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A High-Torque Magnetless Axial-Flux Doubly Salient Machine for In-Wheel Direct Drive Applications

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In this paper, the concept of dc-field excitation is newly incorporated into the axial-flux doubly salient (AF-DS) machine, leading to create the new magnetless AF-DSDC machine. With the external field excitation, the proposed machine not only can enjoy its improved torque density, but also the flux-weakening capability for the wide-speed range operation. With these characteristics, the AF-DSDC is favorable for the in-wheel direct drive application. In particular, the proposed machine is designed and compared based on the requirement of a typical passenger electric vehicle. To have a better illustration, a radial-flux DSDC machine is also designed under the fair environment. The corresponding performances of both machines are analyzed by using the 2-D and 3-D finite element method.

Index Terms—Axial-flux, dc-field excitation, doubly salient, in-wheel direct drive, magnetless machine.

I. INTRODUCTION

WITH ever increasing concerns on energy efficiency and hence environmental protection, the development of the electric machines has been speeding up [1]. In general, the electric machines have to offer high efficiency, high power density, high controllability, wide-speed range, and maintenance-free operation [2], [3]. To achieve these goals, the permanent magnet (PM) machines have been actively developed [4]–[6]. However, in recent years, the supply of the PM materials is limited and fluctuating, leading to the soared material cost [7]. Thus, the advanced magnetless radial-flux doubly salient (RF-DS) machines are becoming attractive [8]. Compared with the PM machines, the magnetless RF-DS machines take the definite merit of low material cost, but suffer from relatively low torque density.

The axial-flux (AF) DS machine, which employs the radial length as active part for torque production, can improve the torque density as compared with its RF counterpart [9]–[11], and it is particular favorable for the application of the in-wheel motor drive [12]. However, similar as the RF-DS machine, the AF-DS machine uses only half of its torque producing zone, resulting with degraded torque performances. Meanwhile, the concept of dc-field excitation, which allows the RF-DS to utilize the remaining torque zones, was proposed [13], [14], but the possibility for implementing the dc winding into the AF-DS machine has rarely addressed yet.

The purpose of this paper is to newly incorporate the dc winding into the AF-DS machine, hence creating the new AF-DSDC machine, purposely for the application of in-wheel direct drive. To achieve a better illustration purpose, an RF-DSDC machine is also designed. The design criteria and operating principles of the two magnetless machines will be discussed. Their performances will be analyzed by using

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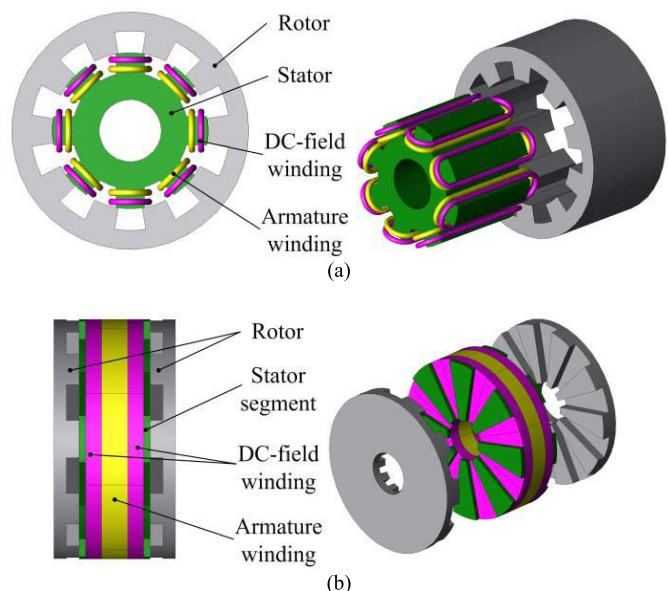


Fig. 1. Proposed machines. (a) RF-DSDC. (b) AF-DSDC.

the 2-D and 3-D finite element methods (FEMs), and then quantitatively compared with the requirements of the typical passenger electric vehicle (EV) applications.

