

Evaluation of DC-link Decoupling Using Electrolytic or Polypropylene Film Capacitors in Three-Phase Grid-Connected Photovoltaic Inverters

Baburaj Karanayil⁽¹⁾, Vassilios G. Agelidis⁽¹⁾ and Josep Pou⁽¹⁾⁽²⁾

⁽¹⁾Australian Energy Research Institute, The University of New South Wales, UNSW Sydney, NSW 2052, Australia

⁽²⁾Terrassa Industrial Electronics Group, Technical University of Catalonia, Catalonia, Spain

Email: k.baburaj@unsw.edu.au, vassilios.agelidis@unsw.edu.au, j.pou@unsw.edu.au

Abstract— The life expectancy and long term reliability of grid-connected three-phase photovoltaic (PV) inverters can be increased by replacing the conventional electrolytic film capacitors by metallized polypropylene film capacitors.

This paper presents a detailed evaluation of a three-phase grid-connected PV inverter performance when replacing the electrolytic capacitor with a minimum value of metallized polypropylene film capacitor-one. The minimum dc bus capacitance leads to larger voltage ripples. However, such ripples were found to be within acceptable limits to run the inverter satisfactorily. Simulation results are presented for a 15-kW grid-connected inverter at nominal voltage of 700V dc and experimental results are provided for a 3.0-kW system at a nominal voltage of 400V dc, built in the laboratory.

Index Terms — Photovoltaic systems, grid-connected inverters, electrolytic capacitor

I. INTRODUCTION

The typical lifetime of commercial photovoltaic (PV) panels is about 25 years [1],[2]. However, the typical operating life of the PV inverters processing energy generated by PV panels is limited by the operating life of their individual components. The operating life of commercial electrolytic capacitors used as dc-link intermediate energy storage is only about 10 years presenting a reliability related weak link in the PV system [3],[4].

The life expectancy and long term reliability of grid-connected three-phase PV-inverters can be increased by replacing the conventional electrolytic film capacitors by metallized polypropylene film capacitors [5]. Film capacitors have a lower capacitance per unit volume compared to the electrolytic counterparts, and therefore a direct replacement will lead to a very large and expensive solution. The biggest design limitation for electrolytic capacitors in the inverter has been the amount of ripple current that the electrolytic capacitor can sustain. Typically, any inverter design ends up with an electrolytic capacitor many times larger because of this restriction.

Electrolytic capacitors have been used in industrial applications for hard switched inverters for many years. The dc-link capacitor C_{dc} in the grid-connected PV inverter shown

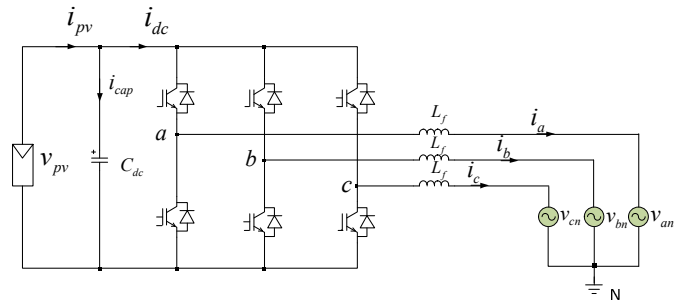


Fig. 1. Three-phase grid-connected PV-inverter.

in Fig. 1 is a load balancing energy storage element between the PV panel and the three-phase grid. This capacitor is connected in parallel to the PV panel to maintain a stiff dc-link voltage across the voltage source inverter (VSI). An electrolytic capacitor mostly used in industrial applications offers a greater capacitance per unit volume compared to a polyester / polypropylene film capacitor. It is easy and affordable to install electrolytic capacitors for decoupling between the input and output of the inverter as they cost less per Joule. They produce more heating inside as they have high equivalent series resistance (ESR). The major drawback of these capacitors is their limited operating life. Over long time, the liquid electrolyte evaporates through the rubber sleeves of the capacitor, degrading the capacitance. The effect can be compensated by oversizing the capacitors by design. However, the lifetime limitation still exists. The film capacitors have lower ESR thus producing less heating and having longer lifetime. The disadvantage of film capacitors is their low capacitance per volume ratio (approximately 20 times lower than electrolytic one) and thus the higher cost per Joule. A direct replacement is therefore not feasible in terms of cost and size. It is hence critical to find out solutions to improve the operating lifetime of inverters using film capacitors with a compromise.

In PV applications, the voltage ripple at the inverter input has to be kept small in order to assure stable operation in the maximum power point (MPP) of the PV modules. Many efforts have been reported in the technical literature to replace electrolytic capacitors by film counterparts. A very general solution for decreasing the capacitance in a single-phase PV-

inverter, which is independent of the inverter topology is the parallel active filter presented in [6]. The dc-link capacitor is separated into two parts. The capacitors are connected via bi-directional dc-dc converter, allowing a different voltage at both of them. The dc/dc converter is operated in a way which keeps the dc-link voltage constant, while the voltage of the second capacitor with a larger ripple with larger part of the stored energy. The dc-link capacitance requirements for a single-phase multi-string PV inverter are compared in [7]. In the two-stage inverter, the boost converter decouples the PV string from the output inverter, allowing use of a smaller dc-link capacitor. The dc-link capacitor for the single-stage inverter was found at least five times the size of the required capacitor in the two-stage inverter in order to retain its advantage of having higher conversion efficiency [7]. The bulky electrolytic capacitors in ac PV interface can be replaced with small film capacitors with dc distribution to fully guarantee the lifetime of the PV system. A boundary conduction mode (BCM) operated single-phase boost converter at 100 kHz with soft switching of semiconductor devices has been presented in [8]. Only a small capacitance which is one thousandth of the ac case is enough to keep the voltage variation within 0.7V. However, the above schemes were studied and reported for single-phase PV-inverters.

