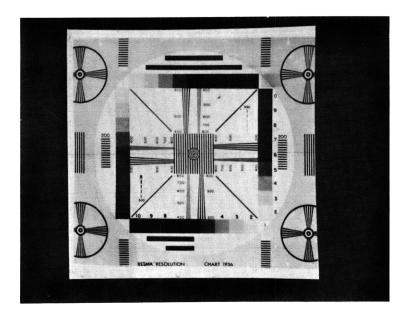
IX. PROCESSING AND TRANSMISSION OF INFORMATION*

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A. PICTURE PROCESSING RESEARCH

We have now made computer tapes from original pictures and played the tapes back for photography. Operations on the IBM 709 computer will begin during the next reporting period.

An example of a 1024×1024 element picture is shown in Fig. IX-1.





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(IX. PROCESSING AND TRANSMISSION OF INFORMATION)

B. NEW SYSTEMATIC DECODING FOR MEMORYLESS CHANNELS

In this report a sequential decoding scheme for random convolutional codes, which is to be used for memoryless channels, is described. The average number of computations does not grow exponentially with n; it is upper-bounded by a quantity proportional to n^2 , for all rates below some cutoff rate R_{comp} (n is the block length).

When this decoding scheme is averaged over a suitably defined ensemble of code words it has an average probability of error with an upper bound whose logarithm is -nE(R). E(R) is dependent on the data rate. (E(R) > 0 for rates below channel capacity.) The decoding scheme is different from other effective decoding schemes such as sequential decoding¹ and low-density parity-check codes.²

The lower bound on R_{comp} of the systematic decoding scheme that is presented in this report is the same as the R_{comp} of sequential decoding for asymmetric channels.

However, in the case of sequential decoding, R_{comp} is valid only for the incorrect subset of code words. The existence of R_{comp} for the correct subset has not yet been established.

Thus, the systematic decoding scheme yields a better bound on the average number of computations for asymmetric channels. (This is not the case when the channel is symmetric, since the modified sequential decoding scheme after Gallager³ may be used.)

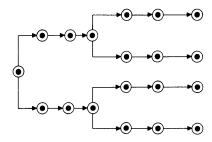


Fig. IX-2. A convolutional tree code.

A convolutional tree code may be viewed topologically as shown in Fig. IX-2. The words are all the directed paths from the input node to the output nodes of the tree (there are no closed paths). From all nontrivial intermediate nodes there emerge ℓ directed links, one for each of ℓ nodes. Let the number of input symbols per code word be n. Let the number of input symbols per link be d. Then the number of links per word is m = n/d (m is the number of information digits per word). In Fig. IX-2, n = 6; m = 2; d = 3; $\ell = 2$.

Reiffen⁴ has shown that the convolutional codes may be generated sequentially.

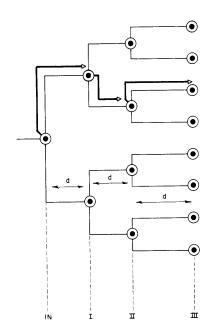
1. The Decoding Procedure

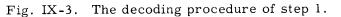
The decoding procedure consists of the following successive operations:

Step 1: The a posteriori probability of each one of the ℓ links of length d that

emerge from the input node to the first ℓ nodes in the tree is computed. The one link that yields the largest a posteriori probability is chosen to represent the corresponding part of the transmitted code word. This detected link connects the input node with one of the ℓ nodes of the next set of nodes (set I in Fig. IX-3).

The same procedure is then repeated with the detected node of set I as a starting point. Thus, the a posteriori probability of each one of the ℓ links emerging from the





node that was previously detected, is now computed, and a decision is made. This procedure is then repeated again and again until termination (i.e., until the detected path reaches one of the output nodes). A metric of the form $D(u, v) = \ln \left[\frac{P(v/u)}{P(v)}\right]$ is then computed. Here, P(v/u) is the a posteriori probability of the complete detected word, u, and P(v) is the probability of the output symbol v.

If D(u, v) is larger than some preset threshold D_0 , a final decision is made and the detection of the first information digit is completed. If $D(u, v) \leq D_0$, the computation procedure is then to go to step 2.

Step 2: The a posteriori probability of each one of the l^2 links of length 2d that emerge from the input node to set II (that consist of l^2 nodes) is computed (Fig. IX-4). The one link that yields the largest a posteriori probability is chosen to represent the corresponding part of the transmitted code word.

