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# Array imaging of localized objects in homogeneous and heterogeneous media

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Abstract. We present a comprehensive study of the resolution and stability properties of sparse promoting optimization theories applied to narrow band array imaging of localized scatterers. We consider homogeneous and heterogeneous media, and multiple and single scattering situations. When the media is homogeneous with strong multiple scattering between scatterers, we give a non-iterative formulation to nd the locations and re ectivities of the scatterers from a nonlinear inverse problem in two steps, using either single or multiple illuminations. We further introduce an approach that uses the top singular vectors of the response matrix as optimal illuminations, which improves the robustness of sparse promoting optimization with respect to additive noise. When multiple scattering is negligible, the optimization problem becomes linear and can be reduced to a hybrid- $_1$  method when optimal illuminations are used. When the media is random, and the interaction with the unknown inhomogeneities can be primarily modeled by wavefront distortions, we address the statistical stability of these methods. We analyze the uctuations of the images obtained with the hybrid-1 method, and we show that it is stable with respect to di erent realizations of the random medium provided the imaging array is large enough. We compare the performance of the hybrid- $_1$  method in random media to the widely used Kirchho migration and the multiple signal classi cation methods.

AMS classi cation scheme numbers: 34B27, 78A46, 78A48

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# 1. Introduction

The recent mathematical theory of compressed sensing [18, 19, 20, 11, 12] has been shown to be very promising in a number of areas as diverse as medicine [29], biomedicine [37], geophysics [39], radar [3], astronomy [6], or microscopy [40]. Most inverse problems in these areas are considered to be underdetermined, meaning that we do not have unique solutions and, therefore, it is apparently impossible to identify which one is indeed the correct one. What makes compressed sensing at once interesting is that, often, the sought solution is known to be structured in the sense that it is sparse or compressible, which means that it depends upon a small number of parameters. This additional information changes the imaging problem dramatically because we can exploit the sparsity of the image and look for the simplest one that tends to be the right one.

In this paper, we study narrow-band, active array imaging of a small number of localized scatterers using both single and multiple illuminations. The goal is to determine the positions and re ectivities of the scatterers from the echoes recorded at an array of sensors when a few narrow band signals are sent to probe the medium. By localized scatterers we mean scatterers whose diameter is small compared to the wavelength. Hence, the Foldy-Lax approximation to the wave equation can be used to model wave propagation in the medium [26, 27, 28]. The number of scatterers is small because only a small portion of the region of interest is occupied by scatterers and, thus, the image we wish to recover is sparse. We study the case in which the interaction between the scatterers is strong so that multiple scattering is important, and the case in which the interaction is small so that multiple scattering is negligible. We consider imaging in homogeneous media and imaging in randomly inhomogeneous media with signi cant scattering from the inhomogeneities. We restrict this study to the case in which the full waveform at the array is available for imaging, which means that in the frequency domain both amplitudes and phases can be measured and recorded. For the case in which the phases cannot be recorded we refer to [15, 22, 23, 13, 14, 32, 31].

In this work we consider narrow-band array systems and, therefore, the frequency diversity content of the data measured at the array is very limited. There is an extensive literature on imaging techniques that deal with this problem. Kirchho migration [4], matched eld imaging [1], and Multiple Signal Classi cation (MUSIC) [36] are among the most used techniques. As in narrow-band array imaging the data is scarce, and hence, there are in nitely many con gurations of scatterers that match the data set, we formulate active array imaging as an optimization problem with constraints [16, 17]. When multiple scattering between the scatterers is important, the resulting problem is nonlinear, and therefore, it is apparently impossible to solve the optimization problem non-iteratively [17]. We show, however, that the nonlinearity can be avoided through a two-step process that e ectively linearizes the inverse problem. In the rst step, we treat the scatterers as equivalent sources and we recover their locations and strengths. In the second step, once the locations of the scatterers are xed, we recover their true re ectivities using a known relationship between the source strengths and the scatterer re ectivities. This is an explicit relation that comes from the Foldy-Lax equations, given the scatterer locations and the illumination.

When multiple scattering is signi cant some scatterers may be obscured due to screening e ects. Therefore, not all the scatterers may be recovered from data generated by a single illumination. Indeed, given an array illumination, multiple scattering may reduce the e ective illumination at certain locations due to destructive interferences of secondary sources coming from all the scatterers. To mitigate this additional problem of multiple scattering, we will discuss the use of multiple illuminations. The resulting optimization problem with multiple illuminations will be formulated as a joint sparsity recovery problem where we seek an unknown matrix whose columns share the same support. Thus, we seek for solution vectors corresponding to di erent illuminations that have a common support but have possibly di erent nonzero values.

The key point of the proposed two-step approach is the possibility of exact recovery of the locations of the equivalent sources in the rst step. We give conditions on the array imaging setup and the measurement noise level under which the locations of the sources can be recovered exactly. The uniqueness and stability of the solution are analyzed, showing that the errors are proportional to the amount of noise in the data with a proportionality factor that depends on the sparsity of the solution and the mutual coherence of the sensing matrix [17]. These conditions are given for general imaging con gurations. Conditions on the resolution of the images that guarantee exact recovery in the paraxial regime are derived in [21] for scatterers whose range is known. A more general paraxial model with scatterers at di erent ranges from the array is considered in [10]. The interesting case of imaging scatterers with small o -grid displacements are studied in [24, 10]. In [24], a simple perturbation method is proposed to reduce the gridding error for o -grid scatterers. The authors in [10] also present a very nice discussion on how to interpret the results obtained with  $_1$  minimization when modeling errors due to o -grid displacements are signi cant.

For the case in which multiple scattering can be ignored, we introduce a hybrid approach that combines the use of the singular value decomposition (SVD) of the data matrix with  $_1$  minimization [16]. We use the top right singular vectors of this matrix as illumination vectors to collect the data. Then, we project the data onto the subspace spanned by the top left singular vectors to lter out the unnecessary data and the noise, and to reduce the dimension of the linear system. Finally,  $_1$  optimization is applied to this reduced linear system to obtain the sparsest solution. This hybrid- $_1$  method turns out to be very useful when imaging in random media.

Imaging in random media is fundamentally di erent from imaging in homogeneous or smoothly varying media. When the medium is inhomogeneous we know, at best, the large scale, but we cannot known the small scale structure. Hence, when the small structure of a medium is important we model it as a random spatial process. In these cases, it is essential to take into consideration the statistical stability of the images, which refers to the robustness of the imaging methods with respect to di erent realizations of the medium. In fact, many of the usual imaging methods used in homogeneous (or smoothly varying) media fail, even for broadband signals and large arrays, because the images become noisy and change unpredictably with the detailed features of the uctuations of the medium. This is the case, for example, of the images obtained with Kirchho migration that depend on the particular realization of the random medium, and thus, they become unstable. Statistical stability typically holds only in broadband, we refer to [7, 8, 9] for details. Here, we consider narrow-band array systems. We show that, in these cases, large arrays are essential to stabilize the images when the medium is random. In particular, we show that the hybrid- 1 method and MUSIC are e cient and robust when the arrays are large enough. We compare, using numerical simulations, the images obtained with these methods with those obtained with Kirchho migration. The numerical simulations show that the hybrid- 1 method becomes stable faster than MUSIC as the array size increases. Kirchho migration is, as expected, unstable even for very large arrays.

The analysis of imaging in random media is done using a relatively simple random phase model for the e ects of the random medium. This model characterizes wave propagation in the high-frequency regime in random media with weak uctuations and small correlation lengths compared to the wavelength. It is widely used, for example, in adaptive optics to compensate for rapidly changing distortions in the received wavefronts due to the atmospheric turbulence caused by changing temperature and wind conditions.

The paper is organized as follows. In section 2, we formulate the array imaging problem in homogeneous media when the multiple scattering is important. In section 3, we describe the optimization methods that determine the locations and re ectivities of scatterers using a two-step non-iterative approach, with and without multiple illuminations. In section 4, we consider the case where multiple scattering is negligible, and we give a hybrid-  $_1$  method that improves the resolution of the image. In section 5, we consider imaging in random media. Using the simple random phase model, we show that the hybrid-  $_1$  and the MUSIC methods are statistically stable provided the arrays are large. The e ectiveness of all the methods is illustrated in various numerical examples with comparisons to other imaging methods in each of the sections. Section 6 contains our conclusions. The proofs of the theoretical results are given in the appendix.

# 2. Data model

Probing of the medium can be done with many di erent types of arrays, transmitters and recording devices. Also transducers that can both sense and transmit are usually employed. Besides, the geometric layout of the arrays depend on the application (acoustics, seismology, radar, ...) and they may be arranged in a transmission or a backscattering con guration. To x ideas, we will consider an active array consisting of N transducers located at positions  $\boldsymbol{x}_i$ , i = 1 N, placed in front of the medium to be probed (see Fig. 1 (a)). Here, and in the rest of the paper, we use boldface lower case letters for vectors, capital letters in boldface for matrices, and the corresponding letters, without boldface, for the entries of the matrices. To ensure that the transducers behave

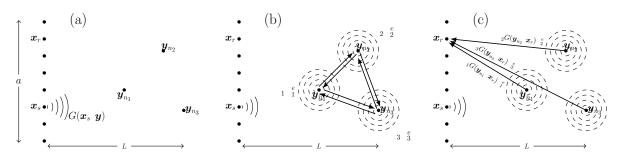


Figure 1. Schematic. (a) A linear array with N = 8 transducers probe a medium sending a spherical wave from  $\boldsymbol{x}_s$ ; the illumination vector is  $\boldsymbol{f} = \begin{bmatrix} 0 & 0 & 0 & 0 & f_6 & 0 & 0 \end{bmatrix}$ . The medium is either homogeneous or inhomogeneous and the corresponding Green s function is  $G(\boldsymbol{x} \ \boldsymbol{y})$ . (b) There are M = 3 point-like scatterers with re ectivities  $_j$  at positions  $\boldsymbol{y}_{n_j}$ ,  $j = 1 \ 2 \ 3$ . The interaction between them is strong creating e ective (or secondary) sources  $_{n_j} = _j \stackrel{e}{_j}$  that depend on all the scatterers positions and their re ectivities. If those are known, the e ective elds  $\stackrel{e}{_j}$  can be computed by solving the Foldy-Lax equations described in Appendix A. (c) The response received at  $\boldsymbol{x}_r$  is the superposition of all the scattered waves  $\stackrel{s}{_j} = _j G(\boldsymbol{y}_{n_j} \ \boldsymbol{x}_r) \stackrel{e}{_j}$  from the scatterers.

like an array of aperture  $a = (N \ 1)h$ , and not like separate entities, they are separated by a distance h of the order of wavelength  $= \frac{2 \ c_0}{2}$  of the probing signals, where  $c_0$  is the wave speed in homogeneous medium and is the corresponding frequency.

We will assume that the object we wish to image consists of M randomly positioned point-like scatterers. The medium can be homogeneous or inhomogeneous. Multiple scattering among the scatterers may or may not be important. All the scatterers, with unknown re ectivities  $_j \quad \mathbb{C}$ , where  $\mathbb{C}$  stands for the complex eld, and positions  $\boldsymbol{y}_{n_j}$ ,  $j = 1 \qquad M$ , are within a region of interest called the image window (IW), which is centered at a distance L from the array. We discretize the IW using a uniform grid of points  $\boldsymbol{y}_j$ ,  $j = 1 \qquad K$ , and we introduce the true *re ectivity vector* 

$$_{0} = \begin{bmatrix} 0 & 0 \end{bmatrix}^{T} \quad \mathbb{C}^{K}$$

such that  $_{0k} = \int_{j=1}^{M} j \mathbf{y}_{n_j} \mathbf{y}_k \quad k = 1$  K where is the classical Kronecker delta and  $^T$  is the transpose only operation, while stands for the conjugate transpose. We further assume that each scatterer is located at one of the K grid points, so  $\mathbf{y}_{n_1} \quad \mathbf{y}_{n_M} \quad \mathbf{y}_1 \quad \mathbf{y}_K$ . For a study of o -grid scatterers we refer to [24, 10].

To write the data received on the array in a compact form, we de ne the Green's function vector

$$\boldsymbol{g}(\boldsymbol{y}; \ ) = [G(\boldsymbol{x}_1 \ \boldsymbol{y}; \ ) \ G(\boldsymbol{x}_2 \ \boldsymbol{y}; \ ) \qquad G(\boldsymbol{x}_N \ \boldsymbol{y}; \ )]^T$$
(2.1)

at location  $\boldsymbol{y}$  in the IW, where  $G(\boldsymbol{x} \ \boldsymbol{y}; \ )$  denotes the free-space Green s function of the (homogeneous or inhomogeneous) medium that characterizes the propagation of a signal of angular frequency from point  $\boldsymbol{y}$  to point  $\boldsymbol{x}$ . When the medium is homogeneous,

$$G(\boldsymbol{x} \ \boldsymbol{y}; \ ) = G_0(\boldsymbol{x} \ \boldsymbol{y}; \ ) = \frac{\exp(\mathbf{i} \ \boldsymbol{x} \ \boldsymbol{y})}{4 \ \boldsymbol{x} \ \boldsymbol{y}} = \frac{1}{c_0}$$
(2.2)

and we have the Green s function vector in a homogeneous medium as

$$m{g}_0(m{y};\ ) = [G_0(m{x}_1\ m{y};\ )\ G_0(m{x}_2\ m{y};\ )\ G_0(m{x}_N\ m{y};\ )]^T$$

If  $\boldsymbol{f} = [f_1 \quad f_N]^T$  is the illumination vector whose entries are the signals sent from the transmitters in the array, then  $\boldsymbol{g}(\boldsymbol{y}; \boldsymbol{\beta})$  gives the eld at position  $\boldsymbol{y}$  in a free-space.

