

Components of Purity to Describe the Polarimetric State of a 3-D Field Within the Reverberating Chamber

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Abstract—Reverberating chambers (RCs) are electrically large microwave enclosures in which a random electromagnetic 3-D, i.e., nonplanar, field is generated. To characterize such 3-D fields, a proper polarimetric decomposition is requested. In this article, a new set of parameters, namely the components of purity (CP), first proposed to the electromagnetic compatibility community, is presented. The CP parameters, which stem from the so-called trivial decomposition, are able to classify any 3-D polarimetric field into two states: regular and nonregular. Such characterization is of practical relevance, e.g., for RC isotropy testing.

Index Terms—Polarimetry, reverberating chamber (RC).

I. INTRODUCTION

REVERBERATING chambers (RC) are reliable and cost-efficient facilities to perform electromagnetic compatibility (EMC) tests [1] since they emulate effective and reliable complex electromagnetic environments [2], [3]. RCs are used for a wide range of applications, e.g., testing antenna performance [4], [5], shielding effectiveness [6], [7], radiated emission test [8], [9], emulating wireless propagation channels [10]–[14], and biomedical measurements [15].

The RC is an electrically large metallic enclosure wherein the electromagnetic field is randomized using a stirring process [16], [17] that generates a random 3-D, i.e., nonplanar, electromagnetic field. To fully characterize its polarimetric properties, specialized descriptors must be used [18]–[23]. In fact, the nonplanar nature of the electromagnetic field calls for an appropriate 3-D decomposition [24]–[28], which leads to physically consistent parameters.

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Any 3-D decomposition starts with the estimate of the 3-D coherency matrix (\mathbf{R}) [29], [30]. Then, compact parameters that carry on information about the state of the electromagnetic field are defined. Among them, the 3-D degree of polarimetric purity P_{3D} [24] (bounded between 0 and 1) can be defined to measure the closeness of the polarization state to a fully polarized one (also known as pure state) [31]. Recently, the so-called index of polarimetric purity (IPP) decomposition was introduced in optics to decompose P_{3D} according to parameters stemming from the trivial decomposition [24]. The IPP parameters, which were first presented and tested for RC electromagnetic field characterization in [29], decouple the information carried out P_{3D} into two parts: first, the stability of the polarization ellipse (P_1) that preserves its physical meaning of 2-D degree of polarization and second, the stability of propagation direction (P_2) that is inherently related to 3-D states [29]. However, the IPP parameters are not able to represent important 3-D electromagnetic behaviors [30]–[37].

With this respect, a new set of parameters are introduced in [30]–[37] and they are hereafter first presented to the EMC community and applied to a set of meaningful microwave measurements carried out in the RC of the Università degli Studi di Napoli Parthenope, Naples, Italy. The new set of parameters is known as component of purity (CP) and splits P_{3D} into three parameters, which now allow to represent a particular class of polarization states, i.e., the nonregular, whose polarimetric behavior cannot be described by the IPP.

In this article, we first show that the nonregular 3-D polarization states may occur in the loaded RC and they can only be represented by means of the CP decomposition. Hence, they can be used to make more accurate field isotropy testing in the RC that, as suggested in [38], should be performed using 3-D polarimetric approaches instead of using the conventional integrated 3-D linear measurements [1]. In fact, the conventional IEC 61000-4-21 standard [1] procedure is considered to be not enough when dealing with loaded RC configurations [38]. Hence, the proposed CP approach represents an important step to trigger the development of more accurate methods to test isotropy within a loaded RC.

The rest of this article is organized as follows. Section II describes the theoretical background on the polarimetric representation of a 3-D field. In Section III, the measurement setup, the procedures for measuring the 3-D field, and the experimental

results are shown and discussed. Finally Section IV concludes this article.

II. THEORETICAL BACKGROUND

In this section, the theoretical background that underpins the description of the polarimetric properties of a generic 3-D electromagnetic field is detailed.

The complete characterization of the polarization properties of an electromagnetic field is given by the 3×3 \mathbf{R} matrix [30]

$$\mathbf{R} = \langle \mathbf{E} \otimes \mathbf{E}^\dagger \rangle = \begin{pmatrix} \langle E_x E_x^* \rangle & \langle E_x E_y^* \rangle & \langle E_x E_z^* \rangle \\ \langle E_y E_x^* \rangle & \langle E_y E_y^* \rangle & \langle E_y E_z^* \rangle \\ \langle E_z E_x^* \rangle & \langle E_z E_y^* \rangle & \langle E_z E_z^* \rangle \end{pmatrix} \quad (1)$$

where \mathbf{E} is the Jones vector associated to the 3-D field whose components in an orthogonal linear reference frame are E_x , E_y , and E_z . \otimes stands for the Kronecker product, \dagger and $*$ stand for complex conjugate transposed and complex conjugate, respectively. Note that $\langle \cdot \rangle$ represents the time averaging. Since, by construction, \mathbf{R} is a complex Hermitian and semidefinite positive matrix, it can be diagonalized as