II. PROPOSED MACHINES

Fig. 1 shows the topologies of the two magnetless machines, namely the 8/10-pole RF-DSDC machine, and the 8/10-pole AF-DSDC machine. The 4-phase topology is chosen because of the considerations of the stability and cost-effectiveness [11]. The RF-DSDC machine adopts the inner-stator outer-rotor structure, while the AF-DSDC adopts the sandwiched-stator sided-rotor structure. With suitable dimensions, both machines can mount its rotors to the tires directly. For instance, one of the most common wheel sizes is 195/65 R15, denoting the axial length is 195 mm and the rim diameter is 15 in, i.e., 381 mm. Hence, the two machines are purposely designed with

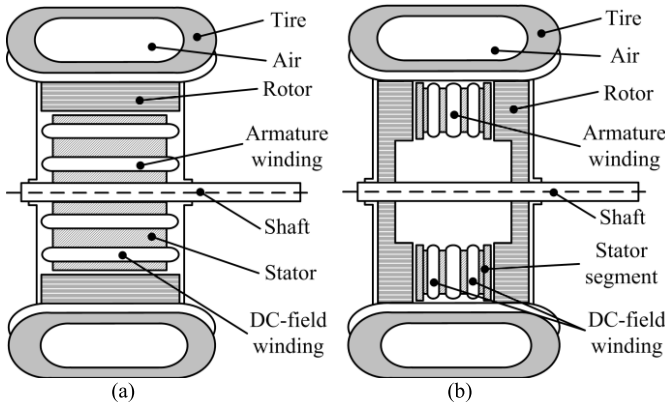


Fig. 2. In-wheel direct drive structures. (a) RF-DSDC. (b) AF-DSDC.

TABLE I

TARGET MOTOR SPECIFICATIONS FOR TYPICAL EV APPLICATION

Item	Value
Peak DC voltage	360 V
Rated power	3.1 kW
Rated torque	100 Nm
Constant-torque operation	0-300 rpm
Constant-power operation	300-900 rpm
Peak power	140% for 60 s
Wheel	195/65 R15

this dimension such that both of them can be implemented into in-wheel motor drive system instinctively, as shown in Fig. 2. Meanwhile, based on the requirements of a typical passenger EV [2], the targeted specifications of the machines are listed in Table I.

Since the proposed AF-DSDC adopts the modular stator construction method [10], the manufacturing complexity becomes its major limitation. Upon this method, the individual stator segments are constructed independently. Then, all the stator segments are consolidated together to form the yokeless stator. With the reduction of the stator yoke, the proposed AF-DSDC machine can further improve its torque density and minimize its material cost. Furthermore, the yokeless stator structure enables more spacing for the winding accommodations.

Since the AF-DSDC machine is derived from the RF-DSDC machine, its design equations such as the pole arrangement [14] can be extended from that of the profound RF-DSDC machine. Thus, the pole arrangements for both the RF-DSDC and AF-DSDC are governed by

$$\begin{cases} N_s = 2mk \\ N_r = N_s \pm 2k \end{cases} \quad (1)$$

where N_s is the number of stator poles, N_r is the rotor poles, m is the armature phases, and k is any integer. By selecting $m = 4$, and $k = 1$, this ends up with $N_s = 8$, $N_r = 10$ and comes up with the proposed structure of the two machines.

To have a fair comparison, the machine dimensions, namely the radial outside diameters, radial inside diameters, axial stack lengths, and airgap lengths are set equal. In addition, the pole arcs, pole heights, slot-fill factors, and current densities are