We further introduce the N - K sensing matrix

$$\boldsymbol{\mathcal{G}}(\ ) = [\boldsymbol{g}(\boldsymbol{y}_1;\ ) \quad \boldsymbol{g}(\boldsymbol{y}_K;\ )]$$
(2.3)

that maps a distribution of sources in the IW to the data received on the array. With this notation, the full response matrix, can be written as

$$\boldsymbol{P} = \boldsymbol{\mathcal{G}} \operatorname{diag}({}_{0}) \mathbf{Z}^{-1}({}_{0}) \boldsymbol{\mathcal{G}}^{T} = \boldsymbol{\mathcal{G}} \operatorname{diag}({}_{0}) \boldsymbol{\mathcal{G}}^{T}_{FL}({}_{0})$$
(2.4)

Here, and in all that follows, we drop the dependence of waves and measurements on the frequency  $\therefore$  In (2.4), **Z**<sup>-1</sup>( $_{0}$ ) denotes the inverse of the K - K Foldy-Lax matrix

$$(Z(_{0}))_{ij} = \begin{array}{c} 1 & i = j \\ \\ _{0j}G(\boldsymbol{y}_{i} \ \boldsymbol{y}_{j}) & i = j \end{array}$$
(2.5)

which depends on the unknown re ectivity vector  $_{0}$  (see Appendix A). To motivate (2.4), consider an illumination vector  $\boldsymbol{f} = [f_{1} \quad f_{N}]^{T}$  (see Fig. 1). Then,  $\boldsymbol{\mathcal{G}}_{FL}^{T}(_{0})\boldsymbol{f}$  gives the total eld at each grid point of the IW, including multiple scattering between the scatterers and the interaction with the unknown inhomogeneities of the medium. The total eld  $\boldsymbol{\mathcal{G}}_{FL}^{T}(_{0})\boldsymbol{f}$  is re ected by the scatterers on the grid that have re ectivities given by the vector  $_{0}$ , and then it is backpropagated to the array by the matrix  $\boldsymbol{\mathcal{G}}$ . All the available information for imaging is contained in the array response matrix (2.4). If transmitters and receivers are located at the same positions, then (2.4) is symmetric.

For a xed array con guration, wave propagation is completely described by the full response matrix P (2.4). Indeed, the data received on the array due to an illumination vector f is given by

$$\boldsymbol{b} = \boldsymbol{P}\boldsymbol{f} \tag{2.6}$$

#### 3. Active array imaging in homogeneous media

In this section, we formulate the inverse problem of active array imaging when the medium is homogeneous and, therefore, the wavefronts are not distorted. In this case, the response matrix can be written as

$$\boldsymbol{P} = \boldsymbol{\mathcal{G}}_{0} \operatorname{diag}({}_{0}) \boldsymbol{Z}^{1}({}_{0}) \boldsymbol{\mathcal{G}}_{0}^{T} = \boldsymbol{\mathcal{G}}_{0} \operatorname{diag}({}_{0}) \boldsymbol{\mathcal{G}}_{0FL}^{T}({}_{0})$$
(3.1)

where  $\mathbf{\mathcal{G}}_0 = [\mathbf{g}_0(\mathbf{y}_1) \quad \mathbf{g}_0(\mathbf{y}_K)]$  denotes the sensing matrix in a homogeneous medium, and  $\mathbf{Z}^{-1}(_0)$  is the inverse of the Foldy-Lax matrix (A.6) with  $G(\mathbf{y}_i \ \mathbf{y}_j) = G_0(\mathbf{y}_i \ \mathbf{y}_j)$ (see Appendix A). The object to be imaged is an ensemble of small but strong scatterers whose mutual interaction cannot be ignored. To determine their positions and re ectivities we use the collected data using a single illumination  $\mathbf{f}$  in subsection 3.1, and using multiple illuminations  $\mathbf{f}^j$ , j = 1, in subsections 3.2 and 3.3. In signal processing literature, the corresponding problems are called *Single Measurement Vector (SMV)* and *Multiple Measurement Vector (MMV)* problems, respectively.

#### 3.1. Imaging using single illumination

For a given illumination vector  $\boldsymbol{f}$ , we de ne the operator  $\boldsymbol{\mathcal{A}}_{f}$  via

$$oldsymbol{\mathcal{A}}_{f}[ \ _{0}] \ _{0} = oldsymbol{P}oldsymbol{f}$$

which maps the re-ectivity vector  $_0$  to the data (2.6). From (3.1) it follows that

$$\boldsymbol{\mathcal{A}}_{f}[] = [g_{f}(\boldsymbol{y}_{1})\boldsymbol{g}_{0}(\boldsymbol{y}_{1}) \qquad g_{f}(\boldsymbol{y}_{K})\boldsymbol{g}_{0}(\boldsymbol{y}_{K})]$$
(3.2)

where  $g_f(\boldsymbol{y}_j) = \boldsymbol{g}_{FL}^T(\boldsymbol{y}_j)\boldsymbol{f}$ , j = 1 K, are scalars meaning the total eld at the grid points  $\boldsymbol{y}_j$  due to the illumination  $\boldsymbol{f}$ , with  $\boldsymbol{g}_{FL}(\boldsymbol{y}_j)$  being the  $j^{\text{th}}$  column of the matrix  $\boldsymbol{\mathcal{G}}_{0FL}$ . Using this notation, active array imaging with a single illumination amounts to nding the unknown re-ectivity vector  $_0$   $\mathbb{C}^K$  from the system of N equations

$$\boldsymbol{\mathcal{A}}_{f}[] = \boldsymbol{b} \tag{3.3}$$

In a typical array imaging conguration, the number of transducers N is much smaller than the number of the grid points K in the IW and, hence, (3.3) is underdetermined. Furthermore, due to the multiple scattering,  $g_f(\boldsymbol{y}_j)$ , j = 1 K, depend on the unknown rejectivity vector  $_0$ , which makes (3.3) nonlinear with respect to  $_$ . Such nonlinearity makes us think that non-iterative inversion is inapplicable to solve (3.3). However, by rearranging the terms in these equations, we can reformulate the problem to solve for the locations of scatterers directly, without any iteration, and then to recover the rejectivities of each scatterer in a second single step, as we explain next.

3.1.1. Support recovery The localization problem is by far much more di cult than the estimation of re ectivities, which is a straightforward inversion if the former is exact. To localize the scatterers without any iteration, we introduce the *e ective source vector* 

$$_{f} = \operatorname{diag}()\mathbf{Z}^{-1}()\boldsymbol{\mathcal{G}}_{0}^{T}\boldsymbol{f}$$

$$(3.4)$$

and seek for its support. Then, according to (3.1)  $\boldsymbol{\mathcal{A}}_{f}[\] = \boldsymbol{\mathcal{G}}_{0}_{-f}$ , and (3.3) becomes

$$\mathcal{G}_{0-f} = \boldsymbol{b} \tag{3.5}$$

which is linear for the new unknown f. Note that in the formulation (3.3) the operator  $\mathcal{A}_{f}[\]$  depends on the illumination f, whereas in (3.5) the unknown f is the one which depends on f.

It is important to emphasize that due to the existence of multiple scattering, the solution of (3.5) may not give all the support of  $_0$ . This is not a law of the new formulation, but an implicit problem of array imaging when multiple scattering is important. Indeed, it is possible that one or several scalars  $g_f(\boldsymbol{y}_j), j = 1$  K, are very small or even zero, and hence, the corresponding scatterers become dark. This is the well-known screening e lect which makes scatterers undetectable, and that is manifested in our formulation making some components of the elective source vector  $f_f$  arbitrarily small.

Since (3.5) is underdetermined and the e ective source vector  $_{f}$  is sparse because M = N = K, we solve the  $_{1}$  minimization problem

$$\min_{f=1} \quad \text{s.t.} \quad \boldsymbol{\mathcal{G}}_{0-f} = \boldsymbol{b} \tag{3.6}$$

when data is noiseless. When the data is contaminated by a noise vector e with nite energy, we then solve the relaxed problem

$$\min_{f=1} \quad \text{s.t.} \quad \boldsymbol{\mathcal{G}}_{0 \ f} \quad \boldsymbol{b}_{2} < \tag{3.7}$$

for some given positive constant . In (3.6) and (3.7),  $_{1} = K_{i=1 \ i}$  and  $_{2} = \frac{K_{i=1 \ i}^{2}}{K_{i=1 \ i}^{2}}$ .

The following theorem gives conditions under which (3.6) recovers the positions and the strengths of the e ective sources exactly if the data is noiseless.

**Theorem 3.1.** Assume that the resolution of the IW is such that

$$\max_{i=j} \frac{\boldsymbol{g}_{0}(\boldsymbol{y}_{i})\boldsymbol{g}_{0}(\boldsymbol{y}_{j})}{\boldsymbol{g}_{0}(\boldsymbol{y}_{i})_{2} \boldsymbol{g}_{0}(\boldsymbol{y}_{j})_{2}} < (3.8)$$

If the number of e ective sources M is such that M < 12, then  $_{0f} = \text{diag}(_{0})\mathbf{Z}^{-1}(_{0})\mathbf{\mathcal{G}}_{0}^{T}\mathbf{f}$  is the unique solution to (3.6) with support fully contained by that of  $_{0}$ .

**Proof.** Let  $T = n_j \ 1 \ j \ M$  be the set of indices corresponding to the support of  $_{0f}$ . Based on (3.8), the submatrix  $\mathcal{G}_{0T}$  formed by the columns  $n_1 \ n_M$  of  $\mathcal{G}_0$ , satis es that  $\mathcal{G}_{0T}\mathcal{G}_{0T}$  is full rank and diagonally dominant. According to the KKT conditions, we need to nd a Lagrange multiplier vector such that

$$\boldsymbol{a}_j = \operatorname{sign}(_{0j}) \quad \text{for} \quad j \quad T \tag{3.9}$$

and

$$\boldsymbol{a}_j < 1 \quad \text{for} \quad j \quad T \tag{3.10}$$

Let  $= \mathcal{G}_{0T}(\mathcal{G}_{0T}\mathcal{G}_{0T})^{-1}\mathbf{1}$  where  $\mathbf{1}$  is the vector whose entries are the sign pattern of  $_{0}$ . Then (3.9) is automatically satisfied because  $\mathcal{G}_{0T} = \mathbf{1}$ . For (3.10), choosing any column of  $\mathcal{G}_{0}$  not in the submatrix  $\mathcal{G}_{0T}$ , we have

$$a_{j}\mathcal{G}_{0T}(\mathcal{G}_{0T}\mathcal{G}_{0T})^{-1}\mathbf{1} \qquad (\mathcal{G}_{0T}\mathcal{G}_{0T})^{-1}\mathcal{G}_{0T}a_{j-1} \qquad (\mathcal{G}_{0T}\mathcal{G}_{0T})^{-1} - \mathcal{G}_{0T}a_{j-1} \qquad \frac{M}{1-M} < 1$$

where the last inequality is due to the condition M < 1 2.

**Remark 3.2.** Note that the condition (3.8) has to be satis ed only on the support of  $_{0f}$ . Hence, if the distance between the e ective sources is known a priori to be large so (3.8) holds for the set of indices corresponding to its support, then the discretization of the IW can be as small as desired.

The next theorem provides an important stability result for the problem (3.7), the proof of which is given by Theorem 4.3 in [17].

**Theorem 3.3.** For a given array imaging conguration, with resolution condition given by (3.8), the solution  $_{f}$  to (3.7) satisfies

$$f \qquad 0f \quad 2 \qquad \boxed{1 \qquad (M \quad 1)} \tag{3.11}$$

provided  $\mathbf{e}_{2}$   $\overline{1 + \frac{M(1 (M-1))}{(1 2M+)^{2}}}$ , where  $\mathbf{e}$  is the noise vector added to the data. Moreover, the support of  $\frac{1}{f}$  is fully contained in that of  $\frac{1}{0f}$ , and all the components satisfying

$$(_{0f})_j > \overline{1 \quad (M \quad 1)} \tag{3.12}$$

are within the support of f.

3.1.2. Re ectivity estimation Optimization (3.6) (or (3.7)) gives the e ective source vector  $_{f}$ . In a second step, we compute the true re ectivities from the solution of this problem at once. Let be the support of the recovered solution such that = M - M, and  $_{fM}$  be the solution vector on  $\therefore$  From (3.4), we have

$$f_{fM} = \operatorname{diag}(\mathbf{Z}^{-1}(\ _{M})\boldsymbol{\mathcal{G}}_{0}^{T}\boldsymbol{f}) \ _{M} = \operatorname{diag}(g_{f}(\boldsymbol{y}_{n_{1}}) \qquad g_{f}(\boldsymbol{y}_{n_{M}})) \ _{M}$$

where  $g_f(\boldsymbol{y}_{n_j}) = \boldsymbol{g}_{0FL}^T(\boldsymbol{y}_{n_j})\boldsymbol{f}$ . Note that the scalars  $g_f(\boldsymbol{y}_{n_j})$  are the exciting elds at the scatterers positions, and that the elds ective sources  $n_j$  are the true relative ectivities  $n_j$  multiplied by the exciting elds (see Fig. 1 (b)). Hence, using (A.3) we can compute  $g_f(\boldsymbol{y}_{n_j})$  explicitly as follows

$$g_{f}(\boldsymbol{y}_{n_{j}}) = \boldsymbol{g}_{0}^{T}(\boldsymbol{y}_{n_{j}})\boldsymbol{f} + \sum_{k=1 \ k=j}^{M} {}_{k}G_{0}(\boldsymbol{y}_{n_{j}} \ \boldsymbol{y}_{n_{k}}) \quad j = 1 \qquad M$$
(3.13)

Then, the true re-ectivities of the scatterers are recovered by

$$n_j = n_j g_f(\boldsymbol{y}_{n_j}) \quad j = 1 \qquad M \tag{3.14}$$

When the data contains additive noise, we choose the support of the solution recovered by (3.7) such that all the components of  $_{fM}$  satisfy (3.12).

