

As far as we are concerned, a *signal* is function of *time* with values in \mathbb{C} . Time can be *discrete* or *continuous* and the signal can be *nonperiodic* or *periodic*:

Continuous Nonperiodic	Discrete Nonperiodic
Continuous Periodic	Discrete Periodic

In the discrete time case, we assume that the time-interval between successive instants is a fixed time-unit. The case of discrete signals measured at irregularly distributed instants is interesting, but we will not cover it.

- **Continuous nonperiodic.** The signal is a function $f: \mathbb{R} \rightarrow \mathbb{C}$.
- **Continuous periodic,** with period $T > 0$. The signal is a function $f: \mathbb{R} \rightarrow \mathbb{C}$ with the property that $f(t+T) = f(t) \forall t \in \mathbb{R}$. Since all that matters is knowledge of f on an interval of length T , we can as well assume that $f: [0, T) \rightarrow \mathbb{C}$. (Any other interval having length T would do).
- **Discrete nonperiodic.** The signal is a function $f: \mathbb{Z} \rightarrow \mathbb{C}$. We choose the time-unit in such a way the distance between successive instants is one unit.
- **Discrete periodic.** Fix a period $N \geq 1$, $N \in \mathbb{N}$. The signal is a function $f: \mathbb{Z} \rightarrow \mathbb{C}$ such that $f(n+N) = f(n) \forall n \in \mathbb{Z}$. We can identify f with a function $f: \{0, 1, \dots, N\} \rightarrow \mathbb{C}$, or with an array (a vector) $(f(j))_{j=0}^N = \begin{pmatrix} f(0) \\ f(1) \\ \dots \\ f(N) \end{pmatrix} \in \mathbb{C}^N$.

Discrete periodic

Linear algebra in \mathbb{C}^N

The case we consider first is that of discrete, periodic signals, which might be seen as a chapter in linear algebra. We start with a review of basic concepts. We only have to take into account that scalars are complex, rather than real.

- If $f, g \in \mathbb{C}^N$ and $\alpha \in \mathbb{C}$, then $(f+g)(j) := f(j) + g(j)$ and $(\alpha f)(j) := \alpha f(j)$ define the sum $f+g$ and the product αf . These operations make \mathbb{C}^N into a vector space. The sum $f+g$ might be seen as the superposition of the two signals and the product αf might be seen as the amplification of a signal f by a factor α .
- We want to measure the size of a signal. There are several ways to do that. The simplest one is using the *standard inner product*: $\langle g, f \rangle = \sum_{j=0}^{N-1} \overline{g(j)} f(j)$. To measure the size, then, we use the *standard norm*: $\|f\| = \sqrt{\langle f, f \rangle} = \sqrt{\sum_{j=0}^{N-1} |f(j)|^2}$.

Properties of the inner product on the vector space $(\mathbb{C}^N, +, \cdot)$.

- $\forall f, g \in \mathbb{C}^N \Rightarrow \langle g, f \rangle = \overline{\langle f, g \rangle}$
- $\forall f, g, h \in \mathbb{C}^N \forall a, b \in \mathbb{C} \Rightarrow \langle h, af + bg \rangle = a \langle h, f \rangle + b \langle h, g \rangle$
- $\forall f, g, h \in \mathbb{C}^N \forall a, b \in \mathbb{C} \Rightarrow \langle af + bg, h \rangle = \bar{a} \langle f, h \rangle + \bar{b} \langle g, h \rangle$
- $\forall f \in \mathbb{C}^N \Rightarrow \langle f, f \rangle \geq 0$ and $\langle f, f \rangle = 0$ if and only if $f = 0$
- $\forall f, g \in \mathbb{C}^N \Rightarrow |\langle g, f \rangle| \leq \|f\| \cdot \|g\|$ (Cauchy-Schwarz inequality)

We can use the norm to define a *distance* between $f, g \in \mathbb{C}^N$. We set it to be $\|g - f\|$.

The only property whose verification is not trivial is the Cauchy-Schwarz inequality. We start with an obvious inequality and do some algebra:

$$\begin{aligned}
0 &\leq \sum_{\substack{j,k=0 \\ N-1}}^{N-1} |f(j)g(k) - g(j)f(k)|^2 \\
&= \sum_{\substack{j,k=0 \\ N-1}}^{N-1} (f(j)g(k) - g(j)f(k)) \overline{(f(j)g(k) - g(j)f(k))} \\
&= \sum_{\substack{j,k=0 \\ N-1}}^{N-1} |f(j)|^2 |g(k)|^2 - \sum_{\substack{j,k=0 \\ N-1}}^{N-1} f(j) \overline{g(j)} \overline{f(k)} g(k) - \sum_{\substack{j,k=0 \\ N-1}}^{N-1} g(j) \overline{f(j)} \overline{g(k)} f(k) + \sum_{\substack{j,k=0 \\ N-1}}^{N-1} |g(j)|^2 |f(k)|^2 \\
&= 2 \sum_{j=0}^{N-1} |f(j)|^2 \sum_{k=0}^{N-1} |g(k)|^2 - 2 \sum_{j=0}^{N-1} f(j) \overline{g(j)} \cdot \sum_{k=0}^{N-1} \overline{g(k)} f(k) \\
&= 2 \left[\sum_{j=0}^{N-1} |f(j)|^2 \sum_{k=0}^{N-1} |g(k)|^2 - \left| \sum_{j=0}^{N-1} f(j) \overline{g(j)} \right|^2 \right] \\
&= 2[\|f\|^2 \|g\|^2 - |\langle g, f \rangle|^2].
\end{aligned}$$

