

CELL-BASED GaAs MMICS FOR SMALL SATELLITE APPLICATIONS

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Highly integrated, reliable microwave components are needed for small satellite applications. The GaAs integrated circuit industry has been developed largely due to military electronic requirements. In particular, the library cell approach by Pacific Monolithics holds promise for making simple and complex microwave systems available at a cost not prohibitive for small satellite applications. This paper outlines the cell library approach, cells now in the library and some microwave systems built by integrating components from the cell library.

INTRODUCTION

The small satellite concept will extend the demand for well-characterized semicustom electronic components for satellite applications. As with any industry, availability of satellite launch opportunities, and lowered life-cycle costs will dramatically increase the commercial demand for satellites. Commercial applications are now under consideration and development that would have been unthinkable only a few years ago. With the proper nurturing both from the logistics and design aspects, this process will accelerate.

The electronics area is key to the lowered costs necessary for the development of the satellite industry. All satellites rely upon electronic systems to provide the backbone of the operational concept. The widespread availability of PC-based information processing, for instance, has made possible very low-cost satellite housekeeping and information storage systems, as well as ground stations.

Microwave electronics is used for all information processing functions with which the satellite communicates with ground-based equipment, or other satellites. These include command and control telemetry reception, data gathering and data transmission. However, a new approach to electronics is necessary for the new satellite industry.

The traditional method of building satellite electronics is to specify and develop a full custom design for each individual satellite system, with a concentration on ground-

based reliability documentation and testing. Each system is complex and expensive for two reasons. First, it is made of expensive microwave components, and second, each component and subassembly, as well as the entire system, is exhaustively tested for reliability reasons before delivery for launch.

While the traditional approach is sufficient for construction of high cost military satellite systems, a new design philosophy is better suited to the small satellite concept, that is GaAs monolithic circuits

Currently available monolithic GaAs cell-based electronics addresses performance, cost and reliability issues. GaAs monolithic technology has been developed to the point where low-cost circuits can be designed, developed and produced with performance which in many cases exceeds that of the hybrid circuitry it replaces in applications such as the small satellites.

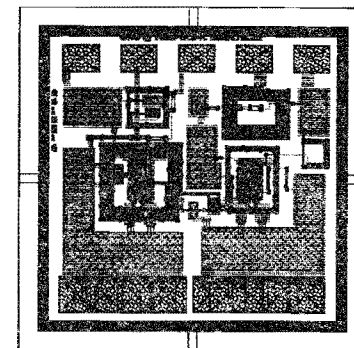
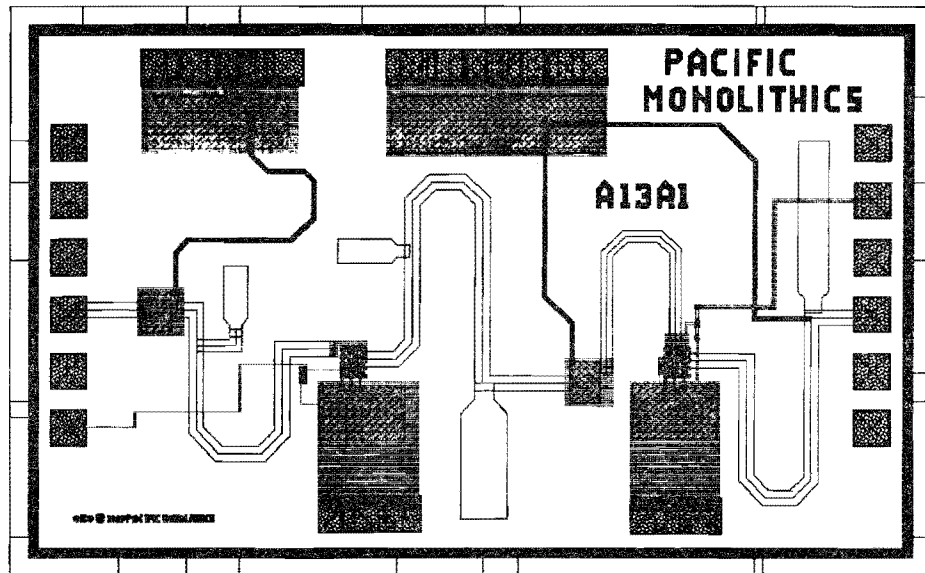
First, the design and recurring costs for highly complex, integrated subsystems is drastically reduced over that for a traditional microwave design. Second, the high integration level, and the high reliability features of the GaAs MMIC architecture cut both the number of test points required (due to the high integration levels possible), and the probable number of failures (due to parts count, high radiation tolerance and high junction temperature tolerance of GaAs MMICs).

