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# A Performance Enhancement for 60 GHz Wireless Indoor Applications

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# Overview of 60 GHz WPAN

## Standards over 60 GHz WPAN

- IEEE 802.15.3c
- WirelessHD
- WiGig
- ECMA-387
- IEEE 802.11ad

## Characteristics of 60 GHz millimeter-wave WPANs

- In-door (<10m)
- Uncompressed HDTV and high rate data transfer
- At least 1 Gbps throughput, 3-4 Gbps preferable

# Operating Modes

- Single Carrier: Low complexity and control information
- OFDM: High performance applications

**Table:** Parameters for OFDM Systems in IEEE 802.11ad

Parameter	Value
Sampling frequency (MHz)	2640
Number of subcarriers	512
Number of data subcarriers	336
Number of pilot subcarriers	16
Subcarrier frequency spacing (MHz)	5.156
Sample duration (ns)	0.38
IFFT and FFT period (ns)	194
OFDM symbol duration (ns)	242

# MIMO-OFDM Communication Model

Let  $y_m$  be the received decision baseband signal for the  $m$ th subcarrier

$$y_m = \tilde{H}_m x_m + n_m, \quad m = 1, \dots, N$$

where  $\mathbf{x}_m$  is the transmitted data symbol,  $\mathbf{n}_m$  is the Gaussian noise vector with zero mean and variance  $\sigma^2$ ,  $\mathbf{N}$  is the number of subcarriers, and  $\tilde{H}_m$  represents the frequency response of the equivalent channel matrix for the  $m$ th subcarrier.

# Maximizing Spatial Diversity

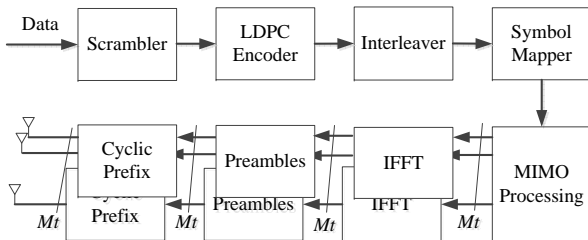


Figure: Block diagram of MIMO-OFDM Transmitter

## Space-Time Block Coding

- Enables linear decoding at the receiver
- Transmission matrix  $\begin{bmatrix} x_1 & -x_2^* \\ x_2 & x_1^* \end{bmatrix}$  for a  $2 \times 2$  architecture

# Increasing Spectral Efficiency

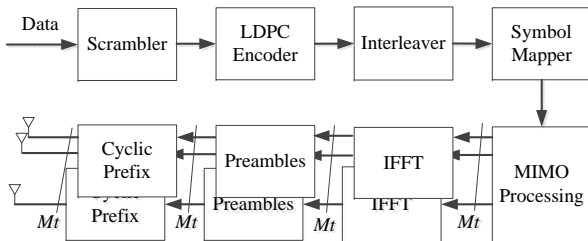


Figure: Block diagram of MIMO-OFDM Transmitter

## Spatial Multiplexing

- Doubles the peak data rate for a  $2 \times 2$  architecture
- Increase the reliability and throughput for lower modes
- Both STBC and SM need an FFT/IFFT per antenna



# Optimization Criteria

Recall

$$y_m = \tilde{H}_m x_m + n_m, \quad m = 1, \dots, N$$

Here the frequency response of the equivalent channel matrix for the  $m$ th subcarrier after beamforming  $\tilde{H}_m$  can be given by:

$$\tilde{H}_m = \mathbf{c}^H \mathbf{H}_m \mathbf{w}, \quad m = 1, \dots, N$$

$\mathbf{w}$  and  $\mathbf{c}$  are the transmitter and the receiver beam steering vector respectively, and  $\mathbf{H}_m$  is the response of the MIMO channel for the  $m$ th subcarrier.

