

50GHz Ge Waveguide Electro-Absorption Modulator Integrated in a 220nm SOI Photonics Platform

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Abstract: We report waveguide-integrated Ge electro-absorption modulators operating at 1615nm wavelength with 3dB bandwidth beyond 50GHz and a capacitance of 10fF. A 2V voltage swing enables 4.6dB DC extinction ratio for 4.1dB insertion loss.

OCIS codes: (250.7360) Waveguide modulators; (230.2090) Electro-optical devices; (200.4650) Optical interconnects.

1. Introduction

Silicon photonics (SiPh) is considered to be a key enabling technology for future CMOS systems since it has the potential to alleviate the bandwidth-power-density bottleneck of electrical interconnects [1]. However, aggressive energy targets for interconnects have to be met before the technology can become mainstream replacement for their electrical counterparts [1]. In addition to stringent energy and footprint limitations, the optical network should have high bandwidth to support the ever increasing data needs of the end user. In this paper, we report compact Ge electro-absorption modulators (EAM) based on Franz-Keldysh effect [2] with a modulation speed of over 50GHz and a capacitance of 10fF. The Ge EAM devices are integrated in SiPh platform on 200nm silicon-on-insulator (SOI) wafers with 220nm top Si thickness, along with various passive and active Si devices. Compared to earlier results on GeSi electro-absorption modulators using a 3 μ m SiPh platform [3], the devices in this work have improved performance including higher modulation speed, higher modulation efficiency and lower capacitance due to strong confinement of optical and electrical field enabled by submicron Ge/Si waveguide platform.

2. Design and Fabrication of the Modulator

The working principle behind bulk Ge electro-absorption modulators is the Franz-Keldysh (FK) effect according to which an applied electric field causes band-tilting thereby increasing the absorption coefficient in the weakly absorbing regime [2]. A significant change in Ge absorption coefficient can be achieved if the electric field difference between OFF and ON state (ΔE) is greater than 10kV/cm as shown in Fig.1(a). A reverse biased p-i-n diode is an obvious choice for realizing such a strong electric field contrast. The dimensions, doping profile and contacting scheme play a crucial role in the overall performance of the modulator. Fig. 1(b) shows a cross-sectional schematic of the Ge EAM devices reported in this paper. A lateral p-i-n diode with moderate doping concentrations ($\sim 5 \times 10^{17} \text{cm}^{-3}$) is implemented in a Ge waveguide. Both the anode and cathode are contacted from highly doped Si layers ($\sim 4 \times 10^{20} \text{cm}^{-3}$), avoiding the need for direct metal contacts on the Ge waveguide, which would otherwise cause excessive optical absorption loss. Extensive simulations were carried out in SentaurusTM to optimize implant conditions for obtaining the strongest optical-electrical interaction. Fig. 1(c) and (d) depict the optical mode and electrical field contrast ΔE respectively in the optimized device, showing a strong overlap between the two.

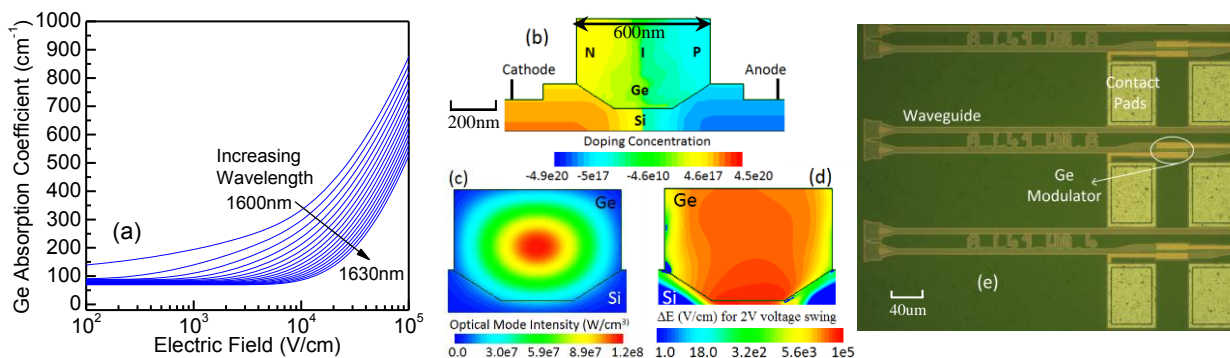


Fig. 1 (a) Modeled absorption coefficient vs. applied electric field. (b) Schematic of the Ge EAM p-i-n diode. (c) Approximated optical field distribution showing good confinement in Ge. (d) Change in electric field (ΔE) between ON and OFF state. (e) Microscope image of the fabricated modulators integrated with Si waveguides and grating couplers.

A set of modulators with varying Ge length (from 10 μm to 80 μm) and width (from 400nm to 1 μm) were fabricated in imec's 200mm Si photonics platform [4]. The Ge thickness was 400nm and the top Si thickness of the SOI wafer was 220nm. A poly-Si taper structure was implemented to ensure low-loss coupling from the Si waveguide into the Ge waveguide and vice versa. TE mode grating couplers with peak-coupling wavelength around 1600nm were used for fiber-chip coupling. Fig. 1(e) shows a top-down microscope image of the Ge modulator EAM array, integrated with Si waveguides.

