

# **FPGA based PWM techniques for controlling Inverter**

A thesis submitted in partial fulfillment of the requirements for the degree

*of*

**Bachelor of Technology**

*in*

**Electronics and Instrumentation Engineering**

*by*

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*Under the guidance of*

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## **National Institute of Technology Rourkela CERTIFICATE**

This is to certify that the thesis entitled, “*FPGA based PWM techniques for controlling Inverter*” submitted by SURYAKANT BEHERA (Roll No.-10607011) in partial fulfilment of the requirements for the degree of Bachelor of Technology in Electronics & Instrumentation Engineering, Session 2006-2010, in the Department of Electronics and Communication Engineering, National Institute of Technology, Rourkela is an authentic work carried out by him under my supervision and guidance.

To the best of my knowledge, the matter embodied in the thesis has not been submitted to any other University / Institute for the award of any Degree.

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**Suryakant Behera**

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## **ABSTRACT**

Pulse Width Modulation has nowadays become an integral part of every electronics system. These techniques have been widely accepted and are researched extensively nowadays. It has found its application in large number of applications as a voltage controller. Its use in controlling output voltage of Inverter is the most frequently used application. There are basically two main techniques of PWM Generation- Analog technique and Digital Technique.

This thesis deals with these two techniques. First Analog techniques were studied in detail but these techniques have some demerits. Due to these demerits digital techniques were studied. Various digital PWM Generator topologies were studied. The VHDL code for each of these topologies was written and synthesized using Xilinx ISE 10.1 software. Behavioral Simulation was performed on the architecture and after verifying the results this VHDL code was downloaded to SPARTAN 3E FPGA. After downloading the code in FPGA real time debugging was done for the architecture. The results were seen in Chipscope Pro software.

Also from Synthesis report generated after synthesizing the VHDL code of each digital PWM Generator topologies comparison was done between these topologies in terms of number of logic blocks used and device utilization of each architecture.

**Key Terms : PWM, FPGA, VHDL, Inverter**

# **Chapter 1**

## **Introduction**



# Chapter 1

## Introduction

### 1.1 Introduction

Pulse Width Modulation (PWM) has now become an integral part of almost all embedded systems. It has been widely accepted as control technique in most of the electronic appliances.

These techniques have been extensively researched during past few years [2]. There are various methods depending upon architecture and requirement of the system. Their design implementation depends upon application type, power consumption, semiconductor devices, performance and cost criteria all determining the PWM method according to N.A. Rahim and Z. Islam [2].

One of the most important application of PWM lies in power electronics applications for controlling power converters (DC/DC, DC/AC, etc.) according to E. Koutroulis, A.Dollas and K.Kalaitzakis in [1]. PWM Inverters are one of those power converters which extensively use concept of PWM for its operation. PWM inverters are recently showing great popularity for industrial applications because of their superior performance.

Advancement in designing technology and development in Semiconductor Electronics has led to this popularity. A numerous PWM schemes are used to obtain variable voltage and frequency supply.

According to N.A. Rahim and Z. Islam in [2], there are two classes of PWM techniques identified optimal PWM and carrier PWM. The optimal PWM requires lot of computation and hence extra hardware and hence extra cost [2]. Carrier PWM techniques require a carrier signal which is modulated with modulating signal to produce desired PWM signal.

## **1.2 PWM Techniques**

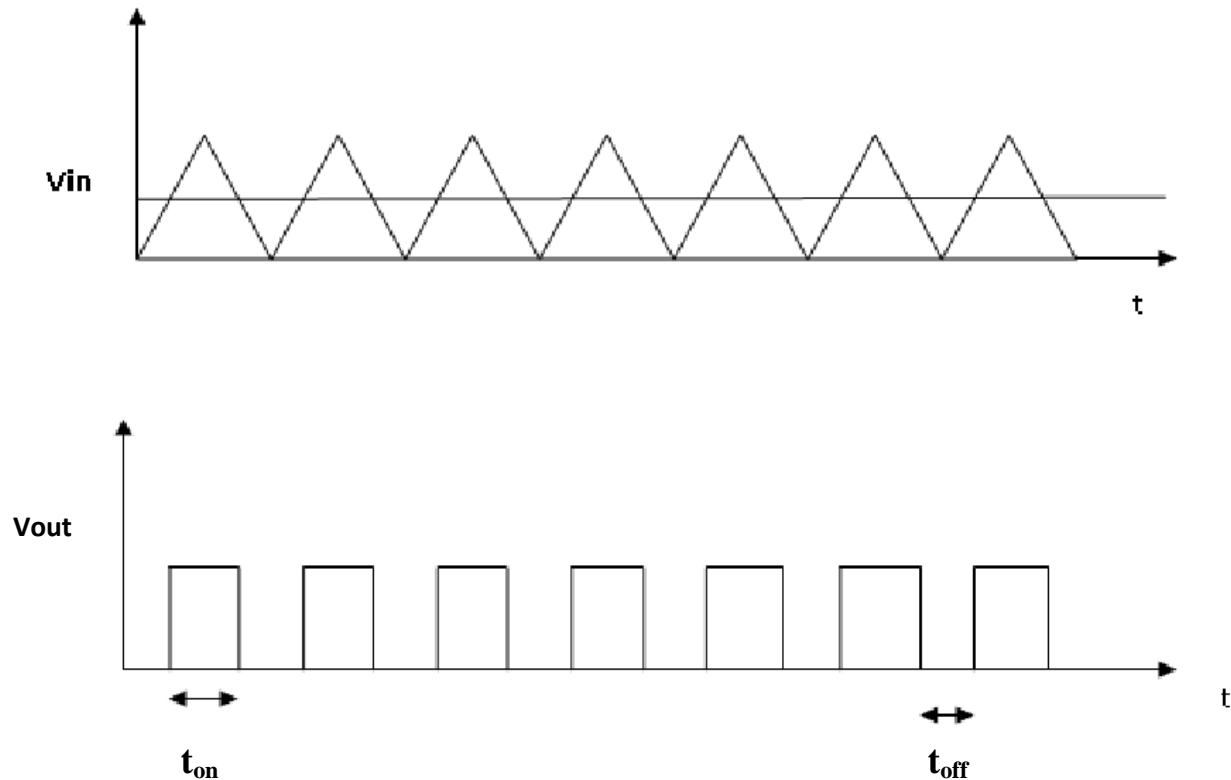
There are basically two PWM techniques –Analog and Digital Techniques. In analog techniques there is a carrier signal and a modulating signal. These two signals are compared using comparator. The output of this comparator is the desired PWM output. There are basically four analog techniques (a) Sinusoidal PWM (b) Modified Sinusoidal (c) Single Pulse Modulation (d) Multiple Pulse Modulation. According to [2] and [4]-[7] the disadvantages of these analog methods are that they are prone to noise and they change with voltage and temperature change. Also they suffer changes due to component variation [1]. They are less flexible as compared to digital methods.

Digital methods are the most suited form for designing PWM Generators. They are very flexible and less sensitive to environmental noise [2]. Also they are simple to construct and can be implemented very fastly. Most of the digital techniques employ counter and comparator based circuits. These techniques are discussed in detail in Chapter-4. Analog techniques are discussed in detail in Chapter-3.

## **1.3 Advantages of FPGA based design**

Field Programmable Gate Array (FPGA) offers the most preferred way of designing PWM Generator for Power Converter Applications. They are basically interconnection between different logic blocks. When design is implemented on FPGA they are designed in such a way that they can be easily modified if any need arise in future. We have to just change the interconnection between these logic blocks. This feature of Reprogramming capability of FPGA makes it suitable to make your design using FPGA [1]. Also using FPGA we can implement design within a short time. Thus FPGA is the best way of designing digital PWM Generators. Also implementation of FPGA-based digital control schemes prove less costly and hence they are economically suitable for small designs [1]. Hence in this thesis FPGA based PWM Generator technique is discussed.

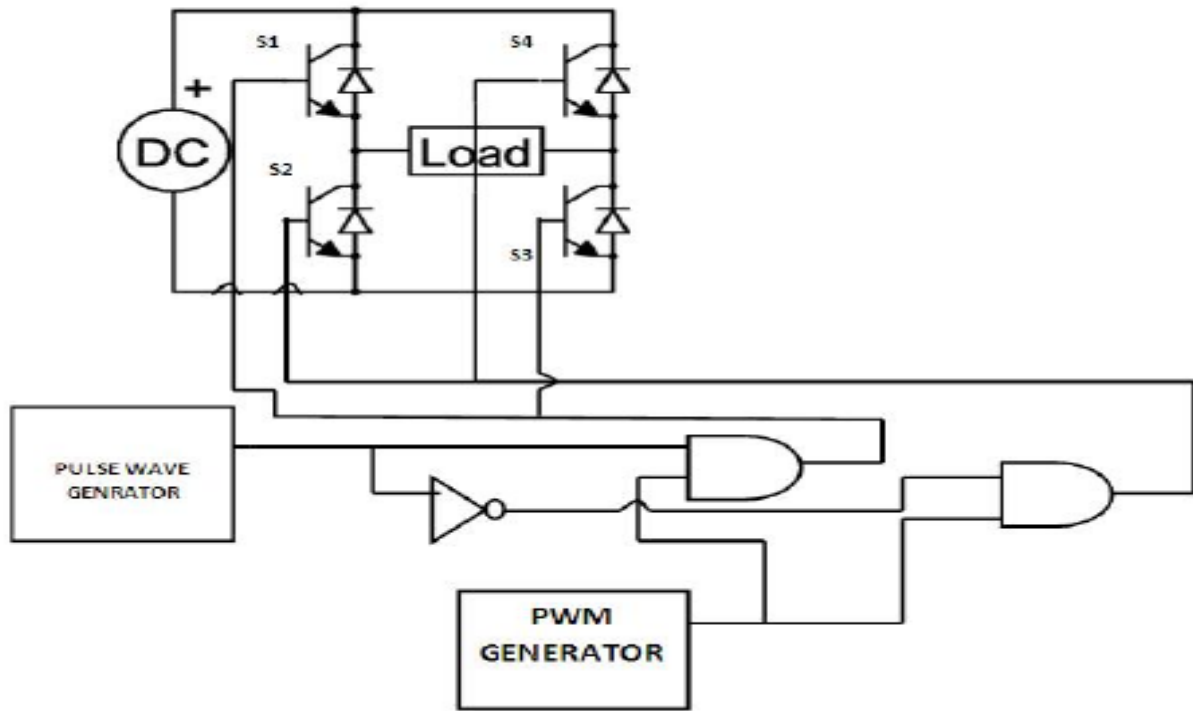
**Fig.1** shows one of the basic method of PWM Generation.



**Fig. 1:** PWM Generation Method

## 1.4 PWM Control of Inverter

The application of PWM control in a Inverter (DC/AC) is shown in **Fig. 2**. The PWM control signal,  $V_{PWM}$  in **Fig. 2**, is generated from PWM generator. This  $V_{PWM}$  is logically ANDED with rectangular pulse waveform coming from pulse generator and is fed to power switches S1 and S3. The inverted rectangular waveform is logically ANDED with PWM waveform and is fed to power switches S2 and S4. Thus ON and OFF time of power switches are controlled by this PWM control signal to modulate input DC voltage to required AC voltage.



**Fig.2** PWM Control of Inverter

The power switch is usually of MOSFET or IGBT. The size of Inverter depends on size of these power switches. Since frequency of operation is inversely dependent upon Inverter size so we have to increase the switching frequency to reduce the Inverter size [1]. So we have to look into the frequency aspect of PWM Generator used so that we get optimized size of Inverter by proper selection of frequency of PWM wave.

# **Chapter 2**

## **Voltage Source Inverters**

# Chapter 2

## Voltage Source Inverters

### 2.1 Definition

Inverters are static power converters that produce an ac output waveform from a dc power supply according to M.H. Rashid [3]. They are applied in adjustable speed drives (ASDs), uninterruptible power supplies (UPS), etc. which are its important applications [3]. For sinusoidal ac outputs, the magnitude, frequency, and phase should be controllable [3]. The voltage source inverters (VSIs), are the inverters whose independently controlled ac output is a voltage waveform.