The same procedure is then repeated with the detected node of set I as a starting point, and so on. This procedure is continued until termination (i.e., until the detected path reaches one of the output nodes).

D(u, v) is then computed for the detected path and compared with the threshold D_0 . If $D(u, v) > D_0$, a final decision is made with respect to the first information digit. If

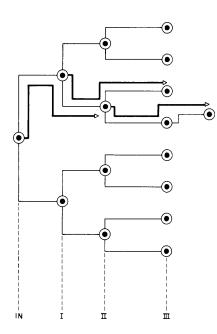


Fig. IX-4. The decoding procedure of step 2.

 $D(u, v) \leq D_0$ the computation procedure is then to go to step 3.

Step 3: This step is similar to step 2, except that the computation length in this case is 3d, and the number of links involved at each stage is l^3 . If no termination occurs at step 3, the computation procedure then reaches step 4, and so on. Following the detection of the first information digit, the whole procedure is repeated for the next digit, and so forth.

a. The Average Number of Computations per Information Digit

The number of computations that are involved in step 1 is equal to $m \cdot l$. The number of computations in step 2 is equal to $(m \cdot l^2)$. The number of computations in step 3 is equal to $(m \cdot l^3)$.

In general, the number of computations in step k is

$$N_k = m \ell^k$$

Let C_k be the condition that no termination occur at step k. Step k will be used only if there were no terminations at all in the previous k - 1 steps. Thus the probability of step k being used is

$$P(k) = P_r(C_1, C_2, C_3, C_4, \dots C_{k-1}).$$
(1)

The average number of computations is given by

$$N = N_{1}P(1) + N_{2}P(2) + \dots + N_{k}P(k) + \dots N_{m}P(m)$$
$$= \sum_{k=1}^{m} N_{k}P(k) = \sum_{k=1}^{m} m \ell^{k}P(k).$$
(2)

b. An Upper Bound on the Average Number of Computations

P(k) may be bounded by

$$P(k) = P_{r}(C_{1}, C_{2}, C_{3}, C_{4}...C_{k-1})$$

$$\leq P_{r}(C_{k-1}).$$
(3)

Thus

$$N = \sum_{k=1}^{m} N_{i} P_{r}(C_{k-1}) = \sum_{k=1}^{m} (m-k+1) \ell^{k} P_{r}(C_{k-1}).$$
(4)

Now let \boldsymbol{u}_k be the code word detected at step k, and let \boldsymbol{u} be the transmitted code word. Then

$$P_{r}(C_{k-1}) = P_{r}(D(u_{k}, v) \leq D_{o})$$

= $P_{r}(D(u_{k}, v) \leq D_{o}; u_{k}=u) + P_{r}(D(u_{k}, v) \leq D_{o}; u_{k}\neq u)$
= $P_{r}(D(u, v) \leq D_{o}; u_{k}=u) + P_{r}(D(u_{k}, v) \leq D_{o}; u_{k}\neq u)$

Thus

$$P_{r}(C_{k-1}) \leq P_{r}[D(u,v) \leq D_{o}] + P_{r}[u_{k}\neq u]$$
(5)

The threshold D_0 is set so as to make

$$P_{r}[D(u,v) \leq D_{o}] \leq e^{-\epsilon(R) \cdot n},$$
(6)

where $-\epsilon(\mathbf{R})$ is a function of the rate **R**.

The number of detected links needed to construct u_k is, as we have shown, (m-k+1). Now, u is equal to u_k only if all of the (m-k+1) links of u_k are equal to the corresponding links of u.

Let e_i be the condition that the a posteriori probability of one of the $(l-1)l^{k-1}$ links emerging from the ith node of u and not including the $(i+1)^{th}$ node of u is greater or equal to that of the corresponding link of u. Then

$$P_r(u_k=u) = P_r(not e_1; not e_2; not e_3; ...; not e_i...; not e_{m-k+1})$$

or

$$P_{r}(u_{k}\neq u) = 1 - P_{r}(u_{k}=u) = P_{r}\begin{bmatrix} m-k+1\\ U\\ i=1 \end{bmatrix}.$$