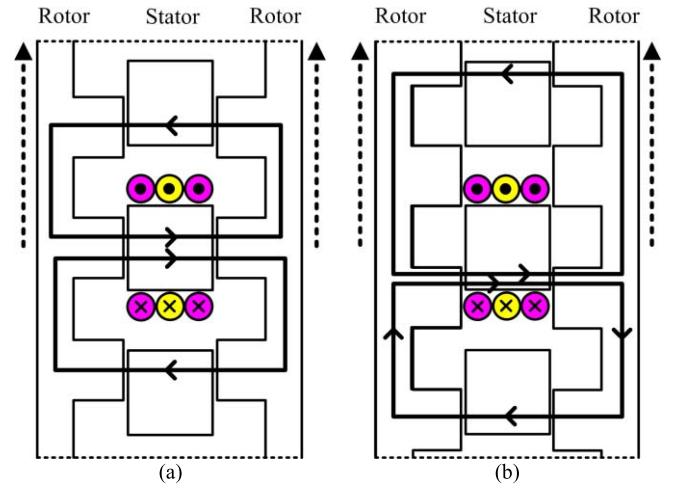


Fig. 3. Flux pattern of AF-DSDC. (a) Position 1. (b) Position 2.

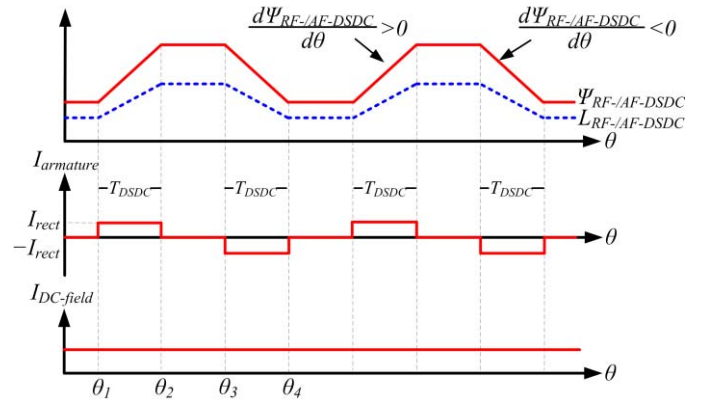


Fig. 4. Theoretical operating waveform.

designed in such a way that the magnetic saturation and hence the core losses can be minimized.

III. OPERATING PRINCIPLE

The proposed AF-DSDC machine adopts the concentrated winding arrangement on the sandwiched-stator in a way that the dc flux-linkages flow along the two sided-rotors via the stator segment as shown in Fig. 3. In addition, it is expected that the flux-linkages vary according to the relative stator-rotor position. Hence, under this winding arrangement, both sided-rotors can be operated in the same sense with the single set of 4-phase armature winding.

With the dc-field excitations, both machines can operate with the bipolar rectangular current operation. When the flux-linkage is increasing, a positive rectangular current I_{rect} is applied to produce a positive torque. Meanwhile, a negative rectangular current $-I_{rect}$ is applied when the flux-linkage is decreasing and the torque produced is also positive. The operating waveforms of the machines are shown in Fig. 4. Under this operation, all the torque producing zones are utilized and the torque ripple can be suppressed. Each phase performs 90° conduction with $\theta_2 - \theta_1 = \theta_4 - \theta_3 = 90^\circ$ and

TABLE II
MACHINE KEY DATA

	RF-DSDC	AF-DSDC
Radial outside diameter	381.0 mm	381.0 mm
Radial inside diameter	100.0 mm	100.0 mm
Airgap length	0.5 mm	0.5 mm
Axial stack length	195.0 mm	195.0 mm
Rotor pole number	10	10
Stator pole number	8	8
No. of phases	4	4
Slot-fill factor	60 %	60 %
No. of armature turns	60	120

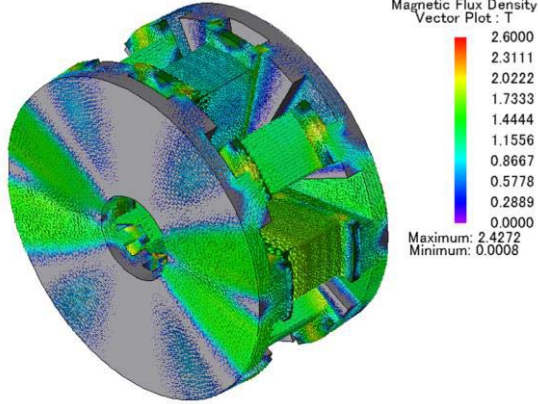


Fig. 5. No-load magnetic field distribution.

