



# **Compact Quantum Random Number Generator with Silicon Nanocrystals Light Emitting Device Coupled to a Silicon Photomultiplier**

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A small-sized photonic quantum random number generator, easy to be implemented in small electronic devices for secure data encryption and other applications, is highly demanding nowadays. Here, we propose a compact configuration with Silicon nanocrystals large area light emitting device (LED) coupled to a Silicon photomultiplier to generate random numbers. The random number generation methodology is based on the photon arrival time and is robust against the non-idealities of the detector and the source of quantum entropy. The raw data show high quality of randomness and pass all the statistical tests in national institute of standards and technology tests (NIST) suite without a post-processing algorithm. The highest bit rate is 0.5 Mbps with the efficiency of 4 bits per detected photon.

Keywords: compact photonic quantum random number generation, silicon nanocrystals LED, silicon photomultiplier, robust methodology, NIST tests

## **1. INTRODUCTION**

Thanks to the quantum properties of light, "truly" random numbers can be produced by photonic quantum random number generators (PQRNG). Cryptographic tasks of encryption and decryption of private data can be executed using secret keys based on high quality random numbers. Even though mathematical algorithms are extensively used to generate random numbers, they suffer from high guessability provided the seed of the algorithm is known. If they have a short periodicity, their repeatability would be a serious flaw, as well.

PQRNGs benefit from the intrinsically random and unpredictable properties of physical processes involving photons as the quanta of light. The randomness in path taken by photons arriving on a beam splitter<sup>1</sup> [1], the comparison of the waiting time for photon arrivals in adjacent time intervals [2] and the combination of both methods [3] have been used to generate random numbers. In some other works, encoding the number of arriving photons in observation windows [4–6] and the randomness in the photon arrival times [7–9] have been used to produce random numbers. Recently, a robust approach based on arrival times of photons has been proposed by our group [9]. It considers all the non-idealities of the source as well as the detector, producing high quality random numbers which pass all the statistical tests in national institute of standards and technology (NIST) tests suite and TestU01 without a post-processing algorithm<sup>1</sup>.

<sup>1</sup>Dataset. http://www.idquantique.com/wordpress/wp-content/uploads/white-paper-understanding-qkd.pdf. (2016).

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However, in all the above-mentioned approaches a bulky setup is used to generate random numbers. A small-sized and compact PQRNG, easy to be implemented in small electronic devices such as mobile phones and cameras for secure data encryption and decryption as well as other applications, is highly essential for facile accessibility to everyone. Here, we present a first step toward this goal: a PQRNG with a novel, compact configuration comprising a silicon nanocrystals large area LED (Si-NCs LLED) coupled with a silicon photomultiplier (SiPM) in free space. Based on some statistical analyses described in section 5, it is proved that the generated hexadecimal random symbols have a very high quality and the corresponding random bits pass all the statistical tests in NIST tests suite with no post-processing operations. The highest bit rate of 0.5 Mbps is achieved with the efficiency of 4-bits per detected photon.

In a previous work [9], we have demonstrated a procedure to extract high quality random numbers from the arrival time of photons emitted by a Si-NCs LED, collected by a fiber bundle and detected by a commercial single photon avalanche diode (SPAD). In this work we use the same robust methodology of random number extraction of Bisadi et al. [9] but we demonstrate a PQRNG where the LLED is directly faced to a SiPM. Moreover, both devices are fabricated by the same silicon pilot line of Bruno Kessler Foundation (FBK) with a dimension allowing optimum coupling which is specifically challenging for the Si-NCs LLED.

This work is organized as follows. In section 2, the Si-NCs LLEDs and their electrical and optical characteristics are described. In section 3, the SiPM is introduced and explained. Section 4 describes the experimental procedure and random numbers extraction. Randomness analyses are discussed in section 5 and at the end the conclusions are presented.

