

An assessment of RTN-induced threshold voltage jitter

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Abstract

Power consumption is a key issue especially for the edge devices/units in an IoT system. Lowering operation voltage is an effective way to reduce power. As the overdrive voltage, $V_g - V_{th}$, becomes smaller, the device is more vulnerable to threshold voltage jitters. One source for the jitter is Random Telegraph Noises (RTN), which cause a fluctuation in both drain current, ΔI_d , and threshold voltage, ΔV_{th} . Early works on RTN were focused on measuring ΔI_d and then evaluate ΔV_{th} from $\Delta I_d/g_m$, where g_m is transconductance. The accuracy of ΔV_{th} obtained in this way is not known. The objective of this work is to assess its accuracy by comparing it with the ΔV_{th} directly measured from pulse $I_d - V_g$. It will be shown that the correlation between these two is poor, so that ΔV_{th} must not be evaluated from $\Delta I_d/g_m$. This is caused by the device-specific localized current distribution near the threshold.

1. Introduction

The instabilities of modern MOSFETs have a number of sources: bias temperature instabilities (BTI) [1-7], hot carrier ageing (HCA) [8-10], and random telegraph noise [11-17]. To increase the instabilities and make them measurable, it is a common practice to use voltage-acceleration [1-10]. This works well for both BTI and HCA and the threshold voltage shift, ΔV_{th} , has been reliably measured at pre-specified time by the measure-stress-measure (MSM) technique [1-10].

The MSM technique, however, is inapplicable for the RTN-induced jitter in V_{th} because of two reasons. First, RTN is dominated by traps near the Fermi-level at the dielectric/substrate interface E_f , as shown in Fig. 1a. Although Fig. 1b shows that there are more traps at high $|V_g|$, V_g -acceleration shifts E_f , so that RTN would be dominated by a different group of traps under V_g -acceleration. Second, the charging-discharging of traps responsible for RTN is highly dynamic in nature. At a pre-specified time, they can be neutral and would be missed if the MSM technique is used.

Because of these difficulties, early works [12,13] focus on measuring the fluctuation in drain current, ΔI_d , under a fixed V_g . This on-the-fly (OTF) ΔI_d is then converted to ΔV_{th} by dividing the transconductance, g_m . The

accuracy of this ΔV_{th} -OTF is not clear, as it was not compared with the real ΔV_{th} measured at $V_g = V_{th}$. The objective of this work is to assess the accuracy of ΔV_{th} -OTF and analyse the source of discrepancies between the ΔV_{th} -OTF and the real ΔV_{th} .

2. Devices and experiments

The devices were fabricated by a 28 nm CMOS process with channel length and width of 27 nm and 90 nm, respectively. The high-k dielectric stack has an equivalent oxide thickness of 1.2 nm and metal gate. To minimize the trap discharge, pulse $I_d - V_g$ (p-IV) was taken and the measurement time is 3 μs . V_{th} was evaluated by using the maximum g_m (Max- g_m) method. The ΔV_{th} -OTF was evaluated from the $\Delta I_d/g_m$ at $|V_g| = 1$ V. The temperature is 125 °C.

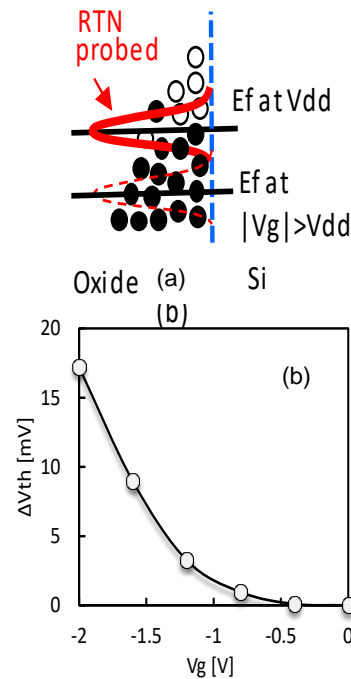


Fig. 1. (a) RTN is dominated by traps near to E_f at the interface. V_g -acceleration shifts E_f and changes the traps probed by RTN. (b) More traps under higher $|V_g|$ result in higher ΔV_{th} .

