_j and L_i can be derived from P as shown in the following two **Subsection 2.2.1** and **Subsection 2.2.2**, further information given e.g., in (Lendek et al., 2010), (Wang et al., 1996) and (Tanaka and Sugeno, 1992).

Hereinafter, the Lyapunov approach **Eq. 10** is denoted as the *global* Lyapunov approach.

2.2.1 Global Controller

LMI derivation

With **Eq. 6** in **Eq. 10** ($\underline{x} \equiv \underline{x}$) it follows (for the *individual* input-matrices B_j)¹:

$$\dot{V} = \underline{x}^T \left(\sum_i^{N_r} \sum_j^{N_r} h_i h_j (A_i^T P - K_j^T B_i^T P + P A_i - P B_i K_j) \right) \underline{x} < 0. \quad (11)$$

With the pre- and postmultiplication $P^{-1} \cdot \square$ and $\square \cdot P^{-1}$, the thereby necessary substitution $X := P^{-1}$ ($\stackrel{\text{Eq. 10}}{=} X$) and the introduction of the slag parameter $M_j = K_j X$ (to avoid the bilinear term $K_j X$ within the resulting inequality) the following inequality for the Lyapunov function dynamics is derived from **Eq. 11**:

$$\dot{V} = \underline{x}^T \left(\sum_i^{N_r} \sum_j^{N_r} h_i h_j (X A_i^T - M_j^T B_i^T + A_i X - B_i M_j) \right) \underline{x} < 0. \quad (12)$$

The inequality **Eq. 12** is solvable with a LMI-solver, if a *discrete* number of LMIs is derived from **Eq. 12**. Therefore, the convex properties of the membership functions $h_i h_j$ are exploited to derive the following LMI set with the intended discrete number of LMIs:

$$X A_i^T - M_j^T B_i^T + A_i X - B_i M_j < 0. \quad (13)$$

If the LMI set **Eq. 13** is solvable, i.e., a positive definite matrix $X (= P^{-1}$ with $P > 0$, see **Eq. 10**) exists, the Lyapunov approach **Eq. 10** (and its derivative **Eq. 12**) is fulfilled and the closed-loop system's stability is guaranteed.

Controller design procedure

Once a *common* matrix P and the slag parameter M_j are found with the LMI solver, the state-feedback gains K_j are defined by $K_j = M_j P$. That is, for *individual* input matrices B_i the state-feedback gain K_j of the j th submodel and subcontroller, respectively, is designed in a way, that the subcontroller holds the LMI **Eq. 10** and LMI set **Eq. 13**, the latter combining¹ the

closed-loop dynamic of the j th submodel with the dynamic of *all* other or the *direct adjacent* submodels.

Number of LMIs

For the discrete number of combined¹ LMIs in the LMI set **Eq. 13** it holds: $n_{global,all}^{LMI,Cntrl} = j_{max} i_{max} = Nr^2$ (with $i_{max} = j_{max} = Nr$). This number can be reduced to $n_{global,adj.}^{LMI,Cntrl} < n_{global,all}^{LMI,Cntrl}$, if just submodels, that are *direct adjacent* to the j th submodel (see **Section 2.1.1**), are blended. Additionally, the stability condition **Eq. 10** comprises a single LMI. That is, for the controller design with *individual* input matrices B_i the total LMI set consists of $n_{global,all/adj.}^{LMI,Cntrl} + 1$ LMIs. For the global Lyapunov approach with the global and *single* P -matrix, respectively this total LMI set is solved with a *single* execution of the LMI solver, i.e., the $n_{global,all/adj.}^{LMI,Cntrl} + 1$ LMIs are solved simultaneously.

2.2.2 Global Observer

LMI derivation

With **Eq. 9** in **Eq. 13** ($\underline{e} \equiv \underline{\mathcal{X}}$) it follows (for the *common* output-matrix $C_i \equiv C$)²:

$$\dot{V} = \underline{e}^T \left(\sum_i^{Nr} h_i (A_i^T P - C^T L_i^T P + P A_i - P L_i C) \right) \underline{e} < 0. \quad (14)$$

With the introduction of the slag parameter $N_i = P L_i$ (to avoid the bilinear term $P L_i$ within the resulting inequality) the following inequality for the error dynamics is derived from **Eq. 14**:

$$\dot{V} = \underline{e}^T \left(\sum_i^{Nr} h_i (A_i^T P - C^T N_i^T + P A_i - N_i C) \right) \underline{e} < 0. \quad (15)$$

Eq. 15 holds (as explained for **Eq. 12**), if the LMI set

$$A_i^T P - C^T N_i^T + P A_i - N_i C < 0 \quad (16)$$

is satisfied, i.e., a positive definite matrix $P > 0$ (see **Eq. 10**) exists, so that the Lyapunov approach **Eq. 10** (and its derivative **Eq. 15**) is fulfilled and the error system's stability is guaranteed.

Observer design procedure

Once a *common* matrix P and the slag parameter N_i are found with the LMI solver, the error-feedback gains L_i are defined by $L_i = P^{-1} N_i$. That is, for a *common* output matrix $C_i \equiv C$ the error-feedback gain L_i of the i th submodel and subobserver, respectively is designed in a way, that the subobserver holds the LMI **Eq. 10** and the LMI set **Eq. 16**, the latter just comprising the error dynamics of *each single* submodel (A_i, B_i) in an individual and single LMI, respectively (and not combining the dynamics of *all* or *direct adjacent* submodels).

Number of LMIs

For the discrete number of individual LMIs in the LMI set **Eq. 16** it holds: $n_{global}^{LMI,Obs} = i_{max} = Nr$. As **Eq. 16** defines a single LMI for each submodel and subobserver, respectively, there is no need to consider *direct adjacent* submodels in the observer design

with a *common* output matrix C . Additionally, the stability condition **Eq. 10** comprises a single LMI. That is, for the observer design with a *common* output matrix C the total LMI set consists of $n_{global}^{LMI,Obs} + 1$ LMIs. For the global Lyapunov approach with the global and *single* P -matrix, respectively this total LMI set is solved with a *single* execution of the LMI solver. i.e., the $n_{global}^{LMI,Obs} + 1$ LMIs are solved simultaneously.

2.3 Local Lyapunov Approach for Takagi-Sugeno Controller and Observer Design

For the proposed *local* Lyapunov approach the *local* stability condition

$$V_i := \underline{\mathcal{X}}^T P_i \underline{\mathcal{X}} > 0 \\ \Leftrightarrow \dot{V}_i = \underline{\dot{\mathcal{X}}}^T P_i \underline{\mathcal{X}} + \underline{\mathcal{X}}^T P_i \underline{\dot{\mathcal{X}}} < 0 \quad \text{with} \quad P_i^T = P_i > 0$$

and $i \in [1, Nr]$ (17)

holds for the *individual* and *local*, respectively, symmetric, positive definite matrix $P_i > 0$ and the system-states $(\underline{\mathcal{X}}, \underline{\dot{\mathcal{X}}}) \equiv (\underline{x}, \underline{\dot{x}})$ or error-states $(\underline{\mathcal{X}}, \underline{\dot{\mathcal{X}}}) \equiv (\underline{e}, \underline{\dot{e}})$. Compared to the *common* and *global* Lyapunov-approach **Eq. 10**, a number of Nr *individual* Lyapunov functions V_i (with $i \in [1, Nr]$) are used instead of *one* common Lyapunov function V (in the global Lyapunov approach **Eq. 10**). That is, for *each* submodel (A_i, B_i) an individual Lyapunov function V_i is defined.

Accordingly, the LMI derived for the closed-loop dynamics **Eq. 13** and error dynamics **Eq. 16** is simplified, as the convex blending becomes obsolete, if the Lyapunov stability condition is defined for *each* submodel individually (compare **Eq. 18** with **Eq. 12** and **Eq. 20** with **Eq. 15**).

2.3.1 Local Controller LMI

For the local Lyapunov approach the inequality derived for the *closed-loop* dynamics (with $(\underline{\mathcal{X}}, \underline{\dot{\mathcal{X}}}) \equiv (\underline{x}, \underline{\dot{x}})$) results in

$$\dot{V} = \underline{x}^T (X_i A_i^T - M_i^T B_i^T + A_i X_i - B_i M_i) \underline{x} < 0 \\ \text{with} \quad X_i := P_i^{-1} \quad \text{and} \quad M_i = K_i X_i. \quad (18)$$

The inequality (**Eq. 18**) holds, if the LMI

$$X_i A_i^T - M_i^T B_i^T + A_i X_i - B_i M_i < 0 \quad (19)$$

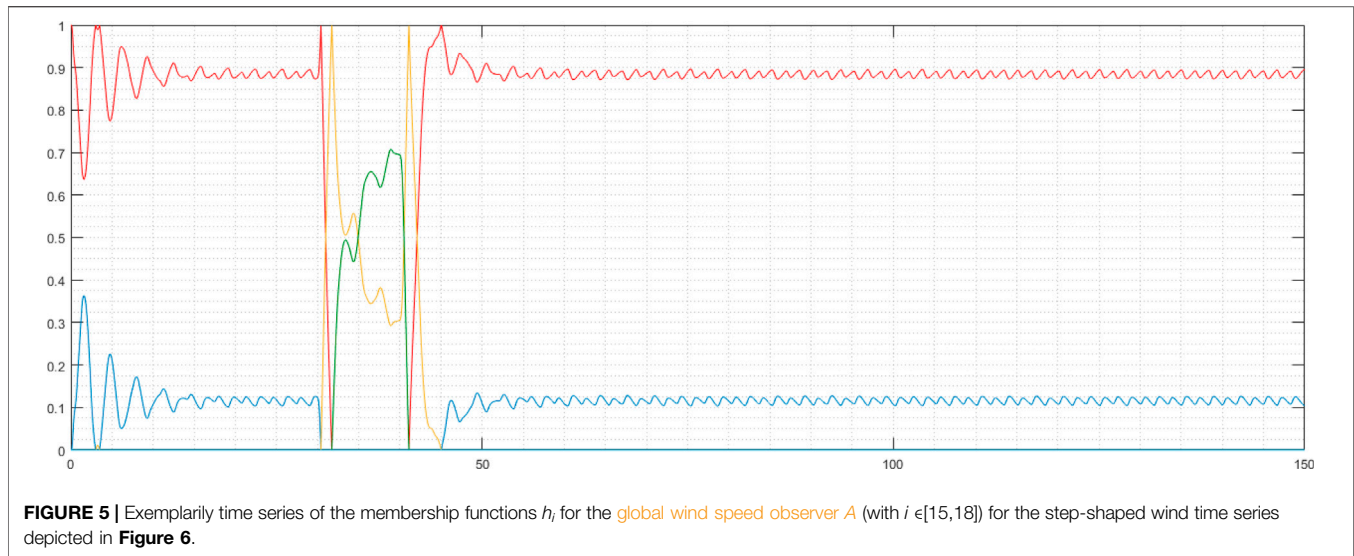
is satisfied, i.e., *individual* and *local*, respectively positive definite matrices X_i exist, which fulfill **Eq. 19**.

2.3.2 Local Observer LMI

The same simplification holds for the *error dynamics*, i.e. for the local Lyapunov approach the inequality derived for the error dynamics (with $(\underline{\mathcal{X}}, \underline{\dot{\mathcal{X}}}) \equiv (\underline{e}, \underline{\dot{e}})$) results in:

$$\dot{V} = \underline{e}^T (A_i^T P_i - C^T N_i^T + P_i A_i - N_i C) \underline{e} < 0 \\ \text{with} \quad N_i = P_i L_i. \quad (20)$$

The inequality (**Eq. 20**) holds, if the LMI



$$A_i^T P_i - C^T N_i^T + P_i A_i - N_i C < 0 \quad (21)$$

is satisfied, i.e. *individual* and local, respectively positive definite matrices P_i exist, which fulfill Eq. 21.

