Single photon multiclock lock-in detection by 1 picosecond time stamping 2

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10 Abstract: Extracting signals at low single-photon count rates from large backgrounds is a 11 challenge in many optical experiments and technologies. Here we demonstrate a single-photon 12 lock-in detection scheme based on continuous photon time stamping to improve the signal-to-13 noise ratio by more than two orders of magnitude. Through time-resolving the signal 14 modulation induced by periodic perturbations, 98% of dark counts are filtered out and the <1 15 count/s contributions from several different nonlinear processes identified. As a proof-of-16 concept, coherent anti-Stokes Raman measurements are used to determine the vibrational 17 lifetime of few molecules in a plasmonic nanocavity. This detection scheme can be applied to 18 all single-photon counting experiments with any number of simultaneous modulation 19 frequencies, greatly increasing signal-to-noise and resolving physical processes with 20 picosecond time resolution while keeping photon dosage small. Open instrumentation provided 21 here enables low-cost implementation.

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24 1. Introduction

25 Transient or time-resolved optical experiments often require elaborate experimental setups 26 designed to measure very low signal intensities over ultrashort time scales, and hence they often 27 suffer from poor signal-to-noise ratios [1]. Typically, the linear response dominates over any 28 perturbation, giving small induced changes in the signal. Increasing the strength of a repetitive 29 perturbation (for instance the optical pulse intensity) to enhance the nonlinear signal is often 30 problematic since this can damage the samples, preventing stroboscopic measurement. For 31 probing single nanostructures or individual quantum systems, the strong perturbation needed 32 to obtain clear signals unfortunately often induces irrevocable structural changes such as bond 33 cleavage [2], atomic displacements [3], reshaping [4] or ablation [5].

34 A technique commonly applied to extract such weak signals from a noisy background is 35 lock-in detection. By introducing a modulation to the sample, the amplitude and phase of the 36 emerging signal can be determined using phase-sensitive heterodyne detection while noise at 37 other frequencies is rejected [6]. For instance, in all-optical experiments such as four-wave-38 mixing in semiconductor optical amplifiers [7] or stimulated emission from single nanocrystals [8], the pump pulse train is amplitude-modulated at high frequency f_{mod} . In 39 40 scanning near-field optical microscopies, this modulation is provided by the vibration of a tip 41 above the sample [9]. Optical lock-in detection is also used for stimulated-emission-depletion 42 microscopy to enhance contrast in fluorescence microscopy [10], and in many other scenarios.



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Fig. 1. Comparison of single-photon detection techniques. (a) Previous single-photon lock-in techniques: photons are counted in time bins to recover an analogue signal for analysis with a standard phase-sensitive lock-in algorithm. (b) Time-correlated single-photon counting: the time interval Δt between a trigger pulse and photon detection events is recorded to resolve the time decay of an optical signal, typically fluorescence. (c,d) Multiclock single-photon lock-in technique developed here: Continuous timestamping of photon arrival times (t_0) and multiple reference signals with frequency f_i allows identification of the phase τ_i of every individual photon compared to all the reference frequencies.

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51 Several versions of phase-sensitive detection have been implemented for single-photon 52 counting detectors [11–16]. However, these rely on binning photon counts in successive time 53 intervals to create a continuous signal for traditional lock-in analysis instead of analysing each 54 photon individually (Fig. 1a). The detection challenge is to identify $\sim 1 \text{ cts/s}$ (counts per 55 second) caused by the linear or nonlinear optical processes of interest under a large background. 56 Here, we present a new approach to single-photon lock-in detection utilising continuous 57 picosecond photon time stamping of each individual photon to temporally resolve the evolution 58 of an optical signal undergoing modulation. By recording the arrival time of each signal photon 59 at the detector alongside time stamps from synchronised reference clocks (Fig. 1c), the time 60 dynamics of the signal is extracted. For each photon, the phase (τ_i) and frequency (f_i) of each 61 reference clock at the photon arrival time (t_0) is accurately determined from a linear fit to the 62 timestamp data of the previous N clock detections (Fig. 1d). In the post-experimental data 63 analysis, experimental noise and background signals can be easily removed through temporal 64 gating, implementing lock-in amplification individually for every photon and each reference 65 clock suitable for increasing the signal-to-noise ratio of single-photon counting experiments by 66 many orders of magnitude (here with factor >100).

