

A CLOSED-LOOP OTOLITH SYSTEM
ASSESSMENT PROCEDURE

By

Dale W. Hiltner

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Signature of Author

Department of Aeronautics and Astronautics
January 13, 1983

Certified by

Prof. Laurence R. Young Thesis Supervisor

Accepted by

Chairman, Departmental Graduate Committee
Professor Harold Y. Wachman

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ABSTRACT

A test procedure that is sensitive to changes in the response of the human otolith system to linear accelerations has been developed. The test is a closed-loop test in which blindfolded subjects are given a sum of sinusoids velocity disturbance in the lateral direction and directed to null their subjective velocity using a joystick controller. The test procedure has been optimized to provide the best possible data for all test subjects. The testing was performed using the M.I.T. Man-Vehicle Laboratory Sled facility.

Classical control theory quasi-linear describing function analysis is used to analyze the test data. Frequency spectrum plots of the velocity and joystick signals, along with velocity and joystick RMS values, are used to measure the velocity nulling performance of the subject. Bode plots relating acceleration input to joystick velocity command output give the transfer function of the subject.

The Bode plots of four of the subjects tested show very good agreement. The one sigma deviations and data scatter are as low or lower than that of most human subject testing. A regression analysis was used to develop a transfer function model, G_{HO} . The model, with the values obtained from one subject, is

$$G_{HO} = \frac{2.02(j\omega)}{(j\omega + 1.42)(j\omega^2 + 2(0.144)(0.540)j\omega + (0.540)^2)}$$

This test procedure will be used in the pre-and post-flight testing of astronauts. Its purpose is to define how humans adapt to weightlessness. The results will help to more fully understand the causes of space motion sickness.

Thesis Supervisor: Dr. Laurence R. Young
Title: Professor of Aeronautics and Astronautics

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May the wind always be on your tail
and your visibility be unlimited.

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CHAPTER 1

INTRODUCTION

The purpose of this work is to develop a test procedure that is sensitive to changes in the response of the human otolith system to linear accelerations. The test is a closed-loop test in which blindfolded subjects are given a motion disturbance in the lateral direction and directed to null their subjective velocity using a joystick controller. This type of test avoids the magnitude estimation problem of open-loop testing. (Ref. 6) However, it also involves more non-linear effects caused by the human operator which will be elaborated upon throughout this work. The experimental hardware used was the M.I.T. Man-Vehicle Laboratory Sled facility which is described in Chapter 3. The test procedure is to be used in pre- and post-flight testing of astronauts. It is expected that the testing will show changes in the way otolith information is processed by the brain following exposure to a weightless environment. This information will then be used to more fully understand space motion sickness.

From previous work with human subjects, Ref. 5,6,7,8, it is known that the acceleration disturbance must not be predictable, as subjects can then learn the disturbance and respond accordingly. To avoid this a sum of sinusoids velocity disturbance is used. The current system used on the Sled has great flexibility in generating these velocity disturbance profiles. This flexibility involves varying the number of sinusoids, the frequency of each sinusoid, and the peak magnitude of the velocity or acceleration at each frequency. Other variables of the system are the gain of the joystick controller and the pole of the digital filter used to filter this joystick signal. The more specific problem, then, is to find the proper disturbance

profile and joystick response by adjusting these parameters.

From previous work in defining otolith system response, Ref. 4,5,9, it was found that good response of the otolith system is obtained in the 0.05-0.5 Hz frequency range. This was the only range considered throughout the testing. The disturbance frequencies are determined by the prime numbers used to multiply a base frequency. The base frequency is determined by the desired period. This allows no harmonic multiples to interfere with the disturbance frequencies. The amplitudes of the disturbance frequencies can be found using many different techniques. These include defining the disturbance by a flat position, velocity, or acceleration amplitude, with or without scaling by a first, second or third order filter. This flexibility was heavily used in developing the final test procedure.

Very little previous work has been done on otolith testing exclusively. Meiry in 1965 attempted a closed-loop otolith test but quickly abandoned it because subjects could not stay within the physical limits of the track. (Ref. 5) This is because the otolith organs are sensitive to acceleration only, and also have an acceleration threshold of approximately 0.005 g's. Thus, constant velocity motion should be undetectable. These limitations make the closed-loop task very difficult as will be shown. Also, the works on human vestibular testing that the author is familiar with do not attempt to rationalize their disturbance time histories. With no known background in this specific area of otolith system testing the test procedure had to be developed from the fundamentals.

Classical control theory describing function techniques are used in the data analysis as the human operator (HO) response is considered to be

quasi-linear. A block diagram of the system under consideration is shown in Fig. 1.01. The final criteria for determining if a particular test profile was acceptable was to look at the frequency response of various signals obtained from the Sled system. The outputs available are position, velocity, acceleration, commanded velocity, and joystick signal. The most important result is found in the transfer function of the HO which is the Bode plot relating acceleration input to joystick output. Of secondary importance, but valuable in qualitative terms, are frequency response plots of velocity amplitude (with and without HO control) and joystick amplitude. While the transfer function gives the overall response of the HO, the amplitude plots give information on individual control differences and qualitative indications of how well the HO performed the velocity nulling task.

The development of the final test procedure has proceeded using experimental techniques. Based on past experience with the Sled some initial velocity disturbance profiles were generated and tested on several subjects. Based on this experience new profiles were developed and tested. Computer simulations were not used in the development phase as most of the problems discovered in the first tests were non-linear and subjective with no previously known quantitative definition. Also, the basic model for the otolith system is linear and would not have shown the non-linear effects seen. Thus, the final procedure was determined based on actual test data from all previous tests. Its justification has been by statistical and qualitative reasoning, rather than by strict mathematical calculations. It is felt that this gives a fully developed profile, as it is based on actual real world experience.

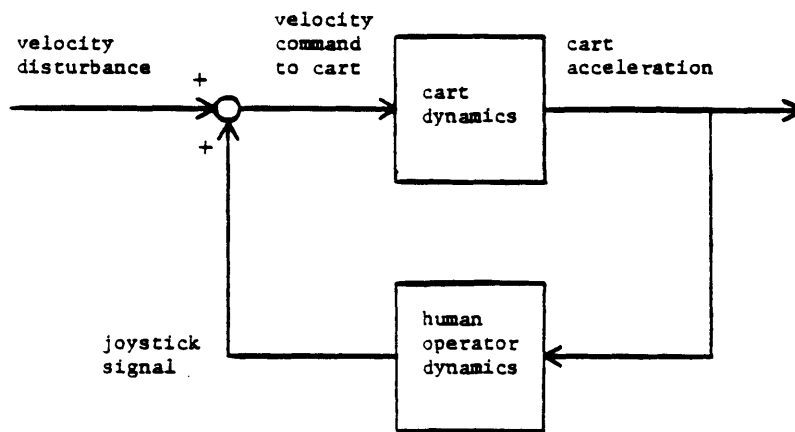


Figure 1.01 The Closed-Loop System

Thesis Organization

Chapter 2 discusses in more detail the space motion sickness problem and shows how the test procedure will be utilized. It also discusses previous work involving the analysis of the human otolith system. Chapter 3 discusses in detail the M.I.T. Sled facility hardware and software and the data reduction techniques used. Chapter 4 is a narrative discussion that reveals the steps taken to achieve the final test procedure. Chapters 5,6, and 7 discuss the final experimental method, the results, and the significant discoveries of this work.

For those interested in only the method and results, it is suggested that Chapters 1,5,6, and 7 be read. Those more interested in the full development process used to obtain the results should read Chapter 2 and 4 also. Those interested in the details of the test facility and the data reduction calculations should also read Chapter 3.

CHAPTER 2

BACKGROUND

2.1 Otolith System Testing for Spacelab

This work is part of the Scientific and Technical Proposal for Vestibular Experiments in Spacelab. (Ref. 1) Its purpose is to define how the human operator changes response to linear accelerations after adapting to weightlessness. This information will then be used to understand more fully the causes of space motion sickness. A brief description of the proposal and the scientific background follows.

The first step to achieve this result is to obtain baseline data in the normal 1 g environment of man. This will be done in the five to six month period before the Space Shuttle flight STS-9. Six test sessions will be held during this period as shown in Table 2.1.01. The tests will be conducted on a quick turnaround basis as the astronauts will be available for only a limited time during each test session. It is also desired for the test results to be obtained in a reasonable time. Baseline data will be obtained for each participating astronaut of the STS-9 mission.

Within eight hours of the astronauts return to earth the first post-flight testing will be done. Subsequent testing will be accomplished over the next two week period as also seen in Table 2.1.01. This testing will show how the HO response has changed due to the intervening weightlessness and will also show a readaptation pattern. In later experiments on the German D-1 Spacelab Mission some sled acceleration tests could be performed in orbit.

FO7 timetable: Baseline Data Collection

F-180	18-20 April 1983	M.I.T. Sled	at	M.I.T.
F-90	28-30 June	U.S. Lab Sled	at	Dryden
F-60	21-22 July	"		"
F-30	31 Aug.-2 September	"		"
F-15	15-16 September	"		"
F-8	21-22 September	"		"
Flight	30 September-8 October			
L+0	8 October	"		"
L+1 to L+7	9-15 October	"		"
L+14	22-23 October	"		"

F = flight
L = landing

TABLE 2.1.01 Spacelab 1 Linear Acceleration Sled Test Timetable

The best theory currently available to define the causes of space motion sickness is the conflict model theory. (Ref.1,2,3) This theory states that upon encountering a weightless environment there is a conflict between visual, tactile, and semi-circular canal sensory perception, and otolith system sensory perception. This conflict is caused by the lack of a 1 g "bias" to the otolith organs. Since the otolith organ output and corresponding brain interpretation is based on millennia of development in a 1 g environment this conflict is easily conceptualized. It is felt that this specific conflict is the cause of space motion sickness.

There are two theories available to explain recovery from space motion sickness based on the conflict model. The primary theory states that since without a constant 1 g "bias" acting on the otolith organ the output is questionable, it is inhibited by the brain. More reliance is then placed on vision to determine orientation. The HO response to linear acceleration is therefore not based on the response of the otolith system and otolith system sensitivity to linear accelerations would be decreased. The secondary theory states that the brain can cancel the 1 g "bias" effects in its processing and concentrate on purely linear acceleration. This would cause an increase in otolith system sensitivity.

These theories must be considered in developing the test procedure to measure changes in the response of the otolith system. The procedure must be able to show an increase or a decrease in otolith system response. The required performance of the HO must not be maximized or minimized so that with varying otolith sensitivities the tests can be completed and precise results obtained.

2.2 The Otolith System Model

Work in defining the otolith system response is found in Ref. 2. This work has resulted in the Young and Meiry model shown in Fig. 2.2.01.

The original data for this model was obtained using a system in which the subject was oscillated at one frequency and indicated the direction of the motion with a joystick. (Ref. 5) The test was therefore an open-loop process in which only phase information was desired. No amplitude information was obtained due to the magnitude estimation problem of open-loop testing. (As stated in Chapter 1, the closed-loop velocity nulling task was attempted but quickly abandoned due to the inability of subjects to stay within the track limits for more than 40 seconds). As expected the Bode plot shows good agreement with the phase data, but the amplitude information is meaningless. It is this amplitude estimation problem that the closed-loop task is expected to resolve.

It is noted that this otolith system transfer function is based on a velocity or acceleration input to the subject and a perceived output indicated by a hand operated joystick. Thus, it is a model for the complete path from the otolith organ output, through the processing of this information by the brain which outputs a signal to the muscles of the hand, and finally to the response of the hand itself. As such, this model can also be used as a basis for the closed-loop task. It is expected that the response of the subject in the closed-loop task will be similar to this complete otolith system model. Possible differences will be discussed in section 2.3.

As is seen from the plots of the otolith system model there is a sharp

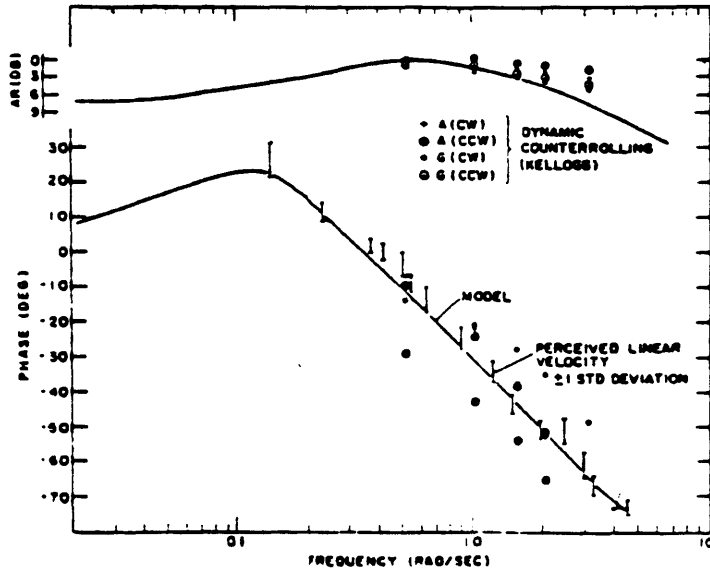
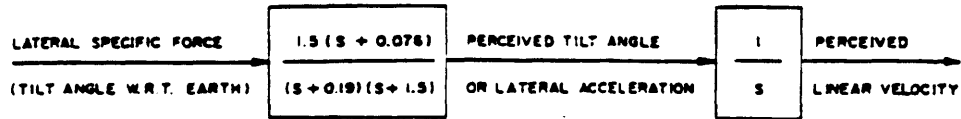


Figure 2.2.01 The Young & Meiry Model (from Ref. 1)

drop-off of the phase at higher frequencies. Assuming the amplitude follows this model it would also show a similar drop-off. This means very little response of the HO to disturbances at the high frequencies. To avoid possible control problems in the closed-loop task a frequency range of 0.05-0.5Hz was chosen. This allows a full octave range and also contains the break frequency of 0.22 Hz (1.5 rad/sec) of the model. Good HO response should be obtained over this frequency range and the break frequency should be indicated to enhance the results.

2.3 The Closed-Loop Task

As stated previously, the main reason for using the closed-loop task is to resolve the magnitude estimation problem. This hopefully will mean more correct magnitude response of the subject as well as correspondingly more correct phase information. However, the closed-loop task contains some additional effects which must be considered.

A block diagram for the closed-loop task is shown in Fig. 2.3.01. With the subject in the loop as shown, the task is not only motion estimation but manual control. As in other manual control tasks different control techniques can be used to achieve the same desired results. This technique, or control strategy, then becomes a part of the HO response. Also, the HO is not a linear system and so does not respond only to the disturbance. The HO will generate some extra response, or remnant, which cannot be linearly correlated with the disturbance. These aspects of the HO control are indicated in the block diagram of Fig. 2.3.02. The $V=0$ summing point indicates the velocity nulling task. The block diagram shows the complete HO system, as considered in this work.

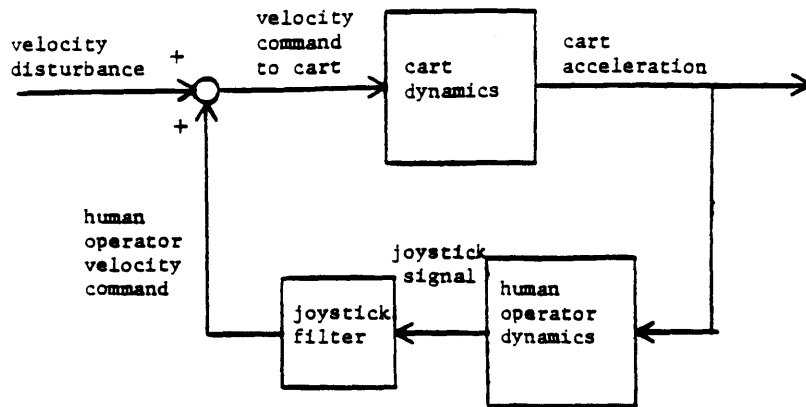


Figure 2.3.01 The Closed-Loop System

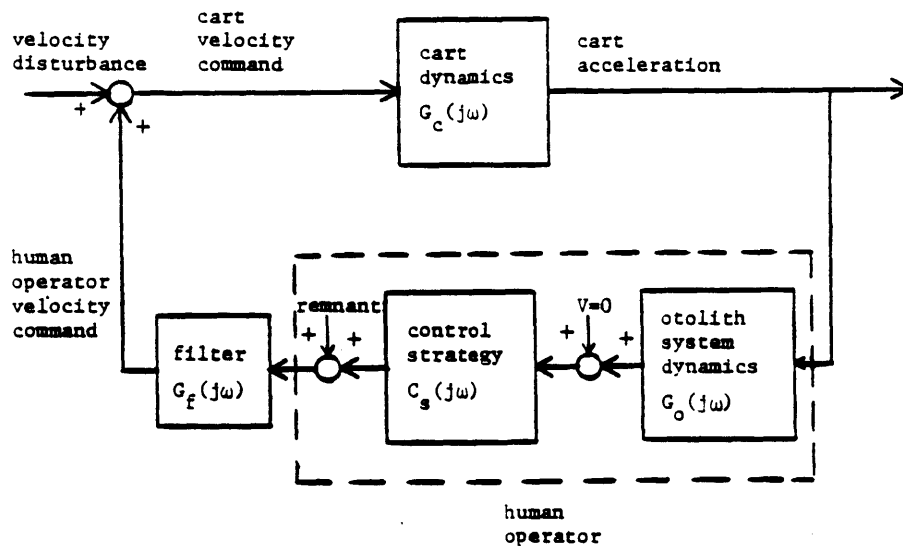


Figure 2.3.02 The Closed-Loop Block Diagram with Details of the Human Operator

The transfer function for the HO is taken across the human operator block shown in Fig. 2.3.02. Thus, the transfer function is not that for the otolith system obtained by open-loop testing. The purpose of this thesis is not to define the control strategy transfer function, but its effects are important and will be elaborated upon throughout this work. The transfer function obtained in this thesis will contain the control strategy effects. This will not effect the desired result, which is to measure HO performance in the closed-loop task, but will effect the analysis and observations of the data.

Of more minor importance from a scientific standpoint but important in a practical sense is the limited track length of the Sled. Because the otolith organs act as accelerometers only, no output will occur for constant velocity motion. (Ref. 5,9) This will cause difficulties for the HO in the closed-loop task. Without an acceleration input deciding on a control input will then be accomplished by guessing. Also, as noted in Ref. 5, subjects often indicate the wrong direction of motion in the open-loop task. For the closed-loop task, then, this could mean initially a wrong control input, as the HO should sense the wrong direction and correct himself. This shows that there is ample opportunity for the HO to input improper control and increase his motion instead of decreasing it. Also, since the HO cannot exactly match the disturbance due to the limitations of the otolith organs, the HO will never stop his motion completely. All this leads to the HO possibly exceeding the limits of the Sled track and ending a run before the disturbance profile is completed. This is of major importance for the data analysis, since a full run is desired for straightforward data reduction, and is one of the major problems to

overcome in developing a satisfactory test procedure.

2.4 Engineering Units

In Ref. 4,5 the otolith system transfer function is shown with velocity or acceleration input and corresponding perceived velocity or acceleration output. Therefore it is possible to construct transfer functions based on the velocity or acceleration disturbance. Since the otolith organs sense only acceleration it seems more correct from a physical viewpoint to use the acceleration input. Therefore, the acceleration input is used in this work.

The disturbance command to the Sled is a velocity command as will be described in Chapter 3. The control of the cart by the HO is added to that of the disturbance command in the feedback loop and is therefore a velocity control. The HO transfer function will have an acceleration input and velocity output. All signals from the Sled are converted to engineering units by the method of Chapter 3; acceleration in m/s^2 and velocity in m/s . In order to use the Young and Meiry model with this input and output it is necessary to add an integrator. This results in the transfer function and Bode plot shown in Fig. 2.4.01. This transfer function was used as a general guideline to verify the form of the Bode plots obtained from all testing.

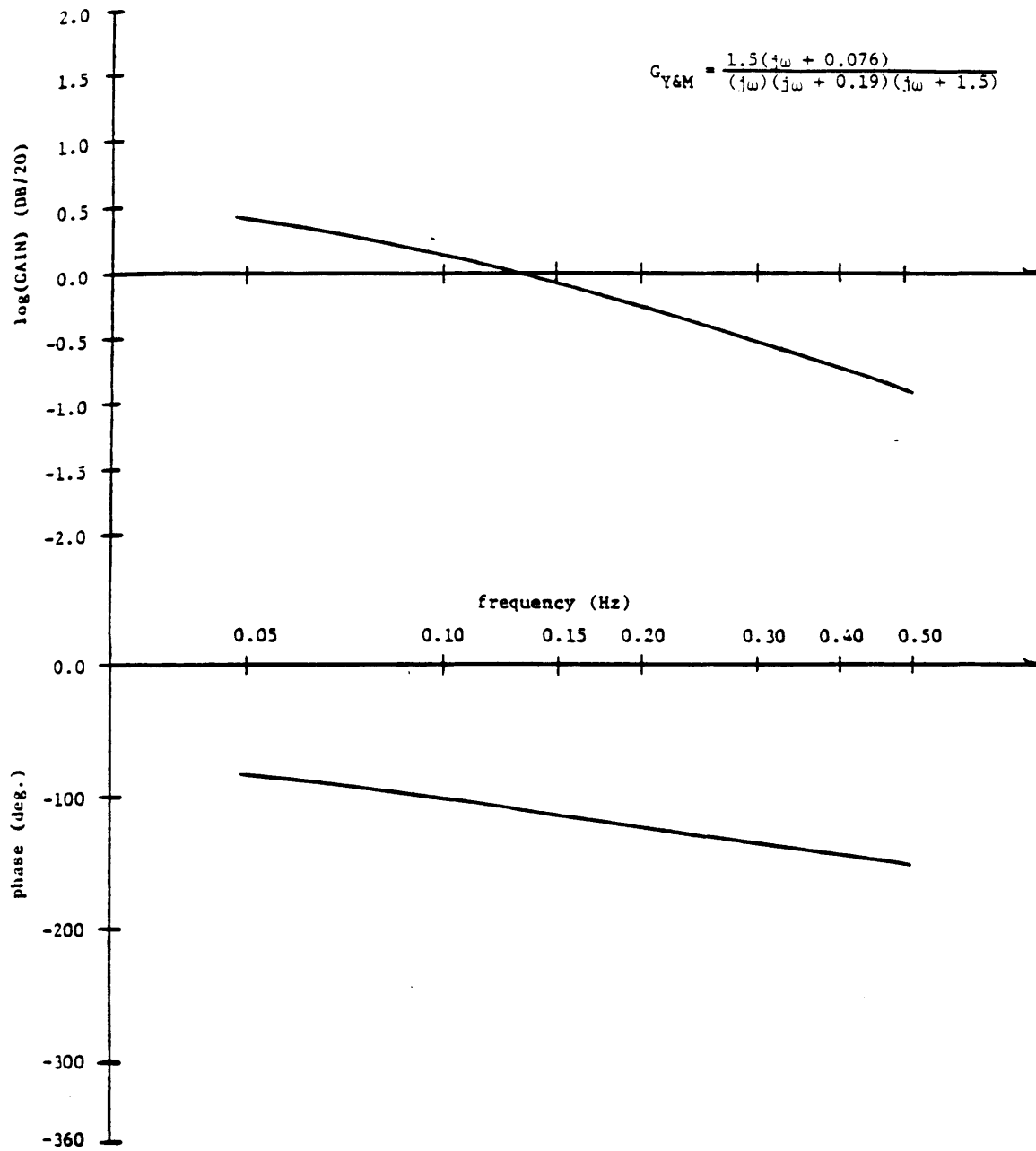


Figure 2.4.01 The Young & Meiry Model
Bode Plot, with Integrator

CHAPTER 3

THE EXPERIMENTAL SYSTEM

3.1 The M.I.T. Sled

The M.I.T. Sled is a rail mounted linear acceleration cart. Four pillow block bushings are mounted to the cart and slide along two circular rails. The cart is aligned for straightness along one rigidly fixed rail while the other rail is held loosely and aligned by the bushings. The total length of travel of the cart is 4.7 m.

A chair is mounted to the cart which can be put at different positions for testing along all three body axes. Lord vibration dampers, which attenuate frequencies below 40 Hz, insulate the chair from the cart frame. The chair is a modified automobile racing seat in which subjects are firmly supported. A lap belt and chest belt are attached to the chair and rigid foam pads are wedged between the shoulders of the subject and the outside chair supports. Two types of head restraints were used in the testing. Both contained foam padding to firmly support the sides and back of the head. One was open-faced, containing no structure in front of the face. This restraint was used in the initial development testing. The other head restraint contained an attachment which is used to take pictures of the subject's eyes in the ocular torsion experiments. This attachment dropped down in front of the subject's face and effectively sealed it from wind generated by the cart motion. Speakers are mounted in both head restraints in which white noise is generated to mask some of the cart motion noise.

A cable attached to both sides of the cart is wound around a pulley at

one end of the rail support structure and a winch drum at the other. The cable is held at 625 lbs. of tension to improve the dynamic response of the cart. The winch drum is driven by a 3.5 horsepower DC permanent magnet torque motor. (Fig. 3.1.01) The motor is controlled by an analog velocity controller. The controller is a PWM (Pulse Width Modulation) controller that uses tachometer feedback. The controller functions as a current generator allowing the velocity of the cart to be proportional to a low current voltage signal applied to the controller. With this controller the maximum acceleration of the cart is 10.0 m/s^2 and the bandwidth is 7 Hz. In addition to the tachometer utilized by the analog controller, a ten-turn position potentiometer is mounted on the motor shaft, and an accelerometer is mounted on the chair near the head of the subject. These transducers give the cart position, velocity, and acceleration signals which are then digitally stored.

Two types of joysticks were used by subjects to control the velocity of the cart. The first joystick consists of a toothed wheel with the axis mounted horizontally and aligned towards the subject. A one turn position potentiometer was mounted to this wheel which gives an output of ± 0.54 volts with full rotation of the wheel. (The ± 15 volt system power supply is used to power the joystick.) This joystick was used in the initial testing only. The joystick used for most of the testing is a standard two-axis joystick similar to the type found on radio control transmitters. The centering spring was removed from the axis used for control allowing no joystick position cue to influence the subject. The output of this joystick with full stick deflection was ± 0.17 volts. This voltage is important as it is used to generate the controller gain. Both joysticks were mounted on

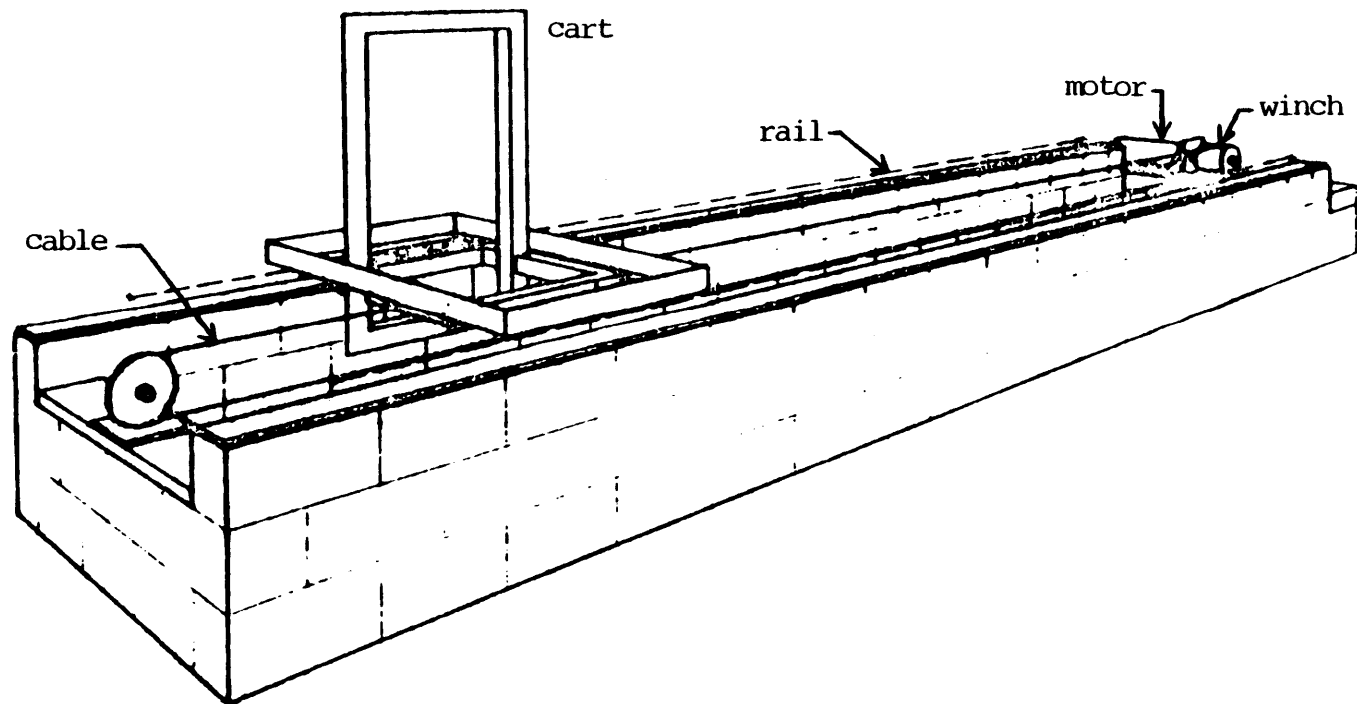


Fig. 3.1.01 The M.I.T. Sled Components

boards which were placed between the cart supports in front of the subject. This allowed the joysticks to be firmly attached to the cart frame. A support for the arm or hand was also mounted to the boards in a convenient position. The joystick output voltage was also recorded. (Appendix A contains pictures of the Sled hardware.)

The hardware safety features on the Sled are numerous. Limit switches are mounted on the Sled support structures near the rail ends. These switches are activated by a probe on the cart frame which stops the system. Shock cords are mounted near the rail ends which contain the cart to the available track when the limit switch is activated. Subjects are given a "panic button" thumb switch which also stops the system and can be activated at any time during a run. The test conductor also has access to two switches which can stop the system.

The Sled system is controlled by a remotely stationed Digital PDP 11/34 minicomputer and a Digital Laboratory Peripheral System (LPS). A fortran program is used to calculate the velocity commands to the cart, which is discussed in section 3.2. These digital commands are stored in a data file and accessed by the test conductor to run the cart. A digital-to-analog converter is used to generate the analog voltage velocity command to the cart controller. If the joystick is used its output is scaled and added to the stored velocity command to determine the final cart velocity command. Analog-to-digital converters are used to convert the analog output signals before they are recorded.

The sled system is controlled by a Sled control panel mounted in the same room as the sled. This panel interacts with the minicomputer. This allows

the test conductor to run any stored velocity command file, set the joystick and data storage to be enabled or disabled, check the digital value of any signal output, and do other operations. The system can also be stopped at any time from this panel. This gives the test conductor full control of the system during the tests.

3.1.1 Calibrations

The D/A converters used in the sled system have 12 bits and a range of ± 10.0 volts. The A/D converters have 12 bits and a range of ± 1.0 volts which gives a gain of 2048 counts/volt. Voltage dividers of 0.1 volts/volt are used to scale the output signals before they are converted by the A/D. This value and the calibrations of the individual transducers have resulted in the following calibrations used to convert the stored digital values to engineering units:

Acceleration	0.01 m/s ² /count
Velocity	0.002722 m/s/count
Position	0.001895 m/count
Commanded Velocity	0.003998 m/s/count
Joystick Velocity Command	0.003998/JSCALE m/s/count

The position calibration was found directly by a system calibration of the position potentiometer. The acceleration calibration was found using the accelerometer calibration. The velocity calibration was found by measuring the tachometer output and the motor RPM. Knowing the drum diameter, in m, the theoretical cart velocity, in m/s, can then be found by

$$\text{velocity} = (\text{RPM}) (\pi) (\text{diameter}) (1/60)$$

to give the required calibration data. The command calibration was found by

injecting a known voltage signal into the controller and measuring the tachometer output. Using the velocity calibration the velocity was found and then the command calibration data could be found. The velocity commanded by the joystick follows this same path as it is also a commanded velocity. The count values of the joystick signal are stored before they are filtered and scaled and added to the stored velocity command. The calibration is therefore the same except for the software scale factor, JSCALE, which is explained in section 3.2. (As noted in section 3.2 the break frequency of the digital joystick filter is 10.54 rad/sec. This is sufficiently far from the maximum disturbance frequency of 3.14 rad/sec so that the filter is not a factor in the calibration.) The A/D and D/A calibrations were used as required to find the final calibrations in engineering units/count.

In order to determine the proper JSCALE value it was decided to scale the maximum commanded velocity to some percentage of the maximum commanded joystick velocity, as described in section 4.5. Using the previously defined calibrations, the following equation was used to find the correct JSCALE:

$$JSCALE = (\text{Volt}_{j\max}) (P \%) (2048) (0.003998) / (V_{\max})$$

where $\text{Volt}_{j\max}$ is the maximum output of the joystick: and V_{\max} is the maximum commanded velocity of the profile in m/s. This results in the maximum commanded velocity being equal to the desired percentage, P, of the maximum joystick commanded velocity.

3.1.2 Cart Transfer Function

The cart system dynamics have been described in Ref. 12. The model developed in this reference was found using bond graph techniques and an assumed cart mass. In order to verify the model, data was taken for a few runs without HO control. The final test profile was used. One run with no subject and one run with a 140 lb. subject were considered. The standard data reduction techniques described in section 3.3 were used with the velocity command as the input and the cart acceleration as the output. A Bode plot of the results is shown in Fig. 3.1.2.01.

This plot shows that the cart transfer function can be approximated by a simple differentiator with a gain of 1.12. Although there is some scatter in the data at the low frequencies it is felt that the more simplified model for the system is more useful for any further work. This plot and model were used as required in all further work. It is also seen that the additional mass of the subject had little effect on the results. This gives assurances that the analog controller is performing satisfactorily with the varying subject mass. It is noted that this model differs from that of Ref. 12.

