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Finite-Time Position and Velocity Estimation Adapted to Noisy Biased Acceleration Measurements from Periodic Motion

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The present work focuses on the problem of velocity and position estimation. A solution is presented for a class of oscillating systems in which position, velocity and acceleration are zero mean signals. The proposed scheme considers that the dynamic model of the system is unknown. Only noisy acceleration measurements, that may be contaminated by zero mean noise and constant bias, are considered to be available. The proposal uses the periodic nature of the signals obtaining *finite-time estimations* while tackling integration drift accumulation.

Keywords: Finite-time, estimation adaptation, periodic signals

1. Introduction

The estimation of position/velocity through the integration of measurements from inertial sensors inherently causes errors to grow with time, commonly known as integration drift. In fact, the process of zero-mean noise integration leads to an output that increases with integration time, even in the case that the accelerometer is at rest (Thong, Woolfson, Crowe, Hayes-Gill, & Jones, 2004). For this reason, *velocity/position estimation based on inertial sensors always requires some additional source of information* providing for the corresponding correct signal value at some time instants. In general this information is updated with intervals significantly larger than the acceleration measurement rate. For instance, in inertial navigation systems (INS) aiding sensors are applied. An INS consists of a triad of orthogonal accelerometers and rate gyroscopes that are used to calculate both the translational and rotational movements of an aircraft over time (Rogers, 2003), a process know as dead reckoning. MEMS technology has made low-cost INS available. However, sensor biases coupled with the time integration, needed for dead reckoning, causes that the solution quickly diverges. That is why externally referenced sensors, such as the global positioning system, are required to regulate the error growth (Grewal, Weill, & Andrew, 2007); Kalman filters are commonly used in order to fuse the two sources of information.

An alternative to the use of aiding sensors is the use of prior-knowledge of motion. In (Liu, Inoue, & Shibata, 2009), for instance, the drift accumulation from gyroscope based orientation estimation, in human walking experiments, is compensated by detecting different gait phases with the aid of accelerometers and the knowledge of the human walking motion. In (Schepers, Asseldonk, Baten, & Veltnik, 2010), foot placement for several strides is estimated. The drift is reduced applying the knowledge of the cyclic walking movement that leads to information on initial and final conditions. Another example of application of prior knowledge of motion is sensing and identification of physiological tremor. Tremor is defined as any involuntary, approximately rhythmic, and roughly sinusoidal movement [19]. In (Riviere, Ang, & Khosla, 2003), a filtering algorithm based on truncated Fourier series, so called weighted frequency Fourier linear combiner (WFLC), is applied for tremor identification. The WFLC models the tremor as a sinusoid and tracks its modulation in frequency, phase, and amplitude. The WFLC is suitable for signals with single dominant frequency, this

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restriction is relaxed in (Veluvolu & Ang, 2011) by means of a band limited multiple Fourier linear combiner (BFMLC) which can be applied to band limited signals consisting of multiple components.

In the present work the prior-knowledge used in the velocity/position estimation solution proposal is periodicity of the motion. Therefore, a period estimation algorithm is fundamental considering that the oscillation period is in general not available. In (Spence & Clarke, 2000), a method for estimating the period and amplitude profile of a signal with multiple periodicities is proposed based on the results reported in (Feder, 1993) which can be seen as an implementation of the so called expectation-maximization algorithm (Dempster, Laird, & Rubin, 1977). In (Spence & Clarke, 2000), a likelihood computation is put in terms of a least mean square (LMS) residual power expression which is evaluated for different test periods. The estimated period is chosen as the one minimising the LMS residual power. In the field of music and speech analysis the pitch period (fundamental frequency) estimation is very important. For short-time estimation of pitch period, the algorithms that find the average fundamental frequency using autocorrelation or linear prediction techniques are the most commonly used (Guangyu & Shize, 2009), (Peeters, 2006), (Bernardin, 2006), (Min, Yingchun, & Zhaohui, 2005).

We propose a period estimation algorithm that takes advantage on the zero mean nature of the signals. This scheme is not restricted to sinusoidal dominant frequency, do not requires a limited frequency band known a priori and is computationally simple. The period information is then used for the solution of velocity/position estimation based only in inertial information with no aiding sensors nor explicit dynamical model. Summarizing, we propose a method that allows to adjust the velocity/position estimations for the case in which the acceleration measurements, which could be biased and contaminated by a zero mean noise, correspond to an oscillating periodical motion with zero mean velocity and position. This proposal compensates the drift using information of the oscillating motion period and can achieve finite-time estimation, for the zero noise and no integration error case.

The rest of the paper is organized as follows, in the next section the problem statement is presented. In section III the possibility of an asymptotic velociy/position estimation is discussed and the idea of using the oscillation period is introduced. Section IV is devoted to the period estimation proposal. In Section V the scheme for finite-time position/velocity estimation is presented. Finally, Section VI presents simulation results.

2. Problem statement

Consider a system with the following acceleration dynamics

$$\ddot{x}(t) = f(x, \dot{x}) \tag{1}$$

and measurement model

$$\Psi(t) = \ddot{x}(t) + \eta + \beta, \tag{2}$$

where η is a bounded zero-mean noise and β is a constant bias. $x \in \mathbb{R}$ and $f(x, \dot{x})$ is an unknown nonlinear function. Mechanical systems are among a wide spectrum of physical systems encompassed by (1). It is assumed that the position x, and hence \dot{x}, \ddot{x} , have a periodic or quasi-periodic behaviour, more precisely it is considered that position is oscillating between two bounds. There exist two constants such that $B_m \leq x^{(i)}(t) \leq B_M$, $\forall t > 0, i = 0, 1, 2$.

