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Control of an underactuated double-pendulum overhead crane using improved model reference command shaping: Design, simulation and experiment



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ABSTRACT

This paper presents a new control scheme based on model reference command shaping (MRCS) for an overhead crane, with double-pendulum mechanism effects. The approach has an advantage in achieving an accurate trolley positioning, with low hook and payload oscillations, under various desired trolley positions and parameter uncertainties, without the requirement for measurement or estimation of system parameters. These are challenging in practice. The previously developed MRCS algorithm is improved in order to reduce its design complexity, as well as to ensure that it can be augmented with a feedback controller so that a concurrent controller tuning can be realised. The combined MRCS and feedback controller is used to achieve both, precise trolley positioning, and low hook and payload oscillations. To evaluate the effectiveness and the robustness of the approach, simulations and experiments using a nonlinear model and a laboratory double-pendulum crane are carried out. Under various desired positions and parameter uncertainties that involve varying the cable lengths (payload hoisting) and the payload mass variations, the superiority of the proposed approach is confirmed by achieving higher hook and payload oscillation reductions when compared with a recently proposed feedback controller. In addition, the desired trolley positions are achieved with smoother responses.

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1. Introduction

As worldwide industrial machine applications, cranes are exclusively used for transferring a variety of massive loads to various locations. In fact, most industrial cranes are considered to be underactuated mechanical systems, which would indicate a lower number of actuators than the degrees of freedom [1]. The control of such an underactuated system is complicated, especially when dealing with double-pendulum mechanism effects [2,3]. For a double-pendulum overhead crane (DPOC), the hook and the payload oscillate with different oscillation modes (known as multimode) during transportation, whereby the simultaneous elimination of both oscillations is difficult [4]. This phenomenon becomes extremely challenging under parameter uncertainties, with varying cable lengths (payload hoisting) and payload mass variations that affect the

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oscillation frequencies, which may lead to positioning errors and significant payload oscillations. These result in a decrease in industrial productivity and they introduce a safety issue [5–7].

Several feedback control approaches have been proposed for various types of crane to enable the cranes to regulate its performance, by ensuring that the actual output is closer to the desired response. These include adaptive control [8], nonlinear control [9–11], sliding mode control [12], time optimal control [13], generalised trajectory modification strategy [14], and fuzzy logic control [15]. For control of the DPOC, feedback control strategies that have been proposed were adaptive [7], linear [16], nonlinear [17,18], and intelligent [19] control approaches. It is known that full state feedback controllers require several sensors for measuring the trolley position, as well as the hook and payload oscillation angles for control action [20]. Nevertheless, the necessary additional sensors require an additional cost, and the payload angle measurements will be difficult, with different payload masses, sizes, and shapes, especially for the double-pendulum mechanism effects [21]. In opposition to the feedback control, a feedforward control has been widely utilised for the oscillation control of a DPOC [2,22,23]. A two-mode input shaping technique has been mostly used when effectively applied to the crane [24–26]. Moreover, feedforward shapers, namely, a command shaper [27,28] and a smoother [29–31], with different design approaches have been investigated. Improved input shapers have also been proposed for an overhead crane [32] and a tower crane [33].

Recently, model reference command shaping (MRCS) was proposed by [34] and was designed by using a critically damped reference model for the oscillation control of a DPOC. As an alternative feedforward control, this approach did not require any oscillation feedback sensor. This certainly saves on the cost, especially for multimode systems [35]. Besides, an MRCS has another great advantage when compared to other feedforward control approaches, as prior knowledge of the system's natural frequencies are not required for the design. Therefore, the difficulties for measurements or estimations of the system parameters can be avoided. Simulations and experiments on a DPOC have shown significant hook and payload oscillation reductions under various crane operating conditions [34]. However, this control approach has a major drawback, as it can only be used for oscillation control, and it does not have the ability to precisely drive the trolley to various desired positions. In addition, the design of an MRCS is challenging, for it involves complicated procedures and mathematical formulations.

In order to utilise the advantage that is offered by an MRCS and to achieve an accurate crane positioning, a method to combine it with a feedback controller can be established. However, for a real-time realisation of the combined control structure, a simpler MRCS algorithm, with less mathematical formulations, is required. In the literature, several efforts have been presented for developing a hybrid (combination of feedforward and feedback) control strategy for a DPOC. These include input shaping with a nominal characteristic trajectory [36], a single input fuzzy logic controller [37], and a model reference control [38]. In [39], the input shaping has been augmented with two feedback controllers (i.e. proportional–derivative (PD) and proportional (P) controllers) for the control of the trolley position and the deflection of the hook. It has been found that most of the hybrid controllers were designed independently. For the control strategies that involved input/command shaping, they required prior knowledge of either the first frequency mode (i.e. the hook), or the second frequency mode (i.e. the payload), or both of the frequency modes (i.e. the hook and the payload) for the design of the control.

This paper presents an MRCS-based control of an underactuated overhead crane, with double-pendulum mechanism effects, under various desired trolley positions and parameter uncertainties. This involved varying the cable lengths (payload hoisting) and the payload mass variations. In order to retain the great advantage offered by an MRCS and to achieve an accurate crane positioning, the previously developed MRCS scheme was improved, in order to reduce the design complexity and to ensure that it could be augmented with a practical feedback controller. To demonstrate the capability of the proposed control structure, a PID controller was utilised as a practical feedback controller, and the concurrent tuning of the control parameters was realised by using particle swarm optimisation (PSO). The proposed controller is called MRCS-PID, and the main difference as compared to the previous MRCS approach is in term of the control structure. In this case, the MRCS is a feedforward control approach, whereas the proposed MRCS-PID is a combined of feedforward and feedback controllers to achieve a precise trolley positioning and satisfactory oscillation control.