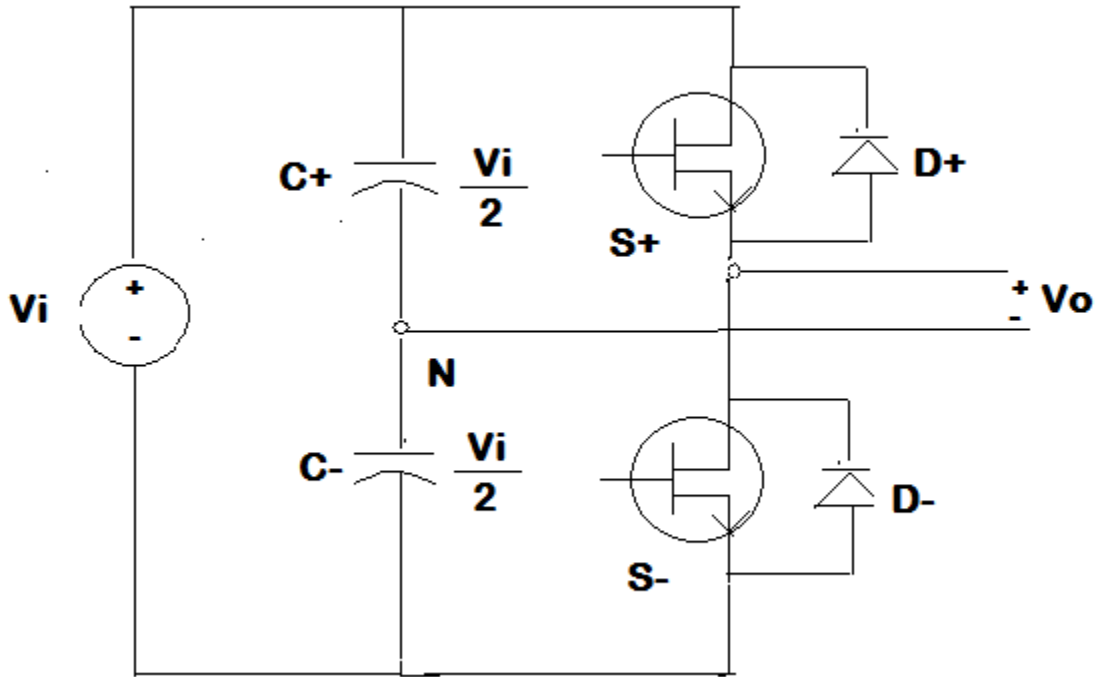
### 2.2 Single Phase Voltage Source Inverter

Single phase inverters are those inverters which produces only single phase of ac output. Single phase inverters can be divided into two categories (a) Half Bridge Single Phase Voltage Source Inverter (b) Full Bridge Single Phase Voltage Source . Although the power range they cover is the low one, they are widely used in power supplies and single phase UPS and multicell configurations according to M.H. Rashid [3] .

#### 2.2.1 Half Bridge Voltage Source Inverter(VSI)

**Fig.3** shows the circuit configuration of a half bridge VSI. These capacitors are required to filter out the low order harmonics produced by operation of inverter according to M.H. Rashid [3]. There are two power switches  $S_+$  and  $S_-$  which in **Fig.3** are MOSFET's. These two switches can not be ON at the same time because a short circuit across the DC voltage source  $V_i$  would be

produced. To avoid the undefined voltage condition and short circuit we should ensure that either of the two switches should remain ON.



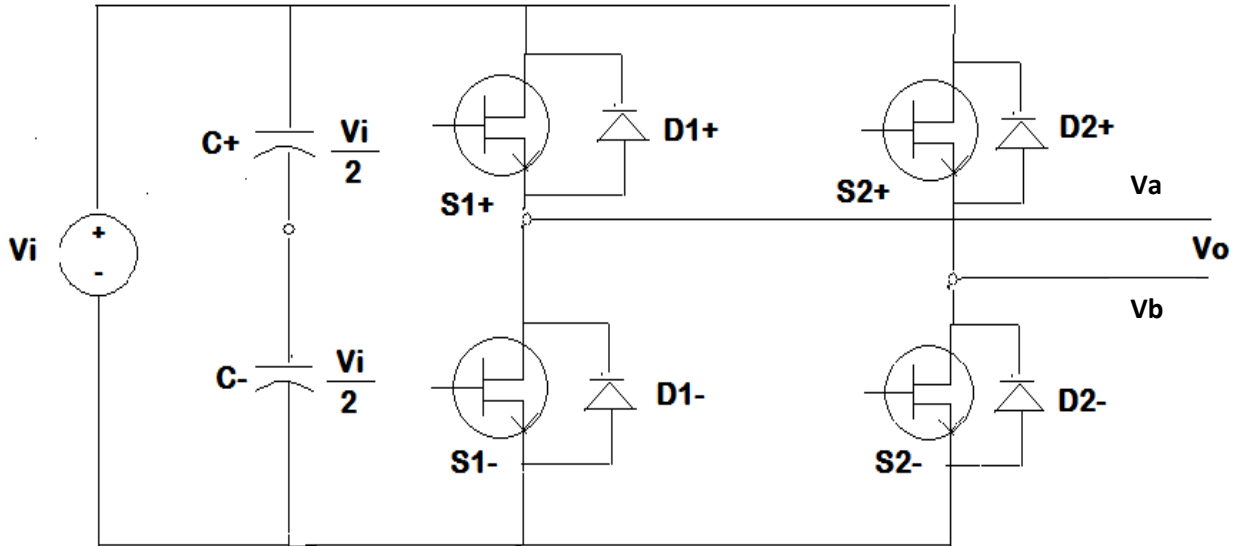
**Fig.3** Single Phase Half Bridge Voltage Source Inverter

When  $S_+$  is ON and  $S_-$  is OFF than inverter power switches, in this case MOSFET, are in State1 Amplitude of the output voltage is  $\frac{V_i}{2}$ . Similarly when  $S_-$  is ON and  $S_+$  is OFF than inverter switches are in State 2. Amplitude of the output voltage is  $-\frac{V_i}{2}$ . State 3 is that state in which both  $S_+$  and  $S_-$  are off.

### 2.2.2 Full Bridge Voltage Source Inverter (VSI)

**Fig.4** shows the circuit configuration of Full Bridge Inverter. There are four power switches  $S_{1+}, S_{1-}, S_{2+}$  and  $S_{2-}$  which are MOSFET in this case.. Switches  $S_{1+}$  and  $S_{1-}$  can not be ON at the same time. Same is the case with switches  $S_{2+}$  and  $S_{2-}$ . They can not be ON at the same time

because short circuit will be produced across DC voltage source  $V_i$ . We should avoid undefined state condition because we always want to define ac output voltage clearly. Output AC voltage can take value up to  $V_i$  which is twice as obtained in Half bridge inverter. This thesis aims at designing PWM circuit for controlling this full bridge inverter.



**Fig.4** Single Phase Full Bridge Voltage Source Inverter

Looking into **Fig.4** when switches  $S_{1+}$  and  $S_{2-}$  are ON and  $S_{1-}$  and  $S_{2+}$  are OFF then inverter is in state 1. In this state  $V_a = V_i/2$  and  $V_b = -V_i/2$ . Since  $V_o = V_a - V_b$  therefore  $V_o = V_i$  for state 1. Similarly when  $S_{1-}$  and  $S_{2+}$  are ON and  $S_{1+}$  and  $S_{2-}$  are OFF then inverter is in state 2 In this case  $V_o = -V_i$ . When  $S_{1+}$  and  $S_{2+}$  are ON and  $S_{1-}$  and  $S_{2-}$  are OFF then inverter is in state 3. In this case output voltage  $V_o$  of inverter is  $V_o = 0$  since both  $V_a$  and  $V_b$  are equal to  $V_i/2$ . Similarly when  $S_{1-}$  and  $S_{2-}$  are ON and  $S_{1+}$  and  $S_{2+}$  are OFF then inverter is in state 4 and in this case also  $V_o = 0$ . Last state, State 5 arises when  $S_{1+}, S_{2+}, S_{1-}$  and  $S_{2-}$ , are all OFF.

### 2.3 Pulse Width Modulation in Inverter

Output Voltage of the Inverter can be modified or controlled by controlling or modifying switching current, or in case of Power Switches, by controlling or modifying Gate current. This control is achieved by PWM control.



In one of the methods of controlling inverter output voltage, a fixed DC voltage is given to the inverter and by varying the ON and OFF time of power switches we get a controlled or modified AC output voltage. This method is popularly called as Pulse Width Modulation (PWM) method.

The advantages of the PWM control are:

- (1) PWM control is very simple and require very less hardware. So they are also cost effective.
- (2) They can easily implemented using DSP or FPGA.

The major disadvantage of PWM control is that power switches associated with PWM switching are very costly as their response time should be very fast. So this increases the total cost of control.

PWM waves are actually pulses of constant amplitude and varying pulse widths. This width can be varied by different modulation schemes. The most famous of these modulation schemes are analog methods which are :

- (a) Single Pulse Modulation
- (b) Sinusoidal Pulse Width modulation
- (c) Modified Sinusoidal PWM
- (d) Multiple Pulse Width Modulation

All these techniques are discussed in detail in Chapter-3. Here we studied about Inverters and how to control them using PWM control.

# Chapter 3

## Analog techniques of PWM generation

# Chapter 3

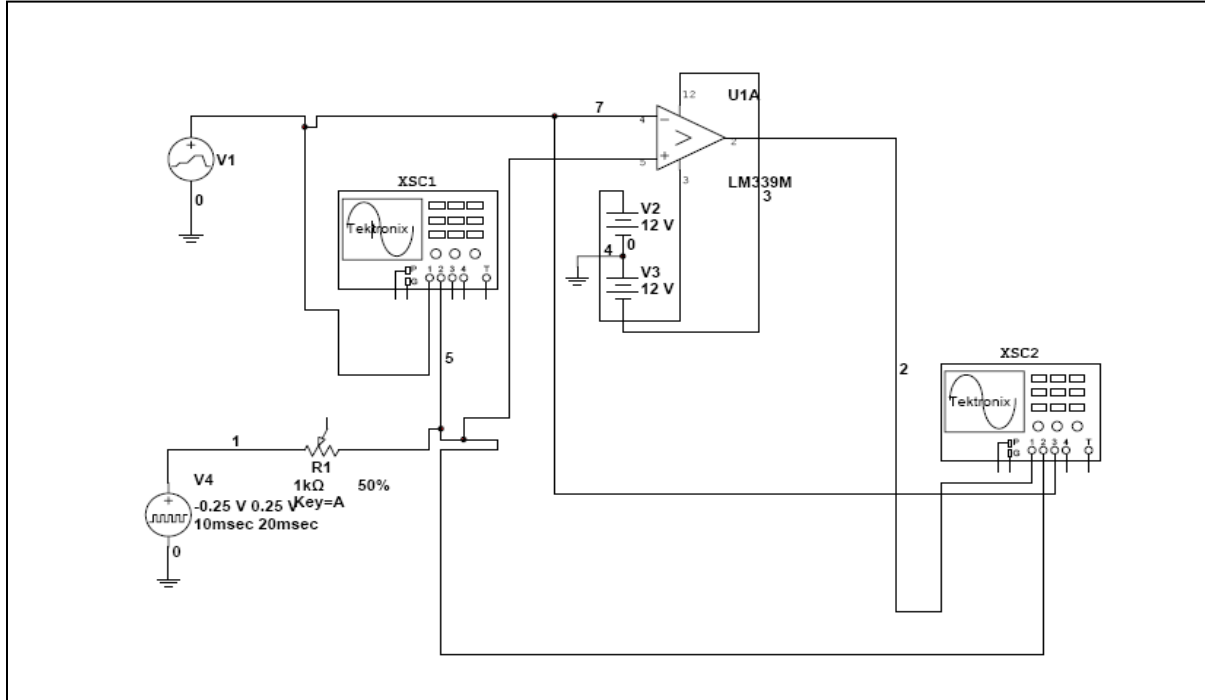
## Analog techniques of PWM generation

### 3 Analog Techniques

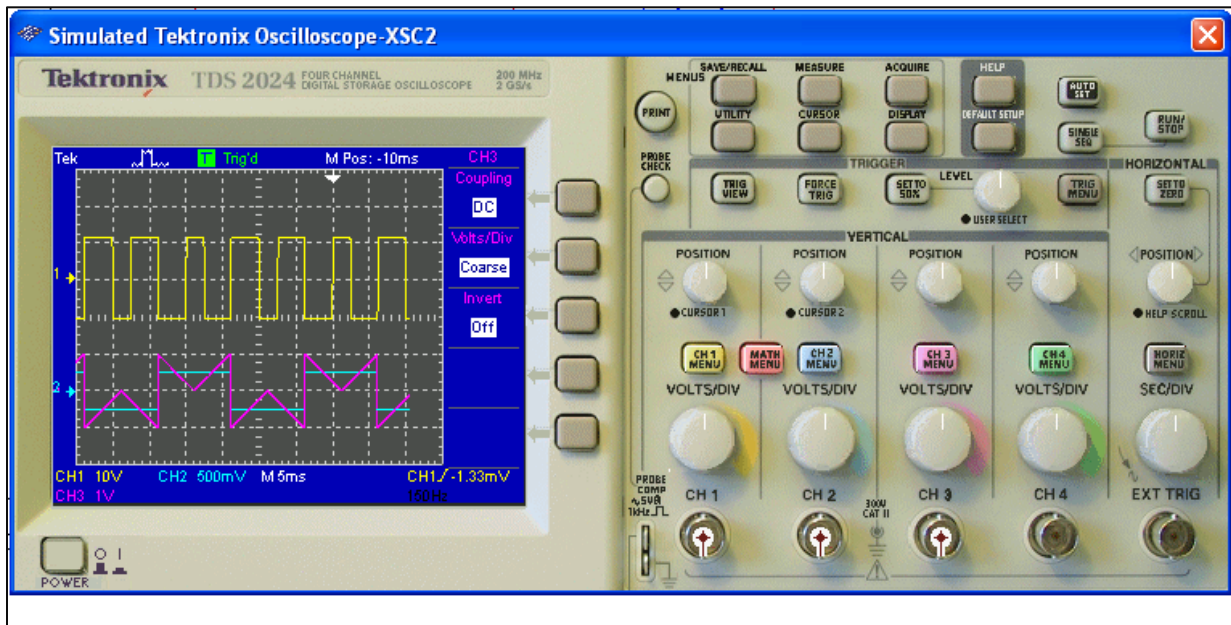
Here we will discuss about various analog methods of generating PWM signals. These method employs the concept of carrier modulation. In this modulation there is one carrier wave which is basically high frequency triangular pulse train and modulating signal which can be sinusoidal, DC signal ,etc depending upon which type of modulation is used. Than these two signals are compared using comparator to give desired PWM output. There are four basic analog modulation methods : (a) Single Pulse Modulation (b)Sinusoidal Pulse Width modulation (c) Modified Sinusoidal PWM (d) Multiple Pulse Width Modulation