#### 2. SI-NCS LARGE AREA LED

Si-NCs are silicon quantum dots which emit light at room temperature in the visible range due to quantum confinement. The emitted photons are emitted independently by a quantum process named spontaneous emission and their statistics obey Poisson statistics (see more in section 4).

Si-NCs LEDs are fabricated by complementary metaloxide-semiconductor (CMOS) processing, they can be easily incorporated in integrated configurations, they emit photons with wavelengths in the spectral range detectable by silicon detectors allowing the fabrication of an all-silicon-based device and since the spontaneous emission of photons in a Si-NCs LED is a non-deterministic, quantum mechanical and random process, they can be used as a quantum source of randomness to generate random numbers. The Si-NCs LEDs were fabricated with a large emitting surface in order to illuminate large area detectors like the SiPM we use here. The matching of the emitter and detector surfaces allows their direct coupling, i.e., without any coupling optics. The Si-NCs LLED (large area LED) has the active layer structure formed by a multilayer structure with 5 periods of silicon rich oxide (SRO)/SiO<sub>2</sub> layers of 3.5-4 nm and 2 nm thicknesses, respectively (Figure 1A). The Si-NCs are grown in a silica matrix through the plasma enhanced chemical vapor deposition (PECVD) technique and annealed at 1150°C for 30 min to form the Si-NCs.

The Si-NCs LLEDs have been prepared in three different sizes: big (b), medium (m) and small (s) with emitting surface area of  $1.3 \text{ mm} \times 0.99 \text{ mm}$ ,  $0.99 \text{ mm} \times 0.82 \text{ mm}$  and  $1.02 \text{ mm} \times 0.11 \text{ mm}$ , respectively (see **Figure 1B**).

The electroluminescence (EL) spectra of the Si-NCs LLEDs can be seen in **Figure 2A** with a high peak at ~ 900 nm attributed to the emission from Si-NCs. Note that all the LLEDs show the same EL lineshape which points to the great uniformity of the fabrication. **Table 1** reports the optoelectronic characteristics of these LLEDs. The figure of merit is the efficiency of the EL which is measured as the ratio between the EL intensity and the driving electrical power. The electrical power density is calculated to be 5.87, 0.93 and 0.08 mW/cm<sup>2</sup> for the (b), (m) and (s) LLEDs, respectively. It should be noted that the applied currents to the (b), (m) and (s) LLEDs are 30, 3 and 3  $\mu$ A, respectively.

At currents lower than  $30\mu$ A to the (b) LLED, no appreciable EL is observed. Therefore, by applying the previously-mentioned currents to the LLEDs, we tried to keep the voltages and hence the electric field through the active area of the LLEDs (with actual thickness of ~22.5 nm) more or less the same. The low current density and high EL intensity of the (m)LLED yield the higher efficiency of this LLED compared with the (b) and (s) LLEDs. In addition to the efficiency, the active area of (m)LLED allows a suitable coupling with the large area SiPM since the SiPM dimensions are of 1 mm × 1 mm. It should be noted that all three LLEDs result in the generation of high quality random numbers since they are fabricated in a very similar way and the detected photons from all of them follow a Poisson distribution.

The current-voltage (I/V) characteristics of Si-NCs LLED are presented in Figure 2B. They show a rectifying behavior with more current density at forward regime-i.e., negative voltage applied to the cathode and zero voltage to the anode-than the reverse regime-i.e., positive voltage applied to the cathode and zero voltage to the anode. It is observed that at a fixed forward voltage, the current density through the (b) LLED is larger than (m) and (s) LLEDs (particularly at 0.5-3 V): this is due to the larger free carrier density flowing through the active area in (b) than (m) and (s) LLEDs. In the reverse bias region (1-6 V), however, the (b), (m), and (s) LLEDs show the same order of magnitude current densities that is related to the inefficient carrier injection to the active area by the accumulation of the charges near the boundaries of the cathode and the anode. This effect blocks the carriers from flowing through, recombining and contributing to the net current and consequently makes the current density independent of the gate areas of the LLEDs<sup>2</sup>. The charge blocking seems to be more effective for (s) LLED with a flat I/V curve at the region of -0.5-1 V (Figure 2B).