The main contributions of this work when in comparison with the existing body of literature are:

- a) To the best of authors' knowledge, this is the first work on designing a hybrid controller (a combined command shaper and feedback controller) for a double-pendulum crane that avoids the requirement for prior knowledge of the system oscillation frequencies.
- b) The hybrid controller was designed, such that a concurrent tuning of the two controllers could be performed, in order to ensure optimal performance in the trolley positioning and the payload oscillation. In addition, the design of the previously developed MRCS algorithm was improved.

Several scenarios of the DPOC that involved the desired trolley positions and the parameter uncertainties were considered. Simulations and experiments using an underactuated nonlinear model and a laboratory DPOC were conducted, so as to investigate the effectiveness of the proposed strategy. The accuracy of the trolley positions and the level of reductions of the hook and payload oscillations were analysed, and these were used to investigate the performance of the controller. Performance comparisons with a recently proposed control algorithm for a DPOC system designed using two PID controllers were also conducted, in order to further verify the robustness of the MRCS-PID control strategy.

2. Model of a DPOC system

Fig. 1 illustrates the schematic diagrams of a DPOC, without and with the varying cable lengths during the payload hoisting. Distance x indicates the trolley path as it moves from the origin, while θ_1 and θ_2 represent the hook and the payload oscillation angles, respectively. In addition, m, m_1 , m_2 , l_1 , l_2 , g, f_x , and u denote the trolley mass, the hook mass, the payload mass, the hook cable length, the payload cable length, the gravitational constant, the friction coefficient of x, and the control force applied on the trolley, respectively. As the work considered for constant and varying cable lengths, dynamic equations of the DPOC under both conditions are presented.

When operating without a varying cable length of l_1 , as illustrated in Fig. 1(a), the dynamic equation of the DPOC can be obtained as [2,23]:

$$(m + m_1 + m_2)\ddot{x} + (m_1 + m_2)[l_1\ddot{\theta}_1\cos\theta_1 - l_1\dot{\theta}_1^2\sin\theta_1] + m_2l_2[\ddot{\theta}_2\cos\theta_2 - \dot{\theta}_2^2\sin\theta_2] = u - f_x\dot{x}$$
(1)

$$(m_1 + m_2)[l_1\ddot{x}\cos\theta_1 + l_1^{2}\ddot{\theta}_1 + gl_1\sin\theta_1] + m_2l_1l_2[\ddot{\theta}_2\cos(\theta_1 - \theta_2) + \dot{\theta}_2^{2}\sin(\theta_1 - \theta_2)] = 0$$
(2)

$$m_2 l_2 [\ddot{x} \cos \theta_2 + l_1 \ddot{\theta}_1 \cos (\theta_1 - \theta_2) + l_2 \ddot{\theta}_2 - l_1 \dot{\theta}_1^2 \sin (\theta_1 - \theta_2) + g \sin \theta_2] = 0$$
(3)

Additionally, the payload was also required to be hoisted (up or down) in practical operations to the desired location. With a varying cable length of l_1 , as illustrated in Fig. 1(b), the dynamic equation of the DPOC can be written as [7,40]:

$$(m + m_1 + m_2)\ddot{x} + (m_1 + m_2)[2\dot{l}_1\dot{\theta}_1\cos\theta_1 + l_1\ddot{\theta}_1\cos\theta_1 - l_1\dot{\theta}_1^2\sin\theta_1 + \ddot{l}_1\sin\theta_1] + m_2l_2[\ddot{\theta}_2\cos\theta_2 - \dot{\theta}_2^2\sin\theta_2]$$

$$= u - f_x\dot{x}$$
(4)

$$(m_1 + m_2)[\ddot{x}l_1\cos\theta_1 + l_1^2\ddot{\theta}_1 + gl_1\sin\theta_1 + 2l_1\dot{l}_1\dot{\theta}_1] + m_2l_1l_2[\ddot{\theta}_2\cos(\theta_1 - \theta_2) + \dot{\theta}_2^2\sin(\theta_1 - \theta_2)] = 0$$
(5)

$$m_2 l_2 [\ddot{x} \cos \theta_2 + l_2 \ddot{\theta}_2 + \ddot{l}_1 \sin(\theta_1 - \theta_2) - l_1 \dot{\theta}_1^2 \sin(\theta_1 - \theta_2) + l_1 \ddot{\theta}_1 \cos(\theta_1 - \theta_2) + 2\dot{l}_1 \dot{\theta}_1 \cos(\theta_1 - \theta_2) + g \sin \theta_2] = 0$$
 (6)

$$(m_1 + m_2)[\ddot{x}\sin\theta_1 + \ddot{l}_1 - l_1\dot{\theta}_1^2 + g(1 - \cos\theta_1)] + m_2l_2[\ddot{\theta}_2\sin(\theta_1 - \theta_2) - \dot{\theta}_2^2\cos(\theta_1 - \theta_2)] = u_l - f_l\dot{l}_1$$
(7)

where u_l and f_l denote the hoisting force and friction coefficient of l_1 .