The objective of this work is to present an alternative solution for velocity and position estimation in the case of only acceleration measurement for a class of periodical signals, i.e., $x^{(i)}(t) = x^{(i)}(t + T_f)$ for some finite T_f .

3. Velocity estimation

3.1 Velocity estimation through averaging

In the following, the super index $<\!i\!>$ is used to indicate the $i-{\rm th}$ time integral of a function $y:R\to R$:

$$y^{}(t) = \int_0^t \dots \int_0^{\tau_2} y(\tau_1) \, d\tau_1 \dots d\tau_i.$$
(3)

While $y^{\langle i \rangle}(t_0, t)$, with $t \geq t_0$, indicates the interval of evaluation:

$$y^{}(t_0,t) = \int_{t_0}^t \dots \int_{t_0}^{\tau_2} y(\tau_1) \, d\tau_1 \dots d\tau_i.$$
(4)

Define

$$\eta_{\beta}^{\langle i \rangle}(t) = \eta^{\langle i \rangle}(t) + \beta^{\langle i \rangle}(t).$$
(5)

Now, in order to obtain velocity estimation we proceed to integrate $\Psi(t)$ directly from the measurement (2). Hence,

$$\Psi^{<1>}(t) = \dot{x}(t) - \dot{x}(0) + \eta_{\beta}^{<1>}(t), \tag{6}$$

$$\Psi^{<2>}(t) = x(t) - x(0) - \dot{x}(0) \cdot t + \eta_{\beta}^{<2>}(t), \tag{7}$$

and dividing (7) by t we obtain

$$\frac{\Psi^{<2>}(t)}{t} = -\dot{x}(0) + \frac{x(t) - x(0) + \eta_{\beta}^{<2>}(t)}{t}.$$
(8)

Then, define the velocity estimation, $\dot{x}(t)$, as follows:

$$\hat{\dot{x}}(t) = \Psi^{<1>}(t) - \frac{\Psi^{<2>}(t)}{t}$$
$$= \dot{x}(t) + \eta_{\beta}^{<1>}(t) - \frac{x(t) - x(0) + \eta_{\beta}^{<2>}(t)}{t}.$$
(9)

Let us recall that the position is bounded. If additionally, it holds that $\forall t \geq 0$, $|\eta_{\beta}^{\leq i>}(t)| \leq \bar{\eta}_{\beta}^{\leq i>}$ for some constant $0 \leq \bar{\eta}_{\beta}^{\leq i>}$ (i.e. $\beta = 0$), the next limit can be obtained

$$\lim_{t \to \infty} \left\{ \frac{x(t) - x(0) + \bar{\eta}_{\beta}^{<2>}}{t} \right\} = 0.$$
 (10)

Therefore, for t "sufficiently" large

$$\hat{x}(t) \simeq \dot{x}(t) + \eta_{\beta}^{<1>}(t).$$
 (11)

A similar averaging approach can be applied for position estimation.

Note that the assumption $\beta = 0$ is a limitation of the described averaging procedure and it is not required for the method presented in this work. The above estimation is *asymptotic* without a possibility to regulate the speed of convergence. The amount of time, which is necessary to obtain a reliable estimation, will depend on the magnitude of the bounds of position and the constant bound for the noise integrals (a very strong and even unrealistic assumption when any measurement bias is present). In addition, integration over an infinite time interval can be technically difficult without a significant accumulation of computation and, as a consequence, estimation errors even in the case of the known initial condition $\dot{x}(0)$.

In the present paper a finite time solution is proposed. In fact, a possibility to find a solution in finite time is closely related with period knowledge. This statement is further discussed in the following subsection.

3.2 Finite time velocity estimation

Assume that \ddot{x}, \dot{x} and x are unbiased periodic functions with period T_f (their integrals over a period are equal to zero). In this paper it is considered that a bias only can be present in the acceleration measurement. In fact, if \dot{x} or \ddot{x} have a bias, then x would not be periodical and bounded. Moreover,

a constant bias in position cannot be identified from acceleration or velocity measurements because it is not reflected in such measurements. Now, consider the equations (6) and (7) evaluated at T_f

$$\Psi^{<1>}(T_f) = \dot{x}(T_f) - \dot{x}(0) + \eta_{\beta}^{<1>}(T_f), \tag{12}$$

$$\Psi^{<2>}(T_f) = x(T_f) - x(0) - \dot{x}(0) \cdot T_f + \eta_{\beta}^{<2>}(T_f),$$
(13)

where $x(T_f) = x(0)$ and $\dot{x}(T_f) = \dot{x}(0)$. Hence, an estimation $\hat{x}(0)$ of $\dot{x}(0)$ can be obtained at $t = T_f$:

$$\hat{x}(0) = -\frac{\Psi^{<2>}(T_f)}{T_f} \tag{14}$$

$$= \dot{x}(0) - \frac{\eta_{\beta}^{<2>}(T_f)}{T_f}.$$
(15)

Then, if period is known, an estimation of velocity can be derived using equations (12) and (15):

$$\hat{\dot{x}}(t) = \Psi^{<1>}(t) + \hat{\dot{x}}(0).$$
 (16)

Next, the same approach can be repeated for position estimation. Consider the following expression for

$$\Psi^{<3>}(T_f) = x^{<1>}(T_f) - x^{<1>}(0) - \dot{x}(0) \cdot T_f - \frac{1}{2}\dot{\hat{x}}(0) \cdot T_f^2 + \eta_{\beta}^{<3>}(T_f),$$

thereby, an estimation of the position initial condition can be obtained as

$$\hat{x}(0) = -\frac{\Psi^{<3>}(T_f)}{T_f} + \frac{1}{2}\hat{x}(0)T_f$$

Remark 1: Note that the estimation would be exact in the absence of noise (a constant measurement bias can be identified and compensated provided the period is available). However, in the presence of noise, a finite-time estimation whose accuracy will depend on the noise magnitude, assumed to be bounded, can be achieved by using the value of the period.

