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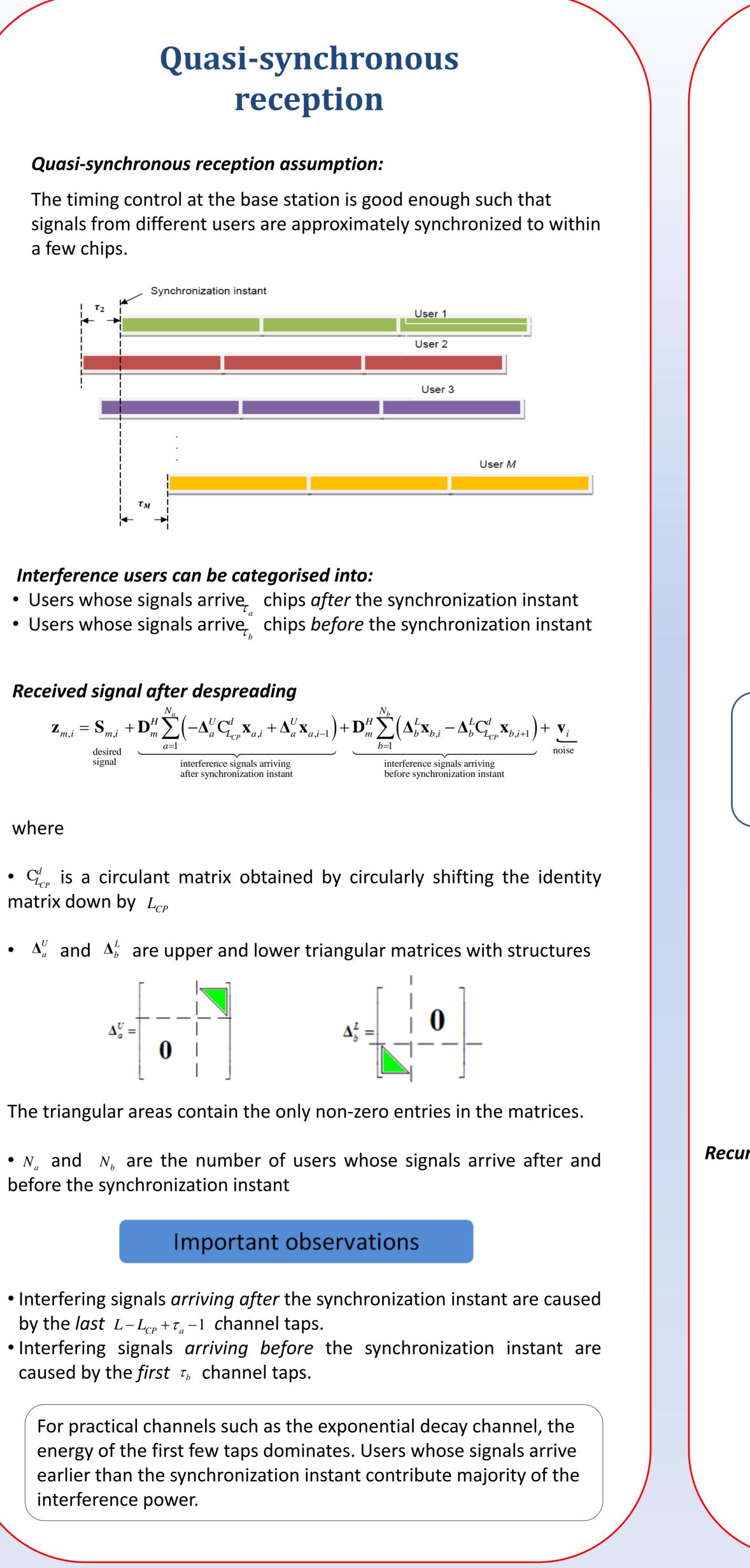
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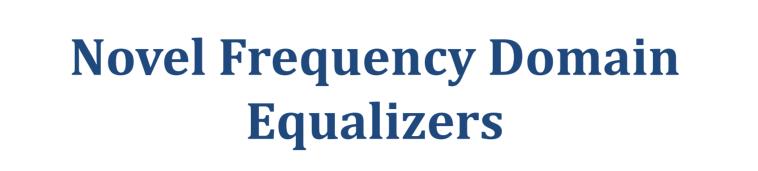
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Overview	
Block spread CDMA (BS-CDMA) [1] is a multi-access technique whereby user-specific precoding along with orthogonal spreading codes are used to achieve multiuser interference (MUI) free reception when signals of all users arrive at the base station synchronously.	
In practice however, imperfect synchronization destroys the orthogonality among users and MUI occurs. This paper investigates the design of linear frequency domain equalizers to reduce the MUI for a quasi-synchronous BS-CDMA system. An optimal frequency domain linear minimum-mean squared error (LMMSE) equalizer is derived. Further simplification leads to a novel sub-optimal equalizer with reduced computational complexity. It is shown through simulation that the proposed equalizers effectively suppress the error floor due to quasi-synchronous reception when channel coding is used.	
BS-CDMA System Model	
BS-CDMA spreads a block of symbols with user specific spreading codes.	
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Transceiver structure [1] :	
bits $\overbrace{\text{interleaver, symbol mapper}}^{\text{Encoder,}} s_{\mu}(k)$ S/P $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ $\overbrace{\text{S/P}$ $\overbrace{\text{S/P}}$ \text	
a) transmitter	
\mathbf{v}_{i} Remove guard interval \mathbf{v}_{i} Block despreading ($\mathbf{c}_{m,i}$) Block decoder ($\Gamma_{m,i}$) Equalizer \mathbf{P}/\mathbf{S} Demapper, deinterleaver, deinterleaver, decoder decoder	
b) receiver	
Features of synchronous BS-CDMA	
Achieves MUI-free reception by using mutually shift-orthogonal spreading codes such as the discrete Fourier transform (DFT) codes.	
Achieves higher bandwidth efficiency by inserting cyclic prefix after block spreading	
Conventional:	
Bandwidth efficient BS-CDMA:	

