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Optimal filtering of the LISA data

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The LISA time-delay-interferometry responses to a gravitational wave signal are rewritten in a form that accounts for the motion of the LISA constellation around the Sun; the responses are given in closed analytic forms valid for any frequency in the band accessible to LISA. We then present a complete procedure, based on the principle of maximum likelihood, to search for stellar-mass binary systems in the LISA data. We define the required optimal filters, the amplitude-maximized detection statistic (analogous to the \mathcal{F} statistic used in pulsar searches with ground-based interferometers), and discuss the false-alarm and detection probabilities. We then test the procedure in numerical simulations of gravitational-wave detection.

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I. INTRODUCTION

The Laser Interferometer Space Antenna (LISA) is a deep-space mission aimed at detecting and studying gravitational radiation in the millihertz frequency band. A joint American and European project, it is expected to be launched in the year 2011, and to start collecting scientific data approximately a year later, after reaching its orbital configuration of operation [1]. LISA consists of three widely separated spacecraft, flying in a triangular, almost equilateral configuration, and exchanging coherent laser beams. In contrast to ground-based, equal-arm gravitational-wave (GW) interferometers, LISA will have multiple readouts, corresponding to the six laser Doppler shifts measured between spacecraft. Modeling each spacecraft as carrying lasers, beam splitters, photodetectors, and drag-free proof masses on each of two optical benches, Armstrong, Estabrook, and Tinto [2-4] showed that it is possible to combine, with suitable time delays, the six time series of the inter-spacecraft Doppler shifts and the six time series of the intra-spacecraft Doppler shifts (measured between adjacent optical benches) to cancel the otherwise overwhelming frequency fluctuations of the lasers $(\Delta \nu / \nu \simeq 10^{-13} / \sqrt{\text{Hz}})$, and the noise due to the mechanical vibrations of the optical benches (which could be as large as $\Delta \nu / \nu \simeq 10^{-16} / \sqrt{\text{Hz}}$). The strain sensitivity level that then becomes achievable, $h \simeq 10^{-21} / \sqrt{\text{Hz}}$, is set by the buffeting of the drag-free proof masses inside each optical bench, and by the shot noise at the photodetectors. Several such laser-noise-free interferometric combinations are possible, and they show different couplings to gravitational waves and to the remaining system noises [2-5]. The technique used to synthesize these combinations is known as time-delay interferometry (TDI); in the case of a stationary array, it was shown that the space of all the possible TDI

observables can be constructed by combining four generators [2,6].

Recently, it was pointed out [7-10] that the rotational motion of the LISA array around the Sun and the time dependence of light travel times introduced by the relative (shearing) motion of the spacecraft have the effect of preventing the suppression of laser frequency fluctuations, at least under the current stability requirements, to the level of the secondary noises in the TDI observables as derived for a stationary array. This problem was addressed by devising new combinations that are capable of suppressing the laser frequency fluctuations below the secondary noises for a rotating LISA array [7,8], and for a rotating and shearing LISA array [9,10]. In this context, the original stationary-array combinations are sometimes known as "TDI 1.0" (or firstgeneration TDI), the rotating-LISA combinations as "TDI 1.5," and the rotating and shearing-LISA combinations as "TDI 2.0;" following Ref. [10], we refer to the last as second-generation TDI. Second-generation combinations are essentially finite differences of first-generation combinations, and as such they appear more complicated. However, they retain the same sensitivity to incoming GWs: this is because the corrections introduced in the original combinations by the changing array geometry are obviously important for laser frequency fluctuations, but they are negligibly small for the GW responses and for the secondary noises; thus, once laser frequency noise is removed, the second-generation observables become finite differences of the corresponding first-generation observables. At a fixed frequency, the ratio of GW response to secondary noises (and hence the sensitivity) is then unchanged.

The GW responses of the TDI combinations depend on the relative orientation of the LISA array with respect to the direction of propagation of the GW signal, on the strength and polarization of the signal, and on its frequency components. Analytic expressions for the TDI responses were first derived by Armstrong, Estabrook and Tinto [2], for a stationary LISA array. A realistic model of LISA must however include the motion of the array around the Sun, which introduces slow modulations in the phase and amplitude of the GW responses (in addition, of course, to the modifications introduced by adopting second-generation TDI). For instance, the LISA responses to the sinusoidal signal emitted

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by a binary system are not simple sinusoids, but rather superpositions of many sinusoids of smaller amplitude. To maximize the likelihood of source detection, these effects must be modeled in GW search algorithms, either by including the modulations in the theoretical models of the signals (i.e., the *templates*), or by demodulating the LISA data for a given set of sky positions as the first step of data analysis [11,12].

In this paper we derive the response of the secondgeneration TDI observables to the GW signals generated by a binary system, and we describe how signal templates based on these responses can be used in a maximum-likelihood, matched-filtering framework to search for binaries and to estimate their parameters once they are found. Other methods to analyze the LISA data for signals from binaries, implemented in the long-wavelength approximation, have been proposed in Refs. [13,14]. We work in the solar-systembaricentric frame, and we follow closely the derivation given by Jaranowski, Królak, and Schutz [15] for continuous sources and ground-based detectors. A similar formalism was used by Giampieri [16] to obtain the antenna pattern of an arbitrary orbiting interferometer, in the long-wavelength approximation. The response of an orbiting equal-arm-Michelson interferometer to a sinusoidal signal was worked out by Cutler [17], again in the long-wavelength limit. Seto [18] extended Cutler's formalism to high frequencies (and to noise-canceling observables), in the context of studying optimal-filtering parameter estimation for supermassiveblack-hole binaries. Cornish, Rubbo, and Poujade [19,20] obtained general expressions valid in the entire LISA frequency band, and for arbitrary GW signals; these expressions are given as integrals over the LISA arms, and they provide the basic building blocks to assemble the TDI observables. By contrast, in this paper we work out explicit time-domain expressions for the LISA response to moderately chirping binary systems, for all the second-generation TDI combinations. These expressions are valid over the entire LISA frequency band, and they are written as linear combinations of four time-dependent functions; this linear structure facilitates the computation of matched filters and the design of optimal filtering algorithms.

This paper is organized as follows. In Sec. II we give a brief overview of the derivation of the TDI responses to GWs for a stationary array, and we argue that the corrections introduced by the motion of the LISA array and by the time dependence of light travel times are negligibly small. Working in the solar-system-baricentric frame, we obtain general expressions for the GW responses of the Michelson $(X_1, X_2,$ X_3), Sagnac (α_1 , α_2 , α_3), and optimal (\overline{A} , \overline{E} , \overline{T} ; see [21]) second-generation TDI observables (expressions for the firstgeneration observables are given in Appendix C); finally, we derive the corresponding closed-form analytic expressions for moderately chirping binary systems, valid at any frequency in the LISA band. In Sec. III we provide expressions for the spectral densities of noise in the TDI combinations. In Sec. IV we combine the results of Secs. II and III to design optimal filters that can be applied to the LISA TDI data to search for binary stars; we take advantage of the linear structure of the responses to define an optimal detection statistic that does not depend on the effective polarization and on the



FIG. 1. Schematic LISA configuration. The spacecraft are labeled 1, 2, and 3; each spacecraft contains two optical benches, denoted by $1,1^*,\ldots$, as indicated. The optical paths are denoted by L_i , where the index *i* corresponds to the opposite spacecraft. The unit vectors $\hat{\mathbf{n}}_i$ point between pairs of spacecraft, with the orientation indicated.

initial phase of the binary, in analogy to the \mathcal{F} statistic [15,22] used in searches for continuous GW sources with ground-based interferometers. In Sec. V we derive the falsealarm and false-dismissal probabilities for our LISA \mathcal{F} statistic. Last, in Sec. 6 we describe an efficient algorithm to compute \mathcal{F} , and we implement it numerically; we perform a simulation of GW detections in both the low and high-frequency part of the LISA band, and for both Michelson and optimal [21] TDI observables, and we show that our algorithm yields very accurate estimates of source parameters. In the rest of this paper we shall use units where c=1.

II. TIME-DELAY INTERFEROMETRY

Figure 1 shows the overall LISA geometry. The spacecraft are labeled 1, 2, and 3; the arms are labeled with the index of the opposite spacecraft (e.g., arm 1 lies between spacecraft 2 and 3). The light travel time (or, loosely, the armlength) along arm *i* is denoted by L_i [7–10]. The basic constituents of the TDI observables are the time series of the relative laser-frequency fluctuations measured between spacecraft, which are denoted by $y_{ii}(t)$, with $i \neq j$: for instance, $y_{31}(t)$ is the time series of relative frequency fluctuations measured for reception at spacecraft 1 with transmission from spacecraft 3 (along arm 2); similarly, $y_{21}(t)$ is the time series measured for reception at spacecraft 1 with transmission from spacecraft 2 (along arm 3), and so on. Six more time series result from comparing the laser beams exchanged between adjacent optical benches within each spacecraft; these time series are denoted by z_{ij} , with $i, j = 1, 2, 3, i \neq j$ (see [3,4,10] for details). Delayed time series are denoted by commas: for instance, $y_{31,2} = y_{31}[t - L_2(t)]$, and so on.

The frequency fluctuations introduced by the lasers, by the optical benches, by the proof masses, by the fiber optics, and by the measurement itself at the photo-detector (i.e., the shot-noise fluctuations) enter the Doppler observables y_{ij} and z_{ij} with specific time signatures; see Refs. [3,4,10] for a detailed discussion. The contribution y_{ij}^{GW} due to GW signals was derived in Ref. [2] in the case of a stationary array. (Note that in Ref. [2], and indeed in all the literature on first-generation TDI, the notation y_{ij} indicates the one-way Doppler measurement for the laser beam received at space-



FIG. 2. Orbital motion of the LISA detector, shown in a solarsystem baricentric ecliptic coordinate system. The trajectories shown correspond to setting $\zeta = -\pi/6$, $\Omega = 2\pi/yr$, and $\xi_0 = \eta_0$ = 0 in Eqs. (5) and (6).

craft *j* and *traveling along arm i*. In this paper we conform to the notation used in Refs. [7-10].)

Since the motion of the LISA array around the Sun introduces a difference between (and a time dependence in) the corotating and counterrotating light travel times, the exact expressions for the GW contributions to the various firstgeneration TDI combinations will in principle differ from the expressions valid for a stationary array [2]. However, the magnitude of the corrections introduced by the motion of the array are proportional to the product between the time derivative of the GW amplitude and the difference between the actual light travel times and those valid for a stationary array. At 1 Hz, for instance, the larger correction to the signal (due to the difference between the corotating and counterrotating light travel times) is two orders of magnitude smaller than the main signal. Since the amplitude of this correction scales linearly with the Fourier frequency, we can completely disregard this effect (and the weaker effect due to the time dependence of the light travel times) over the entire LISA band [10]. Furthermore, since along the LISA orbit the three armlengths will differ at most by $\sim 1\% - 2\%$, the degradation in signal-to-noise ratio introduced by adopting signal templates that neglect the inequality of the armlengths will be at most a few percent. For these reasons, in what follows we shall derive the GW responses of various second-generation TDI observables by disregarding the differences in the delay times experienced by light propagating clockwise and counterclockwise, and by assuming the three LISA armlengths to be constant and equal to $L=5\times10^6$ km ≈ 16.67 s [23]. These approximations, together with the treatment of the moving-LISA GW response discussed at the end of Sec. II C, are essentially equivalent to the rigid adiabatic approximation of Ref. [20], and to the formalism of Ref. [18].

A. Geometry of the orbiting LISA array

We denote the positions of the three spacecrafts by \mathbf{p}_i and the unit vectors along the arms by $\hat{\mathbf{n}}_i$, where $\hat{\mathbf{n}}_1$ points from spacecraft 3 to 2, $\hat{\mathbf{n}}_2$ points from spacecraft 1 to 3, and $\hat{\mathbf{n}}_3$ points from spacecraft 2 to 1. In the coordinate frame where the spacecraft are at rest, we can set without loss of generality

$$\mathbf{p}_i^L = (L/\sqrt{3})(-\cos 2\sigma_i, \sin 2\sigma_i, 0), \qquad (1)$$

and

$$\hat{\mathbf{n}}_i^L = (\cos \sigma_i, \sin \sigma_i, 0), \qquad (2)$$

where

$$\sigma_i = 3\pi/2 - 2(i-1)\pi/3. \tag{3}$$

Because the motion of the LISA guiding center (i.e., the baricenter of the formation) is contained in the plane of the ecliptic, it is convenient to work in a solar-system-baricentric (SSB) ecliptic coordinate system. We take the x axis of this system to be directed toward the vernal point. A realistic set of orbits for the spacecraft [23], shown in Fig. 2, is obtained by setting

$$\mathbf{p}_{i}(t) = \mathbf{r}(t) + \mathbf{O}_{2} \cdot \mathbf{p}_{i}^{L}, \quad \hat{\mathbf{n}}_{i}(t) = \mathbf{O}_{2} \cdot \hat{\mathbf{n}}_{i}^{L}, \quad (4)$$

where \mathbf{r} is the vector from the origin of the SSB coordinate system to the LISA guiding center, as described by the SSB components

$$\mathbf{r} = R(\cos \eta, \sin \eta, 0), \quad R = 1 \text{ AU};$$
 (5)

the function $\eta = \Omega t + \eta_0$ returns the true anomaly of the motion of the LISA guiding center around the Sun. The rotation matrix O_2 models the cartwheeling motions of the spacecraft along their inclined orbits, shown in Fig. 3; it is given by



FIG. 3. Cartwheeling motion of the LISA array, as plotted in a frame with center in the LISA guiding center and axes parallel to the SSB ecliptic frame. We show three snapshots at different times along the LISA orbital period, 1 yr.

$$O_{2} = \begin{pmatrix} \sin \eta \cos \xi - \cos \eta \sin \zeta \sin \xi & -\sin \eta \sin \xi - \cos \eta \sin \zeta \cos \xi & -\cos \eta \cos \zeta \\ -\cos \eta \cos \xi - \sin \eta \sin \zeta \sin \xi & \cos \eta \sin \xi - \sin \eta \sin \zeta \cos \xi & -\sin \eta \cos \zeta \\ \cos \zeta \sin \xi & \cos \zeta \cos \xi & -\sin \zeta \end{pmatrix};$$
(6)

the function $\xi = -\Omega t + \xi_0$ returns the phase of the motion of each spacecraft around the guiding center, while ζ sets the inclination of the orbital plane with respect to the ecliptic. For the LISA trajectory, $\Omega = 2\pi/\text{yr}$ and $\zeta = -\pi/6$ [23]. For simplicity, we can set $\eta_0 = \xi_0 = 0$, so that at time t = 0 the LISA guiding center lies on the positive *x* axis of the SSB system, while \mathbf{p}_1 lies on the negative *y* axis. The spacecraft orbits described by Eq. (4) can be approximately mapped to those used by Cornish and Rubbo [19] by identifying our spacecraft 1, 2, and 3 with their spacecraft 0, 2, and 1, and by setting $\eta_0 = \kappa$, $\xi_0 = 3\pi/2 - \kappa + \lambda$, where κ and λ are the parameters defined below Eqs. (56) and (57) of Ref. [19].