II. CALCULATION OF THE DC LINK CAPACITOR

The selection of dc-link capacitors for high performance inverter applications is reported in [9]. A detailed analysis and evaluation of dc-link capacitors for an 80-kW electric vehicle drive system is reported in [10].

The PV current i_{pv} is the sum of the capacitor current i_{cap} and the dc-link current i_{dc} as shown in Fig. 1. These current waveforms are affected by the load power factor $\cos\phi$ and the inverter modulation index m . Based on the synthesis of inverter current modulation [11], the root-mean-square (RMS) capacitor ripple current can be expressed by

$$I_{cap,RMS} = I_{N,RMS} \sqrt{2m \left\{ \frac{\sqrt{3}}{4\pi} + \cos^2 \phi \left(\frac{\sqrt{3}}{\pi} - \frac{9}{16}m \right) \right\}}, \quad (1)$$

where m is the modulation index, I_N is the output phase current and ϕ is the phase delay of the inverter output current with respect to the fundamental voltage.

A. Simulation Study

For a 15-kW load to be connected to a 415-V three-phase voltage grid, the maximum current stress in the capacitor is calculated as 13.56 A for $m=0.612$, and $\phi=0$.

The electrolytic capacitor 5600 μF / 450V (EPCOS B43456

5568) is selected which has $I_{ac,R} = 22.1$ A at 10 kHz. In order to meet the capacitor ripple current rating to operate at 800V dc, two 5600- μF capacitors should be connected in series. Therefore, the equivalent capacitance is 2800 μF .

The value of the minimum capacitance to meet the 13.56 A ripple current and producing a V_{ripple} will be:

$$C_{dc\min} = \frac{I_{rms}}{V_{ripple} \times 2 \times \pi \times f_s} \quad (2)$$

for V_{ripple} of 5V, $I_{rms} = 13.56$ A, $f_s = 10$ kHz, $C_{dc\min} = 43.16$ μF . The next commercially available capacitor of 47 μF is used in the simulation.

B. Experimental Study

For a 3-kW load at 200V (line-to-line) grid voltage, the maximum current stress in the capacitor is calculated as 5.63 A for $m=0.612$, and $\phi=0$. The electrolytic capacitor 1500 μF / 450V (EPCOS B43456 5158 has $I_{ac,R} = 8.45$ A at 10 kHz. In order to meet the capacitor ripple current rating to operate at 800V dc, the two 1500- μF capacitors should be connected in series. Therefore, the equivalent capacitance is 750 μF . However, two 2200- μF capacitors are connected in the experimental set-up. As per (2), for a V_{ripple} of 4V, $I_{rms} = 5.63$ A and $f_s = 10$ kHz, $C_{dc\min} = 22.4$ μF . AVX brand FFVE6C0476K, 900V dc 47 μF metallized polypropylene film capacitor is used in the experiment set-up.

III. CONTROL STRATEGY FOR THE GRID-CONNECTED PV-INVERTER

Fig. 2 shows the control scheme of the grid-connected PV inverter. In order to decouple the active and reactive power controls, the synchronous rotating $d-q$ reference frame is applied for developing the controllers. A synchronous reference frame phase-locked loop (SRF-PLL) [12] is used to synchronize the d -axis with the grid voltage vector. The d -channel current loop allows the control of active power that is supplied by the PV inverter. The reactive power is controlled by the q -axis current controller. To minimize the inverter power losses, unity power factor at the output of the inverter is desirable. Hence, a null reference i_q^* for the reactive current loop is chosen. The i_d and i_q current references are generated by the outer control loops imposed by the dc voltage and reactive power references respectively. Two inner control loops regulate the i_d and i_q currents respectively where the coupling currents are compensated by feed forward terms as it can be observed in Fig. 2. An outer voltage loop maintains the PV panels' voltage close to a desired reference v_{dc}^* , which is calculated by the Maximum Power Point Tracking (MPPT) algorithm to extract the maximum power from the PV panels.

