

An Investigation of the
Contingent Negative Variation
Using Signal Processing Methods

Martin John Nichols

This thesis is submitted in partial fulfilment of
the requirements of the Council for National Academic
Awards for the degree of Doctor of Philosophy.

Department of Communication Engineering,
Plymouth Polytechnic,

in collaboration with the

Department of Neurological Sciences,
Freedom Fields Hospital,
Plymouth.

August 1982

PLYMOUTH POLYTECHNIC LIBRARY	
Accn. No.	5500170 ⁹
Class. No	T-616.8047547 NIC
Contl. No	x700439535

Declaration

I hereby declare that whilst registered as a candidate for the degree of Doctor of Philosophy with the Council for National Academic Awards I have not been registered for any other qualification of the CNAA or any other examining body.

Signed M. J. Nichols

In accordance with regulations 3.8 and 3.9 I have attended and participated in the following:-

Lectures in Communication Engineering (Intended for B.Sc. Hons. Students) October 1977 - May 1978.

IEE Meeting, Savoy Place, London, "Microprocessors in Medical Instrumentation". 31st May 1978.

BES/HPA, Nottingham, "International Evoked Potentials Symposium". 4th - 6th September 1978.

EPTA, Scientific Meeting, Romford. 25th November 1978.

EURASIP, Lausanne, Switzerland, "First European Signal Processing Conference". 16th - 19th September 1980.

EEG Society, Scientific Meeting, Plymouth. 9th May 1981.

An Investigation of the Contingent Negative Variation
Using Signal Processing Methods

M.J. Nichols

Abstract

The Contingent Negative Variation (CNV) is one of many types of electrical response signals which appear in the electroencephalogram (EEG) of man subsequent to one or more stimuli. Generally these responses are small in comparison to the normal background EEG and had always been thought to consist of a response component which was added to the background EEG. Professor B. MCA. Sayers of Imperial College suggested that the auditory response might actually be due to a temporary ordering of the phases of the components of the background EEG. A model, allowing for additive and ordering effects, is proposed here. This model was tested on both auditory and CNV responses using statistical tests not previously used in evoked potential studies. The tests showed that while the additive model satisfactorily described the auditory responses, it did not explain the CNV responses so well. However, both sets of responses showed a certain amount of phase ordering and this was consistent with the model which showed that a repetitive additional component would always incorporate the phase ordering effect. In the absence of detectable additivity pure phase re-ordering might alternatively occur as proposed by Sayers.

The CNV's of a patient group were also studied and certain tests are proposed as a possible method of diagnosis. The reliability of these tests was not conclusively proved as much larger control and patient groups would be required to do this.

An important part of this work involved the introduction of a quantitative method for assessing the effectiveness of methods of removing eye movement artefact from the EEG. This allowed the development of a more extensive correction method which was tested against two other techniques and found to be superior. This correction method will provide the basis for further research and the development of a corrector to be made commercially.

References, Tables and Diagrams

For each section the references are numbered [1], [2], and are listed at the end of that section.

Tables and diagrams are numbered 'a-b' where 'a' denotes the section number and 'b' the numerical sequence within that section. Diagrams and tables have been inserted after, and as near as possible to the text which first refers to them.

Table of Contents

	Page
Declaration	2
Abstract	3
References, Tables and Diagrams (Annotation of)	4
Table of Contents	5
List of Diagrams	10
1. Introduction	14
1.1 The Origins of Electroencephalography	14
1.2 Technological Change	15
1.3 Evoked Potentials	16
1.4 Additivity and Ordering	18
1.5 An Application of the CNV to Diagnostic Medicine	20
References	21
2. Theoretical Development	24
2.1 Models for Additivity and Phase Ordering in Evoked Responses	24
2.2 Tests of the Models	27
2.2.1 Histograms of Phase and Amplitude	28
2.2.2 Angular Statistical Tests for Phase Ordering	31
2.2.2.1 The Rayleigh Test of Circular Variance	33
2.2.2.2 The Modified Rayleigh Test of Circular Variance	34
2.2.2.3 The Hodges Ajne Test	36
2.2.3 Tests of Additivity	36

	Page	
2.2.3.1	Pre- and Post- Stimulus Energy Tests	36
2.2.3.2	Pre- and Post- Stimulus Mean Amplitude Differences Test	37
2.2.3.3	Nearest and Furthest Mean Amplitude Test	38
2.3	Eye Movement Corrections	39
2.4	Assessment of the Eye Movement Correction Methods	45
2.5	The Modified Correction Method	49
	References	56
3	Experimental Techniques	59
3.1	Initial CNV tests	59
3.2	Data Transfer and Preliminary Processing	61
3.3	Eye Movement Corrections	63
3.4	Processing of Eye Movement Data	65
3.5	CNV Acquisition	68
3.6	Processing of CNV's	70
3.7	AEP Acquisition	75
3.8	Processing of AEP's	77
3.9	Fourier Transform Considerations	79
	References	84
4	Experimental Apparatus	87
4.1	Choice of Apparatus	87
4.2	Analogue Electronics	88
4.2.1	Testing of the Analogue Electronics	96
4.3	High Speed Serial Data Link	102
4.4	Minicomputers and Software	108

	Page
4.5 Other Equipment	110
4.5.1 Eye Movement Correction Screen	110
4.5.2 The CNV Stimulus Generator	112
4.5.3 The AEP Stimulus Generator	114
References	115
5 Results and Discussion	116
5.1 Eye Movement Corrections	116
5.2 Auditory Evoked Potentials	124
5.3 The CNV's of Normal Subjects	131
5.3.1 The Average CNV's	132
5.3.2 Energy Tests	135
5.3.2.1 Broadband Energy Tests	135
5.3.2.2 Amplitude Histograms	138
5.3.2.3 Pre- and Post- Stimulus Mean Amplitude Differences Test	140
5.3.2.4 Nearest and Furthest Mean Amplitude Test	142
5.3.2.5 Discussion of Results of Energy Tests	145
5.3.3 Tests for Phase Ordering	146
5.3.3.1 Phase Histograms	146
5.3.3.2 Rayleigh Test of Circular Variance	146
5.3.3.3 Modified Rayleigh Test of Circular Variance	151
5.3.3.4 Hodges-Ajne Test	151
5.3.4 Discussion of Energy and Phase Results	156
5.4 The CNV's of Abnormal Subjects	160
5.4.1 The Averaged CNV's	160

	Page
5.4.2 Energy Tests	160
5.4.2.1 Broadband Energy Tests	163
5.4.2.2 Amplitude Histograms	163
5.4.2.3 Pre- and Post- Stimulus Mean Amplitude Differences Test	163
5.4.2.4 Nearest and Furthest Mean Amplitude Test	166
5.4.2.5 Discussion of Results of Energy Tests	166
5.4.3 Tests for Phase Ordering	169
5.4.3.1 Phase Histograms	170
5.4.3.2 Rayleigh Test of Circular Variance	170
5.4.3.3 Modified Rayleigh Test of Circular Variance	175
5.4.3.4 Hodges-Ajne Test	175
5.4.4 Discussion of Energy and Phase Results	175
5.5 Distinction Between Patients and Normals on the Basis of their CNV's	181
5.6 The Development of the CNV from trial-to-trial in Normals and Patients	188
References	197
6 Conclusions	199
6.1 Eye Movement Artefact Removal	199
6.2 Evoked Potentials	200
6.3 Auditory Evoked Potentials	200
6.4 The CNV's of Normal Subjects	201
6.5 The CNV's of Abnormal Subjects	202
6.6 Distinctions between the CNV's of Normals and Patients	203

	Page
6.7 Future Work	204
6.7.1 Eye Movement Artefact Removal	204
6.7.2 Auditory Evoked Potentials	205
6.7.3 The CNV's of Patients and Normals	205
References	207
7 Acknowledgements	208
8 Appendices	
8.1 Calculation of Expected Phase Values for Idealized CNV's	A1-1
8.2 Transformation of the Rayleigh Probabilities to those of Circular Variance	A2-1
8.3 Probability Levels for the Modified Rayleigh Test	A3-1
8.4 Probability Levels for the Hodges- Ajne Test	A4-1
8.5 The Paired t-Test	A5-1
8.6 The t-Test	A6-1
8.7 Fortran Programme 'DATAPLOT'	A7-1
8.8 Fortran Programme used in the Analysis of the Eye Movement Correction Methods	A8-1
8.9 Special Instructions for Peripheral Control	A9-1
8.10 PAL 8 Computer Programme Used to Control the Data Acquisition Process	A10-1
8.11 PAL 8 Computer Programme Used to Store the Data onto Disk	A11-1
8.12 Fortran Programmes used in the Analysis of the CNV Data	A12-1
Published Papers (2)	Pocket in Back cover

List of Diagrams

Figure Number		Page
2-1	Phasor diagrams showing phase ordering and additivity.	26
2-2	An Idealized CNV response.	29
2-3	A diagrammatic representation of the Fourier Transform of Figure 2-2.	29
2-4	The Rose diagram.	32
2-5	The eye movement artefact correction method due to McCallum and Walter.	41
2-6	The equivalent circuit of Figure 2-5.	41
2-7	The equivalent circuit of Figure 2-5 with Potentiometer.	42
2-8	The eye movement artefact correction method due to Girton and Kamiya.	44
2-9	A typical vertical EOG signal during the the eye movement correction experiments.	44
2-10	The measurement of the autocorrelation coefficient.	48
2-11	A graph of the variation of the a.c.c. with the EOG amplitude.	51
3-1	The initial attempts at recording CNV responses.	60
3-2	The electrode positions for the EOG signals.	66
3-3	A silver-silver chloride measuring electrode.	66
3-4	The raw CNV waveform and EOG signals.	71
3-5	An averaged 1 second ISI CNV waveform.	72
3-6	The frequency response of the digital low pass filter.	74
3-7	A typical CNV phase histogram.	76

	Page	
3-8	Typical averaged auditory responses.	78
3-9	A typical pair of AEP phasor diagrams.	80
4-1	A block diagram of the data acquisition and processing system.	89
4-2	A block diagram of the input section of the EEG machine.	91
4-3	The circuit of an amplifier and sample hold module.	92
4-4	The crystal oscillator and divider network.	95
4-5a	The amplitude response of the data acquisition system.	97
4-5b	The phase response of the data acquisition system.	98
4-6	The connection of the measurement system for the ramp response tests.	100
4-7	The input and output ramp waveforms.	101
4-8	A block diagram of one of the serial transceivers.	103
4-9a	The circuit diagram of one of the serial transceivers.	105
4-10	The eye movement correction screen.	111
4-11	The circuit of the CNV stimulus generator.	113
5-1	The eye movement correction signals and their a.c.f.'s	117
5-2	The effect of electrode loading.	123
5-3	Typical Pre- and Post- Stimulus AEP phasor diagrams.	127
5-4	The averaged CNV's (1 second ISI) of the normal subjects.	133
5-5	The averaged CNV's (4 second ISI) of the normal subjects.	134
5-6	Typical amplitude histograms of a normal subjects 1 second ISI CNV.	139

	Page.
5-7a	Typical phase histograms of a normal subjects 1 second ISI CNV. 147
5-7b	Typical phase histograms of a normal subjects 4 second ISI CNV. 148
5-8	The averaged CNV's (1 second ISI) of the abnormal subjects. 161
5-9	The averaged CNV's (4 second ISI) of the abnormal subjects. 162
5-10a	Typical phase histograms of an abnormal subjects 1 second ISI CNV. 171
5-10b	Typical phase histograms of an abnormal subjects 4 second ISI CNV. 172
5-11a	Amplitude vs. phase for the first harmonic of the averaged 1 second ISI CNV's. 184
5-11b	Amplitude vs. phase for the first harmonic of the averaged 4 second ISI CNV's. 185
5-12a	Amplitude vs. phase for the second harmonic of the averaged 1 second ISI CNV's. 186
5-12b	Amplitude vs. phase for the second harmonic of the averaged 4 second ISI CNV's. 187
5-13	Amplitude vs. trial number for normal subjects 1 second ISI CNV's. Harmonic 2. 189
5-14	Phase vs. trial number for a normal subjects 1 second ISI CNV's. Harmonic 1. 190
5-15	Phase vs. trial number for an abnormal subjects 1 second ISI CNV's. Harmonic 1. 190
5-16a	Phase vs. trial number for a normal subjects 1 second ISI CNV's. Harmonic 2. 192
5-16b	Phase vs. trial number for an abnormal subjects 1 second ISI CNV's. Harmonic 2. 192
5-17	Amplitude vs. trial number for a normal subjects 4 second ISI CNV's. Harmonic 1. 193

5-18	Phase vs. trial number for the 4 second ISI CNV's of all the normal subjects. Harmonic 1.	194
5-19	Phase vs. trial number for the 4 second ISI CNV's of all the abnormal subjects. Harmonic 1.	195

1. Introduction

1.1 The Origins of Electroencephalography

Electrical engineers are frequently surprised to find that electrical activity of the brain was observed as long ago as 1875 [1]. Their surprise is however understandable when one considers the infant state of electrical engineering at that time. The discovery was due to a British Physiologist called Richard Caton. Caton used Thomson's (moving magnet) reflecting galvanometer to observe electrical fluctuations from the exposed surfaces of the brains of rabbits and monkeys. The potentials were of the order of millivolts and the necessary amplification was provided optically by the galvanometer. Using this galvanometer Caton was able to study and subsequently comment on the electrical activity he observed. His findings were presented at a Conference and published in the British Medical Journal in August 1875 [1]. Despite being unable to make graphic recordings of the activity, Caton detected background and stimulus related potentials.

However it was not until 1929 that Hans Berger [2] discovered the electroencephalogram (EEG) in man. He used a string galvanometer connected to electrodes attached to the scalp. Berger tried many different types of electrodes made from different metals. Unfortunately however Berger's work remained unnoticed for a number of years until Adrian and Matthews [3] (1934) and Jasper and Carmichael [4] (1935) reviewed and confirmed it.

1.2 Technological Change

Technological advances at this time made it possible for the electrical activity to be amplified and displayed on a cathode ray tube (CRT). Then resulting waveforms could then be photographed for a permanent record. These early valve amplifiers were usually a.c. coupled and often suffered from pick-up of external interference. An elegant solution to this problem was the advent and adoption of the differential amplifier which was able to reject the common-mode noise at its input. During the 1940's pen recorders became more widely available and for the first time electroencephalographers could have an immediate permanent record of the brain's electrical activity. Interest in the developing field of EEG analysis grew rapidly. Many workers tried to make objective quantitative analyses of the EEG. A physicist by the name of Dietsch [5] was probably the first worker to examine the frequency of the EEG signals. He performed Fourier analysis on short sections of EEG signals using a mechanical desk calculator in the early 1930's. The method was very tedious and it was not until 1943 that Walter [6] overcame this disadvantage with his frequency analyser. This instrument consisted of a bank of twenty tuned reeds covering the range 1.5 to 30Hz. The movement of each reed was used to switch a charging current into a capacitor and thus to integrate the activity over a ten second period. The outputs of the integrators being roughly proportional to the amount of activity in given frequency band.

During the late 1940's and early 50's a considerable

amount of interest was also shown in topographic EEG displays. Typically these devices employed twenty or more CRT's to display simultaneous EEG signals from different points on the scalp. Like the frequency analyser previously described the 'toposcopes' suffered from the instability of the valve technology of the day.

The development of new concepts in statistical communication theory give rise to the EEG signals being considered as a stochastic process. Brazier and Casby [7,8] (1951,1952) were some of the first workers to apply auto and cross-correlation analysis to the EEG. Subsequently this method became quite popular as a shortcut to obtain the power spectrum of the EEG signal.

The major developments of the late 1950's and early 60's were however in the advent and use of the new solid state technology. The equipment was more reliable and required far less maintenance and calibration than the valve equipment that it replaced. Furthermore the advent of the digital computer made the calculation of the power spectrum a much less awesome prospect although it was not until the introduction of the Fast Fourier Transforms (FFT) by Cooley and Tukey [9] in 1965 that this method became practical for multichannel work.

1.3 Evoked Potentials

In parallel with these developments in the analysis of the background EEG signal came developments in the analysis

of potentials evoked by some external stimulus. Dawson [10] (1947) was the first person to record potentials evoked by stimulation of a peripheral nerve. He used photographic superimposition whereby the waveform was displayed on a CRT whilst a camera, set for a time exposure, recorded the waveform on film. The stimulation was applied many times each causing a single scan of the CRT screen. The evoked potential is thus visible on the developed photograph due to the re-inforcing effect of the overlapping individual responses. This elaborate technique was necessary because the magnitude of the individual evoked responses were considerably smaller than the normal background EEG activity.

Development of the analysis of evoked potentials was limited by the lack of equipment capable of improving the signal to background EEG ratio. It was realised at an early stage that if it were possible to average a number of evoked responses an improvement in signal to background EEG would be achieved. Early analogue averagers were built but were cumbersome and difficult to use. The advent of digital memory and logic devices made averaging a much more attractive technique and during the early 1960's many discoveries were made in the field of evoked responses. Furthermore, the advent of the general purpose laboratory minicomputer meant that evoked responses could be studied without further special equipment (except the stimulator). It was during the study of certain evoked potentials that Walter [11] discovered a new evoked potential which later became known as the Contingent Negative Variation (CNV). This evoked potential was found to occur between a pair of conventional auditory or visual

stimuli provided that the patient was required to perform some action on receiving the second stimuli. The response was found to be a gradual negative shift subsequent to the first, or warning stimulus, and increasing in negativity until the second stimulus when the desired action was performed. Unlike normal visual, auditory or somatosensory evoked responses, the CNV depended on active participation by the subject and thus involves the higher mental processes.

1.4 Additivity and Ordering

Since their discovery, it had always been thought that evoked potentials were due to an additional signal component which was added to the background EEG signal. Professor Sayers [12] questioned this basic assumption and performed several tests to try and establish whether the Auditory Evoked Potential (AEP) could be caused by some other mechanism. One of Sayers tests [12] involving taking a section of normal background EEG and re-arranging the phases of the Fourier components. He found that by doing this the section of background EEG could be made to resemble an AEP. Sayers also found that if the energy contained in an AEP was calculated and compared with that for a section of background EEG, there was no significant difference [12]. This led Sayers to the conclusion that AEP's could in fact be due not to an additional signal, but to some form of phase ordering of the on-going background EEG signal. One could, for example, envisage a number of EEG generators becoming entrained for a short while subsequent to the stimulus presentation. This would give the characteristic shape of the AEP

but would not change the energy content of the EEG.

In order to try and verify this, Professor Sayers obtained the Fourier transforms of a set of AEP's. A diagram was then constructed showing the phase angles of the transformed responses for each harmonic [12]. From these diagrams Sayers was able to detect that the phase angles obtained did not form an uniform pattern between $+\pi$ and $-\pi$ but formed groups.

In subsequent studies Sayers used histograms to show these phase ordering effects [13]. For each of a number of the harmonic frequency components histograms were plotted to show the number of times the phase angle fell into any one of twenty-four frequency intervals of width $2\pi/24$ radians (i.e. 15°). From these histograms Sayers observed grouping in harmonics 2 to 5 (the fundamental frequency in these observations was 1.5625 Hz.) The amount of phase aggregation was found to be dependent on the stimulus level but not on the degree of latency. When no stimulus was applied the phases of the transformed EEG formed a roughly uniform distribution [13].

The importance of Sayers' findings were such that it was considered desirable to carry out similar experiments on CNV's to establish whether they were also due to phase re-ordering as had been previously suggested by Walter [14].

1.5 An Application of the CNV to Diagnostic Medicine

Since it was first reported in 1964 the CNV has not been widely used in routine diagnostic medicine. One possible area in which the CNV may be useful is in the early diagnosis of a hereditary neurological disorder known as Huntington's Chorea (HC). This illness affects several areas of the brain including the cortex, the caudate nucleus and other parts of the corpus striatum [15]. These are all areas thought to be involved in the generation of the CNV [16]. The disease is currently diagnosed only in its later stages, usually in middle age, when the chorea (uncontrollable movement) becomes apparent. The condition becomes progressively worse causing pre-senile dementia and after a distressing long illness, ultimately leads to premature death.

A possible diagnostic procedure would therefore be to compare the CNV's of those people at risk (i.e. those with a known family history of HC) with those obtained from a normal population. Significant deviation from the normal CNV may thus indicate the presence of the disease before the other symptoms become apparent. Clearly the clinicians would have to establish the limits of the normal CNV and also whether other neurological conditions could give similar CNV's to those of HC victims. However before this can take place it is necessary to conceive, design, build and test equipment and processing methods suitable for extracting a reliable quantitative description of the CNV.

References for Section 1

- [1] Caton R, "The Electric Currents of the Brain"
British Medical Journal, Volume 2, page 278, August 1875.
- [2] Berger H, "Über das Elektrenkephalogramm des Menschen"
Archiv für Psychiatrie and Nervenkrankheiten, Volume 87
pp 527-570, 1929.
- [3] Adrian E D, Matthews B H C, "The Berger Rhythm: Potential
Changes from the Occipital Lobes in Man"
Brain, Volume 57, Part 4, pp 355-385, December 1934.
- [4] Jasper H H, Carmichael L, "Electrical Potentials from
the Intact Human Brain"
Science, Volume 81, pp 51-53, January 1935.
- [5] Dietsch G, "Fourier-Analyse von Elektroencephalogrammen
des Menschen"
Pflugers Archiv für die Gesamt Physiologie des Menschen
und der Tiere, Volume 230, pp 106-112, 1932.
- [6] Walter W G, "An Automatic Low Frequency Analyser"
Electronic Engineering, Volume 16, pp 9-13, June 1943.
- [7] Brazier M A B, Casby J U, "An application of the M.I.T.
digital electronic correlator to a problem in the EEG".
Electroencephalography and Clinical Neurophysiology
Volume 3, page 375, 1951.
- [8] Brazier M A B, Casby J U, "Cross correlation and auto-
correlation studies of electroencephalographic
potentials"

- [8] continued
Electroencephalography and Clinical Neurophysiology,
Volume 4, pp 201-211, 1952.
- [9] Cooley J W, Tukey J W, "An Algorithm for the Machine
Calculation of Complex Fourier Series"
Mathematics of Computation, Volume 19, pp 297-301, 1965
- [10] Dawson G D "Cerebral Responses to the Electrical Stim-
ulation of the Peripheral Nerve in Man"
Journal of Neurology, Neurosurgery and Psychiatry,
Volume 10, pp 134-140, 1947.
- [11] Walter W G, Cooper R, Aldridge V J, McCallum W C,
Winter A L, "Contingent Negative Variation: An Electric
Sign of Sensorimotor Association and Expectancy in the
Human Brain".
Nature, Volume 203, pp 380-384, July 1964.
- [12] Sayers B McA, Beagley H A, Henshall W R, "The Mechanism
of Auditory Evoked EEG Responses"
Nature, Volume 247, pp 481-483, February 1974.
- [13] Sayers B McA, Beagley H A, Riha J, "Pattern Analysis of
Auditory-Evoked EEG Potentials"
Audiology, Volume 18, pp 1-16, 1979.
- [14] McCallum W C, Knott J R, (Editors) "Event-Related Slow
Potentials of the Brain: their relation to behaviour"
Electroencephalography and Clinical Neurophysiology,
Supplement No. 33 Part II Discussion, page 128
Published by Elsevier Scientific Publishing Company 1973.

- [15] Scott D F, Heathfield K W G, Toone B,
Margerison J H,
"The EEG in Huntington's chorea: a clinical and
neuropathological study"
Journal of Neurology, Neurosurgery and Psychiatry,
Volume 35, pp 97-102, 1972
- [16] McCallum W C, Papakostopoulos D, Gombi R,
Winter A L, Cooper R, Griffith H B,
"Event Related Slow Potential Changes in Human
Brain Stem"
Nature, Volume 242, pp 465-467, 1973.

2. Theoretical Development

2.1 Models for Additivity and Phase Ordering in Evoked Responses

Although introduced as being mutually exclusive, there is no fundamental reason why an evoked response should not be the result of both additive and phase re-ordered components. Furthermore any added signal giving the characteristic shape of either an AEP or a CNV will have its own well defined phase spectrum. Thus the identification of a phase pattern in an evoked response (CNV or AEP) is not in itself sufficient evidence for phase re-ordering. At this point it is as well to define what is meant by phase ordering and phase re-ordering. Phase ordering is used here to describe the situation where a phase spectrum has a recognizable pattern irrespective of the cause of this pattern, whereas phase re-ordering is used to describe the situation where the phases of an existing signal have been changed so as to cause phase ordering. Since phase patterns are not sufficient evidence for phase re-ordering, it is necessary to consider amplitude characteristics as well.

Consider first a finite realisation of the pre-stimulus background EEG signal which is of the same length as the section of EEG containing the evoked response. Because the signal is random the Fourier harmonic components into which it may be analysed will have random amplitudes and phases. If the n th harmonic is selected it may be very conveniently represented in amplitude and phase by a phasor on a phasor diagram. The same procedure can be carried out for each pre-stimulus realisation recorded in a series of trials, i.e. the

nth harmonic component of each realisation may be plotted as a phasor on the same phasor diagram. This procedure will result in a diagram such as is shown in Fig. 2-1a, in which the amplitudes are random and the phase angles, which are also random, are distributed approximately uniformly about a circle. For ease of comprehension a circle of arbitrary radius has been superimposed on the phasor diagram. The crosses on the circles indicate the directions of the phasors.

Now consider the effects of a stimulus on this diagram. Assume that the phasor diagram for the post-stimulus section of EEG containing the evoked response may be derived directly from the pre-stimulus phasor diagram. If the evoked response were due to phase re-ordering effects its nth harmonic phasor diagram would be obtained by rotating all of the phasors of Fig. 2-1a towards the preferred phase angle. Thus the phasor diagram of Fig. 2-1b would be obtained. The characteristic feature is that although there is phase ordering present the amplitudes are unaltered. The phase ordering present is responsible for the characteristic waveform of the evoked response. This phase re-ordering model may now be compared with the additive signal model. This will be considered for the two cases of low and high level stimulus. In the low level stimulus case it is assumed that a small additive evoked response is produced which may be analysed into its harmonic components. The nth harmonic component, assumed to be the same in each realisation, is represented as a small phasor which has to be added to each of the phasors of Fig. 2-1a to produce the post-stimulus phasor diagram. This produces Fig. 2-1c. This figure shows that a certain amount of phase ordering is produced while the amplitudes of the nth

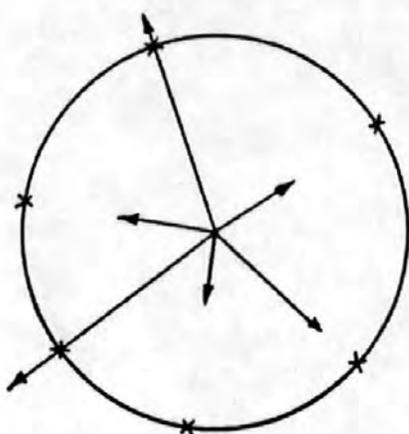


Fig. 2-1a
Background EEG

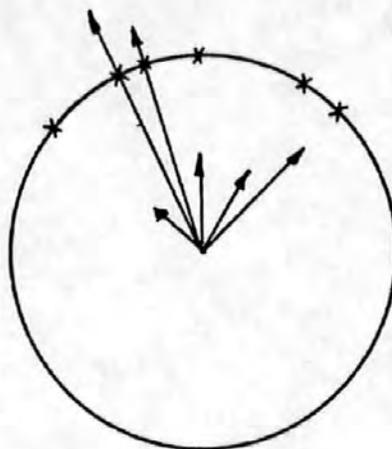


Fig. 2-1b
Phase ordering

Figure 2-1
Phasor diagrams

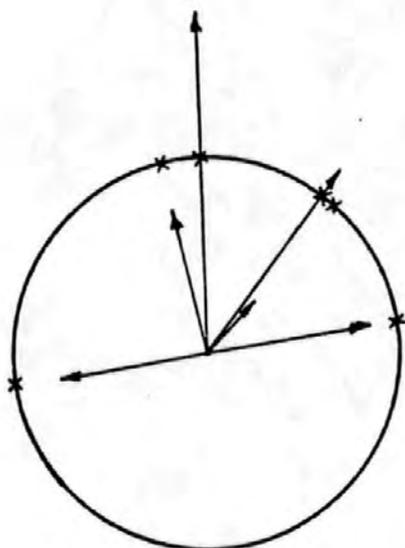


Fig. 2-1c
Small additive component

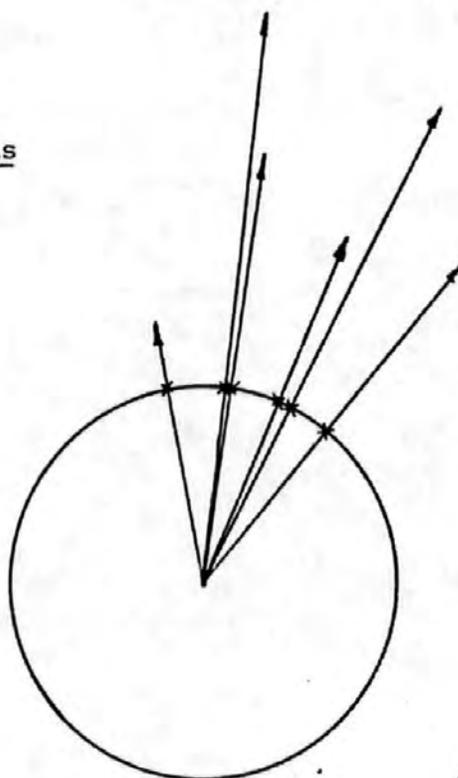


Fig. 2-1d
Large additive component



harmonic are changed. Those phasors directed towards the preferred angle are increased in length, whilst those opposed are decreased. For a small additive signal the average length of the phasors would be virtually unaltered. The same argument may be applied in the case of a high level stimulus. The phasor to be added to Fig. 2-1a is now large and the post-stimulus phasor diagram is as shown in Fig. 2-1d. It is seen that the large additive signal results in pronounced phase ordering and a considerable increase in phasor amplitudes.

The models clearly demonstrate that phase ordering will be produced by either mechanism and will increase with stimulus level. The presence of an additive component may be inferred by changes in amplitude of the nth harmonic. In the next section the statistical tests used to detect phase ordering and amplitude and energy changes will be described.

2.2 Tests of the Models

Various processing methods and tests were devised in an attempt to determine which of the above models was the most appropriate to the CNV. Because of the variable nature of the CNV, both from subject to subject, and also from trial to trial with the same subject, evidence for phase ordering has to be sought on a statistical basis. It is possible to detect phase ordering by Fourier transforming the 'Negative Variation' sections of a sequence of CNV's (i.e. that section of the CNV remaining when the two involuntary stimulus responses were ignored) and testing the phase values. One method of visually detecting phase ordering would be to plot phase histograms, whereas the variability of the responses

could be judged by plotting amplitude histograms.

2.2.1 Histograms of Phase and Amplitude

The phase values for each harmonic frequency component could be grouped into bands of known angular width such that the range $-\pi$ to π was covered in a number of steps. Histograms could then be plotted showing, for a particular frequency, the distribution of the phase values. If phase ordering were present then the phase histograms would be expected to show aggregation about some particular phase value. Indeed if the responses were identical, then in the absence of any noise the histograms would show a given frequency component of every response as having the same phase value. However if no phase ordering were present then the phase histograms would be expected to exhibit an uniform distribution of phase over the range $-\pi$ to π .

By considering the shape of an ideal CNV response [1] it is possible to predict the values, for each harmonic frequency, around which phase grouping should take place. Figure 2-2 shows an idealised CNV of about one second inter-stimulus interval. Ignoring the evoked responses R_1 and R_2 the responses may be considered as a linear function of time. This may be Fourier transformed as follows:-

$$X(n) = \frac{1}{N} \sum_{i=0}^{N-1} \left[x(i) e^{\frac{-2\pi jin}{N}} \right] \dots\dots\dots (1)$$

$$= \frac{1}{N} \sum_{i=0}^{N-1} \left[x(i) \left(\cos\left(\frac{2\pi in}{N}\right) - j \sin\left(\frac{2\pi in}{N}\right) \right) \right] \dots\dots (2)$$

but $x(i) = -ki$ over the range $0 \leq i < N$

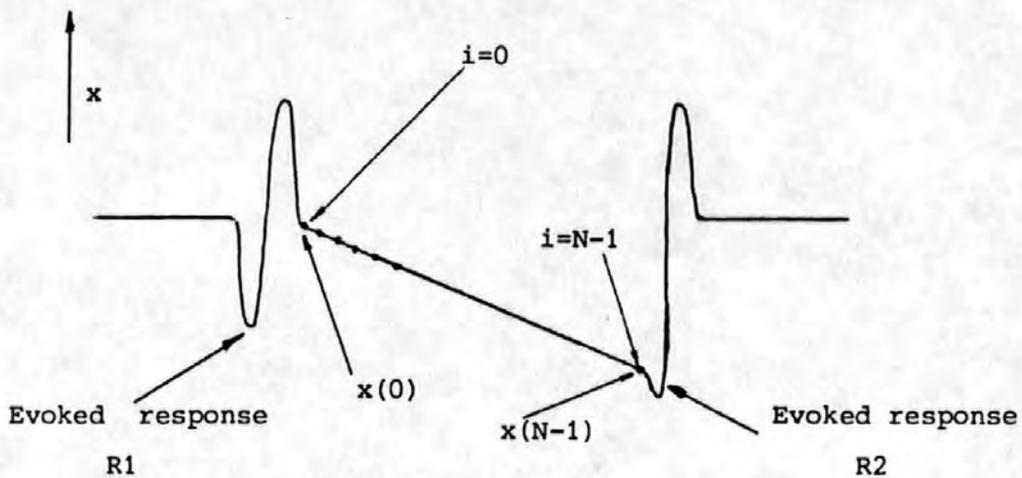


Figure 2-2
An idealized C.N.V response.

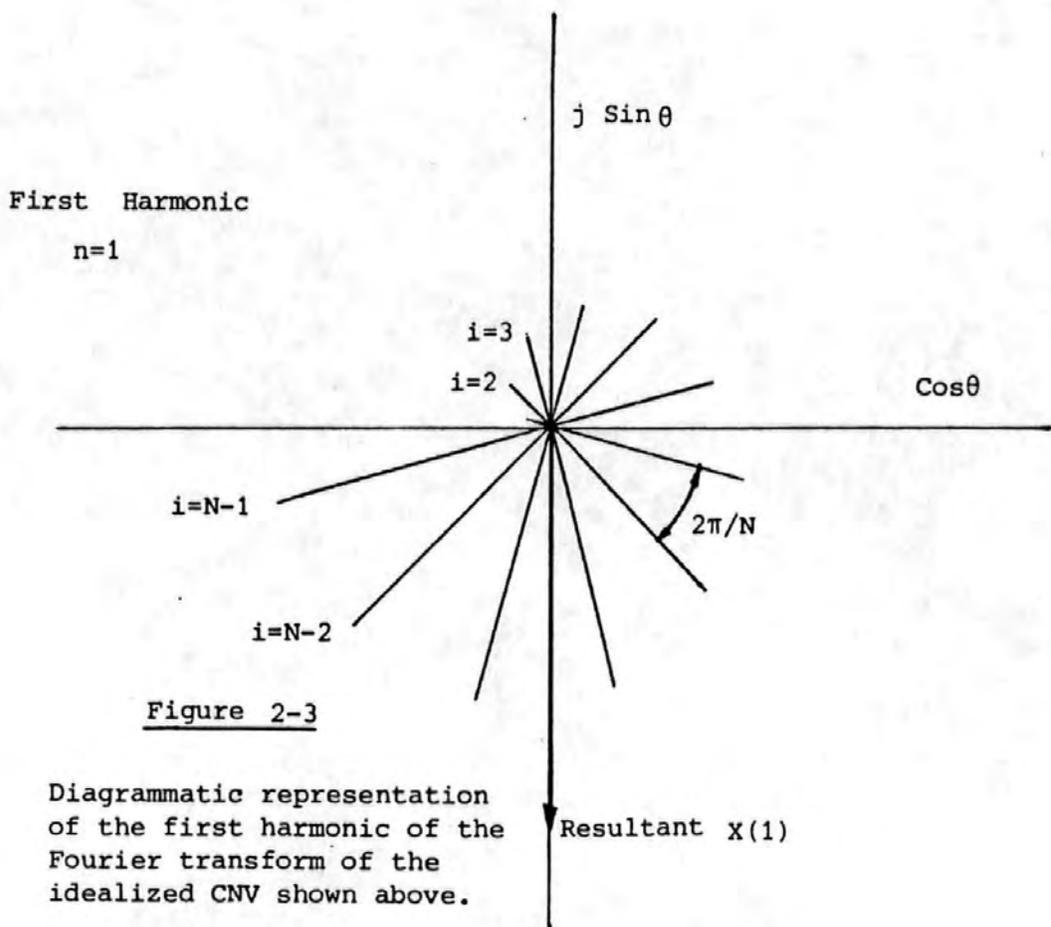


Figure 2-3

Diagrammatic representation of the first harmonic of the Fourier transform of the idealized CNV shown above.

$$X(n) = \frac{1}{N} \sum_{i=0}^{N-1} \left[-ki \cos\left(\frac{2\pi in}{N}\right) \right] + \frac{j}{N} \sum_{i=0}^{N-1} \left[ki \sin\left(\frac{2\pi in}{N}\right) \right] \dots (3)$$

The first harmonic component is thus

$$X(1) = \frac{1}{N} \sum_{i=0}^{N-1} \left[-ki \cos\left(\frac{2\pi i}{N}\right) \right] + \frac{j}{N} \sum_{i=0}^{N-1} \left[ki \sin\left(\frac{2\pi i}{N}\right) \right] \dots (4)$$

The individual terms of these summations may be represented as shown in Figure 2-3. The resultant (i.e. the first harmonic component) is clearly in either the third or fourth quadrant. In fact calculation shows that for large N the angle approaches -90° . Thus the phase histograms would be expected to show aggregation at this phase angle. Additional calculations (see Appendix 8.1) showed that the phase angles of all of the first 6 harmonics would aggregate at -90° . Furthermore, for longer CNV's (i.e. those with a longer ISI) which often change shape after approximately two seconds [2] the phase angles are also in the third and fourth quadrants. This suggests that all normal CNV's should have phase histograms which show aggregation in the third and fourth quadrants.

One disadvantage of phase histograms is that the phase data is cyclic yet the histogram axis is not. This disadvantage can be illustrated by assuming that, for example, an evoked response has a phase which tends towards π . Instead of a group of phase 'bins' at π one would observe a cluster at π and a cluster at $-\pi$ (i.e. at opposite ends of

the angular scale). This problem can be overcome by using either Rose diagrams (see Figure 2-4) or by using the phasor diagrams previously described. Whichever method of display is chosen the question to be resolved remains the same. This is simply whether the phase ordering observed occurs by chance or whether there is some mechanism forcing the phasors to some preferred direction.

Histograms could also be plotted for the amplitude information. Provided the background EEG were negligible, these would give information about the variability of the CNV responses. If, for example, each individual CNV response was identical to the next, then the amplitude histograms would show only one value of amplitude for each frequency component. If, on the other hand, the responses were very variable then the amplitude histograms would show many values of amplitude i.e. a spread of amplitude. This may be useful in the classification of certain subject categories. One difficulty which arises with amplitude histograms is that of the choice of the amplitude scale and interval. Unfortunately this choice might be critical in assessing whether one subjects responses are more variable than another's.

2.2.2 Angular Statistical Tests for Phase Ordering

To determine whether the grouping of a set of phasors could have occurred by chance it is necessary to perform angular statistical tests on the harmonic components of the sample of individual CNV's. These tests are described below. All three tests are non-parametric.

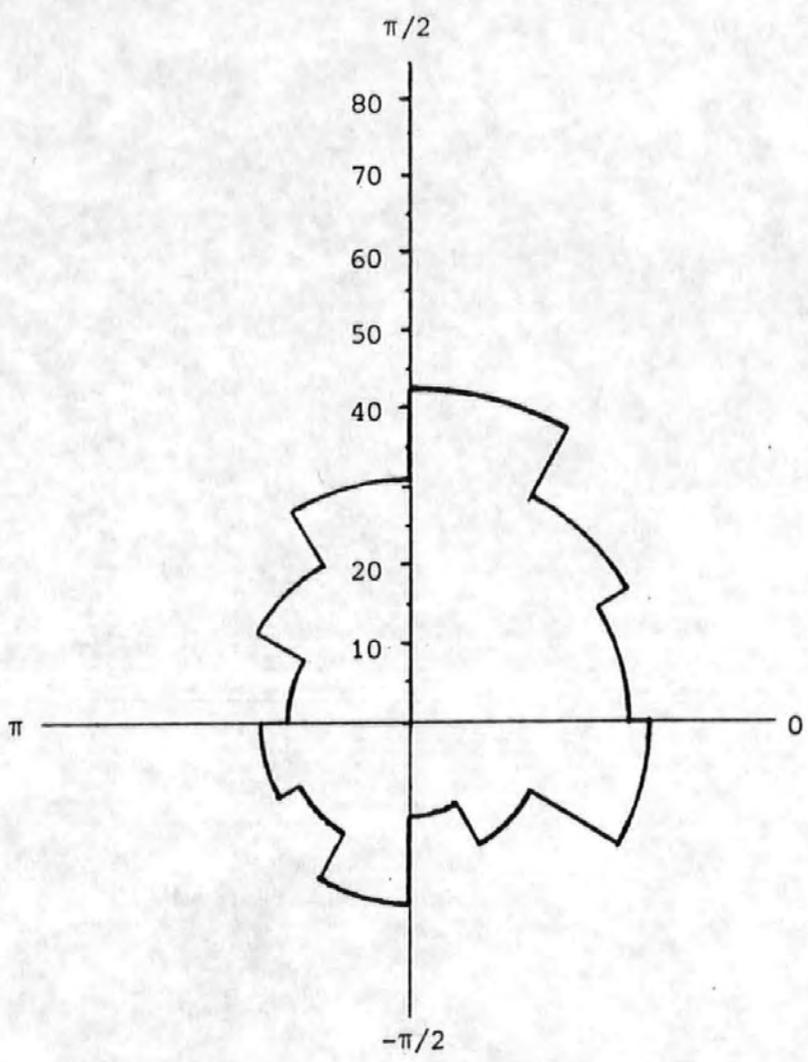


Figure 2-4

The Rose diagram

2.2.2.1 The Rayleigh Test of Circular Variance

This is the first of a number of tests used to determine whether a set of N phase angles $\{\theta_i\}$ are distributed in a non-uniform manner. It could therefore be used to detect phase ordering.

The circular variance can be calculated as follows:- [3]

$$\text{Circular Variance So} = 1 - \bar{R} \quad \dots\dots\dots (5)$$

$$\text{where } \bar{R} = \sqrt{\bar{C}^2 + \bar{S}^2} \quad \dots\dots\dots (6)$$

$$\text{and } \bar{C} = \frac{1}{N} \sum_{i=1}^N \cos \theta_i \quad \dots\dots\dots (7)$$

$$\bar{S} = \frac{1}{N} \sum_{i=1}^N \sin \theta_i \quad \dots\dots\dots (8)$$

Clearly if $\theta_1 = \theta_2 = \dots\dots\dots = \theta_N = \theta$

then $\bar{C} = \cos \theta$ and $\bar{S} = \sin \theta$

$$\text{This gives } \bar{R} = \sqrt{\cos^2 \theta + \sin^2 \theta} = 1 \quad \dots\dots\dots (9)$$

$$\text{So } = 0 \quad \dots\dots\dots (10)$$

This corresponds to the perfect phase ordering situation where all the phase angles are the same.

Alternatively consider the situation where $\theta_i = \frac{2\pi i}{N}$
i.e. $\{\theta_i\}$ are distributed uniformly over the range 0 to 2π .
In this case the summations \bar{C} and \bar{S} are both zero.

Hence $\bar{R} = 0$ (11)

So $= 1$ (12)

This corresponds to the situation where all the phase angles are uniformly distributed. Thus to determine whether a set of angles are distributed in a significantly non-uniform manner, tables of the Rayleigh distribution must be consulted. However the commonly tabulated Rayleigh distribution probabilities [3] are in terms of R not \bar{R} . These two quantities are related by the expression $R = N\bar{R}$. (In the broadest sense \bar{R} is more meaningful than R since \bar{R} always lies between zero and one whereas R lies between zero and N).

Alternatively, however the tables can be transformed to yield significance levels for So in place of R. Details of this transformation and the resulting tables for So are given in Appendix 8.2

2.2.2.2 The Modified Rayleigh Test of Circular Variance

In an attempt to take both the amplitude and the phase angle into consideration Johnson [4] suggested the use of the modified test statistic T_o given by

$$T_o = 1 - \sqrt{\frac{\left[\frac{\sum_{i=1}^N r_i \cos \theta_i}{\sum_{i=1}^N r_i} \right]^2 + \left[\frac{\sum_{i=1}^N r_i \sin \theta_i}{\sum_{i=1}^N r_i} \right]^2}{2}} \quad \dots (13)$$

Where r_i is the length of the i th phasor

θ_i is the phase angle of the i th phasor

If $\{\theta_i\}$ are all aligned (i.e. $\theta_1 = \theta_2 \dots \dots \dots = \theta_N$) then the statistic $T_o = 0$ whereas if the angles are uniformly distributed over the range $0 - 2\pi$ and all the phasors have the same length then $T_o = 1$.

Unfortunately however the distribution of T_o is not easily obtainable and is likely to depend critically on the assumptions made in deriving it. However, Moore [5], uses the rank of the phasor magnitudes rather than their magnitudes and thus avoids this problem. Thus a new statistic U_o may be defined.

$$U_o = 1 - \sqrt{\left[\frac{\sum_{i=1}^N R_i \cos \theta_i}{\sum_{i=1}^N R_i} \right]^2 + \left[\frac{\sum_{i=1}^N R_i \sin \theta_i}{\sum_{i=1}^N R_i} \right]^2} \dots (14)$$

Where R_i is the rank of the i th phasor.

This is closely related to the statistic R^* proposed by Moore [5] and significance levels for U_o may be obtained from those for R^* by use of the formula

$$R^* = \frac{(1 - U_o)(N + 1)}{2\sqrt{N}} \dots \dots \dots (15)$$

Tabulated values of the probabilities for U_0 are given in Appendix 8.3. Where calculated values of U_0 are significantly different from unity then the phasors are non-uniformly distributed.

2.2.2.3 The Hodges-Ajne Test

This is an alternative test used to determine whether a set of angles are distributed in a non-uniform manner. The test statistic, m is given by the minimum number of observations lying in any semi-circle. If the value of m is small in relation to the number of observations, N then the angles are non-uniformly distributed. For given values of m and N the significance level of the test may be calculated from the formula [6].

$$\text{Significance level} = \frac{(N-2m) \binom{N}{m}}{2^{N-1}} \times 100\% \dots\dots\dots (16)$$

Provided $m < \frac{N}{3}$

This test is similar in principle to that used by Sayers [7]. A table of significance levels is given in Appendix 8.4.

2.2.3 Tests of Additivity

These tests were used in an attempt to detect changes in the energy content of the evoked responses. These tests are described below.

2.2.3.1 Pre - and Post - Stimulus Energy Tests

The energy in the pre- and post-stimulus records was

compared by means of a two tailed t-test. For a section of pre-stimulus EEG the mean square value was calculated using the formula

$$E = \frac{1}{N} \sum_{i=1}^N x_i^2 \dots\dots\dots(17)$$

where E is proportional to the signal energy

N is the number of data points over which E is to be calculated

x_i is the ith value of the signal

This calculation was repeated for the section containing the CNV response and the difference between the two values was noted. For one second ISI CNV's the lengths of the pre- and post- stimulus sections were the same (N = 80 or 640 ms) but for the 4 second ISI CNV's only 200 points (1.6 seconds) of pre- stimulus information was available whereas 400 points (3.2 seconds) were included to encompass the CNV response.

The differences for each of the thirty-two trials were then averaged and the mean of the differences was subjected to a two-tailed t-test (see Appendix 8.5) to determine whether it was significantly different from zero. A non-significant mean value indicated no statistical difference between pre- and post- stimulus energy. A significant positive value indicated that the post-stimulus energy exceeded the pre-stimulus energy, whilst a significant negative value indicated the reverse.

2.2.3.2. Pre - and Post - Stimulus Mean Amplitude Differences Test

This test was also a paired two-tailed t-test. For

each of the thirty-two trials the differences between the corresponding pre- and post-stimulus phasor lengths for a given harmonic component were formed. The mean of the differences was calculated and tested to establish whether it was significantly different from zero. A significant positive result indicated that the evoked potential was associated with an additive effect at the nth harmonic, although it gave no indication as to the mechanism involved. Because of the limited amount of pre-stimulus information (see section 2.2.3.1) it was not possible to perform this test on the four second ISI CNV's. (See also Appendix 8.5)

2.2.3.3 Nearest and Furthest Mean Amplitude Test

This test investigated the variation of amplitude with phase angle in the post-stimulus nth harmonic phasor diagram. Increased amplitudes in the direction of preferred phase combined with decreased amplitudes in the opposed direction would be evidence for an additive effect. The mean length of that half of the vectors whose phase angles lay within the smallest arc was calculated as was that of the remaining vectors. A one-tailed t-test was then performed to determine whether the former mean value was greater than the latter. In order to allow for the possibility of unequal variances, a correction was made to the degrees of freedom used in these statistical tests (See Appendix 8.6). A significant result would provide evidence for an additive effect. This test is not infallible, however. The additive signal might combine with an oppositely directed phasor to produce a small phasor in the smallest arc, or alignment of all the phasors in the

preferred direction would also render the test unreliable. For these reasons the results of this test should not be examined without reference to the tests for phase ordering previously described.

2.3 Eye Movement Corrections

At an early stage of the investigation it was found to be necessary to remove the effects of eye movements from the measured EEG signal. These effects are due to a steady potential between the cornea and the retina of each eye. When the eyes are moved the proportion of this potential detected at the scalp electrodes varies and hence an artefact related to the ocular position is superimposed on the EEG signal. Because of the relative magnitudes of the EEG and the corneo-retinal potential the artefact introduced is considerably greater than the EEG. Thus the EEG may be completely obscured by the artefact. A number of workers [8,9,10,11,12,13] had published details of methods for removing the artefact but no record of a comparison of the available techniques could be found.

One method of removing the effects of eye movements from the EEG was proposed and used by McCallum and Walter [8]. The method is based on the use of a potentiometer to balance out the effects of vertical eye movements. One end of the potentiometer was connected to a mid-frontal electrode and the other end to an electrode placed on the mastoid processes as shown in Figure 2-5. The centre tap of the potentiometer was used as a reference for a vertex EEG electrode. The operation of the circuit is best explained by the use of

a simplified equivalent circuit as depicted by Figure 2-6. The EOG signal (V_{EOG}) is shunted by the resistors R_1, R_2, R_3 in series. These resistors represent the tissue etc. surrounding the eye. At point P_1 along this resistance chain the EEG generator may be considered to be connected. Another point (P_2) represents the mastoid process. If the EEG signal is measured between points P_3 and P_2 then the observed signal will be the sum of both the EEG and a fraction of the EOG. The fraction of the EOG signal contained in the EEG will be given by $R_3/(R_1 + R_2 + R_3)$.

$$\therefore V = V_{EEG} + \frac{R_3 V_{EOG}}{R_1 + R_2 + R_3} \dots\dots\dots (18)$$

If, however, the potentiometer is connected as shown in Figure 2-7a then by adjusting the position of the wiper the vertical EOG component can be balanced out. The point of balance can be derived from the re-drawn circuit Figure 2-7b. The current, I , from the EOG generator splits into I_1 and I_2 at the connection of R_5 and R_1

$$I_1 = \frac{R_1 + R_3}{R_1 + R_3 + R_5} I \dots\dots\dots (19)$$

$$I_2 = \frac{R_5}{R_1 + R_3 + R_5} I \dots\dots\dots (20)$$

The potentiometer, R_5 , may be represented as two resistors in series the common point being the wiper. Thus the two resistors have the values kR_5 and $(1-k)R_5$ where

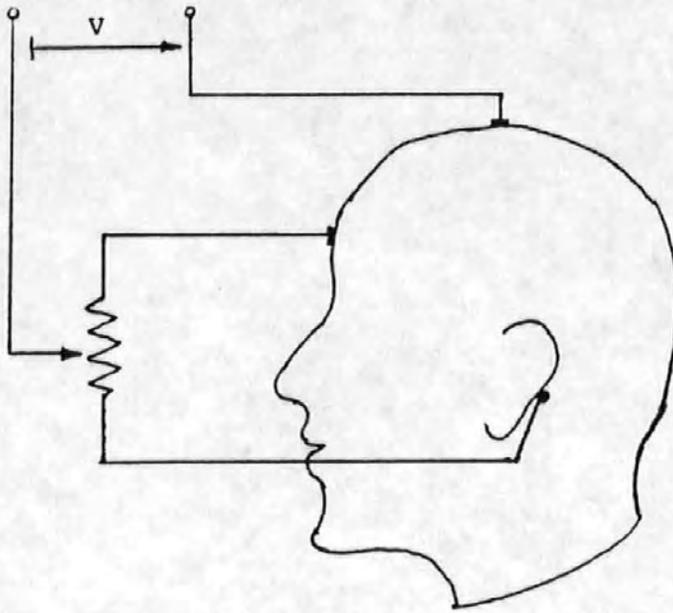


Figure 2-5

Eye movement correction method
due to McCallum and Walter.

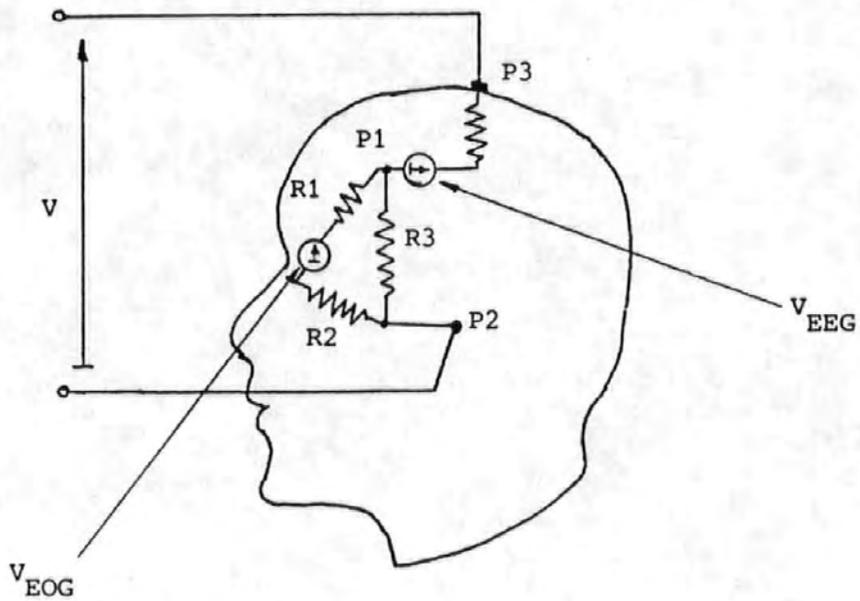


Figure 2-6

The simplified equivalent circuit
without potentiometer

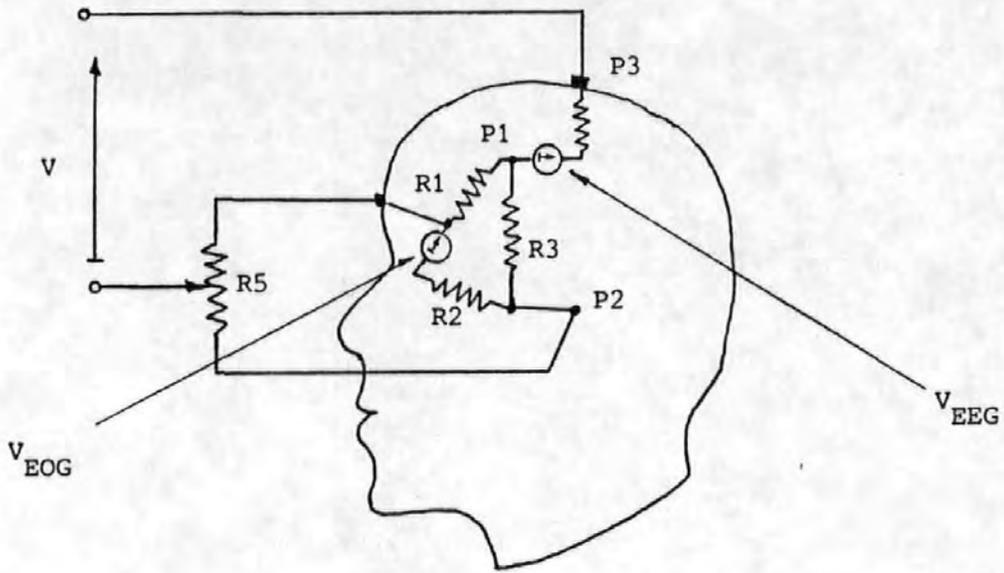


Figure 2-7a

The simplified equivalent circuit with potentiometer.

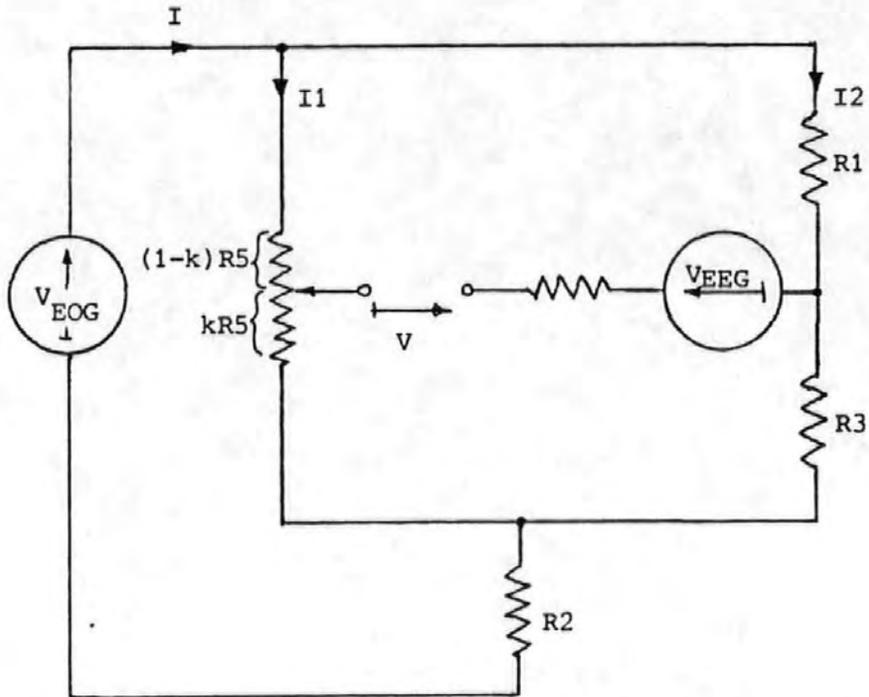


Figure 2-7b The re-drawn equivalent circuit.

k is a fraction between 0 and 1 representing the setting of the potentiometer. When $I_1 k R_5$ equals $I_2 R_3$ then the EOG signal will be eliminated from the EEG.

Thus

$$I_1 k R_5 = I_2 R_3 \dots\dots\dots (21)$$

$$\frac{R_1 + R_3}{R_1 + R_3 + R_5} I k R_5 = \frac{R_5}{R_1 + R_3 + R_5} I R_3 \dots\dots (22)$$

$$\therefore (R_1 + R_3) k = R_3 \dots\dots\dots (23)$$

$$\therefore k = \frac{R_3}{R_1 + R_3} \dots\dots\dots (24)$$

Since $\frac{R_3}{R_1 + R_3}$ is always in the range 0 to 1 for positive values of R_1 and R_3 the balance point can always be attained.

A variation of this technique was proposed by Girton and Kamiya [9] and is shown in Figure 2-8. The observed EEG signal is again assumed to be the sum of the true EEG and a fraction of the EOG signals. Unlike McCallum and Walters method, this technique allows for the independant correction of both horizontal and vertical components of the EOG in the EEG. The horizontal and vertical components of the EOG are obtained from electrodes placed around the eyes and are amplified by the differential amplifiers A_1 and A_2 . The EEG signal is amplified by a further differential amplifier A_3 connected to electrodes at the vertex and at the mastoid processes. Fractions of the amplified EOG components are then tapped off by means of the potentiometers R_1 and R_2 .

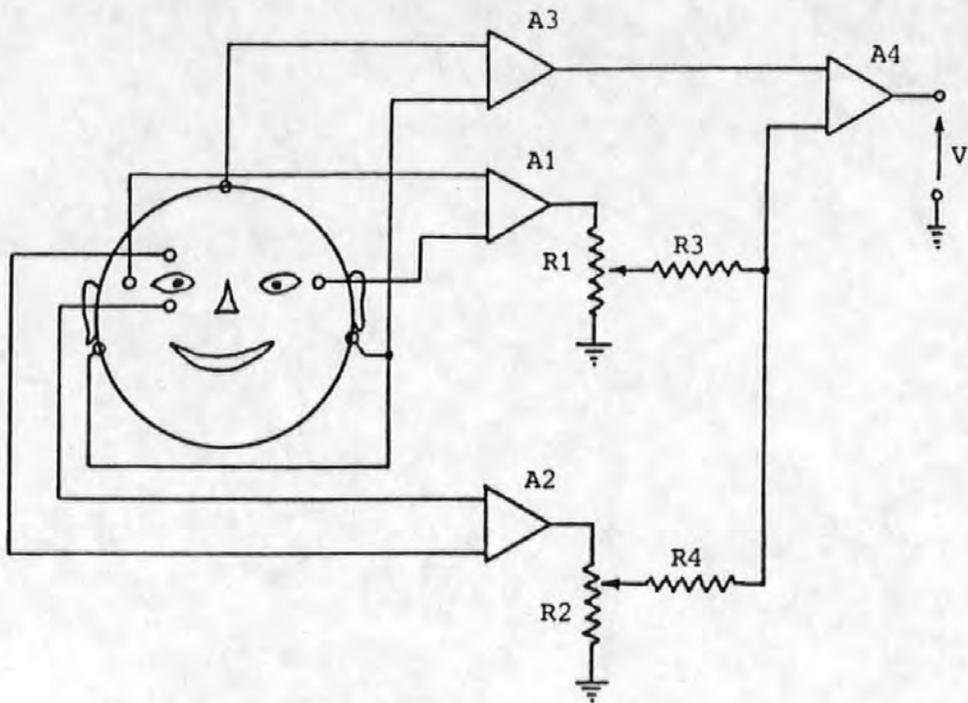


Figure 2-8

The method of Girton and Kamiya

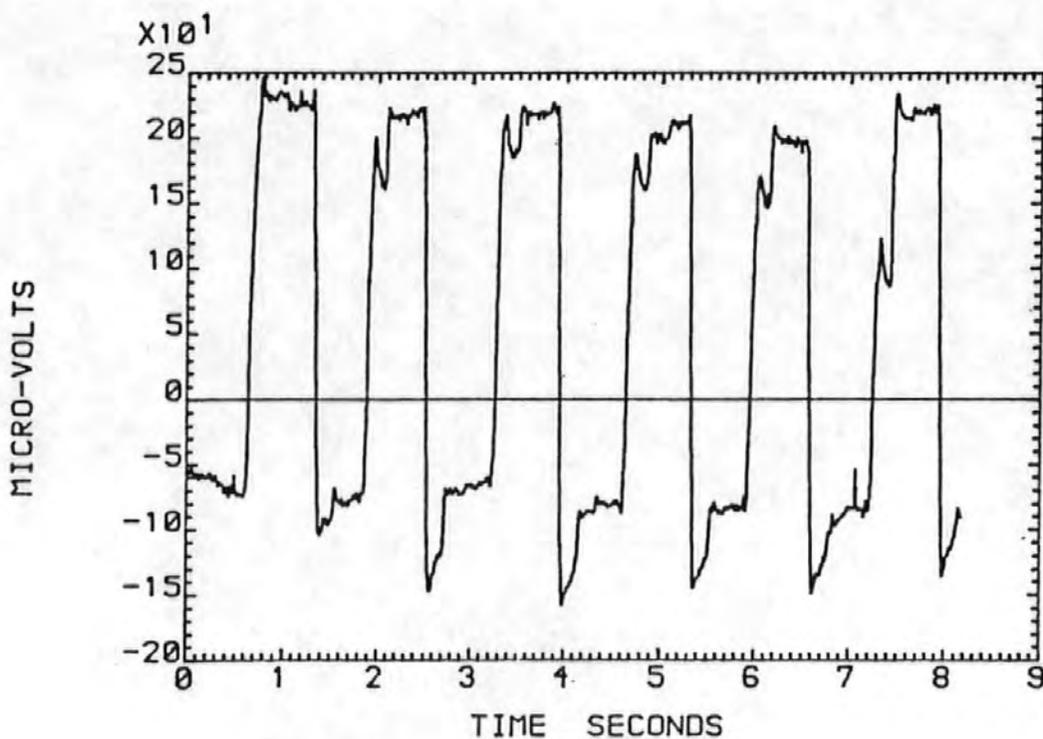


Figure 2-9

A typical vertical left EOG signal during the eye movement correction experiments

These signals are then summed by the resistors R_3 and R_4 . Finally the summed EOG components are subtracted from the EEG by means of a further differential amplifier A_4 . Thus the output signal V consists of the corrupted EEG signal less fractions of the horizontal and vertical components of the EOG. By adjusting the potentiometers R_1 and R_2 these fractions can be chosen so as to remove the corrupting EOG components in the output signal.

Both these techniques (McCallum and Walters and Girton and Kamiya) rely on the manual adjustment of the potentiometer(s) to obtain maximum artefact rejection. These adjustments are normally made whilst observing the chart output and are thus subjective. As the balance point (i.e. the point of maximum artefact rejection) is approached so the effect of further adjustment becomes more difficult to assess because of the masking nature of the background EEG. The process of adjustment is also rather slow. If, for example, it takes one minute to adjust each potentiometer, then for a 16 channel recording it would take about a quarter of an hour to set the potentiometers for McCallum and Walters method or over half an hour for Girton and Kamiya's method. This would normally be quite unacceptable.

2.4 Assessment of the Eye Movement Correction Methods

The quantitative assessment of the different correction procedures was based on the knowledge that the autocorrelation function (a.c.f.) of a rectangular waveform is triangular. [14] In the experiments described in Section 3.3 the almost periodic eye movements produced an EOG which was

nearly a rectangular wave (Figure 2-9) and so the measured EEG's were contaminated by a nearly rectangular wave. It is shown below that incomplete correction may be detected by the presence of a triangular component of the same period as the EEG present in the a.c.f. of the corrected EEG. If the a.c.f. showed no triangular components of similar period to the EOG, then the correction procedure was effective. If the a.c.f. fell rapidly to an average value of zero the efficiency of the corresponding correction was high. This method allowed very small residual EOG signals to be detected in the background EEG activity even in cases where they were visually indiscernable in the corrected waveform.

Let $Q(t)$ be the EEG signal after incomplete correction

$$\text{Then } C_{qq}(\tau) = \epsilon[Q(t)Q(t+\tau)] \dots\dots\dots(25)$$

Where $C_{qq}(\tau)$ is the autocovariance of Q at lag τ , ϵ denotes the expected value and all signals are adjusted to have zero mean value.