B. Generic plane waveform

At the origin of the SSB frame, the transverse-traceless metric perturbation due to a source located at ecliptic latitude β and longitude λ can be written as

$$\mathbf{H}(t) = \mathbf{O}_1 \cdot \mathbf{H}^{\mathcal{S}}(t) \cdot \mathbf{O}_1^{-1},\tag{7}$$

where the metric perturbation in the source frame is taken to be

$$\mathsf{H}^{S}(t) = \begin{pmatrix} h^{+}(t) & h^{\times}(t) & 0\\ h^{\times}(t) & -h^{+}(t) & 0\\ 0 & 0 & 0 \end{pmatrix},$$
(8)

with $h^+(t)$ and $h^{\times}(t)$ the two GW polarizations, and where

$$O_{1} = \begin{pmatrix} \sin\lambda\cos\psi - \cos\lambda\sin\beta\sin\psi & -\sin\lambda\sin\psi - \cos\lambda\sin\beta\cos\psi & -\cos\lambda\cos\beta \\ -\cos\lambda\cos\psi - \sin\lambda\sin\beta\sin\psi & \cos\lambda\sin\psi - \sin\lambda\sin\beta\cos\psi & -\sin\lambda\cos\beta \\ \cos\beta\sin\psi & \cos\beta\cos\psi & -\sin\beta \end{pmatrix};$$
(9)

the dependence of the rotation matrix O_1 on β and λ enforces the transversality of the plane waves, which are propagating from a source located in the direction

$$\hat{\mathbf{k}} = (\cos \lambda \cos \beta, \sin \lambda \cos \beta, \sin \beta); \tag{10}$$

the polarization angle ψ encodes a rotation around the direction of wave propagation, $-\mathbf{\hat{k}}$, setting the convention used to define the two polarizations, + and \times . The polarizations corresponding to $\psi=0$ are shown in Fig. 4 for various source positions in the sky. In the center of the LISA proper frame (the frame where the spacecraft are at rest), the transverse-traceless metric perturbation is given by

$$\mathbf{H}^{L}(t) = \mathbf{O}_{2}^{-1}(t)\mathbf{O}_{1}\mathbf{H}^{S}(t)\mathbf{O}_{1}^{-1}\mathbf{O}_{2}(t).$$
(11)

The time variable *t* that appears in $h^+(t)$ and $h^{\times}(t)$ [and therefore in H(*t*) and H^L(*t*)] is the time at the origin of the SSB frame. It is related to the time in the GW source frame by a relativistic time dilation, due to the proper motion of the source and to cosmological effects. It is however expedient to identify the two times, and to describe GW emission using SSB time; the time dilation is then taken into account by mapping the apparent (measured) physical parameters of a source into its real parameters. The source positional parameters.

eters β , λ , and ψ can be mapped to the parameters θ , ϕ , and ψ used in Ref. [19] by setting $\beta = \pi/2 - \theta$, $\lambda = \phi$, and $\psi = -\psi$.

C. GW response of the LISA array

As derived in Ref. [2] for a stationary, equilateral-triangle LISA array, the one-way Doppler responses y_{21} and y_{31} excited by a plane transverse-traceless GW propagating from the source direction $\hat{\mathbf{k}}$, are given by



FIG. 4. Conventional definition of the GW polarizations + (dashed) and \times (solid) for various ecliptic latitudes β and longitudes λ .

$$y_{21}^{\text{GW}}(t) = \begin{bmatrix} 1 - \hat{\mathbf{k}} \cdot \hat{\mathbf{n}}_3 \end{bmatrix} \begin{bmatrix} \Psi_3(t + \hat{\mathbf{k}} \cdot \mathbf{p}_2 - L) - \Psi_3(t + \hat{\mathbf{k}} \cdot \mathbf{p}_1) \end{bmatrix},$$
(12)
$$y_{31}^{\text{GW}}(t) = \begin{bmatrix} 1 + \hat{\mathbf{k}} \cdot \hat{\mathbf{n}}_2 \end{bmatrix} \begin{bmatrix} \Psi_2(t + \hat{\mathbf{k}} \cdot \mathbf{p}_3 - L) - \Psi_2(t + \hat{\mathbf{k}} \cdot \mathbf{p}_1) \end{bmatrix}$$
(13)

[in the notation of Ref. [2], our y_{21} , y_{31} , and $\hat{\mathbf{k}}$ correspond to y_{31} , y_{21} , and $-\hat{\mathbf{k}}$, respectively] where

$$\Psi_j(t) = \frac{\Phi_j(t)}{1 - (\hat{\mathbf{k}} \cdot \hat{\mathbf{n}}_j)^2}, \quad \Phi_j(t) = \frac{1}{2} \hat{\mathbf{n}}'_j \cdot \mathsf{H}(t) \cdot \hat{\mathbf{n}}_j \quad (14)$$

[the prime denotes vector transposition]. The two Ψ_i terms in each of Eqs. (12) and (13) correspond to the events of emission (at spacecraft 2 and 3, respectively) and reception (at spacecraft 1) of a laser photon packet; the time of the emission event is therefore retarded by an armlength *L*. The $\hat{\mathbf{k}} \cdot \mathbf{p}_i$ terms represent the retardation of the gravitational wavefronts to the positions of the spacecraft. The other four one-way Doppler responses are obtained by cyclical permutation of the indices $(1 \rightarrow 2, 2 \rightarrow 3, 3 \rightarrow 1)$.

Our approximation to the GW response of the moving LISA array is obtained simply by interpreting Eqs. (13) and (14) as written in the SSB ecliptic frame, and by adopting the time-dependent equations (4) for \mathbf{p}_i and $\hat{\mathbf{n}}_i$. Note that $\Phi_j(t)$ can then be written either as $\Phi_j(t) = \frac{1}{2}\hat{\mathbf{n}}'_j(t) \cdot \mathbf{H}(t) \cdot \hat{\mathbf{n}}_j(t)$, or $\Phi_j(t) = \frac{1}{2}(\hat{\mathbf{n}}_j^L)' \cdot \mathbf{H}^L(t) \cdot \hat{\mathbf{n}}_j^L$. The time-dependent rotation of the $\hat{\mathbf{n}}_i(t)$ introduces an amplitude modulation of the responses, generating sidebands at frequency multiples of 1/yr; the time dependence of the wavefront-retardation products $\hat{\mathbf{k}} \cdot \mathbf{p}_i(t)$ introduces a time-dependent Doppler shift caused by the relative motion of the spacecraft with respect to the SSB frame.

D. Chirping-binary waveforms

In the Newtonian limit, the GW signal emitted by a binary system located in the direction $\hat{\mathbf{k}}$ can be written in the form of Eq. (7), with

$$h^{+}(t) = h_{0}^{+} \cos[\phi_{s}(t) + \phi_{0}], \quad h^{\times}(t) = h_{0}^{\times} \sin[\phi_{s}(t) + \phi_{0}].$$
(15)

Here ϕ_0 is an arbitrary constant phase, and the constant amplitudes h_0^+ and h_0^{\times} are given by

$$h_0^+ = h_0 (1 + \cos^2 \iota)/2, \quad h_0^\times = h_0 \cos \iota,$$
 (16)

where ι is the angle between the normal to the orbital plane of the binary and the direction of propagation $-\hat{\mathbf{k}}$, and where

$$h_0 \simeq \frac{4(G\mathcal{M}_c)^{5/3}}{c^4 D} \left[\frac{\omega}{2}\right]^{2/3},$$
 (17)

with $\mathcal{M}_c = m_1^{3/5} m_2^{3/5} / (m_1 + m_2)^{1/5}$ the chirp mass, ω the angular frequency of the GW at t=0, and D the luminosity

distance to the source. Last, the phase $\phi_s(t)$ is given, to the first post-Newtonian order, by [24]

$$\phi_{s}(t) \approx \tilde{\phi}_{s}(t) - \tilde{\phi}_{s}(0),$$

$$\tilde{\phi}_{s}(t) = -\frac{2M}{\mu} \Theta^{5/8}(t) \bigg[1 + \bigg(\frac{3715}{8064} + \frac{55}{96} \frac{\mu}{M} \bigg) \Theta^{-1/4}(t) \bigg],$$

(18)

where

$$\Theta(t) = \frac{\mu c^3}{5GM^2} (t_c - t),$$
(19)

$$t_{c} = \frac{GM^{2}}{\mu c^{3}} \frac{5}{256} \frac{1}{x_{0}^{4}} \bigg[1 + \bigg(\frac{743}{252} + \frac{924}{252} \frac{\mu}{M} \bigg) x_{0} \bigg],$$

$$x_{0} = \bigg[\frac{GM\omega}{2c^{3}} \bigg]^{2/3}, \qquad (20)$$

with $M = m_1 + m_2$ the total mass of the binary, and $\mu = m_1 m_2 / M$ the reduced mass. The time t_c is the time to coalescence of the binary from the initial instant t=0.

In Table I, for binaries consisting of various combinations of white dwarfs (WDs, with $m = 0.35 M_{\odot}$), neutron stars (NSs, with $m = 1.4 M_{\odot}$), and black holes (BHs, with m $= 6M_{\odot}$), and for various fiducial GW frequencies within the LISA band, we show the contributions to the evolution of GW frequency over one year caused by terms at the Newtonian (N) and first post-Newtonian (1PN) order. The table shows that at frequencies smaller or equal to 10^{-3} Hz, the evolution of frequency is negligible. At frequencies approaching 10 mHz, the change in frequency becomes significant, and needs to be included in the model of the signal; however, only the first derivative of the frequency is needed up to about 50 mHz. In binaries with WDs of mass $\sim 0.35 M_{\odot}$, above ~ 20 mHz the WDs fill their Roche lobe, and the dynamical evolution of the system is then determined by tidal interaction between the stars. In binaries with either a NS or a BH, post-Newtonian effects become important at about \sim 50 mHz. At 1 Hz and above, these binaries will coalesce in less than 1 yr; furthermore, population studies [25] suggest that the expected number of binaries above 50 mHz containing neutron stars and black holes is negligible. (The effects of frequency evolution in the LISA response to GW signals from inspiraling binaries are also discussed in Ref. [26].)

Therefore, for sufficiently small binary masses, for sufficiently small GW frequencies (and definitely for all nontidally-interacting binaries that contain WDs), we can approximate the phase of the signal by Taylor-expanding it, and then neglecting terms of cubic and higher order. The resulting expression for the signal phase $\phi_s(t)$ is

$$\phi_s(t) \simeq \omega t + \frac{1}{2} \dot{\omega} t^2, \quad \text{where} \quad \dot{\omega} = \frac{48}{5} \left(\frac{G\mathcal{M}_c}{2c^3} \right)^{5/3} \omega^{11/3}.$$
(21)

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TABLE I. Contributions to the evolution of GW frequency for various types of compact, stellar-mass binaries (white dwarfs with $m = 0.35M_{\odot}$, neutron stars with $m = 1.4M_{\odot}$, and black holes with $m = 6M_{\odot}$), for selected (initial) GW frequencies within the LISA band. The contributions are expressed as GW cycles over one year of evolution, and the effects of Newtonian-order (N) and first post-Newtonian-order (1PN) terms are shown separately. The column labeled "Doppler" reports the integrated phase shift (in cycles) due to the increased Doppler shifting of the source as the frequency increases [see Eq. (45)], where significant. At $f = 10^{-3}$ Hz there is no significant evolution of GW frequency over one year. The symbol \dagger indicates that the Taylor expansion of the phase given by Eq. (21) is accurate to within a quarter of a cycle. Numbers are not shown where a binary of a given class cannot exist at a given frequency. Some of the conclusions that can be drawn from this table are apparent also in Figs. 10 and 12 of Ref. [20]: up to about 1 mHz, LISA cannot differentiate (using 1 yr of data) between a monochromatic binary and a chirping binary (see Fig. 10 of Ref. [20]); above that frequency, chirping becomes appreciable (one additional GW cycle over a year in this table corresponds to a frequency shift of one bin in Fig. 12 of Ref. [20]), but we see that it can still be modeled faithfully by the linear-chirp model of Eq. (21).

| | $f = 10^{-3}$ Hz | | $f = 2 \times 10^{-2} \text{ Hz}$ | | $f = 5 \times 10^{-2} \text{ Hz}$ | | | $f = 10^{-1} \text{ Hz}$ | | |
|--------|------------------|-----|-----------------------------------|------|-----------------------------------|------|---------|--------------------------|-----|---------|
| Binary | Ν | 1PN | Ν | 1PN | Ν | 1PN | Doppler | Ν | 1PN | Doppler |
| WD-WD | 0 | 0 | 24^{\dagger} | 0 | | _ | | | _ | |
| WD-NS | 0 | 0 | 69^{\dagger} | 0 | | _ | | | _ | |
| WD-BH | 0 | 0 | 190^{\dagger} | 0 | | | | | | |
| NS-NS | 0 | 0 | 240^{\dagger} | 0 | 6.9×10^{3} | 3.4 | 0 | 9.3×10^{4} | 78 | 2.7 |
| NS-BH | 0 | 0 | 740 | 0.33 | 2.2×10^{4} | 19.0 | 0.66 | 3.5×10^{5} | 640 | 8.5 |

E. TDI responses

The response of the second-generation TDI observables to a transverse-traceless, plane GW is obtained by setting $y_{ij}(t) = y_{ij}^{\text{GW}}(t)$ [according to Eqs. (12) and (13)] in the TDI expressions of Ref. [9,10]. For instance, the GW response of the second-generation TDI observable X_1 is given by

$$X_{1}^{GW} = \underbrace{\left[(y_{31}^{GW} + y_{13,2}^{GW}) + (y_{21}^{GW} + y_{12,3}^{GW})_{,22} - (y_{21}^{GW} + y_{12,3}^{GW}) - (y_{31}^{GW} + y_{13,2}^{GW})_{,33}\right]}_{X^{GW}(t)} - \underbrace{\left[(y_{31}^{GW} + y_{13,2}^{GW}) + (y_{21}^{GW} + y_{12,3}^{GW})_{,22} - (y_{21}^{GW} + y_{12,3}^{GW}) - (y_{31}^{GW} + y_{13,2}^{GW})_{,33}\right]_{,2233}}_{X^{GW}(t-2L_2-2L_3) \simeq X^{GW}(t-4L)}$$
(22)