#### 3.1 Single Pulse Modulation

In this modulation technique a square wave waveform is compared with triangular waveform and we will get resultant PWM signal. This modulation gives quasi-square wave output. There is single pulse of output voltage during each half cycle. RMS Value of output voltage can be controlled by varying the pulse width. The ratio of triangular wave signal amplitude ( $P_c$ ) and square wave signal ( $P_r$ ) is called modulation index i.e.  $m = \frac{P_r}{P_c}$  . The width of the pulse can be changed by varying the modulation index. When  $m=1$  , Square wave output is obtained. The circuit for this modulation was created in MULTISIM and was simulated in MULTISIM. The circuit and simulation is shown in **Fig.5** and **Fig.6** respectively.



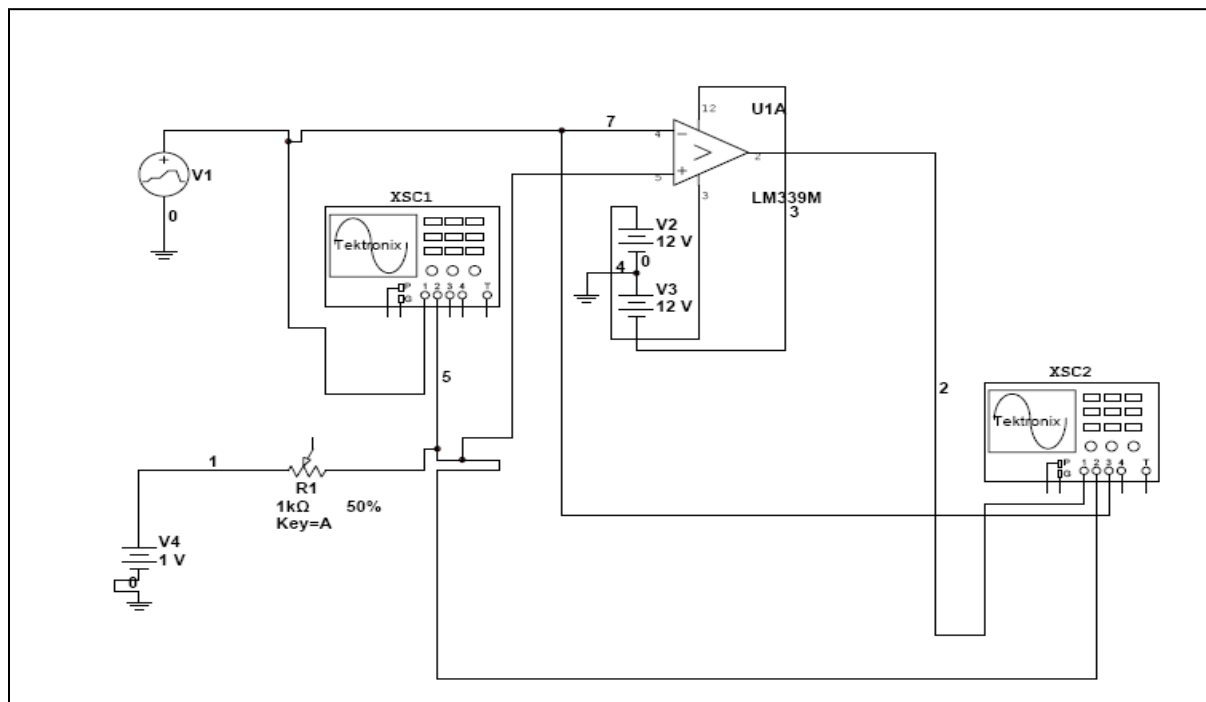
**Fig.5** Circuit for Single Pulse Modulation in MULTISIM



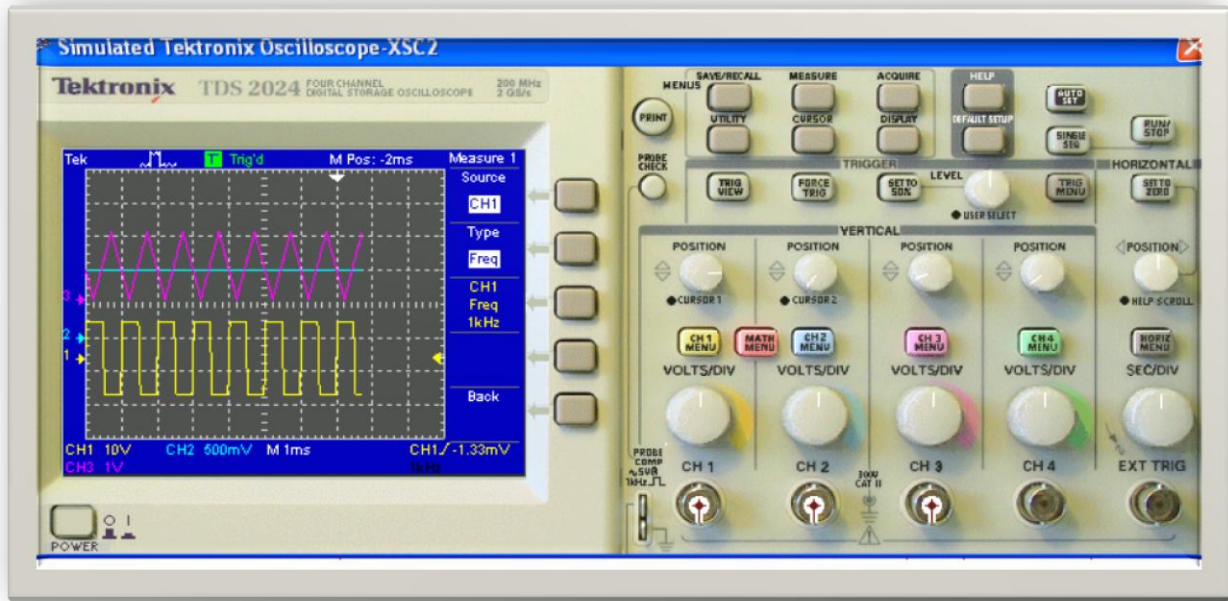
**Fig.6** Simulation of Single Pulse Modulation in MULTISIM

## 3.2 Multiple Pulse Width Modulation

In this modulation technique a DC signal is compared with Triangular waveform and we will get resultant PWM signal. This modulation gives multiple pulses to reduce harmonic content.. RMS Value of output voltage can be controlled by varying the pulse width. The ratio of triangular wave signal frequency ( $F_c$ ) and frequency of output waveform ( $F_o$ ) is called frequency modulation ratio i.e.  $m_f = \frac{F_r}{F_o}$ . The width of the pulse can be changed by varying the amplitude of DC reference. Number of Pulses per half cycle i.e.  $p = \frac{mf}{2}$ . The circuit for this modulation was created in MULTISIM and was simulated in MULTISIM. The circuit and simulation is shown in **Fig.7** and **Fig.8** respectively.



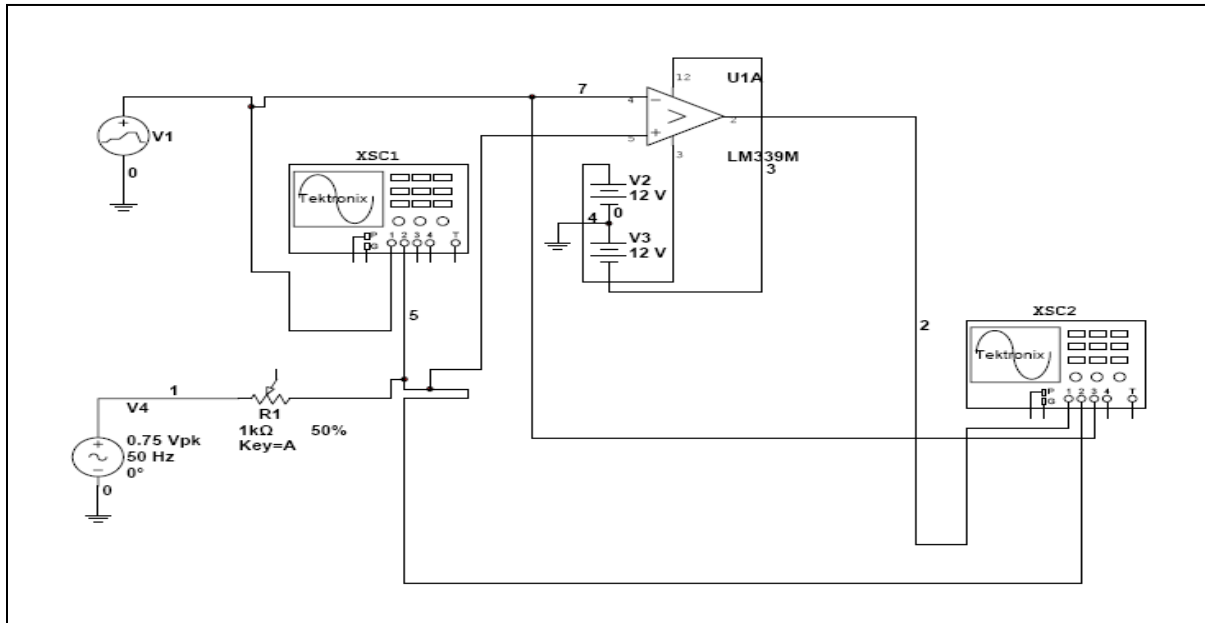
**Fig.7** Circuit for Multiple Pulse Width Modulation in MULTISIM



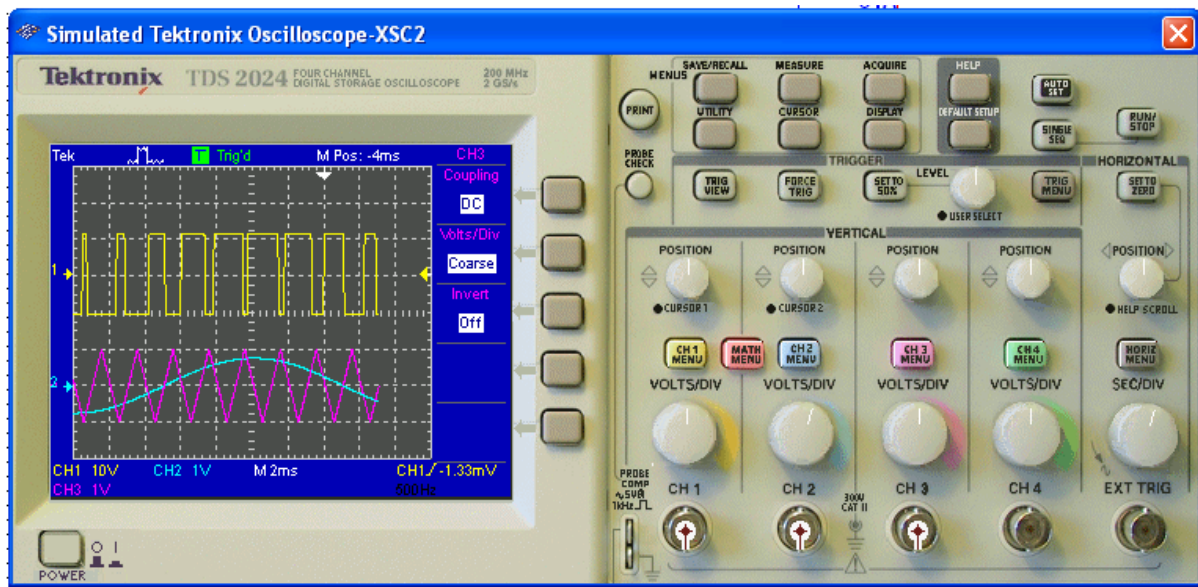
**Fig.8** Simulation of Multiple Pulse Width Modulation in MULTISIM

### 3.3 Sinusoidal Pulse Width Modulation

In this modulation technique a sinusoidal signal is compared with Triangular waveform and we will get resultant PWM signal. The width of each pulse is weighted by the amplitude of sine wave at that instant. RMS Value of output voltage can be controlled by varying the pulse width. The ratio of triangular wave signal frequency ( $F_c$ ) and frequency of output waveform ( $F_o$ ) is called frequency modulation ratio i.e.  $m_f = \frac{F_r}{F_o}$  .. The ratio of triangular wave signal amplitude ( $P_c$ ) and sinusoidal wave signal ( $P_r$ ) is called modulation index i.e.  $m = \frac{P_r}{P_c}$  .It is the most widely used method of voltage control in Inverters. The circuit for this modulation was created in MULTISIM and was simulated in MULTISIM. The circuit and simulation is shown in **Fig.9** and **Fig.10** respectively.