the developed torques of these two machines are shown as

$$T_{\text{RF-DSDC}} = \frac{1}{2\pi} \int_0^{2\pi} \left(I_{\text{rect}} \frac{d\psi_{\text{RF-DSDC}}}{d\theta} + \frac{1}{2} I_{\text{rect}}^2 \frac{dL_{\text{RF-DSDC}}}{d\theta} \right) d\theta \quad (2)$$

$$T_{\text{AF-DSDC}} = \frac{1}{2\pi} \int_0^{2\pi} \left(I_{\text{rect}} \frac{d\psi_{\text{AF-DSDC}}}{d\theta} + \frac{1}{2} I_{\text{rect}}^2 \frac{dL_{\text{AF-DSDC}}}{d\theta} \right) d\theta \quad (3)$$

where $\Psi_{\text{RF-DSDC}}$ and $\Psi_{\text{AF-DSDC}}$ are the flux-linkages of the RF-DSDC and AF-DSDC machines, respectively, and $L_{\text{RF-DSDC}}$ and $L_{\text{AF-DSDC}}$ are the self-inductances of the corresponding armature windings. As illustrated in (2) and (3), the developed torques of both machines are composed of two torque components: 1) the dc-field torque component and 2) the reluctance torque component. Meanwhile, it should be noted that the developed torques are mainly contributed by dc-field torque components, whereas the reluctance torque components resulted as an averaged zero value [14].

IV. PERFORMANCE ANALYSIS

All the two magnetless machines are analyzed under the fair environment and the key design data of the machines are listed in Table II. By using the 2-D and 3-D FEM, all machine performances can be simulated and then quantitatively compared. The no-load magnetic field distribution of the AF-DSDC is shown in Fig. 5.

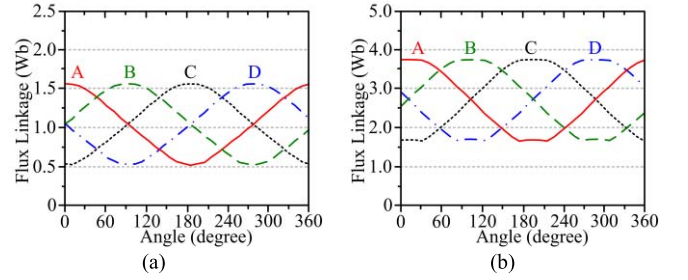


Fig. 6. Flux linkages. (a) RF-DSDC. (b) AF-DSDC.

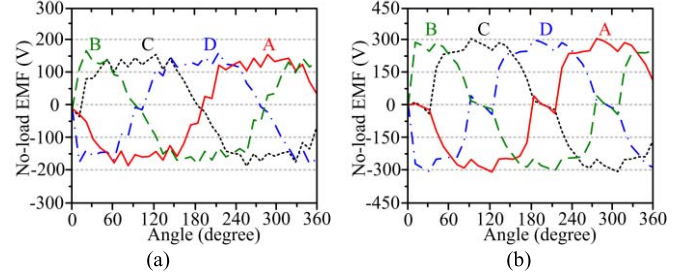


Fig. 7. No-load EMFs at dc-field of 5 A/mm². (a) RF-DSDC. (b) AF-DSDC.

First of all, the flux-linkages of RF-DSDC and AF-DSDC are shown in Fig. 6. It can be seen that both flux-linkages are well balanced among four phases without noticeable distortion. These confirm that the design data of these two machines are correct. In addition, the results confirm that the proposed winding arrangement enables the AF-DSDC to achieve the same flux-linkage pattern as the conventional one.