As anticipated above, here we are disregarding the effects introduced by the time dependence of light travel times, and by the rotation-induced difference between clockwise and counterclockwise light travel times [27]. Each of the two terms delimited by square brackets in Eq. (22) corresponds to the GW response of the first-generation Michelson observable X [2]. The TDI observables X_2 and X_3 are obtained by cyclical permutation of indices in Eq. (22). Likewise, the second-generation Sagnac observables α_1 , α_2 , and α_3 can be written in terms of the first-generation Sagnac observables α , β , and γ [9,10]:

$$\alpha_{1}^{\text{GW}}(t) = \alpha^{\text{GW}}(t) - \alpha^{\text{GW}}(t - L_{1} - L_{2} - L_{3})$$

$$\simeq \alpha^{\text{GW}}(t) - \alpha^{\text{GW}}(t - 3L).$$
(23)

We shall now assemble the Doppler measurements y_{ij}^{GW} from the various ingredients that enter Eqs. (12) and (13). We begin with the functions Φ_j of Eq. (14), which can be rewritten as a linear combination of the two GW polarizations $h^+(t)$ and $h^{\times}(t)$:

$$\Phi_{i}(t) = F_{i}^{+}(t)h^{+}(t) + F_{i}^{\times}(t)h^{\times}(t), \qquad (24)$$

where

$$F_{i}^{+}(t) = u_{i}(t)\cos 2\psi + v_{i}(t)\sin 2\psi, \qquad (25)$$

$$F_j^{\times}(t) = v_j(t) \cos 2\psi - u_j(t) \sin 2\psi.$$
(26)

The modulation functions $u_i(t)$ and $v_i(t)$ depend rather intricately on the LISA-to-SSB (O₂) and source-to-SSB (O₁) rotations; thus, $u_i(t)$ and $v_i(t)$ depend on time through $\eta(t)$ and $\xi(t)$, and on the position of the source in the sky, given by the ecliptic coordinates β and λ . Explicitly, we have

$$u_{i}(t) = U_{0}\cos(-2\gamma_{i}) + U_{1}\cos(\delta - 2\gamma_{i}) + U_{2}\cos(2\delta - 2\gamma_{i}) + U_{3}\cos(3\delta - 2\gamma_{i}) + U_{4}\cos(4\delta - 2\gamma_{i}) + \left(\frac{1}{4} - \frac{3}{8}\cos^{2}\zeta\right)\cos^{2}\beta - \frac{1}{8}\sin 2\zeta\sin 2\beta\cos \delta + \frac{1}{4}\cos^{2}\zeta\left(1 - \frac{1}{2}\cos^{2}\beta\right)\cos 2\delta,$$
(27)

$$v_{i}(t) = V_{0}\sin(-2\gamma_{i}) + V_{1}\sin(\delta - 2\gamma_{i}) + V_{3}\sin(3\delta - 2\gamma_{i})$$
$$+ V_{4}\sin(4\delta - 2\gamma_{i}) - \frac{1}{4}\sin 2\zeta\cos\beta\sin\delta$$
$$+ \frac{1}{4}\cos^{2}\zeta\sin\beta\sin2\delta, \qquad (28)$$

where

$$\delta(t) = \lambda - \eta(t) = \lambda - \eta_0 - \Omega t, \qquad (29)$$

$$\gamma_i = \lambda - \eta_0 - \xi_0 - \sigma_i \tag{30}$$

[see Eq. (3) for the definition of σ_i , and remember that $\eta = \Omega t + \eta_0$, $\xi = -\Omega t + \xi_0$], and where the coefficients U_I and V_I are given by

$$U_0 = \frac{1}{16} (1 + \sin^2 \beta) (1 - \sin \zeta)^2, \qquad (31)$$

$$U_1 = -\frac{1}{8}\sin 2\beta \cos \zeta (1 - \sin \zeta),$$
 (32)

$$U_2 = \frac{3}{8}\cos^2\beta\cos^2\zeta,\tag{33}$$

$$U_3 = \frac{1}{8}\sin 2\beta \cos \zeta (1 + \sin \zeta),$$
 (34)

$$U_4 = \frac{1}{16} (1 + \sin^2 \beta) (1 + \sin \zeta)^2, \tag{35}$$

$$V_0 = -\frac{1}{8}\sin\beta(1-\sin\zeta)^2,$$
 (36)

$$V_1 = \frac{1}{4} \cos\beta \cos\zeta (1 - \sin\zeta), \qquad (37)$$

$$V_3 = \frac{1}{4} \cos\beta \cos\zeta (1 + \sin\zeta), \qquad (38)$$

$$V_4 = \frac{1}{8} \sin \beta (1 + \sin \zeta)^2,$$
 (39)

with $\zeta = -\pi/6$. Expanding the antenna patterns $F_j^+(t)$ and $F_j^{\times}(t)$ of Eq. (24), and using trigonometric identities to absorb the initial phase ϕ_0 into constant coefficients, the functions $\Phi_j(t)$ can be finally written as

$$\Phi_{j}(t) = a^{(1)}u_{j}(t)\cos\phi_{s}(t) + a^{(2)}v_{j}(t)\cos\phi_{s}(t) + a^{(3)}u_{j}(t)\sin\phi_{s}(t) + a^{(4)}v_{j}(t)\sin\phi_{s}(t), \quad (40)$$

where the constant amplitudes $a^{(k)}$ are given by

$$a^{(1)} = h_0^+ \cos \phi_0 \cos 2\psi - h_0^\times \sin \phi_0 \sin 2\psi, \qquad (41)$$

$$a^{(2)} = h_0^+ \cos \phi_0 \sin 2\psi + h_0^\times \sin \phi_0 \cos 2\psi, \qquad (42)$$

$$a^{(3)} = -h_0^+ \sin \phi_0 \cos 2\psi - h_0^\times \cos \phi_0 \sin 2\psi, \qquad (43)$$

$$a^{(4)} = -h_0^+ \sin \phi_0 \sin 2\psi + h_0^{\times} \cos \phi_0 \cos 2\psi.$$
(44)

Because the time scale of detector motion is much longer than the typical GW period (and because we are neglecting the evolution of the GW amplitude h_0), it is sufficient to apply the retardations $\mathbf{\hat{k}} \cdot \mathbf{p}_i(t)$ of Eqs. (12) and (13) to the GW phase:

$$\phi_s(t + \hat{\mathbf{k}} \cdot \mathbf{p}_i) \simeq \omega t + \frac{1}{2} \dot{\omega} t^2 + (\omega + \dot{\omega} t) \hat{\mathbf{k}} \cdot [\mathbf{r} + \mathsf{O}_2(t) \cdot \mathbf{p}_i^L]$$

. . .

$$\simeq \underbrace{\omega t + \frac{1}{2} \dot{\omega} t^{2} + (\omega + \dot{\omega} t) R \cos \beta \cos(\Omega t + \eta_{o} - \lambda)}_{\phi(t)} + \underbrace{\omega \mathbf{\hat{k}} \cdot [\mathsf{O}_{2}(t) \cdot \mathbf{p}_{i}^{L}]}_{2\omega Ld_{i}(t)}, \tag{45}$$

where $\phi(t)$ is the GW phase retarded to the position of the LISA guiding center, and where we have defined

$$d_{i}(t) \equiv \frac{\hat{\mathbf{k}} \cdot [\mathbf{O}_{2}(t) \cdot \mathbf{p}_{i}^{L}]}{2L} = \frac{\sqrt{3}}{8} \cos \beta \cos \gamma_{i} + \frac{1}{4} \sin \beta \cos(\delta - \gamma_{i}) - \frac{\sqrt{3}}{24} \cos \beta \cos(2\delta - \gamma_{i})$$

$$(46)$$

(setting $\zeta = -\pi/6$). Equations (12) and (13) contain also the projection factors $\hat{\mathbf{k}} \cdot \hat{\mathbf{n}}_i(t)$, which are given explicitly by

$$c_{i}(t) \equiv -\hat{\mathbf{k}} \cdot [\mathbf{O}_{2}(t) \cdot \hat{\mathbf{n}}_{i}^{L}] = \frac{3}{4} \cos \beta \sin \gamma_{i}$$
$$-\frac{\sqrt{3}}{2} \sin \beta \sin(\delta - \gamma_{i}) + \frac{1}{4} \cos \beta \sin(2\delta - \gamma_{i}).$$
(47)

The functions $c_i(t)$ and $d_i(t)$ are related by $d_1 = (c_2 - c_3)/6$, $d_2 = (c_3 - c_1)/6$, and $d_3 = (c_1 - c_2)/6$. Substituting the expressions (13) [and similar ones] for the y_{ij}^{GW} into Eq. (22) for X_1^{GW} , we get after some algebra

$$X_1^{\text{GW}}(t) = 4\,\omega L\,\sin(\omega L)\sin(2\,\omega L)\sum_{k=1}^4 a^{(k)}X_1^{(k)}(t); \quad (48)$$

the functions $X_1^{(k)}(t)$ are given by

$$\begin{bmatrix} X_1^{(1)} \\ X_1^{(2)} \end{bmatrix} = -\begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 7x/2] \\ + \operatorname{sinc}[(1-c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 9x/2] \} \\ + \begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 9x/2] \\ + \operatorname{sinc}[(1-c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 7x/2] \}, \quad (49)$$

$$\begin{bmatrix} X_{1}^{(5)} \\ X_{1}^{(4)} \end{bmatrix} = \begin{bmatrix} u_{2}(t) \\ v_{2}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{2})x/2] \cos[\phi(t) - xd_{2} - 7x/2] \\ + \operatorname{sinc}[(1-c_{2})x/2] \cos[\phi(t) - xd_{2} - 9x/2] \} \\ - \begin{bmatrix} u_{3}(t) \\ v_{3}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{3})x/2] \cos[\phi(t) - xd_{3} - 9x/2] \\ + \operatorname{sinc}[(1-c_{3})x/2] \cos[\phi(t) - xd_{3} - 7x/2] \}, \quad (50)$$

where $x = \omega L$ and sinc ... = $(\sin ...)/(...)$. The GW responses for X_2 and X_3 can be obtained by cyclical permutation of the spacecraft indices.

The GW response for α_1 can be written in similar form:

$$\alpha_1^{\rm GW}(t) = 2\,\omega L\,\sin\!\left(\frac{3}{2}\,\omega L\right)\!\sum_{k=1}^{k=4}\,a^{(k)}\alpha_1^{(k)}(t),\qquad(51)$$

where

(2)

$$\begin{bmatrix} \alpha_{1}^{(1)} \\ \alpha_{1}^{(2)} \end{bmatrix} = \begin{bmatrix} u_{1}(t) \\ v_{1}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{1})x/2] \operatorname{cos}[\phi(t) - xd_{1} - 3x] - \operatorname{sinc}[(1-c_{1})x/2] \operatorname{cos}[\phi(t) - xd_{1} - 3x] \} \\ + \begin{bmatrix} u_{2}(t) \\ v_{2}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{2})x/2] \operatorname{cos}[\phi(t) - xd_{2} - 2x] - \operatorname{sinc}[(1-c_{2})x/2] \operatorname{cos}[\phi(t) - xd_{2} - 4x] \} \\ + \begin{bmatrix} u_{3}(t) \\ v_{3}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{3})x/2] \operatorname{cos}[\phi(t) - xd_{3} - 4x] - \operatorname{sinc}[(1-c_{3})x/2] \operatorname{cos}[\phi(t) - xd_{3} - 2x] \},$$
(52)

$$\begin{bmatrix} \alpha_1^{(3)} \\ \alpha_1^{(4)} \end{bmatrix} = \begin{bmatrix} u_1(t) \\ v_1(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x] - \operatorname{sinc}[(1-c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x] \} \\ + \begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 2x] - \operatorname{sinc}[(1-c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 4x] \} \\ + \begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 4x] - \operatorname{sinc}[(1-c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 2x] \}.$$
(53)

The α_2 and α_3 combinations are again obtained by cyclical permutation of the spacecraft indices.

For the second-generation TDI observable ζ_1 (see Ref. [10]; ζ_1 is uniquely determined in the equal-armlength limit, unlike in the general case) we find

$$\zeta_1^{\rm GW}(t) = 2\,\omega L\,\sin\!\left(\frac{1}{2}\,\omega L\right) \sum_{k=1}^4 a^{(k)} \zeta_1^{(k)}(t),\tag{54}$$

with

$$\begin{bmatrix} \zeta_{1}^{(1)} \\ \zeta_{1}^{(2)} \end{bmatrix} = \begin{bmatrix} u_{1}(t) \\ v_{1}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{1})x/2] \operatorname{cos}[\phi(t) - xd_{1} - 3x] - \operatorname{sinc}[(1-c_{1})x/2] \operatorname{cos}[\phi(t) - xd_{1} - 3x] \} \\ + \begin{bmatrix} u_{2}(t) \\ v_{2}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{2})x/2] \operatorname{cos}[\phi(t) - xd_{2} - 3x] - \operatorname{sinc}[(1-c_{2})x/2] \operatorname{cos}[\phi(t) - xd_{2} - 3x] \} \\ + \begin{bmatrix} u_{3}(t) \\ v_{3}(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_{3})x/2] \operatorname{cos}[\phi(t) - xd_{3} - 3x] - \operatorname{sinc}[(1-c_{3})x/2] \operatorname{cos}[\phi(t) - xd_{3} - 3x] \},$$
(55)

$$\begin{bmatrix} \zeta_1^{(3)} \\ \zeta_1^{(4)} \end{bmatrix} = \begin{bmatrix} u_1(t) \\ v_1(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x] - \operatorname{sinc}[(1-c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x] \} \\ + \begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 3x] - \operatorname{sinc}[(1-c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 3x] \} \\ + \begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 3x] - \operatorname{sinc}[(1-c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 3x] \}.$$
(56)

Finally, the *optimal* TDI observables [21], which here we denote as \overline{A} , \overline{E} , and \overline{T} to distinguish them from the optimal combinations A, E, and T derived within first-generation TDI, are defined as linear combinations of α_1 , α_2 , and α_3 :

$$\bar{A} = \frac{1}{\sqrt{2}} (\alpha_3 - \alpha_1),$$

$$\bar{E} = \frac{1}{\sqrt{6}} (\alpha_1 - 2\alpha_2 + \alpha_3),$$

$$\bar{T} = \frac{1}{\sqrt{3}} (\alpha_1 + \alpha_2 + \alpha_3).$$
(57)

It is clear that \overline{A} , \overline{E} , and \overline{T} are also optimal, in the sense discussed in Ref. [21]: this is because they can be written as time-delayed combinations of the first-generation optimal TDI observables, such as $\overline{A} = A(t) - A(t-3L)$, $\overline{E} = E(t)$ -E(t-3L), and $\overline{T} = T(t) - T(t-3L)$; since by construction the noises that enter A, E, and T are uncorrelated, it follows that the noises that enter \overline{A} , \overline{E} , and \overline{T} are also uncorrelated, making these observables optimal.