3. Results and discussions

We start with a large device of $1 \times 3 \mu\text{m}$, where the device-to-device variation is insignificant. The two p-IVs before and after charging are given in Fig. 2a and the ΔV_{th} evaluated from their differences at constant I_d is plotted in Fig. 2b. The $\Delta V_{\text{th-OTF}}$ and the $\Delta V_{\text{th-Max-gm}}$ are shown by the two dashed lines. The $\Delta V_{\text{th-OTF}}$ agrees well with the ΔV_{th} when V_g is close to -1 V in the region ‘A’, while the $\Delta V_{\text{th-Max-gm}}$ agrees with the ΔV_{th} when V_g is close to the V_{th} of -0.45 V in the region ‘B’.

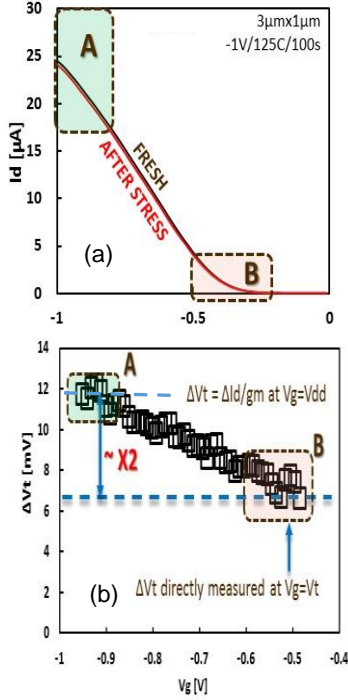


Fig. 2. (a) The pulse IVs before and after charging of a large device ($1 \times 3 \mu\text{m}$). (b) ΔV_{th} evaluated from constant I_d under different V_g . The top dashed line is $\Delta V_{\text{th-OTF}}$ evaluated from $\Delta I_d/g_m$ at $V_{\text{dd}} = -1 \text{ V}$. The bottom dashed line is ΔV_{th} extracted from the max-gm.

$\Delta V_{\text{th-OTF}}$ is twice of $\Delta V_{\text{th-Max-gm}}$ in Fig. 2b. As a result, $\Delta V_{\text{th-OTF}}$ should not be used as a substitute for the real ΔV_{th} . For SRAM, MOSFETs operate close to V_{th} near the trip-point and the ΔV_{th} affects its noise margin [14,16]. For digital circuits, MOSFETs go through V_{th} during the switching and an increase of $|V_{\text{th}}|$ causes delay. As a result, it is important to analyse the sources of the discrepancies between $\Delta V_{\text{th-OTF}}$ and $\Delta V_{\text{th-Max-gm}}$. One source is the mobility degradation, which was neglected in evaluating $\Delta V_{\text{th-OTF}}$. In the following, we will use nano-scale devices to show that the current distribution also contributes to the discrepancies.

Unlike large devices, RTN causes a substantial fluctuation in I_d , as shown in Fig. 3a. As mentioned earlier, if the p-IV was taken at pre-specified time, some traps are neutral and will be missed and one example is given in Fig. 3b. To ensure capturing the trapped charges, the trigger level of the oscilloscope must be adjusted so that it only triggers when the traps are charged, as shown in Fig. 3a. The inset of Fig. 3b illustrates that this method improves the measurement accuracy substantially.

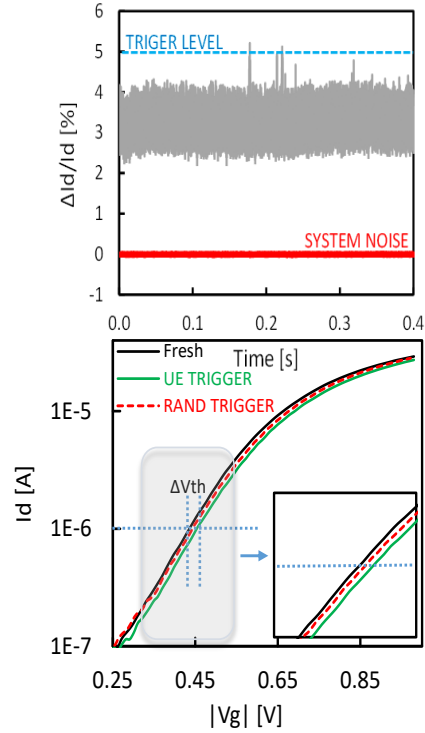


Fig. 3. (a) The p-IV trigger level is set to capture the trapped charges. (b) A comparison between the p-IV(trigger-when-charged) (green line) with the p-IV(trigger-at-pre-specified-time) (dashed red line).