Second, the no-load electromotive forces (EMFs) of the RF-DSDC and AF-DSDC under the dc-field excitation of 5 A/mm² at base speed 300 r/min are shown in Fig. 7. The results show that both machines obtain the trapezoidal patterns, which are suitable for the rectangular current conduction schemes. Meanwhile, the no-load EMFs produced by AF-DSDC are larger than those by RF-DSDC are. Hence, it is expected that the power level of the AF-DSDC should be higher.

In the third place, the airgap flux density distributions of both machines are calculated as shown in Fig. 8. These show that both machines obtain the same airgap flux pattern. For the AF-DSDC, because the flux flows along the two sided-rotors via the sandwiched-stator, the flux patterns of the two airgaps are symmetrical, but in opposite directions.

Fourth, the output torque waveforms of the RF-DSDC and AF-DSDC machines are simulated as shown in Fig. 9. It can be observed that the rated average torques of the RF-DSDC, of the AF-DSDC machines are 80.6 and 161.8 Nm, respectively. Compared with the RF-DSDC machine, the torque enhancement of the AF-DSDC is 100.7%. It can also be observed that the torque ripples of the RF-DSDC and AF-DSDC machines are 33.5% and 19.7%, respectively, which are comparable with the conventional RF machines [3]. The high torque density and low torque ripple value offered by the AF-DSDC machine are particularly favorable for the

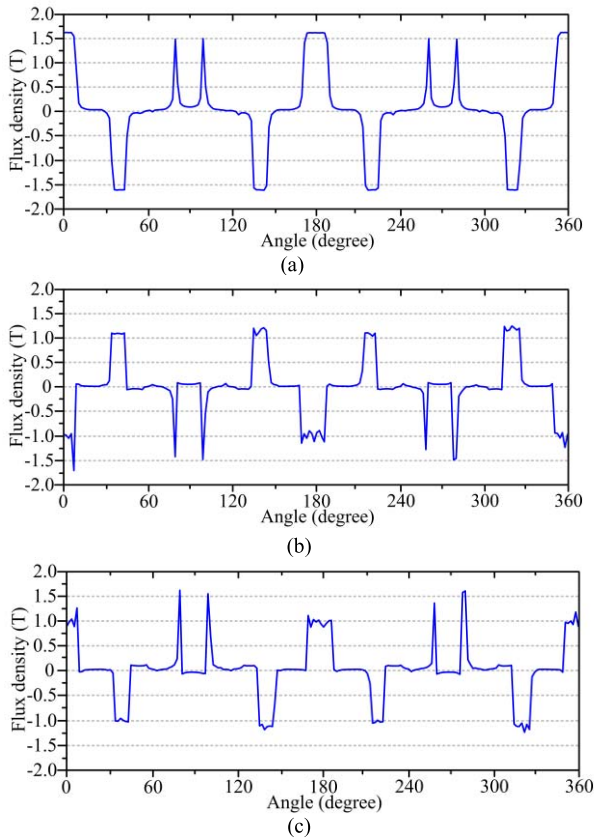


Fig. 8. Airgap flux densities. (a) RF-DSDC. (b) Left airgap of AF-DSDC. (c) Right airgap of AF-DSDC.

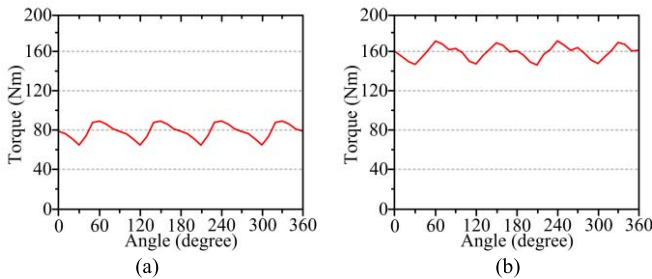


Fig. 9. Rated torque waveforms. (a) RF-DSDC. (b) AF-DSDC.

