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Modeling and Control of a Multiport Converter based EV Charging Station with PV and Battery

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Abstract—As an environmental friendly vehicle, the increasing number of electrical vehicles (EVs) leads to a pressing need of widely distributed charging stations, especially due to the limited on-board battery capacity. However, fast charging stations, especially super-fast charging stations may stress power grid with potential overload at peaking time, sudden power gap and voltage sag. This paper discusses the detailed modeling of a multiport converter based EV charging station integrated with PV power generation, and battery energy storage system, by using ANSYS TwinBuilder. In this paper, the control scheme and combination of PV power generation, EV charging station, and battery energy storage (BES) provides improved stabilization including power gap balancing, peak shaving and valley filling, and voltage sag compensation. As a result, the influence on power grid is reduced due to the matching between daily charging demand and adequate daytime PV generation. Simulation results are presented to confirm the benefits at different modes of this proposed multiport EV charging circuits with the PV-BES configuration. Furthermore, SiC devices are employed to the EV charging station to further improve the efficiency. For different modes and functions, power losses and efficiency are investigated and compared in simulation with conventional Si devices based charging circuits.

Index Terms – EV charging station, PV, BES, power gap balancing, voltage sag compensation.

I. INTRODUCTION

With the growing interest in decreasing the fossil fuel utilization and pollution, electric vehicles (EVs) have emerged as an applicable alternative to conventional gas engine vehicles [1]. The development and increasing utilization of EVs requires widely distributed charging stations due to the limited EV battery capacity [2]. However, large scale of directly grid-connected charging stations, especially fast and superfast charging stations, stress power grid stability and reliability with peak demand overload, voltage sag, and power gap issues [3]. Some researchers have been integrating photovoltaic (PV) generation with EV charging infrastructure [4]; however, the PV integration is still considered as a minor portion of power source for EV charging stations in researches. As for the higher demand of fast-speed charging during daytime, the rapid development of PV generation optimizes power consumption at peak hours with its adequate daytime generations. With respect to the intermittency of solar energy, a battery energy storage (BES) can be employed to regulate the DC bus or load voltage, balance power gap, and smooth PV power [5].

Considering the high power density and high efficiency merits of the multiport power converters [6], a multiport DC/DC converter is employed in this paper for the EV charging station instead of using three separate DC/DC converters. Among the aforementioned research, the charging station architectures can be classified into two topologies: using AC bus or DC bus [7]. As PV output and BES can both be regarded as DC current source [8], DC bus charging station is chosen here to improve the utilization efficiency of solar energy and decrease the cost and losses of converters. Compared with isolated multiport converters, nonisolated multiport converters that are usually derived from buck or boost converters may feature a more compact design, higher power density, and higher efficiency compared with isolated multiport converters [9] [10]. Accordingly, a DC bus nonisolated structure with SiC switches is leveraged in this paper, to improve efficiency and minimize the power losses. To sum up, the works and contributions in this paper can be summarized as follows. First, the PV and BES integration, rather than the power grid, is considered as a predominant power supply for EV charging. Then, detailed operating modes, control scheme, and the interaction among PV, BES, power grid, and EV charging are developed and investigated, in a scenario of high penetration of PV integration and widely spread EV charging infrastructures. Additionally, detailed power losses and efficiency comparison is investigated.

The remainder content of this paper is organized as follows. In Section II, the charging station architecture is proposed. Then the workflows between different power sources, and control designs including mathematical modeling, modulation, and power balance will be introduced. In Section III, the proposed control scheme for different modes is presented. Then the simulation results, power losses and efficiency comparison between Si and SiC EV stations will be presented. Finally, conclusions will be given in Section IV.

II. SYSTEMATIC MODELING AND OPERATING PRINCIPLE

In the conventional architecture of DC bus charging station with PV integration (Fig. 1a), all the three power sources, including PV and EV charger unidirectional sources, and AC grid bi-directional source, are all connected through three separate converters. The proposed DC bus charging station (Fig. 1b), consists of one more bi-directional power source BES sharing the same DC bus. The BES is utilized to maintain the DC link voltage and balance power surplus/insufficiency

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Figure 1. Multiport converter architectures, (a) the conventional architecture of EV charging stations integrated with PV, and (b) the proposed multiport converter based EV charging station architecture integrated with PV and BES.

from the PV (Fig. 4). With this configuration, the function and operating modes can be discussed as follows in detail.