2.3.3 Local Controller and Observer Design Procedure

Once the *individual* matrix P_i and the slag parameter M_i (for local controller design) or N_i (for local observer design) is found with the LMI solver, the state-feedback gain K_i or error-feedback gain L_i is defined by $K_i = M_i P_i$ or $L_i = P_i^{-1} N_i$. That is, the state-feedback gain K_i or error-feedback gain L_i of the i th submodel and subcontroller or subobserver, respectively is designed in a way, that the subcontroller or subobserver holds the LMI Eq. 17 and the LMI set Eq. 19 or Eq. 21, the latter just comprising the state or error dynamics of *each single* submodel (A_i, B_i) in an individual and single LMI, respectively (and not combining the dynamics of *all* or *direct adjacent* submodels).

2.3.4 Number of LMIs

As the local Lyapunov approach simplifies and reduces, respectively the set of combined LMIs (for the local Lyapunov approach the LMI set comprises just *two* LMIs: LMI Eq. 19 or Eq. 21 and the LMI of the positive definite matrix $P_i > 0$, see Eq. 17), the parallel and simultaneously, respectively to be solved LMIs are reduced (from $n_{global,all}^{LMI,Cntrl} + 1 > n_{global}^{LMI,Obs} (= Nr + 1)$ to $n_{Local}^{LMI,Cntrl/Obs} = 2$ LMIs), too. Thereby, the LMI solver is executed Nr times consecutively for the local Lyapunov approach (due to the individual P_i -matrix in Eq. 17), while the LMI solver for the global Lyapunov approach is executed just once, because of the *common* P matrix in Eq. 10. Therefore, the flexibility of the LMI solver in assigning desired pole locations of the closed-loop system is increased for the local Lyapunov approach. The less conservative task for the LMI solver enables the specification of tighter performance constraints for the local Lyapunov approach, e.g., smaller pole regions (by implementing pole region constraints in form of additional LMIs, see Section 2.5), and results in

more flexible and consistent specifications of the desired closed-loop dynamics, denoted as the *general objective* of the local Lyapunov approach.

2.3.5 Subsequent, Global Lyapunov Stability Analysis

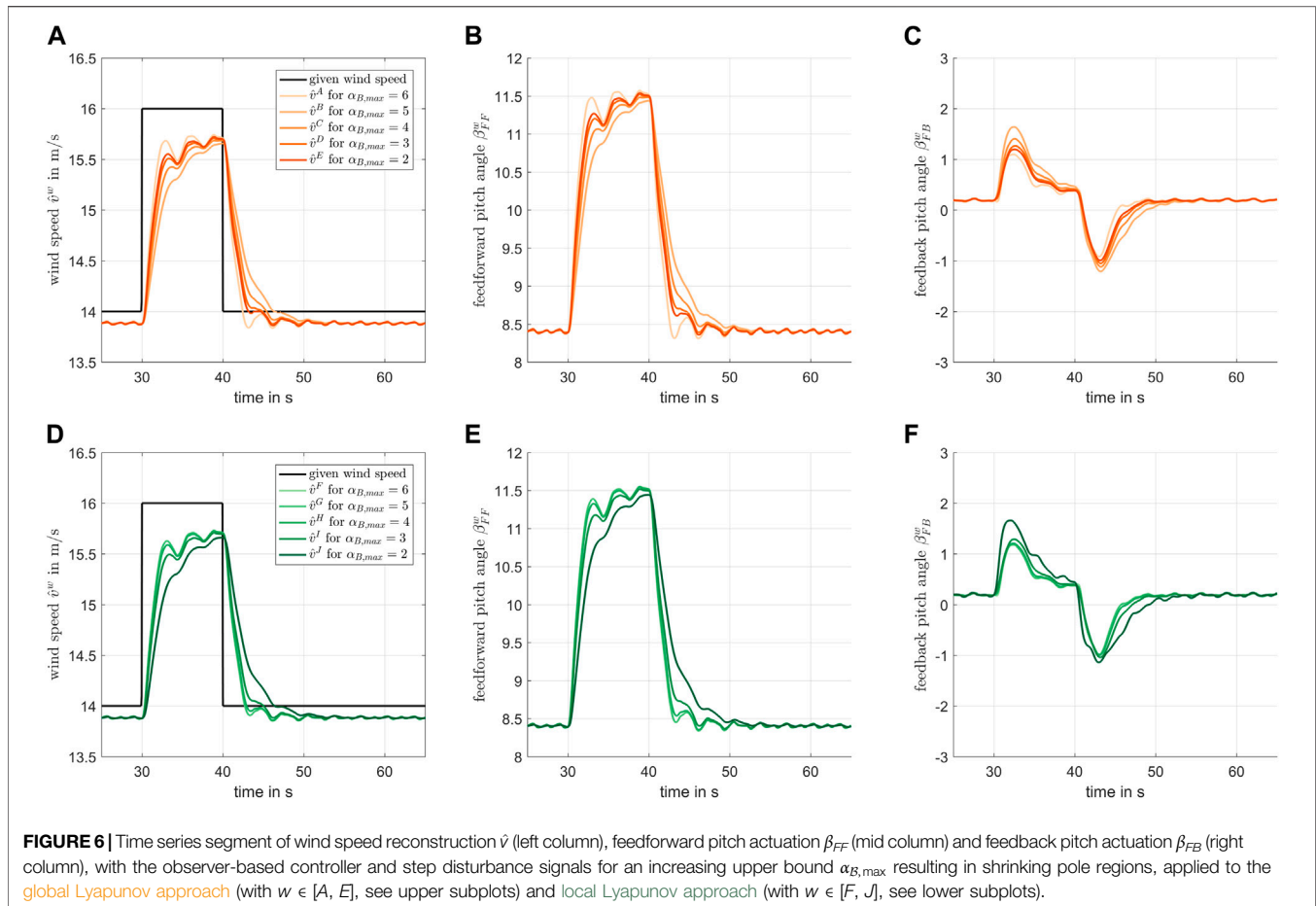
Note: With the local Lyapunov approach and the resulting LMIs Eq. 19 and Eq. 21, just the small signal stability is evaluated, i.e., considering the operating point depicted in the linear submodel, if a positive definite matrix P_i exists. To ensure the global stability of the convexly blended submodels, a final stability analysis has to be performed, enveloping all combined closed-loop submodels. This final stability analysis is executed in the subsequent, global Lyapunov stability analysis described in Section 2.4.4.

2.4 Local Lyapunov Approach in Wind Turbine Control Application

2.4.1 Structural Dynamical Design Model

For the wind turbine (WT) controller design, the authors use simplified structural dynamical models, based on lumped-masses and joined with discrete spring- and damper-elements (e.g., (Bianchi et al., 2007), (Georg, 2015) or (Pöschke et al., 2020)). These models comprise just the essential and costly wind turbine components rotor, drive train and tower. Within this contribution, the rotor blade dynamic is represented, while the tower dynamic is neglected (see Figure 2). The rotor model is composed of the rotor \mathcal{B} blades, rotating as one rigid body in the rotor plane (denoted with *rotor rotation (speed)* $\omega_{\mathcal{R}}$ in Figure 2) and *translating* with $x_{\mathcal{B}}$ in and against the wind direction (denoted by the subscripted index \mathcal{B}). The drive train is represented by its rigid body Rotation (denoted by the subscripted index \mathcal{R}).

After linearising an elaborated WT simulation model for i stationary operating points OP_i (with $i \in [1, Nr]$), the controller design submodels ($A_{i \mathcal{B}/\mathcal{R}}, B_{i \mathcal{B}/\mathcal{R}}$) are composed with the



linearisation coefficients. Thereby, mechanical couplings between those components are specified or neglected, as explained in **Section 2.4.4**. The specification of the design submodels ($A_{iB/R}, B_{iB/R}$) are given in Tables A2, A3 in the appendix.

2.4.2 Wind Turbine Control Objectives, Loading and Operation Concept

In wind turbine (WT) application the controller intends—besides energy yield optimisation—for mechanical load mitigation, due to vast environmental loads acting on the complete WT structure. Besides the ultimate, mechanical loads (that occur for single moments and the corresponding ultimate stress must not exceed the material strength), fatigue, mechanical loads have to be examined. Those fatigue loads, also denoted as *Damage Equivalent Loads (DEL)*, i.e., a mean amplitude with the equivalent damaging effect like constantly or stochastically changing amplitudes, e.g., resulting from turbulent wind time series), characterise the damaging effect resulting from load cycles, occurring all over the components' lifetime (Clemens et al., 2020).

Ascribed to the generator characteristics, two operating modes have to be distinguished: In the *partial load range*, the generator is operated below rated power and the energy yield is optimised by generator torque T_G control, i.e., the

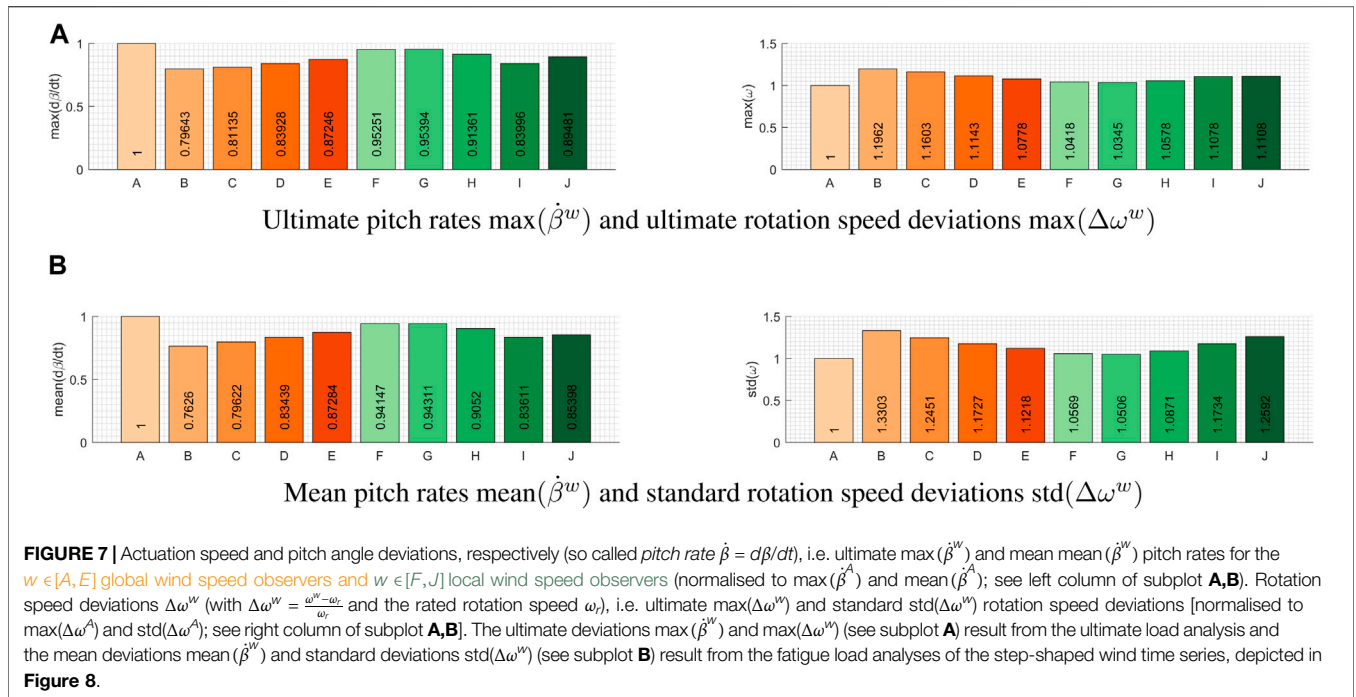
generator torque T_G is one of the two actuating signals. In *full load range*, the generator is operated at rated power. Therefore, the energy extraction from the inflow with the WT rotor is restricted to rated generator power by the pitch angle β control, influencing the blade aerodynamics and mechanical torque T_R generation of the rotor by rotating the complete rotor blade along its longitudinal axis (see **Figure 2**). Hence, the pitch angle β is the second actuating signal of a wind turbine controller, i.e., $\underline{u} = [\beta, T_G]^T$ holds. In this contribution, a collective pitch control algorithm is utilised, i.e., all blades are actuated with the same pitch angle. In Appendix Table A1, the control signals $\underline{u}_{c,i} = [\beta_{c,i}, T_{G,c,i}]^T$ for all 27 stationary operating points in partial and full load range are listed.

2.4.3 Implemented Wind Turbine Controller Structure

Within this contribution an observer-based feedforward-feedback controller in TS framework is utilised and applied to wind turbine control.

