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Study of mm-Wave Microstrip Patch Array on Curved Substrate

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Abstract

A millimetre wave rectangular microstrip patch antenna array for fifth generation communications (5G) applications is presented. With the increase in demand for high data rates and capacity, there is a need to include mm-Wave frequencies for 5G. The 4×2 patch array is simulated and fabricated on a Rogers RT/Duroid 5880 substrate with a thickness of 0.25 mm. The effects of bending the substrate on the antenna performance are also presented. The experimental results show a fractional bandwidth and gain of more than 3% and 16.1 dBi, respectively.

1 Introduction

The expansion of the wireless industry around the globe has enabled people to access the internet and the demand for high-quality wireless communications systems is growing exponentially. This rapid growth demands for high data rates and overcoming challenges including capacity issues to accommodate more users and applications which require quick and quality research in order to develop next generation wireless communications systems. The mm-Wave spectrum is most likely to be the potential candidate to implement the 5G communication applications because of its huge bandwidth [1].

The antenna is an essential component of wireless communication system because of its effects on the receiver's sensitivity. The introduction of the mm-wave band for 5G applications has brought considerable attention to the researchers in order to analyse and design antennas which are reported in several papers [4-7]. In [2], a multi-band split ring resonator was proposed which covers 6 GHz band which is also expected for 5G communications [3]. A 4×1 and 2×2 stacked patch array on LTCC and PCB substrates are presented in [4] with a maximum gain of 13 dBi. Other works include dielectric superstrate and EBG ground structure to improve radiation characteristics in 5G [5] and phased arrays [6]. A review for several 5G antennas was presented in [7].

In this paper, a 4×2 microstrip array with rear feeding has been presented. The spacing between the elements plays a key role in the resonant frequency and the performance (gain and efficiency) of the antenna. The effects of mechanically curving the substrate on the antenna's performance have also been presented in section 4 of this paper. A study of a UWB directional antenna placed conformally on a cylindrical pipe section was reported in [8]. For realistic results, the edgefeeding mechanism is adopted for the curved substrate and results for concave and convex bending are noticed.

2 Antenna Geometry

The geometry of the proposed microstrip patch antenna array for 5G communications is shown in Figure 1. The array is designed on a flexible Rogers RT/Duroid 5880 substrate with relative permittivity of 2.2, loss tangent of 0.0009 and thickness of 0.25 mm having dimensions 19×30 mm². The array consists of 8 elements with a spacing of 0.75 between the row elements. Each patch is connected to a 100 line and then to a matching arrangement using quarter wave transformers. The 90° bends are mitred. The two rows of the array are connected through a one wavelength network including two-quarter wave transformers and a 50 pad to which the connector is soldered from the bottom through a 0.5 mm via hole. The proposed antenna is simulated using CST Microwave Studio Suite 2016 with a 50 modelled SMK (Sub-Miniature version K) connector. Table 1 shows the design parameters of the antenna in detail.

Parameter	Dimension	Parameter	Dimension
	(mm)		(mm)
Ws	30.0	Ls	19.0
Wp	4.48	Lp	3.55
Wi	1.10	Li	0.20
Wq	0.48	Lq	1.99
Wf	0.8	Lf	2.30
W100	0.25	L100	1.11
d	1.52	x	8.10

 Table 1 Design parameters of the proposed antenna array



Fig. 1 The geometry of 4×2 element rear-fed array



Fig. 2 Parametric analysis with element spacing

3 Modelling, Parametric Analysis and Results

In order to achieve high gain results, the single patch antenna was first simulated and optimised at 27.4 GHz for best matching. The addition of an inset feed improves the impedance matching and minimises the input reactance of the antenna. The gain of the single patch antenna was found to be 7.5 dBi.

The effects of spacing between the patch elements in the array were optimised accordingly. Figure 2 shows the effect on the S11 for different element spacing. The frequency response tunes down and the bandwidth decreases as the spacing between the elements increases from 0.75 to 3. The increase in the distance also results in impedance mismatching. From to 3, in the 28 GHz range, there are additional resonances observed.

Figure 3 shows the measured and simulated S11 for the antenna array. The antenna array is measured using Rhode and Schwarz Vector Network Analyzer (ZVA40) and shows good agreement with the simulated results. The simulated results show that the antenna has |S11| < -10 dB between 27.04 GHz and 27.71 GHz centred at 27.4 GHz with a fractional bandwidth of 2.44%. The measured results show that the antenna is operating between 26.97 GHz and 27.82 GHz with a minimum value of S11 at 27.4 GHz.