3.2.1 Importance of period knowledge

Note that x(t) = x(0) may hold for different time instants, but the correct estimation of velocity can only be ensured for $t = \kappa T_f$ with $\kappa \in \mathbb{N}$. It implies that estimation of the initial condition, $\dot{x}(0)$, without knowledge of the period may lead to big errors. To illustrate this idea, consider the next expression for acceleration

$$\ddot{x}(t) = \sin(1.5t) + \sin(4.2t),\tag{17}$$

which is plotted with its corresponding velocity and position signals in Fig. 1. Suppose that measurements starts at t = A. Note that there is a subperiod, between times A and B, for both acceleration and velocity. Given that only the acceleration is available, this zero mean subperiod can be wrongly considered as the real period causing the error term (x(B) - x(A))/(B - A) to be present in the estimation $\hat{x}(0)$, see Eq. (15). A solution to the problem of period estimation is presented in the following Section 4.

4. Period Identification

The algorithm for period estimation proposed in this paper is based on the features of the phase portrait (\dot{x}, \ddot{x}) . To introduce the idea, consider the phase portrait (\dot{x}, \ddot{x}) of (17) depicted in Fig. 2. It is clear that after one period the trajectories pass through each point at least once. Nevertheless, there are a number of points and furthermore trajectory segments, where the trajectory passes more than once over a cycle. If we can select a point which is *reached* only once over a whole period, then an estimation of the period can be obtained by detecting the time elapsed between two consecutive

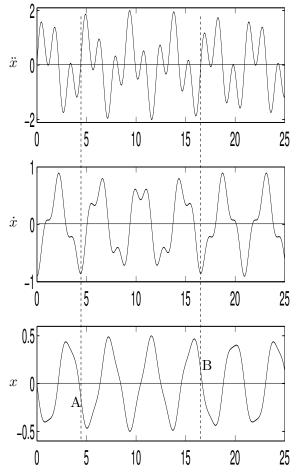


Figure 1. Acceleration, Velocity and Position versus time (sec).

passings over such point. The question is how to properly select this point. Recall that the real phase portrait is not available because there is no velocity measurement. From Equation (6) it can be seen that the velocity estimation, \dot{x} , is shifted by its initial condition. However, it is possible to obtain a period estimation under the next assumptions:

- S1. \dot{x} and \ddot{x} are zero-mean periodical signals with unknown period T_f .
- S2. On the period T_f , \dot{x} has a unique global extremum, either a maximum or a minimum, represented by \dot{x}_M .

Remark 2: Note that, as a consequence of the bounded nature of noise and bias, the time integrals $\eta_{\beta}^{\langle i \rangle}(t)$, i = 1, 2 are bounded $\forall t \in [0, T_f]$.

Remark 3: The Assumption S2 restricts the function \dot{x} to have only a single occurrence of the respective global extremum over one period. In (17), for instance, the maximum for \dot{x} does not fulfills S2.

In (Spence & Clarke, 2000) an equivalent of assumption S2 is also considered while S1 is not required. However, the period estimation uses optimization of a criterion which involves either residual or signal power over a set of data using a set of trial periods, as they mention the basis of its time domain estimator is 'trial and error' using a number of test periods. While not requiring S1 or S2, methods like those in (Peeters, 2006) and (Guangyu & Shize, 2009) are devoted to identify a sinusoidal dominant frequency and the results in (Veluvolu & Ang, 2011) assumes band limited signals consisting of multiple sinusoidal components. The scheme in the present work do not requires that sinusoidal composition paradigm. In this paper it is proposed the detection of two consecutive passings trough an extremum \dot{x}_M which comply with S2. The successful \dot{x}_M detection involves

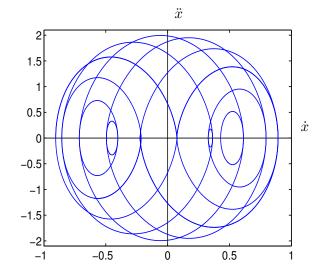


Figure 2. Phase portrait (\dot{x}, \ddot{x}) for (17).

estimation of the drift with respect to the integration time elapsed. This drift estimation is addressed next.

4.1 Estimation of an upper bound for $\eta_{\beta}^{<1>}$

In general, acceleration is a noisy measurement. Additionally, the accuracy of integration is limited by sampling time. The aforementioned factors lead to integration drift. Obviously, such a drift is worsen by the presence of measurement bias. In (Estrada, Efimov, & Perruquetti, 2014), the problem of position and velocity estimation was addressed assuming *unbiased* acceleration measurements and knowledge of a bound for the integration error. In the present proposal the assumptions in (Estrada, Efimov, & Perruquetti, 2014) are relaxed allowing the presence of biased measurements and dropping the requirement of giving an a priori bound for the integration drift. That is achieved thanks to the introduction of a time varying estimation of the integration drift, based on the more realistic assumptions of having bounds for the noise and bias magnitudes. The expressions used for obtaining such an estimation are presented here. Further details regarding derivation of those expressions are included in the Appendix.

