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# Implementation and Evaluation of a 10 Gbps Real-time FSO Link

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**Abstract—** In this paper, we present and experimentally evaluate a real-time 10 Gbps free-space optical (FSO) link under varying atmospheric conditions. In bandwidth-craving wireless technologies due to the ubiquitous consumption by internet of things devices and requiring high data rate online services, unlicensed FSO systems can be a promising candidate to satisfy the network capacity of the existing data communications technologies. In this work, we verify the experiment using small form-factor pluggable transceivers mounted FPGA as a FSO transmitter and receiver. Here, a high-speed single FSO link is proposed and its performance under turbulence and fog conditions using the dedicated indoor atmospheric chamber is evaluated. We show that the proposed system under the turbulence condition with a scintillation index of 0.35 offers the same data rate as the link under a clear channel, while the bit error rate increases from  $10^{-12}$  to  $5 \times 10^{-4}$ .

**Keywords—** FSO, Fog, Turbulence, Last Mile Access

## I. INTRODUCTION

As a branch of optical wireless communications (OWC), free-space optics (FSO) has become increasingly attractive to numerous applications in several environments [1, 2]. This has been driven by the ability to compensate for the required bandwidth that radio frequency (RF) based wireless communications systems are unable to provide as well as the ease, and time of installation that optical fiber (OF) communications cannot achieve. Due to its massive license-free spectrum (up to  $\cong 400$  THz), high data rates in the order of terabits/second, backbone network capability, inherent security at the physical layer, and insignificant inter-channel interference, FSO stands as the preeminent candidate to unlock the mobile spectrum congestion (bandwidth bottleneck) issue, which has been predicted to increase a thousand-fold over the next decade [3]. While 2-10% of the global power consumption is responsible by the data communication technologies, energy-efficient FSO systems are also known for potentially being economically green compared to RF [3]. In the race for reaching high data flow ( $> 10$  Gbps) over short distances, also known as ‘last mile’ or ‘last kilometer’ access, optical communications present significant advantages when compared with the radio even in the millimeter-wave (mmW) band [3]. High performance of data propagation and highly

directional FSO communications systems can be deployed as part of back and front haul traffic in areas where cable-based fixed links are not economical [4]. One of the promising economic reasons for this is the availability of components, which are also extensively used in the matured OF communications technology. In the deployment of future wireless networks (i.e., fifth generation (5G) and beyond), FSO (both visible and infrared bands) have an important role in addressing the growing security threat due to its increase in types of services and devices used in broadband access. Furthermore, FSO can be easily implemented, particularly for the last mile access networks, where high data rates communications over a few kilometers of transmission link span are required [3].

Recently, commercially available technologies of mmW and Terragraph by Facebook offer up to 1 Gbps throughput over a range of 150 m [5]. However, the commercial FSO systems can provide much higher data rates of  $\leq 10$  Gbps over a link span of up to 1.5 km [3, 5]. FSO systems employing advanced modulation and multiplexing schemes, potentially achieving up to hundreds of Tbps employing orbital angular momentum, wave division multiplexing, mode division multiplexing, and space division multiplexing have been reported in [6, 7] and [8]. However, most of the systems are costly and complex with off-line measurements and analysis. In this paper, we present a real-time 10 Gbps FSO system implementation based on small form-factor pluggable (SFP) mounted xilinx field programmable gate arrays (FPGA), which can be used in both last meter/last mile access, and presents a potential replacement of mmW technology, that can promptly be integrated to the backbone OF networks as in Terragraph. The proposed system is cost effective, robust, and easily assembled. We evaluate the performance of the system under various atmospheric conditions using a dedicated indoor atmospheric chamber

The paper is organized as follows: Section 2 describes the FSO system and provides all of the design considerations. Section 3 presents the results and discussion of the measured data. Finally, Section 4 concludes the paper.

## II. SYSTEM DESIGN

The schematic block diagram of the experimental setup is presented in Fig. 1. The transmitter (Tx) unit consists of Zynq ultrascale+ FPGA for generating a 10 Gbps optical link at 1550 nm through a gigabit transceiver (GTH) port wired to a SFP+

2×2 quad-module connector. An erbium doped fiber amplifier (EDFA) is used to amplify the optical signal before being transmitted across the free space channel using a single mode fiber (SMF) integrated kinematic collimator mounting adaptor. Note that, the kinematic adaptor provides pitch and yaw adjustment, thus allowing the collimated beam to be easily steered while aligning the system.

