

Portable Low-Power Electronic Interface For Explosive Detection Using Microcantilevers

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Abstract

Microcantilevers have been recently revealed as highly effective sensing platforms for gas detection at trace level. However, one of the main limitations hindering their widespread deployment for commercial applications is related to the complexity of the read-out systems. A portable low-power electronic interface capable of creating the excitation signal as well as obtaining the response values of resonating Si microcantilevers functionalized with zeolite based coatings has been developed in this work. The accuracy and reliability of the electronic interface have been assessed by comparison with commercial lock-in amplifier's measurements. The prototype performance has been validated for 2-nitrotoluene (o-MNT) detection at ppm level, as an example of an explosive-related molecule. Theoretical limit of detection (LOD) values below 100 ppb have been obtained for Co exchanged BEA modified microcantilevers.

Keywords

Microcantilevers; gas sensing; zeolites; nanoporous coating; explosives; portable lock-in amplifier; low-power.

1. Introduction

Recent advances in chemical microcantilever-based sensors have led to a significant increase in sensitivity, making them a competitive solution in highly demanding applications as food and environmental security or explosives detection due to its ability to determine the presence and concentration of target gas species and volatile organic compounds (VOCs) [1-4]. Standard silicon microcantilevers working as tiny microbalances [5] achieve theoretical mass detection limits of around tens of fg/Hz. For dynamic mode operation, the analyte adsorption over the sensing layer deployed on top of the beam provokes an effective mass variation, which in turn modifies its natural oscillation frequency allowing sensitivities in the order of ppm [6-9]. Even so, its application on real multicomponent gas mixtures is challenging due to selectivity and cross-interference effects. These limitations are usually overcome by employing an array of sensors coated with different sensing materials. Thus, the presence and concentration of the

target compounds can be determined through an appropriate pattern recognition algorithm with suitable accuracy and precision [10].

To accomplish miniaturized chemical microcantilever-based sensor arrays, the use of silicon provides important advantages in terms of shrinking power consumption and size. Our previous works [11-12] have shown that microcantilever type sensors coated with suitable zeolites leads to substantial outperformance in terms of sensitivity and selectivity.

Particularly, for early explosives detection, the main problem is related to their low vapor pressure in the pure form (i.e. between 10^{-2} Pa to 10^{-7} Pa at 25°C). Thus, minor variations in the microcantilever's resonance frequency are expected, making difficult its subsequent readout. In this scenario, lock-in amplifiers (LIAs) are the most common solution on behalf of its versatile function recovering low and noisy signals [13-14]. Lock-in amplifiers are signal processing systems that use the phase sensitive detection (PSD) [15-16] technique to recover the data signal at a specific reference frequency; and reject noise signals at different frequencies without affecting the measurement significantly [17-18]. Nowadays, commercial lock-in amplifiers are easily found in any measurement laboratory. However, weight, cost and power consumption are the main drawbacks hindering their widespread use for portable sensing systems.

Thus, to exploit the truly benefits of the microcantilever-based sensor arrays in terms of portability, reduced size and energy consumption; an electronic readout interface showing similar features is required. In addition, the readout system must be flexible enough to be applied to the different elements constituting the sensor array. In fact, this is the main goal of the work. Firstly, the portable low-power electronic readout interface capable to measure Si microcantilever-based sensor output signals, with a precision up to 20 mHz for the oscillation frequency measurement is presented. Thus, the microcantilever-like sensors used in this work are briefly described in Section 2. This section also highlights the electrical sensor requirements that the prototype must fulfill for a proper function. The design of the electronic interface powered through a universal serial bus (USB) is addressed in Section 3; where the analog lock-in amplifier, denoted as LIA for the remainder of the work, able to process square signals is fully explained. Such key element allows the read out of the low signals provided by the sensing platform. The experimental protocols either for determination or tracking the resonance frequency are described in Section 4 with special emphasis on the Proportional-Integral-Derivative (PID) algorithm for high resolution measurements. The attained results are compared in Section 5 with those obtained with a commercial LIA model 7265 from Signal Recovery. Finally, the developed prototype is applied for 2-nitrotoluene detection, an explosive related molecule, with BEA zeolite coated Si microcantilevers. The evaluated performance demonstrates the potentialities of the sensing platform for detection at ppmV level.

2. Microcantilever-based Gas Sensors

Figure 1.a illustrates the typical layout for a microcantilever electromagnetically actuated. The microdevice is subjected to a constant magnetic field (B_{ext}) originated by a permanent magnet placed below; then, an alternate current (I) passes through the microfabricated actuation wire, generating an alternate deflection force (F_L : Laplace force) that results into the mechanical

oscillation. This oscillation is detected by a strain gauge located in the anchorage of the microcantilever.

The resonant oscillation frequency of a microcantilever tip varies according to its mechanical features. Theoretical frequency f_o for a Si rectangular shape beam (Figure 1.a) can be calculated as follows:

$$f_o = r_n \cdot \frac{t}{L^2} \cdot \sqrt{\frac{E}{\rho}} \quad (1)$$

Where t is the device thickness, L is the length, E is the Young's modulus (for Si <100>, $E = 130$ GPa), ρ is the Si density (2330 kg/m^3) and r_n depends on vibrational mode (for fundamental mode $r_n = 0,1615$). Thus, the cantilever oscillation amplitude reaches a maximum when the frequency of the AC current flowing through the actuation wire corresponds to the cantilever resonant frequency. Otherwise, the oscillation amplitude is progressively damped.