3.2 Sled System Software

All functions of the Sled are controlled by a single program called CART. Individual functions are accessed from the CART program by two letter codes. The hierarchy of the CART program is explained in Ref. 13,14 and will only be described as necessary here. It is noted that the software has been designed to be "user friendly" and has great flexibility in its current capability and potential for future growth. All program parameters, which are used extensively in the software descriptions, are denoted by

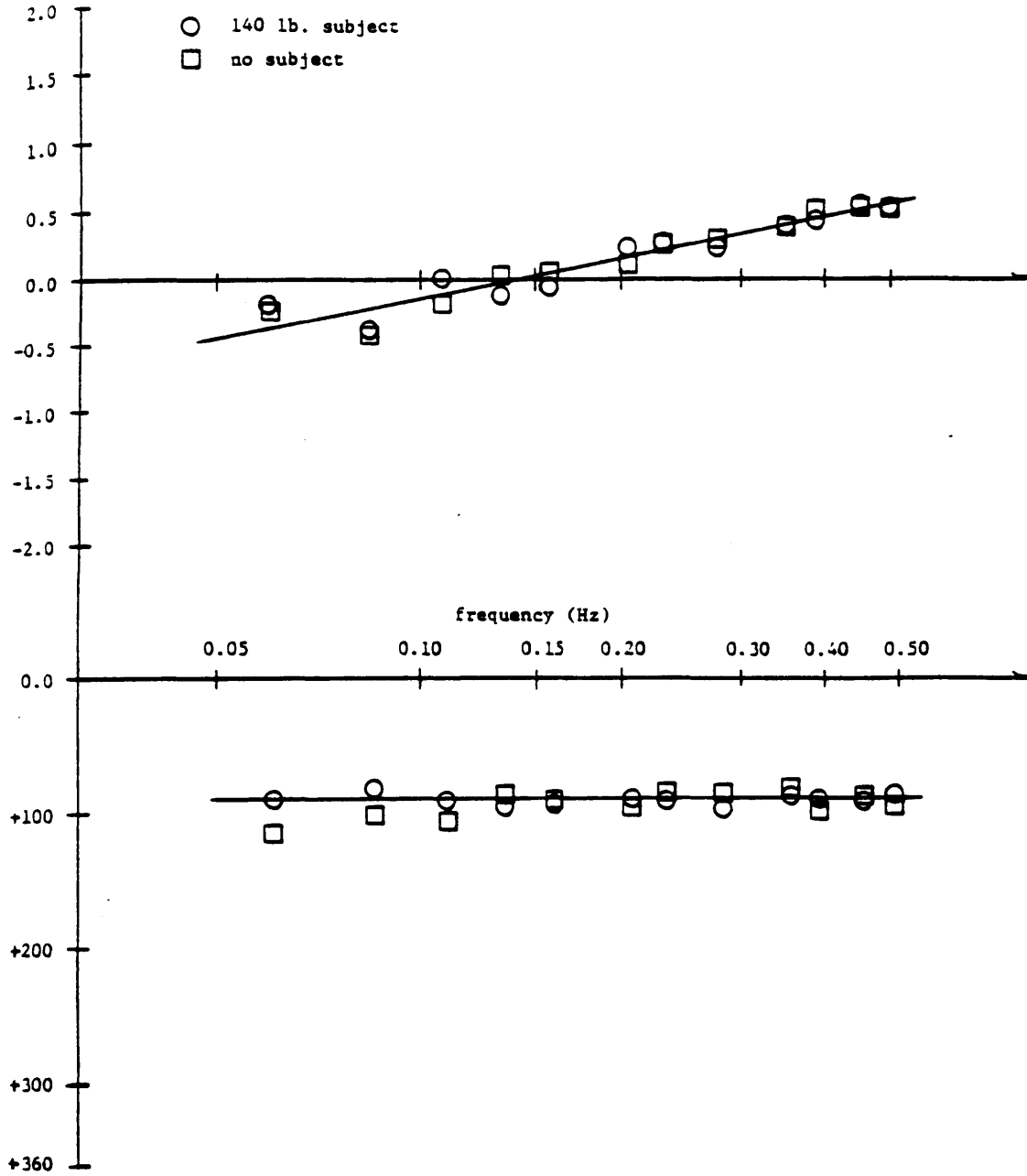


Figure 3.1.2.01 The Cart Transfer Function Bode Plot

capital letters.

safety features have also been incorporated in the cart system software. Limit checks are made on the commanded velocity to prevent an overvoltage to the controller. The cart position and velocity are checked at every sample to determine if the cart could reach the track limits. If so the software decelerates and stops the cart. The deceleration is limited, however, so often the hardware switches are reached before the cart is stopped. These are the principle software safety features.

3.2.1 Disturbance Profile Generation

In order to drive the cart with a sum of sinusoids velocity signal two files have to be created. The first file contains the discrete velocity commands as determined by the sum of sines. These files are generally called velocity command profiles, or profiles, and their generation is described in the next section. Each profile is defined by a different set of parameters. Groups of these profiles are then assigned to files called protocol files. Each protocol file is made up of a series of profiles. A profile is run by accessing it from the protocol file using the Sled control panel described in section 3.1. Ref. 13,14 further describe the file system.

All profiles used to run the cart in this work are sum of sinusoids velocity commands. These profiles are defined by

$$v(t) = \sum A_i \sin(\omega_i nT + \phi_i)$$

where $v(t)$ is the velocity time history in m/s: A_i is the peak amplitude at the i^{th} disturbance frequency, ω_i , in rad/sec : T is the sampling rate

in seconds/sample; n is the consecutive sample number: and t is time = nT .

The program used to generate a sum of sinusoids profile is accessed by the SO command. Ten parameters are needed to generate a sum of sines profile. The profile run time, in seconds, is input as variable TRUN. It is used to determine the fundamental or base frequency in rad/sec by

$$\omega_b = 2\pi/TRUN$$

The fundamental frequency is also input in Hz. This is used by the test conductor for illustrative purposes. The number of sinusoids used in the profile is input as variable NSINES. The disturbance frequencies used in the profile are determined by the h_i numbers stored in a data file. Prime numbers and one even number, if desired, can be used without having the harmonics of the frequencies affect each other. The disturbance frequencies are determined by

$$\omega_i = h_i\omega_b$$

The FLAT input parameter sets the peak amplitudes at the disturbance frequencies of either position, velocity, or acceleration constant. For a constant velocity profile the velocity amplitudes, A_i , are set to

$$A_i = 1.0$$

For a constant position profile the velocity amplitudes are set to

$$A_i = \omega_i$$

For a constant acceleration profile the velocity amplitudes are set to

$$A_i = 1.0/\omega_i$$

The FILTER and FPOLE input parameters can also be used to further scale the

amplitudes before the limit checks are made. This will be discussed in section 3.2.2.

The frequency variation of each sinusoid is adjusted by the DEL input phase angle. This is done to give more flexibility and allow each sinusoid to have a different starting point. The phase angle for each sine is found by

$$\phi_i = i \cdot \text{DEL} \quad i=0,1,\dots,\text{NSINES}-1$$

DEL is chosen so that no phase angle is duplicated.

With all these parameters chosen the sines are completely defined. The amplitudes can now be further adjusted by the input track length limit, FPOS, in m, and the input acceleration limit, FACC, in g's. The track length limit is checked first. The sum of sines velocity is integrated to give the position. With the input sampling time, T, the maximum and minimum position of the run are found using

$$\text{position}(t) = \sum (A_i / \omega_i) \sin(\omega_i nT + \phi_i - \pi)$$

If the maximum position excursion exceeds the FPOS limit then the amplitudes are scaled by

$$A_i = A_i (\text{FPOS} / (\text{pos}_{\max} - \text{pos}_{\min}))$$

Using these newly defined amplitudes the maximum absolute acceleration, in g's, is found by

$$\text{acceleration}(t) = \sum A_i \omega_i \sin(\omega_i nT + \phi_i + \pi) / 9.81$$

If this acceleration exceeds the FACC limit the amplitudes are further scaled by

$$A_i = A_i (FACC / |acc_{max}|)$$

At this point the profile is completely defined. The velocity is then checked to find the first zero crossing, at a time t_0 . The phase angles are then adjusted so the profile will start at this point. This insures that the first velocity commanded by the profile is small. The starting position is then calculated by finding the position at t_0 and then finding

$$\text{starting position} = \text{pos}(t_0) - (\text{pos}_{max} - \text{pos}_{min})/2.0$$

This centers the profile within the cart travel limits.

This completes the profile generation phase. Two more steps are then used to store the profile in a data file, and assign this data file to a protocol file. When these steps are complete the profile can then be used to run the cart.

The profile generation program has been programmed on two different computers. Appendix B contains the program listings and a brief explanation of their use. Only the VAX output calculates the maximum commanded velocity and the histogram values. The histogram data is found by calculating each nT acceleration command value using the equation previously defined. The values of these points are then filed into ranges of multiples of 0.005 g and counted. The maximum value of each 0.005 g range and the number of points in each range is then determined.

3.2.2 Profile Amplitude Scaling

As stated in the previous section the amplitudes of the disturbance frequencies can be scaled by using the FILTER and FPOLE variables. These

parameters define the order and pole location of a low pass filter. The method used to define this scaling will now be developed.

The velocity command is written as

$$v(t) = \sum A_i \sin(\omega_i nT + \theta_i)$$

When filtering is used the amplitudes, A_i , are adjusted so that the power spectral density of the velocity is scaled according to

$$\int_{\omega_{imin}}^{\omega_{imax}} \phi_{vv}(\omega) d\omega = (K / (FPOLE + j\omega)^{FILTER})^2$$

The amplitudes at each frequency are then chosen to be

$$1/2A_i^2 = A_i^2 \int_{\omega_{imin}}^{\omega_{imax}} \phi_{vv}(\omega) d\omega = A_i^2 (g(\omega_{imax}) - g(\omega_{imin}))$$

where $g(\omega)$ is the indefinite integral of $\int \phi_{vv}(\omega) d\omega$ and the ω_{imax} and ω_{imin} are chosen to be the geometric means between the disturbance frequencies. For interior points between disturbance frequencies these frequencies are found by

$$\omega_{imax} = (\omega_i \omega_{i+1})^{0.5}, \text{ and}$$

$$\omega_{imin} = (\omega_i \omega_{i-1})^{0.5}$$

The lowest ω_{imin} frequency, ω_0 , is found by assuming that the lowest disturbance frequency, ω_1 , is the geometric mean of the lowest ω_0 and the next lowest disturbance frequency, ω_2 . Thusly, ω_1 can be found by

$$\omega_1 = (\omega_0 \omega_2)^{0.5}$$

Solving for ω_0 then gives

$$\omega_0 = \omega_1^2 / \omega_2$$

Similarly the highest ω_{imax} frequency, $\omega_{NSINES+1}$, is found by

$$\omega_{NSINES+1} = \omega_{NSINES}^2 / \omega_{NSINES-1}$$

The K value is chosen by specifying the variance of the velocity amplitudes, from 0.0, to be 1.0. This gives:

$$\sigma_d^2 = 1.0 = \sum 1/2A_i^2 = \int_{\omega_{1min}}^{\omega_{1max}} \phi_{vv}(\omega) d\omega + \dots + \int_{\omega_{NSINESmin}}^{\omega_{NSINESmax}} \phi_{vv}(\omega) d\omega = K \int_{\omega_{1min}}^{\omega_{NSINESmax}} \phi_{vv}(\omega) d\omega$$

so

$$K = 1.0 / \int_{\omega_{1min}}^{\omega_{NSINESmax}} \phi_{vv}(\omega) d\omega = 1.0 / [g(\omega_{NSINES+1}) - g(\omega_0)]$$

The final equation for the filtered amplitudes can now be written as

$$A_i = A_i \left[\frac{g(\omega_{imax}) - g(\omega_{imin})}{g(\omega_{NSINES+1}) - g(\omega_0)} \right]$$

The indefinite integrals are readily calculated and will not be elaborated upon here.

3.2.3 The Digital Joystick Filter

As stated in section 3.1 the joystick signal is filtered before it is added to the stored velocity command. A digital first order low pass filter is used. The software implementation of this digital filter is

$$Y(n) = (1-\alpha)Y(n-1) + (\alpha/JSCALE)U(n)$$

where $Y(n)$ is the filtered output: $U(n)$ is the filter input, or raw joystick signal; and n is the sample number. $JSCALE$ and α can be varied by using the JO command in the CART program. In analog form this filter is represented by

$$Y(s) = (K/(1.0/\tau + s))U(s)$$

where τ is the time constant. Comparing the two forms gives

$$K = 1.0/JSCALE, \quad \tau = T/\ln(1.0-\alpha)$$

As is seen the JSCALE variable is used to vary the gain of the filter and the α variable is used to vary the pole of the filter. As is shown in chapter 4, JSCALE is an important parameter in determining the success of a profile. α was set at $\alpha=0.1$ and never varied throughout the profile development.

With $\alpha=0.1$ the equivalent time constant is $\tau=0.095$ sec. From the stripcharts of the cart velocity during full deflection tests of the joystick it was seen that the cart response had no visible delay and no overshoot. To increase the time constant would lower this response time which would be easily noticed by the HO. Also, the human sensory system operates with a 0.20 sec time constant which gives a sufficient safety margin compared to $\tau=0.095$ sec. It would not be desirable to increase τ as this would decrease the safety margin and possibly cause resonance effects similar to pilot induced oscillations. There is no reason to decrease τ , as the response of the cart is quite acceptable. For these reasons α was not varied.

3.3 Data Reduction

A data file is created for every run during which the data storage flag is enabled. The A/D's used to convert the output signals have 8 channels. The 5 outputs available for this work are found in channels 1,2,3,4, and 6. The data points are grouped into blocks of 256 points which gives 32 samples of each channel per block. The PDP 11/34 minicomputer is used to process the data directly from the stored data files.

To reduce this data each channel is accessed individually and stored in a file. This is done by storing every 3th + desired channel # point of the original data file. The data is then concatenated to produce 1024 points to be used to run the Fast Fourier Transform (FFT) algorithm.

The total number of data points of each output of each run is found by

$$\text{run time/sampling rate} = \text{TRUN}/0.01$$

The number of points for each concatenation is then found by

$$N = \text{TRUN}/(0.01)/(1024)$$

From the ensemble of N points an average and standard deviation are determined. Any of the N points which are more than two standard deviations from the average are discarded and a new average is determined. The percentage of the discarded points is printed out as REJECT. This new average is then stored in an array of 1024 points. When all 1024 points are found the average, AVG, square root of the mean squared error, RMS, and the standard deviation, STD, are calculated.

With the concatenation of all channels complete, an FFT is used to find the frequency distribution. A simple fortran FFT program obtained from Ref. 15 was coded into the PDP-11/34 minicomputer. This allows the data to be processed directly from the stored data files obtained during the test runs. Two programs are run sequentially to obtain the final results.

The first amplitude and phase obtained from the FFT are the bias values. The subsequent values are associated with a frequency, f, defined by

$$f(I) = (I-1/1024)*(1.0/\text{TRUN})$$

where I is the array position. The I-1 factor is needed since the first array values are the bias values as stated. The run time, TRUN, is specified to be a multiple of 1024 times the sampling rate. This insures that the disturbance frequencies can be exactly reproduced by the FFT. The remnant frequencies and values are found by averaging the amplitudes, phases, and frequencies of all the points between the disturbance frequencies. Although this is not precisely correct, the real and imaginary parts should be averaged and then the amplitude and phase determined, Ref. 8 shows that there is a negligible difference between the two methods of computation. The log(GAIN) and phases of the transfer function are then determined by

$$\log(\text{GAIN}(N)) = \log \left[\frac{\text{AMPJ}(N+1)}{\text{AMPA}(N+1)} \right], \quad \text{phase}(N) = \text{PHASEJ}(N+1) - \text{PHASEA}(N+1) + 180$$

where AMPJ is the joystick amplitude: AMPA is the acceleration amplitude: PHASEJ is the joystick phase angle: PHASEA is the acceleration phase angle. The 180 deg. correction is added since the subject opposes the cart motion. All count values are converted to engineering units with the calibrations of section 3.1 before entering the FFT program. The desired plots are then created with this data. An explanation of the plots is contained in Appendix E.

It is noted that the FFT does not correct for the run time used. This means that the output is not scaled in a meaningful way. This results in the high amplitude values seen in the frequency spectrum plots. To keep this in mind when looking at these plots, the designation FFT has been placed in with the engineering units notation. This only affects the frequency plot

data as the scaling factors are canceled when the amplitude ratios are taken for the GAIN and phase data.

For runs that are not completed; (i.e. the subject did not stay within the track limits for the full run time) no FFT information was obtained. Since it was desired to work with only completed runs for the procedure little effort was expended on analyzing incomplete runs. A program was written to calculate the RMS, AVG, and STD values of all data points for a run, however. This was used for the initial testing since most of these runs were incomplete. The RMS, AVG, and STD values were computed from only the concatenated 1024 points for all further runs. There is a few percentage points of error between the two methods of computation but it is not of significance for this work.

Listings of all the programs used to reduce data in this work are provided in Appendix C. Brief descriptions of their use, along with input and output samples, are included.

CHAPTER 4

TEST PROCEDURE DEVELOPMENT

4.1 General Concepts

The background of the pre- and post-flight closed-loop otolith system testing has been developed in Chapters 1 and 2. One key factor of this test is that it is to be used on all participating astronauts. Therefore, there could be some variation of otolith sensitivity among the subjects. Also, as stated in Chapter 1, it is expected that adapting to weightlessness will cause a decrease in otolith sensitivity. This increases the range of otolith sensitivity at the less sensitive end. Any test procedure must then have two major goals:

- 1) yielding an accurate description of the HO response,
- 2) yielding this description for a wide range of otolith sensitivities.

As mentioned in Chapter 2 the test period will be of limited duration and data analysis needs to be performed without delay. Because of this it is felt that the test procedure should offer good chances for completing runs with little practice. This also means that some margin for error in control will be available, which should be helpful for subjects with varying otolith sensitivities. Further, it means that the test is not so difficult that results might be in question due to short runs. Finally it gives confidence in the procedure itself.

The data analysis can also be performed in a more straightforward manner with a standard FFT routine when runs are completed. This eliminates the

problems of FFT analysis with incomplete runs and gives more consistent results. The run completion rate is therefore of major practical importance in developing a test procedure.

Initially a computer simulation of the closed-loop system was desired to help determine the general ranges of the system parameters. With the first testing, however, this idea was abandoned. The effects seen were very non-linear and so would not have been evident in a linear simulation. The main thrust of the development was then based on experimental results. Lessons learned from one set of tests were applied to determine the next profiles. This was continued until a profile was found that fit the previous criteria. A description of this development now follows.

4.2 The DHPR02.PRO Series, Part 1, and High Amplitude Problems

As stated previously the authors initial experience with the closed-loop nulling task was as a subject in the tests of Ref. 8. The parameters used to generate this profile are shown in Fig. 4.2.01. This initial experience suggested that a smaller track length be used to help subjects remain within the track limits. Also, it was felt that a run time of 184.32 sec. was too long as in the author's experience fatigue became a factor after about 120 sec.

Using this experience seven profiles were created. The flat velocity calculation was used to scale the amplitudes as this lessened the number of variables required to generate the profiles. The track length was lowered to a range of 1.97-2.38 m while the corresponding maximum accelerations ranged from 0.120-0.204 g. Various numbers and distributions of frequencies

```

=====
SUM OF SINES PROFILE

```

```

1. DURATION OF PROFILE:      184.32 SEC
PARAMETERS OF SINUSOIDS:
2. NUMBER OF SINUSOIDS:      25
3. FUNDAMENTAL FREQUENCY:    0.0054 HZ
4. EQUAL AMPLITUDE DOMAIN:   0 (-1,F;0,U;+1,A)
5. SUCCESSIVE PHASE ANGLE:   247.  DEG
PARAMETERS OF SHAPING FUNCTION:
6. ORDER OF FILTER:          2
7. POLE:                      0.28 HZ
PHYSICAL CONSTRAINTS:
8. LENGTH OF TRACK           3.60 M
9. ALLOWED ACCELERATION      0.41 G
10. TIME INCREMENT:          0.015 SEC

```

```

RESULTING IN THE SUM OF SINUSOIDS:

```

FREQ [HZ]	AMP [M/S]	ACCEL AMP [G]	PHASE [DEG]
0.016	0.08	0.001	0.
0.027	0.09	0.002	247.
0.038	0.11	0.003	134.
0.060	0.11	0.004	21.
0.071	0.10	0.005	268.
0.092	0.10	0.006	156.
0.103	0.10	0.006	43.
0.125	0.12	0.009	290.
0.157	0.10	0.010	177.
0.168	0.09	0.010	64.
0.201	0.10	0.012	311.
0.222	0.07	0.010	198.
0.233	0.06	0.010	85.
0.255	0.08	0.013	332.
0.288	0.08	0.015	220.
0.331	0.08	0.018	107.
0.396	0.07	0.018	355.
0.450	0.06	0.019	242.
0.548	0.05	0.019	130.
0.613	0.04	0.018	17.
0.743	0.04	0.017	265.
0.808	0.02	0.013	153.
0.884	0.02	0.013	40.
0.982	0.02	0.013	288.
1.080	0.02	0.012	176.

```

MAXIMUM ACCELERATION IN SIGNAL:  0.140 G
PERCENT USAGE OF TRACK: 100.00%
STARTING POSITION: 0.00
=====

```

Figure 4.2.01 Profile Parameters of Ref. 8

INPUT PARAMETERS

TRUN- 102.400
 NSINES= 11
 FFREQ- 0.009766
 FLAT= 0
 DEL- 217.000
 FILTER= 0
 FFOLE- 0.310
 FPOS- 3.50
 FACC- 0.300
 TLOOP= 0.010

ROFILE DESCRIPTION

MAX ACCEL- 0.004
 VELMAX- 1.1480
 % USAGE OF TRACK= 57.50
 STARTING POSITION=-0.363
 SCALE= 0.1881

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0488	0.13	0.004	0.
0.1074	0.13	0.009	247.
0.1660	0.13	0.014	134.
0.1855	0.13	0.015	21.
0.2246	0.13	0.019	263.
0.2832	0.13	0.024	155.
0.3027	0.13	0.025	43.
0.3613	0.13	0.030	290.
0.4004	0.13	0.033	177.
0.4090	0.13	0.038	64.
0.5176	0.13	0.043	311.

Figure 4.2.02 DHPRO2.PRO Profile #6 Parameters

were used to test the effects of these variables. The same DEL=247 was used to eliminate the effects of varying this value. One of the profiles is shown in Fig. 4.2.02. Five subjects were tested. The open-faced head restraint was used and subjects were asked to close their eyes during the runs. A blindfold was not used.

The results of these tests were most illuminating qualitatively. Very few runs were completed. The main effects seen were as follows:

- 1) Subjects often lost track of their direction of motion and it was common to make the initial control in the wrong direction. Often this ended a run. This was probably caused by the vibration of the cart giving a velocity cue without an acceleration cue during approximate constant velocity motion.
- 2) There was no tendency to end all runs at the same end of the track. It is noted in Ref. 5 that for angular motion this is not the case.
- 3) Control was sensitive and HO induced oscillations were not uncommon. Often high frequency oscillations were injected by the subject to decrease their response time and to attempt to find the zero input range of the joystick. Also, the high input accelerations at the higher frequencies probably contributed to the HO induced oscillations.

Bode plots were found for each completed run. A typical plot is shown in Fig. 4.2.03. (Appendix E contains a discussion of the plot formats.) The plot shows the general trends expected by the Young and Meiry model. The drop in gain with higher frequencies is clear, but the phase remains relatively flat. There is much data scatter, especially in phase, which signifies a large number of direction reversals. This corresponds with the large amount of HO induced oscillations observed during test runs. This

activity generally means a high joystick remnant and little velocity nulling as will be shown. With this high remnant the Bode plots should tend toward the inverse plant dynamics as suggested in Chapter 6. The nearness of some of these gain plots to a slope of -1 ($\log(\text{GAIN})/\log(\text{freq.}, \text{rad/sec})$) and the flatness of the phase with frequency suggests this tendency. However, the Young and Meiry model is also close to the inverse plant dynamics so it is not clear which transfer function the data is following. No attempt was made to further investigate this problem at this time.

Since very few runs were completed more effort was placed in analyzing the time history data. It was desired to have a time domain measurement of HO performance and RMS data was used for this. The RMS errors of command (or error, since the stored command value is that of the disturbance plus joystick signals), acceleration, velocity, position, and joystick signals were calculated. These were then divided by the corresponding value of the no subject case, except for those of the joystick. The resulting ratios for one subject are shown in Table 4.2.01.

It is difficult to correlate the different ratios for each run. The position and acceleration ratios show the most scatter and indicated no general trends. The error and velocity ratios seem to correlate with each other and with completed runs. There is not as much correlation with run time as expected. This indicates that an HO reaching the track limits is not necessarily caused by poor control during the entire run, but by a few crucial mistakes. The overall poor control is suggested by the ratios being near 1.0, the neutral velocity nulling value. Since little correlation of the RMS errors was seen, consistency among the individual runs of each subject was checked as repeatability is a requirement for the procedure. It

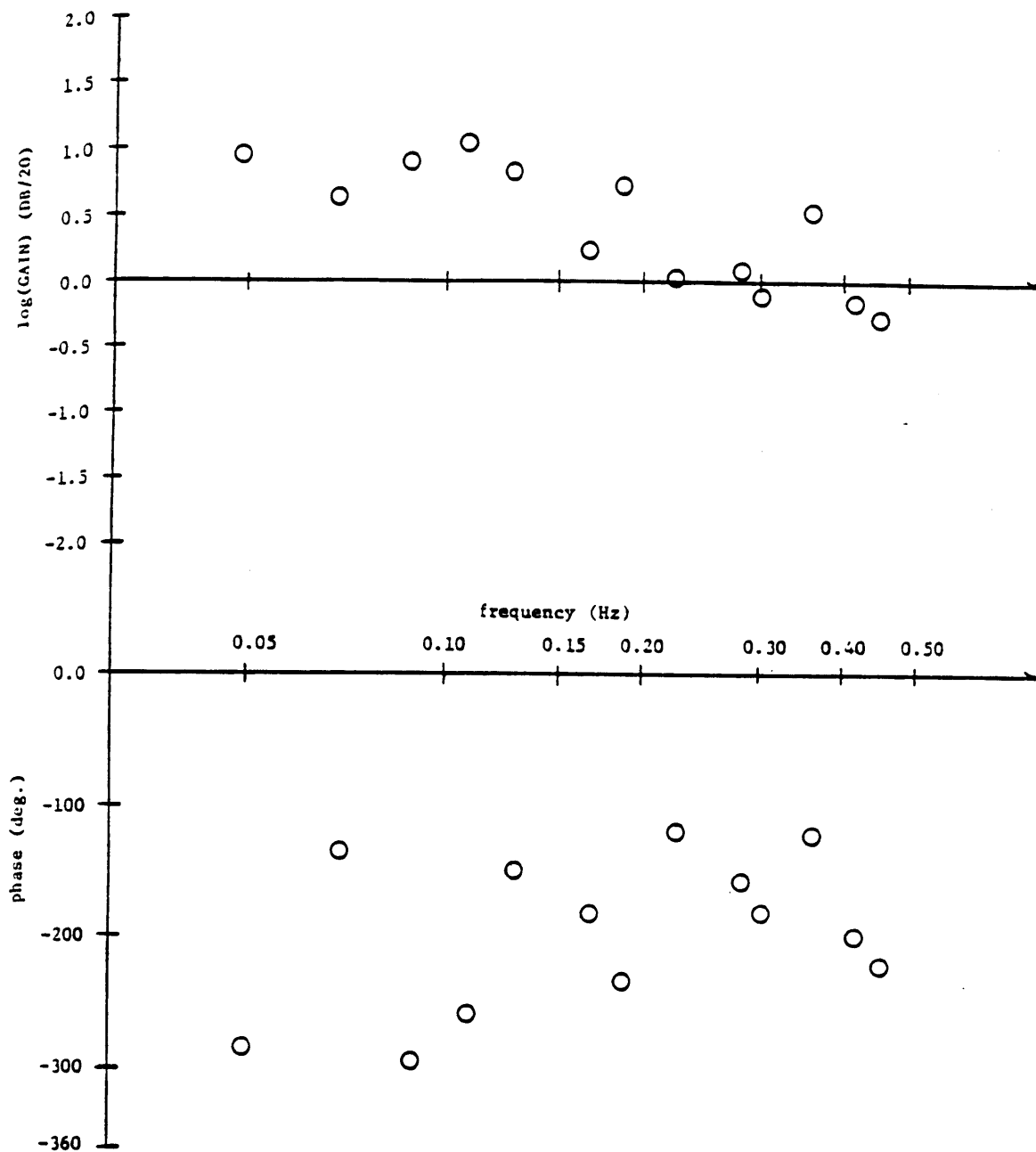


Figure 4.2.03 Bode Plot. Subject JL
DHPRO2.PRO Profile #7

DATA FILE	PR FILE	COMMAND RMS RATIO	POSITION RMS RATIO	VELOCITY RMS RATIO	ACC. RMS RATIO	DURATION (sec.)
005	1	0.911	1.43	1.03	3.19	30.72
006	1	0.70	1.39	0.89	1.78	15.36
007	1	0.77	0.73	0.82	2.08	10.24
010	1	0.98	2.52	1.09	2.60	15.36
008	2	1.12	1.05	0.97	2.25	33.28
009	2	1.10	0.80	1.00	2.60	74.29
011	3	1.29	1.24	1.15	2.67	23.04
012	3	1.22	0.48	0.85	2.88	15.36
013	4	1.24	0.66	1.38	2.70	38.40
014	4	1.52	0.98	1.19	3.30	15.36
015	5	1.30	1.31	0.78	2.50	10.24
016	5	1.26	0.65	1.33	2.33	2.56
017	5	1.86	1.05	1.08	3.47	10.24
018	5	1.52	1.14	1.04	2.33	23.04
019	6	1.45	1.99	0.98	2.84	33.28
021	6	1.32	1.61	0.93	2.51	43.52
022	7	1.41	1.22	0.88	2.61	79.36
023	7	1.44	1.13	0.79	2.57	38.40
024	7	1.34	1.09	0.80	2.22	81.92

TABLE 4.2.01 RMS Ratios. Subject DH, DHPRO2.PRO.

was found that runs 2,6,5 ,and 7 were the most consistent, based on the values of all the ratios. The general lack of consistency and correlation with subject performance will be shown to improve with the lower amplitude profiles.

4.2.1 Summary

Although few runs were completed with this testing some useful information was obtained. During all but one of the test sessions the runs were given to the subject in numerical order. Often the first run was given five or more times until the subject gained some familiarity with the system. Since profiles #6 and #7 were the most consistent and were usually given last it is felt that practice was the main factor for the consistency of these runs. Profiles #2 and #5 were probably consistent due to their low maximum acceleration and corresponding velocity which were the lowest of all the profiles. One of the main reasons for the generally poor run completion rate was disturbance amplitudes that were too large. Another major factor was the gain of the joystick. The JSCALE=2.0 gave a maximum velocity of 1.53 m/s. This allowed total control of the largest maximum velocity of any profile but was much higher than needed for the lower velocity profiles. This high sensitivity caused increased difficulty with the already difficult profiles and contributed to the poor performance.

This initial testing experience also pointed out the many non-otolith cues of the system. The cart motion could always be sensed through the cart noise and vibration which did not hamper the subjects performance but were perceptible. The white noise could not mask all of the cart noise. Wind over the hands and face could also be sensed, especially at the higher

velocities. Subjects were required to wear long sleeves to eliminate this cue as much as possible. Accelerations could also be sensed by tactile cues which were clear since the subjects were firmly strapped to the chair. This points out some of the problems of this type of testing. It is not possible to eliminate all non-otolith cues from the subject. Nevertheless, it is felt that the otolith system provides the predominant motion sensing response.

A list of qualified results follows:

- 1) Subjects needed to be proficient with the joystick and have knowledge of the disturbance before successful runs could be expected. This means more practice is needed.
- 2) Profiles needed to be of lower amplitude and use less track length if subjects were expected to stay within the track limits.
- 3) The joystick gain was too sensitive to precisely control the disturbance.
- 4) The maximum number of disturbance frequencies within the frequency range desired should be used as this gives the largest number of data points in the frequency spectrum.
- 5) The ability of subjects to complete runs was less than expected. This suggested that in future testing more information about the profiles would be needed to determine a successful profile.

4.3 The DHPR02.PRO Series, Part II, and Low Amplitude Problems

With the knowledge gained from the first profiles a second set of profiles was created. Shorter run times were used to reduce possibilities for subject fatigue, which some subjects had noticed. The maximum possible

number of frequencies were used for each profile, which varied with run time. The track length was lowered to a 1.01-1.26 m range, maximum acceleration to 0.102 g, and maximum velocity to a 0.456-0.534 m/s range. A JSSCALE of 3.3 was used which gave a maximum joystick commanded velocity of 0.928 m/s. A typical profile is shown in Fig. 4.3.01.

Four subjects were tested. Again the open-faced head restraint was used and no blindfolds were used but subjects were asked to close their eyes during the runs. Masking noise was set to the highest level the subject could bear. All subjects were given null profile and practice runs with one of these profiles before data was taken. The order the profiles were given in was varied.

The run completion rates were greatly increased. This was mainly due to the lower amplitudes and range of motion of the profiles. It was also felt that the lower amplitude of the disturbance caused the subject to concentrate more on his internal otolith cues rather than his tactile and external cues as a higher level of concentration was required to detect the disturbance. Control was generally better than the first profile set but it was noticed that subjects would often drift along the track with a high frequency motion while trying to find the zero position of the wheel joystick. This "walking" motion was done in an attempt to search out the low disturbance and is the first sign of the control injection control strategy described in section 4.5. The RMS velocity ratios seemed to indicate better control since most values were below 1.0. (Table 4.3.01)

Bode plots of the finished runs were similar to those of the first profile set. Fig. 4.3.02 shows a typical plot. The plot shows the expected

INPUT PARAMETERS

TRUN 81.920
 NSIRFS= 12
 FFREQ= 0.012210
 FLAI= 0
 DELT= 247.000
 FILTER= 0
 FFOLE= 0.310
 FPOS= 2.00
 FACC= 0.150
 TLOOP= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.102
 VELMAX= 0.4900
 % USAGE OF TRACK= 63.19
 STARTING POSITION=-0.163
 SCALE= 0.1124

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0610	0.08	0.003	0.
0.0854	0.08	0.004	247.
0.1099	0.08	0.006	135.
0.1343	0.08	0.007	22.
0.1587	0.08	0.008	269.
0.2075	0.08	0.011	156.
0.2319	0.08	0.012	43.
0.2808	0.08	0.014	290.
0.3540	0.08	0.018	178.
0.3784	0.08	0.019	65.
0.4517	0.08	0.023	312.
0.5005	0.08	0.026	199.

Figure 4.3.01 DHPRO2.PRO Profile #12 Parameters

trends and the data scatter is not significantly less than previously seen. This was unexpected, as it showed that completed runs and low velocity RMS ratios did not always give consistent data. This is a problem of low amplitude profiles.

A possible cause of part of the data scatter was felt to lie in the wheel joystick. Without any zero reference subjects tended to put in high frequency control to help sense their motion. This extra motion caused more scatter in the data and a higher remnant. The lack of positive control caused by the wide range of rotation of the wheel was also a factor.

Because of these problems another joystick was developed for use on the sled. This joystick is shown in Appendix A and is the type found in model aircraft radio control transmitters. The zeroing spring was removed so this extra control cue would not influence the HO. Subjects generally have better knowledge of the input position when using this joystick but not exact knowledge. It is felt that this gives a more effective control without adding additional motion cues. The range of joystick deflection is 40 degrees. The controller for the U.S. Laboratory Sled, which will be used for most of the pre- and post-flight testing, is also a joystick with a 41 degree range. This further supports use of this joystick instead of the wheel. It is felt that this is the best type of joystick to use in this testing.