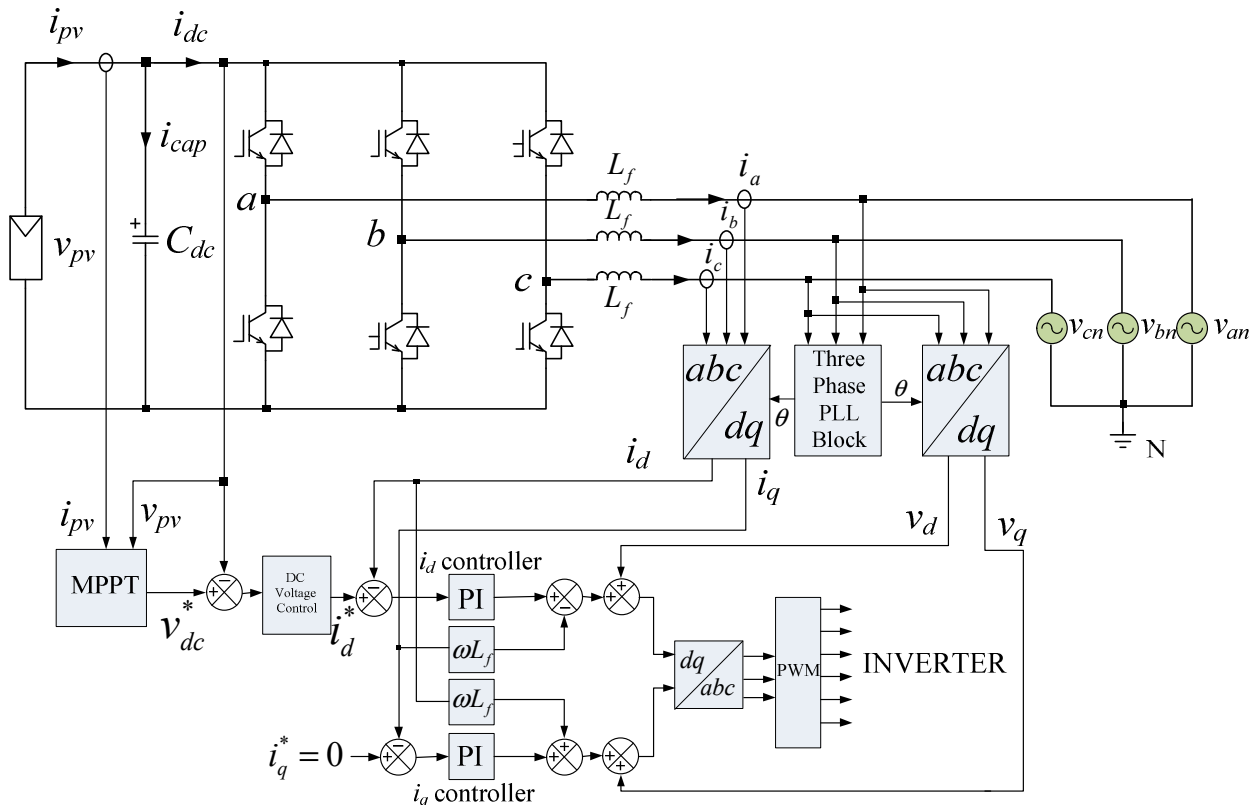


Fig. 2. Schematic of the control block diagram for three-phase grid-connected PV-inverter with dq control.

TABLE I. PARAMETERS OF THE THREE-PHASE PV-INVERTER UNDER SIMULATION STUDY

Rated power of PV array	P_{max}	15600 W
Number of BP365 65W PV modules in the string (series modules)		40
Number of strings in parallel		6
Open circuit voltage of PV array	V_{oc}	868 V
Voltage at maximum power P_{max}	V_{mp}	704 V
Current at maximum power P_{max}	I_{mp}	22.14 A
DC capacitors (electrolytic) EPCOS type 2 x B43456 A5568, 5600 μ F in series	C_{dc}	2800 μ F
DC capacitors (metallized polypropylene film) AVX type FFVE6C0476K, 900V dc	C_{dc}	47 μ F
External line inductor	L_f	5 mH
Inverter switching frequency	f_s	10 kHz

IV. SIMULATION RESULTS

The grid-connected three-phase inverter with a 15-kW PV module has been studied with simulations for steady state and transient behaviors in a MATLAB-PLECS software environment. The parameters of the system used for modeling are listed in Table I. Two different configurations are studied: one with a 2800- μ F electrolytic capacitor and the other one with a 47- μ F film capacitor. The objective is to optimize the ac ripple in the dc bus voltage, the RMS ripple current through the capacitor, the ripple current in the PV module and use the minimum capacitance possible which can still perform as a grid-connected system.

A. Steady State

1. Results with 2800- μ F Electrolytic Capacitor

The 15-kW PV module with specifications in Table I provides the input to the three-phase inverter and the output is connected to the 415 V, 50 Hz grid. The grid voltage v_{an} , the grid current i_a and PV module current i_{pv} are recorded when the PV module feeds 15 kW to the grid, as shown in Fig. 3. The ripple voltage across the capacitor, current i_{cap} and the dc bus voltage v_{pv} are recorded as shown in Fig. 4. The ripple voltage across the capacitor is less than 0.2V at 700V dc as seen in Fig. 4(a). The capacitor current i_{cap} has a peak-to-peak value of 27 A as seen in Fig. 4 (b).

2. Results with 47- μ F Film Capacitor

Two EPCOS 5600- μ F electrolytic capacitors connected in series are replaced by one AVX type FFVE6C0476K, 900V dc 47- μ F film capacitor in the inverter and the steady state measurements are repeated as taken for the case of the 2800- μ F electrolytic capacitor. The grid voltage v_{an} , grid current i_a and the PV module current i_{pv} are recorded when the PV module feeds 15 kW to the three-phase grid as shown in Fig. 5. The dc voltage and ac ripple voltage across the capacitor and the capacitor current i_{cap} are recorded as indicated in Fig. 6. The ripple voltage at 700V dc bus is close to 10V, which is 1.42% as seen in Fig. 6(a). The capacitor current has a larger ripple of 60 A as seen in Fig. 6(b).