These LLEDs can emit light over long hours of operation. **Figure 3** shows the EL of the (m) LLED over  $\sim 16$  h. Note that the EL variation is compensated by the randomness extraction [9] and does not influence the quality of random numbers and therefore no adjustment of the QRNG is needed over a very long working period.

<sup>&</sup>lt;sup>2</sup>Dataset. "http://www.ieee.li/pdf/essay/pin\_diode\_handbook.pdf/" (1998).







FIGURE 2 | (A) The EL spectra of the three different sizes of Si-NCs LLED illustrated in Figure 1A. A high peak at ~900 nm can be seen which is attributed to the emission from Si-NCs. The EL spectra are acquired by a Spectra-Pro 2300i monochromator coupled with a nitrogen-cooled charge coupled device (CCD) camera. The measurements were performed at room temperature in a dark chamber. EL intensity is normalized for the responsivity of the spectrometer. (B) The I/V characteristics of the Si-NCs LLEDs. A quite rectifying behavior is observed in the I/V curves with more current density at forward regime—i.e., negative voltage applied to the n-type contact electrode and zero voltage to the p-type contact electrode—than reverse regime—i.e., positive voltage applied to the n-type contact electrode.

	TABLE 1	The efficiency (E	over injected power	) of (b), (m), and (s)LLED
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LLED	Area (mm <sup>2</sup> )	Voltage (V)	Current Density (mA/cm <sup>2</sup> )	EL (Kcps)	Efficiency (Kcps/W)
b	1.29	2.52	2.33	~533	70
m	0.81	2.34	0.37	~363	517
S	0.11	3	2.67	~364	413

## **3. SILICON PHOTOMULTIPLIER**

The analog SiPM is an array of many (hundreds) of single photon avalanche diodes (SPADs). They are all connected in parallel to a common anode and cathode, each one with its own quenching resistor. Each cell (i.e., SPAD+resistor) is sensitive to a single photon and provides a current pulse at the output. Therefore, the counts at the SiPM output are proportional to the number of triggered cells, thus to the number of detected photons. Different technologies for SiPM have been developed in FBK during last few years, with peak sensitivity in the green part (RGB-SiPM) or in the blue part (NUV-SiPM) of the visible spectrum, and with different cell sizes. The NUV technology, in particular, benefits from an upgraded silicon material [10], employing an epi/substrate structure with a lowerlifetime substrate. This gives particular benefits in terms of correlated noise reduction, i.e., afterpulsing probability (AP) and delayed crosstalk probability (DeCT), which are particularly detrimental for QRNG applications [9].

In this work we employ a 1 mm ×1 mm NUV SiPM (inset in **Figure 4B**), containing 625 cells (SPADs) with the cell size of 40  $\mu$ m and the fill factor (FF) of 60 %. This particular technology has a photon detection efficiency (PDE) (and in particular spectral sensitivity) not matched to the LLED emission. PDE is about 5% at 800 nm, at 4 V of excess bias (i.e., the difference between the applied bias and the breakdown voltage  $(V_{bd})$ ), as shown in **Figure 4A**. However, NUV SiPM has the advantage of a low primary dark count rate (DCR), less than 100 kcps/mm<sup>2</sup> at 5 V of excess bias (see **Figure 4B**), meaning less than 200 cps per single SPAD. In addition, it has a reduced correlated noise probability (overall AP+DeCT probability lower than 5%), which is very important in this kind of application.