The joystick voltage output was less than that of the wheel joystick so it was necessary to change JSCALE. The new value found was based on JSCALE=3.3 for the wheel. Comparing voltage ranges gave JSCALE=1.5. Full deflection of the joystick then resulted in the same velocity as full

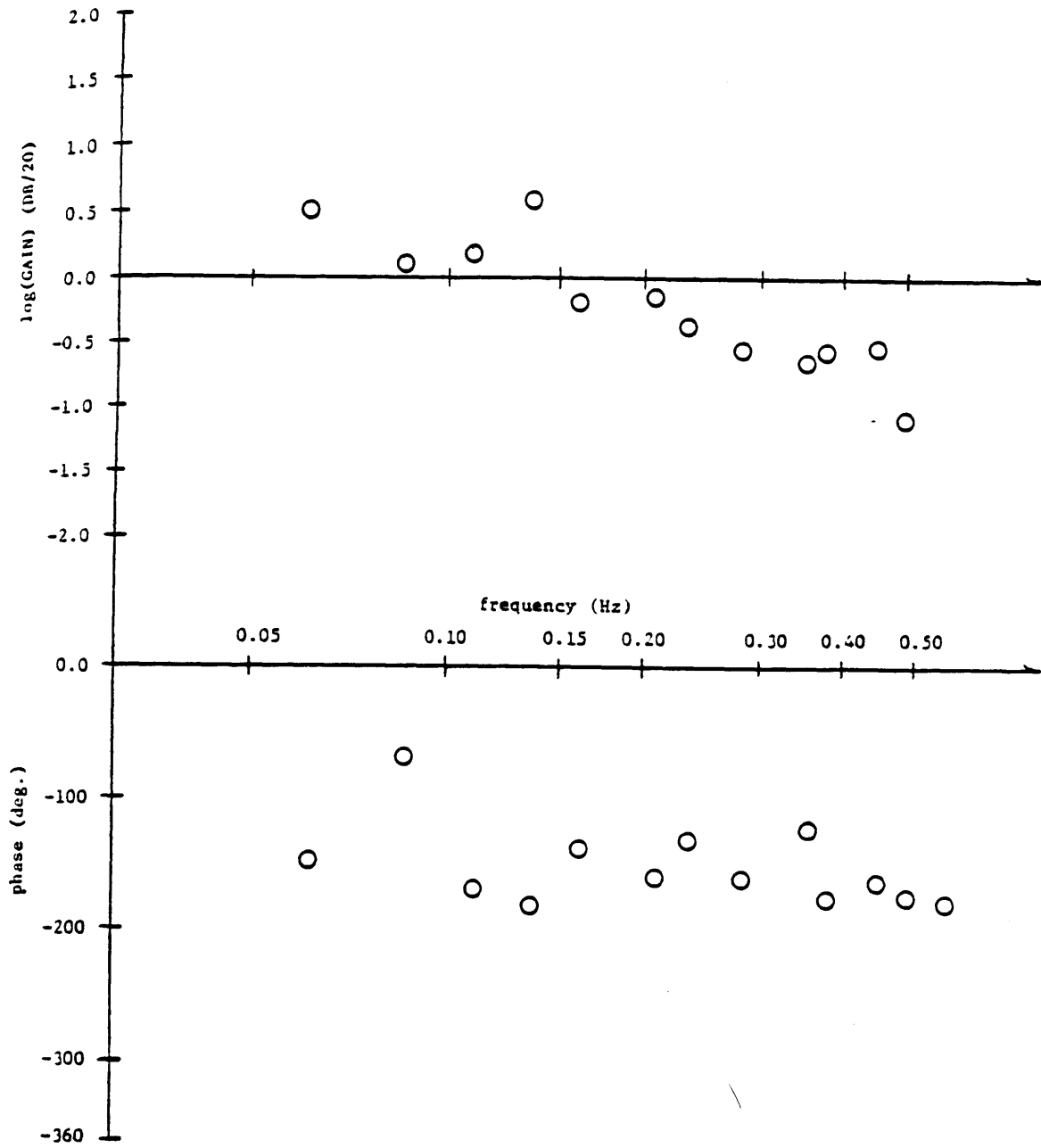


Figure 4.3.02 Bode Plot. Subject DH
DHPRO2.PRO Profile #12

DATA FILE	PR FILE	COMMAND RMS RATIO	POSITION RMS RATIO	VELOCITY RMS RATIO	ACCELERATION RMS RATIO	DURATION (sec.)
03	11	1.08	0.886	0.795	0.863	74.24
04	11	1.40	1.29	0.780	0.779	20.48
06	11	1.07	1.77	0.675	0.759	43.52
16	11	1.05	1.53	0.605	0.659	48.64
07	12	1.08	1.36	0.682	0.890	*81.92
08	12	1.06	1.07	0.723	0.994	*81.92
09	13	1.02	1.15	0.700	0.872	*71.68
10	13	1.73	0.747	0.546	1.063	7.68
11	13	1.02	1.35	0.711	0.900	*71.68
12	14	1.08	1.45	0.790	0.890	51.20
13	14	1.16	0.844	0.811	0.992	*61.44
14	14	1.29	1.65	0.768	0.914	23.04
15	14	1.05	2.35	0.682	0.805	*61.44

* Completed runs.

TABLE 4.3.01 RMS Ratios. Subject DH, DHPRO2.PRO.

deflection of the wheel.

At this time a change in philosophy for further testing occurred. profile #12 had some success in run completion and was a good compromise between run time and number of disturbance frequencies. Because of this it was decided to concentrate on this profile in further testing. More practice would also be given using some of the other profiles. The author would be the principle subject and large population studies would not be done until a successful profile had been found.

The results of testing with these new ideas was indifferent. The run completion rate was similar to that when using the wheel joystick. Individual Bode plots showed similar scatter to previous plots. However, since all the completed runs were of profile #12 they could be averaged and variances and one sigma deviations determined. The resulting Bode plot clearly showed the form of the Young and Meiry model and the average points showed relatively little scatter. The deviations (plotted) were comparable to those seen in the results of the testing of Ref. 3,5,6,7,8.

The subject felt that the joystick gave better control, but that the control was too sensitive. It was easy to put in so much control that the disturbance was masked. This caused more directional activity and more scatter in the phase data. It was then decided to adjust the gain based on the maximum velocity of this specific profile. With this velocity known, the equation of section 3.1.1 was used with $P=90\%$ to find JSCALE. This would give the HO the least sensitive control possible while still giving full control over any part of the disturbance with some margin for calibration and system shifts in joystick voltage output. It was hoped that

this would result in better control and much less data scatter.

Tests on the author were encouraging. After 5 practice runs the next 3 runs were completed. Control was comfortable as the input by the subject did not mask the disturbance and it was easier to input velocity commands near zero.

Data analysis was also encouraging, however, the RMS velocity ratios were greater than 1.0 which caused some concern. (Table 4.3.02) To see if the HO was actually performing the nulling task, frequency spectrum plots of velocity and joystick were made for the three completed runs. A typical plot is shown in Fig. 4.3.04. This type of plot is discussed in Appendix E. The joystick plot shows clearly that the joystick remnant was low indicating that the HO was responding to the disturbance with little control injection. This suggests that the Bode plot data will be precise. The velocity plot shows that there is some velocity nulling, although it is somewhat erratic. This shows that the RMS velocity ratio is not an accurate indicator of the velocity nulling performance. It is rather an indication of the overall velocity activity level, as it includes the remnant effects. The individual Bode plots show less scatter than for most of the previous data. The average Bode plot of the three runs shows phase variances lower than those of the previous test. Fig. 4.3.03 shows this plot. The gain variances are similar but the average points are less scattered as may be expected from the low joystick remnants. Also, there is a flatness in the gain at the low frequencies and some low phase at low frequencies which is suggested in the Young and Meiry model. These results are a major indicator of the better data obtained with the lower joystick gain derived with the P=90% criteria.

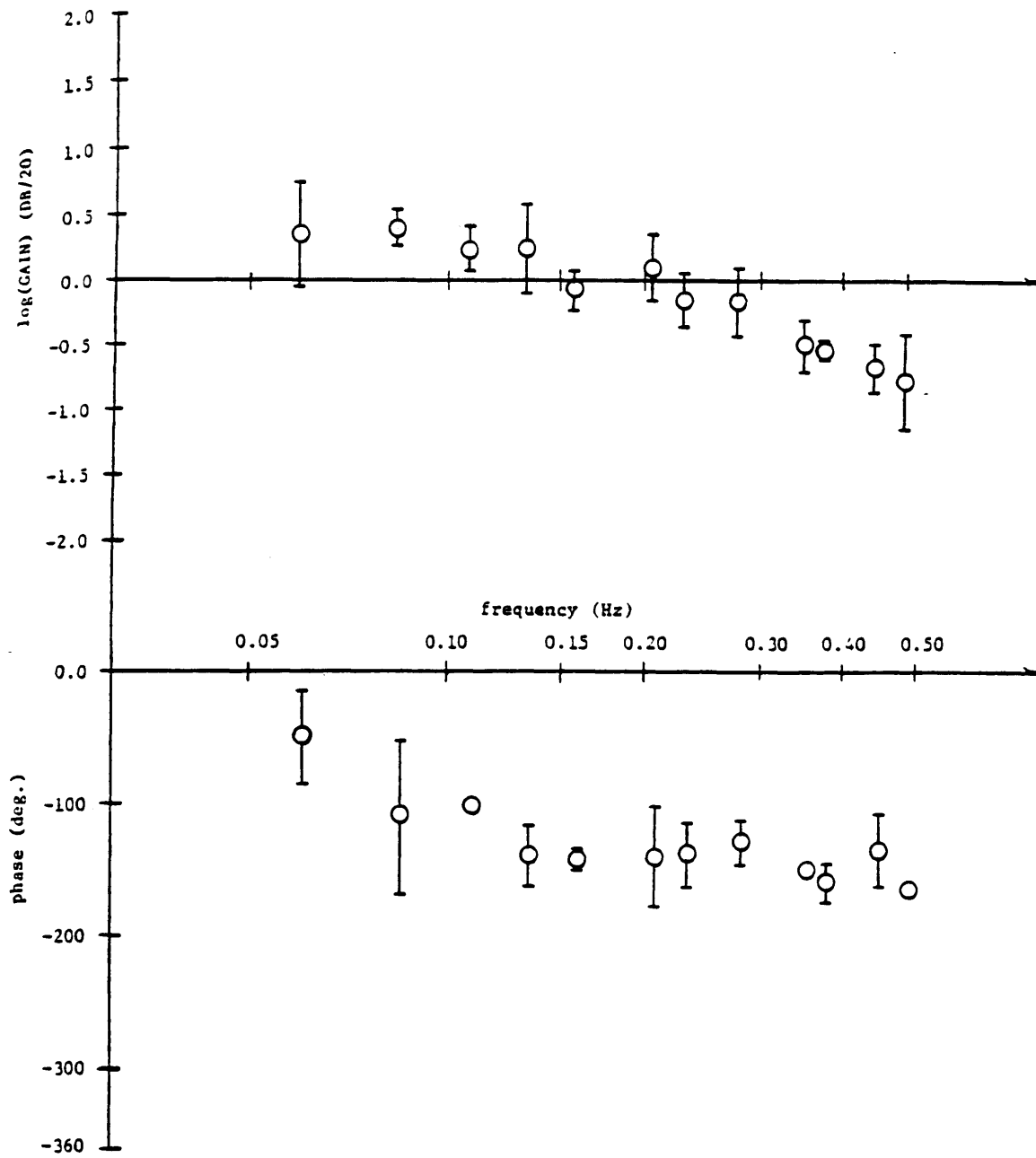


Figure 4.3.03 Bode Plot. Subject DH,
DHPR02.PRO Profile #12, JSCALE=2.56

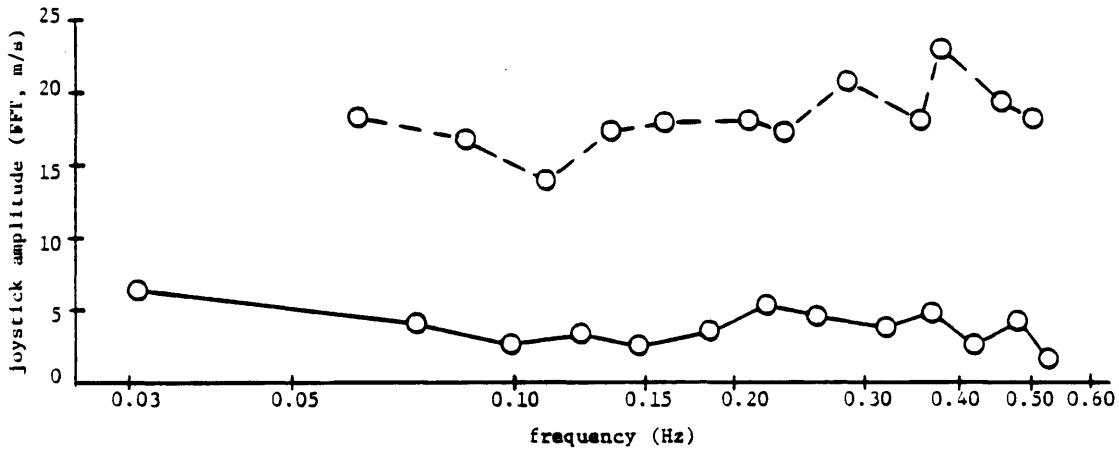
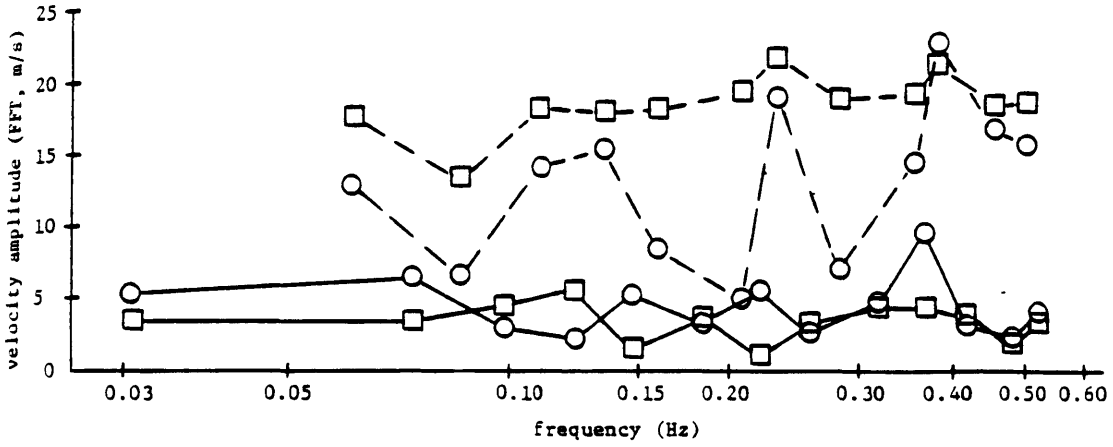


Figure 4.3.04 Frequency Spectrum. Subject DH,
DHPR02.PRO Profile #12, JSCALE=2.56

JSCALE = 1.5

DATA FILE	RMS VELOCITY RATIO	RMS JOYSTICK
11	1.26	0.201
16	1.34	0.164
11	1.34	0.162
18	0.685	0.162
23	0.730	0.162

JSCALE = 2.56

DATA FILE	RMS VELOCITY RATIO	RMS JOYSTICK
01	1.64	0.176 ✓
02	1.63	0.212
03	1.57	0.164

TABLE 4.3.02 RMS Data. Subject DH, DHPR02.PRO
Profile #12

since the RMS velocity ratios were also greater than 1.0 for the previous test, (Table 4.3.02), frequency spectrum plots were made for three of these completed runs. These plots showed that the joystick remnant was much higher and the velocity nulling was less than that for the JSCALE=2.56 case. The average Bode plot data was scattered as a result. It also suggests that the joystick gain was too high for precise control which caused the high joystick remnant. This data helps confirm the use of the $p=90\%$ criterium for finding JSCALE.

4.3.1 Summary

The results obtained for profile #12 were the type of results desired for the final test procedure. However, it was felt that the requirements of the HO were too low to fulfill the test procedure requirements. It would be difficult to see an increase in performance with the low disturbance amplitudes used. The acceleration amplitudes were already so low that to show an increase in performance the HO would have to have more precise control by many orders of magnitude. Also, if an HO's sensitivity were decreased, as expected, it would be very difficult to sense the already low disturbance. This would result in low control input amplitudes which would be difficult to separate from the remnant.

Another problem with this profile is seen by looking at the magnitudes of the acceleration inputs at the disturbance frequencies. Many of these accelerations are below the known threshold of 0.005 g. Thus, it is difficult to justify the result of the transfer function data with an input that the HO is supposedly not able to sense. (It should be noted that the dynamics of the hardware used to determine this threshold were probably not

as good as the Sled dynamics, since it is a more modern system. Also, the threshold for an acceleration disturbance of a sinusoid form, or that of an additional acceleration form, would probably be different. However, it was felt that it would be less controversial to avoid these possible intricacies and assume the 0.005 g threshold as valid.) It was felt that higher amplitudes on the order of 0.015 g at all frequencies should be used to avoid any problems with the threshold. This was the next direction taken.

While these profiles were deemed unusable for the test procedure requirements they were very educational. Some major points discovered were:

- 1) Lower amplitudes of the disturbance result in more complete runs, as expected. They also cause the HO to concentrate more which should help insure that the otolith system is the major contributor to the HO response.
- 2) RMS velocity ratios are an overall activity measure and not necessarily a measure of the velocity nulling task.
- 3) The velocity frequency spectrum gives a clear view of the performance of the velocity nulling task. The joystick frequency spectrum gives a clear view of the control remnant which should be low for precise results. (See section 6.4)
- 4) It is desired to have about 3 complete runs to use in the data analysis. This helps show the subject's consistency.

4.4 The H1PR04.PRO Series and Profile Design Problems.

As stated in the last section the amplitudes of profile #12 were low and it was felt that the data was questionable due to this. It was therefore

desired that the acceleration at all amplitudes be about 0.015 g. To design a profile to meet these conditions the profile generation program on the VAX computer was used. This was necessary as the PDP-11/34 minicomputer was too slow to be used effectively in the trial and error design mode that was utilized. The run times of 81.92 and 92.16 seconds were primarily used with only a few 102.4 second runs checked. The main outputs of interest used to determine if a profile was suitable were the velocity and acceleration amplitudes at each disturbance frequency, and the track length, maximum velocity, and maximum acceleration.

The technique used to search for a profile was straightforward. First the desired run time and frequency distribution was determined. Then, limits were set on track length and maximum acceleration, usually starting with 2.0 and 0.15 respectively, since past profiles had shown these values to be in the proper range. Next the scaling was chosen. Flat velocity, with or without first or second order filtering, and flat acceleration scaling were the types used most often. With these parameters decided upon the only other variable was the DEL frequency. The program was set up to cycle through varying values of DEL and output all the profile parameters. Usually these values were DEL=37 or 33 to DEL=337 or 333 by steps of 30 degrees. The profile that gave the minimum velocity with the desired acceleration or velocity amplitudes was then considered to be the best possible for the frequency range, distribution, and scaling used. Often the DEL variation was restricted and the step decreased to fine tune a promising profile, as changes in the output parameters varied greatly with DEL. The input variables were then adjusted as necessary until the desired output parameters were found.

The five profiles of this series that yielded the most useful information will now be examined. Fig. 4.4.01-.08 show the profiles and the associated histogram data, as well as the time histories of velocity and acceleration taken from runs with no subject.

4.4.1 Profile #1

Profile #1 was developed in an attempt to obtain a profile that had 0.015 g at each disturbance frequency. The flat amplitude scaling was used. A 0.015 g profile was not set up since it was felt that the maximum velocity of this profile was already too high and it would be significantly higher for the 0.015 g case.

The run completion rate of this profile was discouraging. The causes can be seen in the histogram data and the profile time history plots. The time history plots show that there is a part of the profile at the 50-55 second time period where there is a slow change in velocity with an associated low acceleration activity. This lack of acceleration causes confusion for the HO since he can sense the velocity through the cart vibration but not sense its direction. Subjects then tend to apply some low amplitude control input which causes them to drift away and reach the track limits. Also, the low number of velocity and acceleration reversals of the profile do not give enough acceleration input to the subject which makes control more difficult. The histogram data suggests this result as 17% of the profile acceleration command points are below the 0.005 g threshold.

This profile has too much emphasis on the low frequency end of the spectrum in both velocity and acceleration amplitudes. The high maximum

INPUT PARAMETERS

TRUN 92.160
 NSINES= 12
 FFREQ= 0.010850
 FLAT= 1
 DEL= 107.000
 FILTER 0
 FFULE= 0.080
 FFUS= 2.50
 FACC= 0.200
 TLOOP= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.140
 VELMAX 0.8095
 % USAGE OF TRACK=100.00
 STARTING POSITION= 0.734
 SCALE= 0.1333

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0651	0.33	0.014	3.
0.1194	0.18	0.014	113.
0.1411	0.15	0.014	221.
0.1845	0.11	0.014	330.
0.2062	0.10	0.014	78.
0.2496	0.08	0.014	188.
0.3147	0.07	0.014	298.
0.3364	0.06	0.014	46.
0.4015	0.05	0.014	156.
0.4449	0.05	0.014	265.
0.4666	0.05	0.014	373.
0.5100	0.04	0.014	123.

HISTOGRAM DATA

ACC BIN= 0.005	# ACC POINTS= 1742
ACC BIN= 0.010	# ACC POINTS= 1327
ACC BIN= 0.015	# ACC POINTS= 1196
ACC BIN= 0.020	# ACC POINTS= 1024
ACC BIN= 0.025	# ACC POINTS= 831
ACC BIN= 0.030	# ACC POINTS= 600
ACC BIN= 0.035	# ACC POINTS= 455
ACC BIN= 0.040	# ACC POINTS= 373
ACC BIN= 0.045	# ACC POINTS= 286
ACC BIN= 0.050	# ACC POINTS= 176
ACC BIN= 0.055	# ACC POINTS= 208
ACC BIN= 0.060	# ACC POINTS= 168
ACC BIN= 0.065	# ACC POINTS= 141
ACC BIN= 0.070	# ACC POINTS= 179
ACC BIN= 0.075	# ACC POINTS= 133
ACC BIN= 0.080	# ACC POINTS= 78
ACC BIN= 0.085	# ACC POINTS= 59
ACC BIN= 0.090	# ACC POINTS= 49
ACC BIN= 0.095	# ACC POINTS= 27
ACC BIN= 0.100	# ACC POINTS= 43
ACC BIN= 0.105	# ACC POINTS= 15
ACC BIN= 0.110	# ACC POINTS= 19
ACC BIN= 0.115	# ACC POINTS= 27
ACC BIN= 0.120	# ACC POINTS= 6
ACC BIN= 0.125	# ACC POINTS= 8
ACC BIN= 0.130	# ACC POINTS= 8
ACC BIN= 0.135	# ACC POINTS= 11
ACC BIN= 0.140	# ACC POINTS= 22
ACC BIN= 0.145	# ACC POINTS= 5
ACC BIN= 0.150	# ACC POINTS= 0

Figure 4.4.01 H1PRO4.PRO PProfile #1 Parameters

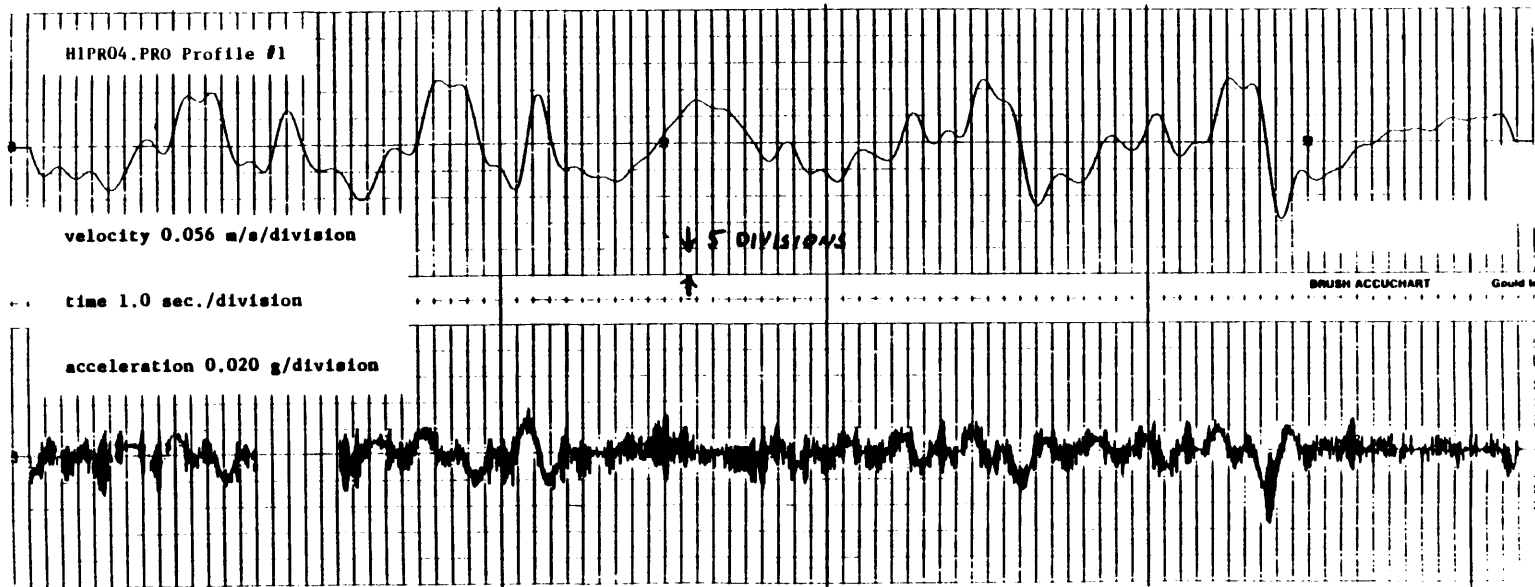


Figure 4.4.02 H1PRO4.PRO Profile #1 Time History

INPUT PARAMETERS

TRUN= 92.160
 NSINES= 12
 FREQ= 0.010850
 FLAI= 0
 DEL= 113.000
 FILIFK= 1
 FPOLE= 0.050
 FFUS= 2.50
 FACC= 0.200
 TLOOP= 0.010

PROFILE DESCRIPTION

MAY ACCEL= 0.156
 VELMAX= 0.7847
 % USAGE OF TRACK=100.00
 STARTING POSITION= 0.696
 SCALE= 0.3279

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0651	0.30	0.012	3.
0.1194	0.21	0.016	119.
0.1411	0.14	0.013	233.
0.1845	0.12	0.014	348.
0.2062	0.10	0.013	102.
0.2496	0.11	0.018	217.
0.3147	0.09	0.017	334.
0.3364	0.07	0.016	88.
0.4015	0.07	0.019	204.
0.4449	0.05	0.014	319.
0.4666	0.05	0.014	73.
0.5100	0.05	0.017	188.

HISTOGRAM DATA

ACC BIN= 0.005	# ACC POINTS= 1109
ACC BIN= 0.010	# ACC POINTS= 1436
ACC BIN= 0.015	# ACC POINTS= 1141
ACC BIN= 0.020	# ACC POINTS= 1017
ACC BIN= 0.025	# ACC POINTS= 782
ACC BIN= 0.030	# ACC POINTS= 670
ACC BIN= 0.035	# ACC POINTS= 459
ACC BIN= 0.040	# ACC POINTS= 462
ACC BIN= 0.045	# ACC POINTS= 407
ACC BIN= 0.050	# ACC POINTS= 248
ACC BIN= 0.055	# ACC POINTS= 230
ACC BIN= 0.060	# ACC POINTS= 210
ACC BIN= 0.065	# ACC POINTS= 192
ACC BIN= 0.070	# ACC POINTS= 102
ACC BIN= 0.075	# ACC POINTS= 141
ACC BIN= 0.080	# ACC POINTS= 148
ACC BIN= 0.085	# ACC POINTS= 120
ACC BIN= 0.090	# ACC POINTS= 84
ACC BIN= 0.095	# ACC POINTS= 35
ACC BIN= 0.100	# ACC POINTS= 38
ACC BIN= 0.105	# ACC POINTS= 29
ACC BIN= 0.110	# ACC POINTS= 26
ACC BIN= 0.115	# ACC POINTS= 13
ACC BIN= 0.120	# ACC POINTS= 12
ACC BIN= 0.125	# ACC POINTS= 15
ACC BIN= 0.130	# ACC POINTS= 20
ACC BIN= 0.135	# ACC POINTS= 19
ACC BIN= 0.140	# ACC POINTS= 6
ACC BIN= 0.145	# ACC POINTS= 8
ACC BIN= 0.150	# ACC POINTS= 10
ACC BIN= 0.155	# ACC POINTS= 16
ACC BIN= 0.160	# ACC POINTS= 11
ACC BIN= 0.165	# ACC POINTS= 0
ACC BIN= 0.170	# ACC POINTS= 0

Figure 4.4.03 H1PRO4.PRO Profile #3 Parameters

INPUT PARAMETERS

TRUN= 92.160
 NSINES= 12
 FFREQ= 0.010850
 FLNF= 0
 DEL= 217.000
 FILTER= 1
 FFOLE= 0.100
 FFIN= 1.80
 FACC= 0.200
 ILIMP= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.134
 VELMAX= 0.6498
 % IN-EDGE OF TRACK= 100.00
 STARTING POSITION= -0.564
 SCALE= 0.2488

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0651	0.19	0.008	1.
0.1194	0.16	0.012	248.
0.1411	0.12	0.010	135.
0.1845	0.10	0.012	23.
0.2062	0.09	0.012	270.
0.2496	0.10	0.016	157.
0.3147	0.08	0.015	45.
0.3364	0.07	0.014	292.
0.4015	0.07	0.017	179.
0.4449	0.05	0.013	67.
0.4666	0.04	0.013	314.
0.5100	0.05	0.015	201.

HISTOGRAM DATA

ACC BIN= 0.005	# ACC POINTS= 1377
ACC BIN= 0.010	# ACC POINTS= 1424
ACC BIN= 0.015	# ACC POINTS= 1243
ACC BIN= 0.020	# ACC POINTS= 1110
ACC BIN= 0.025	# ACC POINTS= 871
ACC BIN= 0.030	# ACC POINTS= 558
ACC BIN= 0.035	# ACC POINTS= 515
ACC BIN= 0.040	# ACC POINTS= 433
ACC BIN= 0.045	# ACC POINTS= 300
ACC BIN= 0.050	# ACC POINTS= 245
ACC BIN= 0.055	# ACC POINTS= 243
ACC BIN= 0.060	# ACC POINTS= 133
ACC BIN= 0.065	# ACC POINTS= 150
ACC BIN= 0.070	# ACC POINTS= 146
ACC BIN= 0.075	# ACC POINTS= 170
ACC BIN= 0.080	# ACC POINTS= 53
ACC BIN= 0.085	# ACC POINTS= 42
ACC BIN= 0.090	# ACC POINTS= 59
ACC BIN= 0.095	# ACC POINTS= 12
ACC BIN= 0.100	# ACC POINTS= 12
ACC BIN= 0.105	# ACC POINTS= 13
ACC BIN= 0.110	# ACC POINTS= 16
ACC BIN= 0.115	# ACC POINTS= 19
ACC BIN= 0.120	# ACC POINTS= 28
ACC BIN= 0.125	# ACC POINTS= 9
ACC BIN= 0.130	# ACC POINTS= 12
ACC BIN= 0.135	# ACC POINTS= 23
ACC BIN= 0.140	# ACC POINTS= 0
ACC BIN= 0.145	# ACC POINTS= 0
ACC BIN= 0.150	# ACC POINTS= 0

Figure 4.4.04 H1PRO4.PRO Profile #6 Parameters

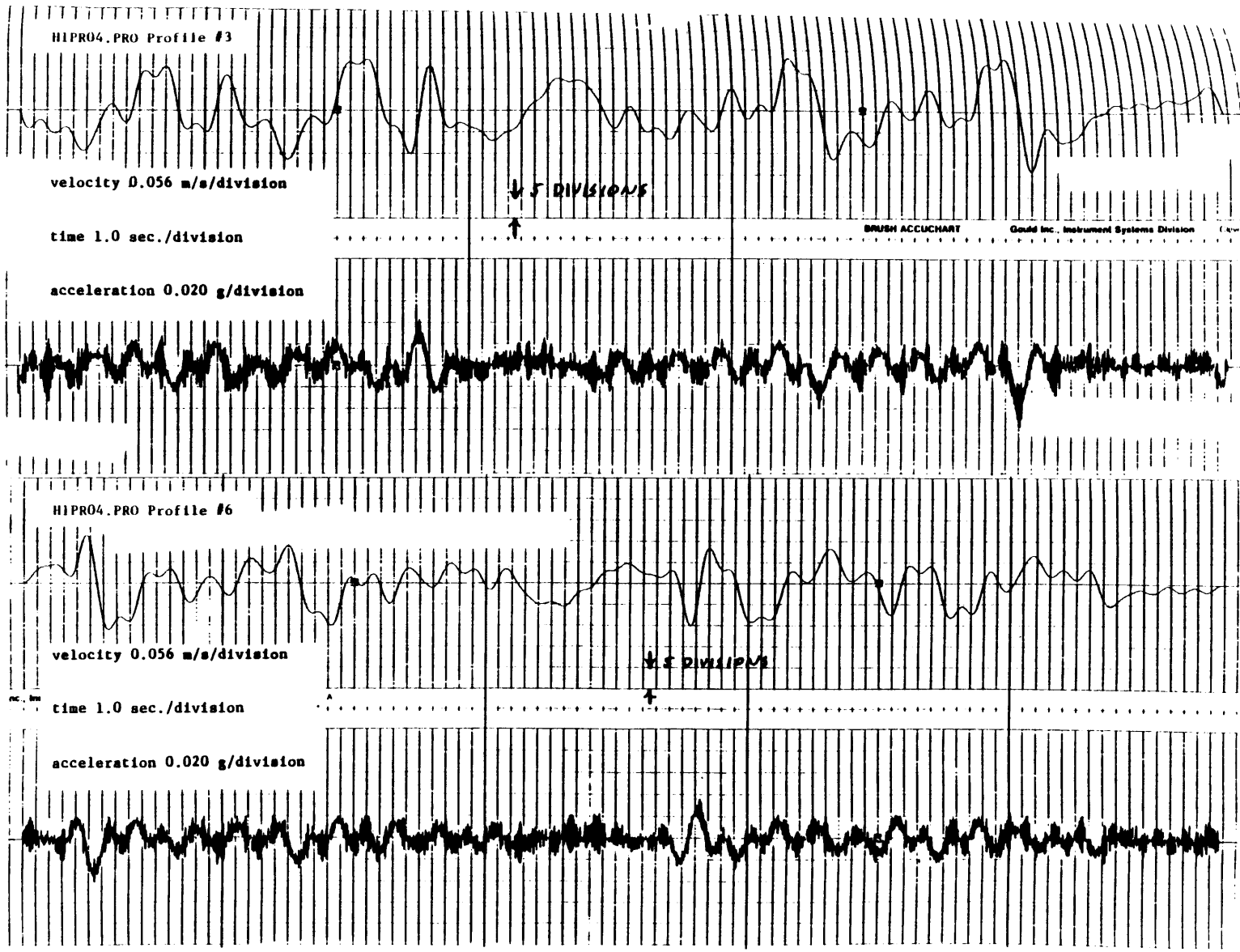


Figure 4.4.05 HPR04.PRO Profile #3 and #6 Time Histories

INPUT PARAMETERS

TRUN= 81.920
 NS(PHS)= 12
 FFPED= 0.012210
 FLAT= 0
 DEL= 100.000
 FILTER= 1
 FFOLE= 0.200
 FFIN= 2.00
 FACU= 0.200
 TLOUP= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.123
 VELMAX= 0.4337
 % USAGE OF TRACK=100.00
 STARTING POSITION= 0.044
 SCALE= 0.2082

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0410	0.11	0.004	1.
0.0804	0.12	0.007	105.
0.1099	0.11	0.008	208.
0.1343	0.11	0.009	312.
0.1587	0.12	0.012	55.
0.2075	0.11	0.015	159.
0.2319	0.10	0.015	263.
0.2808	0.12	0.021	7.
0.3740	0.10	0.022	111.
0.3784	0.08	0.020	215.
0.4217	0.09	0.025	319.
0.5005	0.07	0.023	63.