The probability of the union of events is upper-bounded by the sum of the probabilities of the individual events. Thus

$$P_{r}(u_{k}\neq u) \leq \sum_{i=1}^{m-k+1} P_{r}(e_{i})$$
(7)

The rate per symbol is defined as

$$R = \frac{1}{n} \ln M = \frac{1}{n} \ln \ell^{m} = \frac{1}{n} \ln \ell^{n/d} = \frac{1}{d} \ln \ell,$$

where $\, M \,$ is the number of code words of length n. Thus

$$R = \frac{1}{d} \ln \ell$$
(8)

Fano has shown that for a given rate R and a given word length n the probability of error is bounded by

$$P(e) \leq 2 e^{-E_{opt}(R) \cdot n}.$$
(9)

 $E(R)_{opt}$ is the optimum exponent of the average probability of error and is a function of the rate R. Now, in the case of $P_r(e_i)$, the number of the involved links is $(\ell-1) \ell^{k-1}$ and the length of each link is kd; thus

$$R_k < \frac{1}{dk} \ln \ell^k = \frac{1}{d} \ln \ell = R$$

Therefore

$$P_{r}(e_{i}) \leq 2 e^{-E_{opt}(R)kd}$$
(10)

Thus, by Eqs. 7 and 10,

$$P(u_{k} \neq u) \leq 2m e^{-E_{opt}(R)kd}$$
(11)

Therefore, by Eqs. 5, 6, and 11,

$$P_{r}(C_{k}) \leq e^{-\epsilon(R)n} + 2m e^{-E(R)_{opt}kd}$$
(12)

or

$$P_{r}(C_{k}) \leq 2m e^{-E(R)kd}$$
(13)

because, as we shall show, by Eq. 21, $\epsilon(\mathrm{R}) \leq \mathrm{E}_{\mathrm{opt}}(\mathrm{R})$. Thus

$$P_r(C_{k-1}) \le 2m e^{-E(R)(k-1)d}$$
 (14)

The average number of computations, by Eqs. 4 and 14, is therefore bounded by

$$N \leq 2m^{2} \ell \sum_{k=1}^{m} \ell^{k} e^{-\epsilon(R)(k-1)d}$$
(15)

Now,

$$R = \frac{1}{(k-1)d} \ell^{(k-1)d}.$$

Thus $\ell^{k-1} = e^{(k-1)dR}$. Therefore

$$N \leq 2m^2 \ell \sum_{k=1}^{m} e^{(k-1)d[R-\epsilon(R)]}.$$
(16)

Let R_{comp} be the solution of

$$\mathbf{R} = \boldsymbol{\epsilon}(\mathbf{R}) \tag{17}$$

Then, for all rates below R_{comp} , $R - \epsilon(R) < 0$, and

$$N \leq 2m^{2} \ell \left[\frac{1}{1 - e^{\left[R - \epsilon(R)\right]d}} \right].$$
(18)

The average amount of computations is therefore an algebraic function of m for all rates below $R_{\mbox{comp}}.$

c. Evaluation of R_{comp}

Fano⁵ has shown that

$$E_{opt}(R) = E_{opt}(0) - R; \quad \text{for } R \leq R_{crit}$$
(19)

and

$$E_{opt}(R) \ge E_{opt}(0) - R;$$
 for $R_{crit} \le R < C.$ (20)

Let us set D_0 so as to make ϵ (R) of Eq. 6 equal to

$$\epsilon(\mathbf{R}) = \mathbf{E}_{\text{opt}}(0) - \mathbf{R}_{\text{comp}}.$$
 (21)

Thus, by Eqs. 12 and 21,

$$R_{comp} = E_{opt}(0) - R_{comp}$$

or

$$R_{comp} = \frac{1}{2} E_{opt}(0).$$
⁽²²⁾

Also,

$$\epsilon(\mathbf{R}) = \mathbf{E}_{\text{opt}}(0) - \mathbf{R}_{\text{comp}} = \frac{1}{2} \mathbf{E}_{\text{opt}}(0).$$
(23)

Thus

$$N \leq 2m^{2} \ell \left[\frac{1}{1 - \exp\left\{ d \left[R - \frac{1}{2} E_{opt}(0) \right] \right\}} \right].$$

 $E_{opt}(0)$ is the zero-rate exponent of the upper bound on the average probability of error P_e of the same channel, when the optimal decoding scheme is used.⁵