The propagated signal is coupled into multimode fiber (MMF) with a core size of 105  $\mu\text{m}$  using a condenser/convex lens with the effective focal length (EFL) of 44 mm. The output of the MMF is injected into the SFP module connector mounted on the FPGA board to make a real-time bit by bit comparison to estimate the received bits and, therefore the BER. Extracted information of the link from the FPGA connected to the PC is displayed in Vivado® Xilinx using the customizable LogiCORE™ IP Integrated bit error rate tester (IBERT) core for UltraScale+ architecture GTH transceivers. UltraScale™ GTH transceivers offer up to 16.3 Gbps with a maximum optical transmit power and receiver (Rx) sensitivity of 4 dBm and  $>-16$  dBm, respectively [9]. While the typical plug and play 10GBASE SFPs provide the same transmit and receive power levels with the data rate of 10.312 Gbps [10]. The maximum link margin loss of the proposed FSO link is up to  $\sim 20$  dB, which is dominated by the loss due to misalignment and optics in our case. The green laser is used to estimate the fog attenuation experienced by the system and on off keying (OOK) modulated red laser is used to estimate the scintillation index. The temperature gradient measurement is done by 17 distributed temperature sensors as shown in Fig. 1.

In the context of optical communications, the received signal is given as:

$$y(t) = h(t) * x(t) + n(t), \quad (1)$$

where  $x$  is the transmitted signal,  $n$  is additive white Gaussian noise (AWGN) with variance  $\sigma_n^2$ ,  $*$  is the convolution symbol and  $h$  is the channel gain, which is given by:

$$h = h_a h_t, \quad (2)$$

where  $h_a$  and  $h_t$  are the attenuation constants due to the atmospheric channel and turbulence, respectively. For an FSO link,  $h_a$  is given by Beer's law as [11]:

$$h_a = e^{-\gamma l}, \quad (3)$$

where  $l$  is the link length in m and  $\gamma$  is the attenuation coefficient in  $\text{m}^{-1}$ . The attenuation coefficient  $\gamma$  models the atmosphere and is the combination of absorption and scattering coefficients expressed as a function of wavelength.  $\gamma$  is given by [11]:

$$\gamma(\lambda) = \alpha_m(\lambda) + \alpha_a(\lambda) + \beta_m(\lambda) + \beta_a(\lambda), \quad (4)$$

where,  $\alpha_m(\lambda)$  and  $\alpha_a(\lambda)$  are the molecular and aerosol absorption coefficients, respectively, and  $\beta_m(\lambda)$  and  $\beta_a(\lambda)$  are the molecular and aerosol scattering coefficients, respectively. The transmission wavelength in FSO is selected such that, it coincides with the atmospheric transmission window [12] hence, scattering governs the atmospheric attenuation coefficient and as a result  $\gamma(\lambda) \approx \beta_a(\lambda) \cdot \beta_m(\lambda)$ , which is expressed as [13]:

$$\beta_a(\lambda) = \frac{3.91}{V} \left( \frac{\lambda}{550 \text{ nm}} \right)^{-q}, \quad (5)$$

where,  $V$  is the meteorological visibility in km and  $q$  is the size distribution of scattering particles, for which Kruse model is considered in this paper, as given by [14]:

$$q = \begin{cases} 1.6 & V > 50 \text{ km} \\ 1.3 & 6 \text{ km} < V < 50 \text{ km} \\ 0.16V + 0.34 & 1 \text{ km} < V < 6 \text{ km} \\ V - 0.5 & 1 \text{ km} < V < 1 \text{ km} \\ 0 & V < 0.5 \text{ km} \end{cases}, \quad (6)$$