$Q(t)$ may be considered to be the sum of the uncorrupted EEG signal $E(t)$ and the remaining artefact $I(t)$, so that

$$Q(t) = E(t) + I(t) \dots\dots\dots(26)$$

Hence

$$C_{qq}(\tau) = \epsilon\{[E(t) + I(t)] \cdot [E(t+\tau) + I(t+\tau)]\} \dots\dots(27)$$

$$= \epsilon[E(t) \cdot E(t+\tau)] + \epsilon[I(t) \cdot I(t+\tau)] \\ + \epsilon[E(t) \cdot I(t+\tau)] + \epsilon[I(t) \cdot E(t+\tau)] \dots\dots\dots(28)$$

$$\text{or } C_{qq}(\tau) = C_{ee}(\tau) + C_{ii}(\tau) + C_{ei}(\tau) + C_{ie}(\tau) \quad (29)$$

where the suffices indicate the auto and cross covariances.

Because $E(t)$ and $I(t)$ are statistically independent, $C_{ei}(\tau) = C_{ie}(\tau) = 0$, while $C_{ee}(\tau)$ will fall rapidly with increasing τ .[†]

Hence

$$C_{qq}(\tau) \approx C_{ii}(\tau) \text{ for large } \tau.$$

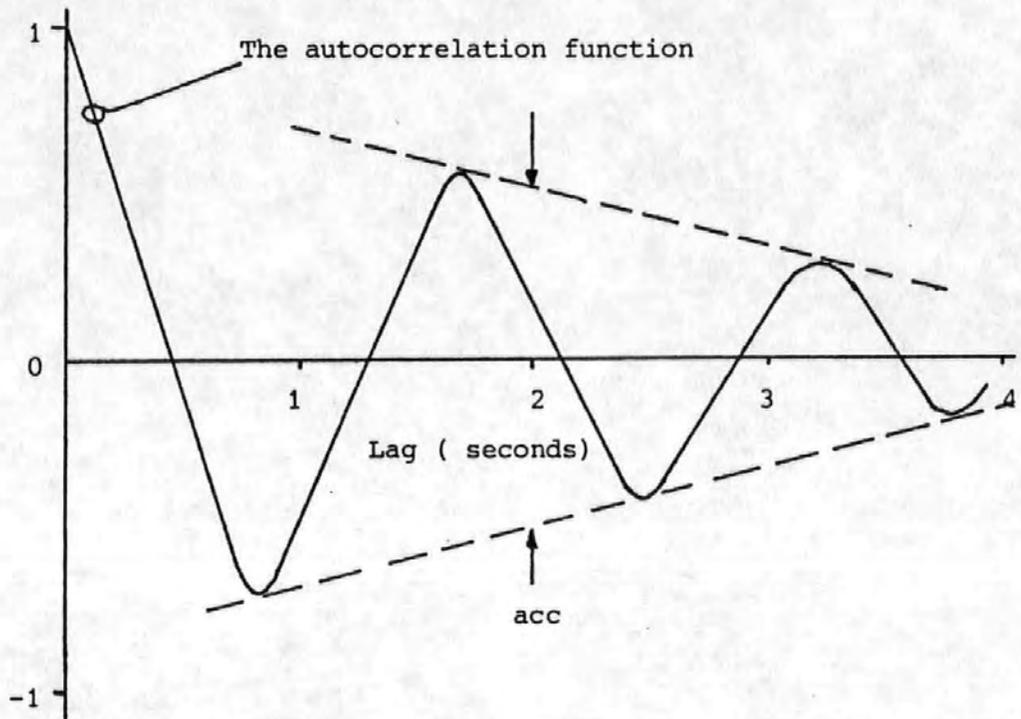
Since the autocorrelation function and the autocovariance function are related by the variance then

$$R_{qq}(\tau) \approx R_{ii}(\tau) \text{ for large } \tau. \text{ (where } R \text{ denotes the autocorrelation function)}$$

Since $I(t)$ is approximately rectangular then $R_{ii}(\tau)$ and hence $R_{qq}(\tau)$ will be triangular.

Thus to assess the effectiveness of the correction method the autocorrelation function of the corrected waveform was calculated and plotted. Those a.c.f.'s which contained any triangularity were examined further and the period, and the peak to peak envelope amplitude of the triangularity were noted at a lag of two seconds. A two second lag was chosen since the initial decrease in the autocorrelation had always been completed within two seconds (by observation of the a.c.f.'s). Figure 2-10 shows graphically how these measurements were made. The peak to peak amplitude of the a.c.f.'s at this lag is referred to as the autocorrelation co-efficient (a.c.c.)

[†]The autocovariance function $C_{ee}(\tau)$ of the EEG signal $E(t)$ will fall rapidly to zero with increasing τ since the EEG contains no regular periodic component.



The autocorrelation co-efficient (acc) is the peak to peak amplitude of the envelope of the autocorrelation function at a lag of two seconds.

Figure 2-10

Measurement of the Autocorrelation Co-efficient

As a practical test of the sensitivity of the method a square wave of known amplitude was added to a section of 'eyes open' background EEG. The frequency of the square wave was chosen to be 0.8 Hz as this was about the most common frequency observed in the EOG signals during the eye movement correction experiments. The resultant signal was analysed in the same way as the corrected EEG signals. Table 2-1 shows the results of these tests with differing amounts of square wave added to the synthetic background EEG. From the table it may be deduced that the autocorrelation coefficient gives a good indication of the magnitude of residual artefact and can detect residual components with amplitudes of only one quarter of the background. Figure 2-11 shows the variation of the a.c.c. with differing amounts of residual square wave present in the EEG.

2.5 The Modified Correction Method

The modified form of Quilters correction method [10] is described here. The fundamental assumption upon which the method is based is that the measured EEG signal can be considered to be formed from a linear combination of the true EEG and the interfering artefact signals. Thus if the artefact signals could be measured independently of the EEG then by subtracting the appropriate fractions of the artefact signals from the observed (corrupted) EEG the true EEG could be established. Fortunately, the artefact signals due to eye movements can be obtained by placing electrodes in close proximity to the eyes, thus if the fractions of these signals present in the observed EEG are known the true EEG can be

Table 2-1

The results of tests performed to determine
the sensitivity of the ACC to residual square wave

SQUARE WAVE BACKGROUND	RATIO dB	ACC OBSERVED	FREQUENCY OBSERVED	COMMENT
100/10	+ 20	1.489	0.8	Triangularity easily visible in ACF Square wave easily visible in EEG.
50/10	+ 14	1.447	0.8	"
40/10	+ 12	1.430	0.8	"
30/10	+ 9.5	1.375	0.8	"
18/10	+ 5.1	1.211	0.8	"
10/10	0	0.885	0.8	Triangularity easily visible in ACF. Square wave visible in EEG but noisy.
7.5/10	- 2.5	0.717	0.8	Triangularity easily visible in ACF. Square wave just visible in EEG.
5/10	- 6.0	0.552	0.82	Triangularity visible in ACF Square wave not visible in EEG.
3.75/10	- 8.5	0.465	0.82	Triangularity visible in ACF
2.5/10	-12.0	0.383	0.78	Triangularity just visible in ACF
0/10	-	0.256	-	No visible periodicity in ACF

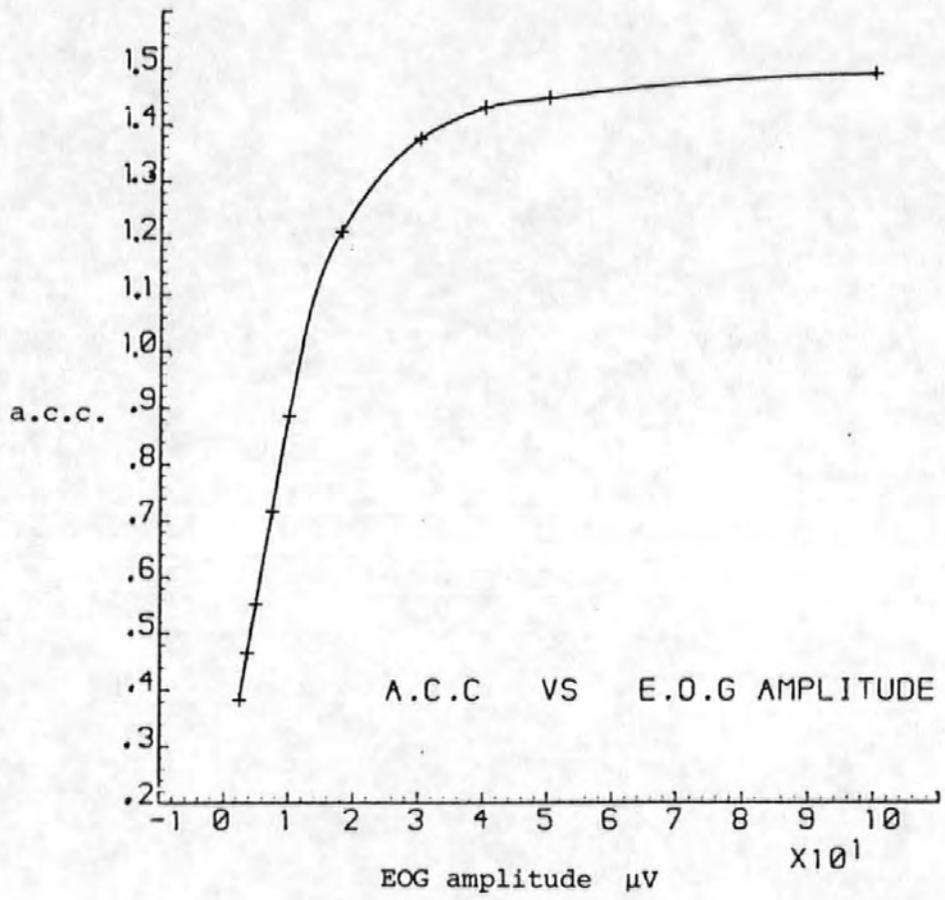


Figure 2-11

A graph showing the variation of the autocorrelation co-efficient with the amplitude of a synthesized square EOG signal.

obtained. Since the eyes are free to swivel about two axes it is necessary to allow for both horizontal and vertical artefact signals. Furthermore since the eyes do not always move in unison or through the same amount of arc, and may have different dipole moments, it may be necessary to allow for horizontal and vertical components from each eye in the EEG.

Assuming that the transmission path between the source and the scalp electrode is linear and $M(t)$ is the measured EEG signal, $E(t)$ is the true EEG signal, K_1, K_2, K_3, K_4 are constants, $V_L(t)$ is the vertical component of the left EOG $H_R(t)$ is the horizontal component of the right EOG etc., then

$$M(t) = E(t) + K_1 V_L(t) + K_2 V_R(t) + K_3 H_L(t) + K_4 H_R(t) \dots (30)$$

or

$$E(t) = M(t) - [K_1 V_L(t) + K_2 V_R(t) + K_3 H_L(t) + K_4 H_R(t)] \dots (31)$$

Thus if the constants $K_1 \dots K_4$ could be found then the uncorrupted EEG signal $E(t)$ could be determined.

Rewriting equation (30) in discrete time form gives

$$M(i) = E(i) + K_1 V_L(i) + K_2 V_R(i) + K_3 H_L(i) + K_4 H_R(i) \dots (32)$$

where i represents the sample number.

Defining

$$M_{VL} = \Sigma [M(i) V_L(i)] \dots (33)$$

$$M_{VR} = \Sigma [M(i) V_R(i)] \dots (34)$$

$$M_{HL} = \Sigma [M(i) H_L(i)] \dots (35)$$

$$M_{HR} = \Sigma [M(i) H_R(i)] \dots (36)$$

Where all the summations are carried out over the range $i = 1$ to N , the number of data samples.

Substituting equation (32) into equation (33) gives:-

$$M_{VL} = \Sigma [\{ E(i) + K_1 V_L(i) + K_2 V_R(i) + K_3 H_L(i) + K_4 H_R(i) \} V_L(i)] \dots (37)$$

$$= \Sigma [E(i) V_L(i) + K_1 V_L^2(i) + K_2 V_R(i) V_L(i) + K_3 H_L(i) V_L(i) + K_4 H_R(i) V_L(i)] \dots (38)$$

$$\text{defining } P_{VL} = \Sigma V_L^2(i) \dots (39)$$

$$B = \Sigma [V_R(i) V_L(i)] \dots (40)$$

$$C_{CL} = \Sigma [V_L(i) H_L(i)] \dots (41)$$

$$C = \Sigma [V_L(i) H_R(i)] \dots (42)$$

gives

$$M_{VL} = \Sigma [E(i) V_L(i)] + K_1 P_{VL} + K_2 B + K_3 C_{CL} + K_4 C \dots (43)$$

Substituting equation (32) into equation (34) and defining

$$P_{VR} = \Sigma V_R^2(i) \dots (44)$$

$$D = \Sigma [V_R(i) H_L(i)] \dots (45)$$

$$C_{CR} = \Sigma [V_R(i) H_R(i)] \dots (46)$$

gives

$$M_{VR} = \Sigma [E(i) V_R(i)] + K_1 B + K_2 P_{VR} + K_3 D + K_4 C_{CR} \dots (47)$$

Substituting equation (32) into equations (35) and (36) in a similar manner and defining

$$P_{HL} = \Sigma H_L^2(i) \dots (48)$$

$$P_{HR} = \Sigma H_R^2(i) \dots (49)$$

$$A = \Sigma [H_L(i) H_R(i)] \dots (50)$$

gives

$$M_{HL} = \Sigma [E(i)H_L(i)] + K_1 C_{CL} + K_2 D + K_3 P_{HL} + K_4 A \dots\dots\dots(51)$$

and

$$M_{HR} = \Sigma [E(i)H_R(i)] + K_1 C + K_2 C_{CR} + K_3 A + K_4 P_{HR} \dots\dots\dots(52)$$

Equations (43), (47), (51), (52) may be more conveniently handled if written in matrix form.

$$\begin{bmatrix} M_{VL} \\ M_{VR} \\ M_{HL} \\ M_{HR} \end{bmatrix} = \begin{bmatrix} \Sigma [E(i)V_L(i)] \\ \Sigma [E(i)V_R(i)] \\ \Sigma [E(i)H_L(i)] \\ \Sigma [E(i)H_R(i)] \end{bmatrix} + \begin{bmatrix} P_{VL} & B & C_{CL} & C \\ B & P_{VR} & D & C_{CR} \\ C_{CL} & D & P_{HL} & A \\ C & C_{CR} & A & P_{HR} \end{bmatrix} \begin{bmatrix} K_1 \\ K_2 \\ K_3 \\ K_4 \end{bmatrix} \dots\dots\dots(53)$$

or

$$\underline{M} = \underline{Z} + \underline{X.K} \dots\dots\dots(54)$$

Where \underline{M} , \underline{Z} , \underline{X} and \underline{K} represent the respective matrices.

The elements of the column matrix \underline{Z} may be considered as the cross co-variance between the true EEG signal and the four artefact signals at zero lag.

Provided $E(i)$ does not affect the artefact signals $V_L(i)$ etc. then the correlation between $E(i)$ and the artefact signals will be small. The matrix \underline{Z} is thus assumed to be zero.

The values $K_1 - K_4$ may now be obtained by solving the equation

$$\underline{M} = \underline{X.K} \dots\dots\dots(55)$$

Since matrices \underline{M} and \underline{X} only involve quantities which

can be obtained from the signals $M(i)$ and $V_L(i)$, $V_R(i)$, $H_L(i)$ and $H_R(i)$.

$$\text{Therefore } \underline{K} = \underline{X}^{-1} \underline{M} \dots\dots\dots (56)$$

The values of $K_1 \dots\dots K_4$ may then be substituted into equation (31) and hence the true EEG signal can be calculated.

There are however a number of limitations to this technique. The most important of these are:-

- (i) A considerable amount of computation is involved in the calculation of the 'sums of products' terms in matrices \underline{X} and \underline{M} .
- (ii) The matrix \underline{Z} may not be sufficiently small to be neglected.
- (iii) The method (as described) cannot be applied on-line since a prior knowledge of the signals over N data points is required.

Offsetting these disadvantages are:-

- (i) No manual setting up is required. (Unlike the methods due to McCallum and Walter and Girton and Kamiya where manual setting of the potentiometers is required. Manual setting up is both time consuming and is subjective and is therefore also inaccurate).
- (ii) The method is self-optimizing i.e. New and optimum values of $K_1 - K_4$ are found for each N point data epoch.

References for Section 2

- [1] McAdam D W, "Physiological Mechanisms"
Published in "Event-Related Slow Potentials of the
Brain: Their relations to Behaviour"
Edited by McCallum W C & Knott J R
Electroencephalography and Clinical Neurophysiology
Supplement No. 33, pp 79-86.
Published by Elsevier Scientific Publishing Company
1973.
- [2] Rohrbaugh J W, Sydulko K, Lindsley D B, "Brain Wave
Components of the Contingent Negative Variation in
Humans"
Science, Volume 191, pp 1055-1057, 1976.
- [3] Mardia K V, "Statistics of Directional Data"
Published by Academic Press (London), 1972.
- [4] Johnson T E, A Private Communication with the Author.
- [5] Moore B R, "A Modification of the Rayleigh test for
Vector data"
Biometrika, Volume 67, pp 175-180, 1980.
- [6] Hodges J L, "A bivariate sign test"
Annals of Mathematical Statistics
Volume 26, pp 523-527, 1955.
- [7] Beagley H A, Sayers B McA, Ross A J, "Fully Objective
ERA by Phase Spectral Analysis"
Acta Oto-laryngologica, Volume 87, pp 270-278, 1979.

- [8] McCallum W C, Walter W G, "The Effects of Attention and Distraction on the Contingent Negative Variation in Normal and Neurotic Subjects"
Electroencephalography and Clinical Neurophysiology, Volume 25, pp 319-329, 1968.
- [9] Girton D G, Kamiya J, "A simple on-line technique for removing Eye Movement Artifacts from the EEG"
Electroencephalography and Clinical Neurophysiology, Volume 34, pp 212-216, 1973.
- [10] Quilter P M, MacGillivray B B, Wadbrook D G, "The Removal of Eye Movement Artefact from EEG Signals using correlation Techniques"
Random Signals Analysis, IEE Conference Publication No. 159, pp 93-100, 1977.
- [11] Whitton J L, Lue F, Moldofsky H, "A spectral Method for removing Eye Movement Artifacts from the EEG"
Electroencephalography and Clinical Neurophysiology, Volume 44, pp 735-741, 1978.
- [12] Hillyard S A, Galambos R, "Eye Movement Artifact in the CNV"
Electroencephalography and Clinical Neurophysiology, Volume 28, pp 173-182, 1970.
- [13] Corby J C, Kopell B S, "Differential Contributions of Blinks and Vertical Eye Movements as Artifacts in EEG Recording"
Psychophysiology, Volume 9, pp 640-644, November 1972.

- [14] Barlow J S, Brazier M A B, "A Note on a Correlator
for Electroencephalographic Work"
Electroencephalography and Clinical Neurophysiology
Volume 6, pp 321-325, 1954.

3. Experimental Techniques

3.1 Initial CNV Tests

In order to gain first hand information about the problems associated with obtaining CNV responses two subjects were examined using conventional evoked response averaging equipment. These tests were carried out using equipment available at Freedom Fields Hospital during the first few months of this investigation. The two stimuli used were an auditory 'click' followed by a flashing lamp. Averaging over 64 trials was performed by a Medelec DAV6 digital averager. The resulting average CNV's are shown in Figure 3-1.

The observations made from these tests shaped the way in which the present measurement system was developed and the tests were performed. These observations are summarized below.

- (1) The dynamic range of the averager[†] was often exceeded causing it to reject some of the trials.
- (2) The onset of the visual stimulus caused the subjects to blink in several of the individual trials. These blinks are highly undesirable since they introduce an artefact which is synchronised to the visual stimulus.

[†]The Medelec DAV6 averager uses an 8 bit analogue-to-digital converter and hence has an inherent dynamic range of one part in 2^8 or 48 dB. at the input. However this range is only fully utilized if the gain of the preceding analogue stage is set such that the largest input signal just fails to overload the converter. Unfortunately the variable nature of most bioelectric signals means that the amplitude of the largest signal cannot be predicted. Thus either a considerable amount of "Headroom" must be allowed (which in turn reduces the effective dynamic range), or a certain degree of overloading of the converter must be tolerated.

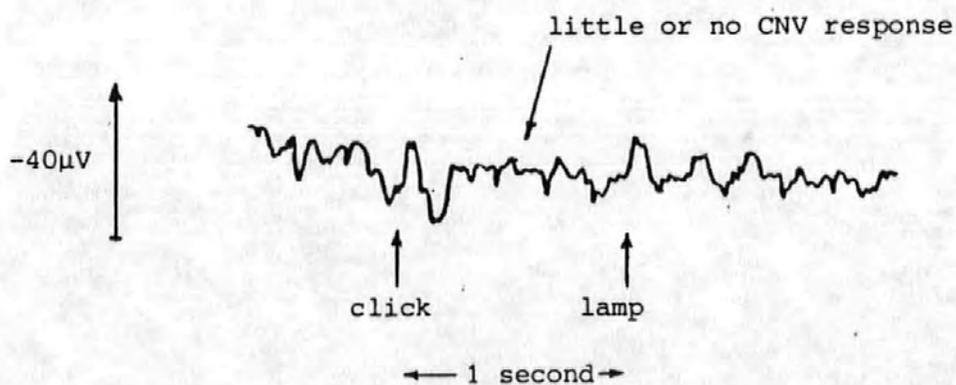
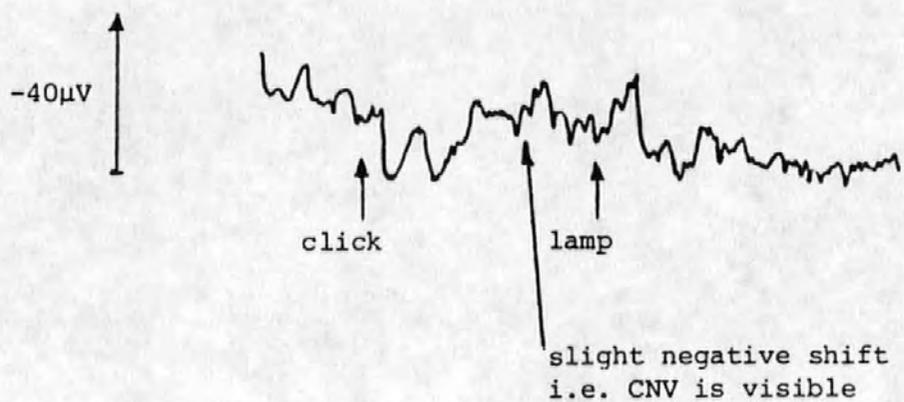


Figure 3-1

The initial attempts at recording CNV responses

- (3) The subjects became tired and showed less motivation in the later trials.
- (4) The subjects frequently moved their eyes which invariably caused the signal averager to overload. This indicated the presence of quite large artefacts induced by eye movements.

As a result of these observations -

- (1) The dynamic range of the proposed data acquisition system was increased to 72 dB.
- (2) Two auditory stimuli were used in place of one auditory and one visual.
- (3) The possibility of using fewer trials in the averaging process or alternative methods to averaging would clearly have to be considered.
- (4) Some method of removing the effects of eye movement artefact would have to be implemented.

3.2 Data Transfer and Preliminary Processing

During transfer from the disk of the minicomputer logging system to a data base on the Polytechnic main computer several tests were performed to verify the authenticity of the data and to ensure that no errors occurred in the transfer. In view of the large amounts of data involved these tests were performed automatically. The eye movement correction experiments involved the transfer of nearly 100,000 twelve bit words for each subject and approximately 400,000 words for each subject for the CNV tests.

For compatibility with conventional hardware each of the twelve bit words was split into two six bit quantities to which parity and marker¹ bits were added. These eight bit quantities (bytes) were initially transferred to the main computer on paper tape although eight inch, single density, single sided, floppy disks were subsequently adopted for speed and ease of handling. To each batch of 1024 words (2048 bytes) were added a batch number, an unique 'start of batch' indicator and a modulo 256 (eight bit) checksum.

When read into the main computer the parity and marker bits were tested for each byte and the checksum re-calculated and compared with that read for each batch. Any errors then found were reported² and the faulty batch was rejected. Rejected batches were re-transferred at a later date. The data for each experiment and subject was stored in a sequential access data file along with certain relevant system and other parameters (e.g. the sample rate, the analogue filter cut-off frequencies, the amplitude calibration constants, subjects name etc.) The final test of the transferred data was performed with the aid of a programme written in Fortran IV and using the Gino [1] graphics subroutines. This programme (Appendix 8.7) was used to plot the data on an interactive graphics terminal where a visual comparison³

¹ The marker bit was used to indicate whether the six bits concerned were the most (marker bit = 1) or the least (marker bit = 0) significant of the original twelve bit word.

² Paper tape error rates $\approx 1:1.8 \times 10^6$ bits
Floppy disk error rates $< 1:34.7 \times 10^6$ bits.

³ An identical waveform was not expected since the chart output of the EEG machine had different filtering to that employed in the acquisition electronics.

was made with the relevant section of the chart output of the EEG machine.

3.3 Eye Movement Corrections

As previously mentioned the eye movement artefact is caused by a standing potential between the cornea and the retina of the eye [2]. Whilst the eyes remain stationary the potential recorded from scalp electrodes will contain no artefact (except possibly a d.c. component) but any change in position will result in an artefact being superimposed on the background EEG activity. The amount of artefact introduced is related to the angular displacement of the eye and the position of the measuring electrodes [2]. Several methods of removing the artefact have been suggested based on the subtraction of a fraction of the electro-oculogram (EOG) i.e. the potential measured at electrodes placed in close proximity to the eyes, from the measured EEG [3,4,5]

Other methods involving a considerable degree of co-operation from the subjects [6,7] were examined but rejected on the grounds that such co-operation cannot always be obtained, particularly with very young, old or diseased subjects.

In order to test the effectiveness of the methods due to McCallum & Walter [3], Quilter [5] and a modified form of Quilters method derived here, eye movement experiments were performed on six volunteers. The volunteers were asked to make periodic eye movements whilst the EOG and EEG data was stored for subsequent analysis. To obtain consistent

eye movements through a known angle (20°) the wooden screen described in section 4.5.1 was utilised.

During these experiments the subjects were asked to fixate on the illuminated centre LED. One of the peripheral LED's was then switched on and the subjects were told to look repeatedly from the centre LED to the illuminated one at the periphery of the screen and back.

Whilst the subject was performing this task eight second epochs of data were digitized at 125 samples per second and stored. This procedure was repeated twice for each of the eight LED's around the periphery of the screen and for ten subjects. The data thus obtained consisted of digitized versions of the following six analogue signals:-

- (i) The vertical component of the left EOG
(electrodes 6 & 7).
- (ii) The vertical component of the right EOG
(electrodes 2 and 3).
- (iii) The horizontal component of the left EOG
(electrodes 4 and 5).
- (iv) The horizontal component of the right EOG
(electrodes 4 & 5).
- (v) The vertex EEG referred to linked earlobes.
- (vi) The analogue corrected vertex EEG (McCallum and Walters method [3]).

The positions of the electrodes for the EOG signals

are shown in Figure 3-2.

Prior to the recordings the subjects were asked to make extreme up and down eye movements in order that the balancing potentiometer utilized in the correction method due to McCallum and Walter [3] could be adjusted for the optimum artefact rejection. It should be noted however that this "optimum setting" was somewhat subjective in that it was based on a visual assessment of the residual eye movement artefact present in the chart output of the EEG machine.

The digitized signals were then transferred from the minicomputer logging system to the Polytechnic main computer for analysis. To assess the effectiveness of the correction techniques the procedure described in section 2.4 was utilized.

3.4 Processing of Eye Movement Data

The computer programme in Appendix 8.8 was used to perform the eye movement corrections described in section 2.5. This programme also calculates the auto-correlation function of the corrected signal for analysis of the effectiveness of the correction method. It was realized at an early stage that although Quilter's original correction method only used single horizontal and vertical EOG components it should be tried to discover if the modified method was significantly better. Furthermore, since visual examination of the early data showed a high degree of correlation between the left and right vertical EOG components it was thought worthwhile to try a correction based on two horizontal EOG components but only one vertical component. (The horizontal components

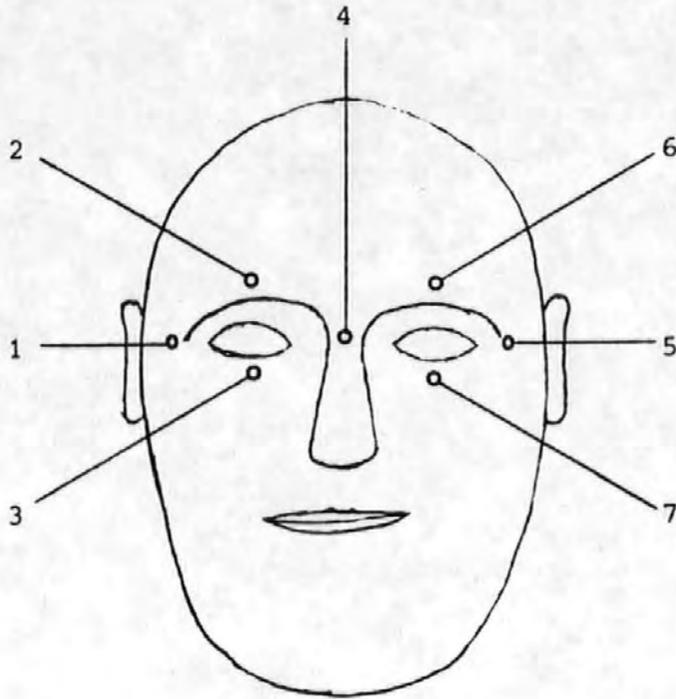


Figure 3-2

Electrode positions for
eye movement corrections

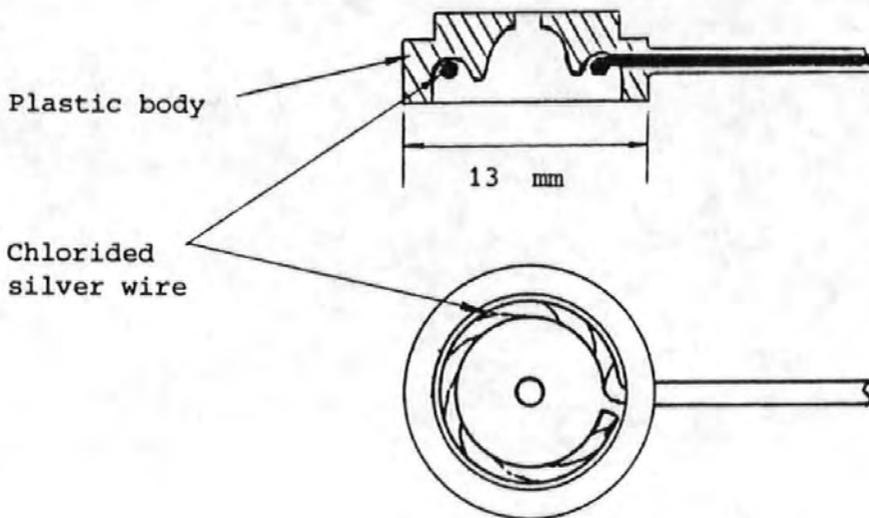


Figure 3-3

Chlorided silver wire
low frequency electrode

of the EOG's were not so highly correlated as the vertical ones). This correction method (the three channel method) may be derived in a similar way to that described in section 2.5 except that the initial equation is

$$M(t) = E(t) + k_1 V(t) + k_2 H_L(t) + k_3 H_R(t) \dots \dots \dots (1)$$

where $V(t)$ is the vertical EOG signal from either the left or the right eye and all other quantities are as described in 2.5

The following processing was thus performed.

- (i) Correction by Quilter's method and calculation of autocorrelation function (acf).
- (ii) Correction by the three channel method and calculation of the acf.
- (iii) Correction by the method described in 2.5 i.e. the four channel method and calculation of the acf.
- (iv) Calculation of the acf for the signal corrected by the method of McCallum and Walter.

In the four channel method electrodes 6-7 (vertical left EOG), 4-5 (horizontal left EOG), 2-3 (vertical right EOG) and 4-1 (horizontal right EOG) were used, (see Figure 3.2). In the three channel method all the above electrodes were used with the exception of those giving the vertical right EOG signal. For Quilter's method electrodes 6-7 and 4-5 were used. (This is very slightly different to the electrode placements used by Quilter who would have used 6-7 and 5-7 instead of 6-7 and 4-5).

3.5 CNV Acquisition

In an attempt to obtain consistent CNV's from the subjects the recording procedure was standardized. The subjects were seated at one end of the recording room facing, and at about 2 metres distant from the end wall. The silver-silver chloride electrodes were attached to the subject with glue (scalp electrodes) or adhesive tape (facial electrodes). Facial electrodes (see Figure 3-2) were used to record the four components of the EOG for subsequent eye movement artefact correction by the four channel method described in Section 2.5. Two channels of CNV information were obtained from electrodes located at the vertex and at a point on the midline approximately 30mm anterior to the vertex. Both electrodes used a common reference which was obtained from a pair of connected electrodes on the left and right earlobes.

After being attached to the subject the electrodes were filled with 'Neptic' electrode gel by means of a syringe with a blunted needle inserted into the hole in the plastic body of the electrode holder (see Figure 3-3). The blunted needle of the syringe was also used to abraid the skin under the electrode whilst the gel was inserted. This procedure ensured a low impedance between the electrode and the scalp (typically $5k\Omega$ or less). When all the electrodes had been similarly treated the impedance between an arbitrary electrode and each of the others was measured using a Specialised Laboratory Equipment model EIT impedance meter.

This instrument measures (approximately) the modulus of the complex impedance at 13Hz. It is important to note that

although the electrode scalp interface impedance is largely resistive at low frequencies, the use of a resistance measuring device with a d.c. internal source must be avoided at all cost when using chlorided silver electrodes [8]. Failure to observe this principle results in a serious degradation of the electrode stability.

If the impedance between any electrode pair was found to be greater than $5k\Omega$ the skin below the offending electrode was further abraded until this value was achieved. When the impedance of all the electrodes was satisfactory they were connected to the EEG machine and the electrode selector switches[†] set for the required electrode pairs. The filters in the data acquisition system (see Section 4.2) were set for a -3dB passband of 0.016 to 30Hz. The sample rate was 125Hz.

In order to familiarize the subject with the stimuli eight presentations were made during which the subject was not required to respond to the second stimulus. The subject was then told to "press the button as quickly as you can when you hear the tone" and a further thirty-two trials were made constituting the CNV run.

Since one particular area of interest was the effect of the inter-stimulus interval (ISI) i.e. the time delay between the click and the tone, two sets of thirty-two CNV trials were made. The first with a one second ISI and the second with a four second ISI. In each case eight presentations

[†]These switches allow each of the machines differential amplifiers to be connected to any two electrodes on the scalp.

were made (with the appropriate ISI) before the run.

As with the previous section the digitized signals were transferred from the logging system to the Polytechnic main computer for processing and analysis.

3.6 Processing of CNV's

Because the CNV is of such a small magnitude in comparison to the normal background EEG activity they are difficult to quantify on an individual basis. Typically the CNV may be of 5-20 μ V in magnitude whereas normal background activity can be 50 μ V or more. However in favourable circumstances the individual CNV's can be observed in chart output of the EEG machine (see Figure 3-4). The normal method employed to improve CNV to background EEG (which may be regarded as signal to noise) ratio is to take a number of individual CNV's and average them. This process produces an average CNV but since the background EEG is not correlated from trial to trial the background EEG in the average is reduced by a factor proportional to the square root of the number of individual CNV's included in the average. In their original report Walter and colleagues [9] used averages of six or twelve trials.

After verification of the stored CNV data a programme was used to calculate the average CNV and plot this waveform on a graph plotter. The averaging was normally carried out over thirty-two individual CNV's. A typical average CNV is shown in Figure 3-5. Prior to averaging, each individual CNV was processed to remove the effects of eye movements as

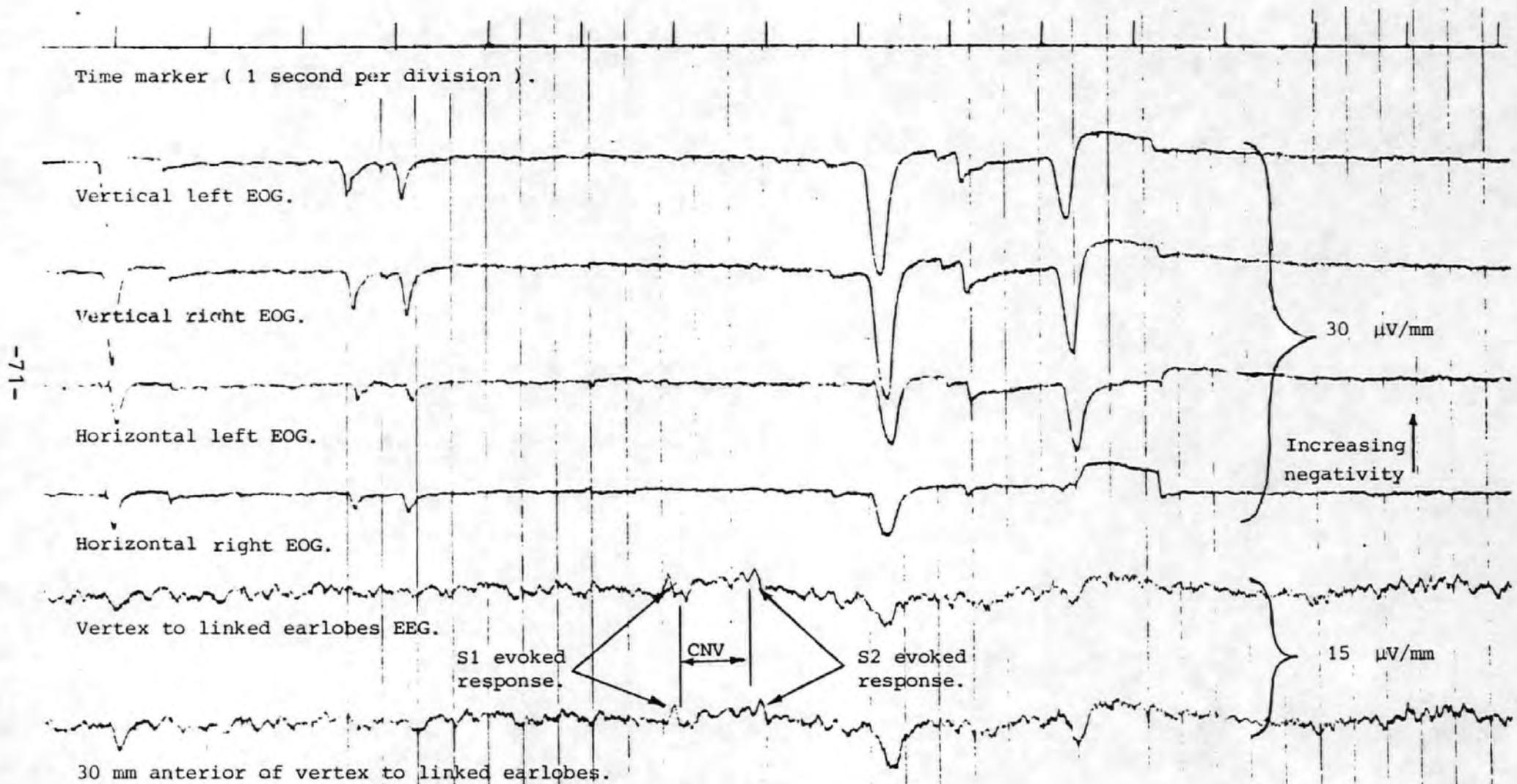


Figure 3-4 The raw CNV waveform

-71-

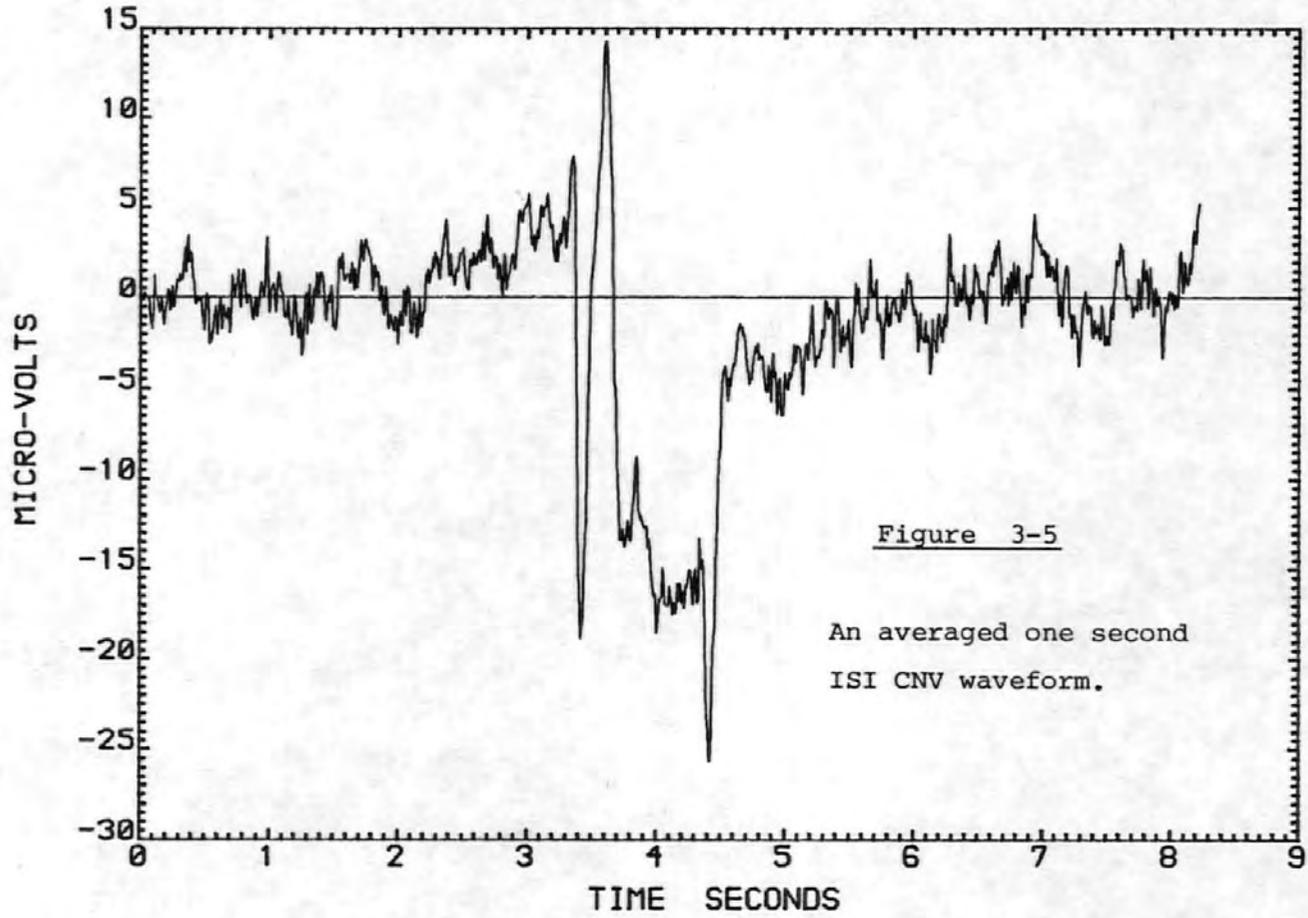


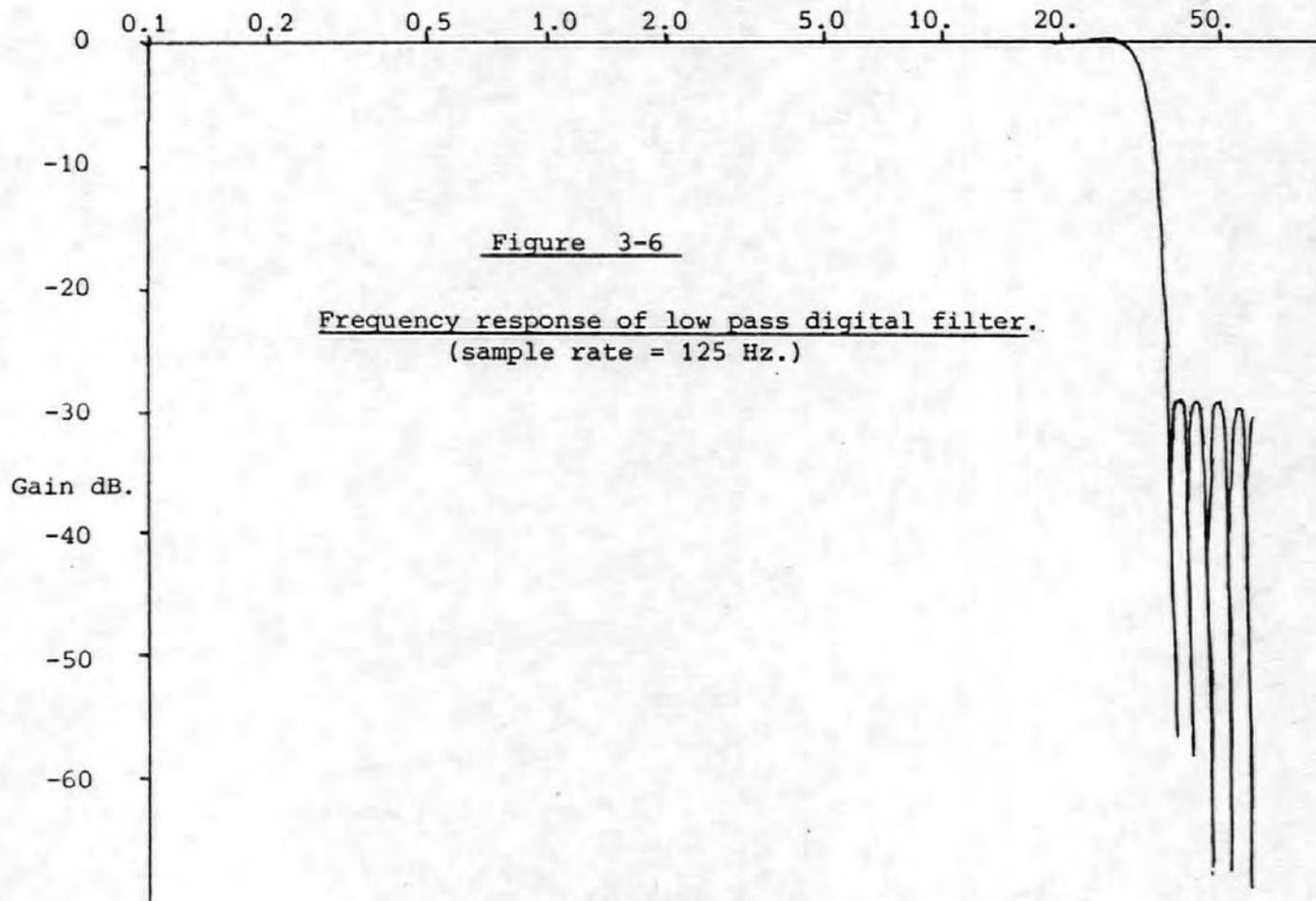
Figure 3-5

An averaged one second
ISI CNV waveform.

previously described. In an attempt to further improve the averaged CNV a linear phase digital lowpass filter was used to remove some of the remaining higher frequency background components. A linear phase filter was chosen in order to preserve the shape of the CNV as much as possible. The response and weighting co-efficients of the filter used most often are given in Figure 3-6. This and other filters have been designed with the aid of the computer programme given in the book by Rabiner and Gold [10].

In view of the work of Professor Sayers and colleagues [11, 12, 13, 14, 15, 16] concerning the nature of the auditory evoked response it was decided to perform similar experiments on the CNV. Sayers showed that the auditory evoked response may be due not to an additional response, but to a re-ordering of the phase spectra of the background, i.e. the phases of certain frequencies of the background EEG generators become entrained by the stimulus and hence reshape the background EEG into the characteristic auditory evoked response. One of Sayers' most important tests was to calculate the energy in the EEG before and during the response to the auditory stimulus. According to Sayers if, as had previously been assumed, the response was an additional signal then the energy during the response would be greater than that before or after, whereas if the response was due to a phase re-ordering then the energy would not change. Sayers offered evidence to suggest that this response may indeed be due to a phase re-ordering. A programme was therefore written to calculate the energies of individual background and CNV sections of the EEG. In addition to this test Fast

Frequency Hz.



Filter co-efficients

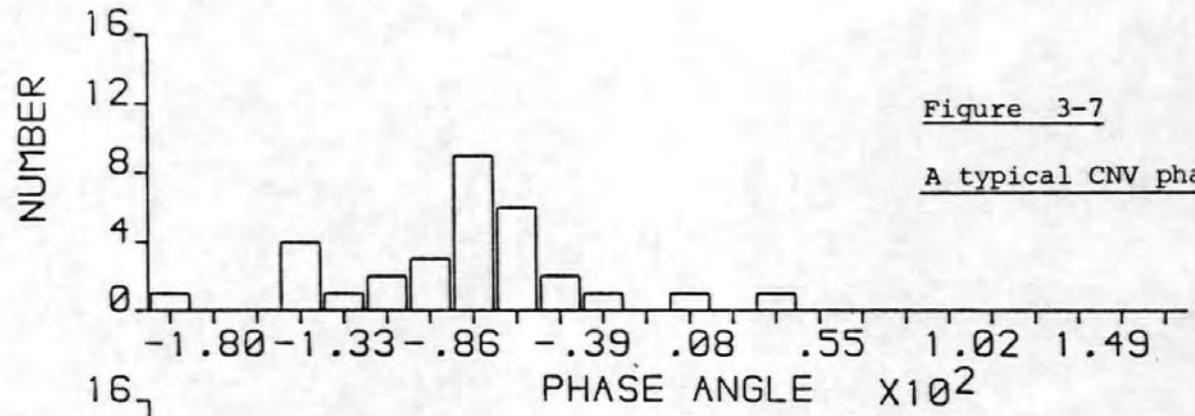
H(1)=H(21) = -0.02232808
H(2)=H(20) = 0.01758909
H(3)=H(19) = 0.02921061
H(4)=H(18) = -0.01742668
H(5)=H(17) = -0.02649306
H(6)=H(16) = 0.04697604
H(7)=H(15) = 0.03685932
H(8)=H(14) = -0.09247865
H(9)=H(13) = -0.03773736
H(10)=H(12) = 0.31495064
H(11) = 0.54097712

Fourier transforms of the CNV's were performed and the phases of the first few frequency components were noted. This procedure was repeated for each of the thirty-two CNV measurements for each subject. The phase information for each frequency components was then sorted into twenty four bands of fifteen degrees each. Histograms (Figure 3-7) were then plotted for each frequency component showing the number of times a phase angle occurred within each band.

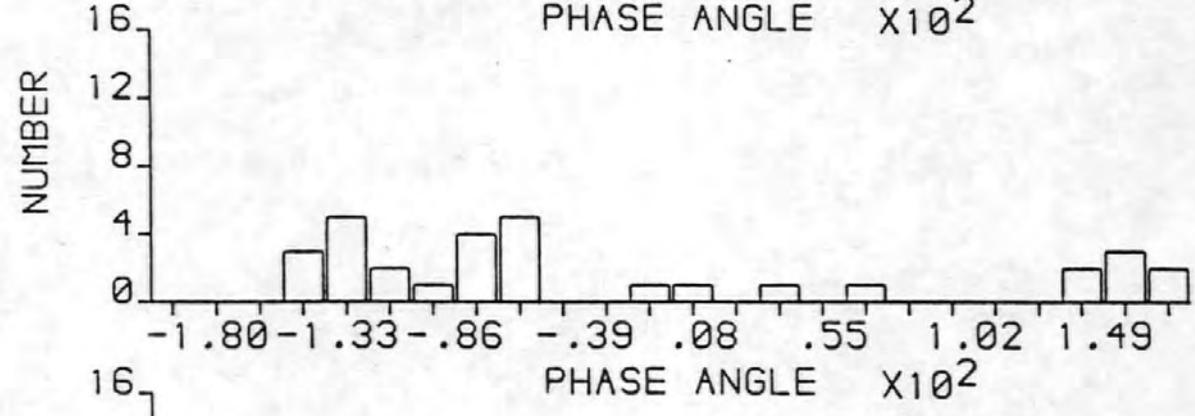
3.7 AEP Acquisition

In view of the important nature of Professor Sayers work [11, 12, 13] on the Auditory Evoked Potential (AEP), and the possible relevance of his findings to other evoked potentials (e.g. the CNV), it was decided to perform a short series of experiments in an attempt to confirm his results. AEP's were recorded from three subjects whilst relaxed and seated in the measurement room. The recording electrodes were placed at the vertex and on the right mastoid processes for the EEG signal, and in the usual facial positions for the four channel eye movement corrections. The auditory stimulus was obtained from an Amplaid stimulus generator set to deliver 1kHz. tone bursts with 100 ms. duration including approximately 10 ms of rise and fall according to a "cosine squared" law. This signal was applied to the right transducer of a pair of Koss K6 stereo headphones. The subjects auditory threshold was then determined by applying the above stimulus with gradually decreasing intensity. When the subject could no longer hear the stimulus tone then the threshold had been found. The experiment then commenced with an

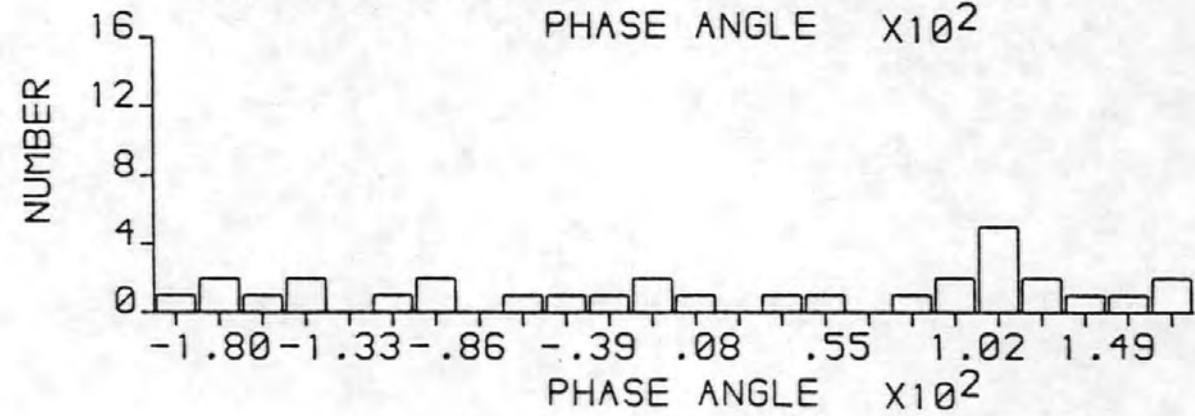
HARMONIC 1



HARMONIC 2



HARMONIC 3



auditory stimulus 70dB (65 for one subject) above the threshold. Thirty-two sections of the EEG/EOG signals were then recorded each with two stimulus presentations at known places within each section. This gave a total of 64 evoked responses at this stimulus level. A further thirty-two trials were then made with the stimulus level set to 40dB above the subjects threshold.

3.8 Processing of AEP's

The stored AEP's were first processed to remove any eye movement artefact by the four channel correction method. The average AEP was then calculated and plotted. A typical pair of high and low stimulus averaged AEP's are shown in Figures 3-8a and 3-8b. A test was then performed to determine whether the energy present in the EEG signal after stimulus was any different to that in the signal prior to the stimulus. The "energy" values were calculated by summing the squares of values of the 64 data points preceeding and succeeding the stimulus as described in Section 2.2.3.1 Since the sampling rate was 125Hz this represented 0.512 seconds of data prior and subsequent to the stimulus. The differences between the pre- and post-stimulus energy values were calculated for each of the 64 individual responses, and the mean of these differences was subjected to a two-tailed t-test. The result of this test indicated whether the mean of the differences was significantly different from zero.

In an attempt to establish whether phase-ordering was present the eye movement corrected sections of EEG containing the AEP's were subjected to Fourier transformation by a

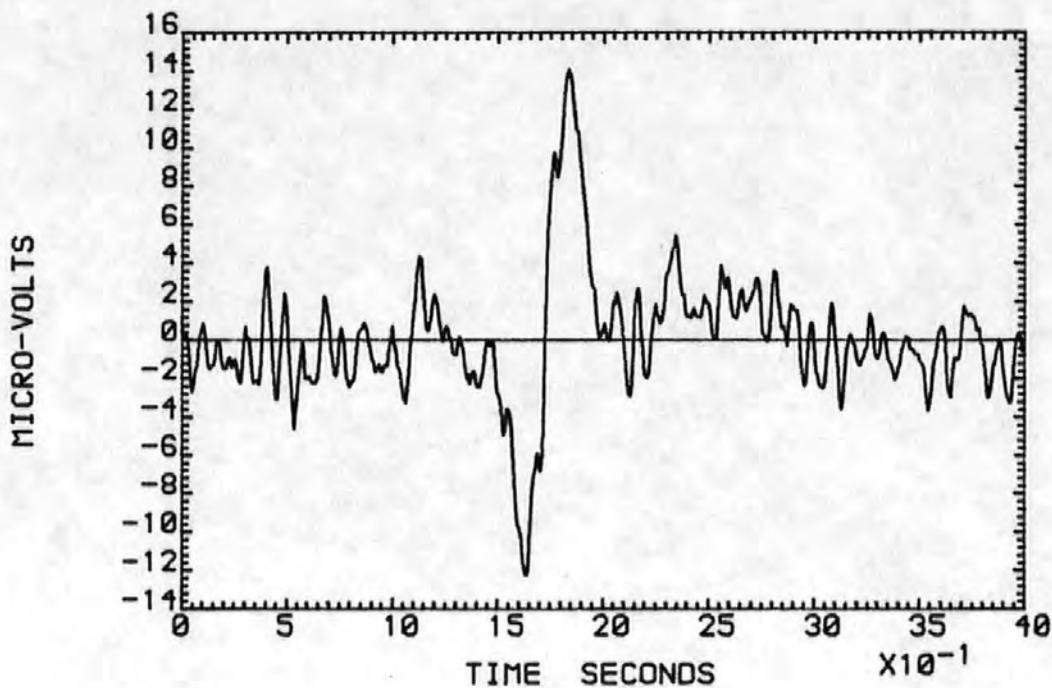


Figure 3-8a An averaged auditory response to a high level stimulus.

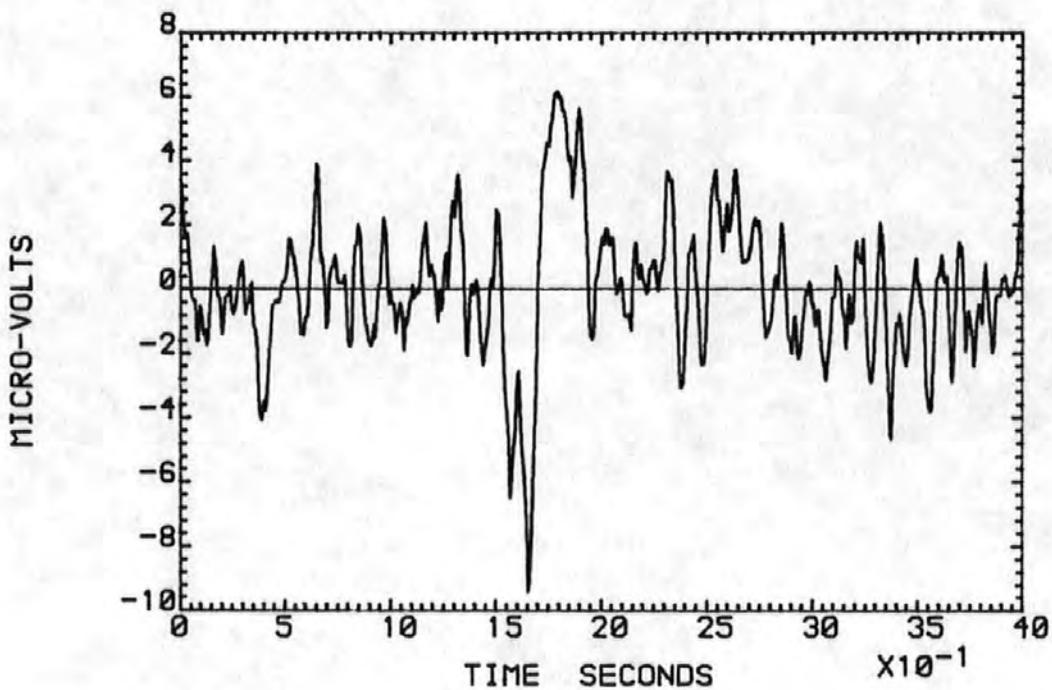


Figure 3-8b. An averaged auditory response to a low level stimulus.

radix two FFT. The phases of the six lowest frequency components (excluding the d.c. term) were stored for each of the 64 AEP's. Histograms were then plotted showing the number of times the phase angle of the transformed AEP's fell within a particular range for each of these six frequency components (sometimes referred to as harmonics).

As an alternative to the phase histogram, phasor diagrams were plotted showing, for a particular harmonic frequency, the amplitudes and phases of the transformed responses. These diagrams were also plotted for the sections of EEG data preceeding the application of the stimulus (i.e. background EEG data). A typical pair of background/AEP phasor diagrams are shown in Figure 3-9. For clarity the phasors were represented by a cross at (what would have been) the tip of the phasor. Also for ease of interpretation a circle of arbitrary radius has been drawn and onto this small triangles have been added to show the phases of the components independently of their amplitudes. Like the phase histograms the phasor diagrams were plotted for each of the first six harmonics. The statistical tests described in Section 2.2.2 were then applied to the phasor diagram information.

3.9 Fourier Transform Considerations

Several of the methods adopted for studying the CNV's and the auditory evoked potentials (AEP's) made use of a time to frequency transformation. A Fast Fourier Transform (FFT) was used to split the responses up into different frequency components which could then be studied in greater detail. A number of important factors were considered in

PRE STIM.

POST STIM.

HARMONIC 1

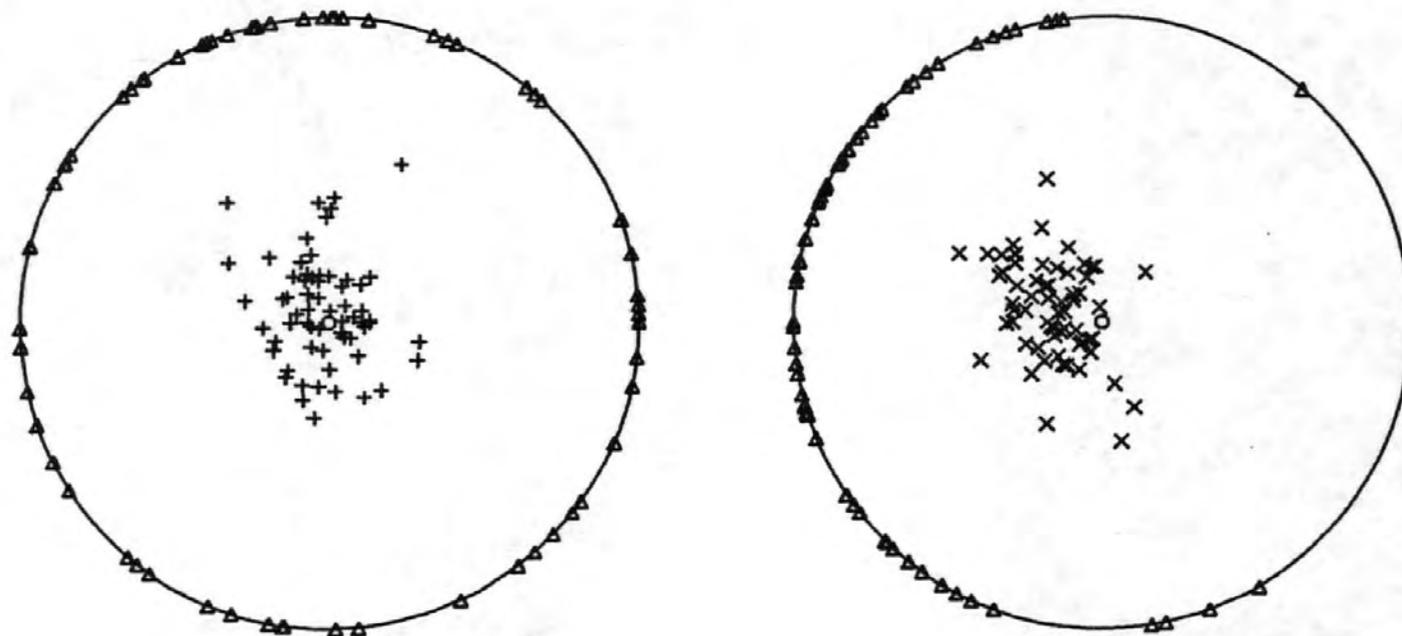


Figure 3-9 A typical AEP phasor diagram

choosing the parameters of the transform. These included the length of the section of data to be transformed, the required spectral resolution, and the time (and cost) of the transformation. However since the lengths of both the CNV's and AEP's were well defined these proved to be the ultimate limitation. In practice the response data was augmented with a number of zero valued "data" points to bring the total number of points up to a value suitable for transformation with the radix 2 FFT.

Because of the finite length (T_m) of the data the spectrum obtained was that of the data convolved with that of the window or truncation function. If, for example, the truncation function was rectangular i.e.

$$\begin{aligned} \text{data to be transformed} &= \text{data multiplied by one} \\ &\quad (0 < t < T_m) \\ &= \text{data multiplied by zero} \\ &\quad \text{elsewhere,} \end{aligned}$$

then the spectrum of the data would be convolved with the well known SINC function with zero crossings at frequencies of n/T_m , where $n = \dots\dots-3, -2, -1, 1, 2, 3\dots\dots$. This would result in distortion of the true spectrum to an extent that would often be unacceptable. For this reason many other truncation functions have been devised which cause much less spectral distortion. The truncation function chosen was the cosine taper (sometimes referred to as the raised cosine) which was applied for 10% of the data length at each end of the data. This function can be described mathematically by the following expressions:-

$$\begin{aligned}
 W(t) = & \frac{1}{2} \left[1 - \cos \left\{ \frac{t\pi}{0.1 T_m} \right\} \right] & 0 < t < 0.1 T_m \\
 & 1 & 0.1 T_m < t < 0.9 T_m \\
 & \frac{1}{2} \left[1 - \cos \left\{ \frac{(1-t)\pi}{0.1 T_m} \right\} \right] & 0.9 T_m < t < T_m
 \end{aligned}$$

This window function has smaller and more rapidly decreasing side lobes than the rectangular window which resulted in much less spectral leakage [17].

Where it was necessary to augment the response data with zeros to fulfil the transform requirements, the windowing was applied prior to the addition of the zeros. Since windowing the zeros does not remove any discontinuity at the end of the true data. Furthermore since the d.c. component was of no interest it was removed by subtracting the mean value of the data from the data.