For the feedforward-feedback controller the extended *parallel distributed compensation (PDC)* controller **Eq. 5** is used.

To attenuate the effect of the disturbing wind speed, a disturbance-observer is used (see **Section 2.1.3** and **Figure 3**)



to determine the rotor effective wind speed v_{eff} . For the disturbance observer, the WT main component model (see **Section 2.4.1** and **Section 2.4.4**) with the highest disturbance sensitivity is used, i.e., for wind turbine systems the downwind and upwind blade deflection x_B^3 is most sensitive to disturbances caused by wind speed fluctuations, due to the very low bending stiffness of the blades (in normal direction to its profile chord). Therefore, the disturbance observer reconstructs the disturbing, effective wind speed $\hat{v} \equiv v_{eff}$ and blade translation speed \hat{x}_B from the blade translation error e_{x_B} determined from the measured and reconstructed blade translation $e_{x_B} = x_B - \hat{x}_B$. For this purpose, the disturbance observer design model is augmented by a simple wind model (Ekelund, 1994), expanding the state vector \hat{x}_B with the reconstructed wind speed \hat{v} (and disturbance signal $\underline{d} \equiv \hat{v}$, respectively) $\hat{\underline{x}}_B = [\hat{x}_B, \hat{v}]^T$ and augmenting the state space model **Eq. 7** and **Eq. 9** accordingly (with \tilde{A}_i, \tilde{B}_i and \tilde{C} ; see also (Pöschke et al., 2020)). Therefore, the used TS disturbance observer, resulting from the integration of a disturbing model in the plant model and augmenting the state vector \hat{x}_B with the disturbing wind state \hat{v} , is classified as an *Extended State Observer* (see (Li et al., 2014)), that is represented in the Takagi-Sugeno framework.

The reconstructed wind speed \hat{v} is defined as premise variable \underline{z} , scheduling the nonlinearity of the system and utilised for the membership function $h_i(\underline{z})$ of the feedforward signal \underline{u}_{FF} and feedback signal \underline{u}_{FB} in **Eq. 5**. Therefore, the disturbance observer

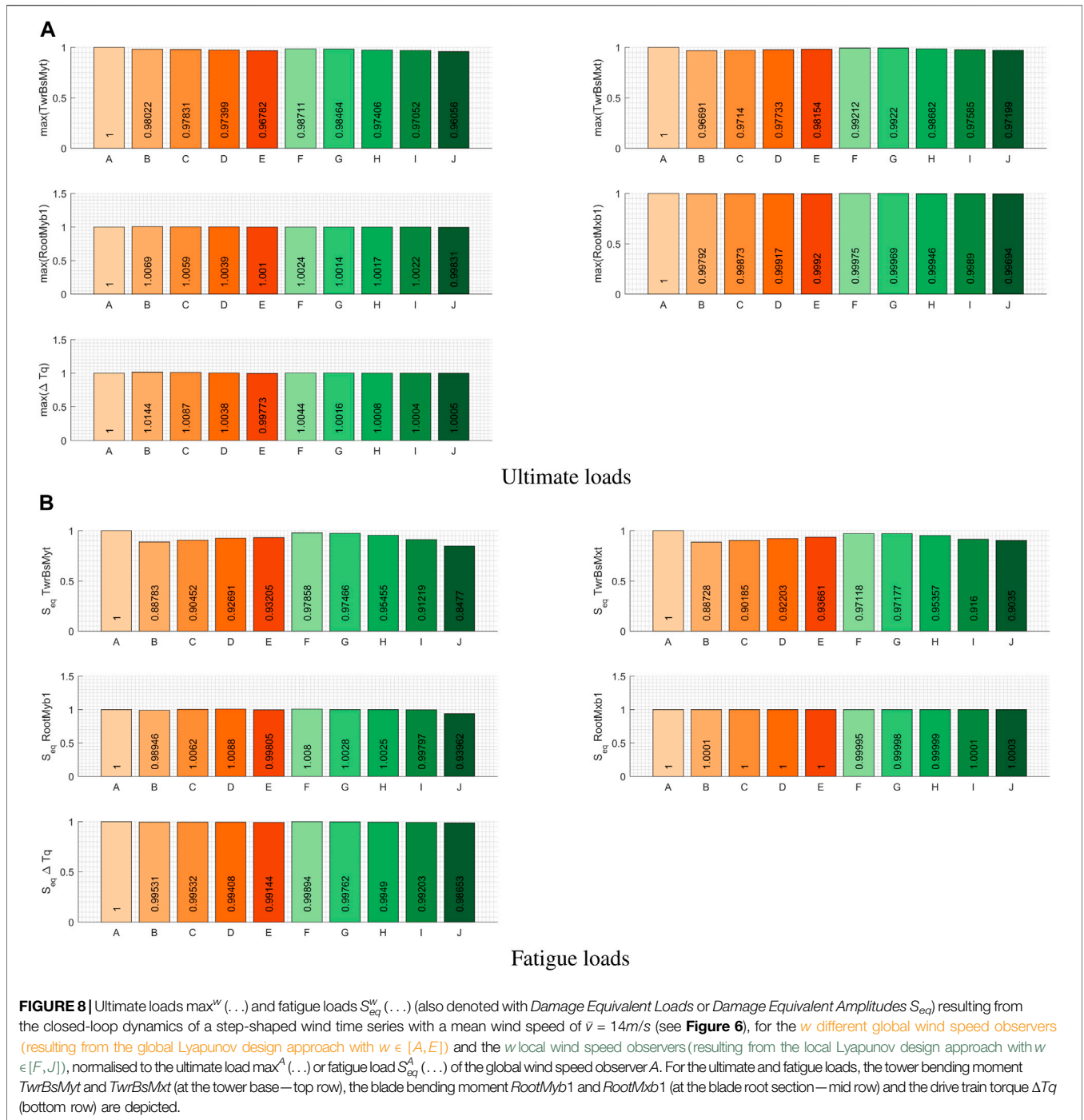
does not intend for a disturbance *rejection*, rather than for an observer-based controller scheme with disturbance *reconstruction*, influencing the controllable system inputs (see also **Section 2.1.3**).

Within this contribution, triangular-shaped⁴ membership functions $h_i(z)$ are used, just blending directly adjacent submodels (A_i, B_i) (see **Figure 1** and the explanations in **Section 2.1.1**). Because of the reduced number of (triangular) membership functions (i.e., just the direct adjacent membership functions), also the number of submodels included in the feedback controller and observer design—based on the *global* Lyapunov approach—is reduced (see **Section 2.3**) (Note: For the *local* Lyapunov approach an individual Lyapunov function V_i is defined for *each* submodel (A_i, B_i), because of the local stability condition **Eq. 17**. Therefore, the convex blending with the membership functions $h_i(\underline{z})$ is obsolete and the form of the membership functions has no influence on the design, as just the i th submodel is included in the observer design—see **Section 2.3**.)

The augmented state, input and output matrices \tilde{A}_i, \tilde{B}_i and \tilde{C} , as well as the augmented steady system states $\tilde{\underline{x}}_{c,i}$ and steady input states $\underline{u}_{c,i}$, used within this contribution, are given in Table A2 to Table A6. Additionally, the state feedback gain matrices K_i and error-feedback gain matrices L_{iB}^w are listed in Table A7 and Table A8.

³Note: Within this contribution just the blade tip translations $x_{B,1/2/3}$ are assumed to be measurable, whereat the mean value $x_B = (1/3)(x_{B,1} + x_{B,2} + x_{B,3})$ is utilised for calculating the error $e_B = x_B - \hat{x}_B$, fed back and amplified with the error gains L_i to reconstruct the blade tip speed \dot{x}_B , blade tip acceleration \ddot{x}_B and temporal wind speed variations \dot{v}_B .

⁴The *triangular* shape of the membership functions holds, if just *one* premise variable $z_k = z$ (with $k = k_{\max} = 1$) is defined. Because in this case, the membership function $h_i(z)$ (**Eq. 1**) is identical to the *triangular*-shaped weighting function $w_{k,i}(z)$, see **Eq. 2**.



2.4.4 Decomposed Wind Turbine System and Subsequent Lyapunov Stability Analysis

Mechanical and mechatronic systems, like wind turbines (WT), are characterised by the pole locations of the coupled mechanical components. Due to the widely spaced, open-loop pole locations of the WT system (see **Figure 4** and **Section 4**), it is advantageous for the controller and observer design models to decompose the

coupled, nonlinear system submodel (A_i, B_i) into decoupled reduced submodels—denoted with *component models*—just comprising the dynamics of the individual, mechanical main components like the rotor blade submodels (A_{iR}, B_{iR}) and drive train submodels (A_{iD}, B_{iD}) (also described in (Pöschke et al., 2020), but now upgraded with the rotor blade submodels (A_{iB}, B_{iB})).

Thereby, couplings between the components' degree of freedom are neglected within the component controller and observer design models, like the missing axial coupling between drive train and main frame/tower depicted in **Figure 2** (just plain ball bearings are defined). Despite this substantial assumption, the component design models achieve satisfying controller performance, while increasing the controller design flexibility significantly.

The corresponding design models are derived from the elaborated NREL FAST 5 MW reference wind turbine (Jonkman, 2016; Jonkman and Buhl (2005) and Jonkman et al. (2009)) simulation model, based on a convex sum of 27 linearised submodels (defined for piecewise equidistant, steady and effective wind speeds from $v \in [3 \text{ m/s}, 25 \text{ m/s}]$; see Table A1), received from the NREL FAST WT linearisation (Jonkman and Jonkman, 2016). Within this linearisation, all aerodynamic and structural dynamic characteristics defined in the elaborated NREL FAST 5 MW reference wind turbine are accounted (see also explanations given in (Pöschke et al., 2020)).

Once the component controllers and observers are designed, the corresponding state-feedback gains (K_{iR} , K_{iB}) and observer gains (L_{iR} , L_{iB}) are superposed for the overall controller gain K_i and observer gain L_i . The stability of the convexly blended i submodels is analysed with the superposed gains K_i and L_i in a subsequent, global Lyapunov stability analysis (see **Section 2.3.5**), separately for the controller and error dynamics (Yoneyama et al. (1998), Ma et al. (1998)). Although, the separation principle and overall stability proof, respectively of separately designed observer and controller is just valid for *measurable* premise variables \underline{z} (Yoneyama et al. (1998), Ma et al. (1998)), it is shown in (Pöschke et al., 2022) for a separated observer and controller design, that the overall stability is also ensured for *reconstructed* premise variables $\hat{\underline{z}}$, under the assumption of a maximum estimation error in the premise variable with $\hat{\underline{z}} \rightarrow \underline{z}$. Therefore, the stability proof and separation principle, respectively holds for $\hat{\underline{z}} \rightarrow \underline{z}$. Even though the maximum error estimation is missing in this contribution, it is assumed, that $\hat{\underline{z}} \rightarrow \underline{z}$ and the separation principle holds. The missing error estimation and overall stability proof corresponding to (Pöschke et al., 2022) has to be given in future work. That is, within this contribution just separated controller and observer syntheses are executed in regard to a pragmatic design process that intends for less conservative state and error state feedback gains K_i and L_i , like explained in **Section 1**. Thus, the design process is focused on an automated controller and observer design, i.e., the Lyapunov approach was selected to achieve the controller and observer gains in a systematic and traceable procedure, rather than to ensure the overall stability of the observer-based system dynamics. Though, the stability condition of the Lyapunov approach is exploited (separately in the controller and observer design) to abbreviate the iterative design process, i. e., all controller and observer design parameter specifications, resulting in unstable dynamics, are eliminated within the controller and observer synthesis to expedite the design process.

2.5 LMI Constraints of the Pole Regions

With the linear matrix inequalities **Eqs 13, 16, 19** and **Eq 21** the pole locations are just restricted to the left half of the complex pole map. Additional constraints need to be defined to tighten the pole location on smaller pole regions. As described in (Chilali and Pascal, 1996), additional bounds specified in the complex pole map can be transformed into LMI, which are applied to wind turbine control e.g., in (Pöschke et al., 2019). Within this contribution, just *vertical upper and lower bounds* (representing the maximum and minimum decay rate $\alpha_{B,\max}$ and $\alpha_{B,\min}$) and a *cone angle* θ (representing the Damping ratio D with $D = \cos(\theta)$) are used and described in the Appendix Section A1.3. With these constraints, pole regions with symmetrical trapezium shape are defined (see **Figure 4**).