3. Controller designs

This section presents the improved MRCS algorithm, which can be utilised and augmented with a feedback controller, in order to achieve an accurate trolley positioning, together with higher oscillation reductions of the hook and the payload for an underactuated DPOC. A practical PID feedback controller was used to demonstrate the proposed MRCS-PID control strategy, and a PSO with a new fitness function was introduced for the concurrent tuning of the controller parameters. Using the

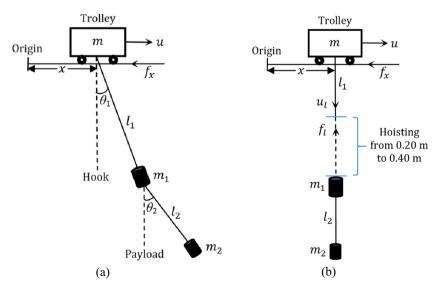


Fig. 1. A DPOC: (a) Without a varying cable length (b) With a varying cable length.

same PSO algorithm, a recently proposed feedback controller using two PID controllers (PID-PID) was designed and implemented for performance comparisons.

3.1. MRCS-PID control strategy

Without a requirement for the crane's oscillation frequencies, the MRCS was previously designed as an alternative command shaper by determining an appropriate shaper, $G_s(s)$, a critically damped reference model, $G_r(s)$, and a DPOC model, $G_c(s)$, that effectively minimised the hook and the payload oscillations, as illustrated in Fig. 2 [34]. The concept of poleszeros cancellation was used in obtaining $G_s(s)$. More specifically, the shaper can be arranged as $G_s(s) = G_r(s)G_c(s)^{-1}$ and formulated as:

$$G_s(s) = \frac{a_n s^n + a_{n-1} s^{n-1} + a_{n-2} s^{n-2} + \dots + a_0}{(s + \omega_r)^n}$$
(8)

where n denotes the system order, ω_r was chosen based upon several tests, and a_n , a_{n-1} , a_{n-2} , ..., a_0 were the design variables of $G_s(s)$. However, the process to obtain the complete design variables of an MRCS for a DPOC involved complicated procedures and mathematical formulations, as this was established in previous work [34].

In an attempt to reduce the design complexity and to ensure that it could be implemented together with a feedback controller, the previously developed MRCS was improved, such that the PSO could be utilised to solve the design variables of the command shaper, $G_s(s)$, as illustrated in Fig. 3. By adopting the same concept as the MRCS approach, appropriate values for the numerator, $G_s(s)$, as described in Eq. (8), were obtained, so that the poles of the DPOC could be cancelled out, and the overall system poles were only from the denominator of the shaper. Hence, the output value of x_r can be expressed as:

$$x_{r}(t) = \sum_{i=1}^{3} a_{2i} x_{2i}(t) \tag{9}$$

where $a_{2i}(i=1,2,3)$ are the numerator coefficients, and $x_{2i}(i=1,2,3)$ are the step responses of $G_s(s)G_c(s)$.

As the MRCS can only handle hook and payload oscillations, a feedback controller was required, in order to achieve an accurate trolley positioning for the various desired positions. In this work, the PID was designated as a feedback controller, due to its low cost, its easy implementation, and because it is widely used in a number of practical systems [41]. Therefore, a hybrid design of an improved MRCS and PID was required to achieve the optimal performance for both of the objectives. As the design variables of $G_s(s)$, and the three PID parameters that needed to be obtained concurrently, the benefits of a PSO algorithm were used to simultaneously tune the MRCS-PID controller parameters, as depicted in Fig. 4, which provided an additional advantage for the proposed hybrid scheme.

With a step input of x_d , and based on the outputs x, θ_1 , and θ_2 , the PSO algorithm was formulated to concurrently calculate the six control parameters, which were a_6 , a_4 , and a_2 for the MRCS, and K_P , K_I , and K_D for the PID controller. When using the MRCS, x_d was shaped as x_r , which was the optimal input reference for the feedback system that theoretically yielded the minimal hook and payload oscillations. For the feedback system, the trolley position of x was the only output used as a feedback signal. In the PSO, two initial parameters of the particle, namely, position, X_{ij}^k and velocity, V_{ij}^k were introduced towards the searching process, where k, i, and j were the iteration numbers, the individual swarm, and the dimension of the particle, respectively. The particle i in dimension j for the searching optimisation process can also be defined as $i \in [1, \rho]$ and $j \in [1, D]$, where ρ and D were the entire population and they were high dimensional in a search space. Conceptually, the new particle velocity, V_{ij}^{k+1} , was adjusted, according to the X_{ij}^k , $pbest_{ij}$, and $gbest_j$ values, and the new particle position, X_{ij}^{k+1} , was updated based on X_{ij}^k and V_{ij}^{k+1} , as in [41-43]:

$$V_{ij}^{k+1} = wV_{ij}^{k} + c_{1}r_{1}(pbest_{ij} - X_{ij}^{k}) + c_{2}r_{2}(gbest_{j} - X_{ij}^{k})$$
(10)

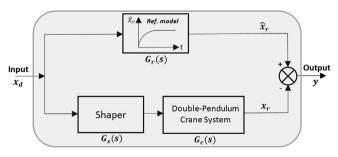


Fig. 2. MRCS block diagram [34].

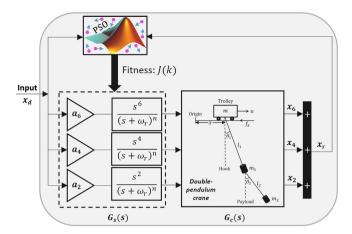


Fig. 3. PSO-based MRCS block diagram.