We recall that in the high-frequency part of the LISA band (i.e., for frequencies equal to or larger than 5 mHz), there exist three independent TDI GW observables (such as \overline{A} , \overline{E} , and \overline{T} , or X_1 , X_2 , and X_3). However, for frequencies smaller than 5 mHz, there are essentially only two independent observables: this is especially obvious if we reason in terms of the optimal combinations, where we observe that for low frequencies the GW signal response of \overline{T} declines much faster than the responses of \overline{A} and \overline{E} [5,21].

F. TDI responses in the long-wavelength limit

The long-wavelength (LW) approximation to the GW responses is obtained by taking the leading-order terms of the generic expressions in the limit of $\omega L \rightarrow 0$. For instance, for X_1 [Eqs. (48)–(50)], we get

$$X_{1,\text{LW}}^{\text{GW}} \approx 16(\omega L)^{3} \{ [u_{3}(t) - u_{2}(t)] \\ \times [a^{(1)} \sin \phi(t) - a^{(3)} \cos \phi(t)] + [v_{3}(t) - v_{2}(t)] \\ \times [a^{(2)} \sin \phi(t) - a^{(4)} \cos \phi(t)] \},$$
(58)

with $a^{(k)}$ given by Eqs. (41)–(44), and $u_i(t), v_i(t)$ by Eqs. (27), (28). The LW responses for X_2 and X_3 can be obtained by cyclical permutation of the indices. Adopting the notation of Ref. [2], we find also that

$$X_{1,\text{LW}}^{\text{GW}}(t) \simeq 8L^3 [(\hat{\mathbf{n}}_3^L)' \cdot \ddot{\mathsf{H}}^L(t) \cdot \hat{\mathbf{n}}_3^L - (\hat{\mathbf{n}}_2^L)' \cdot \ddot{\mathsf{H}}^L(t) \cdot \hat{\mathbf{n}}_2^L],$$
(59)

where the triple overdot denotes the third time derivative, $\mathbf{\hat{n}}_{i}^{L}$ is given by Eq. (2), and $\mathbf{H}^{L}(t)$ is given by Eq. (11). The GW responses of the Sagnac observables $\alpha_{i,\text{LW}}^{\text{GW}}$ are

The GW responses of the Sagnac observables $\alpha_{i,LW}^{GW}$ are equal simply to $\frac{3}{8}X_{i,LW}^{GW}$. From Eqs. (57) we then get the LW GW responses \overline{A}_{LW}^{GW} , \overline{E}_{LW}^{GW} , and \overline{T}_{LW}^{GW} :

$$\overline{A}_{LW}^{GW} \approx 3\sqrt{2}(\omega L)^{3} \{ [2u_{2}(t) - u_{1}(t) - u_{3}(t)] \\ \times [a^{(1)} \sin \phi(t) - a^{(3)} \cos \phi(t)] + [2v_{2}(t) - v_{1}(t) \\ - v_{3}(t)] [a^{(2)} \sin \phi(t) - a^{(4)} \cos \phi(t)] \},$$
(60)

$$\begin{split} \bar{E}_{\rm LW}^{\rm GW} &\simeq 3\sqrt{6}(\omega L)^3 \{ [u_3(t) - u_1(t)] \\ &\times [a^{(1)} \sin \phi(t) - a^{(3)} \cos \phi(t)] \\ &+ [v_3(t) - v_1(t)] [a^{(2)} \sin \phi(t) - a^{(4)} \cos \phi(t)] \}, \end{split}$$
(61)

$$\bar{T}_{\rm LW}^{\rm GW} \simeq O[(\omega L)^4]. \tag{62}$$

III. NOISE SPECTRAL DENSITY

The spectral density of noise for the first-generation TDI observables *X*, *Y*, *Z*, α , β , γ , *A*, *E*, and *T* is given in Refs. [3,21] in the case of an equilateral LISA array, assuming that the noises appearing in all the proof masses and optical paths are uncorrelated. The finite-difference relations between first-and second-generation TDI observables [such as $X_1(t) = X(t) - X(t-4L)$, $\alpha_1(t) = \alpha(t) - \alpha(t-3L)$] imply simple modifications to the first-generation noise densities: for instance,

$$S_{X_1}(\omega) = 4\sin^2(2\omega L)S_X(\omega), \tag{63}$$

$$S_{\alpha_1}(\omega) = 4 \sin^2(3\omega L/2) S_{\alpha}(\omega); \qquad (64)$$

inserting the expression of $S_X = S_Y = S_Z$ from Ref. [3] into Eq. (63) yields



FIG. 5. Spectral densities of noise for the second-generation TDI observables X_1 (continuous), α_1 (dashed), \overline{A} (dotted), and \overline{T} (dash-dotted).

$$S_{X_1} = S_{X_2} = S_{X_3} = 64 \sin^2(\omega L) \sin^2(2\omega L) [2(1 + \cos^2 \omega L)S^{\text{pm}} + S^{\text{op}}]$$
(65)

where $S^{pm}=2.54\times10^{-48}f^{-2}$ Hz⁻¹ and $S^{op}=1.76\times10^{-37}f^2$ Hz⁻¹ are the fractional-frequency-fluctuation spectral densities of proof-mass noise and optical-path noise, respectively [3]. These values correspond to a rms singleproof-mass acceleration noise of 3×10^{-15} m s⁻² Hz^{-1/2}, and to a rms aggregate optical-path noise 20 $\times10^{-12}$ m Hz^{-1/2}, as quoted in the LISA Pre-Phase A Study [23]. For the other TDI observables we find

$$S_{\alpha_1} = S_{\alpha_2} = S_{\alpha_3} = 8 \sin^2(3 \,\omega L/2) \{ [4 \sin^2(3 \,\omega L/2) + 8 \sin^2(\omega L/2)] S^{\text{pm}} + 3 S^{\text{op}} \},$$
(66)

$$S_{\bar{A}} = S_{\bar{E}} = 32 \sin^2(\omega L/2) \sin^2(3\omega L/2) \{ [6+4\cos(\omega L) + 2\cos(2\omega L)] S^{\text{pm}} + [2+\cos(\omega L)] S^{\text{op}} \},$$
(67)

$$S_{\overline{T}} = 8[1 + 2\cos(\omega L)]^{2}\sin^{2}(3\omega L/2) \times [4\sin^{2}(\omega L/2)S^{pm} + S^{op}].$$
(68)

All the noise spectra are shown in Fig. 5. In the longwavelength approximation, the noise expressions simplify to

$$S_{X_1}^{\text{LW}} = S_{X_2}^{\text{LW}} = S_{X_3}^{\text{LW}} \approx 256(\omega L)^2 [4(\omega L)^2 S^{\text{pm}} + (\omega L)^2 S^{\text{op}}], \quad (69)$$

$$S_{\alpha_1}^{\text{LW}} = S_{\alpha_2}^{\text{LW}} = S_{\alpha_3}^{\text{LW}} \approx 18(\omega L)^2 [11(\omega L)^2 S^{\text{pm}} + 3S^{\text{op}}],$$
(70)

$$S_{\bar{A}}^{\rm LW} = S_{\bar{E}}^{\rm LW} \simeq 54(\omega L)^2 [4(\omega L)^2 S^{\rm pm} + (\omega L)^2 S^{\rm op}], \tag{71}$$

$$S_{\bar{T}}^{LW} \simeq 162(\omega L)^2 [(\omega L)^2 S^{pm} + S^{op}].$$
 (72)

IV. OPTIMAL FILTERING OF THE LISA DATA

In this section we develop a maximum-likelihood (ML) formalism to detect GW signals from moderately chirping binaries and to estimate their parameters, by analyzing the time series of the TDI observables. ML detection is based on maximizing the *likelihood ratio* $\Lambda(\theta^{l})$ over the source parameters θ^i ; this ratio is proportional to the probability that the observed detector output could have been produced by a GW source with parameters θ^i , plus instrument noise. The magnitude of the maximum indicates the probability that a signal is indeed present, while its location indicates the most likely parameters (the ML *parameter estimators*). Under the assumption of Gaussian, stationary, additive noise, $\log \Lambda(\theta)$ is computed by correlating the detector output, x(t), with the expected GW detector response $h_{\theta^i}(t)$, while weighting the correlation in the frequency domain by the inverse spectral density of instrument noise, $S_n^{-1}(f)$. The family of GW responses $\{h_{\theta^i}(t)\}$, divided (in the frequency domain) by $S_n(f)$ to incorporate the noise weighting, are known as *opti*mal filters.

In Sec. IV A we describe the computation of Λ and of the ML parameter estimators for the optimal filters derived from the GW responses of Sec. II, and we show how to maximize Λ algebraically over the four source amplitudes $a^{(k)}$. The amplitude-maximized log Λ (known as \mathcal{F}) is then used as a *detection statistic* to search for the most likely GW source, by maximizing it over the remaining source parameters [here denoted by ξ^{μ} ; thus, $\theta^i \equiv (a^{(k)}, \xi^{\mu})$]. In Sec. IV B we study the statistical distribution of $\mathcal{F}(\xi^{\mu})$ in the absence (or presence) of a GW signal of parameters ξ^{μ} ; this distribution determines the statistical significance of observing a certain value of \mathcal{F} , for a fixed ξ^{μ} . In Sec. IV C we study the statistical significance of measuring a certain value of the *completely maximized* statistic max_{\mu}\mathcal{F}(\xi^{\mu}), which leads to the total false-alarm probability for a GW search over a range of

intrinsic parameters. See Helström [28] for an extended discussion of ML detection and parameter estimation.

A. Maximum-likelihood search method

As discussed in Secs. I and II, the LISA Doppler measurements can be recombined into the laser-noise and opticalbench-noise free TDI observables, all of which can be obtained as time-delayed combinations of three generators [2,6]. Thus, in the following we denote the TDI data as the three-vector $\mathbf{x}(t)$ (we shall very shortly specify a convenient vector basis). In the case of additive noise, we can write $\mathbf{x}(t) = \mathbf{n}(t) + \mathbf{h}(t)$, where $\mathbf{n}(t)$ represents detector noise and $\mathbf{h}(t)$ the GW response. Idealizing $\mathbf{n}(t)$ as a zero-mean, Gaussian, stationary, continuous random process, we have [28]

$$\log \Lambda = (\mathbf{x}|\mathbf{h}) - \frac{1}{2}(\mathbf{h}|\mathbf{h}), \qquad (73)$$

where the scalar product $(\ldots | \ldots)$ is defined by

$$(\mathbf{x}|\mathbf{y}) = 4 \operatorname{Re} \int_{0}^{\infty} \widetilde{\mathbf{x}}^{\dagger} \cdot \widetilde{\mathbf{S}}_{n}^{-1} \cdot \widetilde{\mathbf{y}} df;$$
(74)

here the dagger denotes transposition and complex conjugation, the tilde denotes the Fourier transform, and \tilde{S} denotes the one-sided cross spectral density matrix of detector noise, defined by the expectation value

$$\mathcal{E}[\widetilde{\mathbf{n}}(f)\widetilde{\mathbf{n}}^{\dagger}(f')] = \frac{1}{2}\,\delta(f - f')\widetilde{\mathbf{S}}_{n}(f).$$
(75)

The larger the signal with respect to the noise, the higher the probability that a ML search (performed with the appropriate optimal filter) will yield a statistically significant detection, and the better the accuracy of the ML parameter estimators. The accuracy of estimation is also better for the parameters on which the signal is strongly dependent. Signal strength is characterized by the *optimal signal-to-noise ratio* (optimal S/N),

$$\rho^2 = (\mathbf{h}|\mathbf{h}); \tag{76}$$

while the dependence of the instrument response on the parameters is characterized by the Fisher information matrix,

$$\Gamma_{ij} = \left(\frac{\partial \mathbf{h}}{\partial \theta^i} \middle| \frac{\partial \mathbf{h}}{\partial \theta^j} \right). \tag{77}$$

By the Cramèr-Rao inequality [28], the diagonal elements of Γ_{ij}^{-1} provide lower bounds on the variance of any unbiased estimators of the θ^i . In fact, the matrix Γ_{ij}^{-1} is often called the *covariance matrix*, because in the limit of high S/N the ML estimators become unbiased, and their distribution tends to a jointly Gaussian distribution with covariance matrix equal to Γ_{ij}^{-1} .