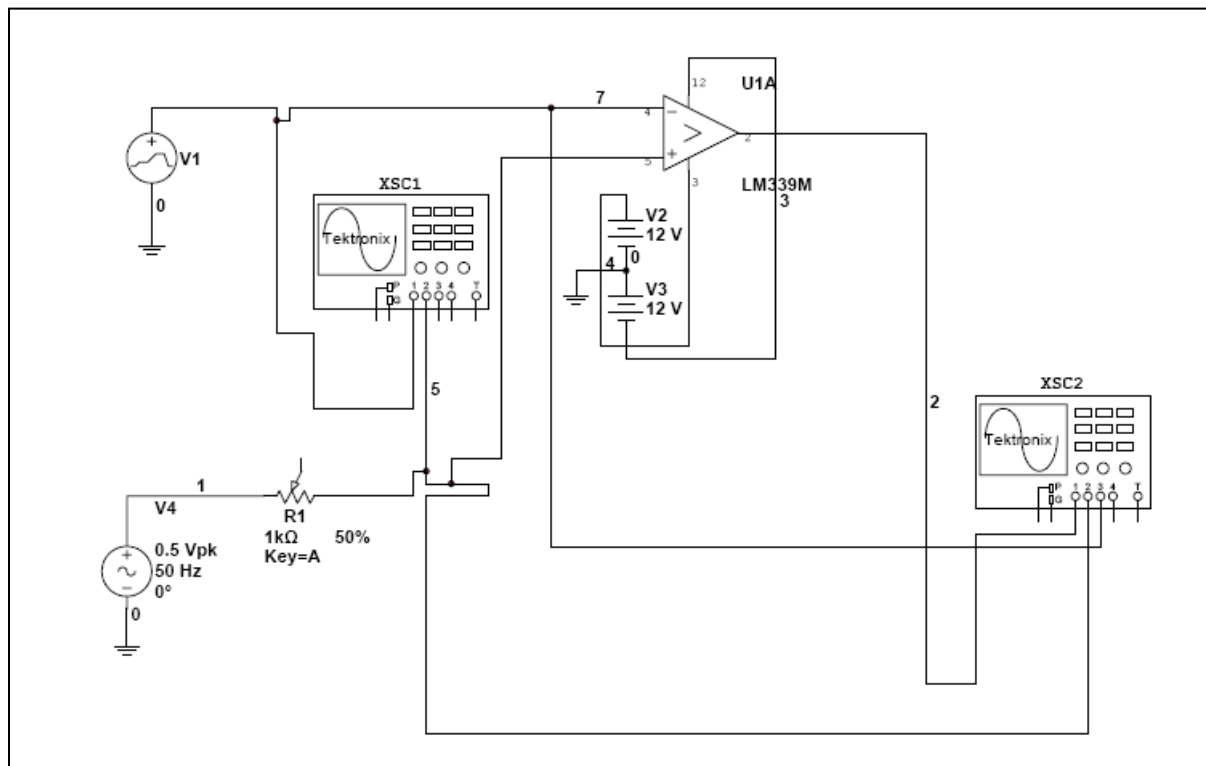
**Fig.9** Circuit for Sinusoidal Pulse Width Modulation in MULTISIM



**Fig.10** Simulation of Sinusoidal Pulse Width Modulation in MULTISIM

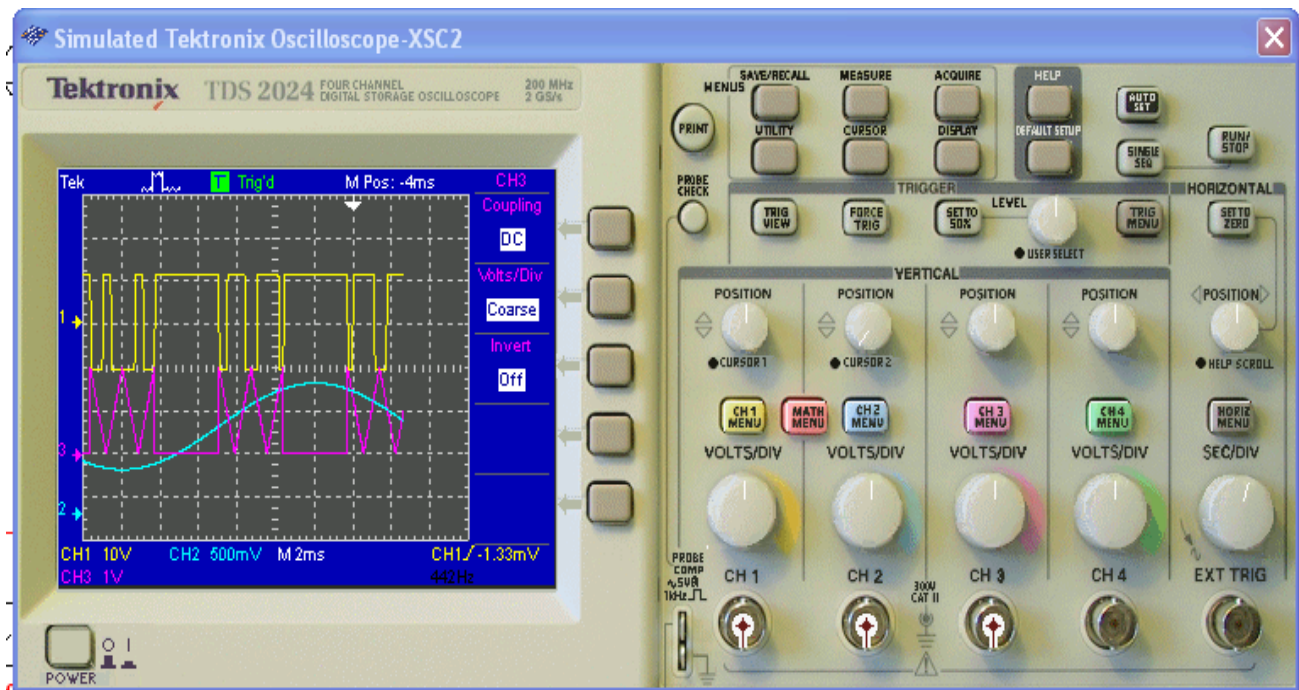
### 3.4 Modified Sinusoidal Pulse Width Modulation

The widths of the pulses near peak of the sine wave do not change much when modulation index is changed. According to M.H. Rashid [3] in this method carrier triangular wave is suppressed at  $30^\circ$  in the neighbourhood of peak of sine wave. Hence triangular wave is present for the period of first  $60^\circ$  and last  $60^\circ$  of the half cycle of sine wave [3]. The middle  $60^\circ$  of the sine wave do not have triangular wave. Hence the generated PWM has less number of pulses [3] as compared to sinusoidal wave. Its RMS value can be changed by changing the amplitude of sinusoidal wave. This modulation scheme reduces harmonic content and switching losses but implementation of this scheme is tougher than sinusoidal PWM technique [3]. The circuit for this modulation was created in MULTISIM and was simulated in MULTISIM. The circuit and simulation is shown in **Fig.11** and **Fig.12** respectively.



**Fig.11** Circuit for Modified Sinusoidal Pulse Width Modulation in MULTISIM





**Fig.12** Simulation of Modified Sinusoidal Pulse Width Modulation in MULTISIM

### 3.5 Disadvantages of Analog Modulation scheme

There are some disadvantages associated with analog modulation scheme which are listed below:

- (1) They are prone to environmental noise and temperature changes as said in [2] and [4]-[7] . Hence they are not suitable where these factors are prominent.
- (2) They also suffer variation due to component variation .e.g. for a variation in comparator there is variation in PWM output.

To overcome these various problems various Digital techniques are there. Few of these techniques are discussed in next chapter.

# **Chapter 4**

## **Digital Techniques of PWM Generation**

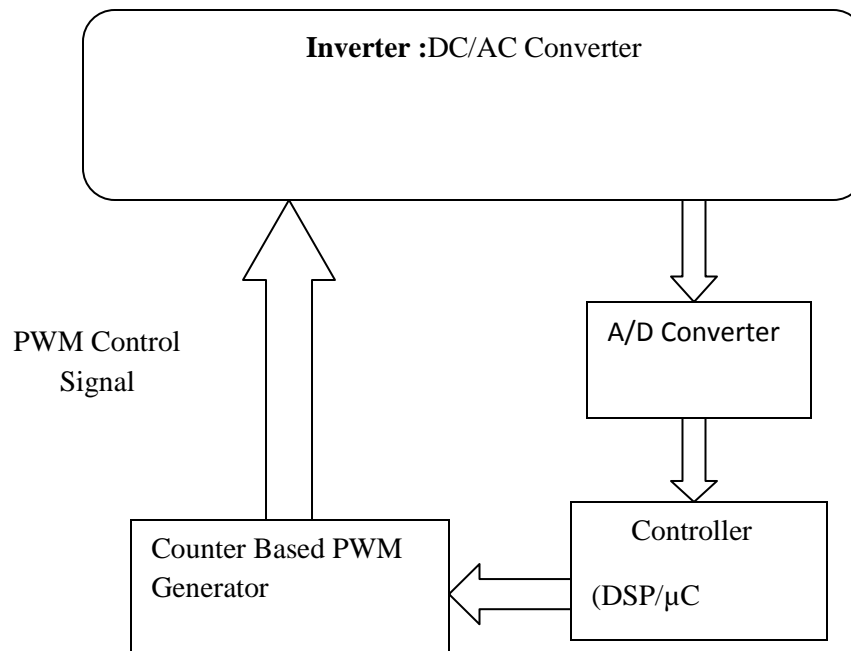
# Chapter 4

## Architecture of PWM Generator

### 4.1 Digital Techniques of PWM Generation

Many digital techniques are based on the use of counter and comparator based design. These digital techniques are easier to implement than analog techniques. Also they are immune to environmental noise and temperature change. Also they do not suffer component variation and switching losses.. For sophisticated control schemes it is desirable to use Digital PWM modulation scheme.