The FFT algorithm adopted was that described by Robinson [18]. The FORTRAN implementation of this algorithm was capable of forward and inverse Fourier transforms. The data to be transformed was submitted to the subroutine as an array of complex numbers. Since all the input data was real, the imaginary components of the array were set to zero before transformation took place. The N output data points were returned to the calling program in the same array and were generally complex. For some of the analysis methods these were then converted by a further subroutine to modulus and phase information arrays. Although N complex points were returned by the FFT only the first N/2+1 were meaningful since

the remaining $N/2-1$ points represent the negative frequency values. (i.e. a mirror image of the first $N/2$ points. The zero frequency or d.c. component was not mirrored).

References for Section 3

- [1] Computer Aided Design Centre, "GINO-F" and "GINO-
GRAF" User Manuals.
Published by the CAD Centre, Madingly Road, Cambridge.
- [2] Geddes L A, Baker L E, "Principles of Applied Bio-
medical Instrumentation"
Chapter 11, pp 509-517. Published by John Wiley &
Sons, New York. 2nd Edition 1975.
- [3] McCallum W C, Walter W G, "The Effects of Attention
and Distraction on the Contingent Negative Variation
in Normal and Neurotic Subjects"
Electroencephalography and Clinical Neurophysiology,
Volume 25, pp 319-329, 1968.
- [4] Girton D G, Kamiya J, "A simple on-line technique for
removing Eye Movement Artifacts from the EEG"
Electroencephalography and Clinical Neurophysiology,
Volume 34, pp 212-216, 1973.
- [5] Quilter P M, MacGillivray B B, Wadbrook D G, "The
Removal of Eye Movement Artefact from EEG Signals
using Correlation Techniques".
Random Signals Analysis, IEE Conference Publication
No. 159, pp 93-100, 1977.
- [6] Hillyard S A, Galambos R, "Eye Movement Artifact in
the CNV"
Electroencephalography and Clinical Neurophysiology,
Volume 28, pp 173-182, 1970.

- [7] Corby J C, Kopell B S, "Differential Contributions of Blinks and Vertical Eye Movements as Artifacts in EEG Recording"
Psychophysiology, Volume 9, pp 640-644, 1972.
- [8] Cooper R, Osselton J W, Shaw J C, "EEG Technology"
Chapter 2, page 20 Third Edition,
Published by Butterworths, London, 1980.
- [9] Walter, W G, Cooper R, Aldridge V J, McCallum W C,
Winter A L, "Contingent Negative Variation": An
Electric Sign of Sensorimotor Association and Ex-
pectancy in the Human Brain".
Nature, Volume 203, pp 380-384, 1964.
- [10] Rabiner L R, Gold B, "Theory and Application of
Digital Signal Processing"
Chapter 3 and pp 194-204 Published by Prentice-
Hall, New Jersey, 1975.
- [11] Sayers B McA, Beagley H A, Henshall W R, "The
Mechanism of Auditory Evoked EEG Responses"
Nature, Volume 247, pp481-483, 1974.
- [12] Sayers B, McA, Beagley H A, "Objective Evaluation
of Auditory Evoked EEG Responses".
Nature, Volume 251, pp 608-609, 1974.
- [13] Inbar G F (Editor), "Signal Analysis and Pattern
Recognition in Biomedical Engineering".
Part 1, pp 7-22 Published by Halsted Press Division,
John Wiley & Sons, New York 1975.

- [14] Sayers B, McA, Beagley H A, Riha J, "Pattern Analysis of Auditory Evoked EEG Potentials" *Audiology*, Volume 18, pp 1-16, 1979.
- [15] Beagley H A, Sayers B McA, Ross A J, "Fully Objective ERA by Phase Spectral Analysis" *Acta Oto-laryngologica* Volume 87, pp 270-278, 1979.
- [16] Ross A J, Beagley H A, Sayers B McA, "Signal Statistics in Objective Auditory Evoked Potential (Detection by the Phase Spectral Method)" *Journal of Biomedical Engineering*, Volume 2, pp 310-314, 1980.
- [17] Otnes R K, Enochson L, "Digital Time Series Analysis" pp 281-285 Published by John Wiley and Sons, 1972.
- [18] Robinson E A, "Multichannel Time Series Analysis with Digital Computer Programs" Page 63 Published by Holden-Day, San Francisco Revised Edition 1978.

4. Experimental Apparatus

In order to investigate the processing of EEG signals a data acquisition and storage system was required. Several constraints were placed on the choice of this system. The most serious constraint was the lack of finance to fund such a system. However a considerable amount of equipment was available both at the Polytechnic and at Freedom Fields Hospital. Thus whilst the apparatus described here may not be the most elegant solution possible, it is a solution which, for the main parts, utilised freely available equipment.

4.1 Choice of Apparatus

Two commonly used methods for the storage of the EEG data were initially considered. The first of these being an analogue magnetic tape system. This method has the advantage that large amounts of data may be recorded and stored for subsequent retrieval and analysis. Typical modern instrumentation tape recorders are capable of storing seven channels of data with a bandwidth extending from d.c. (by the use of frequency modulation) to 20kHz or more depending on the tape speed. Since EEG signals are normally of interest between d.c. and 40 Hz. very low tape speeds may be used. With such low speeds many hours of data may be stored on a 366m (1200 feet) spool of tape.

The second technique considered was on-line digital storage. The signals are first amplified and filtered in the normal way but are then digitised instead of being sent to the usual EEG chart recorder. The digitised signal may be

stored on any convenient medium but magnetic tape or disk is most common due to the high rates at which the data may be transferred. This method has the advantage that once digitised, the signals are subject to no further noise or distortion. Since two minicomputers were available one of which had a high speed analogue-to-digital converter, multiplexer and numerous other interface components, and the other a flying head disk of approximately 2.5 M bytes capacity, the digital method was chosen. Furthermore, since evoked responses lasting but a few seconds were to be studied the disadvantage of being unable to store more than a few minutes of data was not considered important. An added bonus was that the minicomputer could be programmed to provide pulses at the chosen instant to trigger the external stimulus generators necessary for evoked response studies.

4.2 Analogue Electronics

Figure 4-1 is a block diagram of the complete data acquisition system. The EEG signals were obtained from silver-silver chloride electrodes attached to the subject by glue or adhesive tape. These signals were fed into the electrode selector switches and differential amplifiers of an eight channel Elma-Schönander electroencephalograph. In addition to producing the normal paper chart record, the electroencephalograph was coupled to external amplifiers and filters to allow the information from six of the eight channels to be digitised and stored for subsequent analysis.

The point at which the signals were extracted from the EEG machine was chosen such that the electrode selector

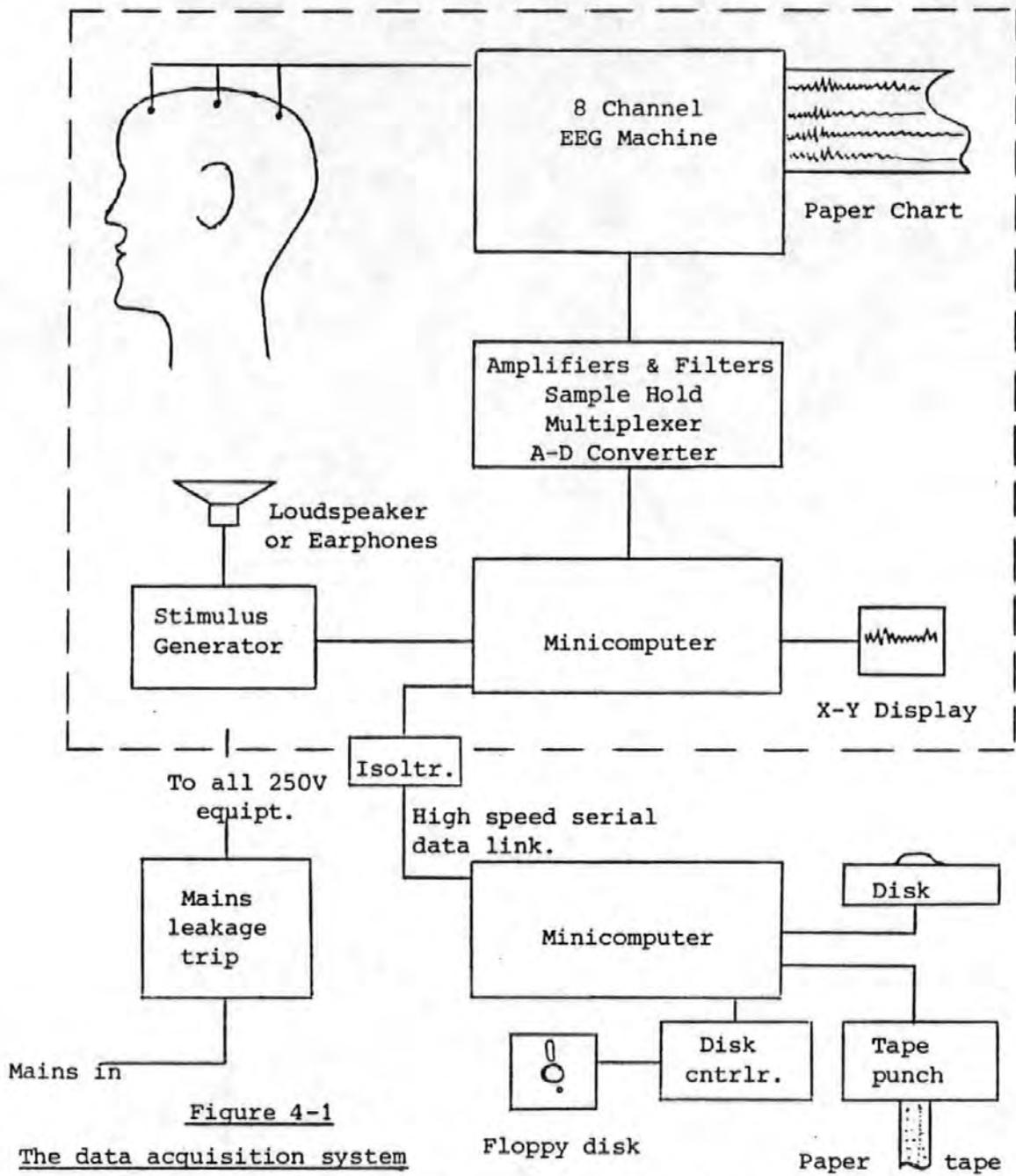
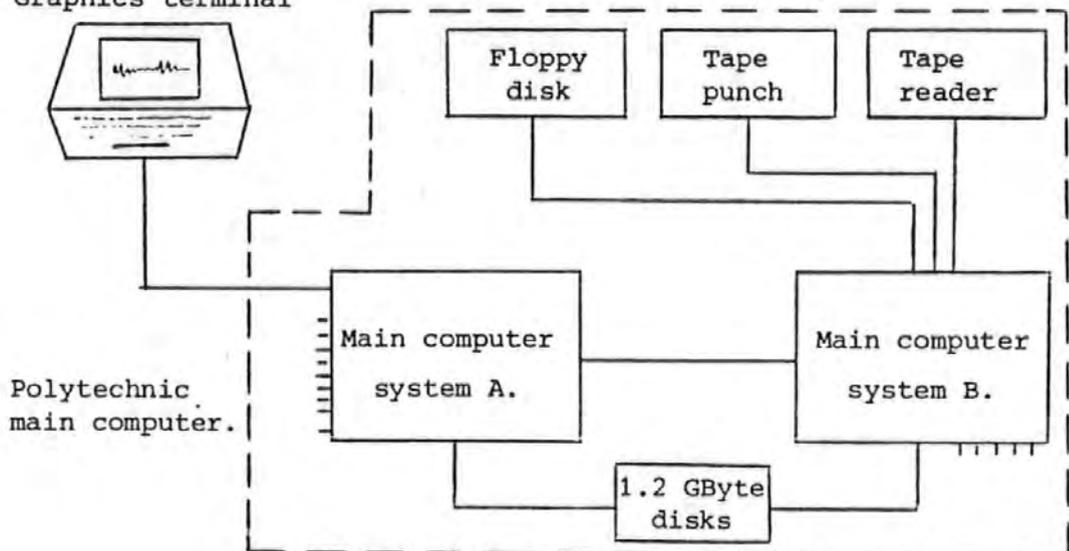


Figure 4-1

The data acquisition system

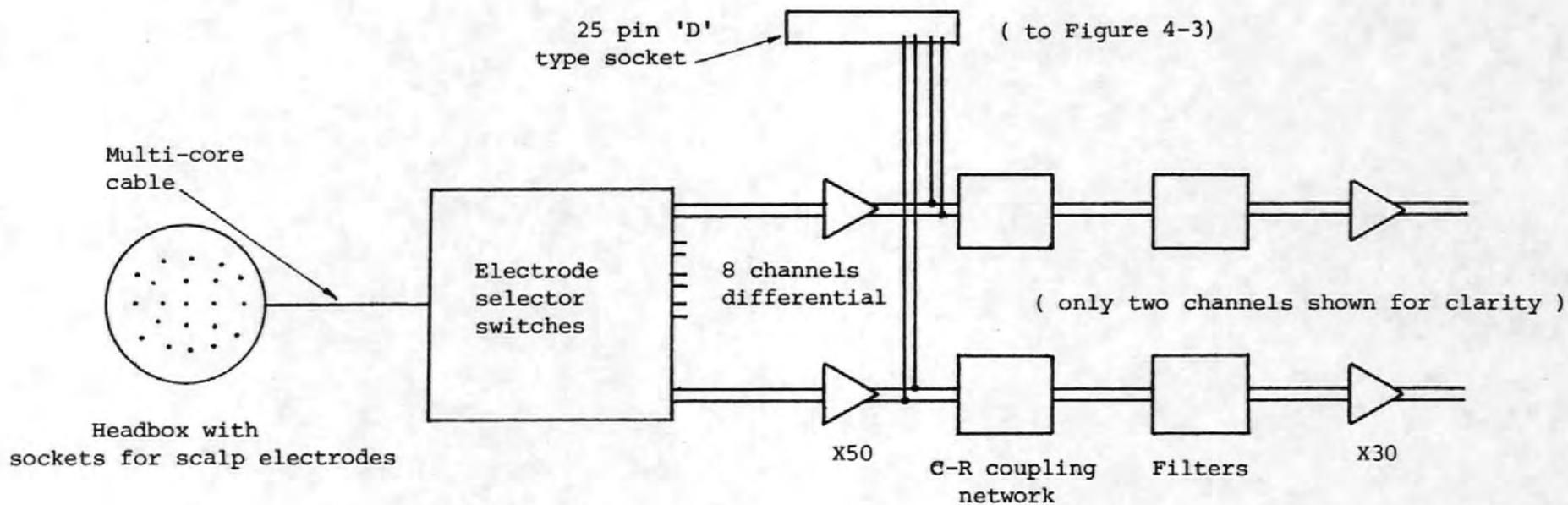
Graphics terminal



switches and the first stage (d.c. coupled x 50) of amplification were utilised (see Figure 4-2). The next stage of amplification in the EEG machine could not be used because the interconnection between the first and second stages formed a high pass C-R network with a cut-off frequency of 0.069 Hz. This frequency was not low enough for CNV recordings where 0.016 Hz. is usually considered to be a more appropriate lower limit (this value corresponds to a time constant of 10 seconds). For this reason the differential outputs of the first stage were wired to a 25 pin 'D type' socket on the rear of the EEG machine. In this way the machine was still able to perform its normal duties and could simply be un-plugged from the additional equipment used to digitise the signals. When used for data logging a fairly short screened multicore cable with a 25 pin 'D type' plug at either end was used to connect the EEG machine to a screened box containing modules for amplifying, filtering and sampling six of the eight EEG machine channels (further modules could be added at a later date should it become necessary to digitise all eight channels).

The input circuit of each module (see Figure 4-3) comprised of a differential high pass C-R network with switchable capacitors to give cut-off frequencies of 0.016, 0.034 and 0.16 Hz. The differential signals were then converted to unbalanced form by an amplifier comprising of two BIFET[†] operational amplifier integrated circuits (RCA type 3140) and a single bipolar operational amplifier (type μ A 741). The gain of this stage was adjusted to 100. The variable resistor VR1 was incorporated to allow optimisation of the

[†]An integrated circuit employing both bipolar and field effect transistors.



16 switches allowing any of the 8 channels to be connected to any pair of scalp electrodes.

Figure 4-2

A block diagram of the input section of the EEG machine.

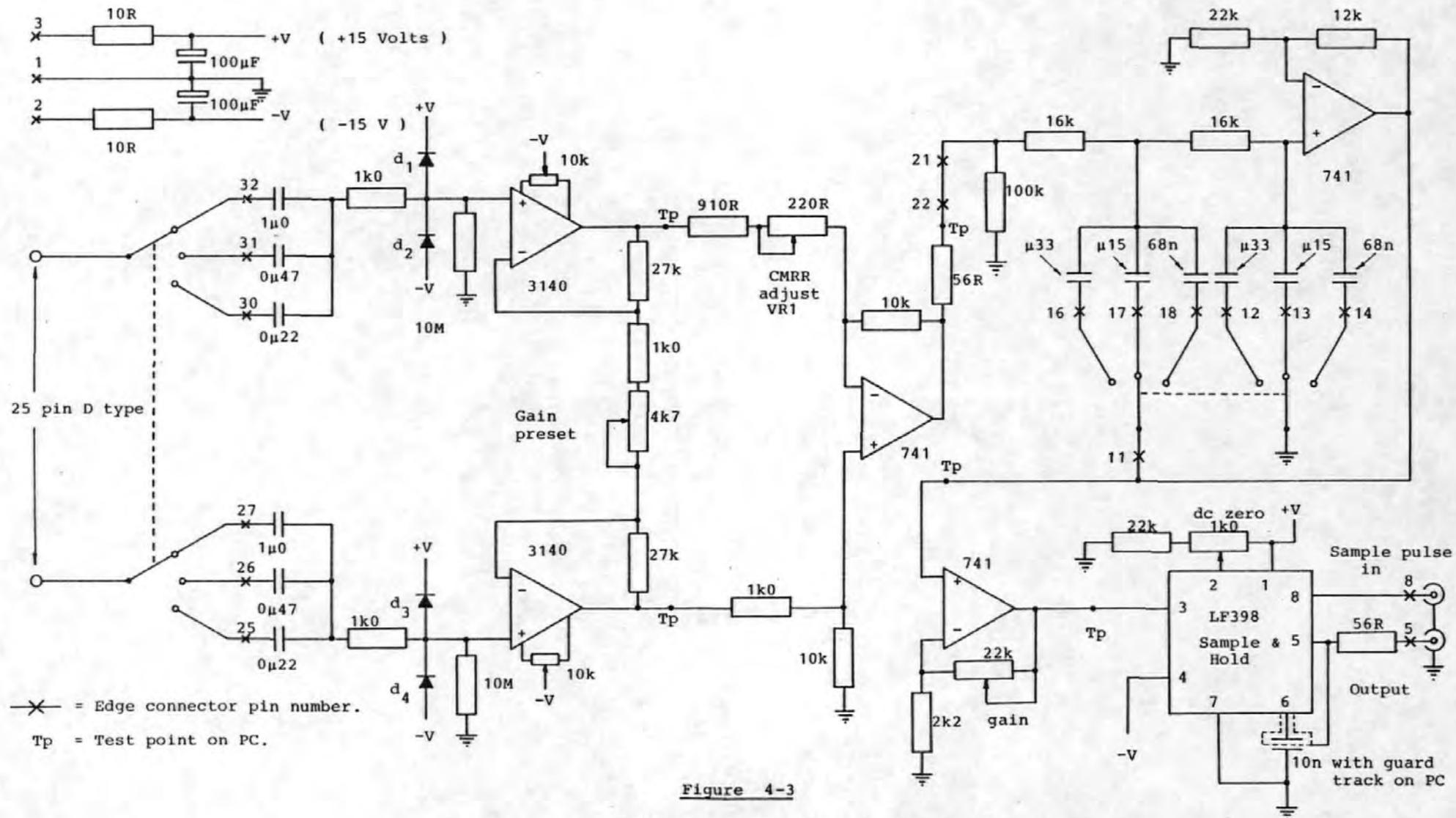


Figure 4-3

The circuit of an amplifier and sample hold module

common mode rejection ratio (CMRR). The diodes $d_1 - d_4$ were incorporated to protect the input circuits of the 3140 operational amplifiers.

The output of this stage was then fed into a second order active filter based on the Sallen and Key design [1]. The calculation of the circuit elements for the desired cut-off frequencies (30, 70 and 130 Hz. switch selectable) was simplified by the use of the tables given in Millman and Halkias [2]. After careful consideration of the various frequency response characteristics the Butterworth filter was chosen because it had a suitably flat amplitude response in the pass-band without undue phase distortion. The cut-off frequencies of 30, 70 and 130 Hz. were chosen because they are commonly used in EEG work and would therefore make any results obtained from our data logging system comparable with results obtained by other workers. Low pass filtering is necessary to prevent aliasing in the subsequent digitisation stage and to attenuate some of the unwanted higher frequency components (e.g. muscle artefact, power frequency interference etc.) High pass filtering (the CR input network) was included to minimise the drift which can occur because of electrode instability. After low pass filtering the signals were further amplified by an operational amplifier (type μA 741) connected in the non-inverting mode with a variable gain. The gain of these amplifiers was set to utilise the complete range of the 12 bit analogue to digital converter with scalp signals of $\pm 700\mu V$ (channels 1-4) and $\pm 350\mu V$ (channels 5 & 6). The final function of the modules was to sample and hold the output of the final amplifier at the required instant in time.

This function was performed by a Signetics type LF 398 integrated circuit.

In order to sample the analogue signals at the required rate a pulse signal was generated by an oscillator and divider network which was then fed into each of the six sample and hold IC's. This pulse signal was also fed into the PDP 8/f Hybrid Computer Interface [3] so as to cause an interrupt at every sample instant. Since the sampling was to be performed at precisely regular intervals a quartz crystal oscillator and a frequency divider were built specially. The oscillator (Figure 4-4) used a circuit recommended by the crystal manufacturers [4] and provided a signal of approximately 1 volt RMS at a frequency of 100kHz. This signal was squared by $T_r 2$ and divided in frequency by two Transistor-Transistor Logic (TTL) 7490 decade dividers and a TTL 7493 4 bit binary divider as shown in Figure 4-4. The outputs of the 7493 give possible sampling frequencies of 500, 250, 125 or 62.5 Hz. Since only two of these frequencies were likely to be needed (125 or 250 Hz.) a two position toggle switch was employed to select the required frequency and feed it to the input of a TTL 74121 monostable multivibrator. This device was employed to convert the square wave output of the divider chain into a pulse waveform, with a 100 μ s active period, necessary for the sample hold circuit. The sampling frequencies are accurate to within approximately one part in 10^5 .

The six sampled analogue waveforms were then passed to an Analogue Devices MPX8A multiplexer and a Hybrid Systems Corp. ADC-591-12A-G 12 bit analogue to digital converter. These two devices were part of the PDP 8/f Hybrid Computer Interface

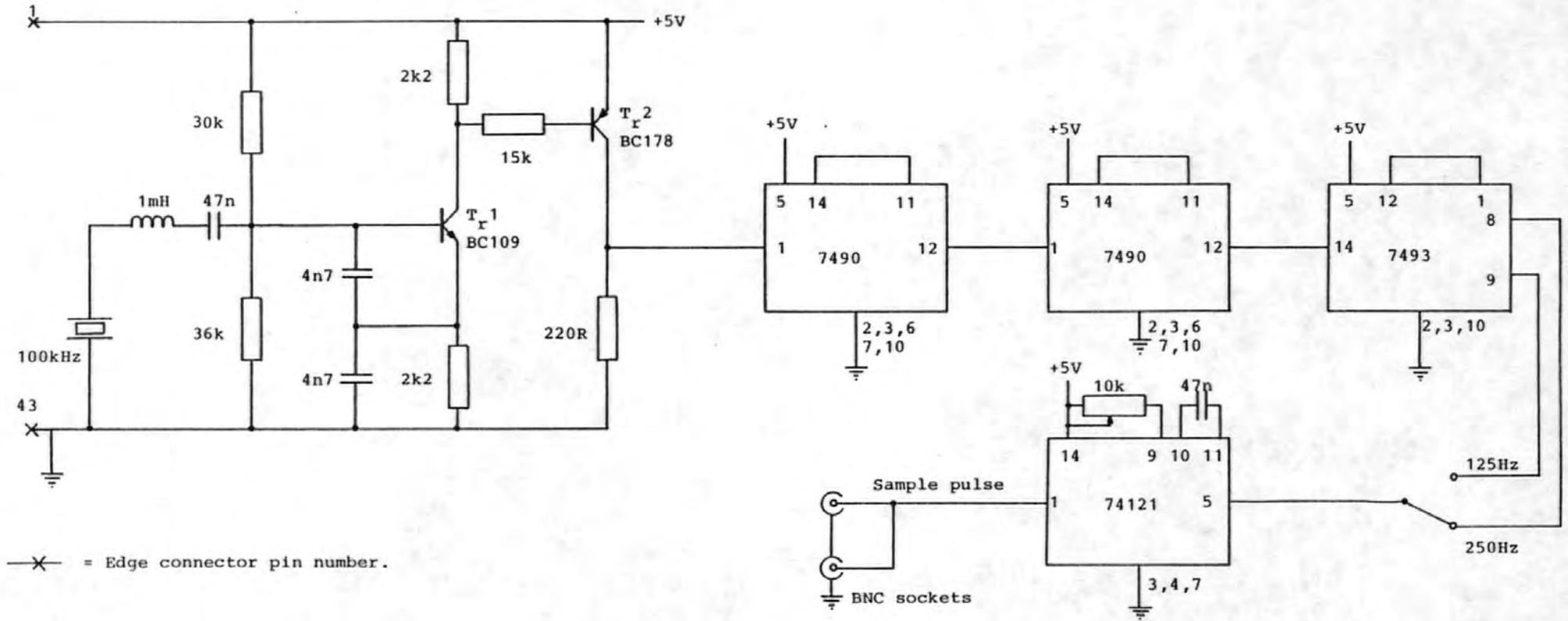


Figure 4-4

The crystal oscillator and divider network

designed and built by Yeats [3].

4.2.1 Testing of the Analogue Electronics

The amplitude and phase responses of one channel of the measurement system are shown in Figures 4-5a and 4-5b. In addition to measuring the amplitude and phase responses other tests were performed to determine the common mode rejection ratio (CMRR) and the effect of the filtering on a ramp waveform (similar to an ideal CNV). The CMRR was measured by connecting the inputs of the channel under test together at the headbox and injecting a 15 Hz. sine wave signal between the connected inputs and earth. The common mode input voltage (V_{ic}) and the output voltage (V_{oc}) were measured and the common mode gain (A_C) found by dividing V_{oc} by V_{ic} . The CMRR was then obtained by dividing the differential gain (A_D) by the common mode gain (A_C). The values of the CMRR for channels 1 and 5 are given in table 4-1.

Table 4-1

The Common Mode Rejection Ratio for Channels 1 and 5

CHANNEL	A_D	A_C	CMRR	CMRR dB
1 †	14290	0.182	78520	97.9
5 †	28570	0.212	134800	102.6

†

The time constant was set to 10 seconds and the low pass filter was set to a cut-off frequency of 30 Hz.

Figure 4-5a

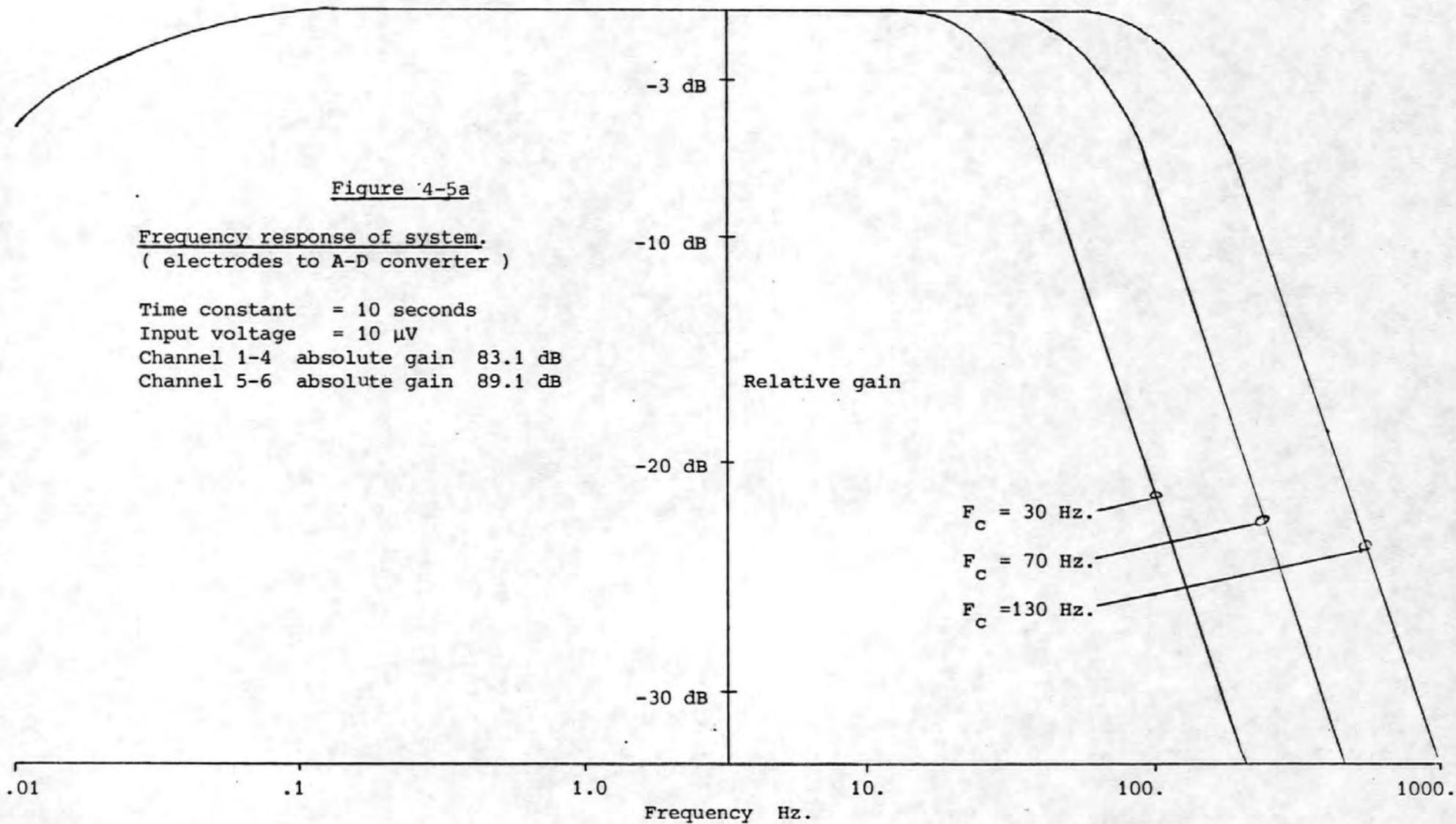
Frequency response of system.
(electrodes to A-D converter)

Time constant = 10 seconds

Input voltage = 10 μ V

Channel 1-4 absolute gain 83.1 dB

Channel 5-6 absolute gain 89.1 dB



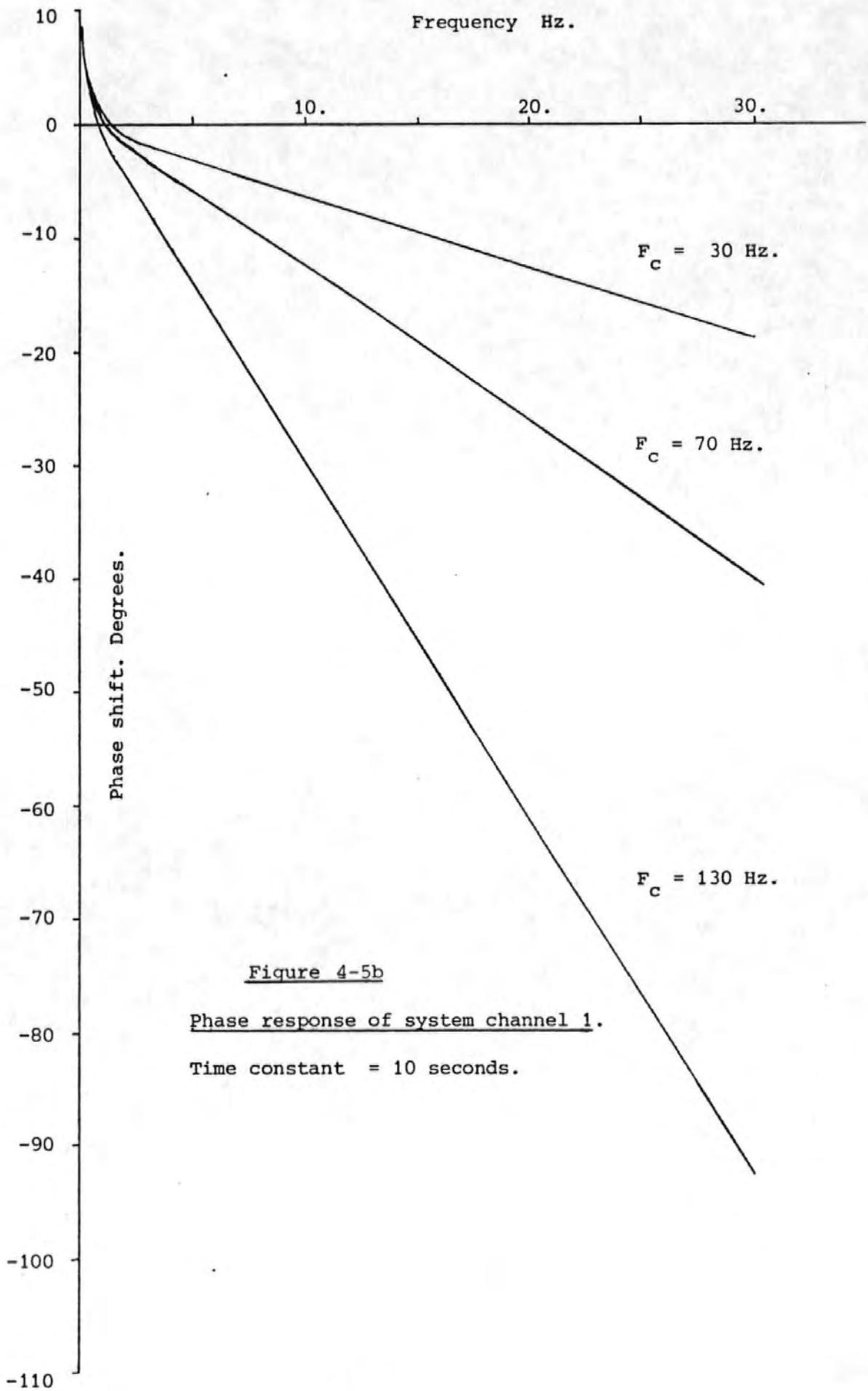


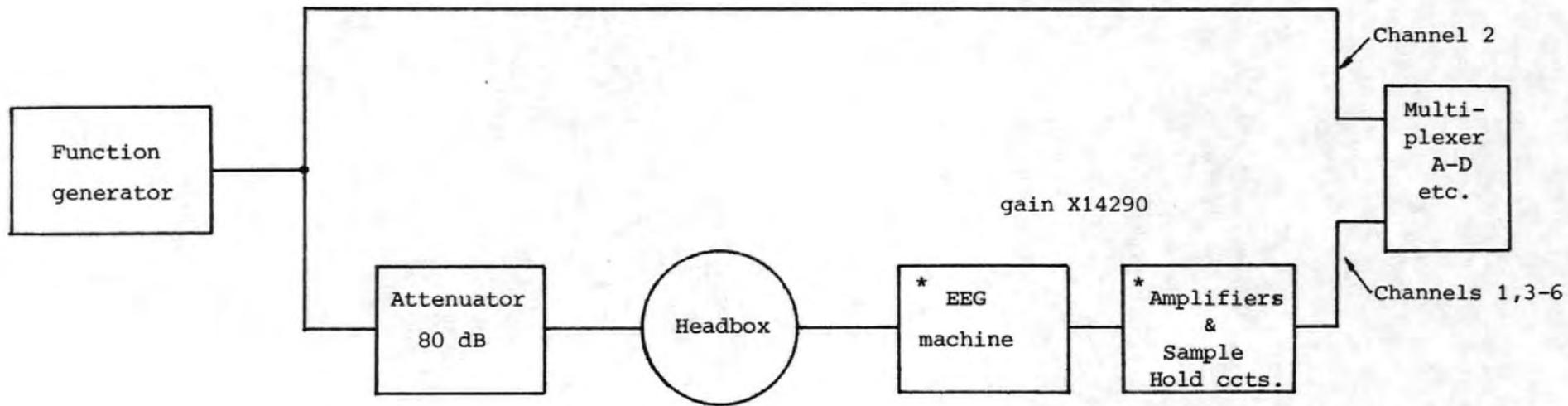
Figure 4-5b

Phase response of system channel 1.

Time constant = 10 seconds.

The circuit of Figure 4-6 was used to show the likely effect of the low and high pass filters on an ideal CNV response. The function generator was set to deliver a ramp waveform which was fed into an attenuator and also to the Channel 2 multiplexer input. The output of the attenuator was fed into the headbox and amplified and filtered in the usual way before being passed on to the multiplexer. Figures 4-7a and 4-7b show the two waveforms, one having been subjected to low and high pass filtering, and the other having undergone no filtering.[†] It may be observed from these diagrams that the distortion introduced is quite small. The settings of the low and high pass filters were 30 Hz. and 0.016 Hz. (time constant = 10 seconds). The amplitude of the ramp signal at the headbox input connections was as indicated in Figure 4-7b. Because of the unusual connection the voltages indicated in Figure 4-7a must be multiplied by the system gain (14290) to give the true voltage levels.

[†] except that inherent in the sampling process.



* See Figures 4-2 & 4-3.

Figure 4-6

The connection of the measurement system for the ramp response tests

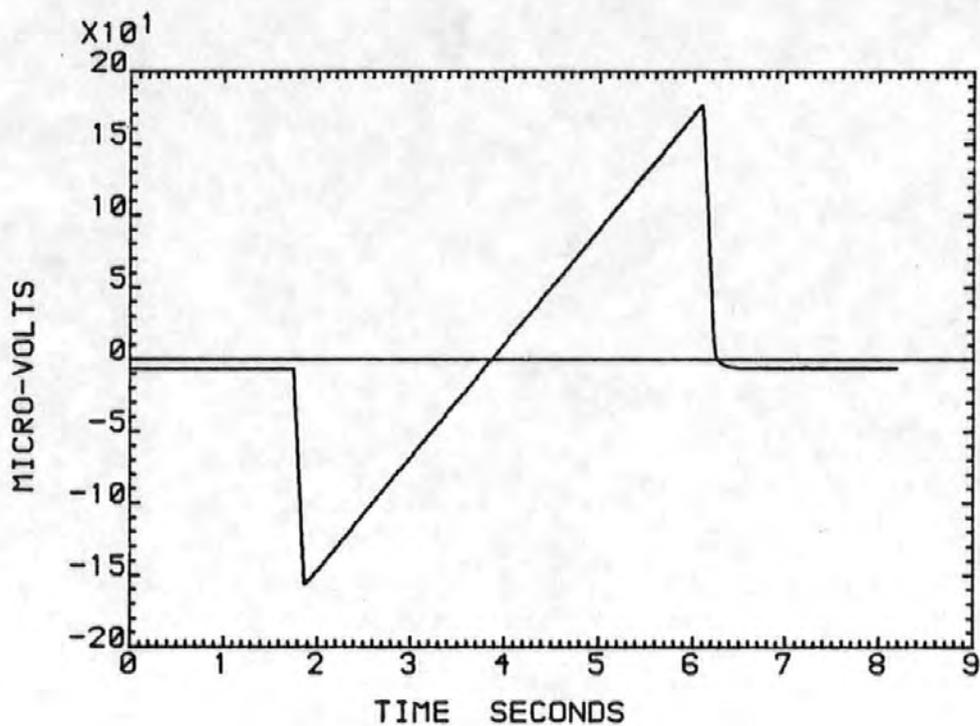


Figure 4-7a

The ramp input waveform (X 14290)

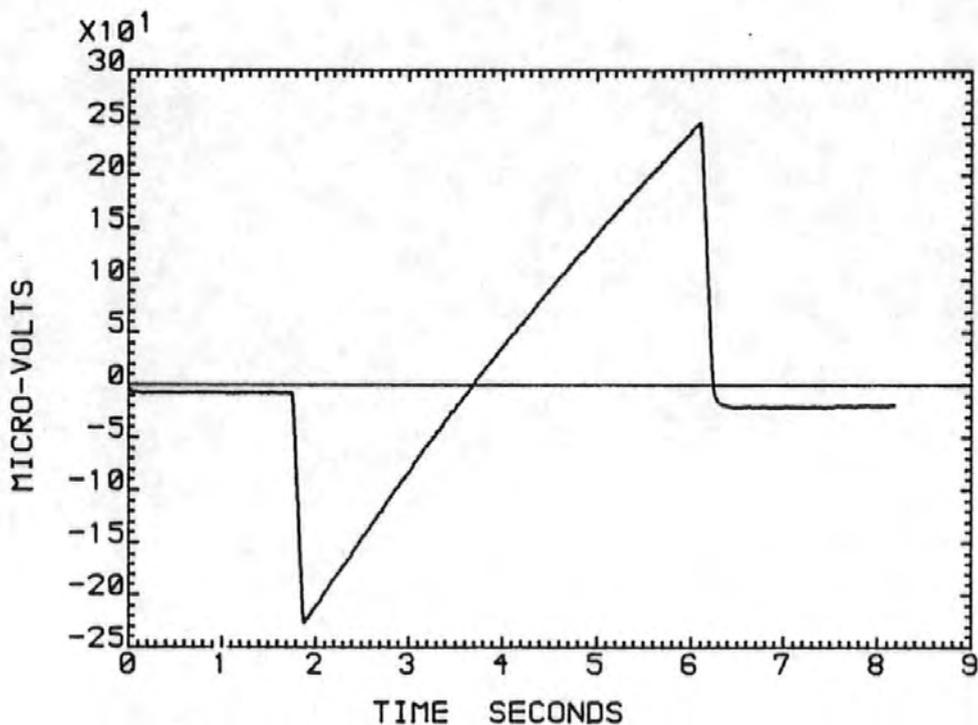


Figure 4-7b

The ramp output waveform.

4.3 High Speed Serial Data Link

In order to transfer the digital data from the mini-computer in the measurement room to the remote minicomputer with the magnetic disk, a pair of high speed serial data transceivers were designed and built. Two identical units were made to allow bi-directional communications to take place over a four-wire link using the 20mA current loop convention. The units were constructed on Vero-cards with 43-way edge connectors suitable for insertion into the extended input/output (I/O) bus racking system with which both the PDP 8's were fitted. This bus comprised of all the input/output timing pulses, the accumulator input and output lines six of the twelve memory buffer lines (from which the I/O devices were addressed), the instruction skip line, the interrupt request line and power supplies of +15, +5, and -15 volts. A block diagram of one of the serial transceivers is given in Figure 4-8. The design of the transceivers was based on two large scale integration integrated circuits. These were the Intersil 6402 Universal Asynchronous Receiver Transmitter (UART) and the Motorola MC 14411 Baud rate generator. The UART is a device capable of translating 8 bit parallel binary data into a serial data stream at a rate determined by an externally supplied clock signal. Simultaneously the device can receive a serial data stream and convert it to an 8 bit parallel binary data word. By means of an internal register, which may be loaded by the user, the device can be instructed to perform many variations on this basic theme. (e.g. the device can be instructed to generate a parity bit and append this to the data being transmitted whilst any received data is

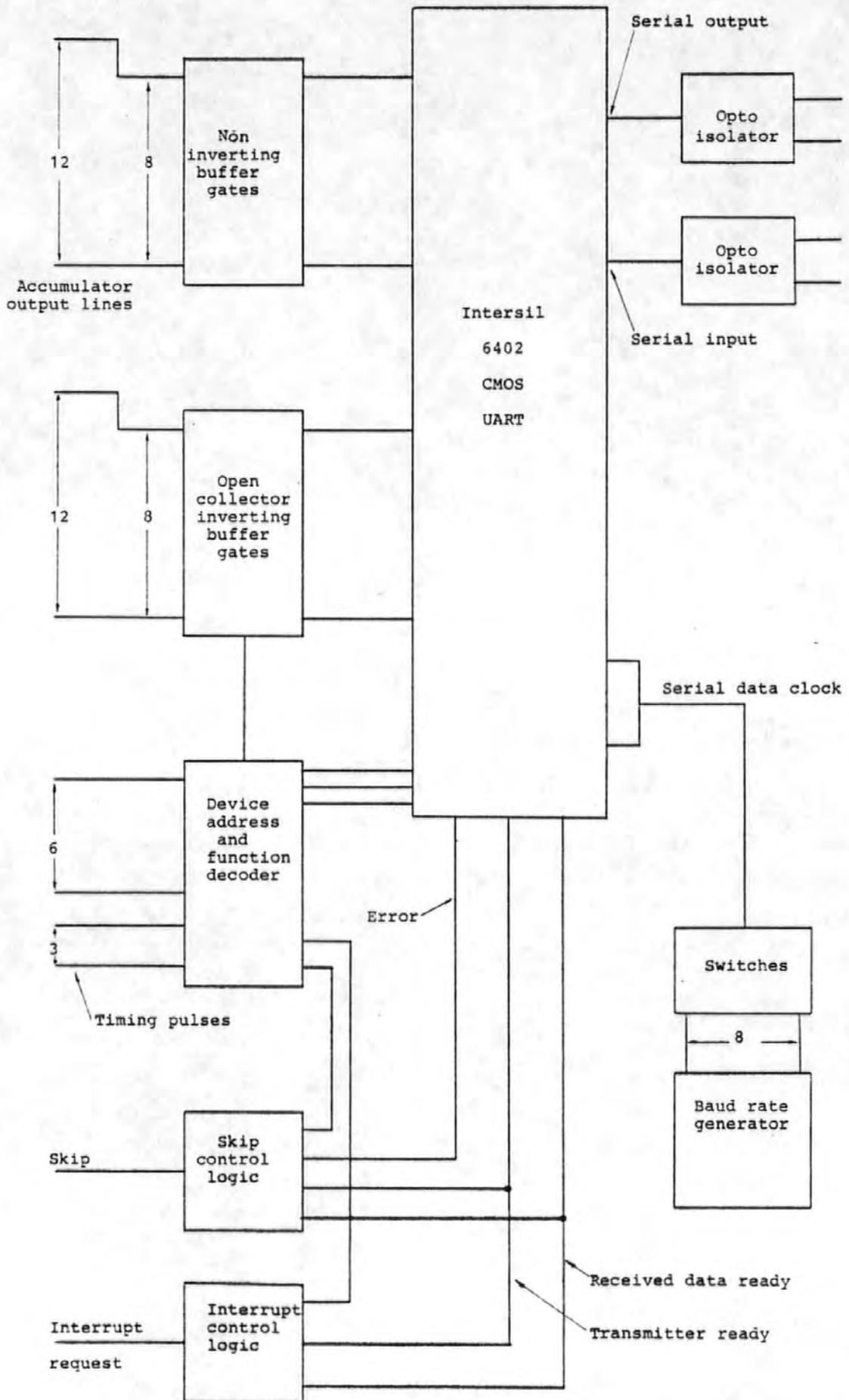
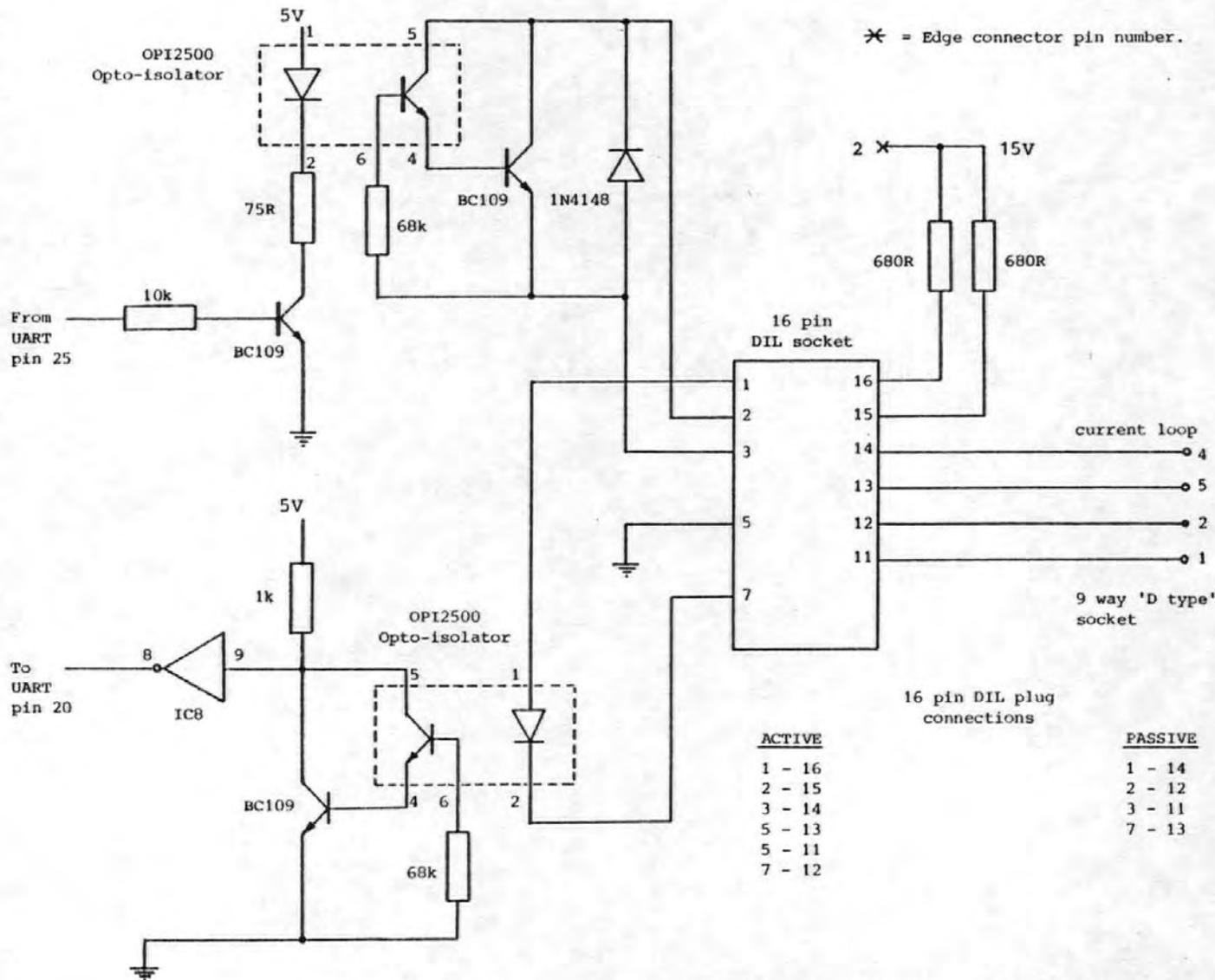


Figure 4-8 A block diagram of the serial transceiver

checked for parity errors). The Baud rate generator was used to generate one of a number of standard frequencies used for serial data transmission. Selection of the particular frequency required was controlled by a group of five dual-in-line switches giving serial data rates from 75 Baud up to 38.4 k Baud. The full circuit diagram of one of the transceivers is given in Figure 4-9. In order to achieve electrical isolation between the equipment in the measuring room and the remote minicomputer, all the data (both transmitted and received) was passed through a pair of opto-isolators.

Numerous Transistor-Transistor Logic (TTL) gates perform the necessary interfacing and decoding between the signals on the 43-way I/O bus and the UART. Additional circuitry, in the form of I/C's 10 and 13, perform the functions associated with the control and generation of programme interrupts. These interrupts may be generated when the UART is ready to transmit another 8 bit word (a transmitter interrupt) or when the UART has just received an 8 bit word (a receiver interrupt). I/C 14 performs an 'inclusive OR' of the three possible receiver error conditions which may then be detected (by the software) upon execution of a 'Skip on error' instruction. Table 4-2 gives a list of the instructions decoded and acted upon by the serial transceivers.



IC types

IC1	7401
IC2	7404
IC3	7430
IC4	7407
IC5	7401
IC6	7430
IC7	7430
IC8	7404
IC9	7401
IC10	7401
IC11	7408
IC12	7400
IC13	7400
IC14	7427
IC15	4051
IC16	6402
IC17	14411

ACTIVE

- 1 - 16
- 2 - 15
- 3 - 14
- 5 - 13
- 5 - 11
- 7 - 12

PASSIVE

- 1 - 14
- 2 - 12
- 3 - 11
- 7 - 13

Figure 4-9b The current loop interface of the serial data transceivers

Table 4-2

Instructions Obeyed By The Serial Data Transceivers

OCTAL INSTRUCTION CODE	MNEMONIC	ACTION
6621	SDR	Skip if the receiver has a data word ready
6631	RUD	Read the data word
6641	SKERR	Skip if an error has been detected
6622	DUI	Disable all interrupts
6632	ERI	Enable receiver interrupts
6642	ETI	Enable transmitter interrupts
6624	STR	Skip if the transmitter is ready for more data
6634	LSTAT	Load the control status register
6644	OUT	Load the data and transmit

A more detailed account of the operation of these instructions and their effect on the serial data transceivers is given in Appendix 8.9.

4.4 Minicomputers and Software

The analogue to digital (A-D) converter and the multiplexer were controlled by a Digital Equipment Co. PDP 8/f minicomputer with 8k words of ferrite core memory. Since evoked potentials were to be studied the computer was also used to present trigger pulses to the external stimulus generators at the appropriate instants. As the digital data was acquired it was stored in the memory of the minicomputer. Simultaneously the data was transmitted over a high speed serial data link to the second PDP 8 minicomputer some distance away from the measurement room for storage on a magnetic disk.

The operation of the complete measuring system can best be described by outlining the sequence of events involved in obtaining a single evoked response.

- (i) The minicomputer waits until the operator pushes a button to start the acquisition process.
- (ii) Under interrupt control the six analogue data channels are digitised and stored in the memory of the minicomputer. The interrupts are generated by a crystal oscillator and divider giving either 125 or 250 samples per second.
- (iii) Under interrupt control the data stored in the memory of the minicomputer is transferred via the serial data link to the second (remote) minicomputer. The interrupts are generated every time the serial data link becomes inactive.

- (iv) When not busy (i.e. when not interrupted) the minicomputer displays via two ten bit digital-to-analogue converters and an X-Y CRT display any one of the six incoming data channels.
- (v) When each sample is taken the minicomputer checks the sample number and if a preset number have been taken it sends a pulse to trigger the first stimulus generator.
- (vi) When each sample is taken the minicomputer checks the sample number and if another preset number have been taken it sends a pulse to trigger the second stimulus generator.
- (vii) After each sample has been taken a test is performed to determine whether the required total number of samples have been taken and if so whether they have all been sent to the remote minicomputer for permanent storage.
- (viii) If both conditions in (vii) are met then the minicomputer waits for the remote minicomputer to acknowledge that the data has been successfully stored before returning to state (i) above.

The sequence of events for the remote minicomputer is as follows;

- (i) Send the 'Ready' signal to the minicomputer in the measurement room.
- (ii) Wait for the data and store it in core memory as it arrives. Perform tests to detect transmission errors.

(iii) When all the data for the evoked response has been received store it on the magnetic disk.

(iv) Return to state (i)

The two programmes to perform the above tasks were written in the PAL 8 assembly language and listings are given in Appendices 8.10 and 8.11.

It should be noted that the speed of the serial data link may be greater or less than the rate at which the data is acquired since the computer memory is used as a temporary storage buffer.

For analysis and processing of the evoked responses the data stored on the magnetic disk of the PDP 8 minicomputer was transferred on either paper tape or floppy disks to a large database on the Polytechnic main computer.

4.5 Other Equipment

4.5.1 Eye Movement Correction Screen

Our experiments comparing the effectiveness of various methods of removing eye movement artefact from the EEG were undertaken with the aid of a simple wooden screen shown in Figure 4-10. The screen was approximately 1.3m square into which a set of nine 5 mm red light emitting diodes (L.E.D.'s) were inserted on a 0.55m grid. A remote control box allowed any of the LED's to be switched on or off as required. The procedure adopted for using the screen is described in Section 3.3.

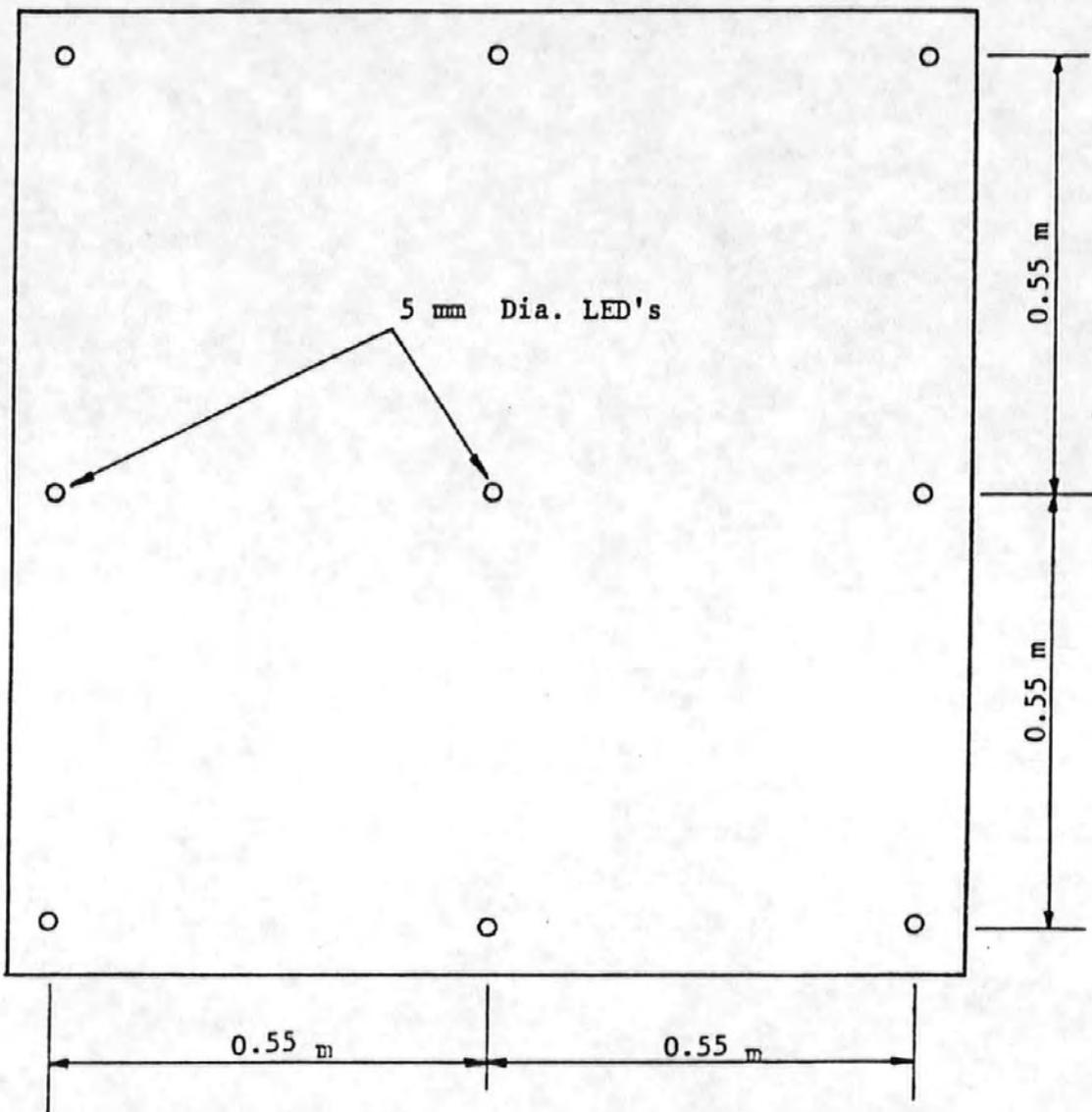


Figure 4-10

The eye movement correction screen.

4.5.2 The CNV Stimulus Generator

In order to record CNV's it is necessary to present the subject with some form of stimulus. The type of stimulus is not important to the final result and the two most commonly used stimuli are auditory and visual. Some workers use two auditory stimuli (e.g. a click followed by a tone [5]) whereas others use combinations of auditory and visual stimuli (e.g. a click followed by a flashing light [6]). Our initial experiments at Freedom Fields Hospital had indicated that the use of a visual stimulus was to be avoided if possible since this often resulted in the subject blinking and hence introducing an artefact in synchronism with this stimulus.

The stimulus paradigm subsequently chosen was that of a click followed by a 1kHz. tone of 90dB. intensity (A weighting). The circuit used for generating this is given in Figure 4-11.

In order to obtain correct CNV responses the subject must perform some action upon receipt of the second stimulus. It is normal practice to use a hand held push button which the subject is required to press.

It is also normal for the push button to be connected to the stimulus generator in order to terminate the second stimulus. This feature gives the subject some motivation for responding to the second stimuli and was incorporated in our stimulus generator.

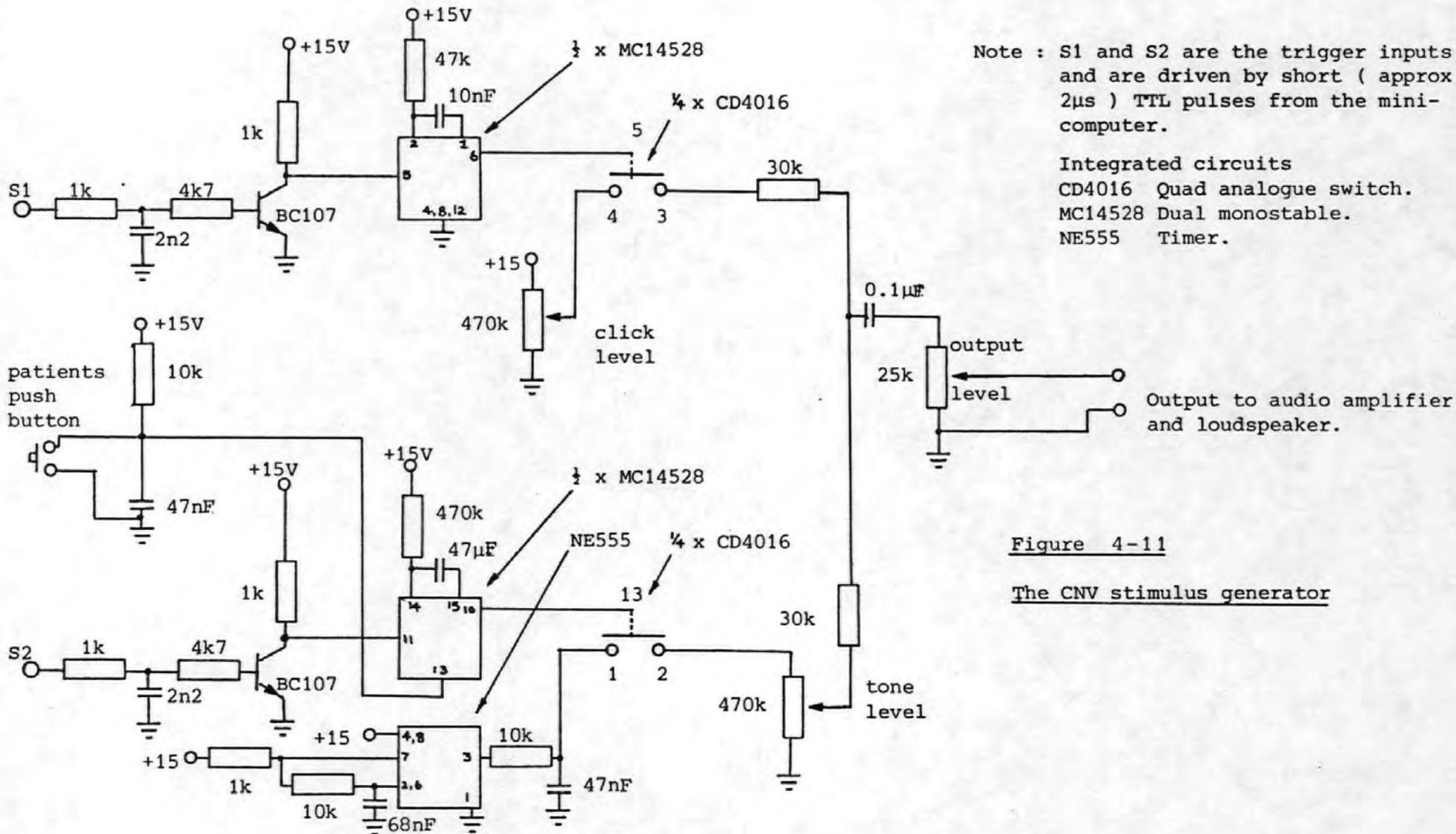


Figure 4-11

The CNV stimulus generator

4.5.3 The AEP Stimulus Generator

The equipment used in the recording of the AEP's was virtually identical to that used for the CNV's. The only difference was in the generation of the stimulus. The auditory stimuli were delivered from an Amplaid stimulus generator to the right transducer of a pair of Koss K6 stereo headphones. Triggering of the stimulus generator was performed at the required instants by the PDP 8 minicomputer. Two different levels of stimulus were applied. A 1kHz. tone of 100ms duration including 10ms of rise and fall according to a cosine squared law was used. The low level stimulus was applied at a level of 40dB. above the subject's auditory threshold, while the high level stimulus was applied 70dB. above it. These stimuli were presented as the subject sat relaxed in a chair.