HISTOGRAM DATA

ACC BIN	ACC POINTS
0.005	809
0.010	882
0.015	1029
0.020	1191
0.025	654
0.030	378
0.035	340
0.040	387
0.045	383
0.050	272
0.055	302
0.060	276
0.065	281
0.070	247
0.075	194
0.080	165
0.085	47
0.090	69
0.095	38
0.100	26
0.105	25
0.110	34
0.115	42
0.120	61
0.125	40
0.130	0
0.135	0
0.140	0
0.145	0
0.150	0

Figure 4.4.06 H1PRO4.PRO Profile #11 Parameters

INPUT PARAMETERS

TRUN= 91.920
 NSINFS= 12
 FFREQ= 0.012210
 FLAT= 0
 DEL= 247.000
 FILTER= 0
 FFOLE= 0.200
 FPOS= 1.60
 FAUG= 0.200
 TLOOF= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.129
 VELMAX= 0.6206
 % USAGE OF TRACK=100.00
 STARTING POSITION=-0.207
 SCALE= 0.0968

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0610	0.10	0.004	0.
0.0854	0.10	0.005	247.
0.1099	0.10	0.007	135.
0.1343	0.10	0.008	22.
0.1587	0.10	0.010	269.
0.2075	0.10	0.013	156.
0.2319	0.10	0.014	43.
0.2808	0.10	0.017	290.
0.3540	0.10	0.022	178.
0.3784	0.10	0.023	65.
0.4517	0.10	0.028	312.
0.5005	0.10	0.031	199.

HISTOGRAM DATA

ACC BIN= 0.005	ACC POINTS= 715
ACC BIN= 0.010	ACC POINTS= 854
ACC BIN= 0.015	ACC POINTS= 888
ACC BIN= 0.020	ACC POINTS= 810
ACC BIN= 0.025	ACC POINTS= 854
ACC BIN= 0.030	ACC POINTS= 577
ACC BIN= 0.035	ACC POINTS= 597
ACC BIN= 0.040	ACC POINTS= 320
ACC BIN= 0.045	ACC POINTS= 311
ACC BIN= 0.050	ACC POINTS= 268
ACC BIN= 0.055	ACC POINTS= 298
ACC BIN= 0.060	ACC POINTS= 221
ACC BIN= 0.065	ACC POINTS= 213
ACC BIN= 0.070	ACC POINTS= 245
ACC BIN= 0.075	ACC POINTS= 276
ACC BIN= 0.080	ACC POINTS= 178
ACC BIN= 0.085	ACC POINTS= 104
ACC BIN= 0.090	ACC POINTS= 113
ACC BIN= 0.095	ACC POINTS= 85
ACC BIN= 0.100	ACC POINTS= 63
ACC BIN= 0.105	ACC POINTS= 22
ACC BIN= 0.110	ACC POINTS= 29
ACC BIN= 0.115	ACC POINTS= 35
ACC BIN= 0.120	ACC POINTS= 56
ACC BIN= 0.125	ACC POINTS= 20
ACC BIN= 0.130	ACC POINTS= 40
ACC BIN= 0.135	ACC POINTS= 0
ACC BIN= 0.140	ACC POINTS= 0
ACC BIN= 0.145	ACC POINTS= 0
ACC BIN= 0.150	ACC POINTS= 0

Figure 4.4.07 H1PRO4.PRO Profile #12 Parameters

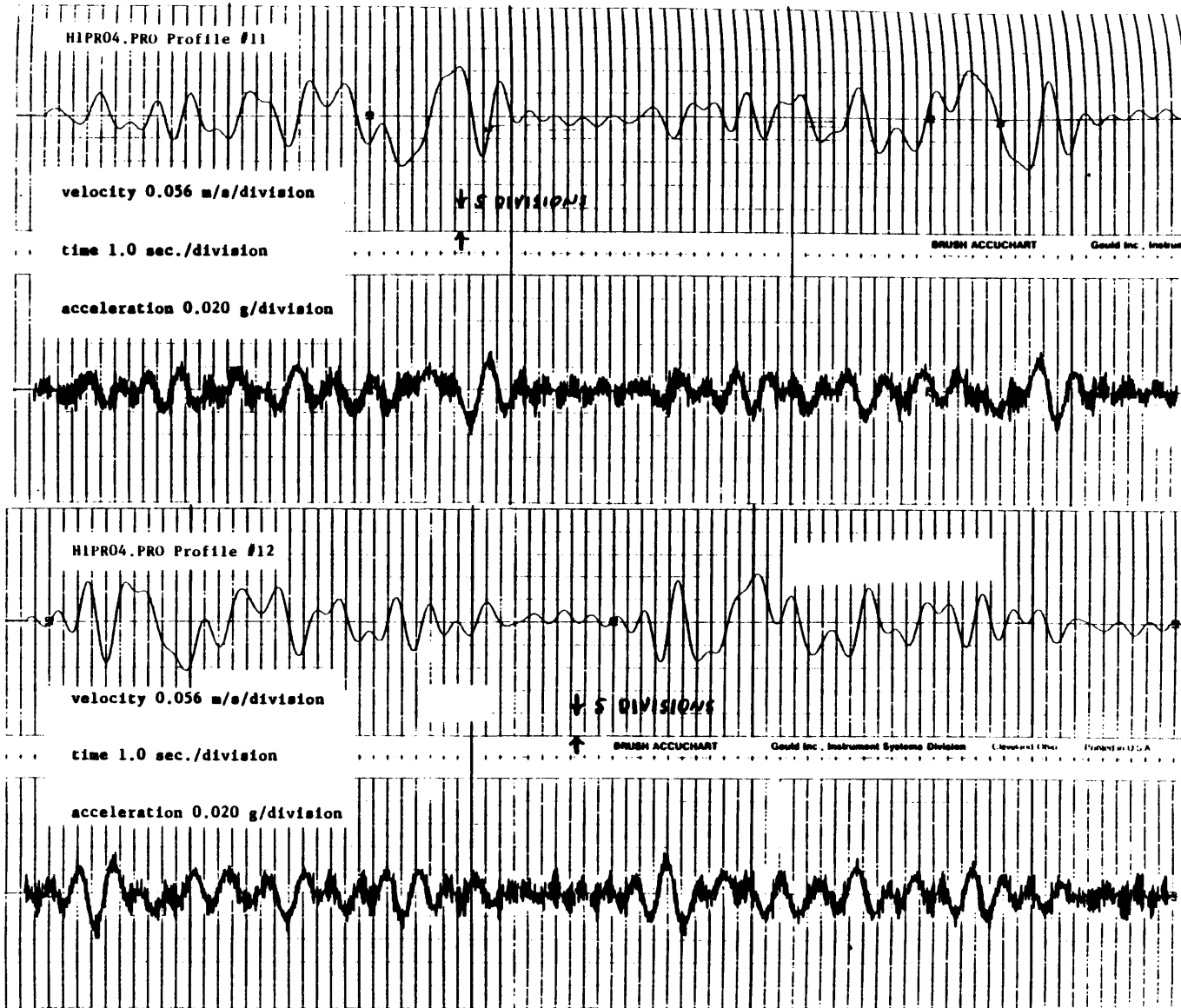


Figure 4.4.08 H1PRO4.PRO Profile #11 and #12 Time Histories

velocity and associated high gain (JSCALE=1.63) may also have contributed to the low run completion rate. Lowering the amplitude would mean an even greater percentage of acceleration command points below threshold and less velocity and acceleration reversals which would only worsen the HO performance. In conclusion, flat amplitude profiles can not be used for the final test procedure.

4.4.2 Profiles #3 and #6

Profiles #3 and #6 were set up using constant velocity and first order filter scaling. Due to the results of profile #1 the acceleration and velocity amplitudes were increased at the higher frequencies and lowered at the lower frequencies. Profile #3's amplitudes are higher to try to stay away from the threshold. The maximum velocity and associated joystick gain are high as a result. Profile #6's amplitudes were determined by a maximum velocity limitation of about 0.66 m/s. (From some testing which will not be elaborated upon here it was decided that JSCALE=2.0 should be the practical lower limit of the joystick gain. This allows a maximum velocity of 0.66 m/s with P=95%, which was used instead of P=90% to achieve a slightly higher velocity with the same joystick sensitivity. This is a compromise between the desire for a higher maximum velocity, to stay away from threshold amplitudes, and the desire for precise HO control without a high remnant.) The histograms show a decrease in the number of points below 0.005 g. The time histories show more zero crossings as expected.

Run completion rates with these profiles were low also. This signifies that the emphasis on the lower frequencies was still too great for subjects to be able to control the profile effectively. One run of each profile was

completed, though, and plots for profile #6 are shown in Fig.4.4.2.01,.02. The Bode plot shows very little scatter and profile #3 showed low scatter also. The joystick frequency spectrum shows a low remnant except at the high frequencies. At these frequencies the velocity disturbance amplitude is so low that it is difficult to separate the joystick remnant from the joystick disturbance frequency amplitudes. The plots tend to show the flat gain and low phase at low frequencies that is characteristic of the HO. These results support the theory that higher input amplitudes give more consistent data, which is desired for the final test profile.

4.4.3 Profile #12

Profile #12 was set up as a limiting case test. It is just an increase of the previously successful DHPR02.PRO profile #12 to a maximum velocity of 0.62 and JSSCALE=2.13 using a P=95% scaling criterion. It represents the greatest amplitudes possible at the high frequencies. (The flat acceleration case may be considered to have the greatest amplitudes possible at the low frequencies so it is also a limiting case.) Profile #11 has some first order filtering and is close to profile #12 in maximum velocity but has less zero crossings. It is included here to illustrate a single point. The time histories of profile #12 show the maximum number of zero crossings possible. The histogram data shows a low number of points below 0.005 g as expected.

The run completion rates for profile #12 were intermediate compared to previous runs. No runs of profile #11 were completed. The time histories show that the maximum velocity activity of profile #12 occurs at the beginning and after the midpoint of the run. Those for profile #11 occur

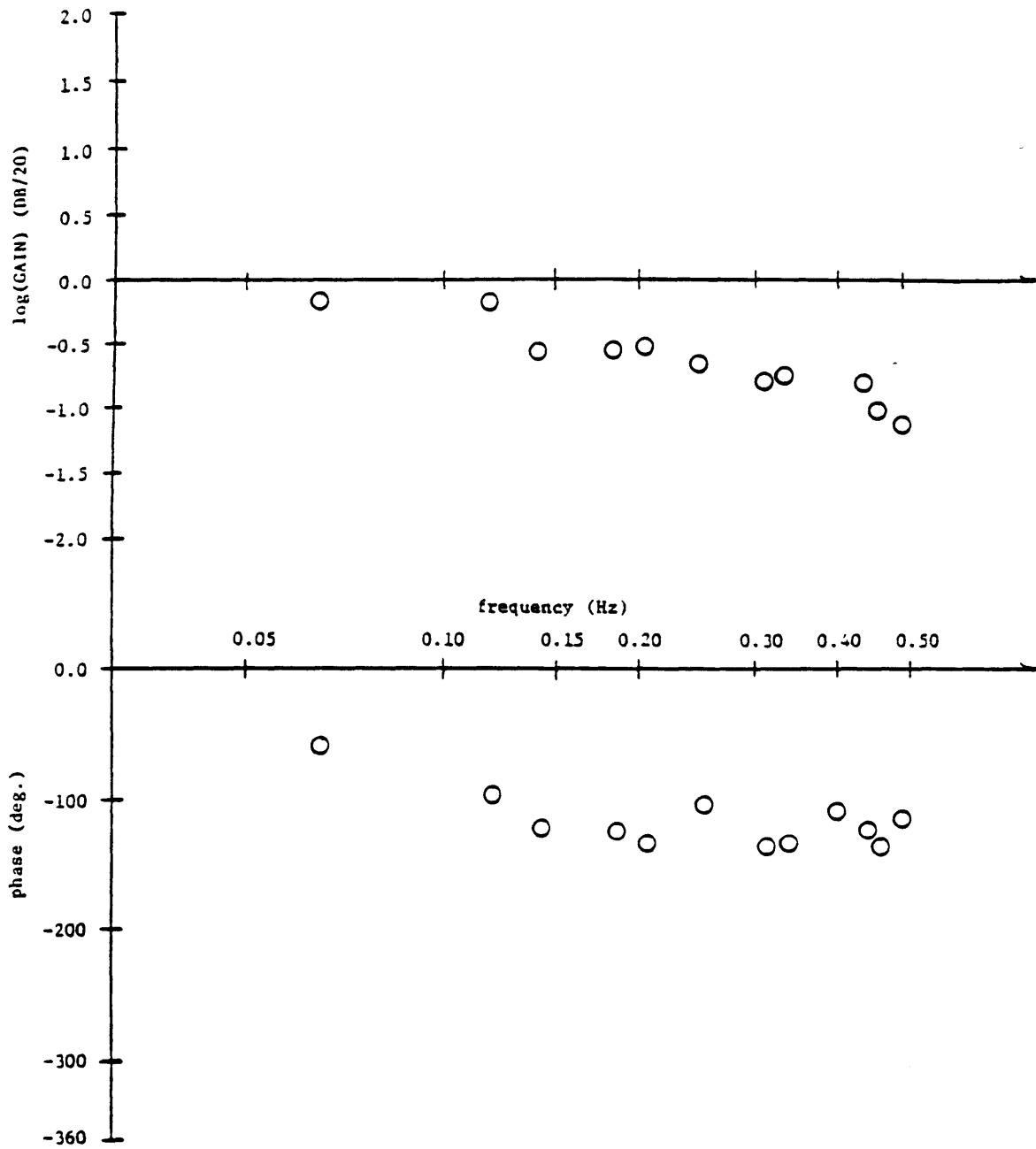


Figure 4.4.2.01 Bode Plot. Subject DH,
H1PR04.PRO Profile #6

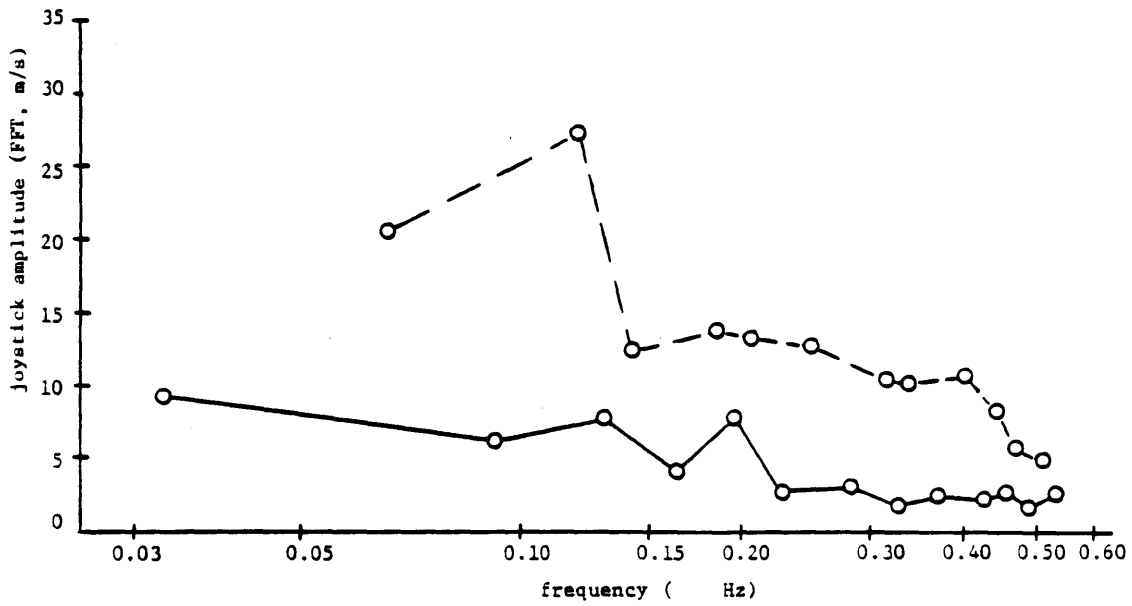
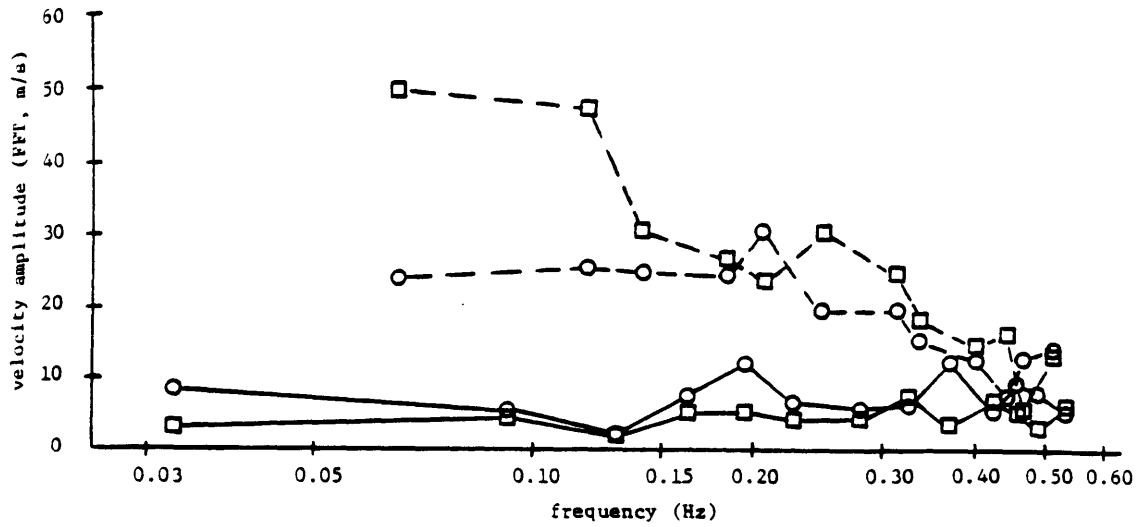


Figure 4.4.2.02 Frequency Spectrum. Subject DH
H1PRO4.PRO Profile #6

before the midpoint and near the end of the run. This is the major difference between the two profiles, and it is felt to contribute to the differences in the run completion rate. It was felt that having the maximum velocity activity near the beginning of the run gave the subject a chance to react to the highest amplitudes of the disturbance while still near the center of the track. This means that the subject was not allowed to drift near the track ends during a low activity period before the first maximum velocity activity. This may also have served to acquaint the subject with the maximum velocity at the beginning of the run and thereby improve his awareness of what to expect. This point is debatable but the idea was useful in other ways as will be shown.

Typical plots are shown in Fig. 4.4.3.01,02. The Bode plot shows much scatter. The joystick frequency spectrum shows a low remnant, in general, but the velocity spectrum shows a high remnant and large amplitude oscillations at the disturbance frequencies. This is caused by the difficulty of the subject to respond to the high frequency motion. It is felt that the HO can not adequately control the profile that has predominant amplitudes at the higher frequencies. This is due to the drop-off in gain associated with the otolith system at high frequencies as shown in the Young and Meiry model. In conclusion, flat velocity profiles do not meet the test requirements.

4.4.4 Summary

The main points illustrated by these profiles are:

- 1) Flat amplitude profiles have input amplitudes that are too high at the low frequencies. This causes a lack of acceleration cues to the HO.

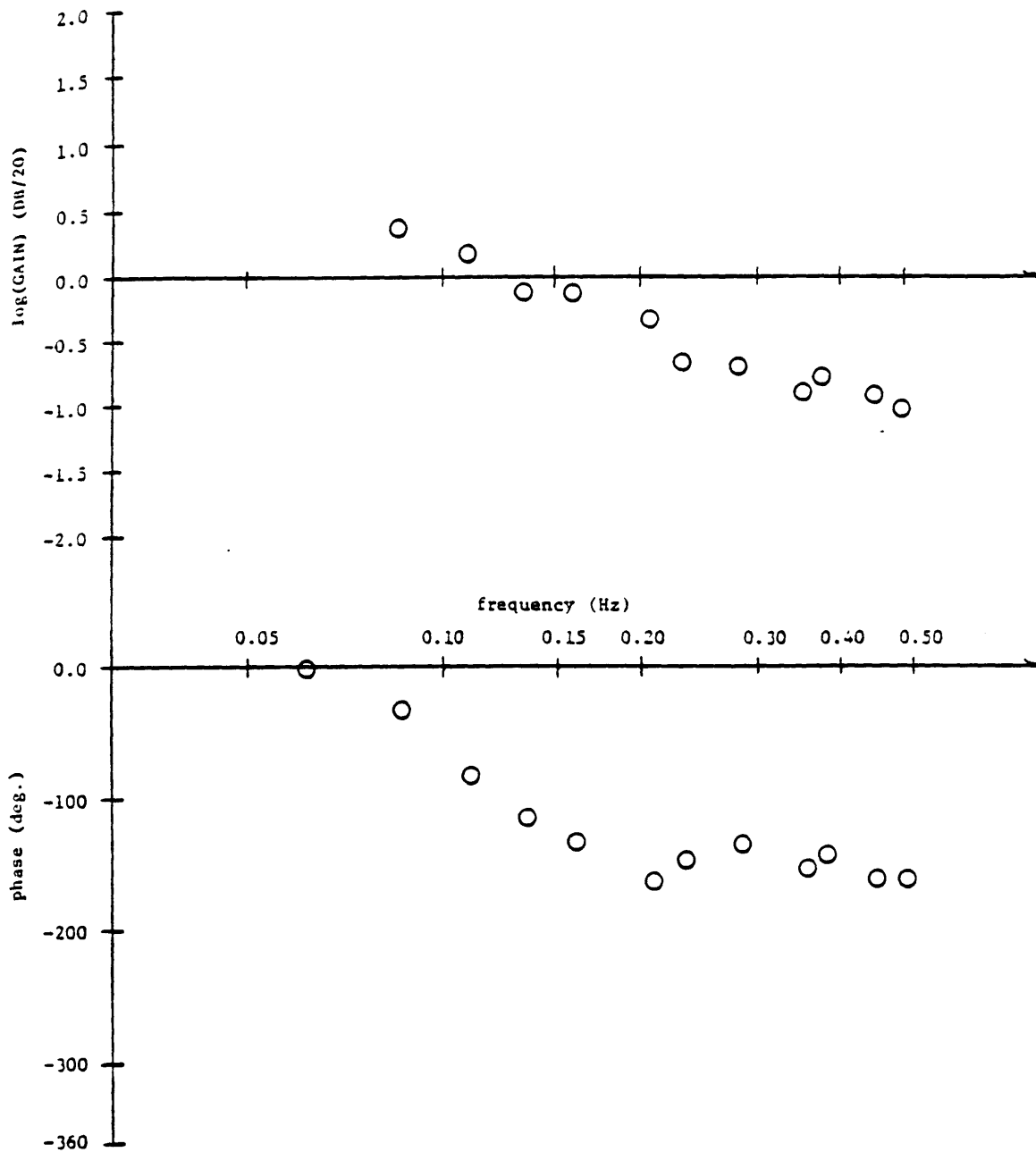


Figure 4.4.3.01 Bode Plot. Subject DH,
H1PRO4.PRO Profile #12

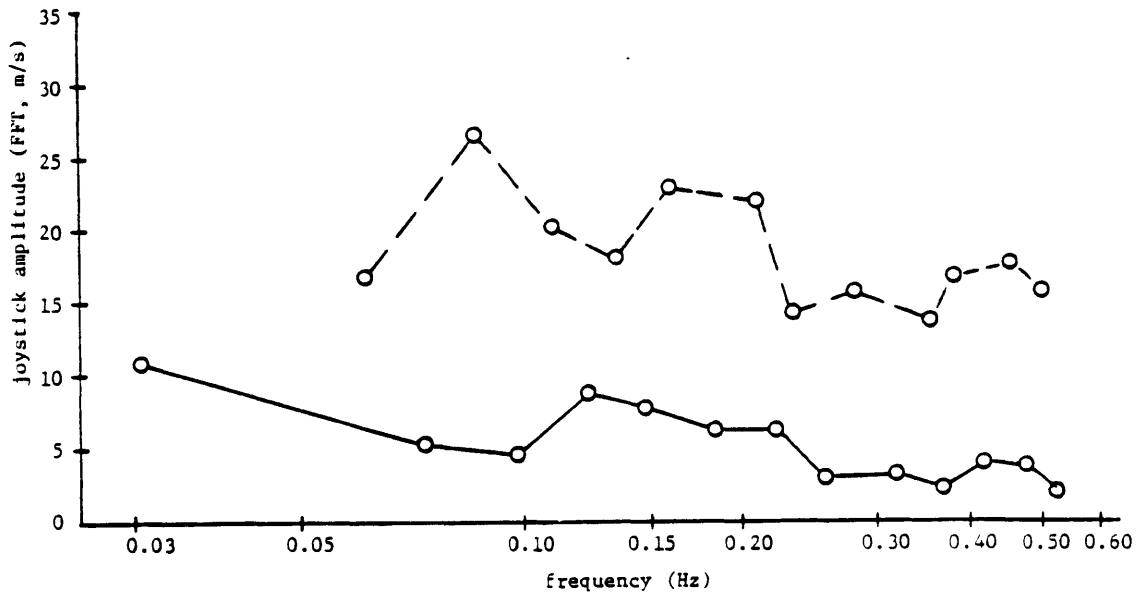
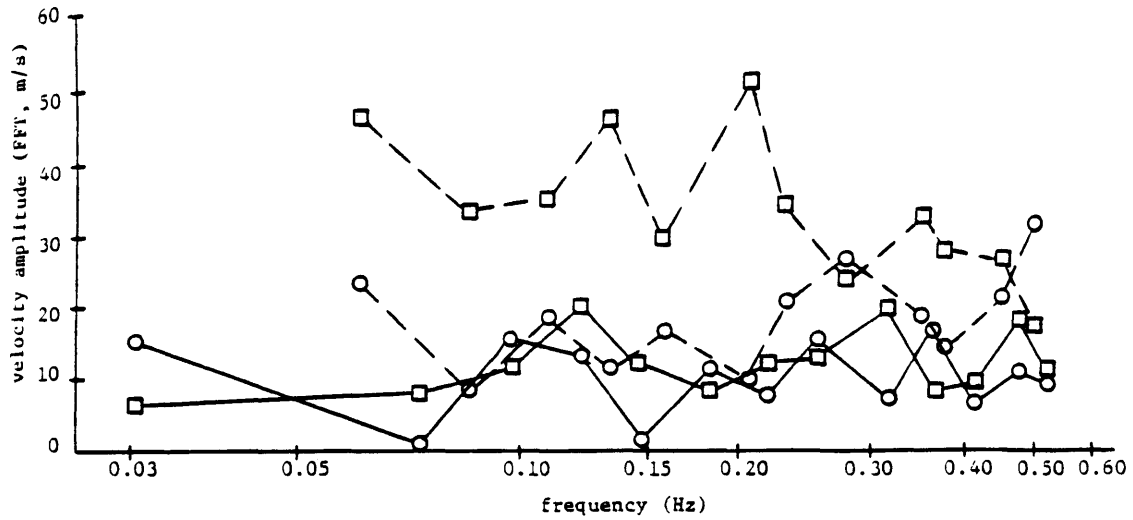


Figure 4.4.3.02 Frequency Spectrum. Subject DH
H1PRO4.PRO Profile #12

- 2) Flat velocity profiles have input amplitudes that are too high at the high frequency end of the spectrum. This causes poor HO performance due to the poor otolith response at higher frequencies.
- 3) The intermediate scaled case shown gave a low run completion rate, but data with little scatter. This is the type of compromise profile, in terms of input amplitudes at high and low frequencies, that is desired for the final profile.

4.5 Further Population Testing

At this point it was decided that population studies should begin to examine the responses seen in more detail. Profiles #3 and #6 had given the best data so far and with their acceleration amplitudes well above threshold they were the best choice to use in further testing. Profile #6 was favored since its maximum velocity points occurred at similar times to those of profile #12. The max velocity of profile #6 was 0.66 m/s with JSCALE=1.97. This maximum velocity was reduced slightly to 0.65 m/s with JSCALE=2.03 to comply with the JSCALE=2.0 limitation stated in the section 4.4.2. The resulting profile is shown in Fig. 4.5.01,.02. It was felt that in this further testing the suggested velocity limitations should be fully complied with to provide more coherency with the previous testing.

Using previous experience, a more procedural method of conducting tests was used. Subjects were seated in the cart and the velocity nulling task was explained. They were told to use any cues and any control strategy desired, but once they felt comfortable with their technique not to change it. A pair of opaque goggles were then put on the subject and kept on

INPUT PARAMETERS

TRUN= 92.160
 NSTRES= 12
 FFREQ= 0.010850
 FLAT= 0
 DEL= 247.000
 FILTER= 1
 FFOLE= 0.100
 FPOS= 1.75
 FACC= 0.200
 TLOOP= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.131
 VELMAX= 0.6512
 % USAGE OF TRACK=100.00
 STARTING POSITION =0.548
 SCALE= 0.7419

FREQ. (HZ)	AMP1	(G/S)	AMP2 (G)	PHASE (DEG)
0.0551		0.19	0.008	1.
0.1144		0.15	0.017	248.
0.1411		0.11	0.010	135.
0.1845		0.10	0.012	23.
0.2062		0.09	0.011	270.
0.2496		0.10	0.015	157.
0.3147		0.07	0.015	45.
0.3364		0.06	0.014	292.
0.4015		0.06	0.017	179.
0.4449		0.05	0.013	67.
0.4666		0.04	0.012	314.
0.5100		0.05	0.015	201.

HISTOGRAM DATA

ACC BIN= 0.005	# ACC POINTS= 1425
ACC BIN= 0.010	# ACC POINTS= 1455
ACC BIN= 0.015	# ACC POINTS= 1259
ACC BIN= 0.020	# ACC POINTS= 1120
ACC BIN= 0.025	# ACC POINTS= 844
ACC BIN= 0.030	# ACC POINTS= 542
ACC BIN= 0.035	# ACC POINTS= 323
ACC BIN= 0.040	# ACC POINTS= 424
ACC BIN= 0.045	# ACC POINTS= 285
ACC BIN= 0.050	# ACC POINTS= 250
ACC BIN= 0.055	# ACC POINTS= 203
ACC BIN= 0.060	# ACC POINTS= 126
ACC BIN= 0.065	# ACC POINTS= 172
ACC BIN= 0.070	# ACC POINTS= 145
ACC BIN= 0.075	# ACC POINTS= 136
ACC BIN= 0.080	# ACC POINTS= 39
ACC BIN= 0.085	# ACC POINTS= 57
ACC BIN= 0.090	# ACC POINTS= 33
ACC BIN= 0.095	# ACC POINTS= 12
ACC BIN= 0.105	# ACC POINTS= 0
ACC BIN= 0.160	# ACC POINTS= 0
ACC BIN= 0.165	# ACC POINTS= 0
ACC BIN= 0.170	# ACC POINTS= 0
ACC BIN= 0.175	# ACC POINTS= 0
ACC BIN= 0.180	# ACC POINTS= 0
ACC BIN= 0.185	# ACC POINTS= 0
ACC BIN= 0.190	# ACC POINTS= 0
ACC BIN= 0.195	# ACC POINTS= 0
ACC BIN= 0.200	# ACC POINTS= 0
ACC BIN= 0.205	# ACC POINTS= 0

Figure 4.5.01 H1PRO4.PRO Profile #17 Parameters

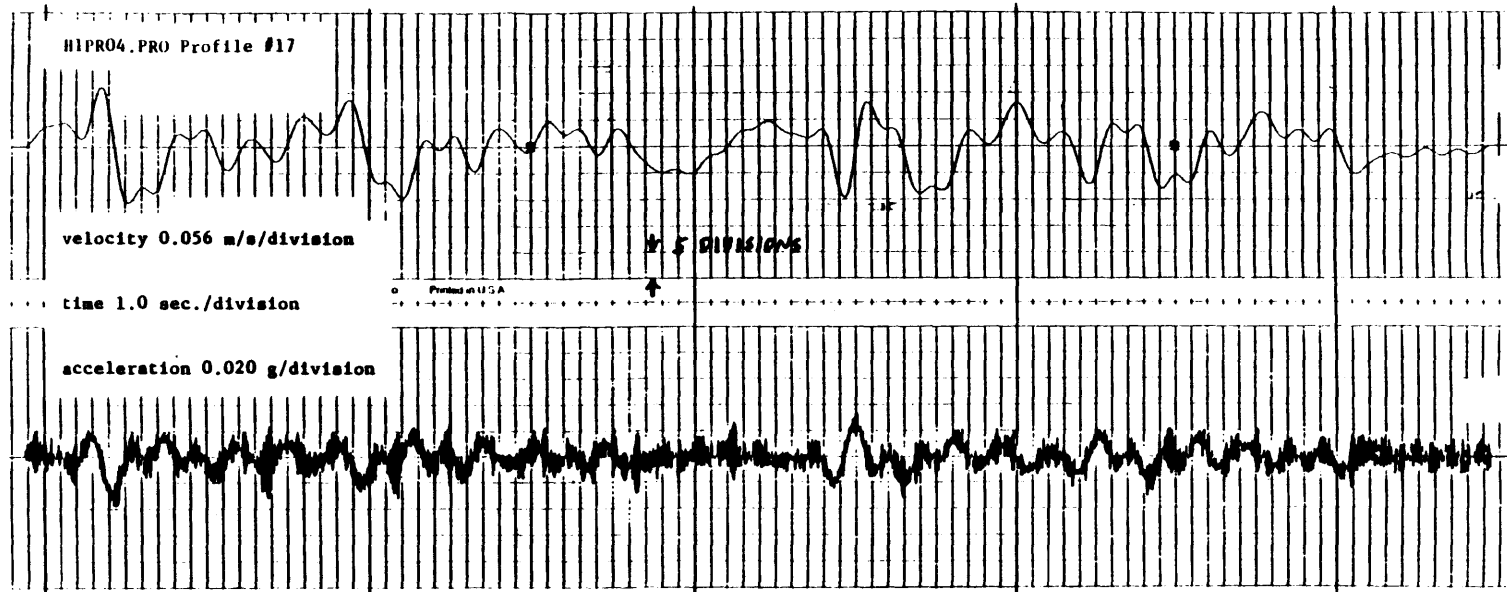


Figure 4.5.02 H1PRO4.PRO Profile #17 Time History

during all runs. The enclosed head restraint was used for this and all further testing. They were then given the null profile run (no disturbance input) and asked to practice with the joystick until they felt comfortable with the control sensitivity. A practice profile was then run with no subject input. This practice profile was then run with subject control until the subject started completing runs or reached his maximum level of performance. The data profile was then run with no control. The data profile was then run with control until 3 runs were completed. Data was stored for data profile runs only. The joystick voltage range was checked after every one or two runs to check for drift. All completed runs were used in the data reduction. Velocity and joystick frequency spectra, and velocity RMS ratio and joystick RMS were obtained for each run. Bode data was obtained from the average of the completed runs.