67 The signal-processing technology to convert the arrival time of an electronic pulse to a 68 digital time stamp is well established and widely used in high-energy physics [17,18]. Field 69 programmable gate array (FPGA) boards are capable of performing logical operations with 69 digital electronic signals in real time with bandwidths exceeding 100 MHz. Time-to-digital 70 converters (TDC) implemented with FPGAs can now achieve timing precision < 10 ps [19] 72 while it is possible to parallelise devices using 264 channels or more on one board [20].

73 Combining picosecond time stamping with pulsed optics holds enormous potential to 74 improve existing and enable new applications. Although the proposed experimental design can 75 increase signal-to-noise in all single-photon experiments, it is particularly advantageous in the 76 areas of quantum correlation, time-of-flight spectroscopy, and scanning near field microscopy. 77 A particular new capability provided by continuous photon time-stamping is the ability to 78 compare photon signals to multiple reference clocks at the same time. For instance here this 79 allows single-photon lock-in synchronisation simultaneously to: (1) the optical pulse repetition 80 rate, (2) the laser power modulation (here two different on-off periodic modulations), and (3) 81 extra triggers (here pulse delay scan). This retrieves the maximum possible information content 82 of each detected photon. In comparison, start-stop photon detection schemes for time-correlated 83 signal photon counting (TCSPC) synchronise only to optical pulses, greatly limiting their use 84 (Fig. 1b) [21]. We also note some similarities to LIDAR technology where multiple laser 85 repetition rates are used to avoid distance ambiguity. However, these measurements are 86 performed in quick succession, while here we resolve a signal modulating with several 87 frequencies at once in our setup.

88 In this paper, coherent anti-Stokes Raman spectroscopy (CARS) is used as an example of a 89 typical nonlinear experiment with low single-photon count rates that benefits from our 90 scheme [22,23]. We thus briefly describe the measurements, while noting the general 91 applicability of the technique. In CARS, molecular vibrations ν are excited by coherently 92 pumping with two laser pulses (pump ω_p and Stokes ω_s , Fig. 2a,b), tuned so their frequency 93 difference matches the vibration, $\omega_p - \omega_s = \nu$. Subsequently anti-Stokes scattering to $\omega_x + \nu$ 94 of a time-delayed separate probe pulse at ω_x is used to identify the relaxing molecular 95 vibrations and measure their vibrational lifetime. To access a domain capable of observing 96 CARS from single molecules [22], nanoscale optical confinement is needed which is accessed 97 here using plasmonic nanocavities [24]. These are based on a nanoparticle-on-mirror geometry 98 (NPoM) in which an ordered molecular monolayer (here biphenyl-4-thiol, BPT) is sandwiched 99 in a nm-thick gap between Au facets [25]. A background of non-resonant four-wave-mixing 100 dominates the detected signal at $\omega_x + \nu$ [3,26] while additional noise is introduced by 101 electronic dark counts and stray light. Our detection scheme here isolates < 1 cts/s rates of this 102 nonlinear CARS signal using picosecond photon time stamping through a low-cost open-103 architecture FPGA board.



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Fig. 2. (a) Schematic experimental setup of coherent anti-Stokes Raman spectroscopy (CARS). Three pulsed lasers with repetition period T_{rep} are focused on a nanoparticle-on-mirror (NPoM) nanocavity sample, with Stokes and probe modulated periodically with period T_{mod} . The emerging signal detected by a single-photon detector (SPAD) contains photons from the nonlinear CARS process, which need to be isolated from other contributions. (b) Excitation scheme of CARS, with pump and Stokes exciting a molecular vibration, probed by anti-Stokes scattering of the probe beam with time delay Δt . (c) Modulation of the detected SPAD signal. Only when all three beams are turned on can the nonlinear 3-pulse CARS signal be observed (green).

112 2. Experimental setup

113 The basis for detecting a nonlinear optical signal above other contributions is to identify and 114 subtract the linear components and separate the multitude of other nonlinear signals. Here, this 115 is achieved by modulating two laser beams exciting the sample (Fig. 2a). For CARS 116 experiments, each of the three laser beams contributes to the total recorded count rate (Fig. 2c). 117 Pump and probe pulses generate broadband contributions across the detected spectral window 118 through electronic anti-Stokes Raman scattering. With pump and Stokes beams exciting the 119 sample, a 2-pulse CARS process is also possible but cannot be time-resolved since the pump 120 then both excites and probes. Only all three pulses together excite emission of the desired 3-121 pulse CARS signal, enabling investigation of the ultrafast vibrational dynamics. In systems 122 with high signal and stability, these contributions can be separated by sequential acquisition of 123 spectra without fast laser modulation. However with a fixed photon budget on samples such as 124 nanoscopic structures and single molecules, a new approach is required to increase signal-to-125 noise and avoid damage.