#### 3.2. Imaging using multiple arbitrary illuminations

Imaging with a single illumination can be very sensitive to additive noise, especially when the noise level is high. Moreover, the *screening e ect* can cause the failure of recovering some of the scatterers. Note that, for a xed imaging con guration, the *screening e ect* depends on the illumination vector  $\mathbf{f}$  and the amount of noise in the data. When the e ective source at  $\mathbf{y}_j$  is below the noise level because  $g_f(\mathbf{y}_j)$  is small, the corresponding scatterer cannot be detected. This motives us to consider active array imaging with multiple illuminations. In this case, active array imaging is modeled as a joint sparsity recovery problem, in which we seek for a matrix solution  $\mathbf{X}$  whose columns share the same support. By increasing the diversity of illuminations, we are able to minimize the *screening e ect* and have higher chance of locating all the scatterers more stably. 3.2.1. Support recovery Instead of simply stacking multiple data vectors  $\boldsymbol{b}^{j}$  obtained from di erent illuminations  $\boldsymbol{f}^{j}$ , j = 1, and solving the corresponding augmented linear system as in section 3.1, we formulate the problem using the *MMV* approach where the unknown vectors j corresponding to each illumination  $\boldsymbol{f}^{j}$  are arranged into a matrix. More precisely, let  $\mathbf{B} = [\boldsymbol{b}^{1} \quad \boldsymbol{b}]$  be the matrix whose columns are the data vectors generated by all the illuminations, and  $\boldsymbol{X} = [1]$  be the unknown matrix whose  $j^{\text{th}}$  column corresponds to the *e ective source vector* j under illumination  $\boldsymbol{f}^{j}$ . Thus, we formulate the problem of active array imaging with multiple measurements as solving the matrix-matrix equation

$$\mathcal{G}_0 \mathbf{X} = \mathbf{B} \tag{3.15}$$

for X. The sparsity of the matrix variable X is characterized by the number of nonzero rows. Thus, we de ne the row-support of a given matrix X by

$$rowsupp(\boldsymbol{X}) = i: \quad j \text{ s.t. } X_{ij} = 0$$

When the matrix X degenerates to a column vector, the row-support reduces to the support of that vector. Similar to the  $_1$  norm relaxation used in the SMV problem in section 3.1, the sparsest solution using multiple illuminations is given by the solution to the convex problem

min 
$$_1(\boldsymbol{X})$$
 s.t.  $\boldsymbol{\mathcal{G}}_0 \boldsymbol{X} = \mathbf{B}$  (3.16)

where  $_1$  is a convenient convex relaxation of the size of rowsupp $(\mathbf{X})$ . As in [17], we take

$$_{1}(\boldsymbol{X}) = \bigwedge_{i=1}^{N} X_{i} \quad _{2} \tag{3.17}$$

where  $X_i$  is the *i*<sup>th</sup> row of the matrix. When the data vectors are contaminated with additive noise vectors  $e^j$ , j = 1, we have the matrix-matrix equation

$$\mathcal{G}_0 X = \mathbf{B} + \mathcal{E} \tag{3.18}$$

where  $\boldsymbol{\mathcal{E}} = [\boldsymbol{e}^1 \quad \boldsymbol{e}]$ , and we seek a solution to

min 
$$_1(\boldsymbol{X})$$
 s.t.  $\boldsymbol{\mathcal{G}}_0\boldsymbol{X}$   $\mathbf{B}_F <$  (3.19)

for a pre-speci ed constant , where the Frobenius norm is given by

$$X_{F} = {\begin{array}{*{20}c} N & K & & \frac{1}{2} \\ & & X_{ij}^{2} \end{array}} X_{ij}^{2}$$

**Remark 3.4.** Similar to Theorems 3.1 and 3.3 for array imaging with a single illumination, we have results regarding the exact recovery and stability of the solution to (3.16) and (3.19) for imaging using multiple illuminations. These results are proved in [17]. We note, however, that these results do not provide a quantitative improvement when the number of measurements increases. Intuitively, this lack of improvement is justi able since the measurements obtained from randomly chosen illuminations could

be rather ine ective. There is no guarantee that randomly picked illuminations bring more information useful for imaging. In practive, however, we do observe a general improvement in the images formed with multiple random illuminations. In section 3.3, we use selective illuminations, obtained from the SVD of the array response matrix  $\mathbf{P}$ to increase the e-ciency of the MMV approach.

3.2.2. Re ectivity estimation Once we obtain the matrix  $\mathbf{X}$  from (3.16) (or (3.19)), whose columns are the e ective sources corresponding to the di erent illuminations, we compute in a second step the true re ectivities as follows. For each component i in the support such that the stability condition in Theorem 4.3 in [17] is satisfied, we compute the re ectivities  $_{i}^{j}$  corresponding to each illumination j by applying (3.13) and (3.14). Then, we take the average  $\frac{1}{j=1}$   $_{j=1}^{j}$  as the estimated re ectivity.

# 3.3. Imaging using optimal illuminations

To further improve the e ciency, and especially the robustness of (3.19), we propose to use selective illuminations as multiple illumination vectors [17]. The optimal set of illumination vectors can be obtained systematically from the SVD of the full response matrix  $\boldsymbol{P}$ . If the full array response matrix is not available, they can be obtained from an iterative time reversal process as discussed in [16]. Let the SVD of  $\boldsymbol{P}$  be

$$\boldsymbol{P} = \boldsymbol{U} \quad \boldsymbol{V} = \begin{bmatrix} M \\ j = 1 \end{bmatrix} {}_{j=1}^{M} {}_{j} U_{j} V_{j}$$

where  $U_j$  and  $V_j$  are the left and right singular vectors, respectively, and the nonzero singular values  $_j$  are given in descending order as  $_1 _2 _M > 0$ , with M - M. When there is no noise in the data, M = M. Now, let the illumination vectors be the top right singular vectors, i.e.  $\mathbf{f}^j = V_j, j = 1$  M. Then, we have

$$\boldsymbol{\mathcal{G}}_{0}\boldsymbol{X}_{V} = \mathbf{B}_{opt} = \boldsymbol{P}\boldsymbol{V}_{1:} = \begin{bmatrix} 1 U_{1} & U \end{bmatrix} + \boldsymbol{\mathcal{E}}$$
(3.20)

where the data matrix  $\mathbf{B}_{opt}$  contains all the essential information for imaging the scatterers. The improvement of the e-ciency, when using optimal illuminations, comes from the fact that illuminations using singular vectors deliver most of the energy around the scatterers, even when multiple scattering is non-negligible. Therefore, taking a few top singular vectors is enough to focus around the scatterers that contribute to the data received on the imaging array.

# 3.4. Numerical experiments

We now present numerical simulations in two dimensions. In all the simulations shown below with a single illumination, we use the iterative shrinkage-thresholding algorithm GelMa proposed in [30] due to its exibility with respect to the choice of the regularization parameter used in the algorithm. In the simulations with multiple illuminations we use an extension of this algorithm described in [17].

Figure 2 shows five scatterers placed within an IW of size  $41\lambda \times 41\lambda$ , which is at a distance  $L = 100\lambda$  from a linear array with 100 transducers that are one wavelength  $\lambda$  apart (the spatial units of all the images is  $\lambda$ ). The amplitudes of the reflectivities of the scatterers  $|\alpha_j|$  are 2.96, 2.76, 2.05, 1.54 and 1.35. Their phases are set randomly in each realization. Note that for a given illumination  $\hat{f}$  and a scatterer configuration  $\rho_0$  with fixed amplitudes, the amount of multiple scattering depends on the realization of the phases in  $\rho_0$ . For the amplitudes of the reflectivities chosen here, the amount of multiple scattering over single scattering ranges between 50% and 100%.

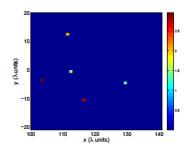


Figure 2. Original configuration of the scatterers in a  $41 \times 41$  image window with grid points separated by 1.

Figure 3 shows the images obtained by  $\ell_1$  norm minimization with 0% (left), 10% (middle), and 20% noise (right) when a single illumination probes the medium. When there is no noise in the data,  $\ell_1$  norm minimization recovers the positions and reflectivities of the scatterers exactly (the true locations are indicated with small white dots). However, when the data is corrupted by 10% and 20% of additive noise (middle and right images of Figure 3),  $\ell_1$  norm minimization fails. Some ghosts appear in the images and some scatterers are obscured because of *screening effect*.

When multiple illuminations are used we expect a general improvement in the images. Figure 4 shows the results of the MMV algorithm when 5 random illuminations are used. By multiple random illuminations we mean a set of illuminations coming, each one, from only one of the transducers on the array at a time, i.e.,  $\hat{f}_p = 1$  and  $\hat{f}_q = 0$  for  $q \neq p$ , with p chosen randomly at a time. The set of random illuminations used in each

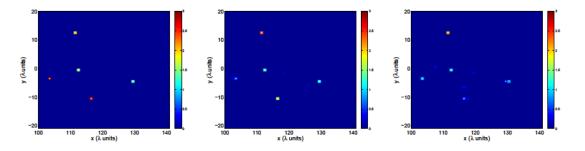


Figure 3. Images reconstructed by solving (3.6) and (3.7) when single illumination is used. From left to right, there is 0%, 10%, and 20% additive noise in the data.

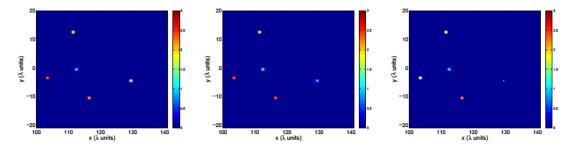


Figure 4. Images reconstructed by solving (3.19) when 5 random illuminations are used. From left to right, there is 10%, 20%, and 50% noise in the data.

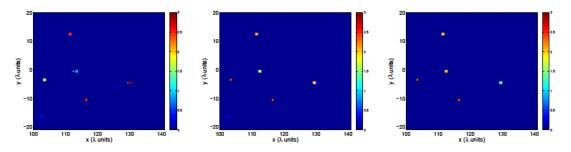


Figure 5. Images reconstructed by solving (3.19) when optimal illuminations are used. From left to right, the images are reconstructed by using 1, 2, and 3 top singular vectors as illumination vectors. There is 50% noise in the data.

image of Figure 4 is kept the same, with the level of additive noise changing from 10% (left), 20% (middle), to 50% (right). We observe that the images obtained with multiple illuminations are more stable with respect to additive noise. However, we still see some missing scatterers in the right image (50% of noise), and therefore, for this particular set of randomly chosen illuminations, imaging is still affected by *screening effect*.

Figure 4 shows the necessity of using "good" illuminations that maximize the information content of the data, especially when the signal-to-noise ratio (SNR) is low. Figure 5 displays the images obtained with 1 (left), 2 (middle), and 3 (right), optimal illuminations associated to the singular vectors  $\hat{f}^{j} = \hat{V}_{j}$ , with j = 1, 2, 3. It is remarkable that only a few of them (2 or 3) are enough to obtain very good images, even with 50% of additive noise in the data.

# 4. Single scattering as a special case: The hybrid- $\ell_1$ method

In situations where the scatterers are weak and/or are very far apart, the incident wave undergoes only one scattering event before coming back to the array. In these cases, the inverse of the Foldy-Lax matrix  $\mathbf{Z}^{-1}(\boldsymbol{\rho})$  in (3.1) becomes the identity matrix, and the array response matrix under the so called Born approximation becomes

$$\widehat{\boldsymbol{\Pi}} = \boldsymbol{\mathcal{G}}_0 \operatorname{diag}(\boldsymbol{
ho}_0) \boldsymbol{\mathcal{G}}_0^T$$
 .

Then, the data received on the array when an illumination vector  $\hat{f}$  probes the medium is given by  $\boldsymbol{b} = \widehat{\Pi} \widehat{f}$ . Similar to (3.3), we can model the data  $\boldsymbol{b}$  as applying the operator

r

 $\mathcal{A}_{f}$  to the unknown re ectivity vector  $_{0}$ . Because multiple scattering is negligible,  $g_{f}(\boldsymbol{y}_{j}) = \boldsymbol{g}_{0}^{T}(\boldsymbol{y}_{j})\boldsymbol{f}$  in (3.2), and thus, (3.3) is now linear in  $_{0}$ , which we denote as  $\mathcal{A}_{0f}$ . Therefore, array imaging of localized scatterers in homogeneous media when multiple scattering is negligible amounts to solving the linear optimization problem

$$\min_{1} \quad \text{s.t.} \quad \boldsymbol{\mathcal{A}}_{0f} = \boldsymbol{b} \tag{4.1}$$

In (4.1) we assume that the data is noiseless. If the data contains noise, we solve

$$\min_{1} \quad \text{s.t.} \quad \boldsymbol{\mathcal{A}}_{0f} \quad \boldsymbol{b}_{2} < \tag{4.2}$$

Because of the linearity of the equations, a single-step process is su-cient to recover both the locations and re-ectivities of the scatterers from (4.1) or (4.2). Moreover, when multiple scattering is negligible, there is no *screening e-ect*, and hence, solving the set of equations generated from a single illumination alone can produce reliable images. Note that Theorems 3.1 and 3.3 can be applied to problems (4.1) and (4.2) as well.

