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# DESIGN OF TERNARY OPERATIONS UTILIZING FLOW-BASED COMPUTING 

by

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M.S. Computer Science, University of Central Florida, 2018

A dissertation submitted in partial fulfilment of the requirements
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#### Abstract

The development of algorithms and circuit designs that exploit devices that have the ability to persist multiple values will lead to alternative technologies to overcome the issues caused by the end of Dennard scaling [1] and slowing of Mooreś Law [2] [3]. Flow-based designs have been used to develop binary adders and multipliers [4] [5] [6]. Data stored on non-volatile memristors are used to direct the flow of current through nanowires arranged in a crossbar. The algorithmic design of the flow-based crossbar is fast, compact, and efficient [5]. In the following paper, we seek to automate the discovery of flow-based designs of ternary circuits utilizing memristive crossbars.


## ACKNOWLEDGMENTS

I recognize Sumit Kumar Jha for his work with crossbar memristors leading to the inspiration of applying multi-valued logic to new technology. The C++ test environment was developed by Jim Pyrich derived from Python code written by Dwaipayan Chakraborty. The NuSMV simulations were performed by Sunny Raj.

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# CHAPTER 1: INTRODUCTION 

Ternary Based Computing

A digital system using a radix of base 3 may be a more efficient implementation of switching circuits than a radix of 2 [Alexander, 1964] [7]. The most economical radix is calculated by finding the relationship of the amount of the equipment required for a certain radix and the digit capacity of the machine. The resulting minimum radix is the natural base ( $e=2.718$ ). Since the radix needs to be an integer, this value is rounded up to 3 which implies that a ternary computer might have an advantage over existing binary based systems. When ternary computers were first proposed, there were no devices in common use that would maintain 3 stable states. The discovery of the missing circuit element [8] [9] and the subsequent fabrication of the memristor [10] provide a device to maintain multiple states required for multi-value operations.

In [11], it is stated that ternary computers (base 3) can compete with binary computers (base 2) in terms of performance, but cost may a more limiting factor. In Table 4.2, there are 18 minterms required to compute a 1-bit ternary adder. This would require 1.62 times more logic to produce a ternary computer when compared to a conventional binary computer. Ternary digits (or trits) carry $\log _{2} 3=1.58$ bits of information.

A design and implementation of 2 bit ternary ALU device in [12] proposed that a ternary system might have an advantage over existing binary based systems since interconnections and chip area would be reduced. Although the ternary ALU (T-ALU) described was designed for 2 bit operations, ALUs could be cascaded resulting in $n$-bit operations. The functions of the T-ALU include addition, subtraction, multiplication, comparison, OR, NOR, AND, NAND, and Ex-OR.

The integration of ternary logic gates into conventional binary structures has been researched [13]
using a carbon nanotube field-effect (CNTFET) circuit. Carbon nanotubes (CNTs) can be used to implement efficient multiple valued logic (MVL) circuits by varying with multiple threshold voltages [14]. The threshold voltages can be adjusted by adjusting the diameter of the CNT [15] with a reference point of [16] as $V_{\pi}(3.033 \mathrm{eV})$ in a SPICE simulation. Based on certain characteristics, this design methodology was shown to be an alternative to metal oxide semiconductor field effect transistor (MOSFET) circuits to efficiently produce MLV circuits.

## Memristive Systems

In 1971, Leon Chua identified a fourth basic circuit element called the memristor [8]. The fundamental circuit elements are current $i$, voltage $v$, charge $i$, and flux $\varphi$. The memristor is defined as the relationship between charge and flux (Memristance $=$ Flux/Charge) [17] [18]. When controlled by charge, the measure of memristance is:

$$
\begin{gather*}
v(t)=M(q(t)) i(t)  \tag{1.1}\\
M(q) \equiv \frac{d \varphi(q)}{d q} \tag{1.2}
\end{gather*}
$$

When controlled by flux, the measure of memductance is:

$$
\begin{gather*}
i(t)=W(\varphi(t)) v(t)  \tag{1.3}\\
W(\varphi) \equiv \frac{d q(\varphi)}{d \varphi} \tag{1.4}
\end{gather*}
$$

In 1976, Chua and Kang broadened the definition of memristors as a class of nonlinear dynamical systems called memristive systems [19]. One of the defining properties was a zero-crossing property where the output was zero when then the input was zero. The output takes the form of a Lissajous figure which can be correlated to a pinched hysteresis loop [20] or bowtie[10].

In 2015, Chua stated that any two terminal device that produces a hysteresis loop and passes through the origin is to be considered a memristor [21]. Classes of memristors are defined as ideal, ideal generic, generic, and extended. In 2018, Biolek et. al. [22] documents other non-memristive systems that produce pinched hysteresis loops. The question is raised whether the newer classes of memristors are actually new circuit elements. Chua states that the definition of a memristive system should be considered a black box since he believes that the internal composition is irrelevant. In [23], Chua states "If it's pinched, it's a memristor." which Biolek considers hyperbole. This paper will focus on ideal memristors and models for general memristive devices [24].

In a presentation titled Fundamentals of Memristors [25], the following equations are presented for ideal, generalized memristors, and memristive systems [8] [19] [23]. Ideal current controlled memristors are defined in equation 1.5. The equation for a current controlled generalized memristive system is 1.6 . For time invariant memristive systems, the properties in 1.7 are needed to maintain a pinched hysteresis loop.

$$
\begin{equation*}
V=R(q) * I \tag{1.5}
\end{equation*}
$$

$$
\begin{align*}
& V=R(x, I) * I  \tag{1.6}\\
& \hat{x}=f(x, I) \\
& R(x, I) \neq \infty  \tag{1.7}\\
& f(x, 0)=0
\end{align*}
$$



Figure 1.1: The Berkeley Model and Algorithm Prototyping Platform

In the above equations, the value $x$ represents the inner state variable.

## Crossbars and Sneak Paths

Memristive crossbars may be effective to overcome limitation of existing architectures [17]. Crossbar networks are organized with a top set of parallel nanowire electrodes perpendicular to a bottom set of the same. At each point where the wires cross is a memristor that can be turned off and on directing current flow to connected wires. A crossbar with $n$ rows and $m$ columns can contain up to $n m$ memristors [4] which results in 2 nm connections between each layer of nanowires. Sneak paths [26] are a problem when undesired paths created by ON memristors allow some current to flow to the attached set of nanowires.

Sneak paths can also be exploited as in [4] [5] [6] [27] [28] [29]. As an example, nanoscale
crossbars have been designed for a one-bit Boolean full adder with three input bits $\left(A, B, C_{i n}\right)$. The input bits determine the state of each memristor located where the crossbar wires intersect. Voltage is applied at the bottom row. The sum and carry-out bits are determined by the sneak current measured on the top two rows. A Boolean value of zero or False relates a turned-off memristor. Conversely, a Boolean value of one or True relates to a turned-on memristor. A crossbar array was fabricated at SUNY Polytechnic Institute using $1 k \Omega$ resistors for the ON state and $1 M \Omega$ for the OFF state. A pulse voltage of 100 mV and 200 mV were applied to the bottom row. The resulting voltages had a noise margin of 55 mV which is a sufficient separation to determine the Sum and $C_{\text {out }}$ Boolean values.

## Multilevel Digital Systems

In [30], mathematical models of memristors are analyzed. Digital circuits based on number systems other than binary are discussed. One method discussed is Memristor Quantization where ranges of resistance are divided evenly by $N$. For ternary options, where $N=3$, each memristor is divided evenly between $R_{o f f}$ and $R_{o n}$ with $\Delta R=\left(R_{o f f}-R_{o n}\right) / 3$. Considering HP's values [9] of $R_{o n}=100 \Omega$ and $R_{o f f}=16 k \Omega$, the center values would be approximately $\{2.5 k \Omega, 7.5 k \Omega$, $12.5 k \Omega\}$. False detection becomes more of an issue with a larger $N$. A gap between regions is a necessity [31] for good accuracy. It may also be beneficial to divide regions unequally based on the probability of results.

## Flow-based Computing

Flow-based designs have been used to develop binary adders and multipliers [4] [5] [6]. Data stored on non-volatile memristors are used to direct the flow of current through nanowires arranged in a
crossbar. The algorithmic design of the flow-based crossbar is fast, compact, and efficient [5]. The data is encoded based on the value of each input, then it is stored in the memristors affecting the flow of current across each nanowire [27] [28]. Some memristors can by permanently turned ON or OFF which can relate to literal values in the particular formula [29].