In this work, the cantilever oscillation movement is measured by a strain gauge composed by two identical piezoresistive elements connected in a half Wheatstone bridge, biased with a constant voltage. One of the piezoresistors is placed on the rigid substrate, while the other is located in the clamped-end of the beam, where stress is maximum. In this way, the oscillation movement of the cantilever acts over the corresponding piezoresistor changing its value. Thus, a variable voltage is generated at the bridge output, which frequency corresponds to the mechanical oscillation frequency of the cantilever. The resonance frequency is detected by measuring the maximum voltage amplitude and its corresponding phase out at the output of the strain gauge. Figure 1.b illustrates the electrical equivalent of this microcantilever sensor. It requires an excitation periodic signal through the actuation wire and two fixed voltages to feed the strain gauges in order to generate the sensor output V_{sens} .

The Si microcantilever chip used in this work (Figure 1c) comprises 8-cantilevers, $200 \mu\text{m}$ wide and $15 \mu\text{m}$ thick, that have been coated with microporous zeolites. The deployment of suitable coatings on the mechanical transducers allows the adsorption of the target analytes. In particular, the commercial BEA type zeolites used in this work were provided by Zeolyst International (Si/Al ratio = 12,5; NH_4^+ form). Cobalt ion exchange was performed following the recipe described in our previous works [9-12]. The Si cantilevers were microfabricated by D+T Microelectrónica A.I.E, according to our own design and specifications [11-12]. One of their main features is the incorporation of a heater wire on the cantilever tip for the fast temperature cycling required to sorbent regeneration. As the length of the beam varies from $500 \mu\text{m}$ to $535 \mu\text{m}$, the exhibited resonance frequency values range from 90 to 65 kHz.

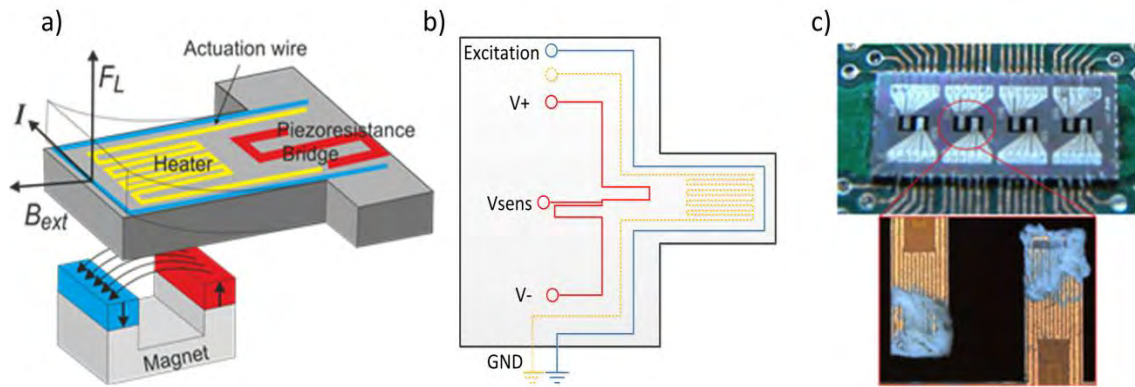


Figure 1. a) Schematic layout of dynamic operation mode for the cantilevers, with electromagnetic actuation, piezoresistive detection and integrated heater; b) Electrical diagram of the microcantilever sensor; c) Optical image of the 8 Si-microcantilevers array (200 μm wide, 15 μm thick, 500 μm to 535 μm length) used in this work, the inset corresponds to the Si microcantilever tip after deposition of zeolite crystals by microdrop coating technique.

3. Readout system

Commercial LIAs are commonly used for both generating the exciting signals as to perform the sensor readout for lab scale characterization. In general, these instruments provide the suitable exciting signal to the actuation wire. In addition, they are capable to perform an accurate reading of small and noisy signals at the sensor output; allowing a precise selection of the operation frequency window. Unfortunately, the use of commercial LIAs as readout devices in real applications using cantilever-based gas sensor arrays becomes unpractical due to size and energy requirements.

In our particular case, a suitable readout system must generate in parallel eight different exciting signals at specific frequency values; which in fact, are continuously modified as a function of the gas phase composition and operating temperature. Furthermore, the sensor's electrical outputs should be measured. The eight sensors could be excited and measured using just one LIA by input and output multiplexing, respectively. However, the selection of the working cantilever among those comprising the array results in de-synchronization between measurements. One possible solution could rely on the use of eight LIAs in parallel; but size, cost and power consumption make it unfeasible. To address this issue, Figure 2 shows the block diagram of the proposed electronic interface prototype. It comprises three main blocks: i) a power supply that generates the DC levels required by both the sensor and the electronic interface; ii) a readout system based on a compact LIA; and, iii) a microcontroller that manages the system operation: DC excitation for the piezoresistors, AC current bias for the actuation wire, system temperature control, frequency selection for the lock-in signals, LIA output reading and host computer communications. Each of these main building units will be discussed next.

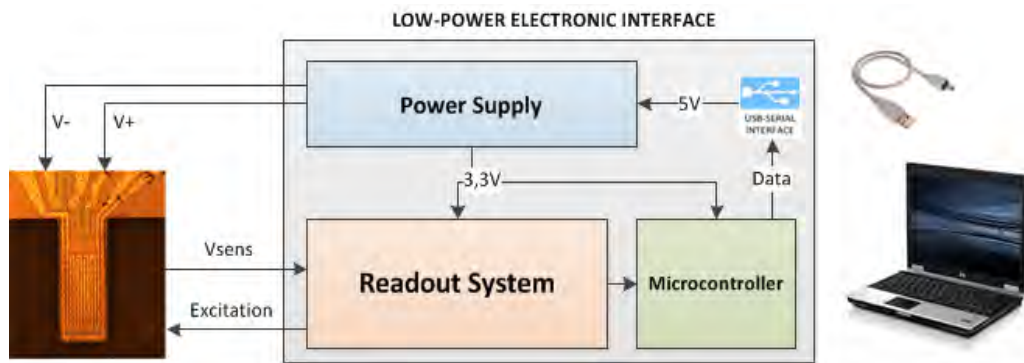


Figure 2. Block Diagram of the implemented low-power electronic interface.