B. Transient

The transient performance of the grid-connected inverter is verified for a solar irradiance change from 0.5 kW/m^2 to 1.0 kW/m^2 . These results are recorded for two cases, when the inverter uses the $2800\text{-}\mu\text{F}$ electrolytic capacitor and then replaced by the $47\text{-}\mu\text{F}$ film capacitor.

1. Results with $2800\text{-}\mu\text{F}$ Electrolytic Capacitor

The grid voltage v_{an} , grid current i_a and the input power supply current i_{pv} are recorded when the solar irradiance increases from

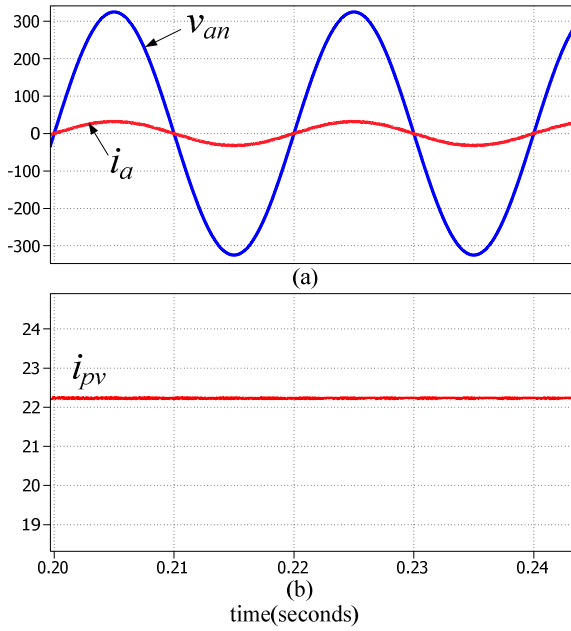


Fig. 3. Simulation results Case I - C_{dc} is $2800\text{-}\mu\text{F}$ electrolytic capacitor: (a) grid voltage v_{an} (V) and grid current i_a (A) (b) PV array current i_{pv} (A).

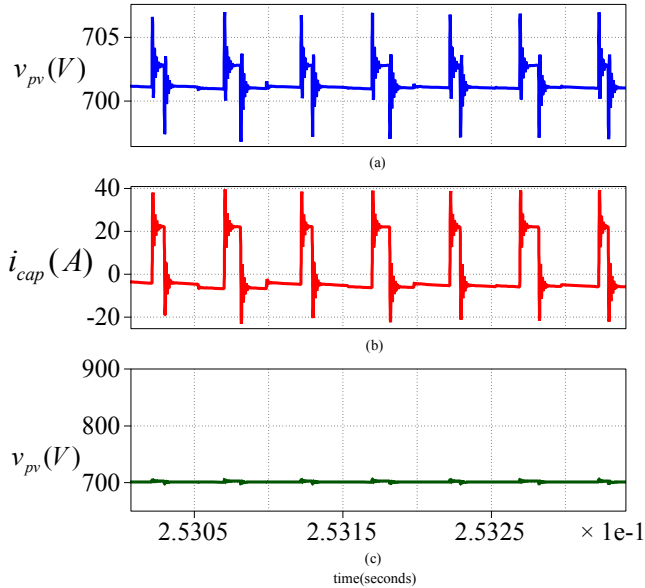


Fig. 4. Simulation results Case I - C_{dc} is $2800\text{-}\mu\text{F}$ electrolytic capacitor: (a) detailed view of dc-link voltage showing the ripple (b) capacitor current and (c) dc-link voltage.

0.5 kW/m^2 to its rated value 1.0 kW/m^2 . The results are shown in Fig. 7. It is observed that the grid current undergoes a smooth transition.

2. Results with $47\text{-}\mu\text{F}$ Film Capacitor

The transient performance of the grid-connected inverter is compared after replacing the $2800\text{-}\mu\text{F}$ electrolytic capacitor with a $47\text{-}\mu\text{F}$ film capacitor. The grid voltage v_{an} , grid current i_a and the input power supply current i_{pv} are recorded when the solar irradiance increases from 0.5 kW/m^2 to its rated value 1.0 kW/m^2 . The results are shown in Fig. 8. The grid current transition is very smooth even in this case.

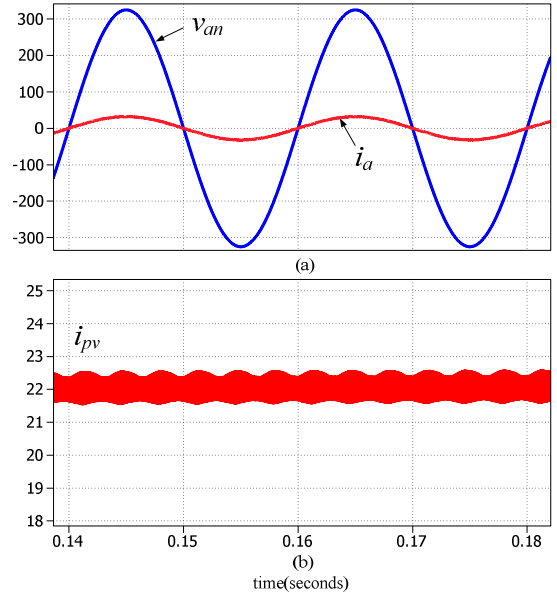


Fig. 5. Simulation results Case II - C_{dc} is $47\text{-}\mu\text{F}$ film capacitor: (a) grid voltage v_{an} (V) and grid current i_a (A) (b) PV array current i_{pv} (A).