Figs. 4a-c show the results from three different devices. The device-to-device variation (DDV) is substantial. In Fig. 4a, the $\Delta V_{\text{th-OTF}}$ at $V_g = -1 \text{ V}$ is ~ 4 times of the ΔV_{th} at $V_g = V_{\text{th}} = -0.45 \text{ V}$, so that the discrepancy is much larger than that of large devices in Fig. 2b. Fig. 4b shows that ΔV_{th} saturates as $|V_g|$ increases, in contrast with the monotonic increase in Fig. 2b. Moreover, Fig. 4c shows that ΔV_{th} turns around: it increases initially and then decreases for higher $|V_g|$. This turn-around behaviour cannot be explained by charge-induced mobility degradation and other mechanisms must be involved.

It has been reported that the current path near to V_{th} is localized, but becomes more evenly distributed as

$|V_g - V_{th}|$ increases. The impact of a charged trap on V_{th} depends on the current density beneath it: the higher the current density, the larger the impact [17,18]. As illustrated in Fig. 5, the trap for Fig. 4a is away from the localized current path at V_{th} . As $|V_g - V_{th}|$ increases, a more evenly distributed current increases the relative local current density, so that the ΔV_{th} increases in Fig. 4a. For Fig. 4c, the trap can be above the localized current path at V_{th} . As $|V_g - V_{th}|$ increases, the initial rise is caused by increased mobility degradation and the subsequent decrease is caused by the relative reduction of current density under this trap, as the distribution spreads. The result of large device in Fig. 2b is the average results of small devices. Under V_{th} , localization of current path leads to many traps being away from it, so that the impact on ΔV_{th} is low. As $|V_g - V_{th}|$ increases, the effect of these traps increases, contributing the higher ΔV_{th} for higher $|V_g - V_{th}|$.

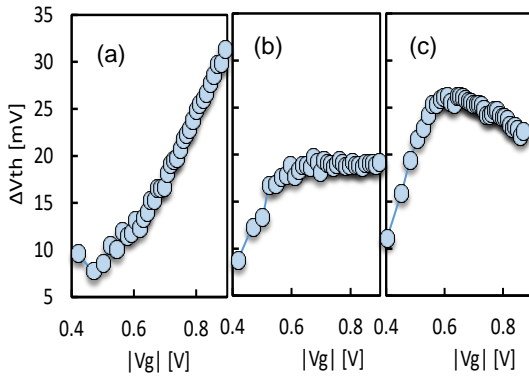


Fig. 4. The dependence of ΔV_{th} on sensing V_g for three nano-scale devices. (a) ΔV_{th} increases with $|V_g|$. (b) ΔV_{th} saturates. (c) ΔV_{th} turns around.

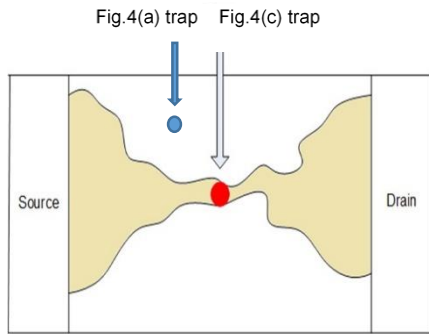


Fig. 5. The localized current path at $V_g = V_{th}$. The trap in Fig. 4a is away from the path, while the trap in Fig. 4c is above it.

The correlation between the $\Delta V_{th}\text{-Max-gm}$ and $\Delta I_d/I_d\text{-OTF}$ is given in Fig. 6. The poor correlation between them confirms that $\Delta I_d/I_d\text{-OTF}$ should not be

used to evaluate the ΔV_{th} .

Fig. 7 shows that the $\Delta I_d/I_d$ against measurement time. Longer time allows slower traps to respond, leading to higher up-envelope (UE). The DDV of $\Delta I_d/I_d$ and ΔV_{th} are given in Figs. 8a&b, respectively and both of them are substantial. Figs. 9a&b show that the statistical distribution follows the defect-centric model well [19].