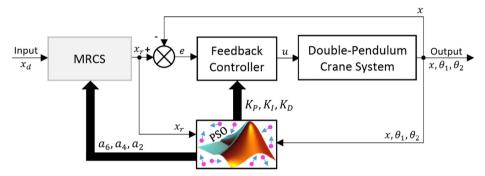


Fig. 4. A hybrid MRCS-PID control block diagram using a PSO algorithm.

$$X_{ij}^{k+1} = X_{ij}^k + V_{ij}^{k+1} \tag{11}$$

where w was chosen for a better exploration and exploitation of the particle that began with 0.9, and linearly decreased to 0.4 at the maximum iteration, N. Furthermore, c_1 and c_2 were cognitive and social acceleration coefficients, respectively, while r_1 and r_2 represented the random function values, $r_1, r_2 \in U(0, 1)$. For the DPOC control, X_{ij}^k can also be defined as $\left\{a_{6_{11}}^k, a_{4_{12}}^k, a_{2_{13}}^k, K_{P_{14}}^k, K_{I_{15}}^k, \text{and } K_{D_{16}}^k\right\}$.

To ensure for a precise trolley positioning and higher oscillation reductions for the hook and payload, a fitness function, J(k) that considers the relationship between the trolley position, the hook and payload movements, and the physical parameters of the crane was designed [16]. This concept was originated from the potential energy, $P = mg\Delta h$ and based on vertical distance oscillations of the DPOC system, where a vertical distance of $\Delta h = h_1 + h_2$ as shown in Fig. 5. The h_1 and h_2 represent the differences in heights of the hook and payload with respect to a reference height without oscillation, respectively. Lower values of h_1 and h_2 indicate low hook and payload oscillations, and directly decrease the potential energy of the system.

The performance of each individual particle was assessed by the fitness function and $pbest_{ij}$ and $gbest_{j}$ were updated, if the particle had a minimum J(k):

$$J(k) = \sum_{k=1}^{N} \left| x_r^{(k)} - x^{(k)} + \Delta h \right| \tag{12}$$

$$\Delta h = (m_1 + m_2)gl_1\left(\cos\theta_{1r} - \cos\theta_1^{(k)}\right) + m_2gl_2\left(\cos\theta_{2r} - \cos\theta_2^{(k)}\right)$$
 (13)

where $\theta_{1r} = \theta_{2r}$ = the reference (zero) angles for the hook and the payload. Therefore, the proposed hybrid control parameters varied according to the particle position movements.

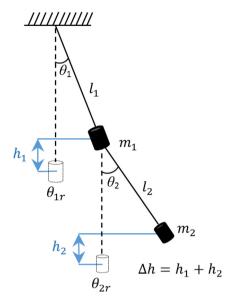


Fig. 5. Vertical distance of hook and payload movements.

3.2. PID-PID control strategy

In this section, two feedback controllers, namely, PID-PID control were combined and implemented for the DPOC control, and they were used for performance comparison. The technique was recently proposed in [16], and it had the advantage of avoiding the measurement of the payload motion, which is challenging in real practice. The simulation results revealed that the performance was comparable with a full state feedback control, thus, the real-time implementation would be desirable and interesting.

Fig. 6 shows the block diagram of the PID-PID control scheme. The separate PID feedback controllers were used for position and oscillation control, respectively, by using the feedback signals, x and θ_1 . Similar to the MRCS-PID control, the PSO algorithm was used to concurrently tune the six PID-PID control parameters, K_{P1} , K_{P1} , K_{P1} , K_{P2} , K_{I2} , and K_{D2} . A similar fitness function, by replacing x_r to x_d in Eq. (12), was used without the payload motion sensor. The condition of $\theta_2 = 2\theta_1$, as was designed in [16], was considered for solving the fitness functions in Eqs. (12) and (13). In this case, X_{ij}^k was defined as $\{K_{P1}^{i}, K_{I1}^{k}, K_{D1}^{k}, K_{D1}^{k}, K_{P2}^{i}, K_{D1}^{k}, K_{D2}^{k}, K_{D2}^{k}$, and K_{D2}^{i} , K_{D1}^{k} .

4. Implementation and results

In this work, three scenarios that involved different operating conditions of a DPOC were considered: (i) various trolley positions, (ii) payload hoisting, and (iii) payload mass variations. The simulations were performed by using the nonlinear underactuated model of a DPOC, as formulated in Eqs. (1)–(7), whereas the experiments were carried out by using a laboratory DPOC, as shown in Fig. 7, in order to investigate the performance of the proposed MRCS-PID and PID-PID control strategies. The laboratory DPOC was equipped with a trolley, two cables, and two cylindrical loads (dimension: diameter \times height) that represented the hook (0.06 m \times 0.076 m) and the payload (0.05 m \times 0.055 m). Three incremental decoders with a resolution of 4096 pulses per rotation were used for measuring the trolley position, cable length and hook oscillation. In addition, a Logitech C170 camera mounted on the trolley was also used to record the payload motion

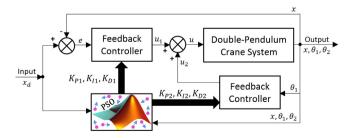


Fig. 6. A PID-PID control block diagram using a PSO algorithm.

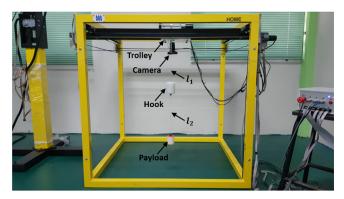


Fig. 7. A laboratory DPOC.

coordinates based on the deflection of a red indicator located at the top of the payload. The parameters of the laboratory DPOC were: m = 1.155 kg, $m_1 = 0.20$ kg, $m_2 = 0.10$ kg, $l_1 = 0.30$ m, $l_2 = 0.20$ m, $f_x = 82$ Ns/m, and g = 9.81 m/s². These parameters were also used in the simulations.