$$\mathbf{R} = \mathbf{U} \text{diag}(\lambda_1, \lambda_2, \lambda_3) \mathbf{U}^\dagger = \sum_{k=1}^3 \lambda_k (\mathbf{u}_k \otimes \mathbf{u}_k^\dagger) \quad (2)$$

where \mathbf{U} is a unitary matrix whose columns are the eigenvectors, i.e., \mathbf{u}_k , of \mathbf{R} , whereas λ_k ($k = 1, 2, 3$) are the nonnegative ordered eigenvalues, i.e., $\lambda_1 \geq \lambda_2 \geq \lambda_3 \geq 0$ and $\text{diag}(\cdot)$ is a 3×3 diagonal matrix where the elements of the main diagonal are specified in the brackets. \mathbf{R} has the structure of a covariance matrix and the polarization state is completely described by means of the eigenvalues. Their sum is such that

$$\text{tr}(\mathbf{R}) = \lambda_1 + \lambda_2 + \lambda_3 = I \quad (3)$$

where I stands for the intensity of the field. Unlike the 2-D case, where the degree of polarization can be physically interpreted as the ratio between the polarized part and the total field intensity; in the 3-D case, this physical rationale in general can no longer be applied. This is due to the fact that \mathbf{R} cannot be generally decomposed into its fully polarized and unpolarized parts [24].

A key parameter to describe the polarimetric state of a generic electromagnetic wave is the 3-D degree of polarization [22] that, in terms of invariants of \mathbf{R} , is given by [30]

$$P_{3D} = \sqrt{\frac{1}{2} \frac{3 \text{tr}(\mathbf{R}^2)}{\text{tr}(\mathbf{R})^2} - 1} \quad (4)$$

where P_{3D} is strictly connected to the eigenvalues of \mathbf{R} and it is bounded into the $[0, 1]$ interval. In fact, a pure 3-D state is characterized by $P_{3D} = 1$, whereas a fully unpolarized 3-D state is characterized by $P_{3D} = 0$ and results from a \mathbf{R} matrix with three nonzero and equal eigenvalues.

Although P_{3D} gives an overall measure of the polarimetric purity of a given 3-D polarization state [36], it does not provide complete information on all invariant polarimetric characteristics associated to the corresponding polarization matrix \mathbf{R} [30]. According to the approach provided in [24] and [27], which grounds on mathematical properties of the eigenvalues of \mathbf{R} , the

structure of the polarization state of a generic 3-D wave can be modeled as a superposition of certain mutually incoherent states. With these respects, it is useful to proceed along the so-called trivial (or characteristic) decomposition of \mathbf{R} [35]

$$\mathbf{R} = P_1 \mathbf{R}_p + (P_2 - P_1) \mathbf{R}_m + (1 - P_2) \mathbf{R}_{u-3D} \quad (5)$$

where

$$\mathbf{R}_p = \mathbf{U} \text{diag}(1, 0, 0) \mathbf{U} \quad (6)$$

$$\mathbf{R}_m = \frac{1}{2} \mathbf{U} \text{diag}(1, 1, 0) \mathbf{U}^\dagger \quad (7)$$

$$\mathbf{R}_{u-3D} = \frac{1}{3} \mathbf{I} \quad (8)$$

with \mathbf{I} being the 3×3 identity matrix. P_1 and P_2 are termed IPP and they weight the elementary states that characterize the generic 3-D state [30]

$$P_1 = \frac{\lambda_1 - \lambda_2}{I} \quad \text{and} \quad P_2 = \frac{\lambda_1 + \lambda_2 - 2\lambda_3}{I}. \quad (9)$$

It is important to note that P_1 deals with the stability of the polarization ellipse, preserving its physical meaning of 2-D degree of polarization, whereas P_2 provides information on the stability of propagation direction [29]. The IPP parameters must satisfy the following inequality:

$$0 \leq P_1 \leq P_2 \leq 1. \quad (10)$$

Also note that by using (3), (8), and (9), P_{3D} can be expressed in terms of P_1 and P_2 according to the following quadratic average:

$$P_{3D} = \sqrt{\frac{3}{4} P_1^2 + \frac{1}{4} P_2^2}. \quad (11)$$

It must be explicitly noted that a deeper physical understanding of the trivial decomposition and, therefore, a better interpretation of the physical features of a 3-D polarization state depend strongly on the matrix \mathbf{R}_m associated with the so-called discriminating state.

It is important to remark that the trivial decomposition physically interpreted by means of the IPP parameters does not allow to represent any 3-D polarimetric state [35]. In order to better clarify this key point, meaningful showcases are illustrated.

The matrices \mathbf{R}_p and \mathbf{R}_{u-3D} always represent a pure 3-D state and a 3-D unpolarized state, respectively. With respect to \mathbf{R}_m , following different cases must be distinguished.