We have then $|\langle g, f \rangle|^2 \leq |\langle g, f \rangle|^2 + \frac{1}{2} \sum_{j,k=0}^{N-1} |f(j)g(k) - g(j)f(k)|^2 \leq \|f\|^2 \|g\|^2$, as wished.

We can also deduce the cases of equality.

Corollary 1. $\forall f, g \in \mathbb{C}^N$: $|\langle g, f \rangle| = \|f\| \cdot \|g\|$ if and only if $0 = \sum_{j,k=0}^{N-1} |f(j)g(k) - g(j)f(k)|^2$, and the latter holds if and only if there are $a, b \in \mathbb{C}$, not both vanishing, such that $af + bg = 0$.

That is, equality holds if and only if f and g are linearly dependent. (**Exercise:** prove the last “if and only if” in the corollary).

A *basis* for \mathbb{C}^N is a family $\{f_0, \dots, f_{N-1}\}$ of elements in \mathbb{C}^N with the property that any element f in \mathbb{C}^N can be written in a unique way as

$$f = \sum_{j=0}^{N-1} \lambda_j f_j,$$

where $\lambda_0, \dots, \lambda_{N-1} \in \mathbb{C}$ are scalars whose value depend on f and on the basis. The family $\{f_0, \dots, f_{N-1}\}$ is a basis for \mathbb{C}^N if and only if, for all $\lambda_0, \dots, \lambda_{N-1} \in \mathbb{C}$,

$$\sum_{j=0}^{N-1} \lambda_j f_j = 0 \text{ if and only if } \lambda_0 = \dots = \lambda_{N-1} = 0.$$

The basis $\{f_0, \dots, f_{N-1}\}$ is *orthonormal* if $\langle f_j, f_k \rangle = 0$ whenever $j \neq k$, and $\|f_j\| = 1$ for $j = 0, \dots, N-1$.

In this case,

$$f = \sum_{j=0}^{N-1} \langle f, f_j \rangle f_j.$$

A special orthonormal basis for \mathbb{C}^N consists of the functions δ_j ($j = 0, \dots, N-1$), defined by

$$\delta_j(k) = \begin{cases} 0 & \text{if } k \neq j \\ 1 & \text{if } k = j \end{cases}$$

The δ_j 's are called in different ways in different communities: Dirac's Deltas, Kroenecker's Delta, unit impulses.... They have a privileged role if we have an interpretation for the parameter $j \in \{0, 1, \dots, N-1\}$. Typically, we see j as "time". Any discrete, periodic signal can be written as linear combination of unit impulses:

$$f = \sum_{j=0}^{N-1} \langle \delta_j, f \rangle \delta_j = \sum_{j=0}^{N-1} f(j) \delta_j.$$

Linear applications

A *linear application* $A: \mathbb{C}^N \rightarrow \mathbb{C}^M$ is a map which satisfies

$$A(af + bg) = aAf + bAg$$

whenever $a, b \in \mathbb{C}$ and $f, g \in \mathbb{C}^N$.

A very common case is $M = 1$, $\Lambda: \mathbb{C}^N \rightarrow \mathbb{C}$, which we might think as the (scalar) *measurement* Af performed on the signal f . The requirement that Λ be linear much restricts the kind of measurements which are taken into account.

Another common case is $M = N$. We might think of this in terms of a system A performing a linear transformation on the input signals $f \in \mathbb{C}^N$, producing an output signal Af .

Theorem 2. (*Riesz-Fisher: finite version, algebraic part*) Let $\Lambda: \mathbb{C}^N \rightarrow \mathbb{C}$ be linear. Then there exists $g \in \mathbb{C}^N$ such that

$$\Lambda(f) = \langle g, f \rangle$$

for $f \in \mathbb{C}^N$. Moreover, $g(j) = \overline{\Lambda(\delta_j)}$ $j = 0, \dots, N-1$, provides a formula to recover g from Λ .

Proof.

$$\begin{aligned} \Lambda(f) &= \Lambda\left(\sum_{j=0}^{N-1} f(j)\delta_j\right) \\ &= \sum_{j=0}^{N-1} f(j)\Lambda(\delta_j) \\ &= \sum_{j=0}^{N-1} f(j)\overline{\Lambda(\delta_j)} \\ &= \langle g, f \rangle, \text{ with } g = \sum_{j=0}^{N-1} g(j)\delta_j = \sum_{j=0}^{N-1} \overline{\Lambda(\delta_j)}\delta_j. \end{aligned}$$

An analogous representation holds in the general case. We only consider $N = M$.