Define $\bar{\eta}^{\langle i \rangle}(t_0, t)$ and $\bar{\beta}^{\langle i \rangle}(t_0, t)$ as upper bounds for $\eta^{\langle i \rangle}(t_0, t)$ and $\beta^{\langle i \rangle}(t_0, t)$, on the interval $t \in [t_0, t]$, respectively (i = 0, 1, 2). Recall that $\bar{\eta}^{\langle 0 \rangle}$ and $\bar{\beta}^{\langle 0 \rangle}$, which do not depend on t, are assumed to be known. The bound of the drift due to the contribution of the zero mean noise integration, over a finite interval $[t_0, t]$, $t_0 \leq t$, is computed as follows

$$\bar{\eta}^{<1>}(t_0, t) = 4\bar{\eta}^{<0>} \frac{\sqrt{t - t_0 + f_s^{-1}}}{\sqrt{f_s}},\tag{18}$$

where f_s is the sampling frequency, t_0 is the initial time of integration. Expression (18) is an estimation of the drift due to integration of noise, see the Appendix section for details. The upper bound for drift due to bias integration is obtained straightforward by integrating its constant upper bound. Therefore,

$$\bar{\beta}^{<1>}(t_0, t) = \bar{\beta}^{<0>} \cdot (t - t_0).$$
⁽¹⁹⁾

Finally, the upper bound of the combined drift is

$$\bar{\eta}_{\beta}^{<1>}(t_0,t) = \bar{\eta}^{<1>}(t_0,t) + \bar{\beta}^{<1>}(t_0,t)$$
(20)

4.2Period Identification Algorithm

Recall that \dot{x}_M denotes the global extremum in velocity. The algorithm for period estimation should perform the following tasks.

- (1) To ensure the detection, once each period, of the extremum \dot{x}_M .
 - Using the $\bar{\eta}_{\beta}^{<1>}(t_0, t)$ bound, discriminate \dot{x}_M from any other local extremum.
 - In addition to possible extrema outside of the $\bar{\eta}_{\beta}^{<1>}(t_0, t)$ -vicinity, there may be local ones, not present on the real \dot{x} but induced by the integration of noise, inside such a vicinity.
 - The above implies that it is necessary to wait until the estimated \dot{x} exits some vicinity of the last \dot{x}_M candidate in order to decide that this candidate can be considered as a new occurrence of \dot{x}_M .
- (2) The integral of acceleration, on which \dot{x} estimation and \dot{x}_M detection is based, should be restarted once each period in order to tackle drift accumulation.
 - This reset can be performed after a new \dot{x}_M occurrence has been confirmed.

If the above conditions are fulfilled then an estimation of the period can be made as the difference between two consecutive detected occurrences of \dot{x}_M . Prior to present the algorithm some definitions are introduced:

- α is a parameter to be selected as 1 or -1 in order to detect a global maximum, or a global minimum respectively.
- \dot{x}_{M^-} is the extremum, different from \dot{x}_M , whose amplitude is the closest to \dot{x}_M . $\Psi^{<1>}(t_a, t_b)$ indicates the integral of Ψ from $t = t_a$ to $t = t_b$.
- $\delta(t)$ is a time varying parameter upper bounding the drift due to noise and bias integration.

Next we present the proposed period detection algorithm:

Algorithm A1. Initialization of variables

 $\alpha = \pm 1, t_0 = 0, t_1 = 0, fl = 0,$ $\delta(t) = \bar{\eta}_{\beta}^{<1>}(t_0, t)$ $\Psi_M^{<1>} = 0,$

C0. If
$$\alpha \Psi^{<1>}(t_0, t) > \alpha \Psi_M^{<1>} - \delta(t) fl$$
 then

$$\Psi_M^{<1>} = \Psi^{<1>}(t_0, t),$$

 $t_1 = t,$ (21)
 $fl = 0.$

C1. If $\alpha \Psi^{<1>}(t_1,t) < -2\delta(t)$ and fl = 0 then

$$\Psi_{M}^{<1>} =0,
\hat{T}_{f} =t_{1} - t_{0},
t_{0} =t_{1},
fl =1,$$
(22)

Remark 4: δ is evaluated according to (20) and updated every sampling time.

We state the next claims:

Lemma 1: Each occurrence of $\dot{x}(T_M^i) = \dot{x}_M$ is followed by only one fulfillment of condition C1 with a value for t_1 such that

$$\dot{x}(t_1) \in [\dot{x}_M - \bar{\eta}_\beta^{<1>}(|T_M^i - t_1|), \dot{x}_M].$$
(23)

Proposition 1: If assumptions S1-S2 are fulfilled then, an estimation of the real period, T_f , can be obtained by

means of the algorithm A1, with $\delta(t) = \bar{\eta}_{\beta}^{<1>}(t_0, t)$, provided that

$$\frac{1}{2}|\dot{x}_M - \dot{x}_{M^-}| > \bar{\eta}_{\beta}^{<1>}(T_f).$$
(24)

Assume that $\alpha = 1$ is selected. Then, the algorithm A1 proceeds in the following manner:

- The condition **C0** at the beginning is $\alpha \Psi^{<1>}(0,t) > 0$, then it is fulfilled for any positive quantity of the integral $\alpha \Psi^{<1>}(0,t)$.
- After the first fulfillment of **C0** the assignment $\Psi_M^{<1>} = \Psi^{<1>}(0,t)$ is done in order to save the new maximum and the time of its occurrence is saved on t_1 .
- The assignment $t_1 = t$, at each new maximum, implies that the integral $\Psi^{<1>}(t_1, t)$ is restarted.
- Due to fl = 1 assignment inside C1, once the latter condition is verified it will not be triggered again before a new C0 verification.
- It can be seen that the condition C1 contains the criteria which decides when a detected maximum is considered as an occurrence of the global maximum x_M. In fact, before the completion of one period, it is possible that a local maximum is considered as x_M. Nevertheless, as it is shown in the proof of Lemma 1, this is anymore the case after the first occurrence of x_M.
 Finally, note that when condition C1 is fulfilled, the integral Ψ^{<1>}(t₀, t) contains the integrated acceleration
- Finally, note that when condition C1 is fulfilled, the integral $\Psi^{<1>}(t_0, t)$ contains the integrated acceleration with an initial time corresponding to the last maximum. Additionally, the condition C0 takes the form $\Psi^{<1>}(t_0, t) > -\delta(t)$. This change in the form of condition C0 is needed in order to detect the occurrence of next maximum taking into account a new drift $\eta_{\beta}^{<1>}(t_0, t) < \delta(t)$.

4.3 Period estimation simulations

The algorithm **A1** was applied to the identification of the period, $T_f = 20.944$, of (17) contaminated only by a zero mean noise, η , fulfilling that $|\eta| \leq 0.024$. The integration step, the simulation time for producing the signals was fixed to 1×10^{-4} s, while the sampling period, the time at which measurements are made, was fixed to 6×10^{-3} s. The parameters $\eta^{<0>} = 0.024$ and $\beta^{<0>} = 0$ are used for $\delta(t)$ evaluation. $\alpha = -1$ for a global minimum detection was chosen. Indeed, the maximum of this signal appears twice over a period, failing to fulfill Assumption S2. The phase portrait of the signals ($\Psi^{<1>}, \Psi$) is depicted in Fig. 3 whereas in Fig. 4 the period estimation is plotted.

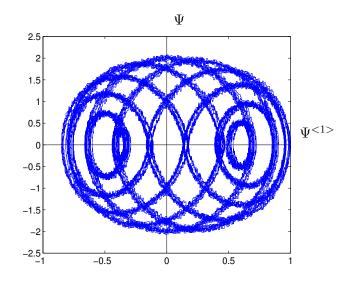


Figure 3. Phase portrait $(\Psi^{<1>}, \Psi)$ for noisy measurements $(|\eta| < 0.024)$ of (17).

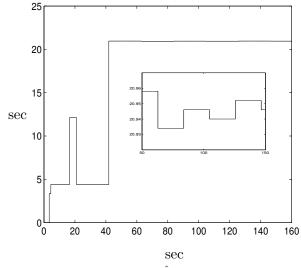


Figure 4. Estimated period \hat{T}_f for signal (17) with noise.

5. Velocity and position estimation based on period estimation

The period estimation, \hat{T}_f , obtained in the previous section can be used to estimate \dot{x}_M and in turn to obtain a real time velocity estimation. The fact that \dot{x}_M is estimated at each period will be used in order to compensate the drift due to errors previously mentioned. Define \hat{T}_f^i as the period estimation corresponding to the time T_M^i , namely the i - th occurrence of \dot{x}_M . Define $\hat{T}_M^{(i-1)}$ and $\hat{T}_M^{(i)}$ as the corresponding extremum occurrence estimated time. Such estimation will correspond to the ones obtained at each fulfillment of the condition **C1** of algorithm **A2**, i.e. $\hat{T}_M^{(i)} = t_1$ and $\hat{T}_M^{i-1} = t_0$. Thus, based on (14) we obtain

$$\hat{x}(\hat{T}_M^{i-1}) = -\frac{\Psi^{<2>}(\hat{T}_M^{i-1}, \hat{T}_M^i)}{(\hat{T}_f^i)}.$$
(25)

Analogously, an estimation $\hat{x}(T_M^i)$ can be obtained by using the estimated period \hat{T}_f^i and the third integral of Ψ that is

$$\Psi^{<3>}(T_M^{i-1}, T_M^i) = x^{<1>}(T_M^i) - x^{<1>}(T_M^{i-1}) - x(T_M^{i-1})(T_M^i - T_M^{i-1}) - \frac{1}{2}\dot{x}(T_M^{i-1})(T_M^i - T_M^{i-1})^2 + \eta^{<3>}(T_M^{i-1}, T_M^i),$$
(26)

then, assuming that $x^{<1>}(T_M^i) = x^{<1>}(T_M^{i-1})$, the next estimation can be obtained

$$\hat{x}(\hat{T}_{M}^{i}) = -\frac{\Psi^{<3>}(\hat{T}_{M}^{i-1}, \hat{T}_{M}^{i})}{\hat{T}_{f}} + \frac{1}{2}\hat{\dot{x}}(\hat{T}_{M}^{i-1})\hat{T}_{f}$$
(27)

In the ideal case of zero noise, no bias, exact integration and exact detection of the extremum instants T_M^i 's, we have that $\Psi^{<1>}(\hat{T}_M^{i-1}, \hat{T}_M^i) = 0$ ($\delta(t) = 0$). Although, in general $\Psi^{<1>}(\hat{T}_M^{i-1}, \hat{T}_M^i) \neq 0$, the assumption $\dot{x}(\hat{T}_M^{i-1}) = \dot{x}(\hat{T}_M^i)$ is made in order to obtain the period estimation \hat{T}_f^i . According to the aforementioned assumption an estimation of the error $\eta_{\beta}^{<1>}$ can be

obtained as

$$\hat{\eta}^{<1>}(\hat{T}_M^i) = \Psi^{<1>}(\hat{T}_M^{i-1}, \hat{T}_M^i).$$
(28)

