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# A Gradient Flow Approach to Robust Pole Assignment in Second-Order Systems

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# Abstract

Robust pole assignment for second-order systems is considered. It is shown that assignment can be achieved by solving a linear matrix equation or linear system. An objective function measuring the robustness of the closed-loop spectrum is minimized via gradient flow.

#### 1 Introduction

Consider a time-invariant, second-order system

$$M\ddot{q} + D\dot{q} + Kq = f \tag{1}$$

where  $q, f \in \mathbb{R}^n$  and  $M, D, K \in \mathbb{R}^{n \times n}$ . Responses of (1) can be altered by applying a feedback control force f = Bu to (1) with  $B \in \mathbb{R}^{n \times m}$  denoting the input matrix and  $u \in \mathbb{R}^m$  the control vector. The state feedback control law is defined as  $u = -F_Kq - F_D\dot{q}$  where  $F_D$  and  $F_K$  are derivative and proportional feedback matrices respectively. Hence, the resulting closed-loop system is

$$M\ddot{q} + (D + BF_D)\dot{q} + (K + BF_K)q = 0.$$
 (2)

In this note, M is assumed to be invertible and (2) can be written in the familiar first-order form

$$\dot{z} = \bar{A}z + \bar{B}u$$

where

$$\bar{A} = \begin{bmatrix} 0 & I \\ -M^{-1}K & -M^{-1}D \end{bmatrix}, \quad \bar{B} = \begin{bmatrix} 0 \\ M^{-1}B \end{bmatrix}.$$

For complete assignability, the pair  $(\overline{A}, \overline{B})$  is assumed to be completely controllable.

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### **2** Pole Assignment

Let  $\Lambda$  be a real pseudo-diagonal matrix containing all the assigned poles, in which  $1 \times 1$  blocks represent real poles and  $2 \times 2$  blocks of the form

$$\begin{bmatrix} \sigma & \omega \\ -\omega & \sigma \end{bmatrix}$$
(3)

represent complex conjugate pairs  $\sigma \pm i\omega$ . Note that (3) is unitary equivalent to a diagonal matrix, that is

$$U^* \left[ \begin{array}{cc} \sigma & \omega \\ -\omega & \sigma \end{array} \right] U = \left[ \begin{array}{cc} \sigma + i\omega & 0 \\ 0 & \sigma - i\omega \end{array} \right]$$

where  $U = \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1 \\ i & -i \end{bmatrix}$  is a unitary matrix.

The idea of state feedback pole assignment is to find  $[F_K \ F_D]$  such that

$$\bar{A} + \bar{B}[F_K \ F_D] = T\Lambda T^{-1} \text{ or } \bar{A} + \bar{B}G = T\Lambda$$
 (4)

where  $G = [F_K \ F_D]T$  and T is some nonsingular matrix. In fact, (4) corresponds to the following linear matrix equation

$$\tilde{A}T\tilde{B} - \tilde{C}T\tilde{D} = \tilde{E}$$
(5)

where

$$\tilde{A} = \begin{bmatrix} 0 & I \\ -K & -D \end{bmatrix}, \quad \tilde{B} = I, \quad \tilde{C} = \begin{bmatrix} I & 0 \\ 0 & M \end{bmatrix},$$
$$\tilde{D} = \Lambda, \quad \tilde{E} = \begin{bmatrix} 0 \\ B \end{bmatrix} \begin{bmatrix} F_K & F_D \end{bmatrix} T = \begin{bmatrix} 0 \\ B \end{bmatrix} G.$$

Since the pencils  $\tilde{A} - \lambda \tilde{B}$  and  $\tilde{D} - \lambda \tilde{B}$  are regular and the spectra  $\rho\left(\tilde{A}, \tilde{C}\right)$  and  $\rho\left(\tilde{D}, \tilde{B}\right)$  are assumed to be disjoint, i.e. no common closed-loop and open-loop poles, (5) is uniquely solvable for a given G [1, 3]. However, by exploiting the special structures

of the coefficient matrices  $\tilde{A}, \tilde{B}, \tilde{C}, \tilde{D}, \tilde{E}$ , we can express (5) into linear systems of equations which can then be solved by some effective algorithms [4]. Define  $\hat{U} = \text{diag}\left\{\hat{U}_1, \ldots, \hat{U}_{n+\frac{p}{2}}\right\}$  where p, an even number, is the number of real poles of (2) and

$$\begin{cases} \hat{U}_j = 1, \quad j = 1, \dots, p, \\ \hat{U}_j = U, \quad j = p+1, \dots, n+\frac{p}{2}. \end{cases}$$