We designed a custom front-end board to amplify and digitalize the analog output signal from the detector (see **Figure 5B**). This is based on a AD8000 amplifier in a transimpedance configuration, followed by a comparator with an adjustable voltage threshold and a monostable, creating pulses of 3.3 V and 100 ns width. This fixes the maximum count



**FIGURE 3** | EL of the (m) LLED vs. time at the applied current density of 0.62 mA/cm<sup>2</sup> (corresponding to an applied voltage of 2.61 V). The right hand side of the figure (blue axis) shows the EL variation percentage.

rate of the detection system, which is anyhow limited by the afterpulsing time constant of the detector, giving an overall time to let all traps to empty, thus an overall time to avoid any possible afterpulsing, of few hundreds of nanoseconds, as seen in the autocorrelation function. As will be explained in section 4, this signal is transferred to a field-programmable gate array (FPGA) processing unit for the generation of random symbols.

#### 4. EXPERIMENTAL

The experimental setup is schematically shown in **Figure 5A**. The (m)LLED is directly facing a SiPM at a distance of  $\sim 1$  mm without any interposed optics or diffuser. The Si-NCs LLED is driven by an Agilent B1500A Semiconductor Device Parameter Analyzer. The TTL output of the SiPM is directly connected to the high speed digital input of an FPGA. A voltage of  $\sim$ 30–36 V is applied to the SiPM by an Agilent E3631A DC Power Supply. The measurement of the arrival times is performed by a fully synchronous logic. The FPGA continuously samples the detector at the frequency of 100 MHz, which is crystal controlled. A valid detection is produced by a high analog logic level heralded by one clock cycle (10 ns) of low analog logic level. A Digilent ATLYS FPGA board has been used with the programming language VHDL.

In order to verify that the Si-NCs emit independent photons, cross correlation measurements can be performed. The measurement is based on the random transmission of the emitted photons from the source (i.e., a Si-NCs LLED) into two arms of a fiber beam splitter (see **Figure 6A**) each connected to a detector (i.e., a SPAD). The longer arms of the fiber splitter are long enough (15 m) to prevent any possible peaks in the correlogram due to the correlation of either backflashs or back reflections and the real signals on the silicon SPADs within 60 ns time lags. Therefore, any visible peaks in the correlogram







**FIGURE 5** | (A) Schematic of the experimental setup with a Si-NCs LLED coupled with a SiPM at a distance of  $\sim$  1 mm. The SiPM board is connected to  $\pm$ 5 V and a bias voltage (Vbias) is applied to the SiPM. The Si-NCs LLED is driven by an applied current (voltage) provided by the power supply. The output signal of the SiPM is transmitted to an FPGA connected to a PC for the generation of random numbers. (B) The front-end board to amplify and digitalize the analog output signal from the detector. This is based on an AD8000 amplifier in a trans-impedance configuration, followed by a comparator with an adjustable voltage threshold (adj.Vth) and a monostable, creating pulses of 3.3 V and 100 ns width.



would be the result of photon bunching. The output signal of the two detectors are then sent to the two channels of a lineartau cross-correlator having the time resolution of 1.3 ns, where the cross-correlation function,  $g^2(\tau)$ , is computed. A peak in the cross-correlogram indicates photon bunching while a dip shows anti-bunching. Photon bunching occurs in the case of chaotic or thermal light which has a super-Poissonian distribution with the mean greater than the variance and photon number fluctuations larger than in a coherent light beam. Photon antibunching refers to a sub-Poissonian distribution with the mean lower than the variance and photon number fluctuations smaller than in a coherent light beam [11]. A flat cross-correlogram demonstrates that the photons are emitted independently with a Poisson distribution [12]. Measurement results are presented in Figure 6B. As can be seen, a flat cross-correlation graph (with no peak or dip) is observed, which demonstrates that the detected photons follow a Poisson distribution. This is another proof, in addition to the  $\chi^2$  statistic [6], that the Poisson distribution is a good match for the distribution of the detected photons which are emitted from the Si-NCs LLED.