# Optimization Criteria

## Maximize Effective SNR

$$\gamma_{\text{eff}} = -\beta \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp(-\gamma_m/\beta) \right]$$

where  $\gamma_m$  is the symbol SNR experienced on the  $m$ th subcarrier,  $\beta$  is a parameter dependent on MCS.

$$\gamma_m = \frac{E \left[ |c^H H_m w x_m|^2 \right]}{E \left[ |n_m|^2 \right]} = \frac{|c^H H_m w x_m|^2}{M_t M_r \sigma^2}$$

where  $M_t$  and  $M_r$  are the number of antenna elements at the transmitter and the receiver respectively. When normalized,  $w^H w = M_t$  and  $c^H c = M_r$

# Optimization Criteria

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where  $M_t$  and  $M_r$  are the number of antenna elements at the transmitter and the receiver respectively. When normalized,  $w^H w = M_t$  and  $c^H c = M_r$ .

# Subcarrier-wise: Maximize SNR on Each Subcarrier

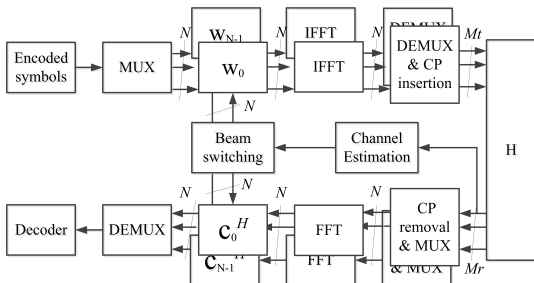


Figure: Block diagram of subcarrier-wise beamforming

$$\gamma_{eff,subcarrier} = -\beta \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp \left( -\frac{\max_{C,w} |C^H H_m w|^2}{\beta M_t M_r \sigma^2} \right) \right]$$

# Subcarrier-wise: Maximize SNR on Each Subcarrier

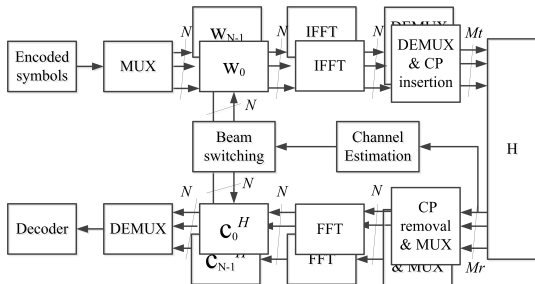


Figure: Block diagram of subcarrier-wise beamforming

Optimal but not practical

- Need full channel state information
- Requires one FFT/IFFT processor per antenna

# Symbol-wise: Applies the Same Weight Vector

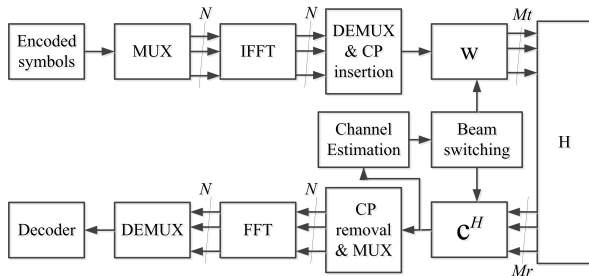


Figure: Block diagram of symbol-wise beamforming

## Pre-defined beam codebook

- Full channel state information is not required
- Depends on the number of antenna elements and beams

# Symbol-wise: Applies the Same Weight Vector

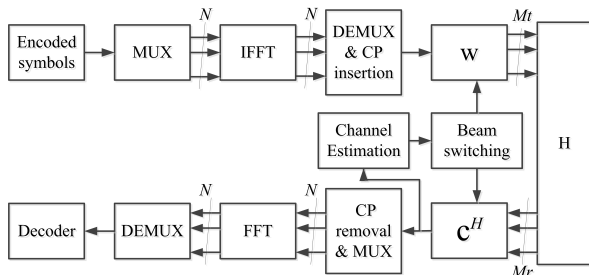


Figure: Block diagram of symbol-wise beamforming

$$\gamma_{eff, symbol} = \max_{c, w \in C} (-\beta) \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp \left( -\frac{|c^H H_m w|^2}{\beta M_t M_r \sigma^2} \right) \right]$$