GaAs CELL LIBRARY DESIGN

The use of GaAs MMICs facilitates the production of extremely small, highly integrated and reliable microwave electronics. A drawback of MMICs, in full custom design situations, is the substantial NRE required for the chip design. The use of a standard cell library design approach can substantially reduce the front-end cost and design time for satellite microwave electronics.

Background - The Lumped-Element Approach

The major technology advance that made a GaAs MMIC microwave cell library practical was the use of the lumped-element circuit components to replace conventional transmission line circuit components. Lumped element circuits are very compact. Fig. 1 is a comparison of a 2 gain stage amplifier realized in transmission line and in lumped element technology. The distributed design is 1.5 by 2.5 mm, and the lumped design is .9 mm square. A substantial reduction in size results from the use of lumped elements. The comparable lumped element design in this case requires



100 MILS

Figure 1. Distributed and Lumped-Element 2-Stage Amplifier chips

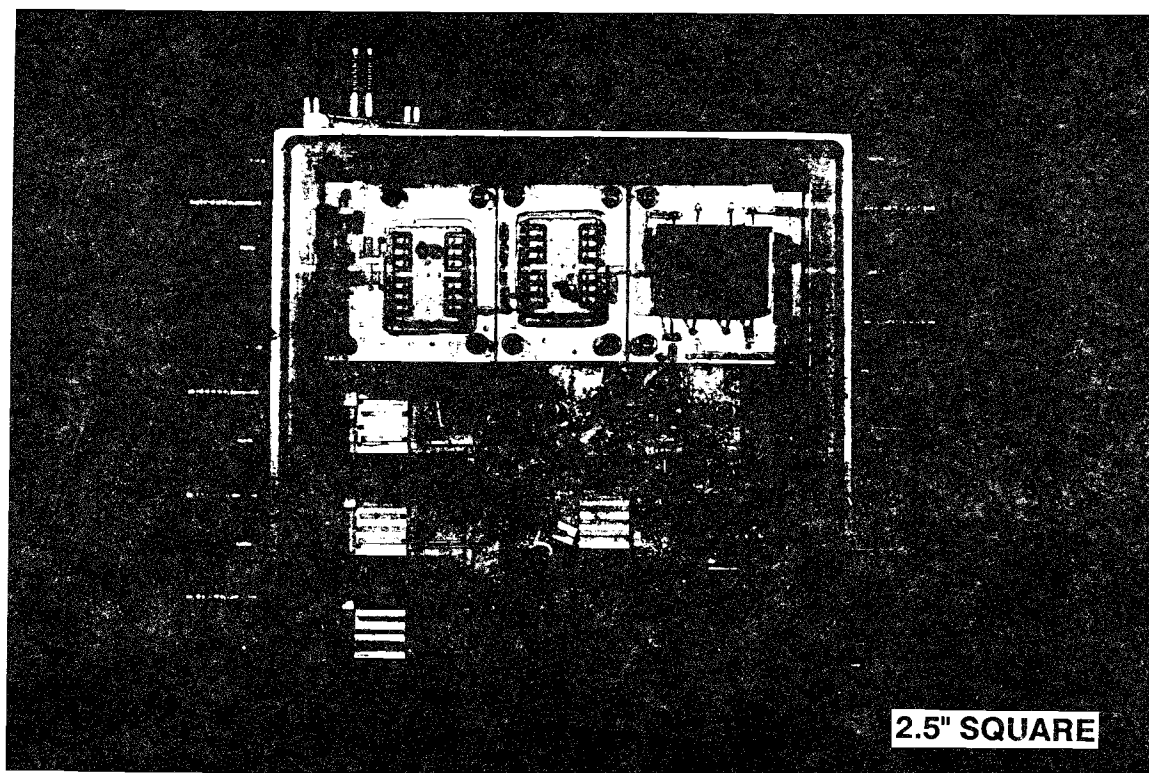
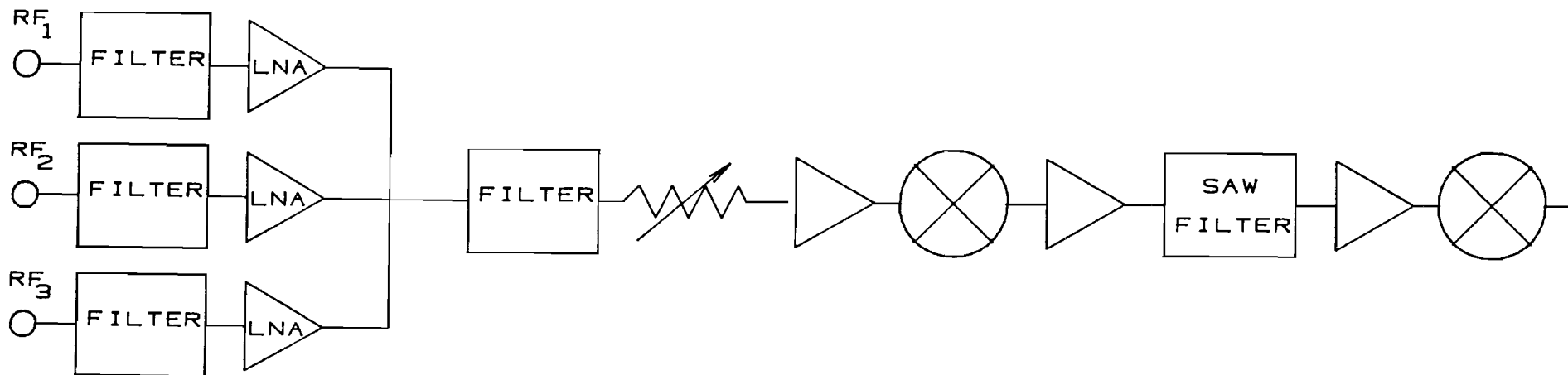