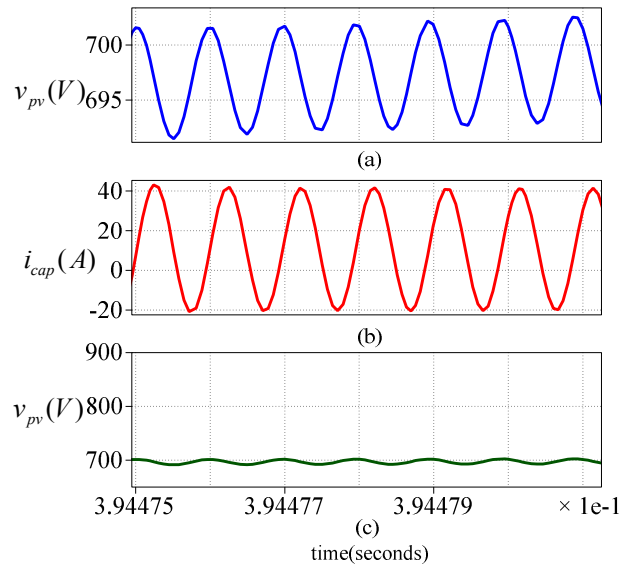


Fig. 6. Simulation results Case II - C_{dc} is $47\text{-}\mu\text{F}$ film capacitor: (a) detailed view of dc-link voltage showing the ripple (b) capacitor current and (c) dc-link voltage.

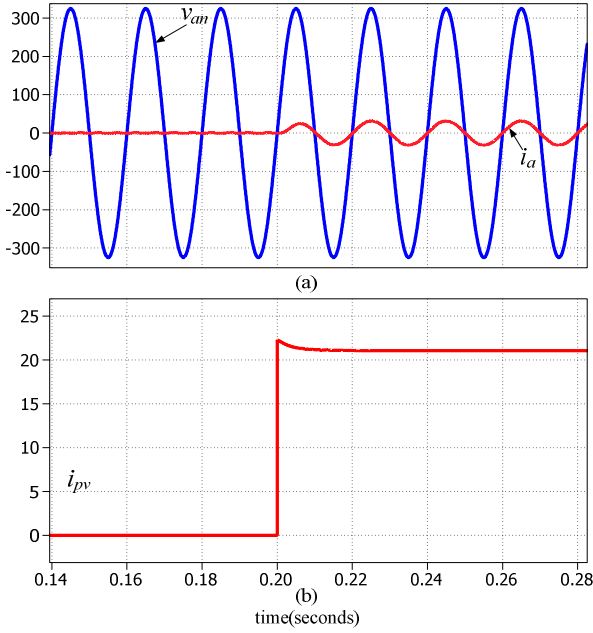


Fig. 7. Simulation results Case I - C_{dc} is 2800- μ F electrolytic capacitor: (a) grid voltage v_{an} (V) and grid current i_a (A) and (b) PV array current i_{pv} (A) for an increase in solar irradiance from 0 kW/m^2 to 1.0 kW/m^2 .

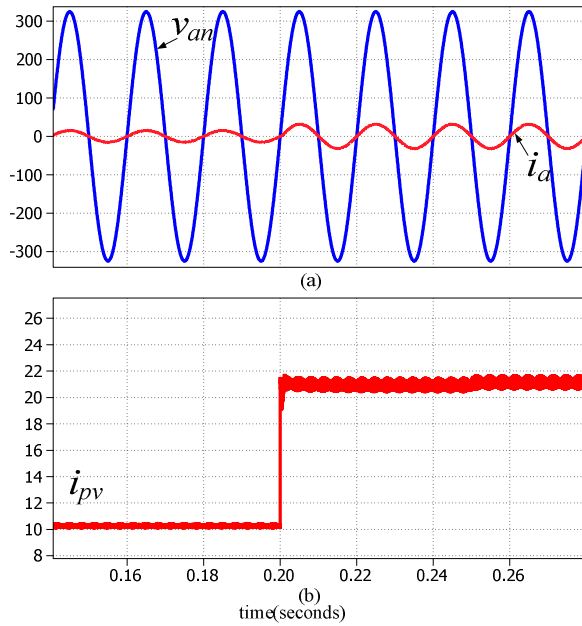


Fig. 8. Simulation results Case II - C_{dc} is 47- μ F film capacitor: (a) grid voltage v_{an} (V) and grid current i_a (A) and (b) PV array current i_{pv} (A) for an increase in solar irradiance from 0 kW/m^2 to 1.0 kW/m^2 .