**Fig. 13** shows the general block diagram of Digital control scheme of Inverter.



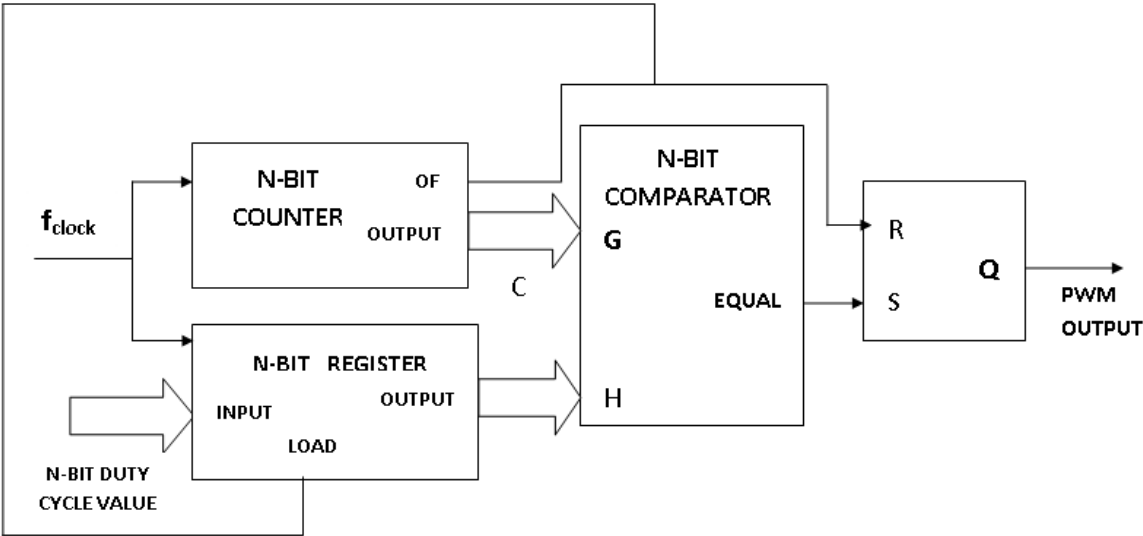
**Fig.13** General block diagram of Digital control scheme of Inverter.

There are many digital techniques available depending upon the arrangement and type of counter used but in this chapter three main PWM Generator topologies are discussed. These are (a) High frequency counter based PWM generator (b) Counter based PWM generator (c) Cascaded Counter based PWM generator.

### 4.2 High frequency counter based PWM Generator

This architecture was proposed by E.Koutroulis , A.Dollas and K.Kalaitzakis in [1]. According to this architecture there is a high speed N-bit free running counter whose output is compared with register output; which stores desired input duty cycle( N-bit value); with the help of comparator. The comparator output is set equal to 1 when both these values are equal. This comparator output is used to set RS latch .The overflow signal from counter is used to reset RS latch. The output of RS latch gives the desired PWM output.. This overflow signal is also used to load new N-bit duty cycle in Register.

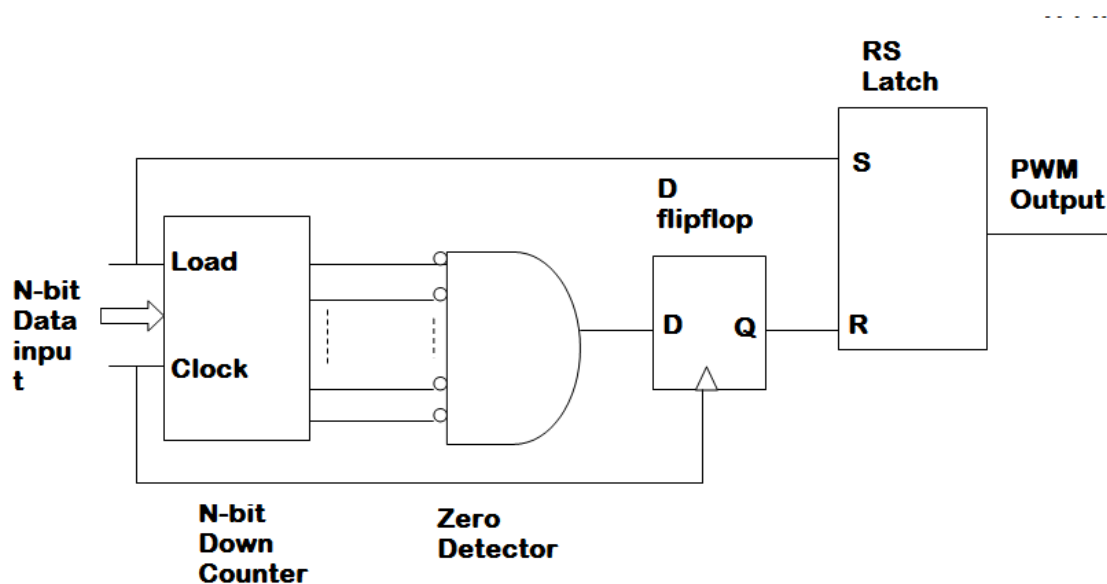
The advantage of these method is that it is used to generate High-frequency PWM output which is not possible in normal counter based approach. **Fig.14** shows the corresponding block diagram of this architecture.



**Fig.14** Architecture of PWM Generator proposed by E. Koutroulis et al. in [1]

### 4.3 Counter based PWM Generator

This PWM Generator architecture is used for low power switching supplies. This architecture was proposed by A.P.Dancy , R.Amirtharajah, and A.P.Chandrakasan in [8]. The architecture shown in **Fig. 15** is based on the principle that due to triggering of a counter by clock signal, clock is set equal to some multiple of switching frequency with help of a counter. The PWM output signal is set high before the clock signal and it remains high until it is reset after the counter value becomes equal to the duty cycle value. Fig.15 shows the above architecture defined.



**Fig.15** Architecture of Counter based PWM Generator proposed by A.P Dancy et al. in [8].

### 4.4 Cascaded Counter based PWM Generator Architecture

In this architecture two 4 bit counters are cascaded together to form 8 bits. Then we apply clock to both of counters. We connect the output of both these counters into the 8-bit A input pin of comparator. We provide 8-bit input duty cycle to B input pin of a counter . The output of comparator is '1' when both A pin value and B pin value matches together. This comparator output is used to reset the RS latch. Higher MSB counter overflow signal is used to set RS latch whenever it becomes activated. The output of RS latch is used to give PWM output.

Here we discussed various topologies of digital PWM Generator. In next chapter we will see what were the steps followed to design and download PWM Generator architecture in FPGA.

# **Chapter 5**

## **Design Procedure on FPGA**

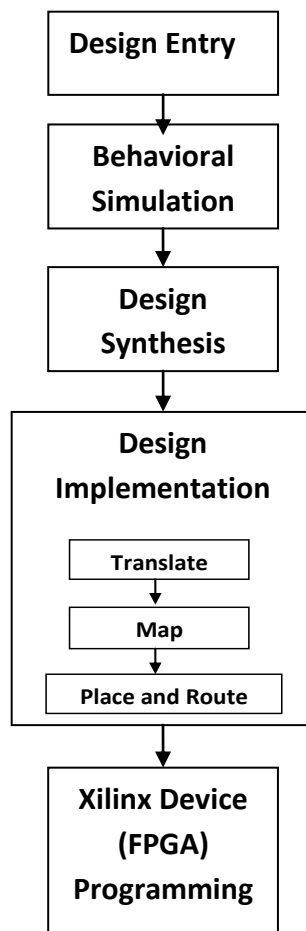
# Chapter 5

## Design Procedure on FPGA

### 5.1 FPGA basics

The Field Programmable Gate Array (FPGA), as the name suggests, is an array of logic cells (or modules) and interconnects, which can be reprogrammed depending upon the requirements of the user. We can design it and make changes in it whenever required. It provides instant manufacturing turnaround and negligible prototype costs which makes it suitable for embedded system design.

### 5.2 FPGA Design Flow



**Fig.16** FPGA Design Flow



**Fig.16** shows the sequence of steps followed when implementing PWM Generator design on FPGA. These steps are discussed in detail here.

### **4.2.1 Design Entry**

This is the first step of implementing a design on FPGA. In this step the VHDL (Very High Speed Integrated Chip Hardware Description Language) code of PWM Generator Architecture was written using software **Xilinx ISE 10.1**. Structural modeling was used for writing the code. After writing the code syntax check was performed on the code to see whether code was properly written using correct syntax

### **4.2.2 Behavioral Simulation**

The next step is behavioral simulation. This step verifies whether the design entered is functionally correct or not. This simulation is called RTL simulation. For this simulation VHDL Testbench was written for PWM Generator architecture and simulation was seen in Xilinx ISE Simulator. After it is verified it is functionally correct we move onto next step.

### **4.2.3 Design Synthesis**

The VHDL code of PWM Generator is then synthesized using Xilinx XST which is a part of Xilinx ISE software. There is a option of Synthesis in process tab of Xilinx ISE which performs the operation of synthesis .The synthesis process is used for optimizing the design architecture selected. The resulting netlist is saved to an NGC file. After design synthesis, synthesis report is generated which gives information about how many logic blocks are used and what is the device utilization of the design architecture synthesized. Synthesis basically maps the behavioral design to gate level design.

### **4.2.4 Design Implementation**

After design synthesis ,design implementation is done which comprises of following three steps

- (a) Translate
- (b) Map
- (c) Place and Route

Before translating the design, User Constrained file (UCF) is written to assign pin configuration of the FPGA to the PWM Generator I/O's. Once this is done Translate merges together this UCF file and netlist generated after synthesis into Xilinx design file

Mapping is done to fit the design into the available resources of target device i.e. FPGA. This is also important step of design.

Last step of Design Implementation is Placing and Routing which places the logic blocks of the design into FPGA and route them together so that they occupy minimum area and meet timing requirements. This operation produces NCD output file.

## 4.2.5 Xilinx Device (FPGA) Programming

There is a option of Generate programming file on the process tab of Xilinx ISE which converts the NCD file generated after routing to BIT file. It produces a bitstream for Xilinx Device (FPGA in this case ) configuration. This BIT file is used to program the FPGA.

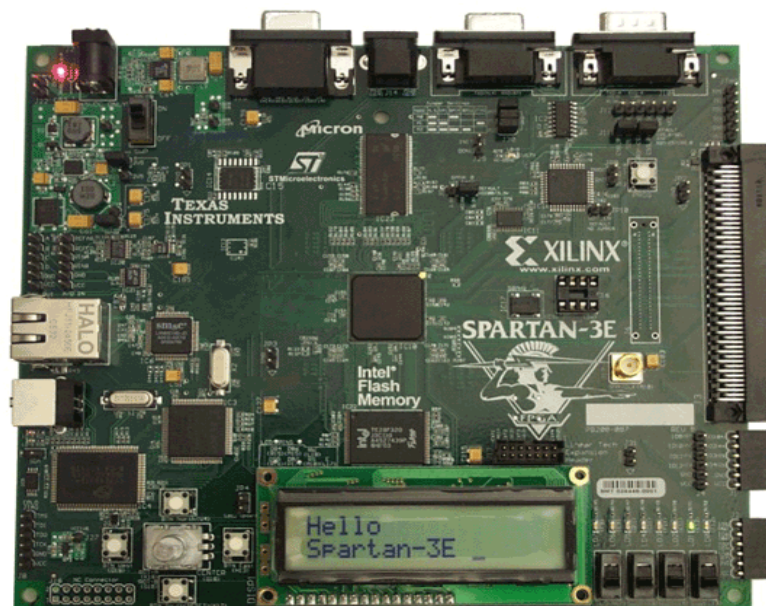
## 4.2.6 Configuring Target Device

There is option of Generate Target PROM/ACE on the process tab of Xilinx ISE which converts the BIT file to the PROM or ACE file. This PROM or ACE file can be downloaded directly into the FPGA's memory cells.

We have to make sure that FPGA is connected to the PC where we are developing this design.

After we download PROM or ACE file into the FPGA the FPGA is ready to be used as PWM Generator. We can give different input combination to see how the output of FPGA varies. This input can be given by switches whose pin number has been assigned to PWM generator input. The output can be seen by LED whose pin number has been assigned to PWM Generator output.

In next chapter different results and simulation obtained are shown for PWM generator. Given below is the figure of SPARTAN 3E FPGA used in this project.



**Fig.17** SPARTAN-3E Starter Kit (FPGA)

# Chapter 6

## Results and Simulation

# Chapter 6

## Results and Simulation

### 6.1 Results

Design steps discussed in last chapter were followed to implement design of PWM generator on FPGA. Different topologies of PWM Generator discussed in Chapter-4 were implemented. Given below are the various tables generated after synthesizing various PWM Generator topologies .

#### 6.1.1 Synthesis Report of High Frequency Counter based PWM Generator

The tables **Table 1** and **Table 2** describe the number of logic blocks used and Device utilization respectively of SPARTAN-3E FPGA when VHDL code for above PWM Generator architecture is written and synthesized using Xilinx XST of Xilinx ISE 10.1 software.

**Selected Device : 3s500efg320-4 (SPARTAN -3E FPGA)**

<b>Logic Blocks</b>	<b>Number of Logic Block used</b>
Registers	15
Flip-Flops	15
Latches	2
1-bit latch	1
8-bit latch	1

**Table.1** Macro Statistics of High Frequency Counter based PWM Generator architecture

Number of Slices	0 out of 4656	0% used
Number of IOs	10	
Number of bonded IOBs	1 out of 232	0% used

**Table.2** Device utilization of High frequency counter based PWM Generator architecture

## 6.1.2 Synthesis Report of Counter based PWM Generator

The tables, **Table 3** and **Table 4** describe the number of logic blocks used and Device utilization respectively of SPARTAN-3E FPGA when VHDL code for above PWM Generator architecture is written and synthesized using Xilinx XST of Xilinx ISE 10.1 software.