# Hybrid: Compromise the Complexity and Performance

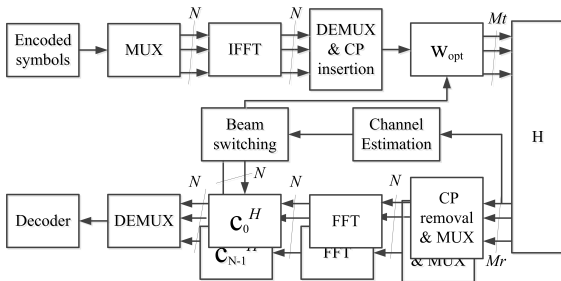


Figure: Block diagram of hybrid beamforming

Symbol-wise at Tx, and subcarrier-wise at Rx

- Optimal each receiver steering vector
- Also use pre-defined codebook



# Hybrid: Compromise the Complexity and Performance

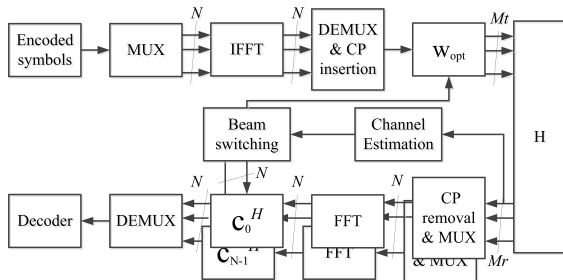


Figure: Block diagram of hybrid beamforming

$$\gamma_{eff, hybrid} = \max_{w \in \mathcal{C}} (-\beta) \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp \left( - \frac{|c_{opt}^H H_m w|^2}{\beta M_t M_r \sigma^2} \right) \right]$$

# Medium Access Control Layer

## Hybrid Access

- CSMA/CA: Lower average latency (web browsing)
- TDMA: Better QoS (video transmission)

## Sources of Overhead

- Preamble
- Header
- Gap Time
- Acknowledgment Frames



# Block ACK and Block NAK

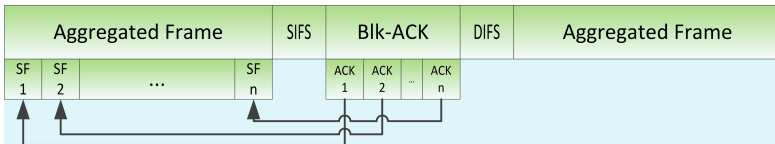


Figure: Blk-ACK

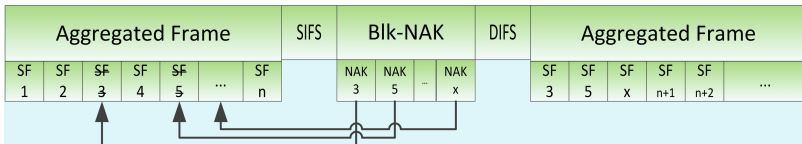


Figure: Blk-NAKs

# Preliminaries

## System Assumptions

- 1D uniform linear array
- $M_t = M_r = 2$  antenna elements
- Half wavelength isotropic radiators

## Channel Assumptions

- Statistic channel from measurements and ray-tracing
- Channel correlation 0.1 (low), 0.5 (medium) and 0.9 (high)
- Both LOS and NLOS

# Preliminaries

## Simulation Setup

- Packet Size: 1KB
- Channel Coding: LDPC
- PER target: 1%
- Cyclic Prefix: 128

**Table:** OFDM Modulation and Coding Schemes

Modulation	Coding Rate	Coded Bits/Symbol	Data Bits/Symbol	Data Rate (Mbps)	SM Data Rate (Mbps)
QPSK	1/2	672	336	1386.00	2772.00
QPSK	5/8	672	420	1732.50	3465.00
QPSK	3/4	672	504	2079.00	4158.00
16-QAM	1/2	1344	672	2772.00	5544.00
16-QAM	5/8	1344	840	3465.00	6930.00
16-QAM	3/4	1344	1008	4158.00	8316.00
16-QAM	13/16	1344	1092	4504.50	9009.00
64-QAM	5/8	2016	1260	5197.50	10395.00
64-QAM	3/4	2016	1512	6237.00	12474.00
64-QAM	13/16	2016	1638	6756.75	13513.50