V. EXPERIMENTAL RESULTS

The performance of the three-phase grid-connected inverter with electrolytic and film capacitors is verified with an experimental set-up in the laboratory. In the absence of a PV simulator, a dc power supply in the voltage source mode is used instead. The software controllers are implemented in dSPACE DS1104 controller residing in a Hewlett Packard workstation. An IGBT inverter with a switching frequency of

10 kHz is used. A three-phase step-down transformer with 415V / 200V is used to connect to the grid in the laboratory. A photograph of the experimental set-up is shown in Fig. 9. The parameters of the experimental set-up are in Table II. Experiments are conducted at 3.0 kW power level. The results are broadly classified as steady state and transient.

TABLE II. PARAMETERS OF THE THREE-PHASE PV-INVERTER UNDER EXPERIMENT STUDY

Rated power of dc power supply	P_{max}	3000 W
Rated voltage of dc power supply	V_{pv}	400 V
DC capacitors (Electrolytic) (EPCOS type 2 x B43564-A5228, 2200 μ F in series)	C_{dc}	1100 μ F
DC capacitors (metallized polypropylene film) AVX type FFVE6C0476K, 900V dc	C_{dc}	47 μ F
External line inductor	L_f	5 mH
Inverter switching frequency	f_s	10 kHz

A. Steady State

1. Results with 1100- μ F Electrolytic Capacitor

The equivalent capacitance of 1100 μ F is achieved with two EPCOS type B43456-A5228 2200 μ F / 450 V capacitors connected in series. The three-phase inverter is connected to the 415-V grid through a 200 V / 415 V transformer and 380 V dc was applied to the dc input. The grid voltage v_{an} , the grid current i_a and power supply current i_{pv} are recorded as shown in Fig. 10. The dc voltage and ripple voltage across the capacitor and the capacitor current are recorded as indicated in Fig. 11. The ripple voltage across the capacitor is close to 2 V at 380 V dc, which is 0.52% as seen in Fig. 11(a). The capacitor current i_{cap} has a peak-peak value of 16 A as seen in Fig. 11(b).

2. Results with 47- μ F Film Capacitor

The two series-connected EPCOS 2200- μ F electrolytic capacitors are replaced by an AVX type FFVE6C0476K, 900V dc 47- μ F film capacitor and the steady state measurements are repeated as taken for 1100- μ F electrolytic capacitor. The grid voltage v_{an} , grid current i_a and power supply current i_{pv} are recorded as shown in Fig. 12. The dc voltage and ripple voltage across the capacitor and the capacitor current i_{cap} are recorded as indicated in Fig. 13. The ripple voltage is close to 9 V for a 380 V dc bus, which is 2.36% as seen in Fig. 13(a). The capacitor current has a peak-peak value of 32 A as seen in Fig. 13(b).

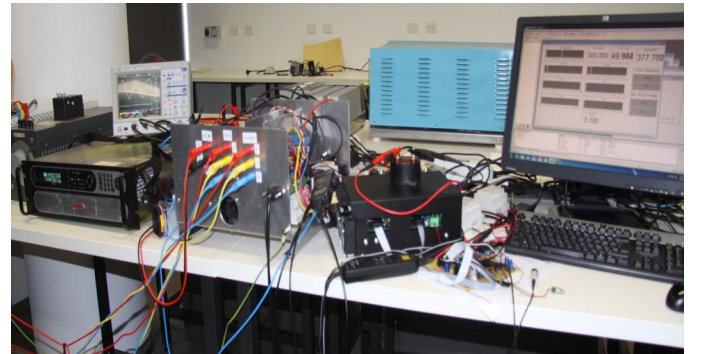


Fig. 9. Experimental set-up of the three-phase grid-connected PV-inverter.

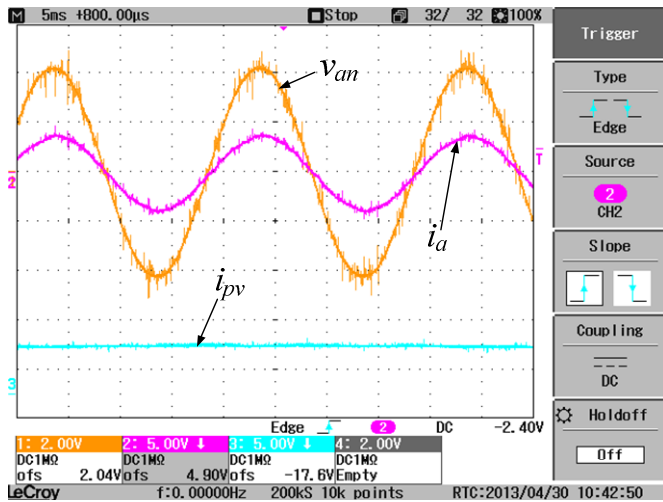


Fig. 10. Experimental results Case I - dc-link capacitor value of 1100 μF electrolytic type: grid voltage v_{an} (80V/div), grid current i_a (15A/div) and PV array current i_{pv} (7.5A/div).