## CHAPTER 2: LITERATURE REVIEW

## SPICE Circuit Simulator

In [32] [33], a memristor device model is proposed to simulate the $I-V$ relationship of a memristor based on changes to a state variable with respect to time. The state variable is a value between 0 and 1 that affects current flow and conductivity of the device. The SPICE sub-circuit in [34] utilizes the hyperbolic sine function to estimate conductivity beyond a voltage threshold. The parameters $a_{1}$ and $a_{2}$ represent amplitudes based on the direction of the current. The intensity of the threshold is determined by the parameter $b$.

$$
I(t)= \begin{cases}a_{1} x(t) \sinh (b V(t)) & V(t) \geq 0 \\ a_{2} x(t) \sinh (b V(t)) & V(t)<0\end{cases}
$$

The conductance is determined using negative and positive thresholds defined as $A_{p}$ and $A_{p}$.

$$
g(V(t))= \begin{cases}A_{p}\left(e^{V(t)}-e^{V_{p}}\right) & V(t)>V_{p} \\ -A_{n}\left(e^{-V(t)}-e^{V_{n}}\right) & V(t)<-V_{n} \\ 0 & -V_{n} \leq V(t) \leq V_{p}\end{cases}
$$

The version of SPICE utilized in this study is the open source version Ngspice, a software application originally written by Berkeley University, currently maintained by the Ngspice project [35]. Ngspice is simulation software used to test circuit designs prior to the fabrication of a physical circuit. Ngspice is used with a model circuit designed specifically for testing memristive devices [36].

Simulated Annealing

Simulated Annealing (SA) can be used to solve optimization problems [37]. The SA algorithm performs a random walk through the search space looking for the lowest energy. The benefit of a random walk is the search will not get stuck in local minima while attempting to find the best solution (i.e. lowest temperature).

SA can be used to find solutions to problems that are considered NP-Hard. One heuristic that can be used to find an optimal solution is to start with a known path followed by iterative improvement based on a cost factor. Another method is start with a fully random solution [38]. Minor changes are made to the path at each step and the cost factor is recomputed.

When annealing occurs in a metal, the rate of the cooling process determines the state of the solid that is formed. In SA, starting/ending temperatures and a rate of cooling are set. To find an optimal solution, the temperature should be reduced slowly. For each iteration (or step), the cost factor (or energy) is computed and returned to the calling algorithm. A set number iterations are performed for each step so the algorithm can explore an area the for lowest energy.

The following constants were used for Simulated Annealing (SA) in the GNU Scientific Library (GSL) [39]:

| Step | 1000000 |
| :--- | ---: |
| Iterations | 1000 |
| Step Size | 1.0 |
| Boltzmann Constant | 1.0 |
| Initial Temp | 2500.0 |
| Damping Factor | 1.002 |
| Minimum Temp | $1 \mathrm{e}-25$ |

The simulated annealing algorithm calculates the probability of taking a step using the Boltzmann
distribution:

$$
p=e^{\left(E_{i+1}-E_{i}\right) / k T}
$$

if $E_{i+1}>E_{i}$ and $p=1$ when $E_{i+1} \leq E_{i}$.

The initial temperature $T$ is set to a high value and is lowered by the damping factor until the minimum temperature is reached. The GSL implementation of simulated annealed uses callback functions for energy and step decisions. The energy callback function calculates the cost or best energy of the current step. If the cost is lower, a step will occur. The GSL function provides a callback function to output its progress. Upon completion, the best result is returned.

## NuSMV

NuSMV is a symbolic model checker based on CMU SMV [40]. NuSMV implements a BDDbased (Binary Decision Diagram) and SAT-based (Boolean Satisfiability Problem) model checking capability. NuSMV was used to prove that a solution existed for implementation of the full-adder using memristive crossbars.

## Modified Nodal Analysis

It is well known that Ohm's Law $(V=I R)$ describes the linear relationship between voltage, current, and resistance. An ideal memristor is a resistor that changes based on the relationship between charge and flux over time [41]. At a snapshot in time, a resistive crossbar network can be analyzed using a modified nodal approach [42].

The following node voltage method can be used to calculate current through a resistive crossbar network [43]. To apply the node voltage method to a circuit with $n$ nodes (with $m$ voltage sources),
perform the following steps [44].

1. Select a reference node (usually ground).
2. Name the remaining $n-1$ nodes and label a current through each passive element and each current source.
3. Apply Kirchhoff's current law to each node not connected to a voltage source.
4. Solve the system of $n-1-m$ unknown voltages.

To solve the system of equations, the matrix formula $A x=z$ is used where $x$ is a $(n+m) x 1$ matrix that contains the unknown $n$ node voltages $m$ represents independent current sources. The solution is found by using an algorithmic MNA formula that takes the inverse of the $A$ matrix multiplied by the $z$ matrix, $x=A^{-1} z$ [43].

## Memristive Device Models

DC analysis of memristive device models can be improved by formatting the simulation algorithms in a differential equation form [24]. Memristors and other devices that exhibit an $i-v$ hyteresis require equations that result in a smooth curve on a graph. Memristive algorithms with piecewise window functions tend not to be continuous. Popular models were modified to address the dynamics of the internal state variable and implement functions that are smooth and safe.

The Berkeley Model and Algorithm Prototyping Platform (MAPP) [45] was used to produce MATLAB graphs and to check results. MAPP utilizes ModSpec [46] to describe and prototype new devices in MATLAB. The MAPP environment is initiated in MATLAB by running startMAPP. The test function memristor ModSpec utilizes two switches, $f 1=1$ (shifting between resistor states) and $f 2=5$ (modified Yakopcic model).


Figure 2.1: University of Berkeley: Yakopcic Model


Figure 2.2: University of Berkeley: 3D Homotopy Analysis

## CHAPTER 3: METHODOLOGY

The research goal is to find an efficient method of simulating ternary operations utilizing memristive crossbars. Initial tests were performed using basinhopping (SciPy) referencing a dictionary of potential memristor state values. A SPICE call was made for each truth table entry. Based on the combined results of each SPICE call, the memristor list was randomly perturbed and the cycle continued until the process timed out or was canceled. The initial program utilized 18 SPICE call since $C_{i n}=2$ was excluded.

The Python program was converted to C++ with the integration of Simulated Annealing (SA) in the GNU Scientific Library (GSL) [39]. The next step was to develop a flow-based method of estimating valid circuits to eliminate the need for external systems calls to SPICE or any other external library. This helped performance greatly although a solution was still elusive. To confirm we were on the right track, NuSMV was used to prove that a potential solution existed. It was then that we realized that we weren't testing the entire truth table for ternary addition. We included all 27 entries and removed the dictionary in lieu of a full memristor array that also included static connections. A half-ternary adder was developed that limited the number of truth table entries to nine which was very fast. A full-ternary adder followed with numerous optimizations for speed and accuracy.

A significant effort was made toward narrowing the gap of resistances for each state (i.e. on, off, mid). It was decided to test resistive networks using SPICE and modified nodal analysis in an attempt to find linear solutions. This research was mildly successful, but the gaps were too tight given the new resistances and the resulting arrays did not port when tested with memristors. There were relatively few outliers that, if resolved, will allow for tighter ranges and larger gaps needed for fabrication.

## CHAPTER 4: SYNTHESIS OF A TERNARY ADDER

## Approximate Synthesis of a One-Bit Adder Using Ternary Logic

In 2014, Velasquez and Jha [17] demonstrated 1-bit addition and other Boolean formulas measuring sneak path voltages on memristive crossbars. The state of each memristor was either $R_{o f f}$ or $R_{o n}$ mapping to 0 or 1 on the truth table shown in Table 4.1. The Boolean formula is for computing the Sum bit is:

$$
\left(A \wedge \neg B \wedge \neg C_{i n}\right) \vee\left(\neg A \wedge B \wedge \neg C_{i n}\right) \vee\left(\neg A \wedge \neg B \wedge C_{i n}\right) \vee\left(A \wedge B \wedge C_{i n}\right)
$$

This work shows that an $n$-ary Boolean function where $n=3$ results in $2^{3}=8$ rows in Table 4.1. A set of memristors are mapped to the values of $A, B$, and $C_{i n}$ producing the values in the Sum column. The simulations relate voltage drops to logical values. A drop of 1 volt indicates true and drops below 0.5 volts indicates the logical value of false. A simulation was run with the current flowing through the voltage source was $10^{-4}$ amperes when the formula was true and less than $10^{-5}$ amperes when the formula was false.

