

1893. Ground-penetrating radar time-frequency analysis method based on synchrosqueezing wavelet transformation

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Abstract. Compared with conventional time-frequency analysis method, synchrosqueezing wavelet transformation (SST) exhibits high resolution capability and good application effect. In this study, SST is introduced to ground-penetrating radar (GPR) processing. This method is applied to analyze a continuous electromagnetic signal. SST can obtain a higher resolution and a better processing effect than conventional wavelet transform and short-time Fourier analysis. In the application of GPR forward analysis data, the transform can correctly distinguish different interfaces and objects. Its resolution increases as frequency increases. However, compression wavelet modulus gradually decays as frequency increases. The proposed method is applied to detect tunnel lining under actual conditions and in a strong noise background. Indeed, the method can efficiently identify interfaces and abnormalities.

Keywords: synchrosqueezing wavelet transformation, time-frequency analysis, ground penetrating radar, short-time Fourier analysis.

1. Introduction

With geophysical technology development, various detection technologies have been invented to solve detection problems in engineering construction. Ground-penetrating radar (GPR) has been commonly applied to archaeology, geologic investigation, and civil engineering because it is efficient, cost effective, and accurate [1-4]. Since 2010, the number of studies on GPR has reached 10,000 per year according to Google Scholar; furthermore, the number of research subjects increases every year. Detection results are directly influenced by the quality of signal processing; however, signal analysis is a key problem. Signal processing involved in GPR is Fourier transform. A nonstationary signal cannot be analyzed correctly because Fourier transform is impeded by time-frequency location. Various methods, such as short-time Fourier transform, Hilbert-Huang transform, and wavelet transform, have been proposed to overcome the limitations of Fourier transform [5-7].

Wavelet transform not only overcomes the defects of Fourier transform but also adjusts the size of time-frequency windows. The former can also be applied to perform a multiscale analysis of signals. Wavelet transform has been employed in GPR signal processing and achieved remarkable effects [8]. In general, the analysis of these remarkable effects influences the selection of basis function in wavelet transform. Huang Norden focused on nonstationary and nonlinear signals and proposed Hilbert-Huang transform, which includes two steps, namely, empirical mode decomposition and Hilbert spectral analysis. Hilbert-Huang transform does not require pre-selected basis functions [9, 10]. This method has been used to process GPR data, but a firm mathematical basis of Hilbert-Huang transform has yet to be developed. However, neither Hilbert-Huang transform nor wavelet transform can solve the modal mixing phenomenon in modal decomposition [11, 12]. Daubechies proposed synchrosqueezing wavelet transformation (SST), which is based on continuous wavelet transform; in SST, wavelet coefficients are squeezed in a time-frequency domain. Even if the function is nonharmonic, an SST algorithm obtains instantaneous frequency [13, 14]. Herrera also proposed and introduced an SST algorithm to the time-frequency analysis of a seismic signal. SST algorithms can achieve higher accuracy than

other algorithms. The influence on analysis results is not evident in basis function selection [15-17]. Therefore, GPR signals should be processed with the SST algorithm and the time-frequency characteristics of abnormalities should be analyzed to improve GPR technology.

In this study, the SST algorithm is introduced to GPR signal analysis. First, the principle and implementation steps of the SST algorithm are provided. The SST algorithm is used to process continuous electromagnetic signals. The proposed algorithm is then compared with a traditional method. Second, the GPR data of a forward simulation model are processed with the SST algorithm to examine its time-frequency characteristics. Finally, the GPR data are considered to evaluate the effect of noise removal on the actual detection of tunnel lining.

2. Theory and methodology

2.1. SST algorithm

In practical engineering, one signal contains multiple components, and each of the components represents one feature. A signal is composed of a group of harmonics, and the remainder is expressed as follows:

$$f(t) = \sum_{i=1}^N A_i(t) \cos[\phi_i(t)] + r(t), \tag{1}$$

where $A_i(t)$ is the amplitude and $r(t)$ is the residue.

The SST algorithm is based on wavelet analysis theory. In this study, $f(t)$ is processed with the following wavelet transform:

$$W_f(a, b) = \int_{-\infty}^{\infty} f(t) \frac{1}{\sqrt{a}} \varphi^* \left(\frac{t-b}{a} \right), \tag{2}$$

where a is the scale factor, b is the time shift, $\varphi^*(t)$ is the complex conjugate of the mother wavelet.