- 1) When $P_1 = P_2$, \mathbf{R}_m does not play any role in (4) that becomes

$$\mathbf{R} = P_1 \mathbf{R}_p + (1 - P_1) \mathbf{R}_{u-3D}. \quad (12)$$

In this particular case, the 3-D polarization state can be considered an incoherent composition of a pure state and an unpolarized 3-D state.

In general, $P_1 \neq P_2$, so that the matrix \mathbf{R}_m plays a key role in the trivial decomposition. Two important cases must be distinguished, which are as follows.

- 1) When \mathbf{R}_m is a real matrix, necessarily $\mathbf{R}_m = \frac{1}{2} \mathbf{I} = \mathbf{R}_{u-2D}$ so that \mathbf{R}_m always corresponds to a 2-D unpolarized state (i.e., the electric field associated to this electromagnetic wave evolves fully randomly in a fixed

plane). Hence, when \mathbf{R}_m is a real matrix, according to (4), the structure of the polarization state can be fully described by means of the incoherent superposition of a pure state (\mathbf{R}_p), a 3-D unpolarized state (\mathbf{R}_{u-3D}), and a 2-D unpolarized state (\mathbf{R}_{u-2D}). The generic states for which this physical rationale applies are termed regular states [35].

- 2) When \mathbf{R}_m is not a real matrix, there is no orthogonal (also known as laboratory) reference frame that diagonalizes the rank-3 \mathbf{R}_m matrix and a unitary rotation matrix (6) must be used. The polarization states such that \mathbf{R}_m is not a real matrix are termed nonregular states [35].

The concept of nonregularity relies on certain properties of the coherency matrix relative to its transformation under arbitrary rotations provided by the orthogonal similarity transformation \mathbf{Q}_0 . Hence, it is instructive to split \mathbf{R}_m into its real (\mathbf{R}_R) and imaginary (\mathbf{R}_I) parts. As a result, we get to the intrinsic coherency matrix \mathbf{R}_0 that contains the same information of \mathbf{R}_m and is given by [30]

$$\begin{aligned} \mathbf{R}_0 &\equiv \mathbf{Q}_0 \mathbf{R} \mathbf{Q}_0^T = \mathbf{Q}_0 \mathbf{R}_R \mathbf{Q}_0^T + i \mathbf{Q}_0 \mathbf{R}_I \mathbf{Q}_0^T \\ &= \begin{pmatrix} a_1 & 0 & 0 \\ 0 & a_2 & 0 \\ 0 & 0 & a_3 \end{pmatrix} + \begin{pmatrix} 0 & -in_{03} & in_{02} \\ in_{03} & 0 & -in_{01} \\ -in_{02} & in_{01} & 0 \end{pmatrix} \end{aligned} \quad (13)$$

where T stands for transpose, i is the imaginary unit, and the nonnegative real parameters $a_{js'}$ are the eigenvalues of \mathbf{R}_R , also known as principal intensities. \mathbf{Q}_0 is defined so as to satisfy $a_1 \geq a_2 \geq a_3$, and $n_{0js'}$ are the components of the intrinsic angular momentum vector $\mathbf{n}_0 \equiv (n_{01}, n_{02}, n_{03})^T$ of the state with respect to the so-called intrinsic reference frame $X_0 Y_0 Z_0$ [35]. CP parameters, namely the degree of linear polarization P_l , the degree of circular polarization P_c , and the degree of directionality P_d are defined from \mathbf{R}_0 as follows [30], [34]:

$$\begin{aligned} P_l &= \frac{a_1 - a_2}{a_1 + a_2 + a_3} \\ P_c &= \frac{\sqrt{n_{01}^2 + n_{02}^2 + n_{03}^2}}{a_1 + a_2 + a_3} \\ P_d &= \frac{a_1 + a_2 - 2a_3}{a_1 + a_2 + a_3}. \end{aligned} \quad (14)$$

The three CP parameters are invariant under orthogonal transformations (rotations in the real space) and provide a meaningful framework to describe any 3-D polarization state in terms of quantities that are intrinsic for each given state [34], [35]. Other mathematical properties associated with CP parameters as well as their expressions in terms of the nine 3-D Stokes parameters are provided in [20]–[28]. It is often useful to combine P_l and P_c into an overall degree of elliptical purity P_e that accounts jointly the degrees of linear and circular polarization [36]

$$P_e = \sqrt{P_l^2 + P_c^2}. \quad (15)$$

The parameter P_d measures how close the polarization state described by \mathbf{R} is to a 2-D polarization state. It is also worth to recall that P_{3D} can be expressed in terms of CP parameters as