A. Mode 1: PV to EV

In this mode, the switches S_{pv} , S_{b1} , and S_{b2} are turned off while S_{EV} is turned on (Fig. 2a). Therefore, PV directly delivers power to the load, as shown in Fig. 2a. The differential equations in this stage can be expressed as follows:

$$i_{PV} = C_1 \frac{dv_{C1}}{dt} + i_{EV} \tag{1}$$

$$C_2 \frac{dv_{C2}}{dt} = \frac{v_{Bat} - v_{C2}}{r_b} - i_{L2} \tag{2}$$

$$i_{EV} = C_3 \frac{dv_{C3}}{dt} + \frac{v_{EV}}{R_{EV}}$$
 (3)

$$v_{C1} - v_{C3} = L_3 \frac{i_{L3}}{dt} \tag{4}$$

$$L_2 \frac{i_{L2}}{dt} = -v_{C2}$$
 (5)

where C_1 , C_2 , C_3 , L_1 , L_2 , L_3 , and r_b represent the capacitance of the PV port capacitor, the capacitance of the BES port capacitor, the capacitance of the EV port capacitor, the inductance of the PV port inductor, the inductance of the BES port inductor, the inductance of the EV load port inductor, and the equivalent resistance between v_{Bat} and C_2 , respectively, as shown in Fig. 1b; i_{PV} , i_{EV} , i_{L2} , and i_{L3} represent the output current from PV panels, the current of EV load, the current through inductor L_2 , and the current through inductor L_3 , respectively; v_{C1} , v_{C2} , v_{C3} , v_{Bat} , and v_{EV} represent the voltage across capacitor C_1 , the voltage across C_2 , the voltage across C_3 , output voltage from BES, and the charger voltage, respectively. The duty cycle for the switch S_{pv} can be obtained with:

$$\frac{V_{DC}}{V_{PV}} = \frac{1}{1 - D_{pv}}$$
(6)

where V_{DC} , V_{PV} , and D_{pv} represent the DC link voltage, voltage of PV array, and duty cycle of switch S_{pv} , respectively.

B. Mode 2: BES to EV

When S_{pv} and S_{EV} are turned on while S_{b1} and S_{b2} are turned off, BES is discharged to the EV load, as shown in Fig. 2b. The differential equations in this mode can be expressed as follows:

$$i_{PV} = C_1 \frac{dv_{C1}}{dt} \tag{7}$$

$$L_2 \frac{i_{L2}}{dt} = v_{DC} - v_{C2} \tag{8}$$

$$v_{DC} - v_{C3} = L_3 \frac{i_{L_3}}{dt} \tag{9}$$

$$C_2 \frac{dv_{C2}}{dt} = \frac{v_{Bat} - v_{C2}}{r_b} - i_{L2} \tag{10}$$

$$i_{EV} = C_3 \frac{dv_{C3}}{dt} + \frac{v_{EV}}{R_{EV}}$$
(11)

where v_{DC} refers to DC link voltage, which equals to the voltage across capacitor C_4 . The duty cycle for switch S_{b1} can be obtained with:

$$\frac{V_{DC}}{V_{Bat}} = \frac{1}{1 - D_{b1}}$$
(12)

where V_{DC} , V_{Bat} , and D_{b1} represent the DC link voltage, voltage of BES, and duty cycle of switch S_{b1} , respectively.

C. Mode 3: PV to BES

When S_{b2} is turned on while S_{b1} , S_{pv} and S_{EV} are turned off, BES is charged from the PV surplus energy, as shown in Fig. 2c. The differential equations in this mode can be expressed as follows:

$$i_{PV} = C_1 \frac{dv_{C1}}{dt} - i_{L2} \tag{13}$$

$$L_2 \frac{i_{L2}}{dt} = v_{C1} + v_{DC} - v_{C2} \tag{14}$$

$$L_3 \frac{i_{L3}}{dt} = v_{DC} - v_{C3} \tag{15}$$

$$C_2 \frac{dv_{C2}}{dt} = \frac{v_{Bat} - v_{C2}}{r_b} - i_{L2} \tag{16}$$

$$i_{EV} = C_3 \frac{dv_{C3}}{dt} + \frac{v_{EV}}{R_{EV}}$$
 (17)

The duty cycle for the switch S_{b2} can be obtained with:



Figure 2. Multiport converter operating modes, (a) PV supplies EV charging when solar energy is sufficient, (b) BES supplies EV charging during PV intermittent, and (c) PV charges BES when solar generation is surplus.



Figure 3. PV outputs when irradiance drops from 1000 to 500 W/m^2 , (a) the I-V and P-V characteristics of the modeled PV panels, (b) the output power from the PV panels, and (c) the output voltage of the PV panels.

$$\frac{V_{Bat}}{V_{DC}} = D_{b2} \tag{18}$$

where D_{b2} represents the duty cycle of the switch S_{b2} .

D. Other Modes: PV to BES, Grid to EV, and PV to Grid

The operating principle of other modes including PV to BES, grid to EV, and PV to grid, are summarized in Table I. Besides, the differential equations can be similarly expressed with the same analysis method in Modes 1 to 3. The detailed simulation analysis will be provided in the following section.