126 A widefield microscope guides these pulses onto the nanocavity sample (Fig. 3), which are 127 generated at 820 nm (Stokes), 726 nm (pump) and 722 nm (probe) from an $f_{rep} = 80$ MHz 128 pumped optical parametric oscillator (Spectra Physics OPO). The three lasers are spectrally 129 tuned so that pump and Stokes pulses can resonantly drive the molecular vibration while the 130 Stokes pulse is off-resonant to vibrations excited by the probe pulse (see supplementary Fig. 131 S2a). With ultrafast pulses of 500 fs duration, the spectral resolution of the exemplar 132 experiment here is 50 cm⁻¹. Stokes and probe beams are each modulated with electro-optic 133 modulators (EOM) driven by function generators (HP 33120A) producing square wave output 134 at variable frequency f_{mod} . The two function generators are phase locked and operate with a fixed phase difference of 90°. For CARS experiments, probe beam pulses are delayed by Δt 135 with respect to the other two laser beams using a delay stage running in a continuous loop 136 137 around $\Delta t = 0$ ($f_{delay} = 1$ Hz). The Stokes pulses are temporally aligned to the pump pulses 138 (by optimising the 2-pulse CARS signal) initially, after which this zero relative delay is fixed. 139 All three beams are spatially overlapped and co-focussed on the sample.

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Fig. 3. Experimental setup for CARS experiments with photon time tagging. Three laser beams are produced by an 80 MHz pulsed pump laser and optical parametric oscillator (OPO). Two of these beams are 50 kHz modulated by electro-optic modulators (EOM) driven with square waves. The probe beam delay is continuously varied by moving a mirror, giving trigger synchronization pulses at the beginning of each loop. All beams are spatially overlapped and focused onto the sample in a microscope. A single-photon avalanche diode (SPAD) detects the spectrally filtered signal photons. A reference signal for the pulsed laser is provided by a fast photodiode (PD). The electronic signals (*V*-*t* diagrams) are passed to a central FPGA board, where the signal and clock timestamps are recorded digitally.

In the detection path, the laser light is blocked using spectral filters and a single-photon avalanche diode (SPAD, Micro Photon Devices PDM \$PD-100-CTD) detects the signal photons. Each arriving photon leads to an electronic pulse with timing accuracy of 35 ps. Typical count rates in our experiment range from 1 to 1000 cts/s obscured by at least 100 cts/s of dark counts and stray light.

156 In addition to the single-photon counts, three reference signals are recorded. A fast 157 photodiode monitors the pulse repetition rate of the Ti:Sapphire pump laser (Spectra Physics 158 MaiTai, photodiode integrated into laser head) producing a digitally-conditioned f_{rep} = 159 80 MHz pulse train. Together with a synchronised TTL square-wave at f_{mod} and TTL trigger pulses at the beginning of each delay stage loop (f_{delay}), all electronic signals are passed to the 160 161 FPGA board (Fig. 3). This board continuously converts the arrival times of all electronic 162 signals to digital timestamps that are streamed to a computer and saved. This time-to-digital conversion is achieved by combining a fine time-to-digital converter (TDC) and a 32-bit coarse 163 164 counter (Fig. 4a). The fine TDC consists of a tapped delay line that provides 30 ps accuracy 165 with a range of 5 ns. The coarse counter simply counts increments of the 200 MHz internal 166 FPGA clock and hence covers the nanosecond to second regime. 167



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Fig. 4. Time-to-digital conversion with an FPGA. (a) Time stamps with ps precision are recorded by combining a fine time-to-digital converter (TDC) and a coarse counter. Data is streamed to the computer via gigabit Ethernet (GbE) when triggered by a signal photon. (b) In the fine TDC, an electronic signal is carried along a tapped delay line connected to a tap register. When the signal is stopped by the FPGA clock, the register entry is recorded.

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174 In a tapped delay line (as commonly applied in time-to-digital conversion [18]), each element 175 is connected to a register that records the state of the carried signal and hence the current 176 position of the signal within the carry line (Fig. 4b). To determine the photon arrival time, an 177 incoming electronic signal launches a pattern that travels along the delay line until it receives a 178 stop signal which causes the position of the signal to be read from the register. Since the stop 179 signal is provided by the FPGA clock, the arrival time of the signal is the time the signal spent 180 propagating along the delay line before the latest clock. The propagation distance is converted 181 to time in post-experimental data processing.