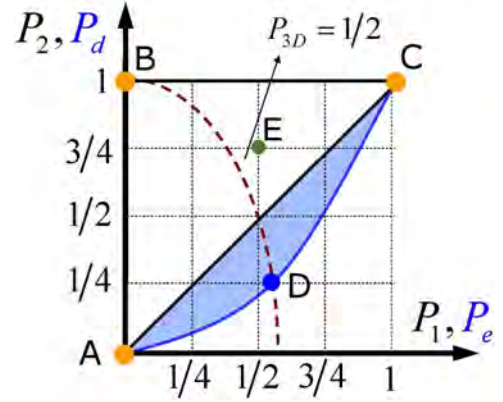


Fig. 1. Feasible region for IPP parameters (P_1, P_2) is determined by the triangle “ABC,” whereas the feasible region for CP parameters (P_e, P_d) also includes the shadowed blue area. For both representations, the points “A” and “C” correspond to regular states. The shadowed blue area (except for the straight line \overline{AC}) is exclusive of nonregular states, whereas the points inside the triangle “ABC” can result from either regular or nonregular states. The point “E” corresponds to the nonregular state measured in the proposed loaded RC configuration.

follows:

$$P_{3D} = \sqrt{\frac{3}{4}(P_l^2 + P_c^2) + \frac{1}{4}P_d^2} = \sqrt{\frac{3}{4}P_e^2 + \frac{1}{4}P_d^2}. \quad (16)$$

The CP parameters always satisfy the following inequalities:

$$P_l^2 \leq P_d^2 \leq 1 \quad (17)$$

$$0 \leq P_e^2 \leq 1. \quad (18)$$

According to the aforementioned rationale, \mathbf{R} represents a regular state if and only if $P_e = P_1$ and $P_d = P_2$. Therefore, the theoretical framework that includes the trivial decomposition and the CP parameters allows a physically consistent interpretation of 3-D states distinguishing between regular and nonregular states.

The peculiar features and differences between IPP and CP indicators can be simply understood by contrasting IPP (P_1, P_2) and CP (P_e, P_d) parameters on the 2-D purity space depicted in Fig. 1. The vertical (horizontal) axis includes P_2 and P_d (P_1 and P_e). The feasible region related to the IPP parameters consists of the triangle “ABC,” whereas the feasible region of the CP parameters consists of the “ABC” triangle augmented by the shadowed blue area.

The points “A” ($P_{3D} = 0$) and “C” ($P_{3D} = 1$) represent regular fully unpolarized and polarized 3-D states, respectively. The straight line \overline{AC} is exclusive of regular states since it calls for $P_1 = P_2$. The shadowed blue area (except for the \overline{AC} segment) is exclusive of nonregular states since it calls for $P_2 \neq P_d$ and $P_1 \neq P_e$. The triangle “ABC” (except for the points “A,” “C,” and the segment \overline{AC}) calls for either regular or nonregular states. In particular, for a given P_{3D} value, IPP and CP parameters describe a common elliptical branch and they coincide only and only if a regular state is in place.

As an illustrative example, the elliptical branch that corresponds to $P_{3D} = 0.5$ is depicted as dashed red line in Fig. 1. One

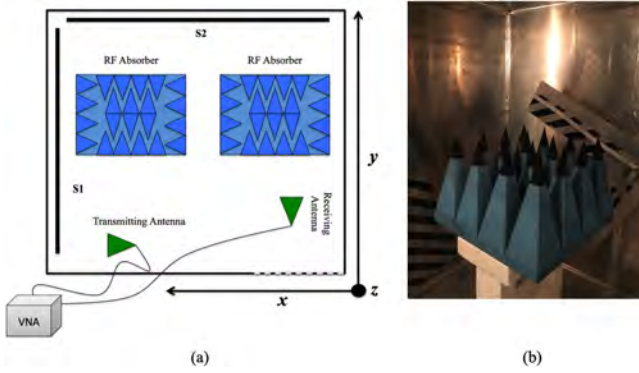


Fig. 2. RC of the UNP. (a) RC configuration, (b) RF standard absorber.

can note that it includes all the discriminating states [i.e., \mathbf{R}_m matrix in (5)] that result from $P_1 = 0$ and $P_2 = 1$. Regular states are achieved if and only if $P_1 = P_e$ and $P_2 = P_d$. It is worth showing the point “D” that, belonging to the shadowed region, results from a nonregular state. In particular, as shown in [37], the point “D” represents a state with maximum nonregularity.

To summarize, the framework that lies on the trivial decomposition together with the CP allows interpreting and classifying polarization states in an objective way and provides a proper way to distinguish between regular and nonregular states.

III. EXPERIMENTAL RESULTS

In this section, a meaningful set of experiments are presented and discussed to show the physical significance and the inter-connection among P_{3D} , IPP and CP parameters, and the limits of IPP.

Measurements are performed using a 8-m³ cubic RC whose setup is depicted in Fig. 2. Within the chamber, three mechanical stirrers, S1, S2, and S3, operate in continuous mode to randomize the input electromagnetic field. S1 and S2 are placed on the left wall and in front of the entrance door, respectively, whereas S3 is placed on the top of the chamber. Two Narda Waveguide Horn Antennas certified to work in X-band, i.e., from 8.5 to 12.4 GHz frequency range, are used. The two antennas do not face each other, i.e., a non-line-of-sight configuration is adopted, and they are placed at about 1 m on the floor on Styrofoam supports that are transparent to the electromagnetic field. A two-port Agilent Technologies vector network analyzer is used to measure the complex parameter S_{21} . It is important to note that the statistics of S_{21} are equivalent to the statistics of the Cartesian field components inside an RC [39], [40].