The optimal TDI observables [21] are obtained by diagonalizing the cross spectrum $\tilde{\mathbf{S}}_n$; it turns out that the eigen-

vectors \overline{A} , \overline{E} , and \overline{T} are independent of frequency. The new observables are the linear combinations of the Sagnac observables $\alpha_1, \alpha_2, \alpha_3$ given by Eq. (57), and by definition their noises are uncorrelated. With reference to Eq. (65), we define $S_{\overline{A}\overline{A}}(f) = S_{\overline{E}\overline{E}}(f) \equiv S_{\overline{A}}(\omega \equiv 2\pi f)$ and $S_{\overline{T}\overline{T}}(f) \equiv S_{\overline{T}}(\omega \equiv 2\pi f)$. It is convenient to use the optimal observables \overline{A} , \overline{E} , and \overline{T} as a basis for the LISA TDI observables, setting

$$\mathbf{x}(t) = \begin{bmatrix} \bar{A}(t) \\ \bar{E}(t) \\ \bar{T}(t) \end{bmatrix}, \quad \mathbf{h}(t) = \begin{bmatrix} \bar{A}^{\mathrm{GW}}(t) \\ \bar{E}^{\mathrm{GW}}(t) \\ \bar{T}^{\mathrm{GW}}(t) \end{bmatrix}; \quad (78)$$

the GW responses \overline{A}^{GW} , \overline{E}^{GW} , and \overline{T}^{GW} are given in Sec. II E for the case of moderately chirping binaries. For these sources, $S_{\overline{A}}(\omega)$ and $S_{\overline{T}}(\omega)$ are approximately constant over the signal bandwidth, so we can expand Eq. (73) as

$$\log \Lambda \cong T_0 \left[(\bar{A} || \bar{A}^{\text{GW}}) - \frac{1}{2} (\bar{A}^{\text{GW}} || \bar{A}^{\text{GW}}) + (\bar{E} || \bar{E}^{\text{GW}}) - \frac{1}{2} (\bar{E}^{\text{GW}} || \bar{E}^{\text{GW}}) \right] / S_{\bar{A}}(\omega) + T_0 \left[(\bar{T} || \bar{T}^{\text{GW}}) - \frac{1}{2} (\bar{T}^{\text{GW}} || \bar{T}^{\text{GW}}) \right] / S_{\bar{T}}(\omega),$$

$$(79)$$

where T_0 is the time of observation, and where we have introduced the time-domain scalar product

$$(B||C) \equiv (2/T_0) \int_0^{T_0} B(t)C(t)dt.$$
(80)

Given the linear dependence [21] of \overline{A} , \overline{E} , and \overline{T} on α_1 , α_2 , and α_3 , from Eq. (51) it follows that

$$\begin{bmatrix} \bar{A}^{GW}(t) \\ \bar{E}^{GW}(t) \\ \bar{T}^{GW}(t) \end{bmatrix}$$

= $2x \sin\left(\frac{3}{2}x\right) \sum_{k=1}^{4} a^{(k)} \begin{bmatrix} \bar{A}^{(k)}(t) \\ \bar{E}^{(k)}(t) \\ \bar{T}^{(k)}(t) \end{bmatrix}$
= $2x \sin\left(\frac{3}{2}x\right) \sum_{k=1}^{4} a^{(k)} \begin{bmatrix} \frac{1}{\sqrt{2}}(\alpha_{3}^{(k)} - \alpha_{1}^{(k)}) \\ \frac{1}{\sqrt{6}}(\alpha_{1}^{(k)} - 2\alpha_{2}^{(k)} + \alpha_{3}^{(k)}) \\ \frac{1}{\sqrt{3}}(\alpha_{1}^{(k)} + \alpha_{2}^{(k)} + \alpha_{3}^{(k)}) \end{bmatrix}$, (81)

where once again $x = \omega L$, the amplitudes $a^{(k)}$ are given by Eqs. (41)–(44), and the functions $\alpha_i^{(k)}$ are given by Eqs. (52) and (53), and by similar equations obtained by cyclical permutation of the indices. Note that the component functions $\overline{A}^{(k)}(t)$, $\overline{E}^{(k)}(t)$, and $\overline{T}^{(k)}(t)$ do not depend on the amplitudes $a^{(k)}$ (or equivalently, on h_0^+ , h_0^{\times} , ϕ_0 , and ψ); they do however depend on the remaining (intrinsic) source parameters, ω , $\dot{\omega}$, β , and λ .

The ML parameter estimators $\hat{\theta}^i$ are found by maximizing log Λ with respect to the source parameters θ^i : that is, by solving

$$\frac{\partial \log \Lambda}{\partial \theta^i} = 0. \tag{82}$$

For the $a^{(k)}$ this is accomplished easily by solving the linear system

$$\sum_{k=1}^{4} M^{(l)(k)} a^{(k)} = N^{(l)}, \quad l = 1, \dots, 4,$$
(83)

where

$$N^{(l)} = 2x \sin\left(\frac{3}{2}x\right) T_0[(\bar{A}||\bar{A}^{(l)})/S_{\bar{A}}(\omega) + (\bar{E}||\bar{E}^{(l)})/S_{\bar{A}}(\omega) + (\bar{T}||\bar{T}^{(l)})/S_{\bar{T}}(\omega)],$$
(84)

and where $M^{(l)(k)}$ is the 4×4 matrix with components

$$M^{(l)(k)} = 4x^{2} \sin^{2} \left(\frac{3}{2}x\right) T_{0}[(\bar{A}^{(l)}||\bar{A}^{(k)})/S_{\bar{A}}(\omega) + (\bar{E}^{(l)}||\bar{E}^{(k)})/S_{\bar{A}}(\omega) + (\bar{T}^{(l)}||\bar{T}^{(k)})/S_{\bar{T}}(\omega)].$$
(85)

The solution of Eq. (83) is simplified by noticing that the component functions $\overline{A}^{(k)}(t)$, $\overline{E}^{(k)}(t)$, and $\overline{T}^{(k)}(t)$ consist of simple sines and cosines with period $\sim 2 \pi/\omega$, modulated by the slowly changing functions $u_i(t)$ and $v_i(t)$ (with periods that are multiples of 1 yr). By the approximate orthogonality of sine and cosine terms, for $T_0 \approx 1$ yr the scalar products $(\overline{A}^{(k)}||\overline{A}^{(l)})$ can be approximated as

$$(\bar{A}^{(1)}||\bar{A}^{(3)}) \simeq (\bar{A}^{(2)}||\bar{A}^{(4)}) \simeq 0,$$
 (86)

and

$$(\bar{A}^{(1)}||\bar{A}^{(1)}) \simeq (\bar{A}^{(3)}||\bar{A}^{(3)}) \equiv \frac{1}{2} U_{\bar{A}},$$
 (87)

$$(\bar{A}^{(2)}||\bar{A}^{(2)}) \simeq (\bar{A}^{(4)}||\bar{A}^{(4)}) \equiv \frac{1}{2} V_{\bar{A}},$$
 (88)

$$(\bar{A}^{(1)}||\bar{A}^{(2)}) \simeq (\bar{A}^{(3)}||\bar{A}^{(4)}) \equiv \frac{1}{2}Q_{\bar{A}},$$
 (89)

$$(\bar{A}^{(1)}||\bar{A}^{(4)}) \simeq -(\bar{A}^{(2)}||\bar{A}^{(3)}) \equiv \frac{1}{2}P_{\bar{A}}, \qquad (90)$$

with similar expressions for the $\overline{E}^{(k)}$ and $\overline{T}^{(k)}$. The matrix $M^{(l)(k)}$ then simplifies to

$$M^{(l)(k)} = \frac{T_0}{2} \begin{pmatrix} U & Q & 0 & P \\ Q & V & -P & 0 \\ 0 & -P & U & Q \\ P & 0 & Q & V \end{pmatrix},$$
(91)

where the elements U, V, Q, and P are given by

$$U = 4x^2 \sin^2\left(\frac{3}{2}x\right) \left[U_{\bar{A}}/S_{\bar{A}}(\omega) + U_{\bar{E}}/S_{\bar{A}}(\omega) + U_{\bar{T}}/S_{\bar{T}}(\omega) \right],$$
(92)

$$V = 4x^{2}\sin^{2}\left(\frac{3}{2}x\right) \left[V_{\bar{A}}/S_{\bar{A}}(\omega) + V_{\bar{E}}/S_{\bar{A}}(\omega) + V_{\bar{T}}/S_{\bar{T}}(\omega)\right],$$
(93)

$$Q = 4x^{2} \sin^{2} \left(\frac{3}{2}x\right) \left[Q_{\bar{A}}/S_{\bar{A}}(\omega) + Q_{\bar{E}}/S_{\bar{A}}(\omega) + Q_{\bar{T}}/S_{\bar{T}}(\omega)\right],$$
(94)

$$P = 4x^{2} \sin^{2} \left(\frac{3}{2}x\right) \left[P_{\bar{A}} / S_{\bar{A}}(\omega) + P_{\bar{E}} / S_{\bar{A}}(\omega) + P_{\bar{T}} / S_{\bar{T}}(\omega) \right].$$
(95)

Then the analytic expressions for the maximum likelihood estimators $\hat{a}^{(k)}$ of the amplitudes $a^{(k)}$ are given by

$$\begin{pmatrix} \hat{a}^{(1)} \\ \hat{a}^{(2)} \\ \hat{a}^{(3)} \\ \hat{a}^{(4)} \end{pmatrix} = \frac{2}{T_0 \Delta} \begin{pmatrix} V & -Q & 0 & -P \\ -Q & U & P & 0 \\ 0 & P & V & -Q \\ -P & 0 & -Q & U \end{pmatrix} \cdot \begin{pmatrix} N^{(1)} \\ N^{(2)} \\ N^{(3)} \\ N^{(4)} \end{pmatrix},$$
(96)

where $\Delta = UV - Q^2 - P^2$.

Substituting the ML amplitude estimators $\hat{a}^{(k)}$ in the likelihood function Λ yields the *reduced likelihood function* Λ_r . The logarithm of Λ_r is known as the \mathcal{F} statistic; using Eqs. (79), (81), (84), and (96), we find

$$\mathcal{F} = \frac{1}{2} \sum_{l=1}^{4} \sum_{k=1}^{4} (M^{-1})^{(l)(k)} N^{(l)} N^{(k)}$$

= $(T_0 \Delta)^{-1} \{ V[(N^{(1)})^2 + (N^{(3)})^2] + U[(N^{(2)})^2 + (N^{(4)})^2] - 2Q[N^{(1)}N^{(2)} + N^{(3)}N^{(4)}] - 2P[N^{(1)}N^{(4)} - N^{(2)}N^{(3)}] \}.$
(97)

We adopt \mathcal{F} as the detection statistic of our proposed search scheme. The statistic \mathcal{F} is already maximized over the amplitudes $a^{(k)}$, which are known in this context as *extrinsic parameters*. By contrast, the ML estimators of the remaining (*intrinsic*) source parameters (ω , $\dot{\omega}$, β , and λ) are found by maximizing \mathcal{F} . In practice, this is done by correlating the detector output with a *bank* of optimal filters precomputed for many values of the intrinsic parameters.

Introducing the complex quantities

$$a^{(u)} = a^{(1)} + ia^{(3)}, (98)$$

$$a^{(v)} = a^{(2)} + ia^{(4)}, \tag{99}$$

$$W = Q + iP, \tag{100}$$

$$N^{(u)} = N^{(1)} + iN^{(3)}, \tag{101}$$

$$N^{(v)} = N^{(2)} + iN^{(4)}, (102)$$

we can write the ML amplitude estimators and the ${\ensuremath{\mathcal F}}$ statistic in the compact form

$$\hat{a}^{(u)} = 2(T_0 \Delta)^{-1} [V N^{(u)} - W^* N^{(v)}], \qquad (103)$$

$$\hat{a}^{(v)} = 2(T_0 \Delta)^{-1} [UN^{(v)} - WN^{(u)}], \qquad (104)$$

(where $\Delta = UV - |W|^2$) and

$$\mathcal{F} = (T_0 \Delta)^{-1} \{ V | N^{(u)} |^2 + U | N^{(v)} |^2 - 2 \operatorname{Re}[WN^{(u)}(N^{(v)})^*] \}.$$
(105)

In Sec. V we shall see that this expression is very suitable for numerical implementation. Equations (103)-(105) summarize the proposed ML data-analysis scheme, which uses all the available LISA data. Similar expressions hold if we analyze a single interferometric combination, such as X_1 . In Appendix A we describe a useful complex representation of the GW TDI responses that simplifies the integrals involved in the computation of \mathcal{F} and of the ML amplitude estimators.

In the LW approximation, Eqs. (103)-(105) simplify somewhat: using the $\overline{A}_{LW}^{GW}(t)$ and $\overline{E}_{LW}^{GW}(t)$ of Eqs. (60) and (61) [and remembering that $\overline{T}_{LW}^{GW}(t) \approx 0$], we go through with our formalism in parallel with Eqs. (84)–(91), and find that $P_{LW}\approx 0$, so W_{LW} is real. The complex variables $N^{(u)}$ and $N^{(v)}$ are given by the integrals

$$N_{\rm LW}^{(u)} = -2i \frac{(\omega L)^3}{S_{\bar{A}}^{\rm LW}(\omega)} \int_0^{T_0} \{3\sqrt{2}[2u_2(t) - u_1(t) - u_3(t)]\bar{A}(t)$$

$$+3\sqrt{6[u_3(t)-u_1(t)]\bar{E}(t)}e^{i\phi(t)}dt,$$
(106)

$$N_{\rm LW}^{(v)} = -2i \frac{(\omega L)^3}{S_{\bar{A}}^{\rm LW}(\omega)} \int_0^{T_0} \{3\sqrt{2}[2v_2(t) - v_1(t) - v_3(t)]\bar{A}(t) + 3\sqrt{6}[v_3(t) - v_1(t)]\bar{E}(t)\}e^{i\phi(t)}dt.$$
(107)

Analogous LW expressions hold for a single TDI observable, such as X_1 .

B. Distribution of the \mathcal{F} statistic

Crucial to a search scheme based on comparing the ML statistic \mathcal{F} with a predefined threshold is the determination of the false-alarm probability P_F (which determines how often

 \mathcal{F} will exceed the threshold in presence of noise alone) and of the detection probability P_D (which determines how often \mathcal{F} will exceed the threshold when a signal is present, resulting in correct detection). In this section we compute the probabilities P_F and P_D for the correlation of detector data against a single optimal filter (i.e., for fixed values of the intrinsic parameters).