3 SIMULATIONS AND RESULTS

With the local Lyapunov approach, the *general objective* of a more flexible specification of the pole locations and increased consistency in the desired and achieved closed-loop system dynamics is intended (see **Section 2.3.4**). This general objective is analysed within this contribution for a wind turbine specific, *particular objective*—the *decreased* observer performance and feedforward actuation—described in **Section 3.1**. The achieved results from wind turbine simulations are presented in **Section 3.2**.

3.1 Simulation Design: Particular Objective of the Local Lyapunov Approach Within this Contribution

Within this contribution, the *general objective* of the local Lyapunov approach (see **Section 2.3.4**) is assessed for a particular wind speed observer performance, hereinafter denoted with the *particular objective*, supposed to be beneficial regarding load mitigation: For turbulent wind excitation, high mechanical loads (especially fatigue loads) often result from brisk *feedforward* actuation⁵. Therefore, the proposed local Lyapunov approach shall yield for a *decreased* disturbance reconstruction \hat{v} performance—affecting the premise variable \hat{v} —to attenuate the premise variable \hat{v} driven feedforward actuation $\underline{u}_{FF}(h_i(\hat{v}))$ (see **Eq. 5**) with $\underline{z} \equiv \hat{v}$) and to increase the feedback compensation $\underline{u}_{FB}(h_i(\hat{v}))$ (see **Section 2.1.2**). That is for this particular objective, the closed-loop dynamics is defined in such a way, that the performance of the wind speed \hat{v} reconstruction and the precision of the feedforward actuation $\underline{u}_{FF}(h_i(\hat{v}))$ is intentionally lowered to achieve rising deviations, which are compensated by increased feedback controller actuation $\underline{u}_{FB}(h_i(\hat{v}))$. As the error-feedback gains L_i govern the observer performance and these gains are achieved from the observer design according to the specified error-feedback pole regions, the pole region specification is varied by shrinking its size. That is, a significant modification of the pole regions is conducted with the

⁵These *brisk* feedforward-actuation result from feedforward specifications without dynamics (i.e., without any *damping* influence, e.g., from the K_i gains), but simple convex blending of the steady states $\underline{u}_{c,j}$ (see \underline{u}_{FF} in **Eq. 5**).

local Lyapunov approach to achieve the *particular objective* of a decreased observer performance, resulting in mitigated mechanical loads, especially mitigated fatigue loads. To lower the wind speed observer performance and error dynamics (i.e., the wind speed \hat{v} depending error-feedback gains L_i in $\sum_i^{Nr} h_i(\hat{v})L_i C(\underline{x} - \hat{\underline{x}}) = L(\hat{v})C\underline{e}$ ⁶), the upper bound $\alpha_{B, \max}$ of the error dynamics' pole region is increased, i.e., shifted towards the imaginary axis in the left half of the complex pole map, while the lower bound $\alpha_{B, \min}$ (located close to the open-loop poles) is kept constant (see **Figure 4**). It is expected, that the error-feedback gains L_i decrease with increasing upper bounds $\alpha_{B, \max}$ ($\alpha_{B, \max} \nearrow \Rightarrow L_i \searrow$), as the distances between open-loop poles and closed-loop poles of the error dynamics $\Delta s_{p,i}^{w,p}(\alpha_{B, \max})$ decrease with increasing upper bounds $\alpha_{B, \max}$ and the quantity of the error-feedback gain depends on this distance $|L_i| = f(\Delta s_{p,i}^{w,p})$ ⁷.

To distinguish the pole locations s_p^w (see **Figure 4**), error-feedback gains L_i^w (see Table A8) or mechanical loads like S_{eq}^w (see **Figure 8**), resulting from a number of w different error-feedback pole region specifications, the superscript index w (denoting the w different error-feedback pole regions and corresponding wind speed observers) is introduced in the following. To analyse the error-feedback gains L_{iB}^w of the Blade model-based wind speed observer, the following state space equation is relevant (see **Eq. 7**), describing the reconstructed blade component dynamics:

$$\underbrace{\begin{bmatrix} \dot{\hat{\underline{x}}}_B \\ \dot{\hat{\underline{x}}}_B \\ \dot{\hat{v}} \end{bmatrix}}_{\hat{\underline{x}}} = \sum_i^{Nr} h_i(\underline{z}) \left(A_i(\hat{\underline{x}} - \hat{\underline{x}}_{c,i}) + B_i(\underline{u} - \underline{u}_{c,i}) \right) + \sum_i^{Nr} h_i(\hat{v}) \underbrace{\begin{bmatrix} L_{iB}^{w,1} \\ L_{iB}^{w,2} \\ L_{iB}^{w,3} \end{bmatrix}}_{L_{iB}^w} \underbrace{\begin{bmatrix} 1 & 0 & 0 \\ \dot{x}_B - \hat{x}_B \\ v - \hat{v} \end{bmatrix}}_{\underline{\epsilon}} \quad (22)$$

$$= \sum_i^{Nr} h_i(\underline{z}) \left(A_i(\hat{\underline{x}} - \hat{\underline{x}}_{c,i}) + B_i(\underline{u} - \underline{u}_{c,i}) \right) + \sum_i^{Nr} h_i(\hat{v}) \begin{bmatrix} L_{iB}^{w,1} & 0 & 0 \\ L_{iB}^{w,2} & 0 & 0 \\ L_{iB}^{w,3} & 0 & 0 \end{bmatrix} \begin{bmatrix} x_B - \hat{x}_B \\ \dot{x}_B - \hat{x}_B \\ v - \hat{v} \end{bmatrix}$$

with the j components \hat{x}^j of the reconstructed and augmented state vector $\hat{\underline{x}}$ and with j matrix elements $L_{iB}^{w,j}$ of the error-feedback gain L_{iB}^w .

Within this contribution, the *general objective* of an increased *flexibility* in the observer-based controller

design and increased *consistency* of the desired and achieved system dynamics (see **Section 2.3.4**) is analysed with the *particular objective* of an decreased observer performance and feedforward actuation, resulting from shrinking pole regions. That is, for both Lyapunov approaches identical upper bound variation and shrinking pole regions are defined and the *flexibility* and *consistency* of both approaches are compared with the help of several metrics (like pole locations, error-feedback gains, reconstructed wind speeds, pitch angles, pitch rate and rotation speed deviations): Regarding the *flexibility*, the *minimum* pole region dimension is determined. That is, for all poles it is examined, if the poles are located *inside* the imposed pole region after executing the error-feedback gain L_i design. The intended, *particular objective* of a decreased observer performance of the *local* Lyapunov approach is fulfilled, if this approach leads to a smaller, minimum pole region than the global Lyapunov approach, enabling the specification of tighter performance constraints for the system dynamics. Regarding the *consistency* of desired and achieved system dynamics, it is analysed, if the shrinking poles regions result in decreased feedforward pitch angles β_{FF} and increased feedback pitch angles β_{FB} with decreased pitch rates $\dot{\beta}$ and increased, resulting rotation speed deviations $\Delta\omega$.

In addition, the control objectives of mitigated ultimate and fatigue loads are analysed for identical WT controller pole region specifications, but related to both disturbance observer design approaches.

3.2 Simulation Results: Disturbance Observer Variation, Resulting System Dynamics and Mechanical Loads

For each Lyapunov approach, five different disturbance observers for the **global Lyapunov approach** (denoted in the following with the superscripted index w in $[A, E]$) and for the **local Lyapunov approach** (denoted with $w \in [F, J]$) are designed—based on five different pole regions with shrinking size—and the resulting closed-loop dynamics as well as the resulting mechanical loads on the WT components are analysed.

Pole regions and submodels

In **Table 1**, the pole region specifications for both design approaches are listed. For the analysed wind time series excitation with prescribed wind speeds $v(t)$ between 14 m/s and 16 m/s, just four of the $i \in [1,27]$, from linearisation achieved submodels (denoted with the subscripted index i) are convexly blended. Therefore, just these four submodels (A_i, B_i) with $i \in [15,18]$ are taken into account in the results and discussion.

Pole locations

Figure 5 shows the resulting p pole locations $s_{p,i}^{OL/w,p}$ of the corresponding open-loop dynamics (denoted with the superscripted index OL), as well as the closed-loop dynamics

⁶with $\sum_i^{Nr} h_i(\hat{v})L_i C(\underline{x} - \hat{\underline{x}}) = \sum_i^{Nr} h_i(\hat{v})L_i C\underline{e} = L(\hat{v})C\underline{e}$, see **Eq. 7** (with $\underline{z} \equiv \hat{v}$) and **Eq. 8**.

⁷As the pole location distance $\Delta s_{p,i}^{w,p}$ depends on the difference between $s_{p,i}^{OL,p}$ and $s_{p,i}^{w,p}$ (i.e., $\Delta s_{p,i}^{w,p} = f(s_{p,i}^{OL,p} - s_{p,i}^{w,p})$, see **Section 3.2** with $s_{p,i}^{OL,p} = eig(A_i)$ and $s_{p,i}^{w,p} = eig(A_i - L_i^w C)$, i.e., $\Delta s_{p,i}^{w,p} = f(eig(A_i) - eig(A_i - L_i^w C))$), this distance $\Delta s_{p,i}^{w,p}$ in-/decreases, if the error-feedback gain L_i^w in-/decreases—and vice versa, i.e., L_i^w in-/decreases for in-/decreasing $\Delta s_{p,i}^{w,p}$.

TABLE 1 | Description of the TS disturbance observers (i.e., wind speed observers) (for the wind speed observers w in A to E , based on the global Lyapunov – approach and for the wind observers w in F to J , based on the local Lyapunov – approach, with the cone angle, and the lower and the upper bounds of the five pole regions (see Figure 4), defined by the Blade error state damping D_B and the Blade decay rates $\alpha_{B,\min}$ and $\alpha_{B,\max}$).

	A	B	C	D	E	F	G	H	I	J
Lyapunov-approach	global	global	global	global	global	local	local	local	local	local
D_B	0.6	0.6	0.6	0.6	0.6	0.6	0.6	0.6	0.6	0.6
$\alpha_{B,\min}$	0.5	0.5	0.5	0.5	0.5	0.5	0.5	0.5	0.5	0.5
$\alpha_{B,\max}$	6.0	5.0	4.0	3.0	2.0	6.0	5.0	4.0	3.0	2.0

TABLE 2 | Distances $\Delta s_{P_j}^{w,p}$ between the open-loop pole location $s_{P_j}^{OL,p}$ and closed-loop pole locations $s_{P_j}^{w,p}$ of each of the $p \in [1,3]$ poles (with $p = 1$ for the real-valued pole of the wind model and $p = 2,3$ for the complex-valued poles of the blade model) of a single submodel i in the pole map for the w th wind speed observer; average distance $\Delta s_{P_j}^{w,p}$ of the single submodel i ; average distance $\Delta s_{P_j}^{w,p}$ of all four submodels.