Most of the reasoning used in this refined procedure is self-explanatory but some points should be made. Subjects were told to use any possible cues so they would not try to avoid cues and so lose concentration on their otolith cues. Subjects were allowed to ride through the profiles without any control so they could be familiarized with the amplitudes and frequency range of the disturbance and the associated motion cues. Often during a run there was confusion as to what was the disturbance and what was the HO control so this helped alleviate this problem. Only three data runs were taken to lessen the amount of data analysis required and to keep the test time limited to roughly one hour. All runs were done with the subject blindfolded so the subject was only concentrating on the cues used in the data runs and not on extraneous cues from other senses.

Four subjects were tested. The rate of completion was not as high as

desired but this was expected. It was felt that a profile that gave good results was the top priority and run completion was secondary. The results were mixed, however. Fig. 4.5.03-06 show typical plots. For two subjects three or more runs were completed and the Bode plots were similar to previous plots. The one sigma deviations for one of these subjects, subject MM, were similar to previous data, but the author's deviations were remarkably low. The author was intimately familiar with the profile, knowing when the high and low velocity disturbances occurred, and this greatly helped his control. Also, the author had the most experience of any of the subjects tested. For the other two subjects the run completion rate was low and the data scatter was high.

The velocity RMS ratios are not consistent for these runs but the joystick RMS errors are. (Table 4.5.01) This suggests some difficulty in maintaining precise control and is probably due to still too few acceleration cues and the associated difficulty in controlling the low frequency velocities. This would allow the subject to slowly wander over the full track length, as was noted during some runs, while still yielding effective overall velocity nulling but a higher velocity RMS ratio.

Upon closer examination of the control strategy observed during the tests of these subjects two distinct types of control were seen. The two subjects with the most scattered data were using the control injection technique. Subject MS was one of these subjects and his data is shown in Fig. 4.5.03,04. With this technique the subject attempts to determine what the disturbance is by inputting some high frequency control of significant amplitude and noting the response. The control position is then adjusted by noting if the motion is increased or decreased. As is seen from the

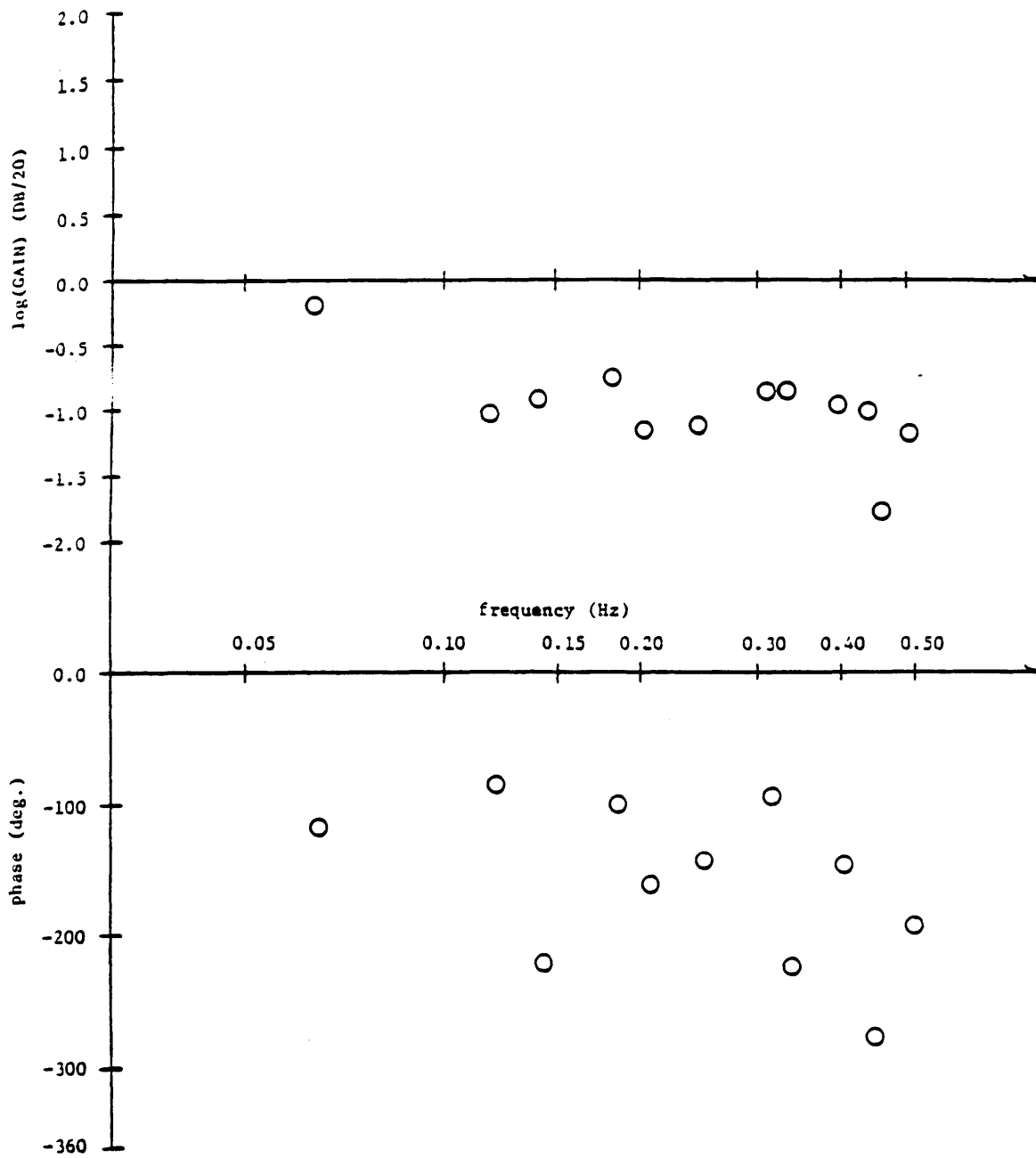


Figure 4.5.03 Bode Plot. Subject MS,
H1PRO4.PRO PProfile #17
(average of two runs,
deviations not shown)

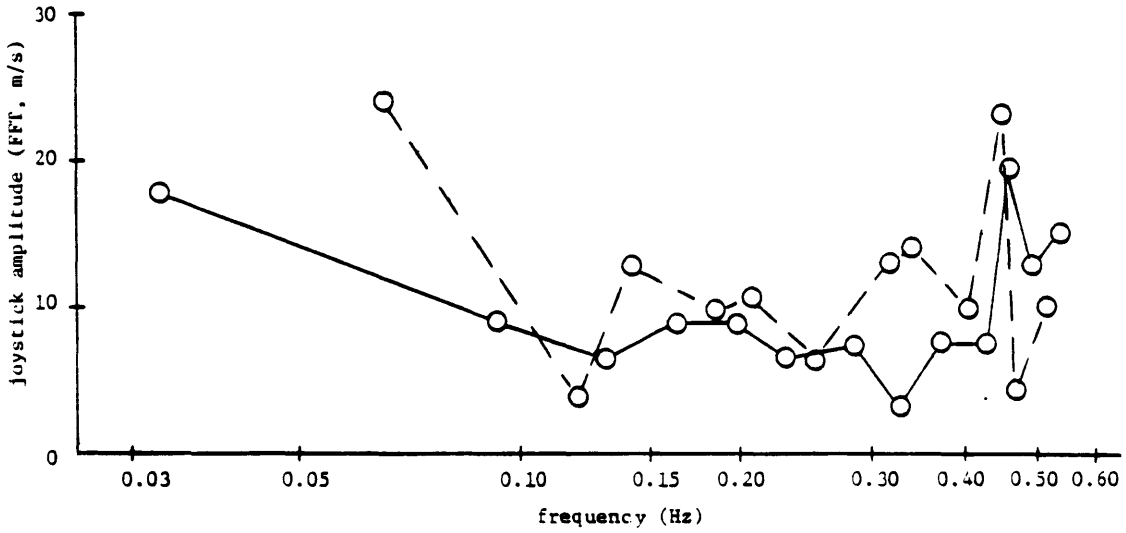
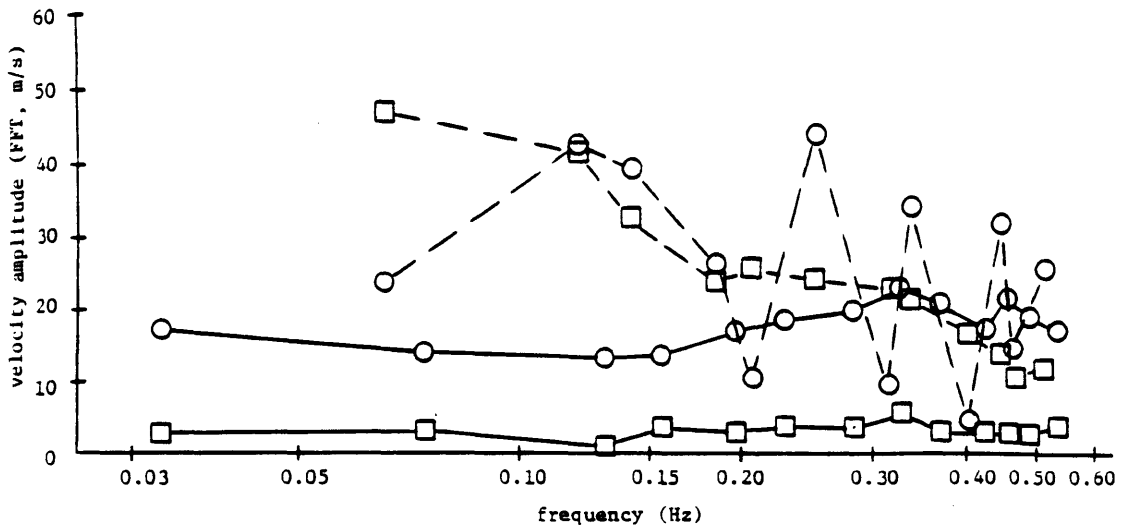


Figure 4.5.04 Frequency Spectrum. Subject MS, H1PRO4.PRO Profile #17

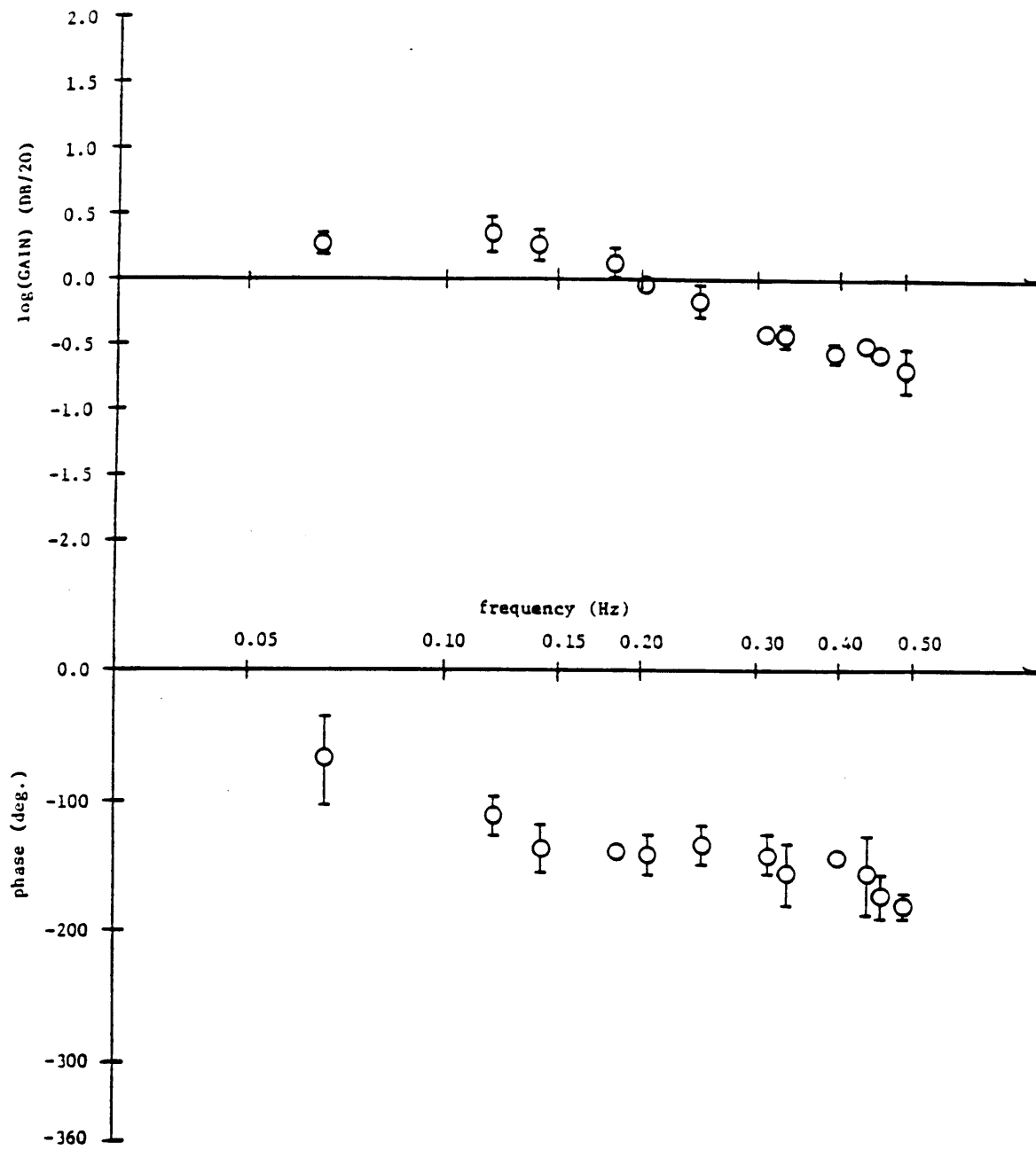


Figure 4.5.05 Bode Plot. Subject DH,
H1PRO4.PRO Profile #17

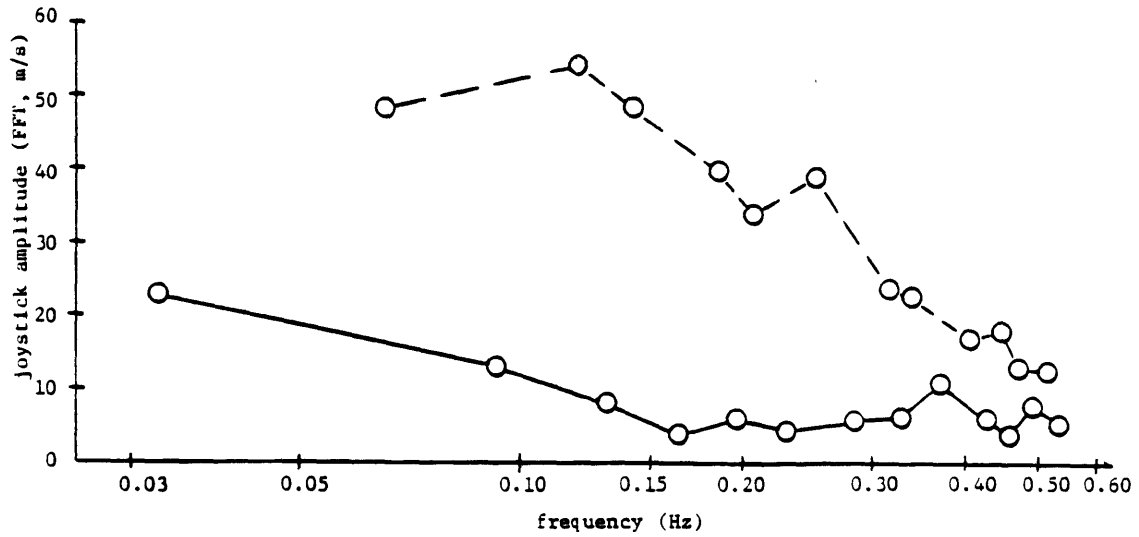
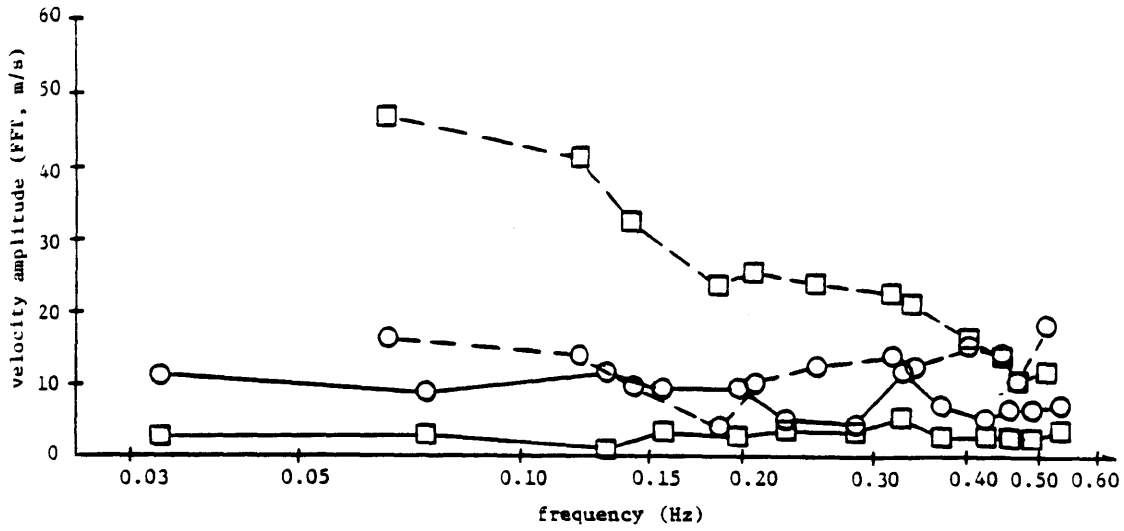


Figure 4.5.06 Frequency Spectrum. Subject DH,
H1PRO4.PRO Profile #17

SUBJECT	DATA FILE	RMS VELOCITY RATIO	RMS JOYSTICK
AA	03	3.63	0.346
MS	04	1.19	0.173
	11	3.06	0.241
MM	02	0.790	0.221
	03	1.22	0.223
	05	2.01	0.242
DH	05	1.01	0.191
	08	0.642	0.180
	12	1.76	0.207

TABLE 4.5.01 RMS Data. H1PRO4.PRO Profile #17

plots this type of control results in high remnants of velocity and joystick and also does not have much effect on the disturbance velocity.

Subjects DH and MM used a more passive, reactionary control. This technique is to wait for the disturbance before responding with the control. Some minimal control injection may also be used but it is not a significant amount of the control. The primary cue is then the acceleration sensed during the acceleration reversals. As is seen in the data for subject DH the results are much more consistent. It is also noticed that although these Bode plots show the same general trends, there are differences in the break frequencies of the gain plots. This was a satisfying and desirable result as it showed that the transfer functions could detect differences in subject performance.

4.5.1 Summary

The conclusions of this testing are as follows:

- 1) For data with little scatter and overall better velocity nulling the HO should try to REACT to the disturbance and not attempt to search out the disturbance.
- 2) This profile still has input amplitudes that are too high at the lower frequencies.
- 3) The data for the reacting subjects was similar in quality to that for other HO experiments. Individual differences could also be seen which is an indication of the effectiveness of the profile and the closed-loop test.
- 4) Practice and knowledge of the profile can greatly improve the subjects performance and the quality of the data.

5) The test procedure and data analysis procedure are basically sound.

There were still some major problems with this profile. One was the low run completion rate. It was felt that the amplitudes at the low frequencies were still too high, resulting in too few acceleration cues. Another problem was the low velocity amplitude at the high frequencies. As is seen from the disturbance velocity frequency spectrum the velocity of the disturbance and the remnant will often mesh at the high frequencies. It is difficult to distinguish between the two and this makes the remnant and control effectiveness question more difficult. Also, in order to get higher amplitudes at the lower frequencies while maintaining the track length and maximum velocity constraints it was necessary to limit the number of low frequency disturbance frequencies. This resulted in only one very low frequency point at 0.065 Hz with the next at 0.11 Hz. It was felt that a profile with more evenly distributed frequencies on the log(rad/sec) scale was desirable. Another profile was needed.

CHAPTER 5

THE FINAL EXPERIMENT

5.1) The Experimental Method

All of the lessons of the previous testing were used to determine the final profile. Consistent data had been obtained from a very low amplitude flat velocity profile, DHPR02.PRO profile #12. However, the input amplitudes of this profile were too low to fulfill the test procedure requirements. Flat acceleration profiles yielded low run completion rates and were not favorable due to the portions of the profile that had minimal acceleration disturbances and few zero-crossings.

Profiles with acceleration amplitudes in the 0.008-0.017 g range gave very consistent Bode plots. However, these profiles were still emphasizing the amplitudes at low frequencies too much. A flat velocity profile with increased amplitudes to 0.004-0.031 g gave a high number of zero crossings and a high run completion rate. The Bode plots from this profile showed some inconsistency and it was felt that the results would suffer due to the low amplitudes at the low frequencies and corresponding emphasis on the high frequencies. This is particularly important since better velocity nulling generally occurs at the low frequencies. This also points out the need for more low frequency points than had been used in the previous test. As is so often the case in engineering work a compromise was needed.

In generating a final profile the desire was to lower the input amplitudes at the low frequencies and raise them at the high frequencies. An average range between H1PR04.PRO profiles #17 and #12 was felt to yield

the best values of input acceleration amplitudes. This range was 0.06-0.24 g. It was also felt that the 81.92 second run time should be used as it had given the desired track length of about 2 m and a maximum velocity of about 0.65 m/s while allowing more low frequency points.

It was impossible to obtain these input amplitude requirements while maintaining the track length and maximum velocity limits. The input amplitudes had to be lowered to resolve this situation as higher input amplitudes had given poorer results in the past. The final profiles are shown in Fig. 5.1.01-.03. As is seen the range of acceleration amplitudes is 0.005-0.021 g. The lowest amplitude is still not below threshold so it was felt to be acceptable. The maximum velocity has also been lowered to 0.63 m/s giving JSCALE=2.13, to allow slightly less joystick sensitivity than that for the profile of section 4.5. The overall emphasis of the input amplitudes have been shifted to the higher frequencies as desired. A frequency determined by an even number multiplying the base frequency was included to prevent the profile from being made up of two identical halves. Two profiles were found to meet these conditions, one with DEL=253 and the other with DEL=103 degrees. In order to chose one profile for obtaining data the time histories were checked. (Fig. 5.1.03) As is seen profile #2 has its maximum velocity activity at the beginning and after the midpoint of the run. Profile #1 is about 180 degrees out of phase with this activity but is almost a mirror image of profile #2. Using the theory of section 4.5, on the occurrence of the maximum velocity activity, profile #2 was chosen to be the data profile with profile #1 being the practice profile. A further advantage of profile #2, and similar profiles, is that at the end of the run when all the sinusoids are converging to zero amplitude, the

disturbance is low. This means that the HO's control input will be low so there should not be a bias in the joystick data that could influence the FFT data reduction. This should result in cleaner data than with profile #1, and is a more valid reason for choosing profile #2 to be the data profile.

5.2) The Formal Test Procedure

The same basic test procedure used for population tests with the last profile was used for these population tests. However, a major modification was made in the instruction given to the subject. The basic test procedure has been described in section 4.5, but its main features will be repeated here.

The enclosed head restraint and joystick controller are used for all testing. The instruction given to the subjects is to maintain zero velocity, or keep their motion stopped, by REACTING to the disturbance. To do this subjects must sense their motion and respond with a joystick input. The clearest sense of motion seems to occur during acceleration changes and this should be the primary motion cue to the subject. For the cleanest data subjects should REACT to this acceleration. That is, they should clearly sense their acceleration before responding with a control input. This input should then not be changed until the subjects sense their acceleration changing again. If no acceleration is sensed the joystick should be moved to the zero input position. Above all, subjects should not try to search out the disturbance by inputting high frequency control. This is a tendency noted with many subjects and should be avoided as it leads to poor data. This explanation should be told to the subjects and their control technique

INPUT PARAMETERS

TRUN= 81.920
 NSINES= 12
 FFREQ= 0.012210
 FLAG= 0
 DEL= 103.000
 FILTER= 1
 FFOLE= 0.150
 FPOS= 2.15
 FACC= 0.200
 TLOGF= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.117
 VELMAX= 0.6285
 % USAGE OF TRACK= 100.00
 STARTING POSITION= 0.158
 SCALE= 0.2551

FREQ. (HZ)	AMP1 (M/S)	AMP2 (G)	PHASE (DEG)
0.0610	0.13	0.005	2.
0.0854	0.13	0.007	106.
0.1099	0.12	0.008	210.
0.1343	0.11	0.010	314.
0.1587	0.12	0.012	57.
0.2075	0.11	0.015	162.
0.2319	0.10	0.014	266.
0.2808	0.11	0.020	11.
0.3540	0.09	0.021	116.
0.3906	0.07	0.019	220.
0.4517	0.07	0.021	325.
0.5005	0.06	0.020	70.

HISTOGRAM DATA

ACC BIN= 0.005	# ACC POINTS= 815
ACC BIN= 0.010	# ACC POINTS= 1034
ACC BIN= 0.015	# ACC POINTS= 1095
ACC BIN= 0.020	# ACC POINTS= 848
ACC BIN= 0.025	# ACC POINTS= 698
ACC BIN= 0.030	# ACC POINTS= 555
ACC BIN= 0.035	# ACC POINTS= 509
ACC BIN= 0.040	# ACC POINTS= 463
ACC BIN= 0.045	# ACC POINTS= 344
ACC BIN= 0.050	# ACC POINTS= 344
ACC BIN= 0.055	# ACC POINTS= 307
ACC BIN= 0.060	# ACC POINTS= 193
ACC BIN= 0.065	# ACC POINTS= 217
ACC BIN= 0.070	# ACC POINTS= 173
ACC BIN= 0.075	# ACC POINTS= 177
ACC BIN= 0.080	# ACC POINTS= 114
ACC BIN= 0.085	# ACC POINTS= 107
ACC BIN= 0.090	# ACC POINTS= 31
ACC BIN= 0.095	# ACC POINTS= 25
ACC BIN= 0.100	# ACC POINTS= 28
ACC BIN= 0.105	# ACC POINTS= 46
ACC BIN= 0.110	# ACC POINTS= 37
ACC BIN= 0.115	# ACC POINTS= 14
ACC BIN= 0.120	# ACC POINTS= 18
ACC BIN= 0.125	# ACC POINTS= 0
ACC BIN= 0.130	# ACC POINTS= 0
ACC BIN= 0.135	# ACC POINTS= 0
ACC BIN= 0.140	# ACC POINTS= 0
ACC BIN= 0.145	# ACC POINTS= 0
ACC BIN= 0.150	# ACC POINTS= 0

Figure 5.1.01 H1PRO5.PRO Profile #2 Parameters

INPUT PARAMETERS

TRUM 81.920
 NSINES 12
 FFFREQ 0.012210
 FLAT 0
 DEL 203.000
 FILTER 1
 FPOLE 0.160
 FENS 2.05
 FACC 0.200
 TLUMP 0.010

PROFILE DESCRIPTION

MAX ACCEL 0.113
 VELMAX 0.6290
 % LINEAR OF TRACK 100.00
 STARTING POSITION -0.236
 SCALE 0.2502

FREQ. (HZ)	AMP1	(M/S)	AMP2 (G)	PHASE (DEG)
0.0610		0.12	0.005	0.
0.0854		0.13	0.007	253.
0.1099		0.12	0.008	147.
0.1343		0.11	0.009	40.
0.1587		0.12	0.012	293.
0.2075		0.11	0.015	186.
0.2319		0.10	0.014	79.
0.2808		0.11	0.019	332.
0.3540		0.09	0.021	226.
0.3906		0.07	0.019	119.
0.4517		0.07	0.021	12.
0.5005		0.06	0.020	265.

HISTOGRAM DATA

ACC BIN= 0.005	# ACC POINTS=	862
ACC BIN= 0.010	# ACC POINTS=	1059
ACC BIN= 0.015	# ACC POINTS=	1048
ACC BIN= 0.020	# ACC POINTS=	908
ACC BIN= 0.025	# ACC POINTS=	664
ACC BIN= 0.030	# ACC POINTS=	551
ACC BIN= 0.035	# ACC POINTS=	559
ACC BIN= 0.040	# ACC POINTS=	426
ACC BIN= 0.045	# ACC POINTS=	355
ACC BIN= 0.050	# ACC POINTS=	342
ACC BIN= 0.055	# ACC POINTS=	313
ACC BIN= 0.060	# ACC POINTS=	210
ACC BIN= 0.065	# ACC POINTS=	207
ACC BIN= 0.070	# ACC POINTS=	175
ACC BIN= 0.075	# ACC POINTS=	180
ACC BIN= 0.080	# ACC POINTS=	126
ACC BIN= 0.085	# ACC POINTS=	83
ACC BIN= 0.090	# ACC POINTS=	41
ACC BIN= 0.095	# ACC POINTS=	26
ACC BIN= 0.100	# ACC POINTS=	30
ACC BIN= 0.105	# ACC POINTS=	49
ACC BIN= 0.110	# ACC POINTS=	37
ACC BIN= 0.115	# ACC POINTS=	21
ACC BIN= 0.120	# ACC POINTS=	0
ACC BIN= 0.125	# ACC POINTS=	0
ACC BIN= 0.130	# ACC POINTS=	0
ACC BIN= 0.135	# ACC POINTS=	0
ACC BIN= 0.140	# ACC POINTS=	0
ACC BIN= 0.145	# ACC POINTS=	0
ACC BIN= 0.150	# ACC POINTS=	0

Figure 5.1.02 H1PR05.PRO Profile #1 Parameters

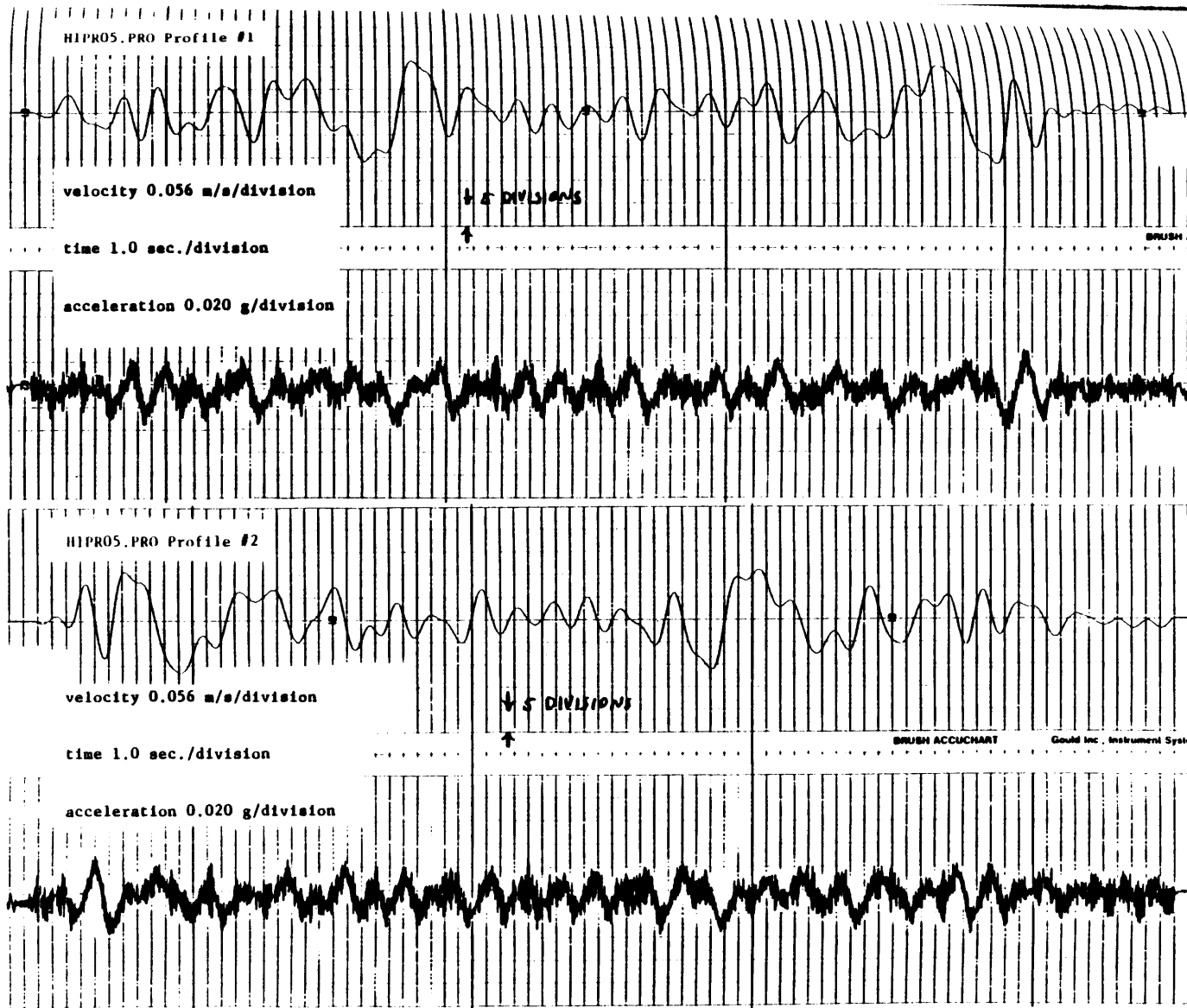


Figure 5.1.03 H1PRO5.PRO Profile #1 and #2 Time Histories

closely monitored. The explanation to REACT to the disturbance should be repeated if necessary.

Once this instruction is given subjects are given a ride through the practice profile with the joystick disabled. This is done to familiarize them with the motion disturbance. The joystick is then enabled and the practice profile run until subjects reach their maximum performance level or start completing runs. Subjects are then given a ride through the data profile without control. The data profile is then run with subject control until 4 or 5 runs are completed. This completes the test session. Data is stored for data runs only, but all runs should be logged on the run data sheets.