Power supply

The 5 V line provided by the USB connector powers the prototype. Voltage levels are adapted using a voltage regulator that provides the 3,3 V suitable for the system electronics. Strain gauges use positive and negative DC bias voltage. Both levels are obtained using a LM2662 voltage doubler from Texas Instruments (± 9 V DC output) connected to an ASA00CC18-L DC-DC Converter from Emerson Network Power to generate DC levels of -15 V and +15 V (V^- and V^+ , Figure 2) to bias the Wheatstone bridge.

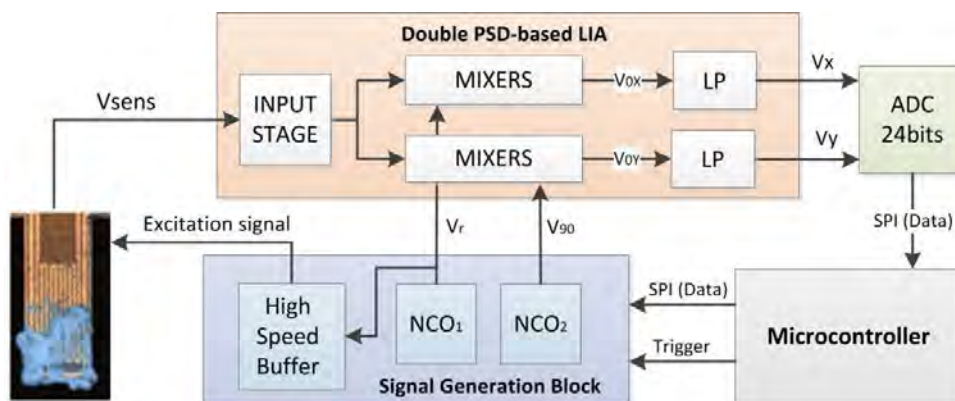


Figure 3. Block diagram of the readout block.

Readout Block

The core of the proposed portable readout block (Figure 3) is a low-voltage low-power dual PSD-based LIA [19-20] that carry out the measurement of the sensor output signal. An in-depth development of the dual PSD technique for processing of square signals, showing recovery equations for both amplitude and phase; and its experimental verification have been recently attempted by our group [21]. Besides, it comprises a signal generation block that generates the periodic signals at the operating frequency of the cantilever, and an ADC Stage for a precise digitization of the LIA outputs.

Some commercial integrated circuits, such as analog multipliers or demodulators [22], are amenable for assembly on a lock-in amplifier with appropriate additions. However, the complex configuration of the circuit, the supply voltages and consumption levels make them inappropriate for portable systems. Diverse interfaces suitable for sensor conditioning systems

[23] can be found in the published literature. In general, these systems are designed on ad-hoc basis [24], that means specific operation frequency windows and voltage output. To the best of our knowledge, a general purpose portable lock-in amplifier has not been previously attempted. A preliminary version of the novel architecture based on the application of a dual PSD technique processing square signals instead of sinusoidal counterparts, was advanced by the authors earlier in [19-20]. Additional efforts have resulted in the herein used quasi-digital LIA capable of recovering of both amplitude and phase by means of a novel fully tested and validated algorithm [21]. The use of square signals for exciting sensor reduces electronics complexity, and accordingly, size and power consumption. Eventually, depending on the required frequency resolution, the system microcontroller can act as a square wave generator. All these features are clearly advantageous for the LIA implementation in portable systems that utilize single-supply battery cells.

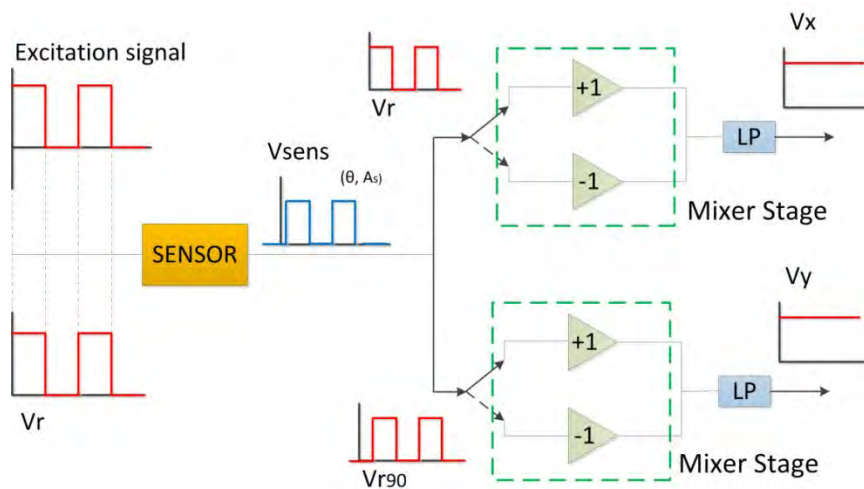


Figure 4. Block diagram of a two branches quasi-digital PSD system, each branch consisting of a mixer and low-pass filter.