The calibration procedure is carried out using the CalKit 85400 and the response and isolation procedure. All the experiments refer to a 12-GHz electromagnetic field and different RC configurations are considered, which are obtained by varying the number and the position of the microwave absorbers. The measurements are performed by aligning the transmitting antenna to each of the three axis of an orthogonal Cartesian reference system centered into the transmitting antenna. For each transmitted linearly polarized wave, the receiving antenna is aligned to each of the three axis of an orthogonal Cartesian

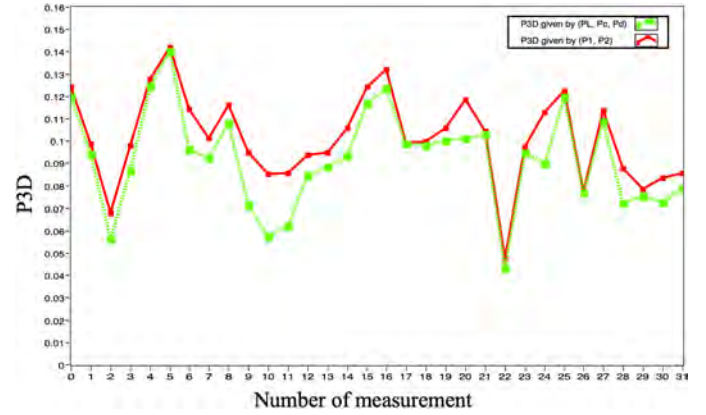


Fig. 3. Unloaded RC. P_{3D} evaluated measuring CPs (green dotted line) and IPPs (red dot continuous line). The transmitting antenna is polarized along the x-axis.

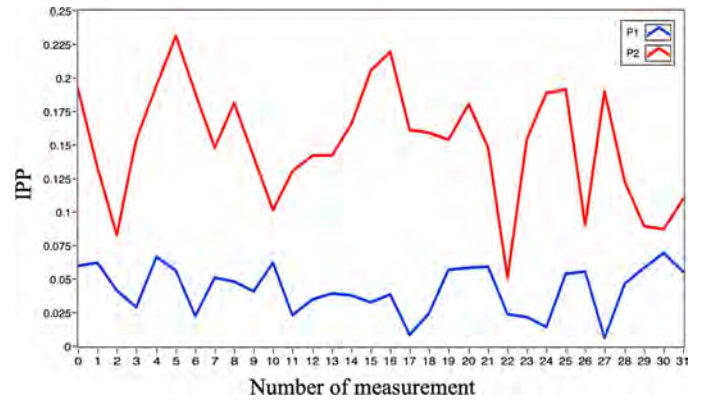


Fig. 4. Unloaded RC. P_1 (blue line) and P_2 (red line) are shown when the transmitting antenna is polarized along the x-axis.

reference system centered into the receiving antenna [29]. For each receiving antenna position, 32 independent measurements are obtained by averaging 500 field samples. It is worth noting that since the continuous stirring mechanism is adopted, the geometry of the RC and its boundary conditions are continuously changing during the measurement time. Hence, rotating the antenna is not expected to significantly affect the measurement process [17]. In addition, the average field measured within a well-stirred RC has been shown in [41] to be independent of the antennas used. The IPP and CP parameters are then offline evaluated and analyzed using the LabVIEW environment.

The first experiment refers to the unloaded RC configuration and aims at verifying that IPP (11) and CP (16) parameters allow the same physical interpretation of the field within the RC when a regular state is in place. In Fig. 3, the transmitting antenna is aligned with the x-axis of a Cartesian reference system and P_{3D} is shown in red continuous line (IPP) and green dotted line (CP). As expected, the two lines are almost completely overlapped resulting in P_{3D} values (mean \pm standard deviation) equal to 0.079 ± 0.038 and 0.076 ± 0.038 for the IPP and CP cases, respectively (see Table I). P_{3D} values obtained in this experiment, according to what formerly experienced in [23]

TABLE I
 P_{3D} (MEAN VALUES \pm STD) OBTAIN USING IPP AND CP AND RANK OF THE COHERENCY MATRIX \mathbf{R} RELATED TO DIFFERENT RC CONFIGURATION AND MEASURED WHEN THE TRANSMITTING ANTENNA IS ALIGNED WITH THE X, Y, AND Z AXES