The correctness of the nonlinear underactuated DPOC model was previously verified in [34], where satisfactory agreements between the simulation and the experimental results were obtained. All of the MRCS-PID and PID-PID control parameters that were designed in Sections 3.1 and 3.2 were used in the simulations and in the real-time implementations, respectively. In order to evaluate the oscillation control performances, maximum amplitudes of the hook, θ_{1m} , and the payload, θ_{2m} , angles were measured. In addition, a mean squared error (MSE) was used as a performance index, where low MSE $_{\theta_1}$ and MSE $_{\theta_2}$ values were desirable, as they indicated low overall hook and payload oscillation responses, respectively.

4.1. Optimal parameters of the MRCS-PID and PID-PID control strategies

The MRCS-PID and PID-PID control parameters that were based on x_d = 0.40 m were optimally obtained when using the PSO algorithm with a fitness function, as designed in Sections 3.1 and 3.2, respectively, upon the completion of the iteration process. The best practice is to select ρ in between 20 and 50 particles, and with no specific limit of k numbers [44]. On the other hand, in [45,46], the values of c_1 and c_2 were suggested as 2 to ensure that the particles attract and influence on the average of $pbest_{ij}$, and $gbest_j$ values. Therefore, in this work, ρ , k, c_1 , and c_2 were set as 20, 100, 2, and 2, respectively. It can be seen that the control parameters varied and remained unchanged, once they reached the 51st and 54th iterations, as shown in Figs. 8 and 9. For the results, the final optimal MRCS-PID control parameters were obtained as a_6 = 0.1957, a_4 = 43.6252, a_2 = 1063.6992, a_2 = 1063.6992, a_3 = 1.0401, a_4 = 1.0401, a_5 = 1.1480, a_5 = 1.1480, a_5 = 0.1314.

It was important to ensure that all of the poles were in a stable region of the s-plane, indicating a closed-loop stable system for the implementations. From Fig. 10, the poles of the proposed MRCS-PID control strategy were located at

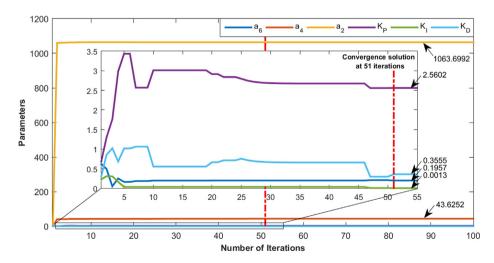


Fig. 8. The optimal MRCS-PID control parameters.

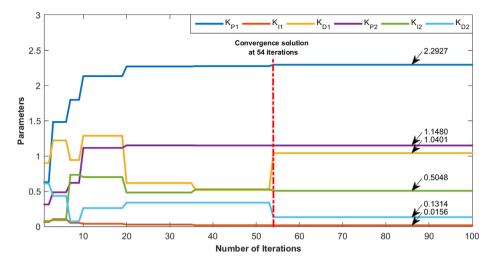


Fig. 9. The optimal PID-PID control parameters.

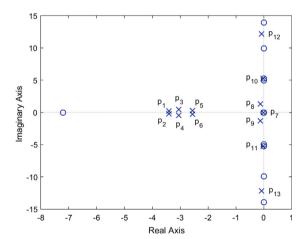


Fig. 10. Pole locations of the MRCS-PID control strategy.

 $p_{1,2} = -3.4 \pm j0.213$, $p_{3,4} = -3.04 \pm j0.482$, $p_{5,6} = -2.56 \pm j0.293$, $p_7 = -0.0000517$, $p_{8,9} = -0.118 \pm j1.31$, $p_{10,11} = -0.0142 \pm j5.25$, and $p_{12,13} = -0.0723 \pm j12.20$. On the other hand, for the PID-PID control strategy, the poles were located at $p_1 = -0.006826$, $p_{2,3} = -0.3315 \pm j1.156$, $p_{4,5} = -0.1412 \pm j5.399$, and $p_{6,7} = -0.3775 \pm j12.26$, as shown in Fig. 11. These demonstrated that the MRCS-PID and PID-PID control parameters that were obtained by using the PSO algorithm provided a stable system, where all of the poles lie on the left half of the *s*-plane. Thus, the optimal control parameters, as shown in Figs. 8 and 9, were ready to be used through the simulation and experimental exercises.

4.2. Various trolley positions

Initially, the performance of the proposed MRCS-PID control was investigated by the unshaped control that was generated only by using the PID controller (bypassing the MRCS in Fig. 4), in order to ensure that the trolley reached the desired position. This methodology was also presented in [39], so as to observe the effectiveness of an MRCS as an oscillations control. In Fig. 12, with the desired trolley position, $x_d = 0.40$ m, both of the simulations (Sim) and the experiments (Exp) yielded a similar pattern of trolley position responses for the unshaped and MRCS-PID controllers. In the experiments, the trolley reached the destination within 4.57 s and 6.62 s, by using the unshaped and MRCS-PID controllers, respectively. It was noted that the unshaped control provided a 31% faster response than the proposed strategy.

Unfortunately, the fast motion response of the unshaped control suffered from higher hook and payload oscillation responses, as presented in Figs. 13 and 14. Meanwhile, the hook and the payload continuously oscillated with large oscilla-

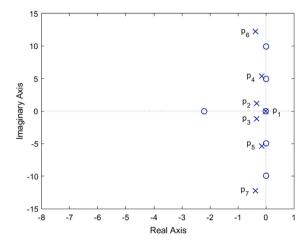


Fig. 11. Pole locations of the PID-PID control strategy.

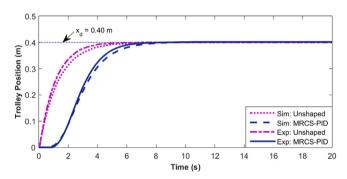


Fig. 12. Trolley movement when $x_d = 0.40$ m.