In the designed LIA (Figure 4), the input signal provided by the sensor V_{sens} enters to the Input Stage for pre-conditioning. This signal reaches both Mixer Stages, which are controlled by two reference square signals in quadrature (V_r , V_{r90}). Each mixer consists of two ADG719 2:1 multiplexers and a unity gain differential amplifier AD8476 from Analog Devices. Thus, V_{ox} and V_{oy} are the outputs from the mixers at each branch. The final stage corresponds to two low pass filters providing the DC signals V_x and V_y that are required to obtain both magnitude A_s and phase shift θ . Low-pass filters are RC circuits with a cut-off frequency in the Hz range. These DC outputs V_x and V_y are fed to identical 24-bit LTC2400 low-power ADC converters from Linear Technology. The designed algorithm uses V_x and V_y to accurately calculate the output sensor signals parameters.

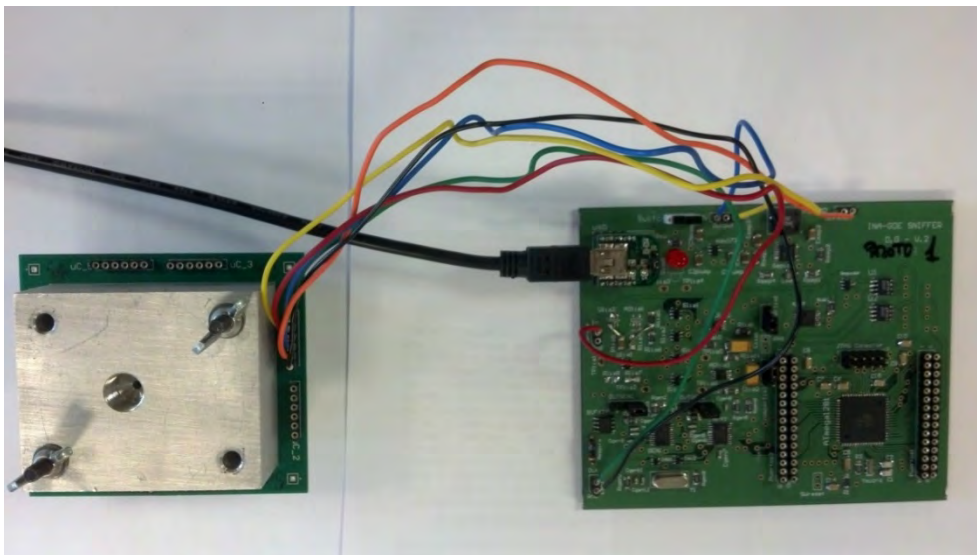
For microcantilever-based sensors operating in dynamic mode, an extremely accurate frequency determination becomes pivotal to ensure gas detection at low concentrations. Consequently, a specific signal generation block has been implemented to select the operation frequency window in a wide range with suitable accuracy and resolution. Signal generation is due to two AD9834 Numerically Controlled Oscillators (NCOs) from Analog Devices included in the prototype. NCOs are capable of generating square signals with frequencies from 1 to 100

kHz with a resolution of 20 mHz using a 10 MHz quartz crystal as external reference. NCOs parameters are configured by the system microcontroller. Outputs from NCOs are two square wave signals in quadrature (V_r , V_{r90} , Figure 4) that control the Mixer Stage of the implemented LIA. V_r is also used as excitation signal of the sensor. Due to the low input impedance presented by the microcantilevers, a high speed unity-gain open-loop buffer BUF634 from Burr-Brown is used as output stage (see Figure 3).

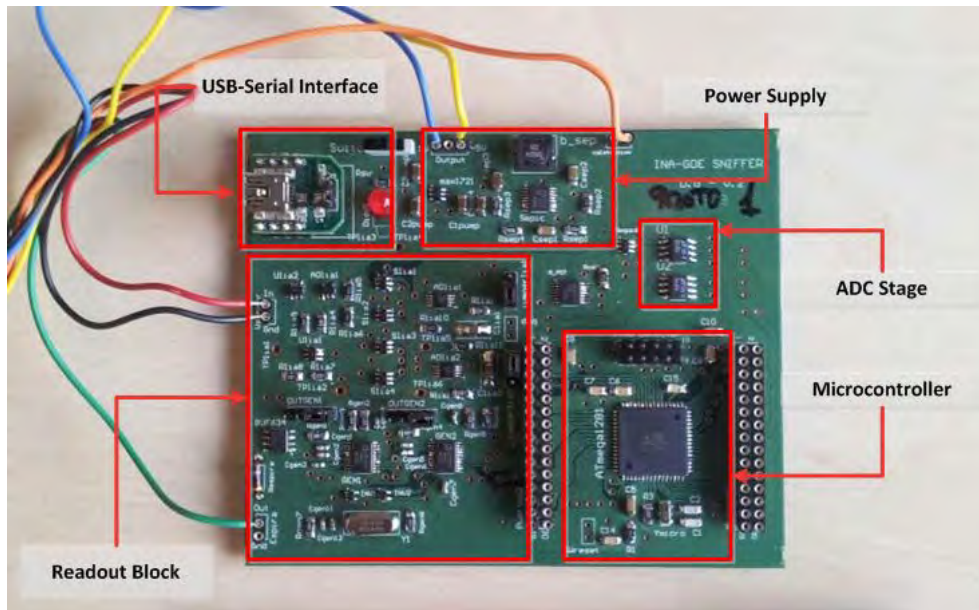
Microcontroller

The electronic interface is controlled by a low-cost low-power Atmega1281 microcontroller from Atmel. It manages both ADCs and NCOs operation by SPI communication [26], selecting the frequency and phase values to the NCOs output signals and reading the data provided by the ADCs. After every measuring cycle the data are transferred from the microcontroller to the host computer through a UB232R [25] Serial-to-USB adaptor from FTDI Chip.