**Selected Device : 3s500efg320-4 (SPARTAN -3E FPGA)**

Logic Blocks	Number of Logic Block used
Counters	1
4-bit down counter	1
Registers	1
1-bit register	1
1-bit latch	1
8-bit latch	1

**Table.3** Macro Statistics of Counter based PWM Generator architecture

Number of Slices	8 out of 4656	0% used
Number of Slice Flip Flops	5 out of 9312	0% used
Number of GCLKs	1 out of 24	4% used
Number of 4 input LUTs	15 out of 9312	0% used
Number of bonded IOBs	7 out of 232	3% used

**Table.4** Device utilization of Counter based PWM Generator architecture

### 6.1.3 Synthesis Report of Cascaded Counter based PWM Generator

The tables **Table 5** and **Table 6** describe the number of logic blocks used and Device utilization respectively of SPARTAN-3E FPGA when VHDL code for above PWM Generator architecture is written and synthesized using Xilinx XST of Xilinx ISE 10.1 software.

Logic Blocks	Number of Logic Block used
Registers	8
Flip-Flops	8

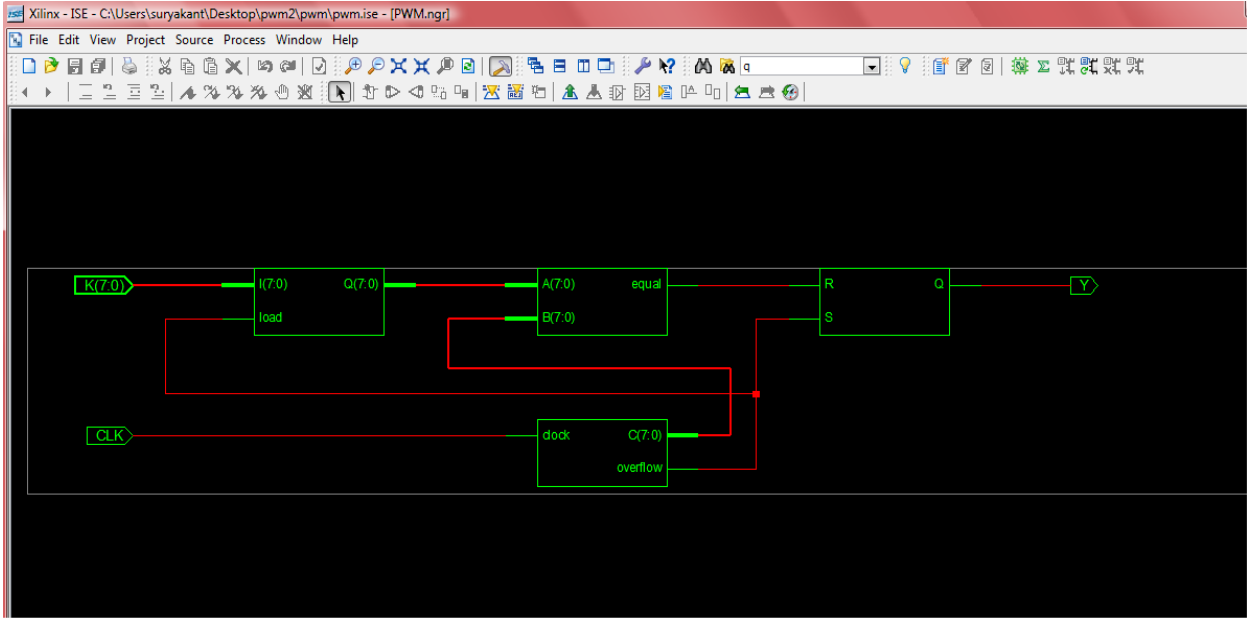
**Table.5** Macro Statistics of Cascaded Counter based PWM Generator architecture

Number of Slices	8 out of 4656	0% used
Number of Slice Flip Flops	9 out of 9312	0% used
Number of GCLKs	1 out of 24	4% used
Number of 4 input LUTs	13 out of 9312	0% used
Number of bonded IOBs	10 out of 232	4% used
Number of IO's	10	
IOB Flip Flops	1	

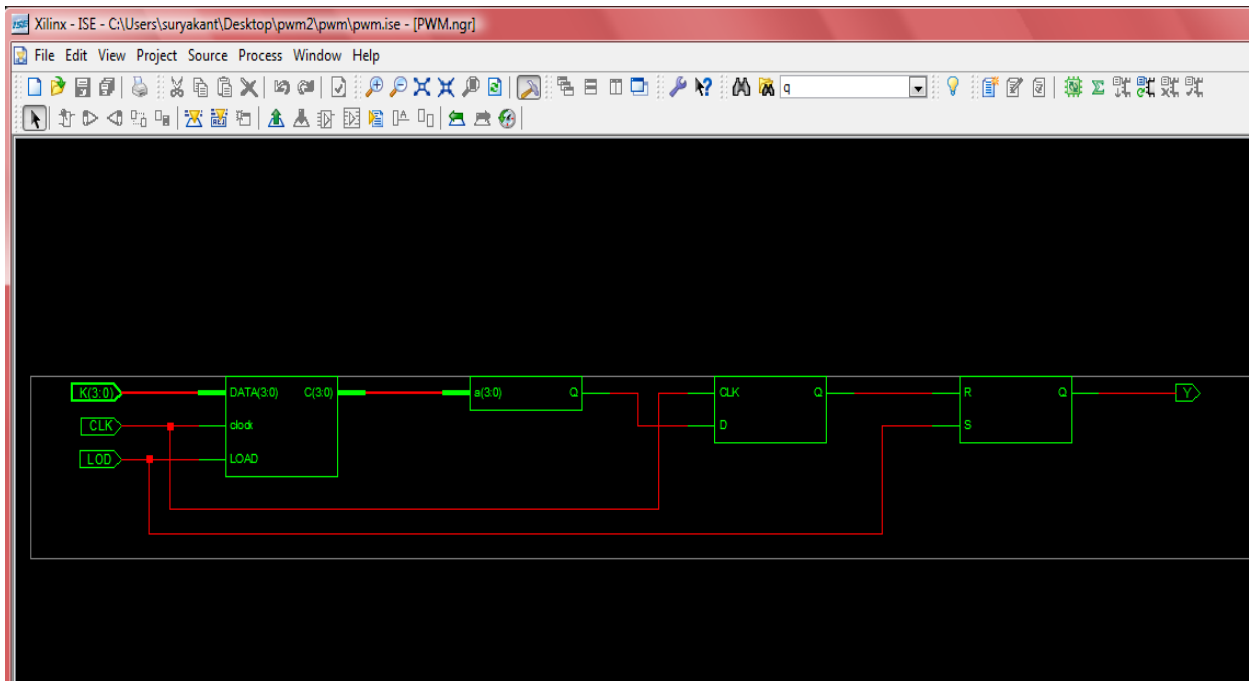
**Table.6** Device utilization of Cascaded Counter based PWM Generator architecture

## 6.2 RTL Schematic

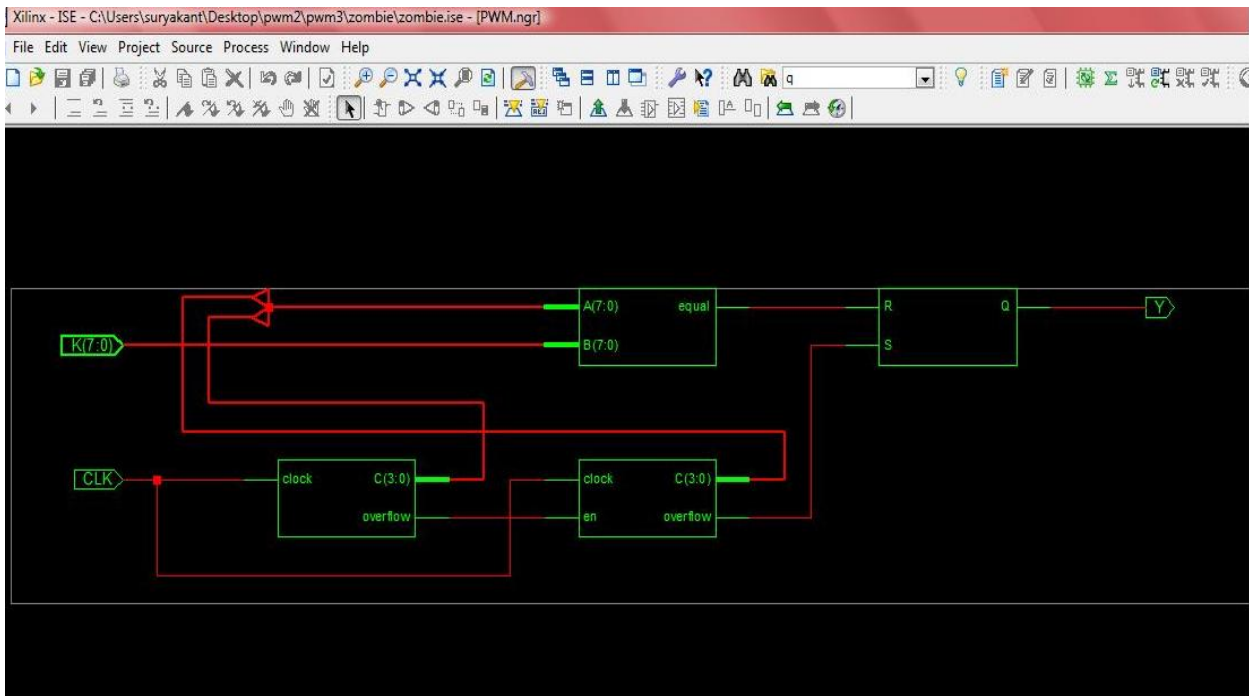
Next shown are the RTL schematic generated after synthesis of different PWM Generator topologies