	A	B	C	D	E	F	G	H	J	J
$\Delta s_{P_{15}}^{w,1}$	1.51	0.53	0.68	0.81	0.90	1.22	1.21	1.01	0.77	0.50
$\Delta s_{P_{15}}^{w,2}$	2.39	5.65	1.32	0.75	0.81	2.56	2.07	1.20	0.88	1.78
$\Delta s_{P_{15}}^{w,3}$	2.39	5.65	1.32	0.75	0.81	2.56	2.07	1.20	0.88	1.78
$\Delta s_{P_{16}}^{w,1}$	1.52	0.53	0.68	0.88	1.22	1.21	1.02	0.77	0.50	
$\Delta s_{P_{16}}^{w,2}$	2.41	5.68	1.34	0.76	0.81	2.59	2.12	1.26	0.91	1.79
$\Delta s_{P_{16}}^{w,3}$	2.41	5.68	1.34	0.76	0.81	2.59	2.12	1.26	0.91	1.79
$\Delta s_{P_{17}}^{w,1}$	1.52	0.53	0.68	0.81	0.88	1.22	1.21	1.02	0.77	0.50
$\Delta s_{P_{17}}^{w,2}$	2.42	5.68	1.35	0.77	0.81	2.60	2.13	1.28	0.92	1.80
$\Delta s_{P_{17}}^{w,3}$	2.42	5.68	1.35	0.77	0.81	2.60	2.13	1.28	0.92	1.80
$\Delta s_{P_{18}}^{w,1}$	1.52	0.53	0.67	0.80	0.87	1.21	1.22	1.02	0.78	0.50
$\Delta s_{P_{18}}^{w,2}$	2.44	5.70	1.37	0.78	0.80	2.62	2.18	1.34	0.95	1.81
$\Delta s_{P_{18}}^{w,3}$	2.44	5.70	1.37	0.78	0.80	2.62	2.18	1.34	0.95	1.81
$\Delta s_{P_{15}}^{w,p}$	2.10	3.94	1.11	0.77	0.84	2.11	1.78	1.14	0.85	1.36
$\Delta s_{P_{16}}^{w,p}$	2.11	3.96	1.12	0.78	0.83	2.13	1.82	1.18	0.86	1.36
$\Delta s_{P_{17}}^{w,p}$	2.12	3.97	1.12	0.78	0.83	2.14	1.83	1.19	0.87	1.37
$\Delta s_{P_{18}}^{w,p}$	2.14	3.98	1.14	0.78	0.82	2.15	1.86	1.24	0.89	1.37
$\Delta s_{P_j}^{w,p}$	2.12	3.96	1.12	0.78	0.83	2.13	1.82	1.19	0.87	1.36

(denoted with the superscripted index w for the w in $[A, E]$ global wind speed observers and $w \in [F, J]$ local wind observers) and error dynamics, respectively, achieved within the disturbance observer design, in the complex pole map. As the disturbance observer design model and wind speed observer design model, respectively, consists of the vibrating Blade model (with its two states \hat{x}_B, \hat{x}_B see Eq. 22) and the non-vibrating wind model (with the wind speed state \hat{v}), the design model posses three poles (i.e., $p \in [1,3]$): the conjugated-complex blade poles $s_{P_i}^{OL/w,2 \vee 3}$ (inside the pole map, for $p = 2 \vee p = 3$) and the real-valued wind model pole $s_{P_i}^{OL/w,1}$ (on the real axis of the pole map, for $p = 1$). In Table 2 the distances

$$\Delta s_{P_i}^{w,p} = \sqrt{(\text{Re}(s_{P_i}^{OL,p}) - \text{Re}(s_{P_i}^{w,p}))^2 + (\text{Im}(s_{P_i}^{OL,p}) - \text{Im}(s_{P_i}^{w,p}))^2} \quad (23)$$

TABLE 3 | Mean Euclidean norm $\|L_{iB}^{w,j}\|_2$ and average, mean Euclidean norm $\|L_{iB}^{w,j}\|_2$ of the error-feedback gain matrices $L_i(v_i)$ of the global wind speed observers A to E (i.e., $w \in [A, E]$) and local wind speed observers F to J (i.e., $w \in [F, J]$, see Table A8) for increasing upper bounds $\alpha_{B,\max}$ (with $\|L_{iB}^{w,3}\|_2$ for the average, mean Euclidean norm of the wind model error-feedback gains ($j = 3$) and $\|L_{iB}^{w,(1,2)}\|_2$ for the average, mean Euclidean norm of the Blade error-feedback gains ($j = 1,2$))

	A	B	C	D	E	F	G	H	I	J
$\ L_{15B}^{w,j}\ _2$	14.1	15.6	10.1	6.2	4.4	14.9	13.4	9.1	5.4	4.7
$\ L_{16B}^{w,j}\ _2$	14.2	15.7	10.2	6.3	4.5	15.1	13.6	9.5	5.7	5.0
$\ L_{17B}^{w,j}\ _2$	14.3	15.8	10.3	6.4	4.5	15.1	13.7	9.6	5.8	5.2
$\ L_{18B}^{w,j}\ _2$	14.4	15.9	10.4	6.5	4.5	15.2	13.9	10.0	6.1	5.5
$\ L_{iB}^{w,j}\ _2$	14.2	15.7	10.3	6.4	4.5	15.1	13.7	9.6	5.7	5.1
$\ L_{iB}^{w,3}\ _2$	2.6	0.9	1.2	1.4	1.3	2.3	1.9	1.6	1.1	0.4
$\ L_{iB}^{w,(1,2)}\ _2$	14.0	15.7	10.2	6.2	4.3	14.9	13.5	9.4	5.6	5.1

between open-loop poles $s_{P_i}^{OL,p}$ and closed-loop poles $s_{P_i}^{w,p}$ in the pole map are given for each of the $p \in [1,3]$ poles of the wind and blade model. Additionally, the related average distance

$$\Delta s_{P_i}^{w,\bar{p}} = \frac{1}{3} \sum_{p=1}^3 \Delta s_{P_i}^{w,p} = \frac{1}{3} (\Delta s_{P_i}^{w,1} + \Delta s_{P_i}^{w,2} + \Delta s_{P_i}^{w,3}) \quad (24)$$

of the single submodel i and the average distance

$$\Delta s_{P_j}^{w,\bar{p}} = \frac{1}{4} \sum_{i=15}^{18} \Delta s_{P_i}^{w,\bar{p}} = \frac{1}{4} (\Delta s_{P_{15}}^{w,\bar{p}} + \Delta s_{P_{16}}^{w,\bar{p}} + \Delta s_{P_{17}}^{w,\bar{p}} + \Delta s_{P_{18}}^{w,\bar{p}}) \quad (25)$$

of all four submodels are listed in Table 2.

Error feedback gains

The related error-feedback gain matrices $L_{iB}^{w,j}$ (with their j elements $L_{iB}^{w,j}$ described in Eq. 22) are summarised in Table A8. To assess the $L_{iB}^{w,j}$ deviations resulting from the shrinking pole regions, two different mean Euclidian norms $\|L_{iB}^{w,j}\|_2$ and $\|L_{iB}^{w,j_3 / j_{1,2}}\|_2$ are calculated as metrics of the error-feedback gains $L_{iB}^{w,j}$ for each of the $w \in [A, E]$ global wind speed observers and $w \in [F, J]$ local wind speed observers and both metrics averaging all, four incorporated submodels

$i \in [15, 18]$:

The metric $\|L_{iB}^{w,j}\|_2$ cumulates all three $L_{iB}^{w,j}$ elements (with $j \in [1, 3]$)

$$\begin{aligned} \|L_{iB}^{w,j}\|_2 &= \frac{1}{4} \sum_{i=15}^{18} \|L_{iB}^{w,j}\|_2 \\ &= \frac{1}{4} (\|L_{15B}^{w,j}\|_2 + \|L_{16B}^{w,j}\|_2 + \|L_{17B}^{w,j}\|_2 + \|L_{18B}^{w,j}\|_2) \end{aligned} \quad (26)$$

with $\|L_{iB}^{w,j}\|_2 = \sqrt{(L_{iB}^{w,1})^2 + (L_{iB}^{w,2})^2 + (L_{iB}^{w,3})^2}$. The metrics $\|L_{iB}^{w,j_3/\bar{j}_{1,2}}\|_2$ are calculated separately for L_{iB}^{w,j_3} (to gain the wind speed state \hat{v} , see Eq. 22) and $L_{iB}^{w,j_{1,2}}$ (to gain the blade states \hat{x}_B and \ddot{x}_B):

$$\begin{aligned} \|L_{iB}^{w,j_3/\bar{j}_{1,2}}\|_2 &= \frac{1}{4} \sum_{i=15}^{18} \|L_{iB}^{w,j_3/\bar{j}_{1,2}}\|_2 \\ &= \frac{1}{4} \left(\|L_{15B}^{w,j_3/\bar{j}_{1,2}}\|_2 + \|L_{16B}^{w,j_3/\bar{j}_{1,2}}\|_2 + \|L_{17B}^{w,j_3/\bar{j}_{1,2}}\|_2 + \|L_{18B}^{w,j_3/\bar{j}_{1,2}}\|_2 \right) \\ \text{with } \|L_{iB}^{w,j_3}\|_2 &= \sqrt{(L_{iB}^{w,3})^2} \\ \text{and } \|L_{iB}^{w,j_{1,2}}\|_2 &= \sqrt{(L_{iB}^{w,1})^2 + (L_{iB}^{w,2})^2}. \end{aligned} \quad (27)$$

The results are depicted in **Table 3**.

Membership functions

To visualise the convex blending of direct adjacent submodels (A_i, B_i), the time series of the membership functions h_i for a step-shaped wind time series are depicted in **Figure 5**. Note: As the premise variable in this contribution is defined by *one* parameter ($z \equiv \hat{v}$), the membership function $h_i(z)$ is identical to the weighting function $w_{k,i}(z)$ (see **Section 2.4.3**, especially footnote⁴). Therefore, just *two* submodels are blended at the same time (for all operating points between two steady state operating points OP_i ; see also the triangular-shaped $h_i(z)$ progression (with $h_i(z) \equiv w_{k,i}(z_k)$ for $z_k = z_1 = z$ with $k_{\max} = 1$) in **Figure 1**).

Wind turbine simulations, actuation signals, pitch rate deviations, rotation speed deviations and resulting mechanical loads

To assess the closed-loop system dynamics (see **Figure 6**), a disturbing step-shaped time series of 150 s duration is applied to the system, visualising the effect of the particular wind speed observer design (with decreased performance, see **Section 3.1**) on the actuation signals. The resulting time series $\hat{v}^w(t)$, $\beta_{FF}^w(t)$ and $\beta_{FB}^w(t)$ were achieved with NREL FAST 5 MW reference wind turbine simulations, embedded in a Matlab Simulink model, using the wind speed observer-based PDC controller **Eq. 5** (i.e., for all simulations the same PDC feedback-controller⁸ is utilised) with step-shaped wind time series as disturbing excitation. For clarity, the simulations are restricted to the full load range with pitch angle actuation, i.e., the generator torque is kept constant (see **Section 2.4.2** and Table A1).

The given wind speeds v and the reconstructed wind speeds \hat{v}^w of the step-shaped wind time series are depicted in the left column of **Figure 6**, the actuated feedforward and feedback pitch angles $\beta_{FF}^w(t)$ and $\beta_{FB}^w(t)$ are depicted in the mid and right column of **Figure 6**.

⁸PDC feedback controller: *ct1010001*

The resulting ultimate and mean pitch angle deviations, i.e. the pitch rates $\max(\dot{\beta})$ and $\text{mean}(\dot{\beta})$ are depicted in the left column of **Figures 7A,B**; the closed-loop system dynamics, i.e. the ultimate and standard drive train rotation speed deviations $\max(\Delta\omega) = \max(\frac{\omega-\omega_r}{\omega_r})$ and $\text{std}(\Delta\omega) = \text{std}(\frac{\omega-\omega_r}{\omega_r})$ (with the rated rotation speed ω_r), are depicted in the right column of **Figures 7A,B**.

In addition, the step-shaped wind time series – representing a typical wind increase and wind decrease event (i.e. a disturbing step up and step down signal) within a turbulent wind excitation time series – is used to assess the ultimate and fatigue loads resulting from the closed-loop system dynamics of both approaches. For this load assessment, five different loads are analysed that are crucial or at least very important for the wind turbine's main component design of tower, blades and the drive train:

- The tower bending moment *in wind direction* (also denoted with the *fore-aft* bending moment, *TwrBsMyt*) and *normal* to the wind direction (also denoted with the *side-to-side* bending moment, *TwrBsMxt*), calculated for the tower base,
- the blade bending moment *in wind direction* (i.e., resulting from blade bending out of the rotor plane, also denoted with the *flapwise* or *out-of-plane* bending moment, *RootMyb1*) and blade bending moment *inside the rotor plane* (*edgewise* or *in-plane* bending moment, *RootMxb1*), calculated for the blade root (next to the hub body) of the first blade, and
- the torsional torque (ΔTq) along the drive train axis.