Table 4.1: Truth Table: Binary Addition

| $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Sum |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 1 |
| 0 | 1 | 0 | 0 | 1 |
| 0 | 1 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 | 0 |
| 1 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 1 | 1 |

Table 4.2: Truth Table: Ternary Addition (partial)

| $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Sum |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 1 |
| 0 | 1 | 0 | 0 | 1 |
| 0 | 1 | 1 | 0 | 2 |
| 0 | 2 | 0 | 0 | 2 |
| 0 | 2 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 | 2 |
| 1 | 1 | 0 | 0 | 2 |
| 1 | 1 | 1 | 1 | 0 |
| 1 | 2 | 0 | 1 | 0 |
| 1 | 2 | 1 | 1 | 1 |
| 2 | 0 | 0 | 0 | 2 |
| 2 | 0 | 1 | 1 | 0 |
| 2 | 1 | 0 | 1 | 0 |
| 2 | 1 | 1 | 1 | 1 |
| 2 | 2 | 0 | 1 | 1 |
| 2 | 2 | 1 | 1 | 2 |

Simulations

A test application was designed to use Ngspice (v.28) to test each possible circuit in the search for a solution. The program was written in Python (Chakraborty, 2018) simulating solutions of the ternary adder in Table 4.2 by using Simulated Annealing (SA) [39]. A circuit file was generated as input to Ngspice, followed by an external program call with subsequent analysis of the output. This process was repeated for each row of the truth table. The Python program didn't result in a solution initially. To speed up the process, a testing program was written in $\mathrm{C}++$ (Pyrich and Jha, 2018). Some of the processes were faster, but system calls to SPICE were very inefficient. A different approach was needed to simulate the crossbar and memristor states to search for a possible solution or approximation.

It was determined that numerous system calls to SPICE was causing a bottleneck, so the energy function was refactored to a flow-based design with SPICE calls occurring only when the free energy equaled zero, indicating a possible design of a crossbar adder. During each SA callback to the energy function, the memristors on the crossbar were randomly perturbed. A SPICE circuit was then generated based on the dictionary in Table 4.3.

Table 4.3: Simulated Annealing Dictionary

| Memristor | Variable | Value |
| :---: | :---: | :---: |
| 0 | Constant | 0 |
| 1 | Constant | 1 |
| 2 | Constant | 2 |
| 3 | $A$ | truth table |
| 4 | $A^{\prime}$ | $(A+1) \bmod 3$ |
| 5 | $\neg A$ | $(A+2) \bmod 3$ |
| 6 | $B$ | truth table |
| 7 | $B^{\prime}$ | $(B+1) \bmod 3$ |
| 8 | $\neg B$ | $(B+2) \bmod 3$ |
| 9 | $C_{i n}$ | truth table |
| 10 | $C_{\text {in }}^{\prime}$ | $\left(C_{i n}+1\right) \bmod 3$ |
| 11 | $\neg C_{\text {in }}$ | $\left(C_{i n}+2\right) \bmod 3$ |

Before proceeding with a redesign of the ternary adder application, NuSMV was used to prove that a memristive crossbar existed (Raj, 2018). The ternary full-adder was used ( $3^{3}=27$ entries) to produce the solution in Table 4.4.

Table 4.4: NuSMV $5 \times 5$ design of the sum bit

| $m$ | 0 | 1 | 2 |
| :---: | :---: | :---: | :---: |
| 1 | 0 | 0 | 1 |
| 2 | 0 | 1 | 0 |
| 3 | 0 | 0 | 2 |
| 4 | 0 | 2 | 0 |
| 5 | 0 | 1 | 2 |
| 8 | 0 | 2 | 2 |
| 9 | 1 | 0 | 0 |
| 12 | 1 | 0 | 2 |
| 13 | 1 | 2 | 0 |
| 18 | 2 | 0 | 0 |
| 19 | 2 | 0 | 1 |
| 21 | 2 | 0 | 2 |



Figure 4.1: Crossbar design for computing the sum bit of ternary addition obtained from NuSMV

# CHAPTER 5: DESIGN OF A TERNARY ADDER 

Flow-Based Computing


#### Abstract

A flow-based methodology was used to compute memristive crossbars based on the state of the memristor and resulting sneak paths shown in Table 5.1 [47].


Table 5.1: Crossbar Simulation

| Memristor | Row | Column | Result |
| :---: | :---: | :---: | :---: |
| on | on | off | col = on |
| on | on | mid | col = on |
| on | mid | off | col = mid |
| on | mid | mid | col = mid+ |
| on | off | on | row = on |
| on | mid | on | row = on |
| on | off | mid | row = mid |
| on | mid | mid | row = mid |
| mid | on | off | col = mid |
| mid | on | mid | col = mid+ |
| mid | mid | off | col = mid+ |
| mid | mid | mid | col = mid+ |
| mid | off | on | row = mid |
| mid | mid | on | row = mid+ |
| mid | off | mid | row = mid+ |
| mid | mid | mid | row = mid+ |

In the previous chapter, it was shown that the SA energy function required a faster method to find potential circuits. The revised SA energy function estimates current flow based on the state of each row and column based on the state of the memristor. The state of the memristor is determined by each row in Table 5.2. For example, if the memristor value is B13, the table entry is $1,2,0$ which maps to $B=[o n$, mid, of $f)$ where $o f f=0$, mid $=2$,on $=1$. Note that in the experiments listed in the chapter on Findings, values are sometimes mapped as off $=0$, mid $=1$, on $=2$ for
reasons of better performance or separation of output voltage ranges.
Table 5.2: Design of a 1-bit ternary adder

| $m$ | 0 | 1 | 2 |
| :---: | :---: | :---: | :---: |
| 1 | 0 | 0 | 1 |
| 2 | 0 | 1 | 0 |
| 3 | 0 | 0 | 2 |
| 4 | 0 | 2 | 0 |
| 7 | 0 | 1 | 1 |
| 8 | 0 | 2 | 2 |
| 13 | 1 | 2 | 0 |
| 16 | 1 | 2 | 1 |
| 18 | 2 | 0 | 0 |
| 21 | 2 | 0 | 2 |
| 22 | 2 | 2 | 0 |

## Ternary Adder Algorithm

The SA energy callback function is described in the Ternary Adder search Algorithm 1. The mem value passed into the ENERGY function from SA is an array of possible memristor values. Flowbased analysis of each row in the ternary truth table is performed based on each value in the mem array. The resulting value is read from the last row for Sum and the next to the last row for $C_{\text {out }}$. When the energy factor bestenergy $=$ mismatches/num_values is zero, a possible design has been found. A SPICE circuit is prepared for each table entry, storing the values in spiceresults. If spiceresults match Sum and $C_{o u t}$, the ENERGY function returns success and the program ends, otherwise the SA procedure continues searching for another possible design.

```
Algorithm 1 Ternary Adder
    procedure SEARCH
            Initialize rows, columns, resistances
            Load ternary truth table \(A, B, C_{\text {in }}\)
            Randomize initial memristor states
            Initialize Simulated Annealing (SA)
            loop
                function EnERGY(mem) \(\triangleright\) mem suggested memristor array
                    Simulate Circuit Design
                    Apply Current to row \(=0\)
                    Logically Determine Flow Based on Crossbar Design
                    Read Resulting Voltages for \(C_{\text {out }}\) and Sum
                    Calculate Energy Factor bestenergy
                    if bestenergy \(>0\) then
                return bestenergy
                    Prepare SPICE Circuit
                    Compare SPICE Results to Acceptable Ranges spiceresults
                    if spiceresults \(=\) sum and \(C_{\text {out }}\) then
                    Report Results
                    return success
                    else
                    return continue
```


## Ternary Operations

Binary half-adders have two input bits that produce a $S u m$ and $C_{\text {out }}$ bit [48]. Ternary half-adders have two ternary inputs that outputs ternary values $(0,1,2)$ for $S u m$ and $C_{o u t}$. Since the carry-in value $C_{i n}$ is excluded for a half-adder, the truth table 5.3 is simplified.