Figure 5.a depicts the full system including the microcantilevers array connected to the herein presented interface. Figure 5.b shows a detailed image of the electronic interface. The consumption of the whole interface is 24 mA allowing the connection of more than 10 prototypes in parallel to the same USB port.



(a)



(b)

Figure 5. a) Portable Low-Power Electronic Interface for Si-microcantilever based sensors: a) gas chamber for the sensing platform connected to the implemented interface; b) detailed image of the electronic interface(110 mm x 90 mm).

4. Measurement Protocols

Natural frequency determination

According to the behavior of a resonant oscillating system, the oscillation amplitude is maximal at the natural frequency f_o , where the characteristic phase change is also located. Consequently, the experimental f_o value could be easily determined by carrying out a frequency sweep routine of both amplitude and phase parameters around the theoretical resonant frequency. Figure 6 shows a comparison of the amplitude (Figure 6.a) and phase (Figure 6.b) measurements of the microcantilever A5 (Table 1) performed with the proposed electronic interface and the commercial 7265 lock-in amplifier. Both systems have been configured to work at 1 step (data acquisition plus frequency actuation) per second. Note that in the proximity of the natural resonance, phase tracking is more sensitive than amplitude tracking for f_o determination. It is also quite remarkable the similarity between both results. The natural frequency evaluated with the commercial LIA processing sinusoidal signals is 81531 Hz and the measured by our prototype working with square waves is 81529 Hz. This deviation could be related to the differences in the sensor excitation signal: 1 Vrms vs. 3,3 V for commercial and portable LIA respectively. Furthermore, phase deviation out of the resonance region is due to the drastically reduced response of the sensor (Figure 6.a), that makes difficult a proper phase measurement.

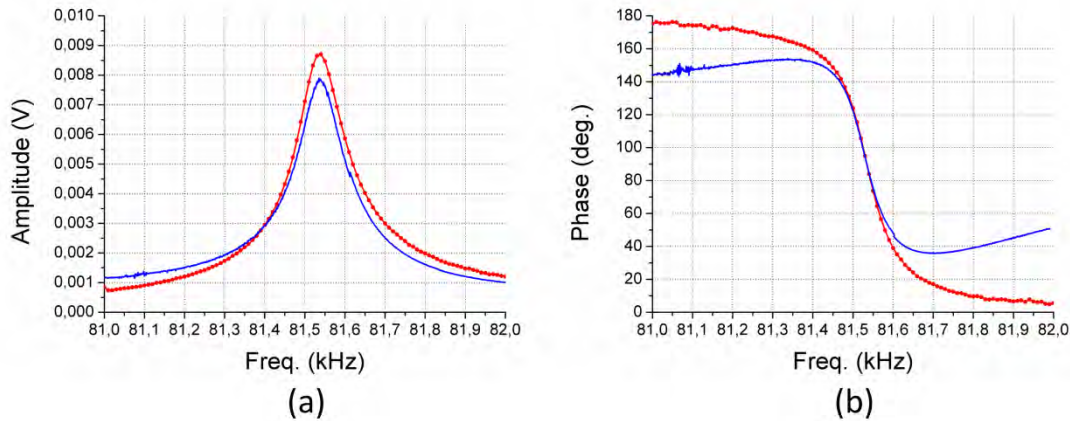


Figure 6. Amplitude (a) and phase (b) measurement with prototype (red dot) and commercial 7265 (blue line) LIAs near the natural resonance frequency for the microcantilever A5 (Table 1).

Frequency tracking

For gas sensing measurements, once the fundamental frequency has been determined in inert atmosphere, the system tracks the cantilever behavior and keeps its oscillation. To accomplish this, it modifies the excitation frequency in accordance to the effective mass variation of the beam due to the chemical interaction between the sensing material and the gas phase. Afterwards, these frequency shifts are correlated to the concentration of target analytes present in the atmosphere.

Frequency tracking for a given cantilever is performed by means of a classical PID algorithm. At the resonance, the sensitivity with respect to frequency changes is minimum for the amplitude and maximum for the phase (Figure 6). For that reason, we have implemented a PID algorithm to keep its phase constant during the whole gas sensing experiment by properly triggering the corresponding NCO.

The electronic interface is controlled by a graphical software (Figure 7) running on the host computer. For natural frequency determination in inert atmosphere (Figure 7.a), the software requires three inputs from the user: initial and final frequency values and step size. Once the frequency sweep is accomplished, data are processed for the accurate determination of the natural frequency of the cantilever (amplitude and phase). In fact, such information (frequency value and target phase) has to be provided by the user to perform the tracking routine required for gas detection. By controlling the phase shift variations (Figure 7.b bottom graph), the implemented algorithm modifies the sensor excitation signal to ensure cantilever oscillation at its natural resonance mode. As an example, Figure 7.b represents the typical resonance frequency evolution (top graph) with gas phase composition due to desorption/adsorption on target analytes on the cantilever tip.