According to Plancherel theorem, $W_f(a, b)$ is the wavelet coefficient $f(t)$. The equation can be written as follows:

$$\begin{aligned} W_f(a, b) &= \frac{1}{2\pi} \int \hat{f}(\xi) \sqrt{a} \hat{\varphi}^* e^{ib\xi} d\xi \\ &= \frac{A}{4\pi} \int [\delta(\xi - \omega) + \delta(\xi + \omega)] \sqrt{a} e^{ib\xi} d\xi = \frac{A}{4\pi} \sqrt{a} \hat{\varphi}^*(a\xi) e^{ib\xi}, \end{aligned} \tag{3}$$

where $\hat{\varphi}(t)$ and $\hat{f}(t)$ are the mother wavelet function $\varphi(t)$ and signal $f(t)$, respectively, after Fourier transform is performed.

In $(W_x(a, b))^{-1} \partial_b W_x(a, b) = i\omega$, the instantaneous frequency of $f(t)$ can be preliminarily estimated as follows:

$$\omega_f(a, b) = \begin{cases} \frac{-i \partial_b W_f(a, b)}{W_f(a, b)}, & |W_f(a, b)| > 0, \\ \infty, & |W_f(a, b)| = 0. \end{cases} \tag{4}$$

In the synchrosqueezing phase, where $(a, b) \rightarrow (\omega_x(a, b), b)$, the energy is converted from a time-scale plane to a time-frequency plane. The synchrosqueezing value of wavelet coefficients are obtained by determining the wavelet coefficients in the central frequency domain:

$$T_f(\omega, b) = \int_{A(b)} W_f(a, b) a_i^{-3/2} da, \tag{5}$$

where $T_f(\omega, b)$ is the wavelet synchrosqueezing coefficient.

When the signal is discrete, the wavelet synchrosqueezing coefficients are expressed as follows:

$$T_f(\omega_l, b) = \sum_{a_i: |\omega_f(a, b) - \omega_l| \leq \frac{\Delta\omega}{2}} W_f(a, b) a^{-\frac{3}{2}} (\Delta a)_i, \tag{6}$$

where ω_l is the central frequency in $[\omega_l - (\Delta\omega/2), \omega_l + (\Delta\omega/2)]$.

2.2. Comparative analysis of the continuous electromagnetic signal sample

We apply conventional wavelet transform, short-time Fourier transform, and SST algorithm to analyze the transient electromagnetic wave signal and to determine the solution of the SST algorithm for time-varying signal processing. The time-varying signal $f(t)$ of analytical expressions is represented as follows [18]:

$$f(t) = \begin{cases} \sin(2\pi \times 10^7 \times t), & 0 \leq t \leq 273, \\ \sin(2\pi \times 10^7 \times t) + \sin(4\pi \times 10^7 \times t), & 274 \leq t \leq 449, \\ \sin(4\pi \times 10^7 \times t), & 450 \leq t \leq 549, \\ \sin(4\pi \times 10^7 \times t) + \sin(6\pi \times 10^7 \times t), & 550 \leq t \leq 723, \\ \sin(6\pi \times 10^7 \times t), & 724 \leq t \leq 1024. \end{cases} \tag{7}$$