Using (28) we can obtain an estimation of the errors $\eta^{<2>}$ and $\eta^{<3>}$ allowing to compensate them in $\Psi^{<2>}$ and $\Psi^{<3>}$, before computing $\hat{x}(\hat{T}_M^{(i)})$ and $\hat{x}(\hat{T}_M^{(i)})$ respectively. In order to do that, we consider $\hat{\eta}^{<1>}(T_M^i)$ as the result of the integration of the constant magnitude $\frac{\hat{\eta}^{<1>}(\hat{T}_M^i)}{\hat{T}_f^i}$ over the interval $[\hat{T}_M^{i-1}, \hat{T}_M^i]$. Thereby, the next error estimations are obtained

$$\hat{\eta}^{<2>}(\hat{T}_M^i) = \frac{1}{2} \Psi^{<1>}(\hat{T}_M^{i-1}, \hat{T}_M^i) \hat{T}_f$$
(29)

$$\hat{\eta}^{<3>}(\hat{T}_M^i) = \frac{1}{6} \Psi^{<1>}(\hat{T}_M^{i-1}, \hat{T}_M^i) \hat{T}_f^2.$$
(30)

Compensating (25) and (27) with the above errors yields

$$\hat{x}(\hat{T}_{M}^{i}) = -\frac{\Psi^{<2>}(\hat{T}_{M}^{i-1}, \hat{T}_{M}^{i})}{(\hat{T}_{f}^{i})} + \frac{1}{2}\Psi^{<1>}(\hat{T}_{M}^{i-1}, \hat{T}_{M}^{i}),$$
(31)

$$\hat{x}(\hat{T}_{M}^{i}) = -\frac{\Psi^{<3>}(\hat{T}_{M}^{i-1}, \hat{T}_{M}^{i})}{\hat{T}_{f}} - \left[\frac{1}{2}\hat{x}(\hat{T}_{M}^{i-1}) - \frac{1}{6}\Psi^{<1>}(\hat{T}_{M}^{i-1}, \hat{T}_{M}^{i})\right]\hat{T}_{f}.$$
(32)

Finally, the estimation of velocity and position is obtained through the next expressions

$$\hat{x}(t) = \Psi^{<1>}(\hat{T}_M^i, t) + \hat{x}(\hat{T}_M^i), \tag{33}$$

$$\hat{x}(t) = \Psi^{<2>}(\hat{T}_M^i, t) + \hat{x}(\hat{T}_M^i),$$
(34)

using (31) and (32) for the evaluation of $\hat{x}(\hat{T}_M^i)$ and $\hat{x}(\hat{T}_M^i)$ respectively.

Remark 5: While the assumptions S1 and S2, altogether with condition (24), restrict the signals and the signal to noise ratio that the estimation algorithm can manage, it is worth mentioning that no a priori bandwidth is assumed, no sinusoidal dominant frequency is required, noise and constant bias are allowed to be present and knowledge of the system model is prescinded.

6. Example

The velocity and position scheme presented in previous section was tested through simulations, using the estimations obtained with the algorithm A1. The following Van der Pol oscillator was used for generation of the position, velocity and acceleration signals

$$\ddot{x} + 0.1(x+1)(x-1)(x-2.2)\dot{x} + x = 0$$
(35)

$$\Psi(t) = \ddot{x}(t) + \eta(t) + \beta$$
(36)

$$=\ddot{x}(t) + \eta(t) + \beta \tag{36}$$

The parameters in (35) where proposed in (Karreman & Prood, 1995) as a mathematical model for an isometric contraction of the mammalian cardiac muscle. The acceleration signal was contaminated in simulation by a bounded zero mean noise, with $|\eta(t)| \leq 0.11$, and a constant bias, $\beta = 0.05$, hence, $\bar{\eta}^{<0>} = 0.11$ and $\bar{\beta}^{<0>}$ are used for simulations. The phase portrait (\ddot{x}, \dot{x}) of (35), with initial conditions $x_0 = 3$ and $\dot{x}_0 = -2$, is displayed in Fig. 5. The estimation results are shown in Fig. 6 and Fig. 7.

7. Conclusions

For a class of oscillatory motion, a scheme for velocity and position estimation using the information of noisy and biased acceleration measurements was presented. The inherent error accumulation, due

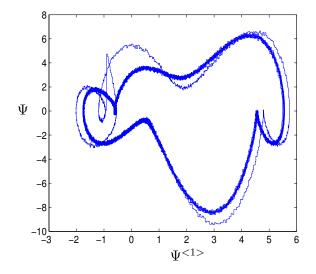


Figure 5. Phase portrait $(\Psi^{<1>}, \Psi)$ for system (35)-(36).

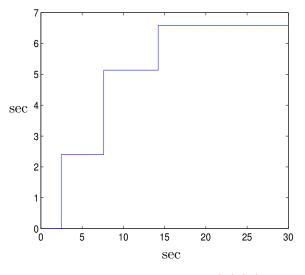


Figure 6. Period estimation for (35)-(36).

to integration drift, is compensated using the value corresponding to the period of oscillation. An algorithm for estimation of the aforementioned period of oscillation is proposed. No model of the motion nor a priori information on the period of the signals is required. As a result, the proposed scheme is suitable as a standalone solution for estimation of position and velocity using only inertial information without any aiding measurements. Further research is directed to the application of the proposed solution on a practical scenario with real measurement data.