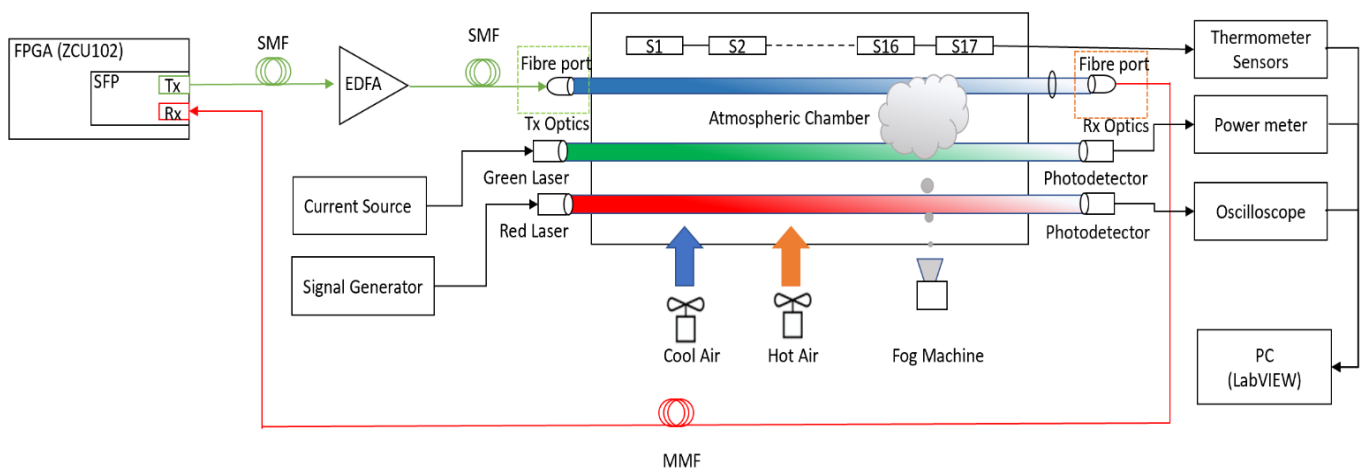


Fig. 1. Experimental setup.

The optical intensity  $I$  of a wave propagating in a turbulence channel undergoes a random fading effect with the normalized variance or the scintillation index, which is given as [15]:

$$\sigma_I^2 = \frac{\langle I^2 \rangle - \langle I \rangle^2}{\langle I \rangle^2}, \quad (7)$$

where  $\langle \cdot \rangle$  denotes the ensemble average equivalent to long-time averaging with the assumption of an ergodic process. Assuming plane wave propagation,  $\sigma_I^2$  is given by [16]:

$$\sigma_I^2(D) = \exp \left[ \frac{0.49 \sigma_R^2}{\left( 1 + 0.653d^2 + 1.11\sigma_R^{\frac{12}{5}} \right)^{\frac{7}{6}}} + \frac{0.51 \sigma_R^2 \left( 1 + 0.69 \sigma_R^{\frac{12}{5}} \right)^{-\frac{5}{6}}}{\left( 1 + 0.9d^2 + 0.621 d^2 \sigma_R^{\frac{12}{5}} \right)^{\frac{12}{5}}} \right] - 1, \quad (8)$$

where  $d = \frac{D}{2} \sqrt{\frac{k}{l}}$  is the circular aperture scaled by Fresnel zone provided,  $k$  is the wavenumber and  $D$  is the Rx aperture diameter.  $\sigma_R^2$  is Rytov variance showing the strength of the turbulence is expressed by:

$$\sigma_R^2 = 1.23 C_n^2 k^{7/6} l^{11/6}, \quad (9)$$

The strength of turbulence can be classified as weak ( $\sigma_R^2 < 1$ ), moderate ( $\sigma_R^2 \cong 1$ ) and strong ( $\sigma_R^2 > 1$ ) [17]. For weak to moderate we can assume that  $\sigma_I^2 = \sigma_R^2$ .  $C_n^2$  is the refractive index structure parameter and is expressed as [18]:

$$C_n^2 = \left[ 79 \times 10^{-6} \frac{P}{T^2} \right] C_T^2, \quad (10)$$

where,  $P$  is the atmospheric pressure in mbar,  $T$  is temperature in kelvin and  $C_T$  is the temperature structure parameter, which is given by [18]:

$$C_T = \sqrt{\langle \delta T^2 \rangle} r^{-1/3}, \quad (11)$$

where  $\delta T$  refers to the temperature differential between two points separated by a distance  $r$ . For weak turbulence, the turbulence induced fading gain is given as [19]:

$$h_t = e^{\chi + iS}, \quad (12)$$

where  $S$  is the phase and  $\chi$  is the fading log-amplitude having normal distribution with mean  $\mu$  and variance  $\sigma_\chi^2$ . The signal to noise ratio at the Rx for OOK is expressed as [20]:

$$\text{SNR} = \frac{(RP_{\text{rx}})^2}{\sigma_n^2}, \quad (13)$$

where  $R$  is the responsivity of the Rx in A/W and  $\sigma_n^2$  is the combination of shot noise and thermal noise modelled as AWGN variance. The received power can be expressed as a function of the transmit power and the system losses, which is given by:

$$P_{\text{rx}} = 10^{\frac{(L_{\text{Geo}})}{10}} \times 10^{\frac{(L_{\text{Atm}})}{1000} \frac{l}{1000}} \times 10^{\frac{(L_{\text{Misc}})}{10}} P_{\text{tx}}, \quad (14)$$

where  $P_{\text{tx}}$  is the transmitted power,  $L_{\text{Geo}}$  is the geometrical loss,  $L_{\text{Atm}}$  is the atmospheric loss and  $L_{\text{Misc}}$  is the miscellaneous loss including coupling losses due to optics to

**Table 1. Parameters and Specification of the system.**

Parameter		Specification
<b>Transmitter</b>	Tx power ( $P_{\text{tx}}$ )	5 dBm
	Tx optics (Threaded Kinematic adapter)	$D_{\text{tx}}$ : 6.5 mm
	SFP+ Tx power	4 dBm
	Tx beam divergence ( $\theta$ )	0.279 mrad
<b>Receiver</b>	FPGA, SFP+ modules (2×2 quad-connector) - ZCU102 (Ultrascale +) evaluation kit	Throughput: 10.3125Gbps Clock frequency: 156.2 MHz
	Rx power	-6 dBm
	Rx optics - Lens	Effective focal length: 44 mm $D_{\text{rx}}$ : 50 mm
	SFP+ sensitivity	-16 dBm
<b>Other parameters</b>	SFP+ wavelength	1550 nm
	Tx core diameter (SMF)	8.2 $\mu\text{m}$
	Rx core diameter (MMF)	105 $\mu\text{m}$

fiber and optics coupling. Geometric loss occurs due to the spreading of the transmitted beam, increasing to a size larger than the Rx aperture  $D_{rx}$ . The geometric loss in dB is given by [21]:

$$L_{Geo}(dB) = 20 \log_{10} \left( \frac{D_{rx}}{D_{tx} + l \times \theta} \right), \quad (15)$$

where  $D_{tx}$ ,  $D_{rx}$  are the Tx and the Rx aperture diameters in meters, respectively and  $\theta$  is the Tx beam divergence.  $L_{Atm}$  is the atmospheric attenuation in dB/km and is given by [20]:

$$L_{Atm} \left( \frac{dB}{km} \right) = 4.343 \beta_a(\lambda). \quad (16)$$

### III. RESULTS AND DISCUSSIONS

We have experimentally investigated the 10 Gbps FSO link shown in Fig. 2. Fig. 2(a) pictures a foggy chamber and temperature sensors. Fog is generated using the fog machine and the turbulence regime is created using hot and cool fans injected into the chamber. The temperature gradient

measurement within the atmospheric chamber was taken by using 17 distributed temperature sensors S1 to S17 with a spacing of  $\sim 25$  cm each along the chamber. Utilization of temperature sensors is to estimate the channel refractive index structure coefficient,  $C_n^2$  within the chamber by following (10). The alignment and performance analysis of the 10 Gbps link is carried out in 7.2-meter atmospheric chamber. Fig. 2(b) depicts the Tx unit consisting of a signal generator coupled to a red laser (685 nm). Both red and green lasers are powered independently. A 10 Gbps non-return to zero on/off keying (NRZ-OOK) signal format is generated via the FPGA and an EDFA is used to amplify the optical signal power before being transmitting through the SMF cable, see Fig. 2(c). At the Rx side, MMF has been used to couple the signal into the SFP transceiver module. Additionally, two photodetectors (PD) were used at the end of the chamber to capture a red laser and a green laser. A red laser beam is transmitted through the chamber and after detection captured by sampling oscilloscope to estimate the turbulence effect on channel in terms of scintillation index  $\sigma_I^2$ .