2. The Average Probability of Error

Let u be the transmitted code word. Let v be the corresponding channel output vector. Let u' be one of the $\frac{M(\ell-1)}{\ell}$ code words which starts with an information letter other than that of u. The probability of error is bounded by

$$P_{e} \leq P_{r}(D(u, v) \leq D_{o}) + \frac{M(\ell-1)}{\ell} P_{r}(D(u', v) > D_{o}; D(u, v) > D_{o})$$

$$(24)$$

Thus

$$P_{e} \leq P_{r}(D(u,v) \leq D_{o}) + e^{nR} P_{r}(D(u',v) > D_{o}; D(u,v) > D_{o})$$
(25)

or

$$P_{e} \leq P_{r}(D(u, v) \leq D_{o}) + e^{nR} P_{r}(D(u', v) > D; D(u, v) > D_{o})$$
 (26)

Let

$$D(u, v) = \ln \frac{P(v/u)}{P(v)}$$
$$D(u', v) = \ln \frac{P(v/u')}{P(v)}$$

where

$$P(v/u) = \prod_{i=1}^{n} P(y_i/x_i)$$

 and

$$P(v) = \prod_{i=1}^{n} P(y_i).$$

 $P(\textbf{y}_i/\textbf{x}_i)$ is the probability that the output is $\textbf{y}_i,$ given that the input symbol \textbf{x}_i was transmitted.

$$\sum_{\mathbf{x}_{i}} \mathbf{P}(\mathbf{x}_{i}) \mathbf{P}(\mathbf{y}_{i}/\mathbf{x}_{i}) = \mathbf{P}(\mathbf{y}_{i})$$

Thus

$$D(u, v) = \sum_{i=1}^{n} d_{i}(x_{i}, y_{i}),$$

where

$$d_{i}(x_{i}, y_{i}) = \ln \frac{P(y_{i}/x_{i})}{P(y_{i})}.$$

Thus by the use of the Chernoff Bound, 6

$$\begin{split} & P_{r}(D(u, v) \leq D_{o}) \leq e^{n(\mu(s) - s\mu'(s))}; \quad s \leq 0 \\ & P_{r}(D(u', v) \leq D_{o}) \leq e^{n(\gamma(t) - t\gamma'(t))}; \quad t \geq 0 \end{split}$$

where

$$\mu(s) = \sum_{xy} P(x) P(y/x) e^{sd(x, y)}$$
$$\gamma(t) = \sum_{xxy} P(x) P(x') P(y/x) e^{sd(x', y)}$$

or

$$\mu(s) = \sum_{xy} P(x) P(y/x)^{1+s} P(y)^{-s}$$

 and

$$\gamma(t) = \sum_{x'y} P(x') P(y)^{1-t} P(y/x')^{t}$$
$$= \sum_{xy} P(x) P(y)^{1-t} P(y/x)^{t}.$$

Also,

$$\mu'(s) = \gamma'(t) = \frac{D_o}{n}.$$

If we let t = 1 + s, it can be shown that $\gamma'(t) = \mu'(t)$ (also $\gamma(t) = \mu(s)$). Thus

$$\begin{split} & P_r(D(u, v) \leq D_o) \leq e^{n(\mu(s) - s\mu'(s))} \\ & P_r(D(u', v) \geq D_o) \leq e^{n(\mu(s) - (1+s)\mu'(s))}, \end{split}$$

where

$$\mu'(s) = \frac{D_o}{n}$$

$$\mu(s) = \sum_{xy}^{\infty} P(x) P(y/x)^{1+s} P(y)^{-s}.$$

Thus

$$P_{e} \leq e^{n(\mu(s) - s\mu'(s))} + e^{n(R + \mu(s) - (1 + s)\mu'(s))}.$$
(27)

Now, by Eqs. 6 and 23, $D_0^{}$ is set so as to get

$$P_{r}(D(u,v) \le D_{o}) \le e^{n(\mu(s)-s-'(s))} = e^{-n\epsilon(R)},$$
 (28)

where

$$\epsilon(\mathbf{R}) = \frac{1}{2} \mathbf{E}_{opt}(0); \text{ for all } \mathbf{R} \leq \mathbf{R}_{comp}$$
 (28a)

and, as shown by Fano,⁵

$$E_{opt}(0) = E_{opt}(R) \Big|_{R=0} = -\ln \sum_{y} \left[\sum_{x} P(x) P(y/x) \frac{1}{2} \right]^{2}.$$
 (28b)