Since the laser pulse reference clock is at 80 MHz, recording every pulse arrival time would require a data transfer rate >25 GB/s. Therefore, a trigger system was implemented to only record reference time stamps when a signal photon is detected: every time a photon arrives at the SPAD, a defined (and tuneable) number of most recent timestamps from all reference clocks are sent to the host PC by the readout controller alongside the photon timestamp. This allows the user to control the number of reference timestamps recorded for each signal event, and thus optimise signal-to-noise by maximising the accuracy of the laser reference given the maximum data transfer rate at each SPAD count rate. As we show below, even a stable repetitive laser system experiences variations of <0.1% in cavity length, which if not tracked greatly reduce the timing precision. Here, ten timestamps per reference clock sufficed to reach the optimum time resolution and thus minimise the required data size.

The TDC was implemented on a Digilent Arty Z7 development board specifically programmed for this application. We provide the FPGA software online [will be provided on acceptance of the paper] under an open source licence to allow for low-cost implementation. Here, we demonstrate the function of this system with three reference clocks and one signal channel. However, the FPGA board supports up to eight input signals which can be assigned to record either reference or signal channels, allowing for more complex experimental setups.

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201 3. Results

With this setup, data files are acquired containing a list of arrival timestamps. For the first demonstrations here, two reference clock signals are chosen, f_{rep} and f_{mod} . In general, any periodic reference signal can be chosen as a clock enabling a plethora of different applications. As an initial calibration experiment, a single laser pump beam modulated at $f_{mod} = 1$ kHz is focused on the sample and spontaneous Stokes scattering from the nanocavity is recorded by the SPAD (see Fig. S1 for spectrum detected here).



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210Fig. 5. Analysis of timestamp data from spontaneous Raman scattering. (a,b) Histograms of the clock frequency f for211(a) the laser pulses and (b) the EOM modulation, over 100 s. (c) Reconstruction of the laser pulse shape discriminating212signal (green) and dark counts (grey). Inset: 0.2 ns time window $\Delta \tau$ including 60% of signal counts (dashed). Bin size21325 ps. (d) Reconstruction of the laser modulation cycle, recording Raman scattering from a nanocavity with single laser214pump turned off and on (laser power 2 μ W). Bin size 2 μ s.

From the periodic clock timestamps, the frequency f of the clock at the time of arrival of each photon at the detector is determined. For each of the detected photons, the clock frequency slightly varies due to fluctuations in the period seen in the histograms of Fig. 5a,b. Over this 100 s data set, the average pulse repetition rate of the laser was $f_{rep} = 79.64$ MHz with a standard deviation of 25 kHz. For the EOM modulation of the laser, a frequency of 1.00001 kHz is found \pm 0.78 mHz as expected for this source. These vary little between successive data sets, demonstrating the accuracy of our global clock.

The point in the clock cycle τ when each signal photon was detected is now calculated as well. Extracting this value for all photons detected during a measurement allows us to reconstruct the signal modulation through the clock cycle, of length set by the inverse of the clock frequency. This projects all counts into a single clock cycle and hence repetitive but extremely low count rates can be analysed by simply increasing the integration time.

228 We first apply this temporal reconstruction to the periodically pulsed laser. Using the laser repetition rate to wrap the signal within the $T_{rep} = 12.5$ ns long clock period places most counts 229 230 near a specific time which depends on a system electronic delay from optical paths, cables and 231 latency (Fig. 5c). All photons emerging from the sample due to excitation by the periodic laser 232 pulses are detected within a $\Delta \tau \sim 200$ ps wide window, because Raman is a prompt process. 233 Even though the photons are emitted within the 500 fs optical pulse width, this peak is 234 broadened by the detection electronics. In contrast, electronic dark counts and stray light 235 photons are uncorrelated with the excitation pulses and hence give a constant background signal 236 spread over the whole period. Exploiting this property of dark counts allows us to distinguish 237 them from photons emitted by the sample and remove them from the data. This results in a 238 >98% reduction in dark counts and thus an increase of $T_{rep}/\Delta \tau > 60$ in signal-to-noise for low 239 count rate experiments (Fig. 5c). The approach to remove dark counts in the time domain is 240 similar to previous reports of time-gated single-photon counting which has been demonstrated 241 both with actively quenched detectors [27–29] and digitally by TCSPC [30].