Figure 3. Three-Channel Converter System

approximately 20 percent of the chip area of the distributed design.

The reduction in size permits a microwave cell library approach on a practical basis. It should be noted that a cell library approach uses more real estate than full custom. It is also important to realize that often, in order to obtain optimum microwave performance, active components must be placed very close to each other. By using a lumped element circuit design approach, the real estate inefficiency caused by using a cell library methodology does not cause a significant impact on chip size, and therefore, circuit performance or cost.

Fig.1(a) is a 2 stage amplifier using transmission line interconnects. Fig. 1(b) shows a comparable 2 stage amplifier using lumped element interconnects. The lumped element design is .8 x .8 mm. The lumped element design is 13.5 times smaller.

Microwave Signal Processing Functions - The Cell Families

Table 1 shows the major classes of microwave signal-processing functions. The cell library has been designed to be very flexible in order to address a number of microwave signal processing systems and subsystems such as command and control telemetry, data links, radar transmitters and receivers, EW receivers, communications receivers, GPS receivers, Microwave Landing System receivers, and expendable counter measure systems.

Key elements in these families have been developed and tested, and are used routinely to develop semicustom GaAs MMICs. New cells are constantly being generated. Table II is a listing of the major cells in the PM Library at this time.

GaAs CELL-BASED ELECTRONIC SYSTEMS

A number of microwave electronic systems have been built at Pacific Monolithics using the GaAs cell library for military and satellite applications.

Fig. 2 shows a single-chip FM-CW radar subsystem developed by integrating microwave circuit cells onto a GaAs chip. The circuit includes many gain stages, a mixer, and a voltage controlled oscillator. The oscillator is externally controlled to sweep frequency, resulting in the FM-CW transmitted signal. The power amplifier for the transmitter as well as the receiving amplifier are both on chip. The transmitted frequency is used for the LO for the received