Transmitted Polarization	RC Configuration	$\langle P_{3D} \rangle \pm \sigma_{P_{3D}}$ in terms of (P_e, P_d)	$\langle P_{3D} \rangle \pm \sigma_{P_{3D}}$ in terms of (P_1, P_2)	Rank of \mathbf{R}
x	No Abs	0.076 ± 0.038	0.079 ± 0.038	3
	1 Abs	0.812 ± 0.030	0.714 ± 0.029	2
	2 Abs	0.850 ± 0.035	0.854 ± 0.034	2
y	No Abs	0.134 ± 0.052	0.142 ± 0.056	3
	1 Abs	0.820 ± 0.045	0.823 ± 0.046	2
	2 Abs	0.826 ± 0.048	0.829 ± 0.051	2
z	No Abs	0.137 ± 0.035	0.138 ± 0.038	3
	1 Abs	0.580 ± 0.078	0.432 ± 0.076	3
	2 Abs	0.609 ± 0.099	0.606 ± 0.099	2

TABLE II
 CP AND IPP VALUES (MEAN VALUES \pm STD) RELATED TO DIFFERENT RC CONFIGURATION AND MEASURED WHEN THE TRANSMITTING ANTENNA IS ALIGNED WITH THE X, Y, AND Z AXES

Transmitted Polarization	RC Configuration	$\langle P_l \rangle \pm \sigma_{P_l}$	$\langle P_c \rangle \pm \sigma_{P_c}$	$\langle P_d \rangle \pm \sigma_{P_d}$	$\langle P_e \rangle \pm \sigma_{P_e}$	$\langle P_1 \rangle \pm \sigma_{P_1}$	$\langle P_2 \rangle \pm \sigma_{P_2}$
x	No Abs	0.079 ± 0.009	0.010 ± 0.007	0.135 ± 0.008	0.080 ± 0.006	0.055 ± 0.006	0.125 ± 0.008
	1 Abs	0.807 ± 0.040	0.062 ± 0.008	0.962 ± 0.006	0.654 ± 0.017	0.646 ± 0.055	0.886 ± 0.072
	2 Abs	0.872 ± 0.046	0.066 ± 0.052	0.975 ± 0.010	0.876 ± 0.059	0.889 ± 0.045	0.989 ± 0.072
y	No Abs	0.100 ± 0.062	0.085 ± 0.038	0.086 ± 0.066	0.139 ± 0.067	0.069 ± 0.034	0.260 ± 0.121
	1 Abs	0.782 ± 0.060	0.118 ± 0.056	0.895 ± 0.024	0.803 ± 0.095	0.848 ± 0.055	0.892 ± 0.104
	2 Abs	0.795 ± 0.061	0.167 ± 0.062	0.856 ± 0.049	0.815 ± 0.111	0.831 ± 0.054	0.972 ± 0.094
z	No Abs	0.068 ± 0.042	0.083 ± 0.032	0.167 ± 0.095	0.144 ± 0.042	0.070 ± 0.033	0.231 ± 0.075
	1 Abs	0.353 ± 0.123	0.330 ± 0.077	0.750 ± 0.058	0.506 ± 0.095	0.151 ± 0.065	0.805 ± 0.079
	2 Abs	0.476 ± 0.157	0.155 ± 0.077	0.892 ± 0.056	0.521 ± 0.050	0.551 ± 0.059	0.953 ± 0.075

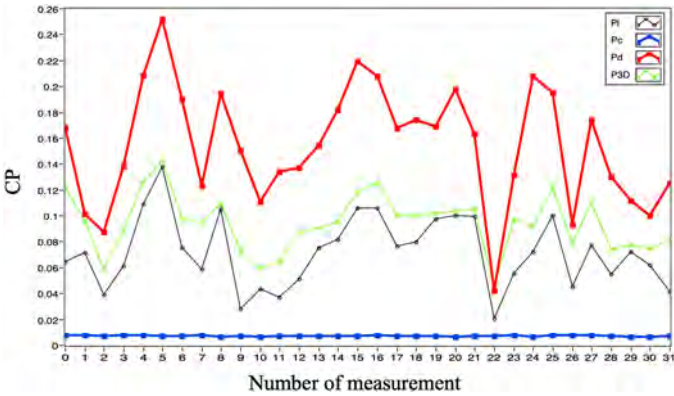


Fig. 5. Unloaded RC. P_l (dotted black line), P_c (blue line), P_d (red line), and P_{3D} (green dotted line) are shown when the transmitting antenna is polarized along the x-axis.

through independent measurements, are compatible with a fully unpolarized 3-D state. This is further confirmed by the analysis of IPP (see Fig. 4), which shows that as expected, $P_1 < P_2$ and both the parameters result in very low values. In particular, P_1 and P_2 values (mean \pm standard deviations) are 0.055 ± 0.006 and 0.125 ± 0.008 , respectively (see Table II). Hence, a 3-D unpolarized state is in place. The behavior of CP parameters is depicted in Fig. 5 where one can see that they result, as expected, in very low values. In addition, by using P_e (12), one can note that $P_e \cong P_l \cong P_1$, i.e., $P_c \cong 0$, whereas $P_d \cong P_2$. Hence, the rank-3 \mathbf{R} matrix represents a 3-D regular state [35]. The mean values \pm standard deviations of both IPPs and CPs

values are listed in Table II. Similar results are obtained when the transmitting antenna is polarized along the y and z axes (see Tables I and II).