The ultimate and fatigue loads, resulting from the closed-loop system dynamics, are depicted in **Figure 8**. To eliminate effects, resulting from simulation initialisation, the time period $0s \leq t_{CutOff} \leq 10s$ of all time series is not taken into account in the load evaluation. (Note: If the fatigue loads in **Figure 8** just differ even in the third or fourth decimal place, these differences are of significance. Because, the listed fatigue loads in **Figure 8** result from a wind times series of extremely short duration (of 150 s), while wind turbines typically operate for 20 years with approximately 97% availability. That is, also fourth decimal place differences in the fatigue loads, depicted in **Figure 8**, do have an enormous impact on the expected wind turbine lifetime, as all damages cumulate over the complete lifetime. This cumulation does not apply to ultimate loads, as these loads occur rarely in a wind turbine's lifetime).

4 DISCUSSION

Both Lyapunov approaches fulfill the *particular objective* to decrease the disturbance observer performance and feedforward actuation, while increasing the feedback actuation for shrinking pole region dimensions, resulting in mitigated mechanical loads (see **Section 3.1**). But the local approach achieves significantly increased flexibility and consistency of the desired and resulting system dynamics than the global approach:

Pole locations

The open-loop poles of the wind turbine system, depicted with gray symbols in **Figure 4**, are widely spaced, as significant distances between the real-valued wind model poles $s_{p,i}^{OL,1}$ (on

the real axis of the pole map) and the conjugated-complex blade poles $s_{P,i}^{OL,2 \vee 3}$ (in the pole map) exist. With increasing upper bounds $\alpha_{B,\max}$, resulting in shrinking pole regions⁹, the closed-loop pole locations $s_{P,i}^{w,p}$ are shifted for both design approaches inside the left half of the pole map in direction of the imaginary axis and towards the open-loop poles $s_{P,i}^{OL,p}$. Thereby, the average distances $\Delta s_{P,i}^{w,\bar{p}}$ in the complex pole map decrease (i.e., $\Delta s_{P,i}^{w,\bar{p}} > \Delta s_{P,i}^{(w+1),\bar{p}}$, e.g., $\Delta s_{P,i}^{B,\bar{p}} > \Delta s_{P,i}^{C,\bar{p}}$, see **Figure 4** and **Table 2**). Just for the last global and local wind speed observers the distances increase, i.e., $\Delta s_{P,i}^{D,\bar{p}} < \Delta s_{P,i}^{E,\bar{p}}$ and $\Delta s_{P,i}^{I,\bar{p}} < \Delta s_{P,i}^{J,\bar{p}}$, as the closed-loop poles (and their real-values $\text{Re}(s_{P,i}^{w,p})$, respectively) are shifted beyond the open-loop poles $\text{Re}(s_{P,i}^{OL,p})$ in direction of the imaginary axis. Therefore, those two wind speed observers *E* and *J* are not accounted in the following evaluations, but they are relevant for the local approach assessment, as described in the following with regard to the pole region violation.—For the wind speed observer poles, based on the local Lyapunov approach (see $w \in [F, I]$ in **Table 2**), the pole distances $\Delta s_{P,i}^{w,\bar{p}}$ decrease steadily, while the distances $\Delta s_{P,i}^{w,\bar{p}}$ for the wind speed observer poles, based on the global Lyapunov approach (see $w \in [A, D]$), show a local maximum for the wind speed observer *B* (i.e., $\Delta s_{P,i}^{B,\bar{p}} > \Delta s_{P,i}^{A,\bar{p}} > \Delta s_{P,i}^{C,\bar{p}} > \Delta s_{P,i}^{D,\bar{p}}$). Thus, the local Lyapunov approach is rated to be more coincident in regard to a desired pole location specification. Additionally, it is obvious from the pole locations depicted in **Figure 4**, that the local Lyapunov approach is more effective and flexible, respectively, as all poles are located *inside* the specified pole regions, while the global Lyapunov approach is not able to satisfy the pole region specifications for the smallest pole region, resulting in poles located *outside* the specified pole regions (see the pole location $s_{P,i}^{E,p}$ of the global wind speed observer *E* in **Figure 4E**). That is, the local Lyapunov approach is more flexible in assigning desired pole locations than the global Lyapunov approach.

(Note, that the LMI solver is capable to find solutions, even if the pole region LMIs are violated. However, the resulting poles are still located in the left half of the complex pole map and fulfill the basic stability condition for the corresponding closed-loop or error dynamics.)

Error-feedback gains

With the decreasing pole distances $\Delta s_{P,i}^{w,\bar{p}}$, also the error-feedback gains $L_{iB}^{w,j}$ —representing a measure or the effort of all error states acting on the w th observer—are mitigated. For all¹⁰ local wind speed observer designs these gains decrease expectedly with increasing upper bound $\alpha_{B,\max}$ and with the reduced pole distances $\Delta s_{P,i}^{w,\bar{p}}$ (i.e., $\|L_{iB}^{F,j}\|_2 > \|L_{iB}^{G,j}\|_2 > \|L_{iB}^{H,j}\|_2 > \|L_{iB}^{I,j}\|_2$).

Corresponding to the pole distances $\Delta s_{P,i}^{w,\bar{p}}$, the local wind speed observer gains $L_{iB}^{w,j}$ are mitigated steadily (and yield lower gains $\|L_{iB}^{B,j}\|_2 > \|L_{iB}^{G,j}\|_2, \|L_{iB}^{C,j}\|_2 > \|L_{iB}^{H,j}\|_2$ and $\|L_{iB}^{D,j}\|_2 > \|L_{iB}^{I,j}\|_2$), while the global wind speed observer gains show a local maximum of $\|L_{iB}^{w,j}\|_2$ for the wind speed observer *B* (i.e., $\|L_{iB}^{B,j}\|_2 > \|L_{iB}^{A,j}\|_2 > \|L_{iB}^{C,j}\|_2 > \|L_{iB}^{D,j}\|_2$), demonstrating the increased consistency of the desired and achieved error-feedback gains $L_{iB}^{w,j}$ of the local Lyapunov approach.

Wind speed reconstruction and actuation signals

With the mitigated error-feedback gains $L_{iB}^{w,j}$, the reconstructed states \hat{x} , especially the reconstructed wind speeds \hat{v} , are mitigated, too (see **Eq. 7¹¹**): While the reconstructed wind speed $\hat{v}^w(t_1)$ of a single and arbitrary time point $t = t_1$ decreases *steadily* for the wind speed observer design with local Lyapunov approach (i.e., $\hat{v}^F(t_1) \approx \hat{v}^G(t_1) > \hat{v}^H(t_1) > \hat{v}^I(t_1) > \hat{v}^J(t_1)$ ¹⁰, see left column in **Figure 6**), the reconstructed wind speed $\hat{v}^w(t_1)$ for the wind speed observer design with global Lyapunov approach decreases *unsteadily* (i.e., $\hat{v}^A(t_1) > \hat{v}^D(t_1) > \hat{v}^E(t_1) > \hat{v}^C(t_1) > \hat{v}^B(t_1)$, corresponding to the unsteady decrease of the mean Euclidean norm of the wind error state gains $\|L_{iB}^{w,\bar{3}}\|_2$ of the global wind speed observers (with $w \in [A, E]$, see **Table 3**; i.e., $\|L_{iB}^{A,\bar{3}}\|_2 > \|L_{iB}^{D,\bar{3}}\|_2 > \|L_{iB}^{C,\bar{3}}\|_2 > \|L_{iB}^{B,\bar{3}}\|_2$)¹⁰.

The mitigated, reconstructed wind speeds \hat{v}^w of both approaches result in the following actuation signals: The mitigated wind speeds \hat{v}^w lead to less dominant feedforward pitch actuation β_{FF} —just governed by the wind speed driven feedforward actuation $\underline{u}_{FF}(h_j(\underline{z}))$ (with $u_{FF} \equiv \beta_{FF}$ and $\underline{z} \equiv \hat{v}^w$) according to **Eq. 5** and the steady state pitch angles $\beta_{c,i}$ (listed in **Table A1**)—and increase the feedback pitch angles β_{FB} (see mid and right column in **Figure 6**). That is, corresponding to the decreased, reconstructed wind speeds \hat{v}^w (and mean wind state feedback gains $\|L_{iB}^{w,\bar{3}}\|_2$), the feedforward pitch angles of the local wind speed observers β_{FF}^w (with $w \in [F, I]$)¹⁰ are mitigated steadily (i.e., $\beta_{FF}^F \approx \beta_{FF}^G > \beta_{FF}^H > \beta_{FF}^I > \beta_{FF}^J$), see mid column in **Figure 6**) and yield steadily increased feedback pitch angles (i.e., $\beta_{FB}^F < \dots < \beta_{FB}^I < \beta_{FB}^J$), see right column in **Figure 6**), while the feedforward pitch angles of the global wind speed observers β_{FF}^w (with $w \in [A, D]$) decrease unsteadily (i.e., $\beta_{FF}^A < \beta_{FF}^D < \beta_{FF}^C < \beta_{FF}^B$) and yield correspondingly increasing feedback pitch angles β_{FB}^w (with $w \in [A, D]$). Thus, the intended particular objective of a steadily decreasing feedforward-actuation with increasing feedback-compensation

⁹to realise the shrinking pole region size, compare the pole regions' size (and location) in **Figure 4** for the wind speed observer design A and F with E and J

¹⁰The global and local wind speed observers *E* and *J* are not taken into account, because of their (closed-loop) pole locations, which are moved beyond the open-loop pole locations, as explained before in the subsection Pole locations.

¹¹If **Eq. 7** is evaluated for a single and arbitrary time point, it is obvious, that $\dot{\hat{x}}$ (and \hat{x}) decreases, if $L_{iB}^{w,j}$ is mitigated, while all other parameters and states do not change.

(see **Section 3.1**) is achieved with the local, not with the global Lyapunov wind speed observer design approach.

Pitch rate deviations and rotation speed deviations

The pitch angle deviations result in corresponding pitch and rotation speed metrics: With the steadily mitigated premise variable $\hat{v}^F > \hat{v}^G > \hat{v}^H > \hat{v}^I$ of the local wind speed observer F to I , the ultimate pitch rates $\max(\dot{\beta}) (= \max(\dot{\beta}_{FF} + \dot{\beta}_{FB}))$ with $\max(\dot{\beta}^F) \approx \max(\dot{\beta}^G) > \max(\dot{\beta}^H) > \max(\dot{\beta}^I)$ and mean pitch rates $\text{mean}(\dot{\beta})$ (with $\text{mean}(\dot{\beta}^F) \approx \text{mean}(\dot{\beta}^G) > \text{mean}(\dot{\beta}^H) > \text{mean}(\dot{\beta}^I)$) decrease steadily, too (see left column in **Figures 7A,B**), while the unsteadily decreasing wind speeds $\hat{v}^A > \hat{v}^D > \hat{v}^C > \hat{v}^B$ of the global wind speed observers A to D lead to unsteadily decreasing ultimate pitch rates $\max(\dot{\beta}^A) > \max(\dot{\beta}^D) > \max(\dot{\beta}^C) > \max(\dot{\beta}^B)$ and mean pitch rates $\text{mean}(\dot{\beta}^A) > \text{mean}(\dot{\beta}^D) > \text{mean}(\dot{\beta}^C) > \text{mean}(\dot{\beta}^B)$. As the pitch rates $\dot{\beta}$ decrease, increasing system dynamic deviations appear, e.g. represented by the metric of the drive train rotation speed deviations $\max(\Delta\omega)$ and $\text{std}(\Delta\omega)$, that increase steadily for the local wind speed observers (with $\max(\Delta\omega^F) \approx \max(\Delta\omega^G)$ and $\text{std}(\Delta\omega^F) \approx \text{std}(\Delta\omega^G)$) and unsteadily for the global wind speed observers (see right column in **Figure 7 A,B**). Thus, the local Lyapunov approach achieves the expected system dynamics with increased consistency between the desired and achieved pitch rate deviations and rotation speed deviations, compared to the global Lyapunov approach.

Load mitigation

The ultimate and fatigue loads resulting from closed-loop dynamics with the step-shaped wind excitation (depicted in **Figure 8**) are satisfying for both approaches, but especially for the local Lyapunov approach. In the following, the ultimate and fatigue loads are first evaluated separately for each of the two Lyapunov approaches (see **Table A9**), then the loads from both approaches are compared with each other (see **Table A10**).