Table 1
MICROWAVE SIGNAL PROCESSING FUNCTIONS

Amplifiers

Low Noise
Power

Converters

Up/Down
Image Reject
High Dynamic Range

Oscillators

SAW
DRO
VCO
Phase Locked Oscillators

Modulators

Amplitude
Frequency
Phase

Special Functions

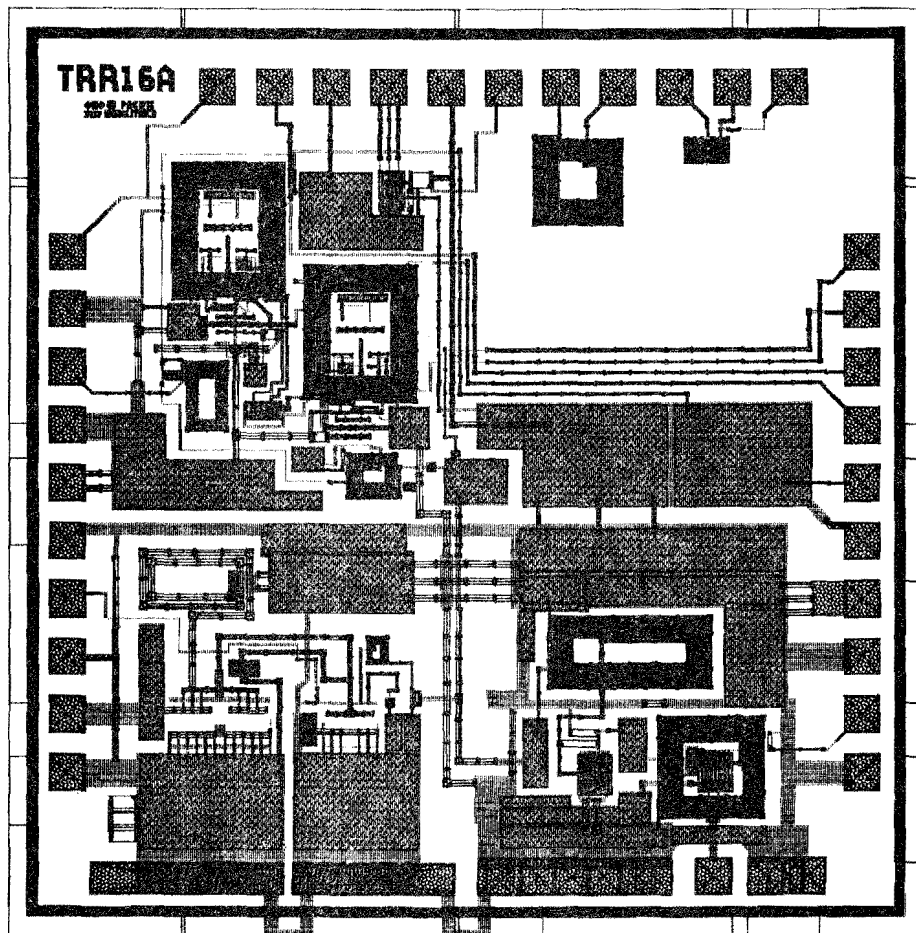
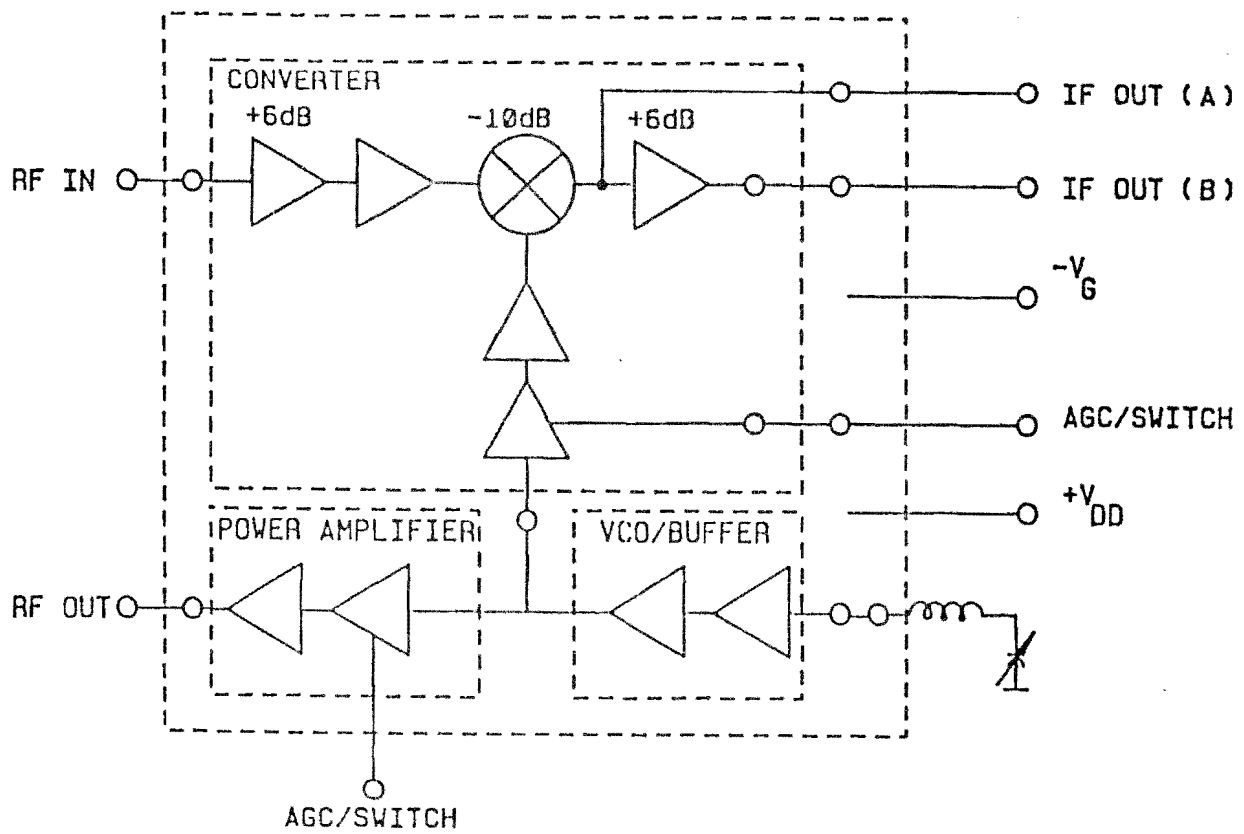
Detectors
Switches
Limiters
Couplers
Phase Shifters