**Fig.18** RTL Schematic of High Frequency Counter based PWM Generator



**Fig.19** RTL Schematic of Counter based PWM Generator



**Fig.20** RTL Schematic of Cascaded Counter based PWM Generator

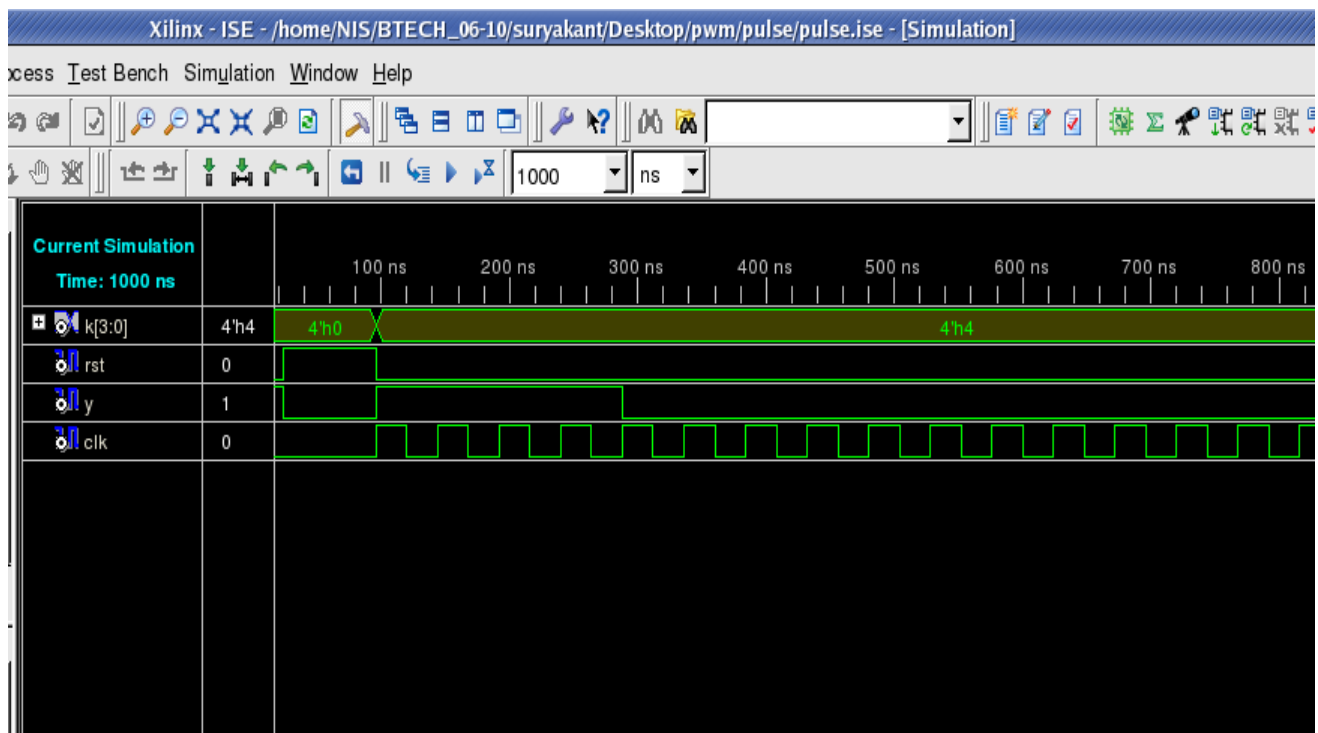


## 6.3 Simulation

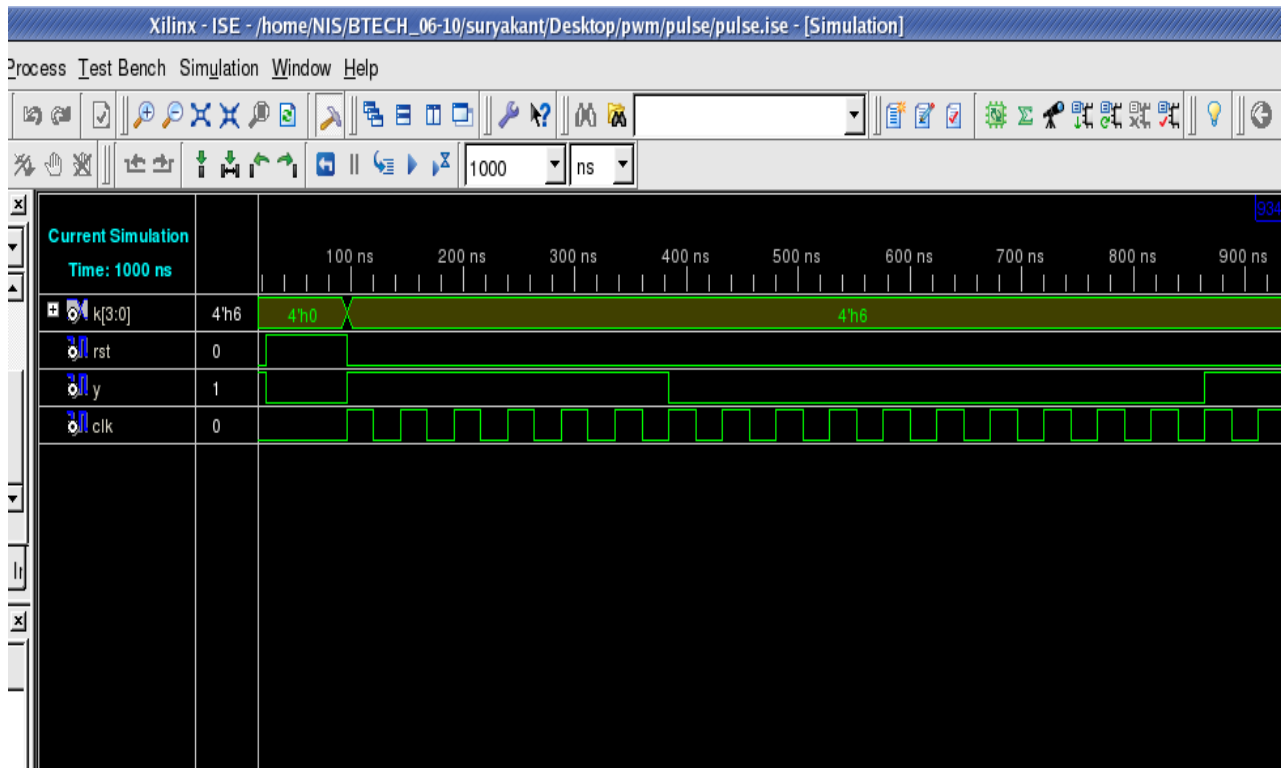
The Xilinx ISE simulator was used for functional verification of PWM Generator topologies. The resulting simulations for 4-bit data input configuration of PWM generation unit for different duty cycle is shown below in **Fig.21** to **Fig.23**. The duty cycle is calculated as below

$$\text{Duty Cycle} = \frac{\text{Integer Value of 4-bit } K \text{ word}}{16}$$

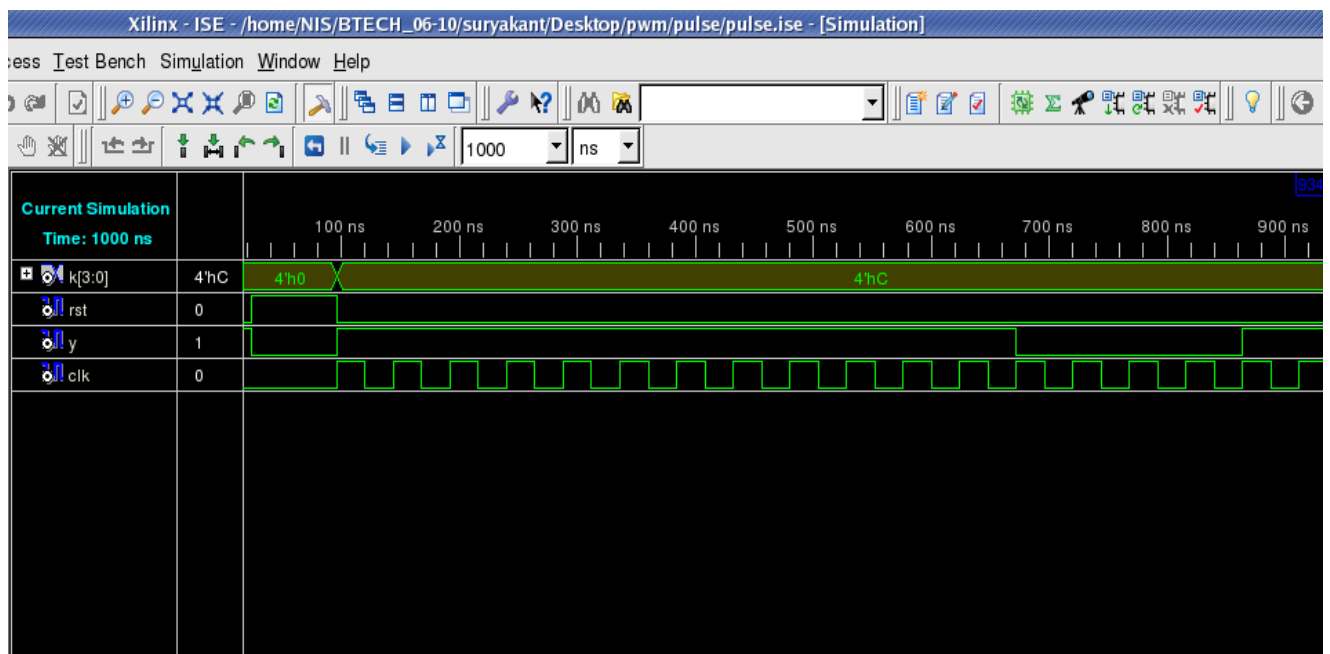
Where K is the 4-bit input to PWM Generator. Suppose the value of input data word to PWM Unit K='1001' than Duty Cycle =  $\frac{9}{16} = 0.5625$  OR 56.25% duty cycle.



**Fig .21** For input value K ='0100' or duty cycle=25%

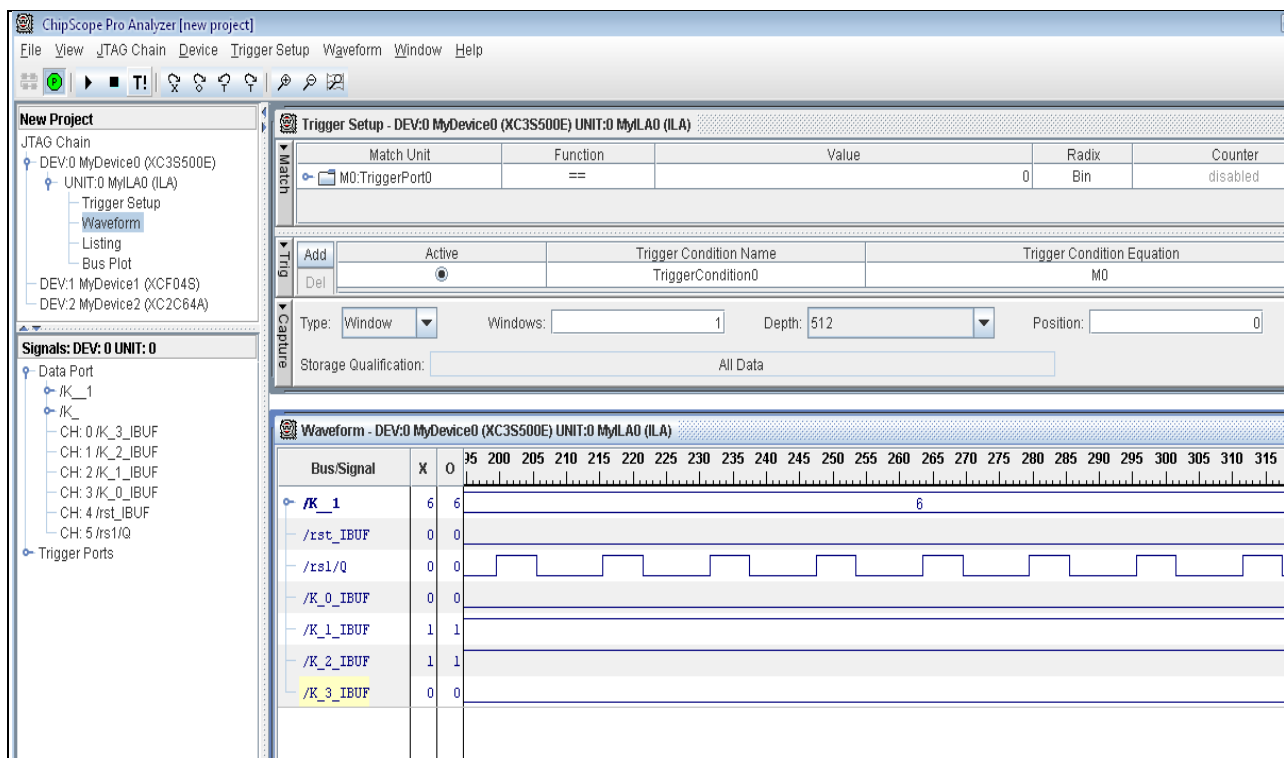


**Fig .22** For K='0110' or duty cycle =37.5%



**Fig .23** For K='1100' or duty cycle= 75%

After doing Placing and Routing the VHDL code of PWM Generator was downloaded into the Spartan 3E FPGA board and Real time debugging was done for the architecture. The input word K was given by 4 switches present in FPGA board. The clock for the architecture was provided by clock (C9) present in FPGA board. The frequency of the clock was 50 MHz. One LED was assigned to see PWM output. The software ChipScope Pro Analyser was used to see the real timing waveform of PWM output as we change the ‘K’ value on the FPGA board by different configuration of switches. There was also reset button which can stop the operation of PWM generator. The various timing waveforms for different value of input data integer ‘K’ as seen in ChipScope Pro Analyser is shown in **Fig .24** to **Fig.28**