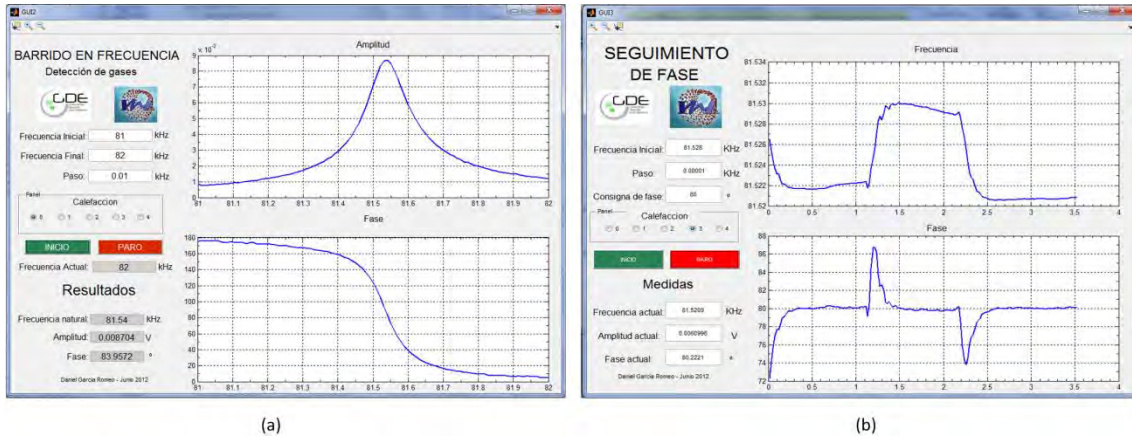


Figure 7. Implemented graphical user interfaces: (a) frequency sweep: amplitude (top graph) and phase (bottom graph) versus frequency, (b) frequency tracking: operating frequency (top graph) and phase (bottom graph) versus time on stream.

5. Experimental results

Several experiments have been carried out for prototype operation and validation. Firstly, resonance frequency values have been obtained for the 8-microcantilever chip shown in Figure 1. The results have been assessed by comparison with a commercial lock-in amplifier. The so obtained values prove the accuracy and reliability of the herein presented prototype. Secondly, the prototype performance for gas sensing has been demonstrated on zeolite functionalized cantilever exposed to different concentrations of 2-nitrotoluene (o-MNT), an explosive related molecule, in presence of humidity.

Table 1. Fundamental resonance frequency values for the 8 Si-microcantilevers estimated by commercial LIA 7265 and the prototype herein developed.

	Length (μm)	Theoretical Frequency	LIA7265 Freq. (Hz)	Prototype Freq. (Hz)
Cantilever A1	500	86868	84624	84621
Cantilever A2	535	75874	73415	73411
Cantilever A3	505	85156	83552	83555
Cantilever A4	530	77312	75325	75330
Cantilever A5	510	83495	81531	81529
Cantilever A6	525	78792	76748	76744
Cantilever A7	515	81881	80212	80217
Cantilever A8	520	80314	77532	77535

LIA Comparison

The mechanical response of each sensor from the 8-microcantilever chip has been completely characterized by gathering both amplitude and phase at their fundamental frequencies with the custom LIA prototype. Table 1 comparatively shows these experimental frequencies values along with theoretically predicted ones and those acquired by a commercial LIA 7265 from Signal Recovery.

The theoretical frequency calculated with formula (1) is about 2% higher than the experimental value. Such deviation could be explained by a 2% variation of the Si beam thickness with respect to the nominal value. On the other hand, an excellent agreement between both LIA amplifiers is achieved.

Vapor Explosive detection

The herein described LIA prototype has been applied to gas sensing experiments using zeolite coated microcantilever. When microcantilevers are used as mass sensors, higher mass sensitivity (S) (Hz/ng) factors are pursued. Thus, little variations in the effective cantilever mass (Δm), due to specific analyte adsorption or sensing material deposition, provokes noticeable frequency shifts (Δf), (see equation 2, where f_0 is the fundamental mode frequency; m is the cantilever effective mass; w denotes the width beam, i.e. 200 μm ; L is the length; E is the Si Young's modulus, i.e. 130 GPa; ρ is the Si density, i.e. 2330 kg/m^3 ; and $r_n = 0,1615$ for fundamental mode r_n).

$$S = \frac{\Delta f}{\Delta m} = -\frac{1}{2} \frac{f_0}{m} = -r_n^{-1} \cdot L^3 w \sqrt{\frac{\rho^3}{E}} \quad (2)$$

Accounting from our previous works [11-12] on zeolite coated cantilevers for explosives detection; some Si microcantilevers have been functionalized with BEA type zeolites by microdropping technique and tested with the LIA prototype. Cobalt (II) exchange procedure was applied on commercial BEA zeolite crystals in the ammonium form (Zeolyst CP814E) following the procedure described by [27] to modify its affinity towards 2-nitrotoluene vapors in presence of interferences. The exchanged powders exhibit a Co/Al atomic ratio up to 0,76 (measure by EDX – energy dispersive X-ray spectroscopy), i.e. higher than the corresponding to 100% of exchanged NH_4^+ , (Co/Al=0,5); indicating the existence of extra Co (II) species on the external surface of zeolite crystals. Table 2 summarizes the main characteristics of the selected modified Si-microcantilevers for the purposes of this work. Uncoated bare Si cantilever has also been included for comparison purposes.

The 8 Si-microcantilever chip was allocated in a Teflon® chamber with a total volume of 1 mL, where gas composition measurements have been performed at 30°C. Gas streams containing mixtures of 2-nitrotoluene vapors and nitrogen as a carrier gas were fed to the chamber at set flowrates by means of mass-flow controllers through explosive saturator trains at controlled temperature. The further dilution with a second mass-flow controlled stream of nitrogen allowed to reach the desired concentration levels (from 1 ppmV to 18 ppmV).