Table 2
GaAs MMIC Cell Library

MMIC Cell #	Freq (GHz)	Gain (dB)	Cell Function	MMIC Cell #	FREQ (GHz)	Gain (dB)	Cell Function
MMIC Amplifiers				Digital IC's			
AG01A	2-6	10	Amplifier, Gain Block	DD01A	DC-1.5	—	Divide by Two
AL18A	2-6	~10	Amplifier, Low Noise	DT01A	DC-3	—	2:1 MUX
AG1001G	2-6	10	Amplifier, Higher Power	DT01A	DC-3	—	2 Input NOR Gate
AG1001A	2-6	10	Amplifier AGC	DT01A	DC-3	—	Inverter
AG10A	2-6	10	Amplifier, Temp Comp	DD01A	DC-1.5	—	Master/Slave
AP04C	3-7	10	Power Amplifier				D-Flip/Flop
AP04A	2-6	10	Power Amplifier	DCJ121A	DC-2	—	Dual Complementary
AG03A	2-6	10	Amplifier, High Yield				Clock Driver
AG1003A	2-6	10	Amplifier, High Yield	DDJ121A	DC-2	—	Complementary
AC161A	2-7	10	Power Amplifier				Divide by Two
AGC151A	2-7	10	Amplifier, Ultra Flat	DDJ162A	DC-1.3	—	Lower Power
AGC152A	2-7	10	Amplifier, Tilt				Divide by Two
AGC153A	2-7	10	Amplifier, Higher Tilt	DDJ163A	DC-1.2	—	Lowest Power
AGC154A	2-7	10	Amplifier, Negative Slope				Divide by Two
AGC155A	2-7	10	Amplifier, Low Current	DP02A	DC-100 MHz	—	7 to 20 Line
AGC156A	2-7	10	Amplifier, Unmatched				Phase Shifter
AL95A	1-6	10-16	Unmatched LNA				Controller
ALC121A	2-8	>10	Low Noise Amplifier				
AG1006A	2-6	20	Amplifier, Univ Gain Block				
Other MMICs				Special Building Blocks			
FDG12A	1-6	6	Active Doubler	AL62B	0.1-8	15	Low Noise Amp
PDG12A	1-4	0	Active Splitter	AGO6A	2-6	20	Limiting Amp
CG141A	+5	10	Converter	AL82C	2-12	8	Amplifier Distributed
CG142A	5	0	Converter	AL82A	2-18	5	Amplifier Distributed
6-18 GHz MMIC Amplifiers				CV0301	1-3	30	Downconverter
AGR12A3	6-18	8±1	Amplifier, Universal Block	CV0201	1-3	25	Downconverter
AGR12A2	5-28	10±1	Amplifier, Broadband	C1070A	7-15	15	Downconverter
AGR12A1	6-18	9±1	Amplifier, Cascadable	CV0801	5-8	25	Downconverter
AGR12A4	6-18	12±2	Amplifier, Cascadable	M971B	0.01-2	25	VHF, UHF Converter
AG1072DR	6-12	6±0.5	Amplifier, Univ Gain Block	CV0601	3-6	40	Downconverter
AG1073A	7-17	9±1	Amplifier, High Efficiency	CV0602	3-6	20	Downconverter
AG1073B	6-18	+ Slope	Amplifier, High Producibility	DD01B	0-2	—	Flip Flop/Latches
AG1073C	7-17	11±2	Amplifier, Low Noise	DT01C	0-3.0	—	Random (glue)
AG1073D	7-17	12±0.5	Amplifier, Low Noise				Logic Cells (NOR, NAND, INVERT, AND-OR)
AG1073E	7-17	9±1	Amplifier, High Efficiency	AL94C	8-12	—	Oscillator
AG1073ER	6-18	9±1	Amplifier, High Efficiency	M80A	0.01-2	-4	UHF, VHF Mixer
AG1073F	7-17	9±1	Amplifier, Matched	M82A	6-12	-10	Diode Mixer
AG1073G	7-17	5±1	Amplifier, Matched	M83A	10-18	-10	Diode Mixer
AG1073I	7-17	6±1	Amplifier, High Gain Match	M84A	12-18	-10	Diode Mixer
AG1073IR	7-18	~6 dB±2	Amplifier	SW1201A	0.3-12	-3	SPDT Switch
AGR12A5	6-18	~8 dB±2	Amplifier	AT1202A	1-12	-3.5	Attenuator
ALF103A	6-12	5 dB±1	Amplifier, High Power	AT	2-18	-3	Attenuator
ALF104A	6-10	7±0.5	Amplifier	HWC	2-10	-4	Power Splitter
AGF14A1	6-16	7±1	Amplifier	HVM	5-10	-6	Vector Modulator
AGF14A2	8-14	8±0.5	Amplifier	P01A	5-10	-6	QPSK Modulator
AGR12A5	6-18	8 Min	Amplifier	H18	2-12	-6	BPSK Modulator
6-18 GHz MMIC Converters				IS01	2-12	-3	Active Isolator
C1071A	7-17	12±4	Converter	B971A	1-3	-2	Active Balun
C1071B	7-17	12±4	Converter	B02A	6-12	-2	Passive Balun
C1071I	7-17	0	Converter	BC01A	2-6	-2	90° Hybrid
CR12A1	10-15	-2	Converter	B03A	3-6	-3	Passive Balun
				P1037A	1-6	-8	4-Bit Generic Phase Shifter