References for Section 4

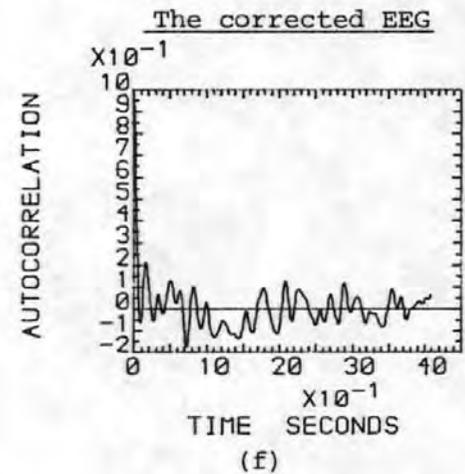
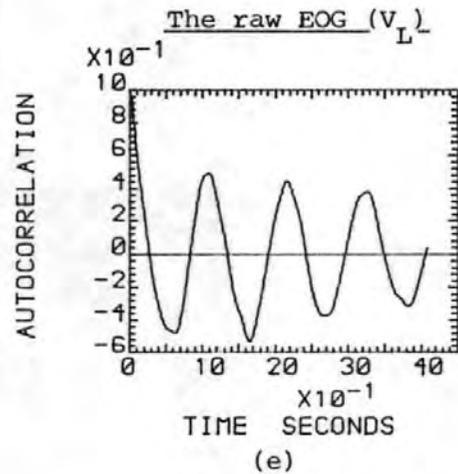
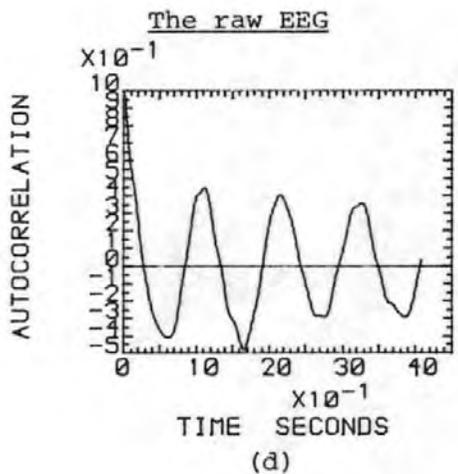
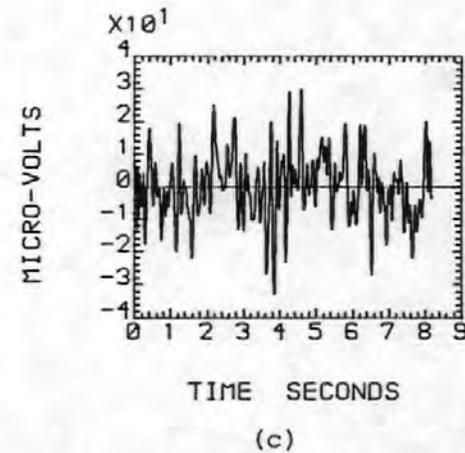
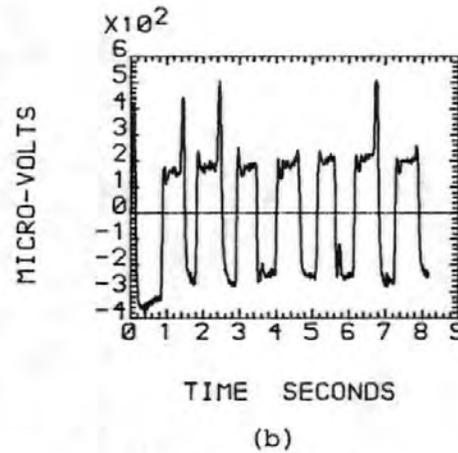
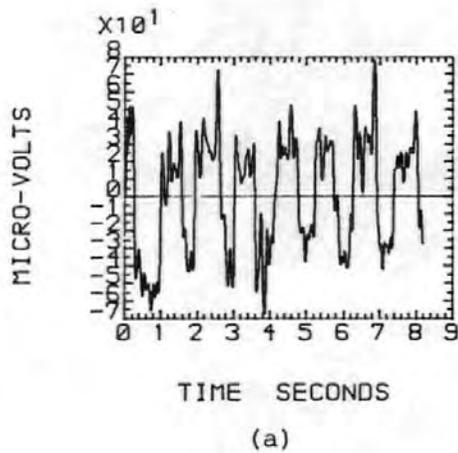
- [1] Sallen R P, Key E L, "A Practical Method of Designing R C Active Filters"
Institute of Radio Engineers, Transactions on Circuit Theory, Volume CT2, pp 74-85, 1955.
- [2] Millman J, Halkias C C, "Integrated Electronics: Analog and Digital Circuits and Systems"
Chapter 16, page 550. Published by McGraw-Hill Kogakusha Ltd., 1972.
- [3] Yeats R W, "Hybrid Computing and its Applications"
Internal Report, School of Electrical Engineering, Plymouth Polytechnic, January 1975.
- [4] Interface Quartz Devices Ltd., "Typical Oscillator Circuits"
Available from Interface Quartz Devices Ltd., 29 Market Street, Crewkerne, Somerset TA18 7JU.
- [5] Hillyard S A, Galambos R, "Eye Movement Artifact in the CNV"
Electroencephalography and Clinical Neurophysiology, Volume 28, pp 173-182, 1970.
- [6] Walter W G, Cooper R, Aldridge V J, McCallum W C, Winter A L, "Contingent Negative Variation: An Electric Sign of Sensorimotor Association and Expectancy in the Human Brain"
Nature, Volume 203, pp 380-384, July 1964.

5. Results and Discussion

5.1 Eye Movement Corrections

A set of typical waveforms depicting the raw EEG, the raw (vertical left) EOG and the corrected EEG are given in Figure 5-1 (a-c). Also shown are the autocorrelation functions (a.c.f.'s) of the EEG, the EOG and the corrected EEG signals Figure 5-1 (d-f). Table 5-1 gives the results as evaluated from the plots derived from four experiments. This table gives both the frequencies and the autocorrelation co-efficients (a.c.c.'s, see Section 2.4 for details) of each of the corrected EEG's along with the frequency of the EOG deduced from its own a.c.f. Some of the frequencies present appear to be harmonics of the EOG which suggests that the path between the eye and the scalp electrode may be frequency selective or non-linear. If this is so then the proportions of the harmonics making up the rectangular EOG will not be maintained at the scalp electrode. This will result in some harmonics being over or under corrected and thus remaining in the output waveform.

When the artefact contains components due to both eye movements and blinks (which do not have the same effect on the EEG [1]), then the computerised correction will attempt to achieve an optimum correction so that as much as possible of both components is removed. This is still a compromise however, since neither artefact signal can be completely removed by this method whilst the other is present in the same section of signal record. The methods of McCallum and



The ACF of (a)

The ACF of (b)

The ACF of (c)

Figure 5-1

Table 5-1

QUANTITATIVE RESULTS OF EYE MOVEMENT
EXPERIMENTS

EYE MOVEMENT	PARAMETER	2 CHANNEL CORRECTION (± 0.02)	3 CHANNEL CORRECTION (± 0.02)	4 CHANNEL CORRECTION (± 0.02)	ANALOGUE CORRECTION (± 0.02)	EOG FREQ. (± 0.02)	INTERPRETATION
DOWN	FREQ.	0.70*	N.P.	N.P.	0.38*	0.38	A B
	r	0.09	N.P.	N.P.	0.38		
LEFT	FREQ.	N.P.	N.P.	N.P.	0.27 +	0.36	A
	r	N.P.	N.P.	N.P.	0.16		
UP	FREQ.	N.P.	N.P.	N.P.	0.33 +	0.38	A
	r	N.P.	N.P.	N.P.	0.22		
RIGHT	FREQ.	0.55*	0.50*	0.50*	0.55*	0.52	A
	r	0.22	0.15	0.15	0.28		

FREQ. = frequency of autocorrelation function (Hz)

r = autocorrelation coefficient at T=2 sec.

N.P. denotes a non-periodic autocorrelation function indicating complete correction.

A means correlation technique better than analogue technique.

B means 4 and 3 channel correction are better than 2 channel.

* denotes a frequency related to the EOG frequency.

+ denotes a frequency unrelated to the EOG frequency.

Walter, and Girton [2] and Kamiya [3] make no allowances at all for the differential contributions of blinks and eye movement artefacts.

The deductions made from forty-eight experiments are summarized in Table 5-2. Due to the amount of computer time involved only the vertical and horizontal eye movements were analysed (i.e. The Forty-eight other experiments involving the subjects looking to the corners of the eye movement screen were, with a few exceptions, not analysed).

Table 5-2

Summary of Results of Quantitative
Eye Movement Experiments

INTERPRETATION	VERTICAL EYE MOVEMENT	HORIZONTAL EYE MOVEMENT
A	20	17
D	3	7
E	3	3
F	2	0
G	1	0
H	1	1

- A means the correlation correction method was better than the analogue method.
- D means both the correlation correction and the analogue correction were equally effective.
- E means the 3-channel correction was better than the 2 channel correction.
- F means the 2 channel correction was better than the 3 channel correction.
- G means the 4 channel correction was not the best.
- H means the correlation correction was incomplete.

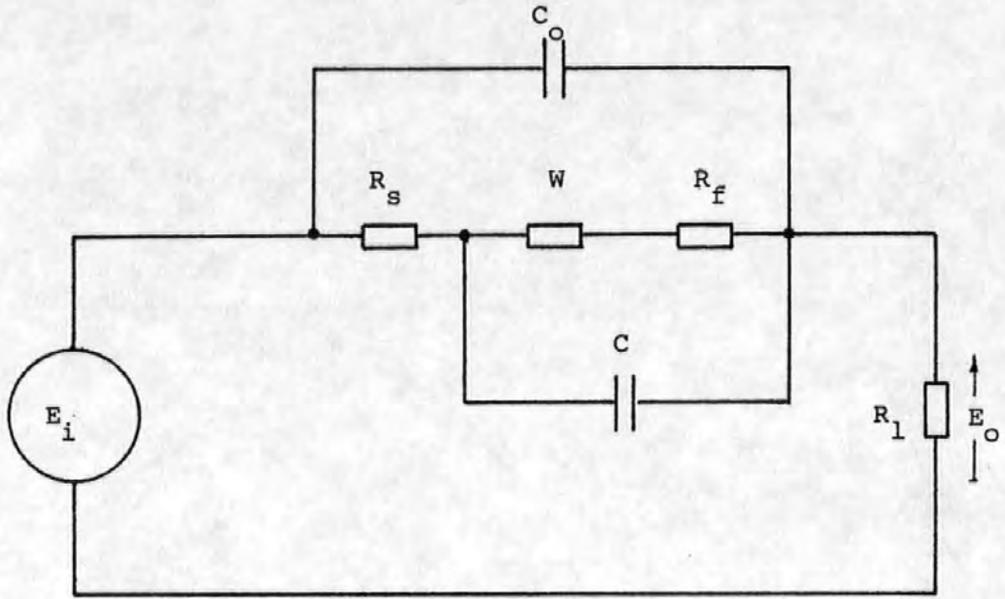
From Table 5-2 it was deduced that:

- (i) The correlation correction technique was complete in 96% of the cases.
- (ii) In the majority of cases (77%) the correlation correction technique was superior to the analogue correction technique.
- (iii) In 21% of the cases it was not possible to distinguish between the effectiveness of the analogue and correlation techniques.
- (iv) The evidence from interpretations E and F suggests that to correct for horizontal eye movements it may be necessary always to use the 3-channel correction rather than only a 2-channel correction. This is to be expected since the ocular dipoles tend to oppose each other during horizontal eye movements whereas they re-inforce each other during vertical eye movements.
- (v) Interpretation G indicates that in 98% of cases the 4-channel correction is as good as or better than the other methods. However if computing time is at a premium (e.g. in an on-line situation) it would probably be adequate to rely on the 3-channel correction method.
- (vi) Interpretation D shows that the analogue technique was as good as the correlation technique more often for horizontal eye movements than for vertical eye movements. (7 times against 3 times out of 48 experiments). Since the analogue technique is not intended to correct for horizontal eye movements, this suggests that it may

actually be an unsuitable one to use to correct for vertical eye movements.

In addition to the points previously mentioned, two other disadvantages of the analogue techniques were noted. Firstly it was difficult to optimise the potentiometer setting required to minimise the artefact and analogue correction would therefore be very time consuming in multi-channel recording. Second, in all the subjects tested there was a degree of coupling between the EOG and EEG when the EOG changed rapidly. This was probably due to the R.C. network formed by the potentiometer resistance ($25k\Omega$) and the electro-chemical capacitance of the electrode-skin interface. This effect was reduced when a larger value potentiometer ($1M\Omega$) was used. Unfortunately however this solution adversely affects the common mode rejection ratio of the EEG machine's differential amplifier (because of the increased source impedance of the input with the $1M\Omega$ potentiometer in circuit). The presence of this R.C. time constant will result in distortion of the EOG and hence false correction of the artefact. This may be the reason behind the evidence in (Vi) above that the analogue technique may even be erroneous. This is illustrated in Figure 5-2.

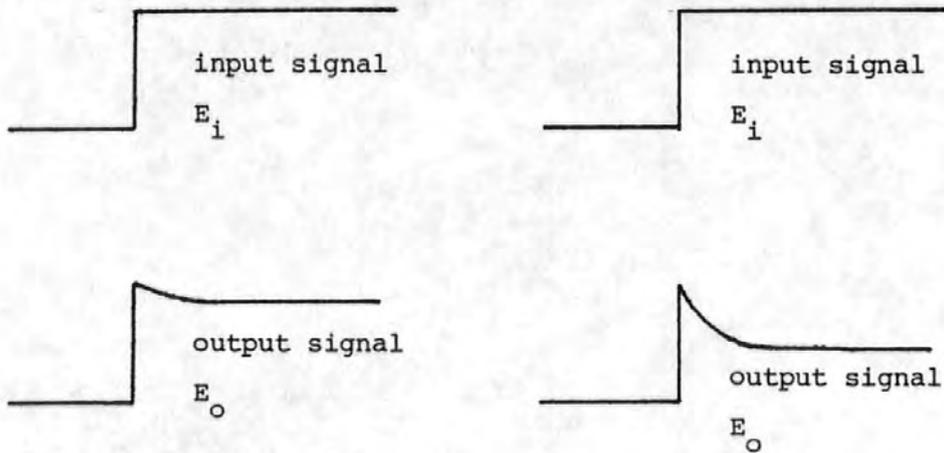
In conclusion it was found that the 4-channel computerised correlation method of correcting the EEG for eye movement artefact gave complete correction in 96% of the cases studied. However, in practice the 3-channel method would normally be adequate. The analogue technique was never the best. Furthermore it was very time consuming and quite possibly erroneous.



- E_i scalp EEG signal.
- C_o stray capacitance (usually negligible).
- R_s resistance of the electrode jelly.
- R_f Faradic resistance.[†]
- W Warburg impedance.[†]
- C electro-chemical capacitance.

see [11] for further details.

[†] These quantities are usually resistive and small for Ag-AgCl electrodes.



Normal situation when R_l is large in comparison with $R_f + W$.

The situation when R_l is small.

Figure 5-2 The effect of electrode loading.

Some of the above work has been published [4].

A more detailed account has also been written as a longer paper with the title "Comparison of Methods for Removing Eye Movement Artefact from the EEG" for submission to the IEEE Transactions on Biomedical Engineering. This paper has not yet been submitted at the request of the National Research Development Corporation who have awarded a grant towards the development of a Commercial EEG Eye Movement Artefact Corrector based on the above work which they intend to patent.

5.2 Auditory Evoked Potentials

64 pre-stimulus and 64 post-stimulus realisations each of length 0.512 s were recorded at the two levels of stimulus for three different subjects.

Following the earlier work [5-8], the pre- and post-stimulus realisations were given the statistical test described in 2.2.3.1 above, the test for changes in signal energy, and phase histograms were also plotted for the first six harmonic frequency components. The statistical test showed that only one out of the six sets of results (3 subjects x 2 stimulus levels) exhibited a significant energy difference and in the sense that the energy in the post-stimulus case exceeded that in the pre-stimulus case.

The results of these tests are shown in table 5-3.

Table 5-3

Comparison of Pre- and Post- Stimulus AEP Energy

	Subject	Mean of difference $\times 10^{-8}$	ν (degrees of freedom)	T Statistic	% Significance
LOW LEVEL STIMULUS (+ 40dB)	1	0.0075	63	0.129	N-S
	2	- .1305	63	-0.709	N-S
	3	0.6817	63	2.729	1.0
HIGH LEVEL STIMULUS (+ 70dB)	1	0.1265	63	1.548	20
	2	0.2238	63	1.129	N-S
	* 3	0.2558	63	0.886	N-S

* 65dB. above threshold

The results obtained were in accord with Sayers' findings [5] since he could not establish any consistent difference between the pre- and post- stimulus energies of the averaged waveform. This result led Sayers to the important conclusion that if there were no additional energy in the evoked response, then the characteristic shape of the AEP must be due to some form of phase alignment of certain background components.

However, this test was not considered very reliable since the contribution of every harmonic component was included in the energy calculation and the higher harmonics were probably representative of background noise. Thus the effects of background noise were likely to mask the presence of the AEP. Furthermore, while the phase histograms indicated phase ordering they did not reveal the cause. For these reasons the findings of the previous workers [5-8] although confirmed, were not considered conclusive.

The data was also subjected to the other statistical tests described in sections 2.2.2 and 2.2.3. These tests held potential advantages over the previous test since they were applied to individual harmonic components. They could therefore be used to investigate the first few harmonics which would be expected to constitute the major components of the AEP. The higher harmonics which were expected to be associated with the random background could be ignored. Typical pre- and post- stimulus phasor diagrams are shown in Figure 5-3. The triangles on the circles indicate the directions in which the phasors lie, while the crosses indicate the locations of the phasor tips. The results of the statistical tests carried out for harmonic numbers 1-6 are shown in Table 5-4. Each broad column of results contains the value of the test statistic, the number of degrees of freedom ν and the level at which the result is significant. A significant positive result in column (B) means that the nearest mean amplitude is greater than the furthest. A significant positive result in column (C) means that the post-stimulus amplitude is greater than the pre-stimulus

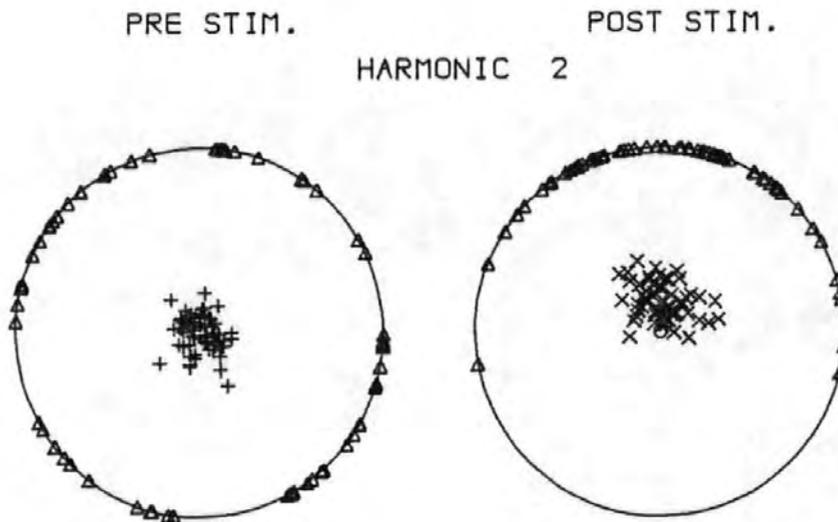
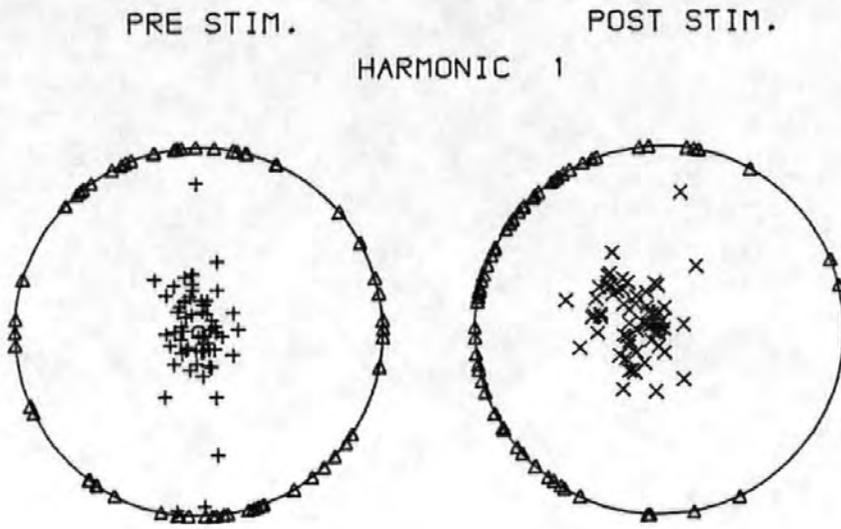


Figure 5-3

Typical Pre- and Post-Stimulus AEP phasor
diagrams

Table 5-4

TEST STATISTICS AND LEVELS OF SIGNIFICANCE FOR STATISTICAL TESTS
ON AEP PHASOR DIAGRAMS BY HARMONIC NUMBER,
STIMULUS LEVEL AND SUBJECT

HARMONIC NUMBER & FREQUENCY	STIMULUS LEVEL	SUBJECT	NEAREST-FURTHEST MEAN AMPLITUDE			PRE-POST MEAN AMPLITUDE DIFFERENCES			RAYLEIGH CIRCULAR VARIANCE		HODGES-AJNE		MODIFIED RAYLEIGH CIRCULAR VARIANCE	
			(B)			(C)			(D)		(E)		(F)	
			T	v	%	T	v	%	So	%	m	%	Do	%
1 1.953 Hz	LOW	1	-0.294	63	N-S	-0.273	63	N-S	0.735	5	19	2.5	0.677	1
		2	3.67	64	.05	3.96	63	0.1	0.612	0.1	12	0.001	0.439	0.1
		3	1.69	63	5	1.01	63	N-S	0.398	0.1	7	0	0.345	0.1
	HIGH	1	1.55	62	10	2.80	63	1	0.538	0.1	13	0.005	0.477	0.1
		2	0.550	61	N-S	5.35	63	0.1	0.278	0.1	4	0	0.224	0.1
		3	-0.136	38	N-S	2.50	63	2	0.332	0.1	5	0	0.336	0.1
2 3.906 Hz	LOW	1	1.68	64	5	2.27	63	5	0.584	0.1	14	0.019	0.512	0.1
		2	0.992	58	N-S	4.13	63	0.1	0.487	0.1	11	0	0.461	0.1
		3	0.700	64	N-S	6.78	63	0.1	0.204	0.1	2	0	0.198	0.1
	HIGH	1	-0.513	59	N-S	4.96	63	0.1	0.440	0.1	10	0	0.340	0.1
		2	5.09	55	.05	3.84	63	0.1	0.445	0.1	12	0.001	0.308	0.1
		3	0.851	64	N-S	5.85	63	0.1	0.266	0.1	6	0	0.205	0.1
3 5.893 Hz	LOW	1	2.33	62	2.5	3.57	63	0.1	0.370	0.1	9	0	0.264	0.1
		2	0.480	64	N-S	2.16	63	5	0.605	0.1	17	0.45	0.504	0.1
		3	0.498	63	N-S	0.632	63	N-S	0.643	0.1	16	0.17	0.594	0.1
	HIGH	1	2.83	64	0.5	3.50	63	0.1	0.303	0.1	5	0	0.199	0.1
		2	0.0	63	N-S	1.60	63	20	0.580	0.1	14	0.019	0.350	0.1
		3	-0.763	42	N-S	1.91	63	10	0.572	0.1	13	0.005	0.502	0.1
4 7.812 Hz	LOW	1	1.51	64	10	0.0	63	N-S	0.735	5	20	5.1	0.648	1
		2	-0.362	63	N-S	-0.310	63	N-S	0.757	5	20	5.1	0.789	N-S
		3	3.08	61	0.5	1.75	63	10	0.542	0.1	14	0.019	0.421	0.1
	HIGH	1	1.75	63	5	1.81	63	10	0.403	0.1	7	0	0.282	0.1
		2	-1.18	61	N-S	-1.40*	63	20	0.777	5	22	20	0.797	N-S
		3	0.995	64	N-S	1.75	63	10	0.458	0.1	11	0	0.417	0.1
5 9.765 Hz	LOW	1	-1.21	47	N-S	1.90	63	10	0.787	10	23	20	0.812	N-S
		2	0.07	64	N-S	-0.092	63	N-S	0.848	N-S	24	N-S	0.763	10
		3	0.814	64	N-S	1.81	63	10	0.713	1	17	0.45	0.662	1
	HIGH	1	1.25	59	N-S	-0.772	63	N-S	0.705	1	20	5.1	0.612	0.1
		2	-0.02	64	N-S	-0.983	63	N-S	0.885	N-S	26	N-S	0.891	N-S
		3	0.551	63	N-S	1.81	63	10	0.463	0.1	11	0	0.456	0.1
6 11.718 Hz	LOW	1	1.65	64	10	2.96	63	1	0.686	1	17	0.45	0.566	0.1
		2	0.821	59	N-S	-2.98*	63	1	0.957	10	28	N-S	0.932	N-S
		3	1.49	64	10	1.83	63	10	0.638	0.1	17	0.45	0.523	0.1
	HIGH	1	1.19	57	N-S	-1.18	63	N-S	0.875	N-S	26	N-S	0.839	N-S
		2	1.37	64	10	-1.64*	63	20	0.799	10	21	9.8	0.708	5
		3	-1.31*	61	10	1.79	63	10	0.861	N-S	25	N-S	0.934	N-S

amplitude. Positive significant results in the remaining columns indicate phase ordering.

N-S means non-significant. While all levels of significance were recorded in order to extract maximum information, levels in excess of 5% were regarded as non-significant. Some of the test statistics in columns (B) and (C) are negative. The ones marked by an asterisk were significant in a negative direction. This occurred in column (C) and meant that the post-stimulus amplitude was less than the pre-stimulus one. This also occurred once in column (B) and indicated that the more widely spaced phasors were larger than the "grouped" ones. However, examination of the corresponding columns (D), (E) and (F) reveals that no phase ordering i.e. grouping had in fact occurred and the test would therefore be unreliable. (See also section 2.2.3.3). The level of significance between columns (B) and (C) do not always agree. The Nearest and Furthest Mean Amplitude Test (one-tailed) depends specifically upon the details of the assumed additive model and is not infallible, vide statistical Test B. By comparison the Pre- and Post- Stimulus Mean Amplitude Differences Test (two-tailed) tests for an additive effect irrespective of the mechanism by which it occurs. Inspection of column (C) showed that thirteen out of the eighteen results obtained for harmonics 1-3 were significant at the 5% level (i.e. additive signal detected in 72% of cases), while for harmonics 4-6 only two out of the eighteen results were significant. For column (B) the corresponding figures were six out of eighteen and two out of eighteen. Inspection

of column (C) showed that only in one case was the mean of the amplitude differences significantly less than zero (i.e. the pre-stimulus phasors were larger than the post-stimulus ones). The remaining twenty-one cases were not significant in either direction. On an individual basis and ignoring significance levels nine cases showed a post-stimulus decrease in energy. Taken together these results indicated that additive energy could be detected in a large percentage of the cases for harmonics 1-3. Interestingly the Pre- and Post-Stimulus Mean Amplitude Differences Test applied to the second harmonic gave a positive significant result for each subject and for both levels of stimulus, and was the only test to do this. This may be a useful result audiologically but further investigation would be advised.

Columns (D), (E) and (F) revealed that the Rayleigh Test of Circular Variance, the Hodges-Ajne, and the Modified Rayleigh Test of Circular Variance were all in good agreement for the first three harmonics. However, the modified Rayleigh Test statistic was less than the Rayleigh statistic in 17 cases out of 18 which suggested that the amplitudes were orientated towards the preferred direction. This was not true for harmonics 4-6. In these cases the Rayleigh and Hodges-Ajne tests agreed but the modified Rayleigh Test did not always agree with them. Inspection of the tabulated results revealed that harmonic components 1-3 were strongly phase ordered while harmonics 4-6 showed less phase ordering. Overall, post-stimulus phase ordering was seen to be significant at the 5% level in all but seven of the thirty six cases observed.

Reflection upon all the observations so far made it apparent that harmonics 1-3 exhibited different amplitude and phase properties to harmonics 4-6. Harmonics 4-6 showed less evidence of phase ordering and even less evidence of additive energy. All of this was regarded as evidence that harmonics 4-6 were more representative of the noisy background EEG than of the evoked response. This in turn lent further support to the previously stated opinion that the comparison of pre- and post- stimulus realisation energies was unlikely to provide a reliable means for the investigation of additive energy, due to the masking effects of the higher harmonic noise.

Returning to Table 5-4, Column (C) taken together with Column (D) showed that when the AEP contained an additive component there was also phase ordering. This finding agreed well with the proposed additive model. Thus out of 29 cases exhibiting phase ordering 14 (or 48.3%) contained an additive component. There was also a pronounced tendency for additive and phase ordering effects to be concentrated in the first three harmonics. In other respects the occurrence of ordering or additivity appeared random. For example, for an individual the occurrence of ordering or an additive component in a particular harmonic at one level of stimulus did not necessarily mean that it would be found in the same harmonic for a different level of stimulus.

5.3 The CNV's of Normal Subjects

CNV responses were recorded from a total of five normal

subjects between the ages of 18 and 60 years. Both one and four second interstimulus intervals were used according to the procedure described in section 3.5. Prior to further processing all the individual CNV responses were corrected for the effects of eye movement artefact as described in section 2.5.

5.3.1 The Averaged CNV's

After processing to remove eye movement artefact the CNV's were averaged, filtered with the low pass digital filter described in section 3.6 and plotted. The resulting averaged CNV's are shown in Figures 5-4 a-e (1 second ISI) and Figures 5-5 a-e (4 second ISI). The averaged CNV waveforms were then used to determine the section of the response to be analysed in the subsequent sections i.e. the 'Negative Variation' section excluding the evoked responses. This section was determined in terms of 'sample numbers' and for the one second ISI was found to lie between samples 472 and 536 (The stimuli being given concurrently with samples 407 and 532), whereas for the four second ISI it was found to lie between samples 295 and 695 (The stimuli being given concurrently with samples 219 and 719). Since the digital filter was used in most of the subsequent analysis, the sample numbers obtained from the averaged CNV responses included a delay due to the filter identical to that which would be obtained in the subsequent processing. Any processing carried out on unfiltered data used points 462 and 526 (1 second ISI) or points 285 and 685 (4 second ISI).

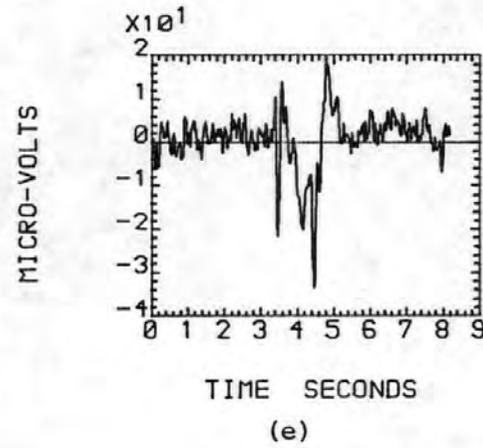
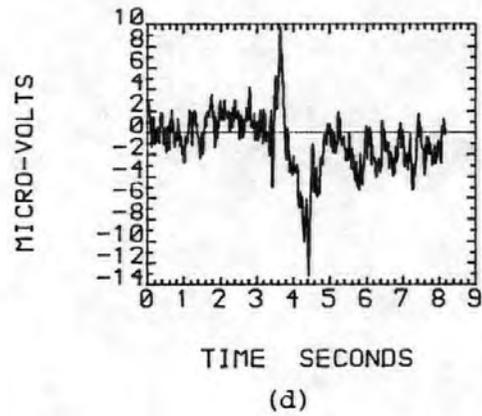
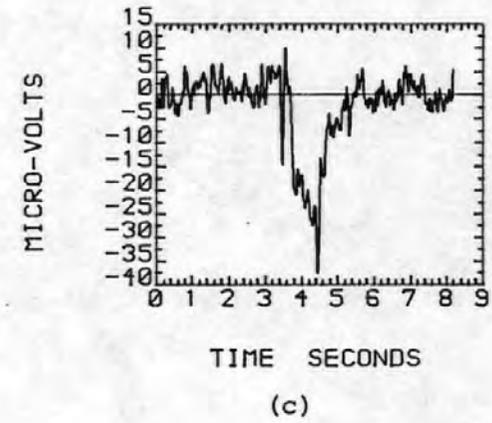
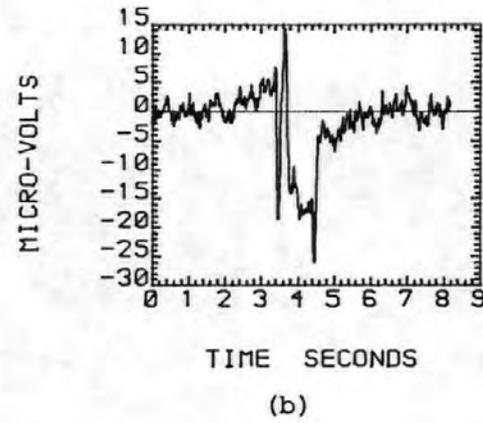
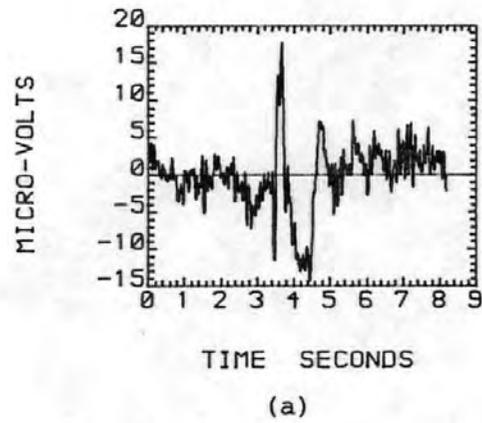
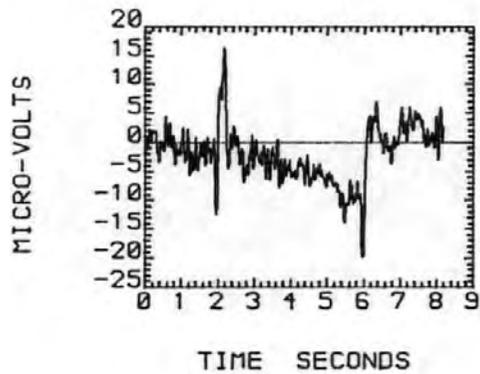
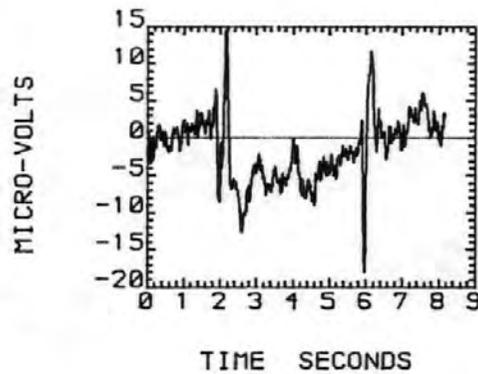


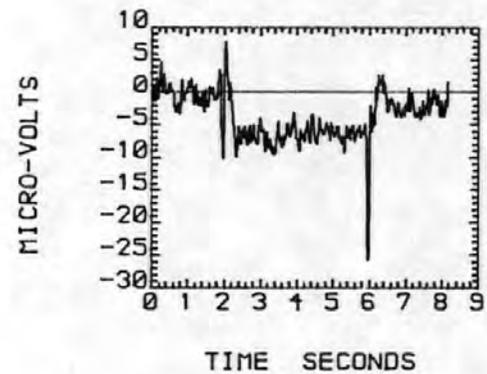
Figure 5-4
The averaged one second ISI CNV's
of normal subjects.



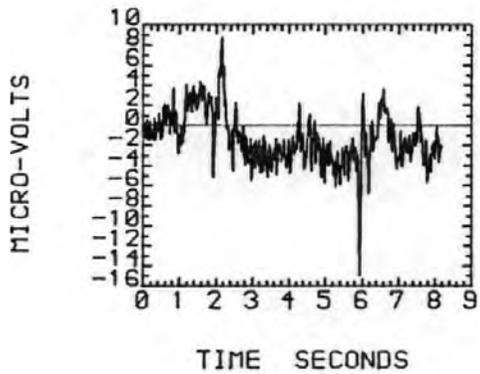
(a)



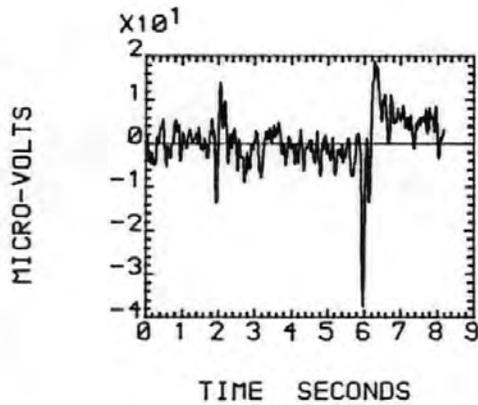
(b)



(c)



(d)



(e)

Figure 5-5
The averaged four second ISI CNV's
of normal subjects.

5.3.2 Energy Tests

These tests were performed to establish whether any additive component could be detected in the CNV response.

5.3.2.1 Broadband Energy Tests

Prior to the calculation of the mean energy for the pre- and post- stimulus sections of the CNV's, the data was corrected for the effects of eye movements and filtered in the usual manner. One of the side effects of the eye movement correction procedure was to remove any d.c. component present in the data epoch and, where an individual CNV response contained a marked negative shift, the effect of this was to cause a positive shift of the pre- and post-stimulus baselines. Thus prior to the calculations described in section 2.2.3.1 it was necessary to re-establish the true baseline. This was achieved by subtracting the mean signal level, calculated over that section of the data prior to the S1 stimulus, from the data. Furthermore, to allow for any small d.c. drift during the acquisition of the data, the mean signal level was also calculated for that section of the data from a point one second after the S2 stimulus (to allow the S2 evoked response and the CNV to return to the zero level) to the end of the data record. This value was subtracted from this section of the data. Between these two mean values the data was corrected by subtracting the appropriate fraction of the difference between these values. The overall correction was thus:-

Section of data
(sample numbers)

Corrections used

$1 \leq k \leq S1$

$$x_k = x_k - \frac{1}{S1} \sum_{i=1}^{S1} x_i$$

$S1 < k \leq S2+D$

$$x_k = x_k - \left[\frac{\left(\frac{1}{(N-S2-D)} \sum_{i=S2+D}^N x_i \right) - \left(\frac{1}{S1} \sum_{i=1}^{S1} x_i \right)}{S2 + D - S1} \right] (k-S1)$$

$S2+D < k \leq N$

$$x_k = x_k - \frac{1}{(N-S2-D)} \sum_{i=S2+D}^N x_i$$

Where $S1$ is the sample number corresponding to the instant of the $S1$ stimulus application.

$S2$ as above for $S2$.

D is the delay after $S2$ to allow the responses to settle. (a figure of one second ($D = 125$) was used).

x_i is the i^{th} data point.

N is the total number of data points.

This correction is subsequently referred to as the baseline correction.

After applying this correction to each CNV response the mean energy values were calculated and subjected to a T-test as previously described. The results of these tests are shown in Tables 5-5a and 5-5b.

Table 5-5a

Broadband energy tests of normal subjects

One second ISI CNV's

Subject	Number of Responses	T Statistic	Significance Level %	Number of* Responses Post < pre
1	31	2.91	1.0	7
2	32	7.05	<<0.1	1
3	19	5.22	<<0.1	3
4	32	2.18	5.0	10
5	32	4.69	<<0.1	6

Table 5-5b

Broadband energy tests of normal subjects

Four second ISI CNV's

Subject	Number of Responses	T Statistic	Significance Level %	Number of* Responses Post < pre
1	32	2.47	2.0	11
2	32	3.47	0.2	3
3	32	4.39	<<0.1	5
4	32	3.72	0.1	7
5	32	5.55	<<0.1	5

* The number of individual responses where the post-stimulus energy was less than the pre-stimulus energy.

They show that for both ISI's and to a level of significance of at least 5%, and mostly considerably better, the CNV response contained more energy than the background EEG prior to stimulation. The conclusion drawn from this was that the CNV must contain additional energy. On an individual basis however, 58 out of 306 (18.9%) of the CNV responses contained less energy than the background EEG. This possibly reflected the variability of the background EEG and of the CNV rather than it implied the CNV energy might have been less than the background.

5.3.2.2 Amplitude Histograms

The background EEG has a virtually random amplitude and phase but an added signal component due to the CNV might be expected to have a constant amplitude and phase. If the amplitude of the added component were the same for each stimulus and were large in comparison with the background EEG, then the n^{th} harmonic component of each response would be similar. Hence the n^{th} harmonic amplitude histogram for a sequence of CNV's should exhibit a peak about some preferred value. Amplitude histograms were plotted to test this hypothesis. It was found that the amplitude histogram of the first harmonic of the one second ISI CNV's showed some similarities. In particular most exhibited a lower amplitude limit while the most probable amplitudes were different for different subjects. A typical histogram is shown in Figure 5-6. None of the other harmonics showed any obvious patterns. It was thought that the absence of definite patterns in the histograms of the higher harmonics might indicate the masking effects of the background EEG. Accordingly a comparison of

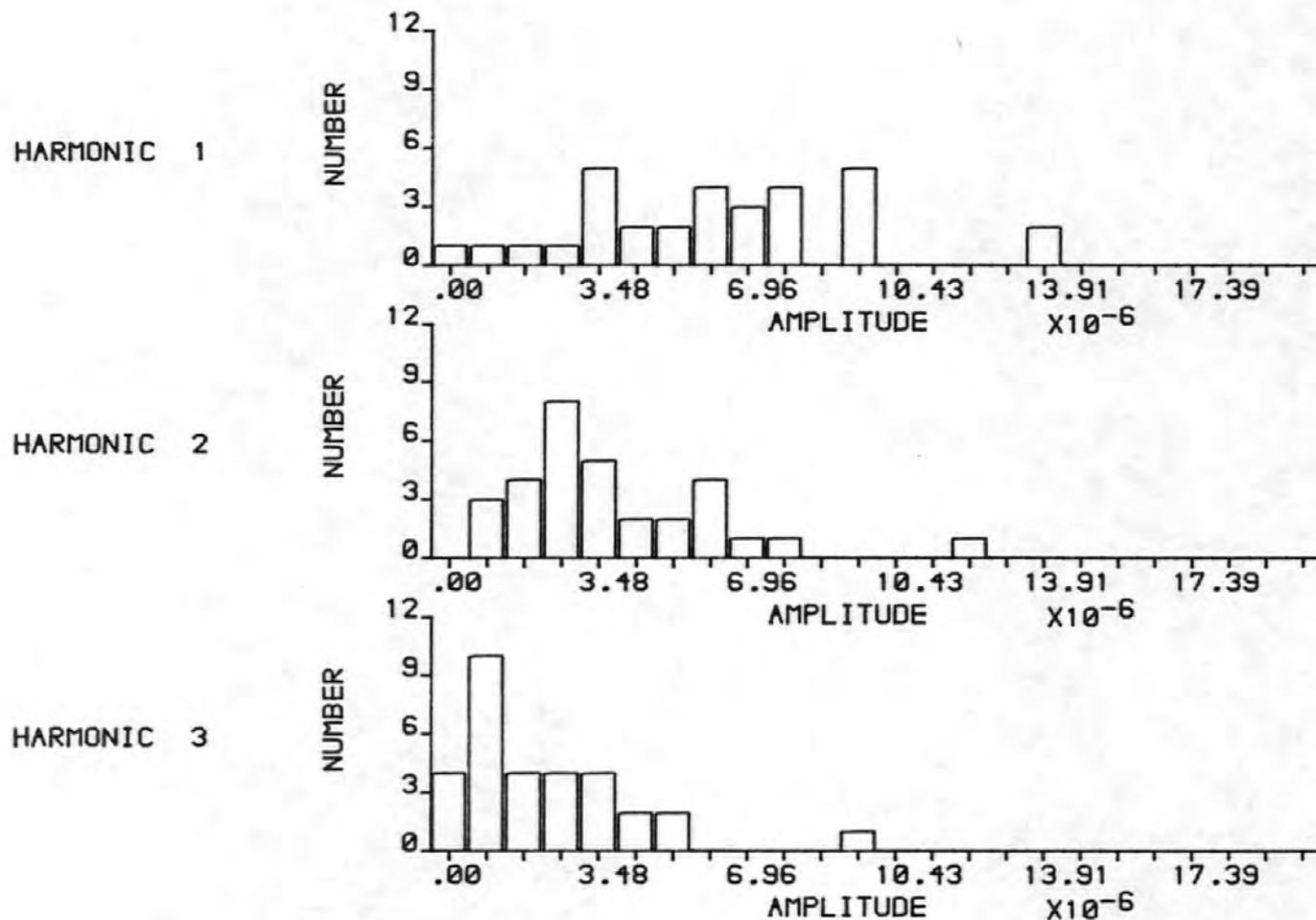


Figure 5-6 Typical amplitude histograms of a one second ISI CNV, Harmonics 1 to 3.

the spectra of the background EEG was made with the spectrum of the averaged CNV. The latter spectrum would be relatively free of background EEG due to the averaging process.

The spectrum of the averaged CNV's displayed the presence of signal power at frequencies of up to 12 Hz. Over this frequency range the signal levels were about 0.5 - 6.5 $\mu\text{V Hz.}^{-\frac{1}{2}}$. By comparison the background EEG contained signal levels of 1.5 - 13.8 $\mu\text{V Hz.}^{-\frac{1}{2}}$ over the same range. Thus the frequency components of the background EEG were indeed compatible to or greater in amplitude than those of the CNV, and were therefore capable of masking it.

5.3.2.3 Pre- and Post- Stimulus Mean Amplitude Differences

Test

These tests were performed on the mean of the amplitude differences of the pre- and post- stimulus phasors as described in section 2.2.3.2. Because of the limited amount of pre-stimulus information available these tests were not carried out on the four second ISI CNV's. The results of the tests are shown in Table 5-6. The column giving the significance level of any detected difference shows that additivity was only detected in 11 of the 30 results. Furthermore, only 2 of these 11 were significant at the 5% or higher level. Surprisingly however, 13 of the 30 results showed decreased amplitudes in the post-stimulus case (denoted by asterisks in the table) although only 2 of these results were statistically significant at the 5% level.

Table 5-6

Pre- and Post-stimulus Mean Amplitude Differences Tests

For Normal Subjects

Subject Number	Mean of Differences $\times 10^{-5}$	Stand.Dev. of Differences $\times 10^{-5}$	Harmonic Number	T	v	Significance %
1	0.1586	0.4696	1	1.88	30	10
	0.0144	0.3228	2	0.248	30	N-S
	-0.0381	0.2387	3	-0.888*	30	N-S
	0.0709	0.2889	4	1.37	30	20
	0.0974	0.4162	5	1.30	30	N-S
	-0.1399	0.2229	6	-3.49*	30	0.2
2	0.1561	0.4827	1	1.83	31	10
	0.0651	0.2817	2	1.31	31	20
	0.0449	0.1978	3	1.28	31	N-S
	0.0110	0.1883	4	0.33	31	N-S
	-0.0079	0.1341	5	-0.33*	31	N-S
	-0.0261	0.1218	6	-1.21*	31	N-S
3	-0.0974	0.4229	1	-1.00*	18	N-S
	0.0362	0.4806	2	0.33	18	N-S
	0.0884	0.2425	3	1.59	18	20
	-0.0780	0.2308	4	-1.47*	18	20
	-0.0451	0.1543	5	-1.27*	18	N-S
	0.0292	0.1971	6	0.65	18	N-S
4	0.0880	0.3578	1	1.39	31	20
	0.0797	0.2526	2	1.79	31	10
	0.0435	0.1776	3	1.39	31	20
	-0.0532	0.1367	4	-2.20*	31	5
	-0.0355	0.2242	5	-0.896*	31	N-S
	-0.0412	0.1477	6	-1.58*	31	20
5	0.2115	0.8769	1	1.36	31	20
	0.2965	0.6157	2	2.72	31	2
	0.2428	0.4536	3	3.03	31	1
	-0.1074	0.3109	4	-1.95*	31	10
	-0.0951	0.3153	5	-1.71*	31	10
	-0.0671	0.2202	6	-1.72*	31	10

* denotes that the post-stimulus amplitudes were smaller than the pre-stimulus amplitudes.

5.3.2.4 Nearest and Furthest Mean Amplitude Test

These tests were performed on the amplitudes of post-stimulus phasors for the one and four second ISI CNV's described in section 2.2.3.3. The results are shown in Tables 5-7a and 5-7b for the one and four second ISI's respectively. For the one second ISI the significance column shows that additivity was detected in 7 cases out of 30, but of these only 4 were significant at the 5% level. In one case (i.e. subject 4, harmonic 3), the furthest phasors had larger amplitudes than the nearest. For the four second ISI CNV's additivity was detected in 4 cases and of these only 2 were significant at the 5% level. The furthest phasors had larger amplitudes than the nearest in 2 cases.

Table 5-7a

Nearest and Furthest Mean Amplitude Test

For 1 second ISI CNV's of Normal Subjects

Subject Number	Nearest		Furthest		Harmonic	T	v	Significance %
	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$				
1	1.04	0.353	0.761	0.523	1	1.76	28	5
	0.650	0.361	0.514	0.269	2	1.19	28	N-S
	0.380	0.251	0.382	0.328	3	-0.017	30	N-S
	0.485	0.311	0.560	0.380	4	-0.609	31	N-S
	0.807	0.431	0.831	0.535	5	-0.137	30	N-S
	0.418	0.168	0.354	0.234	6	0.882	29	N-S
2	1.24	0.457	0.769	0.363	1	3.23	30	0.5
	0.643	0.368	0.456	0.301	2	1.58	31	10
	0.439	0.237	0.415	0.192	3	0.322	31	N-S
	0.308	0.202	0.309	0.217	4	-0.017	32	N-S
	0.293	0.134	0.211	0.114	5	1.86	31	5
	0.276	0.107	0.216	0.107	6	1.58	32	10
3	0.805	0.434	0.781	0.337	1	0.131	17	N-S
	0.660	0.288	0.675	0.426	2	-0.09	18	N-S
	0.575	0.258	0.554	0.318	3	0.165	19	N-S
	0.362	0.200	0.241	0.149	4	1.48	16	10
	0.274	0.156	0.303	0.162	5	-0.396	19	N-S
	0.329	0.219	0.253	0.257	6	0.690	19	N-S
4	0.813	0.388	0.641	0.383	1	1.26	32	N-S
	0.525	0.337	0.419	0.254	2	1.00	30	N-S
	0.332	0.168	0.432	0.171	3	-1.65*	32	10
	0.243	0.162	0.259	0.101	4	-0.337	27	N-S
	0.298	0.110	0.359	0.258	5	-0.868	21	N-S
	0.202	0.142	0.239	0.125	6	-0.777	32	N-S
5	2.16	0.813	1.52	0.800	1	2.24	32	2.5
	1.17	0.463	1.47	0.817	2	-1.26	25	N-S
	0.985	0.541	0.973	0.607	3	0.061	32	N-S
	0.603	0.247	0.506	0.274	4	1.05	32	N-S
	0.548	0.276	0.456	0.305	5	0.895	32	N-S
	0.340	0.216	0.335	0.164	6	0.08	30	N-S

Table 5-7b

Nearest and Furthest Mean Amplitude Test

For 4 second ISI CNV's of Normal Subjects

Subject	Nearest		Furthest		Harmonic	T	v	Significance
	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$				
1	0.740	0.304	0.491	0.267	1	2.47	32	1
	0.370	0.219	0.448	0.243	2	-0.958	32	N-S
	0.451	0.246	0.324	0.120	3	1.85	23	5
	0.357	0.125	0.308	0.163	4	0.954	30	N-S
	0.430	0.176	0.460	0.221	5	-0.429	30	N-S
	0.342	0.252	0.432	0.190	6	-1.03	30	N-S
2	1.119	0.554	0.912	1.13	1	0.663	23	N-S
	0.943	1.05	0.744	0.349	2	0.721	19	N-S
	0.347	0.230	0.624	0.583	3	-1.77*	20	5
	0.721	0.744	0.462	0.260	4	1.32	19	N-S
	0.531	0.514	0.327	0.152	5	1.52	18	10
	0.383	0.221	0.354	0.213	6	0.369	32	N-S
3	0.806	0.382	0.879	0.373	1	-0.547	32	N-S
	0.715	0.324	0.741	0.372	2	-0.214	31	N-S
	0.510	0.323	0.606	0.280	3	-0.903	31	N-S
	0.475	0.237	0.401	0.176	4	1.00	29	N-S
	0.437	0.177	0.341	0.152	5	1.65	31	10
	0.370	0.152	0.333	0.171	6	0.648	32	N-S
4	0.554	0.226	0.503	0.259	1	0.615	31	N-S
	0.428	0.235	0.449	0.268	2	-0.227	31	N-S
	0.419	0.177	0.378	0.141	3	0.738	30	N-S
	0.330	0.176	0.279	0.160	4	0.845	32	N-S
	0.331	0.192	0.435	0.219	5	-1.43*	31	10
	0.312	0.184	0.320	0.124	6	-0.147	28	N-S
5	1.06	0.951	1.09	0.806	1	-0.114	31	N-S
	0.864	0.499	0.930	0.488	2	-0.382	32	N-S
	0.789	0.444	0.684	0.479	3	0.642	32	N-S
	0.887	0.353	0.866	0.645	4	0.110	24	N-S
	0.728	0.418	0.663	0.330	5	0.485	30	N-S
	0.577	0.377	0.602	0.452	6	-0.166	31	N-S

* denotes that the furthest phasors were larger than the nearest phasors.

5.3.2.5. Discussion of Results of Energy Tests

The results of the Pre- and Post- Stimulus Mean Amplitude Differences tests and of the Nearest and Furthest Mean Amplitude tests, which indicated additivity for only a minority of the harmonic components, were in contrast to those of the pre-stimulus and CNV broadband energy tests which offered strong evidence of additivity. The differences may be attributed to either the effects of the background EEG or to the suitability of the tests. For example, the Nearest and Furthest Mean Amplitude test could be prone to error and while positive results supported the additivity model, negative ones did not necessarily disprove it. It was assumed that for each trial the noise and signal amplitudes were the same and that the response was the same in both amplitude and phase. These assumptions may not have been sufficiently true and if, for example, the response had a random phase there would be no detectable preferred direction even though added energy was present. Also in the additive model a phasor could have been reversed from a large phasor in one direction to a small phasor in the preferred direction. This would have been contrary to the general assumption that phasors would be larger in the preferred direction. Thus the test was not infallible and there may have been added energy in some of the other cases. The Pre- and Post-Stimulus Mean Amplitude Differences test could have been affected by variations in the background EEG and thus made unreliable. Alternatively the extra energy detected by the broadband tests could have been at frequencies other than those examined.

5.3.3 Tests for Phase Ordering

The following tests were performed to establish whether any phase ordering could be detected in the CNV response.

5.3.3.1 Phase Histograms

It was found that the phase histograms of the first harmonic of the CNV's were visually similar for all subjects and for both one second and four second ISI's as predicted in Section 2.2.1 and Appendix 8.1. Typical phase histograms for both ISI's are shown in Figures 5-7a and 5-7b. The patterns observed in these histograms are evidence for phase ordering. Similar histograms, for the higher harmonic frequency components, did not exhibit any noticeable phase patterns or ordering. Because of the limited number of observations made and the relatively large number of phase intervals necessary to observe patterns and ordering, the phase histograms were not considered to be very reliable indicators of these effects. Furthermore, since it was necessary to judge visually the presence (or absence) of phase patterns only limited credence could be given to these histograms.

5.3.3.2 Rayleigh Test of Circular Variance

The results of this test are given in tables 5-8a and 5-8b for the one second and four second ISI CNV's respectively. For the one second ISI the test gave a significant result in 10 out of the 30 cases studied while of these only 6 were significant at the 5% level. In the case of the

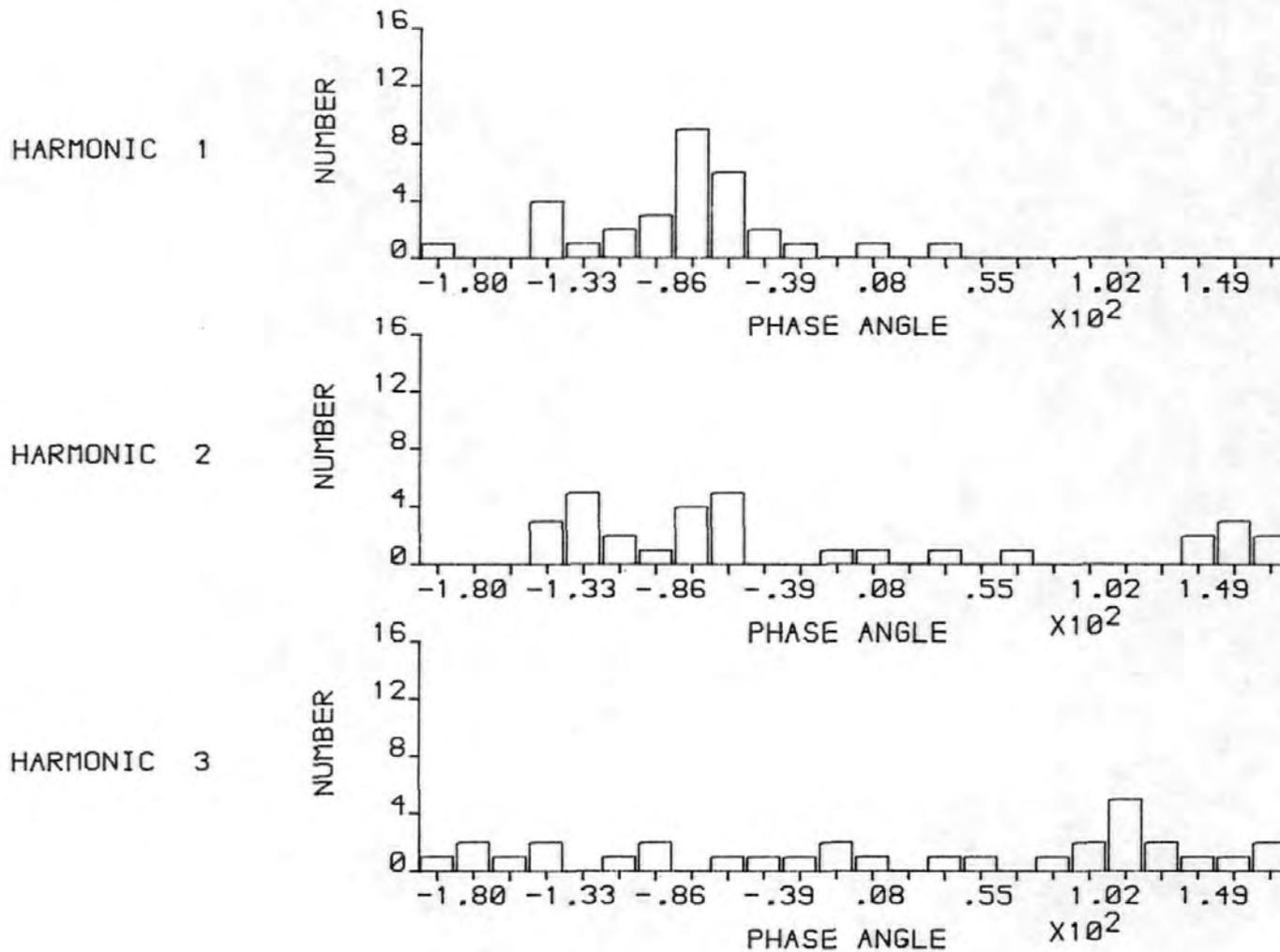


Figure 5-7a A typical one second ISI phase histogram of a normal subject

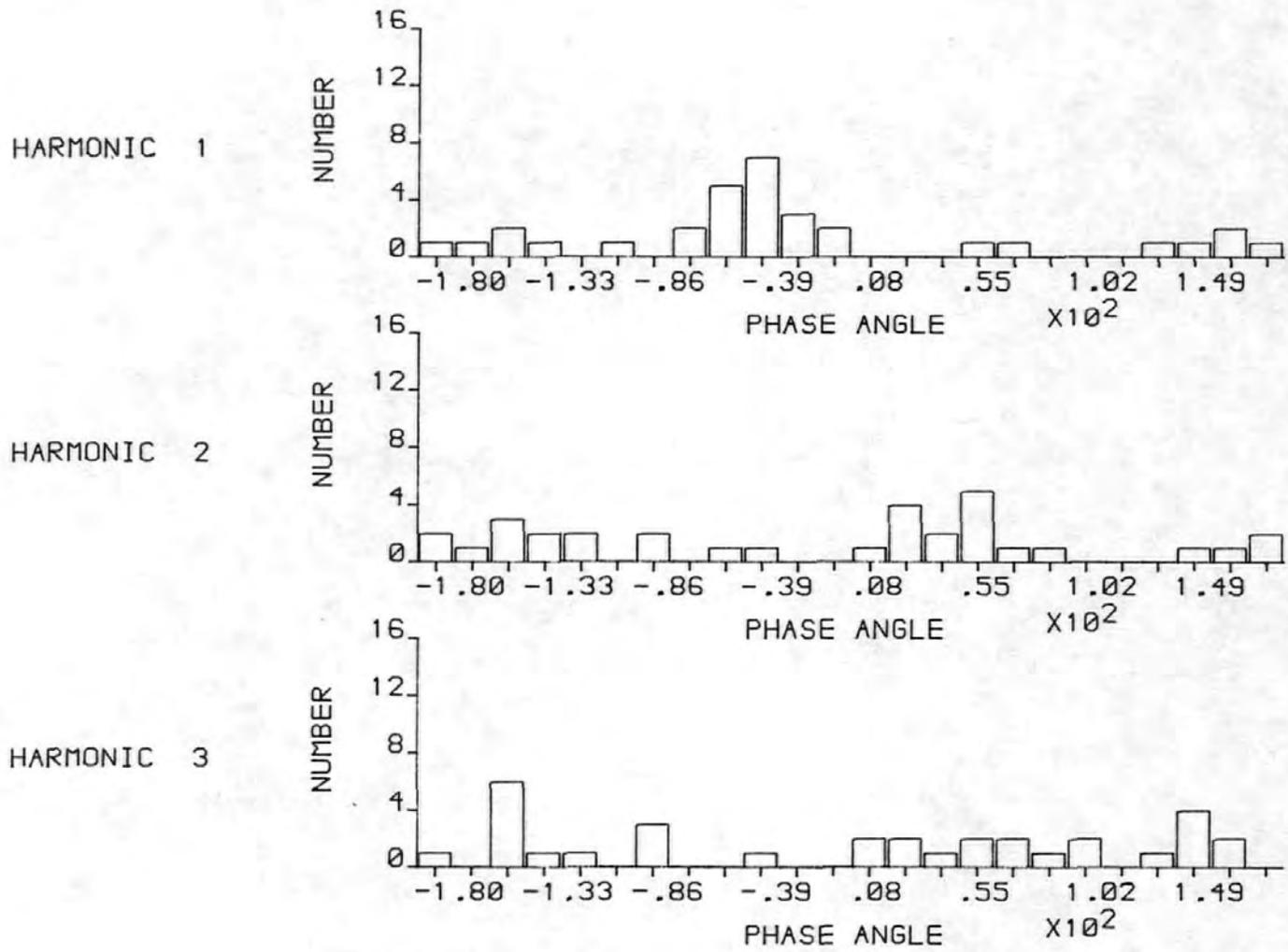


Figure 5-7b A typical four second ISI phase histogram of a normal subject

Table 5-8a

Circular Variance for 1 second ISI CNV's of Normal Subjects

Subject	Harmonic	Number of Trials	So	Significance %
1	1	31	0.202	0.1
	2	31	0.554	1.0
	3	31	0.787	N-S
	4	31	0.916	N-S
	5	31	0.713	10.0
	6	31	0.806	N-S
2	1	32	0.683	5.0
	2	32	0.893	N-S
	3	32	0.940	N-S
	4	32	0.804	N-S
	5	32	0.848	N-S
	6	32	0.796	N-S
3	1	19	0.642	10
	2	19	0.699	N-S
	3	19	0.781	N-S
	4	19	0.425	1.0
	5	19	0.694	N-S
	6	19	0.855	N-S
4	1	32	0.747	N-S
	2	32	0.695	10.0
	3	32	0.841	N-S
	4	32	0.700	10.0
	5	32	0.902	N-S
	6	32	0.780	N-S
5	1	32	0.354	0.1
	2	32	0.990	N-S
	3	32	0.661	5.0
	4	32	0.792	N-S
	5	32	0.877	N-S
	6	32	0.771	N-S

Table 5-8b

Circular Variance for 4 second ISI CNV's of Normal Subjects

Subject	Harmonic	Number of Trials	So	Significance %
1	1	32	0.599	1.0
	2	32	0.973	N-S
	3	32	0.807	N-S
	4	32	0.935	N-S
	5	32	0.926	N-S
	6	32	0.945	N-S
2	1	32	0.867	N-S
	2	32	0.882	N-S
	3	32	0.779	N-S
	4	32	0.699	10.0
	5	32	0.799	N-S
	6	32	0.916	N-S
3	1	32	0.822	N-S
	2	32	0.877	N-S
	3	32	0.993	N-S
	4	32	0.897	N-S
	5	32	0.914	N-S
	6	32	0.896	N-S
4	1	32	0.957	N-S
	2	32	0.724	N-S
	3	32	0.856	N-S
	4	32	0.790	N-S
	5	32	0.861	N-S
	6	32	0.883	N-S
5	1	32	0.808	N-S
	2	32	0.793	N-S
	3	32	0.835	N-S
	4	32	0.896	N-S
	5	32	0.949	N-S
	6	32	0.811	N-S

four second ISI CNV's only 2 out of the 30 cases were significant and of them only 1 was significant at the 5% level. Thus phase ordering appeared to be less pronounced in the case of the four second ISI CNV's. This difference reflected the more complex composition of the longer ISI CNV's.

5.3.3.3 Modified Rayleigh Test of Circular Variance

The results of these tests are shown in tables 5-9a and 5-9b for the two ISI's used. For the one second ISI phase ordering was significant in 10 out of the 30 cases, being significant at the 5% level in 7 cases. For the four second ISI the figures were 4 and 3 respectively. Thus the phase again appeared to be less pronounced in the four second ISI CNV's.

5.3.3.4 Hodges-Ajne Test

Tables 5-10a and 5-10b show the results of these tests for the one and four second ISI CNV's respectively. The one second ISI CNV's exhibited ordering in 5 cases out of 30 and of these 4 were significant at the 5% level. There were only 2 cases of ordering in the results for the four second ISI CNV's and only 1 was significant at the 5% level.