Under the assumption of zero-mean Gaussian noise, the weighted correlations $N^{(k)}$ [Eq. (84)] are Gaussian random variables; since \mathcal{F} is a quadratic form in the $N^{(k)}$ [see Eq. (97)], it must follow the χ^2 distribution. Following Sec. III B of Ref. [22], we can diagonalize the quadratic form to find that, in the absence of the signal, $2\mathcal{F}$ follows the χ^2 distribution with $n=4 \times n_c$ degrees of freedom, where n_c is the number of independent observables included in \mathcal{F} [29]. For instance, if we use \overline{A} , \overline{E} , \overline{T} then n=12, while if we use only X_1 , then n=4. In presence of the signal, $2\mathcal{F}$ follows a noncentral χ^2 distribution with $n=4 \times n_c$ degrees of freedom and with noncentrality parameter κ equal to the optimal $(S/N)^2 = \rho^2$ [see Eq. (76)]. For instance, if we use \overline{A} , \overline{E} , \overline{T} ,

$$\kappa = \rho^2 = (\bar{A}^{\mathrm{GW}} | \bar{A}^{\mathrm{GW}}) + (\bar{E}^{\mathrm{GW}} | \bar{E}^{\mathrm{GW}}) + (\bar{T}^{\mathrm{GW}} | \bar{T}^{\mathrm{GW}})$$
(108)

(which agrees with the result derived in Ref. [21]), while if we use only X_1 ,

$$\kappa = \rho^2 = (X_1^{\text{GW}} | X_1^{\text{GW}}). \tag{109}$$

The χ^2 probability density function is

$$p_0(\mathcal{F}) = \frac{\mathcal{F}^{n/2-1}}{(n/2-1)!} \exp(-\mathcal{F})$$
(110)

for $\kappa = 0$, or

$$p_1(\rho;\mathcal{F}) = \frac{(2\mathcal{F})^{(n/2-1)/2}}{\rho^{n/2-1}} I_{n/2-1}(\rho\sqrt{2\mathcal{F}}) \exp\left(-\mathcal{F} - \frac{1}{2}\rho^2\right)$$
(111)

for $\kappa = \rho^2$, where $I_{n/2-1}$ is the (n/2-1)th-order modified Bessel function of the first kind. Thus, the false-alarm probability for a threshold \mathcal{F}_0 is

$$P_{F}(\mathcal{F}_{0}) = \int_{\mathcal{F}_{0}}^{\infty} p_{0}(\mathcal{F}) d\mathcal{F} = \exp(-\mathcal{F}_{0}) \sum_{k=0}^{n/2-1} \mathcal{F}_{0}^{k}/k!$$
(112)

(for even *n*; for odd *n* the result involves the error function) while the detection probability, in the presence of a $S/N = \rho$ signal (using the correct optimal filter), is

$$P_D(\rho;\mathcal{F}_0) = \int_{\mathcal{F}_0}^{\infty} p_1(\rho,\mathcal{F}) d\mathcal{F}; \qquad (113)$$

this integral cannot be evaluated in closed form in terms of known special functions, but it is clear that the higher the optimal S/N, the higher the detection probability.

C. False-alarm and detection probabilities for GW searches

In actual GW searches, the detector output will be correlated to a bank of optimal filters corresponding to different values of the intrinsic parameters ξ^{μ} . For a given set of detector data, the statistic $\mathcal{F}(\xi^{\mu})$ is a generalized multiparameter random process known as *random field* (see Adler's monograph [30] for a comprehensive discussion): we can use the theory of random fields to get a handle on the total falsealarm and detection probabilities for the entire filter bank.

We define the *autocovariance* C of the random field $\mathcal{F}(\xi^{\mu})$ as

$$\mathcal{C}(\xi^{\mu},\xi'^{\mu}) = \mathcal{E}_0[\mathcal{F}(\xi^{\mu})\mathcal{F}(\xi'^{\mu})] - \mathcal{E}_0[\mathcal{F}(\xi^{\mu})]\mathcal{E}_0[\mathcal{F}(\xi'^{\mu})],$$
(114)

where the expectation value \mathcal{E}_0 is computed over an ensemble of realizations of noise (in absence of the signal). In Ref. [15] the total false-alarm probability was estimated by noticing that the autocovariance function tends to zero as the displacement $\Delta \xi^{\mu} = \xi'^{\mu} - \xi^{\mu}$ increases (and in fact, it is maximum for $\Delta \xi^{\mu} = 0$). The space of intrinsic parameters may then be partitioned into a set of *elementary cells*, whereby the autocovariance is appreciably different from zero for within each cell, but negligible between cells. The number of elementary cells needed to cover the parameter space gives an estimate of the number of independent realizations of the random field (i.e., the number of statically independent ways that pure noise can be strongly correlated with one or more of the optimal filters).

There is of course some arbitrariness in choosing the boundary of the elementary cells; we define them by requiring that the autocovariance between the center and the surface be one half of the autocovariance *at the center*:

$$C(\xi^{\mu}, \xi'^{\,\mu}) = \frac{1}{2}C(\xi^{\mu}, \xi^{\mu}), \qquad (115)$$

for ξ^{μ} at cell center, ξ'^{μ} on cell boundary. Taylor-expanding the autocovariance to second order in $\Delta \xi^{\mu}$, we obtain the approximate condition

$$\mathcal{C}(\xi^{\mu},\xi'^{\mu}) \simeq \mathcal{C}(\xi^{\mu},\xi^{\mu}) + \frac{1}{2} \left. \frac{\partial^2 \mathcal{C}(\xi^{\mu},\xi'^{\mu})}{\partial \xi'^{\rho} \partial \xi'^{\sigma}} \right|_{\xi'^{\mu} = \xi^{\mu}} \Delta \xi^{\rho} \Delta \xi^{\sigma}$$
$$= \frac{1}{2} \mathcal{C}(\xi^{\mu},\xi^{\mu}), \qquad (116)$$

with implicit summation over ρ and σ . Within the approximation (necessary to obtain results in simple analytical form), the cell boundary is the (hyper-)ellipse defined by $G_{\rho\sigma}\Delta\xi^{\rho}\Delta\xi^{\sigma}=1/2$, where [31]

$$G_{\rho\sigma} = -\frac{1}{2} \frac{1}{\mathcal{C}(\xi^{\mu}, \xi^{\mu})} \frac{\partial^2 \mathcal{C}(\xi^{\mu}, \xi'^{\mu})}{\partial \xi'^{\rho} \partial \xi'^{\sigma}} \bigg|_{\xi'^{\mu} = \xi^{\mu}}$$
(117)

(in Appendix B we shall derive a relation between this $G_{\rho\sigma}$ and the Fisher information matrix). The volume V_{cell} of the elementary cell is then

$$V_{\text{cell}} = \frac{(\pi/2)^{K/2}}{\Gamma(K/2+1)\sqrt{\det G_{\rho\sigma}}},$$
(118)

where *K* is the number of intrinsic parameters, and Γ is the gamma function. The total number of elementary cells within the parameter volume *V* is given by

$$N_{\text{cell}} = \frac{\Gamma(K/2+1)}{(\pi/2)^{K/2}} \int_{V} \sqrt{\det G_{\rho\sigma}} dV.$$
(119)

As discussed above, we consider the values of the statistic \mathcal{F} within each cell as independent random variables, which in the absence of signal are distributed according to Eq. (110). By our definition of false alarms, the probability that \mathcal{F} will not exceed the threshold \mathcal{F} in a given cell is just $1 - P_F(\mathcal{F}_0)$; the probability that \mathcal{F} will not exceed the threshold \mathcal{F}_0 in any of the cells is

$$1 - P_{F, \text{tot}}(\mathcal{F}_0) \simeq [1 - P_F(\mathcal{F}_0)]^{N_{\text{cell}}};$$
(120)

this $P_{F,tot}(\mathcal{F}_0)$ is therefore the total false-alarm probability for our detection scheme.

When the signal is present, a precise calculation of the probability distribution function of \mathcal{F} is nontrivial, since the presence of the signal makes the random process $\mathbf{x}(t)$ non-stationary. However, we can still use the detection probability given by Eq. (113) for known intrinsic parameters as a substitute for the detection probability when the parameters are unknown. This is correct if we assume that, when the signal is present, the true values of the intrinsic parameters fall within the cell where \mathcal{F} is maximum. This approximation is accurate for sufficiently large S/N.

V. FAST COMPUTATION OF THE $\mathcal F$ STATISTIC

The detection statistic \mathcal{F} [Eq. (105)] involves integrals of the general form

$$\int_{0}^{T_{0}} x(t)m(t;\omega,\beta,\lambda)\exp[i\phi_{\mathrm{mod}}(t;\omega,\dot{\omega},\beta,\lambda)]\exp[i\omega t]dt$$
(121)

where *m* is a combination of the complex modulation functions defined in Appendix A, while the phase modulation ϕ_{mod} is given by

$$\phi_{\rm mod}(t) = \frac{1}{2} \dot{\omega} t^2 + \omega R \cos \beta \cos(\Omega t + \eta_0 - \lambda) \quad (122)$$

[Eq. (45)]. We see that the integral (121) can be interpreted as a Fourier transform [and computed efficiently with a fast Fourier transform (FFT)], if ϕ_{mod} and *m* do not depend on the frequency ω . In fact, even in that case we can still use FFTs by means of the procedure that we now present.

From the original data we generate several band-passed data sets, choosing the bandwidth of each set so that $m \exp[i\phi_{mod}]$ is approximately constant over the band. We then search for GW signals in each band-passed data set: this is done by computing the \mathcal{F} statistic over a grid in the pa-

rameter space (ω, β, λ) , set finely enough that we do not miss any signal. We follow the grid-construction procedure presented in Sec. III A of Ref. [32]. The phase modulation can be usefully reparametrized as

$$\phi_{\text{mod}} = p_1 t^2 + A \cos(\Omega t) + B \sin(\Omega t), \qquad (123)$$

where

$$p_{1} = \frac{1}{2}\dot{\omega},$$

$$A = \omega R \cos\beta \cos(\lambda - \eta_{0}),$$

$$B = \omega R \cos\beta \sin(\lambda - \eta_{0}).$$
(124)

Since *m* is a slowly changing function of time, we consider it constant for the purpose of constructing a grid over the parameter space. The result is a uniform grid of prisms with hexagonal bases, where the parameter subspace *A*-*B* is tiled by regular hexagons. The grid in the parameters $\dot{\omega}$, β , and λ is then derived by applying the inverse transformation,

$$\omega = 2p_1,$$

$$\beta = \pm \arccos(\sqrt{A^2 + B^2}/\omega R),$$

$$\lambda = \eta_0 + \arctan(B/A),$$
(125)

where for each band-passed data set we set the unknown frequency ω to the maximum frequency of the band. The computation of the \mathcal{F} statistic includes both phase- and amplitude-modulation effects, even if these were neglected in the construction of the grid [in fact, the sign degeneracy for β in Eq. (125) is resolved by amplitude modulation, which distinguishes between sources in opposite directions with respect to the plane of the ecliptic].

Once we have a detection, the accurate estimation of signal parameters requires a second step. Since the coarse signal search described above is performed by evaluating the function $m \exp[i\phi_{mod}]$ at the maximum frequency of each band, our filters are not perfectly matched to the signal, and thus are not optimal; as a consequence, the location of the maximum of \mathcal{F} does not correspond to the correct ML estimators. We therefore refine the coarse search by maximizing \mathcal{F} near the coarse-search maximum, this time without any approximation.

We have performed a few numerical simulations to assess the performance of our optimal-filtering algorithm. Here we report on three of them. In the first simulation we analyzed the X_1 TDI data corresponding to two simultaneous monochromatic signals, of frequency f=3 mHz and S/Ns of 24 and 10, emitted from sources at opposite positions with respect to the plane of the ecliptic. We generated a one-yearlong time series for X_1 by implementing Eq. (58) numerically, and we included noise by adding a Gaussian random process (as realized by a random number generator) with spectral density given by Eq. (65). We narrowbanded the X_1 data to a bandwidth of 0.125 mHz around 3 mHz, and we analyzed the resulting data by implementing the two-step procedure described above, using the Nelder-Mead maximization algorithm [33] for the second step. The angular grid for the all-sky search consisted of about 900 points. We then performed the following operations: (i) detecting the stronger signal and estimating its parameters; (ii) reconstructing the stronger signal and subtracting it from the data; (iii) detecting the weaker signal and estimating its parameters; (iv) subtracting it from the data. Figure 6 shows the amplitude spectrum of X_1 before and after the subtraction of the two signals, as compared with the spectrum of noise alone. Figure 7 shows a comparison of the input signals with the reconstructed signals (built with the parameters specified by the ML estimators). We see that the amplitude modulations in the GW response enable us to determine the sky location of two sources of the same frequency, and also to resolve the two GW polarizations. Signal resolution will degenerate rapidly as more sources of the same frequency are added, so the steps described above cannot be used as a general signal subtraction procedure [34].

In the second simulation we analyzed the X_1 TDI data corresponding to a single signal of frequency f=25 mHz, S/N=9.5, and \dot{f} = 6.5×10⁻¹³ Hz s⁻¹ (corresponding to a binary of chirp mass $\mathcal{M}_c = 0.9 M_{\odot}$). We generated a one-yearlong time series for X_1 by implementing numerically the exact GW response, Eq. (48), and we added noise as described above. We narrowbanded the X_1 data to a bandwidth of 0.5 mHz, and again we analyzed the resulting data with the two-step procedure described above. The sky search was performed on a small grid (~ 300 gridpoints) around the true values of the signal parameters. In the third simulation we analyzed the \overline{A} , \overline{E} , and \overline{T} TDI data corresponding to the same signal, for a total S/N=19. The ML search procedure was performed as in the second simulation. The top panel of Fig. 8 shows the TDI observable $\overline{A}(t)$ for the signal alone, superimposed on the TDI observable for signal plus noise: we see that the signal is more than one order of magnitude weaker than the noise. The bottom panel of Fig. 8 shows the \mathcal{F} statistic (already maximized over f, β , and λ) near the input signal frequency. We see that the statistical significance is higher for the multiple-observable search than for the X_1 search. Figure 9 shows a comparison of the input signals with the reconstructed signals. We see that reconstruction is more accurate for the multiple-observable search, but in both searches our procedure resolves the two GW polarizations successfully.

We conclude that our proposed algorithm performs satisfactorily, detecting the simulated signals, accurately estimating their parameters, and resolving the two GW polarizations, both in the low-frequency regime (first simulation) and high-frequency regime (second and third simulation). In a future paper, we plan to discuss in detail the expected errors in parameter estimation for a source with given frequency, sky position, and S/N.

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FIG. 6. Quality of signal reconstruction, as seen in the Fourier domain, in the first simulation. The original X_1 time series contains noise, plus two monochromatic signals of equal frequency (3 mHz) and ecliptic longitude, but opposite ecliptic latitudes. We show the X_1 amplitude spectrum before and after the subtraction of the reconstructed signals, compared with the amplitude spectrum of noise alone.