Similarly to the case where multiple scattering was important, we can use multiple illuminations and the MMV approach to nd the re ectivity vector  $_0$ . Furthermore, imaging using optimal illuminations has a simpler form and weaker requirements on the imaging con guration when multiple scattering is negligible. Indeed, assume that the data is noiseless, and let the SVD of array response matrix be  $= \boldsymbol{U} \cdot \boldsymbol{V}$ , with M singular values  $_j \stackrel{M}{_{j=1}}$  greater than 0. Following the MMV formulation in section 3.2, we have  $\boldsymbol{b}^j = V_j = _j U_j$  for j = 1 M. On the other hand,  $\boldsymbol{b}^j = \mathcal{A}_{0V_j} = \mathcal{G}_0 \stackrel{j}{_0}$ , where  $_0^j = \text{diag}(g_{V_j}(\boldsymbol{y}_k)) = \mathbf{f}_0$  is linear in  $_0$ , which is the major difference from (3.4). Now consider (3.20) with  $\boldsymbol{\mathcal{E}} = \mathbf{0}$  for symplicity, and project  $\boldsymbol{B}_{opt}$  onto the space spanned by the top left singular vectors corresponding to the significant singular values. Denoting  $\boldsymbol{\mathcal{U}} = [U_1 \quad U_M]$  and  $\boldsymbol{X}_V = [ \ 1 \ M ]$ , we get

$$\mathcal{U} \ \mathcal{G}_{0} \mathbf{X}_{V} = \begin{pmatrix} U_{1} \mathcal{G}_{0} & ^{1} & U_{1} \mathcal{G}_{0} & ^{2} & U_{1} \mathcal{G}_{0} & ^{M} \\ U_{2} \mathcal{G}_{0} & ^{1} & U_{2} \mathcal{G}_{0} & ^{2} & U_{2} \mathcal{G}_{0} & ^{M} \\ \vdots & \vdots & \vdots & \vdots \\ U_{M} \mathcal{G}_{0} & ^{1} & U_{M} \mathcal{G}_{0} & ^{2} & U_{M} \mathcal{G}_{0} & ^{M} \\ = \mathcal{U} \ \mathbf{B}_{opt} = \operatorname{diag}(\begin{array}{cc} 1 & & _{M} \end{pmatrix}$$

$$(4.3)$$

In (4.3), there are M unknowns  $^{1}$   $^{M}$ , and  $M^{2}$  equations. However, when multiple scattering is negligible we can associate a nonzero singular value to each scatterer and the corresponding singular vector is given by

$$U_{j} = \overline{V_{j}} \quad \frac{\boldsymbol{g}_{0}(\boldsymbol{y}_{j})}{\boldsymbol{g}_{0}(\boldsymbol{y}_{j})_{2}} \qquad j = 1 \qquad M$$

$$(4.4)$$

up to an arbitrary complex constant of modulus 1. Therefore, the rank-1 matrices  $V_j U_i$  have a (non-trivial) eigenvalue di erent from zero only when i = j. That is, only the diagonal blocks of the matrix  $\mathcal{U} \mathcal{G}_0 X_V$  contribute in (4.3), and hence, we can simplify (4.3) further as a set of M equations. Using (4.4) and

$$U_{i}\boldsymbol{\mathcal{G}}_{0}^{\ j} = \sum_{k=1}^{N} (U_{i}\boldsymbol{g}_{0}(\boldsymbol{y}_{k}))(\boldsymbol{g}_{0}(\boldsymbol{y}_{k}) \ V_{j})_{0k} = \sum_{k=1}^{N} \boldsymbol{g}_{0}(\boldsymbol{y}_{k})(V_{j}U_{i})\boldsymbol{g}_{0}(\boldsymbol{y}_{k})_{0k}$$

we can write (4.3) as a set of M equations linear in  $_0$  such that

$$\boldsymbol{\mathcal{B}}_{0} = \begin{bmatrix} 1 & M \end{bmatrix}^{T}$$
(4.5)

where

$$\boldsymbol{\mathcal{B}} = \frac{\overline{\boldsymbol{g}_{0}(\boldsymbol{y}_{1})U_{1}}\boldsymbol{g}_{0}^{T}(\boldsymbol{y}_{1})V_{1}}{\overline{\boldsymbol{g}_{0}(\boldsymbol{y}_{K})U_{1}}\boldsymbol{g}_{0}^{T}(\boldsymbol{y}_{K})V_{1}} \\ \boldsymbol{\mathcal{B}} = \frac{\ddots}{\overline{\boldsymbol{g}_{0}(\boldsymbol{y}_{1})U_{M}}\boldsymbol{g}_{0}^{T}(\boldsymbol{y}_{1})V_{M}} \frac{\overline{\boldsymbol{g}_{0}(\boldsymbol{y}_{K})U_{M}}\boldsymbol{g}_{0}^{T}(\boldsymbol{y}_{K})V_{M}}{\overline{\boldsymbol{g}_{0}(\boldsymbol{y}_{K})U_{M}}\boldsymbol{g}_{0}^{T}(\boldsymbol{y}_{K})V_{M}}$$
  
he hybrid- 1 method is to solve for \_\_\_0 from the SMV problem

Therefore, th

min <sub>1</sub> s.t. 
$$\boldsymbol{\mathcal{B}} = \begin{bmatrix} 1 & M \end{bmatrix}^T$$
 (4.6)

The advantage of hybrid-<sub>1</sub> method compared to the MMV formulation is its simpler form and lower dimensionality. Moreover, the hybrid  $_1$  method guarantees exact recovery, the su cient condition for which also implies that it can form images with higher resolution (see [16] for details).

**Theorem 4.1.** Assume that the scatterers are far part such that (4.4) is satis ed. Let E be the submatrix of  $\mathcal B$  formed by normalized columns corresponding to the scatterers location and S be the submatrix formed by the remaining normalized columns of  $\mathcal{B}$ . If  $oldsymbol{S}_{1-1} < 1$   $oldsymbol{E}_{1-1}$ , where  $oldsymbol{I}_{1-1}$  is the matrix 1-norm induced by the vector  $oldsymbol{I}_1$ norm, and I is the identity matrix, then  $_0$  is the unique solution to (4.6).

#### 4.1. Numerical experiments

Now we present the results of a numerical experiment that shows the performance of the hybrid-<sub>1</sub> method compared to those obtained with KM and MUSIC. We consider the same imaging set-up as in subsection 3.4. KM is a simple and robust imaging method (with respect to additive noise) which propagates the imaging data received on the array back to the medium. Under the setup of imaging in section 2, the KM imaging functional can be written as

$$_{KM} = \mathcal{A}_{0f} \boldsymbol{b} \tag{4.7}$$

MUSIC, on the other hand, is a subspace projection algorithm that uses the SVD of the array response matrix. The idea of MUSIC is to project the reference illumination vectors  $\boldsymbol{g}_0(\boldsymbol{y}_j), j = 1$  M, onto the noise space using the projection operator

$$\mathcal{P}\boldsymbol{g}_{0}(\boldsymbol{y}) = \prod_{j=1}^{M} (U_{j}\boldsymbol{g}_{0}(\boldsymbol{y}))U_{j} \quad \boldsymbol{g}_{0}(\boldsymbol{y})$$
(4.8)

Then, according to (4.4), the normalized functional

$$\mathcal{I}_{MUSIC}(\boldsymbol{y}_k) = \frac{\min_{1 \ j \ K} \mathcal{P}\boldsymbol{g}_0(\boldsymbol{y}_j)}{\mathcal{P}\boldsymbol{g}_0(\boldsymbol{y}_k)} k = 1 \qquad K$$
(4.9)

will peak at the search points  $\boldsymbol{y}_k$  only at locations where there is a scatterer. Imaging with MUSIC (4.9) only locates of the scatterer s positions. Their re ectivities are usually obtained using a separate procedure after the locations are known.

Figure 6 shows the simulation results with 100% of additive noise in the data. From left to right, we display the reference image, and the images obatined with KM, MUSIC and the hybrid- $_1$  method. In the simulations for MUSIC and hybrid- $_1$  we assume that the top 5 singular vectors have been obtained. In the simulations for KM we use 5 random illuminations. It is apparent that the image created by the hybrid- $_1$  method is better than the other two. The hybrid- $_1$  method forms a clear image with perfect recovery of the location of the scatterers, and accurate estimates of their re ectivities. The robustness of the hybrid- $_1$  method relies on the selection of the top singular values so that the noise in the subspace formed by singular vectors associated with small singular values is ltered out.

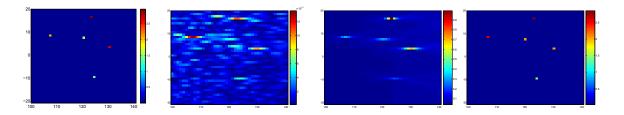


Figure 6. From left to right: original conguration of the scatterers, KM image (using 5 illuminations), MUSIC image, and hybrid- $_1$  image. There is 100% noise in the data.

# 5. Active array imaging in random media

Many natural media vary randomly in space, and therefore, waves propagating in these media also uctuate in space. Thus the data collected on the array inherit the uncertainty of the uctuations of the medium resulting in wave distortion. Because wave distortion e ects induced by the inhomogeneities are very di erent from additive uncorrelated noise, imaging through random media is very di erent from that in homogeneous media. For example, the images obtained by Kirchho migration become noisy and change unpredictably with the detailed features of the medium, and thus, are useless. In this section, we use a simple random phase model to analyze the performance of the hybrid- 1 and MUSIC methods in random media. Both methods use the SVD of the response matrix. The random phase model distorts the wavefronts but keeps the amplitude of the waves unchanged and is valid in the regime of geometrical optics.

#### 5.1. Random phase model

In random media, the Green s function that characterizes the wave propagation from  $\boldsymbol{x}$  to  $\boldsymbol{y}$  is given by the wave equation

$$G(\boldsymbol{x} \ \boldsymbol{y}) + {}^{2}n^{2}(\boldsymbol{x})G(\boldsymbol{x} \ \boldsymbol{y}) = (\boldsymbol{x} \ \boldsymbol{y})$$
(5.1)

where  $n(\boldsymbol{x}) = \frac{c_0}{c(\boldsymbol{x})}$  is the random index of refraction of the medium with local wave speed  $c(\boldsymbol{x})$ . In a homogeneous medium, we have  $c(\boldsymbol{x}) = c_0$  for any location  $\boldsymbol{x}$ , and

hence,  $G_0(\boldsymbol{x} \ \boldsymbol{y})$  defined in (2.2) is the solution to (5.1). In random media, however, the wave speed  $c(\boldsymbol{x})$  depends on the position  $\boldsymbol{x}$ . In this paper, we consider random uctuations of the wave speed satisfying the model

$$\frac{1}{c^2(\boldsymbol{x})} = \frac{1}{c_0^2} \ 1 + \quad (\frac{\boldsymbol{x}}{l})$$
(5.2)

where l is the correlation length of the medium, determines the strength of the ucutation around the constant speed  $c_0$ , and () is a stationary random process with zero mean and normalized autocorrelation function  $R(\mathbf{x} \ \mathbf{x}) = \mathbb{E}((\mathbf{x}) \ (\mathbf{x}))$ , so R(0) = 1. Here, we only consider weak uctuations such that 1.

When the index of refraction  $n(\mathbf{x})$  is random, there is no analytic solution to (5.1), and hence, numerical schemes are often used to obtained an approximated numerical solution. This can be time consuming, especially for large distances L. Instead, the random phase model provides an analytical approximation for the Green s function between two points at a distance of order L = l from each other, given by

$$G(\boldsymbol{x} \ \boldsymbol{y}) = G_0(\boldsymbol{x} \ \boldsymbol{y}) \exp i \quad \boldsymbol{x} \quad \boldsymbol{y} \quad \begin{bmatrix} \mathbf{x} \\ \mathbf{z} \end{bmatrix} + \frac{s}{l}(\boldsymbol{y} \ \boldsymbol{x}) ds \quad (5.3)$$

This approximation is valid when (i) the wavelength is much smaller than the correlation length l so the geometric optics approximation holds, (ii) the correlation length l is much smaller than the propagation distance L so the statistics of the phase is Gaussian, and (iii) the strength of the uctuation is small so the amplitude of the wave is kept unchanged, but large enough to ensure that the perturbations of the phases are not negligible. The last condition holds when  $\frac{2L^3}{l^3} = \frac{2}{2lL} = 1$  (for details, see [38, 35, 9]). Note that although we take weak uctuations, the distortion of the wavefronts by the inhomogeneities of the medium is observable because the wave travels long distances.

Comparing (5.3) to the homogeneous Green s function (2.2) we see that, in this regime, only the phase is perturbed by the random medium and the magnitude remains unchanged. The following result shows that the second order moment of (5.3) is close to the expected value.

**Proposition 5.1.** Assume that the autocorrelation function  $R \quad \mathbb{L}_2(\mathbb{R}_+)$  is di erentiable and its derivative R satis es  $\frac{R(t)}{t} \quad \mathbb{L}(\mathbb{R}_+)$  with exponential decay. Let  $\mathbf{y}_1$  and  $\mathbf{y}_2$  be two points on the same plane at a distance L from point  $\mathbf{x}$ , such that  $\mathbf{y}_1 \quad \mathbf{y}_2 \quad L$ . Then, for the Greens function (5.3) we have

$$\mathbb{E} \ G(\boldsymbol{x} \ \boldsymbol{y}_1) \overline{G(\boldsymbol{x} \ \boldsymbol{y}_2)} \qquad G_0(\boldsymbol{x} \ \boldsymbol{y}_1) \overline{G_0(\boldsymbol{x} \ \boldsymbol{y}_2)} e^{-\frac{2a_e^2}{2L^2}} \boldsymbol{y}_1 \ \boldsymbol{y}_2^2 \qquad (5.4)$$

with

$$a_e = L = 1 - \frac{2L}{3l} - \frac{R(t)}{t} dt^{\frac{1}{2}}$$
(5.5)

According to Proposition 5.1, the back-propagated signal in random media is equal to that in a homogeneous medium times a Gaussian factor with variance  $\frac{L^2}{2a_z^2}$ . The

length  $a_e$  is an intrinsic property of the random media. It depends on the propagation distance L and the statistics of the random uctuations of the medium. According to (5.4), as  $a_e$  increases, the refocused spot size for time reversal in random media is tighter. Using the moment estimate (5.4), we immediately have the following variance estimate.