A ternary full-adder has three ternary inputs $A, B$, and $C_{i n}$ that produce output values of $S u m$ and $C_{\text {out }}$ based on Table 5.4.

The ternary multiplier has three ternary inputs $A, B$, and $C_{i n}$ that produces output values of Product and $C_{\text {out }}$ based on Table 5.5.

Table 5.3: Truth Table: Ternary Half-Adder

| $m$ | $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Sum |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 |
| 2 | 0 | 1 | 0 | 0 | 1 |
| 4 | 0 | 2 | 0 | 0 | 2 |
| 9 | 1 | 0 | 0 | 0 | 1 |
| 11 | 1 | 1 | 0 | 0 | 2 |
| 13 | 1 | 2 | 0 | 1 | 0 |
| 18 | 2 | 0 | 0 | 0 | 2 |
| 20 | 2 | 1 | 0 | 1 | 0 |
| 22 | 2 | 2 | 0 | 1 | 1 |

Table 5.4: Truth Table: Ternary Full-Adder

| $m$ | $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Sum |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 1 | 0 | 1 |
| 2 | 0 | 1 | 0 | 0 | 1 |
| 3 | 0 | 0 | 2 | 0 | 2 |
| 4 | 0 | 2 | 0 | 0 | 2 |
| 5 | 0 | 1 | 2 | 1 | 0 |
| 6 | 0 | 2 | 1 | 1 | 0 |
| 7 | 0 | 1 | 1 | 0 | 2 |
| 8 | 0 | 2 | 2 | 1 | 1 |
| 9 | 1 | 0 | 0 | 0 | 1 |
| 10 | 1 | 0 | 1 | 0 | 2 |
| 11 | 1 | 1 | 0 | 0 | 2 |
| 12 | 1 | 0 | 2 | 1 | 0 |
| 13 | 1 | 2 | 0 | 1 | 0 |
| 14 | 1 | 1 | 1 | 1 | 0 |
| 15 | 1 | 1 | 2 | 1 | 1 |
| 16 | 1 | 2 | 1 | 1 | 1 |
| 17 | 1 | 2 | 2 | 1 | 2 |
| 18 | 2 | 0 | 0 | 0 | 2 |
| 19 | 2 | 0 | 1 | 1 | 0 |
| 20 | 2 | 1 | 0 | 1 | 0 |
| 21 | 2 | 0 | 2 | 1 | 1 |
| 22 | 2 | 2 | 0 | 1 | 1 |
| 23 | 2 | 1 | 1 | 1 | 1 |
| 24 | 2 | 1 | 2 | 1 | 2 |
| 25 | 2 | 2 | 1 | 1 | 2 |
| 26 | 2 | 2 | 2 | 2 | 0 |

Table 5.5: Truth Table: Ternary Multiplication

| $m$ | $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Product |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 1 | 0 | 1 |
| 2 | 0 | 1 | 0 | 0 | 0 |
| 3 | 0 | 0 | 2 | 0 | 2 |
| 4 | 0 | 2 | 0 | 0 | 0 |
| 5 | 0 | 1 | 2 | 0 | 2 |
| 6 | 0 | 2 | 1 | 0 | 1 |
| 7 | 0 | 1 | 1 | 0 | 1 |
| 8 | 0 | 2 | 2 | 0 | 2 |
| 9 | 1 | 0 | 0 | 0 | 0 |
| 10 | 1 | 0 | 1 | 0 | 1 |
| 11 | 1 | 1 | 0 | 0 | 1 |
| 12 | 1 | 0 | 2 | 0 | 2 |
| 13 | 1 | 2 | 0 | 0 | 2 |
| 14 | 1 | 1 | 1 | 0 | 2 |
| 15 | 1 | 1 | 2 | 1 | 0 |
| 16 | 1 | 2 | 1 | 1 | 0 |
| 17 | 1 | 2 | 2 | 1 | 1 |
| 18 | 2 | 0 | 0 | 0 | 0 |
| 19 | 2 | 0 | 1 | 0 | 1 |
| 20 | 2 | 1 | 0 | 0 | 2 |
| 21 | 2 | 0 | 2 | 0 | 2 |
| 22 | 2 | 2 | 0 | 1 | 1 |
| 23 | 2 | 1 | 1 | 1 | 0 |
| 24 | 2 | 1 | 2 | 1 | 1 |
| 25 | 2 | 2 | 1 | 1 | 2 |
| 26 | 2 | 2 | 2 | 2 | 0 |

## CHAPTER 6: FINDINGS

Ternary Adder

The initial search for a ternary design was a $4 \times 4$ memristive crossbar that outputs its $S u m$ value on the last row the crossbar and a $6 \times 5$ array that outputs $S u m$ on the last row and $C_{o u t}$ on the next to the last row.

An input array of 11 values $\left(0,1,2, A, A^{\prime}, \neg A, B, B^{\prime}, \neg B, C_{i n}, \neg C_{i n}\right)$ is prepared. The values of $A$ and $B$ are $(0,1,2)$ and $C_{i n}$ are $(0,1)$. This results in a total of $3 * 3 * 2=18$ possible outcomes.

The voltage values of 0 and 1 are the same as for a binary system as for the ternary design. The value of 2 maps to a mid-range value on the crossbar and memristor. The goal is to search for a design that matches the Sum column on Table 4.2. The initial attempt resulted in a best energy of 0.333333 which represents 12 matches and 6 mismatches totaling 18 Sum values.

| k | 0 | 1 | 9 | 13 | 16 | 17 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| a | 0 | 0 | 1 | 2 | 2 | 2 |
| b | 0 | 0 | 1 | 0 | 2 | 2 |
| cin | 0 | 1 | 0 | 0 | 0 | 1 |
| sum | 0 | 1 | 2 | 2 | 1 | 2 |
| xbar | 0 | 0 | 2 | 2 | 2 | 2 |
| diff |  | X |  |  | X |  |


|  | Rows: | Columns: | Best Energy: <br> 0.333333 |
| :--- | :--- | :--- | :--- |
|  | 4 | 3 |  |
| Memristor: |  |  | 5 |
|  | 1 | 0 | 4 |
|  | 5 | 4 | 8 |
|  | 7 | 10 | 6 |
|  | 0 | 3 |  |
| k=0 |  |  | $5: 2$ |
|  | $1: 1$ | $0: 0$ | $4: 1$ |
|  | $5: 2$ | $4: 1$ | $8: 2$ |
|  | $7: 1$ | $10: 1$ | $6: 0$ |
|  | $0: 0$ | $3: 0$ | $5: 2$ |
|  | $1: 1$ | $0: 0$ | $4: 1$ |
|  | $5: 2$ | $4: 1$ | $8: 2$ |
|  | $7: 1$ | $10: 0$ | $6: 0$ |
|  | $0: 0$ | $3: 0$ | $5: 1$ |
|  | $1: 1$ | $0: 0$ | $4: 0$ |
|  | $5: 1$ | $4: 0$ | $8: 1$ |
|  | $7: 0$ | $10: 1$ | $6: 2$ |
|  | $0: 0$ | $3: 2$ | $5: 1$ |
|  | $1: 1$ | $0: 0$ | $4: 0$ |
|  | $5: 1$ | $4: 0$ | $8: 1$ |
|  | $7: 0$ | $10: 0$ | $6: 2$ |

The value $k$ represents the truth table row. In the example above, the approximation was correct where $A=1, B=1$, and $C_{i n}=0$, but was incorrect where $A=2, B=2$, and $C_{i n}=0$. A number of changes were required achieve a better result including changing resistances, simulated annealing settings, and varying the mid, neg, and constant input values.

The values displayed for each $k$ are in the form $x: y$. The value $x$ indicates the memristor number and the value $y$ is the best energy state returned by the GSL simulated annealing function. The input values are mapped in Table 6.1.

Table 6.1: Input values to SA function

| Item | Variable | Value |
| :---: | :---: | :---: |
| 0 | Constant | 0 |
| 1 | Constant | 1 |
| 2 | Constant | 2 |
| 3 | A | truth table |
| 4 | $\mathrm{~A}^{\prime}$ | $(\mathrm{A}+1) \bmod 3$ |
| 5 | $\neg \mathrm{~A}$ | $(\mathrm{~A}+2) \bmod 3$ |
| 6 | B | truth table |
| 7 | B | $(\mathrm{~B}+1) \bmod 3$ |
| 8 | $\neg \mathrm{~B}$ | $(\mathrm{~B}+2) \bmod 3$ |
| 9 | $C_{\text {in }}$ | truth table |
| 10 | $\neg C_{\text {in }}$ | $\neg C_{\text {in }}$ |

In some experiments, the adjustment to the constants and variables were randomized. In the table, $A^{\prime}$ and $\neg A$ are adjusted by 1 and 2 , respectively. In one randomized version, the values were flipped (i.e. 2 and 1). In another case, we allowed duplicate values to see if we could achieve a better result.