242 The same concept of temporal reconstruction is now applied to determine which lasers are 243 exciting the sample when each signal photon was generated. This is crucial for separating the 244 linear and nonlinear components of the signal. To demonstrate this functionality in our 245 calibration case, the single laser pump is switched on and off at frequency 1 kHz and 50% duty 246 cycle while recording Raman scattering from the nanocavity sample (average laser power on 247 sample $2 \mu W$). Consequently during the laser modulation cycle, two temporal regions with 248 constant count rates are observed, with the background for laser-off being simply identified 249 (Fig. 5d).

250 When the power of the laser is reduced by near hundred-fold (50 nW on sample), the 251 modulation depth of the signal is greatly reduced since the dark count rate is now larger than 252 the Raman signal (Fig. 6a). Dark counts outside the 200 ps window centred on the laser pulse 253 in Fig. 5c are now removed, reducing the count rate outside the laser window close to zero (Fig. 254 6b) with only a residual 2% unfiltered dark counts remaining. This dark count rejection 255 increases the modulation visibility by 400%, but in measurements with even lower laser power 256 or more stray light/dark counts, this enhancement can exceed 4000%. Analysing the 257 distribution of the noise in the modulation reveals two Poisson distributions for the laser on and 258 off states (Fig. 6c). These arise from the single-photon counting statistics in the experiment and 259 show that the signal-to-noise is now only limited by photon shot noise, which can thus be 260 improved by increasing the integration time to collect more photons.



Fig. 6. Demonstration of dark count removal. (a) Reconstruction of laser-modulated spontaneous Raman scattering from a nanocavity at low average laser power (50 nW) over 100 s, with Raman signal smaller than the dark count rate.
(b) Reconstruction of modulated signal after dark count removal. Bin size for (b, c) 2 μs. (c) Distribution of noise in

266 (b), with fitted Poisson distributions (lines) indicating the single-photon statistics of the experiment.



Fig. 7. Time-resolved CARS experiments. (a) CARS signal during 50 kHz laser modulation period (dark counts removed). Contributions from the different lasers are marked in orange (pump), red (Stokes) and yellow (probe), with the nonlinear signal CARS signal in green. Bin size 40 ns. (b) Extracted contributions from (a) locked-in to the delay stage sweep as the probe (Pr) pulses are delayed with respect to pump (Pu) and Stokes pulses. (c) Time-resolved CARS signal with exponential decay fit to extract a vibrational lifetime of 1595 fs of the BPT molecules in the nanocavity. For negative delays, the signal follows a Gaussian matching the 500 fs pulse width. Average laser power on sample per beam is 0.2 µW.

278 In order to demonstrate the ability of the setup to detect a small nonlinear signal upon a large background, CARS experiments are then carried out. The detected CARS spectra are shown in 279 supplementary Fig. S2b. Since two beams are now modulated, each at $f_{mod} = 50$ kHz (to 280 281 improve signal-to-noise) but with one phase shifted by 90°, with addition of a third beam of 282 constant intensity, the signal modulation shows four distinct windows of different height (Fig. 283 7a) as expected from the experimental design (Fig. 2c). Dark counts in the CARS experiments 284 throughout the entire modulation period were removed as detailed above. The measured count 285 rate is highest when all three beams illuminate the sample and lowest when only the pump beam 286 is on. Analysing the different windows allows the contributions from each individual laser 287 combination to be determined (yellow, orange and red in Fig. 7a). Subtracting these values 288 from the count rate when all lasers illuminate the sample allows the nonlinear 3-pulse CARS 289 count rate to be extracted (green in Fig. 7a).

290 With this method to extract the nonlinear signal, time-resolved CARS measurements can 291 now be performed by delaying the probe pulse compared to pump and Stokes pulses. This is 292 achieved by mechanically scanning a delay line for the probe pulse. Conventionally slow scans 293 are performed, integrating until sufficient signal-to-noise is achieved at each time point, 294 however this produces strong artefacts in the delay scan due to transient changes in the emission 295 spectrum (caused by movement of Au atoms on the nanoparticle facet [31,32]) and damage to 296 the nanostructure. Hence, the delay stage is continuously scanned back and forth at $f_{delay} =$ 297 1 Hz and a third reference trigger is introduced into the FPGA from the scanning delay line. 298 This allows each photon detected to also be tagged with the time-delay at which it was 299 measured, thus building up the entire time-delay curve simultaneously, without any artefacts.