Note that  $\hat{U} = U \bigotimes I_n$  if p = 0 and  $\hat{U} = I_{2n}$  if p = 2n. By comparing both sides of (5), we eventually obtain the linear system

$$\left(\lambda_j^2 M + \lambda_j D + K\right) w_j = -B\hat{g}_j, \quad j = 1, \dots, 2n.$$
 (6)

where  $\hat{g}$  is the *j*-th column of  $\hat{G} = G\hat{U}$ . Then for a given G, we can solve for  $W = [w_1, \ldots, w_{2n}]$  from (6) and  $T = \begin{bmatrix} W\hat{U}^* \\ W\hat{U}^*\Lambda \end{bmatrix}$ . Consequently,

$$[F_K \quad F_D] = GT^{-1} \tag{7}$$

which is the augmented derivative and proportional feedback matrix such that the closed-loop system (2) have the assigned spectrum.

#### **3** Robust Pole Assignment

In measuring the robustness of the closed-loop spectrum, an objective function is defined as

$$\phi(T(G)) = ||T(G)||_F^2 + ||T(G)^{-1}||_F^2.$$
(8)

It can be shown that [5] the objective function (8) has to be minimized in order to obtain a robust closed-loop system. For notational simplicity, we write  $\phi(T(G))$  as  $\phi(G)$  throughout.

For a given G, T is solved via the linear matrix equation (5) or linear system (6) at each minimization iteration for  $\phi(G)$ . When an optimum solution, say  $G^*$ , is reached such that  $\phi(G^*)$  attains its minimum, the required augmented feedback matrix  $F = [F_K \ F_D] =$  $G^*T^{*-1}$  is recovered from (7), where  $G^*$  and  $T^*$  satisfy (4).

With gradient flow analysis, this minimization problem comes down to solving the following system of ordinary differential equations

$$\dot{G}(t) = 2 \left[ \operatorname{trace} \left\{ \frac{\partial T}{\partial g_{jk}} \left( T^{-1} T^{-T} T^{-1} - T^T \right) \right\} \right]_{m \times 2n}$$
  

$$G(0) = G_0 \in \mathcal{G}.$$
(9)

where  $\mathcal{G} = \{G \mid T \text{ is a nonsingular solution of (4)}\}$  is open and dense in  $\mathbb{R}^{m \times 2n}$ . The solution, say  $G^*$ , to (9) will be a minimum of  $\phi(G)$  in (8). Important issues on

the existence and convergence of the solution to (9) on  $[0, \infty)$  are discussed in [5].

In the ODE (12),  $T = S\hat{U}^*$  where

$$s_l = - \begin{bmatrix} I \\ \lambda_l I \end{bmatrix} \mathcal{C}_l^{-1} B \hat{g}_l$$

with  $C_l = \lambda_l^2 M + \lambda_l D + K$ . And also  $\frac{\partial T}{\partial g_{jk}} = \frac{\partial S}{\partial g_{jk}} \hat{U}^*$ where

$$\frac{\partial s_l}{\partial g_{jk}} = \begin{cases} -\delta_{lk} \begin{bmatrix} I \\ \lambda_l I \end{bmatrix} C_l^{-1} b_j, \\ k = 1, \dots, p, \\ -\frac{\delta_{lk} + \delta_l(k+1)}{\sqrt{2}} \begin{bmatrix} I \\ \lambda_l I \end{bmatrix} C_l^{-1} b_j, \\ k = p + 1, p + 3, \dots, 2n - 1, \\ -i \frac{\delta_l(k-1) - \delta_{lk}}{\sqrt{2}} \begin{bmatrix} I \\ \lambda_l I \end{bmatrix} C_l^{-1} b_j, \\ k = p + 2, p + 4, \dots, 2n \end{cases}$$

where  $\delta_{jk}$  the Kronecker delta. It is important to realize that  $\frac{\partial T}{\partial g_{jk}}$  is a constant matrix and is only required to evaluate once in the computation process.

# 4 Conclusion

In this note, the problem of robust pole assignment for second-order systems by state feedback is examined. It has been shown that pole assignment can be achieved by either solving a linear matrix equation or linear system. The assigned spectrum is made optimally robust by minimizing an objective function via gradient flow which involves the solution of a system of ordinary differential equation.

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