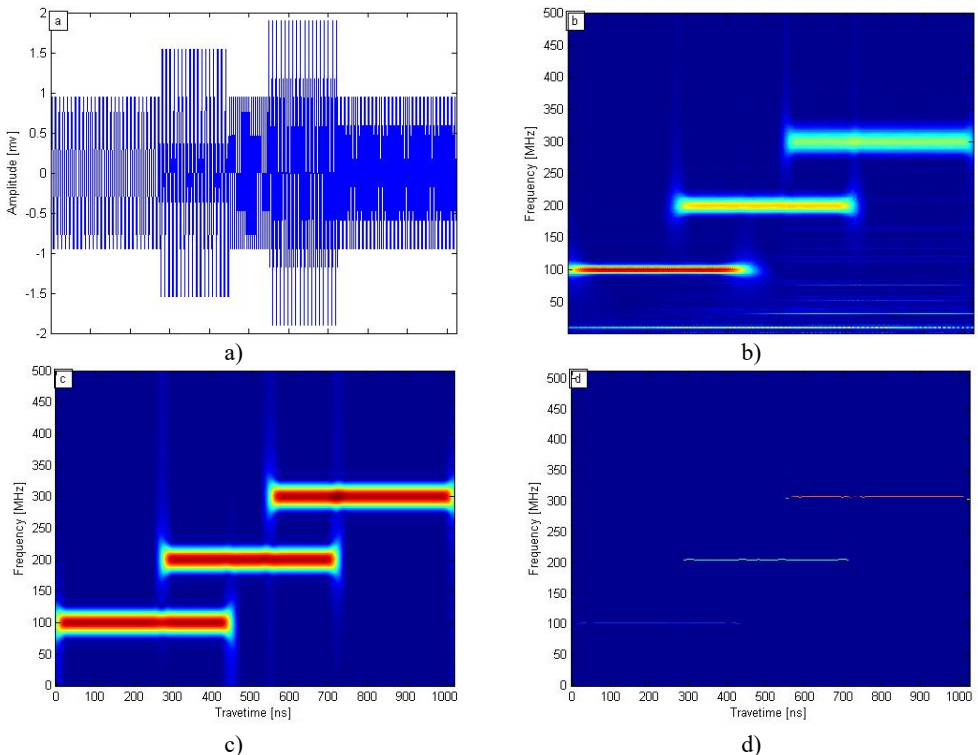


Fig. 1. a) Synthetic signal of time-varying electromagnetic signal, b) conventional wavelet transform, c) short-time Fourier transform, d) SST results

The signal $f(t)$ mainly contains three main frequency components: 100, 200, and 300 MHz. The respective time sections are 0-499, 274-723, and 550-1024 ns. In the continuous signal $f(t)$ sampling, the sampling rate is 1000 and the time window is 1024 ns. Fig. 1(a) illustrates $f(t)$.

To analyze the effect of the SST algorithm on electromagnetic signal processing, we transformed $f(t)$ with SST, conventional wavelet transform, and short-time Fourier transform. Figs. 1(b), 1(c), and 1(d) correspond to the conventional wavelet transform, short-time Fourier transform, and SST frequency distribution contour, respectively. In Fig. 1(b), the wavelet transform results show that the frequencies of 100 and 300 MHz are detected at the highest and lowest resolutions, respectively; the resolution time-frequency decreases as the frequency increases. In Fig. 1(c), the short-time Fourier transform results demonstrate that frequency slightly influences resolution. In Fig. 1(d), the SST results reveal three straight lines corresponding exactly to three different frequencies: 100, 200, and 300 MHz. In Figs. 1(b), 1(c), and 1(d), the conventional wavelet transform and short-time Fourier transform results indicate an evident banded range, whereas the SST results accurately display the frequency and time location. Therefore, SST can distinguish time-frequency more precisely and yield a higher resolution than the two conventional conversion methods.

3. Engineering applications

3.1. Forward numerical simulation model

GPR forward is its data interpretation basis. The finite-difference time-domain method (FDTD) is the most common numerical technique for solving the problems of the electromagnetic field. The FDTD method can be applied directly to the finite differential instead of Maxwell local field differential equations in order to get the field components of the finite difference formula. We use space grid to simulate purpose body having the same electrical parameters and set the appropriate initial and boundary conditions for field computing space to obtained the numerical solution of variables in Maxwell's equations [19].

We use a three-layer structure model. The uppermost layer is air, the middle layer is concrete, and the bottom layer is dry sand. A rectangular void is placed in the dry sand layer. The model size and location are shown in Fig. 2. Both the concrete and sand layers have thicknesses of 0.15 m, while the internal hole size is 0.05×1.3 m. The GPR signal of the model can be obtained through numerical simulation.