**Fig .24** For K='1100' or duty cycle= 75%

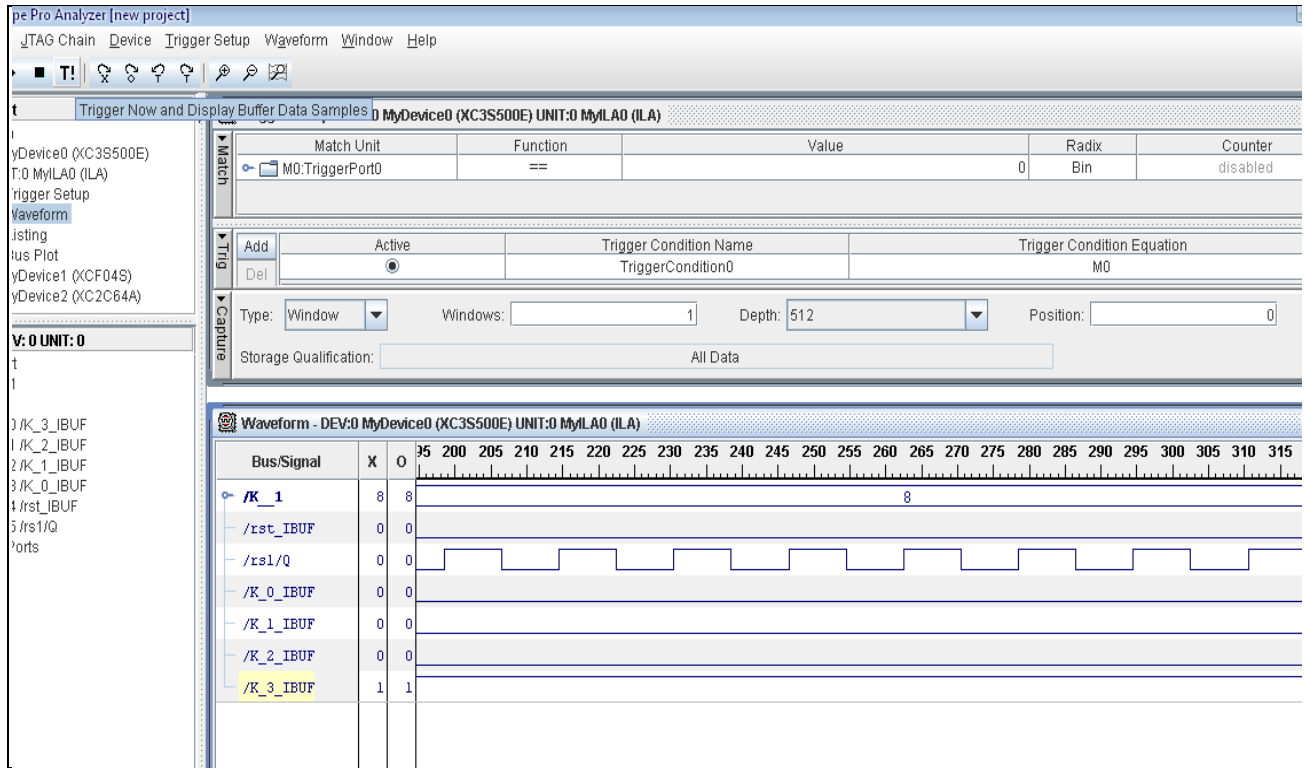


Fig 25 For K='1000' or duty cycle =50%

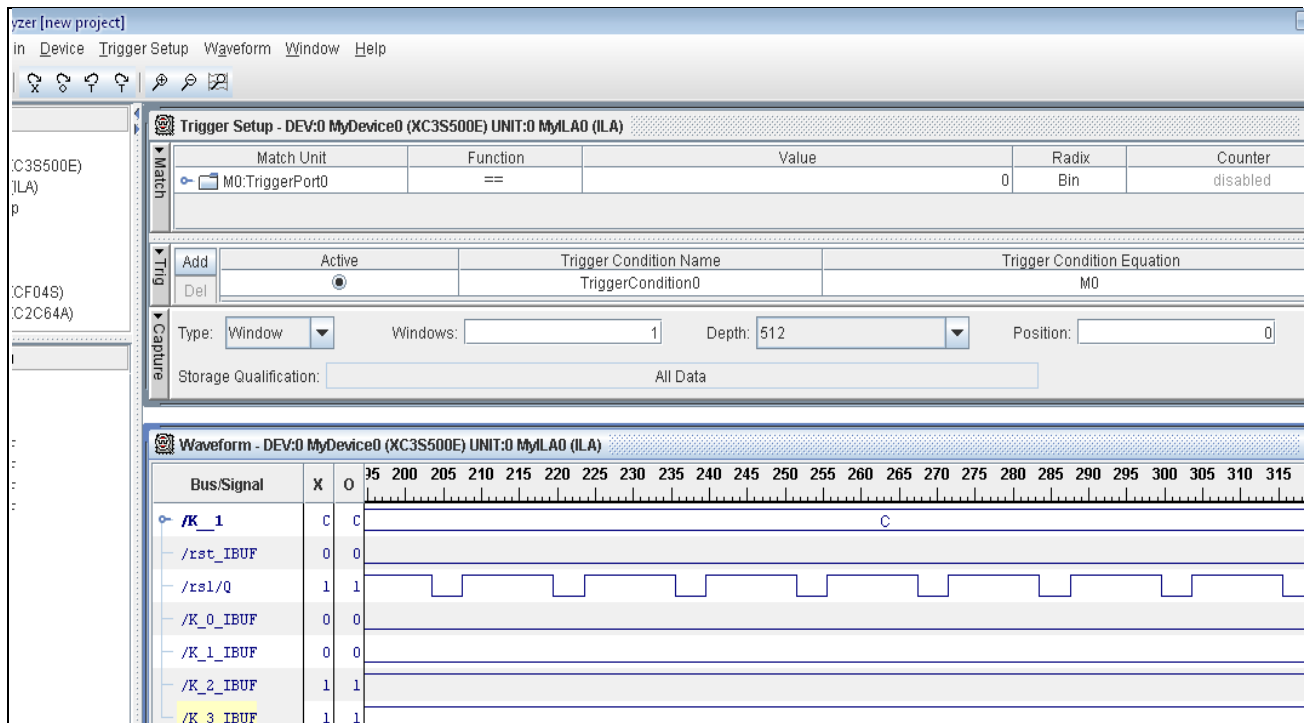
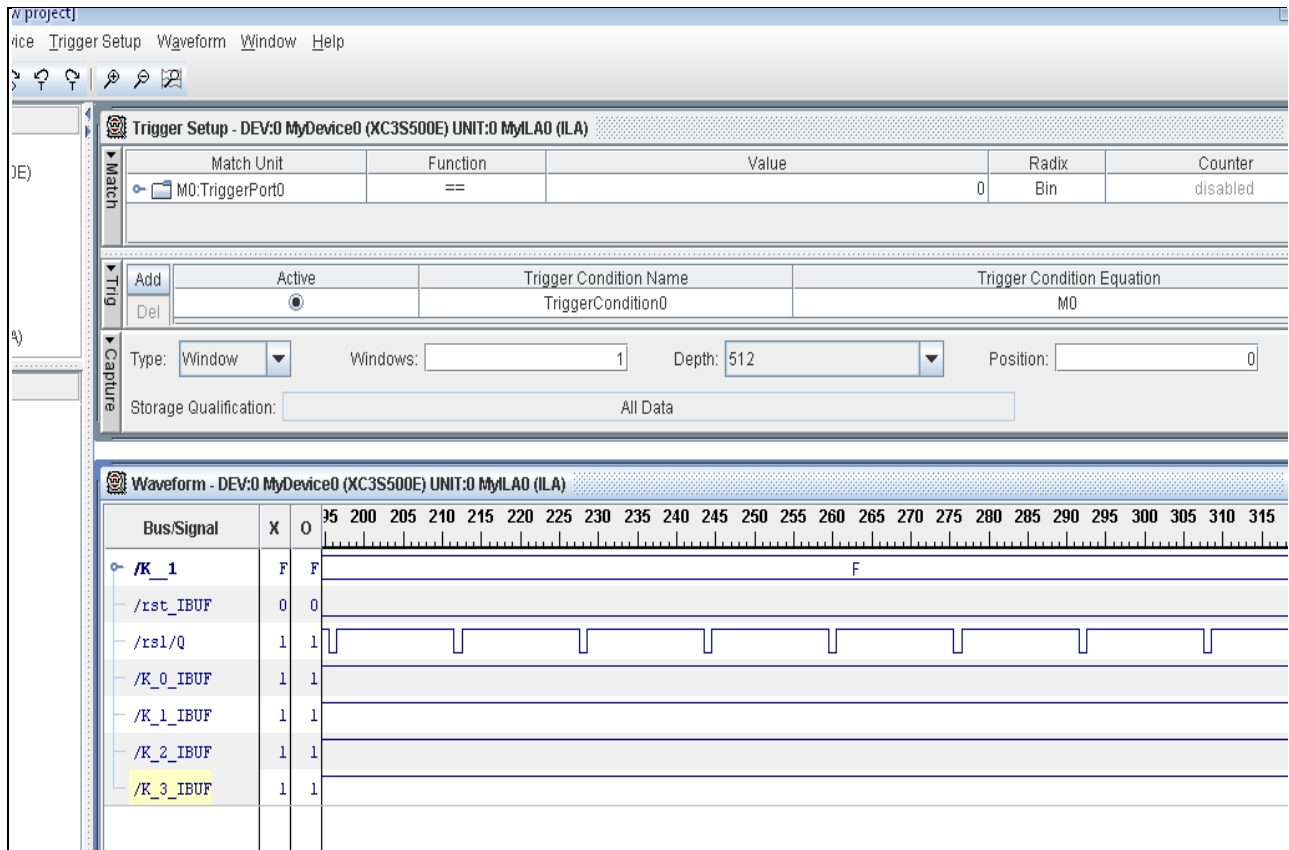
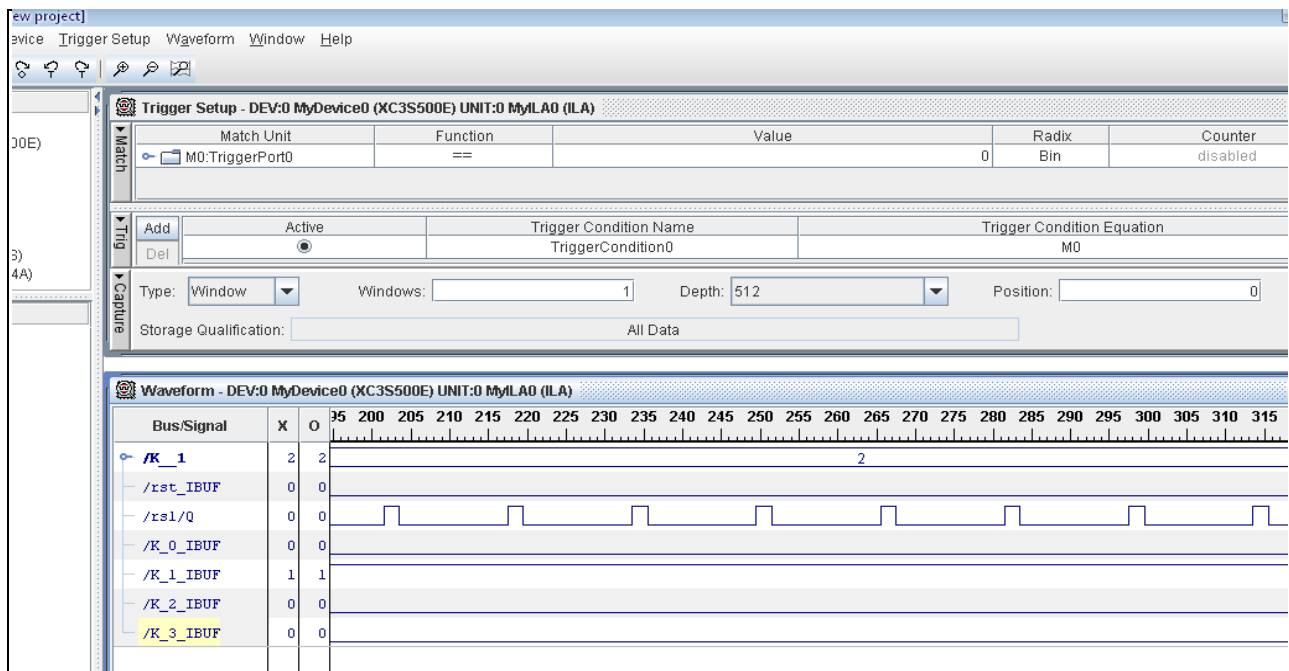


Fig.26 For K='1100' or duty cycle =75%



**Fig.27** For K='1111' or duty cycle =93.75%



**Fig.28** For K='0010' or duty cycle =12.5%

# **Chapter 7**

## **Conclusion and Future Work**

# Chapter 6

## Conclusion and Future Work

### 7.1 Conclusion

Various PWM Generator topologies were studied and they were synthesized using Xilinx ISE software. Device Utilization and No of Logic Blocks of each topology were noted down and compared with each other. It was seen that number of logic blocks used in FPGA for design of PWM Generator is minimum in case of Counter based PWM Generator and maximum in case of High frequency Counter based PWM generator. Device (FPGA) utilization is best for Cascaded Counter based PWM Generator and minimum for High frequency Counter based PWM generator. Functional verification was also performed on VHDL code of PWM Generator and simulation was observed as given in **Fig 21** to **Fig.23**. After downloading the design into the SPARTAN 3E FPGA board Real Time Debugging was performed on this architecture and results were observed in Chipscope Pro Analyzer. These timing waveforms are shown in **Fig.24** to **Fig.28**

### 7.2 Future Work

PWM Generator architecture which was downloaded onto FPGA can be used to control Gate signal of Power Switches of Inverter. By this way we can be able to control ON and OFF time of Inverter. Hence we can use this architecture for controlling Inverter which produces AC from DC source. Hence we can implement PWM Inverter from this FPGA based PWM Generator.

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