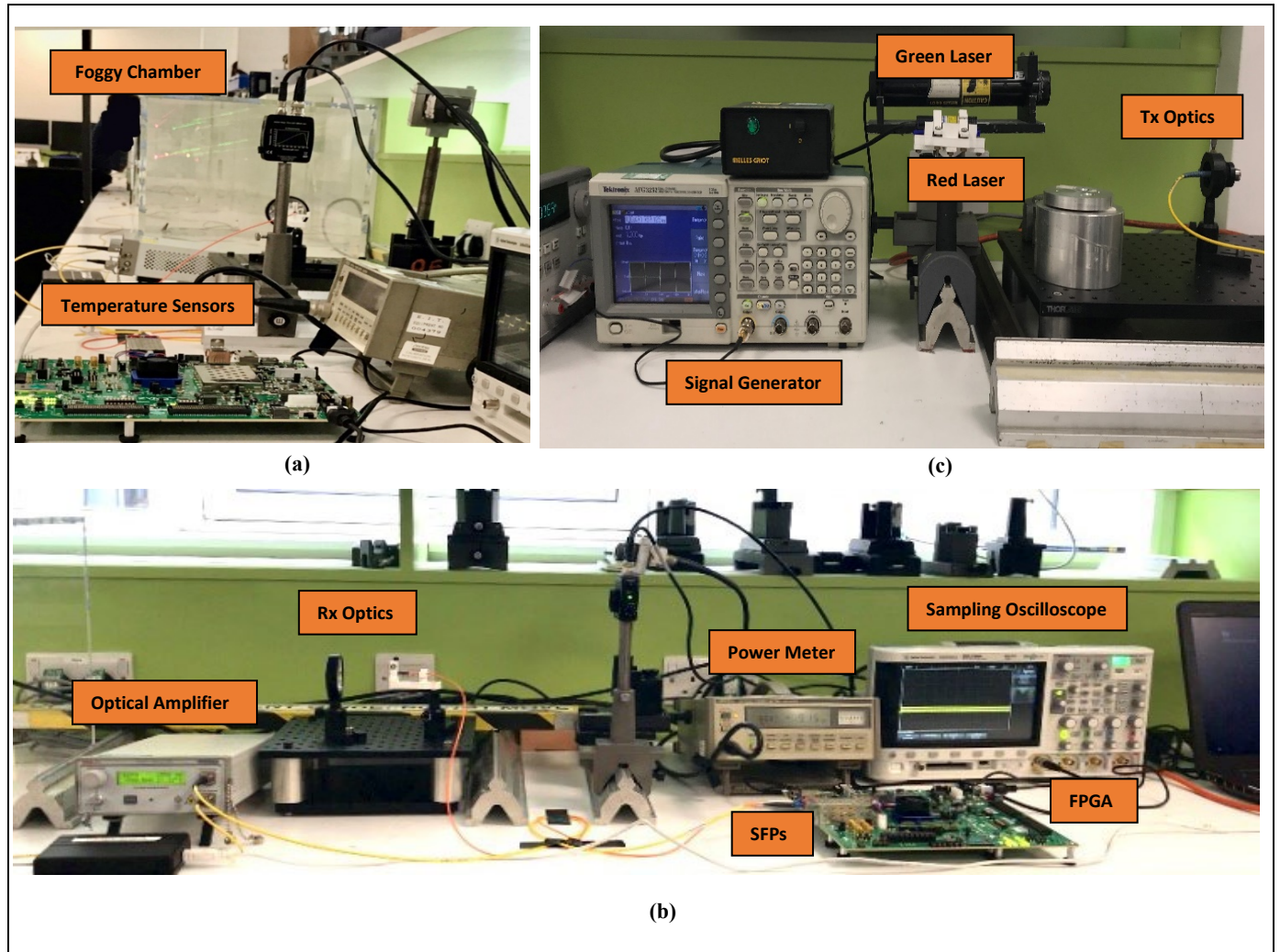


Fig. 2 (a) Channel setup when fog is injected, and (b) Rx setup with power meter connected to the green laser photodetector to estimate visibility and sampling oscilloscope connected to measure scintillation index through red laser photodetector. Oscilloscope, power meter, and temperature sensors are connected to PC to read the measurements via LabVIEW. SFP transceiver mounted FPGA is connected to PC. (c) Tx setup with green and red laser to estimate the atmospheric and turbulent condition in the chamber.