Before adsorption experiments, a zeolite regeneration step is carried out by using the heater integrated on tip. It enables the water or volatiles content removal from the zeolite

micropores. In particular, average temperature values on the sensing coating above 150°C [11-12] are easily attained with the printed heater on tip. This procedure ensures the zeolite adsorption capabilities refreshment; and enlarges the life expectancy of our sensor device. Moreover, the heating of the cantilever during sensing could provide with a valuable tool to discriminate the presence of interferences, and consequently increase the final selectivity towards explosive related molecules.

Table 2. Main characteristics of the zeolite modified Si-microcantilever used in this work for 2-nitrotoluene detection in vapor phase.

	Sensitivity Hz/ng*	Frequency (Hz) Before coating	Sensing Layer	Frequency (Hz) After coating	Sensing material loading (ng)
Cantilever B2	16.24	74020	NH ₄ BEA	67450	404,5
Cantilever B3	19.31	84190	CoBEA	77140	365,0
Cantilever B8	17.69	78300	None	None	None

*Calculated from equation 2.

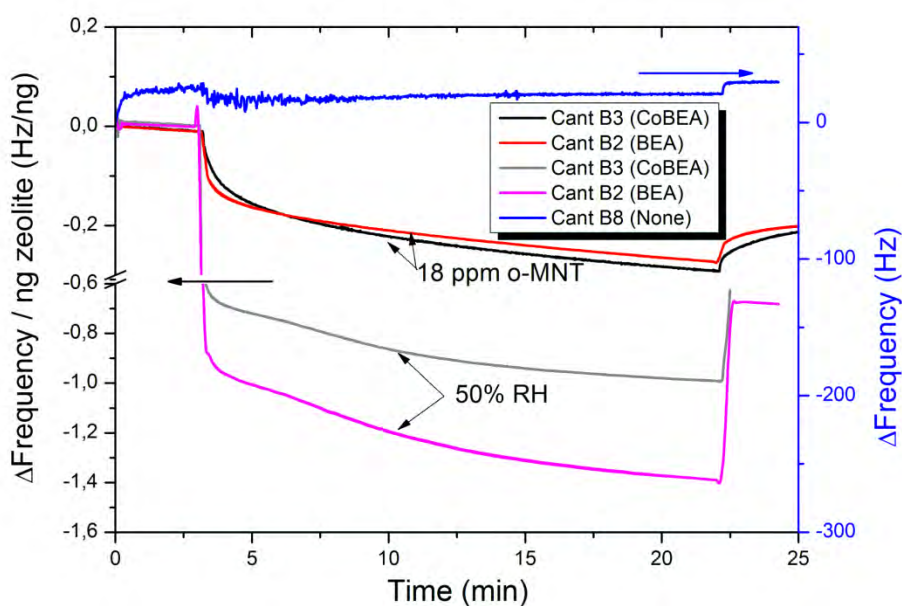


Figure 8. Evolution of the sensor response upon the introduction of 2-nitrotoluene (o-MNT) vapors and humidity at room temperature for cantilevers B3 (coated with Co-BEA) and B2 (coated with NH₄BEA). Response of bare Si cantilever B8 is also included as reference.

Figure 8 illustrates the frequency shift of Si microcantilevers modified with commercial NH₄BEA (Cantilever B2) and CoBEA (Cantilever B3) type zeolites respectively in presence of humidity (50% of relative humidity (RH) at ambient conditions) and o-MNT as single analytes. As it can be observed, the sensor response has been normalized per sensing material loading

for a proper data comparison. The performance of bare Si cantilever under identical conditions has also been included as reference. In spite of the huge differences in the gas phase concentration for the tested analytes (20000 ppmV H₂O vs 18 ppmV o-MNT), the zeolite modified sensor responses upon the introduction of explosive related molecule are only two fold lower than the obtained after water exposure. This behavior highlights the specific nitroaromatic sorption over the selected zeolites. It is also noteworthy the mechanical response of bare Si cantilever, almost unaffected by the gas composition. Received NH₄BEA zeolites exhibit higher water adsorption capability than Co exchanged counterparts. This differential behavior is attributed to extra-framework Co²⁺ cations and also extra cobalt species on the external zeolite surface. This presence modifies the water adsorption properties; and, eventually, hinders the diffusion of water molecules to the micropores by steric effect. On the contrary, 2-nitroderivate sorption is rather similar for both BEA type zeolites. Thus, it can be concluded that o-MNT/water selectivity has been notably improved by the Co exchange process.

Figure 9 shows the resonant frequency evolution for the CoBEA modified microcantilever upon the introduction of 2-nitrotoluene (o-MNT) steps 3 min in duration varying from 12 ppmV to 1 ppmV. These experiments were performed to evaluate the kinetics and sorption capability on the basis of specific target-zeolite interactions. Once the feed is switched to inert N₂ gas, the frequency response of the resonating microcantilever does not recover the initial value due to strong interactions between BEA type and 2-nitrotoluene vapors whatever the concentration tested. This behavior is clearly indicating the importance of an adequate sensor conditioning step when zeolites are used as sensing material.

From this set of vapor related explosive experiments, the as calculated o-MNT gas sensitivity (expressed in Hz/ppmV) by proper fitting of the experimental points (not shown here) is 40 Hz/ppm. In this work, the lowest concentration of o-MNT experimentally measured has been 1 ppmV due to practical limitations in the vapor generation system. However, accounting from the experimental signal noise (about 1 Hz), theoretical limit of detection (LOD) values below 100 ppb are obtained for B3 cantilever.