In order to characterize afterpulsing and crosstalk in the SiPM, autocorrelation,  $g^2(\tau)$ , measurements of its signal were performed via a multitau digital correlator with 4 ns resolution [13].  $g^2(\tau)$  exhibits a main peak within ~140 ns from the main autocorrelation peak at  $\tau = 0$  (**Figure** 7). The plateau in  $g^2(\tau)$  approaches the normalization value of 1 at about 950 ns.

The measurements for random number generation were performed on the (m)LLED with an active area of  $\sim 0.99 \text{ mm} \times 0.82 \text{ mm}$  (see **Figure 1A**) which matches the SiPM dimension of 1 mm×1 mm. The applied forward current to LLED was kept below  $\sim 45 \ \mu A$  corresponding to the forward voltage of 3 V (see **Figure 2B**) in order to avoid degradation of



the oxide layer in the active area of the Si-NCs LLED [14]. The V<sub>bias</sub> to SiPM was 32 V corresponding to an excess bias of ~6 V ( $V_{bd} = 26 V$ ) with the DCR of ~80 kcps/mm<sup>2</sup>.

The methodology to generate random numbers is similar to our recent work [9]. It is based on the property of the Poisson statistics that, if an event is observed in a time interval T, the probability that the observation is performed in any subinterval of fixed length of T is uniform. Based on this property and to account for the system limitations, we can describe the methodology by defining "double length" periodic time intervals with an associated fully deterministic "target function". In the case of 16 random number generating subintervals, the alphabet of the symbols is N, 0, 1, ... F, that reads N (nonumber), and the hexadecimal numbers 0 to F. Each interval has 32 N subintervals in the first half and an alternation of N subintervals and the full set of numeric symbols in the second half, with a total of 64 subintervals (inset of Figure 7). Only if one single detection (event) hits the target function associated with an interval, a random symbol is generated. A mitigation technique is developed in order to improve the non-uniformity in the probability distribution of the generated random symbols. It is called "super interval structure" and is made before the possible arrival of a photon in an interval. As can be seen in the inset of Figure 7, it is composed of 16 "double length" intervals in which the random number generating symbols are ordered as 0, 1,... F in the first interval, F, 0, ... E in the second one and so on [9]. This approach results in the nearly uniform probability distribution of the generated symbols (**Figure 9**).

The duration of the "double length" interval is determined by the afterpulsing and crosstalk distribution of the SiPM (**Figure** 7). In order to mask the afterpulsing and crosstalk distribution of SiPM, we needed to set the first half of "double length" interval to ~950 ns (see **Figure** 7). We considered target functions with "super interval" of lengths 640, 1280, and 1920 ns and studied the autocorrelation coefficient of the generated hexadecimal symbols at time lag 1. It is observed to decrease as the length of "super interval" increases with the values of  $1.29 \times 10^{-4}$ ,  $1.05 \times 10^{-4}$ and  $1.45 \times 10^{-5}$  corresponding to the "double length" interval of 640, 1280, and 1920 ns, respectively. Therefore, we fixed the "double length" interval to 1920 ns and acquired sequences of random symbols.

## 5. EVALUATION OF RANDOMNESS

A very straightforward way to detect an observable pattern among the random symbols or codes is to create a 2-dimensional map of them. A  $512 \times 512$  map of the 16 hexadecimal symbols generated by our methodology from a recording of our system is presented in **Figure 8**. As can be seen clearly, no particular, periodic pattern is observed among the symbols.

Figure 9 shows the probability of generated hexadecimal symbols in a sequence of 1 G symbols. It is seen to follow an uniform distribution (the theoretical value for the probability







distribution of 16 bin symbols (1/16) is indicated by a solid red line in **Figure 9**).

The high quality of random symbols is proved through the evaluation of the joint probability mass function (JPMF) [15], which shows a very low deviation of  $\sim 5 \times 10^{-6}$  from the expected theoretical value of  $(1/16) \times (1/16) = 0.00390625$  (**Figure 10**), and the evaluation of the mutual information (MI) [16], which is calculated to be  $\sim 1.5 \times 10^{-7}$  bits considering 1 G random symbols.