The green laser and a photodetector connected to the power meter is then utilized to calculate the visibility ( $V$ ) of the chamber is determined by following the method outlined in [19] and (5) under foggy conditions within the chamber. The channel condition and parameters are analysed using Vivado, which provides important characteristics of the link such as BER, number of errors, the data rate, and link stability. We observed the clear channel and the link parameters for approximately 90 seconds while an average of  $3.8 \times 10^{13}$  bits were transmitted within this period. While observing the system we carried out the BER measurement of the link, which was  $2.6 \times 10^{-12}$  and the maximum throughput of 10.312 Gbps. Subsequently, we performed the measurements under different

conditions of fog and turbulence. Considering the clear channel  $V > 20$  km, a light fog regime was generated within the chamber to increase the attenuation of the link to  $\sim 0.2$  dB, thus resulting in increased BER to  $5 \times 10^{-5}$ . We reduced  $V$  further to  $\sim 8.5$  m while the attenuation was increased to 13 dB with the BER of  $1.9 \times 10^{-4}$ .

In the next experiment, we tested turbulence within the chamber by injecting cool and hot air using two fans and two heaters. With weak turbulence of 0.35 of  $\sigma_I^2$ , the availability of the link and data rate remains intact, however, BER increased. In consideration of the realistic link length for last mile access and short/mid-range distance, we have made further analysis of the link spans of 50, 160, and 200 m. With the observation in

**Table 2: Parameters and specification of the system.  $C_n^2$  and  $\sigma_I^2$  denotes the refractive index structure coefficient and the scintillation index, respectively.**

Link Span	Parameter	Value	BER
$l = 7.2$ m (Experimental)	Visibility (clear)	20 km	$2.6 \times 10^{-12}$
	Attenuation	0 dB	
	Visibility	385 m	$5 \times 10^{-5}$
	Attenuation	0.2 dB	
	$C_n^2$ $\sigma_I^2$	$6.8 \times 10^{-11} \text{m}^{-\frac{2}{3}}$ 0.16	$1.5 \times 10^{-5}$
Visibility Attenuation	$\sim 8.5$ m 13 dB		
$l = 50$ m (Simulation)	$C_n^2$ $\sigma_I^2$	$1.45 \times 10^{-10} \text{m}^{-\frac{2}{3}}$ 0.35	$5 \times 10^{-4}$
	Visibility Attenuation	4.2 km 0.2 dB	
	$C_n^2$ $\sigma_I^2$	$1.95 \times 10^{-12} \text{m}^{-\frac{2}{3}}$ 0.16	$1.5 \times 10^{-5}$
	Visibility Attenuation	65 m 13 dB	
$C_n^2$ $\sigma_I^2$	$4.27 \times 10^{-12} \text{m}^{-\frac{2}{3}}$ 0.35	$5 \times 10^{-4}$	
Visibility Attenuation	13.5 km 0.2 dB		
$l = 160$ m (Simulation)	$C_n^2$ $\sigma_I^2$	$2.31 \times 10^{-13} \text{m}^{-\frac{2}{3}}$ 0.16	$6.6 \times 10^{-4}$
	Visibility Attenuation	208 m 13 dB	
	$C_n^2$ $\sigma_I^2$	$5.06 \times 10^{-13} \text{m}^{-\frac{2}{3}}$ 0.35	$5.2 \times 10^{-3}$
	Visibility Attenuation	16.9 km 0.2 dB	
$C_n^2$ $\sigma_I^2$	$1.54 \times 10^{-13} \text{m}^{-\frac{2}{3}}$ 0.16	$6.7 \times 10^{-3}$	
Visibility Attenuation	261 m 13 dB		
$l = 200$ m (Simulation)	$C_n^2$ $\sigma_I^2$	$3.36 \times 10^{-13} \text{m}^{-\frac{2}{3}}$ 0.35	$2.1 \times 10^{-2}$
	Visibility Attenuation	$\approx 1$ (0.24)	
	$C_n^2$ $\sigma_I^2$	$5.06 \times 10^{-13} \text{m}^{-\frac{2}{3}}$ 0.35	$5.2 \times 10^{-3}$
	Visibility Attenuation	16.9 km 0.2 dB	
$C_n^2$ $\sigma_I^2$	$1.54 \times 10^{-13} \text{m}^{-\frac{2}{3}}$ 0.16	$6.7 \times 10^{-3}$	
Visibility Attenuation	261 m 13 dB		
$C_n^2$ $\sigma_I^2$	$3.36 \times 10^{-13} \text{m}^{-\frac{2}{3}}$ 0.35	$2.1 \times 10^{-2}$	
Visibility Attenuation	$\approx 0.5$ (0.32)		