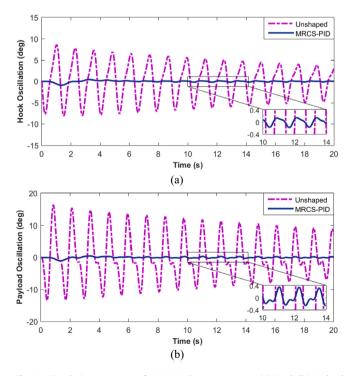


Fig. 13. Simulation responses of a DPOC when x_d = 0.40 m: (a) Hook (b) Payload.

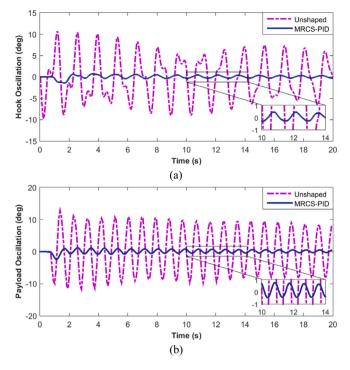


Fig. 14. Experimental responses of a DPOC when $x_d = 0.40$ m: (a) Hook (b) Payload.

tions, even after the trolley reached the desired location, which then brought forth a safety concern. In the experiments, the unshaped control provided θ_{1m} , θ_{2m} , MSE $_{\theta_1}$, and MSE $_{\theta_2}$ being as large as 10.635°, 12.810°, 20.984, and 46.855, respectively, whereas the proposed method provided 1.046°, 2.430°, 0.146, and 0.328, as shown in Fig. 14. It was found that the MRCS-PID control provided a significant oscillation reduction of 86.8% and 81% in θ_{1m} and θ_{2m} , respectively, when compared to the unshaped control. Furthermore, the hook and the payload oscillations were efficiently attenuated and eliminated within 3 s. This has indicated that the implementation of an MRCS into a hybrid control strategy contributes to a significant impact in minimising both of the oscillations.

The effectiveness of the MRCS-PID tracking control against various x_d was next investigated. The trolley was required to move ± 0.20 m from the position in the previous investigation, which were the lower ($x_d = 0.20$ m) and higher ($x_d = 0.60$ m) positions. The proposed MRCS-PID successfully positioned the trolley accurately, even though the controller parameters were kept constant, as plotted in Fig. 15. The results also showed a similar pattern of trolley positions throughout the simulations and the experiments. In the experiments, the trolley successfully reached the lower and higher positions within 5.81 s and 8.13 s, respectively.

As a higher desired position results in a higher control input, the hook and the payload oscillations when $x_d = 0.60$ m were higher than in the case when $x_d = 0.20$ m. This can be seen in Figs. 16 and 17. The overall performances of the hook and the payload using different trolley positions are summarised in Fig. 18. The effectiveness and the robustness of the MRCS-PID control were further investigated, by comparing with the PID-PID control under scenarios (ii) and (iii), in Sections 4.3 and 4.4, respectively.

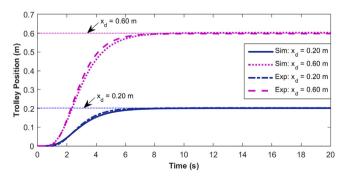


Fig. 15. Trolley position responses with various x_d

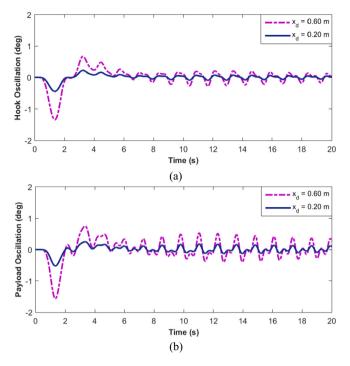


Fig. 16. Simulation responses of a DPOC at various x_d : (a) Hook (b) Payload.

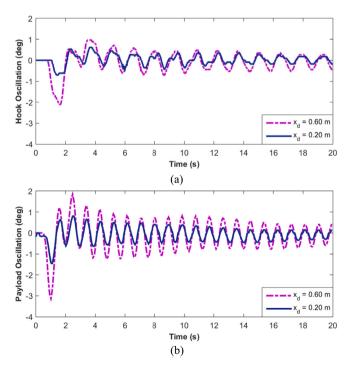


Fig. 17. Experimental responses of a DPOC at various x_d : (a) Hook (b) Payload.

4.3. Payload hoisting

In this work, the effectiveness of the proposed hybrid control was evaluated for a case of varying cable lengths during the payload hoisting, as this operation is essential in industries. The same DPOC parameters as in Section 4 were used, but l_1 was

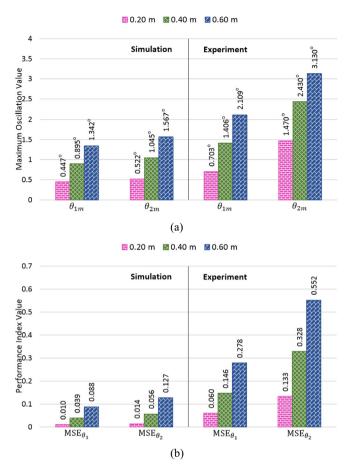


Fig. 18. Performance of the controller at various x_d : (a) Maximum oscillation (b) Overall oscillation.

continuously hoisted from 0.20 m to 0.40 m (l_1 = 0.20–0.40 m) and f_i = 75 Ns/m. It was clearly noted that the payload reached 0.40 m in 2.09 s and 2.24 s in the simulation and in the experiment, respectively, as depicted in Fig. 19.