Theorem 3. (*Representation of linear operators on \mathbb{C}^N : the algebraic part*) Let $A: \mathbb{C}^N \rightarrow \mathbb{C}^N$ be a linear application from \mathbb{C}^N to itself. Then, there exists an $N \times N$ matrix $\tilde{A} = [a(k, j)]_{k=0, \dots, N-1; j=0, \dots, N-1}$ such that

$$A(f) = \tilde{A}f,$$

where on the right we perform row-times-column multiplication, as usual. Moreover, $a(k, j) = \langle \delta_k, A(\delta_j) \rangle$.

Proof. We can write

$$\begin{aligned}
A(f) &= A\left(\sum_{j=0}^{N-1} f(j)\delta_j\right) \\
&= \sum_{j=0}^{N-1} f(j)A(\delta_j) \\
\text{hence,} \\
\langle \delta_k, A(f) \rangle &= \sum_{j=0}^{N-1} f(j) \langle \delta_k, A(\delta_j) \rangle \\
&= \sum_{j=0}^{N-1} a(k, j)f(j),
\end{aligned}$$

as wished.

From now on we identify $\tilde{A} = A$.

Assuming that $\|f\|$ is a good way to measure the size of the signal $f \in \mathbb{C}^N$, a linear measurement Λ or a linear transformation A will be relatively stable if for $\|f\|$ small, the size $|\Lambda(f)|$ of the measured quantity or the size $\|A(f)\|$ of the transformed signal are small as well. Observe that we have scalar homogeneity w.r.t. f :

$$|\Lambda(af)| = |a| |\Lambda(f)| \text{ and } \|A(af)\| = |a| \|A(f)\|.$$

We can then rescale everything and just consider $\|f\| = 1$, if we wish so.

The *norm* of Λ and A will vbe defined accordingly.

$$\begin{aligned}
\|\Lambda\| &= \text{Max}_{\|f\|=1} |\Lambda(f)| = \text{Max}_{f \neq 0} \frac{|\Lambda(f)|}{\|f\|} \\
\text{and} \\
\|A\| &= \text{Max}_{\|f\|=1} \|A(f)\| = \text{Max}_{f \neq 0} \frac{\|A(f)\|}{\|f\|}.
\end{aligned}$$

The equality between the expressions on the right in both formulas follows from homogeneity.

Next questions is: is there any reasonable way to compute, or just estimate, the norm of Λ , or of A ?

It turns out that the case of Λ ($M=1$) is easy, that of A ($M=N$) much less so.

Theorem 4. *Let $\Lambda: \mathbb{C}^N \rightarrow \mathbb{C}$, $\Lambda(f) = \langle g, f \rangle$. Then, $\|\Lambda\| = \|g\|$.*

Proof. By the Cauchy-Schwarz inequality,

$$\begin{aligned}
\frac{|\Lambda(f)|}{\|f\|} &= \frac{|\langle g, f \rangle|}{\|f\|} \\
&\leq \frac{\|g\| \cdot \|f\|}{\|f\|} = \|g\|,
\end{aligned}$$

then $\|\Lambda\| \leq \|g\|$. In the other direction, let $f = g$ if $g \neq 0$ (if $g = 0$ then $\Lambda = 0$ and $\|0\| = 0$ trivially). Then,

$$\|\Lambda\| \geq \frac{|\Lambda(g)|}{\|g\|} = \frac{|\langle g, g \rangle|}{\|g\|} = \frac{\|g\|^2}{\|g\|} = \|g\|.$$

This finishes the proof.

When the operator A maps \mathbb{C}^N to \mathbb{C}^N , there is no such simple formula, and we have to consider specific classes of operators. The simplest case is that of the *diagonal operators*.

Let $\lambda_0, \lambda_1, \dots, \lambda_N \in \mathbb{C}$ and form the matrix

$$\text{Diag}(\lambda_0, \dots, \lambda_N) = \begin{pmatrix} \lambda_0 & 0 & 0 & \dots & 0 \\ 0 & \lambda_1 & 0 & \dots & 0 \\ \dots & \dots & \dots & \dots & \dots \\ 0 & 0 & \dots & 0 & \lambda_{N-1} \end{pmatrix}$$

The effect on a unit impulse at time j is $\text{Diag}(\lambda_0, \dots, \lambda_N)\delta_j = \lambda_j\delta_j$: the numbers λ_j are the *eigenvalues*, and the unit impulses the *eigenvectors*, of the matrix $\text{Diag}(\lambda_0, \dots, \lambda_N)$.

Proposition 5. $\|\text{Diag}(\lambda_0, \dots, \lambda_N)\| = \text{Max}\{|\lambda_0|, \dots, |\lambda_{N-1}|\}$.

Proof. Exercise.