accordance with  $V$ ,  $\sigma_I^2$  and  $C_n^2$ , we considered similarly the attenuated channel with 0.2 and 13 dB as the reference and calculated  $V$ . For the turbulence channel, assuming that the scintillation index (i.e., the intensity of turbulence) in both links under weak and moderate turbulence regimes are 0.16 and 0.35, respectively, we determined  $C_n^2$  for the links.

Table 2 describes the channel under the different controlled atmospheric experimental conditions and the parameters of the channel impairments. Even though, the attenuation of the clear channel is 0 dB, note that the additional losses due to the Tx and the Rx optics and misalignment are kept to the minimum level. During the experiment, we observed a number of sources, which contributed additional losses including (i) coupling free space optical source into the optical fibre at the Rx unit; and (ii) light coupling from SFP to SMF (Tx) and MMF to the PD at the Rx. Note, SFP requires the use of an attenuator to stabilize the optical power and to make a connection with the module. In this work, we used a 105  $\mu\text{m}$  core size MMF at the Rx, see Table 1, which is disadvantageous in the turbulent channel as the smaller the core size the higher the loss due to beam wandering misalignment [17]. In the experimental setup (i.e., a length of 7.2 m) and up to 150 m links, the beam spot size compared with the Rx aperture is small. Therefore, the geometrical loss  $L_{\text{Geo}}$  is not considered. In addition, we assumed that the BER for the FSO link for up to 150 m is the same as the 7.2 m link span as described in Table 2. For the 7.2 m link span, the Tx power  $P_{\text{Tx}}$  and Rx power  $P_{\text{Rx}}$  are 5 and -6 dBm, respectively, thus resulting in the total loss (i.e., coupling, misalignment, and optics) of 11 dB. For the 160 m link,  $L_{\text{Geo}}$  is introduced and  $L_{\text{Geo}}$  is increased accordingly with the link span, see Table 2. With the introduction of fog with 13 dB of attenuation, both 160 and 200 m links are intolerable. This is due to the additional  $L_{\text{Geo}}$  values of 0.2 and 1.9 dB for the 160 and 200 m links, respectively, which exceeds the system margin. The link unavailability is attributable to the aperture of the Rx diameter  $D_{\text{Rx}}$  size being smaller than the Tx beams footprint, which increases with distance, thus wasted "overfill" energy. Therefore, the coupling fibre core size and larger  $D_{\text{Rx}}$  or smaller beam transmitted divergence (collimated) according to the turbulence intensity and link distance are suggested to be considered to reduce this additional loss. Increasing  $P_{\text{Tx}}$  will also ease the attenuation due to the fog over a longer link span. Considering that, the misalignment loss is negligible,  $P_{\text{Rx}}$  for 160 and 200 m is 4.8 and 3.1 dBm, respectively under clear channel with geometrical loss. With the misalignment loss included in the system,  $P_{\text{Rx}}$  will then change on -6.2 and -7.9 dBm, respectively.

#### IV. CONCLUSION

In this work, we demonstrated an experimental real-time FPGA implementation of a 10 Gbps FSO link and investigated its performance under different atmospheric conditions. We made the bit by bit comparison using customizable LogiCORE™ IP IBERT core for UltraScale+ architecture GTH transceivers. We discussed the causes incurring additional loss of the system which can be addressed by using different or larger Rx aperture and smaller collimated Tx beam divergence.

Additionally, we considered 50, 160, and 200 m link spans under the same atmospheric conditions and investigated the link performances. We showed that the proposed FSO link can be improved by reducing misalignment and geometrical losses. We outlined that, the proposed FSO link under heavy fog condition offers the same data rate but of course increased BER (i.e., increasing from  $10^{-12}$  to  $10^{-4}$ ) over a 7.2 m link range. Under weak turbulence, we observed that the link BER is increased from  $10^{-12}$  to  $10^{-4}$  due to both phase and amplitude variations.

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