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APPENDIX A: COMPLEX REPRESENTATION OF THE RESPONSE

From Eqs. (52) and (53) it is easy to see that the Sagnac TDI observables α_i can be rewritten in the complex form

$$\alpha_{i}^{\text{GW}} = 2x \sin\left(\frac{3}{2}x\right) \text{Re}[a^{(u)} * m_{\alpha_{i}}^{(u)}(t)e^{i\phi(t)} + a^{(v)} * m_{\alpha_{i}}^{(v)}(t)e^{i\phi(t)}], \quad (A1)$$

where $x = \omega L$, the complex amplitudes $a^{(u)}$ and $a^{(v)}$ are defined in Eqs. (98) and (99), the phase $\phi(t)$ by Eq. (45), and the complex *modulation functions* $m_{\alpha_i}^{(u)}$ and $m_{\alpha_i}^{(v)}$ are given by

$$\begin{bmatrix} m_{\alpha_{i}}^{(u)}(t) \\ m_{\alpha_{i}}^{(v)}(t) \end{bmatrix} = \sum_{j=1}^{3} \begin{bmatrix} u_{j}(t) \\ v_{j}(t) \end{bmatrix} e^{-ixd_{j}} \{ \operatorname{sinc}[(1+c_{j})x/2]n_{\alpha_{i}}^{+j} + \operatorname{sinc}[(1-c_{j})x/2]n_{\alpha_{i}}^{-j} \},$$
(A2)

with $u_j(t)$, $v_j(t)$ given by Eqs. (27) and (28), $c_j(t)$ and $d_j(t)$ by Eqs. (47) and (46), and with the constants $n_{\alpha_i}^{\pm j}$ given in the left part of Table II.

The optimal combinations \overline{A}^{GW} , \overline{E}^{GW} , \overline{T}^{GW} are given by formulas similar to Eq. (A1), with modulation functions



$$m_{\bar{A}}^{(u)} = \frac{1}{\sqrt{2}} (m_{\alpha_3}^{(u)} - m_{\alpha_1}^{(u)}), \tag{A3}$$

$$m_{\bar{E}}^{(u)} = \frac{1}{\sqrt{6}} (m_{\alpha_1}^{(u)} - 2m_{\alpha_2}^{(u)} + m_{\alpha_3}^{(u)}), \tag{A4}$$

$$m_{\bar{T}}^{(u)} = \frac{1}{\sqrt{3}} (m_{\alpha_1}^{(u)} + m_{\alpha_2}^{(u)} + m_{\alpha_3}^{(u)}),$$
(A5)

and similar expressions for $m_{\overline{A}}^{(u)}$, $m_{\overline{E}}^{(u)}$, and $m_{\overline{T}}^{(u)}$. The quantities $N^{(u)}$, $N^{(v)}$, U, V, and W [Eqs. (101), (102), (92), (93), (100)] that are needed to compute the ML amplitude estima-

FIG. 7. Quality of signal reconstruction, as seen in the time domain, in the first simulation. The panels show the input signals (the stronger on the left, the weaker on the right), compared with the reconstructed signals; the two GW polarizations are plotted together (top row), and separately (middle and bottom rows).

tors $\hat{a}^{(i)}$ and the \mathcal{F} statistic [Eq. (105)] can be written in terms of the complex modulation functions as

$$N^{(u)} = 4x \sin\left(\frac{3}{2}x\right) \int_0^{T_0} \left[\frac{\overline{A}(t)m_{\overline{A}}^{(u)}(t) + \overline{E}(t)m_{\overline{E}}^{(u)}(t)}{S_{\overline{A}}(\omega)} + \frac{\overline{T}(t)m_{\overline{T}}^{(u)}(t)}{S_{\overline{T}}(\omega)}\right] e^{i\phi(t)}dt, \qquad (A6)$$

$$N^{(v)} = 4x \sin\left(\frac{3}{2}x\right) \int_{0}^{T_{0}} \left[\frac{\bar{A}(t)m_{\bar{A}}^{(v)}(t) + \bar{E}(t)m_{\bar{E}}^{(v)}(t)}{S_{\bar{A}}(\omega)} + \frac{\bar{T}(t)m_{\bar{T}}^{(v)}(t)}{S_{\bar{T}}(\omega)}\right] e^{i\phi(t)}dt,$$
(A7)



FIG. 8. Maximum-likelihood detection in the second and third simulations. In the top panel we plot $\overline{A}(t)$ for the input signal alone, superimposed on the same observable for signal plus noise. In the bottom panel we plot the \mathcal{F} statistic (already maximized over \dot{f} , β , and λ) near the input signal frequency, for a single-observable search using X_1 , and for a multiple-observable search using \overline{A} , \overline{E} , \overline{T} . The frequency of the input signals is correctly estimated in both cases, but the statistical significance of the multipleobservable detection is higher.

and

$$U = 4x^{2} \sin^{2} \left(\frac{3}{2}x\right) (2/T_{0}) \int_{0}^{T_{0}} \left[\frac{|m_{\bar{A}}^{(u)}(t)|^{2} + |m_{\bar{E}}^{(u)}(t)|^{2}}{S_{\bar{A}}(\omega)} + \frac{|m_{\bar{T}}^{(u)}(t)|^{2}}{S_{\bar{T}}(\omega)}\right] dt,$$
(A8)

$$V = 4x^{2} \sin^{2} \left(\frac{3}{2}x\right) (2/T_{0}) \int_{0}^{T_{0}} \left[\frac{|m_{\bar{A}}^{(v)}(t)|^{2} + |m_{\bar{E}}^{(v)}(t)|^{2}}{S_{\bar{A}}(\omega)} + \frac{|m_{\bar{T}}^{(v)}(t)|^{2}}{S_{\bar{T}}(\omega)}\right] dt,$$
(A9)

$$W = 4x^{2} \sin^{2} \left(\frac{3}{2} x \right)$$

$$\times (2/T_{0}) \int_{0}^{T_{0}} \left[\frac{m_{\bar{A}}^{(u)*}(t) m_{\bar{A}}^{(v)}(t) + m_{\bar{E}}^{(u)*}(t) m_{\bar{E}}^{(v)}(t)}{S_{\bar{A}}(\omega)} + \frac{m_{\bar{T}}^{(u)*}(t) m_{\bar{T}}^{(v)}(t)}{S_{\bar{T}}(\omega)} \right] dt.$$
(A10)

The X_i TDI observables can be written in the complex form as

$$X_{i}^{\text{GW}} = 4x \sin(x)\sin(2x)\text{Re}[ia^{(u)}*m_{X_{i}}^{(u)}(t)e^{i\phi(t)} + ia^{(v)}*m_{X_{i}}^{(v)}(t)e^{i\phi(t)}]; \qquad (A11)$$



FIG. 9. Quality of signal reconstruction, as seen in the time domain, in the second and third simulations. The panels show the input signals (X_1 on the left, \overline{A} on the right), compared with the reconstructed signals; the two GW polarizations are plotted together (top row), and separately (middle and bottom rows). Signal reconstruction is more successful for the multiple-observable search (right) than for the singleobservable search (left).

the modulation functions $m_{X_i}^{(u)}(t)$ and $m_{X_i}^{(v)}(t)$ have exactly the same functional form as the functions $m_{\alpha_i}^{(u)}(t)$, $m_{\alpha_i}^{(v)}(t)$ defined in Eq. (A2), except that the coefficients $n_{X_i}^{\pm j}$ are those given in the right part of Table II. For the single X_1 observable, the ML estimators for the amplitudes and for \mathcal{F} are again given by

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Time [s]

 $\hat{a}^{(u)} = 2(T_0 \Delta_{X_1})^{-1} [V_{X_1} N_{X_1}^{(u)} - W_{X_1}^* N_{X_1}^{(v)}], \quad (A12)$

$$\hat{a}^{(v)} = 2(T_0 \Delta_{X_1})^{-1} [U_{X_1} N_{X_1}^{(v)} - W_{X_1} N_{X_1}^{(u)}]$$
(A13)

(where $\Delta_{X_1} = U_{X_1}V_{X_1} - |W_{X_1}|^2$) and

6360 6370

TABLE II. Constants that appear in the complex representation of the GW responses of the TDI observables. The constants $n_{\alpha_2}^{\pm j}$ and $n_{\alpha_3}^{\pm j}$ are obtained from $n_{\alpha_1}^{\pm j}$ by cyclical permutation of the index *j*, as are $n_{X_2}^{\pm j}$ and $n_{X_3}^{\pm j}$ from $n_{X_1}^{\pm j}$.

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Time [s]

| j | $n_{\alpha_1}^{+j}$ | $n_{\alpha_1}^{-j}$ | $n_{\alpha_2}^{+j}$ | $n_{\alpha_2}^{-j}$ | $n_{\alpha_3}^{+j}$ | $n_{\alpha_3}^{-j}$ | $n_{X_1}^{+j}$ | $n_{X_1}^{-j}$ | $n_{X_2}^{+j}$ | $n_{X_2}^{-j}$ | $n_{X_{3}}^{+j}$ | $n_{X_3}^{-j}$ |
|---|---------------------|---------------------|---------------------|---------------------|---------------------|---------------------|----------------|----------------|----------------|----------------|------------------|----------------|
| 1 | e^{-i3x} | $-e^{-i3x}$ | e^{-i2x} | $-e^{-i4x}$ | e^{-i4x} | $-e^{-i2x}$ | 0 | 0 | $e^{-i7x/2}$ | $-e^{-i9x/2}$ | $-e^{-i9x/2}$ | $-e^{-i7x/2}$ |
| 2 | e^{-i2x} | $-e^{-i4x}$ | e^{-i4x} | $-e^{-i2x}$ | e^{-i3x} | $-e^{-i3x}$ | $e^{-i7x/2}$ | $e^{-i9x/2}$ | $-e^{-i9x/2}$ | $-e^{-i7x/2}$ | 0 | 0 |
| 3 | e^{-i4x} | $-e^{-i2x}$ | e^{-i3x} | $-e^{-i3x}$ | e^{-i4x} | $-e^{-i2x}$ | $-e^{-i9x/2}$ | $-e^{-i7x/2}$ | 0 | 0 | $e^{-i7x/2}$ | $e^{-i9x/2}$ |

$$\mathcal{F} = (T_0 \Delta)^{-1} \{ V_{X_1} | N_{X_1}^{(u)} |^2 + U_{X_1} | N_{X_1}^{(v)} |^2 - 2 \operatorname{Re}[W_{X_1} N_{X_1}^{(u)} (N_{X_1}^{(v)})^*] \}, \qquad (A14)$$

with

$$N_{X_{1}}^{(u)} = 8x \sin(x)\sin(2x) \int_{0}^{T_{0}} \left[\frac{X_{1}(t)m_{X_{1}}^{(u)}(t)}{S_{X_{1}}(\omega)} \right] ie^{i\phi(t)} dt,$$
(A15)

$$N_{X_1}^{(v)} = 8x\sin(x)\sin(2x) \int_0^{T_0} \left[\frac{X_1(t)m_{X_1}^{(v)}(t)}{S_{X_1}(\omega)} \right] ie^{i\phi(t)} dt,$$
(A16)

and

$$U_{X_1} = 16x^2 \sin^2(x) \sin^2(2x) (2/T_0) \int_0^{T_0} \left[\frac{|m_{X_1}^{(u)}(t)|^2}{S_{X_1}(\omega)} \right] dt,$$
(A17)

$$V_{X_1} = 16x^2 \sin^2(x) \sin^2(2x) (2/T_0) \int_0^{T_0} \left[\frac{|m_{X_1}^{(v)}(t)|^2}{S_{X_1}(\omega)} \right] dt,$$
(A18)

$$V_{X_1} = 16x^2 \sin^2(x) \sin^2(2x)$$

$$\times (2/T_0) \int_0^{T_0} \left[\frac{m_{X_1}^{(u)*}(t) m_{X_1}^{(v)}(t)}{S_{X_1}(\omega)} \right] dt.$$
(A19)

APPENDIX B: REDUCED INFORMATION MATRIX

It is interesting to examine the relation between the matrix $G^{\mu\nu}$ defined by Eq. (117) and the Fisher information matrix Γ^{ij} . We consider the case of a single TDI observable; multiple observables can be treated in similar fashion. As seen in Sec. II, the generic TDI GW response h(t) can be written as the linear combination

$$h(t) = \sum_{k=1}^{4} a^{(k)} h^{(k)}(t, \xi^{\mu});$$
(B1)

as discussed in Sec. IV A, the amplitudes $a^{(k)}$ are extrinsic parameters, while all the other parameters (denoted together as ξ^{μ}) are intrinsic [all the parameters are denoted together as $\theta^i \equiv (a^{(k)}, \xi^{\mu})$]. Note that in the case of the TDI observables X_1^{GW} or \overline{A}^{GW} [Eqs. (48),(81)], the component functions $h^{(k)}(t)$ would include the factors $4x \sin(x)\sin(2x)$ and $2x \sin(\frac{3}{2}x)$, respectively.