The data analysis is started by determining frequency spectrum plots for all completed runs. The velocity RMS ratio and the joystick RMS should also be calculated to provide additional information to that contained in the plots. The three best runs are then chosen from these plots and RMS data and used to determine the Bode plots. To choose the best runs the following criteria should be used as a guide. The joystick remnant should be low, below or about 10 units. The joystick remnant should be 25-30% or less of the joystick amplitudes at the disturbance frequencies. The velocity remnant with control should be roughly that of the no subject velocity remnant. The velocity disturbance frequency amplitudes with control should be less than the amplitude with no subject. This will vary depending upon the capability of the subject to perform the velocity nulling task so no definite value can be stated. All curves should be smooth with no erratic oscillations. The RMS data should be consistent for the three chosen runs, but this is of secondary importance. If the frequency plots and RMS data do

not conform to the above criteria, the data should not be used in further analysis.

A formal test procedure checklist is shown in Appendix D. This lists all the important steps required for acceptable results. It covers the entire spectrum from setting up the Sled system to plotting data. This checklist should be used in all future testing.

It should be emphasized that a very important part of the procedure is the instruction to the subject. It should be made clear to the subject to try to REACT to the disturbance. This should be discussed with subject/test conductor dialogue during the practice runs. Practice is also important and data runs should not be made until all involved are sure that no improvement in subject performance can occur. These are the two critical subjective elements of the test procedure that must be monitored closely.

It is expected that some subjects will give desirable results with little variance, while others will not. It is felt that most of the astronaut participants, due to their skill at operating complex man-machine systems, will yield desirable results. However, subjects who have difficulty in performing the task should not be used in the final analysis.

CHAPTER 6
RESULTS AND DISCUSSION

6.1) Results

Five subjects were tested with the procedure described in Chapter 5. However, in all but the first test an unwanted vibration hampered the subjects performance. This vibration occurred at the low velocity portion of velocity zero-crossings. It was of such a magnitude and frequency that it tended to mask the change in acceleration. As this change in acceleration is the major cue to the subject, it was difficult for the subjects to determine the proper control input. Either much practice was needed for the subject to sense the acceleration changes more readily, or the low velocity acceleration was ignored since it could not be sensed clearly. Although this vibration hampered the testing, it is felt that the major goals of this work have been accomplished.

The plotted results of the tests are shown in Appendix E, which also contains an explanation of the plot formats. Two typical plots are shown in Fig. 6.1.01,.02. RMS data is shown in Table 6.1.01. As is seen the data shows little scatter for most subjects. Also, the one sigma deviations are as low or lower than those of other human subject experiments. The general trends seen in previous testing are also shown. There is a flatness or peak of gain and phase at low frequencies, and a drop-off at high frequencies. Individual plots show differences which is desired. Table 6.1.01 shows the RMS data for all subjects. The velocity ratios tend to show the relative level of velocity nulling performance. For most of the subjects tested the velocity and joystick remnants were low, and the data smooth. Under these

conditions, the velocity RMS ratio can indicate general performance levels. It is noted that there is some inconsistency among individual subjects, however. The joystick RMS shows the level of joystick activity, as always.

Fig. 6.1.03 shows a Bode plot of the average values of the four subjects with acceptable data. (Subject LR did not have valid data as will be explained in the next section.) The consistency between the subjects is encouraging. An acceptable range of values can encompass all of the valid runs. Fig. 6.1.04 shows the average values of all the significant tests for subject DH. The consistency of this data is also encouraging.

6.2) Discussion of Individual Subject Results

The results of subject MS are shown in Fig. E6.1.01,.08-.11. This was the first subject tested with the final procedure. As noted in Chapter 4 subject MS had done poorly in the previous testing because he used the control injection technique. With this testing, and the instruction to REACT to the disturbance his results have greatly improved. The frequency spectrum plots show a clear separation between the joystick disturbance frequency amplitudes and the remnant, with the remnant at 30-40% of the disturbance. The velocity nulling has been somewhat effective, being 50-90% of the velocity disturbance. Runs 07, 08, and 09 yielded the best joystick disturbance frequency amplitude/remnant separation and the best velocity nulling. Run 02 was not included due to the high velocity remnant. The RMS data is also consistent for these three runs and improving. These three runs were used to generate the Bode plot. The plot shows acceptable deviations and scatter in $\log(\text{GAIN})$, 0.0-0.25 log units,

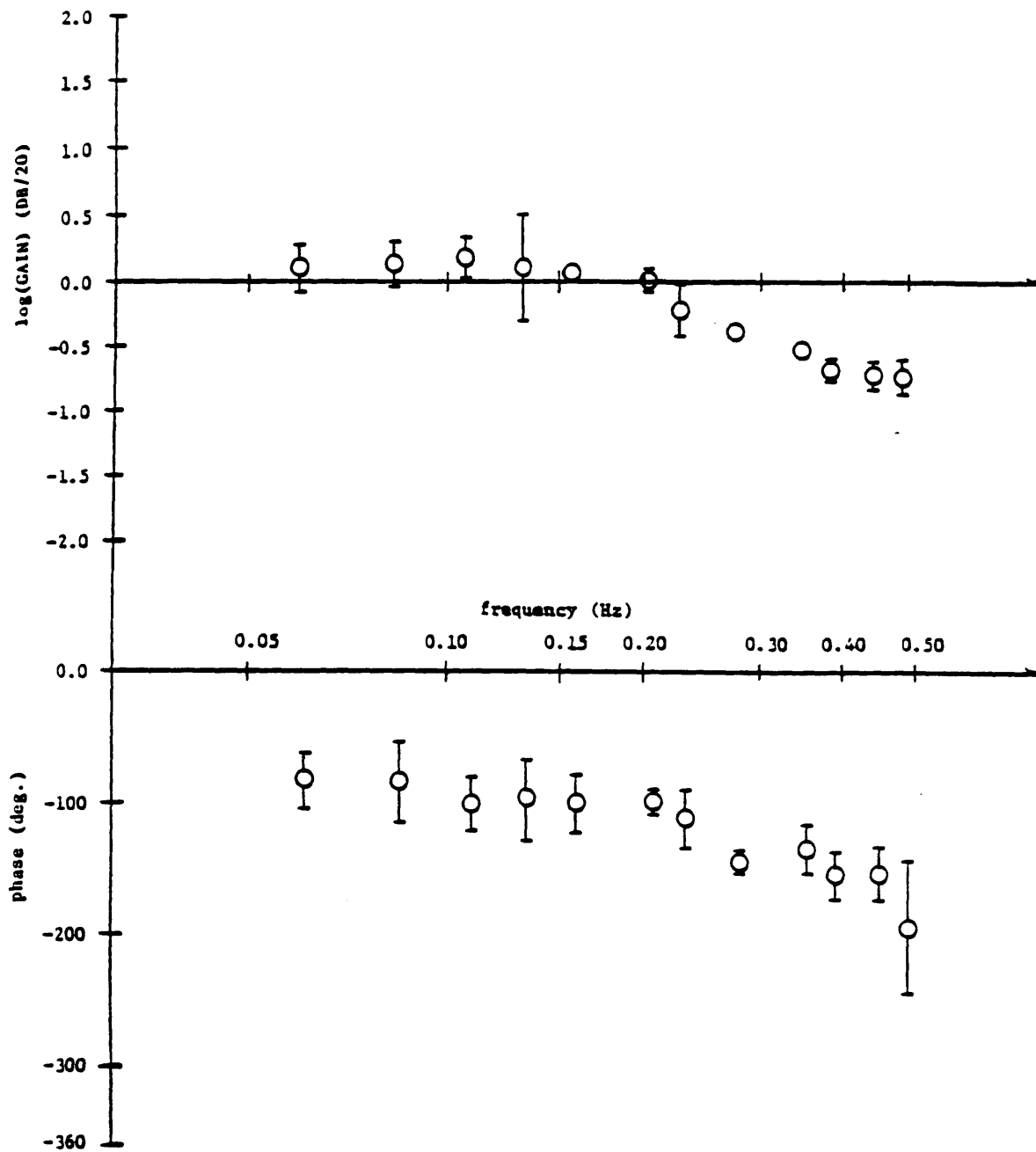


Figure 6.1.01 Bode Plot. Subject DH, Final Profile
Average of Runs 15,21,22

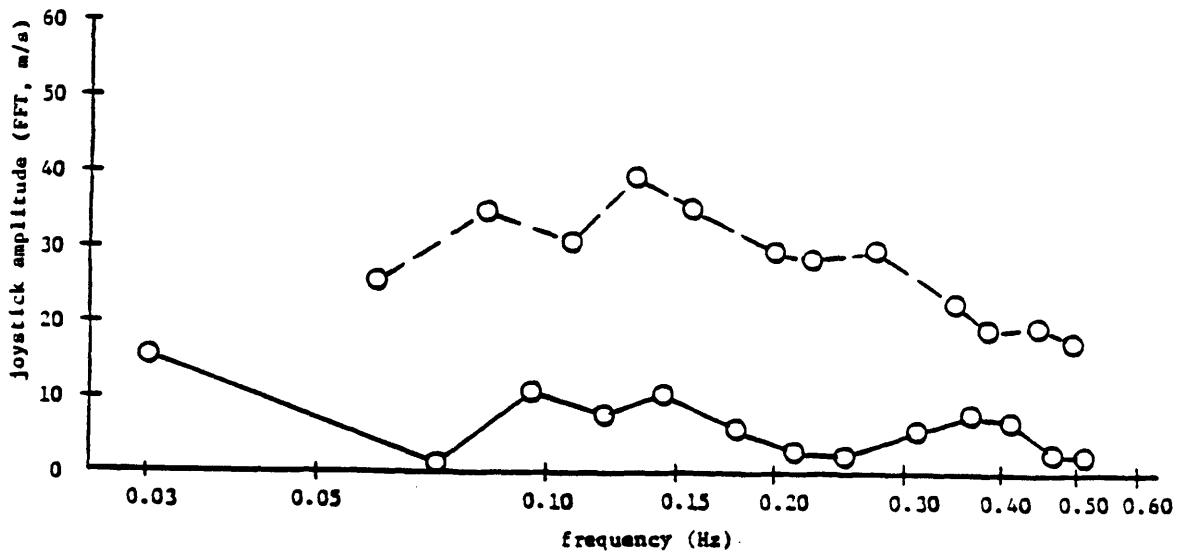
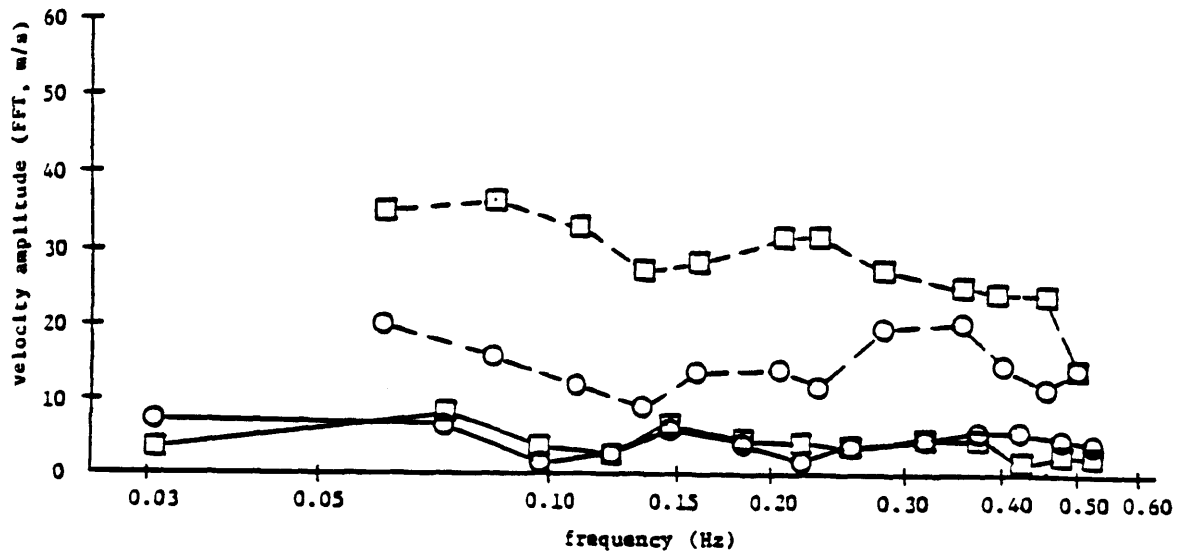


Figure 6.1.02 Frequency Spectrum. Subject DH, Run 15, Final Profile

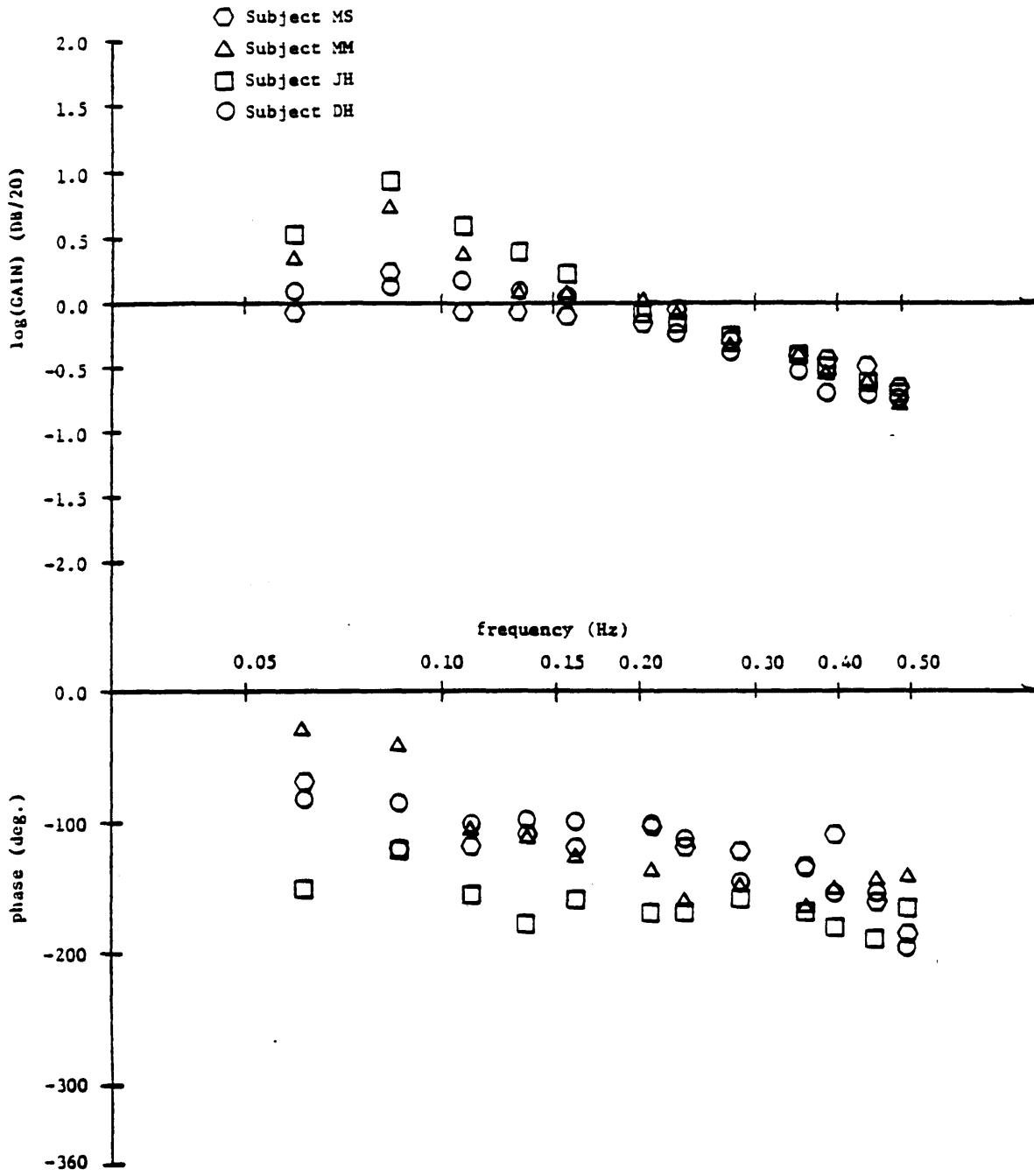


Figure 6.1.03 Summary Bode Plot, All Subjects, Final Profile

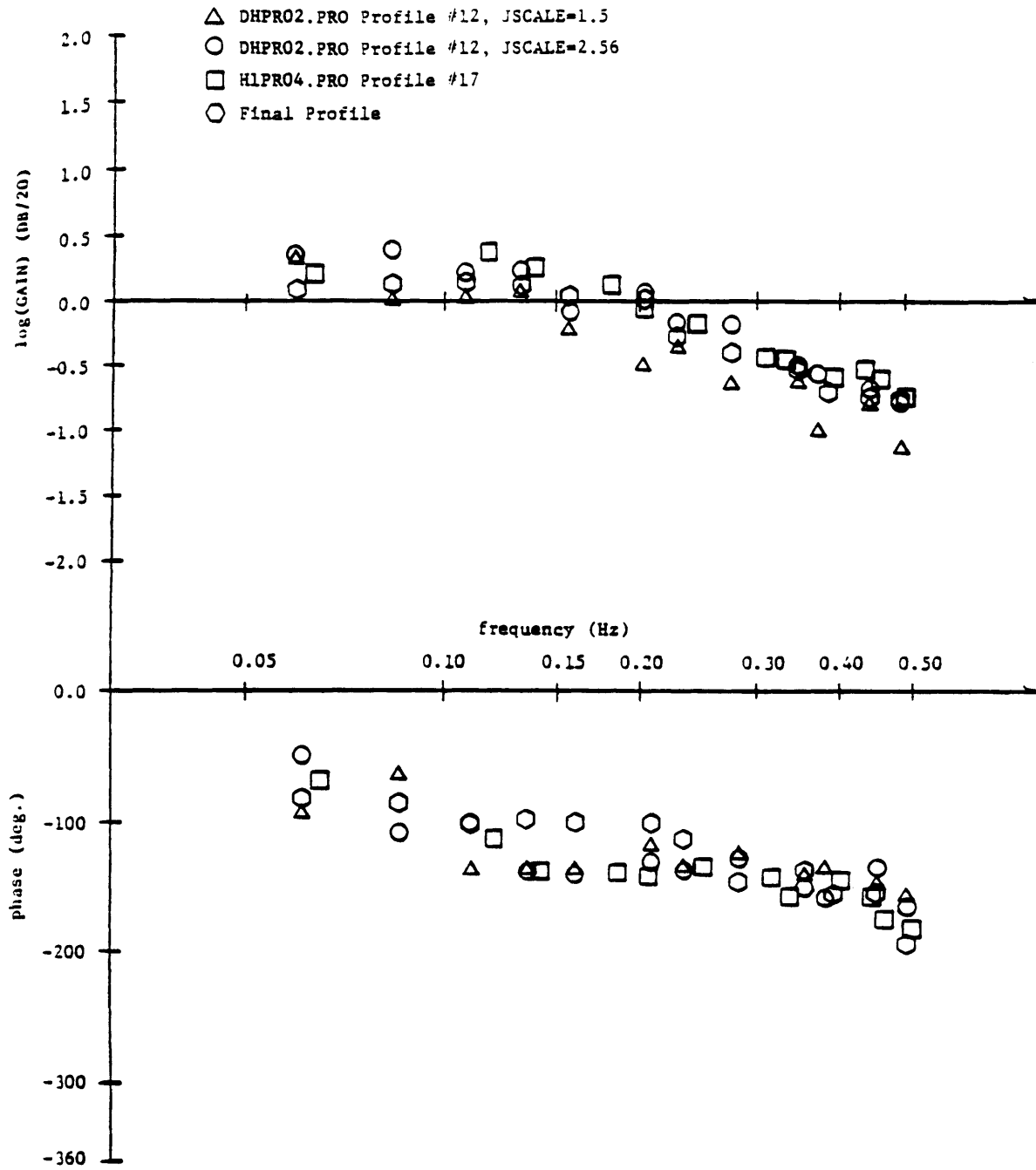


Figure 6.1.04 Summary Bode Plot, Subject DH

SUBJECT	DATA FILE	RMS VELOCITY RAT IO	RMS JOYST ICK
MS	02	2.33*	0.206
	07	1.56	0.210
	08	1.29	0.216
	09	1.11	0.189
LR	02	3.51	0.284
	03	2.47	0.304
	06	1.24	0.284
JH	03	0.885*	0.242
	06	0.627	0.253
	08	0.979	0.245
	09	0.720	0.251
DH	12	1.39*	0.190
	15	0.870	0.170
	21	0.699	0.175
	22	0.735	0.195
	23	0.777*	0.160
MM	03	0.697	0.215
	09	0.702	0.206
	12	0.812	0.213
	15	0.716*	0.212
	18	0.759*	0.205

* Not used for Bode plots.

TABLE 6.1.01 RMS Data. H1PRO5.PRO Profile #2.

and phase, 0.0-50.0 deg. There is also an approximately zero gain at low frequencies, and a break at about 0.2 Hz. These are the characteristics of this subject which will be quantified further in section 6.3.

subject MM was tested twice with this procedure. The run completion rate was 100% during the first testing. Unfortunately the data was not stored. The vibration was also not prevalent during this test. This gives strong support to the fulfillment of the test requirements, however.

Subject MM had always given acceptable results in past testing and had never used the control injection technique. The overall high quality of his data is readily seen. (Fig. E6.1.02,.12-.16) Runs 03,09, and 12 were used to generate the Bode plot. As is seen, there is a higher gain at low frequencies than for subject MS. This is also shown in the frequency spectrum plots which show velocity nulling at 50% of the disturbance values. The nulling is not effective at the high frequencies, being 70-100% of the disturbance values. This is shown as a drop-off in gain at the higher frequencies. The phase data also follows the same trends as the gain data. The data scatter and deviations are quite acceptable, at roughly 0.3 log(GAIN) and 50 deg. phase. The RMS data shows roughly the same level of joystick activity, but much lower velocity activity than subject MS. This is as suggested by the plots.

Subject JH had a large amount of experience as a sled subject in the experiments of Ref. 8. The data shows a joystick disturbance frequency amplitude/remnant separation that is the highest seen at low frequencies. (Fig. E6.1.03,.17-.20) The velocity remnant is low, approximating the no-subject remnant, and the nulling is the best seen at low frequencies, being

pushed down into the remnant at about 25% of the disturbance. This is reflected in the Bode plot which shows the highest gain seen at low frequencies. The nulling is poor at high frequencies, 80-100%, which seems to be a general characteristic of the HO. The phase data is also the lowest seen at low frequencies. The joystick RMS shows the next to the highest values, reflecting the high joystick disturbance frequency amplitudes. The RMS velocity ratios are some of the lowest seen, as expected.

Subject DH had the most experience with the type of testing of this work. The data shows average joystick disturbance frequency amplitude/remnant separation, 25-30% of the disturbance, but there is some fluctuation in the disturbance frequency amplitude values. (Fig. E6.1.04,.21-.25) The joystick remnant is the lowest observed being at or below 10 units. The velocity nulling is effective through the mid-frequency range, being about 60-100% of the disturbance. This results in the flatness in gain and phase through this same frequency range seen in the Bode plot. The RMS data shows the lowest joystick values, which is due to the low remnant. The velocity ratios are also some of the lowest seen. The experience of the subject was felt to have resulted in the lowest remnants observed. The experience, then, has caused the subject to be able to sense and respond only to the disturbance, while clearly showing his specific level of performance. This is an encouraging result and shows that practice leads to better data and not to the best velocity nulling performance.

The DH summary Bode plot shows some encouraging results. (Fig. 6.1.04,E6.1.07) All the individual tests show acceptable agreement except for DHPR02.PRO profile #12, JSCALE=1.5 and the phase of H1PR05.PRO profile #2. It is noted that all tests used a P=90 or 95% JSCALE criterion except

for the profile #12 JSCALE=1.5 test. Since the joystick voltage is stored before it is scaled the JSCALE directly effects the magnitude of the $\log(\text{GAIN})$. The JSCALE=1.5 data was obtained from the same profile as the DHPR02.PRO profile #12, JSCALE=2.56 data. If a correction of $\log(2.56/1.5)=0.23$ is applied to the JSCALE=1.5 data it would then agree with all the other data. The JSCALE=1.5 data does have the most scatter, however, which was caused by the joystick gain being too sensitive. The H1PR05.PRO profile #2 data shows a difference in low frequency phase, and the lowest $\log(\text{GAIN})$ of all tests. It is suspected that this was mainly caused by the cart vibration problems as noted in Chapter 5. The overall agreement of the different tests is still acceptable, though, which supports the validity of the closed-loop test.

Subject LR was the only female and non-graduate student tested. She also had the least experience with the Sled, as this was her second test session. Only three runs were completed during her testing. The plots show high remnant activity of joystick and velocity. (Fig. E6.1.05,.26-.28) The control is only effective at the low frequencies. The Bode plots show the general trends, but there is much scatter, particularly at the high frequencies, and some larger than desirable deviations, 80 deg. in phase. The RMS data shows the highest joystick and velocity activity, which is suggested by the high remnant. As will be shown in section 6.4, the joystick disturbance frequency amplitude/remnant separation is not sufficient to yield valid results. Thusly, the Bode data is not included in the plot of Fig. 6.1.03,E6.1.06. The data is shown as an illustration only. The data does show steady improvement, though, so in future test sessions better data may be obtained.

6.3) The Transfer Function Model

A transfer function model is desired so that the characteristics seen in the HO responses can be quantified. A BMDP non-linear regression analysis program was used to determine the model. (Ref. 16) It was originally desired to fit a model with two real poles, as in the Young and Meiry model. However, the BMDP program contains numerical problems which does not allow this restriction. Rather, a 2nd order system equation must be used as this allows the poles to become complex. Various structures were tried in attempts to define a structure that would be sufficiently general to include all the variations seen in the HO responses, while being as simple as possible. The final structure used was as follows:

$$\frac{K(j\omega + a)}{(j\omega + b)(j\omega^2 + 2\zeta\omega_n j\omega + \omega_n^2)}$$

The following limits were placed on the parameters:

parameter	min	max
K	0.0	100.0
a	0.0	25.0
b	0.0	25.0
ζ	0.0	10.0
ω_n	0.0	100.0

The HO response data was converted to GAIN (amplitude ratio) and frequency (rad/sec) units before inputting them to the regression analysis. Only the GAIN data was used in the fit. No phase data was used in the analysis.

Fig. 6.3.01-.04, E6.3.01-.04 show the models and their curves plotted against the Bode plots. As is seen the gain fits are very precise but the

corresponding phase fits are poor. For all but subject DH a shift of about 50 deg. would be needed for the phase curve to conform to the data. Also, more phase adjustment is needed at the low frequencies than at the high frequencies. No other simple phase adjustment, without a corresponding gain adjustment, would yield a better fit. The data therefore does not conform to a minimum phase system with this model.

On all the models the zero has been placed at the limit value of 0.0, resulting in a differentiator. This is the lead term as suggested in the Young and Meiry model by the zero at 0.0076 rad/sec. That the zero is 0.0 reflects the accuracy of the regression analysis and also the problems of trying to define the system over a narrow range of frequencies.

The pole shows some variation and is the main factor in determining the individual structure for each subject. In two cases the pole has been placed at the limit of 25 rad/sec. This suggests that for these cases the pole is not needed. This is seen in the high frequency slopes which are approximately -1.0 ($\log(\text{GAIN})/\log(\text{freq.}, \text{rad/sec})$). The large pole has also caused the K, gain, values of these cases to be larger than necessary. In the range of the disturbance frequencies, 0.383-3.145 rad/sec, the magnitude of this pole remains at approximately 25.0. The K values should therefore be lowered by a factor of 25.0. The corrected values are shown in brackets.

For subject DH the pole has been placed at 0.0 resulting in an integrator. This implies that for this case neither the pole nor zero are needed. This is due to the flatness in gain at the low frequencies and the slope of approximately -2.0 ($\log(\text{GAIN})/\log(\text{freq.}, \text{rad/sec})$) at the high