We shall now prove that, once D_0 is set so as to make $-\mu(s) + s\mu'(s) = \epsilon(R) = \frac{1}{2} E_{opt}(0)$, as in Eq. 28, then

$$-\mu(s) + (1+s) \ \mu'(s) \ge \frac{1}{2} E_{opt}(0).$$
⁽²⁹⁾

PROOF: The minimum of $\{(2s+1)\mu'(s)-2\mu(s)\}$ occurs at that s for which

$$\frac{\partial}{\partial s} \left[(2s+1)\mu'(s) - 2\mu(s) \right] = 0$$
(2s+1) \mu'(s) + 2\mu'(s) - 2\mu'(s) = 0
Thus s = $-\frac{1}{2}$.

Also,

$$\left[(1+2s)\mu'(s)-2\mu(s)\right]'' \bigg|_{s=-1/2} = 2\mu''\left(-\frac{1}{2}\right) \ge 0,$$

since $\mu''\left(-\frac{1}{2}\right)$ is the variance (see Fano⁵) of a random variable. Thus $s = -\frac{1}{2}$ is indeed a minimum point. Thus

$$[(2s+1)\mu'(s)-2\mu(s)] \ge -2\mu(-\frac{1}{2}).$$
(30)

Now

$$\mu\left(-\frac{1}{2}\right) = \ln \sum_{xy} P(x) P(y)^{1/2} P(y/x)^{1/2}$$

and therefore

$$2\mu\left(-\frac{1}{2}\right) = \ln\left[\sum_{xy} P(x) P(y)^{1/2} P(y/x)^{1/2}\right]^2.$$

Thus

$$2\mu\left(-\frac{1}{2}\right) = \ln\left\{\sum_{y} f(y) P(y)^{1/2}\right\}^{2}$$

where $f(y) = \sum_{x} P(x) P(y/x)^{1/2}$. By the Schwarz inequality,

$$\left\{\sum_{y} f(y) P(y)^{1/2}\right\}^2 \leq \sum_{y} f(y)^2 \sum_{y} P(y)$$

Thus

$$2\mu\left(-\frac{1}{2}\right) \leq \sum_{\mathbf{y}} \mathbf{P}(\mathbf{y})^2 \sum_{\mathbf{y}} \mathbf{P}(\mathbf{y}) = \sum_{\mathbf{y}} \mathbf{P}(\mathbf{y})^2,$$

since $\sum_{y} P(y) = 1$. Therefore, by Eq. 28b,

$$2\mu\left(-\frac{1}{2}\right) \leq \ln \sum_{y} \left[\sum_{x} P(x) P(y/x)^{1/2}\right]^2 = -E_{opt}(0).$$

Thus $-2\mu\left(-\frac{1}{2}\right) \ge E_{opt}(0)$, and therefore, by Eq. 30,

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$$-[\mu(s)-s\mu'(s)+\mu(s)-(1+s)\mu'(s)] \ge E_{opt}(0).$$
(31)

But, by Eq. 28,

$$-E(R) = \mu(s) - s\mu'(s) = -\frac{1}{2}E_{opt}(0).$$

Therefore, by Eq. 31,

Thus, by Eqs. 27, 28, and 32,

$$P_{e} \leq e^{-\frac{1}{2}E_{opt}(0)n} + e^{n\left(R - \frac{1}{2}E_{opt}(0)\right)}$$
$$P_{e} \leq 2 e^{n\left(R - \frac{1}{2}E_{opt}(0)\right)} = 2 e^{n\left(R - R_{comp}\right)}$$
$$P_{e} \leq 2 e^{n\left(R - R_{comp}\right)}.$$

If D_0 is set so as to minimize the probability of error by making $R + \mu(s) - (1+s)\mu'(s) = \mu(s) - s\mu'(s) = E(R)$, Shannon⁷ has shown that

 $P_{o} \leq 2 e^{-nE(R)}$

where E(R) > 0 for R < C, and $E(0) > \frac{1}{2} E_{opt}(0)$. However, R_{comp} is then lower-bounded by $R_{comp} \ge \frac{1}{2} E(0) > \frac{1}{4} E_{opt}(0)$.

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J. Ziv

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