Linear Equalizers for Quasi-Synchronous Block Spread CDMA Systems

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Optimal LMMSE Equalizer

• Objective: to minimise the mean square error (MSE) between the transmitted and recovered signal, given by

$$= \operatorname{Tr}\left\{ \operatorname{E}\left[\left(\mathbf{s}_{m,i} - \mathbf{w}_{m,i} \right) \left(\mathbf{s}_{m,i} - \mathbf{w}_{m,i} \right)^{H} \right] \right\}$$

where $\mathbf{W}_{m,i}$ is the data block after equalization.

• Obtained by solving $\partial \dot{o} / \partial \mathbf{G}_{out}^* = 0$

Calculation of the optimal LMMSE equalizer requires matrix inverse -- high computational complexity!

Suboptimal Equalizer

- *Objective:* To reduce the computational complexity of the optimal LMMSE equalizer
- Obtained by taking expectation over the channel taps as well
- Can be computed recursively by using the matrix inverse lemma

Matrix inverse lemma:

If a matrix is defined as: $C = A + xy^{H}$

Then its inverse can be computed as: $C^{-1} = A^{-1} - \frac{A^{-1}xy^{H}A^{-1}}{1+y^{H}A^{-1}x}$

$$\mathbf{G}_{subopt}^{*} = M \boldsymbol{\sigma}_{m,i}^{2} \boldsymbol{\Xi}_{m}^{H} \left[\boldsymbol{\Pi} + \sum_{k=1}^{K} \boldsymbol{\alpha}_{k} \mathbf{f}_{k} \mathbf{f}_{k}^{H} \right]^{-1}$$

The number of users
Signal energy
$P \times P$ Diagonal matrices determined by the
channel impulse response
Random variables determined by channel
statistics
<i>kth</i> column of the DFT matrix

Recursive computation of the matrix inverse:

Initialization: $\mathbf{A} = \mathbf{\Pi}$ For k = 1: K• $\mathbf{C} = \mathbf{A} + \alpha_k \mathbf{f}_k \mathbf{f}_k^H$ • $\mathbf{C}^{-1} = \mathbf{A}^{-1} - \alpha_k \frac{\mathbf{A}^{-1} \mathbf{f}_k \mathbf{f}_k^H \mathbf{A}^{-1}}{1 + \mathbf{f}_k^H \mathbf{A}^{-1} \mathbf{f}_k}$ • $\mathbf{A} = \mathbf{C}^{-1}$ End For

Complexity decreases from $\mathcal{O}(P^3)$ to $\mathcal{O}(P^2)$

Simulation parameters: QPSK modulation Block length of 16 **Observations:** • 8 active users the coded systems **Observations:** coding