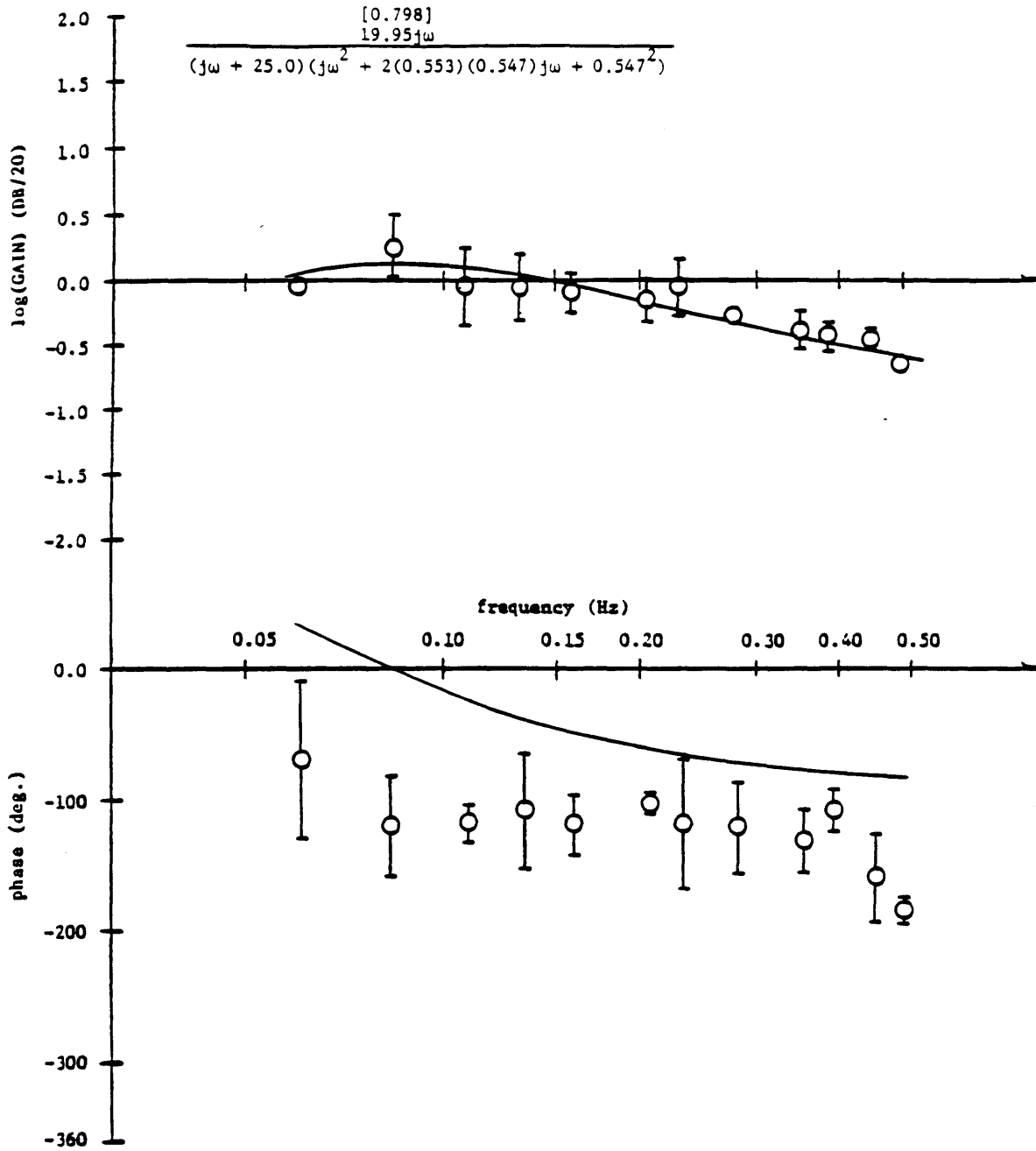


Figure 6.3.01 Model and Curve Fit. Subject MS

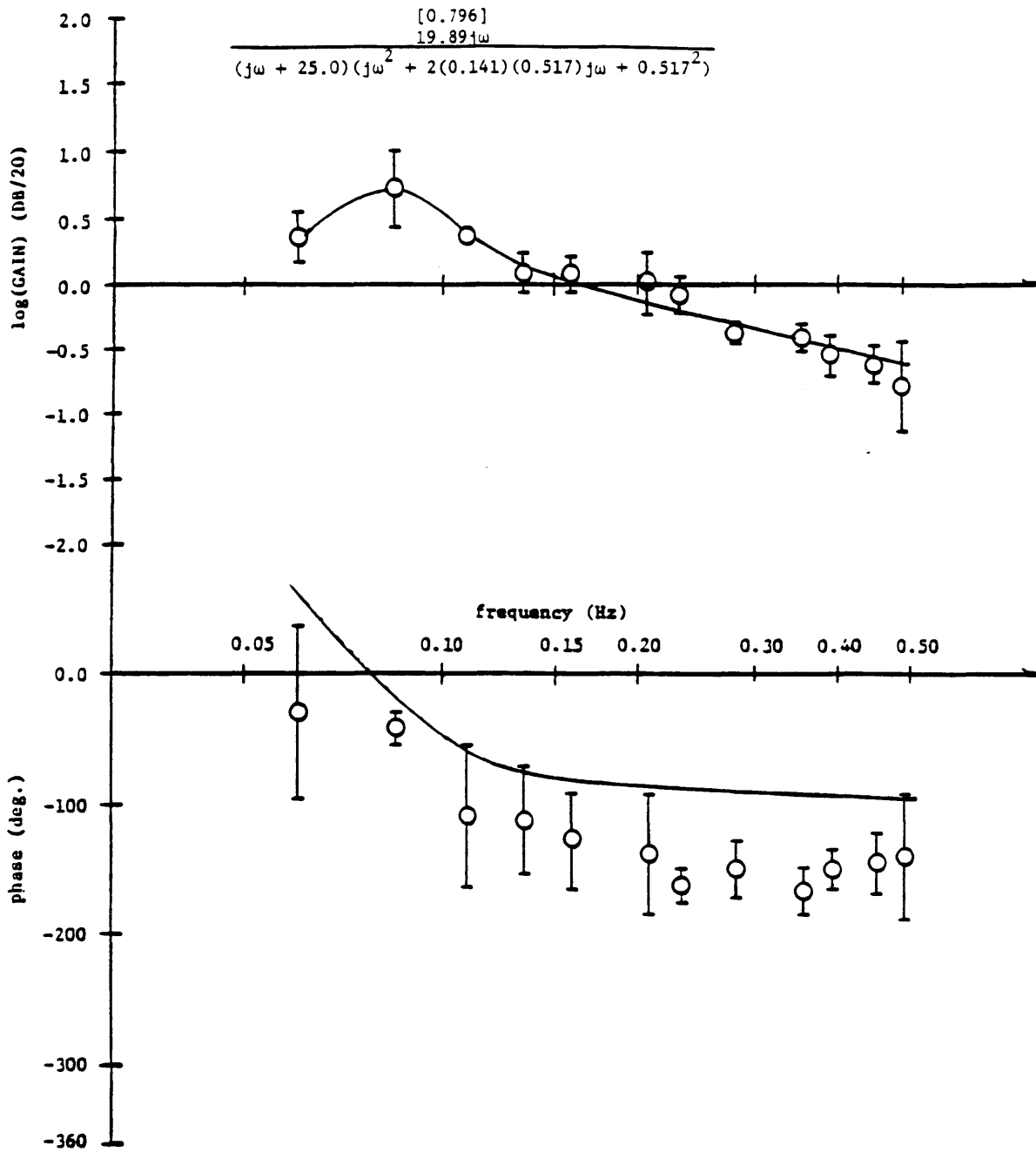


Figure 6.3.02 Model and Curve Fit. Subject MM

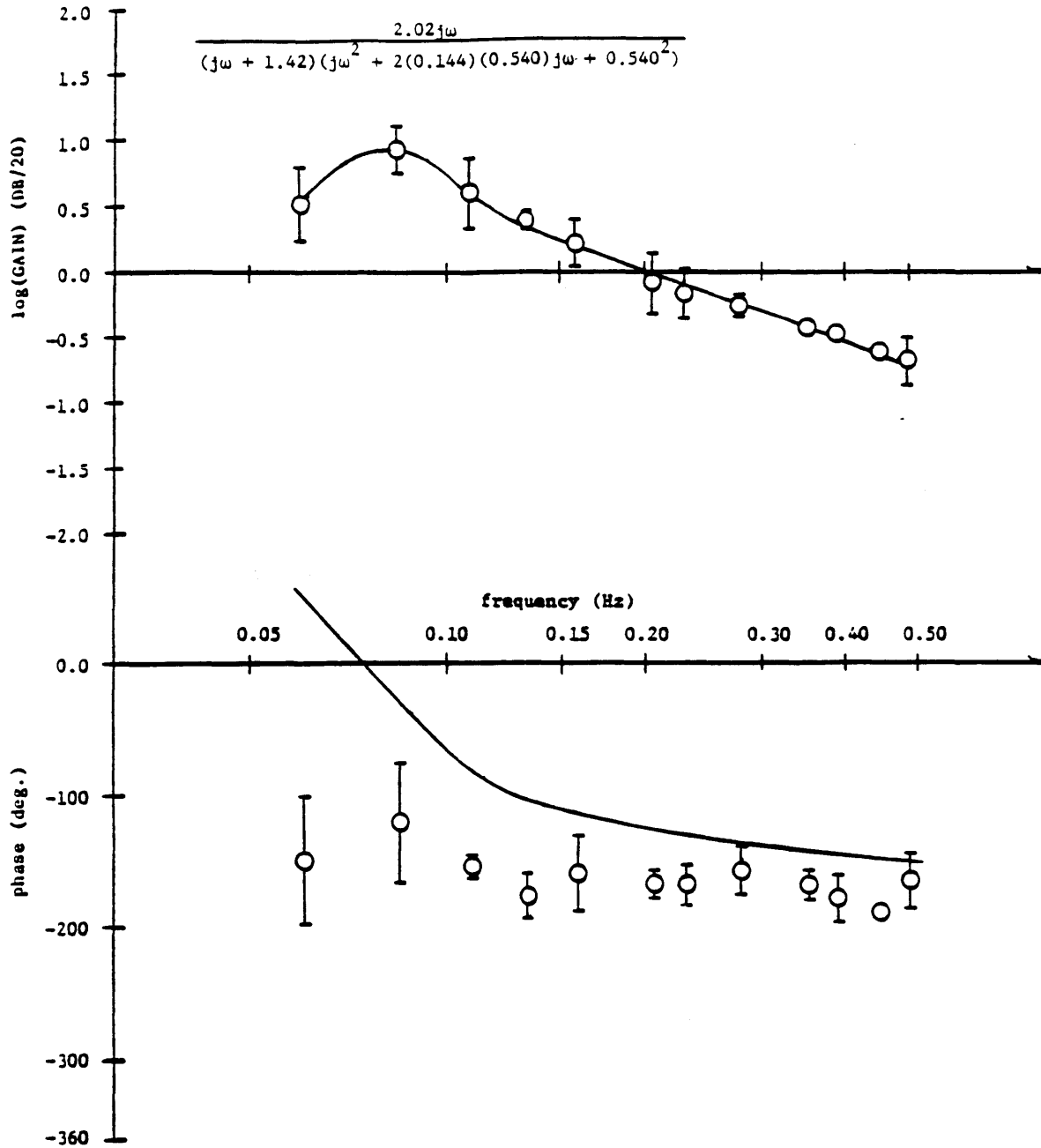


Figure 6.3.03 Model and Curve Fit. Subject JH

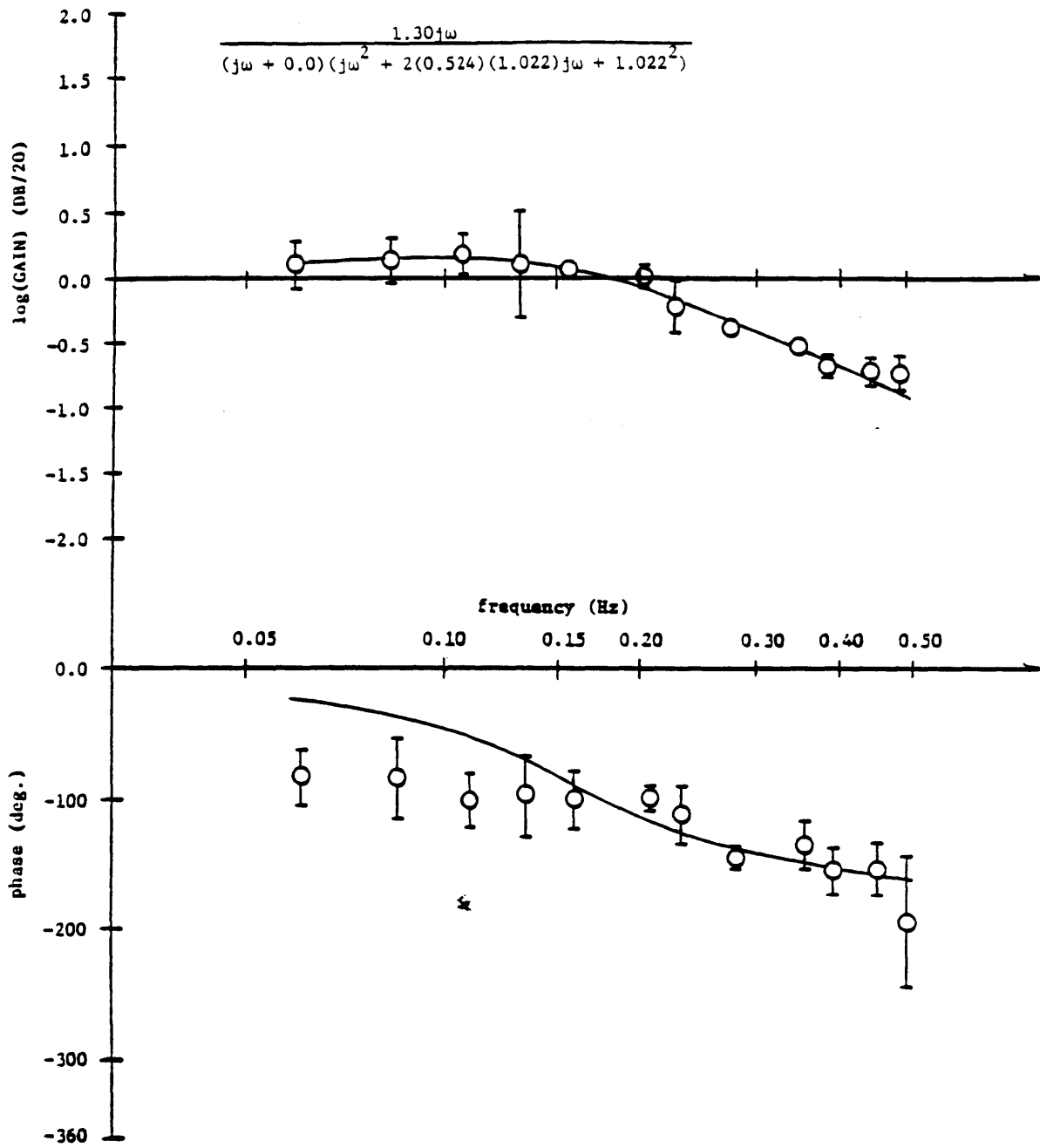


Figure 6.3.04 Model and Curve Fit. Subject DH

frequencies. The resulting model would then be a 2nd order system. In the case for subject JH, the pole has been precisely placed at 1.42 rad/sec. This accurately shows that the high frequency slope is between -1.0 and -2.0 ($\log(\text{GAIN})/\log(\text{freq.}, \text{rad/sec})$).

The values of most importance, in terms of reflecting the task performance, are K , ζ , and ω_n . The ω_n precisely defines the break frequency of the response. For cases where the low frequency $\log(\text{GAIN})$ shows no peak, $\zeta=0.5$ or less, the ω_n is the break frequency below which the velocity nulling has been most effective. (subject DH) The model for subject MS emphasizes the low peak too much and does not follow this trend, as it seems it should from looking at the plot. This may show the limitations of the regression analysis. For cases where ζ is much lower than 0.5, there is a resonance peak. For these cases the plots show that the velocity nulling has been very effective, roughly 25-30% of the disturbance in the region of ω_n . (subjects MM, JH)

There does not seem to be a correlation between all K values and the task performance. Subject MM had much better velocity nulling than subject MS. Their K values are the same but their ζ values and low frequency responses are different. However, for cases with similar ζ values, yielding similar low frequency responses, K corresponds to the level of velocity nulling performance. In this way K can be used to determine varying performance levels. (Compare subjects DH and MS, and subjects MM and JH.) In summary, the zero defines the low frequency response. The K , ζ , and ω_n define the task performance in the frequency range of the disturbance. And the pole shows modifications needed for the high frequency response at the higher

frequencies of the disturbance.

6.4) Remnant Analysis

No remnant correction has been applied to the data in this work. However, a criteria can be used to determine if the joystick remnant is too large to yield valid results. The full development of this criteria is found in Ref. 7. Only the final result will be shown here.

The system under consideration is shown in Fig. 6.4.01 with the joystick remnant shown. The required parameters are also labeled. The analysis of Ref. 7 results in a function based on the power spectrum of these parameters. This function is:

$$|\alpha| = \{[\phi_{\lambda\lambda}/\phi_{AA} - \widehat{\phi}_{\lambda\lambda}/\phi_{AA}]/[1 - |G_c|^2(\widehat{\phi}_{\lambda\lambda}/\phi_{AA})]\}^{1/2}$$

where; $\phi_{\lambda\lambda}$ is the power of the joystick output, including the remnant: $\widehat{\phi}_{\lambda\lambda}$ is the power of the joystick remnant at the disturbance frequencies, found by linear interpolation: ϕ_{AA} is the power of the disturbance acceleration: $|G_c|^2$ is the power of the cart dynamics taken from the plot of Chapter 3.

The limiting behavior of this $|\alpha|$ function is based on the remnant power:

$$\begin{aligned} |\alpha| &\Rightarrow (\phi_{\lambda\lambda}/\phi_{AA})^{1/2} & G_{HO} & \text{as} & \widehat{\phi}_{\lambda\lambda} &\Rightarrow 0.0 \\ |\alpha| &\Rightarrow 1.0/G_c = \text{inverse plant dynamics} & & \text{as} & \widehat{\phi}_{\lambda\lambda} &= \infty \end{aligned}$$

This says that for the transfer function of the HO as defined in this work to be valid, the joystick remnant must be low. How low is not suggested, so a typical test run was used to find specific values for the above

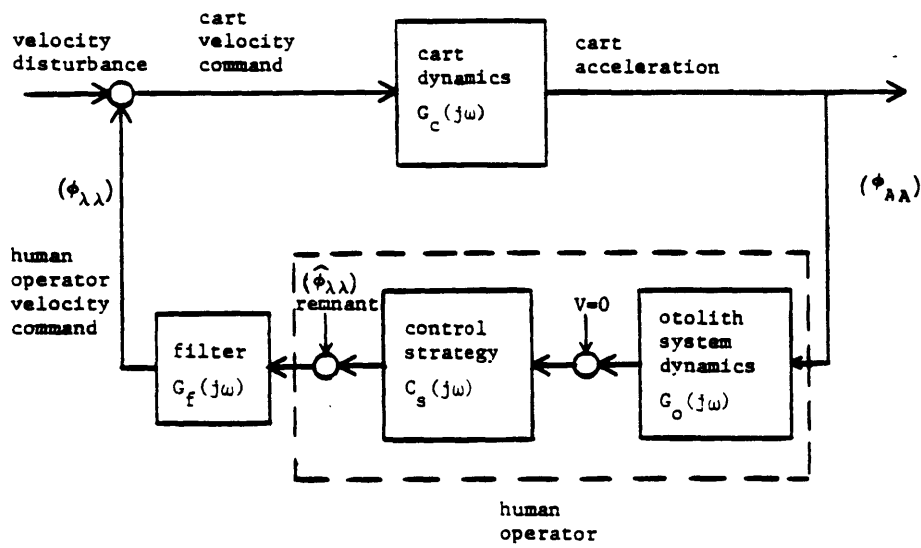


Figure 6.4.01 The Closed-Loop System with Remnant Analysis Parameters Labeled

equations. The data from subject DH, H1PR05.PRO profile #2, run 22 was used as it showed a middle range of joystick disturbance frequency amplitude/remnant separation. The data is shown in Table 6.4.01.

The table shows the transfer function GAIN as well as $|\alpha|$ and the other parameters needed in the calculation. As is seen, for all but two points there is no major difference between $|\alpha|$ and the GAIN. The two points noted are seen to have a joystick remnant value of about 80% of the joystick amplitudes at the disturbance frequencies. For all other points the joystick remnant is 25-30% of the joystick disturbance frequency amplitudes. The resulting $|\alpha|$ and GAIN values have a maximum difference of 12%. A 12% error in the GAIN data of this type of experiment is considered quite acceptable. Therefore the joystick remnant of 25-30% of the disturbance frequency amplitudes can be considered as a guideline to determining the quality of the data. All of the results obtained in this work have joystick disturbance frequency amplitude/remnant ratios in this range, except those of subject LR, which confirms the validity of the data.

In the work of Ref. 10 any point with a high joystick remnant was discarded. In this work, the velocity and joystick frequency spectra were checked and if the remnants were low for most frequencies the entire run was used. This may have introduced some unwanted error at a few frequencies. The overall quality of the data appears to be acceptable, however. Possibly, the above remnant analysis should be done for all runs and bad points discarded in the future.

As described previously, if the joystick remnant is high the HO transfer function should approximate the inverse plant dynamics. However, if the HO

FREQ. (Hz)	JOYSTICK REMNANT	JOYSTICK AMPLITUDE	ACC. AMPLITUDE	G	α	GAIN	%
0.061	13.69	37.49	24.86	0.575	1.48	1.51	1.89
0.085	7.15	25.65	25.77	0.390	0.961	0.995	3.54
0.110	5.14	43.84	21.15	0.637	2.08	2.07	-0.53
0.134	9.81	18.52	32.50	1.07	0.511	0.570	11.55
0.159	12.14	38.50	33.26	1.13	1.21	1.16	4.14
0.207	11.99	40.59	32.38	1.28	1.36	1.25	-7.93
0.232	11.95	25.79	31.11	1.78	1.01	0.829	17.59*
0.281	8.43	30.59	64.89	1.96	0.469	0.471	0.43
0.354	12.13	29.54	86.10	2.36	0.332	0.343	3.31
0.391	12.96	19.15	104.47	3.19	0.147	0.183	24.49*
0.452	6.73	24.83	114.89	3.30	0.212	0.216	1.89
0.500	6.61	15.82	116.64	3.29	0.126	0.136	7.94

* Out of tolerance points.

TABLE 6.4.01 Remnant Analysis. Subject DH, H1PRO5.PRO
Profile #2, Run 22.

transfer function approximates the inverse plant dynamics it can mean effective velocity nulling. The closed-loop transfer function of the system, neglecting the filter, is:

$$G_{cl} = \frac{-G_{HO}G_{cl}}{1.0 - G_c G_{HO}} \quad \frac{\text{Acceleration}}{\text{Disturbance}}$$

If the HO matches the inverse plant dynamics exactly, except for the opposite sign, the result is:

$$G_{cl} = 1/2 \Rightarrow \text{Acceleration} = 1/2 \text{ Disturbance}$$

The HO has therefore effectively nulled 50% of the disturbance acceleration or approximately 50% of the disturbance velocity. Since the HO is very adaptable, it is not unlikely that his control strategy would cause him to approximate the inverse cart dynamics. However, if the remnant is high and the velocity nulling is not effective, the remnant analysis also shows that the HO transfer function should approach the inverse cart dynamics. In short, with this type of testing it is difficult to separate the HO transfer function from the inverse plant dynamics. It is clear, however, that when the joystick remnant is low, the data is accurate. The remnant analysis, then, is the most important criteria for determining the data quality.

CHAPTER 7

CONCLUSIONS AND RECOMMENDATIONS

7.1) Conclusions

The requirements of the test procedure have been fulfilled by the procedure developed in this work. It is felt that the test procedure is the best that can be obtained using the simplest disturbance profiles and data reduction techniques possible. The procedure yields high run completion rates for most subjects. It is expected to yield near 100% completion rates for most of the participating astronauts as suggested in Chapter 5. The procedure also clearly shows different levels of velocity nulling task performance with different subjects. Therefore, there is a high probability that the procedure will be able to reveal any differences in otolith sensitivity that may occur between the pre- and post-flight testing.

The procedure also yields very accurate results. The one sigma deviation and data scatter of the valid results are at or below those of other human subject testing. This is shown for individual results as well as the results of all subjects compared together. Consistency of response with varying profiles is also shown. This proves that the closed-loop test used in this work is an accurate test method. This response consistency is the most encouraging result discovered in this work.

The major effort of this work has been to optimize the test procedure so that this consistency could be obtained as readily as possible. A major factor in the success of the procedure has been restricting the disturbance profile to a maximum velocity of approximately 0.63 m/s and a track length

to roughly 2.0 m. This allows some margin for error and varying levels of performance while still yielding a high run completion rate. Another major factor has been scaling the joystick gain to the maximum velocity using the $p=95\%$ criterion. (Chapter 3) This gives a quantitative value for the lowest practical sensitivity of the joystick control for any profile. Perhaps the most important factor has been the scaling of the velocity amplitudes at the disturbance frequencies. With this scaling, the disturbance profile has been tailored to the capabilities of the human operator. This has significantly helped the run completion rate and the quality of the results. These are the most important factors discovered in the development of the profile itself. It is felt that a successful profile is not unique, but that the factors described above are. Therefore, there are many possibilities for further work with similar profiles.

A final major factor in the success of the results has been the instruction to the subject to REACT to the disturbance. With subjects following this instruction the results have been improved. This improvement not only shows in the data, but in the subjects performance during the test sessions. This was a major discovery for obtaining consistent data.

No washout filtering was used in the procedure development as it was desired to keep the procedure as simple as possible. A desirable washout would add a below threshold velocity to the cart if it was moving toward the track center and subtract this same velocity if it was moving away from the track center. This would tend to help keep the cart at the center of the track. It would be especially useful during periods of small oscillations when the cart is near the track limits. This type of motion was often seen during the testing and usually led to a short run. It is

felt that a washout would not allow more than a 15% increase in the maximum velocity or a disturbance frequency below 0.03 Hz. Larger changes in these requirements would require larger washout amplitudes. This would tend to destroy the quasi-linearity of the sum of sines disturbance and the describing function analysis. A non-linear analysis would then be needed which would be much more difficult to use and interpret.

The procedure reveals differences in subject response which is desirable. These differences are difficult to quantify with the model used, however. Due to numerical problems a simpler transfer function with non-complex poles could not be obtained, as noted in Chapter 6. Although the structure used yielded models with excellent agreement to the log(GAIN) data, it was difficult to compare the parameter values of the models with the associated task performance. The model parameters do not show a simple direct comparison to task performance in all cases, which would be desirable.

The models also do not show good agreement with the phase data. This is primarily due to the emphasis on the low frequency peak, seen in the Bode plot log(GAIN) data, which adds more phase lead. Since such peaks were not clearly shown until the final testing, it is suspected that they are caused by the influence of the excessive cart vibration on the subjects performance. Also, for the small disturbance frequency range used in this testing it is difficult to find a model which will show good log(GAIN) and phase agreement by using only the GAIN data. Since the log(GAIN) showed less variance, however, it is the best data to use for modeling purposes. Either the peaks are not valid and therefore should not be emphasized, or the HO is not a minimum phase system. Further population testing, without the cart vibration, should resolve this discrepancy.

That the phase data shows more variance is not surprising. It has been stated that the otolith organs act as accelerometers and would not sense a constant velocity motion. This often occurred in testing. Subjects usually could sense their motion, through the cart noise and vibration, but were often unsure of their direction. During the acceleration reversals the magnitude and direction could be clearly sensed, but between these reversals there was confusion. Subjects often used small control inputs to determine their motion during these parts of the profile. Such inputs would have little influence on the GAIN data, due to their low amplitudes, but would have much influence on the phase data, due to their sign reversals. It is probably this effect which results in more phase variance.

7.2) Recommendations for Further Work

The most important recommendation is the need for further population testing. A larger subject sample is needed to confirm that the trends seen in this work are valid. This will also help determine the effects of the cart vibration.

Another model structure should be investigated. A more consistent structure for all subject responses would be desirable. A more simple and direct performance indication from the model parameters is also desired. A better phase fit may also be obtained by a different model structure. Possibly both the GAIN and the phase data could be used in the modeling to achieve this.

A washout filter should also be investigated. The main result desired would be to increase the run completion rate. If this occurred perhaps the

effects of a higher maximum velocity and/or a lower disturbance frequency range could be investigated. This would help to more fully define the low frequency response.

Since the cart dynamics have not been clearly separated from the HO response in this work, it is felt that this should be investigated. It would be interesting to see what effects varying cart dynamics had on the HO response. It would be desirable to use a system with cart dynamics clearly different than the HO response discovered in this work. This would hopefully reveal more clearly the capability and adaptability of the HO in this particular task.

APPENDIX A

SLED SYSTEM PICTURES

The following pictures show the various head restraint and joystick configurations used in the testing.

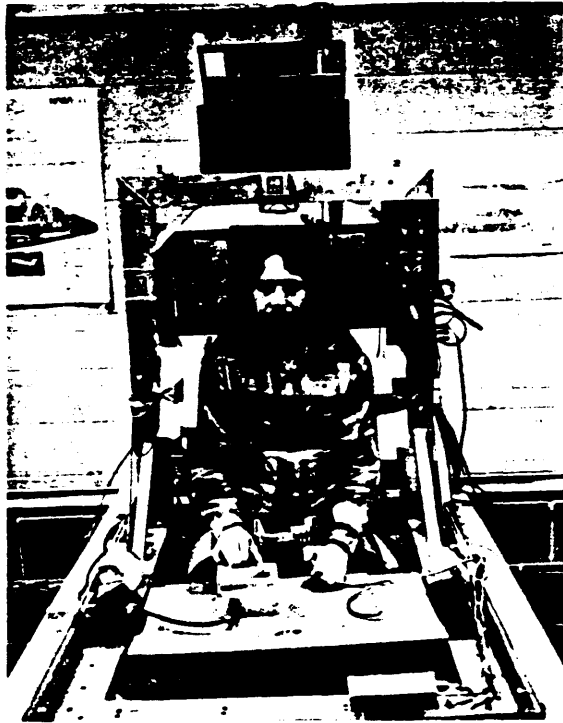


Figure A.1 Seated Subject with Open-Faced Head Restraint Used in Initial Tests

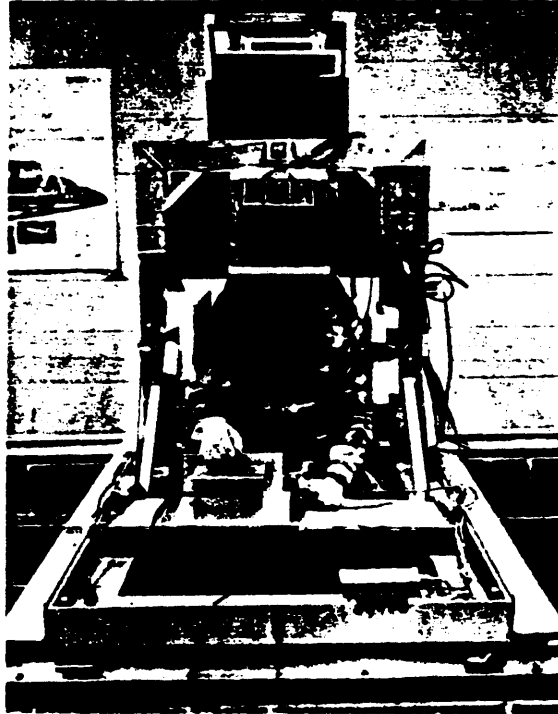


Figure A.2 Seated Subject with Enclosed Head Restraint

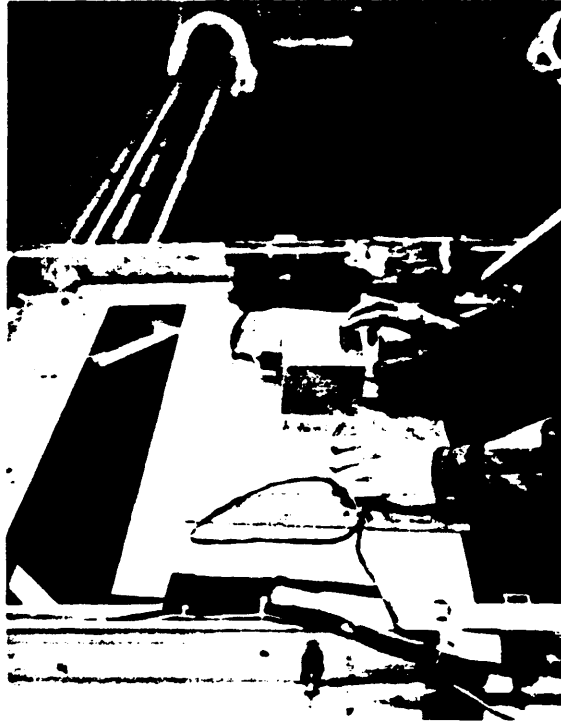


Figure A.3 Seated Subject with Wheel Joystick
Used in Initial Tests

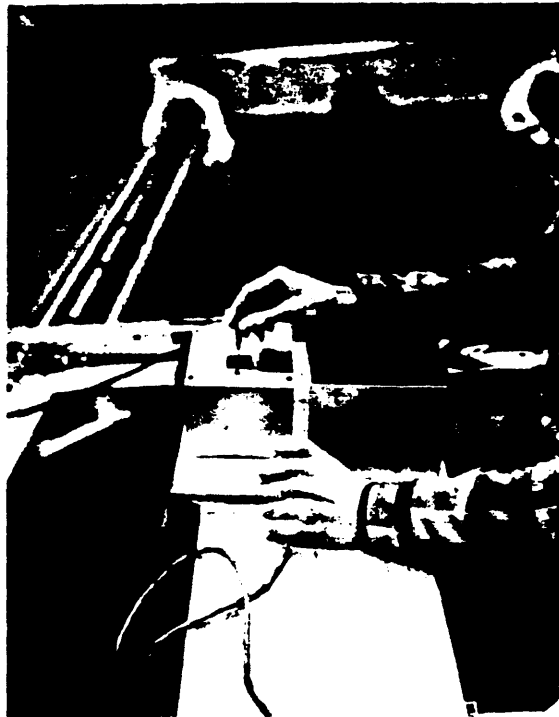


Figure A.4 Seated Subject with Joystick Controller

APPENDIX B

PROFILE GENERATION PROGRAMS

The two versions of the profile generation program used in this work are listed. The hierarchy is as follows:

Program	Subroutines	Data Files	Page
QSOS.FUN	SUMSIN.FUN, POWER.FUN	PRIMES.DAT	131-139
DSOS.FOR	SUMSIN.FOR, FPOWER.FOR	PRIMES.DAT, DSOS.DAT AMPAG.DAT	140-145

Program QSOS.FUN is a CART program subroutine accessed by the SO command. It is used with the PDP 11/34 minicomputer. Program DSOS.DAT is used on the VAX computer and is a simplified copy of QSOS.FUN. (DSOS.DAT is also a subroutine, as a simple command program (not listed) is used to run it.)

Both programs are the same except for their input and output formats. Program QSOS.FUN is user friendly. Only the prime numbers are input from a separate data file. The program prompts the user for all other inputs. The inputs and outputs are displayed in a self-explanatory format. A sample input and output is shown.

Program DSOS.FOR uses inputs from separate files. PRIMES.DAT contains the prime numbers as in the QSOS.FUN program. DSOS.DAT contains all other parameter inputs. The program prompts the user to determine if single or multiple DEL frequencies are to be used. The multiple DEL frequencies must be set in the program itself. It also prompts the user to determine if arbitrary amplitudes will be used as contained in file AMPAG.DAT, and to determine if histogram data is required. The input and output parameters

are displayed by their variable names as described in Chapter 3. The profile amplitudes and frequencies are then displayed as in QSOS.FUN. Input and output is also shown for the same example run used for the QSOS.FUN program example.

```

CART SYSTEM U03
0001   INTEGER FUNCTION QSOS (TRUN,STPOS,TLOOP,RSINES,TO)
C
C   Sets parameters for sum of sinusoids profile.
C   Author: A.P. Arrott
C   Adapted from RANDOM.SUB (Arrott,4-May-79)
C   Required subprograms:
C     FUNCTION POWER (which shares COMMON BLOCK/FILCOM/)
C     FUNCTION SUMSIN
C     FUNCTION ACCEPT
C     FUNCTION IACCEP
C
C   =====
C   DECLARATIONS
C   =====
C
0002   INTEGER PCODE,PRIME(50),FLAMPA
0003   LOGICAL ANSWER,YES
0004   REAL LVEL
0005   REAL W(50),WT(50),BOXW(51),COSAMP(50),COSPHI(50)
C
0006   INTEGER ZNSINE,ZTO,FLAT
0007   INTEGER FILTER
0008   REAL TL(5)
0009   REAL KCMD
0010   REAL AMP(50),WDELTA(50),PHI(50),AMPAG(50)
C
0011   COMMON/GLOBAL/ TRACK,GMAX,DECHAX,TL
0012   COMMON/UNITS/  KCMD
0013   COMMON/SOS/   AMP,WDELTA,PHI,ZTRUN,ZSTPOS,ZTLOOP,ZNSINE,ZTO,
+                 DELPHI,POSIM,ACCLIM,FLAT
0014   COMMON/FILCOM/ FILTER,POLE
C
0015   DATA PI/3.14159/,YES/'Y',INFLAG/0/
0016   DATA PRIME/3,5,7,11,13,17,19,23,29,31,37,41,43,47,53,61,73,83,
+         101,113,137,149,163,181,199,233,263,293,317,353,383,421,457,
+         499,547,587,619,661,691,739,787,823,863,911,947,997,1051,
+         1091,1163,1193/
C
0017   QSOS=4
0018   FLAMPA=0
C
C   Convert phase angles from phase at t=TO-TLOOP to phase at t=0.
0019   DO 112 I=1,ZNSINE
0020   112 PHI(I)=PHI(I)+(ZTO/ZTLOOP-1)*WDELTA(I)
C   Convert amplitudes from m/s to cart command units.
0021   DO 114 I=1,ZNSINE
0022   114 AMP(I)=AMP(I)/KCMD
C
C   Previous parameters
0023   TRUN=ZTRUN
0024   STPOS=ZSTPOS
0025   TLOOP=ZTLOOP
0026   NSINES=ZNSINE
0027   TO=ZTO
C
C   =====
C   OBTAIN PRIME NUMBER SERIES
C   =====
C
0028   CALL TTYOUT('Use prime number table ? ')
0029   READ(5,116) ANSWER
0030   116 FORMAT(A1)
0031   IF(ANSWER.NE.YES) GOTO 120
0032   CALL ASSIGN(4,'PRIMES.DAT')
0033   READ(4,117) NPRIME
0034   117 FORMAT(I5)
0035   DO 118 I=1,NPRIME
0036   118 READ(4,117) IPRIME
0037   PRIME(I)=IPRIME
0038   DO 119 I=NPRIME+1,50
0039   119 PRIME(I)=0
0040   CALL CLOSE(4)
0041   WRITE(7,1195) PRIME
0042   1195 FORMAT(2X,5I5)
0043   C

```

```

C =====
C DISPLAY TABLE OF PARAMETERS
C =====
0044 120 WRITE(7,121) TRUN,NSINES,1./TRUN,FLAT,360.*DELPHI/(2.*PI),
+ FILTER,POLE/(2.*PI),POSLIM*2.,ACCLIM/9.812,TLOOP,FLAMPA
0045 121 FORMAT ('O=====')
+ / SUM OF SINES PROFILE'//
+ / 1. DURATION OF PROFILE: ',F7.2,' SEC'/
+ / PARAMETERS OF SINUSOIDS: '//
+ / 2. NUMBER OF SINUSOIDS: ',I6,/'
+ / 3. FUNDAMENTAL FREQUENCY: ',F9.4,' HZ'/
+ / 4. EQUAL AMPLITUDE DOMAIN: ',I6,' (-1,P10,U+1,A)'/
+ / 5. SUCCESSIVE PHASE ANGLE: ',F7.0,' DEG'/
+ / PARAMETERS OF SHAPING FUNCTION: '//
+ / 6. ORDER OF FILTER: ',I6,/'
+ / 7. POLE: ',F9.2,' HZ'/
+ / PHYSICAL CONSTRAINTS: '//
+ / 8. LENGTH OF TRACK: ',F9.2,' M'/
+ / 9. ALLOWED ACCELERATION: ',F9.2,' G'/
+ / 10. TIME INCREMENT: ',F10.3,' SEC'/
+ / ARBITRARY ACCELERATION AMPLITUDE: '//
+ / 11. ARBITRARY AMP FLAG: ',I6,' (0,NO;1,YES)'/
0046 IF (INFLAG.EQ.0) GO TO 450
0048 WRITE(7,122)
0049 122 FORMAT ('O RESULTING IN THE SUM OF SINUSOIDS: '//
+ / FREQ AMP ACCEL AMP PHASE'/
+ / [HZ] [M/S] [G] [DEG]')
C
C =====
C GENERATE FREQUENCY TABLE
C =====
0050 200 CONTINUE
C
C FIRST METHOD: USE PRIME NUMBERS STORED IN FILE, 'PRIMES.TAB'
C
C CALCULATE HARMONIC FREQUENCIES
0051 FUNDA=2.*PI/TRUN ! FUNDAMENTAL FREQUENCY
0052 DO 220 J=1,NSINES
0053 W(J)=PRIME(J)*FUNDA ! HARMONIC FREQUENCIES
0054 WDELTA(J)=W(J)*TLOOP ! INCREMENT OF SINE ARGUMENT FOR W(J)
0055 220 CONTINUE
0056 GO TO 230
C
C SECOND METHOD: DETERMINE EQUAL SPACING OF LN(PRIMES)
C (ALLOWS INDEPENDENT SETTING OF HIGH,LOW,
C AND NO. OF FREQUENCIES IN SIGNAL)
C
C ** TO BE DEVELOPED **
C
C =====
C GENERATE AMPLITUDE TABLE
C =====
C
C USE ARBITRARY ACCELERATION AMPLITUDES FROM FILE 'AMPAG.DAT'
0057 230 IF(FLAMPA.EQ.0)GO TO 240
0059 CALL ASSIGN(10,'AMPAG.DAT')
0060 READ(10,*)(AMPAG(I),I=1,NSINES)
0061 DO 232 I=1,NSINES
0062 232 AMP(I)=AMPAG(I)*9.81/W(I)
0063 GO TO 260
C
C
0064 240 CONTINUE
0065 IF(FLAT) 241,242,243
C
0066 241 DO 2415 J=1,NSINES ! EQUAL AMPLITUDES OF POSITION
0067 2415 AMP(J)=W(J)
0068 GO TO 244
C
0069 242 DO 2425 J=1,NSINES ! EQUAL AMPLITUDES OF VELOCITY
0070 2425 AMP(J)=1.
0071 GO TO 244
C
0072 243 DO 2435 J=1,NSINES ! EQUAL AMPLITUDES OF ACCELERATION
0073 2435 AMP(J)=1./W(J)
C
0074 244 CONTINUE
C