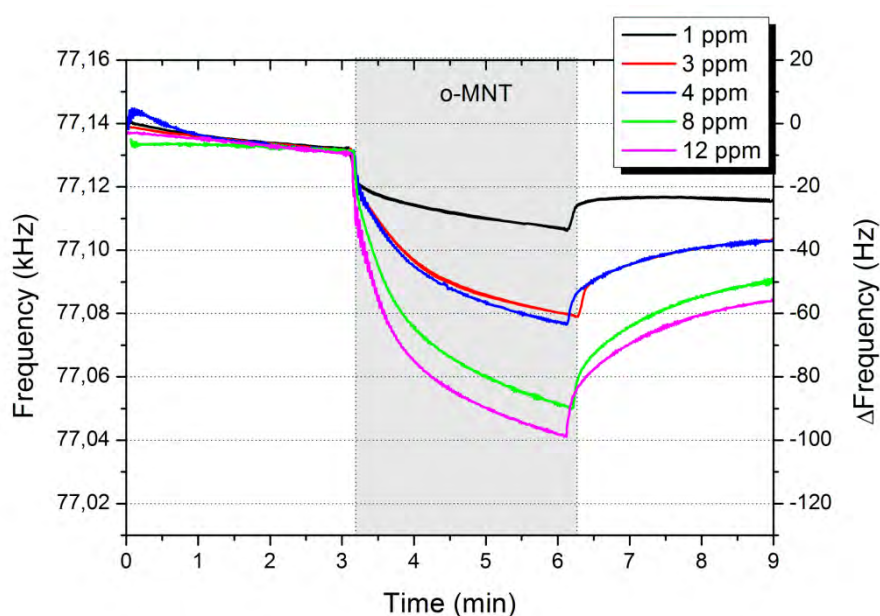


Figure 9. Evolution of the resonant frequency upon the introduction of 2-nitrotoluene (o-MNT) vapors at room temperature for B3 cantilever.

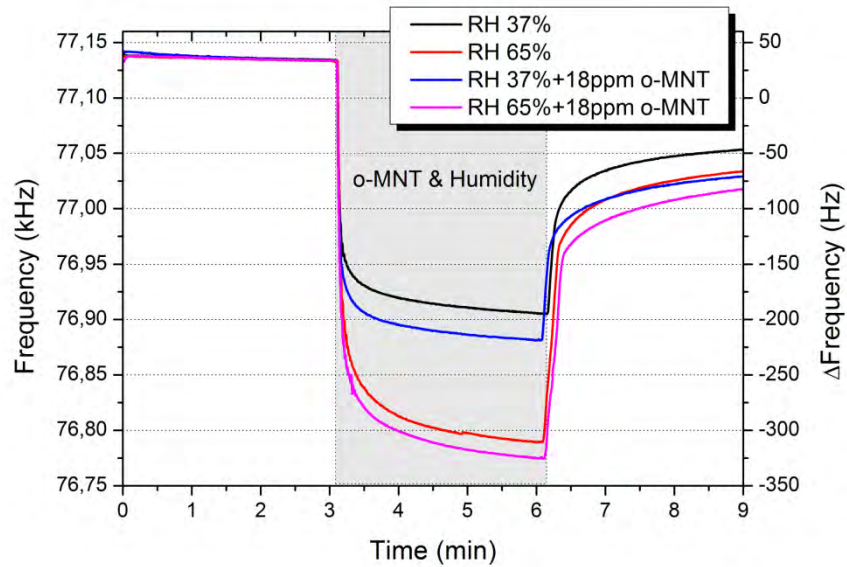


Figure 10. Evolution of the resonant frequency upon the introduction of binary mixtures of 18 ppmV of o-MNT vapors at two different relative humidity (RH) values as interference for B3 cantilever.

The characterization of the sensors herein prepared in presence of humidity as interference is crucial due to the hydrophilic properties of zeolites. Therefore, additional experiments with mixtures containing o-MNT (18 ppmV) and H₂O vapors (15000 ppmV circa 7% RH, or 26000 ppmV circa 65% RH) have been accomplished on B3 cantilever. The obtained results are depicted in Figure 10 where single H₂O vapors (37% RH and 65% RH) detection experiments are also included for comparison. The most outstanding feature is the frequency shift, higher than 100 Hz; and attributed to the presence of 18 ppmV of o-MNT even at 65% RH. This observation is again underlying the specific adsorption properties of the zeolite based sensing materials towards the target analytes in presence of interferences.

6. Conclusions

A portable low-power electronic interface for a multi-sensing platform comprising 8-microcantilever has been successfully demonstrated in this work. The core element is a square signal based analog lock-in amplifier based on a dual phase technique that uses two reference signals in quadrature which enables the acquisition of both amplitude and phase measurements regardless of phase deviations produced by the sensors. This configuration together with a new algorithm capable of handling square signals constitutes a clear innovation for portable sensing systems. Specific software implementation has been developed to accomplish frequency sweeps at the desired frequency range and to enable dynamic mode operation with the microcantilever oscillating at its natural frequency. The accuracy of the LIA prototype has been proved by comparison with a commercial 7265 lock-in

amplifier from Signal Recovery at reference conditions. The portable electronic interface, solely fed by the 5V line from a USB connection to a host computer, has been validated for 2-nitrotoluene detection, an explosive related molecule, by using BEA type zeolite modified microcantilevers. In particular, LOD values below 100 ppb for CoBEA modified cantilevers have been obtained.

The LIA prototype described in this work presents clear advantages in terms of versatility, computational simplicity, reduced size and energy consumption. Overall, the simplification of the cumbersome instrumentation paves the way for field experiments under relevant conditions. All these features also facilitate an easy implementation of our customized LIA to any other portable multi-sensing platform whatever the transduction schemes involved.

Acknowledgements

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