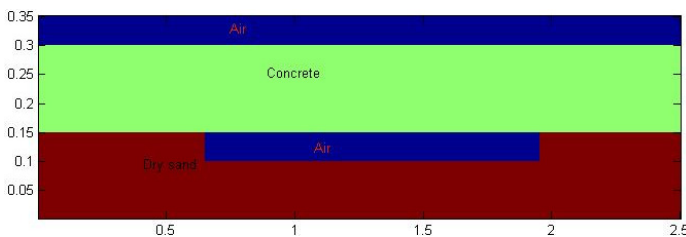


Fig. 2. Schematic of geometric model

In the forward process, the parameters are dry sand permittivity Eq. (3), conductivity (0.0001 S/m), and concrete permittivity Eq. (6), conductivity (0.005 S/m). The set calculation area is 2.5 m×0.35 m, the computational grid size is 0.0025 m, and the time window is 12 ns. Ricker wave is employed as excitation, the dominant frequency is at 900 MHz, the transeiver antenna distance is 0.025 m, and the analogue sampling is 125 traces. The result was simulated by FDTD, as shown in Fig. 3. The FDTD method is performed for GPR signals of time series at all traces. Fig. 3 shows the GPR signal at first trace. The measurement line direction is the x -axis, and time is the y -axis in ordinate. The time-series signal of 125 traces is synthesized, and the result is displayed using image mode in Fig. 4.

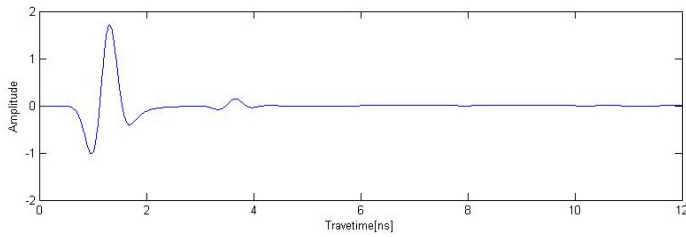


Fig. 3. GPR signal at first trace

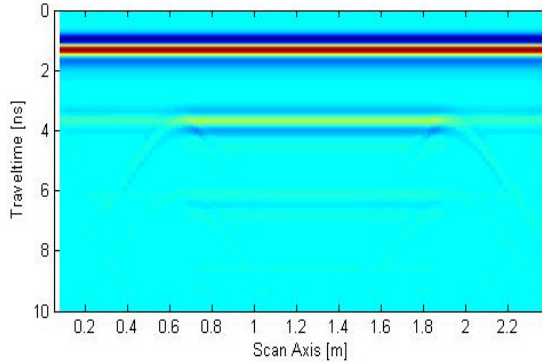


Fig. 4. Forward numerical simulation signal of GPR

The wavelet modulus is obtained in the time-frequency domain and is synchrosqueezed for the wavelet synchrosqueezing modulus in the central frequency domain by utilizing the SST algorithm of the forward signal transform. Four frequency moduli, namely, 500 MHz, 900 MHz, 1.5 GHz, and 3.0 GHz are then analyzed (Fig. 5).

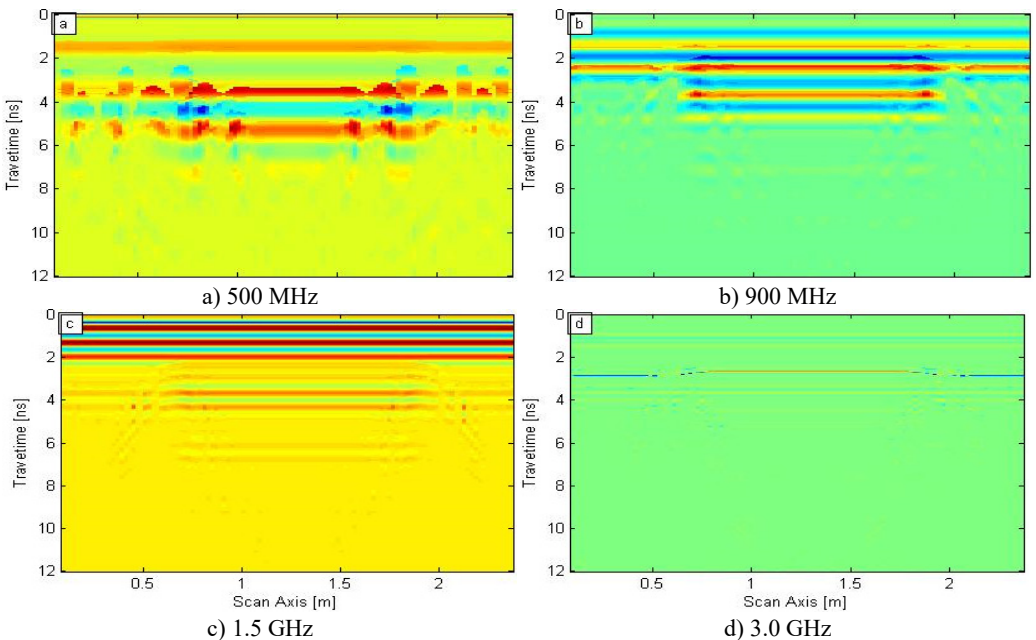


Fig. 5. Different frequency wavelet synchrosqueezing moduli

In Fig. 5, the synchrosqueezing modulus exhibits low resolution when the frequency is