- **Ultimate Loads** (see **Figure 8A**): The ultimate tower bending moments $\max(TwrBsMyt)$ and $\max(TwrBsMxt)$, as well as the ultimate in-plane blade bending moments $\max(RootMxb1)$ are mitigated for both Lyapunov approaches, and for the local Lyapunov approach these ultimate loads decrease even steadily (see lines 1 to 3 in **Table A9**; with one marginal exception for $\max^F(TwrBsMxt) < \max^G(TwrBsMxt)$). The ultimate out-of-plane blade bending moments $\max(RootMyb1)$ and the ultimate drive train torque $\max(\Delta Tq)$ increase slightly, but not significantly (see **Figure 8A** and lines 4 to 5 in **Table A9**). – Comparing both approaches with each other, for most pole regions lower ultimate loads are achieved with the local Lyapunov design approach (see **Figure 8B** and lines 1 to 5 in **Table A10**). If a controller variation yields comparable or even lower ultimate loads, the controller (design approach) assessment depends on the fatigue loads.
- **Fatigue Loads** (see **Figure 8B**): The fatigue tower bending moments $S_{eq}(TwrBsMyt)$ and $S_{eq}(TwrBsMxt)$, as well as

the fatigue drive train torque $S_{eq}(\Delta T)$ are mitigated for both Lyapunov approaches, and for the local Lyapunov approach these fatigue loads decrease steadily (see lines 6 to 8 in **Table A9**; with one exception for $S_{eq}^F(TwrBsMxt) < S_{eq}^G(TwrBsMxt)$). The fatigue out-of-plane blade bending moments $S_{eq}(RootMyb1)$ and in-plane blade bending moments $S_{eq}(RootMxb1)$ increase slightly, but for the local Lyapunov approach the out-of-plane blade bending fatigue loads $S_{eq}^w(RootMyb1)$ decrease steadily (see **Figure 8B** and lines 9 to 10 in **Table A9**). – Comparing both approaches with each other, for most pole regions lower fatigue loads are achieved with the local Lyapunov design approach, again (see **Figure 8B** and lines 6 to 10 in **Table A10**).

Additionally, the local wind speed observer design approach with the smallest pole region (see wind speed observer J) achieves the lowest fatigue loads for those bending moments and torsional torque compared to the fatigue loads, resulting from the global wind speed observer design (compare the fatigue loads of local wind speed observer J with the fatigue loads of all global wind speed observers for $S_{eq}(TwrBsMyt)$, $S_{eq}(TwrBsMxt)$ ¹², $S_{eq}(RootMyb1)$ and $S_{eq}(\Delta Tq)$ in **Figure 8**). – Just the in-plane blade fatigue bending moments $S_{eq}(RootMxb1)$ increase for the local wind speed observer design approach (as well as for the global approach). This fatigue load increase might result from the increasing rotation speed deviations $\max(\Delta\omega)$ and $\text{std}(\Delta\omega)$. Because, for identical wind times series and system excitations, respectively, increasing fatigue loads result from increasing amplitude and/ or increasing load cycles. As the rotation speed deviations $\max(\Delta\omega)$ and $\text{std}(\Delta\omega)$ increase with the shrinking pole region dimension (see $\max(\Delta\omega)$ and $\text{std}(\Delta\omega)$ for WindObs w with $w \in [G, J]$ in the right column of **Figures 7A,B**), it seems reasonable, that the fatigue load increase results from the increasing rotation speed, rather than from an amplitude increase. This assumption is also confirmed by the ultimate loads that decrease with the shrinking pole region size (see $\max(RootMxb1)$ for WindObs w with $w \in [F, J]$ in **Figure 8**), i.e. not the amplitudes of ultimate and fatigue loads, but the load cycles of the fatigue loads increase. Additionally, with the increasing rotor speed deviations, the number of blade passages through the tower shadow¹³ increases, resulting in increasing load cycles and leading to increased fatigue loads.

Therefore, it seems promising as a subject of future work, to split the wind speed observer, based on a local Lyapunov approach, into two, separated wind speed observers: A

¹²with two exceptions for the tower *side-to-side*-bending moments $S_{eq}^B(TwrBsMxt) < S_{eq}^G(TwrBsMxt)$ and $S_{eq}^C(TwrBsMxt) < S_{eq}^J(TwrBsMxt)$ (see **Figure 8B** and line 7 in **Table A9** as well as line 7 in **Table A10**).

¹³Tower shadow (effect): The tower poses as an obstacle in the inflow that increases the dynamic pressure and decreases the wind speed in front, i.e. in upwind direction of the tower.

feedforward wind speed observer—specified with a similar pole region (like wind speed observer I , used within this contribution)—with decreased reconstruction performance to assign the premise variable \hat{v}_{FF} to the feedforward actuation $\underline{u}_{FF}(h(\hat{v}_{FF}))$. To compensate the decreased feedforward pitch rates $\hat{\beta}_{FF}$, an additional *feedback* wind speed observer with increased performance will be implemented, to assign its reconstructed wind speed \hat{v}_{FB} to the feedback actuation. For this feedback wind speed observer, a tight pole region will be specified, located left from the open-loop poles, resulting in significantly increased error-feedback gains $L_{i,FB}$ and reconstructed wind speeds \hat{v}_{FB} , so that the feedback pitch rates $\hat{\beta}_{FB}$ increase, resulting in mitigated rotation speed deviations and fatigue loads. Summing up the load evaluation, the local Lyapunov approach is rated to gain an increased consistency between intended and achieved load mitigation compared to the global approach. While the ultimate loads, achieved with the global and local wind speed observer approach are comparable, the local approach achieves an increased fatigue load mitigation. Because, for steadily shrinking pole regions with steadily decreasing feedforward actuation, the fatigue loads decrease steadily, too.

5 CONCLUSION

With this contribution a local Lyapunov approach is introduced for controller and observer design in the Takagi-Sugeno framework. Compared to the common global Lyapunov approach, a more dedicated closed-loop dynamic is intended with the local Lyapunov approach, i.e., an increased flexibility in the design process and increased consistency between desired and achieved system dynamics is aspired.

The applicability and effectiveness of the local Lyapunov approach to wind turbine control is analysed with wind turbine simulations. The simulation results show, that the local Lyapunov approach makes it possible to influence the pole locations, the resulting error-feedback gains and closed-loop system dynamics more flexible and with an increased consistency between desired and achieved system dynamics

than the global Lyapunov approach. That is, with the local Lyapunov approach, the assignment of smaller pole regions is possible, enabling a higher flexibility in assigning desired system dynamics. Additionally, the local Lyapunov approach reaches an improved similarity between the desired and achieved closed-loop system dynamics, indicating the increased consistency of the local Lyapunov approach. As the local Lyapunov approach is less conservative, but more dedicated to the desired system dynamics, i.e., the design process achieves an increased flexibility and increased consistency, it is rated to have a higher potential in fulfilling primary and secondary control objectives, like energy yield optimisation and load mitigation in wind turbine application.

The evaluation of the local Lyapunov approach will be continued in oncoming studies, with the controller structure adapted to the new possibilities arising from the local design approach, like the separated reconstruction of the feedforward and the feedback premise variable in a split wind speed observer.

DATA AVAILABILITY STATEMENT

The original contributions presented in the study are included in the article/Supplementary Material, further inquiries can be directed to the corresponding author.

AUTHOR CONTRIBUTIONS

Eckhard Gauterin and Horst Schulte devised the basic Takagi-Sugeno observer-based feedforward control concept for wind turbine control and conceptualized this study. Florian Pöschke developed the linearised wind turbine model in Takagi-Sugeno framework and conceived the local Lyapunov approach. Florian Pöschke expedited the implementation and development of the control concept in the simulation software, while Eckhard Gauterin conducted the simulation studies. All three authors analysed the results and Eckhard Gauterin created the manuscript, that was reviewed by all three authors.

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APPENDIX

Specification of all steady operation points OP_i

In Table A1 all steady Operation Points OP_i of this work are listed.

Specification of the component submodels

In Tables A2–A8, all matrices and steady states are listed, necessary to replicate the achieved results of this work with the help of NREL FAST 5MW reference wind turbine simulation software (see NREL (XXXX), Jonkman and Buhl (2005) and Jonkman et al. (2009)). Note: As the disturbing wind time series and excitation, respectively is restricted to wind speeds from $v = 14m/s$ to $v = 16m/s$, the corresponding i submodels are restricted to $i \in$ (Lendek et al., 2010; Jonkman and Jonkman, 2016) (see also explanation in Section 3.2).

All states and parameters are expressed in SI units, despite the generator torque T_G (that is expressed in kNm), pitch angles (that are expressed in (angular) degree) and the rotation speed (expressed in revolution per minute).

Specification of the LMI constraints

Restrictions for the decay rates $\alpha_{\min} / \alpha_{\max}$ and natural frequencies of the system response (characterised by the real part and imaginary part of the closed-loop poles $\text{Re}(s_{p,i})$ and $\text{Im}(s_{p,i})$) can be specified with bounds, e.g. with horizontal lines (with $\alpha_{\max} < \text{Re}(s_{p,i}) < \alpha_{\min}$) and diagonal origin lines (with $\tan(\theta) < \frac{\text{Im}(s_{p,i})}{\text{Re}(s_{p,i})} \Leftrightarrow \theta < \arctan \frac{\text{Im}(s_{p,i})}{\text{Re}(s_{p,i})}$, also denoted as *cone angle*) in the complex pole map, resulting in bounded pole map segments and pole regions, respectively (Pöschke et al., 2019).

In (Chilali and Pascal, 1996), LMI representations for these bounded pole map segments, i.e. the restricted location of the poles and eigenvalues of a linear system, respectively, are derived.

For the upper vertical bound $\alpha_{B,\max}$ of the error dynamics, based on the blade design model, the following LMI holds (with $N_i^w = P_i^w L_{iB}^w$, defined in (20)):

$$A_{iB}^T P_i - C_B^T N_i^{wT} + P_i A_{iB} - N_i^w C_B > -2\alpha_{B,\max} P_i, \quad (23)$$

and for the lower vertical bound $\alpha_{B,\min}$ it holds:

$$A_{iB}^T P_i - C_B^T N_i^{wT} + P_i A_{iB} - N_i^w C_B < -2\alpha_{B,\min} P_i. \quad (24)$$

For the cone angle bound θ of the error dynamics, based on the blade design model, the following LMI holds:

$$\begin{bmatrix} \sin(\theta) \left(-A_{iB}^T P_i - C_B^T N_i^{wT} + P_i A_{iB} - N_i^w C_B \right) \\ \cos(\theta) \left(-A_{iB}^T P_i + C_B^T N_i^{wT} + P_i A_{iB} - N_i^w C_B \right) \\ \cos(\theta) \left(-A_{iB}^T P_i - C_B^T N_i^{wT} - P_i A_{iB} + N_i^w C_B \right) \\ \sin(\theta) \left(-A_{iB}^T P_i - C_B^T N_i^{wT} + P_i A_{iB} - N_i^w C_B \right) \end{bmatrix} < 0. \quad (25)$$

The matrices used in (23) to (25) are listed in Tables A2, A4, A8.

All simulations were executed with the controller *ct1210013*. The following global Lyapunov approach based wind speed observers were used: $A \equiv ot1210027$, $B \equiv ot1210030$, $C \equiv ot1210033$, $D \equiv ot1210051$, $E \equiv ot1210049$, $F \equiv ot1210036$, $G \equiv ot1210039$, $H \equiv ot1210042$, $I \equiv ot1210043$, $J \equiv ot1210044$.

The state observer *ot1010001* was utilised.

For the wind excitation the *Imp14.hh* wind time series is used.

To calculate the *mean* Euclidian norm $\|L_{iB}^{w,j}\|_2$ of the error-feedback gains $L_{iB}^{w,j}$ [see (26)] and the *average, mean* Euclidian norm $\|L_{iB}^{w,j}\|_2$ [see (27)] the worksheet *Uebersicht_L_Matrizen_Pitchwinkel-YYYY_MM_DD.xlsx* is used.