300 Time-resolved tracks for all contributions identified by the laser modulation are compared 301 in Fig. 7b. The 2-pulse CARS signal induced by pump and Stokes beams (red) stays constant 302 as well as the electronic Raman scattering from the pump (orange). While electronic Raman 303 scattering from the probe also leads to a constant signal, vibrational pumping by surface-304 enhanced Stokes scattering of pump and probe photons adds a time-dependent signal to the 305 yellow contribution in Fig. 7a. This signal decays equally to both positive and negative delays 306 as pump or probe excite the sample identically. On the other hand, the 3-pulse CARS count rate 307 decreases for both positive and negative delay, but is not symmetric around 0 ps (Fig. 7c). When 308 the probe pulse arrives before the molecules are excited, the signal vanishes quickly with a rise 309 time corresponding to the pulse length of 500 fs (orange dashed, Fig. 7c). For probe pulses 310 arriving after pump and Stokes, the signal decreases exponentially. From an exponential fit, the 311 lifetime of the 1585 cm⁻¹ vibration of BPT is estimated as 1595 ± 420 fs. This signal is emitted 312 from only an estimated 100 molecules in the nanocavity gap, billions of times fewer than for solution measurements. Previous attempts to measure this without single-photon detection 313 314 required average laser powers $>4 \mu W$ per beam, at least ten-fold more than in this photon-315 counting lock-in mode, and which is enough to perturb and destroy the nanocavity structures. 316 With safe powers of $I=0.2 \,\mu$ W per beam employed here, the CARS signal is 1000-times smaller 317 (since it scales as I^3), below the limits of integrating detectors. As the signal strongly varies 318 from nanoparticle to nanoparticle, further ongoing experiments and theory are required for full 319 analysis, but beyond the scope of this article.

320 4. Discussion

321 To quantify the improvements made by eliminating dark counts, the window (within which 322 counts are not removed) is centred on the pulse and increased in width (Fig. 5a inset). For each 323 width, the percentage of dark counts that are removed is calculated as well as the photon events 324 preserved (Fig. 8a). A window width of $\Delta \tau = 200$ ps retains ~60% of real photon events and 325 only 1.6% of all dark counts. To remove a higher percentage of dark counts, the window width 326 has to be reduced thus decreasing the signal counts. The ratio of signal counts to dark counts 327 (Fig. 8b) shows that by filtering out dark counts, this ratio can be increased by almost two 328 orders of magnitude from 1.3 in the unfiltered data to >100. Since a compromise between best 329 signal-to-noise and maximum retained signal counts is demanded, we thus choose a window 330 width of 200 ps which retains signal counts most efficiently (before green curve in Fig. 8a 331 saturates) that increases the signal-to-noise by 60. 332



Fig. 8. Characterisation of dark count removal while varying filter window width $\Delta \tau$ (see Fig. 5c). (a) Percentage of signal counts preserved and dark counts removed for increasing window widths. (b) Ratio of signal counts to dark counts, compared to unfiltered data (horizontal line at 1.3).

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338 Currently, limitations of the timing precision in this setup are observed in the distorted pulse 339 shape. Even though the laser pulses are < 1 ps, the reconstructed pulse shape (Fig. 5c) has a 340 width of ~ 200 ps. The SPAD has a nominal jitter of only 35 ps and the FPGA board has a 341 timing precision of 30 ps (see Fig. S3), but further inaccuracies are introduced by the fast 342 photodiode (internal in Spectra Physics Maitai pump laser) and amplification of the MHz clock 343 signal by two amplifiers (Mini-Circuits ZFL-1000LN). Additionally, the detector electronic 344 response adds a shoulder to the peak, thus decreasing the percentage of preserved signal counts 345 after filtering out dark counts. Improvements in the detection electronics can thus further 346 increase the fraction of dark counts removed by this technique. For instance, without FPGA 347 noise and reaching the SPAD resolution of 35 ps would give another ~10-fold improvement.

We presented a scheme here of modulated lasers for the example of CARS. Typically laser modulation and lock-in detection is also used for stimulated Raman scattering (SRS). Since SRS requires detection of small changes in the pump laser intensity, the detected powers are far above the single-photon regime. In an optimised configuration, our FPGA setup can only record up to 1 Mcts/s count rates (pW) and is therefore not suitable for SRS without upgrading electronics to deal with much higher count rates.