The following results were obtained when the randomizer was used for each truth table row. The result of zero best energy was achieved after running simulated annealing for nearly 500 million steps. These results were considered an approximation since only a subset of the truth table was utilized.

| SUCCESS: | Rows: | Columns: |  |
| :--- | :--- | :--- | :--- |
|  | 4 | 3 | Best Energy: |
|  |  |  |  |
| Memristor: | 6 | 1 | 3 |
|  | 10 | 3 | 9 |
|  | 6 | 7 | 3 |
|  | 1 | 1 | 1 |
|  |  |  |  |
| k=0 | $6: 0$ | $1: 0$ | $3: 0$ |
|  | $10: 1$ | $3: 0$ | $9: 0$ |
|  | $6: 0$ | $7: 1$ | $3: 0$ |
|  | $1: 0$ | $1: 0$ | $1: 0$ |
|  | $6: 0$ | $1: 1$ | $3: 0$ |
|  | $10: 0$ | $3: 0$ | $9: 1$ |
|  | $6: 0$ | $7: 2$ | $3: 0$ |
| $\mathrm{k}=16$ | $1: 1$ | $1: 1$ | $1: 1$ |
|  | $6: 2$ | $1: 1$ | $3: 2$ |
|  | $10: 1$ | $3: 2$ | $9: 0$ |
|  | $6: 2$ | $7: 1$ | $3: 2$ |
| $\mathrm{k}=17$ | $1: 1$ | $1: 1$ | $1: 1$ |
|  | $6: 2$ | $1: 2$ | $3: 2$ |
|  | $10: 0$ | $3: 2$ | $9: 1$ |
|  | $6: 2$ | $7: 1$ | $3: 2$ |
|  | $1: 2$ | $1: 2$ | $1: 2$ |

The cross reference Table 6.2 was discovered using NuSMV. The search took over 7 hours to complete. Once this table was loaded into the C++ program and a two level lookup was implemented, simulated annealing started to find simulated designs (i.e. best energy $=0$ ) in 20 minutes. We then input the potential memristor into our Python-based SPICE checker to verify that the circuit simulation produced the same result. This method seemed to work well for $4 \times 4$ memristors with no carry-out bit.

Testing for the carry-out value was added resulting in possible solutions for $6 \times 5$ crossbars with some variation simulated in SPICE. The SPICE simulation was migrated to $\mathrm{C}++$ to automate and speed up the process. The final step was to test for acceptable ranges of output from SPICE.

Table 6.2: Cross Reference derived from NuSMV

| mem | $0 \quad 12$ | mem | 0 | 1 | 1 | 2 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 000 | 14 | 1 |  | 1 | 1 |
| 1 | $0 \begin{array}{lll}0 & 0 & 1\end{array}$ | 15 | 1 | 1 | 1 | 2 |
| 2 | $0 \begin{array}{lll}0 & 1\end{array}$ | 16 | 1 | 2 | 2 | 1 |
| 3 | $0 \begin{array}{lll}0 & 0 & 2\end{array}$ | 17 | 1 | 2 | 2 | 2 |
| 4 | $0 \quad 20$ | 18 | 2 |  | 0 | ) |
| 5 | $0 \begin{array}{lll}0 & 1\end{array}$ | 19 | 2 |  | 0 |  |
| 6 | $0 \quad 21$ | 20 | 2 |  | 1 |  |
| 7 | $0 \begin{array}{lll}0 & 1\end{array}$ | 21 | 2 | 0 | 0 | 2 |
| 8 | 022 | 22 | 2 |  | 2 | 0 |
| 9 | 100 | 23 | 2 |  | 1 | 1 |
| 10 | $1 \begin{array}{lll}1 & 0 & 1\end{array}$ | 24 | 2 |  | 1 | 2 |
| 11 | 110 | 25 | 2 |  | 2 | 1 |
| 12 | 1002 | 26 |  |  |  |  |
| 13 | 120 |  |  |  |  |  |

This check was moved to the energy function when the best energy was calculated as zero (i.e. every value matched to the truth table). The last change was to continue the SA process when the SPICE checker didn't fall into a preset acceptable range for each memristor state. This produced multiple simulations resulting in the selection of memristive crossbar designs that might perform more efficiently with less potential for error.

In the tests using NuSMV, it was discovered that better results were achieved when using a full 3-bit truth table containing $3^{3}=27$ rows. When implementing SA, any truth table entry where $C_{\text {in }}=2$ was excluded from the energy calculation, so those items would not appear in the final solution. Figure 6.1 is an example of memristor placement for a $4 x 4$ ternary adder with no carry-out bit.

Once the design in Figure 6.1 was found, it was tested using SPICE. The SPICE column shows the simulated voltage of the last row given the inputs $A, B$, and $C_{i n}$. The resistance and voltage ranges were varied in testing. The SPICE results can be clearly mapped to the Sum bit in Table 6.3.

Table 6.3: SPICE: Ternary Adder $4 \times 4$ with no carry-out bit

| $A$ | $B$ | $C_{\text {in }}$ | Sum | SPICE |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0.000080 |
| 0 | 0 | 1 | 1 | 0.007803 |
| 0 | 1 | 0 | 1 | 0.007828 |
| 0 | 2 | 0 | 2 | 0.200024 |
| 0 | 2 | 1 | 0 | 0.000060 |
| 0 | 1 | 1 | 2 | 0.200339 |
| 1 | 0 | 0 | 1 | 0.007862 |
| 1 | 0 | 1 | 2 | 0.201286 |
| 1 | 1 | 0 | 2 | 0.200028 |
| 1 | 2 | 0 | 0 | 0.000060 |
| 1 | 1 | 1 | 0 | 0.000080 |
| 1 | 2 | 1 | 1 | 0.007823 |
| 2 | 0 | 0 | 2 | 0.200027 |
| 2 | 0 | 1 | 0 | 0.000080 |
| 2 | 1 | 0 | 0 | 0.000060 |
| 2 | 2 | 0 | 1 | 0.007816 |
| 2 | 1 | 1 | 1 | 0.007823 |
| 2 | 2 | 1 | 2 | 0.200336 |



Figure 6.1: Crossbar design for $4 \times 4$ ternary adder with no carry-out bit

To implement the carry-out value, the size of the crossbar array was increased to $6 x 5$ and SA energy function was modified to check for the carry-out bit. The last row is checked for the Sum bit while the next to the last row is checked for $C_{\text {out }}$. Figure 6.2 is the design of a crossbar for a 1-bit adder. It is observed that the SPICE results are within acceptable ranges.

Ternary Full-Adder

Based on the success of the initial tests, the next step was to create a ternary full-adder using the entire ternary truth table as shown in 5.4 and defined in Algorithm 2.