**Corollary 5.2.** Under the same conditions given in Proposition 5.1, we have for (5.3)

$$\mathbb{E} G(\boldsymbol{x} \ \boldsymbol{y}_1) \overline{G(\boldsymbol{x} \ \boldsymbol{y}_2)} \quad \mathbb{E} G(\boldsymbol{x} \ \boldsymbol{y}_1) \overline{G(\boldsymbol{x} \ \boldsymbol{y}_2)}^2$$

$$\frac{1}{16^{-2} \boldsymbol{x} \ \boldsymbol{y}_1^{-2} \boldsymbol{x} \ \boldsymbol{y}_2^{-2}} \quad 1 \quad e^{-\frac{2a_e^2}{L^2} \boldsymbol{y}_1 \ \boldsymbol{y}_2^{-2}}$$
(5.6)

This property of the Green s function (5.3) is generally referred to as *self-averaging*, which means that a time-reversed, back-propagated signal in a random medium refocuses near the source independently of the particular realization of the random medium. Self-averaging properties of back-propagated signals in random media have been shown under broad-band settings, see for example [25, 5, 2, 33, 34], but are not true, in general, for narrow-band signals. Corollary 5.2 states that the Green s function corresponding to the phase random model (5.3) exhibits the same properties, even for narrow-band pulses.

The next result states that a single realization of a back-propagated signal via the Green s function vector g(y) (2.1) is stable when the aperture of the imaging array becomes in nity, in the sense that its value is close to the back-propagated signal averaged over multiple realizations.

**Proposition 5.3.** For a large aperture size a, the signal sent from  $y_1$ , recorded at the array, and back-propagated at  $y_2$  is statistically stable for (5.3) in the sense that

$$\frac{\mathbb{E} \boldsymbol{g} (\boldsymbol{y}_1) \boldsymbol{g}(\boldsymbol{y}_2) \quad \mathbb{E}(\boldsymbol{g} (\boldsymbol{y}_1) \boldsymbol{g}(\boldsymbol{y}_2))^2}{\mathbb{E} \boldsymbol{g}(\boldsymbol{y}_1) \ ^2 \mathbb{E} \boldsymbol{g}(\boldsymbol{y}_2) \ ^2} \quad 0 \qquad a$$
(5.7)

Furthermore, as shown in Appendix B, the decay rate of (5.7) is bounded by

1 e 
$$\frac{l^2}{L^2} \boldsymbol{y}_1 \, \boldsymbol{y}_2^{2} a_e^2 = \frac{l^2}{L^2 \log 1 + \frac{a^2}{4L^2}}$$
 (5.8)

for any aperture size a. Thus, for short distances so L is of order a, the decay rate is only logarithmic in a. However, when the distance is large such that L = a (e.g. the remote sensing regime), we can use a linear approximation on the logarithm function and obtain, up to a constant, the approximated decay rate

$$1 \quad e^{-\frac{2}{L^2}} \boldsymbol{y}_1 \quad \boldsymbol{y}_2^{-2} a_e^2 \qquad \frac{l}{a} \quad (5.9)$$

which implies a quadratic decay rate in a. Such quadratic decay rate in a can also be justi ed by using the paraxial approximation when L = a. As it is noted in Remark B.2, the decay rate in the paraxial approximation is given by (B.8), so the bound in (5.9) coincides with the paraxial approximation up to oscillations caused by the sinc function.

We emphasize that, although (5.8) indicates little improvement in the refocused spot size when the physical aperture of the array a is large, Proposition 5.3 implies that large arrays stabilize refocusing of narrow-band pulses in random media. The fact that large arrays improve the quality of imaging in cluttered media has been recognized for long time in radar and seismic imaging. The theoretical results obtained here using the random phase model support this empirical observation.

Next, we give a result that is essential for the performance of hybrid-  $_1$  and MUSIC methods in random media. When imaging in random media, the Green's function vector  $\boldsymbol{g}(\boldsymbol{y})$  is random and not known. The best we can do is to use the deterministic Green's function vector  $\boldsymbol{g}_0(\boldsymbol{y})$  which is, in general, quite di erent from the random one. Furthermore, replacement of  $\boldsymbol{g}(\boldsymbol{y})$  by  $\boldsymbol{g}_0(\boldsymbol{y})$  might give good results for some realizations of the random medium but not for others. The next result tells us that a single realization of a narrow-band backpropagated signal  $\boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2)$ , with the Green's function corresponding to the phase random model, is stable as the aperture of the array becomes in nity, which means that its value is close to the back-propagated signal averaged over multiple realizations.

**Proposition 5.4.** Let  $y_1$  and  $y_2$  be two points satisfying  $y_1$   $y_2$  L. Then, for the Greens function (5.3), we asymptotically have

$$\boldsymbol{g} (\boldsymbol{y}_1) \boldsymbol{g}(\boldsymbol{y}_2) \quad 0 \qquad a \tag{5.10}$$

and for the mixed inner products

$$\boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2) \quad 0 \qquad a \tag{5.11}$$

where the limit is under probability measure  $\mathbb{P}$  induced by (). Moreover, the mixed inner product is also statistically stable in the sense that

$$\frac{\mathbb{E} \boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2) - \mathbb{E}(\boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2))^2}{\boldsymbol{g}_0(\boldsymbol{y}_1)^{-2}\mathbb{E} \boldsymbol{g}(\boldsymbol{y}_2)^{-2}} = 0 \quad a$$
(5.12)

Proposition 5.4 implies that imaging in random media, i.e., a narrow-band pulse  $g(y_2)$  back-propagated via  $g_0(y_1)$  in the medium, is statistically stable for large arrays, and the pulse heading to the point  $y_1$  will focus around the correct location  $y_2$ .

#### 5.2. The hybrid- $_1$ method

Assume that multiple scattering between the scatterers is negligible, and wave distortion is well described by the random phase model (5.3). Then, the response matrix has the form

$$\boldsymbol{P} = \int_{j=1}^{M} {}_{j}\boldsymbol{g}(\boldsymbol{y}_{n_{j}})\boldsymbol{g}^{T}(\boldsymbol{y}_{n_{j}})$$
(5.13)

with random Green s function vectors  $\boldsymbol{g}(\boldsymbol{y}) = [G(\boldsymbol{x}_1 \ \boldsymbol{y}) \quad G(\boldsymbol{x}_N \ \boldsymbol{y})]^T$ , where  $G(\boldsymbol{x} \ \boldsymbol{y})$ is given by (5.3). Because multiple scattering is not important, the transformation  $\boldsymbol{\mathcal{A}}_f = \boldsymbol{b}$  that relates the re ectivity vector  $_0$  with the data measured at the array  $\boldsymbol{b}$  is linear. However, when the medium contains random inhomogeneities, this linear transformation is random because the column vectors of the operator  $\boldsymbol{\mathcal{A}}_f$  are random functions. Therefore, to solve for the re ectivity vector  $_0$ , we have to approximate the unknown Green's function vectors  $\boldsymbol{g}(\boldsymbol{y})$  by the ones for a homogeneous medium, after which,  $\boldsymbol{\mathcal{A}}_f$  becomes the linear operator  $\boldsymbol{\mathcal{A}}_{0f}$  given in section 4. The substitution of  $\boldsymbol{g}(\boldsymbol{y})$  by  $\boldsymbol{g}_0(\boldsymbol{y})$  introduces a discrepancy between model and data, and thus, we must solve the  $_1$  minimization problem (4.2) with an inequality constraint when data from a single illumination is available. We note that to estimate the error bound of such discrepancy is nontrivial due to the existing correlation in the measurement noise in random media.

According to Proposition 5.4, when the scatterers are far apart, the random Green s function vectors  $\boldsymbol{g}(\boldsymbol{y}_{n_j}), j = 1$  M, are approximately orthogonal in probability as the size of the imaging array becomes large. In this case, we can associate to each of the M scatterers a nonzero singular value  $_j = _j \boldsymbol{g}(\boldsymbol{y}_{n_j}) _2^2$  with singular vectors

$$U_{j} = \frac{\boldsymbol{g}(\boldsymbol{y}_{n_{j}})}{\boldsymbol{g}(\boldsymbol{y}_{n_{j}})_{2}} \quad V_{j} = \frac{\boldsymbol{g}(\boldsymbol{y}_{n_{j}})}{\boldsymbol{g}(\boldsymbol{y}_{n_{j}})_{2}} \qquad j = 1 \qquad M$$
(5.14)

If the full array response matrix  $\boldsymbol{P}$  (or its SVD) is available, we can use the hybrid- $_1$ as follows. We use the right singular vectors of  $\boldsymbol{P}$  as the illumination vectors, and we project the data to the space spanned by the left singular vectors in the same way as shown in section 4. Thus, we form the same hybrid- $_1$  optimization given in (4.6) and there is no need to estimate the error bound of the discrepancy between model and data. Recall that  $(ij)^{\text{th}}$  entry of the hybrid- $_1$  matrix  $\boldsymbol{\mathcal{B}}$  is related to the singular vectors in the form of  $\boldsymbol{g}_0(\boldsymbol{y}_j)U_i\boldsymbol{g}_0^T(\boldsymbol{y}_j)V_i$ , and the singular vectors of the response matrix  $\boldsymbol{P}$  are related to the random Green s function vectors as in (5.14). Hence, each entry is the mixed inner product (5.11) between  $\boldsymbol{g}_0(\boldsymbol{y}_i)$  and  $\boldsymbol{g}(\boldsymbol{y}_i)$ .

Without loss of generality, assume that the M scatterers correspond to the rst M columns in  $\mathcal{B}$ , i.e.  $n_j = j, j = 1$  M. We can write  $\mathcal{B}$  in the form of block matrices as  $\mathcal{B} = [\mathcal{B}_M \ M \ S]$ , where the columns in S correspond to grid points where there is no scatterers. Due to Proposition 5.4, the submatrix  $\mathcal{B}_M \ M$  is a diagonal matrix D perturbed by a matrix E with  $E_{1-1}$  1. The value of  $E_{1-1}$  depends on the minimal distance between any two scatterers, and is close to zero if they are well separated. Due to the incomplete phase cancellation in the mixed inner products, D is composed of complex values but, as long as the scatterers are far apart, D + E is diagonal dominated. Furthermore, because of the statistical stability of the mixed inner product (5.12), if  $D^{-1}S_{1-1} < 1$   $D^{-1}E_{1-1}$ , (4.6) gives the right solution even though there exists (correlated) noise in the data caused by the random media. This saves computational time by eliminating the need for di cult error bounds testing.

#### 5.3. MUSIC

In random media, the exact knowledge of the reference illumination g(y) is unknown, and therefore imaging using MUSIC amounts to approximate the unknow illumination with that in homogeneous context, i.e.

$$\mathcal{I}_{MUSIC}^{RM}(\boldsymbol{y}_k) = \frac{\min_{1 \ j \ K} \mathcal{P}\boldsymbol{g}_0(\boldsymbol{y}_j)}{\mathcal{P}\boldsymbol{g}_0(\boldsymbol{y}_k)} k = 1 \qquad K$$
(5.15)

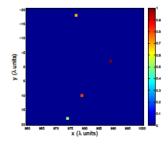


Figure 7. Original configuration of the scatterers in a  $41 \times 41$  image window with grid points separated by 1.

According to (5.14) and (5.11), the proxy  $\hat{g}_0(y)$  will focus around the scatterers when the imaging array is large enough. However, the image obtained by (5.15) will have lower resolution compared to the ideal case of (4.9), since the random phase in  $\hat{g}(y)$ cannot be fully cancelled out using the homogenous Green's function vector  $\hat{g}_0(y)$ .

#### 5.4. Numerical experiments

We now present numerical simulations in two dimensions to illustrate the performance of the MUSIC and hybrid- $\ell_1$  imaging methods in random media. We consider a random medium with correlation length  $l = 20\lambda$ , and standard deviation of the fluctuations  $\sigma = 0.1\%$ . For comparison purposes we also show images obtained by Kirchhoff migration with a single illumination sent from the central transducer of the array.

We consider a small and a large array, both consisting of 501 transducers uniformily distributed over the aperture. The small array has an aperture of 25l, and the large array an aperture of 100*l*. Four scatterers are placed within an IW of size  $41\lambda \times 41\lambda$ , which is at a distance L = 50l from the linear array, see Figure 7. We discretize the IW using a uniform grid with points separated by one wavelength  $\lambda$  (the spatial unit in all the figures is  $\lambda$ ). The amplitudes of the reflectivities of the scatterers,  $|\alpha_j|$ , are 0.8, 1.0, 0.5, and 0.7. The phases are set randomly in each realization. We compute the array data (5.13) using the Green's function given by (5.3). The line integral of the random field in (5.3) is approximated by a quadrature rule.

As a reference, we show in Figure 8 the images obtained in a homogeneous medium using KM (left column), MUSIC (center column), and hybrid- $\ell_1$  methods (right column) with noiseless data. The top and bottom rows show the results for the small array and the large array, respectively. As expected, the resolution of the KM images improves greatly for large arrays. On the other hand, in a homogeneous medium, MUSIC and hybrid- $\ell_1$  achieve an excellent resolution, even for small arrays. The images shown in Figure 8 do not change too much when the data is corrupted with up to 100% of additive noise [16].