8. Appendix

Proofs.

Consider that the unique absolute extremum for \dot{x} is a global maximum denoted as \dot{x}_M (the proof in the case of a global minimum follows the same rationale). For this case $\alpha = 1$ is selected and the

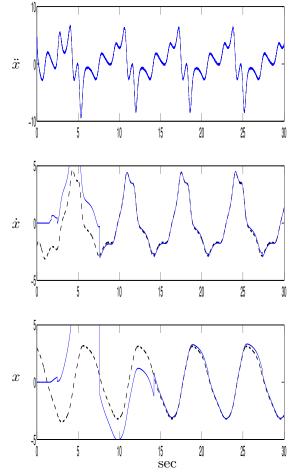


Figure 7. Velocity and position estimation for (35)-(36).

corresponding forms of conditions ${\bf C0}$ and ${\bf C1}$ are rewritten as

$$\Psi^{<1>}(t_0,t) > \Psi_m^{<1>} \tag{37a}$$

$$\Psi^{<1>}(t_0,t) > -\delta(t)$$
 (37b)

and

$$\begin{array}{ccc} \Psi^{<1>}(t_1,t) &< & -2\delta(t), \\ fl &= & 0. \end{array} \right\}$$
 (38)

Note that condition C0 can take two forms, (37a) for fl = 0 and (37b) for fl = 1. Proof of Lemma 1. This proof is divided into two parts:

• Localization of the global maximum. This part is focused on the fact that the first fulfillment of Lemma 1:

$$\dot{x}(t_1) \in [\dot{x}_M - \bar{\eta}_\beta^{<1>}(|T_M - t_1|), \dot{x}_M].$$
(39)

is not affected for the initial conditions $\ddot{x}_{\circ} = \ddot{x}(0), \dot{x}_{\circ} = \dot{x}(0)$ and $x_{\circ} = x(0)$. Define T_M^1 as the time of first appearance of \dot{x}_M , i.e., $\dot{x}(T_M^1) = \dot{x}_M$. Suppose that the initial condition \dot{x}_{\circ} is such that

$$\dot{x}_M - \dot{x}_\circ > \bar{\eta}_\beta(0, T_M^1). \tag{40}$$

Then, considering (40), if C1, i.e. (38), is not fulfilled between t = 0 and $t = T_M^1$, implying \dot{x} is not decreasing, we have

$$\Psi^{<1>}(t_0, T_M^1) = \Psi^{<1>}(0, T_M^1)$$

= $\dot{x}_M - \dot{x}_\circ + \eta_\beta(0, T_M^1)$
> $\bar{\eta}_\beta(0, T_M^1) - |\eta_\beta(0, T_M^1)|$
> 0.

Thereby, C0, that is (37a), will be necessarily true for some $t \in [t_0, T_M^1]$. This implies that the time t_1 captured due to C0 verifies (39).

Captured due to CO vermes (59). On the other hand, suppose that C1, see (38), is true **before** T_M^1 , at a time which will be denoted as T_{C1}^1 . This implies that $\Psi^{<1>}(t_0, T_{C1}^1)$ has decreased more than $\bar{\eta}_{\beta}^{<1>}(t_0, T_{C1}^1)$, because $\eta_{\beta}(t_0, T_{C1}^1) < \delta(t) = \bar{\eta}_{\beta}^{<1>}(t_0, T_{C1}^1)$. The assignments $\Psi_M^{<1>} = 0$ and fl = 1 due to C1 implies that condition C0 takes the form (37b). Now, note that the latter must be fulfilled before a new true event of C1: Since t_0 is reseted to the time of occurrence of the last largest value of $\Psi^{<1>}(t_0, T_{C1}^1)$, one obtains that

$$\Psi^{<1>}(t_0, T_{C1}^1) < -2\delta(t_0, T_{C1}^1),$$

holds, after the assignments in C1, with x_M still to appear. Then, there exist a time t^* such that

$$\dot{x}(t^*) = \dot{x}(t_0);$$
 with $t^* < T_M^1 < T_f$

and for $t = t^*$ one obtains

$$\Psi^{<1>}(t_0, t^*) = \dot{x}(t^*) - \dot{x}(t_0) + \eta_{\beta}^{<1>}(t_0, t^*)$$

$$\geq -|\eta_{\beta}^{<1>}(t_0, t^*)| > -\delta(t).$$

Then, (37b) is fulfilled and **C0** takes the form (37a) again. From the above, it is concluded that condition (37) is verified, independently of the initial conditions, for some time t_1 such that (39) holds.

• Successive global maximum detections. Once t_1 satisfies (39), condition **C1** is fulfilled only once after each \dot{x}_M occurence. This ensures that no local extremum is mistaken as the global one. Consider that **C1** is true after (39) is satisfied then,

$$\dot{x}(t_1) \in [\dot{x}_M - \bar{\eta}_\beta^{<1>}(|T_M - t_1|), \dot{x}_M].$$
(41)

and note that **C0** must be true before **C1** can be satisfied again. Recall that the local extremum closest in magnitude to \dot{x}_M is \dot{x}_M^- . Then, for $\dot{x}(t) = \dot{x}_M^-$, after (39) holds, we obtain

$$\Psi^{<1>}(t_0,t) = \dot{x}_M^- - \dot{x}(t_0) + \eta_{\beta}^{<1>}(t_0,t)$$

$$\Psi^{<1>}(t_0,t) \le \dot{x}_M^- - \dot{x}_M + \eta_{\beta}^{<1>}(t_0,t) + \bar{\eta}_{\beta}^{<1>}(Tf).$$

From (24) we have that $\dot{x}_M - \dot{x}_M < -2\delta(T_f)$, hence

$$\Psi^{<1>}(t_0,t) < -2\delta(T_f) + \delta(T_f)$$

$$\Psi^{<1>}(t_0,t) < -\delta(T_f)$$

thus, $\dot{x}(t)$ must be larger than \dot{x}_{M} in order to fulfill again **C0**. Thus, after the first time (39) is fulfilled, **C0** cannot be triggered due to a local extremum in the original \dot{x} . This concludes the proof of Lemma 1.