Figure 6.2: Crossbar design for 1-bit ternary addition

```
Algorithm 2 Ternary Adder
        global values
            \(a[] \leftarrow[0,0,0,0,0,0,0,0,0,1,1,1,1,1,1,1,1,1,2,2,2,2,2,2,2,2,2] \quad \triangleright\) ternary truth table
            \(b[] \leftarrow[0,0,1,0,2,1,2,1,2,0,0,1,0,2,1,1,2,2,0,0,1,0,2,1,1,2,2]\)
            \(c[] \leftarrow[0,1,0,2,0,2,1,1,2,0,1,0,2,0,1,2,1,2,0,1,0,2,0,1,2,1,2]\)
            \(u[] \leftarrow[0,0,0,0,0,1,1,0,1,0,0,0,1,1,1,1,1,1,0,1,1,1,1,1,1,1,2]\)
            \(s[] \leftarrow[0,1,1,2,2,0,0,2,1,1,2,2,0,0,0,1,1,2,2,0,0,1,1,1,2,2,0]\)
    end global values
```

The MAIN procedure in Algorithm 3 calculates an initial array of memristors that is passed to the GSL Simulated Annealing (SA) function. If the SA function is returned to MAIN, a suboptimal solution was found. The best solution is checked with a SPICE sub-circuit and the output is reported.

```
Algorithm 3 Ternary Adder- Main
    procedure MAIN
            Call random number generator with seed value
            for \(i \leftarrow 1\) to rows do
            for \(j \leftarrow 1\) to cols do
                \(k \leftarrow\) rand \(\bmod\) dict
                if \(k \geq\) static then
                        mem \(\leftarrow(k-\) static \()\) mod inputs
                    else
                \(k=k \bmod\) bits
                    initstate \([i *\) cols \(+j]=k\)
            Call gsl_siman_solve
\(\triangleright\) GSL Simulated Annealing
            Call OutputReport
            Call OutputCheck \(\quad\) SPICE Simulation
            return "Suboptimal solution found"
```

The SA energy callback function in Algorithm 4 is used to check whether an input array of memristors results in a valid circuit. The function iterates through each row on the truth table using flow-based analysis based on the value of each memristor. The resulting Sum and $C_{\text {out }}$ values are determined by the resulting value on the last two rows. A best energy value between 0 and 1 is returned to the calling function. The value is determined by the formula currentenergy $\leftarrow$ mismatchcount / (matchcount + mismatchcount. A value of currentenergy $=0$ represents a potential circuit which is then checked using a SPICE sub-circuit based on the values of each memristor.

The memristor array that is passed into the energy function is split into two parts, an input variable (ex. $A, B, C_{i n}$ and truth table position between 0 and 26. Static values of 0,1 , and 2 are also possible. The state of the memristor is determined by the truth table value

```
Algorithm 4 Ternary Adder- GSL SA Energy Function
    procedure EnERGY(*memrs)
        for \(k \leftarrow 1\) to inputs do
            Calculate the value of each memristor based on truth table entry \(k\)
            for each \(i=\) row and \(j=\) column do
                \(m \leftarrow\) Memristor variable \(A\), \(B\), or \(C_{i n}\)
                \(s e l \leftarrow\) Truth table value at row \(k\) for variable \(m\)
                val \(\leftarrow\) Numeric part of memristor
                \(x b a r[i, j] \leftarrow\) Value of truth table entry sel, val
                row \([1] \leftarrow\) memon \(\quad \triangleright\) simulate voltage applied to first row
                    loop while changes are occurring to circuit
                        for each \(i=\) row and \(j=\) column do
                    Estimate current for each row and column using flow-based analysis
                    If memristor state is mem_hi, the highest row/column current is used
                    If memristor state is mem_mid, the maximum current is mem_mid
                    If memristor state is mem_low, no changes are made to current at crossbar
            currentenergy \(\leftarrow\) mismatchcount / (matchcount + mismatchcount)
            if currentenergy \(:=0\) then
            Call OutputReport
            mismatchcount \(\leftarrow\) Call OutputCheck \(\triangleright\) SPICE Simulation
            if mismatchcount \(:=0\) then
                EXIT success
            Return currentenergy
```

The following are successful tests of a ternary adder using resistances of mem_hi $=1 e+06$, mem_mid $=10000$, and mem_low $=100$. The voltage midpoints of 0.003 and 0.25 were determined by finding the average delta between each desired result.

Table 6.4: SPICE Simulation: 7x7 Crossbar

| ABC-in | Sum | SPICE-Sum | C-out | SPICE-C-out |
| :---: | :---: | :--- | :---: | :--- |
| 000 | 0 | 0.001685 | 0 | 0.000230 |
| 001 | 1 | 0.201411 | 0 | 0.000321 |
| 010 | 1 | 0.200552 | 0 | 0.000635 |
| 002 | 2 | 0.024407 | 0 | 0.000295 |
| 020 | 2 | 0.010261 | 0 | 0.001198 |
| 012 | 0 | 0.001970 | 1 | 0.333405 |
| 021 | 0 | 0.001061 | 1 | 0.250621 |
| 011 | 2 | 0.010826 | 0 | 0.000882 |
| 022 | 1 | 0.080744 | 1 | 0.156926 |
| 100 | 1 | 0.200174 | 0 | 0.000603 |
| 101 | 2 | 0.019225 | 0 | 0.000418 |
| 110 | 2 | 0.010313 | 0 | 0.001077 |
| 102 | 0 | 0.001059 | 1 | 0.147804 |
| 120 | 0 | 0.001115 | 1 | 0.199882 |
| 111 | 0 | 0.001273 | 1 | 0.199771 |
| 112 | 1 | 0.102247 | 1 | 0.303276 |
| 121 | 1 | 0.150553 | 1 | 0.150228 |
| 122 | 2 | 0.005414 | 1 | 0.168877 |
| 200 | 2 | 0.019054 | 0 | 0.000645 |
| 201 | 0 | 0.001162 | 1 | 0.333047 |
| 210 | 0 | 0.001173 | 1 | 0.199828 |
| 202 | 1 | 0.028480 | 1 | 0.124897 |
| 220 | 1 | 0.103217 | 1 | 0.057031 |
| 211 | 1 | 0.044280 | 1 | 0.304228 |
| 212 | 2 | 0.003602 | 1 | 0.333539 |
| 221 | 2 | 0.009425 | 1 | 0.358887 |
| 222 | 0 | 0.000304 | 2 | 0.011560 |



Figure 6.3: Ternary Adder: 7x7 Crossbar


Figure 6.4: SPICE Results: 7x7 Ternary Full Adder- SUM


Figure 6.5: SPICE Results: 7x7 Ternary Full Adder- Carry

In the next 8 x 8 crossbar test, the values for 1 and 2 were flipped. In this experiment, $1=$ mem_mid and $2=$ mem_hi. Since $C_{\text {out }}$ only has one possible result equal to 2 , it was thought that this might result in a larger tolerance between resulting values. The resistances used were mem_hi $=1 e+06$, mem_mid $=10000$, and mem_low $=100$ and the voltage midpoints were 0.00275 and 0.080 .

Table 6.5: Ternary Adder: 8x8 Crossbar (012)

| ABC-in | Sum | SPICE-Sum | C-out | SPICE-C-out |
| :---: | :---: | :--- | :---: | :--- |
| 000 | 0 | 0.001170 | 0 | 0.000211 |
| 001 | 1 | 0.010416 | 0 | 0.000360 |
| 010 | 1 | 0.010204 | 0 | 0.000454 |
| 002 | 2 | 0.199779 | 0 | 0.001312 |
| 020 | 2 | 0.201326 | 0 | 0.001311 |
| 012 | 0 | 0.001268 | 1 | 0.004581 |
| 021 | 0 | 0.001674 | 1 | 0.003733 |
| 011 | 2 | 0.199869 | 0 | 0.001366 |
| 022 | 1 | 0.010280 | 1 | 0.016106 |
| 100 | 1 | 0.020008 | 0 | 0.000339 |
| 101 | 2 | 0.204043 | 0 | 0.000869 |
| 110 | 2 | 0.201852 | 0 | 0.001334 |
| 102 | 0 | 0.001887 | 1 | 0.005810 |
| 120 | 0 | 0.001891 | 1 | 0.003381 |
| 111 | 0 | 0.002170 | 1 | 0.005224 |
| 112 | 1 | 0.020407 | 1 | 0.005617 |
| 121 | 1 | 0.020719 | 1 | 0.005096 |
| 122 | 2 | 0.200488 | 1 | 0.007451 |
| 200 | 2 | 0.144609 | 0 | 0.001170 |
| 201 | 0 | 0.001561 | 1 | 0.010075 |
| 210 | 0 | 0.001930 | 1 | 0.019376 |
| 202 | 1 | 0.011755 | 1 | 0.017998 |
| 220 | 1 | 0.010883 | 1 | 0.010872 |
| 211 | 1 | 0.004618 | 1 | 0.007216 |
| 212 | 2 | 0.137306 | 1 | 0.024590 |
| 221 | 2 | 0.196218 | 1 | 0.023525 |
| 222 | 0 | 0.001506 | 2 | 0.236781 |



Figure 6.6: Ternary Adder: 8x8 Crossbar (012)


Figure 6.7: SPICE Results: 8x8 Ternary Full Adder (012)- SUM


Figure 6.8: SPICE Results: $8 \times 8$ Ternary Full Adder (012)- Carry

In the following $8 \times 8$ crossbar test, the values for 1 and 2 were set back to the original $2=$ mem_mid and $1=$ mem_hi. The resistances used were mem_hi $=1 e+06$, mem_mid $=10000$, and mem_low $=100$ and the voltage midpoints were 0.003 and 0.038 .