The situation changes when there is correlated noise in the data because the signals propagate through a random medium with a complex structure. This is illustrated in Figure 9, where we show the images produced by these imaging methods using a small

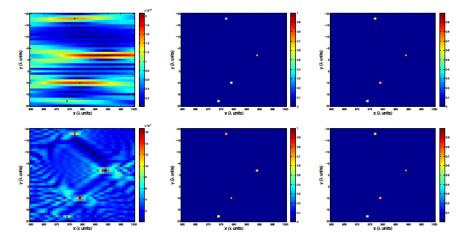


Figure 8. Top row: small array. Bottom row: large array. KM (left column), MUSIC (center column) and hybrid- $\ell_1$  (right column) images in homogeneous media. There is no additive noise in the data.

array (a = 25l). The three imaging methods show different behaviors though. Kirchhoff migration completely fails to image the scatterers, as can be seen in the top row the figure. There is not only degradation in the resolution, but also loss of stability. Observe that the images obtained with KM are significantly different from one realization to another.

The images obtained with MUSIC (middle row of Figure 9) are also blurred compared to those obtained in a homogeneous medium. Furthermore, the images change from one realization of the random medium to another and, therefore, MUSIC is also unstable if the array size is small. The hybrid- $\ell_1$  method also produces images that change from one realization to another (bottom row of Figure 9), but tries to keep a good resolution to provide a sparse solution. Observe that the detected scatterers dance along the cross-range direction around the true locations indicated in the figure with white dots. The images obtained with the hybrid- $\ell_1$  method also show some ghosts.

These problems are overcome when the array is large (a = 100l), see Figure 10. As expected, the resolution of the images obtained by KM and MUSIC improve a lot. However, KM still fails to image in random media as it produces clutter noise in the images from which it is hard or impossible to identify the location of the four scatterers. On the other hand, MUSIC and the hybrid- $\ell_1$  method are able to recover the sparse solution. We observe, though, that the performance of the hybrid- $\ell_1$  method is better. The locations of the four scatterers are found exactly by this method.

#### 6. Conclusions

We present a comprehensive study of optimization based methods applied to narrow band array imaging of localized scatterers. We have considered homogeneous and heterogeneous media. When the media is homogeneous but multiple scattering between

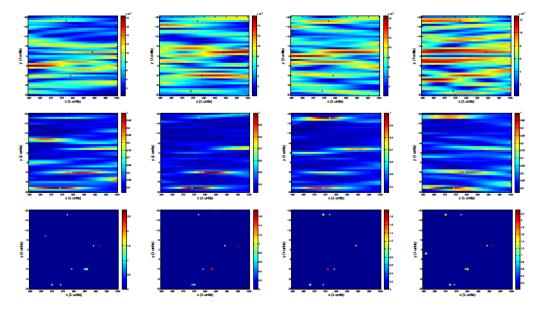


Figure 9. Small array. KM (top row), MUSIC (center row) and hybrid- $\ell_1$  (bottom row) images in the same four realizations of a random medium.

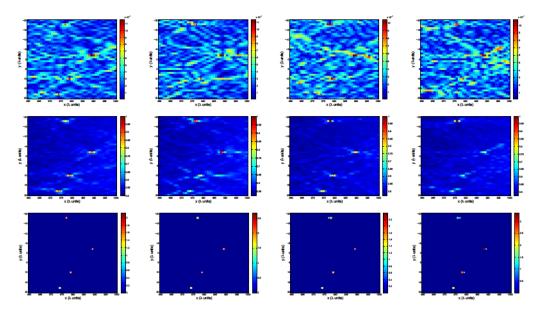


Figure 10. Large array. KM (top row), MUSIC (center row) and hybrid- $\ell_1$  (bottom row) images in the same four realizations of a random medium.

the scatterers is important, we give a non-iterative formulation of the nonlinear inverse problem that allows us to determine the locations and reflectivities of the scatterers non-iteratively using sparsity promoting optimization. We also propose to apply optimal illuminations to improve the robustness of the imaging methods and the resolution of the images. When multiple scattering is negligible, the optimization problem becomes linear. In this case, our formulation can be reduced to a hybrid- $\ell_1$  method that uses the optimal

illuminations and  $_1$  minimization. This method reduces the dimensionality of the problem, lters out the noise, and keeps all the essential properties of  $_1$  minimization.

When the media is random, we study the important concept of statistical stability which relates to the robustness of the imaging methods with respect to di erent realizations of the random media. Provided the imaging array is large enough, we show that the hybrid- $_1$  method gives very accurate results and is statistically stable.

We illustrate the theoretical results with various numerical examples and compared the performance of the proposed optimization based methods to the widely used Kirchho migration and the MUSIC methods.

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# Appendix

#### A. Foldy-Lax model

Multiple scattering between point-like scatterers is modeled by means of the Foldy-Lax equations [26, 27, 28]. Under this framework, the scattered wave received at transducer  $\boldsymbol{x}_r$  due to a narrow band signal of angular frequency — sent from  $\boldsymbol{x}_s$  can be written as the superposition of all scattered waves from the M scatterers at  $\boldsymbol{y}_{n_1} = \boldsymbol{y}_{n_M}$  (see Fig. 1 (c)), so

$$P(\boldsymbol{x}_r \ \boldsymbol{x}_s) = \int_{j=1}^{M} \int_{j=1}^{s} (\boldsymbol{x}_r; \boldsymbol{y}_{n_1} \quad \boldsymbol{y}_{n_M})$$
(A.1)

Here,  ${}^{s}_{j}(\boldsymbol{x}_{r}; \boldsymbol{y}_{n_{1}} \quad \boldsymbol{y}_{n_{M}})$  represents the scattered wave observed at  $\boldsymbol{x}_{r}$  due to the emaniting wave from the scatterer at position  $\boldsymbol{y}_{n_{j}}$ . It depends on the positions of all the scatterers and is given by

$${}^{s}_{j}(\boldsymbol{x}_{r};\boldsymbol{y}_{n_{1}}) = {}_{j}G(\boldsymbol{x}_{r} \; \boldsymbol{y}_{n_{j}}) \; {}^{e}_{j}(\boldsymbol{y}_{n_{1}} \; \boldsymbol{y}_{n_{M}})$$
(A.2)

where  ${}^{e}_{j}(\boldsymbol{y}_{n_{1}} \quad \boldsymbol{y}_{n_{M}})$  represents the exciting eld at the scatterer located at  $\boldsymbol{y}_{n_{j}}$ . Ignoring the self-interacting elds, the exciting eld at  $\boldsymbol{y}_{n_{j}}$  is equal to the sum of the incident eld  ${}^{inc}_{j} := {}^{inc}(\boldsymbol{y}_{n_{j}} \ \boldsymbol{x}_{s})$  at  $\boldsymbol{y}_{n_{j}}$  and the scattered elds at  $\boldsymbol{y}_{n_{j}}$  due to all scatterers except for the one at  $\boldsymbol{y}_{n_{j}}$ . Hence, it is given by

$${}_{j}^{e}(\boldsymbol{y}_{n_{1}} \quad \boldsymbol{y}_{n_{M}}) = {}^{inc}(\boldsymbol{y}_{n_{j}} \ \boldsymbol{x}_{s}) + {}_{k=j} {}_{k}G(\boldsymbol{y}_{n_{j}} \ \boldsymbol{y}_{n_{k}}) {}_{k}^{e}(\boldsymbol{y}_{n_{1}} \quad \boldsymbol{y}_{n_{M}})$$
(A.3)

for j = 1 2 M. This is a self-consistent system of M equations for the M unknown exciting elds

$$egin{array}{cccc} e_1 := & e_1(oldsymbol{y}_{n_1} & oldsymbol{y}_{n_M}) & & e_M := & e_M(oldsymbol{y}_{n_1} & oldsymbol{y}_{n_M}) \end{array}$$

We write (A.3) in matrix form as

$$\mathbf{Z}_M(\ ) \quad ^e = \quad ^{inc} \tag{A.4}$$

where  $e = \begin{bmatrix} e & e \\ 1 & M \end{bmatrix}^T$  and  $inc = \begin{bmatrix} inc & inc \\ 1 & M \end{bmatrix}^T$  are vectors whose components are the exciting and incident elds on the M scatterers, respectively, and

$$(Z_M(\ ))_{ij} = \begin{array}{cc} 1 & i = j \\ {}_{j}G(\boldsymbol{y}_{n_i} \ \boldsymbol{y}_{n_j}) & i = j \end{array}$$
(A.5)

is the M M Foldy-Lax matrix which depends on the re ectivities  $=\begin{bmatrix} 1 & M \end{bmatrix}^T$ , and on the positions of the scatterers  $\boldsymbol{y}_{n_j}$ . With the solution of (A.4), we use (A.2) and (A.1) to compute the scattered data received at the array.

We extend the M M matrix  $\mathbf{Z}_M()$ , de ned only for pairwise combinations of scatterers at  $\boldsymbol{y}_{n_j}$ , to a larger K K matrix

$$(Z(_{0}))_{ij} = \begin{bmatrix} 1 & i = j \\ & \\ & \\ & \\ & 0_{j}G(\boldsymbol{y}_{i} \ \boldsymbol{y}_{j}) & i = j \end{bmatrix}$$
(A.6)

which includes all pairwise combinations of the K grid points  $\boldsymbol{y}_j$  in the IW. Using (A.6) and (2.3), we can write the response matix as (2.4).

# B. Proof of results in Section 5

In order to prove Proposition 5.1, we need to show the following lemma rst.

**Lemma B.1.** Assume the autocorrelation function R of the random eld in  $\mathbb{L}_2(\mathbb{R}_+)$ with derivative R such that  $\frac{R(t)}{t} \quad \mathbb{L}(\mathbb{R}_+)$  and both R and R decay exponentially. Then

$$\frac{R(t)}{t}\,\mathrm{d}t < 0$$

**Proof.** Since  $R \quad \mathbb{L}_2(\mathbb{R}_+)$ , the Fourier transform exists, i.e.

$$R(\boldsymbol{k}) = R(\boldsymbol{x}) e^{i \boldsymbol{k} \cdot \boldsymbol{x}} d\boldsymbol{x}$$

Moreover, R() = 0. Using spherical coordinate (r = 0) we obtain

$$R(\mathbf{k}) = \begin{bmatrix} 2 & R(r)r^2 \sin e^{i\mathbf{k} r \cos} dr d d \\ 0 & 0 & 0 \end{bmatrix}$$
$$= 2 \qquad R(r)r^2 \sin e^{i\mathbf{k} r \cos} dr d d$$
$$= 4 \qquad \begin{bmatrix} 0 & R(r)r \frac{\sin \mathbf{k} r}{\mathbf{k}} dr \end{bmatrix}$$

Since the correlation function R() is defined only on [0 + ), we extend it symmetrically to the whole real line by R() = R() for any value in (0). Then, the above integral becomes

$$R(\mathbf{k}) = \frac{2}{\mathbf{k}} \qquad R(r)r\sin\mathbf{k} r\,\mathrm{d}r = \frac{2}{\mathbf{k}^3} \qquad R(r)(\sin\mathbf{k} r - \mathbf{k} r\cos\mathbf{k} r)\,\mathrm{d}r$$

where the last equality holds due to integration by parts and the exponential decay of R(). Then, with Riemann Lebesgue lemma, the following limit is

$$\lim_{K} \int_{0}^{K} \boldsymbol{k}^{3} R(\boldsymbol{k}) d\boldsymbol{k} = 2 \qquad \frac{R(r)}{r} dr$$

The symmetric extension of R() implies that R(r) = -R(-r) when r < 0. Therefore,

$$\int_{0} r^{3}R(r) \, \mathrm{d}r = \lim_{K} \int_{0}^{K} k^{3}R(k) \, \mathrm{d}k = 4 \int_{0}^{K} \frac{R(r)}{r} \, \mathrm{d}r$$

which implies that the integral on the right hand side is not positive.  $\blacksquare$ 

Note that the conditions on the autocorrelation function R in Lemma B.1 are not restrictive. Many autocorrelation functions satisfy those conditions, as for example,  $R(\boldsymbol{x}) = e^{-\frac{1}{2}\boldsymbol{x}^2}$  for a Gaussian random eld and the power law function  $R(\boldsymbol{x}) = (1 + \boldsymbol{x})e^{-\boldsymbol{x}}$ . With this lemma, we can now prove Proposition 5.1.