In these cases, the MRCS-PID control was further investigated by comparing with a PID-PID feedback control, as shown in Fig. 6. The parameters of the MRCS-PID control remained unchanged, as in Fig. 8, while the PID-PID control parameters that were obtained in Fig. 9 were used for both the simulation and the experiment implementations. Since the higher desired position induced the higher oscillation responses, as demonstrated in Figs. 16-18, $x_d = 0.60$ m was selected as an extreme condition, in order to investigate the effectiveness and the robustness of the proposed method under payload hoisting. The MRCS-PID and PID-PID control strategies yielded the control inputs, u, as shown in Fig. 20.

As expected, both of the MRCS-PID and PID-PID control strategies that were using optimal control parameters, successfully moved the trolley to the desired location accurately, as shown in Fig. 21. For the MRCS-PID control when using a smoother control input, as shown in Fig. 20, the experimental results showed that the trolley was able to reach the desired position within 8.39 s. Noticeably, this response was 15.9% faster than the PID-PID control that required 9.98 s with a visible jerk motion, as zoomed-in Fig. 21. In addition, the smoother control input in the MRCS-PID control strategy resulted in a less actuator effort.

Moreover, both of the oscillation results in the simulation and in the experimental were also recorded in Figs. 22 and 23, respectively. The jerk motion in the PID-PID control disrupted the hook oscillation over the first four seconds, as exhibited in Fig. 23(a). In contrast to the case in Section 4.2, the payload motion was measured experimentally in centimetres. In fact, the PID-PID control resulted in higher θ_{1m} , θ_{2m} , MSE $_{\theta_1}$, and MSE $_{\theta_2}$ that reached 4.922°, 5.030 cm, 2.830, and 5.192, respectively, as recorded in Table 1. With further observations, the MRCS-PID achieved 2.022°, 3.920 cm, 0.217, and 0.488, with reductions of 58.9%, 22.1%, 92.3%, and 90.6% in θ_{1m} , θ_{2m} , MSE $_{\theta_1}$, and MSE $_{\theta_2}$, respectively, when compared to the results of the PID-PID control.

4.4. Payload hoisting with payload mass variations

From a practical perspective, the payload changed under the circumstances of payload mass variations. Using $x_d = 0.60$ m, the robustness of the MRCS-PID and the PID-PID control schemes was further investigated, by considering a payload,

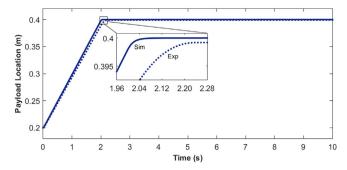


Fig. 19. Varying cable lengths during the payload hoisting ($l_1 = 0.20-0.40 \text{ m}$).

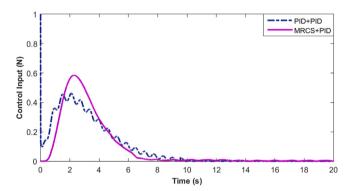


Fig. 20. Control inputs for the MRCS-PID and PID-PID control strategies.

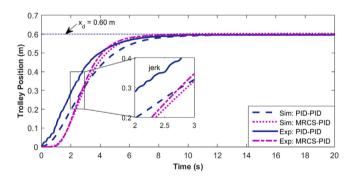


Fig. 21. Trolley position at $l_1 = 0.20-0.40$ m.

 m_2 = 0.20 kg, which was twice the amount of the previous investigation. From Fig. 24, even using a different payload mass, the proposed method successfully positioned the trolley at the desired position. Similarly, the simulation and the experimental responses showed that the PID-PID control resulted in much larger hook and payload oscillations, as shown in Figs. 25 and 26, respectively. The overall effects of the payload, m_2 = 0.20 kg are recorded in Table 2. Specifically, the experimental results revealed that the MRCS-PID control was found to be superior at θ_{1m} , θ_{2m} , MSE $_{\theta_1}$, and MSE $_{\theta_2}$, with reductions of 52.7%, 36.6%, 86.4%, and 83.8%, respectively, under payload hoisting with a higher payload mass.

4.5. Other control issues

For all scenarios, it was noted that the hook and payload oscillations obtained using the MRCS-PID control in experiments (Figs. 14, 17, 23 and 26) were slightly larger than the simulation results (Figs. 13, 16, 22 and 25). This might be due to the friction which was ignored in the simulation. Nevertheless, the experimental hook and payload oscillations were considerably small, less than 1 degree.

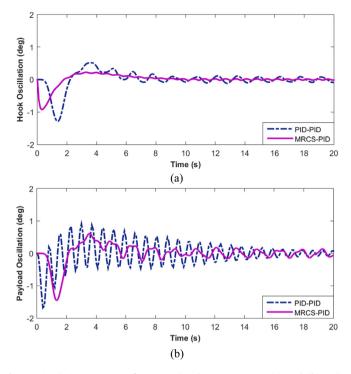


Fig. 22. Simulation responses of a DPOC when $l_1 = 0.20-0.40$ m: (a) Hook (b) Payload.

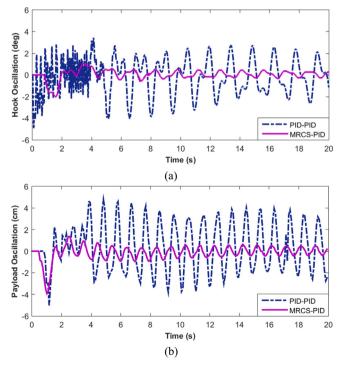


Fig. 23. Experimental responses of a DPOC when $l_1 = 0.20-0.40$ m: (a) Hook (b) Payload.