Table 5-9a

The Modified Circular Variance Tests for the 1 second
ISI CNV's of Normal Subjects

Subject	Harmonic	Number of Trials	Uo	Significance %
1	1	31	0.166	0.1
	2	31	0.453	0.1
	3	31	0.739	N-S
	4	31	0.961	N-S
	5	31	0.721	N-S
	6	31	0.672	10.0
2	1	32	0.530	1.0
	2	32	0.795	N-S
	3	32	0.807	N-S
	4	32	0.802	N-S
	5	32	0.727	N-S
	6	32	0.715	N-S
3	1	19	0.526	5.0
	2	19	0.806	N-S
	3	19	0.828	N-S
	4	19	0.307	0.1
	5	19	0.726	N-S
	6	19	0.818	N-S
4	1	32	0.672	10.0
	2	32	0.607	5.0
	3	32	0.915	N-S
	4	32	0.743	N-S
	5	32	0.863	N-S
	6	32	0.731	N-S
5	1	32	0.216	0.1
	2	32	0.949	N-S
	3	32	0.678	10.0
	4	32	0.777	N-S
	5	32	0.892	N-S
	6	32	0.772	N-S

Table 5-9b

The Modified Circular Variance Tests for the 4 second
ISI CNV's of Normal Subjects

Subject	Harmonic	Number of Trials	Uo	Significance %
1	1	32	0.506	1.0
	2	32	0.929	N-S
	3	32	0.717	N-S
	4	32	0.848	N-S
	5	32	0.873	N-S
	6	32	0.832	N-S
2	1	32	0.743	N-S
	2	32	0.970	N-S
	3	32	0.857	N-S
	4	32	0.651	5.0
	5	32	0.623	5.0
	6	32	0.848	N-S
3	1	32	0.959	N-S
	2	32	0.912	N-S
	3	32	0.953	N-S
	4	32	0.788	N-S
	5	32	0.827	N-S
	6	32	0.790	N-S
4	1	32	0.947	N-S
	2	32	0.679	10.0
	3	32	0.845	N-S
	4	32	0.777	N-S
	5	32	0.966	N-S
	6	32	0.921	N-S
5	1	32	0.816	N-S
	2	32	0.770	N-S
	3	32	0.746	N-S
	4	32	0.796	N-S
	5	32	0.962	N-S
	6	32	0.758	N-S

Table 5-10a

The Hodges-Ajne test for the 1 second ISI CNV's of normals

Subject	Harmonic	Number of Trials	M	Significance %
1	1	31	2	0.001
	2	31	4	0.067
	3	31	10	N-S
	4	31	13	N-S
	5	31	9	N-S
	6	31	11	N-S
2	1	32	10	N-S
	2	32	12	N-S
	3	32	14	N-S
	4	32	10	N-S
	5	32	12	N-S
	6	32	11	N-S
3	1	19	4	16.0
	2	19	5	N-S
	3	19	5	N-S
	4	19	2	0.9
	5	19	6	N-S
	6	19	6	N-S
4	1	32	10	N-S
	2	32	9	N-S
	3	32	12	N-S
	4	32	9	N-S
	5	32	12	N-S
	6	32	12	N-S
5	1	32	3	0.006
	2	32	13	N-S
	3	32	9	N-S
	4	32	10	N-S
	5	32	12	N-S
	6	32	10	N-S

Table 5-10b

The Hodges-Ajne test for the 4 second ISI CNV's of normals

Subject	Harmonic	Number of Trials	M	Significance %
1	1	32	6	0.8
	2	32	13	N-S
	3	32	10	N-S
	4	32	12	N-S
	5	32	13	N-S
	6	32	14	N-S
2	1	32	12	N-S
	2	32	12	N-S
	3	32	11	N-S
	4	32	8	7.8
	5	32	11	N-S
	6	32	13	N-S
3	1	32	11	N-S
	2	32	12	N-S
	3	32	14	N-S
	4	32	12	N-S
	5	32	13	N-S
	6	32	13	N-S
4	1	32	13	N-S
	2	32	10	N-S
	3	32	11	N-S
	4	32	12	N-S
	5	32	12	N-S
	6	32	12	N-S
5	1	32	11	N-S
	2	32	10	N-S
	3	32	11	N-S
	4	32	12	N-S
	5	32	13	N-S
	6	32	12	N-S

5.3.4. Discussion of Energy and Phase Results

Table 5-11 summarises the additivity and phase ordering results which were found to be significant at the 5% level.

That part of the table concerned with phase ordering was inspected in order to compare the results of the different tests. Asterisks were used to indicate similar results which were given by either two or three of the tests. Scanning of the rows showed that the results of the phase ordering tests were mainly consistent and therefore probably reliable. The Modified Rayleigh test tended to produce more results which were significant than either of the other two tests and this was because it took into account both amplitude and phase information. For the one second ISI CNV's phase ordering was detected in 8 cases out of 30 at the 5% level. For the four second ISI CNV's only 3 of the 30 were significant at the 5% level.

It was stated above (Section 5.3.3.1) that the phase histograms of the first harmonic for both the one and four second ISI CNV's had similar patterns which were evidence of phase ordering, but that the effect was less obvious for the higher harmonics. Inspection of the results of the statistical tests for phase ordering partly confirms those findings. Table 5-12 shows the mean directions of the phasors and the levels of significance for some of the results. It can be seen that for the one second ISI CNV's the mean direction of the phasors falls in the theoretically predicted

Table 5-11

Summary of Additivity and Phase Ordering Results Significant
at the 5% Level for Normal Subjects

ISI	SUBJECT NUMBER	ADDITIVITY		PHASE ORDERING		
		NEAREST & FURTHEST MEAN AMPLITUDE	PRE & POST MEAN AMPLITUDE DIFFERENCES	RAYLEIGH CIRCULAR VARIANCE	MODIFIED RAYLEIGH CIRCULAR VARIANCE	HODGES-AJNE
1	1	H1		H1*, H2*	H1*, H2*	H1*, H2*
	2	H1, H5+		H1*	H1*	
	3			H4*	H1, H4*	H4*
	4				H2	
	5	H1	H2+, H3	H1*, H3	H1*	H1*
4	1	H1, H3+	X	H1*	H1*	H1*
	2				H4, H5	

H_n indicates that the nth harmonic had a significant result.

+ indicates additivity with no accompanying ordering.

* indicates agreement between two or more tests.

Table 5-12

Mean Directions and Significance for Selected Cases

ISI	HARMONIC NUMBER	SUBJECT NUMBER	NUMBER OF RESPONSES	MEAN DIRECTION	SIGNIFICANCE OF PHASE ORDERING†
1	1	1	31	-70	S
		2	32	-75	S
		3	19	106	S
		4	32	-82	N-S
		5	32	-81	S
	2	1	31	-102	S
		2	32	114	N-S
		3	19	116	N-S
		4	32	-76	S
		5	32	51	N-S
	3	1	31	140	N-S
		2	32	-24	N-S
		3	19	-125	N-S
		4	32	-36	N-S
		5	32	57	S
4	1	1	32	-57	S
		2	32	108	N-S
		3	32	138	N-S
		4	32	-129	N-S
		5	32	-159	N-S
	2	1	32	-5	N-S
		2	32	-125	N-S
		3	32	7	N-S
		4	32	10	N-S
		5	32	131	N-S
	3	1	32	155	N-S
		2	32	104	N-S
		3	32	95	N-S
		4	32	-150	N-S
		5	32	36	N-S

† S indicates phase ordering detected by any of the three tests significant at the 5% level.

N-S indicates no significant phase ordering.

range of 0° to -90° in 4 out of 5 cases and for 3 of these there was also significant phase ordering. Hence a distinctive pattern would indeed be anticipated in the phase histogram. The mean direction is less likely to be in the predicted range and the degree of ordering is also seen to be decreased as the harmonic number increases. This is in agreement with the reduced observable phase ordering in the phase histograms of the higher harmonics. The results for the four second ISI CNV's showed little phase ordering and many of the angles were outside the predicted range. This did not agree with the conclusion reached by visual inspection of the phase histograms. This discrepancy was probably due to the unreliability of the phase histograms for the reasons previously mentioned.

Table 5-11 was also inspected to check whether additivity was always accompanied by phase ordering as the additivity model would suggest. It was noted that in three cases additivity was detected in the absence of detectable phase ordering. This may have been due to the limitations of the tests or it may indicate that the CNV response cannot be described by the additivity model.

Overall it appears that the CNV responses of normal subjects may contain a certain amount of added random energy and are therefore probably less deterministic in nature than, for example, an auditory evoked response.

5.4 The CNV's of Abnormal Subjects

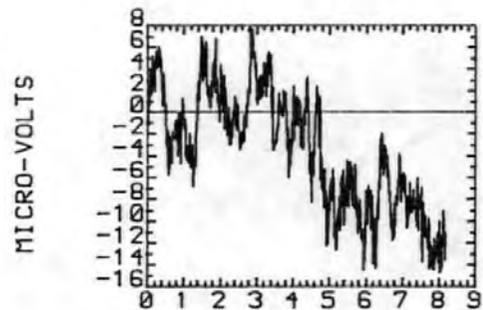
CNV responses were also recorded from a total of five patients for whom Huntington's chorea had been diagnosed. Once again two interstimulus intervals were used (one and four seconds) and prior to further processing the individual responses were corrected for the effects of eye movement as described in Section 2.5

5.4.1 The Averaged CNV's

After processing to remove eye movement artefact the CNV's were averaged, filtered with the low pass digital filter described in Section 3.6 and plotted. The results of this procedure are shown in Figures 5-8 a-e and 5-9 a-e. A characteristic feature of some of these CNV's was the slow return to the baseline subsequent to the S2 stimulus. This feature, known as the Post Imperative Negative Variation, has been described elsewhere [9] although not in connection with Huntington's chorea. Some of the averaged CNV's were generally similar to those obtained from normal subjects (e.g. Figure 5-8c) whilst others (e.g. Figure 5-8b) showed very little evidence of a CNV response at all. The differences in the averaged waveforms of normals and patients were exploited in distinguishing between the two categories as described in Section 5.5

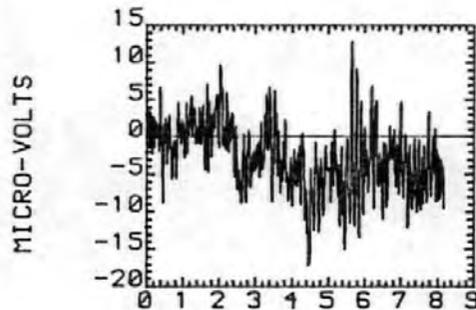
5.4.2. Energy Tests

These tests were performed to establish whether any additive component could be detected in the CNV response of the H.C. patient group.



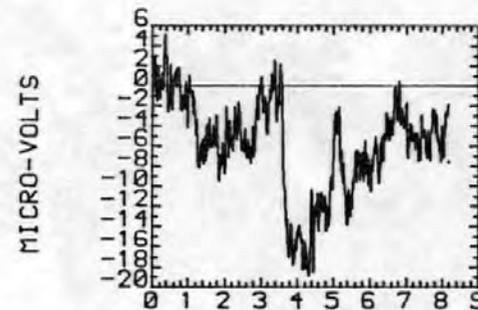
TIME SECONDS

(a)



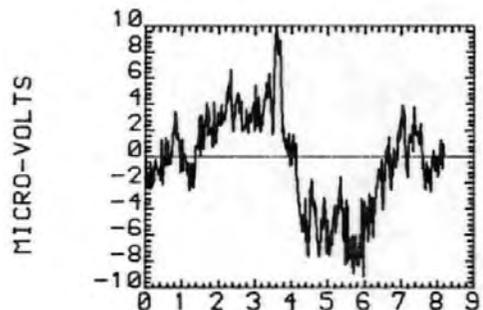
TIME SECONDS

(b)



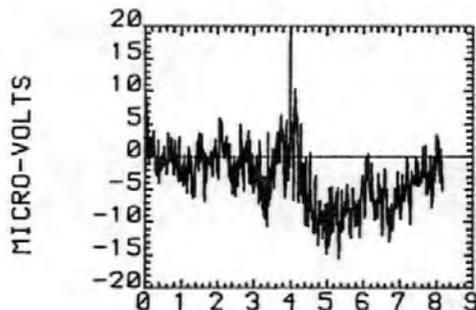
TIME SECONDS

(c)



TIME SECONDS

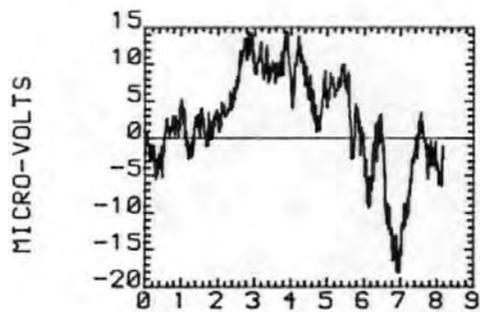
(d)



TIME SECONDS

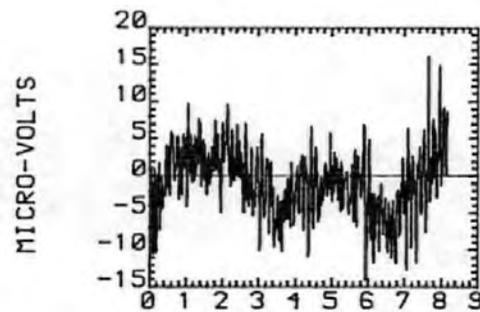
(e)

Figure 5-8
The averaged one second ISI CNV's
of abnormal subjects.



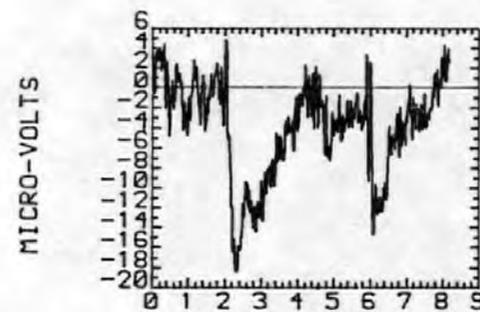
TIME SECONDS

(a)



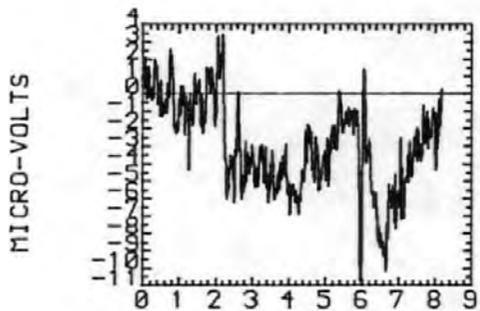
TIME SECONDS

(b)



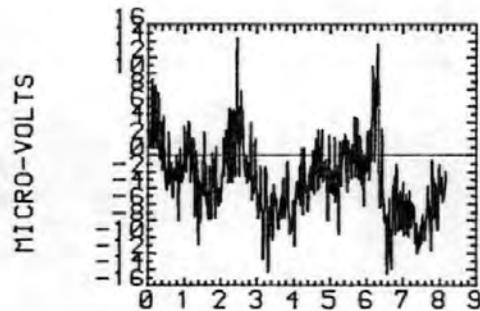
TIME SECONDS

(c)



TIME SECONDS

(d)



TIME SECONDS

(e)

Figure 5-9
The averaged four second ISI CNV's
of abnormal subjects.

5.4.2.1 Broadband Energy Tests

The same processing procedure was used in these tests as that described in Section 5.3.2.1. The results of the tests are shown in Tables 5-13a and 5-13b. In the case of the one second ISI CNV's there was evidence of added energy at the 5% level of significance in 3 cases out of the 5 subjects investigated. There was considerably more added energy in the case of subject 5. For the four second ISI CNV's 4 of the subjects showed evidence of added energy at the 5% significance level. Thus for both ISI's the majority of the subjects showed increased energy in the post stimulus realizations. However, as indicated by the right most column in Tables 5-13a and 5-13b, there were a number of individual responses where the pre-stimulus energy was greater than the post-stimulus energy.

5.4.2.2 Amplitude Histograms

Some amplitude histograms were produced as it was thought that they might have shown some differences to those of normal subjects. However, they did not exhibit any characteristic feature which could be of diagnostic value.

5.4.2.3 Pre- and Post-stimulus Mean Amplitude Differences Test

The results of this test on the one second ISI CNV's are shown in Table 5-14. All but one of the results showed evidence of added energy in at least one harmonic at the 5% level. It was not possible to discover any pattern in the occurrence of significant results. Once more it was noted

Table 5-13a

Broadband Energy Tests of Abnormal Subjects one second ISI CNV's

Subject	Number of Responses	T Statistic	Significance Level %	Number of Responses * Post < Pre
1	32	1.89	10	10
2	32	1.81	10	10
3	32	3.75	0.1	8
4	32	3.47	0.2	7
5	32	4.82	<<0.1	4

Table 5-13b

Broadband Energy Tests of Abnormal Subjects four second ISI CNV's

Subject	Number of Responses	T Statistic	Significance Level %	Number of Responses * Post < Pre
1	30	3.73	0.1	4
2	32	2.56	2.0	9
3	32	1.64	20	7
4	32	5.16	<<0.1	6
5	32	3.16	1.0	8

* The number of individual responses where the post-stimulus energy was LESS than the pre-stimulus energy.

Table 5-14
Pre- and Post-Stimulus Mean Amplitude Differences Test
for Abnormal Subjects
1 second ISI

Subject Number	Mean of Differences $\times 10^{-5}$	St.Dev. of Differences $\times 10^{-5}$	Harmonic Number	T	v	Significance %
1	0.178	1.02	1	0.990	31	N-S
	0.179	0.403	2	2.52	31	2.0
	-0.003	0.318	3	-0.059*	31	N-S
	-0.053	0.260	4	-1.15 *	31	N-S
	0.019	0.252	5	0.434	31	N-S
	0.069	0.209	6	1.86	31	10
2	0.189	0.576	1	1.94	31	10
	0.065	0.509	2	0.727	31	N-S
	-0.064	0.646	3	-0.557*	31	N-S
	0.379	0.887	4	2.42	31	5.0
	-0.168	0.724	5	-1.31 *	31	20
	-0.010	0.504	6	-0.118*	31	N-S
3	-0.067	0.666	1	-0.568*	31	N-S
	0.029	0.555	2	0.297	31	N-S
	-0.139	0.356	3	-2.21 *	31	5
	0.035	0.336	4	0.585	31	N-S
	-0.089	0.337	5	-1.49 *	31	20
	-0.070	0.252	6	-1.58 *	31	20
4	-0.016	0.502	1	-0.184*	31	N-S
	0.147	0.242	2	3.44	31	0.2
	0.05	0.136	3	2.08	31	5
	-0.017	0.113	4	-0.85 *	31	N-S
	-0.051	0.117	5	-2.46 *	31	2
	-0.021	0.159	6	-0.747*	31	N-S
5	0.202	0.622	1	1.84	31	10
	0.213	0.572	2	2.11	31	5
	0.409	0.664	3	3.49	31	0.2
	0.901	0.731	4	6.98	31	<<0.1
	0.507	0.677	5	4.23	31	<<0.1
	0.246	0.580	6	2.40	31	5

* denotes a reduction in the post stimulus amplitude

that subject 5 showed enhanced evidence of additivity in each of the harmonics. Those cases in which the energy showed a decrease were marked by an asterisk in the table.

5.4.2.4 Nearest and Furthest Mean Amplitude Test

The results of these tests are shown in Tables 5-15a and 5-15b. The single tailed paired t-test produced only one result which was significant for the one second ISI CNV's and this was evidence that the nearest phasors were smaller than the furthest. Thus this test provided no evidence for increased energy. In the case of the four second ISI CNV's there were 5 significant cases and 4 of these were significant at the 5% level.

5.4.2.5 Discussion of Results of Energy Tests

Comparison of the results in Tables 5-13a, 5-14 and 5-15a for the one second ISI CNV's did not reveal any noticeable correlations. However, it was noted that the Pre- and Post-Stimulus Mean Amplitude Differences tests offered more evidence of additivity than the Nearest and Furthest Mean Amplitude test. The results obtained from it were therefore more in agreement with the detection of added energy by the Broadband Energy test. However, there were discrepancies between these tests although they were based on exactly the same pre- and post-stimulus data. For example according to Table 5-14 there was no evidence of additivity for subject 3 but according to the Broadband Energy test additivity was detected at a level of significance of 0.1%. It was thought that this discrepancy may have occurred because the added energy of the response was located in higher harmonics than those considered.

Table 5-15a

Nearest and Furthest Mean Amplitude Test for Abnormal Subjects

1 second ISI

Subject Number	Nearest		Furthest		Harmonic	T	v	Significance %
	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$				
1	1.24	0.657	1.07	0.977	1	0.599	28	N-S
	0.564	0.496	0.550	0.290	2	0.100	25	N-S
	0.312	0.192	0.267	0.213	3	0.631	32	N-S
	0.191	0.090	0.265	0.133	4	-1.84	28	* 5
	0.204	0.149	0.229	0.231	5	-0.367	27	N-S
	0.233	0.232	0.226	0.094	6	0.110	21	N-S
2	0.874	0.386	0.802	0.398	1	0.519	32	N-S
	0.596	0.540	0.569	0.395	2	0.166	29	N-S
	0.483	0.309	0.581	0.497	3	-0.673	26	N-S
	1.02	0.857	0.869	0.799	4	0.504	32	N-S
	0.423	0.368	0.482	0.473	5	-0.399	30	N-S
	0.439	0.454	0.326	0.219	6	0.894	23	N-S
3	0.667	0.236	0.681	0.393	1	-0.116	26	N-S
	0.476	0.196	0.437	0.270	2	0.466	29	N-S
	0.220	0.137	0.220	0.118	3	0.014	31	N-S
	0.354	0.282	0.309	0.159	4	0.553	25	N-S
	0.307	0.163	0.269	0.134	5	0.713	31	N-S
	0.176	0.126	0.242	0.191	6	-1.15	28	N-S
4	0.498	0.352	0.498	0.319	1	0.00	32	N-S
	0.430	0.200	0.364	0.214	2	0.899	32	N-S
	0.251	0.159	0.224	0.137	3	0.511	31	N-S
	0.129	0.088	0.166	0.088	4	-1.16	32	N-S
	0.133	0.078	0.154	0.062	5	-0.821	30	N-S
	0.143	0.079	0.173	0.099	6	-0.947	31	N-S
5	0.972	0.534	0.908	0.486	1	0.352	32	N-S
	0.803	0.555	0.595	0.415	2	0.620	30	N-S
	0.980	0.682	0.776	0.500	3	0.966	29	N-S
	1.55	0.730	1.25	0.650	4	1.20	32	N-S
	1.00	0.601	1.10	0.628	5	-0.453	32	N-S
	0.625	0.324	0.725	0.746	6	-0.493	21	N-S

* denotes that the furthest phasors were larger than the nearest phasors.

Table 5-15b

Nearest and Furthest Mean Amplitude Test for Abnormal Subjects

a second ISI

Subject Number	Nearest		Furthest		Harmonic	T	v	Significance %
	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$	Mean $\times 10^{-5}$	St.Dev. $\times 10^{-5}$				
1	1.59	0.774	1.00	0.451	1	2.52	24	1
	1.17	0.736	1.49	0.584	2	-1.33	28	* 10
	1.18	0.504	1.07	0.471	3	0.613	30	N-S
	0.857	0.600	0.952	0.508	4	-0.467	29	N-S
	0.814	0.392	0.919	0.382	5	-0.741	30	N-S
	0.783	0.344	0.752	0.422	6	0.225	29	N-S
2	1.51	0.857	1.42	1.25	1	0.233	28	N-S
	1.46	1.44	1.06	0.583	2	1.02	20	N-S
	0.909	0.670	0.727	0.467	3	0.895	28	N-S
	0.567	0.416	0.609	0.326	4	-0.318	30	N-S
	0.541	0.390	0.482	0.327	5	0.467	31	N-S
	0.403	0.181	0.475	0.258	6	-0.908	28	N-S
3	1.30	0.869	0.847	0.317	1	1.96	19	5
	1.10	0.578	0.574	0.252	2	3.33	21	0.5
	0.781	0.541	0.794	0.477	3	-0.069	32	N-S
	0.685	0.529	0.581	0.450	4	0.600	31	N-S
	0.493	0.428	0.560	0.377	5	-0.474	32	N-S
	0.616	0.232	0.555	0.256	6	0.703	32	N-S
4	0.457	0.225	0.667	0.356	1	-1.99	27	* 5
	0.617	0.361	0.484	0.284	2	1.16	30	N-S
	0.399	0.211	0.459	0.274	3	-0.693	30	N-S
	0.381	0.230	0.449	0.151	4	-0.989	27	N-S
	0.288	0.166	0.373	0.220	5	-1.24	30	N-S
	0.409	0.261	0.312	0.168	6	1.24	27	N-S
5	1.05	0.504	0.949	0.416	1	0.632	31	N-S
	1.19	0.557	0.870	0.409	2	1.87	29	5
	0.699	0.273	0.811	0.400	3	-0.922	28	N-S
	0.876	0.399	0.655	0.342	4	1.69	31	10
	0.647	0.270	0.608	0.348	5	0.350	30	N-S
	0.421	0.334	0.742	0.409	6	-2.43	31	* 2.5

* denotes that the furthest phasors were larger than the nearest phasors.

However it was considered unlikely that CNV energy would be found at higher harmonics without also being present in the lower harmonics (i.e. those more near to the fundamental frequency of the response). Hence the added energy may have been associated with a random element of the response or to an artefact. In other cases the test results on the harmonic components may have been subject to errors in either the transformation process or the subsequent testing (e.g. the limitations of the Nearest and Furthest Mean Amplitude test described in Section 2.2.3.3). While much of this remains conjecture a clear conclusion must be that although the responses of abnormal subjects may sometimes contain added energy it is not possible to exploit this in any useful way because the results are essentially random.

Comparison of the results for the four second ISI CNV's given in Tables 5-13b and 5-15b confirmed the lack of correlation between the Broadband Energy test and the Nearest and Furthest Mean Amplitude test. The former of these tests offered far more evidence of additivity, again suggesting that the added energy may have been present mainly in the higher harmonics.

5.4.3 Tests for Phase Ordering

The tests described below were carried out in an attempt to detect any phase ordering present in the CNV responses of the patient group.

5.4.3.1 Phase Histograms

From the one second ISI CNV phase histograms it was noticed that angles of about 10° tended not to occur. A similar minimum in the phase pattern had been observed in the case of normal subjects. By contrast to the histograms of normals, those of the abnormals showed an accumulation of values in the vicinity of $+180^{\circ}$. They also exhibited a smaller peak at negative angles (-80° to 180°). Thus it appeared that there might be a method, based on phase angles, to at least partially differentiate between patients and normals. A typical one second ISI CNV phase histogram for an abnormal is shown in Figure 5-10a.

The four second ISI CNV phase histograms of the abnormals showed a larger peak in the vicinity of 150° than either the one second ISI CNV phase histograms of abnormals or normals. A typical phase histogram is shown in Figure 5-10b.

Thus whilst for normal subjects the one and four second ISI CNV phase histograms were similar those for the patients were not.

5.4.3.2 Rayleigh Test of Circular Variance

Tables 5-16a and 5-16b show that there was very little evidence for phase ordering for either ISI CNV. At the 5% significance level there were only 2 (out of 30) positive results for the one second ISI and only 4 (out of 30) positive results for the four second ISI CNV's.

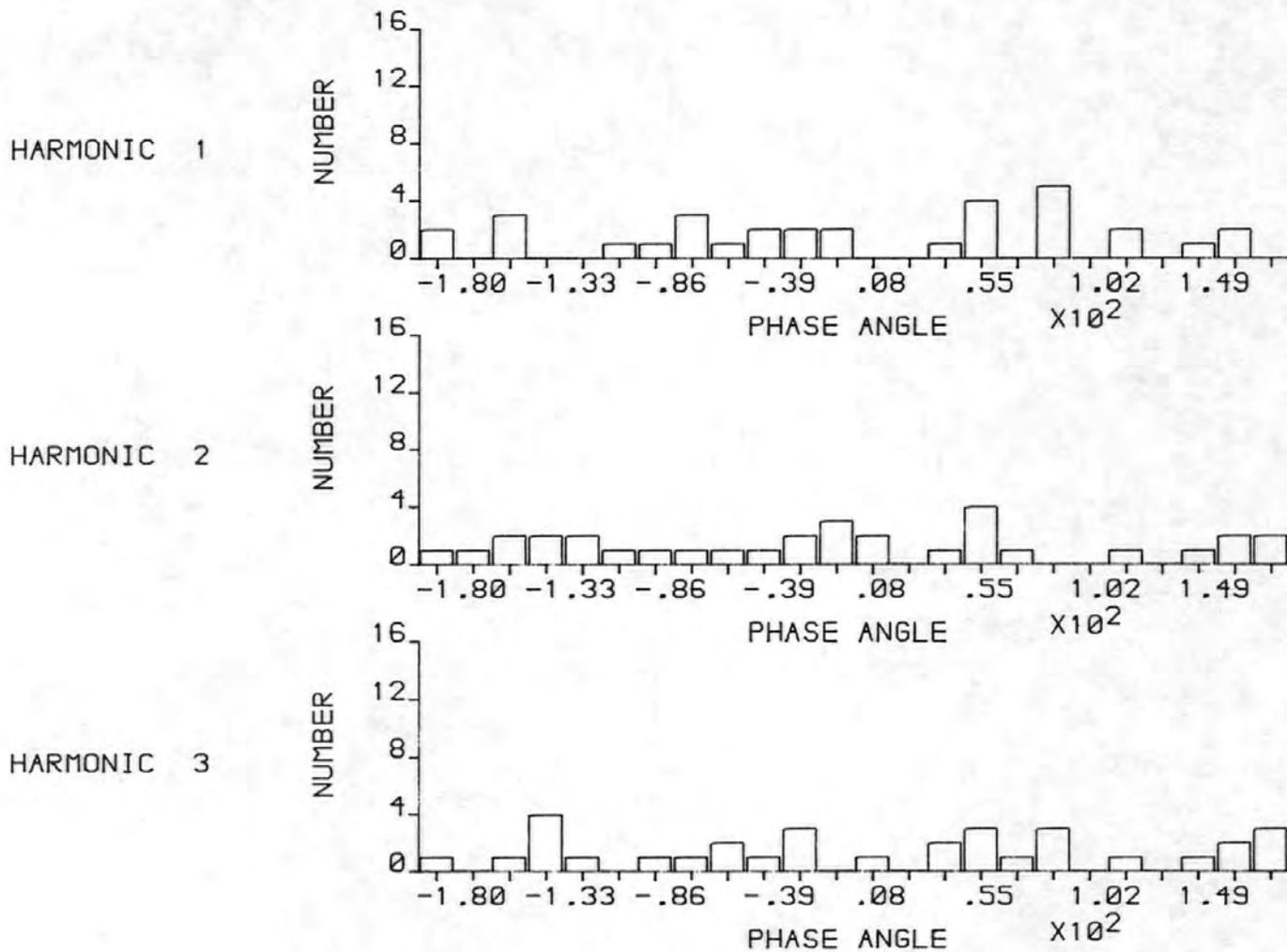


Figure 5-10a A typical one second ISI phase histogram of an abnormal subject

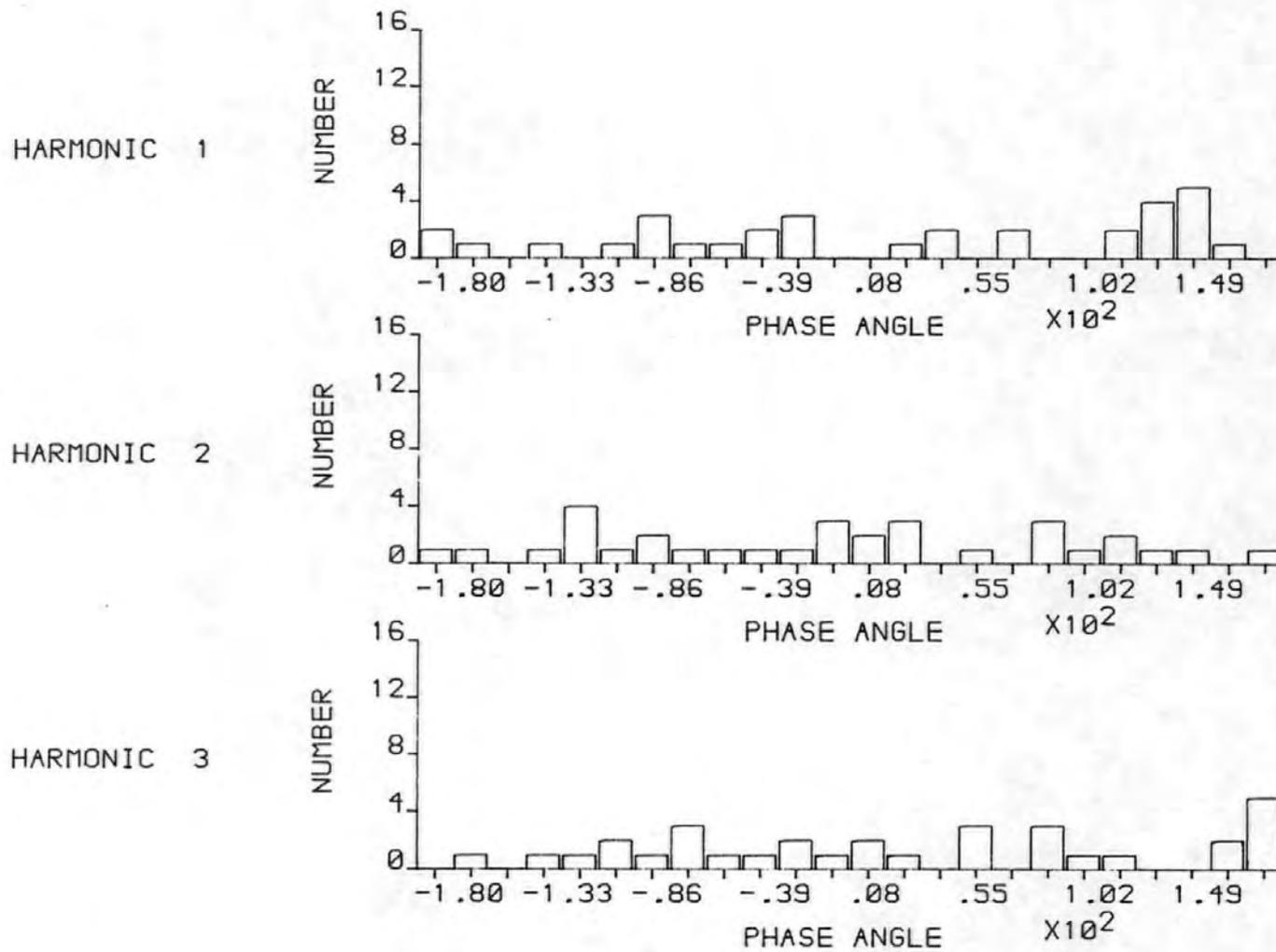


Figure 5-10b A typical four second ISI phase histogram of an abnormal subject

Table 5-16a

Circular Variance for 1 second ISI CNV's
of Abnormal Subjects

Subject	Harmonic	Number of Trials	So	Significance %
1	1	32	0.868	N-S
	2	32	0.873	N-S
	3	32	0.737	N-S
	4	32	0.829	N-S
	5	32	0.624	5
	6	32	0.869	N-S
2	1	32	0.889	N-S
	2	32	0.896	N-S
	3	32	0.991	N-S
	4	32	0.807	N-S
	5	32	0.874	N-S
	6	32	0.808	N-S
3	1	32	0.966	N-S
	2	32	0.744	N-S
	3	32	0.946	N-S
	4	32	0.847	N-S
	5	32	0.870	N-S
	6	32	0.895	N-S
4	1	32	0.805	N-S
	2	32	0.771	N-S
	3	32	0.858	N-S
	4	32	0.788	N-S
	5	32	0.681	5
	6	32	0.995	N-S
5	1	32	0.694	10
	2	32	0.809	N-S
	3	32	0.751	N-S
	4	32	0.800	N-S
	5	32	0.794	N-S
	6	32	0.762	N-S

Table 5-16b
Circular Variance for 4 second ISI CNV's
of Abnormal Subjects

Subject	Harmonic	Number of Trials	So	Significance %
1	1	30	0.844	N-S
	2	30	0.901	N-S
	3	30	0.761	N-S
	4	30	0.934	N-S
	5	30	0.812	N-S
	6	30	0.885	N-S
2	1	32	0.874	N-S
	2	32	0.853	N-S
	3	32	0.954	N-S
	4	32	0.720	10
	5	32	0.925	N-S
	6	32	0.983	N-S
3	1	32	0.403	0.1
	2	32	0.653	5
	3	32	0.859	N-S
	4	32	0.771	N-S
	5	32	0.963	N-S
	6	32	0.857	N-S
4	1	32	0.738	N-S
	2	32	0.811	N-S
	3	32	0.897	N-S
	4	32	0.868	N-S
	5	32	0.784	N-S
	6	32	0.784	N-S
5	1	32	0.657	5
	2	32	0.689	5
	3	32	0.910	N-S
	4	32	0.815	N-S
	5	32	0.863	N-S
	6	32	0.858	N-S

5.4.3.3 Modified Rayleigh Test of Circular Variance

These results (Tables 5-17a and 5-17b) were similar to those of the Rayleigh Test. There were no cases of significance at the 5% level for the one second ISI CNV's and only 3 cases for the four second ISI CNV's.

5.4.3.4 Hodges-Ajne Test

This test indicated only one case of phase ordering at the 5% level for the one second ISI CNV's and only one case for the four second ISI CNV's. The results are shown in Tables 5-18a and 5-18b.

5.4.4 Discussion of Energy and Phase Results

Table 5-19 summarises the results of energy and phase tests applied to the harmonics which were significant at the 5% (or better) level. Harmonic numbers which are asterisked indicate those cases in which phase ordering was detected by at least two of the tests. The Rayleigh test indicated some results to be significant which the other two tests did not substantiate. Examination of the additivity results shows that the Nearest and Furthest Mean Amplitude test did not reveal any cases of significant additivity for the one second ISI CNV's although the Pre- and Post- Stimulus Mean Amplitude Differences test did. It is even more interesting to note that the former test however did indicate cases of significant additivity for the four second ISI CNV's. This contrasting result probably reflected the different composition of the one and four second ISI CNV's. In fact the averaged four second ISI CNV responses were more like the averaged four

Table 5-17a

The Modified Rayleigh Test of Circular Variance for
1 second ISI CNV's of Abnormal Subjects

Subject	Harmonic	Number of Trials	Uo	Significance %
1	1	32	0.764	N-S
	2	32	0.856	N-S
	3	32	0.662	10
	4	32	0.894	N-S
	5	32	0.682	10
	6	32	0.874	N-S
2	1	32	0.773	N-S
	2	32	0.805	N-S
	3	32	0.925	N-S
	4	32	0.799	N-S
	5	32	0.906	N-S
	6	32	0.801	N-S
3	1	32	0.909	N-S
	2	32	0.671	10
	3	32	0.900	N-S
	4	32	0.829	N-S
	5	32	0.879	N-S
	6	32	0.958	N-S
4	1	32	0.806	N-S
	2	32	0.672	10
	3	32	0.771	N-S
	4	32	0.833	N-S
	5	32	0.695	N-S
	6	32	0.947	N-S
5	1	32	0.691	10
	2	32	0.810	N-S
	3	32	0.706	N-S
	4	32	0.726	N-S
	5	32	0.805	N-S
	6	32	0.696	N-S

Table 5-17b

The Modified Rayleigh Test of Circular Variance for
4 second ISI CNV's of Abnormal Subjects

Subject	Harmonic	Number of Trials	Uo	Significance %
1	1	30	0.665	10
	2	30	0.858	N-S
	3	30	0.698	N-S
	4	30	0.867	N-S
	5	30	0.846	N-S
	6	30	0.861	N-S
2	1	32	0.835	N-S
	2	32	0.901	N-S
	3	32	0.920	N-S
	4	32	0.765	N-S
	5	32	0.813	N-S
	6	32	0.914	N-S
3	1	32	0.338	0.1
	2	32	0.484	0.1
	3	32	0.860	N-S
	4	32	0.674	10
	5	32	0.939	N-S
	6	32	0.812	N-S
4	1	32	0.831	N-S
	2	32	0.787	N-S
	3	32	0.914	N-S
	4	32	0.913	N-S
	5	32	0.823	N-S
	6	32	0.695	N-S
5	1	32	0.704	N-S
	2	32	0.590	5
	3	32	0.886	N-S
	4	32	0.736	N-S
	5	32	0.821	N-S
	6	32	0.932	N-S

Table 5-18a

Results of the Hodges-Ajne Test for 1 second ISI CNV's
of Abnormal Subjects

Subject	Harmonic	Number of Trials	M	Significance %
1	1	32	12	N-S
	2	32	12	N-S
	3	32	9	18
	4	32	12	N-S
	5	32	7	2.8
	6	32	12	N-S
2	1	32	12	N-S
	2	32	13	N-S
	3	32	14	N-S
	4	32	11	N-S
	5	32	11	N-S
	6	32	11	N-S
3	1	32	13	N-S
	2	32	9	18
	3	32	13	N-S
	4	32	12	N-S
	5	32	12	N-S
	6	32	12	N-S
4	1	32	11	N-S
	2	32	9	18
	3	32	13	N-S
	4	32	11	N-S
	5	32	8	7.8
	6	32	14	N-S
5	1	32	8	7.8
	2	32	11	N-S
	3	32	10	N-S
	4	32	11	N-S
	5	32	10	N-S
	6	32	10	N-S

Table 5-18b

Results of the Hodges-Ajne Test for 4 second ISI CNV's
of Abnormal Subjects

Subject	Harmonic	Number of Trials	M	Significance %
1	1	30	10	N-S
	2	30	11	N-S
	3	30	10	N-S
	4	30	13	N-S
	5	30	10	N-S
	6	30	12	N-S
2	1	32	11	N-S
	2	32	12	N-S
	3	32	12	N-S
	4	32	8	7.8
	5	32	13	N-S
	6	32	13	N-S
3	1	32	4	0
	2	32	9	18
	3	32	12	N-S
	4	32	10	N-S
	5	32	12	N-S
	6	32	12	N-S
4	1	32	9	18
	2	32	9	18
	3	32	12	N-S
	4	32	10	N-S
	5	32	11	N-S
	6	32	11	N-S
5	1	32	8	7.8
	2	32	10	N-S
	3	32	13	N-S
	4	32	10	N-S
	5	32	11	N-S
	6	32	11	N-S

Table 5-19

Summary of Additivity and Phase Ordering Results Significant
at the 5% level for the Patient Group

ADDITIVITY				PHASE ORDERING		
ISI	SUBJECT NUMBER	NEAREST & FURTHEST MEAN AMPLITUDE	PRE-POST MEAN AMPLITUDE DIFFERENCES	CIRCULAR VARIANCE	MODIFIED CIRCULAR VARIANCE	HODGES-AJNE
1	1		H2 ⁺	H5 *		H5 *
	2		H4 ⁺			
	3					
	4		H2 ⁺ H3 ⁺	H5		
	5		H2 ⁺ H3 ⁺ H4 ⁺ H5 ⁺ H6 ⁺			
4	1	H1 ⁺	X			
	2					
	3	H1 H2		H1 ⁺ H2 ⁺	H1 ⁺ H2 ⁺	H1 *
	4					
	5	H2		H1 H2*	H2*	

H_n indicates that the nth harmonic had a significant result.

+ denotes additivity without phase ordering.

* denotes two or more ordering tests are in agreement.

second ISI CNV responses of normals than were the one second ISI responses. The table also shows that a number of instances of additivity without phase ordering occurred. For the one second ISI CNV's there were 9 spread between 4 different subjects and for the four second ISI CNV's there was only 1 case.

The conclusions must therefore be that (i) the CNV responses of abnormals may contain added energy but that this may be random in nature rather than part of a true CNV response, (ii) the responses show very little phase ordering and (iii) additivity may occur unaccompanied by phase ordering.

5.5 Distinction Between Patients and Normals on the Basis of their CNV's

The previous two sections have shown that the CNV responses of patients tend to be more random than those of normals. It has also been shown that their averaged CNV waveforms were generally different (Section 5.4.1). Further attempts were therefore made to distinguish between the two subject categories in a quantitative manner on the basis of their averaged CNV responses. In particular (i) the distribution of amplitude and phase of the individual harmonics of the averaged CNV were compared and (ii) plots of amplitude verses phase angle with subject as a parameter were also compared. Attempts to distinguish between patients and normals on the basis of the trial-by-trial development of the CNV's are described in the next section.

In the case of the first three harmonics of the one second ISI CNV's of normals (see Table 5-20) it was found that with

Table 5-20

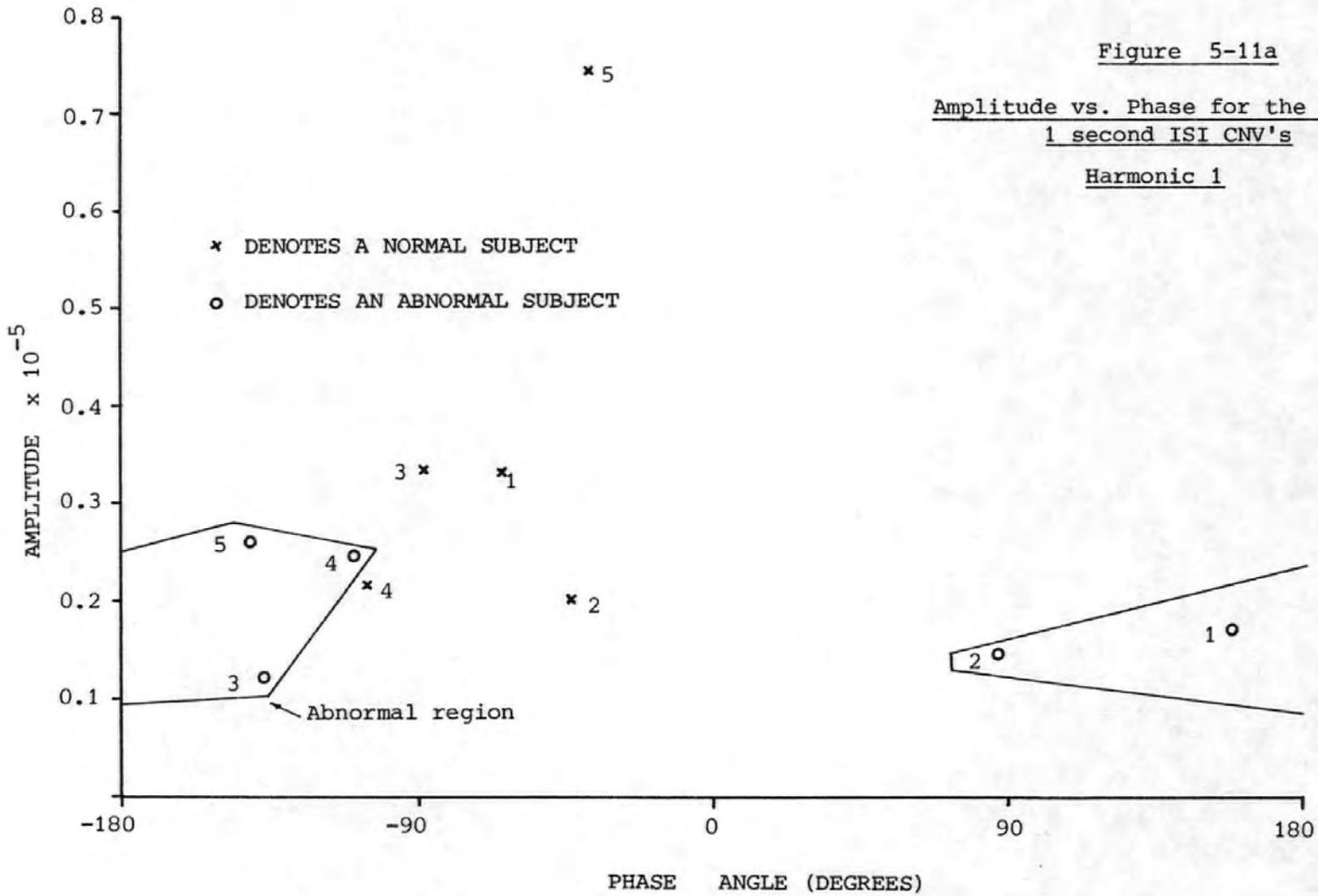
Phase Angles of the Fourier Components of the Averaged one
second ISI CNV's of Normal and Abnormal Subjects

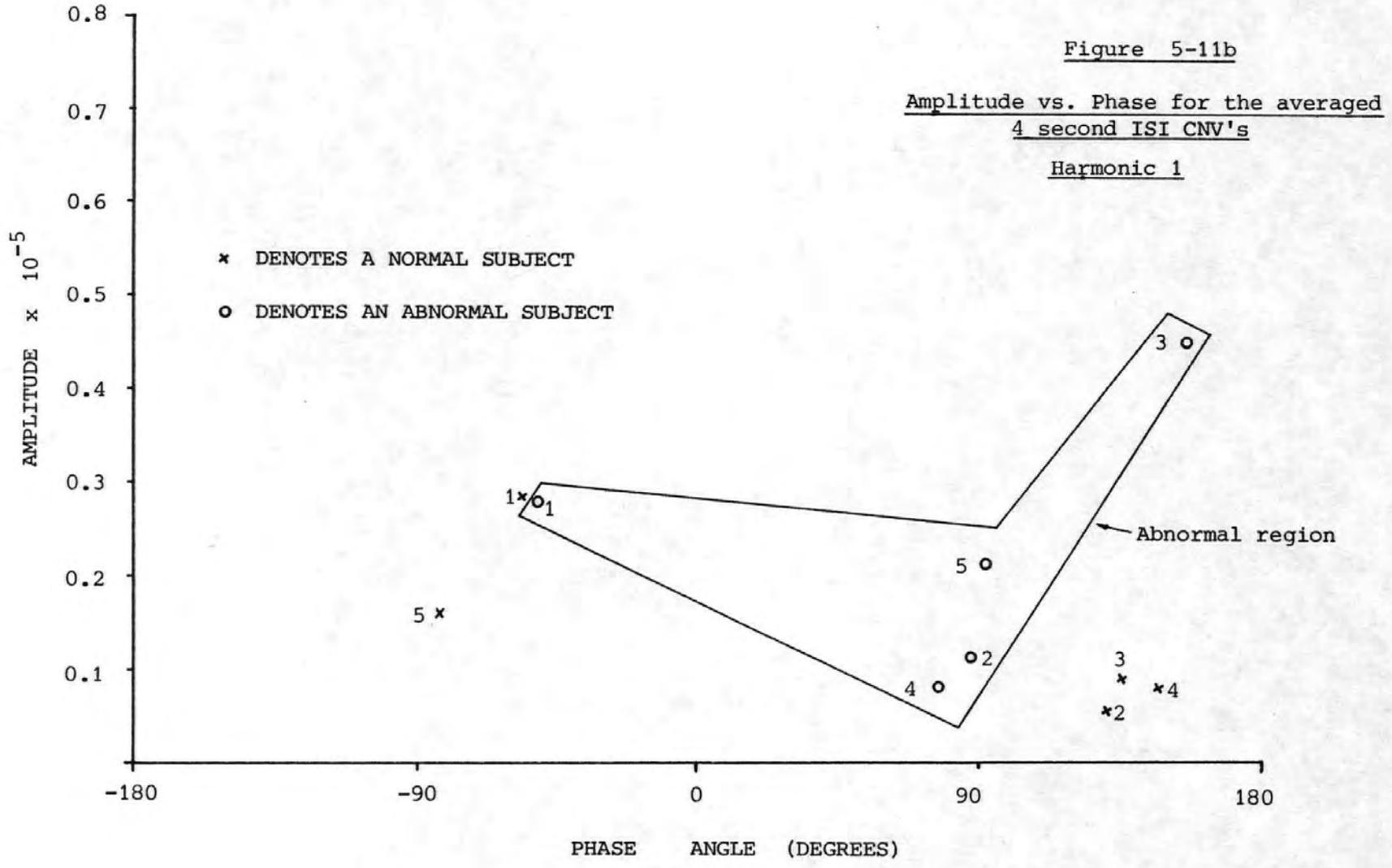
Subject No.	Harmonic	NORMAL SUBJECTS	ABNORMAL SUBJECTS
		Phase Angle °	Phase Angle °
1	1	- 63.9	159.6
	2	- 79.4	40.0
	3	- 99.4	25.0
2	1	- 43.1	88.0
	2	-103.7	73.6
	3	-175.7	88.9
3	1	- 87.8	-136.1
	2	-104.6	67.5
	3	12.2	- 72.9
4	1	-103.5	-109.5
	2	- 42.3	- 36.0
	3	- 92.8	- 99.8
5	1	- 78.9	-139.6
	2	-157.6	- 54.4
	3	-141.2	44.4

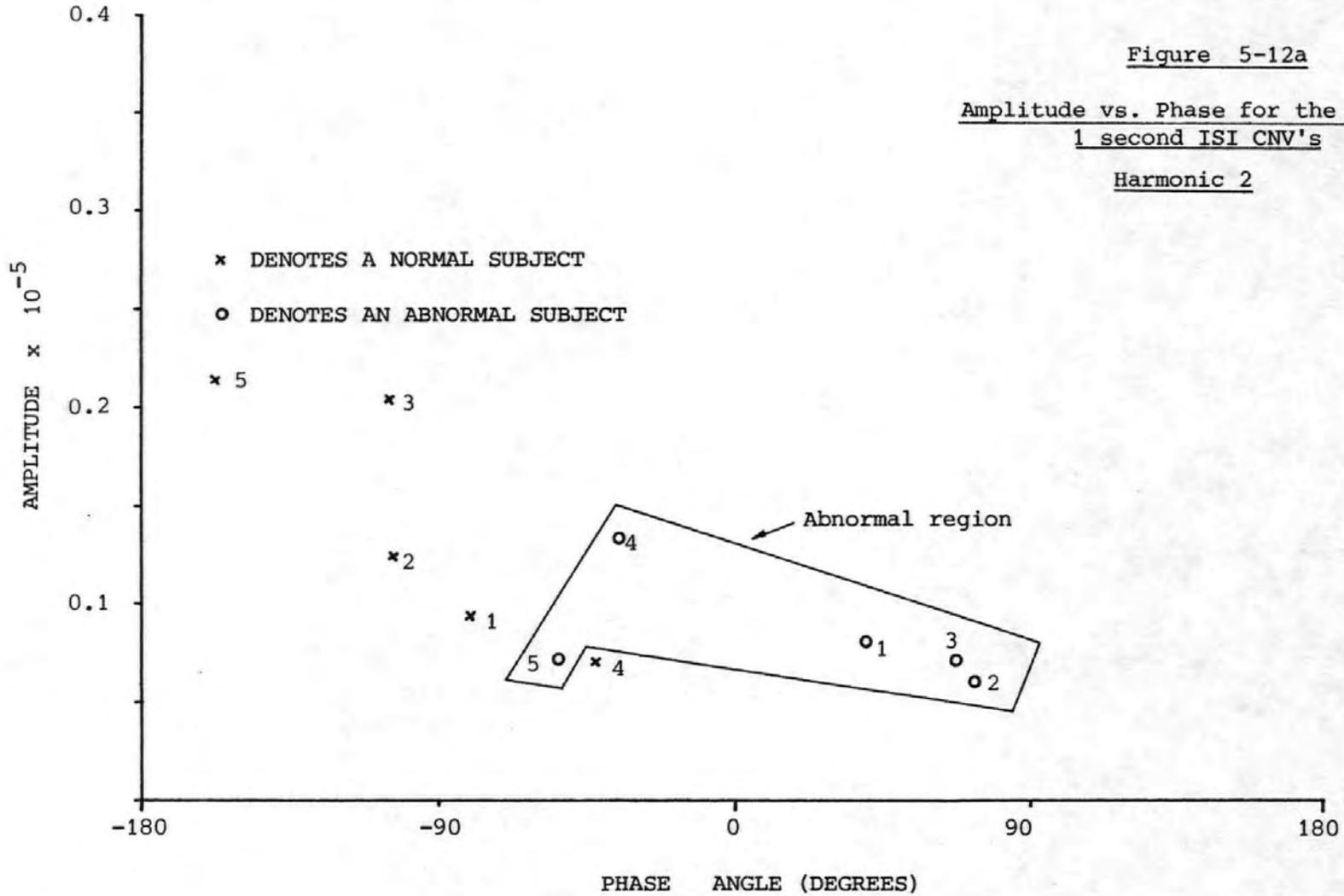
only one exception the phase angles were negative. By contrast the patients had phase angles covering a wider range ($+160^{\circ}$ to -140° compared with -176° to $+12^{\circ}$ for normals).

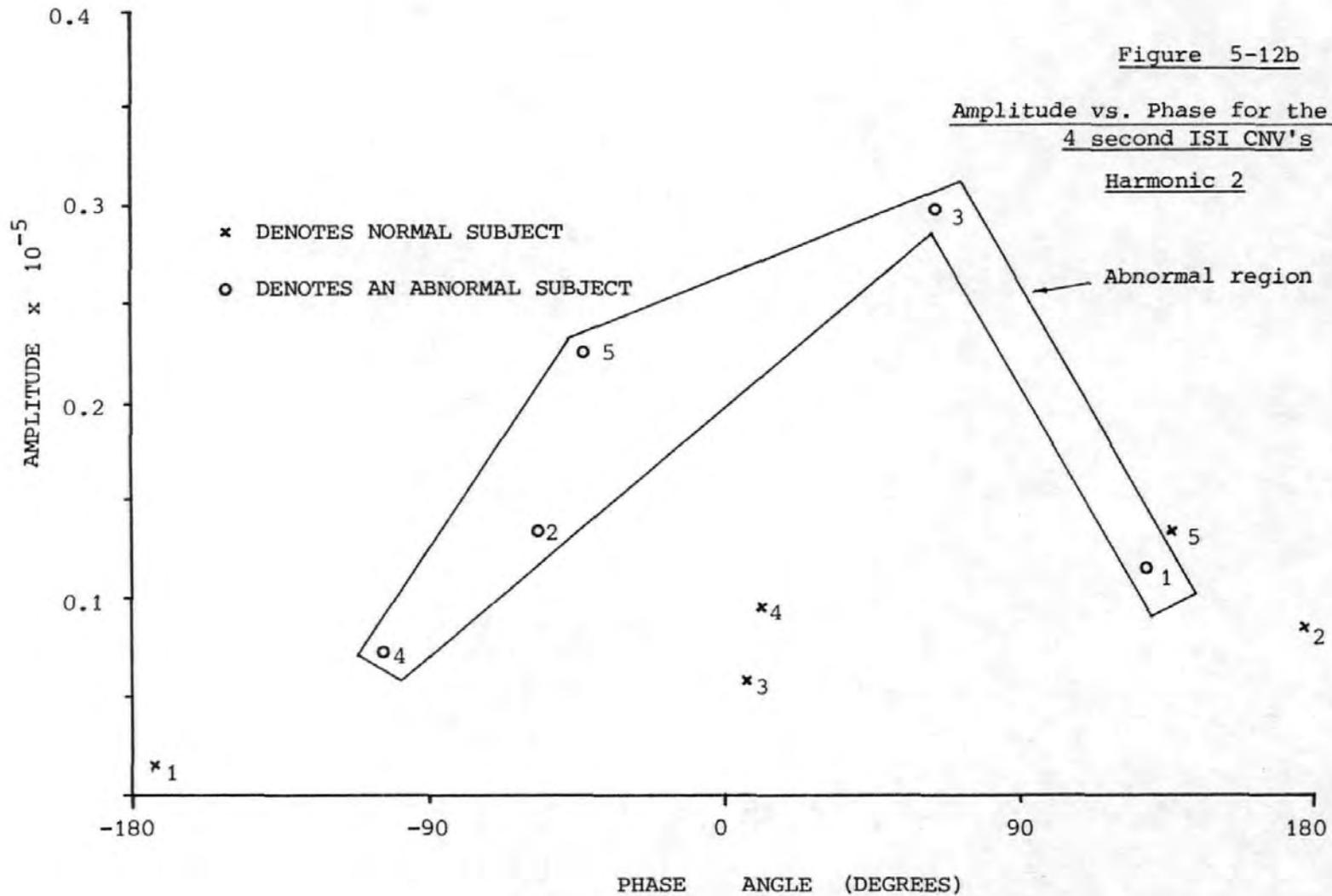
Plots of amplitude verses phase angle with subject as a parameter indicated on the graphs are shown in Figures 5-11a and 5-11b for the one and four second ISI's respectively. The baseline correction procedure used in obtaining these plots was different to that described in Section 5.3.2.1 and used previously. The reason for this change was that the slow return to baseline found in some of the abnormal subjects averaged CNV's made it difficult to obtain a meaningful post S2 average background EEG level. Therefore the averaged CNV's of both patients and normals were baseline adjusted by subtracting the mean value calculated over a one second length of background EEG preceding the S1 stimulus. This is a technique which has been adopted previously by other workers [10].

Inspection of Figures 5-11a and 5-11b reveals that separate areas of the graphs may be ascribed to the normals and patients, although there are instances where the points for the two groups are close together. This method of distinguishing between normals and patients therefore requires more research in order to establish its reliability and usefulness. Similar comments apply in the case of the second harmonic where there is slightly more of an overlap between the areas ascribed to patients and normals (Figures 5-12a and 5-12b). It is possible that such plots may offer useful additional diagnostic evidence for the clinician. It is really necessary first,









however, to investigate much larger populations of patients and normals.

5.6 The Development of the CNV from trial-to-trial in Normals and Patients

In order to examine the trial-by-trial development of the CNV plots of harmonic amplitude and phase verses trial number were produced for both normal and abnormal subjects. In the case of the one second ISI CNV's of normal subjects the plots for harmonic two, subjects 1 and 2 exhibited slight evidence for increased amplitudes i.e. a plateau, over the range of trial numbers 9 to 20 and 12 to 22 respectively. These plots are shown in Figures 5-13a and 5-13b. None of the other amplitude plots, either of normal or abnormal subjects, showed any obvious patterns.

There was some evidence in the plots of phase verses trial number that the development of the one second ISI CNV response was different in abnormal subjects to that in normal subjects. Thus whilst for the normals there was a tendency in the case of harmonic 1 for only negative phase angles to occur in the earlier trials with positive angles only occurring in the subsequent trials, (see Figure 5-14), the abnormal subjects showed that both positive and negative phase angles could occur throughout the acquisition sequence (see Figure 5-15). There was also some evidence from the amplitudes verses phase plots for the averaged CNV's to suggest that those normals whose amplitude and phase co-ordinates were furthest from those of patients had trial-by-trial phase angles which were negative up to the final few trials. Thus the phase angles

AMPLITUDE VS TRIAL NUMBER

HARMONIC 2

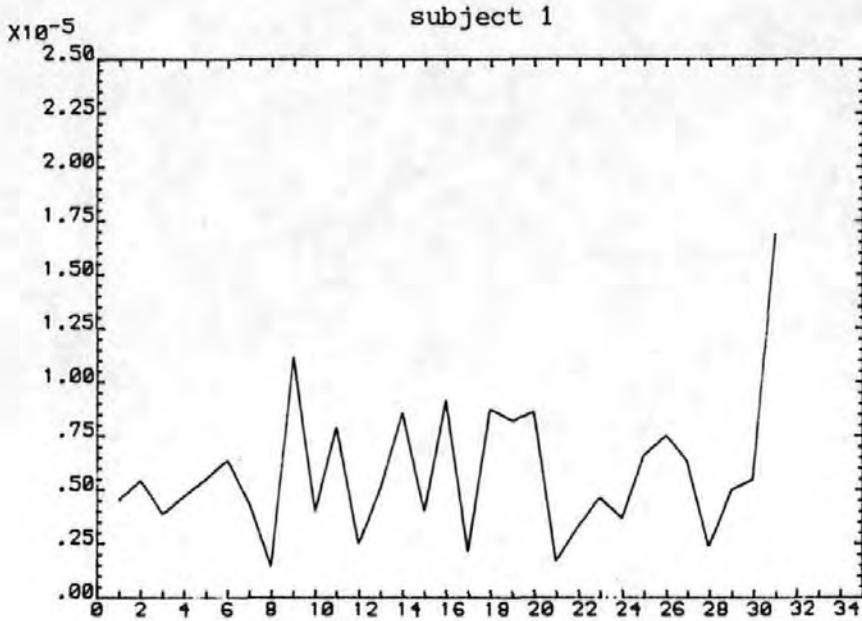


Figure 5-13a

Amplitude vs Trial number for a 1sec. ISI CNV of a normal subject. Harmonic 2.

AMPLITUDE VS TRIAL NUMBER

HARMONIC 2

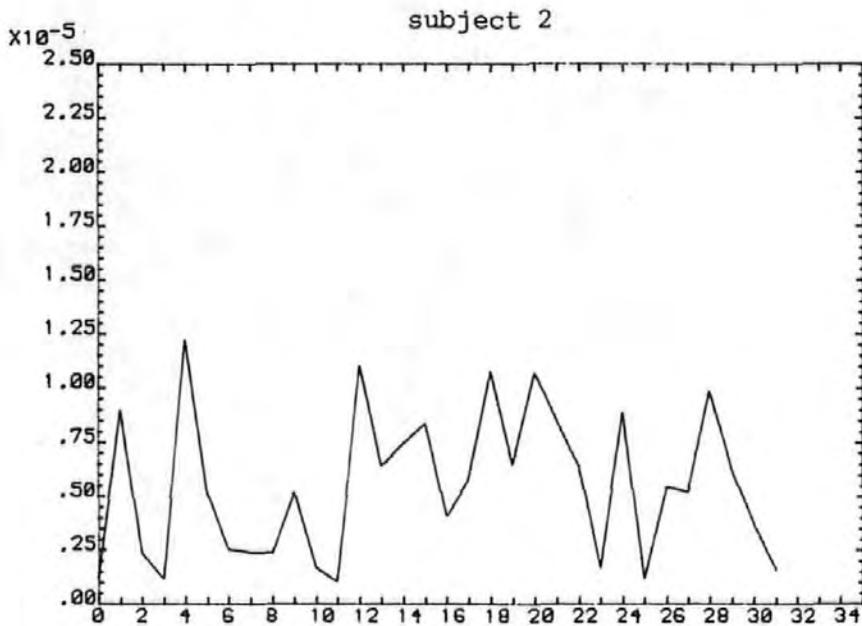


Figure 5-13b

Amplitude vs Trial number for a 1sec. ISI CNV of a normal subject. Harmonic 2.

PHASE VS TRIAL NUMBER

HARMONIC 1

subject 2

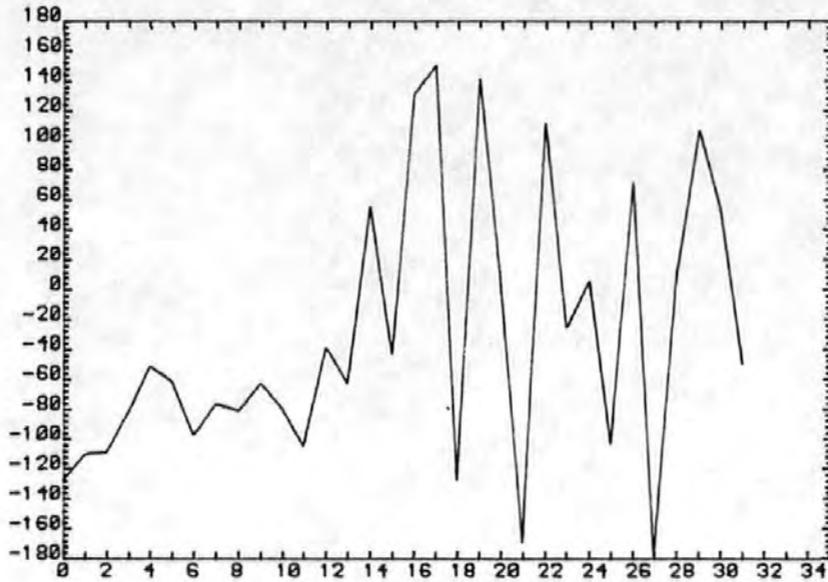


Figure 5-14

Phase vs Trial number for a 1sec. ISI CNV of a normal subject.