Table I THE EV CHARGING OPERATING MODES

S_{pv}	S_{b1}	S_{b2}	S_{EV}	Power flow
off	off	off	on	PV to EV
off	off	on	off	PV to BES
on	off	off	on	BES to EV
—	on/off	off/on	on	Grid to EV
off	off	off	off	PV to grid

III. SIMULATION RESULTS

To evaluate the proposed charging station functions and control schemes, a simulation model following the structure in Fig. 1b is established in ANSYS TwinBuilder. The PV array is modeled with Suntech STP235-20-Wd [11], with 14 strings in series and 5 strings in parallel. For the open circuit, the modeled PV array is able to supply 16kW 500V to feed the



Figure 4. The block diagram for the BES controller and the PV controller with MPPT.

EV charger and a MPPT controller (Fig. 3a) is developed to extract the maximum power from PV panels while maintains a constant DC voltage.

For the BES controller, the objective is to regulate the power gap and support the load voltage (Fig. 4). When the solar is sufficient, EV charging is supplied from PV panels. If the PV generation is surplus, BES will be charged and consume locally. If PV is insufficient such as partial shading and other intermittent conditions, BES starts to discharge and fill in the power gap between PV and EV charging.





Figure 5. The simulation results of EV charging, (a) the demand and consumed power of EV charging, (b) the terminal voltage of the EV charger.

Figure 6. The simulation results of the BES, (a) the output power from BES, (b) the terminal voltage of the BES.

A case of this control scheme is simulated with ANSYS TwinBuilder. At 400ms of the simulation time, the irradiance drops from $1000k/W^2$ to $500k/W^2$, and the output power of the PV panels drops from 11kW to 5.7kW at 425V (Fig. 3), with the MPPT strategy implemented. At 700ms of the simulation time, EV charging demand suddenly goes up from 5.7kW to 7.7kW.

In this scenario, between the simulation time of 0 to 400ms, the EV charging demand is low while the PV generation is sufficient. Therefore, both PV-to-EV and PV-to-BES modes are triggered, and the surplus PV generation charges the BES. Between the simulation time of 400ms to 700ms, the PV panels can provide 5.7kW which meets the EV charging amount. As a result, the system is operated in PV-to-EV mode and no BES charging/discharging is required. After the charging demand increase at 700ms, the PV panels are not able to supply all the required 7.7kW charging power under the condition of $500k/W^2$ irradiance. Therefore, the BES starts to discharge and supply EV charging with 2kW and provides voltage support, as shown in Fig. 6.

Additionally, to improve the efficiency and power density of the multiport converter, SiC MOSFETs (CREE 900V/36A C3M0065090D [12]) are utilized. The comparison of the

conduction losses and switching losses in each semiconductor device for different operating modes at different load percentage is simulated, as shown in Fig. 7. It can be seen that both the conduction (Fig. 7a) and switching losses (Fig. 7b) are reduced in the proposed SiC based EV charging stations. Particularly, the switching loss of the conventional Si converters is much larger than the one in the proposed SiC counterparts. As a result, the efficiency comparison based on the simulated device losses are shown in Fig. 8. At nominal load condition, the proposed EV charging station has an efficiency of 98.41% for PV-to-EV mode, which is 5.67% higher than the conventional Si (Infineon 900V/36A IPW90R120C3 Si MOSFET [13]) based converter at the same condition. For PV-to-BES mode at nominal rating, the proposed EV charging station has an efficiency of 98.37%, which is 4.46% higher than the conventional Si based converter under the same condition. Under the condition of the nominal rating of the BES-to-EV mode, the efficiency of the proposed charging station is 6.00% higher than the Si counterpart.

IV. CONCLUSIONS

In this paper, a multiport converter based EV charging station with PV and BES is proposed. A BES controller is



Figure 7. Comparison of the power device losses between the Si converters and SiC counterpart, at various load percentage and different operating modes (PV to EV, PV to BES, and BES to EV), (a) conduction losses, (b) switching losses with the same labels and orders as in (a).



Figure 8. Comparison of the peak efficiency between the Si converters and SiC counterpart, at various load percentage and different operating modes (PV to EV, PV to BES, and BES to EV).

developed to regulate the voltage sag, and balance the power gap between PV generation and EV charging demand. With the proposed control design, BES starts to discharge when PV is insufficient for local EV charging, and starts to charge when PV generation is surplus or power grid is at valley demand, such as during nighttime. As a result, the combination of EV charging, PV generation, and BES enhances the stability and reliability of the power grid. Different operating modes and their benefits are investigated and then, simulation and thermal models of the multiport converter based EV charging stations and the proposed SiC counterpart are developed in ANSYS TwinBuilder. Simulation results show that the efficiency can be improved by 5.67%, 4.46%, and 6.00%, respectively, for PV-to-EV mode, PV-to-BES, and BES-to-EV mode at nominal operating condition, compared to Si based EV charging stations under the same operating conditions.

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