To highlight the potential of the presented technology, Fig. 9 compares different photon detection techniques. Depending on the photon count rates, different photodetectors must be selected. For single-photon experiments, SPADs are suitable while traditional photodiodes (PD) are needed for spectroscopies delivering higher light intensities (> nW). In between, avalanche photodiodes (APD) and photomultiplier tubes (PMT) provide detection of photocurrents with high gain.

The crucial experiment signal-to-noise ratio strongly depends on signal detection and amplification. A SPAD delivers one voltage pulse for every detected photon and thus the only source of noise is photon shot noise and dark events. For photodiodes, where an electrical
current induced by light is produced, additional noise sources include the electronic shot noise
due to the diode dark current and thermal detector noise as well as noise from amplifiers
necessary to record the small output currents.



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Fig. 9. Comparison of different photon detection techniques. Single-photon avalanche diodes (SPAD), avalanche photo diodes (APD), and photomultiplier tubes (PMT) can detect single-photon count rates whereas photodiodes (PD) operate at higher light intensities. The signal-to-noise ratio of the detectors can either be improved by lock-in amplifiers in the high signal regime, or with the photon time stamping setup (SPAD + Timetag) presented here for single-photon detection, which can be extended to higher count rates using a combination of several SPADs (multiSPAD + Timetag). A detailed description of the comparison is provided in the SI.

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375 For photodiodes, a typical way to improve this signal-to-noise is to modulate the excitation 376 light source (or sample perturbation) at a fixed frequency and use a lock-in amplifier to record 377 the amplitude of the modulated detected photocurrent. Analogously, our FPGA system can 378 enhance the signal-to-noise ratio in single-photon experiments by more than two orders of 379 magnitudes. By resolving this modulation of the signal, it is possible to filter out unwanted 380 background photons and those from other contributions. In the experiment presented here, we 381 demonstrate reliable detection of 3-pulse CARS at count rates of 1 cts/s within a background 382 of more than 300 cts/s. This ratio of $>10^2$ is remarkable for a single-photon experiment, but can 383 increase even further for applications with high backgrounds such as stray light.

Improving the signal-to-noise even further would be possible by enhancing the timing precision to narrow the electronic pulse as discussed above. Increasing the detector count rates beyond the saturation of a single SPAD can be handled by splitting the light intensity over multiple SPADs and connecting them to different channels of the FPGA. In combination with a spectrometer grating, a SPAD array could even then resolve the spectral dependence of the signal.

390 Here, we have demonstrated the working principle of our setup using laser pulses, laser 391 modulation and the delay stage sweep as reference clocks. However, any periodic signal can 392 act as a reference, making our setup attractive for a wide range of experimental research fields. 393 In particular the additional time resolution of the signal during a reference period can enable a 394 plethora of new applications. For example, the light emission from an optoelectronic device 395 induced by an alternating voltage can be tracked in time to characterize the response time of 396 the device. In scanning near-field optical microscopy, an oscillating tip above the sample can 397 provide a reference frequency, both drastically increasing the signal-to-noise and recording the signal as a function of tip-sample distance. Moreover with a dispersive fibre, photons can be
delayed depending on their colour enabling optical time-of-flight spectroscopy [33] with the
80 MHz lock-in frequency. Finally, the switching of photoactive molecules can be resolved in
time with modulated lasers. We suggest this is of particular interest for single-molecule
fluorescence spectroscopy and microscopy.

403 In conclusion, we developed a method to separate different contributions to a signal by 404 resolving the variation of the detected single-photons over the period of a reference signal. The 405 technique relies on continuously recording the arrival time of each photon at a single-photon 406 detector with an FPGA board and comparing it to reference clock time stamps. With this setup, 407 we reconstructed the periodic modulation of a sub-ps excitation laser allowing 98% of stray 408 light and dark counts to be filtered out. The capability of this method was demonstrated in a 409 CARS experiment, where single-photon per second count rates of a nonlinear signal were 410 detected. Due to the high flexibility for different reference clock signals from Hz to MHz 411 frequencies this concept can be applied universally to all single-photon experiments, drastically 412 increasing their signal-to-noise.

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421 **Data availability.** Data underlying the results presented in this paper are available at DOI: (will be provided on acceptance of the paper).

- 423 Supplemental document. See Supplement 1 for supporting content.
- 424

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