PHASE VS TRIAL NUMBER

HARMONIC 1

subject 1

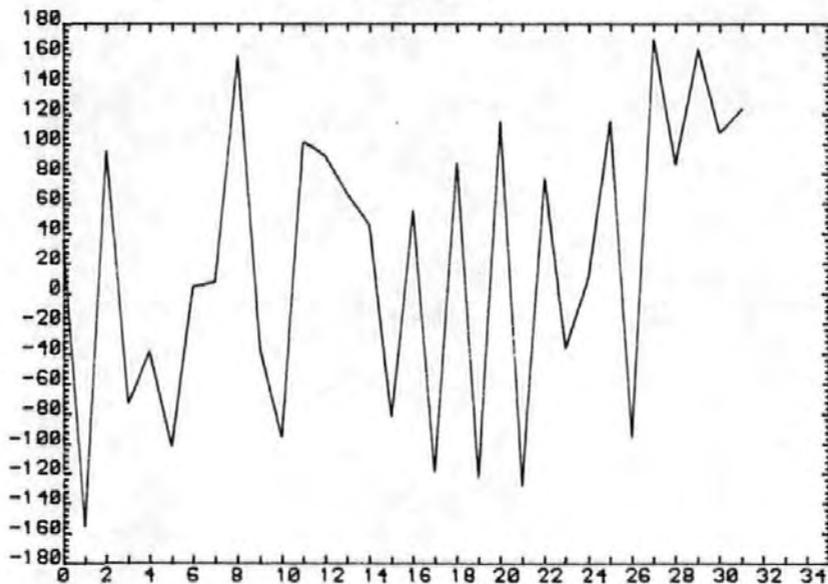


Figure 5-15

Phase vs Trial number for a 1sec. ISI CNV of an abnormal subject.

of the most "normal" subjects were characteristically negative for the earlier trials whilst those of the abnormals were characteristically randomly positive or negative. This difference may provide another means for differentiating between the two groups but again more results would be required to confirm these findings. In the case of harmonic 2 both positive and negative phase angles were found throughout the acquisition sequence and for both normals and patients (see Figure 5-16a and 5-16b).

Plots of amplitude verses trial number for the first harmonic of the four second ISI CNV's of the normals did not reveal any definite pattern, although in some cases the mid-trial values may be slightly larger (see Figure 5-17). No similar evidence could be found for the abnormal subjects.

The plots of phase angle verses trial number for the first harmonic of the four second ISI CNV's were inspected. The phase angles for the initial trials were either negative or positive for the normal subjects (see Figure 5-18) but were positive for each of the abnormal subjects (see Figure 5-19). Again this may be useful but more results would be required to confirm these findings.

The corresponding plots of amplitude and phase for the second harmonic of the four second ISI CNV's of both patients and normals revealed little further information.

In conclusion it would appear that the phase properties of the harmonic components are more significant than those of the amplitude. In particular the phase properties of the

PHASE VS TRIAL NUMBER

HARMONIC 2

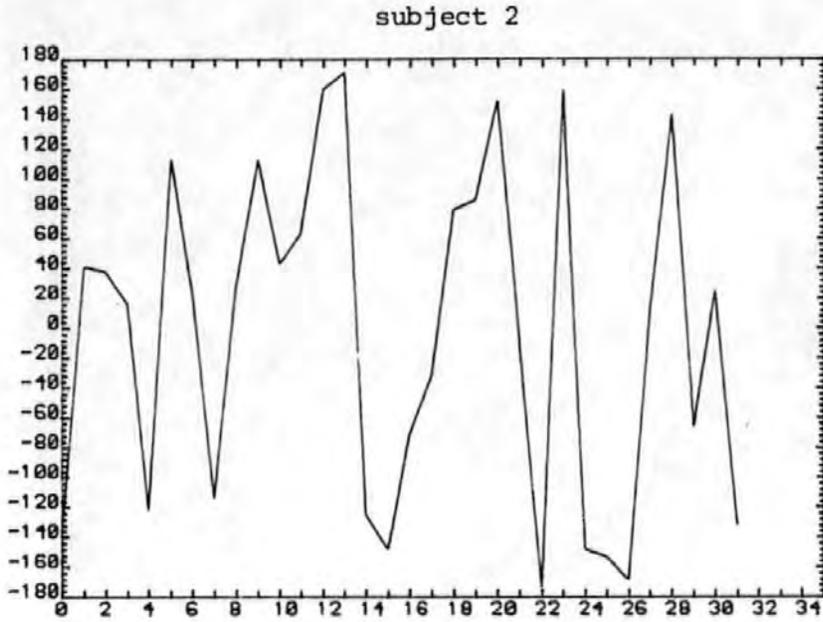


Figure 5-16a

Phase vs Trial number for a 1sec. ISI CNV of a normal subject.

PHASE VS TRIAL NUMBER

HARMONIC 2

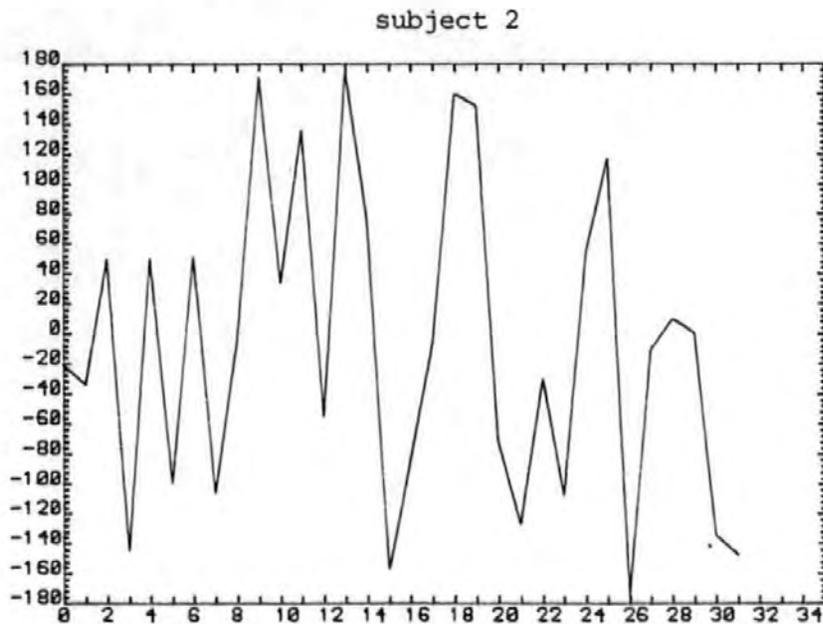


Figure 5-16b

Phase vs Trial number for a 1sec. ISI CNV of an abnormal subject.

AMPLITUDE VS TRIAL NUMBER

HARMONIC 1

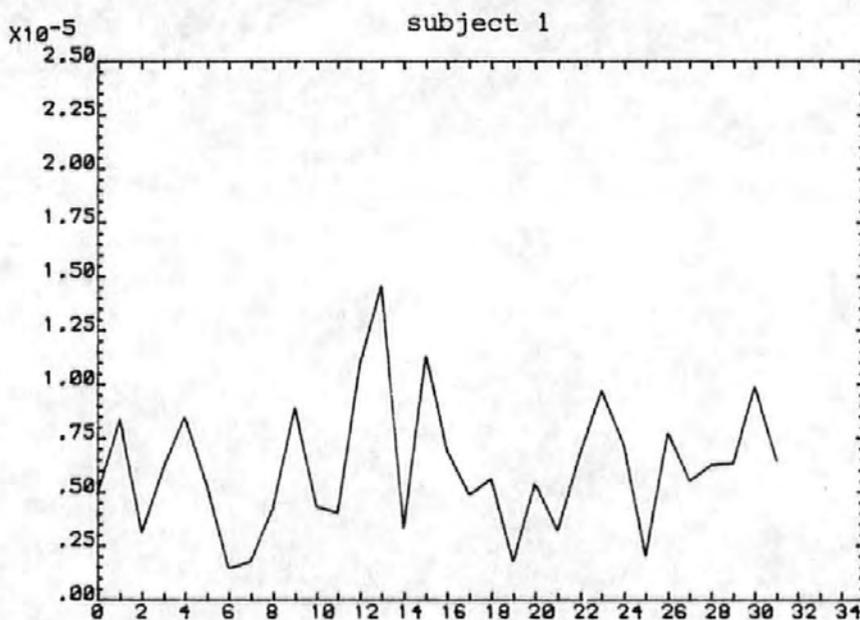
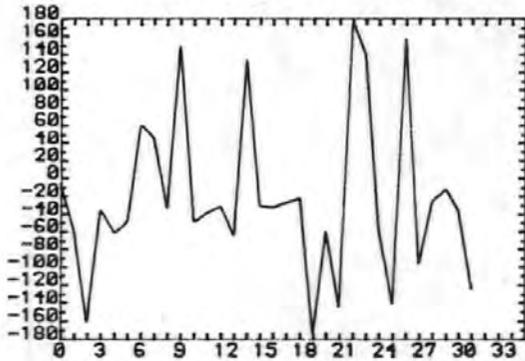


Figure 5-17

Amplitude vs Trial number for a 4sec. ISI CNV of
a normal subject.

PHASE VS TRIAL NUMBER

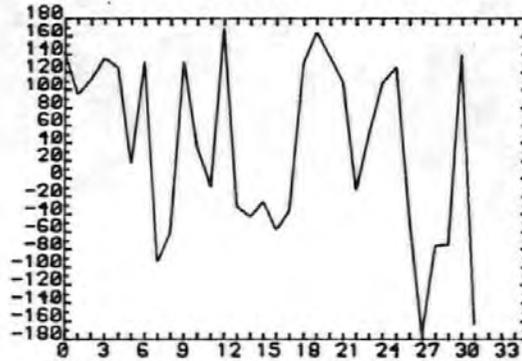
HARMONIC 1



subject 1

PHASE VS TRIAL NUMBER

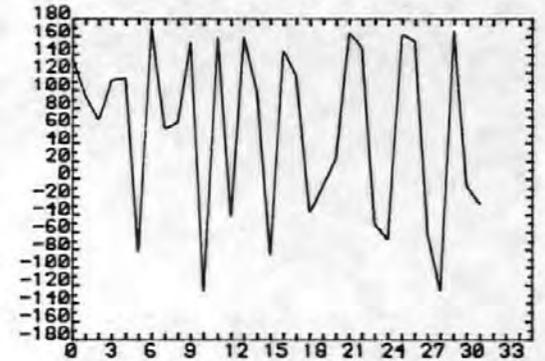
HARMONIC 1



subject 2

PHASE VS TRIAL NUMBER

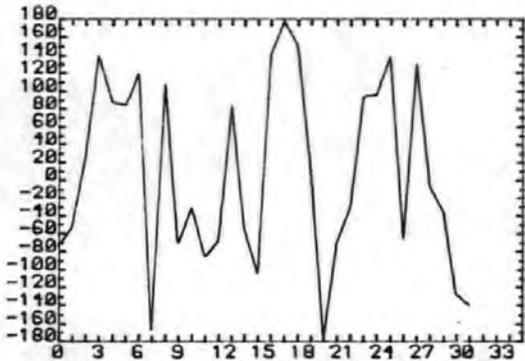
HARMONIC 1



subject 3

PHASE VS TRIAL NUMBER

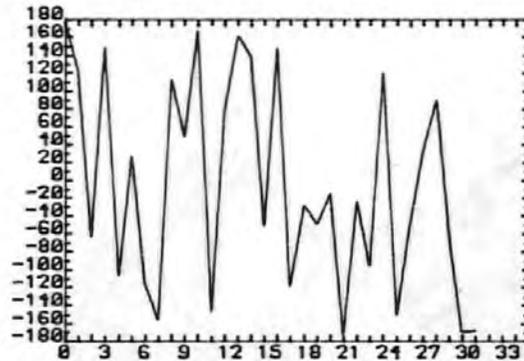
HARMONIC 1



subject 4

PHASE VS TRIAL NUMBER

HARMONIC 1



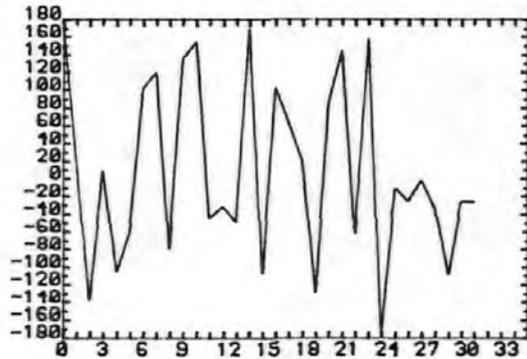
subject 5

Figure 5-18

Phase vs Trial number for the 4sec.
ISI CNV's of the normal subjects.

PHASE VS TRIAL NUMBER

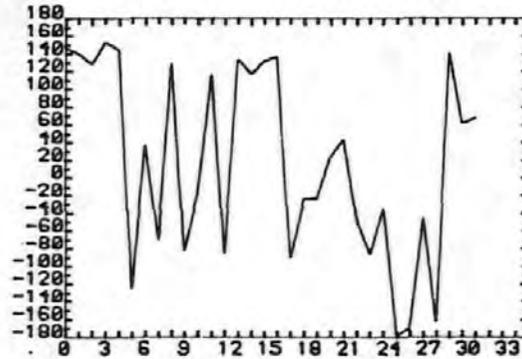
HARMONIC 1



subject 1

PHASE VS TRIAL NUMBER

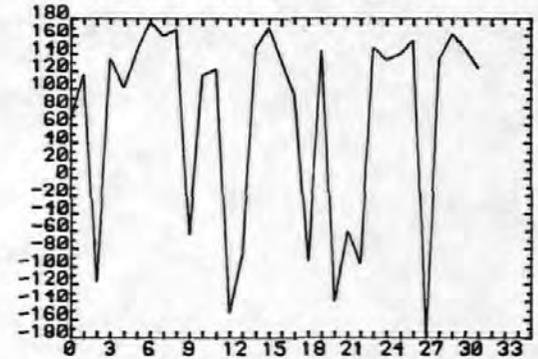
HARMONIC 1



subject 2

PHASE VS TRIAL NUMBER

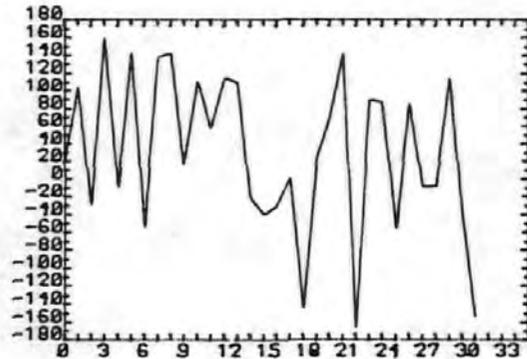
HARMONIC 1



subject 3

PHASE VS TRIAL NUMBER

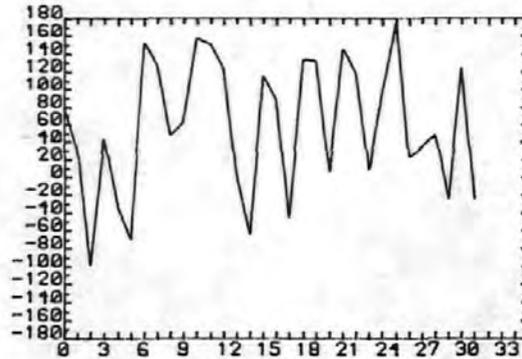
HARMONIC 1



subject 4

PHASE VS TRIAL NUMBER

HARMONIC 1



subject 5

Figure 5-19

Phase vs Trial number for the 4sec.
ISI CNV's of the abnormal subjects.

first harmonic of the one second ISI CNV response may, subject to confirmation, allow differentiation between the normal and patient groups. In this case the initial phase angles are negative for the normals and either positive or negative for the patients. The onset of positive phase values occurs at much later trial numbers for normals than for the patients.

References for Section 5

- [1] Corby J C, Kopell B S, "Differential Contributions of Blinks and Vertical Eye Movements as Artifacts in EEG Recording"
Psychophysiology, Volume 9, pp 640-644, November 1972.
- [2] McCallum W C, Walter W G, "The Effects of Attention and Distraction on the Contingent Negative Variation in Normal and Neurotic Subjects"
Electroencephalography and Clinical Neurophysiology, Volume 25, pp 319-329, 1968.
- [3] Girton D G, Kamiya J, "A Simple on-line Technique for Removing Eye Movement Artifacts from the EEG"
Electroencephalography and Clinical Neurophysiology, Volume 34, pp 212-216, 1973.
- [4] Jervis, B W, Nichols M J, Allen E, Hudson N G, Johnson T E, "The Quantitative Assessment of Electroencephalograms Corrected for Eye Movement Artefact"
EUSIPCO-80, Short communication and poster digest, Swiss Federal Institute of Technology, Lausanne, Switzerland 1980.
- [5] Sayers B McA, Beagley H A, Henshall W R, "The Mechanism of Auditory Evoked EEG Responses"
Nature, Volume 247, pp 481-483, 1974.
- [6] Sayers B McA, Beagley H A, Riha J, "Pattern Analysis of Auditory-Evoked EEG Potentials"
Audiology, Volume 18, pp 1-16, 1979.

- [7] Beagley H A, Sayers B McA, Ross A J, "Fully objective ERA by Phase Spectral Analysis" Acta Otolaryngology, Volume 87, pp 270-278, 1979.
- [8] Sayers B McA, "Science and Judgement in Biological Signal Analysis" in "Signal analysis and pattern recognition in biomedical engineering" Edited by Inbar G F, Halstead Press, New York, pp 3-32, 1975.
- [9] Dongier M, "Separation of the various independant phenomena among the slow potential changes (Contingent Negative Variations)" Electroencephalography and Clinical Neurophysiology, Volume 27, pp 108-109, 1969.
- [10] Loveless N E, Sanford A J, "The CNV Baseline: Considerations of Internal Consistency of Data" Published in "Event-Related Slow Potentials of the Brain: Their relations to Behaviour", Edited by McCallum W C & Knott J R. Electroencephalography and Clinical Neurophysiology Supplement No. 33, pp 19-23. Published by Elsevier Scientific Publishing Company, 1973.
- [11] Cooper R, Osselton J W, Shaw J C, "EEG Technology" page 13. Published by Butterworths, 1969.

6. Conclusions

6.1 Eye Movement Artefact Removal

A method of testing the effectiveness of eye movement artefact removal techniques has been derived. This procedure was used to test four methods of removing eye movement artefact from the EEG. These four methods were (i) the analogue method using a potentiometer due to McCallum and Walter [1], (ii) The computational method due to Quilter et al [2], (iii) our extension of Quilters method to incorporate three EOG components, (iv) our extension of Quilters to incorporate four EOG components. Recordings were made from normal volunteer subjects who were asked to make deliberate eye movements and the various methods of removing the artefact from the EEG were compared.

It was found that the computational correction method gave excellent results particularly when the horizontal and vertical components of the EOG of both eyes were taken into account. However it was possible to obtain good artefact removal using both horizontal, but only one vertical EOG component (i.e. the three EOG components method). The reduction in the computational effort obtained by this simplification may be important in an on-line multi-channel situation.

The four EOG component method of eye movement artefact removal was used in all the subsequent investigations.

6.2 Evoked Potentials

New models have been developed to describe the fundamental nature of evoked potentials in terms of additive and phase re-ordered components. The models showed that a repetitive additional component or an alignment of existing background EEG components could give rise to observable phase ordering in the post stimulus response. Tests were then contrived in an attempt to determine which of these mechanisms was operative in both auditory evoked potentials and the much longer Contingent Negative Variations.

6.3 Auditory Evoked Potentials

The tests mentioned above were applied to the auditory evoked potentials recorded from normal volunteer subjects. Of the 36 sets of results (3 subjects, 2 stimulus levels, 6 harmonics) 29 exhibited phase ordering. Some of the harmonic components were found to contain additional energy and all of these exhibited phase ordering. This finding was consistent with the proposed additivity model although a combination of the additive and phase re-ordered models could give the same results. However, 15 of the 29 results showing phase ordering did not show any additive effects. Thus either the additivity tests were not sufficiently sensitive to detect small additional components, or pure phase re-ordering was also present.

For all the subjects tested, and for both levels of auditory stimulation, the first three harmonic components all showed significant phase ordering. Furthermore, in all but

one harmonic for one subject, the results of all three phase ordering tests were significant at 0.1% or better. This result may be useful in the diagnosis of audiological defects or establishing auditory thresholds, since the statistical results give a quantitative indication of the presence or absence of a response whereas inspection of the averaged waveform is rather subjective.

6.4 The CNV's of Normal Subjects

The pre- and post- stimulus (i.e. broadband) energy tests showed that all the CNV's of the normal subjects contained additional energy. However only a small proportion of the harmonics examined showed any evidence of this feature. Thus the additional energy must have been at frequencies other than those studied. Some of the harmonics exhibited phase ordering although this feature was much less pronounced than the phase ordering observed in the auditory evoked potentials. The one second ISI CNV's showed more ordering than the four second and most of the ordering observed was confined to the lower harmonics. There were some examples of phase ordering without any accompanying additivity. Thus like the auditory responses the CNV's might be explained in terms of a mixture of the additive and phase re-ordered models.

However some results were in conflict with this conclusion. There were instances where additivity was detected in the absence of any phase ordering. This may have been the effect of a non-repetitive additional component.

The trial-by-trial analysis of the phase components of the CNV's revealed that for the first harmonic negative phase angles tended to occur in the early trials. Some of the second harmonic amplitude components showed a plateau in the middle of the acquisition sequence.

6.5 The CNV's of Abnormal Subjects

The broadband energy tests revealed fewer instances of additivity in the CNV's of the abnormal subjects than for the normal subjects. However the individual harmonics showed a slightly higher number with detectable additivity than those of the normals.

There were fewer cases of phase ordering than for the CNV's of the normal subjects. Thus the CNV's of the abnormal subjects did not fit either the additive or phase re-ordered models so well as those of the normal subjects. This may have been due to the limitations of the models or it may have indicated the differing nature of the responses obtained from the Huntingtons Chorea group.

The averaged CNV waveforms tended to support the latter theory since the one second CNV's showed little similarity with those of the normals. The four second CNV's however, particularly those for subjects three and five, did appear slightly more like the normal CNV's. These observations are supported by the phase ordering tests. The one second ISI CNV's exhibited no phase ordering in the lower harmonics whereas those of subjects three and five for the four second

ISI CNV's did.

Generally the CNV's of the patient group were most easily characterised by their averaged waveforms. These waveforms sometimes showed a slow return to baseline.

6.6 Distinctions between the CNV's of Normals and Patients

As stated above the averaged CNV waveforms of the HC patients tended to be rather different to those of the normal group. This may be a sufficient difference to aid the detection of H.C. However, the patients for whom H.C. had only recently been diagnosed (i.e. those at an early stage of the illness) showed more normal averaged CNV's than did those for whom the disease was at an advanced stage. (Compare the recently diagnosed H.C. of Figure 5-8c with Figures 5-4a-e).

Another feature of the averaged waveforms of the patient group was that of the slow return to baseline subsequent to the S2 stimulus exhibited by some of the subjects. This may also be a useful diagnostic feature.

An alternative to the averaged waveform was provided by the plots of amplitude verses phase for the various normal and abnormal subjects. It may be possible to ascribe certain areas of these graphs to each of the two populations.

The analysis of the phase angle of the first harmonic on a trial-by-trial nature may also be of diagnostic value since all the normal subjects tested produced negative phase

angles for the initial trials whereas the patient group did not.

6.7 Future Work

6.7.1 Eye Movement Artefact Removal

As previously mentioned hardware to perform the eye movement artefact removal procedure is to be developed and produced commercially. There are a number of improvements and tests which should be carried out before this is done. The method should first be tested to ensure that any frontal EEG activity which may be present in the EOG signals is not superimposed on the EEG by the correction procedure. One possible way of achieving this would be to experiment with the placement of EOG electrodes in order to find electrode positions which maximise the EOG and minimise the EEG amplitudes.

The presence of harmonics of the EOG in the corrected EEG signal may be due to a frequency selective path between the eye and the scalp. The most likely characteristic of such a path would be that of a low pass filter. The correction procedure assumes that the path between the eye and the scalp is linear and has an infinite bandwidth. The correction method may thus overcorrect the higher harmonics of the EOG and hence leave traces of these components in the corrected signal. A possible cure for this phenomena would be to introduce a filter (either electrical or digital) into the EOG signal path. The filter would have to have the same

properties as those of the electrical path between the eye and the scalp. This may simply be of a frequency selective nature or may include non-linearities. It might even be possible for the correction procedure to have a 'set-up' mode whereby the nature of this network is determined by the correction system before the corrector is used.

6.7.2 Auditory Evoked Potentials

The statistical tests for phase ordering should be applied to a much larger sample of auditory evoked potentials covering a wide range of auditory stimulation levels. The results of the statistical tests should then be correlated with the findings of audiologists to determine whether the statistical tests for phase ordering would be audiologically useful.

6.7.3 The CNV's of Patients and Normals

Although several possible diagnostic procedures have been suggested, much larger samples of normals, patients with H.C. and patients with other neurological defects must be examined before these procedures could safely be adopted.

Further research should be aimed at establishing why the broadband energy tests do show additional energy whereas the tests on the individual harmonics sometimes do not. It should be possible to account for this extra energy and hence balance the energy figures. The extra energy may be contained in the d.c. term or in the higher harmonics.

The time to return to the baseline should be measured to establish whether any correlation can be detected between this and the severity of the disease.

References for Section 6

- [1] McCullum W C, Walter W G, "The Effects of Attention and Distraction on the Contingent Negative Variation in Normal and Neurotic Subjects"
Electroencephalography and Clinical Neurophysiology,
Volume 25, pp 319-329, 1968.
- [2] Quilter P M, MacGillivray B B, Wadbrook D G, "The Removal of Eye Movement Artefacts from EEG Signals using Correlation Techniques" Random Signals Analysis,
IEE Conference Publication No. 159, pp 93-100.

7. Acknowledgements

I would like to thank the many people who have given me help and advice throughout the course of this research project. In particular special thanks must go to my Supervisors, Drs. Barrie Jervis (Dept. of Communication Engineering, Plymouth Polytechnic), Elaine Allen (Dept. of Neurological Sciences, Freedom Fields Hospital, Plymouth), and John Fryer (Dept. of Mathematics Statistics and Operational Research, Exeter University) for their unwavering support and encouragement. Special thanks must also go to Dr. Terry Johnson (Dept. of Mathematics, Statistics and Computing, Plymouth Polytechnic) for his advice on the use of angular statistics and allowing me to use his computer subroutines to calculate these statistics. I would particularly like to thank Mr. Nigel Hudson (Dept. of Neurological Sciences, Freedom Fields Hospital), for applying and removing the measuring electrodes from the subjects and for liaising with the consultants and next of kin of the Huntington's chorea patients.

Thanks are also due to the members of staff, both academic and technical of the Departments of Communication Engineering and Electrical & Electronic Engineering who have given their advice so freely. Special mention must be made to the advice given by Mr. Alan Roberts and Mr. Adrian Jerram concerning the PDP8 minicomputers and the hybrid computer interface utilized in the data acquisition system.

I would also like to thank the past and present members of staff of the Plymouth Polytechnic Computer Centre for their

help and advice on programming matters.

For financial support I would like to thank the Devon County Local Education Authority and the Nuffield Foundation.

For permission to use their facilities I would like to thank Mr. B.R. Webster, Head of the Department of Electrical & Electronic Engineering.

For showing me their facilities and for several hours of useful and interesting discussion I would like to thank Drs. W.C. McCallum and R. Cooper of the Burden Neurological Institute, Bristol and Dr. P. Fenwick of St. Thomas' Hospital and The Institute of Psychiatry, London.

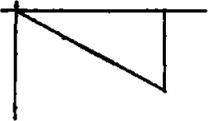
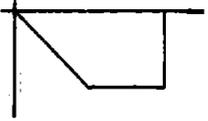
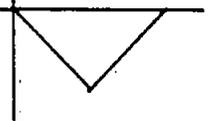
For her accurate and painstaking typing of this thesis I would like to thank Mrs. L. Mortimore.

Finally I would like to thank all those volunteers who participated in the tests and without whom this research would not have been possible.

APPENDIX 8.1

Calculation of Expected Phase Values for Idealized CNV's

By considering the CNV as one of the simple shapes shown below it is possible to calculate the resulting phase angles for each of the harmonic frequency components. These calculations were performed using a microcomputer to compute the discrete Fourier transform of each of the possible CNV shapes. The listing of the computer programme and the results are given below.

<u>WAVESHAPE</u>	<u>HARMONIC</u>					
	1	2	3	4	5	6
Type 0 	-90	-90	-90	-90	-90	-90
Type 1 	-57.4	-89.6	-77.5	-89.3	-81.9	-88.9
Type 2 	0	*	0	*	0	*

* Phase angles are indeterminate due to zero amplitudes of these harmonic components.

```

10 PI=3.14159265
20 INPUT " NUMBER OF POINTS ";N
30 DIM FT(N-1)
40 INPUT " MAXIMUM AMPLITUDE ";K
50 INPUT " WHICH TYPE OF CNV WAVESHAPE ";Q
60 IF Q=0 THEN SL=K/N
70 IF Q>0 THEN SL=2*K/N
80 INPUT " HARMONIC NUMBER ";H
90 FOR I=0 TO N-1
100 IF I < N/2 THEN FT(I)=SL*I : GOTO 140
110 IF Q=0 THEN FT(I)=SL*I
120 IF Q=1 THEN FT(I)=K
130 IF Q=3 THEN FT(I)=-SL*I+2*K
140 NEXT I
150 C=0
160 S=0
170 FOR I=0 TO N-1
180 C=C+FT(I)*COS(2*PI*I*H/N)
190 S=S-FT(I)*SIN(2*PI*I*H/N)
200 NEXT I
210 C=C/N
220 S=S/N
230 PRINT "REAL=";C;" " ;"IMAG=";S
240 TH=180/PI*ATN(S/C)
250 IF C>0 GOTO 290
260 IF S>0 THEN TH=TH+180 : GOTO 290
270 TH=TH-180
280 GOTO 290
290 PRINT "THETA=";TH
300 GOTO 80

```

APPENDIX 8.2

Transformation of the Rayleigh Probabilities to those of Circular Variance

Using the definitions given in section 2.2.2.1 a relationship may be derived between the Rayleigh statistic R and the circular variance statistic S_0 . For a set of N angles $\{\theta_i\}$ the Rayleigh statistic is given by

$$R = \sqrt{\left[\begin{array}{c} N \\ \sum_{i=1} \cos \theta_i \end{array} \right]^2 + \left[\begin{array}{c} N \\ \sum_{i=1} \sin \theta_i \end{array} \right]^2}$$

Clearly this is simply N times \bar{R} as defined by equation 6 of section 2.2.2.1.

$$\text{since } S_0 = 1 - \bar{R}$$

$$\text{then } S_0 = 1 - \frac{R}{N}$$

Hence a table of critical values of S_0 may be compiled from a table of critical values of R by dividing each 'R' value by N and subtracting the resulting value from unity.

A table thus obtained is given overleaf.

Table of Critical Values of S_0

N	P			
	0.10	0.05	0.01	0.001
5	0.333	0.246	0.121	0.009
6	0.382	0.310	0.175	0.060
7	0.428	0.358	0.229	0.109
8	0.465	0.398	0.275	0.153
9	0.496	0.431	0.313	0.192
10	0.552	0.460	0.335	0.225
11	0.544	0.484	0.373	0.257
12	0.563	0.506	0.398	0.284
13	0.580	0.525	0.420	0.308
14	0.595	0.542	0.440	0.331
15	0.609	0.557	0.458	0.351
16	0.621	0.571	0.475	0.370
17	0.633	0.583	0.490	0.387
18	0.643	0.595	0.540	0.403
19	0.652	0.606	0.516	0.417
20	0.661	0.615	0.528	0.431
21	0.669	0.625	0.539	0.444
22	0.677	0.633	0.549	0.456
23	0.684	0.641	0.559	0.467
24	0.691	0.649	0.568	0.478
25	0.697	0.656	0.577	0.488
30	0.723	0.685	0.613	0.530
35	0.744	0.708	0.641	0.564
40	0.760	0.727	0.664	0.591
45	0.774	0.743	0.682	0.614
50	0.786	0.756	0.699	0.633
64	0.810	0.784	0.732	0.672

APPENDIX 8.3

Probability Levels for the Modified Rayleigh Test

The following table gives the critical values of U_0 as derived from Moore [section 2 reference 5]. The derivations are calculated from

$$U_0 = 1 - \frac{2\sqrt{N} R^*}{(N+1)}$$

Table of Critical Values of U_0

N	P			
	0.1	0.05	0.01	0.001
5	0.264	0.192	0.094	
6	0.320	0.248	0.139	
7	0.365	0.295	0.181	
8	0.404	0.334	0.219	
9	0.436	0.368	0.253	
10	0.462	0.397	0.282	0.176
12	0.506	0.444	0.332	0.224
14	0.541	0.483	0.375	0.266
16	0.570	0.515	0.412	0.304
18	0.594	0.541	0.443	0.336
20	0.614	0.563	0.469	0.365
22	0.631	0.583	0.491	0.389
24	0.647	0.600	0.512	0.413
30	0.684	0.641	0.560	0.470
32	0.692	0.652	0.573	0.485
40	0.725	0.688	0.617	0.535
60	0.774	0.744	0.684	0.617
64	0.782	0.752	0.695	0.629
80	0.804	0.778	0.726	0.667
100	0.826	0.802	0.755	0.700

APPENDIX 8.4

Probability Levels for the Hodges-Ajne Test

The following table gives the critical values of m for the Hodges-Ajne test.

Table of Critical Values of m

N	P			
	0.10	0.05	0.025	0.01
9	0	0	0	0
10	1	0	0	0
11	1	0	0	0
12	1	1	0	0
13	1	1	1	0
14	2	1	1	0
15	2	2	1	1
16	2	2	1	1
17	3	2	2	1
18	3	3	2	2
19	3	3	2	2
20	4	3	3	2
21	4	4	3	2
22	5	4	3	3
23	5	4	4	3
24	5	5	4	3
25	6	5	4	4
30	7	7	6	5
35	9	9	8	7
40	11	10	10	9
50	15	14	13	12
60	19	18	14	13

APPENDIX 8.5

The Paired t-Test

Where a set of experimental results fall naturally into pairs the above statistical test may be employed to establish whether a consistent difference between each of the paired values exists. An example will best illustrate the method.

Two students, X and Y, measured the resistance of twelve resistors. Each student measured all the resistors and the results they obtained are shown in the table.

	RESISTOR											
STUDENT	1	2	3	4	5	6	7	8	9	10	11	12
X	100	90	130	110	117	75	32	88	41	57	18	67
Y	105	91	128	109	119	77	32	87	36	60	21	72

Are one students results consistently higher than the others?

The differences are first calculated;

$$x = \quad -5 \quad -1 \quad 2 \quad 1 \quad -2 \quad -2 \quad 0 \quad 1 \quad 5 \quad -3 \quad -3 \quad -5$$

If there was no consistent difference between the students measurements then the differences should form a zero mean normal distribution. The t-test is used to test this hypothesis.

$$\begin{aligned} \text{Mean value} &= \bar{x} = \frac{\Sigma x}{N} \\ &= \frac{-12}{12} \end{aligned}$$

$$\bar{x} = -1$$

An estimate of the variance is given by

$$\hat{\sigma}^2 = \frac{1}{N-1} \left[\Sigma x^2 - \frac{(\Sigma x)^2}{N} \right]$$

$$= \frac{1}{12-1} \left[108 - \frac{(-12)^2}{12} \right]$$

$$= 8.73$$

$$\therefore \hat{\sigma} = 2.95$$

The t statistic is given by

$$\begin{aligned} t &= \frac{|\bar{x} - \mu|}{\frac{\hat{\sigma}}{\sqrt{N}}} \\ &= \frac{|-1 - 0|}{2.95} \quad 12 \end{aligned}$$

$$t = 1.17 \text{ with } (N-1) \text{ degrees of freedom.}$$

Tables of the t statistic show that with 11 degrees of freedom a value of 1.17 is not significant. Hence there is no evidence that one student's measurements are higher than the others.

The t-Test

Because of the relatively small number of observations made in some of our tests the t statistic was used in preference to the normal Z statistic. Frequently the question to be resolved was "could these two sets of observations have come from the same parent population?"

Where this was so the test statistic was calculated according to the formula

$$t = \frac{\bar{X}_A - \bar{X}_B}{\sqrt{\frac{S_A^2}{N_A} + \frac{S_B^2}{N_B}}}$$

where \bar{X}_A was the calculated mean of the N_A observations

\bar{X}_B was the calculated mean of the N_B observations

S_A was the standard deviation of the 'A' observations

S_B was the standard deviation of the 'B' observations

Furthermore since the variances S_A^2 and S_B^2 were only estimates of the variances of the entire population, a 'pooled sum of squares' method was used to estimate the number of degrees of freedom to be used with the t statistic. This was calculated from the formula

$$v = \frac{\left[\frac{S_A^2}{N_A} + \frac{S_B^2}{N_B} \right]^2}{\frac{\left[\frac{S_A^2}{N_A} \right]^2}{N_A + 1} + \frac{\left[\frac{S_B^2}{N_B} \right]^2}{N_B + 1}} \quad - 2$$

which although rather complex, caused no extra work since both the t statistic and the number of degrees of freedom were calculated by computer. (Had the calculations been made by hand then the F test would have been used to establish whether the variances S_A^2 and S_B^2 were sufficiently different to warrant calculation of v by the formula given in place of $N_A + N_B - 2$).

When the appropriate values of t and v had been ascertained then the tables were consulted to determine whether the test was significant or not. If the test was "is \bar{X}_A less than \bar{X}_B ?" or "is \bar{X}_A greater than \bar{X}_B ?" then a 'one tailed' test was performed whereas if the test was "is \bar{X}_A different from \bar{X}_B ?" then a two tailed test was performed. The only difference between these two tests is that the tables give areas (i.e. probabilities) for one tail only. Thus for two tailed tests the probabilities must be doubled.

APPENDIX 8.7

FORTRAN Programme 'DATAPLOT'

This programme was used for plotting the raw EOG/EEG data on a graphics terminal or graph plotter. The programme uses the GINO graphics subroutines. Listings of subroutines 'DATIN' and 'GETNAM' which are used to read the data in from a data file and to obtain a filename from the user are given in Appendix 8.12.

The listing of programme 'DATAPLOT' follows.

```

C          A PROGRAMME TO PLOT THE EOG / EEG DATA
C          ON THE TEKTRONIX 4010.
C
INTEGER*2 INP(1024),RNAME(20),IBATNO
LOGICAL CSE
DATA CSE/.FALSE./
1 WRITE(1,2)
2 FORMAT('TEKTRONIX(0) OR CALCOMP(1) OUTPUT')
C READ(1,*,ERR=1)IDEVIC
  NOMINATE THE REQUIRED DEVICE
  IF(IDEVIC.EQ.1)CALL CC906
  IF(IDEVIC.EQ.0)CALL T4010
  IBAUD=1200
  CALL DEVSPE(IBAUD)
  CALL UNITS(0.24)
  CALL GETNAM(RNAME)
18 WRITE(1,20)
20 FORMAT('WHICH BATCH')
22 READ(1,*,ERR=18)IBATNO
  IF(IBATNO.LT.0)GO TO 999
  IF(FLOAT(IBATNO/6).NE.FLOAT(IBATNO)/6.)GO TO 18
24 WRITE(1,27)
27 FORMAT('SCALE FACTOR')
  READ(1,*,ERR=24)SCALE
C      SET FOR SOLID LINES.
C      AND CLEAR THE SCREEN.
  CALL PICCLE
  CALL BROKEN(0)
  STX=150.
  X=STX
  STY=70.
  Y=0.
  CALL MOVTO2(X,Y)
  DO 200 I=1,6
  Y=Y+20.
  CALL MOVTO2(X,Y)
  Y=Y+100.
200 CALL LINTO2(X,Y)
  Y=70.
  DO 300 I=1,6
  X=STX
  CALL MOVTO2(X,Y)
  X=950.
  CALL LINTO2(X,Y)
300 Y=Y+120.
  KBAT=IBATNO
  CALL MOVTO2(40.,740.)
  CALL CHAHOL('DATA FROM FILE *.')
  CALL CHAARR(RNAME,20,2)
  CALL MOVTO2(0.,690.)
  CALL CHAHOL('VL EOG*.')
  CALL MOVTO2(20.,650.)
  CALL CHAINT(KBAT,3)
  CALL MOVTO2(0.,570.)
  CALL CHAHOL('VR EOG*.')
  KBAT=KBAT+1
  CALL MOVTO2(20.,530.)
  CALL CHAINT(KBAT,3)
  CALL MOVTO2(0.,450.)
  CALL CHAHOL('HL EOG*.')
  KBAT=KBAT+1
  CALL MOVTO2(20.,410.)
  CALL CHAINT(KBAT,3)
  CALL MOVTO2(0.,330.)
  CALL CHAHOL('HR EOG*.')
  KBAT=KBAT+1
  CALL MOVTO2(20.,290.)
  CALL CHAINT(KBAT,3)
  CALL MOVTO2(0.,210.)
  CALL CHAHOL('M1 EEG*.')
  KBAT=KBAT+1
  CALL MOVTO2(20.,170.)
  CALL CHAINT(KBAT,3)
  CALL MOVTO2(0.,90.)
  CALL CHAHOL('M2 EEG*.')
  KBAT=KBAT+1
  CALL MOVTO2(20.,50.)
  CALL CHAINT(KBAT,3)
  YD=600.
  LE=IBATNO

```

```

DO 400 L=1,6
Y=STY+YD
X=STX
CALL MOVTO2(X,Y)
CALL DATIN(LE,INP,RNAME,SF1,SF2,SAMRAT,CSE)
DO 350 I=1,1024
Y=FLOAT(INP(I))*SF1*SCALE
IF(L .GE. 5)Y=Y*SF2/SF1
Y=Y/10. + STY + YD
X=X+0.78125
350 CALL LINTO2(X,Y)
LE=LE+1
400 YD=YD-120.
CALL MOVTO2(0.,0.)
CALL CHAMOD
GO TO 22
999 CALL PICCLE
CALL DEVEND
CSE=.TRUE.
CALL DATIN(IBATNO,INP,RNAME,SF1,SF2,SAMRAT,CSE)
CALL EXIT
END

```

APPENDIX 8.8

FORTRAN Programme used in the Analysis of the Eye

Movement Correction Methods

This programme was used to calculate the correction constants for the four channel eye movement correction method. The calculated constants were printed and applied to the data to give the corrected EEG signal. The autocorrelation function of the corrected signal was subsequently calculated and plotted on a graphic terminal. Subroutine 'MES1' was then used to measure both the a.c.c. and the frequency of the a.c.f. by means of the graphics cursor which was set by the user to the appropriate points on the graph.

The autocorrelation function of the signal corrected by the method of McCallum and Walter (and stored as the sixth data channel) was finally calculated and plotted. Once again subroutine 'MES1' was used to estimate the a.c.c. and the frequency of the a.c.f.

The subroutines used are listed after the main programme. Although having similar names these subroutines are not necessarily the same as some of those listed in Appendix 8.12.

The programme listing follows:-

```

C      THIS PROGRAM IS INTENDED TO MINIMISE THE AMOUNT OF
C      E.O.G POWER IN THE E.E.G. IT USES THE MODIFIED
C      QUILTER TECHNIQUE. HORIZONTAL AND VERTICAL COMPONENTS
C      OF BOTH EYES ARE TAKEN INTO CONSIDERATION.
C      THE PROGRAM REMOVES ANY D.C OFFSET ON ANY OF
C      THE INPUT DATA CHANELS.
REAL M1M,M2M
REAL VL(1024),VR(1024),HL(1024),HR(1024),M1(1024),M2(1024)
DIMENSION X1(4,5),X2(4,5),RM1(4),RM2(4)
DIMENSION COR1(1024)
DIMENSION COVAR1(512),COVAR2(512)
INTEGER*2 BATNO,INP(1024),FNAME(7)
LOGICAL CSE
CSE=.FALSE.
N=1024
NCORL=512
C      READ THE NUMBER OF THE FIRST BATCH OF DATA
C      TO BE PROCESSED.
C      THE DATA IS ASSUMED TO BE IN THE FOLLOWING
C      ORDER ; VL , VR , HL , HR , M1 , M2
C      M1 IS THE CHANNEL TO BE CORRECTED BY THE
C      MODIFIED QUILTER TECHNIQUE.
C      M2 IS THE CHANNEL CORRECTED BY THE BURDEN TECHNIQUE.
C      GET THE SCALE FACTORS FROM THE DATA FILE.
C      SF1 FOR EOG DATA. SF2 FOR EEG DATA.
5 WRITE(1,6)
6 FORMAT('4 CHANNEL EYE MOVEMENT CORRECTION PROGRAM')
CALL GETNAM(FNAME)
WRITE(1,10)
10 FORMAT('WHICH BATCH OR -1 TO QUIT')
READ(1,*)BATNO
IF(BATNO .LT. 0)GO TO 5000
C      CHECK THAT THE BATCH NUMBER IS VALID.
IF(BATNO .GT. 191)GO TO 2999
IF(FLOAT(BATNO/6) .NE. FLOAT(BATNO)/6.)GO TO 2999
WRITE(1,14)
14 FORMAT('FILTER THE BURDEN SIGNAL(0) NO (1) YES')
READ(1,*)IBRDFL
WRITE(1,16)
16 FORMAT('FILTER THE CORRECTED SIGNAL(0) NO (1) YES')
READ(1,*)ICRRFL
WRITE(1,18)
18 FORMAT('PLOT THE CORRECTED WAVEFORM')
READ(1,*)IPLTCR

20 WRITE(1,22)BATNO
22 FORMAT('EYE MOVEMENT CORRECTIONS 4 CHAN. BATCH',I5)
WRITE(1,24)FNAME
24 FORMAT('DATA FILE ',7A2,///)
L=BATNO
C      READ THE DATA AND CONVERT TO REAL FORMAT.
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 32 I=1,N
32 VL(I)=FLOAT(INP(I))*SF1
L=L+1
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 34 I=1,N
34 VR(I)=FLOAT(INP(I))*SF1
L=L+1
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 36 I=1,N
36 HL(I)=FLOAT(INP(I))*SF1
L=L+1
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 38 I=1,N
38 HR(I)=FLOAT(INP(I))*SF1
L=L+1
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 40 I=1,N
40 M1(I)=FLOAT(INP(I))*SF2
L=L+1
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 52 I=1,N
52 M2(I)=FLOAT(INP(I))*SF2
WRITE(1,56)SF1,SF2
56 FORMAT('SF1= ',F8.6,' SF2= ',F8.6)
C      SUBTRACT THE MEAN OF EACH DATA BATCH FROM THE DATA.
CALL SMEAN(N,VL,VLM)
CALL SMEAN(N,VR,VRM)
CALL SMEAN(N,HL,HLM)

```

```

CALL SMEAN(N,HR,HRM)
CALL SMEAN(N,M1,M1M)
C      FORM THE CORRELATION SUMS OF PRODUCTS.
A=0.
B=0.
C=0.
D=0.
E=0.
F=0.
G=0.
P=0.
Q=0.
R=0.
S=0.
T=0.
U=0.
V=0.
DO 100 I=1,N
A=A+VL(I)*VL(I)
B=B+VL(I)*VR(I)
C=C+VL(I)*HL(I)
D=D+VL(I)*HR(I)
E=E+VR(I)*VR(I)
F=F+VR(I)*HL(I)
G=G+VR(I)*HR(I)
P=P+HL(I)*HL(I)
Q=Q+HL(I)*HR(I)
R=R+HR(I)*HR(I)
S=S+M1(I)*VL(I)
T=T+M1(I)*VR(I)
U=U+M1(I)*HL(I)
V=V+M1(I)*HR(I)
100 CONTINUE
X1(1,1)=A
X1(1,2)=B
X1(1,3)=C
X1(1,4)=D
X1(2,2)=E
X1(2,3)=F
X1(2,4)=G
X1(3,3)=P
X1(3,4)=Q
X1(4,4)=R
X1(1,5)=S
X1(2,5)=T
X1(3,5)=U
X1(4,5)=V
C      SET UP SYMMETRICAL MATRIX.
X1(2,1)=X1(1,2)
X1(3,1)=X1(1,3)
X1(4,1)=X1(1,4)
X1(3,2)=X1(2,3)
X1(4,2)=X1(2,4)
X1(4,3)=X1(3,4)
C      SOLVE THE SIMULTANEOUS EQUATIONS BY THE
C      GAUSS PIVOTAL METHOD.
MSIZE=4
CALL GAUSS(MSIZE,X1,RM1)
IF(IBRDFL.EQ.1)CALL FILTER(N,M2)
WRITE(1,150)
150 FORMAT(' THE CORRECTION CONSTANTS ARE ',//)
WRITE(1,160)
160 FORMAT(10X,'VL * ',10X,'VR * ',10X,'HL * ',10X,'HR * ')
WRITE(1,200)(RM1(I),I=1,4)
200 FORMAT(4(3X,F12.5))
C      APPLY THE CORRECTIONS TO THE DATA.
DO 220 I=1,N
COR1(I)=M1(I) - (RM1(1)*VL(I) + RM1(2)*VR(I) + RM1(3)*HL(I)
1+ RM1(4)*HR(I))
220 CONTINUE
C      PLOT THE CORRECTED DATA.
C      CALCULATE AND PRINT THE CROSS CORRELATIONS
IF(ICRRFL.EQ.1)CALL FILTER(N,COR1)
IF(IPLTCR.EQ.1)CALL GRAPH3(N,COR1,SAMRAT)
SC=200.
CALL AUTCO(N,COR1,NCORL,COVAR1)
CALL GRAPH3(NCORL,COVAR1,SAMRAT)
WRITE(1,240)
240 FORMAT(' MEASURE(0) NO (1) YES')
READ(1,*)IMES

```

```
IF(IMES .EQ. 1)CALL MES1
CALL AUTO(N,M2,NCORL,COVAR2)
CALL TNOUA('WAITING, BURDEN ACF NEXT',24)
READ(1,*)IWT
CALL GRAPH3(NCORL,COVAR2,SAMRAT)
WRITE(1,240)
READ(1,*)IMES
IF(IMES .EQ. 1)CALL MES1
GO TO 5
2999 WRITE(1,4000)BATNO
4000 FORMAT(' BATCH NUMBER INCORRECT',I6)
5000 CSE=.TRUE.
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
CALL DEVEND
CALL EXIT
END
```

SUBROUTINE MES1

C
C
CMEASURES THE ACC & FREQUENCY
BY MEANS OF THE GRAPHICS CURSOR.WRITE(1,10)
10 FORMAT('GIVE CO-ORDS. FOR ACF')CENLIN=0.
XM=400.
XS=100.
YS=390.
CALL CURSOR(ICOM,X,Y)
X=X-XS
Y=Y-YS
CALL CURSOR(ICOM,X1,Y1)
X1=X1-XS
Y1=Y1-YS
S1=(Y-Y1)/(X-X1)
C1=Y1-S1*X1C
C

DRAW FIRST CONSTRUCTION LINE

XL1=50.
XL2=750.
YL1=S1*XL1 + C1
CALL MOVTO2(XL1+XS,YL1+YS)
YL2=S1*XL2 + C1
CALL LINTO2(XL2+XS,YL2+YS)
CALL CURSOR(ICOM,X,Y)
X=X-XS
Y=Y-YS
CALL CURSOR(ICOM,X1,Y1)
X1=X1-XS
Y1=Y1-YS
S2=(Y-Y1)/(X-X1)
C2=Y1-S2*X1C
C

DRAW SECOND CONSTRUCTION LINE

YL1=S2*XL1 + C2
CALL MOVTO2(XL1+XS,YL1+YS)
YL2=S2*XL2 + C2
CALL LINTO2(XL2+XS,YL2+YS)

C

A=(ATAN(S1)+ATAN(S2))/2.
S=SIN(A)/COS(A)
C=(C1+C2)/2.
IF(CENLIN .EQ. 1.)CALL MOVTO2(XL1+XS,YS+S*XL1+C)
IF(CENLIN .EQ. 1.)CALL LINTO2(XL2+XS,YS+S*XL2+C)
TEST FOR ZERO SLOPEC
CIF(S .EQ. 0.)GO TO 100
S3=-1./S
YM=S*XM+C
C3=YM-S3*XM
XP1=(C1-C3)/(S3-S1)
XP2=(C2-C3)/(S3-S2)
YP1=S3*XP1 + C3
YP2=S3*XP2 + C3
DY=SQRT((XP1-XP2)*(XP1-XP2) + (YP1-YP2)*(YP1-YP2))/400.
GO TO 200100 YP1=S1*XM + C1
YP2=S2*XM + C2
DY=ABS(YP1-YP2)/400.
200 CALL MOVTO2(0.,700.)
CALL CHAMODWRITE(1,210)
210 FORMAT('GIVE CO-ORDS. FOR TIME MEASUREMENT')CALL CURSOR(ICOM,X,Y)
CALL CURSOR(ICOM,X1,Y1)
DX=(ABS(X-X1)/800.)*4.096
F=1./DX
CALL MOVTO2(0.,650.)
CALL CHAMOD
WRITE(1,240)DY,F
240 FORMAT('ACF = ',F8.4,10X,'FREQ = ',F9.6)
RETURN
ENDC
C

SUBROUTINE GETNAM(NAME)

GETS A FILENAME FROM THE USER

```

C
  INTEGER*2 NAME(7),TBUF(7)
  DATA TBUF / 'NO','NA','ME','GI','VE','N','E' /
6  WRITE(1,8)
8  FORMAT('GIVE NAME OF FILE TO BE PROCESSED')
  READ(1,10)NAME
10 FORMAT(7A2)
  IF(NAME(1) .EQ. ' ')GO TO 100
  DO 20 I=1,7
20  TBUF(I)=NAME(I)
  RETURN
100 DO 120 I=1,7
120 NAME(I)=TBUF(I)
  RETURN
  END

```

```

C
C
C
  SUBROUTINE AUTCO(NPOINT,DATA,NCOREL,AUTOCF)
      CALCULATES THE ACF
  DIMENSION DATA(NPOINT),AUTOCF(NCOREL),DATAZM(1024)
  INTEGER*2 Z1
  AMEAN=0.
  DO 10 I=1,NPOINT
10  AMEAN=AMEAN+DATA(I)
  AMEAN=AMEAN/FLOAT(NPOINT)
  DO 15 I=1,NPOINT
15  DATAZM(I)=DATA(I)-AMEAN
  DO 25 I=1,NCOREL
  II=I-1
  STORE=0.
  Z1=NPOINT-II
  DO 20 J=1,Z1
20  STORE=STORE+DATAZM(J)*DATAZM(J+II)
25  AUTOCF(I)=STORE
  STORE=ABS(AUTOCF(I))
  DO 30 I=1,NCOREL
30  AUTOCF(I)=AUTOCF(I)/STORE
  RETURN
999 WRITE(1,40)NPOINT,NCOREL
40  FORMAT('ERROR IN NUMBER OF CORRELATIONS',2I8)
  STOP
  END

```

```

C
C
C
  SUBROUTINE SMEAN(NPTS,DATA,RMEAN)
      SUBTRACTS THE MEAN VALUE FROM THE DATA.
  DIMENSION DATA(NPTS)
  RMEAN=0.
  DO 20 I=1,NPTS
20  RMEAN=RMEAN+DATA(I)
  RMEAN=RMEAN/NPTS
  DO 30 I=1,NPTS
30  DATA(I)=DATA(I)-RMEAN
  RETURN
  END

```

```

C
C
C
  SUBROUTINE FILTER(NPTS,XT)
      LOW PASS FILTER THE DATA.
  DIMENSION XT(NPTS),ZA(11),ZB(21),DATOUT(1024)
  ZA(1)=0.03125
  ZA(2)=0.
  ZA(3)=-0.09375
  ZA(4)=0.
  ZA(5)=0.3125
  ZA(6)=0.5
  ZA(7)=0.3125
  ZA(8)=0.
  ZA(9)=-0.09375
  ZA(10)=0.
  ZA(11)=0.03125
  ZB(1)=0.03125
  ZB(2)=0.
  ZB(3)=0.
  ZB(4)=0.
  ZB(5)=-0.09375
  ZB(6)=0.

```

```

ZB(7)=0.
ZB(8)=0.
ZB(9)=0.3125
ZB(10)=0.
ZB(11)=0.5
ZB(12)=0.
ZB(13)=0.3125
ZB(14)=0.
ZB(15)=0.
ZB(16)=0.
ZB(17)=-0.09375
ZB(18)=0.
ZB(19)=0.
ZB(20)=0.
ZB(21)=0.03125
M=11
DO 21 I=1,NPTS
J=M
STORE=0.
IF(M .LT. I)GO TO 19
STORE=XT(I)
GO TO 21
19 DO 20 K=1,J
20 STORE=STORE + XT(I-K+1)*ZA(K)
21 DATOUT(I)=STORE
DO 30 I=1,NPTS
30 XT(I)=DATOUT(I)
M=21
DO 41 I=1,NPTS
J=M
STORE=0.
IF(M .LT. I)GO TO 39
STORE=XT(I)
GO TO 41
39 DO 40 K=1,J
40 STORE=STORE + XT(I-K+1)*ZB(K)
41 DATOUT(I)=STORE
DO 50 I=1,NPTS
50 XT(I)=DATOUT(I)
RETURN
END

```

```

SUBROUTINE GAUSS(K,B,X)
C          A SUBROUTINE TO SOLVE SIMULTANEOUS
C          EQUATIONS BY THE GAUSS PIVOTAL METHOD
DIMENSION A(4,5),X(4),B(4,5)
INTEGER*2 Z1,Z2,Z3,Z4,Z5,Z6,Z7
Z1=K+1
DO 10 I=1,K
DO 10 J=1,Z1
10 A(I,J)=B(I,J)
C          SAVE INPUT DATA
Z2=K-1
DO 35 I=1,Z2
L=I
DO 15 J=I,K
15 IF(ABS(A(L,I)) .LT. ABS(A(J,I)))L=J
IF(ABS(A(L,I)) .EQ. 0.)GO TO 60
IF(L .EQ. I)GO TO 21
Z3=K+1
DO 20 N=I,Z3
SAVE=A(I,N)
A(I,N)=A(L,N)
20 A(L,N)=SAVE
C          PIVOTAL REDUCTION
21 Z4=I+1
DO 35 M=Z4,K
D=A(M,I)/A(I,I)
Z5=I+1
Z6=K+1
DO 35 J=Z5,Z6
35 A(M,J)=A(M,J)-D*A(I,J)
C          BACK SUBSTITUTION
DO 50 L=1,K
J=K+1-L
IF(J .EQ. K)GO TO 45
Y=A(J,K+1)
Z7=K-1
DO 40 M=J,Z7
40 Y=Y-A(J,M+1)*X(M+1)

```

```

X(J)= Y/A(J,J)
IF(J .NE. K)GO TO 50
45 X(J)=A(J,K+1)/A(J,J)
50 CONTINUE
RETURN
60 WRITE(1,65)
65 FORMAT(' ERROR MESSAGE ZERO COLUMN FOUND')
STOP
END

```

SUBROUTINE DATIN(IBATNO, IDATA, RNAME, SF1, SF2, SAMRAT, CSE)

C
C
C

GETS DATA FROM THE DATA FROM SPECIFIED
DATA FILE.

§INSERT SYSCOM>ASKEYS

```

INTEGER*2 IDATA(1024), ONAME(7), RNAME(7), TITLE(36), NME(6)
INTEGER*2 RWKEY, NLEN, NLEN2, PRIMNO, NBAT, IBATNO, IAA1(4)
LOGICAL OPEN, NEOPEN, CSE
DATA NEOPEN /.FALSE./
RWKEY=1
PRIMNO=1
NLEN=14
NLEN2=NLEN/2
IF(NEOPEN)GO TO 1300
IF(CSE)RETURN
900 OPEN=OPEN$(RWKEY, RNAME, NLEN, PRIMNO)
IF(.NOT. OPEN)GO TO 1700
DO 1000 IC=1, NLEN2
1000 ONAME(IC)=RNAME(IC)
NEOPEN=.TRUE.
READ(5, 3000, END=1800, ERR=1900)NME, SF1, SF2
READ(5, 3010, END=1800, ERR=1900)MAXBAT
READ(5, 3020, END=1800, ERR=1900)TITLE
DO 234 IL=1, 4
234 IAA1(IL)=TITLE(IL+20)
DECODE(8, 236, IAA1)SAMRAT
236 FORMAT(F8.4)
1100 READ(5, END=1800, ERR=1900)NBAT
READ(5, END=1800, ERR=1900)(IDATA(I), I=1, 1024)
IF(NBAT .NE. IBATNO)GO TO 1100
NBAT=NBAT+1
RETURN
1200 OPEN=CLOSS$(PRIMNO)
IF(.NOT. OPEN)GO TO 1600
NEOPEN=.FALSE.
RETURN
1300 IF(CSE)GO TO 1200
DO 1400 IC=1, NLEN2
1400 IF(ONAME(IC) .NE. RNAME(IC))GO TO 1500
IF(IBATNO .GE. NBAT)GO TO 1100
1500 OPEN=CLOSS$(PRIMNO)
IF(OPEN)GO TO 900
1600 WRITE(1, 1610)ONAME
1610 FORMAT(' *** CANT CLOSE FILE ', 7A2, ' ***')
STOP 1
1700 WRITE(1, 1710)RNAME
1710 FORMAT(' *** CANT OPEN FILE ', 7A2, ' ***')
STOP 2
1800 WRITE(1, 1810)RNAME
1810 FORMAT(' *** END OF FILE ', 7A2, ' ***')
STOP 3
1900 WRITE(1, 1910)IBATNO, RNAME
1910 FORMAT(' *** ERROR TRYING TO READ BATCH ', I5, ' FROM FILE ', 7A2,
+ ' ***')
STOP 4
3000 FORMAT(6A2, 2F8.6)
3010 FORMAT(I4)
3020 FORMAT(36A2)
END

```

SUBROUTINE GRAPH3(N, DATA, SAMRAT)

C
C
C

PLOTS A GRAPH ON THE REQUESTED GRAPHICS DEVICE.

```

DIMENSION DATA(N), XARRY(1024), YARRY(1024)
INTEGER*2 IYLAB(7), IYLAB2(9)
LOGICAL INIT
DATA INIT/.FALSE./
DATA IYLAB / 'MI', 'CR', 'O-', 'VO', 'LT', 'S ', '*.' /

```

```

DATA IYLAB2 / 'AU', 'TO', 'CO', 'RR', 'EL', 'AT', 'IO', 'NS', '*.' /
IACF=0
IF(INIT)GO TO 50
CALL T4010
IBAUD=1200
CALL DEVSPE(IBAUD)
CALL UNITS(0.24)
50 CALL PICCLE
CALL WINDOW(0)
CALL CHASIZ(15.,15.)
DO 60 I=1,N
XARRY(I)=FLOAT(I)/SAMRAT
YARRY(I)=DATA(I)
60 CONTINUE
CALL MOVTO2(450.,0.)
CALL CHAHOL('TIME SECONDS*.')
CALL MOVTO2(50.,300.)
CALL CHAANG(90.)
IF(IACF .EQ. 0)GO TO 100
CALL CHAARR(IYLAB2,9,2)
GO TO 150
100 CALL CHAARR(IYLAB,7,2)
150 CALL CHAANG(0.)
CALL GRAF(XARRY,YARRY,N,0)
CALL MOVTO2(0.,779.)
CALL CHAMOD
INIT=.TRUE.
RETURN
END

```

APPENDIX 8.9

Special Instructions for Peripheral Control

The following instructions were used to control the peripheral devices connected to the PDP8 minicomputer. Since these peripherals were not part of the standard PDP8 equipment, the instructions controlling them are described in detail in this appendix.

OCTAL CODE	MNEMONIC	DESCRIPTION
6056	OUTX	Load the 10 least significant bits of the accumulator into the X 10 bit D/A converter and convert to analogue. Pulse the Z modulation output. This instruction was used for X-Y display of stored data.
6066	OUTY	As above but for the Y D/A converter.
6412	SION*	Enable the interrupt facility from the hybrid computer interface Skip/Interrupt inputs.
6414	SIOF*	Disable the interrupt facility from the the hybrid computer interface Skip/Interrupt inputs.
6441	ADINP*	Load the 12 bit word from the analogue to digital converter into the accumulator.
6442	STC *	Start the conversion process on the 12 bit A/D converter. (Conversion takes approximately 3.5 μ s)
6451	MXR *	Set the multiplexer to select the channel number given by the 3 least significant bits of the accumulator. (i.e. channels 0-7).
6501	ACOUT*	Load the accumulator into the output register available on the hybrid computer Patch Panel.

OCTAL CODE	MNEMONIC	DESCRIPTION
6504	ACIN*	Load the accumulator from the input register on the hybrid computer interface Patch Panel.
6511	SK1*	Skip the next instruction if the Patch Panel number one input is at a logical '1'.
6221	SDR	Skip the next instruction if the serial data transceiver has received a byte of data.
6622	DUI	Disable interrupts from the serial data transceiver.
6624	STR	Skip the next instruction if the serial data transceiver is ready to transmit a new data byte.
6631	RUD	Read the received data from the serial data transceiver into the 8 ⁺¹ least significant bits of the accumulator. Clear the SDR flag.
6632	ERI	Enable the received data interrupt facility on the serial data transceiver (i.e. allow interrupts to occur when serial data is received).
6634	LSTAT	Load the 5 least significant bits of the accumulator into the serial data transceiver status control register ⁺² .
6641	SKERR	Skip the next instruction if either a framing, parity or over-run error has occurred in receiving serial data.
6642	ETI	Enable the transmitter interrupt facility on the serial data transceiver. (i.e. allow interrupts to occur whenever the serial data transmitter is not busy).
6644	OUT	Load the 8 ⁺¹ least significant bits of the accumulator into the serial data transmitter and transmit. Clear the STR flag.

Notes

- * These peripherals are part of the Hybrid Computer Interface [see Section 4 reference 3].
- ⁺¹ May be less than 8 bits. This depends on the setting of the Universal Asynchronous Receiver Transmitter status control register.
- ⁺² The 5 control bits have the following functions

ACCUMULATOR
BIT

FUNCTION

11 Parity inhibit. A '1' disables the generation and checking of the parity bit.

10 Stop bit select. A '1' gives two stop bits. A '0' gives one stop bit.

9 Character length select. Bits 8 and 9 allow characters of either 5,6,7 or 8 bits to be transmitted and received.

8	bit	8	9	
		0	0	5 bits
		0	1	6 bits
		1	0	7 bits
		1	1	8 bits

7 Even parity enable. A logical '1' selects even parity. A logical '0' selects odd parity. (Subject to bit 11 being a logical '1'.

APPENDIX 8.10

PAL 8 Computer Programme Used to Control the Data Acquisition Process

This programme was used to control the A-D converter, the multiplexer, the serial interface, the stimulus presentation and an X-Y display during the acquisition of the EEG data. The programme was started and controlled by means of the console switches. The starting address is 200_8 in field zero. Two locations (131_8 and 132_8) must be set, by means of the console switch register, to contain the two's complement of the sample numbers at which time the stimulus pulses are to be presented.

e.g. If the S1 stimulus is to be presented after 200_{10} samples have been taken, then location 131_8 must be set to contain 7470_8 . ($200_{10} \equiv 310_8$)

Whilst running the programme will show on an X-Y display whichever channel is selected by the three least significant console switches.

The assembly language listing follows.