Next experiments refer to the RC loaded with absorbers and aim at providing, on one side, an understanding of the difference between IPP and CP parameters; on the other side, they demonstrate that the RC is a very versatile environment that allows generating any polarimetric state, including the nonregular one. Hence, an Emerson and Cuming RF standard absorber block, consisting of 16 conical pieces arranged as a 4×4 array, is used [see Fig. 2(b)] to load the RC. The absorber block is placed in the RC corner, in front of the transmitting antenna and about 1 m above the floor [see Fig. 2(a)]. The taper length is about 20 cm.

The second experiment refers to the transmitting antenna polarized along the x-axis. P_{3D} values evaluated using the IPP and CP parameters are equal to 0.714 ± 0.030 and 0.812 ± 0.029 , respectively (see Table I). The degree of purity increased significantly with respect to the unloaded case witnessing that the absorbers reduce the randomness of the field within the RC. The behavior of P_1 and P_2 is depicted in Fig. 6. P_1 and P_2 values (0.646 ± 0.055 and 0.886 ± 0.072 , respectively) increased with respect to the unloaded case (see Table II). In particular, P_2 is closer to 1, witnessing that the direction of propagation is practically fixed and well defined. Moreover, \mathbf{R} is now a rank-2 matrix, witnessing that a 2-D polarization state is in place. Such result is confirmed by CP parameters (see Fig. 7), where P_l (black curve), P_c (blue curve), and P_d (red curve) are shown. Even in this case, P_l and P_d increased with respect to the unloaded case, whereas P_c is still negligible. The mean value and the standard

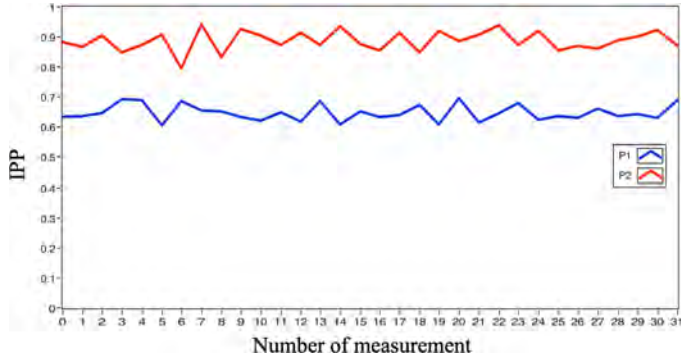


Fig. 6. RC loaded with one piece of absorber. P_1 (blue line) and P_2 (red line) are shown when the transmitting antenna is polarized along the x-axis.

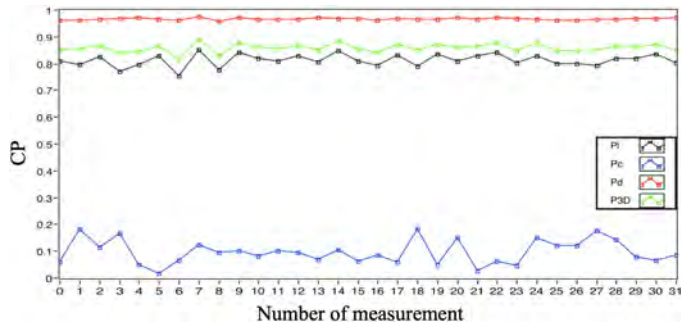


Fig. 7. RC loaded with one piece of absorber. P_1 (dotted black line), P_c (blue line), P_d (red line), and P_{3D} (green dotted line) are shown when the transmitting antenna is polarized along the x-axis.

deviation of IPP and CP parameters are listed in Table II. In summary, one can note that $P_d \cong P_2 \cong 1$, $0 \leq P_e \cong P_1 < 1$, and $1/2 \leq P_{3D} < 1$. The field received within the well-stirred RC loaded with one block of absorbers is a 2-D regular polarization state and both IPP and CP parameters still provide the same information. Similar results are obtained when the transmitting antenna is aligned with the y-axis (see Tables I and II).

The third experiment refers to the transmitting antenna polarized along the z-axis. P_{3D} values evaluated using the IPP and CP parameters are equal to 0.432 ± 0.076 and 0.580 ± 0.078 , respectively (see Table I). It can be noted that P_{3D} decreased with respect to the previous cases witnessing that a more random configuration is achieved when the transmitting antenna is z-polarized. The behavior of P_1 and P_2 is depicted in Fig. 8 (see blue and red continuous lines, respectively). It is worth pointing out that in this case, the IPP parameters do not offer additional information on the polarimetric properties of this electromagnetic field. In fact, as usual, $0 \leq P_1 \leq P_2 \leq 1$; hence, since $1/2 \leq P_{3D} \leq 1$, we can only say that a partially polarized wave is in place. However, the polarimetric state can be either 2-D or 3-D.

To further shed light on this ambiguous case, CP parameters are used (see Fig. 9). The mean values of the CP parameters are listed in Table II, which show that $0 \leq P_d < P_2 \cong 1$ and $0 \leq P_1 < P_e < 1$; in addition, a rank-3 of \mathbf{R} matrix is obtained.