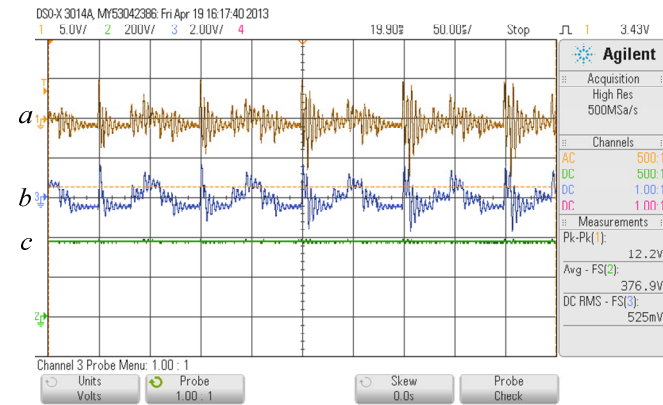


Fig. 11. Experimental results Case I - dc-link capacitor value of 1100 μF electrolytic type: detailed view of the dc-link voltage v_{pv} showing the ripple (5V/div), capacitor current i_{cap} (20A/div) and dc-link voltage v_{pv} (200V/div).

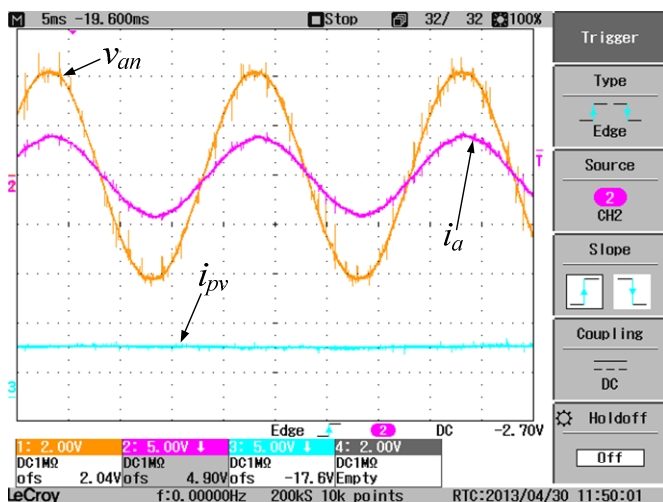


Fig. 12. Experimental results Case II - dc-link capacitor value of 47 μF film type: grid voltage v_{an} (80V/div), grid current i_a (15A/div) and dc input current i_{pv} (7.5A/div).

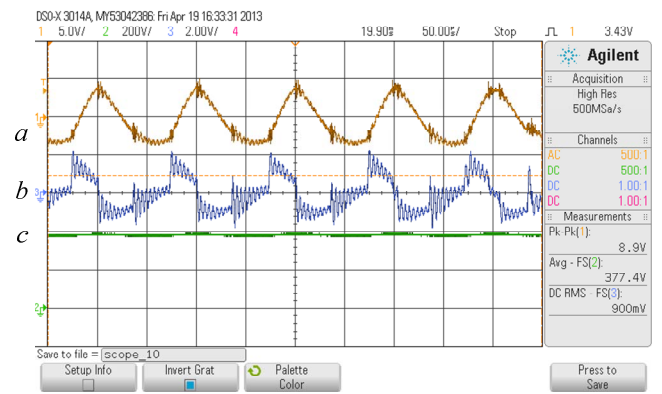


Fig. 13. Experimental results Case II - dc-link capacitor value of 47 μF film type: detailed view of dc-link voltage v_{pv} showing the ripple (5V/div), capacitor current i_{cap} (20A/div) and dc-link voltage v_{pv} (200V/div).

B. Transient

The transient performance of the grid-connected inverter is verified after replacing the two series-connected 2200- μF electrolytic capacitors in the inverter with one 47- μF film capacitor.

1. Results with 1100- μF Electrolytic Capacitor

The grid voltage v_{an} , grid current i_a and the input power supply current i_{pv} are recorded for a step change in the dc input voltage from 0 V to 380 V. The results are shown in Fig. 14. Also, the transient results are recorded for a step change in the dc input voltage from 380 V to 0 V as shown in Fig. 15. A smooth transition in the grid current is observed as seen in Fig. 15(b).

2. Results with 47- μF Film Capacitor

The transient performance of the grid-connected inverter is compared after replacing the two series-connected 2200- μF electrolytic capacitors with one 47- μF film capacitor. One result is recorded for a step change in the dc input voltage

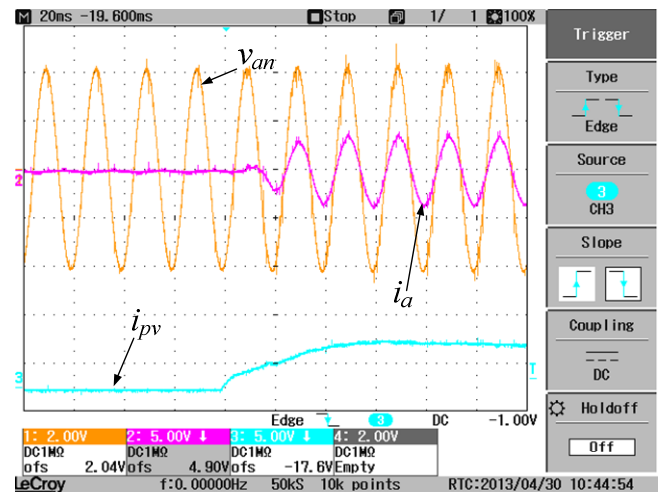


Fig. 14. Experimental results Case I - dc-link capacitor value of 1100 μF electrolytic type: grid voltage v_{an} (80V/div), grid current i_a (15A/div) and dc input current i_{pv} (7.5A/div) when the 380V dc is switched on.