Table 2
GaAs MMIC Cell Library
(cont.)

MMIC Cell #	FREQ (GHz)	Gain (dB)	Cell Function	MMIC Cell #	FREQ (GHz)	Gain (dB)	Cell Function
Low Frequency MMIC Amplifiers				MMIC Control Elements & Phase Shifters (Cont'd)			
PM06				T01A	.5-12	-3	Single Ended Attenuator
AI01A	.2-2	20	Low Noise Gain Block	T01C	.5-12	-3	Single Ended Attenuator
AG971A	.03-1	20	Amplifier, Gain Block	SL124A	.5-12	2	Single Ended Switch
AG971B	.03-1.5	26	Amplifier, Gain Block	SL125A	.5-12	-3	Single Ended Switch
AG971C	.03-1.5	40	Amplifier, Gain Block	TA124A	.5-12	2	Single Ended Attenuator
AL971A	.03-1	11	Amplifier, Gain Block	TA125A	.5-12	.	Single Ended Attenuator
AL971B	.9-3	20	Amplifier, Gain Block	Other MMICs			
AL971C	0.3-1.1	15	Amplifier, Gain Block	HDG	6-10	-4-20	Seg. Dual Gate FET Attenuator
AF971A	.3-1	25	Amplifier, Gain Block	HDV	6-10	-4	Seg. Dual Gate Vector Modulator
AF971B	.5-1	25	Amplifier, Gain Block	AGS122A	2-6	11	Balanced, Amplifier
AGB121A	.02-1	20	Low Noise Amplifier	AGS162A	2-6	11	Balanced, Amplifier
AGB122A	.02-2	29	Low Noise Amplifier	AGS121A	1-3	10	Balanced, Amplifier
ALB121A	.9-2.5	27	Low Noise Amplifier	AGS123A	3-9	10	Balanced, Amplifier
APB121A	.3-3	25	Low Noise Amplifier	AGS124A	6-12	8	Balanced, Amplified
APB122A	.3-1.1	25	Low Noise Amplifier	AIP121E	.05-2	7	IF Amplifier
ALB141A	.9-2.5	22	Low Noise Amplifier	MMIC Oscillator			
AGB141A	.05-2	7	Amplifier/Attenuator	V971A	.05-2	—	Oscillator, for Use w/SAW
AGB142A	.05-2	22	Amplifier/Switch	V971B	.05-2	—	Oscillator/Low Phase Noise
ZB141A	dc-.1		Operational Amplifier	GA9	3.5-6	—	Oscillator, Negative Bias
APB151A	.1-2	19	Power Amplifier	GA13S10	3.5-6	—	Oscillator, Negative Bias
AGB161A	.03-2	16	Amplifier	V81A	16-18	—	Oscillator, Balanced
AGB161A	.95-1.05	20	Amplifier	V02A	3-6	—	Oscillator, Balanced GA9
ALB161A	.03-1.1	21	Amplifier	VH101B	3-6	—	Oscillator, No Neg Bias
Low Frequency MMIC Mixers				VH102A	2-3	—	Oscillator-R, Low AM Noise
PM08A				VH103A	6-12	—	Oscillator
M81A	200 X 0.5 μM FET QUAD		Mixer QUAD	VH103B	12-18	—	Oscillator Idss
M971A	.01-1	25	Mixer/Amplifier	VH120A	2-4	—	Oscillator/VCO Choke
M971B	.01-2	25	Mixer/Amplifier	VH122A	11-18	—	Oscillator/Power Detector
MB121AM	.01-1.2	11	Mixer/Amplifier	VH120B	2-4	—	Oscillator/Common Drain
MB122AM	0.1-1.8	20	Mixer/Gilbert Cell	VH1402	1-2	—	Oscillator/Common Drain
MB151A	.01-2	14	Mixer/Gilbert Cell	VH1404	2-4	—	Oscillator/Common Drain
MB161A	.01-2	12	Mixer/Gilbert Cell	VH1408	4-8	—	Oscillator, Low AM Noise
MB162A	.01-1.8	20	Mixer/Gilbert Cell	VH1412	6-12	—	Oscillator, Low AM Noise
Low Frequency MMIC Components				VH1418	9-18	—	Oscillator, Low AM Noise
V971A	.2-2	11	Oscillator/Limiter	VBM1418	12-18	—	Oscillator, Balanced
V971B	.2-2	11	Oscillator/AGC Amp	MMIC Converter/Mixer			
1/2 of ALB161A	.2-2	16	Oscillator	C01A	.8-3	—	Converter, Balanced
PSB141A	.1-8	0	Phase Splitter, 90°	C02A	1.2-2.6	—	Converter, Balanced
B971A	.03-1	0	Phase Splitter, 180°	C03A	1.4-2.8	—	Converter, Single-Ended
ZB144A	.02-1.5	0	Phase Splitter, 180°	C04A	3-6	—	Converter, Balanced
ZB141A	.03-2	-6	Buffer Amplifier	CR05A	3-6	—	Converter, LNB
ZB164A	.03-4	0	Buffer Amplifier	CI06A	3-5	—	Converter
ZB143A	.001-6	-1	Switch	CV07A	3-5	—	Converter, Dual "HV Chip"
ZB165A	.001-6	-1	Switch, Ultra Flat	C11A	5-8	—	Converter, Balanced
MMIC Control Elements & Phase Shifters				CI06C	3-5	—	Image Reject Mixer
H90	6-10	-4	90° Phase Shifter	CQ08B	2-4	—	QPSK Demodulator
H190	2-18	-3-4	180° Phase Shifter	CQ09B	2-4	—	QPSK Demodulator
HAT	2-12	-1.5	Push-Pull Attenuator	CQ10B	4-5	—	QPSK Demodulator
HSW	2-12	-2.5	Push-Pull Switch	C12A	5-8	—	High Efficiency
H2B	6-10	-10	2-bit 0 Shifter	MH126B	2-6	—	4 Diode Mixer
P01A	6-10	-10	Modified 2-Bit 0 Shifter	CIH140A	3-4	—	Hi Dynamic Range IFM
P02A	6-10	-4	Modified 90° 0 Shifter				
S01A	.5-12	-2.5-3	Single Ended Switch				



(100 mils, edge-to-edge)

Figure 2. FM-CW Radar Chip