# NLOS Scenario

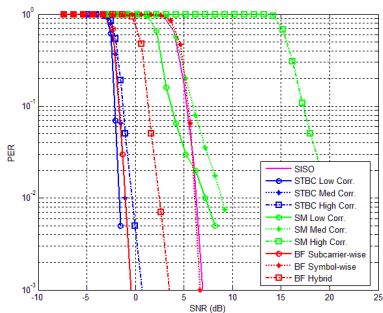


Figure: PER comparison with NLOS

- STBC and SM performance varies depending on the correlation factors
- STBC offers a PER gain of 7-8.5 dB
- SM requires higher SNR than SISO but doubles the data rate
- Hybrid beamforming achieves 4 dB gain

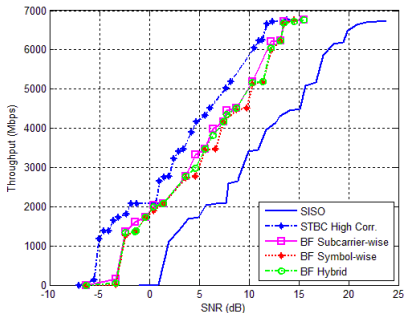


# Link Throughput in LOS

## Link Adaptation Scheme

- The PHY mode with highest throughput will be selected:

$$\text{Throughput} = R(1 - \text{PER})$$



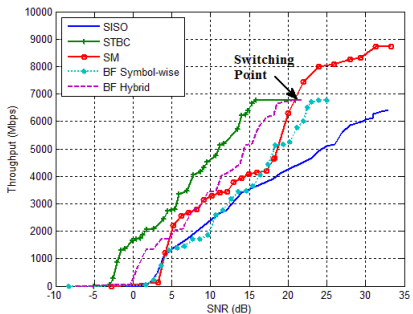
- The throughput envelope is the ideal adaptive MCS based on the optimum switching point
- At a certain SNR, MIMO systems outperform SISO system

# Link Throughput in NLOS

## Link Adaptation Scheme

- The PHY mode with highest throughput will be selected:

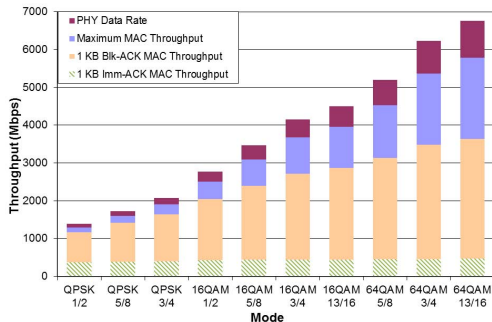
$$\text{Throughput} = R(1 - \text{PER})$$



- STBC and hybrid beamforming provide 2-6 dB gain
- More gain can be achieved for very high throughput (>4500 Mbps)
- After the switching point at 21 dB, SM is the best



# Max Throughput Achieved for Each Mode



- Imm-ACK does not depend on the mode
- While Blk-ACK varies depending on PHY mode
- Imm-ACK efficiency is 6.9%-26%, and Blk-ACK improves by 3-8 times

Figure: Max Throughput for Each Mode

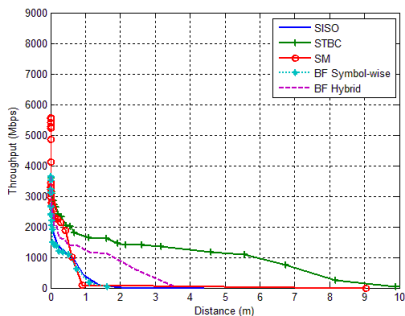




# Operation Range in NLOS

## Link Budget Model

$$P_T - PL \geq kTB + NF + ReceiverSNR$$



- The SISO system could not provide service beyond 1m
- Hybrid beamforming extend the achievable range to 3.5m, and STBC is possible to provide service up to 10m

Figure: Operation range in NLOS

# Summary

- STBC produces the best performance due to its robustness in all conditions; While SM doubles the error-free data rate and increase the reliability for lower MCS modes;
- Beamforming increases the performance significantly. In NLOS, hybrid beamforming provides considerable improvements while maintaining reasonable hardware complexity.
- Frame aggregation and Blk-ACK increase the MAC throughput 3-8 times compared to Imm-ACK



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# Thank you! and Questions?

please Email to <x.zhu@Bristol.ac.uk>