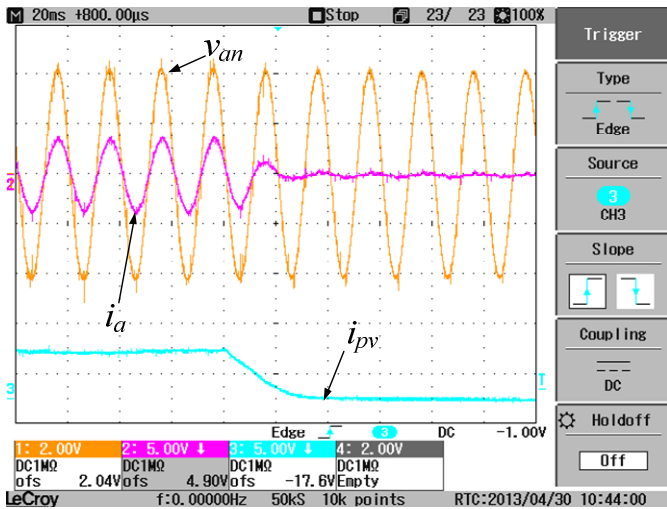


Fig. 15. Experimental results Case I - dc-link capacitor value of 1100 μF electrolytic type: grid voltage v_{an} (80V/div), grid current i_a (15A/div) and dc input current i_{pv} (7.5A/div) when the 380 V dc is switched off.

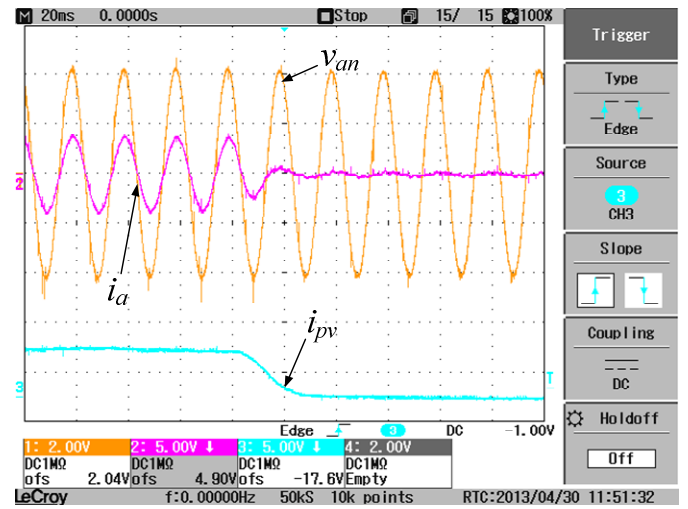


Fig. 17. Experimental results Case II - dc-link capacitor value of 47 μF film type: grid voltage v_{an} (80V/div), grid current i_a (15A/div) and dc input current i_{pv} (7.5A/div) when the 380V dc is switched off.

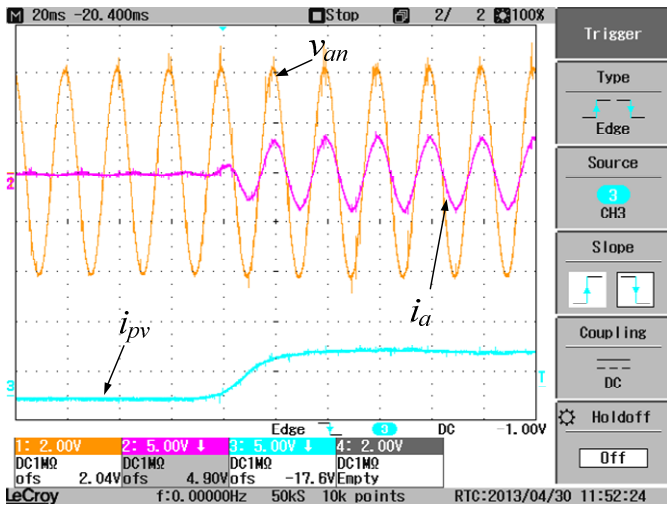


Fig. 16. Experimental results Case II - dc-link capacitor value of 47 μF electrolytic type: grid voltage v_{an} (80V/div), grid current i_a (15A/div) and dc input current i_{pv} (7.5A/div) when the 380 V dc is switched on.

from 0 V to 380 V as shown in Fig. 16. The test was repeated for a step change in the dc input voltage from 380 V to 0 V as shown in Fig. 17. The 47- μF film capacitor is found adequate for this grid-connected three-phase inverter system.

VI. CONCLUSION

A three-phase grid-connected PV-inverter with polypropylene film capacitor for dc-link decoupling has been presented in this paper. The aim is to extend the operating lifetime of the PV-inverter to match that of the PV panel.

The steady state and transient measurements carried out with grid-connected PV systems with electrolytic and film capacitors are found to be in close agreement that large dc-link electrolytic capacitors can be replaced by small film capacitors without compromising the system's performance.

In this case, the small increase in ripple in the dc current i_{pv} and the ripple in the dc voltage v_{pv} are within acceptable limits.

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