Proof of Proposition 1. Proposition 1 is a straightforward result of Lemma 1. Note that after (39) holds, for each subsequent C1 verification, the variables t_0 and t_1 will contain the time of

detection of two consecutive global maxima of $\Psi^{<1>}$. The condition (39) cannot be ensured before a whole period has elapsed, then at least two periods are needed in order to ensure that a correct estimation of T_f is obtained. As a consequence, the first period estimation may occur during the third period after launching the algorithm (C1 verification after two periods will occur on the third one).

Time Varying $\bar{\eta}^{<1>}$

We mentioned before that integration of noise leads to a drift. The magnitude of this drift increases with the time of integration, on the contrary to the mean value of the noise measurements, that approaches zero as time increases due to the zero mean assumption. In (Thong et al., 2002) it is validated that, for the sampled case, the value of the integral of zero mean noise, with uncorrelated measurements, is a random value with Gaussian distribution. The variance of such Gaussian distribution increases with time. In fact, the selection of a constant δ , in Algorithm A1, implies in some sense knowledge of both the bound of noise and some upper bound for the period.

For the sake of completeness a brief description of the procedure in (Thong et al., 2002) to obtain the expression for the variance of the drift due to integration of noise at a time T is included next. Consider an arbitrary time interval T, and the N values of the noise obtained over this interval as a result of sampling at frequency f_s , i.e. $N = Tf_s$. Define $\eta[i]$ as the sampled noise values. The average of the noise values over the N sample points is:

$$D_N = \frac{1}{N} \sum_{i=1}^{N} \eta[i].$$
(42)

 D_N approaches zero as $N \to \infty$, but for finite N will in general be nonzero. The square of the mean of D_N is given by

$$E[(D_N)^2] = \frac{1}{N^2} E[(\eta[1] + \ldots + \eta[N]) \times (\eta[1] + \ldots + \eta[N])]$$
(43)

where $E[(D_N)^2]$ denotes the expected value. Expanding the brackets, using the symmetry condition $E[\eta[i] \cdot \eta[j]] = E[\eta[j] \cdot \eta[i]]$ and using the *p*-th lag of the autocorrelation function, under the assumption of stationary data, defined as

$$r[p] = E[\eta[i] \cdot \eta[i+p]]$$

the next expression is obtained

$$E[(D_N)^2] = R_1 + R_2$$
where
$$R_1 = \frac{r[0]}{N}$$

$$R_2 = \frac{2}{N^2} \sum_{j=1}^{N-1} r[j](N-j).$$
(44)

The term R_1 is the contribution of noise samples in the absence of correlations between samples and R_2 is the respective contribution from correlations. For the white noise model no correlation between samples is assumed, that is $R_2 = 0$. Let σ_{η} be the standard deviation of the noise. Using $R_2 = 0$ and $r[0] = \sigma_{\eta}^2$ in equation (44) one gets

$$E[(D_N)^2] = \frac{\sigma_\eta^2}{N}.$$
(45)

Taking square roots on both sides of the above equation yields (Thong et al., 2002)

$$RMS(D_N) = \frac{\sigma_{\eta}}{\sqrt{N}},$$

$$= \frac{\sigma_{\eta}}{\sqrt{Tf_s}}.$$
 (46)

By integrating the above average value over T, the standard deviation of the integral of noise at time T, denoted as $S_v(T)$, can be obtained:

$$S_{v}(T) = \int_{0}^{T} D_{N} dt,$$

$$RMS(S_{v}(T)) = \int_{0}^{T} RMS(D_{N}) dt,$$

$$= \sigma_{\eta} \frac{\sqrt{T}}{\sqrt{f_{s}}}.$$
(47)

Thus, for the case in which a bound for $\eta^{<1>}$, cannot be given a priori but there is some knowledge of the bounds of noise we propose to use a time varying δ in algorithm A1 defined as follows

$$\eta^{<1>}(T) = 4RMS(S_v(T))$$
$$= 4\sigma_\eta \frac{\sqrt{T}}{\sqrt{f_s}}$$
(48)

The reason for the above criterion is that $P(|S_v(T)| \le 4RMS(S_v(T))) \approx 0.99994$, giving a good reliability that the drift due to noise integration will be below such value. The time T of integration is evaluated as

$$T = t - t_0 + f_s^{-1}, (49)$$

the term f_s^{-1} is introduced for practical reasons, in this way in δ is considered at least one sampling time of integration.

Remark 6: Instead of the standard deviation σ_{η} , which may be difficult to know, and in order to overestimate the threshold $\eta^{<1>}$, we propose to use an upper bound bound of the noise as σ_{η} . Thus, using $\sigma_{\eta} = \bar{\eta}^{<0>}$ and (49) in (48), the expression (18) can be obtained.

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