3. Modulator Performance

(a) *Static Characteristics:* Fig. 2(a) shows I-V characteristics of fabricated EAM devices with varying length. The low dark current of the Ge/Si diodes ($\sim 10\text{nA}$ at -1V for a 40 μm long device) signify high quality Ge and controlled electrostatics thereby ensuring very low static energy consumption in the modulator. Fig. 2(b) shows the measured and simulated absorption spectrum with different bias voltages applied to the Ge modulator. The measured change in absorption with bias is clearly evident and is reasonably well matched with theoretical predictions from FK theory and simulation. The absorption spectrum for different bias voltages intersect at $\sim 1580\text{nm}$ which corresponds to the HH- Γ band gap for epitaxial Ge on Si. Fig. 2(c) shows the static extinction ratio and insertion loss for a 40 μm long and 600nm wide device. For a 2V swing at 1615nm, the insertion loss (IL) is 4.14dB and the extinction ratio (ER) is 4.58dB. It should be noted that all the optical data is normalized to a neighboring straight waveguide, but it includes the coupling loss between the Si waveguide and the Ge modulator.

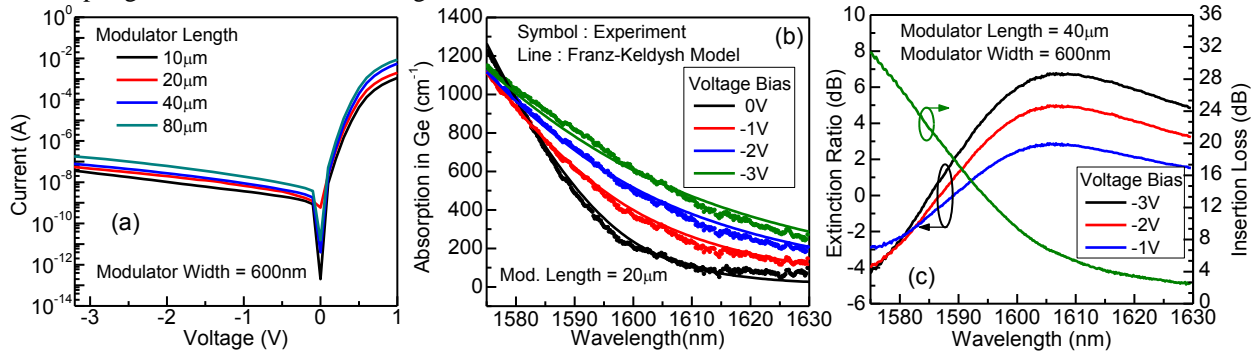


Fig. 2 (a) Dark I-V of modulators with varying length. (b) measured change in absorption coefficient of Ge is well matched by simulations. (c) extinction ratio and insertion loss of a representative device.

An additional benchmark, comparing the ratio between ER and IL, is shown in figure 3(a). It is clear that the modulator of 40 μm length and 600nm width has the best performance. Longer devices suffer of increased IL while shorter devices don't exhibit sufficient ER. Similarly, wider devices suffer from lower ΔE while narrower devices have excessive coupling loss. Another important figure of merit is the Link Power Penalty (LPP) defined as $(P_{\text{out}}(1) - P_{\text{out}}(0)) / (2xP_{\text{in}})$, where $P_{\text{out}}(1)$ and $P_{\text{out}}(0)$ are the high and low level of output optical power and P_{in} is the input optical power [3]. Fig. 3(b) shows that the minimum LPP for the best device is 8.2dB for 2V swing. The usable optical bandwidth is defined as the wavelength range between which the LPP remains within 1dB of its minimum value. Due to the limitations in available laser wavelengths, we could measure only half of this bandwidth which came out to be 17.5nm (1612.5 – 1630nm) for 2V swing. Based on simulations, we can extrapolate the full 1dB optical bandwidth to be greater than 35nm. Fig. 3(c) shows the variation of minimum LPP with modulator length and voltage swing. The trends for LPP are very similar to that of ER/IL seen before.

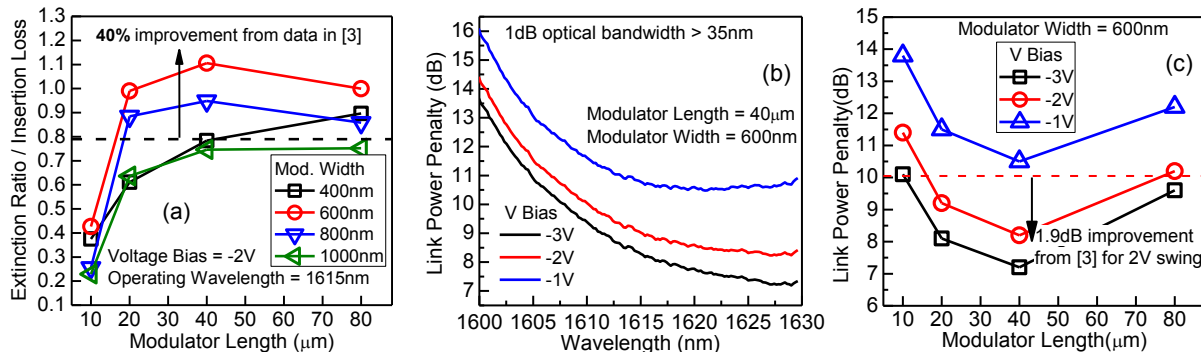


Fig. 3 (a) Figure of merit (ER/IL) versus modulator length and width. (b) Link Power Penalty (LPP) of the best device. (c) Variation of minimum LPP with modulator length and voltage swing.