relatively low; by contrast, the frequency is relatively high, the synchrosqueezing modulus yields high resolution, but the modulus strength is weak. Figs. 5(b) and 5(c) correspond to the frequencies of 900 MHz and 1.5 GHz, respectively; the void significantly affected the synchrosqueezing modulus. Figs. 5 (b) and 5(c) show that the modulus within a certain frequency range can identify the interface and defects. Therefore, the SST transformation can distinguish different interfaces and exceptions; furthermore, resolution increases as frequency increases. However, the modulus strength decreases as frequency increases.

3.2. Application analysis of tunnel lining detection

A tunnel lining detection signal is used as an example to evaluate the denoising effect of SST on practical engineering. The tunnel detection site is shown in Fig. 6(a). During the tunnel lining survey, there were three are three detection lines respectively located at the vault and the left and right waist. The three survey lines were arranged along the tunnel in the longitudinal direction. The three survey lines and the lining thickness for detecting defects, as well as the survey line layout are shown in Fig. 6(b).

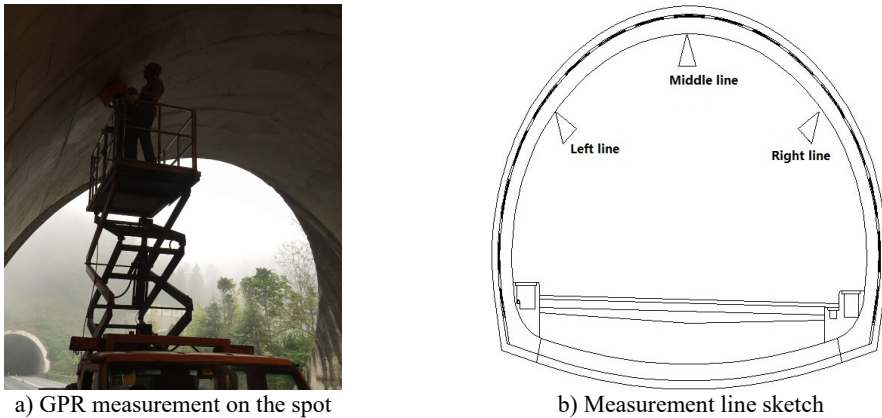


Fig. 6. GPR tunnel lining detection

The instrument employed in this analysis is LTD2100 GPR. The sampling center frequency is 0.4 GHz, the time window is 35 ns, and the trace spacing is 0.015 m. The chosen length of the target section is 8 m in left survey line. The detection results are shown in Fig. 7(a). The reflected signal of the secondary lining bottom interface appears within 10 ns of the simulation. The reflected signal is also a steel arch, which is shown in the initial lining. In general, the containing noise signal is denoising data minus original noiseless data, and the ratio of signal to noise energy is regarded as signal to noise ratio (SNR). We can use this parameter to measure the denoising effect. The formula is as follows:

$$SNR = \frac{\|M\|_2^2}{\|S - M\|_2^2} \tag{8}$$

where S_{NR} is SNR, $\|\cdot\|_2^2$ is square norm, M is original noiseless data, and S is containing noise data.

In the SST filtering process, referred to as SNR in this study, is 0.290 (-5.369 dB). The original signal in Fig. 7(a) joins the strong interference. White noise is added, and the result is shown in Fig. 7(b). With white noise interference, the interface between the second lining and the initial lining is extremely fuzzy, and the steel arch reflection signal in the primary lining is also not evident. The signal of the GPR is processed through SST filtering to verify whether the noise effect is removed [Fig. 7(b)].

3) The SST algorithm correctly separates the interference from the signal even in a strong-interference environment; this algorithm also effectively removes noise.

The SST algorithm is more complex than conventional algorithm; this algorithm also exhibits a relatively low processing efficiency. Hence, the algorithm should be improved to upgrade its efficiency.

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