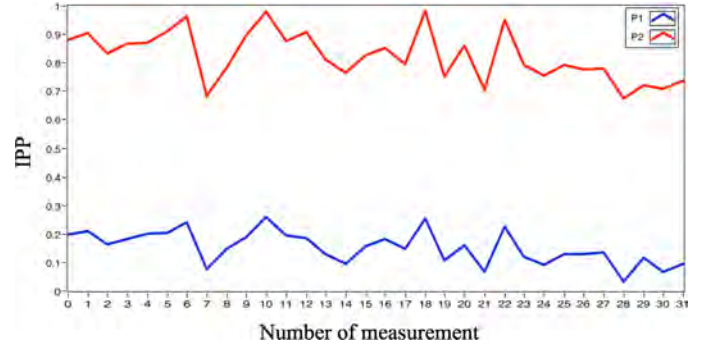


Fig. 8. RC loaded with one piece of absorber. P_1 (blue line) and P_2 (red line) are shown when the transmitting antenna is polarized along the z-axis.

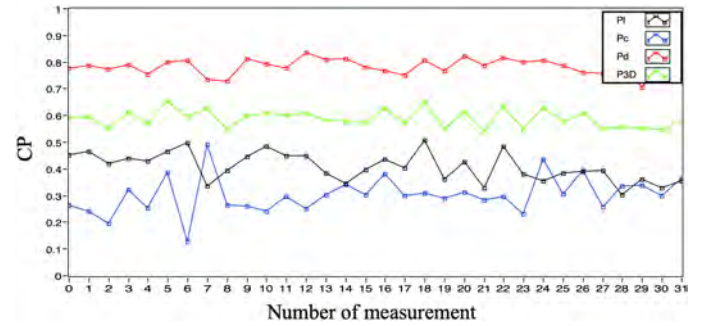


Fig. 9. RC loaded with one piece of absorber. P_1 (dotted black line), P_c (blue line), P_d (red line), and P_{3D} (green dotted line) are shown when the transmitting antenna is polarized along the z-axis.

Hence, the CP parameters witness that a 3-D nonregular polarization state is achieved.

From a physical view point, the generation of this nonregular state is likely due to the fact that the loaded RC is no longer an isotropic environment. In detail, when the antenna is moved in the azimuth plane (e.g., when the antenna is aligned along the x and y axes), the RC configuration is still isotropic since both the antenna opening and the RF absorber are at the same height. When the antenna points along the z-direction, a different scenario is in place since the antenna opening is at a different height with respect to the RF absorber block and it faces to the RC metallic floor. This scenario breaks the RC symmetry resulting in a nonisotropic environment, which allows obtaining 3-D electromagnetic fields whose polarimetric state belongs to the nonregular class.

The last set of measurements is made loading the RC with two RF absorbers. In particular, an additional absorber (of the same type of the previous one) is placed in the RC corner, in front of the receiving antenna and about 1 m above the floor [see Fig. 2(a)]. In this case, CP and IPP mean values confirm the results achieved when only one RF absorber is used, i.e., a partially polarized wave is in place.

When the transmitting antenna is aligned with the x-axis the mean P_{3D} values obtained using IPP and CP are equal to 0.850 ± 0.035 and 0.854 ± 0.034 , respectively. The mean value and the standard deviation of IPP and CP parameters are listed in

Table II. In summary, one can note that $P_d \cong P_2 \cong 1$, $0 \leq P_e \cong P_1 < 1$, and $1/2 \leq P_{3D} < 1$. Hence, the polarimetric structure is now a 2-D one [35]. Similar results can be obtained when the transmitting antenna is aligned with the y-axis (see Tables I and II). It is interesting to note that when the antenna is aligned with the z-axis, the mean P_d value is equal to 0.892 ± 0.056 . Hence, $P_d \cong P_2 \cong 1$ and $P_e \cong P_1 \cong 1$ (see Table II). Moreover, $P_e \cong P_1$ (see Table II). Hence, unlike the previous experiment, in this case, a regular state is obtained and both CP and IPP parameters provide the same information. This result shows that the number and the position of the absorbers significantly affect the polarimetric behavior of the field inside the RC.

IV. CONCLUSION

In this article, a new set of 3-D polarimetric CP parameters was first presented to the EMC community. The CP decomposition was first experimentally tested in this article where it is proven to be of physical interest. The main outcomes can be summarized as follows.

- 1) The CP decomposition provided a step forward with respect to the IPP decomposition in distinguishing between regular and nonregular states, overcoming the IPP decomposition limits.
- 2) The RC allowed emulating both regular and nonregular states. Nonregular states were associated to loaded RC.
- 3) Nonregular states occurring in the loaded RC were likely connected to the isotropy loss. Hence, the CP decomposition could be used to support more accurate field isotropy testing in the RC.

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within reverberation chamber, and characterization of the wireless propagation channels in controlled environments.