/ ON LINE CNV DATA AQUISITION PROGRAM

/-----
 /
 / THIS PROGRAM CONTROLS
 / THE MULTIPLEXER, THE A-D CONVERTER
 / THE S1 AND S2 SIGNALS AND THE SERIAL
 / DATA OUTPUT DURING THE ACQUISITION OF 6
 / CHANNELS OF ANALOGUE DATA.
 /

/ MEMORY MAP

/ FIELD ZERO

/-----
 / 0000 TO 2777 AREA FOR PROGRAM.
 / 3000 TO 3577 SOFTWARE STACK AREA
 / 3600 TO 5577 CHANNEL 4
 / 5600 TO 7577 CHANNEL 5
 / 7600 TO 7777 DEC SOFTWARE (RIM & BIN)

/ FIELD ONE

/-----
 / 0000 TO 1777 CHANNEL 0
 / 2000 TO 3777 CHANNEL 1
 / 4000 TO 5777 CHANNEL 2
 / 6000 TO 7777 CHANNEL 3
 /-----

00000 0000 *0
 00001 5440
 0020 *20

0
 JMP I ISERV

/ SUBROUTINE POINTERS

00020 0517 IUART, UARTTX
 00021 1101 ITTYRX, TTYRX
 00022 1046 ITTYX, TTYX
 00023 0464 ITMR, TIMER
 00024 1000 ISAM, SAMPLE
 00025 1070 ICVRT, CVRT
 00026 1034 IMPXR, MPXR
 00027 0600 IINC, INCR1
 00030 0436 IRET, RETN
 00031 0632 IPOIN1, POIN1
 00032 0673 IPOIN2, POIN2
 00033 0615 IINCR2, INCR2
 00034 1102 IRSTIN, RSTIN
 00035 1121 IRSTOU, RSTOU
 00036 1047 INTBL, WTBL
 00037 1061 IRNDWT, RNDWT
 00040 0400 ISERV, SERV
 00041 0421 ISTRT, STRT
 00042 1200 IDISP, DISPLA
 00043 1157 IINC3, INC3
 00044 1232 IPOIN3, POIN3
 00045 1140 IRSTD, RSTD

/ CONSTANTS ETC.

00046	0000	TAC,	0	
00047	0000	TFG,	0	
00050	0000	PRIDR,	0	
00051	0000	P0,	0	
00052	2000	P1,	2000	
00053	4000	P2,	4000	
00054	6000	P3,	6000	
00055	3600	P4,	3600	
00056	5600	P5,	5600	
00057	0000	ZEROP,	0	
00060	0000	ONEP,	0	
00061	0000	TWOP,	0	
00062	0000	THREEP,	0	
00063	0000	FOURP,	0	
00064	0000	FIVEP,	0	
00065	0000	D0P,	0	
00066	0000	D1P,	0	
00067	0000	D2P,	0	
00070	0000	D3P,	0	
00071	0000	D4P,	0	
00072	0000	D5P,	0	
00073	0000	D0P,	0	
00074	0000	D1P,	0	
00075	0000	D2P,	0	
00076	0000	D3P,	0	
00077	0000	D4P,	0	
00100	0000	D5P,	0	
00101	0000	SAVE,	0	
00102	0000	MADD,	0	
00103	0000	DUCHN,	0	
00104	0000	CURPNT,	0	
00105	0000	CURDP,	0	
00106	0000	SWSP,	0	
00107	4400	SWSPL,	4400	
00110	0001	D1,	1	
00111	0003	D3,	3	
00112	0007	D7,	7	
00113	7777	M1,	7777	
00114	7776	M2,	7776	
00115	7775	M3,	7775	
00116	7774	M4,	7774	
00117	7773	M5,	7773	
00120	6000	NOSAMS,	6000	
00121	7772	LIMIT,	7772	
00122	0000	TRIAL,	0	
00123	7740	TRIALS,	7740	
00124	0000	SW,	0	
00125	0026	STATUS,	26	
00126	3000	SP,	3000	
00127	0052	STCODE,	52	
00130	0001	LGT,	1	
00131	7604	MS1,	7604	
00132	6226	MS2,	6226	
00133	0003	S1,	3	
00134	0005	S2,	5	
00135	0000	DISNUM,	0	
00136	0000	XDIS,	0	
00137	0000	CPTR,	0	
	0200	*200		
00200	6002	START,	IDF	/ INITIALIZE CONSTANTS ETC.
00201	6622		DUI	
00202	6414		SIOF	
00203	4434		JMS I IRSTIN	
00204	4435		JMS I IRSTOU	
00205	7300		CLA CLL	
00206	5102		DCA MADD	
00207	3050		DCA PRIDR	
00210	3122		DCA TRIAL	
00211	3124		DCA SW	
00212	3103		DCA DUCHN	
00213	1126		TAD SP	
00214	3106		DCA SWSP	
00215	1125		TAD STATUS	

00216	6634		LSTAT	
00217	6631		RUD	/ DUMMY READ TO CLEAR UART BUFFER.
00220	6631		RUD	
00221	6030		KCF	
00222	6042		TCF	
00223	4436		JMS I IWTBL	
00224	4437		JMS I IRNDWT	
00225	6412		SION	
00226	6642		ETI	
00227	6001		ION	
00230	5442	WAIT,	JMP I IDISP	
00231	7200	TRIEND,	CLA	
00232	1126		TAD SP	
00233	3105		DCA SWSP	
00234	3102		DCA MADD	
00235	3103		DCA OUCHN	
00236	4434		JMS I IRSTIN	
00237	4435		JMS I IRSTOU	
00240	2122		ISZ TRIAL	
00241	1122		TAD TRIAL	
00242	1123		TAD TRIALS	
00243	7650		SNA CLA	
00244	5254		JMP FINSH	
00245	4436		JMS I IWTBL	
00246	4437		JMS I IRNDWT	
00247	7300		CLA CLL	
00250	3050		DCA PRIOR	
00251	6412		SION	
00252	6642		ETI	
00253	5442		JMP I IDISP	
00254	6002	FINSH,	IOF	
00255	5442		JMP I IDISP	
	0400			
00400	3506	*400	DCA I SWSP	/ THIS SECTION SAVES THE
00401	6004	SERV,	GTF	/ AC, LINK AND STATUS.
00402	2106		ISZ SWSP	
00403	3506		DCA I SWSP	
00404	2106		ISZ SWSP	
00405	1000		TAD 0	/ GET RETURN ADDRESS
00406	3506		DCA I SWSP	
00407	2106		ISZ SWSP	
00410	1106		TAD SWSP	
00411	1107		TAD SWSPL	
00412	7700		SMA CLA	
00413	7402		HLT	
00414	1050		TAD PRIOR	
00415	1041		TAD ISTRT	
00416	3220		DCA .+2	
00417	5620		JMP I .+1	
00420	0000		0	
00421	6511	STRT,	SK1	/ SKIP ON TIMER
00422	5224		JMP .+2	
00423	5423		JMP I ITIMR	
00424	6624		STR	/ SKIP ON UART TX
00425	5227		JMP .+2	
00426	5420		JMP I IUARTT	
00427	6031		KSF	
00430	5232		JMP .+2	
00431	5421		JMP I ITTYRX	
00432	6041		TSF	
00433	5235		JMP .+2	
00434	5422		JMP I ITTYTX	
00435	7402		HLT	/ SHOULD NOT GET HERE !!
			/	
			/	/ THIS SECTION RESTORES THE AC LINK
			/	/ AND RETURNS TO THE INTERRUPTED
			/	/ SECTION.
			/	
00436	6002	RETN,	IOF	
00437	6203		CDIO	
00440	7340		CLA CLL CMA	
00441	1106		TAD SWSP	
00442	3106		DCA SWSP	

```

00443 1506 TAD I SWSP
00444 3000 DCA 0
00445 7340 CLA CLL CMA
00446 1106 TAD SWSP
00447 3106 DCA SWSP
00450 1506 TAD I SWSP
00451 3047 DCA TFG
00452 7340 CLA CLL CMA
00453 1106 TAD SWSP
00454 3106 DCA SWSP
00455 1506 TAD I SWSP
00456 3046 DCA TAC
00457 1047 TAD TFG
00460 6005 RTF
00461 7200 CLA
00462 1046 TAD TAC
00463 5400 JMP I 0
/
/ THIS SECTION SAMPLES AND STORES THE
/ DATA IN THE CORE BUFFER.
/
00464 7300 TIMER, CLA CLL
00465 6451 MXR / SET MULTIPLEXER TO CHAN. 0
00466 4424 JMS I ISAM / SEND S1 OR S2 IF NECESSARY
00467 4425 NXCHNL, JMS I ICVRT / CONVERT CURRENT CHAN.
00470 3101 DCA SAVE
00471 7327 CLA CLL CML IAC RTL
00472 6451 MXR / GROUND MULTIPLEXER INPUT TO
00473 6442 STC / MINIMISE CROSSTALK.
00474 7300 CLA CLL
00475 1102 TAD MADD
00476 4431 JMS I IPDINI
00477 3104 DCA CURPNT
00500 1101 TAD SAVE
00501 3504 DCA I CURPNT
00502 4426 JMS I IMPXR
00503 7620 SNL CLA
00504 5267 JMP NXCHNL
00505 7327 CLA CLL CML IAC RTL
00506 6451 MXR
00507 4427 JMS I IINC
00510 7420 SNL
00511 5430 JMP I IRETN
00512 6414 SIOF / TAKEN ALL SAMPLES SO DISABLE
00513 7300 CLA CLL / TIMER INTERRUPT.
00514 1111 TAD 03
00515 3050 DCA PRIOR
00516 5430 JMP I IRETN
/
/ THIS SECTION TAKES DATA FROM THE
/ CORE BUFFER AND TRANSMITS IT OVER
/ THE SERIAL DATA LINE.
/
00517 6622 UARTTX, DUI
00520 6001 ION
00521 7300 CLA CLL / FIRST OR SECOND SIX BITS ?
00522 1124 TAD SW
00523 7640 SZA CLA
00524 5346 JMP SECND
00525 1057 TAD ZEROP / WAIT TILL AT LEAST ONE SAMPLE
00526 7450 SNA / HAS BEEN TAKEN.
00527 5343 JMP TXEND
00530 7041 CIA
00531 1065 TAD DOP / HAVE ALL THE SAMPLES SO FAR
00532 7700 SMA CLA / TAKEN BEEN SENT ?
00533 5343 JMP TXEND
00534 1103 TAD OUCHN
00535 4432 JMS I IPDIN2
00536 3105 DCA CUROP
00537 1505 TAD I CUROP
00540 6644 OUT

```

00541	7201		CLA IAC	/ SET SWITCH FOR 2nd SIX BITS.
00542	3124		DCA SW	
00543	6002	TXEND,	IOF	
00544	6642		ETI	
00545	5430		JMP I IRETN	
00546	3124	SECND,	DCA SW	/ RESET SWITCH.
00547	1103		TAD OUCHN	
00550	4432		JMS I IPOIN2	
00551	3105		DCA CUROP	
00552	1505		TAD I CUROP	
00553	7002		BSW	
00554	6644		OUT	
00555	7300		CLA CLL	
00556	2103		ISZ OUCHN	
00557	1103		TAD OUCHN	
00560	1121		TAD LIMIT	
00561	7710		SPA CLA	
00562	5343		JMP TXEND	
00563	3103		DCA OUCHN	
00564	4433		JMS I IINCR2	
00565	7420		SNL	
00566	5343		JMP TXEND	
00567	5777		JMP TRIEND	
00577	0231			
	0600	*600		
00500	0000	INCR1,	0	
00601	7300		CLA CLL	
00602	2057		ISZ ZEROP	
00603	2060		ISZ ONEP	
00604	2061		ISZ TWOP	
00605	2062		ISZ THREEP	
00606	7000		NOP	
00607	2063		ISZ FOURP	
00610	2064		ISZ FIVEP	
00611	1057		TAD ZEROP	
00612	1120		TAD NOSAMS	
00613	7200		CLA	
00614	5600		JMP I INCR1	
00615	0000	INCR2,	0	
00616	7300		CLA CLL	
00617	2065		ISZ OOP	
00620	2066		ISZ 01P	
00621	2067		ISZ 02P	
00622	2070		ISZ 03P	
00623	7000		NOP	
00624	2071		ISZ 04P	
00625	2072		ISZ 05P	
00626	1055		TAD OOP	
00627	1120		TAD NOSAMS	
00630	7200		CLA	
00631	5615		JMP I INCR2	
00632	0000	POIN1,	0	
00633	7041		CIA	
00634	6211		CDF+10	
00635	7450		SNA	
00636	5271		JMP ZERO	
00637	7001		IAC	
00640	7450		SNA	
00641	5267		JMP ONE	
00642	7001		IAC	
00643	7450		SNA	
00644	5265		JMP TWO	
00645	7001		IAC	
00646	7450		SNA	
00647	5263		JMP THREE	
00650	6201		CDF+0	
00651	7001		IAC	
00652	7450		SNA	
00653	5261		JMP FOUR	
00654	7001		IAC	
00655	7440		SZA	
00656	7402		HLT	
00657	1064		TAD FIVEP	

00660	5632		JMP I POIN1
00661	1063	FOUR,	TAD FOURP
00662	5632		JMP I POIN1
00663	1062	THREE,	TAD THREEP
00664	5632		JMP I POIN1
00665	1061	TWO,	TAD TWOP
00666	5632		JMP I POIN1
00667	1060	ONE,	TAD ONEP
00670	5632		JMP I POIN1
00671	1057	ZERO,	TAD ZEROP
00672	5632		JMP I POIN1
00673	0000	POINZ,	0
00674	7041		CIA
00675	6211		CDF+10
00676	7450		SNA
00677	5332		JMP ZER
00700	7001		IAC
00701	7450		SNA
00702	5330		JMP 0
00703	7001		IAC
00704	7450		SNA
00705	5326		JMP T
00706	7001		IAC
00707	7450		SNA
00710	5324		JMP TH
00711	6201		CDF+0
00712	7001		IAC
00713	7450		SNA
00714	5322		JMP F
00715	7001		IAC
00716	7440		SZA
00717	7402		HLT
00720	1072		TAD 05P
00721	5673		JMP I POIN2
00722	1071	F,	TAD 04P
00723	5673		JMP I POIN2
00724	1070	TH,	TAD 03P
00725	5673		JMP I POIN2
00726	1067	T,	TAD 02P
00727	5673		JMP I POIN2
00730	1066	D,	TAD 01P
00731	5673		JMP I POIN2
00732	1065	ZER,	TAD 00P
00733	5673		JMP I POIN2
	1000	*1000	
01000	0000	SAMPLE,	0
01001	7300		CLA CLL
01002	1233		TAD LOGOUT
01003	7040		CMA
01004	6501		ACOUT
01005	7340		CLA CLL CMA
01006	6501		ACOUT
01007	7300		CLA CLL
01010	1057		TAD ZEROP
01011	1131		TAD MS1
01012	7650		SNA CLA
01013	5225		JMP TS1
01014	1130		TAD LGT
01015	3233		DCA LOGOUT
01016	1057		TAD ZEROP
01017	1132		TAD MS2
01020	7650		SNA CLA
01021	5230		JMP TS2
01022	1130		TAD LGT
01023	3233		DCA LOGOUT
01024	5600		JMP I SAMPLE
01025	1133	TS1,	TAD S1
01026	3233		DCA LOGOUT
01027	5600		JMP I SAMPLE
01030	1134	TS2,	TAD S2
01031	3233		DCA LOGOUT
01032	5600		JMP I SAMPLE
01033	0001	LOGOUT,	1

/ AC TO PATCH PNL. 6501

01034	0000	MPXR,	0	
01035	7300		CLA CLL	
01036	2102		ISZ MADD	
01037	1102		TAD MADD	
01040	1121		TAD LIMIT	
01041	7650		SNA CLA	
01042	3102		DCA MADD	
01043	1102		TAD MADD	
01044	6451		MXR	/ AC TO MULTIPLEXER. 6451
01045	5634		JMP I MPXR	
01046	7402	TTYTX,	HLT	
01047	0000	WTBL,	0	
01050	7300		CLA CLL	
01051	6621		SDR	
01052	5251		JMP .-1	
01053	6631		RUD	
01054	7041		CIA	
01055	1127		TAD STCODE	
01056	7640		SZA CLA	
01057	5251		JMP .-6	
01060	5647		JMP I WTBL	
01061	0000	RNDWT,	0	
01062	7300		CLA CLL	
01063	6504		ACIN	
01064	0110		AND 01	
01065	7650		SNA CLA	
01066	5263		JMP .-3	
01067	5661		JMP I RNDWT	
01070	0000	CVRT,	0	
01071	6442		STC	/ START AD CONVERSION. 6442
01072	7300		CLA CLL	
01073	7000		NOP	
01074	7000		NOP	
01075	7000		NOP	
01076	7000		NOP	
01077	6441		ADINP	/ GET DATA FROM AD. 6441
01100	5670		JMP I CVRT	
01101	7402	TTYRX,	HLT	
01102	0000	RSTIN,	0	
01103	7300		CLA CLL	
01104	1051		TAD P0	
01105	3057		DCA ZEROP	
01106	1052		TAD P1	
01107	3060		DCA ONEP	
01110	1053		TAD P2	
01111	3061		DCA TWOP	
01112	1054		TAD P3	
01113	3062		DCA THREEP	
01114	1055		TAD P4	
01115	3063		DCA FOURP	
01116	1056		TAD P5	
01117	3064		DCA FIVEP	
01120	5702		JMP I RSTIN	
01121	0000	RSTOU,	0	
01122	7300		CLA CLL	
01123	1051		TAD P0	
01124	3065		DCA OOP	
01125	1052		TAD P1	
01126	3066		DCA O1P	
01127	1053		TAD P2	
01130	3067		DCA O2P	
01131	1054		TAD P3	
01132	3070		DCA O3P	
01133	1055		TAD P4	
01134	3071		DCA O4P	
01135	1056		TAD P5	
01136	3072		DCA O5P	
01137	5721		JMP I RSTOU	
01140	0000	RSTD,	0	
01141	7300		CLA CLL	
01142	1051		TAD P0	

01143	3073		DCA DOP
01144	1052		TAD P1
01145	3074		DCA D1P
01146	1053		TAD P2
01147	3075		DCA D2P
01150	1054		TAD P3
01151	3076		DCA D3P
01152	1055		TAD P4
01153	3077		DCA D4P
01154	1056		TAD P5
01155	3100		DCA D5P
01156	5740		JMP I RSTD
01157	0000	INCS,	0
01160	7300		CLA CLL
01161	2073		ISZ DOP
01162	2074		ISZ D1P
01163	2075		ISZ D2P
01164	2076		ISZ D3P
01165	7000		NOP
01166	2077		ISZ D4P
01167	2100		ISZ D5P
01170	1073		TAD DOP
01171	1120		TAD NOSAMS
01172	7200		CLA
01173	5757		JMP I INCS
	1200	*1200	
01200	7300	DISPLA,	CLA CLL
01201	7604		LAS
01202	0112		AND 07
01203	3135		DCA DISNUM
01204	3136		DCA XDIS
01205	1135	NXLOC,	TAD DISNUM
01206	4444		JMS I IPOIN3
01207	2137		DCA CPTR
01210	7330		CLA CLL CML RAR
01211	1537		TAD I CPTR
01212	6201		CDF+0
01213	7110		CLL RAR
01214	7110		CLL RAR
01215	6062		OUTY
01216	7300		CLA CLL
01217	1136		TAD XDIS
01220	6052		OUTX
01221	2138		ISZ XDIS
01222	4443		JMS I IINCS
01223	1138		TAD XDIS
01224	7041		CIA
01225	1052		TAD P1
01226	7640		SZA CLA
01227	5205		JMP NXLOC
01230	4445		JMS I IRSTD
01231	5201		JMP NXLOC-4
01232	0000	POIN3,	0
01233	7041		CIA
01234	6211		CDF+10
01235	7450		SNA
01236	5271		JMP DZ
01237	7001		IAC
01240	7450		SNA
01241	5267		JMP DO
01242	7001		IAC
01243	7450		SNA
01244	5265		JMP DTWO
01245	7001		IAC
01246	7450		SNA
01247	5263		JMP DTHREE
01250	6201		CDF+0
01251	7001		IAC
01252	7450		SNA
01253	5261		JMP DFOUR
01254	7001		IAC
01255	7440		SZA
01256	7402		HLT

01257	1100		TAD D5P
01260	5632		JMP I POIN3
01261	1077	DFOUR,	TAD D4P
01262	5632		JMP I POIN3
01263	1076	DTHREE,	TAD D3P
01264	5632		JMP I POIN3
01265	1075	DTWO,	TAD D2P
01266	5632		JMP I POIN3
01267	1074	DO,	TAD D1P
01270	5632		JMP I POIN3
01271	1073	DZ,	TAD DOP
01272	5632		JMP I POIN3

/ SPECIAL INSTRUCTIONS
/

6062	DUTY=6062
6052	DUTX=6052
6511	SK1=6511
6621	SDR=6621
6624	STR=6624
6442	STC=6442
6451	MXR=6451
6501	ACDUT=6501
6504	ACIN=6504
6644	OUT=6644
6634	LSTAT=6634
6642	ETI=6642
6632	ERI=6632
6622	DUI=6622
6641	SKERR=6641
6441	ADINP=6441
6412	SION=6412
6414	SIOF=6414
6631	RUD=6631
6203	CDIO=6203

ACIN	6504	LIMIT	0121	SKERR	6641
ACDUT	6501	LOGOUT	1033	SK1	6511
ADINP	6441	LSTAT	6634	SP	0126
CDIO	6203	MADD	0102	START	0200
CPTR	0137	MPXR	1034	STATUS	0125
CUROP	0105	MS1	0131	STC	6442
CURPNT	0104	MS2	0132	STCODE	0127
CVRT	1070	MXR	6451	STR	6624
DFOUR	1261	M1	0113	STRT	0421
DISNUM	0135	M2	0114	SW	0124
DISPLA	1200	M3	0115	SWSP	0106
DO	1267	M4	0116	SWSP	0107
DTHREE	1263	M5	0117	S1	0133
DTWO	1265	NOSAMS	0120	S2	0134
DUI	6622	NXCHNL	0467	T	0726
DZ	1271	NXLOC	1205	TAC	0046
DOP	0073	O	0730	TFG	0047
D1P	0074	ONE	0667	TH	0724
D2P	0075	ONEP	0060	THREE	0563
D3P	0076	OUCHN	0103	THREEP	0062
D4P	0077	OUT	6644	TIMER	0464
D5P	0100	OUTX	6052	TRIAL	0122
ERI	6632	DUTY	6062	TRIALS	0123
ETI	6642	DOP	0065	TRIEND	0231
F	0722	O1	0110	TS1	1025
FINSH	0254	O1P	0066	TS2	1030
FIVEP	0064	O2P	0067	TTYRX	1101
FOUR	06E1	O3	0111	TTYTX	1046
FOURP	0053	O3P	0070	TWO	0665
ICVRT	0025	O4P	0071	TWOP	0061
IDISP	0042	O5P	0072	TXEND	0543
IINC	0027	O7	0112	UARTTX	0517
IINCR2	0033	POIN1	0632	WAIT	0230
IINC3	0043	POIN2	0673	WTBL	1047
IMPXR	0025	POIN3	1232	XDIS	0136

INCR1	0600	PRIOR	0050	ZER	0732
INCR2	0615	P0	0051	ZERO	0671
INCR3	1157	P1	0052	ZEROP	0057
IPOIN1	0031	P2	0053		
IPOIN2	0032	P3	0054		
IPOIN3	0044	P4	0055		
IRETN	0030	P5	0056		
IRNDWT	0037	RETN	0436		
IRSTD	0045	RNDWT	1061		
IRSTIN	0034	RSTD	1140		
IRSTOU	0035	RSTIN	1102		
ISAM	0024	RSTOU	1121		
ISERV	0040	RUD	6631		
ISTR1	0041	SAMPLE	1000		
ITIMR	0023	SAVE	0101		
ITTYRX	0021	SDR	6621		
ITTYTX	0022	SECND	0546		
IUARTT	0020	SERV	0400		
IWTBL	0036	SIOF	6414		
LGT	0130	SION	6412		

ERRORS DETECTED: 0
 LINKS GENERATED: 1

APPENDIX 8.11

PAL 8 Computer Programme Used to Store the Data onto Disk

This programme was used to store the data on a magnetic disk. It made use of the OS8 operating system User Service Routine (USR) to create core image files into which the data were transferred. Details of the USR may be found in the OS8 Software Support Handbook available from the Digital Equipment Company.

The programme consists of two sections:-

- a) Memory locations 0_8 - 3652_8 (Disk store programme).

This section stores the data received over the high speed serial interface in a memory buffer, checks for transmission errors and calls the USR to create data files and store the data. Each data file is automatically given a new filename based on a "seed" name specified by the operator. (e.g. If the operator specified the "seed" name as MJN000 then the subsequent files would have the names MJN001, MJN002, MJN003.....)

- b) Memory locations 16000_8 - 16777_8 (Data examine and transfer programme).

This section is stored on disk with the data in every data file. Subsequent to the completion of the data acquisition process it allows the operator to display (via D-A converters and an X-Y display) any of the six stored channels in the data file. It also allows the

operator to send the data to floppy disk (or paper tape) for transfer to the main computer. These functions are controlled by simple commands typed at the console terminal.

e.g. .R XYZ007 would cause OS8 to load the data in file XYZ007.SV and start the Data examine and transfer programme.

D 0 would display channel zero on the X-Y display.

H 48 would set the initial batch number transferred onto floppy disk (or paper tape) to 48.

P would transfer the data to floppy disk or paper tape (depending on which of these was connected).

Control & C would transfer control back to OS8.

The programme is started by giving the OS8 command "R DDS44".

The assembly language listing of the programme follows.

/ DISK DATA STORE.
/ -----
/

/ THIS PROGRAM STORES THE DATA RECEIVED FROM
/ THE SERIAL DATA INTERFACE ON AN RK05 DISK.
/

/ MEMORY MAP
/ -----
/

/ FIELD ZERO
/ -----
/

/ 0000 TO 0777 DISK STORE PROG.
/ 1000 TO 1377 CORE CONTROL BLOCK
/ 1400 TO 3377 CHANNEL 0

/ OVERLAY FOR FILENAME AT 2000
/ THIS IS ONCE ONLY CODE.
/

/ 3400 TO 5377 CHANNEL 1
/ 5400 TO 7377 CHANNEL 2
/ 7400 TO 7577 DEVICE HANDLER
/ 7600 TO 7777 DEC SOFTWARE

/ FIELD ONE
/ -----
/

/ 0000 TO 1777 CHANNEL 3
/ 2000 TO 3777 CHANNEL 4
/ 4000 TO 5777 CHANNEL 5
/ 6000 TO 6777 DATA EX. AND PUNCH PROG.
/ 7600 TO 7777 DEC SOFTWARE
/ -----
/

/ SPECIAL INSTRUCTIONS.
/

6634	LSTAT=6634	/ LOAD UART CONTROL REGISTER.
6631	RUD=6631	/ READ UART DATA.
6644	OUT=6644	/ OUTPUT SERIAL DATA.
6621	SDR=6621	/ SKIP IF UART DATA READY.
6641	SKERR=6641	/ SKIP IF UART RX. ERROR.
6056	OUTX=6056	/ OUTPUT X TO D-A AND BR. UP.
6066	OUTY=6066	/ OUTPUT Y TO D-A AND BR. UP.
6624	STR=6624	/ SKIP IF TX. BUFFER MT.
00000	0000 *0	0
00001	7402	HLT / BEWARE OF INTERRUPTS.
00020	0020 *20	
00021	7700	USR, 7700 / FIELD 1. USR CALL ADDRESS.
00022	7600	MONST, 7600 / MONITOR RESTART ADDRESS.
00023	0026	STATUS, 26 / UART CONTROL WORD.
00024	0000	TMPST, 0
00025	0000	TRIALS, 0
00026	0052	STCODE, 52 / INDICATES TO BF THAT BL IS READY.
00027	0040	TRILIM, 40 / NUMBER OF TRIALS TO BE STORED.
00030	0324	ZT, 324
00031	0325	ZU, 325
	0330	ZX, 330

00032	0303	ZC,	303	
00033	0305	ZE,	305	
00034	7775	MCNC,	-3	
00035	0336	UPARRO,	336	
00036	0215	CR,	215	
00037	0212	LF,	212	
00040	0007	D7,	7	
00041	0016	D16,	16	
00042	0017	D17,	17	
00043	0060	D60,	60	
00044	0077	D77,	77	
00045	0177	D177,	177	
00046	0377	D377,	377	
00047	4000	D4000,	4000	
00050	7700	D7700,	7700	
00051	7777	M1,	-1	
00052	7776	M2,	-2	
00053	7775	M3,	-3	
00054	7774	M4,	-4	
00055	7773	M5,	-5	
00056	7772	M6,	-6	
00057	7771	M7,	-7	
00060	7766	M12,	-12	
00061	7746	M32,	-32	
00062	7720	M50,	-60	
00063	7706	M72,	-72	
00064	7677	M101,	-101	
00065	0000	MADD,	0	
00066	0000	CURPNT,	0	
00067	6000	NDSAMS,	6000	/ - NUMBER OF SAMPLES TO BE TAKEN.
00070	0000	SVE,	0	
00071	0000	TEMP,	0	
00072	0000	TEMP2,	0	
00073	0000	TEMP3,	0	
00074	0000	ACMLTR,	0	
00075	7400	HANADR,	7400	/ ADDRESS AT WHICH THE DEVICE HANDLER / MAY BE PLACED. NOTE THAT THIS IS A DUMMY / ADDRESS SINCE THE DEVICE (BEING THE / SYSTEM DEVICE) IS ALWAYS (?) RESIDENT.
00076	0317	DEVNME,	DEVICE COPY	
00077	2031			
00100	0101	NAMEP,	NAM1	/ POINTS TO FILENAME.
00101	0401	NAM1,	401	/ FILENAME DEFAULTS TO
00102	2460	NAM2,	2460	/ DAT000.SV
00103	6060	NAM3,	6060	
00104	2326	NAM4,	2326	
00105	1400	HK5,	1400	/ START OF BUFFER AREA (FIELD 0).
00106	0000	CPYNUM,	0	/ DEVICE NUMBER. DETERMINED BY 'FETCH'.
00107	0000	CPYNTY,	0	/ HANDLER ENTRY ADDRESS. DETERMINED / BY 'FETCH'.
00110	0033	NUMBLK,	33	/ NUMBER OF BLOCKS FOR EACH FILE.
00111	0000	STBLK,	0	/ NEXT FREE BLOCK ON DISK.
00112	1000	CCBARA,	1000	/ LOCATION AT WHICH THE CORE / IMAGE OF THE CCB IS STORED.
00113	4200	CCBFCW,	4200	/ FUNCTION CONT. WORD FOR STORING / THE CCB.
00114	7410	FCW1,	7410	/ FUNCTION CONTROL WORD.
00115	7000	FCW2,	7000	/ SEE OSB SOFTWARE SUPPORT MANUAL / PAGE 4-2.
00116	4000	PNTR,	4000	
00117	0000	CARRY,	0	
00120	0000	XDIS,	0	
00121	0000	DISNUM,	0	/ INITIAL BUFFER ADDRESSES.
00122	1400	P0,	1400	/ FIELD 0
00123	3400	P1,	3400	/ FIELD 0
00124	5400	P2,	5400	/ FIELD 0
00125	0000	P3,	0	/ FIELD 1
00126	2000	P4,	2000	/ FIELD 1
00127	4000	P5,	4000	/ FIELD 1
00130	0000	ZERP,	0	/ CURRENT BUFFER POINTERS.

```

00131 0000 ONEP, 0
00132 0000 TWOP, 0
00133 0000 THREEP, 0
00134 0000 FOURP, 0
00135 0000 FIVEP, 0
/ SUBROUTINE POINTERS.
00136 0347 ICHRD, ZCHRD
00137 0660 INPRT, NUMPNT
00140 0250 IRENME, RENAME
00141 0400 IDSK, DSK
00142 0542 IINCR, INCR1
00143 0556 IMAINC, MAINC
00144 0600 IPOINI, PDINI
00145 0566 IGTWRD, GETWRD
00146 0703 IGETBY, GETBY
00147 0641 IRSTIN, RSTIN
00150 0000 SAVE1, 0
00151 0000 SAVE3, 0
00152 0240 ZSPCE, 240
00153 7000 NOPER, 7000
00154 0000 PTR, 0
00155 3000 PPTR, 3000
00156 0000 TPTR, 0
00157 4773 LIMIT, -3005
0200 0200 *200
00200 7300 CLA CLL / INITIALISE PROGRAM.
00201 6046 TLS
00202 1022 TAD STATUS
00203 6634 LSTAT
00204 6631 RUD
00205 6631 RUD / ENSURE UART BUFFERS EMPTY.
00206 4777 GNME, JMS FNME
00207 7300 CLA CLL
00210 3024 DCA TRIALS
00211 3065 TRL, DCA MADD / GET DATA AND STORE IN CORE.
00212 4547 JMS I IRSTIN
00213 4537 JMS I INPRT / PRINT NEXT BATCH NUMBER
00214 1025 TAD STCODE
00215 6644 OUT / SEND "READY" SIGNAL.
00216 4545 NEXT, JMS I IGTWRD
00217 3151 DCA SAVE3
00220 1065 TAD MADD
00221 4544 JMS I IPOINI
00222 3066 DCA CURPNT
00223 1151 TAD SAVE3
00224 3466 DCA I CURPNT
00225 6201 CDF+0
00226 4543 JMS I IMAINC / SKIP IF ALL CHANNELS DONE.
00227 7620 SNL CLA
00230 5216 JMP NEXT
00231 3065 DCA MADD
00232 4542 JMS I IINCR / INCREMENT ALL POINTERS.
00233 7420 SNL / SKIP IF ALL SAMPLES TAKEN.
00234 5216 JMP NEXT
00235 4541 JMS I IDSK / TRANSFER CORE BUFFER TO DISK.
00236 4540 JMS I IRENME / UPDATE FILE NAME.
00237 7301 CLA CLL IAC
00240 1024 TAD TRIALS
00241 3024 DCA TRIALS
00242 1026 TAD TRILIM
00243 7041 CIA
00244 1024 TAD TRIALS
00245 7640 SZA CLA / SKIP IF ALL TRIALS COMPLETE.
00246 5211 JMP TRL
00247 5421 JMP I MONST / JUMP BACK TO MONITOR.
/
/ THIS SUBROUTINE UPDATES THE FILENAME.
/
00250 0000 RENAME, 0
00251 7300 CLA CLL
00252 1103 TAD NAM3
00253 4325 JMS INC
00254 7640 SZA CLA

```

00255	5261		JMP LBY
00256	1043		TAD 060
00257	3071		DCA TEMP
00260	2117		ISZ CARRY
00261	1103	LBY,	TAD NAM3
00262	0050		AND 07700
00263	1071		TAD TEMP
00264	3103		DCA NAM3
00265	1117		TAD CARRY
00266	7650		SNA CLA
00267	5650		JMP I RENAME
00270	1103		TAD NAM3
00271	4335		JMS PPSW
00272	4325		JMS INC
00273	7640		SZA CLA
00274	5300		JMP MBY
00275	1043		TAD 060
00276	3071		DCA TEMP
00277	2117		ISZ CARRY
00300	1071	MBY,	TAD TEMP
00301	4335		JMS PPSW
00302	3071		DCA TEMP
00303	1103		TAD NAM3
00304	0044		AND 077
00305	1071		TAD TEMP
00306	3103		DCA NAM3
00307	1117		TAD CARRY
00310	7650		SNA CLA
00311	5650		JMP I RENAME
00312	1102		TAD NAM2
00313	4325		JMS INC
00314	7640		SZA CLA
00315	5320		JMP HBY
00316	1043		TAD 060
00317	3071		DCA TEMP
00320	1102	HBY,	TAD NAM2
00321	0050		AND 07700
00322	1071		TAD TEMP
00323	3102		DCA NAM2
00324	5650		JMP I RENAME
00325	0000	INC,	0
00326	0044		AND 077
00327	7001		IAC
00330	3071		DCA TEMP
00331	3117		DCA CARRY
00332	1071		TAD TEMP
00333	1063		TAD M72
00334	5725		JMP I INC
			/ THIS SUBROUTINE SIMULATES THE
			/ PDP8E 'BSW' INSTRUCTION.
			/ IT WAS WRITTEN BY A. JERRAM.
00335	0000	PPSW,	0
00336	7100		CLL
00337	3150		DCA SAVE1
00340	1150		TAD SAVE1
00341	0050		AND 07700
00342	1150		TAD SAVE1
00343	7006		RTL
00344	7006		RTL
00345	7006		RTL
00346	5735		JMP I PPSW
00347	0000	ZCHRO,	0
00350	6041		TSF
00351	5350		JMP .-1
00352	6046		TLS
00353	7300		CLA CLL
00354	5747		JMP I ZCHRO
00377	2000		
	0400	*400	
			/ THIS SUBROUTINE USES THE USR
			/ TO STORE THE DATA ON THE
			/ RK05 DISK.
00400	0000	DSK,	0

00401	7300	CLA CLL	
00402	1076	TAD DEVNME	/ GET DEVICE NAME FROM CONSTANT
00403	3224	DCA FARG1	/ AREA AND PREPARE TO FETCH THE
00404	1077	TAD DEVNME+1	/ DEVICE HANDLER.
00405	3225	DCA FARG2	
00406	1075	TAD HANADR	/ GET THE LOCATION AT WHICH THE
00407	3226	DCA FARG3	/ HANDLER MAY BE PLACED.
00410	1100	TAD NAMEP	/ POINTS TO THE FILENAME.
00411	3247	DCA EARG1	
00412	3250	DCA EARG2	/ CLEAR 2ND. ARG.
00413	1100	TAD NAMEP	
00414	3336	DCA CARG1	
00415	1110	TAD NUMBLK	/ GET THE FILE LENGTH.
00416	3337	DCA CARG2	/ (IN BLOCKS).
00417	7300	CLA CLL	
00420	6201	CDF+0	
00421	6212	CIF+10	
00422	4420	JMS I USR	/ JUMP TO USR.
00423	0001	1	/ DENOTES "FETCH" FUNCTION.
00424	0000	FARG1, 0	
00425	0000	FARG2, 0	
00426	0000	FARG3, 0	
00427	7402	HLT	/ ERROR IN GETTING HANDLER.
00430	7300	CLA CLL	
00431	1225	TAD FARG2	/ NOW THE DEVICE NUMBER.
00432	0042	AND 017	/ MASK OFF UPPER BITS.
00433	3106	DCA CPYNUM	
00434	1226	TAD FARG3	/ ENTRY POINT OF HANDLER.
00435	3107	DCA CPYNTY	
00436	1110	TAD NUMBLK	/ CALCULATE VALUE IN AC
00437	0046	AND 0377	/ ON ENTRY TO USR IN THE
00440	7106	CLL RTL	/ "ENTER" MODE. SEE OSB
00441	7006	RTL	/ SOFTWARE SUPPORT MANUAL
00442	1106	TAD CPYNUM	/ PAGE 2-7.
00443	6201	CDF+0	
00444	6212	CIF+10	
00445	4420	JMS I USR	
00446	0003	3	/ DENOTES "ENTER" FUNCTION.
00447	0000	EARG1, 0	
00450	0000	EARG2, 0	
00451	7402	HLT	/ ERROR RETURN.
00452	7300	CLA CLL	
00453	1247	TAD EARG1	/ NOW THE FIRST EMPTY
00454	3111	DCA STBLK	/ BLOCK ON THE DEVICE.
00455	1113	TAD CCBFCW	/ CONTROL WORD FOR THE CCB AREA.
00456	3306	DCA CHARG1	
00457	1112	TAD CCBARA	/ LOCATION AT WHICH THE CCB
00460	3307	DCA CHARG2	/ IS STORED IN CORE.
00461	1111	TAD STBLK	/ FIRST M.T. BLOCK.
00462	3310	DCA CHARG3	
00463	2111	ISZ STBLK	/ ALLOW ONE BLOCK FOR THE CCBB.
00464	7000	NOP	
00465	1114	TAD FCW1	/ GET FIELD 1 FCW.
00466	3315	DCA HARG1	
00467	3316	DCA HARG2	/ FIELD 1 BUFFER STARTS AT 0.
00470	1111	TAD STBLK	/ NEXT M.T. BLOCK AFTER CCB.
00471	3317	DCA HARG3	
00472	1115	TAD FCW2	/ GET FIELD 0 FCW.
00473	3324	DCA HARG4	
00474	1105	TAD HK5	/ START OF FIELD 0 BUFFER.
00475	3325	DCA HARG5	
00476	1111	TAD STBLK	/ ALLOW FOR FIELD 1 DATA.
00477	1041	TAD 016	/ =14 DATA, 2 PROG.=16 (8.).
00500	3111	DCA STBLK	
00501	1111	TAD STBLK	
00502	3326	DCA HARG6	
00503	6201	CDF+0	
00504	6202	CIF+0	
00505	4507	JMS I CPYNTY	/ SAVE CCB.
00506	0000	CHARG1, 0	
00507	0000	CHARG2, 0	
00510	0000	CHARG3, 0	
00511	7402	HLT	

```

00512 6201 CDF+0
00513 6202 CIF+0
00514 4507 JMS I CPYNTY / SAVE FIELD 1 DATA AND PROG.
00515 0000 HARG1, 0
00516 0000 HARG2, 0
00517 0000 HARG3, 0
00520 7402 HLT
00521 6201 CDF+0
00522 6202 CIF+0
00523 4507 JMS I CPYNTY / SAVE FIELD 0 DATA.
00524 0000 HARG4, 0
00525 0000 HARG5, 0
00526 0000 HARG6, 0
00527 7402 HLT
00530 7300 CLA CLL
00531 1106 TAD CPYNUM
00532 6201 CDF+0
00533 6212 CIF+10
00534 4420 JMS I USR
00535 0004 4
00536 0000 CARG1, 0
00537 0000 CARG2, 0
00540 7402 HLT
00541 5600 JMP I DSK
/ THIS SUBROUTINE INCREMENTS ALL THE
/ DATA POINTERS AND SETS THE LINK
/ IF ALL THE SAMPLES HAVE BEEN TAKEN.

00542 0000 INCR1, 0
00543 7300 CLA CLL
00544 2130 ISZ ZEROP
00545 2131 ISZ ONEP
00546 2132 ISZ TWOP
00547 2133 ISZ THREEP
00550 2134 ISZ FOURP
00551 2135 ISZ FIVEP
00552 1133 TAD THREEP
00553 1067 TAD NOSAMS
00554 7200 CLA
00555 5742 JMP I INCR1
00556 0000 MAINC, 0
00557 7301 CLA CLL IAC
00560 1065 TAD MADD
00561 3065 DCA MADD
00562 1056 TAD M6
00563 1065 TAD MADD
00564 7200 CLA
00565 5756 JMP I MAINC
00566 0000 GETWRD, 0
00567 4546 JMS I IGETBY
00570 3070 DCA SVE
00571 4546 JMS I IGETBY
00572 7106 CLL RTL
00573 7106 CLL RTL
00574 7106 CLL RTL
00575 1070 TAD SVE
00576 5766 JMP I GETWRD
0600 *600 / THIS SUBROUTINE RETURNS THE CURRENT
/ POINTER FOR THE CHANNEL REQUESTED BY
/ THE AC CONTENTS. IT ALSO SETS THE
/ DATA FIELD.

00600 0000 PDIN1, 0
00601 7041 CIA
00602 6201 CDF+0
00603 7450 SNA
00604 5237 JMP ZERO
00605 7001 IAC
00606 7450 SNA
00607 5235 JMP ONE
00610 7001 IAC
00611 7450 SNA
00612 5233 JMP TWO
00613 6211 CDF+10
00614 7001 IAC

```

00615	7450		SNA
00616	5231		JMP THREE
00617	7001		IAC
00620	7450		SNA
00621	5227		JMP FOUR
00622	7001		IAC
00623	7440		SZA
00624	7402		HLT
00625	1135		TAD FIVEP
00626	5600		JMP I POIN1
00627	1134	FOUR,	TAD FOURP
00630	5600		JMP I POIN1
00631	1133	THREE,	TAD THREEP
00632	5600		JMP I POIN1
00633	1132	TWO,	TAD TWOP
00634	5600		JMP I POIN1
00635	1131	ONE,	TAD ONEP
00636	5600		JMP I POIN1
00637	1130	ZERO,	TAD ZEROP
00640	5600		JMP I POIN1
00641	0000	RSTIN,	O
00642	7300		CLA CLL
00643	1122		TAD P0
00644	3130		DCA ZEROP
00645	1123		TAD P1
00646	3131		DCA ONEP
00647	1124		TAD P2
00650	3132		DCA TWOP
00651	1125		TAD P3
00652	3133		DCA THREEP
00653	1126		TAD P4
00654	3134		DCA FOURP
00655	1127		TAD P5
00656	3135		DCA FIVEP
00657	5641		JMP I RSTIN
00660	0000	NUMPNT,	O
00661	7300		CLA CLL
00662	1035		TAD CR
00663	4536		JMS I ICHRO
00664	1037		TAD LF
00665	4536		JMS I ICHRO
00666	1102		TAD NAM2
00667	0044		AND 077
00670	4536		JMS I ICHRO
00671	1103		TAD NAM3
00672	7012		RTR
00673	7012		RTR
00674	7012		RTR
00675	0044		AND 077
00676	4536		JMS I ICHRO
00677	1103		TAD NAM3
00700	0044		AND 077
00701	4536		JMS I ICHRO
00702	5660		JMP I NUMPNT
00703	0000	GETBY,	O
00704	7300		CLA CLL
00705	6621	WAIT,	SBR
00706	5325		JMP CNTRLC
00707	6631		RUD
00710	0044		AND 077
00711	6641		SKERR / TEST FOR ERRORS.
00712	5703		JMP I GETBY
00713	7300		CLA CLL
00714	1152		TAD ZSPCE
00715	4536		JMS I ICHRO
00716	1027		TAD ZT
00717	4536		JMS I ICHRO
00720	1031		TAD ZX
00721	4536		JMS I ICHRO
00722	1033		TAD ZE
00723	4536		JMS I ICHRO
00724	5777		JMP TRL
00725	6031	CNTRLC,	KSF

00726	5305		JMP WAIT	
00727	6036		KRB	
00730	0045		AND 0177	
00731	1034		TAD MCNC	
00732	7640		SZA CLA	
00733	5305		JMP WAIT	
00734	1035		TAD UPARRO	
00735	4536		JMS I ICHRO	
00736	1032		TAD ZC	
00737	4536		JMS I ICHRO	
00740	6201		CDF+0	
00741	5421		JMP I MONST	
00777	0211			
	1000	*1000		
			/ CORE CONTROL BLOCK	
01000	7776		7776	/ 2 SEGMENTS
01001	6213		6213	/ STARTING FIELD
01002	6000		6000	/ STARTING ADDRESS
01003	1400		1400	/ JOB STATUS WORD
01004	0000		0000	/ CORE ORIGIN
01005	3410		3410	/ 34 PAGES, FIELD 1
01006	1400		1400	/ CORE ORIGIN
01007	3000		3000	/ 30 PAGES, FIELD 0
	2000	*2000		
			/ CHARACTER INPUT ROUTINE	
			/ HANDLES ^U, ^C, ^C, ^J & RUBOUT.	
02000	0000	FNME,	0	
02001	7300		CLA CLL	
02002	1153		TAD NOPER	
02003	3777		DCA GNME	/ RESET JMS TO NOP
02004	6046		TLS	
02005	7300	START,	CLA CLL	
02006	1360		TAD PSTRG	
02007	4776		JMS PRINT	/ PRINT PROMPT
02010	7300		CLA CLL	
02011	3357		DCA FLAG	/ CLEAR RUBOUT FLAG
02012	1155		TAD PPTR	/ GET START OF BUFFER
02013	3154		DCA PTR	/ AND SAVE IN PTR
02014	4775	GTCR,	JMS ZCHRI	/ GET A CHARACTER
02015	3073		DCA TEMP3	
02016	1073		TAD TEMP3	
02017	1355		TAD MCRET	/ = C.R. ?
02020	7650		SNA CLA	
02021	5305		JMP OK5	
02022	1073		TAD TEMP3	
02023	1352		TAD MRBOUT	/ WAS IT A RUBOUT ?
02024	7650		SNA CLA	/ SKIP IF NOT
02025	5330		JMP ROUT	/ JUMP TO RUBOUT IF WAS
02026	1357		TAD FLAG	
02027	7650		SNA CLA	/ WAS RUBOUT FLAG SET ?
02030	5234		JMP NOFLG	/ NO SO JUMP OVER NEXT BIT.
02031	1356		TAD SL5H	
02032	4535		JMS I ICHRO	
02033	3357		DCA FLAG	/ AND CLEAR THE FLAG
02034	1073	NOFLG,	TAD TEMP3	
02035	1353		TAD MCNTU	/ WAS IT A ^U ?
02036	7640		SZA CLA	
02037	5245		JMP NOCU	/ NO SO JUMP OVER NEXT BIT.
02040	1035		TAD UPARRO	/ YES SO ECH ^
02041	4536		JMS I ICHRO	
02042	1030		TAD ZU	
02043	4536		JMS I ICHRO	/ AND "U"
02044	5205		JMP START	/ AND JUMP BACK.
02045	1073	NOCU,	TAD TEMP3	
02046	1354		TAD MCNTJ	
02047	7650		SNA CLA	/ WAS IT A ^J ?
02050	5313		JMP LFEED	
02051	1073		TAD TEMP3	/ NO SO TEST FOR ^C
02052	1034		TAD MCNC	
02053	7640		SZA CLA	
02054	5263		JMP NOCC	/ NO SO JUMP OVER NEXT BIT.
02055	1035		TAD UPARRO	/ YES SO ECHO ""
02056	4536		JMS I ICHRO	
02057	1032		TAD ZC	/ AND "C"

02060	4536		JMS I ICHRO	
02061	6203		CDF CIF+0	/ ENSURE I/F AND D/F = 0
02062	5421		JMP I MONST	
02063	1073	NOCC,	TAD TEMP3	
02064	1062		TAD M60	/ NOW CHECK FOR NUMBER 0-9 OR
02065	7510		SPA	/ LETTER A-Z (UPPER CASE).
02066	5214		JMP GTCR	
02067	1060		TAD M12	
02070	7510		SPA	
02071	5300		JMP OK	/ WAS A NUMBER
02072	1057		TAD M7	
02073	7510		SPA	
02074	5214		JMP GTCR	/ CODE WAS 72-100 OCTAL SO ERROR.
02075	1061		TAD M32	
02076	7500		SMA	
02077	5214		JMP GTCR	/ CODE WAS > 132 (Z) SO ERROR.
02100	7300	OK,	CLA CLL	
02101	1073		TAD TEMP3	
02102	3554		DCA I PTR	
02103	1073		TAD TEMP3	
02104	4536		JMS I ICHRO	/ ECHO TO TTY.
02105	2154	OKS,	ISZ PTR	/ INCREMENT POINTER.
02106	1073		TAD TEMP3	
02107	1355		TAD MCRET	/ TEST FOR RETURN.
02110	7440		SZA	
02111	5214		JMP GTCR	
02112	5774		JMP TSTNME	
02113	4773	LFEED,	JMS CRLF	/ PRINT RETURN AND LINEFEED.
02114	1155		TAD PPTR	/ SET TEMP3 POINTER=START
02115	3156		DCA TPTR	
02116	1154	NXCH,	TAD PTR	
02117	7041		CIA	
02120	1156		TAD TPTR	
02121	7700		SMA CLA	
02122	5214		JMP GTCR	/ JUMP BACK IF FINISHED
02123	1556		TAD I TPTR	/ GET CURRENT CHAR.
02124	4536		JMS I ICHRO	
02125	7300		CLA CLL	
02126	2156		ISZ TPTR	/ INCREMENT POINTER.
02127	5316		JMP NXCH	
02130	1357	ROUT,	TAD FLAG	/ GET FLAG
02131	7440		SZA	
02132	5337		JMP FLSET	
02133	7001		IAC	
02134	3357		DCA FLAG	/ SET FLAG
02135	1356		TAD SLSH	/ AND PRINT "/"
02136	4536		JMS I ICHRO	
02137	7340	FLSET,	CLA CLL CMA	/ SET AC=-1
02140	1154		TAD PTR	/ DEC POINTER.
02141	3154		DCA PTR	
02142	1155		TAD PPTR	
02143	7041		CIA	
02144	1154		TAD PTR	
02145	7710		SPA CLA	/ BACK TO START OF BUFFER.
02146	5205		JMP START	/ YES SO BACK TO START.
02147	1554		TAD I PTR	/ GET CHAR.
02150	4536		JMS I ICHRO	/ AND ECHO IT.
02151	5214		JMP GTCR	
02152	7601	MRBOUT,	-177	
02153	7753	MCNTU,	-25	
02154	7766	MCNTJ,	-12	
02155	7763	MCRET,	-15	
02156	0057	SLSH,	57	
02157	0000	FLAG,	0	
02160	3500	PSTRG,	STRING	
02173	2200			
02174	2224			
02175	2206			
02176	2214			
02177	0206			
	2200	PAGE		
02200	0000	CRLF,	0	
02201	1036		TAD CR	

02202	4536		JMS I ICHRO	
02203	1037		TAD LF	
02204	4536		JMS I ICHRO	
02205	5600		JMP I CRLF	
02206	0000	ZCHRI,	0	
02207	6031		KSF	
02210	5207		JMP .-1	
02211	6036		KRB	
02212	0045		AND 0177	
02213	5606		JMP I ZCHRI	
02214	0000	PRINT,	0	
02215	3072		DCA TEMP2	
02216	1472		TAD I TEMP2	
02217	7450		SNA	
02220	5614		JMP I PRINT	
02221	4536		JMS I ICHRO	
02222	2072		ISZ TEMP2	
02223	5216		JMP PRINT+2	
02224	7300	TSTNME,	CLA CLL	/ TEST THE FILENAME IS VALID.
02225	1155		TAD PPTR	/ CHECK THE LENGTH FIRST.
02226	7041		CIA	
02227	1154		TAD PTR	
02230	1057		TAD M7	
02231	7640		SZA CLA	
02232	5265		JMP ERMESS	/ IS THE FIRST CHARACTER ALPHA.
02233	1555		TAD I PPTR	
02234	1064		TAD M101	
02235	7710		SPA CLA	
02236	5265		JMP ERMESS	
02237	1100		TAD NAMEP	
02240	3156		DCA TPTR	/ NAME LOOKS OK SO PACK INTO
02241	1155		TAD PPTR	/ 6 BIT OCTAL.
02242	3154		DCA PTR	
02243	1554	NXCHRP,	TAD I PTR	
02244	0044		AND 077	
02245	7106		CLL RTL	
02246	7006		RTL	
02247	7006		RTL	
02250	3071		DCA TEMP	
02251	2154		ISZ PTR	
02252	1554		TAD I PTR	
02253	0044		AND 077	
02254	1071		TAD TEMP	
02255	3556		DCA I TPTR	
02256	2154		ISZ PTR	
02257	2156		ISZ TPTR	
02260	1154		TAD PTR	
02261	1157		TAD LIMIT	
02262	7710		SPA CLA	
02263	5243		JMP NXCHRP	
02264	5777		JMP GNME	
02265	7300	ERMESS,	CLA CLL	
02266	1271		TAD PSTRIN	
02267	4214		JMS PRINT	
02270	5776		JMP START	
02271	3632	PSTRIN,	STRIN2	
02376	2005			
02377	0206			
	3600	*3600		
03600	0212	STRING,	212	
03601	0215		215	
03602	0320		"P	
03603	0314		"L	
03604	0305		"E	
03605	0301		"A	
03606	0323		"S	
03607	0305		"E	
03610	0240		"	
03611	0324		"T	
03612	0331		"Y	
03613	0320		"P	
03614	0305		"E	
03615	0240		"	

03616	0306		"F
03617	0311		"I
03620	0314		"L
03621	0305		"E
03622	0316		"N
03623	0301		"A
03624	0315		"M
03625	0305		"E
03626	0212		212
03627	0215		215
03630	0276		">
03631	0000		0
03632	0212	STRIN2,	212
03633	0215		215
03634	0311		"I
03635	0316		"N
03636	0326		"U
03637	0301		"A
03640	0314		"L
03641	0311		"I
03642	0304		"D
03643	0240		"
03644	0316		"N
03645	0301		"A
03646	0315		"M
03647	0305		"E
03650	0212		212
03651	0215		215
03652	0000		0
	0001	FIELD 1	
	6000	*6000	
16000	7300		CLA CLL / THIS IS STORED AT THE SAME TIME
16001	6026		PLS / AS THE DATA. IT ENABLES THE USER TO
16002	6046		TLS / EXAMINE (ON AN X-Y DISPLAY) OR TRANSFER
16003	1350		TAD STUS/ THE DATA TO THE SERIAL LINE OUTPUT
16004	6634		LSTAT / BY USING THE OSB 'RUN DEV:FILENAME'
			/ COMMAND. ONCE RUNNING THE USER MAY TYPE
			/ THE FOLLOWING COMMANDS:
			/ "D 0", "D 1", ETC TO DISPLAY THE DATA.
			/ "H NNN" TO SET THE BATCH NUMBER TO NNN.
			/ "P" TO TRANSFER THE DATA TO THE SERIAL
			/ LINE.
16005	7300	NXCMD,	CLA CLL
16006	6211		CDF+10
16007	4777		JMS PCRLF
16010	4776		JMS GETCHR
16011	3356		DCA CMND
16012	1356		TAD CMND
16013	1357		TAD MH
16014	7650		SNA CLA
16015	5233		JMP HDR
16016	1356		TAD CMND
16017	1360		TAD MP
16020	7650		SNA CLA
16021	5775		JMP PUNCH
16022	1356		TAD CMND
16023	1361		TAD MD
16024	7650		SNA CLA
16025	5774		JMP DISP
16026	1362	ERROR,	TAD SPCE
16027	4773		JMS CHROUT
16030	1363		TAD QMK
16031	4773		JMS CHROUT
16032	5205		JMP NXCMD
16033	1362	HDR,	TAD SPCE
16034	4773		JMS CHROUT
16035	1365		TAD BUFF
16036	3366		DCA BUFFP
16037	4776	GET,	JMS GETCHR
16040	3355		DCA TMPSTR
16041	1355		TAD TMPSTR
16042	1352		TAD M260
16043	7710		SPA CLA
16044	5256		JMP ENDCHK
16045	1355		TAD TMPSTR

16046	1353		TAD	M272
16047	7700		SMA	CLA
16050	5226		JMP	ERROR
16051	1355		TAD	TMPSTR
16052	1352		TAD	M260
16053	3766		DCA	I BUFFP
16054	2366		ISZ	BUFFP
16055	5237		JMP	GET
16056	1355	ENDCHK,	TAD	TMPSTR
16057	1354		TAD	MCR
16060	7640		SZA	CLA
16061	5226		JMP	ERROR
16062	3364		DCA	BATNO
16063	1366		TAD	BUFFP
16064	7041		CIA	
16065	1365		TAD	BUFF
16066	7650		SNA	CLA
16067	5314		JMP	ENDHDR
16070	4324		JMS	BUFDEC
16071	3364		DCA	BATNO
16072	1366		TAD	BUFFP
16073	7041		CIA	
16074	1365		TAD	BUFF
16075	7650		SNA	CLA
16076	5314		JMP	ENDHDR
16077	4324		JMS	BUFDEC
16100	4332		JMS	TENS
16101	1354		TAD	BATNO
16102	3364		DCA	BATNO
16103	1366		TAD	BUFFP
16104	7041		CIA	
16105	1365		TAD	BUFF
16106	7650		SNA	CLA
16107	5314		JMP	ENDHDR
16110	4324		JMS	BUFDEC
16111	4341		JMS	HUNS
16112	1364		TAD	BATNO
16113	3364		DCA	BATNO
16114	7300	ENDHDR,	CLA	CLL
16115	1364		TAD	BATNO
16116	7510		SPA	
16117	5226		JMP	ERROR
16120	1351		TAD	PM193
16121	7710		SPA	CLA
16122	5205		JMP	NXCMD
16123	5226		JMP	ERROR
16124	0000	BUFDEC,	0	
16125	7340		CLA	CLL CMA
16126	1366		TAD	BUFFP
16127	3366		DCA	BUFFP
16130	1766		TAD	I BUFFP
16131	5724		JMP	I BUFDEC
16132	0000	TENS,	0	
16133	7041		CIA	
16134	3371		DCA	T1
16135	1370		TAD	KTEN
16136	2371		ISZ	T1
16137	5335		JMP	.-2
16140	5732		JMP	I TENS
16141	0000	HUNS,	0	
16142	7041		CIA	
16143	3371		DCA	T1
16144	1367		TAD	KHUN
16145	2371		ISZ	T1
16146	5344		JMP	.-2
16147	5741		JMP	I HUNS
			/	CONSTANTS
16150	0017	STUS,	17	
16151	7477	PM193,	7477	
16152	7520	M260,	-260	
16153	7506	M272,	-272	
16154	7563	MCR,	7563	
16155	0000	TMPSTR,	0	

16156	0000	CMND,	0
16157	7470	MH,	7470
16160	7460	MP,	7460
16161	7474	MD,	7474
16162	0240	SPCE,	240
16163	0277	QMK,	277
16164	0000	BATND,	0
16165	7000	BUFF,	7000
16166	0000	BUFFP,	0
16167	0144	KHUN,	144
16170	0012	KTEN,	12
16171	0000	T1,	0
16173	6317		
16174	6254		
16175	6200		
16176	6600		
16177	6325		
	6200	*6200	
16200	4777	PUNCH,	JMS LLDR
16201	4777		JMS LLDR
16202	3352		DCA CHND
16203	1776	STRT,	TAD P0377
16204	4333		JMS UADUT
16205	1776		TAD P0377
16206	4333		JMS UADUT
16207	1775		TAD BATNO
16210	4774		JMS WRDOUT
16211	7300		CLA CLL
16212	3355		DCA CKSM
16213	1353		TAD MCNT
16214	3351		DCA CNTR
16215	1352		TAD CHND
16216	4773		JMS POINT
16217	3345		DCA CPTR
16220	6214		RDF
16221	3354		DCA DFLD
16222	1354	GDAT,	TAD DFLD
16223	7650		SNA CLA
16224	5201		CDF+0
16225	1745		TAD I CPTR
16226	6211		CDF+10
16227	4774		JMS WRDOUT
16230	2345		ISZ CPTR
16231	7000		NOP
16232	2351		ISZ CNTR
16233	5222		JMP GDAT
16234	1346		TAD P0300
16235	4333		JMS UADUT
16236	1355		TAD CKSM
16237	4333		JMS UADUT
16240	2352		ISZ CHND
16241	1352		TAD CHND
16242	1350		TAD PM6
16243	7700		SMA CLA
16244	5251		JMP FIN
16245	2775		ISZ BATNO
16246	7000		NOP
16247	4777		JMS LLDR
16250	5203		JMP STRT
16251	4777	FIN,	JMS LLDR
16252	4777		JMS LLDR
16253	5772		JMP NXCMD
16254	7300	DISP,	CLA CLL / DISPLAY SLECTED CHANNEL.
16255	1771		TAD SPCE
16256	4317		JMS CHRUT
16257	4770		JMS GETND
16260	3343		DCA CHAN
16261	3344		DCA PXDIS
16262	1343		TAD CHAN
16263	4773		JMS POINT
16264	3345		DCA CPTR
16265	7330	NXLOC,	CLA CLL CML RAR
16266	1745		TAD I CPTR

16267	7110		CLL RAR
16270	7110		CLL RAR
16271	6066		OUTY
16272	7300		CLA CLL
16273	1344		TAD PXDIS
16274	6056		OUTX
16275	2345		ISZ CPTR
16276	7300		CLA CLL
16277	2344		ISZ PXDIS
16300	1344		TAD PXDIS
16301	1347		TAD PM2000
16302	7710		SPA CLA
16303	5265		JMP NXLOC
16304	6031		KSF
16305	5261		JMP DISP+5
16306	5772		JMP NXCMD
16307	0000	CHRIN,	0
16310	7300		CLA CLL
16311	6031		KSF
16312	5311		JMP .-1
16313	6036		KRB
16314	0767		AND PO177
16315	1766		TAD PO200
16316	5707		JMP I CHRIN
16317	0000	CHROUT,	0
16320	6041		TSF
16321	5320		JMP .-1
16322	6046		TLS
16323	7300		CLA CLL
16324	5717		JMP I CHROUT
16325	0000	PCRLF,	0
16326	1341		TAD PCR
16327	4317		JMS CHROUT
16330	1342		TAD PLF
16331	4317		JMS CHROUT
16332	5725		JMP I PCRLF
16333	0000	UAOUT,	0
16334	6624		STR
16335	5334		JMP .-1
16336	6644		OUT
16337	7300		CLA CLL
16340	5733		JMP I UAOUT
16341	0215	PCR,	215
16342	0212	PLF,	212
16343	0000	CHAN,	0
16344	0000	PXDIS,	0
16345	0000	CPTR,	0
16346	0300	PO300,	300
16347	6000	PM2000,	6000
16350	7772	PMS,	7772
16351	0000	CNTR,	0
16352	0000	CHNO,	0
16353	6000	MCNT,	6000
16354	0000	DFLD,	0
16355	0000	CKSM,	0
16366	6557		
16367	6556		
16370	6532		
16371	6162		
16372	6005		
16373	6471		
16374	6411		
16375	6164		
16376	6560		
16377	6400		
	6400	*6400	
16400	0000	LLDR,	0
16401	7300		CLA CLL
16402	1362		TAD KLD
16403	3346		DCA T2
16404	4777		JMS UAOUT

16405	2346		ISZ T2
16406	5204		JMP .-2
16407	5600		JMP I LLDR
16410	0100	PO100,	100
16411	0000	WRDOUT,	0
16412	3347		DCA T3
16413	7300		CLA CLL
16414	1347		TAD T3
16415	0361		AND PQ7700
16416	7012		RTR
16417	7012		RTR
16420	7012		RTR
16421	1210		TAD PO100
16422	4245		JMS EPTY
16423	3346		DCA T2
16424	1346		TAD T2
16425	1776		TAD CKSM
16426	0360		AND PO377
16427	3776		DCA CKSM
16430	1346		TAD T2
16431	4777		JMS UAOUT
16432	1347		TAD T3
16433	0355		AND PO77
16434	4245		JMS EPTY
16435	3346		DCA T2
16436	1346		TAD T2
16437	1776		TAD CKSM
16440	0360		AND PO377
16441	3776		DCA CKSM
16442	1346		TAD T2
16443	4777		JMS UAOUT
16444	5611		JMP I WRDOUT
16445	0000	EPTY,	0
16446	0356		AND PO177
16447	3351		DCA T6
16450	1354		TAD PM7
16451	3352		DCA T7
16452	3353		DCA BITS
16453	1351		TAD T6
16454	7010		RAR
16455	7430		SZL
16456	2353		ISZ BITS
16457	2352		ISZ T7
16460	5254		JMP .-4
16461	7300		CLA CLL
16462	1353		TAD BITS
16463	7010		RAR
16464	7200		CLA
16465	1351		TAD T6
16466	7430		SZL
16467	1357		TAD PO200
16470	5645		JMP I EPTY
16471	0000	POINT,	0
16472	7041		CIA
16473	6201		CDF+0
16474	7450		SNA
16475	5330		JMP DZ
16476	7001		IAC
16477	7450		SNA
16500	5326		JMP DD
16501	7001		IAC
16502	7450		SNA
16503	5324		JMP DTW
16504	6211		CDF+10
16505	7001		IAC
16506	7450		SNA
16507	5322		JMP DTH
16510	7001		IAC
16511	7450		SNA
16512	5320		JMP DF
16513	7001		IAC
16514	7440		SZA
16515	7402		HLT

16516	1353		TAD FIP
16517	5671		JMP I POINT
16520	1364	DF,	TAD FOP
16521	5671		JMP I POINT
16522	1365	DTH,	TAD THP
16523	5671		JMP I POINT
16524	1366	DTW,	TAD TWP
16525	5671		JMP I POINT
16526	1357	DO,	TAD OP
16527	5671		JMP I POINT
16530	1370	DZ,	TAD ZP
16531	5671		JMP I POINT
16532	0000	GETNO,	0
16533	4775		JMS GETCHR
16534	1774		TAD M260
16535	3350		DCA T4
16536	1350		TAD T4
16537	7510		SPA
16540	5773		JMP ERROR
16541	1772		TAD PMS
16542	7700		SMA CLA
16543	5773		JMP ERROR
16544	1350		TAD T4
16545	5732		JMP I GETNO
16546	0000	T2,	0
16547	0000	T3,	0
16550	0000	T4,	0
16551	0000	T6,	0
16552	0000	T7,	0
16553	0000	BITS,	0
16554	7771	PM7,	7771
16555	0077	PO77,	77
16556	0177	PO177,	177
16557	0200	PO200,	200
16560	0377	PO377,	377
16561	7700	PO7700,	7700
16562	7730	KLD,	7730
16563	4000	FIP,	4000
16564	2000	FDP,	2000
16565	0000	THP,	0
16566	5400	TWP,	5400
16567	3400	OP,	3400
16570	1400	ZP,	1400
16572	6350		
16573	6026		
16574	6152		
16575	5600		
16576	6355		
16577	6333		
	6600	*16600	
16600	0000	GETCHR,	0
16601	4777		JMS CHRIN
16602	3222		DCA SAVE
16603	1222		TAD SAVE
16604	1223		TAD MCC
16605	7550		SNA CLA
16606	5213		JMP EXIT
16607	1222		TAD SAVE
16610	4776		JMS CHROUT
16611	1222		TAD SAVE
16612	5500		JMP I GETCHR
16613	1224	EXIT,	TAD UPARR
16614	4776		JMS CHROUT
16615	1225		TAD CHC
16616	4776		JMS CHROUT
16617	6203		CDF CIF+0
16620	5621		JMP I .+1
16621	7600		7600
16622	0000	SAVE,	0
16623	7575	MCC,	7575
16624	0336	UPARR,	336
16625	0303	CHC,	303
16776	6217		

16777 6307

ACMLTR	0074	FLAG	2157	MCRET	2155	PCR	6341
BATND	6164	FLSET	2137	MD	6161	PCRLF	6325
BITS	6553	FNME	2000	MH	6157	PLF	6342
BUFDEC	6124	FOP	6564	MONST	0021	PM193	6151
BUFF	6165	FOUR	0627	MP	6160	PM2000	6347
BUFFP	6166	FOURP	0134	MRBOUT	2152	PM5	6350
CARG1	0536	GDAT	6222	M1	0051	PM7	6554
CARG2	0537	GET	6037	M101	0064	PNTR	0116
CARRY	0117	GETBY	0703	M12	0060	POINT	6471
CCBARA	0112	GETCHR	6600	M2	0052	POIN1	0600
CCFCW	0113	GETND	6532	M260	6152	PD100	6410
CHAN	6343	GETWRD	0566	M272	6153	PD177	6556
CHARG1	0506	GNME	0206	M3	0053	PD200	6557
CHARG2	0507	GTCR	2014	M32	0061	PD300	6346
CHARG3	0510	HANADR	0075	M4	0054	PD377	6560
CHC	6625	HARG1	0515	M5	0055	PO77	6555
CHNO	6352	HARG2	0516	M5	0056	PO7700	6561
CHRIN	6307	HARG3	0517	M50	0062	PPTR	0155
CHROUT	6317	HARG4	0524	M7	0057	PRINT	2214
CKSM	6355	HARG5	0525	M72	0063	PSTRG	2160
CMND	6156	HARG5	0526	NAMEP	0100	PSTRIN	2271
CNTR	6351	HBY	0320	NAM1	0101	PTR	0154
CNTRLC	0725	HDR	6033	NAM2	0102	PUNCH	6200
CPTR	6345	HK5	0105	NAM3	0103	PXDIE	6344
CPYNTY	0107	HUNS	6141	NAM4	0104	P0	0122
CPYNUM	0106	ICHRD	0136	NEXT	0216	P1	0123
CR	0036	IDSK	0141	NOCC	2063	P2	0124
CRLF	2200	IGETBY	0146	NCCU	2045	P3	0125
CURPNT	0066	IGTWRD	0145	NOFLG	2034	P4	0126
DEVNME	0076	IINCR	0142	NOPER	0153	P5	0127
DF	6520	IMAINC	0143	NOSAMS	0067	QMK	6163
DFLD	6354	INC	0325	NUMBLK	0110	RENAME	0250
DISNUM	0121	INCR1	0542	NUMPNT	0660	ROUT	2130
DISP	6254	INPRT	0137	NXCH	2116	RSTIN	0641
DO	6526	IPOIN1	0144	NXCHRP	2243	RUD	6631
DSK	0400	IRENME	0140	NXCMD	6005	SAVE	6622
DTH	6522	IRSTIN	0147	NXLOC	6265	SAVE1	0150
DTW	6524	KHUN	6167	OK	2100	SAVE3	0151
DZ	6530	KLD	6562	OK5	2105	SDR	6621
EARG1	0447	KTEN	6170	ONE	0635	SKERR	6641
EARG2	0450	LBY	0261	ONEP	0131	SLSH	2156
ENDCHK	6056	LF	0037	OP	6567	SPCE	6162
ENDHDR	6114	LFEED	2113	OUT	6644	START	2005
EPTY	6445	LIMIT	0157	OUTX	6056	STATUS	0022
ERMESS	2265	LLDR	6400	OUTY	6066	STBLK	0111
ERROR	6026	LSTAT	6634	016	0041	STCODE	0025
EXIT	6513	MADD	0065	017	0042	STR	6624
FARG1	0424	MAINC	0556	0177	0045	STRING	3600
FARG2	0425	MBY	0300	0377	0046	STRING2	3632
FARG3	0426	MCC	6623	04000	0047	STRT	6203
FCW1	0114	MCNC	0034	050	0043	STUS	6150
FCW2	0115	MCNT	6353	07	0040	SVE	0070
FIN	6251	MCNTJ	2154	077	0044	TEMP	0071
FIP	6563	MCNTU	2153	07700	0050	TEMP2	0072
FIVEP	0135	MCR	6154	PBSW	0335	TEMP3	0073
TENS	6132						
THP	6565						
THREE	0631						
THREEP	0133						
TMPST	0023						
TMPSTR	6155						
IPTR	0156						
TRIALS	0024						
TRILIM	0026						
TRL	0211						
TSTNME	2224						
TWO	0633						

TWOP	0132
TWP	6566
T1	6171
T2	6546
T3	6547
T4	6550
T6	6551
T7	6552
UAOUT	6333
UPARR	6624
UPARRD	0035
USR	0020
WAIT	0705
WRDOUT	6411
XDIS	0120
ZC	0032
ZCHRI	2206
ZCHRD	0347
ZE	0033
ZERO	0637
ZEROP	0130
ZP	6570
ZSPCE	0152
ZT	0027
ZU	0020
ZX	0021

ERRORS DETECTED: 0
LINKS GENERATED: 52

APPENDIX 8.12

FORTRAN Programmes used in the Analysis of the CNV Data

The listings of the three major analysis programmes are given in this appendix together with their associated subroutines. These programmes perform (i) The Pre- and Post- Stimulus (Broadband) Energy Test as described in section 2.2.3.1; (ii) The tests described in sections 2.2.2.1 (The Rayleigh Test of Circular Variance), 2.2.2.2 (The Modified Rayleigh Test), 2.2.2.3 (The Hodges-Ajne Test) and 2.2.3.3 (The Nearest and Furthest Mean Amplitude Test); (iii) All the tests in (ii) and in addition the test described in section 2.2.3.2 (The Pre- and Post- Stimulus Mean Amplitude differences test). All the CNV data were tested using programme (i). The one second ISI CNV's were tested using Programme (iii) whilst the four second ISI CNV's could only be tested using Programme (ii) because of the limited amount of pre-stimulus information available.