By using the optimal parameters of the command shaper and the feedback controller obtained with the PSO, the MRCS-PID control was shown to be robust to the crane parameter uncertainties involving varying cable lengths and payload mass variations. Lower and satisfactory hook and payload oscillations were achieved under these cases as compared to other

Table 1 Performance of the control strategies when l_1 = 0.20–0.40 m.

	Method	θ_{1m}	θ_{2m}	$MSE_{ heta_1}$	$MSE_{ heta_2}$
Sim	PID-PID MRCS-PID	1.286° 0.929°	1.678° 1.454°	0.078 0.035	0.149 0.092
	% Reduction	27.8%	13.3%	55.1%	38.3%
Exp	PID-PID	4.922°	5.030 cm	2.830	5.192
	MRCS-PID	2.022°	3.920 cm	0.217	0.488
	% Reduction	58.9%	22.1%	92.3%	90.6%

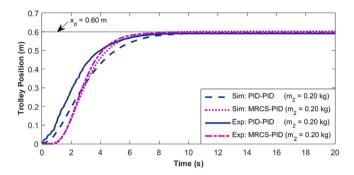


Fig. 24. Trolley position, with payload hoisting and a different payload mass.

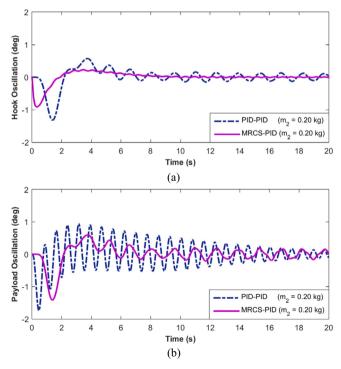


Fig. 25. Simulation responses of a DPOC when $m_2 = 0.20$ kg: (a) Hook (b) Payload.

approaches. However, the proposed control structure especially relating to the MRCS is not robust to the influence of external disturbances such as wind. To handle the disturbances, a scheme that can automatically update the hybrid controller has to be developed. One of the techniques which can be explored in the future is as presented in [5,46], in which a neural network was used to predict and update the input shaping parameters online.

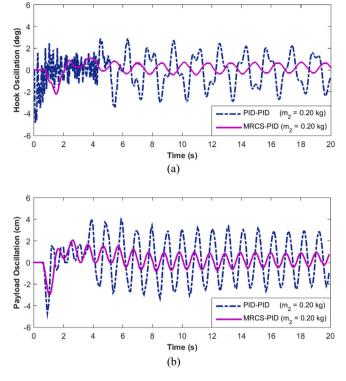


Fig. 26. Experimental responses of a DPOC when $m_2 = 0.20$ kg: (a) Hook (b) Payload.

Table 2 Performance of the controllers when $m_2 = 0.20$ kg.

	Method	θ_{1m}	θ_{2m}	$MSE_{ heta_1}$	$MSE_{ heta_2}$
Sim	PID-PID	1.321°	1.745°	0.085	0.176
	MRCS-PID	0.916°	1.415°	0.035	0.095
	% Reduction	30.7%	18.9%	58.8%	46%
Exp	PID-PID	4.834°	4.700 cm	2.190	4.034
	MRCS-PID	2.285°	2.980 cm	0.297	0.653
	% Reduction	52.7%	36.6%	86.4%	83.8%

In this work, the cable of DPOC system is considered to be rigid, with the assumption that it does not bend during payload swing. As described in [47], in certain cases, cable flexibility occurs that will exhibit flexible characteristics and introduce bending deformation during payload oscillation. This may result in a larger amplitude payload oscillation, similar to the effect of external disturbance. In addition, the natural frequency of the DPOC may also be affected. As the proposed MRCS-PID is designed based on optimal constant command shaper parameters and feedback controller gains, the system performance will be affected. In future, it will be interesting to explore the changes in the system dynamics of the DPOC due to the cable flexibility, and to design an improved MRCS controller to handle this issue.

5. Conclusion

A new control structure using an improved MRCS, together with a practical feedback controller (MRCS-PID), was designed for accurate trolley positioning and oscillation control for an underactuated DPOC under parameter uncertainties. The existing MRCS algorithm was improved, in order to reduce the design complexity and to ensure concurrent tuning with the feedback controller can be performed. This was realised using the PSO algorithm, which has not been implemented in the existing literature. In addition, the advantage of an MRCS, which does not require prior knowledge of crane oscillation frequencies, was retained. The proposed approach was demonstrated to be robust under the parameter uncertainties, with respect to the changes in crane dynamics that resulted from the various desired trolley positions, the payload hoisting, and the payload mass variations. Under these conditions, higher reductions in the maximum and overall oscillations were attained when compared to the PID-PID control, in both of the simulations and the experiments. It is envisaged that the

improved MRCS can be further utilised for control of other underactuated systems with a higher system order. These include a multi-link flexible robot manipulator, various types of double-pendulum crane and a drone with a cable-suspension payload. The vibrations and oscillations occur in these systems involve with several modes of frequencies and damping ratios.

CRediT authorship contribution statement

H.I. Jaafar: Conceptualization, Methodology, Software, Investigation, Writing - original draft. **Z. Mohamed:** Conceptualization, Validation, Writing - review & editing, Supervision. **M.A. Ahmad:** Conceptualization, Writing - review & editing. **N.A. Wahab:** Validation, Writing - review & editing. **L. Ramli:** Conceptualization, Validation. **M.H. Shaheed:** Conceptualization.

Declaration of Competing Interest

The authors declare that they have no known competing financial interests or personal relationships that could have appeared to influence the work reported in this paper.

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