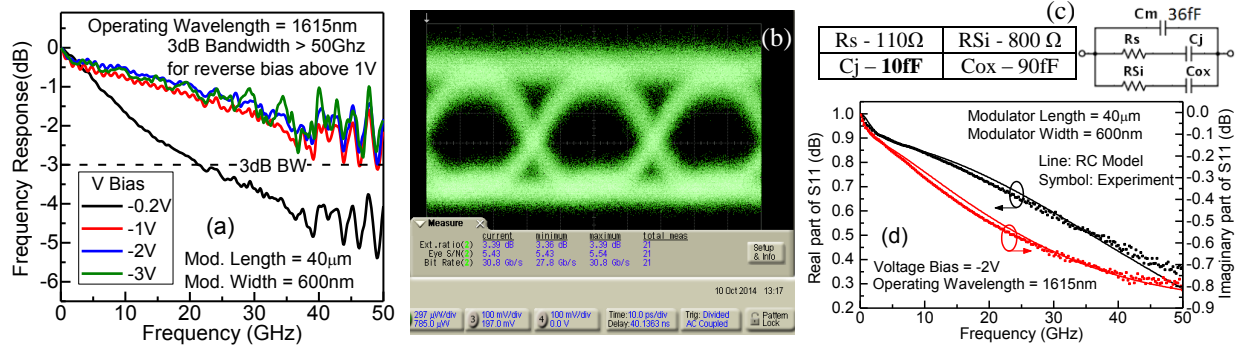


Fig. 4 (a) RF S21 measurements showing >50 GHz bandwidth. (b) Eye diagram showing clear opening at 28Gbps with a dynamic ER of 3.39dB for a nominal 2Vpp swing. (c) Equivalent Circuit and (d) S11 parameter fit showing $C_j=10\text{fF}$ and $R_s = 110\Omega$.

(b) *High-Speed Performance:* The 3dB bandwidth (BW) of the modulator is greater than 50GHz for reverse bias above 1V, as extracted from electro-optic S21 measurements shown in Fig. 4(a). The FK effect is known to be a sub-picosecond effect [5] and therefore the speed of the EAM modulators is expected to be RC limited. Fig. 4(b) shows the measured eye diagram at 1610nm for a 28Gbps data stream (PRBS31) generated by a pattern generator and delivered to the device by a 50Ω terminated RF probe. An open eye diagram with dynamic ER of 3.39 dB at 2V swing can be observed. Since the BW of the modulator is more than 50GHz, one can expect it to operate at data rates as high as 56Gbps. The RC model of the EAM can be extracted by fitting the S11 parameters of the equivalent circuit [6] shown in Fig. 4(c). A reasonable fit yields a junction capacitance of 10fF and series resistance of 110Ω for 2V reverse bias, as shown in Fig. 4(d). This translates to an estimated dynamic energy dissipation of just 10fJ/bit for a 2V swing. The relatively small height and length of the Ge modulator leads to a low junction capacitance and hence better high speed performance with minimal energy consumption. For benchmarking this work, Table 1 summarizes the key figures of merit with other representative modulators, illustrating the relative advantages of the presented Ge EAM versus other Si-based modulator types and earlier Ge(Si) EAM demonstrations.

4. Conclusion

We have demonstrated a CMOS compatible, waveguide integrated Ge EAM with a bandwidth greater than 50GHz, capacitance of 10fF and link power penalty of 8.2dB for a 2Vpp drive swing. The demonstrated device shows good potential for enabling power-efficient and compact 28G, 40G and 56G electro-optical transmitters.

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Table 1: Performance metrics for different representative modulators reported experimentally

Modulator Type	3dB BW (GHz)	Modulator Capacitance (fF)	V Swing (V)	Active Footprint (μm^2)	ER (dB)	IL (dB)	Optical BW (nm)
This work	50	10	2	24	4.6	4.1	>35
GeSi FK [3]	34	69.6	2	40	3.8	4.8	40
GeSi FK [7]	1.2	11	3	30	8	3.7	14
GeSi QCSE* [8]	3.5	3	1	8	3.2	15	20
GeSi QCSE* [9]	6.3	200	3	400	4	3	20
Si MZI** [10]	30	~ 800	1.5	~ 1500	3.4	7.1	>80
Si Ring [11]	21	~ 17	0.5	20	6.4	1.2	<0.1

*QCSE: Quantum Confined Stark Effect; **MZI: Mach-Zehnder interferometer

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