Programmes similar to (i) and (iii) were used in the analysis of the auditory responses.

The programmes listed in this appendix and in appendices 8.7 and 8.8 use a number library subroutines and functions specific to the Prime 750/550 system installed at Plymouth Polytechnic.

Brief details of these are as follows:-

1) Input and Output

a) On the Prime computer system the FORTRAN input / output unit numbers are as follows:-

unit 1 = Input / output from / to the users terminal.

unit 5 = Input / output from / to PRIMOS file unit 1.

unit 6 = Input / output from / to PRIMOS file unit 2.

Thus READ(1,*)A,I would perform a READ operation from the users terminal. The '*' infers free format whereby any numbers separated by spaces or commas are assigned to the elements of the variable list.

2) Applications Library and Operating System Subroutines.

The following routines are part of the PRIMOS applications library.

CLOS\$A[†] Closes a file by its PRIMOS file unit number.

EXST\$A Tests whether the file specified exists or not.

FILL\$A Fills an array with a specified ASCII character.

MSUB\$A	Copies a string of characters into another string.
NLEN\$A	Returns the number of non-blank characters in a string array.
OPEN\$A [†]	Opens a file for reading or writing on a specified PRIMOS file unit.
TREE\$A	Tests a string of characters to establish whether they constitute a valid file treename. The file need not exist.
YSNO\$A	Prompts the user with a specified string and requests a YES or NO answer. Returns a logical .TRUE. or .FALSE. according to the users reply.

† May be treated as Functions or Subroutines.

The following routines are part of the PRIMOS operating system.

EXIT	Returns from FORTRAN to PRIMOS.
TNOUA	Transfers a character string to the users terminal without appending a 'newline' sequence.
TODEC	Writes the argument as a decimal number on the users terminal.

T1IN Obtains a single character from
the users terminal.

3) Graphics subroutines

The following subroutines form part of the
GINO graphics package. For further details
see the GINOGRAF and GINO-F manuals or contact
The GINO-F Support Team, C.A.D. Centre, Madingley
Road, Cambridge, CB3 0HB.

ARCTO2	DEVEND
BROKEN	DEVSPE
CC81	LINTO2
CC906	MOVTO2
CHAANG	PENSEL
CHAARR	PICCLE
CHAFLO	S5660
CHAINT	SYMBOL
CHAHAR	
CHAHOL	T4010
CHAMOD	UNITS
CHASIZ	WINDO2
CURSOR	WINDOW

The following subroutine is part of the
GINOGRAF graphics package.

GRAF


```

WRITE(1,540)RESNAM
540 FORMAT('OVERWRITE EXISTING FILE : ',20A2)
IF(.NOT. YSNO$A('YES OR NO ',10,ASDNO))FILE=.FALSE.
IF(.NOT. FILE)GO TO 560
550 IF(.NOT. OPEN$A(2,RESNAM,RESLEN,2))GO TO 2900
560 DO 600 IC=1,MBAT
DIFF=SSSI(IC)-SSPRS1(IC)
WRITE(1,730)BATNRS(IC),SSPRS1(IC),SSSI(IC),DIFF
IF(FILE)WRITE(6,735)BATNRS(IC),SSPRS1(IC),SSSI(IC),DIFF
600 CONTINUE
730 FORMAT('BATCH ',I3,' PRE S1 ',E16.8,5X,' POST S1 ',
*E16.8, DIFFERENCE ',E16.8)
735 FORMAT(I3,3(E16.8,2X))
IF(FILE)CALL CLOS$A(2)
1900 CALL CLOS$A(1)
CALL EXIT
2900 WRITE(1,2901)RESNAM
2901 FORMAT('CANT OPEN RESULTS FILE : ',20A2)
STOP
END

```

PHASOR DIAGRAM ANALYSIS OF CNV

C
C
C
C
C
C
C
C
C
C
C

INPUT FILES: DATA
OUTPUT FILES: NONE

§INSERT SYSCOM>A\$KEYS
C

```
REAL DATA(1024)
COMPLEX TDATA(512), S1(6, 32) /* 6 HARMONICS, 32 TRIALS
DIMENSION TRDAT(512), ANGLE(32), RAD(32)
INTEGER*2 FNAME(20), CHANAN, BATNRS(32), SSR, BLINE
INTEGER*2 TDLEN, BATNO
INTEGER*2 LAB1(40), EP1, ODEV, DEL
COMMON ODEV
DATA ODEV / 99 /
PI=3.14159265.
WRITE(1, 1)
```

C
C
C
C

1 FORMAT('CNV PHASOR DIAGRAM ANALYSIS')

FIND OUT WHICH GRAPHICAL OUTPUT
DEVICE IS TO BE USED.

```
WRITE(1, 2)
2 FORMAT('WHICH OUTPUT DEVICE FOR GRAPHIC DATA')
3 WRITE(1, 5)
5 FORMAT('TEKTRONIX(0), CALCOMP(1) OR SIGMA(2)')
READ(1, *, ERR=3) IDEV
6 WRITE(1, 7)
7 FORMAT('CNV PARADIGM 1 OR 4 SECONDS')
READ(1, *, ERR=6) IPARA
IF(IPARA .EQ. 1 .OR. IPARA .EQ. 4) GO TO 10
GO TO 6
10 N=1024
```

C
C
C
C
C
C
C
C
C

THESE PARAMETERS ARE GOVERNED
BY THE CNV PARADIGM
EP1 IS THE START OF THE POST- S1 DATA.
NP IS THE LENGTH OF THE POST- S1 DATA.
IS1 IS THE S1 STIMULUS POINT
IS2 IS THE S2 STIMULUS POINT
DEL IS A DELAY FROM S2 TO ALLOW THE AEP & CNV TO SETTLE.
ALL VALUES ARE IN TERMS OF 'SAMPLE NUMBER' 1 N 1024

```
NP=64
TDLEN=64
DEL=125
```

C
C
C

FOR ONE SECOND CNV'S

```
EP1=472
IS1=407
IS2=532
IF(IPARA .EQ. 1) GO TO 20
NP=400
TDLEN=512
```

C
C
C

FOR FOUR SECOND CNV'S

```
EP1=295
IS1=219
IS2=719
```

C
C
C

GET THE NAME OF THE INPUT DATA FILE.

```
20 CALL GETNAM(FNAME)
LGTDLN=IFIX(ALOG10(FLOAT(TDLEN))/ALOG10(2.) + 0.5)
ITL=TDLEN/2 + 1
CHANAN=4
```

C
C

STUDY CHANNEL 4 I.E. VERTEX

```
CHANAN=CHANAN-3
25 WRITE(1, 30)
30 FORMAT('PERFORM EYE MOVEMENT CORRECTIONS')
READ(1, *, ERR=25) E CORR
31 WRITE(1, 35)
35 FORMAT('PERFORM BASELINE CORRECTIONS')
READ(1, *, ERR=31) BLINE
38 WRITE(1, 40)
```

```

40 FORMAT('FILTER')
  READ(1,*,ERR=38)IFILT
C
C
C
      ALLOW VARIABLE SCALE FACTORS SO AS TO GET
      ALL THE POINTS ON THE PHASOR DIAGRAM.
45 WRITE(1,50)
50 FORMAT('MULTIPLICATION FACTOR')
  READ(1,*,ERR=45)SF
  SF=SF*1.5E6
C
C
C
      FIND OUT WHICH BATCHES ARE TO BE ANALYSED.

  CALL BATNOS(BATNRS,MBAT)
  SSR=1
  N1=NP*SSR
  CALL FILLSA(LAB1,40,' ')
  ENCODE(57,52,LAB1)TDLEN,NP,SSR,EP1,BLINE
52 FORMAT(I3,' PT FFT',I3,' DATA PTS, SSR=',I2,', POST =',I3,
+4X,' B. LINE=',I2)
C
C
C
      NOW THE ITERATIVE BIT

  DO 640 IC=1,MBAT
  BATNO=BATNRS(IC)
54 WRITE(1,55)BATNO
55 FORMAT('PROCESSING BATCH ',I4)
  CALL EYECOR(FNAME,BATNO,CHANAN,DATA,SF1,SF2,SAMRAT,ECORR)
  IF(BLINE.EQ.1)CALL BASLINE(DATA,N,I,IS1,IS2+DEL,N)
  IF(IFILT.EQ.1)CALL FILTER(N,DATA)
  ISTRT=EP1-1
C
C
C
      EXTRACT DATA TO BE ANALYSED AND PUT
      INTO TRDAT.

  L1=0
  DO 135 I=1,N1,SSR
  L1=L1+1
135 TRDAT(L1)=DATA(I+ISTRT)
C
C
C
      TAPER THE DATA.

  CALL TAPER2(TRDAT,NP)
  DO 136 I=1,TDLEN
  TDATA(I)=CMPLX(0.,0.)
136 IF(I.LE.NP)TDATA(I)=CMPLX(TRDAT(I),0.)
C
C
C
      CALL FFT SUBROUTINE

151 CALL NLOGN(LGTDLN,TDATA,-1.)
  DO 160 IHAR=1,6 /* STUDY HARMONICS 1-6
C
C
C
      NOT INTERESTED IN DC TERM
      I.E. TDATA(1)

160 S1(IHAR,IC)=TDATA(IHAR+1)*2./FLOAT(NP)
640 CONTINUE
  CALL PHASR1(S1,MBAT,SF,SAMRAT,FNAME,LAB1,IDEV)
  CALL PHASR1(S1,MBAT,SF,SAMRAT,FNAME,LAB1,-1)
  IF(IDEV.EQ.0.OR. IDEV.EQ.2)CALL TIIN(IXYZ)
  DO 1650 IHAR=1,6
  DO 1630 K=1,MBAT
  ANGLE(K)=ATAN2(AIMAG(S1(IHAR,K)),REAL(S1(IHAR,K)))
  RAD(K)=CABS(S1(IHAR,K))
1630 CONTINUE
C
C
C
      NOW DO THE STATISTICAL TESTS.

  WRITE(1,1635)IHAR
1635 FORMAT(20X,'ANGULAR STATISTICS FOR POST-STIMULUS CNV HARMONIC',
+I3,/)
  CALL RSTAT1(RAD,MBAT)
  CALL ASTAT1(ANGLE,MBAT)
  CALL ASTAT2(ANGLE,MBAT)
  CALL VSTAT1(ANGLE,RAD,MBAT)
  CALL VSTAT3(ANGLE,RAD,MBAT)
  CALL VSTAT2(ANGLE,RAD,MBAT)
  WRITE(1,1640)
1640 FORMAT(////)
1650 CONTINUE

```

1900 CALL CLOSSA(1)
CALL EXIT
END


```

135 TRDAT(L1)=DATA(I+ISTR)
CALL TAPER2(TRDAT,NP)
DO 136 I=1,TDLEN
  TDATA(I)=CMPLX(0.,0.)
136 IF(I.LE.NP)TDATA(I)=CMPLX(TRDAT(I),0.)
151 CALL NLOGN(LGTDLN,TDATA,-1.)
DO 160 IHAR=1,6 /* STUDY HARMONICS 1-6
160 PRES1(IHAR,IC)=TDATA(IHAR+1)*2./FLOAT(NP)
C
C POST STIMULUS
ISTR=POSS-1
L1=0
DO 200 I=1,N1,SSR
  L1=L1+1
200 TRDAT(L1)=DATA(I+ISTR)
CALL TAPER2(TRDAT,NP)
DO 220 I=1,TDLEN
  TDATA(I)=CMPLX(0.,0.)
220 IF(I.LE.NP)TDATA(I)=CMPLX(TRDAT(I),0.)
CALL NLOGN(LGTDLN,TDATA,-1.)
DO 230 IHAR=1,6
230 S1(IHAR,IC)=TDATA(IHAR+1)*2./FLOAT(NP)
640 CONTINUE
CALL PHASOR(PRES1,S1,MBAT,SF,SAMRAT,FNAME,LAB1,IDEV)
CALL PHASOR(PRES1,S1,MBAT,SF,SAMRAT,FNAME,LAB1,-1)
IF(IDEV.EQ.0)CALL TIIN(IXYZ)
DO 1650 IHAR=1,6
DO 1600 K=1,MBAT
  ANGLE(K)=ATAN2(AIMAG(PRES1(IHAR,K)),REAL(PRES1(IHAR,K)))
  RAD(K)=CABS(PRES1(IHAR,K))
1600 CONTINUE
WRITE(1,1610)IHAR
1610 FORMAT(20X,'ANGULAR STATISTICS FOR PRE-STIMULUS HARMONIC ',I3)
CALL RSTAT1(RAD,MBAT)
CALL ASTAT1(ANGLE,MBAT)
CALL ASTAT2(ANGLE,MBAT)
CALL VSTAT1(ANGLE,RAD,MBAT)
CALL VSTAT3(ANGLE,RAD,MBAT)
CALL VSTAT2(ANGLE,RAD,MBAT)
WRITE(1,1620)
1620 FORMAT(7////)
DO 1630 K=1,MBAT
  ANGLE(K)=ATAN2(AIMAG(S1(IHAR,K)),REAL(S1(IHAR,K)))
  RAD(K)=CABS(S1(IHAR,K))
  RADDIF(IHAR,K)=CABS(S1(IHAR,K))-CABS(PRES1(IHAR,K))
1630 CONTINUE
WRITE(1,1635)IHAR
1635 FORMAT(20X,'ANGULAR STATISTICS FOR POST-STIMULUS HARMONIC ',I3)
CALL RSTAT1(RAD,MBAT)
CALL ASTAT1(ANGLE,MBAT)
CALL ASTAT2(ANGLE,MBAT)
CALL VSTAT1(ANGLE,RAD,MBAT)
CALL VSTAT3(ANGLE,RAD,MBAT)
CALL VSTAT2(ANGLE,RAD,MBAT)
WRITE(1,1640)
1640 FORMAT(7////)
1650 CONTINUE
1900 CALL CLOSSA(1)
WRITE(1,2200)
2200 FORMAT(7,'RESULTS OF A PAIRED T-TEST ON THE PRE-POST RADIUS
LENGTHS',7)
MBAT1=MBAT-1
DO 3000 IHAR=1,6
  SUM=0.
  DO 2000 I=1,MBAT
    SUM=SUM+RADDIF(IHAR,I)
  RMEAN=SUM/FLOAT(MBAT)
  SUMSQ=0.
  DO 2100 I=1,MBAT
    SUMSQ=SUMSQ+(RADDIF(IHAR,I)-RMEAN)*(RADDIF(IHAR,I)-RMEAN)
  STDEV=SQRT(SUMSQ/FLOAT(MBAT1))
  TSTAT=RMEAN/(STDEV/SQRT(FLOAT(MBAT)))
  WRITE(1,2210)IHAR,RMEAN,STDEV,TSTAT,MBAT1
2210 FORMAT('HARMONIC=',I2,5X,'MEAN=',E14.8,5X
&'ST. DEV=',E14.8,5X,'T=',F8.4,5X,'WITH',5X,I3,' DF')
3000 CONTINUE
CALL EXIT
END

```

```

SUBROUTINE ASTAT1(ANGLE,N)
C
C THIS SUBROUTINE CALCULATES SUMMARY STATISTICS FOR THE N ANGULAR
C VALUES (RADIANS) STORED IN ARRAY 'ANGLE'.
C
C MEAN DIRECTION .... THETA=ATAN(S/C)
C
C WHERE C = AVERAGE COSINE VALUE
C S = AVERAGE SINE VALUE
C
C CIRCULAR VARIANCE .... VO=1-SQRT(C*C+S*S)
C
C VO HAS A VALUE 1 FOR COMPLETE UNIFORMITY ON THE CIRCLE
C 0 FOR A SET OF IDENTICAL ANGLES
C.....
C
C WRITTEN BY TERRY JOHNSON. DEPT. OF MATHS. STATS. & COMPUTING
C PLYMOUTH POLYTECHNIC.
C
C DIMENSION ANGLE(N)
C DATA PI/3.1415926536/
C C=0
C S=0
C DO 1 I=1,N
C C=C+COS(ANGLE(I))
1 S=S+SIN(ANGLE(I))
C C=C/N
C S=S/N
C THETA=ATAN2(S,C)
C VO=1-SQRT(C*C+S*S)
C WRITE(1,100) THETA
C THETA=THETA*180/PI
C WRITE(1,102) THETA
C WRITE(1,101) VO
100 FORMAT(/'MEAN DIRECTION',13X,'=',F10.5,' RADIANS')
101 FORMAT('CIRCULAR VARIANCE',10X,'=',F10.5)
102 FORMAT(27X,'=',F10.5,' DEGREES')
C RETURN
C END

```

```

SUBROUTINE ASTAT2(ANGLE,N)
C
C THIS SUBROUTINE TESTS THE N ANGULAR VALUES IN ARRAY 'ANGLE'
C FOR UNIFORMITY OF DISTRIBUTION ON THE CIRCLE USING THE
C HODGES-AJNE TEST. THE TEST STATISTIC IS M WHERE M IS THE
C MINIMUM NUMBER OF OBSERVATIONS FOUND IN ANY SEMI-CIRCLE.
C.....
C
C WRITTEN BY TERRY JOHNSON. DEPT. OF MATHS. STATS. & COMPUTING
C PLYMOUTH POLYTECHNIC.
C
C DIMENSION ANGLE(N)
C DOUBLE PRECISION S
C DATA PI/3.1415926536/
C M=N
C DO 1 I=1,N
C NR=0
C IF(ANGLE(I).GT.0.0) GO TO 3
C AMIN=ANGLE(I)
C AMAX=AMIN+PI
C DO 2 J=1,N
C IF(ANGLE(J).GT.AMIN.AND.ANGLE(J).LE.AMAX) NR=NR+1
2 CONTINUE
C GO TO 5
3 AMAX=ANGLE(I)
C AMIN=AMAX-PI
C DO 4 J=1,N
C IF(ANGLE(J).GE.AMIN.AND.ANGLE(J).LT.AMAX) NR=NR+1
4 CONTINUE
5 IF(NR.LT.M) M=NR
C IF(N-NR.LT.M) M=N-NR
1 CONTINUE
C WRITE(1,100) M
C
C FOR M LESS THAN (N/3) EXACT SIGNIFICANCE IS GIVEN BY
C
C 
$$S = (N-2*M) \cdot C \cdot 2^{\frac{N}{M} - (1-N)}$$

C

```

C
C

M

```
IF(M.GT.N/3) RETURN
S=(N-2*M)/(2.0**(N-1))
IF(M.EQ.0) GO TO 7
DO 6 I=1,M
S=S*(N+1-I)/I
7 WRITE(1,101) S
100 FORMAT('HODGES-AJNE TEST STATISTIC = ',I4)
101 FORMAT('LEVEL OF SIGNIFICANCE = ',F10.5)
RETURN
END
```

C
C
C
C
C
C
C

SUBROUTINE BASLNE(DATA,N,NP1,NP2,NP3,NP4)

CORRECT THE BASELINE OF A SECTION OF EEG BY AVERAGING BETWEEN NP1&NP2 AND ALSO BETWEEN NP3&NP4. CALCULATE THE DIFFERENCE BETWEEN THE AVERAGES AND SUBTRACT THE APPROPRIATE FRACTION FROM THE DATA. FINALLY ADJUST THE SECTIONS NP1-NP2 AND NP3-NP4 TO HAVE A ZERO MEAN.

```
INTEGER*2 N,NP1,NP2,NP3,NP4,Z1,Z2
DIMENSION DATA(N)
CALL SECTAV(NP1,NP2,DATA,SAV1)
CALL SECTAV(NP3,NP4,DATA,SAV2)
GRAD=(SAV2-SAV1)/(FLOAT(NP3-NP2))
DO 70 I=1,NP2
70 DATA(I)=DATA(I)-SAV1
Z1=NP2+1
DO 72 I=Z1,NP3
72 DATA(I)=DATA(I)-SAV1-GRAD*(I-NP2)
Z2=NP3+1
DO 74 I=Z2,N
74 DATA(I)=DATA(I)-SAV2
RETURN
END
```

SUBROUTINE BATNOS(BATS,MAX)

GET A SEQUENCE OF BATCH NUMBERS FROM THE USER.

C
C
C
C

```
INTEGER*2 BATS(32)
LOGICAL YSNO$A
10 WRITE(1,20)
20 FORMAT('HOW MANY BATCHES TO BE PROCESSED')
READ(1,*,ERR=10)MAX
IF(MAX.GT.32.OR.MAX.LT.1)GO TO 10
IF(MAX.EQ.32)GO TO 100
ITBAT=0
I=1
40 CALL TNOUA('BATCH',5)
CALL TODEC(ITBAT)
IF(.NOT.YSNO$A('TO BE INCLUDED ',16,1))GO TO 50
BATS(I)=ITBAT
I=I+1
50 ITBAT=ITBAT+6
IF(I.LE.MAX)GO TO 40
RETURN
100 DO 120 I=1,MAX
BATS(I)=(I-1)*6
120 CONTINUE
RETURN
END
```

SUBROUTINE DATIN(IBATNO, IDATA, RNAME, SF1, SF2, SAMRAT, CSE)

THIS SUBROUTINE READS DATA FROM A DATA FILE SPECIFIED BY ARGUMENT RNAME. THE BATCH SPECIFIED IS READ INTO IDATA. THE PROGRAM ALSO RETURNS THE SCALE FACTORS SF1, SF2, AND THE SAMPLE RATE. (SAMRAT) SETTING CSE = -1 CLOSES ANY OPEN FILE AND RETURNS.

C
C
C
C
C
C
C

```
LOGICAL CLOSSA,OPEN$A
INTEGER*2 IDATA(1024),ONAME(20),RNAME(20),TITLE(36),NME(6)
INTEGER*2 RWKEY,NLEN,NLEN2,PRIMNO,NBAT,IBATNO,IAA1(4)
```

```

LOGICAL NEOPEN,CSE
DATA NEOPEN /.FALSE./
RWKEY=1
PRIMNO=1
NLEN=40
NLEN2=NLEN/2
IF(NEOPEN)GO TO 1300
IF(CSE)RETURN
900 IF(.NOT. OPENS(A(RWKEY,RNAME,NLEN,PRIMNO)))GO TO 1700
DO 1000 IC=1,NLEN2
1000 ONAME(IC)=RNAME(IC)
NEOPEN=.TRUE.
READ(5,3000,END=1800,ERR=1900)NME,SF1,SF2
READ(5,3010,END=1800,ERR=1900)MAXBAT
READ(5,3020,END=1800,ERR=1900)TITLE
DO 234 IL=1,4
234 IAA1(IL)=TITLE(IL+20)
DECODE (8,236,IAA1)SAMRAT
236 FORMAT(F8.4)
1100 READ(5,END=1800,ERR=1900)NBAT
READ(5,END=1800,ERR=1900)(IDATA(I),I=1,1024)
IF(NBAT .NE. IBATNO)GO TO 1100
NBAT=NBAT+1
RETURN
1200 IF(.NOT. CLOS(A(PRIMNO)))GO TO 1600
NEOPEN=.FALSE.
RETURN
1300 IF(CSE)GO TO 1200
DO 1400 IC=1,NLEN2
1400 IF(ONAME(IC) .NE. RNAME(IC))GO TO 1500
IF(IBATNO .GE. NBAT)GO TO 1100
1500 IF(CLOS(A(PRIMNO)))GO TO 900
1600 WRITE(1,1610)ONAME
1610 FORMAT('*** CAN NOT CLOSE FILE ',20A2,' ***')
STOP 1
1700 WRITE(1,1710)RNAME
1710 FORMAT('*** CAN NOT OPEN FILE ',20A2,' ***')
STOP 2
1800 WRITE(1,1810)RNAME
1810 FORMAT('*** END OF FILE ',20A2,' ***')
STOP 3
1900 WRITE(1,1910)IBATNO,RNAME
1910 FORMAT('*** ERROR TRYING TO READ BATCH ',I5,' FROM FILE ',
+20A2,' ***')
STOP 4
3000 FORMAT(6A2,2F8.6)
3010 FORMAT(I4)
3020 FORMAT(36A2)
END

```

```

SUBROUTINE EYECOR(FNAME,BATNO,ICHAN,COR1,SF1,SF2,SAMRAT,ECORR)
C THIS PROGRAM IS INTENDED TO MINIMISE THE AMOUNT OF
C E.O.G POWER IN THE E.E.G. IT USES THE MODIFIED
C QUILTER TECHNIQUE. HORIZONTAL AND VERTICAL COMPONENTS
C OF BOTH EYES ARE TAKEN INTO CONSIDERATION.
C THE PROGRAM REMOVES ANY D.C OFFSET ON ANY OF
C THE INPUT DATA CHANNELS.
REAL VL(1024),VR(1024),HL(1024),HR(1024),E1(1024),COR1(1024)
DIMENSION XI(4,5),RM1(4),RHS(4)
INTEGER*2 BATNO,INP(1024),FNAME(20)
LOGICAL CSE
DATA CSE /.FALSE./
N=1024
C THE DATA IS ASSUMED TO BE IN THE FOLLOWING
C ORDER ; VL , VR , HL , HR , M1 , M2
C M1 OR M2 ARE THE CHANNELS TO BE CORRECTED BY THE
C MODIFIED QUILTER TECHNIQUE.
C SF1 FOR EOG DATA. SF2 FOR EEG DATA.
C CHECK THAT THE BATCH NUMBER IS VALID.
IF (BATNO .LT. 0 .OR. BATNO .GT. 191)GO TO 2999
IF(FLOAT(BATNO/6) .NE. FLOAT(BATNO)/6.)GO TO 2999
L=BATNO
C READ THE DATA AND CONVERT TO REAL FORMAT.
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 22 I=1,N
22 VL(I)=FLOAT(INP(I))*SF1*1.E-06
L=L+1
CALL DATIN(L,INP,FNAME,SF1,SF2,SAMRAT,CSE)
DO 24 I=1,N
24 VR(I)=FLOAT(INP(I))*SF1*1.E-06

```

```

L=L+1
CALL DATIN(L, INP, FNAME, SF1, SF2, SAMRAT, CSE)
DO 26 I=1, N
26 HL(I)=FLOAT(INP(I))*SF1*1.E-06
L=L+1
CALL DATIN(L, INP, FNAME, SF1, SF2, SAMRAT, CSE)
DO 28 I=1, N
28 HR(I)=FLOAT(INP(I))*SF1*1.E-06
L=L+1

C
C
C
C
C
CHECK WHICH CHANNEL
IS TO BE CORRECTED.
IF CHANNEL 5 INCREMENT L.

IF(ICHAN .GE. 2)L=L+1
CALL DATIN(L, INP, FNAME, SF1, SF2, SAMRAT, CSE)
DO 30 I=1, N
30 EI(I)=FLOAT(INP(I))*SF2*1.E-06
C
SUBTRACT THE MEAN OF EACH DATA BATCH FROM THE DATA.
CALL SMEAN(N, EI, E1AV)
IF(ECORR .NE. 1.)GO TO 1000
CALL SMEAN(N, VL, VLM)
CALL SMEAN(N, VR, VRM)
CALL SMEAN(N, HL, HLM)
CALL SMEAN(N, HR, HRM)
C
FORM THE CORRELATION SUMS OF PRODUCTS.
A=0.
B=0.
C=0.
D=0.
E=0.
F=0.
G=0.
P=0.
Q=0.
R=0.
S=0.
T=0.
U=0.
V=0.
DO 100 I=1, N
A=A+VL(I)*VL(I)
B=B+VL(I)*VR(I)
C=C+VL(I)*HL(I)
D=D+VL(I)*HR(I)
E=E+VR(I)*VR(I)
F=F+VR(I)*HL(I)
G=G+VR(I)*HR(I)
P=P+HL(I)*HL(I)
Q=Q+HL(I)*HR(I)
R=R+HR(I)*HR(I)
S=S+EI(I)*VL(I)
T=T+EI(I)*VR(I)
U=U+EI(I)*HL(I)
V=V+EI(I)*HR(I)
100 CONTINUE
X1(1,1)=A
X1(1,2)=B
X1(1,3)=C
X1(1,4)=D
X1(2,2)=E
X1(2,3)=F
X1(2,4)=G
X1(3,3)=P
X1(3,4)=Q
X1(4,4)=R
RHS(1)=S
RHS(2)=T
RHS(3)=U
RHS(4)=V
C
SET UP SYMMETRICAL MATRIX.
X1(2,1)=X1(1,2)
X1(3,1)=X1(1,3)
X1(4,1)=X1(1,4)
X1(3,2)=X1(2,3)
X1(4,2)=X1(2,4)
X1(4,3)=X1(3,4)
C
SOLVE THE SIMULTANEOUS EQUATIONS BY THE
GAUSS PIVOTAL METHOD.
CALL GAUSS(4,X1,RHS,RM1)

```

```

C          APPLY THE CORRECTIONS TO THE DATA.
DO 220 I=1,N
COR1(I)=E1(I) - (RM1(1)*VL(I) + RM1(2)*VR(I) + RM1(3)*HL(I)
1+ RM1(4)*HR(I))
220 CONTINUE
RETURN
1000 DO 1010 I=1,N
1010 COR1(I)=E1(I)
RETURN
2999 WRITE(1,4000)BATNO
4000 FORMAT(///, ' BATCH NUMBER INCORRECT',I6)
STOP 8
END

```

SUBROUTINE FILTER(NPTS,XT)

C
C
C

DIGITAL FILTER THE DATA.

```

DIMENSION XT(NPTS),DATOUT(1024),H(128)
REAL H
LOGICAL GOTEM,OPENSA,CLOSSA
INTEGER*2 WTFILE(20),NLENSA
DATA GOTEM / .FALSE. /
IF(GOTEM)GO TO 60
4 WRITE(1,5)
5 FORMAT('NAME OF FILTER CO-EFFICIENT FILE')
READ(1,8,ERR=4)WTFILE
8 FORMAT(20A2)
NLEN=NLENSA(WTFILE,40)
IF(.NOT. OPENSA(1,WTFILE,NLEN,2))GO TO 100
READ(6,10)
10 FORMAT(/)
READ(6,15)ICASE
15 FORMAT(6X,I2)
CASE 1 = ODD LENGTH, SYMMETRICAL
CASE 2 = EVEN LENGTH, SYMMETRICAL
CASE 3 = ODD LENGTH, ANTI-SYMMETRICAL
CASE 4 = EVEN LENGTH, ANTI-SYMMETRICAL
ISGN=1
IF(ICASE .EQ. 3 .OR. ICASE .EQ. 4)ISGN=-1
READ(6,20)N
20 FORMAT(16X,I4)
N2=(N+1)/2
DO 50 I=1,N2
READ(6,40)H(I)
40 FORMAT(9X,E15.8)
50 H(N+1-I)=H(I)*FLOAT(ISGN)
GOTEM=.TRUE.
IF(.NOT. CLOSSA(2))GO TO 110
60 DO 80 I=1,NPTS
STORE=0.
IF(N .LT. I)GO TO 70
STORE=XT(I)
GO TO 80
70 DO 75 K=1,N
75 STORE=STORE + XT(I-K+1)*H(K)
80 DATOUT(I)=STORE
DO 85 I=1,NPTS
85 XT(I)=DATOUT(I)
RETURN
100 WRITE(1,105)WTFILE
105 FORMAT('CANT OPEN FILE',10A2)
STOP
110 WRITE(1,115)WTFILE
115 FORMAT('CANT CLOSE FILE',10A2)
STOP
END

```

```

SUBROUTINE GAUSS(K,B,RHS,X)
C          A SUBROUTINE TO SOLVE SIMULTANEOUS
C          EQUATIONS BY THE GAUSS PIVOTAL METHOD
DIMENSION A(7,8),X(K),B(K,K),RHS(K)
INTEGER*2 Z2,Z3,Z4,Z5,Z6,Z7

```

C
C
C
C

AUGMENT INPUT MATRICES
AND SAVE INPUT DATA.

```

DO 12 I=1,K
DO 10 J=1,K
10 A(I,J)=B(I,J)

```

```

12 A(I,K+1)=RHS(I)
   Z2=K-1
   DO 35 I=1,Z2
     L=I
     DO 15 J=I,K
15  IF(ABS(A(L,I)) .LT. ABS(A(J,I)))L=J
     IF(ABS(A(L,I)) .EQ. 0.)GO TO 60
     IF(L .EQ. 1)GO TO 21
     Z3=K+1
     DO 20 N=I,Z3
       SAVE=A(I,N)
       A(I,N)=A(L,N)
20  A(L,N)=SAVE
C      PIVOTAL REDUCTION
21  Z4=I+1
     DO 35 M=Z4,K
       D=A(M,I)/A(I,I)
       Z5=I+1
       Z6=K+1
       DO 35 J=Z5,Z6
35  A(M,J)=A(M,J)-D*A(I,J)
C      BACK SUBSTITUTION
     DO 50 L=1,K
       J=K+1-L
       IF(J .EQ. K)GO TO 45
       Y=A(J,K+1)
       Z7=K-1
       DO 40 M=J,Z7
40  Y=Y-A(J,M+1)*X(M+1)
       X(J)= Y/A(J,J)
       IF(J .NE. K)GO TO 50
45  X(J)=A(J,K+1)/A(J,J)
50  CONTINUE
     RETURN
60  WRITE(1,65)
65  FORMAT(' ERROR MESSAGE ZERO COLUMN FOUND')
     STOP
     END

SUBROUTINE GETNAM(NAME)
C      GET A VALID FILENAME FROM THE USER.
C
C
INTEGER*2 NAME(20),TBUFF(20),NLEN$A
LOGICAL EXST$A
DATA TBUFF /'NONAMEGIVEN
6  WRITE(1,8)
8  FORMAT(' GIVE NAME OF FILE TO BE PROCESSED')
   READ(1,10)NAME
10  FORMAT(20A2)
   IF(NAME(1) .EQ. ' ')GO TO 100
   DO 20 I=1,20
20  TBUFF(I)=NAME(I)
   GO TO 130
100 DO 120 I=1,20
120 NAME(I)=TBUFF(I)
130 IF(EXST$A(NAME,NLEN$A(NAME,40)))RETURN
   WRITE(1,140)NAME
140 FORMAT(' INVALID OR NON-EXISTANT FILE : ',20A2)
   GO TO 6
   END

SUBROUTINE NLOGN(N,X,SIGN)
C      THIS PROGRAM PERFORMS THE FFT.
C      N=BASE 2 LOG OF NO. OF POINTS.
C      X= COMPLEX ARRAY OF DATA FOR TRANSFORMATION.
C      SIGN= -1. FOR FFT.
C      SIGN= +1. FOR IFFT.
C      TRANSFORMED DATA IS RETURNED IN X.
C
C      WRITTEN BY E. A. ROBINSON.
C
DIMENSION M(12)
COMPLEX WK,HOLD,Q,X(2048)
LX=2**N
DO 1 I=1,N
1  M(I)=2**(N-I)
DO 4 L=1,N
NBLOCK=2**(L-1)
LBLOCK=LX/NBLOCK

```

```

LBHALF=LBLOCK/2
K=0
DO 4 IBLOCK=1,NBLOCK
FK=K
FLX=LX
V=SIGN*6.283185308*FK/FLX
WK=CMPLX(COS(V),SIN(V))
ISTART=LBLOCK*(IBLOCK-1)
DO 2 I=1, LBHALF
J=ISTART+I
JH=J+LBHALF
Q=X(JH)*WK
X(JH)=X(J)-Q
X(J)=X(J)+Q
2 CONTINUE
DO 3 I=2,N
II=I
IF(K.LT.M(I))GO TO 4
3 K=K-M(I)
4 K=K+M(II)
K=0
DO 7 J=1,LX
IF(K.LT.J)GO TO 5
HOLD=X(J)
X(J)=X(K+1)
X(K+1)=HOLD
5 DO 6 I=1,N
II=I
IF(K.LT.M(I))GO TO 7
6 K=K-M(I)
7 K=K+M(II)
IF(SIGN.LT.0.)RETURN
DO 8 I=1,LX
8 X(I)=X(I)/FLX
RETURN
END

```

SUBROUTINE PHASOR(PRE,POST,NMAX,SF,SAMRAT,FNAME,ILAB1,DEVICE)

C
C
C

PRE- AND POST-STIMULUS PHASOR DIAGRAMS.

COMPLEX PRE(6,64),POST(6,64)
INTEGER*2 DEVICE,ODEV,ILAB1(40),FNAME(20)
COMMON ODEV

C
C
C
C

TEST FOR A VALID DEVICE CODE
CURRENTLY 0=TEKTRONIX
1=CALCOMP 536 VIA 906
2=SIGMA 5670 COLOUR.

IF(DEVICE .GT. 3)GO TO 999

C
C
C
C

SKIP THE INITIALISATION IF THIS
DEVICE ALREADY BEEN INITIALISED

IF(DEVICE .EQ. ODEV)GO TO 50
IF(DEVICE .LT. 0)GO TO 600

C
C
C
C

DE-ASSIGN CURRENT DEVICE IF
DIFFERENT DEVICE REQUESTED.

IF(ODEV .NE. 99)CALL DEVEND

C
C
C

ASSIGN REQUESTED DEVICE

IF(DEVICE .EQ. 0)CALL T4010
IF(DEVICE .EQ. 1)CALL CC906
IF(DEVICE .EQ. 2)CALL S5660
IF(DEVICE .EQ. 3)CALL CC81
IBAUD=1200
CALL DEVSPE(IBAUD)
IF(DEVICE .EQ. 0)CALL UNITS(1.0)
IF(DEVICE .EQ. 1)CALL UNITS(1.0)
IF(DEVICE .EQ. 2)CALL UNITS(1.5)
IF(DEVICE .EQ. 3)CALL UNITS(0.75)
ODEV=DEVICE

50 CALL CHAHAR(1,0)

XS=46.
YS=70.
XD=100.
CRAD=40.
CALL WINDO2(0.,250.,0.,200.)

```

CALL WINDOW(1)
DO 400 IHAR=1,6
CALL TITLE(FNAME,ILAB1,DEVICE)

C
C
C
PRE STIM CIRCLE DIAGRAM OUTLINE.

IF(DEVICE .EQ. 2)CALL PENSEL(6,0.,0) /* RED
CALL MOVTO2(XS,YS)
CALL SYMBOL(7)
CALL MOVTO2(XS-CRAD,YS)
CALL ARCTO2(XS,YS,XS-CRAD,YS,0)

C
C
C
POST STIM CIRCLE DIAGRAM OUTLINE.

IF(DEVICE .EQ. 2)CALL PENSEL(5,0.,0) /* BLUE
CALL MOVTO2(XS+XD,YS)
CALL SYMBOL(7)
CALL MOVTO2(XS+XD-CRAD,YS)
CALL ARCTO2(XS+XD,YS,XS+XD-CRAD,YS,0)
IF(DEVICE .EQ. 2)CALL PENSEL(7,0.,0) /* WHITE
CALL MOVTO2(XS-15.,135.)
CALL CHAHOL('PRE STIM.*')
CALL MOVTO2(XS+XD-15.,135.)
CALL CHAHOL('POST STIM.*')
CALL MOVTO2(XS+XD/2-15.,125.)
CALL CHAHOL('HARMONIC *')
CALL CHAINT(IHAR,2)

C
C
C
C
C
DRAW PRE STIMULUS DIAGRAM
RADSM=SUM OF RADII.
RADMN=AVERAGE RADII.
RADSTD=STANDARD DEVIATION OF RADII.

RADSM=0.
RADSMS=0.
IF(DEVICE .EQ. 2)CALL PENSEL(3,0.,0) /* GREEN
DO 250 NTR=1,NMAX
X=REAL(PRE(IHAR,NTR))
Y=AIMAG(PRE(IHAR,NTR))

C
C
C
CALCULATE SUMS AND SUMS OF SQUARES.

RAD=CABS(PRE(IHAR,NTR))
RADSM=RADSM+RAD
RADSMS=RADSMS+RAD*RAD

C
C
C
CALCULATE ANGLE.

PHI=ATAN2(Y,X)
X1=CRAD*COS(PHI)
Y1=CRAD*SIN(PHI)

C
C
C
PUT TRIANGLE ON CIRCLE AT CORRECT ANGLE.

CALL MOVTO2(X1+XS,Y1+YS)
CALL SYMBOL(1)

C
C
C
PUT CROSS AT END OF PHASOR.

CALL MOVTO2(SF*X+XS,SF*Y+YS)
250 CALL SYMBOL(3)
IF(DEVICE .EQ. 2)CALL PENSEL(7,0.,0) /* WHITE

C
C
C
CALCULATE STATISTICS.

RADMN=RADSM/FLOAT(NMAX)
RADSTD=SQRT((RADSMS-RADSM*RADSM/FLOAT(NMAX))/FLOAT(NMAX-1))

C
C
C
PUT RESULTS ON GRAPH.

CALL MOVTO2(10.,10.)
CALL CHAHOL('AV. RAD= *')
CALL MOVTO2(40.,10.)
CALL CHAFLO(RADMN,10)
CALL MOVTO2(10.,4.)
CALL CHAHOL('ST. DEV= *')
CALL MOVTO2(40.,4.)
CALL CHAFLO(RADSTD,10)

C
C
DRAW POST STIMULUS DIAGRAM

```

```

C
RADSM=0.
RADSMS=0.
IF(DEVICE .EQ. 2)CALL PENSEL(2,0.,0) /* YELLOW
DO 300 NTR=1,NMAX
X=REAL(POST(IHAR,NTR))
Y=AIMAG(POST(IHAR,NTR))
RAD=CABS(POST(IHAR,NTR))
RADSM=RADSM+RAD
RADSMS=RADSMS+RAD*RAD
PHI=ATAN2(Y,X)
X1=CRAD*COS(PHI)
Y1=CRAD*SIN(PHI)
CALL MOVTO2(X1+XS+XD,Y1+YS)
CALL SYMBOL(1)
CALL MOVTO2(SF*X+XS+XD,SF*Y+YS)
300 CALL SYMBOL(4)
IF(DEVICE .EQ. 2)CALL PENSEL(7,0.,0) /* WHITE
RADMN=RADSM/FLOAT(NMAX)
RADSTD=SQRT((RADSMS-RADSM*RADSM/FLOAT(NMAX))/FLOAT(NMAX-1))
CALL MOVTO2(108.,10.)
CALL CHAHOL('AV. RAD= *.*')
CALL MOVTO2(138.,10.)
CALL CHAFLO(RADMN,10)
CALL MOVTO2(108.,4.)
CALL CHAHOL('ST. DEV= *.*')
CALL MOVTO2(138.,4.)
CALL CHAFLO(RADSTD,10)
CALL MOVTO2(0.,180.)
CALL CHAMOD
400 IF(DEVICE .EQ. 0 .OR. DEVICE .EQ. 2)CALL TIIN(IXYZ)
CONTINUE
ODEV=DEVICE
RETURN
600 ODEV=99
CALL DEVEND
RETURN
999 WRITE(1,1000)DEVICE
1000 FORMAT('INVALID PLOTTING DEVICE CODE',I5)
STOP 1
END

```

```

SUBROUTINE PHASR1(DATA,NMAX,SF,SAMRAT,FNAME,ILAB1,DEVICE)

```

C
C
C

SINGLE PHASOR DIAGRAM

```

COMPLEX DATA(6,32)
INTEGER*2 DEVICE,ODEV,ILAB1(40),FNAME(20)
COMMON ODEV

```

C
C
C
C

```

TEST FOR A VALID DEVICE CODE
CURRENTLY 0=TEKTRONIX
          1=CALCOMP 536 VIA 906
          2=SIGMA 5670 COLOUR.

```

```

IF(DEVICE .GT. 2)GO TO 999

```

C
C
C
C

```

SKIP THE INITIALISATION IF THIS
DEVICE ALREADY BEEN INITIALISED

```

```

IF(DEVICE .EQ. ODEV)GO TO 50
IF(DEVICE .LT. 0)GO TO 600

```

C
C
C
C

```

DE-ASSIGN CURRENT DEVICE IF
DIFFERENT DEVICE REQUESTED.

```

```

IF(ODEV .NE. 99)CALL DEVEND

```

C
C
C

ASSIGN REQUESTED DEVICE

```

IF(DEVICE .EQ. 0)CALL T4010
IF(DEVICE .EQ. 1)CALL CC906
IF(DEVICE .EQ. 2)CALL S5660
IBAUD=1200
CALL DEVSPE(IBAUD)
IF(DEVICE .EQ. 0)CALL UNITS(1.0)
IF(DEVICE .EQ. 1)CALL UNITS(1.0)
IF(DEVICE .EQ. 2)CALL UNITS(1.5)
ODEV=DEVICE

```

```

50 CALL CHAHAR(1,0)
XS=100.
YS=70.

```



```

C
  DIMENSION RAD(N)
  SUM=0.
  SUMSQ=0.
  DO 10 I=1,N
  SUM=SUM+RAD(I)
  SUMSQ=SUMSQ+RAD(I)*RAD(I)
10 CONTINUE
  RADMN=SUM/FLOAT(N)
  RADSTD=SQRT((SUMSQ-SUM*SUM/FLOAT(N))/FLOAT(N-1))
  WRITE(1,20)RADMN,RADSTD
20 FORMAT(/,'MEAN RADIUS LENGTH',9X,'=',E11.4,
+/, 'STANDARD DEV. OF RADIUS',4X,'=',E11.4)
  RETURN
  END

```

```

  SUBROUTINE SECTAV(N1,N2,A,AV)

```

```

C
C
C
      FIND THE MEAN VALUE OF THE DATA
      BETWEEN POINTS N1 AND N2.

```

```

  DIMENSION A(1024)
  AV=0.
  DO 10 I=N1,N2
10 AV=AV+A(I)
  AV=AV/FLOAT(N2-N1)
  RETURN
  END

```

```

  SUBROUTINE SMEAN(NPTS,DATA,RMEAN)

```

```

C
C
C
      REMOVE THE MEAN VALUE FROM THE DATA.

```

```

  DIMENSION DATA(NPTS)
  RMEAN=0.
  DO 20 I=1,NPTS
20 RMEAN=RMEAN+DATA(I)
  RMEAN=RMEAN/NPTS
  DO 30 I=1,NPTS
30 DATA(I)=DATA(I)-RMEAN
  RETURN
  END

```

```

  SUBROUTINE SSQRE(N,DATA,SUMSQ)

```

```

C
C
C
C
      CALCULATE THE SUMS OF THE SQUARES
      OF THE SIGNAL AND DIVIDE BY THE
      NUMBER OF POINTS.

```

```

  DIMENSION DATA(N)
  SUMSQ=0.
  DO 200 I=1,N
200 SUMSQ=SUMSQ+(DATA(I))*(DATA(I))
  SUMSQ=SUMSQ/FLOAT(N)
  RETURN
  END

```

```

  SUBROUTINE STDMN(N,DATA,MEAN,STDEV)

```

```

C
C
C
      CALCULATE THE MEAN AND SD.

```

```

  REAL MEAN
  DIMENSION DATA(N)
  SUM=0.
  DO 100 I=1,N
100 SUM=SUM+DATA(I)
  MEAN=SUM/FLOAT(N)
  SUM=0.
  DO 200 I=1,N
200 SUM=SUM+(DATA(I)-MEAN)*(DATA(I)-MEAN)
  STDEV=SQRT(SUM/FLOAT(N-1))
  RETURN
  END

```

```

  SUBROUTINE TAPER2(X,N)

```

```

C
C
C
      TAPERS ARRAY X AND SUBTRACTS MEAN

```

```

  DIMENSION X(N)
  DOUBLE PRECISION SUM1,SUM2,SUM3

```

```

C
SUM1=0.0
SUM2=0.0
SUM3=0.0
DO 1 I=1,N
W=WINDY(I,N)
SUM1=SUM1+W
SUM3=SUM3+W*W
1 SUM2=SUM2+W*X(I)
SUM2=SUM2/SUM1
SUM3=SQRT(N/SUM3)
DO 2 I=1,N
2 X(I)=(X(I)-SUM2)*SUM3*WINDY(I,N)
RETURN
END

FUNCTION WINDY(J,N1)
PARAMETER (PI=3.1415926536)
TL=0.125
AN=FLOAT(N1)
AN1=AN-1

C
C
C TL IS THE TAPER LENGTH ( PER UNIT )

N1L=IFIX(TL*AN + 0.5)
WINDY=1.0
AJ=J-0.5
IF(J .GE. N1L .AND. J .LE. (N1-N1L))RETURN
WINDY=(1. - COS(PI*AJ/(AN1*TL)))/2.
IF(J .GT. (N1-N1L))
+WINDY=(1. + COS(PI*(AJ+N1L-AN1)/(AN1*TL)))/2.
RETURN
END

SUBROUTINE TITLE(FNAME,GENINF,DEVICE)
C
C
C PUTS A TITLE AND FILENAME ON A GRAPH.

INTEGER*2 FNAME(20),GENINF(40),ODEV,DEVICE,NLENSA,GL
COMMON ODEV

C
C
C TEST FOR A VALID DEVICE CODE
CURRENTLY 0=TEKTRONIX
1=CALCOMP 536 VIA 906
2=SIGMA 5660 (5670)
3=CALCOMP 81

IF(DEVICE .GT. 3)GO TO 999

C
C
C SKIP THE INITIALISATION IF THIS
DEVICE ALREADY BEEN INITIALISED

IF(DEVICE .EQ. ODEV)GO TO 50
IF(DEVICE .LT. 0)GO TO 200

C
C
C DE-ASSIGN CURRENT DEVICE IF
DIFFERENT DEVICE REQUESTED.

IF(ODEV .NE. 99)CALL DEVEND

C
C
C ASSIGN REQUESTED DEVICE

IF(DEVICE .EQ. 0)CALL T4010
IF(DEVICE .EQ. 1)CALL CC906
IF(DEVICE .EQ. 2)CALL S5660
IF(DEVICE .EQ. 3)CALL CC81
IBAUD=1200
CALL DEVSPE(IBAUD)
IF(DEVICE .EQ. 0)CALL UNITS(1.0)
IF(DEVICE .EQ. 1)CALL UNITS(1.0)
IF(DEVICE .EQ. 2)CALL UNITS(1.5)
IF(DEVICE .EQ. 3)CALL UNITS(0.75)
50 CALL PICCLE
CALL WINDO2(0.,240.,0.,185.)
CALL WINDOW(1)
CALL MOVTO2(80.,180.)
CALL CHAHAR(1,0)
CALL CHAHOL('DATA FILE: *.')
NL=IFIX(FLOAT(NLENSA(FNAME,40))/2. + 0.5)
CALL CHARR(FNAME,NL,2)

```

```

CALL MOVTO2(0.,174.)
GL=IFIX(FLOAT(NLENSA(GENINF,80))/2. + 0.5)
CALL CHAARR(GENINF,GL,2)
CALL MOVTO2(0.,170.)
CALL CHAMOD
ODEV=DEVICE
RETURN
200 ODEV=99
CALL DEVEND
RETURN
999 WRITE(1,1000)DEVICE
1000 FORMAT('INVALID PLOTTING DEVICE CODE ',I5)
STOP 1
END

```

```

SUBROUTINE VSTAT1(ANGLE,RAD,N)

```

```

C THIS SUBROUTINE CALCULATES SUMMARY STATISTICS FOR THE N VECTORS
C WHOSE DIRECTIONS ARE STORED IN ARRAY 'ANGLE' AND WHOSE MAGNITUDES
C ARE STORED IN ARRAY 'RAD'.

```

```

C RESULTANT DIRECTION .... THETA=ATAN(S/C)

```

```

C WHERE S = WEIGHTED AVERAGE OF SINE VALUES
C C = WEIGHTED AVERAGE OF COSINE VALUES
C AND THE WEIGHTING FACTORS ARE THE VECTOR MAGNITUDES.

```

```

C DISPERSION FACTOR .... UO=1-SQRT(S*S+C*C)

```

```

C UO HAS VALUE 1 FOR A ZERO MAGNITUDE RESULTANT VECTOR
C 0 FOR A SET OF ALIGNED VECTORS

```

```

C WRITTEN BY TERRY JOHNSON. DEPT. OF MATHS. STATS. & COMPUTING
C PLYMOUTH POLYTECHNIC.

```

```

C DIMENSION ANGLE(N),RAD(N)
C DATA PI/3.1415926536/
C C=0
C S=0
C SUMRAD=0.0
C DO 1 I=1,N
C SUMRAD=SUMRAD+RAD(I)
C C=C+RAD(I)*COS(ANGLE(I))
1 S=S+RAD(I)*SIN(ANGLE(I))
C C=C/SUMRAD
C S=S/SUMRAD
C THETA=ATAN2(S,C)
C UO=1-SQRT(S*S+C*C)
C WRITE(1,100) THETA
C THETA=THETA*180/PI
C WRITE(1,101) THETA
C WRITE(1,102) UO
100 FORMAT('RESULTANT DIRECTION',8X,'=',F10.5,' RADIANS')
101 FORMAT(27X,'=',F10.5,' DEGREES')
102 FORMAT('DISPERSION FACTOR',10X,'=',F10.5)
C RETURN
C END

```

```

SUBROUTINE VSTAT2(ANGLE,RAD,N)

```

```

C THIS SUBROUTINE EXAMINES THE N VECTORS WHOSE DIRECTIONS ARE STORED
C IN ARRAY ANGLE AND WHOSE MAGNITUDES ARE STORED IN ARRAY RAD. THE
C PROGRAM IDENTIFIES THE (N/2) VECTORS WHICH LIE IN THE SMALLEST ARC
C AND COMPARES THE AVERAGE LENGTH OF THESE VECTORS WITH THE AVERAGE
C LENGTH OF THE REMAINING VECTORS. COMPARISON IS MADE BY A T-TEST
C WITH AN APPROXIMATE CORRECTION FOR (POSSIBLE) UNEQUAL VARIANCES.

```

```

C WRITTEN BY TERRY JOHNSON. DEPT. OF MATHS. STATS. & COMPUTING
C PLYMOUTH POLYTECHNIC.

```

```

C DIMENSION ANGLE(N),RAD(N),TEMP(100)
C DOUBLE PRECISION SUM11,SUM12,SUM21,SUM22
C DATA PI/3.1415926536/
C STORES ANGLES IN A TEMPORARY ARRAY AND ARRANGES IN ASCENDING
C ORDER USING A BUBBLE SORT.

```

```

C
1 DO 1 I=1,N
  TEMP(I)=ANGLE(I)
  DO 2 I=2,N
    MAX=1
    AMAX=TEMP(1)
    NMAX=N-I+2
    DO 3 J=2,NMAX
      IF(TEMP(J).LE.AMAX) GO TO 3
      AMAX=TEMP(J)
      MAX=J
    3 CONTINUE
    TEMP(MAX)=TEMP(NMAX)
    TEMP(NMAX)=AMAX
  2
C
C FIND THE (N/2) VALUES LYING IN THE SMALLEST ARC.
  N2=N/2
  DMIN=TEMP(N2)-TEMP(1)
  AMIN=TEMP(1)
  AMAX=TEMP(N2)
  DO 4 I=2,N
    IU=I+N2-1
    IF(IU.LE.N)UPPER=TEMP(IU)
    IF(IU.GT.N)UPPER=TEMP(IU-N)+2.*PI
    IF((UPPER-TEMP(I)).GE.DMIN)GO TO 4
    DMIN=UPPER-TEMP(I)
    AMIN=TEMP(I)
    AMAX=UPPER
  4 CONTINUE
C
C CALCULATES MEAN AND VARIANCE FOR BOTH SETS OF DATA
  SUM11=0.0
  SUM12=0.0
  SUM21=0.0
  SUM22=0.0
  N2=0
  N21=0
  CONST=0.
  IF(AMAX.GT.PI) CONST=2*PI
  DO 5 I=1,N
    A=ANGLE(I)
    IF(A.LT.0.)A=A+CONST
    IF(A.LT.AMIN.OR.A.GT.AMAX)GO TO 6
    N2=N2+1
    SUM11=SUM11+RAD(I)
    SUM12=SUM12+RAD(I)*RAD(I)
    GO TO 5
  6 SUM21=SUM21+RAD(I)
    N21=N21+1
    SUM22=SUM22+RAD(I)*RAD(I)
  5 CONTINUE
  SUM11=SUM11/N2
  SUM21=SUM21/N21
  SUM12=(SUM12/N2-SUM11*SUM11)/(N2-1)
  SUM22=(SUM22/N21-SUM21*SUM21)/(N21-1)
C
C CALCULATES T-STATISTIC AND OUTPUTS RESULTS
  TSTAT=SNGL((SUM11-SUM21)/DSQRT(SUM12+SUM22))
  DF=(SUM12+SUM22)*(SUM12+SUM22)
  DF=DF/(SUM12*SUM12/(N2+1)+SUM22*SUM22/(N21+1))-2
  SUM12=DSQRT(SUM12*N2)
  SUM22=DSQRT(SUM22*N21)
  WRITE(1,100)N2,SUM11,SUM12
  WRITE(1,101)N21,SUM21,SUM22
  WRITE(1,102)TSTAT,DF
100 FORMAT(/'MEAN AND S.D. OF LENGTHS OF THE',I4,' CLOSEST '
+ 'VECTORS =',E16.8,E16.8)
101 FORMAT('MEAN AND S.D. OF LENGTHS OF THE',I4,' REMAINING '
+ 'VECTORS =',E16.8,E16.8)
102 FORMAT(/'T-STATISTIC =',F10.5,' WITH ',F6.1,
+ ' DEGREES OF FREEDOM')
  RETURN
  END
  SUBROUTINE VSTAT3(ANGLE,RAD,N)
C
C THIS SUBROUTINE CALCULATES A 'MODIFIED DISPERSION' STATISTIC FOR

```

```

C 'N' VECTORS WHOSE DIRECTIONS ARE STORED IN ARRAY 'ANGLE' AND WHOSE
C MAGNITUDES ARE STORED IN ARRAY 'RAD'.
C
C MODIFIED DISPERSION FACTOR ..... UM=1-SQRT(S*S+C*C)
C
C WHERE S= WEIGHTED AVERAGE OF SINE VALUES
C        C= WEIGHTED AVERAGE OF COSINE VALUES
C
C AND THE WEIGHTING FACTORS ARE THE RANK ORDERS OF THE VECTORS
C
C UM HAS VALUE 1 FOR A ZERO MAGINTUDE RESULTANT VECTOR
C           0 FOR A SET OF ALIGNED VECTORS
C
C .....
C
C WRITTEN BY TERRY JOHNSON. DEPT. OF MATHS. STATS. & COMPUTING
C PLYMOUTH POLYTECHNIC.
C
C   DIMENSION ANGLE(N),RAD(N),TEMPA(100),TEMPR(100)
C   DO 1 I=1,N
C   TEMPA(I)=ANGLE(I)
1  TEMPR(I)=RAD(I)
C   C=0.0
C   S=0.0
C   DO 2 I=2,N
C   MAX=1
C   AMAX=TEMPR(1)
C   NMAX=N-I+2
C   DO 3 J=2,NMAX
C   IF (TEMPR(J).LE.AMAX) GO TO 3
C   AMAX=TEMPR(J)
C   MAX=J
3  CONTINUE
C   C=C+NMAX*COS(TEMPA(MAX))
C   S=S+NMAX*SIN(TEMPA(MAX))
C   TEMPR(MAX)=TEMPR(NMAX)
2  TEMPA(MAX)=TEMPA(NMAX)
C   C=C+COS(TEMPA(1))
C   S=S+SIN(TEMPA(1))
C   SUMN=N*(N+1)/2
C   C=C/SUMN
C   S=S/SUMN
C   UM=1.0-SQRT(S*S+C*C)
100 WRITE(1,100) UM
C   FORMAT(/,'MODIFIED DISPERSION FACTOR = ',F10.5)
C   RETURN
C   END

```