**Proof of Proposition 5.1.** To simplify the notation, we rst de ne the path integral of the random process function () as

$$(\boldsymbol{x} \ \boldsymbol{y}) = \begin{bmatrix} 1 & \mathbf{x} \\ 0 & \frac{1}{l} + s \frac{\boldsymbol{y} \ \boldsymbol{x}}{l} \end{bmatrix} ds$$

and

$$(\boldsymbol{x} \ \boldsymbol{y}) = \ \boldsymbol{x} \quad \boldsymbol{y} \ (\boldsymbol{x} \ \boldsymbol{y})$$

We need to compute moment estimation

$$\mathbb{E}(G(\boldsymbol{x} \ \boldsymbol{y}_1)\overline{G(\boldsymbol{x} \ \boldsymbol{y}_2)}) = G_0(\boldsymbol{x} \ \boldsymbol{y}_1)\overline{G_0(\boldsymbol{x} \ \boldsymbol{y}_2)}E$$

where

$$E = \mathbb{E} \quad e^{-i} \quad (\boldsymbol{x} \quad \boldsymbol{y}_1 \quad (\boldsymbol{x} \quad \boldsymbol{y}_1) \quad \boldsymbol{x} \quad \boldsymbol{y}_2 \quad (\boldsymbol{x} \quad \boldsymbol{y}_2)) = e^{-\frac{1}{2} \quad 2 \quad \mathbb{E} \left( -(\boldsymbol{x} \quad \boldsymbol{y}_1) \quad (\boldsymbol{x} \quad \boldsymbol{y}_2) \right)^2}$$
(B.1)

Because  $\boldsymbol{y}_1 \quad \boldsymbol{y}_2 \quad L$  we can write

and then

$$(\boldsymbol{x} \ \boldsymbol{y}_2) = (\boldsymbol{x} \ \boldsymbol{y}_1) + \begin{array}{cc} T \ \boldsymbol{y}_1 \end{array} (\boldsymbol{x} \ \boldsymbol{y}_1) (\boldsymbol{y}_2 \ \boldsymbol{y}_1) + o(\ \boldsymbol{y}_2 \ \boldsymbol{y}_1)$$

Dropping terms of order 2 or higher, we estimate the expectation in the exponent as  $\mathbb{E} \quad (\boldsymbol{x} \ \boldsymbol{y}_1) \qquad (\boldsymbol{x} \ \boldsymbol{y}_2)^2$   $\mathbb{E} \quad (\boldsymbol{x} \ \boldsymbol{y}_1) \frac{(\boldsymbol{x} \ \boldsymbol{y}_1)^T}{\boldsymbol{x} \ \boldsymbol{y}_1} (\boldsymbol{y}_2 \ \boldsymbol{y}_1) + \boldsymbol{x} \quad \boldsymbol{y}_1 \quad \overset{T}{\boldsymbol{y}}_1 \ (\boldsymbol{x} \ \boldsymbol{y}_1) (\boldsymbol{y}_2 \ \boldsymbol{y}_1)^2$ 

For simplicity, we consider  $\boldsymbol{y}_1$  and  $\boldsymbol{y}_2$  on the same plane, at a distance L from point  $\boldsymbol{x} = (0 \ 0 \ 0)$ , so  $\boldsymbol{y}_1 = (0 \ 0 \ L)$  and  $\boldsymbol{y}_2 = (L)$ . Then,  $(\boldsymbol{y}_2 \quad \boldsymbol{y}_1)^T (\boldsymbol{x} \quad \boldsymbol{y}_1) = 0$  and the

 $\mathbb E$ 

variance is reduced to the estimation of

 $\mathbf{2}$ 

$$\begin{array}{rcl}
\overset{T}{\boldsymbol{y}} & (\boldsymbol{x} \ \boldsymbol{y})(\boldsymbol{y} \quad \boldsymbol{y}) \\
&= & {}^{2}\mathbb{E} & -\frac{1}{y_{1}} & (\boldsymbol{x} \ \boldsymbol{y}) & {}^{2} + & {}^{2}\mathbb{E} & -\frac{1}{y_{2}} & (\boldsymbol{x} \ \boldsymbol{y}) & {}^{2} + & {}^{2}\mathbb{E} & -\frac{1}{y_{1}} & (\boldsymbol{x} \ \boldsymbol{y}) -\frac{1}{y_{2}} & (\boldsymbol{x} \ \boldsymbol{y}) \\
&= & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{1}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) + & {}^{2} -\frac{1}{y_{2}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
&+ & {}^{2} & -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
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& & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
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& & {}^{2} -\frac{1}{y_{1}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{2}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{2}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{2}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{2}} -\frac{1}{y_{2}}\mathbb{E}( & (\boldsymbol{x} \ \boldsymbol{y}) & (\boldsymbol{x} \ \boldsymbol{y})) \\
& & {}^{2} -\frac{1}{y_{2}} -\frac{1}{y_{$$

where  $\frac{y_i}{y_i}$  means the derivative with respect to the  $i^{\text{th}}$  component of vector  $\boldsymbol{y}$ . Denote  $y = \begin{bmatrix} \\ \end{bmatrix}^T$  and  $y = \begin{bmatrix} 0 \\ 0 \end{bmatrix}^T$ , such that  $\boldsymbol{y} = \begin{bmatrix} y^T \\ L \end{bmatrix}^T$  and  $\boldsymbol{y} = \begin{bmatrix} y^T \\ L \end{bmatrix}^T$ , and let  $= \frac{1}{l^2} sy \quad s \ y^{-2} + \frac{L^2}{l^2} (s - s)^2$ . We compute the derivatives of the expectations

$$\mathbb{E}((\boldsymbol{x} \boldsymbol{y}) (\boldsymbol{x} \boldsymbol{y})) = \begin{bmatrix} 1 & 1 \\ & 1 \\ & 0 \end{bmatrix} R() \, \mathrm{d}s \, \mathrm{d}s$$

in (B.2) as follows:

$$-\underbrace{y_1}{\mathbb{E}}((\boldsymbol{x} \ \boldsymbol{y}) \ (\boldsymbol{x} \ \boldsymbol{y})) = \begin{bmatrix} 1 & 1 \\ & 1 \end{bmatrix} R() \frac{s}{l^2}(sy_1 \ sy_1) \, \mathrm{d}s \, \mathrm{d}s$$

$$\frac{1}{y_1 \ y_1} \mathbb{E}(\ (\boldsymbol{x} \ \boldsymbol{y}) \ (\boldsymbol{x} \ \boldsymbol{y})) = \frac{1}{0} \frac{1}$$

Taking  $\boldsymbol{y} = \boldsymbol{y} = (0 \ 0 \ L)$  in the above second order derivatives, we obtain that

$$\mathbb{E} - \frac{y_1}{y_1} (\boldsymbol{x} \ \boldsymbol{y}) - \frac{y_2}{y_2} (\boldsymbol{x} \ \boldsymbol{y}) = 0$$

so that

$$\mathbb{E}$$
  $\stackrel{T}{\boldsymbol{y}}$   $(\boldsymbol{x} \ \boldsymbol{y})(\boldsymbol{y} \ \boldsymbol{y})^{2} = (2+2)\mathbb{E} - \frac{1}{y_{1}} (\boldsymbol{x} \ \boldsymbol{y})^{2}$ 

Next, we compute

$$\mathbb{E} \quad \frac{1}{y_1} \left( \boldsymbol{x} \; \boldsymbol{y} \right)^2 = \frac{1}{0} \frac{1}{R} \frac{L s \cdot s}{l} \frac{1}{R} \frac{ss}{lL s \cdot s} ds ds$$
$$= \frac{2l^2}{3L^4} \frac{L l}{0} sR(s) ds \quad \frac{2}{3lL} \frac{L l}{0} \frac{R(t)}{t} dt + \frac{2}{3L^2} R \frac{L}{l} \frac{R(0)}{L^2}$$
$$\frac{1}{L^2} \frac{1}{2} \frac{2L}{3l} \frac{R(t)}{0} \frac{R(t)}{t} dt$$

where the last approximation is based on the condition l = L and R() has exponential decay with normalization R(0) = 1. Now, let  $a_e^2 = {}^2L^4\mathbb{E} - \frac{1}{y_1} (\boldsymbol{x} \ \boldsymbol{y})^2$ . We then have

 $\mathbb{E}((\boldsymbol{x} \ \boldsymbol{y}_1) \ (\boldsymbol{x} \ \boldsymbol{y}_2))^2 \ \boldsymbol{x} \ \boldsymbol{y}_1^{\ 2} \mathbb{E}(\stackrel{T}{\boldsymbol{y}} \ (\boldsymbol{x} \ \boldsymbol{y}_1)(\boldsymbol{y}_2 \ \boldsymbol{y}_1))^2 \ \frac{\boldsymbol{y}_1^{\ 2} \boldsymbol{y}_2^{\ 2}}{{}^2 L^2} a_e^2$ (B.3) and using (B.1)

$$\mathbb{E}(G(\boldsymbol{x} \ \boldsymbol{y}_1)\overline{G(\boldsymbol{x} \ \boldsymbol{y}_2)}) = G_0(\boldsymbol{x} \ \boldsymbol{y}_1)\overline{G_0(\boldsymbol{x} \ \boldsymbol{y}_2)}E \quad G_0(\boldsymbol{x} \ \boldsymbol{y}_1)\overline{G_0(\boldsymbol{x} \ \boldsymbol{y}_2)}e^{-\frac{2a_e^2}{2L^2}}\boldsymbol{y}_1 \ \boldsymbol{y}_2^{\ 2}$$
  
i.e. the moment estimate (5.4).

Proof of Corollary 5.2. The result is a direct application of Proposition 5.1. Let  $\begin{aligned}
\mathcal{C} &= \frac{1}{(4 \ x \ y_1)^2(4 \ x \ y_2)^2}. \text{ Using (5.4), we have} \\
\mathbb{E} \ G(x \ y_1)\overline{G(x \ y_2)} & \mathbb{E} \ G(x \ y_1)\overline{G(x \ y_2)} & ^2 \\
\mathbb{E} \ G_0(x \ y_1)\overline{G_0(x \ y_2)} & e^i \ (\ (x \ y_1) \ (x \ y_2)) & e^{-\frac{1}{2} 2\frac{a_e^2}{L^2}} y_1 \ y_2 & ^2 \\
&= \mathcal{C}\mathbb{E}(1 + e^{-\frac{2a_e^2}{L^2}} y_1 \ y_2 & ^2 & e^{-\frac{2a_e^2}{2L^2}} y_1 \ y_2 & ^2(e^i \ (\ (x \ y_1) \ (x \ y_2)) + e^{-i} \ (\ (x \ y_1) \ (x \ y_2)))) \\
&= \mathcal{C}(1 + e^{-\frac{2a_e^2}{L^2}} y_1 \ y_2 & ^2 & e^{-\frac{2a_e^2}{2L^2}} y_1 \ y_2 & ^2(e^{-\frac{1}{2} 2 - 2\mathbb{E}(\ (x \ y_1) \ (x \ y_2))^2}) \\
&= \mathcal{C}(1 - e^{-\frac{2a_e^2}{L^2}} y_1 \ y_2 & ^2) & (B.4)
\end{aligned}$ 

where we use the estimate (B.3) in the last step.

**Proof of Proposition 5.3.** For simplicity, we use the same con guration as in the proof of Proposition 5.1, where  $\boldsymbol{y}_1 = \begin{bmatrix} 0 & 0 & L \end{bmatrix}^T$  and  $\boldsymbol{y}_2 = \begin{bmatrix} & L \end{bmatrix}^T$ . We rest look at the denominator of the ratio in (5.7)

$$\boldsymbol{g}(\boldsymbol{y}_1)^{2} = \sum_{j=1}^{N} G(\boldsymbol{x}_j \ \boldsymbol{y}_1)^{2} = \sum_{j=1}^{N} G_0(\boldsymbol{x}_j \ \boldsymbol{y}_1)^{2} = \boldsymbol{g}_0(\boldsymbol{y}_1)^{2}$$
 (B.5)

Using the same approach as in the proof in [16], under continuous limit, we have

$$g(y_1)^2 = \frac{1}{(4 \ h)^2} = \frac{1}{(x)} \frac{\mathrm{d}x}{x \ y^2} = \frac{1}{16 \ h^2} \log 1 + \frac{a^2}{4L^2}$$

On the other hand, the numerator can be computed as follows

$$\mathbb{E} \ \boldsymbol{g} \ (\boldsymbol{y}_{1})\boldsymbol{g}(\boldsymbol{y}_{2}) = \mathbb{E} \left( \boldsymbol{g} \ (\boldsymbol{y}_{1})\boldsymbol{g}(\boldsymbol{y}_{2}) \right)^{2} = \mathbb{E} \left( \begin{array}{c} N \\ I = \mathbb{E} \\ I = I \\ I = I$$

$$= \int_{j j=1}^{N} G_0(\boldsymbol{x}_j \ \boldsymbol{y}_1) \overline{G_0(\boldsymbol{x}_j \ \boldsymbol{y}_2)} G_0(\boldsymbol{x}_j \ \boldsymbol{y}_1)} \overline{G_0(\boldsymbol{x}_j \ \boldsymbol{y}_2)} G_0(\boldsymbol{x}_j \ \boldsymbol{y}_2)$$
  
e  $\frac{\frac{2a_e^2}{L^2}}{L^2} \boldsymbol{y}_1 \ \boldsymbol{y}_2^2$  e  $^{2-2\mathbb{E}(-(\boldsymbol{x}_j \ \boldsymbol{y}_1) - (\boldsymbol{x}_j \ \boldsymbol{y}_2))(-(\boldsymbol{x}_j \ \boldsymbol{y}_1) - (\boldsymbol{x}_j \ \boldsymbol{y}_2))(-(\boldsymbol{x}_j \ \boldsymbol{y}_1) - (\boldsymbol{x}_j \ \boldsymbol{y}_2))}$  1

From the above expression, we can see the variance of  $\boldsymbol{g}(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2)$  is close to the denominator up to the factor

$$e^{-\frac{2a_e^2}{L^2}}\boldsymbol{y}_1 \boldsymbol{y}_2^2 e^{-2\mathbb{E}(\boldsymbol{x}_j \boldsymbol{y}_1)} (\boldsymbol{x}_j \boldsymbol{y}_2))(\boldsymbol{x}_j \boldsymbol{y}_1) (\boldsymbol{x}_j \boldsymbol{y}_2)) = 1 \qquad (B.6)$$

The expectation

$$\mathbb{E}(\begin{array}{cccc} (\boldsymbol{x}_j \ \boldsymbol{y}_1) & (\boldsymbol{x}_j \ \boldsymbol{y}_2))( \begin{array}{cccc} (\boldsymbol{x}_j \ \boldsymbol{y}_1) & (\boldsymbol{x}_j \ \boldsymbol{y}_2)) \end{array}$$

is nonzero only when there is strong correlation between the path starting from  $\boldsymbol{x}_j$  and  $\boldsymbol{x}_j$ . This is controlled by the correlation length l of the random medium. When those paths are within l, the value of (B.6) is reduced to 1 e  $\frac{2a_e^2}{L^2} \boldsymbol{y}_1 \boldsymbol{y}_2^2$  and otherwise (B.6) is equal to zero. Thus in the continuous limit, the value of numerator can be approximated by

$$\mathbb{E} \boldsymbol{g} (\boldsymbol{y}_1) \boldsymbol{g}(\boldsymbol{y}_2) = \mathbb{E} \left( \boldsymbol{g} (\boldsymbol{y}_1) \boldsymbol{g}(\boldsymbol{y}_2) \right)^2 = \frac{1}{h^4} \left( 1 - e^{-\frac{2a_e^2}{L^2}} \boldsymbol{y}_1 - \boldsymbol{y}_2 \right)^2$$

$$G_0(\boldsymbol{x} | \boldsymbol{y}_1) \overline{G_0(\boldsymbol{x} | \boldsymbol{y}_2)} \, \mathrm{d} \boldsymbol{x} = \frac{G_0(\boldsymbol{x} | \boldsymbol{y}_1)}{B(\boldsymbol{0}|_l)} \left( \frac{G_0(\boldsymbol{x} | \boldsymbol{y}_2)}{G_0(\boldsymbol{x} | \boldsymbol{y}_2)} \, \mathrm{d} \boldsymbol{x} \right)$$
(B.7)

where  $B(\mathbf{0} \ l)$  is a ball centered at  $\mathbf{0}$  with radius l within which factor (B.6) is not equal to zero.