Table 6.6: Ternary Adder: 8x8 Crossbar (021)

| ABC-in | Sum | SPICE-Sum | C-out | SPICE-C-out |
| :---: | :---: | :--- | :---: | :--- |
| 000 | 0 | 0.001290 | 0 | 0.000103 |
| 001 | 1 | 0.200937 | 0 | 0.000295 |
| 010 | 1 | 0.145380 | 0 | 0.000260 |
| 002 | 2 | 0.011023 | 0 | 0.001084 |
| 020 | 2 | 0.019923 | 0 | 0.000427 |
| 012 | 0 | 0.001611 | 1 | 0.208595 |
| 021 | 0 | 0.001318 | 1 | 0.333050 |
| 011 | 2 | 0.011687 | 0 | 0.000300 |
| 022 | 1 | 0.144542 | 1 | 0.156188 |
| 100 | 1 | 0.115627 | 0 | 0.000822 |
| 101 | 2 | 0.011205 | 0 | 0.000605 |
| 110 | 2 | 0.005918 | 0 | 0.001531 |
| 102 | 0 | 0.001114 | 1 | 0.290627 |
| 120 | 0 | 0.001884 | 1 | 0.200131 |
| 111 | 0 | 0.001762 | 1 | 0.199599 |
| 112 | 1 | 0.143393 | 1 | 0.242766 |
| 121 | 1 | 0.045034 | 1 | 0.303812 |
| 122 | 2 | 0.004108 | 1 | 0.245840 |
| 200 | 2 | 0.005601 | 0 | 0.000756 |
| 201 | 0 | 0.001806 | 1 | 0.144311 |
| 210 | 0 | 0.001360 | 1 | 0.143828 |
| 202 | 1 | 0.145150 | 1 | 0.150409 |
| 220 | 1 | 0.097161 | 1 | 0.161720 |
| 211 | 1 | 0.099343 | 1 | 0.160879 |
| 212 | 2 | 0.004573 | 1 | 0.259182 |
| 221 | 2 | 0.006581 | 1 | 0.329366 |
| 222 | 0 | 0.000130 | 2 | 0.029392 |



Figure 6.9: Ternary Adder: 8x8 Crossbar (021)


Figure 6.10: SPICE Results: 8x8 Ternary Full Adder (021)- SUM


Figure 6.11: SPICE Results: 8x8 Ternary Full Adder (021)- Carry

## Ternary Multiplication

The methodology to create a ternary adder can be applied to other operations, including multiplication. A ternary multiplier was designed using truth table 6.7 defined in Algorithm 5.

```
Algorithm 5 Ternary Multiplication
    global values
        \(\begin{array}{rlr}a[] & \leftarrow[0,0,0,0,0,0,0,0,0,1,1,1,1,1,1,1,1,1,2,2,2,2,2,2,2,2,2] & \triangleright \text { ternary truth table } \\ b[] & \leftarrow[0,0,1,0,2,1,2,1,2,0,0,1,0,2,1,1,2,2,0,0,1,0,2,1,1,2,2] & \\ c] & \leftarrow[0,1,0,2,0,2,1,1,2,0,1,0,2,0,1,2,1,2,0,1,0,2,0,1,2,1,2] & \\ v[] & \leftarrow[0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,1,1,1,0,0,0,0,1,1,1,1,2] & \\ p[] & \leftarrow[0,1,0,2,0,2,1,1,2,0,1,1,2,2,2,0,0,1,0,1,2,2,1,0,1,2,0] & \\ \text { end global values } & \end{array}\)
```

Table 6.7: Truth Table: Ternary Multiplication

| $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Product |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 1 |
| 0 | 1 | 0 | 0 | 0 |
| 0 | 0 | 2 | 0 | 2 |
| 0 | 2 | 0 | 0 | 0 |
| 0 | 1 | 2 | 0 | 2 |
| 0 | 2 | 1 | 0 | 1 |
| 0 | 1 | 1 | 0 | 1 |
| 0 | 2 | 2 | 0 | 2 |
| 1 | 0 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 | 1 |
| 1 | 1 | 0 | 0 | 1 |
| 1 | 0 | 2 | 0 | 2 |
| 1 | 2 | 0 | 0 | 2 |


| $A$ | $B$ | $C_{\text {in }}$ | $C_{\text {out }}$ | Product |
| :---: | :---: | :---: | :---: | :---: |
| 1 | 1 | 1 | 0 | 2 |
| 1 | 1 | 2 | 1 | 0 |
| 1 | 2 | 1 | 1 | 0 |
| 1 | 2 | 2 | 1 | 1 |
| 2 | 0 | 0 | 0 | 0 |
| 2 | 0 | 1 | 0 | 1 |
| 2 | 1 | 0 | 0 | 2 |
| 2 | 0 | 2 | 0 | 2 |
| 2 | 2 | 0 | 1 | 1 |
| 2 | 1 | 1 | 1 | 0 |
| 2 | 1 | 2 | 1 | 1 |
| 2 | 2 | 1 | 1 | 2 |
| 2 | 2 | 2 | 2 | 0 |

The same tests were run for multiplication as for ternary addition. In the 7 x 7 crossbar test, the resistances used were mem_hi $=1 e+06$, mem7_mid $=10000$, and mem_low $=100$ and the voltage midpoints were 0.005 and 0.073 .

Table 6.8: Ternary Multiplication: 7x7 Crossbar

| ABC-in | Prod | SPICE-Prod | C-out | SPICE-C-out |
| :---: | :---: | :--- | :---: | :--- |
| 000 | 0 | 0.000441 | 0 | 0.000257 |
| 001 | 1 | 0.375242 | 0 | 0.000342 |
| 010 | 0 | 0.001200 | 0 | 0.000834 |
| 002 | 2 | 0.010409 | 0 | 0.000011 |
| 020 | 0 | 0.001247 | 0 | 0.001006 |
| 012 | 2 | 0.015699 | 0 | 0.000645 |
| 021 | 1 | 0.334094 | 0 | 0.000723 |
| 011 | 1 | 0.333481 | 0 | 0.000613 |
| 022 | 2 | 0.011033 | 0 | 0.000319 |
| 100 | 0 | 0.000432 | 0 | 0.000268 |
| 101 | 1 | 0.252281 | 0 | 0.000380 |
| 110 | 1 | 0.200764 | 0 | 0.000976 |
| 102 | 2 | 0.010277 | 0 | 0.000016 |
| 120 | 2 | 0.027418 | 0 | 0.000849 |
| 111 | 2 | 0.010495 | 0 | 0.001033 |
| 112 | 0 | 0.001388 | 1 | 0.210561 |
| 121 | 0 | 0.001638 | 1 | 0.199815 |
| 122 | 1 | 0.109146 | 1 | 0.122912 |
| 200 | 0 | 0.000653 | 0 | 0.000511 |
| 201 | 1 | 0.258248 | 0 | 0.000463 |
| 210 | 2 | 0.010434 | 0 | 0.000798 |
| 202 | 2 | 0.010663 | 0 | 0.000109 |
| 220 | 1 | 0.139452 | 1 | 0.136182 |
| 211 | 0 | 0.001248 | 1 | 0.200110 |
| 212 | 1 | 0.166118 | 1 | 0.087771 |
| 221 | 2 | 0.006807 | 1 | 0.207492 |
| 222 | 0 | 0.000136 | 2 | 0.019944 |



Figure 6.12: Ternary Multiplication: 7x7 Crossbar


Figure 6.13: SPICE Results: 7x7 Ternary Multiplication- SUM


Figure 6.14: SPICE Results: 7x7 Ternary Multiplication- Carry

In the 8 x 8 crossbar test for multiplication, the values for 1 and 2 were again flipped. $C_{\text {out }}$ only has one possible result equal to 2 . The resistances used were mem_hi $=1 e+06$, mem_mid $=10000$, and mem_low $=100$ and the voltage midpoints were 0.0029 and 0.080 .