Fig. 3 Simulated and measured S11 at 27.4 GHz



Fig. 4 Total and radiation efficiency

The measured fractional bandwidth is more than 3.1%.

The radiation and total efficiencies of the array are presented in Figure 4. At 27.4 GHz, the antenna has the maximum efficiency and the percentage of the power to the antenna to the power radiated from the antenna is more than 85%. The 90° mitring helps to improve the efficiency of the antenna by 4%.

Figure 5 shows the measured and simulated gain of the array antenna. From the figure, it can be seen that the array has a high measured gain between 25 GHz and 29 GHz with a peak gain of 16.1 dBi at 27.4 GHz. The simulated gain at 27.4 GHz is 15.7 dBi. The feeding network is optimised in such a way that the feeding arrangement and the power divider network constitute to only 0.4 dBi attenuation of the total gain.

The simulated and measured radiation pattern plots are shown in Figure 6. Figure 6(a) shows the results in the yz-plane (E-Plane). The radiation pattern is tested in an anechoic chamber. The antenna has a directive radiation pattern with a half power beamwidth (HPBW) of 35° in yz-plane and 20° in xz-plane. It further verifies the peak gain at 27.4 GHz to be 16.1 dBi. The squint visible in the yz-plane of the rear-fed antenna array is most likely a result of the addition of quarterwave transformers or microstrip lines [9]. Figure 6(b) shows the polar plot for xz-plane (H-plane) of the antenna. The measured and simulated results are in good agreement.



Fig. 5 Simulated and measured gain for the 4×2 array



Fig. 6 Measured and simulated radiation patterns (a) yz-plane (b) xz-plane

4 Curved Substrate: Modelling and Results

The effects of mechanically curving the substrate on matching and radiation properties were studied to determine tolerances. With the configuration shown in Figure 1, it is not realistically possible to curve the substrate because of the soldered connector at the bottom of the antenna. However, some arrays may have edge fed connectors and an array shown in Figure 7 was prototyped and tested. In place of the feed line connecting the two rows of the array shown in Figure 1, a 50 line is introduced which connects the two rows with a one wavelength line and extends to the edge of the substrate. The edges of the line are tapered to avoid shortening the connector. The dimensions of this antenna are 23×30 mm². The rest of the configurations are same as in Figure 1.



Fig. 7 Geometry of the curved substrate edge-fed antenna

Figure 8 shows the effects of concave and convex bending of the substrate on the S11. From the graph, it can be noticed that there are minute resonant frequency shifts with the concave curving but there is little change in bandwidth. The impedance mismatching increases with the increase in concave bending. On the other hand, the frequency tunes higher than the reference 27.4 GHz when the antenna is curved in a convex shape. At $+3^{\circ}$ bend, another resonance emerges at 25.23 GHz and shifts to the right as the convexity increases. This is because of the increased patch mutual coupling with reduced distance between patches. Similar to concave bending, the antenna impedance matching degrades with increased curve angle but little changes are observed in the bandwidth.



Fig. 8 Simulated S11 with concave and convex bending

Figure 9 shows the gain plot of the array on curved substrate. With reference to the gain for 0° bending, the gain does not vary much with the concave bending. The antenna gain profile for 25-30 GHz is persistent and maximum variation in peak gain is 1 dBi for bending from 0° to -30° (concave). The gain profile changes for convex bending greater than 0° . Although the peak gain is stable with this design, it tunes up in frequency with convex bending.



Fig. 9 Simulated realized gain with concave and convex bending



Fig. 10 Polar plots with curved substrate (a) yz-plane (b) xz-plane

The 2-D polar plots on the curved substrate are shown in Figure 10. Figure 10(a) shows the pattern for the yz-plane of the antenna. It can be seen in the figure that the increase in the concave angle does not affect the overall gain pattern but has small variations in the gain. The plots of additional resonances with convex bending can be seen distinctly with reduced gain. The xz-plane in Fig. 10(b) clearly shows the main lobe splits at 3° for 25.23 GHz, which is the additional resonance. The increase in bending in concave direction increases the side lobe level, which leads to a wider beamwidth but reduced gain. The prototyped antenna is shown in Figure 11.



Fig. 11 Prototype of the proposed antennas

5 Conclusion

A 4×2 element microstrip antenna array of dimension $19\times30\times0.25$ mm³ for 5G, providing a 16.1 dBi peak gain, is

described. The antenna is analysed with rear probe feed and edge feed arrangements. The effects of substrate concave and convex bending are investigated using the edge feed. Concave curving of the substrate was found to have minimal effect whereas performance is shown to be heavily dependent on convex curve angle.

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