Therefore the ratio on the left handside of (5.7) is bounded by

When size of array *a* increases, we have  $\frac{\mathbb{E} g(y_1)g(y_2) \mathbb{E}(g(y_1)g(y_2))^2}{g(y_1)^2 g(y_2)^2}$  goes to 0 logarithmically.

**Remark B.2.** The result in Proposition 5.3 holds for any regime no matter a = L or  $L \leq a$ . When a = L, i.e. in the paraxial regime, we can use parabolic approximation to compute the approximate value for the ratio on the lefthand side of (5.7). Let  $\boldsymbol{x} = [x_1 \ x_2 \ 0]^T$ . Then we have

$$\boldsymbol{x} \quad \boldsymbol{y}_1 = \overline{x_1^2 + x_2^2 + L^2} \quad L + \frac{x_1^2 + x_2^2}{2L}$$

and

$$\boldsymbol{x} \quad \boldsymbol{y}_2 = \overline{L^2 + (x_1)^2 + (x_2)^2} = L + \frac{(x_1)^2 + (x_2)^2}{2L}$$

Since  $\boldsymbol{y}_1 \quad \boldsymbol{y}_2 \quad a \quad L$ , we approximate in denominator

$$oldsymbol{x}$$
  $oldsymbol{y}_1$   $oldsymbol{x}$   $oldsymbol{y}_2$   $L$ 

and for the two integrals in (B.7),

$$\begin{array}{c} (\boldsymbol{x}) \\ (\boldsymbol{x}) \\ (\boldsymbol{x}) \end{array} \overset{G_0(\boldsymbol{x} \ \boldsymbol{y}_2)}{=} \mathrm{d} \boldsymbol{x} & \frac{\mathrm{e}^{-\mathrm{i} \frac{2+2}{2L}}}{L^2} & \mathrm{e}^{\mathrm{i} \frac{x_1 + x_2}{L}} \mathrm{d} \boldsymbol{x} \\ &= \frac{\mathrm{e}^{-\mathrm{i} \frac{2+2}{2L}}}{L^2} & \overset{a \ 2}{=} \mathrm{e}^{\mathrm{i} \frac{x_1}{L}} \mathrm{d} x_1 & \overset{a \ 2}{=} \mathrm{e}^{\mathrm{i} \frac{x_2}{L}} \mathrm{d} x_2 \\ &= \frac{a^2 \mathrm{e}^{-\mathrm{i} \frac{2+2}{2L}}}{L^2} \mathrm{sinc} & \frac{a}{L} \mathrm{sinc} & \frac{a}{L} \end{array}$$

and similarly,

$${}_{B(\mathbf{0}\,l)}G_0(\mathbf{x}\ \mathbf{y}_1)\overline{G_0(\mathbf{x}\ \mathbf{y}_2)}\,\mathrm{d}\mathbf{x} \quad \frac{a^2\mathrm{e}^{-\mathrm{i}\ \frac{2}{2L}}}{L^2}\mathrm{sinc}\ \frac{l}{L}\ \mathrm{sinc}\ \frac{l}{L}$$

Note that when a = L, we can approximate linearly log  $1 + \frac{a^2}{4L^2} = \frac{a^2}{4L^2}$  Therefore, the lefthand side of (5.7) is approximately equal to

$$\frac{\mathbb{E} \ g \ (y_1)g(y_2) \quad \mathbb{E} \left(g \ (y_1)g(y_2)\right)^2}{g(y_1)^2 \ g(y_2)^2} \\ \frac{a^2l^2 \ 1 \ e^{-\frac{2a_e^2}{L^2}} y_1 \ y_2^2 \ \operatorname{sinc} \ \frac{a}{L} \ \operatorname{sinc} \ \frac{a}{L} \ \operatorname{sinc} \ \frac{l}{L} \ \operatorname{sinc} \ \frac{l}{L} \ \operatorname{sinc} \ \frac{l}{L} \\ \frac{\frac{L^4 \ ^2}{256} \ ^2 \ a^2}{4L^2} \\ 256 \ ^2 \ 1 \ e^{-\frac{2a_e^2}{L^2}} y_1 \ y_2^2 \ \frac{l}{a}^2 \ \operatorname{sinc} \ \frac{a}{L} \ \operatorname{sinc} \ \frac{a}{L} \ \operatorname{sinc} \ \frac{l}{L} \ \operatorname{sinc} \ \frac{l}{L}$$

This matches the result of logarithmic decay bound in general.

**Proof of Proposition 5.4.** We rst look at (5.10) for back-propagation in the true random media scenario. It has been shown in Proposition 5.3 that when size of array is large, the single realization will approach the average value. Therefore, it is enough to show the average value goes to zero as points  $y_1$  and  $y_2$  are far apart and the result is true due to the Chebyshev inequality under probability measure  $\mathbb{P}$  induced from random eld (). Using the moment formula, we have

$$\mathbb{E}(\boldsymbol{g} (\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2)) = \prod_{j=1}^{N} \overline{G_0(\boldsymbol{x}_j \ \boldsymbol{y}_1)} G_0(\boldsymbol{x}_j \ \boldsymbol{y}_2) e^{-\frac{2a_e^2}{2L^2}} \boldsymbol{y}_1 \ \boldsymbol{y}_2^2$$

Since the multiplier factor e  $\frac{2a_e^2}{2L^2} y_1 y_2^2 < 1$  and goes to zero as  $y_1 y_2$  increases, we thus have

$$\mathbb{E}(\boldsymbol{g} \ (\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2)) = 0 \qquad \underline{\boldsymbol{y}_1 \quad \boldsymbol{y}_2}$$

especially when size of array a becomes large. When  $a_e$  is large in random media, average value will decay to zero faster than that in homogeneous media. This is why in random media better resolution can be achieved.

Next, we look at (5.11). First, we compute the expectation of  ${}^{2}(x \ y)$ 

$$\mathbb{E}^{-2}(\boldsymbol{x} \ \boldsymbol{y}) = 2 \int_{0}^{1} \int_{0}^{s} R \ \frac{\boldsymbol{y} \ \boldsymbol{x}}{l}(s \ s) \ \mathrm{d}s \ \mathrm{d}s = \frac{2l}{\boldsymbol{y} \ \boldsymbol{x}} \int_{0}^{1} \int_{0}^{1} R(s) \ \mathrm{d}s \ \mathrm{d}s$$

Under l = L and symmetric extension of R() to negative real line, we have

$$\mathbb{E}^{-2}(\boldsymbol{x} \ \boldsymbol{y}) \quad \frac{2l}{\boldsymbol{y} \ \boldsymbol{x}} \quad {}_{0} \quad R(s) \, \mathrm{d}s = \frac{lR(0)}{\boldsymbol{y} \ \boldsymbol{x}}$$

Then it is easy to calculate the expectation of mixed inner product as

$$\mathbb{E}(\boldsymbol{g}_{0}(\boldsymbol{y}_{1})\boldsymbol{g}(\boldsymbol{y}_{2})) = \prod_{\substack{j=1\\j=1\\N}}^{N} \overline{G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{1})} G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) \mathbb{E}e^{i \boldsymbol{x}_{j} \ \boldsymbol{y}_{2} \ (\boldsymbol{x}_{j} \ \boldsymbol{y}_{2})}$$
$$= \prod_{\substack{j=1\\N\\G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{1})}^{N} \overline{G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2})} e^{-\frac{1}{2} \cdot 2 \cdot \frac{2}{0} \boldsymbol{x}_{j} \ \boldsymbol{y}_{2} \cdot 2\mathbb{E} \cdot 2(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2})}$$
$$= \prod_{\substack{j=1\\N\\G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{1})}^{N} \overline{G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2})} e^{-\frac{1}{2} \cdot 2 \cdot \frac{2}{0} lR(0) \ \boldsymbol{x}_{j} \ \boldsymbol{y}_{2}}$$

Because R(0) = 0, it implies  $\mathbb{E}(\boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2))$  is approximately equal to  $\boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}_0(\boldsymbol{y}_2)$  multiplied by a factor less that 1 and therefore

 $\mathbb{E}(\boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2)) = 0 \quad \underline{\boldsymbol{y}_1 \quad \boldsymbol{y}_2}$ 

To show the statistical stability of the mixed inner product, we rst compute the bound of the numerator similar to that in the proof of Proposition 5.4.

$$\mathbb{E} \boldsymbol{g}_{0}(\boldsymbol{y}_{1})\boldsymbol{g}(\boldsymbol{y}_{2}) \quad \mathbb{E} \boldsymbol{g}_{0}(\boldsymbol{y}_{1})\boldsymbol{g}(\boldsymbol{y}_{2})^{-}$$

$$\mathbb{E} \int_{j=1}^{N} \overline{G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{1})} G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) \quad e^{i} \quad (\boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) \quad e^{-\frac{1}{2} \ 2 \ 2 lR(0)} \ \boldsymbol{x}_{j} \ \boldsymbol{y}_{2} \qquad 2^{2}$$

$$= \int_{j=1}^{N} \overline{G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{1})} G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{1}) \overline{G_{0}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2})}$$

$$= e^{-\frac{1}{2} \ 2 \ 2 lR(0)} (\boldsymbol{x}_{j} \ \boldsymbol{y}_{2} + \boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) \quad e^{-2 \ 2 \mathbb{E}} \ (\boldsymbol{x} \ \boldsymbol{y}_{2}) \ (\boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) \quad 1$$

The correlation term  $(\boldsymbol{x}_j \ \boldsymbol{y}_2) \ (\boldsymbol{x}_j \ \boldsymbol{y}_2)$  is nonzero only when the paths connecting  $\boldsymbol{x}_j$ ,  $\boldsymbol{x}_j$  with  $\boldsymbol{y}_2$  are within the correlation length l. Also according to Cauchy-Schwartz inequality,

$$\mathbb{E}(\begin{array}{ccc} (\boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) \ (\boldsymbol{x}_{j} \ \boldsymbol{y}_{2})) & \frac{1}{2} \end{array} \mathbb{E}^{-2}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2}) + \mathbb{E}^{-2}(\boldsymbol{x}_{j} \ \boldsymbol{y}_{2})$$

Therefore the numerator is approximately bounded by continuous limit

$$\begin{split} \mathbb{E} \ \boldsymbol{g}_{0}(\boldsymbol{y}_{1})\boldsymbol{g}(\boldsymbol{y}_{2}) & \mathbb{E} \boldsymbol{g}_{0}(\boldsymbol{y}_{1})\boldsymbol{g}(\boldsymbol{y}_{2})^{2} \\ & \lesssim \frac{1}{h^{4}} \ 1 \quad \mathrm{e}^{-\frac{2-2}{LlR(0)}} (\boldsymbol{x}) \overline{G_{0}(\boldsymbol{x} \ \boldsymbol{y}_{1})} \overline{G_{0}(\boldsymbol{x} \ \boldsymbol{y}_{2})} \, \mathrm{d} \boldsymbol{x} \\ & \lesssim \frac{1}{256^{-2}h^{4}} \ 1 \quad \mathrm{e}^{-\frac{2-2}{LlR(0)}} \log \ 1 + \frac{a^{2}}{4L^{2}} \ \log \ 1 + \frac{l^{2}}{4L^{2}} \end{split}$$

The denominator is norm of the Green function vector which has been calculated in the proof of Proposition 5.3. Thus the ratio in (5.12) is bounded by

$$\frac{\mathbb{E} \ \boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2) - \mathbb{E}(\boldsymbol{g}_0(\boldsymbol{y}_1)\boldsymbol{g}(\boldsymbol{y}_2))^2}{\boldsymbol{g}_0(\boldsymbol{y}_1)^{-2} \ \boldsymbol{g}(\boldsymbol{y}_2)^{-2}} \lesssim 1 \quad \mathrm{e}^{-\frac{2-2}{2}LlR(0)} \ \frac{\log -1 + \frac{l^2}{4L^2}}{\log -1 + \frac{a^2}{4L^2}}$$

When a, the right handside goes to zero. The decay rate is again controlled by the logrithmic factor of a and when a L, the decay rate is quadratic of a which is the same as that in Proposition 5.3. Due to the statistical stability, the mixed inner product satis es (5.11).

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