To further analyze the quality of generated random numbers, each symbol is replaced with its corresponding 4-bit binary values. In this way, we obtain a binary sequence of random bits. We then apply the 15 statistical tests in the NIST tests suite to the





<b>TABLE 2</b>   NIST tests results for 2 G random bits $(2 \times 10^9 \text{ bits})$ . The significance
level is $\alpha = 0.01$ . In order to pass, the p-value should be larger than 0.01 and the
proportion should be more than 0.983.

Statistical test	P-value	Proportion	Result
Frequency	0.2861	0.9930	Passed
Block frequency	0.2868	0.9935	Passed
Cumulative sum	0.1657	0.9920	Passed
Runs	0.3298	0.9935	Passed
Longest run	0.4817	0.9910	Passed
Rank	0.3611	0.9860	Passed
FFT	0.0401	0.9910	Passed
Non-overlapping template	0.5666	0.9905	Passed
Overlapping template	0.4064	0.9900	Passed
Universal	0.1404	0.9850	Passed
Approximate entropy	0.2854	0.9930	Passed
Random excursions	0.5310	0.9938	Passed
Random excursions variant	0.3127	0.9883	Passed
Serial	0.3376	0.9870	Passed
Linear complexity	0.2550	0.9905	Passed

generated raw data. Various datasets with 1 and 2 Gbits length at different applied currents to the (m)LLED were obtained. They all passed the NIST tests without the application of a post-processing algorithm irrespective of the EL variations of the (m)LLED during data acquisition. The results for a dataset of 2 Gbits are tabulated in **Table 2**. The highest experimental bit-rate is calculated to be 0.5 Mbps at the EL intensity of  $\sim$  550 Kcps.

## 6. CONCLUSIONS

We realized a compact quantum random number generator with a novel configuration comprising a Si-NCs LLED directly interfaced with a SiPM without any coupling optics. This paves the way to the further integration of the photon source and the single photon detectors in a single integrated circuits. Indeed, both the devices were fabricated by using the FBK technology. Our research is currently focusing to define a single fabrication process allowing the fabrication of both devices in a single silicon chip.

Remarkably the Si-NCs LLED have similar emission properties and statistics as the standard small Si-NCs which we have previously developed for PQRNG application [6, 9]. Therefore we could use the same robust methodology to extract high quality random numbers which is implemented on a FPGA. The methodology considers the non-idealities of the detector and the source of photons, including parameters (like EL) drifts. The generated high quality random numbers pass all the statistical tests in NIST tests suite without any post-processing. The highest bit rate is 0.5 Mbps with the efficiency of 4-bits per detected photon.

This compact QRNG, with the capability of producing high quality random numbers, can be implemented in small electronic devices providing utmost security accessible to everyone. The proposed device configuration has several advantage with respect to what we already reported in Bisadi et al. [9], particularly with respect to the simplicity of the system (no optics, no thermoelectrical cooler, simple power supply). Still the proposed

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device can be further optimized by improving the spectral overlap between the LLED and the SiPM and by increasing the parallelization of the Si-NCs LLED/SiPM which can further improve the bit rate to higher values if required by some applications.

## AUTHOR CONTRIBUTIONS

ZB prepared the setup, conducted the experiments and analyzed the acquired data. She also made some suggestions to adapt the robust methodology to generate high quality random numbers with the structure. FA partially designed the silicon photomultiplier (SiPM) and characterized it. GF created the robust methodology to generate high quality random numbers and implemented the target function on the FPGA. NZ designed most part of the SiPM. CP supervised the development of the technology used in the SiPM. GP designed and fabricated the large area silicon nanocrystals LED (Si-NCs LLED). LP supervised the whole research work.

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