Table 6.9: Ternary Multiplication: 8x8 Crossbar (012)

| ABC-in | Prod | SPICE-Prod | C-out | SPICE-C-out |
| :---: | :---: | :--- | :---: | :--- |
| 000 | 0 | 0.000544 | 0 | 0.000252 |
| 001 | 1 | 0.019782 | 0 | 0.000038 |
| 010 | 0 | 0.000594 | 0 | 0.000204 |
| 002 | 2 | 0.259399 | 0 | 0.000467 |
| 020 | 0 | 0.000583 | 0 | 0.000215 |
| 012 | 2 | 0.178732 | 0 | 0.000906 |
| 021 | 1 | 0.015880 | 0 | 0.000234 |
| 011 | 1 | 0.015014 | 0 | 0.000220 |
| 022 | 2 | 0.157622 | 0 | 0.000881 |
| 100 | 0 | 0.001189 | 0 | 0.000801 |
| 101 | 1 | 0.020779 | 0 | 0.000221 |
| 110 | 1 | 0.019892 | 0 | 0.001898 |
| 102 | 2 | 0.259852 | 0 | 0.000611 |
| 120 | 2 | 0.149120 | 0 | 0.001652 |
| 111 | 2 | 0.200641 | 0 | 0.001424 |
| 112 | 0 | 0.001164 | 1 | 0.004803 |
| 121 | 0 | 0.001601 | 1 | 0.006089 |
| 122 | 1 | 0.004848 | 1 | 0.013958 |
| 200 | 0 | 0.001017 | 0 | 0.000378 |
| 201 | 1 | 0.016866 | 0 | 0.000174 |
| 210 | 2 | 0.153724 | 0 | 0.001317 |
| 202 | 2 | 0.209636 | 0 | 0.000544 |
| 220 | 1 | 0.019433 | 1 | 0.006247 |
| 211 | 0 | 0.001861 | 1 | 0.003981 |
| 212 | 1 | 0.007555 | 1 | 0.010164 |
| 221 | 2 | 0.143934 | 1 | 0.006056 |
| 222 | 0 | 0.001134 | 2 | 0.212312 |



Figure 6.15: Ternary Multiplication: 8x8 Crossbar (012)


Figure 6.16: SPICE Results: 8x8 Ternary Multiplication (012)- SUM


Figure 6.17: SPICE Results: 8x8 Ternary Multiplication (012)- Carry

In the following $8 \times 8$ crossbar test, the values for 1 and 2 were set back to the original $2=$ mem_mid and $1=$ mem_hi. The resistances used were mem_hi $=1 e+06$, mem_mid $=10000$, and mem_low $=100$ and the voltage midpoints were 0.0048 and 0.0248 .

Table 6.10: Ternary Multiplication: 8x8 Crossbar (021)

| ABC-in | Prod | SPICE-Prod | C-out | SPICE-C-out |
| :---: | :---: | :--- | :---: | :--- |
| 000 | 0 | 0.000797 | 0 | 0.000001 |
| 001 | 1 | 0.437246 | 0 | 0.000536 |
| 010 | 0 | 0.001138 | 0 | 0.000231 |
| 002 | 2 | 0.010367 | 0 | 0.000018 |
| 020 | 0 | 0.001192 | 0 | 0.000203 |
| 012 | 2 | 0.010767 | 0 | 0.000225 |
| 021 | 1 | 0.369672 | 0 | 0.000936 |
| 011 | 1 | 0.334825 | 0 | 0.000884 |
| 022 | 2 | 0.010741 | 0 | 0.000227 |
| 100 | 0 | 0.000796 | 0 | 0.000002 |
| 101 | 1 | 0.213973 | 0 | 0.000590 |
| 110 | 1 | 0.112541 | 0 | 0.000989 |
| 102 | 2 | 0.007651 | 0 | 0.000208 |
| 120 | 2 | 0.010903 | 0 | 0.001311 |
| 111 | 2 | 0.011178 | 0 | 0.001192 |
| 112 | 0 | 0.000450 | 1 | 0.115094 |
| 121 | 0 | 0.002129 | 1 | 0.200214 |
| 122 | 1 | 0.102857 | 1 | 0.078814 |
| 200 | 0 | 0.000797 | 0 | 0.000001 |
| 201 | 1 | 0.331818 | 0 | 0.000551 |
| 210 | 2 | 0.015180 | 0 | 0.001286 |
| 202 | 2 | 0.010164 | 0 | 0.000029 |
| 220 | 1 | 0.180130 | 1 | 0.026985 |
| 211 | 0 | 0.001965 | 1 | 0.200338 |
| 212 | 1 | 0.103446 | 1 | 0.078853 |
| 221 | 2 | 0.011687 | 1 | 0.249980 |
| 222 | 0 | 0.000581 | 2 | 0.019489 |



Figure 6.18: Ternary Multiplication: 8x8 Crossbar (021)


Figure 6.19: SPICE Results: 8x8 Ternary Multiplication (021)- SUM


Figure 6.20: SPICE Results: 8x8 Ternary Multiplication (021)- Carry

In the previous experiments, acceptable range limits were calculated based on the average SPICE values of valid flow-based designs. For the following test, the algorithm was altered to include fixed ranges and multiple SPICE tests to reduce outliers and produce tighter ranges.

This 8 x 8 ternary adder used the values $1=$ mem_mid and $2=$ mem_hi. The resistances used were mem_hi $=1 e+06$, mem_mid $=10000$, and mem_low $=100$ and the voltage midpoints were 0.0021 and 0.0805 .

Table 6.11: Enhanced Ternary Addition: 8x8 Crossbar (012)

| ABC-in | Prod | SPICE-Prod | C-out | SPICE-C-out |
| :---: | :---: | :--- | :---: | :--- |
| 000 | 0 | 0.000182 | 0 | 0.000066 |
| 001 | 1 | 0.007695 | 0 | 0.000066 |
| 010 | 1 | 0.007841 | 0 | 0.000031 |
| 002 | 2 | 0.144499 | 0 | 0.000098 |
| 020 | 2 | 0.200315 | 0 | 0.000174 |
| 012 | 0 | 0.000170 | 1 | 0.007994 |
| 021 | 0 | 0.000155 | 1 | 0.004060 |
| 011 | 2 | 0.200034 | 0 | 0.000105 |
| 022 | 1 | 0.007898 | 1 | 0.011768 |
| 100 | 1 | 0.007810 | 0 | 0.000069 |
| 101 | 2 | 0.202304 | 0 | 0.000067 |
| 110 | 2 | 0.199991 | 0 | 0.000109 |
| 102 | 0 | 0.000168 | 1 | 0.007739 |
| 120 | 0 | 0.000161 | 1 | 0.007161 |
| 111 | 0 | 0.000189 | 1 | 0.004050 |
| 112 | 1 | 0.007804 | 1 | 0.011854 |
| 121 | 1 | 0.007907 | 1 | 0.011204 |
| 122 | 2 | 0.207165 | 1 | 0.011880 |
| 200 | 2 | 0.145351 | 0 | 0.000089 |
| 201 | 0 | 0.000171 | 1 | 0.005780 |
| 210 | 0 | 0.000118 | 1 | 0.007887 |
| 202 | 1 | 0.013365 | 1 | 0.032476 |
| 220 | 1 | 0.014953 | 1 | 0.015480 |
| 211 | 1 | 0.007874 | 1 | 0.007903 |
| 212 | 2 | 0.197981 | 1 | 0.016423 |
| 221 | 2 | 0.210937 | 1 | 0.008040 |
| 222 | 0 | 0.000144 | 2 | 0.238636 |



Figure 6.21: Enhanced Ternary Adder: 8x8 Crossbar (012)


Figure 6.22: SPICE Results: 8x8 Enhanced Ternary Adder (012)- SUM


Figure 6.23: SPICE Results: 8x8 Enhanced Ternary Adder (012)- Carry

## CHAPTER 7: CONCLUSION AND FUTURE WORK

Some goals for future development might include producing circuit designs with a tighter range of resistances, narrower voltage thresholds, and reducing the occurrence of strays or outliers. Further modeling of the internal state variable needs to be done prior to fabrication.

Additional memristive crossbar designs that implement a ternary n-bit adder should be a goal. Full integration of Modified Nodal Analysis to complement SPICE including input, output, wire resistances, and parasitic voltage loss needs to be addressed.

Refinement of the simulated annealing energy function will result in enhanced performance and more accurate results. The final process should continue to run to find a solution where all results are close to the midpoint of each voltage range.

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