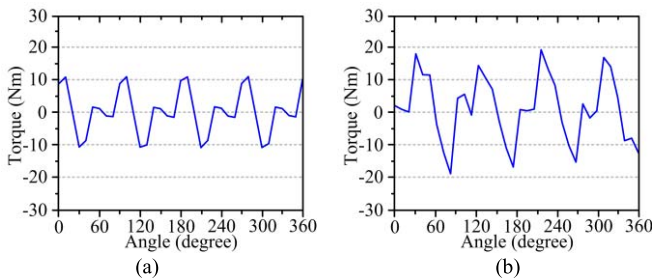


Fig. 10. Cogging torque waveforms. (a) RF-DSDC. (b) AF-DSDC.

in-wheel direct drive applications [12]. Meanwhile, the cogging torques of the RF-DSDC and AF-DSDC machines are shown in Fig. 10. It can be shown that the peak amplitudes of the cogging torques of the RF-DSDC and AF-DSDC are 10.7 and 19.6 Nm, respectively. These values are still acceptable,

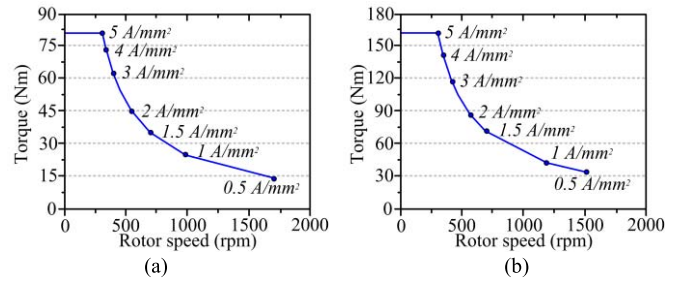


Fig. 11. Torque-speed characteristics. (a) RF-DSDC. (b) AF-DSDC.

TABLE III
MACHINE PERFORMANCE COMPARISON

	RF-DSDC	AF-DSDC
Power	2500 W	5050 W
Base speed	300 rpm	300 rpm
Field excitation	5 A/mm ²	5 A/mm ²
No-load EMF	150 V	305 V
Rated torque	80.6 Nm	161.8 Nm
Torque enhancement	N/A	100.7 %
Torque density	4920 Nm/m ³	9850 Nm/m ³
Cogging torque	10.7 Nm	19.6 Nm
Torque ripple	33.5 %	19.7 %

which are only 13.3% and 12.1% of their rated torques. By using pole optimization, the cogging torques can be further minimized; leading to further improve the torque ripples [2]. Such optimization will be the substance of our future papers.

Finally, the flux-weakening performances of the RF-DSDC and AF-DSDC are simulated and shown in Fig. 11. The dc-field excitations of both machines can be independently controlled easily. Hence, both machines can obtain high flux-weakening capabilities to achieve wide-speed range operations. With the 0.5 A/mm² field excitations, the operating speeds of the RF-DSDC and AF-DSDC are 1691 and 1517 r/min, respectively. These confirm the proposed machines can fulfill the requirements of the normal EV applications.

V. CONCLUSION

In this paper, two magnetless DS machines, namely the RF-DSDC and AF-DSDC have been analyzed and quantitatively compared, while the comparisons are summarized in Table III. By applying the radial area as active part for torque production, the AF-DSDC exhibits better torque performances than its RF counterpart does. To be specific, the AF-DSDC can produce satisfactory power and torque densities compared with the PM machines do [3]. Compared with the PM machines, the proposed magnetless AF-DSDC machine may result in lower efficiency, power factor, and torque density [14]. Meanwhile, the manufacturing cost of the PM machines is dominated by the PM material cost so that the proposed magnetless machine can offer better cost-effectiveness than the PM machines do [7], [11]. Meanwhile, with the independent dc-field windings, the proposed machines both enjoy the flux-weakening capability for the wide-speed range operations. Based on the analysis and comparisons, the proposed AF-DSDC can fulfill all the requirements for the

in-wheel direct drive application, while the RF-DSDC fails in some tasks from the expectations.

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