LOAD ANALYSIS

For the ultimate loads max^w and fatigue loads S_{eq}^w , resulting from five different wind speed observers (i.e. for the $w \in [A, E]$ global wind speed observers and $w \in [A, E]$ local wind speed observers; see Figure 8), the steady increase or decrease of the loads is evaluated separately for each of the two observer approaches (see Table A9) and in comparison to each other (see Table A10).

TABLE A1 | States of the i steady state operations points OP_i of the NREL FAST 5MW reference wind turbine with the wind speed $v_{c,i}$, rotor rotational speed $\omega_{R,c,i}$, generator torque $T_{G,c,i}$ and pitch angle $\beta_{c,i}$.

i	$v_{c,i}$	$\beta_{c,i}$	$T_{G,c,i}$	$\omega_{R,c,i}$
1	3	0	2.912	3.4
2	4	0	5.193	4.6
3	5	0	8.079	5.7
4	6	0	11.646	6.9
5	7	0	15.843	8.0
6	8	0	20.671	9.2
7	9	0	26.128	10.3
8	9.5	0	29.094	10.9
9	10	0	32.267	11.4
10	10.5	0	35.547	12.0
11	11	0	40.433	12.1
12	11.5	2.2	43.094	12.1
13	12	4.1	43.094	12.1
14	12.5	5.5	43.094	12.1
15	13	6.6	43.094	12.1
16	14	8.6	43.094	12.1
17	15	10.4	43.094	12.1
18	16	12.0	43.094	12.1
19	17	13.4	43.094	12.1
20	18	14.8	43.094	12.1
21	19	16.1	43.094	12.1
22	20	17.4	43.094	12.1
23	21	18.6	43.094	12.1
24	22	19.7	43.094	12.1
25	23	20.8	43.094	12.1
26	24	22.0	43.094	12.1
27	25	23.0	43.094	12.1

TABLE A2 | State matrices A_{iB} and augmented state matrices \tilde{A}_{iB} of the B lade model (for the submodels $i \in [15,18]$).

A_{15B}	0 -21.82 -	1 -5.41 -	\tilde{A}_{15B}	0 -21.82 0	1 -5.41 0	0 9.58 -0.25
A_{16B}	0 -21.88 -	1 -5.36 -	\tilde{A}_{16B}	0 -21.88 0	1 -5.36 0	0 9.51 -0.25
A_{17B}	0 -21.92 -	1 -5.35 -	\tilde{A}_{17B}	0 -21.92 0	1 -5.35 0	0 9.48 -0.25
A_{18B}	0 -21.95 -	1 -5.30 -	\tilde{A}_{18B}	0 -21.95 0	1 -5.30 0	0 9.38 -0.25

TABLE A3 | Input matrices B_{iB} and augmented input matrices \tilde{B}_{iB} of the B lade model (for the submodels $i \in [15,18]$).

B_{15B}	0 -563.53 -	0 0 -	\tilde{B}_{15B}	0 -563.53 0	0 0 0
B_{16B}	0 -589.16 -	0 0 -	\tilde{B}_{16B}	0 -589.16 0	0 0 0
B_{17B}	0 -606.41 -	0 0 -	\tilde{B}_{17B}	0 -606.41 0	0 0 0
B_{18B}	0 -628.21 -	0 0 -	\tilde{B}_{18B}	0 -628.21 0	0 0 0

TABLE A4 | Common output matrix C_B and augmented common output matrix \tilde{C}_B of the *Blade* model (for all submodels).

C_B	1	0	\tilde{C}_B	1	0	0
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TABLE A5 | Steady states $\underline{x}_{c,iB}$ and augmented steady states $\tilde{\underline{x}}_{c,iB}$ of the *Blade* model (for the submodels $i \in [15,18]$).

$\underline{x}_{c,15B}$	3.61 0 -	$\tilde{\underline{x}}_{c,15B}$	3.61 0 13
$\underline{x}_{c,16B}$	3.15 0 -	$\tilde{\underline{x}}_{c,16B}$	3.15 0 14
$\underline{x}_{c,17B}$	2.73 0 -	$\tilde{\underline{x}}_{c,17B}$	2.73 0 15
$\underline{x}_{c,18B}$	2.44 0 -	$\tilde{\underline{x}}_{c,18B}$	2.44 0 16

TABLE A6 | Steady state pitch angle $\beta_{c,i}$ and generator torque $T_{G,i}$ (for the submodels $i \in [15,18]$).

$\beta_{c,15}$ $T_{G,15}$	6.6 43.094
$\beta_{c,16}$ $T_{G,16}$	8.6 43.094
$\beta_{c,17}$ $T_{G,17}$	10.4 43.094
$\beta_{c,18}$ $T_{G,18}$	12.0 43.094

TABLE A7 | State feedback matrices K_{iR} of the (rigid body) *R*otation drive train model (for the submodels $i \in [15,18]$).

K_{15R}	-1.31 0
K_{16R}	-1.03 0
K_{17R}	-0.85 0
K_{18R}	-0.71 0

TABLE A8 | Error state feedback gain matrices $L_{1B}^{w,j}$ of the blade model based wind speed observers B for:

- global Lyapunov approach with $w \in [A, E]$
- local Lyapunov approach with $w \in [F, J]$
- submodels $i \in [15, 18]$
- matrix elements $j \in [1, 3]$.

	A	B	C	D	E	F	G	H	I	J
$L_{15B}^{w,1}$	3.87	2.81	2.04	1.28	0.30	4.06	3.02	1.79	0.38	-1.35
$L_{15B}^{w,2}$	-13.31	-15.30	-9.80	-5.92	-4.22	-14.19	-12.90	-8.78	-5.23	-4.46
$L_{15B}^{w,3}$	2.56	0.88	1.21	1.43	1.31	2.27	1.90	1.61	1.11	0.39
$L_{16B}^{w,1}$	3.87	2.81	2.04	1.28	0.30	4.07	3.07	1.84	0.43	-1.30
$L_{16B}^{w,2}$	-13.44	-15.45	-9.95	-6.05	-4.28	-14.33	-13.14	-9.18	-5.57	-4.85
$L_{16B}^{w,3}$	2.56	0.88	1.20	1.43	1.31	2.26	1.91	1.61	1.12	0.40
$L_{17B}^{w,1}$	3.88	2.81	2.04	1.28	0.30	4.07	3.08	1.85	0.44	-1.29
$L_{17B}^{w,2}$	-13.49	-15.50	-10.01	-6.10	-4.31	-14.39	-13.22	-9.30	-5.68	-4.98
$L_{17B}^{w,3}$	2.56	0.88	1.20	1.43	1.31	2.26	1.92	1.61	1.12	0.40
$L_{18B}^{w,1}$	3.88	2.81	2.04	1.28	0.29	4.08	3.13	1.90	0.50	-1.23
$L_{18B}^{w,2}$	-13.60	-15.62	-10.13	-6.19	-4.33	-14.51	-13.43	-9.70	-6.00	-5.39
$L_{18B}^{w,3}$	2.56	0.88	1.20	1.43	1.31	2.26	1.93	1.61	1.13	0.40

TABLE A9 | Analysis of the ultimate loads \max^w and fatigue loads S_{eq}^w resulting from five different wind speed observers regarding the steady increase or decrease of the loads (evaluated separately for each of the two Lyapunov approaches with $w \in [A, E]$ for the global wind speed observers and with $w \in [F, J]$ for the local wind speed observers; based on the loads depicted in Figure 8).

Loads	Global Lyapunov approach	Local Lyapunov approach
<i>TwrBsMxt</i>	$\max^A > \max^B > \max^C > \max^D > \max^E$	$\max^F > \max^G > \max^H > \max^I > \max^J$
<i>TwrBsMxt</i>	$\max^A > \max^B < \max^C < \max^D < \max^E$	$\max^F \approx \max^G > \max^H > \max^I > \max^J$
<i>RootMxb1</i>	$\max^A > \max^B < \max^C < \max^D < \max^E$	$\max^F > \max^G > \max^H > \max^I > \max^J$
<i>RootMyb1</i>	$\max^A < \max^B > \max^C > \max^D > \max^E$	$\max^F > \max^G < \max^H < \max^I > \max^J$
ΔT	$\max^A < \max^B > \max^C > \max^D > \max^E$	$\max^F > \max^G > \max^H > \max^I \approx \max^J$
<i>TwrBsMxt</i>	$S_{eq}^A > S_{eq}^B < S_{eq}^C < S_{eq}^D < S_{eq}^E$	$S_{eq}^F > S_{eq}^G > S_{eq}^H > S_{eq}^I > S_{eq}^J$
<i>TwrBsMxt</i>	$S_{eq}^A > S_{eq}^B < S_{eq}^C < S_{eq}^D < S_{eq}^E$	$S_{eq}^F < S_{eq}^G > S_{eq}^H > S_{eq}^I > S_{eq}^J$
ΔT	$S_{eq}^A > S_{eq}^B \approx S_{eq}^C > S_{eq}^D > S_{eq}^E$	$S_{eq}^F > S_{eq}^G > S_{eq}^H > S_{eq}^I > S_{eq}^J$
<i>RootMyb1</i>	$S_{eq}^A > S_{eq}^B < S_{eq}^C < S_{eq}^D > S_{eq}^E$	$S_{eq}^F > S_{eq}^G > S_{eq}^H > S_{eq}^I > S_{eq}^J$
<i>RootMxb1</i>	$S_{eq}^A < S_{eq}^B = S_{eq}^C = S_{eq}^D = S_{eq}^E$	$S_{eq}^F \approx S_{eq}^G \approx S_{eq}^H < S_{eq}^I < S_{eq}^J$

TABLE A10 | Analysis of the ultimate loads \max^w and fatigue loads S_{eq}^w resulting from five different wind speed observers regarding the steady increase or decrease of the loads (comparing both Lyapunov approaches with each other with $w \in [A, E]$ for the global wind speed observers and with $w \in [F, J]$ for the local wind speed observers; based on the loads depicted in Figure 8).

Loads	A \Leftrightarrow F	B \Leftrightarrow G	C \Leftrightarrow H	D \Leftrightarrow I	E \Leftrightarrow J
<i>TwrBsMxt</i>	$\max^A > \max^F$	$\max^B < \max^G$	$\max^C < \max^H$	$\max^D > \max^I$	$\max^E > \max^J$
<i>TwrBsMxt</i>	$\max^A > \max^F$	$\max^B < \max^G$	$\max^C < \max^H$	$\max^D > \max^I$	$\max^E > \max^J$
<i>RootMyb1</i>	$\max^A < \max^F$	$\max^B > \max^G$	$\max^C > \max^H$	$\max^D > \max^I$	$\max^E > \max^J$
<i>RootMxb1</i>	$\max^A > \max^F$	$\max^B < \max^G$	$\max^C < \max^H$	$\max^D > \max^I$	$\max^E > \max^J$
ΔT	$\max^A < \max^F$	$\max^B > \max^G$	$\max^C > \max^H$	$\max^D > \max^I$	$\max^E < \max^J$
<i>TwrBsMxt</i>	$S_{eq}^A > S_{eq}^F$	$S_{eq}^B < S_{eq}^G$	$S_{eq}^C < S_{eq}^H$	$S_{eq}^D > S_{eq}^I$	$S_{eq}^E > S_{eq}^J$
<i>TwrBsMxt</i>	$S_{eq}^A > S_{eq}^F$	$S_{eq}^B < S_{eq}^G$	$S_{eq}^C < S_{eq}^H$	$S_{eq}^D > S_{eq}^I$	$S_{eq}^E > S_{eq}^J$
<i>RootMyb1</i>	$S_{eq}^A < S_{eq}^F$	$S_{eq}^B < S_{eq}^G$	$S_{eq}^C > S_{eq}^H$	$S_{eq}^D > S_{eq}^I$	$S_{eq}^E > S_{eq}^J$
<i>RootMxb1</i>	$S_{eq}^A > S_{eq}^F$	$S_{eq}^B > S_{eq}^G$	$S_{eq}^C > S_{eq}^H$	$S_{eq}^D < S_{eq}^I$	$S_{eq}^E < S_{eq}^J$
ΔT	$S_{eq}^A > S_{eq}^F$	$S_{eq}^B < S_{eq}^G$	$S_{eq}^C > S_{eq}^H$	$S_{eq}^D > S_{eq}^I$	$S_{eq}^E > S_{eq}^J$