In this notation, it is easy to show that the optimal S/N and the Fisher matrix can be written as

$$\rho^2 = \mathbf{a}^T \cdot \mathbf{M} \cdot \mathbf{a}, \tag{B2}$$

and

$$\Gamma^{ij} = \begin{pmatrix} \mathsf{M} & \mathsf{F} \cdot \mathbf{a} \\ \mathbf{a}^T \cdot \mathsf{F}^T & \mathbf{a}^T \cdot \mathsf{S} \cdot \mathbf{a} \end{pmatrix}, \tag{B3}$$

where the top and left blocks correspond to the extrinsic parameters, while the bottom and right blocks correspond to the intrinsic parameters. The superscript *T* denotes transposition over the extrinsic parameter indices. Furthermore, $\mathbf{a} \equiv (a^{(1)}, a^{(2)}, a^{(3)}, a^{(4)})$, and the matrices M, F, and S are given by

$$M^{(k)(l)} = (h^{(k)}|h^{(l)}), \tag{B4}$$

$$F_{\mu}^{(k)(l)} = \left(h^{(k)} \middle| \frac{\partial h^{(l)}}{\partial \xi^{\mu}} \right), \tag{B5}$$

$$S_{\mu\nu}^{(k)(l)} = \left(\frac{\partial h^{(k)}}{\partial \xi^{\mu}} \middle| \frac{\partial h^{(l)}}{\partial \xi^{\nu}} \right).$$
(B6)

The covariance matrix C^{ij} , which expresses the expected variance of the ML parameter estimators, is defined as $(\Gamma^{-1})^{ij}$. Using the standard formula for the inverse of a block matrix [35] we have

$$\mathbf{C} = \begin{pmatrix} \mathbf{M}^{-1} + \mathbf{M}^{-1} \cdot (\mathbf{F} \cdot \mathbf{a}) \cdot \overline{\Gamma}^{-1} \cdot (\mathbf{F} \cdot \mathbf{a})^T \cdot \mathbf{M}^{-1} & -\mathbf{M}^{-1} \cdot (\mathbf{F} \cdot \mathbf{a}) \cdot \overline{\Gamma}^{-1} \\ -\overline{\Gamma}^{-1} \cdot (\mathbf{F} \cdot \mathbf{a})^T \cdot \mathbf{M}^{-1} & \overline{\Gamma}^{-1} \end{pmatrix},$$
(B7)

where

$$\overline{\Gamma} = \mathbf{a}^T \cdot (\mathbf{S} - \mathbf{F}^T \cdot \mathbf{M}^{-1} \cdot \mathbf{F}) \cdot \mathbf{a}.$$
(B8)

We shall call $\overline{\Gamma}^{\mu\nu}$ (the *Schur complement* of M) the *projected Fisher matrix* (onto the space of intrinsic parameters). Because the projected Fisher matrix is the inverse of the intrinsic-parameter submatrix of the covariance matrix C^{ij} , it expresses the information available about the intrinsic parameters once the extrinsic parameters are set to their ML

estimators. Note that $\overline{\Gamma}^{\mu\nu}$ is still a function of the putative extrinsic parameters. Using Eq. (B2) we define the *normalized projected Fisher matrix*

$$\overline{\Gamma}_{n} \equiv \overline{\Gamma}/\rho^{2} = \frac{\mathbf{a}^{T} \cdot (\mathbf{S} - \mathbf{F}^{T} \cdot \mathbf{M}^{-1} \cdot \mathbf{F}) \cdot \mathbf{a}}{\mathbf{a}^{T} \cdot \mathbf{M} \cdot \mathbf{a}}.$$
 (B9)

From the Rayleigh principle [35], it follows that the minimum value of the component $\overline{\Gamma}_n^{\mu\nu}$ is given by the smallest eigenvalue (taken with respect to the extrinsic parameters) of the matrix $[(S - F^T \cdot M^{-1} \cdot F) \cdot M^{-1}]^{\mu\nu}$. Similarly, the maximum value of the component $\overline{\Gamma}_n^{\mu\nu}$ is given by the largest eigenvalue of that matrix. Because the trace of a matrix is equal to the sum of its eigenvalues, the matrix

$$\widetilde{\Gamma} = \frac{1}{4} \operatorname{Tr}[(\mathbf{S} - \mathbf{F}^T \cdot \mathbf{M}^{-1} \cdot \mathbf{F}) \cdot \mathbf{M}^{-1}], \qquad (B10)$$

where the trace is taken over the extrinsic-parameter indices, expresses the information available about the intrinsic parameters, averaged over the possible values of the extrinsic parameters. Note that the factor $\frac{1}{4}$ is specific to the case of four extrinsic parameters. We shall call $\tilde{\Gamma}^{\mu\nu}$ the *reduced Fisher matrix*. This matrix is a function of the intrinsic parameters alone.

Let us now compute the components of $G^{\mu\nu}$, defined by Eq. (117). We start from log Λ , which in our notation is given by

$$\log \Lambda = \mathbf{a}^T \cdot \mathbf{N} - \frac{1}{2} \mathbf{a}^T \cdot \mathbf{M} \cdot \mathbf{a}, \qquad (B11)$$

where $N^{(k)} = (x|h^{(k)})$, with x(t) = n(t) + h(t), and n(t) a zero-mean Gaussian random process. The ML estimators $\hat{a}^{(k)}$ are given by $\hat{\mathbf{a}} = \mathbf{M}^{-1} \cdot \mathbf{N}$, so for the \mathcal{F} statistic we have $\mathcal{F} = \frac{1}{2} \mathbf{N}^T \cdot \mathbf{M}^{-1} \cdot \mathbf{N}$. Using the relations [22]

$$\mathcal{E}\{(n|s_1)(n|s_2)\} = (s_1|s_2), \tag{B12}$$

$$\mathcal{E}\{(n|s_1)(n|s_2)(n|s_3)(n|s_4)\}$$

= $(s_1|s_2)(s_3|s_4) + (s_1|s_3)(s_2|s_4) + (s_1|s_4)(s_2|s_3),$
(B13)

where s_1 , s_2 , s_3 , and s_4 are deterministic functions, we find that the autocovariance function $C(\xi^{\mu}, \xi'^{\mu})$ of Eq. (114) is given by

$$\mathcal{C}(\xi^{\mu},\xi'^{\mu}) = \frac{1}{2} \operatorname{Tr}[\mathbf{Q}^T \cdot \mathbf{M}^{-1} \cdot \mathbf{Q} \cdot \mathbf{M}'^{-1}], \qquad (B14)$$

where

$$Q_{kl} = (h^{(k)}|h'^{(l)}),$$
 (B15)

and the primes denote functions of the primed parameters ξ'^{μ} . Inserting Eq. (B14) into Eq. (117), after some lengthy algebra (omitted here) we come to the final result

$$G_{\mu\nu} = \frac{1}{4} \sum_{k,l,m,n} \left[S_{\mu\nu}^{(k)(l)} - F_{\mu}^{(m)(k)} (M^{-1})^{(m)(n)} F_{\nu}^{(n)(l)} \right] \times (M^{-1})^{(l)(k)} = \widetilde{\Gamma}_{\mu\nu}.$$
(B16)

Thus, the \mathcal{F} -statistic metric $G^{\mu\nu}$ [31] is found to be exactly equal to the reduced Fisher matrix $\tilde{\Gamma}^{\mu\nu}$; that this should be the case is understandable, since both matrices contain information about the relatedness of waveforms with nearby values of their intrinsic parameters (while both assume that the extrinsic parameters are being set to their ML estimators). For a related argument about the placement of templates for a partially maximized detection statistic, see Ref. [36].

APPENDIX C: FIRST-GENERATION TDI RESPONSES

The GW response of the first-generation TDI observable X is given by [2]

$$X^{\text{GW}} = (y_{31}^{\text{GW}} + y_{13,2}^{\text{GW}}) + (y_{21}^{\text{GW}} + y_{12,3}^{\text{GW}})_{,22} - (y_{21}^{\text{GW}} + y_{12,3}^{\text{GW}}) - (y_{31}^{\text{GW}} + y_{13,2}^{\text{GW}})_{,33};$$
(C1)

after some algebra we get to

$$X^{\rm GW}(t) = 2\,\omega L\,\sin(\omega L) \sum_{k=1}^{4} a^{(k)} X^{(k)}(t), \qquad (C2)$$

where the functions $X^{(k)}(t)$ are given by

$$\begin{bmatrix} X^{(1)} \\ X^{(2)} \end{bmatrix} = \begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \cos[\phi(t) - xd_2 - 3x/2] \\ + \operatorname{sinc}[(1-c_2)x/2] \cos[\phi(t) - xd_2 - 5x/2] \} \\ - \begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \cos[\phi(t) - xd_3 - 5x/2] \\ + \operatorname{sinc}[(1-c_3)x/2] \cos[\phi(t) - xd_3 - 3x/2] \}, \quad (C3)$$

$$\begin{bmatrix} X^{(3)} \\ X^{(4)} \end{bmatrix} = \begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 3x/2] \\ + \operatorname{sinc}[(1-c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 5x/2] \} \\ - \begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 5x/2] \\ + \operatorname{sinc}[(1-c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 3x/2] \}, \quad (C4)$$

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(C5)

 $\alpha^{\text{GW}}(t) = \omega L \sum_{k=1}^{k=4} a^{(k)} \alpha^{(k)}(t),$

where $x = \omega L$ and sinc ... = $(\sin ...)/(...)$. The GW responses for *Y* and *Z* can be obtained by cyclical permutation of the spacecraft indices.

The GW response for α can be written in similar form:

where

$$\begin{bmatrix} \alpha^{(1)} \\ \alpha^{(2)} \end{bmatrix} = \begin{bmatrix} u_1(t) \\ v_1(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x/2] - \operatorname{sinc}[(1-c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x/2] \} \\ + \begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - x/2] - \operatorname{sinc}[(1-c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 5x/2] \} \\ + \begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 5x/2] - \operatorname{sinc}[(1-c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - x/2] \}, \quad (C6)$$

$$\begin{bmatrix} \alpha^{(3)} \\ \alpha^{(4)} \end{bmatrix} = -\begin{bmatrix} u_1(t) \\ v_1(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_1)x/2] \operatorname{cos}[\phi(t) - xd_1 - 3x/2] - \operatorname{sinc}[(1-c_1)x/2] \operatorname{cos}[\phi(t) - xd_1 - 3x/2] \} \\ -\begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{cos}[\phi(t) - xd_2 - x/2] - \operatorname{sinc}[(1-c_2)x/2] \operatorname{cos}[\phi(t) - xd_2 - 5x/2] \} \\ -\begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{cos}[\phi(t) - xd_3 - 5x/2] - \operatorname{sinc}[(1-c_3)x/2] \operatorname{cos}[\phi(t) - xd_3 - x/2] \}.$$
(C7)

The β and γ combinations are again obtained by cyclical permutation of the spacecraft indices.

For the TDI observable ζ we find

$$\zeta^{\rm GW}(t) = \omega L \sum_{k=1}^{4} a^{(k)} \zeta^{(k)}(t), \tag{C8}$$

with

$$\begin{bmatrix} \zeta^{(1)} \\ \zeta^{(2)} \end{bmatrix} = \begin{bmatrix} u_1(t) \\ v_1(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x/2] - \operatorname{sinc}[(1-c_1)x/2] \operatorname{sin}[\phi(t) - xd_1 - 3x/2] \} \\ + \begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 3x/2] - \operatorname{sinc}[(1-c_2)x/2] \operatorname{sin}[\phi(t) - xd_2 - 3x/2] \} \\ + \begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 3x/2] - \operatorname{sinc}[(1-c_3)x/2] \operatorname{sin}[\phi(t) - xd_3 - 3x/2] \},$$
 (C9)

$$\begin{bmatrix} \zeta^{(3)} \\ \zeta^{(4)} \end{bmatrix} = -\begin{bmatrix} u_1(t) \\ v_1(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_1)x/2] \operatorname{cos}[\phi(t) - xd_1 - 3x/2] - \operatorname{sinc}[(1-c_1)x/2] \operatorname{cos}[\phi(t) - xd_1 - 3x/2] \} \\ -\begin{bmatrix} u_2(t) \\ v_2(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_2)x/2] \operatorname{cos}[\phi(t) - xd_2 - 3x/2] - \operatorname{sinc}[(1-c_2)x/2] \operatorname{cos}[\phi(t) - xd_2 - 3x/2] \} \\ -\begin{bmatrix} u_3(t) \\ v_3(t) \end{bmatrix} \{ \operatorname{sinc}[(1+c_3)x/2] \operatorname{cos}[\phi(t) - xd_3 - 3x/2] - \operatorname{sinc}[(1-c_3)x/2] \operatorname{cos}[\phi(t) - xd_3 - 3x/2] \}.$$
(C10)

Finally, the optimal TDI observables A, E, and T [21] are defined as linear combinations of α , β , and γ :

$$A = \frac{1}{\sqrt{2}} (\gamma - \alpha),$$

$$E = \frac{1}{\sqrt{6}} (\alpha - 2\beta + \gamma),$$
 (C11)

$$T = \frac{1}{\sqrt{3}} (\alpha + \beta + \gamma).$$

The long-wavelength approximation to the GW responses is obtained by taking the leading-order terms of the generic expressions in the limit of $\omega L \rightarrow 0$. For instance, for *X* [Eqs. (C2)–(C4)], we get

$$X_{\rm LW}^{\rm GW} \approx 4(\omega L)^2 \{ [u_2(t) - u_3(t)] [a^{(1)} \cos \phi(t) + a^{(3)} \sin \phi(t)] + [v_2(t) - v_3(t)] [a^{(2)} \cos \phi(t) + a^{(4)} \sin \phi(t)] \},$$
(C12)

with $a^{(k)}$ given by Eqs. (41)–(44), and $u_i(t)$, $v_i(t)$ by Eqs. (27), (28). The LW responses for *Y* and *Z* can be obtained by cyclical permutation of the indices. Adopting the notation of Ref. [2], we find also that

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$$X_{\text{LW}}^{\text{GW}}(t) \simeq 2L^2[(\hat{\mathbf{n}}_3^L)' \cdot \ddot{\mathsf{H}}^L(t) \cdot \hat{\mathbf{n}}_3^L - (\hat{\mathbf{n}}_2^L)' \cdot \ddot{\mathsf{H}}^L(t) \cdot \hat{\mathbf{n}}_2^L],$$
(C13)

where the double dot denotes the second time derivative, $\hat{\mathbf{n}}_{i}^{L}$ is given by Eq. (2), and $\mathsf{H}^{L}(t)$ is given by Eq. (11).

The GW response of the Sagnac observable α_{LW}^{GW} is equal simply to $\frac{1}{2}X_{LW}^{GW}$. From Eqs. (C11) we then get the LW GW responses A_{LW}^{GW} , E_{LW}^{GW} , and T_{LW}^{GW} :

$$A_{\rm LW}^{\rm GW} \approx \sqrt{2} (\omega L)^2 \{ [-2u_2(t) + u_1(t) + u_3(t)] \\ \times [a^{(1)} \cos \phi(t) + a^{(3)} \sin \phi(t)] \\ + [-2v_2(t) + v_1(t) + v_3(t)] \\ \times [a^{(2)} \cos \phi(t) + a^{(4)} \sin \phi(t)] \},$$
(C14)

$$E_{\rm LW}^{\rm GW} \simeq \sqrt{6} (\omega L)^2 \{ [u_1(t) - u_3(t)] \\ \times [a^{(1)} \cos \phi(t) + a^{(3)} \sin \phi(t)] \\ + [v_1(t) - v_3(t)] [a^{(2)} \cos \phi(t) \\ + a^{(4)} \sin \phi(t)] \},$$
(C15)

$$T_{\rm LW}^{\rm GW} \simeq O[(\omega L)^3]. \tag{C16}$$

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ables $Z^{(i)}$ as $N^{(i)} = \sqrt{M}^{(i)(j)} Z_{(j)}$ (implicit summation intended); thus, the expression $\mathcal{F} \propto M^{(i)(j)} N_{(i)} N_{(j)}$ collapses to $Z^{(i)} Z_{(i)}$. For n_c (uncorrelated) TDI observables, each $N^{(i)}$ is the sum of n_c uncorrelated Gaussian random variables, leading to the final count of $4 \times n_c$ degrees of freedom for the χ^2 distribution.

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