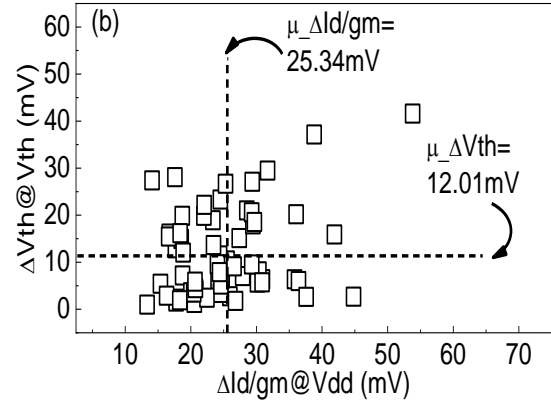


Fig. 6. The poor correlation $\Delta I_d/gm$ and the $\Delta V_{th}\text{-Max-gm}$. [17]

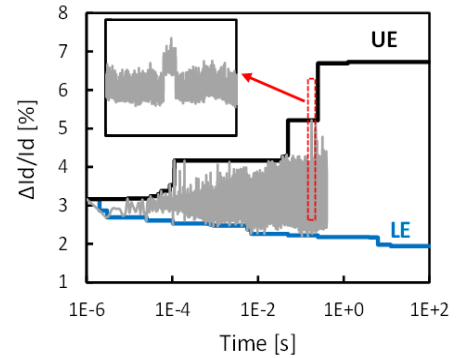


Fig. 7. Larger time window captures slower traps, leading to higher up-envelope (UE).

4. Conclusions

$\Delta V_{th}\text{-OTF}$ measured under V_{dd} is compared with the real ΔV_{th} extracted by maximum-gm method. On average the former is twice of the latter. This discrepancy is caused not only by the increased mobility degradation for higher $|V_g - V_{th}|$, but also caused by a more evenly distributed current. The RTN-induced device-to-device variations follow the defect-centric model for both $\Delta I_d/I_d$ and ΔV_{th} . There is a poor correlation between $\Delta I_d/I_d\text{-OTF}$ and ΔV_{th} .

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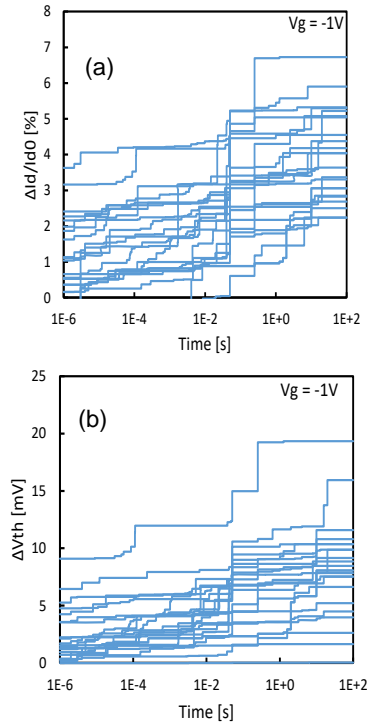


Fig. 8. The device-to-device variations of (a) $\Delta I_d/I_{d0}$ -OTF and (b) ΔV_{th} -Max-gm.

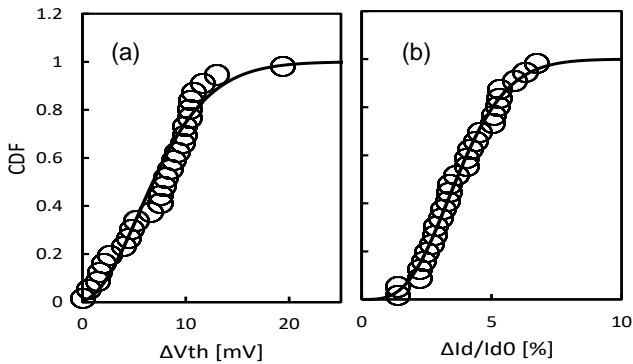


Fig. 9. The statistical distribution follows Defect-centric model (Lines) (a) ΔV_{th} -Max-gm and (b) $\Delta I_d/I_{d0}$ -OTF.

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