```

```

0075     IF(FILTER.EQ.0) GO TO 260
0077     IF(NSINES.EQ.1) GO TO 260
C
C   S H A P I N G   F U N C T I O N
C       (adapted from LMKRUN by G.L.Zacharias)
C
C   CALCULATE FICTITIOUS END FREQUENCIES
0079     W0=W(1)*W(1)/W(2)
0080     W(NSINES+1)=W(NSINES)*W(NSINES)/W(NSINES-1)
C
C   CALCULATE BOX FREQUENCIES
0081     DO 245 J=NSINES+1,2,-1
0082     BOXW(J)=SQRT(W(J-1)*W(J))
0083     BOXW(1)=SQRT(W0*W(1))
C
C   CALCULATE AMPLITUDES
0084     GAIN=2./(POWER(BOXW(NSINES+1))-POWER(BOXW(1)))
0085     DO 248 J=1,NSINES
0086     248  AMP(J)=AMP(J)*SQRT(GAIN*(POWER(BOXW(J+1))-POWER(BOXW(J))))
C
C   =====
C   GENERATE PHASE ARRAY
C   =====
C
0087 260  CONTINUE
C
C   FIRST METHOD: CONSTANT PHASE DIFFERENCE
0088     PHI(1)=0.
0089     DO 264 J=2,NSINES
0090     PHI(J)=PHI(J-1)+DELPHI
0091     PHI(J)=AMOD(PHI(J),2.*PI)  ! REMAINDER FUNCTION
0092 264  CONTINUE                ! (ADJUSTS PHASES > 2*PI)
0093     GO TO 400
C
C   SECOND METHOD: SET PHASES INDIVIDUALLY
C
C   ** TO BE DEVELOPED **
C
C   =====
C   FIND FIRST ZERO CROSSING      (THEREBY ESTABLISHING THE
C   =====                      BEGINNING OF THE SIGNAL)
C
0094 400  CONTINUE
C
C   GENERATE VELOCITY SIGNAL
C   INITIALIZE
0095     T0=0.
0096     DO 420 J=1,NSINES
0097 420  WT(J)=-WDELTA(J)
0098     LVEL=SUMSIN(NSINES,AMP,WT,WDELTA,PHI)  ! VALUE AT T=0
C                                             (ALGORITHM IGNORES POSSIBILITY THAT T=0 IS
C                                             A ZERO CROSSING)
C
C   ITERATE
0099 430  T0=T0+TLOOP
0100     VEL=SUMSIN(NSINES,AMP,WT,WDELTA,PHI)
C
C   COMPARE VALUE WITH VALUE OF PREVIOUS ITERATION
0101     IF (VEL.GE.0.AND.LVEL.LE.0.) GO TO 440
0103     IF (VEL.LE.0.AND.LVEL.GE.0.) GO TO 440
0105     LVEL=VEL  ! ZERO CROSSING NOT FOUND: UPDATE 'LVEL',
0106     IF (T0.GT.TRUN) GO TO 434  ! CHECK FOR END OF SIGNAL,
0108     GO TO 430  ! AND ITERATE AGAIN.
C
C   ERROR: NO ZERO CROSSING FOUND IN SIGNAL
0109 434  WRITE(7,435)
0110 435  FORMAT (' ==>Error[QSOS] Unable to find zero crossings of',
+             ' signal.')
0111     GOTO 120
C
C   ZERO CROSSING DETECTED
0112 440  CONTINUE
C

```

```

C =====
C DETERMINE SIGNAL SCALE FACTOR
C =====
C
C GENERATE POSITION PROFILE
C INITIALIZE
0113 DO 520 J=1,NSINES
0114 WT(J)=T0*W(J)-WDEL(T(J) ! START SIGNAL AT TO
0115 COSAMP(J)=AMP(J)/W(J) ! INTEGRATION COEFFICIENT
0116 COSPHI(J)=PHI(J)-.5*PI ! INTEGRATION PHASE (1/4 CYCLE LAG)
0117 520 CONTINUE
0118 NSTEPS=TRUN/TLOOP
0119 POSMAX=0.
0120 POSMIN=0.
C ITERATE
0121 DO 540 IT=1,NSTEPS
0122 POS=SUMSIN(NSINES,COSAMP,WT,WDEL(T,COSPHI)
C
C FIND MAX/MIN
0123 IF (POS.GT.POSMAX) POSMAX=POS
0125 IF (POS.LT.POSMIN) POSMIN=POS
0127 540 CONTINUE
C
C SCALE SIGNAL TO LENGTH OF TRACK
0128 PSF=POSLIM*2./(POSMAX-POSMIN) ! POSITION SCALE FACTOR
C
C GENERATE ACCELERATION PROFILE
C INITIALIZE
0129 DO 570 J=1,NSINES
0130 WT(J)=T0*W(J)-WDEL(T(J)
0131 COSAMP(J)=PSF*AMP(J)*W(J) ! DERIVATIVE COEFFICIENT
0132 COSPHI(J)=PHI(J)+.5*PI ! DERIVATIVE PHASE (1/4 CYCLE LEAD)
0133 570 CONTINUE
0134 ACCMAX=0.
0135 ACCMIN=0.
C ITERATE
0136 DO 580 IT=1,NSTEPS
0137 ACC=SUMSIN(NSINES,COSAMP,WT,WDEL(T,COSPHI)
C
C FIND MAX/MIN ACCELERATION
0138 IF (ACC.GT.ACCMAX) ACCMAX=ACC
0140 IF (ACC.LT.ACCMIN) ACCMIN=ACC
0142 580 CONTINUE
C
C DETERMINE ACCELERATION SCALE FACTOR
0143 IF (ABS(ACCMIN).GT.ACCMAX) ACCMAX=-ACCMIN
0145 ASF=ACCLIM/ACCMAX
0146 IF (ACCMAX.LE.ACCLIM) ASF=1.
C
C CALCULATE SIGNAL SCALE FACTOR AND SCALE AMPLITUDES
0148 USAGE=ASF*100. ! PERCENT USAGE OF TRACK
0149 SCALE=PSF*ASF ! SIGNAL SCALE FACTOR
0150 DO 590 J=1,NSINES
0151 590 AMP(J)=SCALE*AMP(J)
0152 ACCMAX=ASF*ACCMAX ! MAXIMUM ACCELERATION IN SIGNAL
C
C CALCULATE STARTING POSITION
0153 DO 594 J=1,NSINES
0154 WT(J)=T0*W(J)-WDEL(T(J)
0155 COSAMP(J)=AMP(J)/W(J)
0156 COSPHI(J)=PHI(J)-.5*PI !CHANGED + TO - DWH 25/AUG/82
0157 594 CONTINUE
0158 STPOS=SUMSIN(NSINES,COSAMP,WT,WDEL(T,COSPHI)
+ -(POSMAX+POSMIN)*.5*SCALE
C
C =====
C DISPLAY SIGNAL CHARACTERISTICS
C =====
C
0159 600 CONTINUE
C
C DISPLAY FREQUENCIES, AMPLITUDES, AND PHASES OF SINUSOIDS
0160 DO 620 J=1,NSINES
0161 WRITE(7,611) W(J)/(2.*PI), AMP(J), AMP(J)*W(J)/9.81,
+ W(J)*T0+360.*PHI(J)/(2.*PI)
0162 611 FORMAT(F11.3,F11.2,F11.3,F11.0)
0163 620 CONTINUE
C

```



```

C   DISPLAY OTHER SIGNAL CHARACTERISTICS
0164   WRITE(7,631) ACCMAX/9.81      ! CONVERT TO UNITS OF GRAV.ACC.
0165 631  FORMAT('OMAXIMUM ACCELERATION IN SIGNAL:',F7.3,' G')
0166   WRITE(7,635) USAGE           ! PERCENT USAGE OF TRACK
0167 635  FORMAT(' PERCENT USAGE OF TRACK:',F7.2,'%')
C   WRITE(7,637) STPOS,TO*100/TRUN
C637  FORMAT(' INITIAL POSITION:',F5.2,' FT
C   +
C       [SIGNAL SHIFT:',F6.2,' XJ')
C
0168   WRITE(7,638) STPOS
0169 638  FORMAT(' STARTING POSITION:',F7.2)
C   WRITE(7,6383) SCALE,SCALE*POSMAX,SCALE*POSMIN
C6383  FORMAT(' SCALE:',F5.2,' PMAX:',F6.2,' PMIN:',F6.2)
C   WRITE(7,6385) SCALE*(POSMIN+POSMAX)*.5,SCALE*(POSMAX-POSMIN)*.5
C6385  FORMAT(' AVG:',F7.2,' TRACK USED:',F7.2)
0170   WRITE(7,639)
0171 639  FORMAT(' =====')
C
C   =====
C   INTERACTIVE PARAMETER CHANGES
C   =====
C
0172 640  WRITE(7,641)
0173 641  FORMAT('$OK? ')
0174   READ(5,643) ANSWER
0175 643  FORMAT(A1)
0176   IF (ANSWER.EQ.YES) GO TO 750
C
C   CHANGE PARAMETERS
0178 650  INFLAG=1
0179   WRITE(7,651)
0180 651  FORMAT('$ PARAMETER #')
0181   READ(5,653,ERR=650) PCODE
0182 653  FORMAT(I9)
0183   IF (PCODE.EQ.0) GO TO 120 ! REDISPLAY SIGNAL PARAMETERS
0185 660  IF (PCODE.LT.1.OR.PCODE.GT.15) GO TO 650
0187   GO TO (701,702,703,704,705,706,707,708,709,710,711,712,
C   +
C       713,714,715) PCODE
C
C   DURATION OF RUN
0188 701  WRITE(7,7011) TRUN
0189 7011  FORMAT('$DURATION OF RUN:',F5.0,' SEC.      ENTER NEW VALUE: ')
0190   READ(5,7013,ERR=701) TEMP
0191 7013  FORMAT(F15.6)
0192   IF (ACCEPT(TEMP,1.,10000.)) 660,650,7016
0193 7016  TRUN=TEMP
0194   W(1)=2.*PI/TRUN
0195   W(NSINES)=PRIME(NSINES)*W(1)
0196   GO TO 650
C
C   NUMBER OF SINES
0197 702  WRITE(7,7021) NSINES
0198 7021  FORMAT('$NUMBER OF SINUSOIDS:',I5,'      ENTER NEW VALUE: ')
0199   READ(5,7023,ERR=702) ITEMP
0200 7023  FORMAT(I9)
0201   IF (IACCEP(ITEMP,1,50)) 660,650,7026
0202 7026  NSINES=ITEMP
0203   W(NSINES)=PRIME(NSINES)*W(1)
0204   GO TO 650
C
C   FUNDAMENTAL FREQUENCY
0205 703  WRITE(7,7031) W(1)/(2.*PI)
0206 7031  FORMAT('$FUNDAMENTAL FREQUENCY:',F7.4,' HZ. ENTER NEW VALUE: ')
0207   READ(5,7013,ERR=703) TEMP
0208   IF (ACCEPT(TEMP,.0001,1.)) 660,650,7036
0209 7036  W(1)=TEMP*2.*PI
0210   TRUN=1./TEMP
0211   W(NSINES)=PRIME(NSINES)*W(1)
0212   GO TO 650
C
C   EQUAL AMPLITUDE DOMAIN
0213 704  WRITE(7,7041) FLAT
0214 7041  FORMAT('$POS=-1,VEL=0,ACC=+1, NOW=',I2,'      ENTER NEW VALUE: ')
0215   READ(5,7023,ERR=704) ITEMP
0216   IF (IACCEP(ITEMP,-1,1)) 660,7046,7046
0217 7046  FLAT=ITEMP
0218   GO TO 650
C

```

```

C SUCCESSIVE PHASE ANGLES
0219 705 WRITE(7,7051) DELPHI*360./(2.*PI)
0220 7051 FORMAT('SUCCESSIVE PHASE ANGLE:',F5.0,' DEG. ENTER NEW VALUE: ')
0221 READ(5,7013,ERR=705) TEMP
0222 IF (ACCEPT(TEMP,0.,36000.)) 660,650,7056
0223 7056 DELPHI=TEMP*2.*PI/360.
0224 GO TO 650

C
C ORDER OF FILTER
0225 706 WRITE(7,7061) FILTER
0226 7061 FORMAT('ORDER OF FILTER:',I5,' ENTER NEW VALUE: ')
0227 READ(5,7023,ERR=706) ITEMP
0228 IF (IACCEPT(ITEMP,1,3)) 660,7066,7066
0229 7066 FILTER=ITEMP
0230 GO TO 650

C
C POLE OF FILTER
0231 707 WRITE(7,7071) POLE/(2.*PI)
0232 7071 FORMAT('POLE OF FILTER:',F5.2,' ENTER NEW VALUE: ')
0233 READ(5,7013,ERR=707) TEMP
0234 IF (ACCEPT(TEMP,.001,10.)) 660,650,7076
0235 7076 POLE=TEMP*2.*PI
0236 GO TO 650

C
C LENGTH OF TRACK
0237 708 WRITE(7,7081) POSLIM*2.
0238 7081 FORMAT('LENGTH OF TRACK:',F5.2,' M. ENTER NEW VALUE: ')
0239 READ(5,7013,ERR=708) TEMP
0240 IF (ACCEPT(TEMP,1.,15.)) 660,650,7086
0241 7086 POSLIM=TEMP*.5
0242 GO TO 650

C
C MAXIMUM ALLOWED ACCELERATION
0243 709 WRITE(7,7091) ACCLIM/9.812
0244 7091 FORMAT('ALLOWED ACCELERATION:',F5.3,' ENTER NEW VALUE: ')
0245 READ(5,7013,ERR=709) TEMP
0246 IF (ACCEPT(TEMP,.001,1.)) 660,650,7096
0247 7096 ACCLIM=TEMP*9.812
0248 GO TO 650

C
C TIME INCREMENT
0249 710 WRITE(7,7101) TLOOP
0250 7101 FORMAT('TIME INCREMENT:',F5.3,' SEC ENTER NEW VALUE: ')
0251 READ(5,7013,ERR=705) TEMP
0252 IF (ACCEPT(TEMP,.001,.500)) 660,650,7106
0253 7106 TLOOP=TEMP
0254 GO TO 650

C
C ARBITRARY ACCELERATION AMPLITUDE FLAG
0255 711 WRITE(7,7111) FLAMPA
0256 7111 FORMAT('ARBITRARY ACCEL AMP FLAG:',I5,' ENTER NEW VALUE: ')
0257 READ(5,7023,ERR=711) ITEMP
0258 IF (IACCEPT(ITEMP,0,1)) 660,7116,7116
0259 7116 FLAMPA=ITEMP
0260 GO TO 650
0261 712 CONTINUE
0262 713 CONTINUE
0263 714 CONTINUE
0264 715 CONTINUE
0265 GO TO 650

C
0266 750 RSINES=NSINES ! Float no. of sines.
C Convert from phase angles at t=0 to phase angles at t=T0-TLOOP
0267 DO 755 I=1,NSINES
0268 755 PHI(I)=PHI(I)+(T0/TLOOP-1)*WDELTA(I)
C Convert amplitudes from m/s to cart command units.
0269 DO 760 I=1,NSINES
0270 760 AMP(I)=KCMD*AMP(I)
C
0271 WRITE(7,762) KCMD
0272 762 FORMAT(5X,'KCMD=',F10.4)
C
C Save parameter values.
0273 ZTRUN=TRUN
0274 ZSTPOS=STPOS
0275 ZTLOOP=TLOOP
0276 ZNSINE=NSINES
0277 ZT0=T0
C
0278 RETURN
0279 END
RS05

```

```

CART SYSTEM V03
0001     FUNCTION SUMSIN (N,A,WT,WDELTA,PHI)
C
C   AUTHOR: ARROTT
C   CREATION DATE: 18-JAN-79
C
C   PURPOSE: ITERATES SUM OF SINES SIGNAL
C
C   N:      NUMBER OF SINES IN SUM.
C   A:      ARRAY OF SINE AMPLITUDES.
C   WT:     ARRAY OF PRODUCTS OF SINE FREQUENCIES AND CURRENT TIME.
C           (IN RADIANS)
C   WDELTA: ARRAY OF PRODUCTS OF SINE FREQUENCIES AND ITERATION
C           INTERVAL.
C   PHI:    ARRAY OF PHASE ANGLES
C
0002     DIMENSION A(50),WT(50),WDELTA(50),PHI(50)
0003     SUMSIN=0.
0004     DO 1 J=1,N
0005         WT(J)=WT(J)+WDELTA(J)
0006         SUMSIN=SUMSIN+A(J)*SIN(WT(J)+PHI(J))
0007 1     CONTINUE
0008     END
SUMSIN

0001     FUNCTION POWER (W)
C
C   Used in calculation of amplitudes of discrete frequencies
C   in the shaping function algorithm in subprogram QSOS.FUN
C
C   Author: G.L.Zacharias (originally FUNCTION PWR in LNKRUN program).
C   Adapted by Arrott for use by module RANDOM.SUB in the SLED program.
C   Adaptation date: 18-Jan-79
C   No changes necessary for use by module QSOS.FUN in CART program.
C
0002     INTEGER FILTER
0003     COMMON/FILCOM/FILTER,POLE
C
0004     TEMP=(ATAN(W/POLE))/POLE
0005     GO TO (1,2,3) FILTER
C
C   FIRST ORDER FILTER
0006 1     POWER=TEMP
0007     RETURN
C
C   SECOND ORDER FILTER
0008 2     PP=POLE*POLE
0009     POWER=(W/(PP+W*W)+TEMP)/(2.*PP)
0010     RETURN
C
C   THIRD ORDER FILTER
0011 3     PP=POLE*POLE
0012     PI=PP+W*W
0013     POWER=-((W*W*W)/(PI*PI)-(5.*W/PI+3*TEMP)*.5)/(4.*PP*PP)
0014     RETURN
C
0015     END
POWER

```

PRIMES.DAT

12
5
7
9
11
13
17
19
23
29
31
37
41

AMPAG.DAT

0.08,0.09,0.09,0.10,0.10,0.13,0.13,0.12,0.13,0.11,0.11,0.13

RUN HICAR7

CART CONTROLLER V04.7 (9-NOV-81) 06-JAN-83 16:26:06
CART:SO

Use Prime number table ? YES

5	7	9	11	13
17	19	23	29	32
37	41	0	0	0
0	0	0	0	0
0	0	0	0	0
0	0	0	0	0
0	0	0	0	0
0	0	0	0	0
0	0	0	0	0
0	0	0	0	0
0	0	0	0	0

=====

SUM OF SINES PROFILE

1. DURATION OF PROFILE: 184.32 SEC
 PARAMETERS OF SINUSOIDS:
 2. NUMBER OF SINUSOIDS: 25
 3. FUNDAMENTAL FREQUENCY: 0.0054 HZ
 4. EQUAL AMPLITUDE DOMAIN: 0 (-1,P;0,V;+1,A)
 5. SUCCESSIVE PHASE ANGLE: 247. DEG
 PARAMETERS OF SHAPING FUNCTION:
 6. ORDER OF FILTER: 2
 7. POLE: 0.31 HZ
 PHYSICAL CONSTRAINTS:
 8. LENGTH OF TRACK: 4.00 M
 9. ALLOWED ACCELERATION: 0.41 G
 10. TIME INCREMENT: 0.015 SEC
 ARBITRARY ACCELERATION AMPLITUDE:
 11. ARBITRARY AMP FLAG: 0 (0,NO;1,YES)

PARAMETER #1
 DURATION OF RUN: 184. SEC. ENTER NEW VALUE: 81.92
 PARAMETER #2
 NUMBER OF SINUSOIDS: 25 ENTER NEW VALUE: 12
 PARAMETER #3
 FUNDAMENTAL FREQUENCY: 0.0122 HZ. ENTER NEW VALUE: 0.0122
 PARAMETER #5
 SUCCESSIVE PHASE ANGLE: 247. DEG. ENTER NEW VALUE: 253.
 PARAMETER #6
 ORDER OF FILTER: 2 ENTER NEW VALUE: 1
 PARAMETER #7
 POLE OF FILTER: 0.31 ENTER NEW VALUE: 0.16
 PARAMETER #8
 LENGTH OF TRACK: 4.00 M. ENTER NEW VALUE: 2.05
 PARAMETER #9
 ALLOWED ACCELERATION: 0.408, ENTER NEW VALUE: 0.20
 PARAMETER #10
 TIME INCREMENT: 0.015 SEC ENTER NEW VALUE: .010
 PARAMETER #

 SUM OF SINES PROFILE

1. DURATION OF PROFILE: 81.92 SEC
 PARAMETERS OF SINUSOIDS:
 2. NUMBER OF SINUSOIDS: 12
 3. FUNDAMENTAL FREQUENCY: 0.0122 HZ
 4. EQUAL AMPLITUDE DOMAIN: 0 (-1,PI/0,V;+1,A)
 5. SUCCESSIVE PHASE ANGLE: 253. DEG
 PARAMETERS OF SHAPING FUNCTION:
 6. ORDER OF FILTER: 1
 7. POLE: 0.14 HZ
 PHYSICAL CONSTRAINTS:
 8. LENGTH OF TRACK: 2.05 M
 9. ALLOWED ACCELERATION: 0.20 G
 10. TIME INCREMENT: 0.010 SEC
 ARBITRARY ACCELERATION AMPLITUDE:
 11. ARBITRARY AMP FLAG: 0 (0:NO;1:YES)

RESULTING IN THE SUM OF SINUSOIDS:

FREQ [HZ]	AMP [M/S]	ACCEL AMP [G]	PHASE [DEG]
0.061	0.12	0.005	0.
0.085	0.13	0.007	253.
0.110	0.12	0.008	147.
0.134	0.11	0.009	40.
0.159	0.12	0.012	293.
0.208	0.11	0.015	186.
0.232	0.10	0.014	79.
0.281	0.11	0.019	332.
0.354	0.09	0.021	226.
0.391	0.07	0.019	119.
0.452	0.07	0.021	12.
0.500	0.06	0.020	265.

MAXIMUM ACCELERATION IN SIGNAL: 0.113 G
 PERCENT USAGE OF TRACK: 100.00%
 STARTING POSITION: -0.24

 OK? YES

```

0001      SUBROUTINE DSOS
0002      C
0003      C
0004      C Profile generation for VAX. By Dale W. Hiltner Sept-82
0005      C Copied from CART program subroutine DSOS.FUN
0006      C
0007      C
0008      REAL LVEL,TL(5),KCMD,AMP(50),WDEL(50),PHI(50)
0009      REAL W(50),WT(50),BOXW(51),COSAMP(50),COSPHI(50)
0010      REAL AMPA(50)
0011      INTEGER PCODE,PRIME(50),ZNSINE,ZTO,FLAT,FILTER
0012      INTEGER A1,A2,IB,NSIN,NBIN,NABIN(41)
0013      COMMON AMP,WDEL,PHI,TLOOP,NSINES,NSTEPS,TRUN
0014      COMMON/FILCOM/FILTER,POLE
0015      DATA FI/3.14159/
0016      C SET PRIME NUMBERS
0017      CALL ASSIGN(4,'PRIMES.DAT')
0018      READ(4,500)NPRIME
0019      500  FORMAT(15)
0020      DO 10 I1=1,NPRIME
0021      READ(4,500)IFRIME
0022      10  PRIME(I1)=IFRIME
0023      DO 20 I2=NPRIME+1,50
0024      20  PRIME(I2)=0.0
0025      C SET OTHER INPUT PARAMETERS
0026      CALL ASSIGN(6,'DSOS.DAT',0)
0027      READ(6,*)TRUN,NSINES,FFREQ,FLAT,DEL,FILTER,FPOLE,FPOS,FACC,TLOOP
0028      C
0029      C INPUT ACCELERATION AMPLITUDES ARBITRARILY AND OTHER CHECKS
0030      WRITE(5,212)
0031      212  FORMAT(5X,'SET ACCEL AMPS PER AMPAG.DAT? (1=Y/0=N)')
0032      READ(5,213)A1
0033      213  FORMAT(I4)
0034      WRITE(5,214)
0035      214  FORMAT(5X,'MULTIPLE DEL FREQUENCIES? (1=YES/0=NO)')
0036      READ(5,215)A2
0037      215  FORMAT(I4)
0038      WRITE(5,110)
0039      110  FORMAT(5X,'CALCULATE HISTOGRAM DATA? (YES=1/NO=0)')
0040      READ(5,112)IBC
0041      112  FORMAT(I4)
0042      C
0043      IF(A2.EQ.0)GO TO 501
0044      C
0045      C SET DEL FREQUENCIES FOR MULTIPLE RUNS
0046      C
0047      DO 900 III=1,12
0048      DEL=(III-1)*30.0+13.0
0049      501  WRITE(7,499)
0050      499  FORMAT(1X,'INPUT PARAMETERS',/)
0051      WRITE(7,502)TRUN,NSINES,FFREQ,FLAT,DEL,FILTER,FPOLE,FPOS,FACC,TLOOP
0052      502  FORMAT(5X,'TRUN=',F8.3,/,5X,'NSINES=',I3,/,5X,'FFREQ=',F9.6,/,
0053      +      5X,'FLAT=',I2,/,5X,'DEL=',F8.3,/,5X,'FILTER=',I3,/,
0054      +      5X,'FPOLE=',F8.3,/,5X,'FPOS=',F6.2,/,5X,'FACC=',F6.3,/,
0055      +      5X,'TLOOP=',F6.3)
0056      C
0057      DELPHI=DEL*2.0*PI/360.0
0058      POLE=FPOLE*2.0*PI
0059      POSLIN=FPOS*0.5
0060      ACCLIN=FACC*9.812
0061      C FIND HARMONIC FREQUENCIES
0062      FUNDA=2.0*PI/TRUN
0063      DO 30 I3=1,NSINES
0064      W(I3)=PRIME(I3)*FUNDA
0065      WDEL(I3)=W(I3)*TLOOP
0066      30  CONTINUE
0067      C
0068      C SET ACCELERATION AMPLITUDES ARBITRARILY
0069      IF(A1.EQ.0)GO TO 240
0070      CALL ASSIGN(10,'AMPAG.DAT',0)
0071      READ(10,*)(AMPA(I),I=1,NSINES)
0072      CALL CLOSE(10)
0073      DO 220 I=1,NSINES
0074      220  AMP(I)=AMPA(I)*9.81/W(I)
0075      GO TO 260
0076      C
0077      C GENERATE AMPLITUDE TABLE
0078      240  IF(FLAT)241,242,243
0079      241  DO 2415 I=1,NSINES
0080      2415  AMP(I1)=W(I1)
0081      GO TO 244

```

```

0082      242 DO 2425 I2=1,NSINES
0083      2425 AMP(I2)=1.0
0084          GO TO 244
0085      243 DO 2435 I3=1,NSINES
0086      2435 AMP(I3)=1.0/W(I3)
0087      C
0088      244 IF(FILTER.EQ.0)GO TO 260
0089          IF(NSINES.EQ.1)GO TO 260
0090      C
0091      C SHAPING FUNCTION
0092      C
0093      C CALCULATE FICTITIOUS END FREQUENCIES
0094      C
0095          W0=W(1)*W(1)/W(2)
0096          W(NSINES+1)=W(NSINES)*W(NSINES)/W(NSINES-1)
0097      C
0098      C CALCULATE BOX FREQUENCIES
0099          DO 245 J=NSINES+1,2,-1
0100      245 BOXW(J)=SQRT(W(J-1)*W(J))
0101          BOXW(1)=SQRT(W0*W(1))
0102      C
0103      C CALCULATE AMPLITUDES
0104          GAIN=2.0/(FPOWER(BOXW(NSINES+1))-FPOWER(BOXW(1)))
0105          DO 248 J=1,NSINES
0106      248 AMP(J)=AMP(J)*SQRT(GAIN*(FPOWER(BOXW(J+1))-FPOWER(BOXW(J))))
0107      C
0108      C
0109      260 CONTINUE
0110      C CONSTANT PHASE DIFFERENCE
0111          PHI(1)=0
0112          DO 264 J=2,NSINES
0113          PHI(J)=PHI(J-1)+DELPHI
0114          PHI(J)=AMOD(PHI(J),2.0*PI)
0115      264 CONTINUE
0116      C GENERATE VELOCITY SIGNAL
0117          T0=0.0
0118          DO 420 J=1,NSINES
0119      420 WT(J)=-WDEL(T,J)
0120          LEVEL=SUMSIN(NSINES,AMP,WT,WDEL(T,PHI))
0121      430 T0=T0+TLOOP
0122          VEL=SUMSIN(NSINES,AMP,WT,WDEL(T,PHI))
0123          IF(VEL.GE.0.AND.LEVEL.LE.0)GO TO 440
0124          IF(VEL.LE.0.AND.LEVEL.GE.0)GO TO 440
0125          LEVEL=VEL
0126          IF(T0.GT.TRUN)GO TO 434
0127          GO TO 430
0128      434 WRITE(7,435)
0129      435 FORMAT('NO ZERO CROSSING FOUND. TERMINATE PROGRAM.')
0130          GO TO 99999
0131      C FIND MAX, MIN VELOCITY
0132      440 NSTEPS=TRUN/TLOOP
0133          VELMAX=0.0
0134          VELMIN=0.0
0135          DO 445 I4=1,NSINES
0136      445 WT(J)=T0*W(J)-WDEL(T,J)
0137          DO 450 I5=1,NSTEPS
0138          VEL=SUMSIN(NSINES,AMP,WT,WDEL(T,PHI))
0139          IF(VEL.GT.VELMAX)VELMAX=VEL
0140          IF(VEL.LT.VELMIN)VELMIN=VEL
0141      450 CONTINUE
0142          IF(ABS(VELMIN).GT.VELMAX)VELMAX=-VELMIN
0143      C
0144      C GENERATE POSITION PROFILE
0145          DO 520 J=1,NSINES
0146          WT(J)=T0*W(J)-WDEL(T,J)
0147          COSAMP(J)=AMP(J)/W(J)
0148          COSPHI(J)=PHI(J)-0.5*PI
0149      520 CONTINUE
0150      C FIND MAX AND MIN POSITION
0151          NSTEPS=TRUN/TLOOP
0152          POSMAX=0.0
0153          POSMIN=0.0
0154          DO 540 I7=1,NSTEPS
0155          POS=SUMSIN(NSINES,COSAMP,WT,WDEL(T,COSPHI))
0156          IF(POS.GT.POSMAX)POSMAX=POS
0157          IF(POS.LT.POSMIN)POSMIN=POS
0158      540 CONTINUE
0159      C SCALE SIGNAL TO LENGTH OF TRACK
0160          PSF=POSLIM*2.0/(POSMAX-POSMIN)

```

```

0161 C GENERATE ACCELERATION PROFILE
0162 DO 570 J=1,NSINES
0163 WT(J)=T0*W(J)-WDELTA(J)
0164 COSAMP(J)=PSF*AMP(J)*W(J)
0165 COSPHI(J)=PHI(J)+0.5*PI
0166 570 CONTINUE
0167 C FIND MAX AND MIN ACCELERATION
0168 ACCMAX=0.0
0169 ACCMIN=0.0
0170 DO 580 I=1,NSTEPS
0171 ACC=SUMSIN(NSINES,COSAMP,WT,WDELTA,COSPHI)
0172 IF(ACC.GT.ACCMAX)ACCMAX=ACC
0173 IF(ACC.LT.ACCMIN)ACCMIN=ACC
0174 580 CONTINUE
0175 C DETERMINE ACCELERATION SCALE FACTOR
0176 IF(ABS(ACCMIN).GT.ACCMAX)ACCMAX=-ACCMIN
0177 ASF=ACCLIM/ACCMAX
0178 IF(ACCMAX.LE.ACCLIM)ASF=1.0
0179 C CALCULATE SIGNAL SCALE FACTOR AND SCALE AMPLITUDES
0180 USAGE=ASF*100.0
0181 SCALE=PSF*ASF
0182 DO 590 J=1,NSINES
0183 ANF(J)=SCALE*AMP(J)
0184 ACCMAX=ASF*ACCMAX
0185 C CALCULATE STARTING POSITION
0186 DO 594 J=1,NSINES
0187 WT(J)=T0*W(J)-WDELTA(J)
0188 COSAMP(J)=AMP(J)*W(J)
0189 COSPHI(J)=PHI(J)-0.5*PI
0190 594 CONTINUE
0191 STPOS=SUMSIN(NSINES,COSAMP,WT,WDELTA,COSPHI)
0192 + (POSMAX+POSMIN)*0.5*SCALE
0193 C HISTOGRAM DATA CALCULATION
0194 IF(IBC.EQ.0)GO TO 831
0195 DO 120 J=1,NSINES
0196 WT(J)=T0*W(J)-WDELTA(J)
0197 COSAMP(J)=AMP(J)*W(J)
0198 COSPHI(J)=PHI(J)+0.5*PI
0199 120 CONTINUE
0200 ACC=0.0
0201 DO 124 I=1,41
0202 124 NABIN(I)=0
0203 DO 122 I=1,NSTEPS
0204 ACC=SUMSIN(NSINES,COSAMP,WT,WDELTA,COSPHI)
0205 VACC=ABS(ACC/9.81)/0.005
0206 NBIN=VACC+1
0207 NABIN(NBIN)=NABIN(NBIN)+1
0208 122 CONTINUE
0209 C
0210 C DISPLAY OUTPUT
0211 WRITE(7,829)
0212 829 FORMAT(1X,/, 'PROFILE DESCRIPTION',/)
0213 831 WRITE(7,830)ACCMAX/9.81,VELMAX*SCALE,USAGE,STPOS,SCALE
0214 830 FORMAT(5X, 'MAX ACCEL=' ,F6.3,/, 5X, 'VELMAX ' ,F8.1,/,
0215 + 5X, ' % USAGE OF TRACK=' ,F6.2,/,
0216 + 5X, ' STARTING POSITION=' ,F6.3,/,
0217 + 5X, ' SCALE=' ,F7.4,/)
0218 WRITE(7,800)
0219 800 FORMAT(1X, 'FREQ. (HZ)',3X, 'AMP1 (K/S)',3X, 'AMP2 (G) ',
0220 + 2X, 'PHASE (DEG)')
0221 DO 620 J=1,NSINES
0222 WRITE(7,810)W(J)/(2.0*PI),AMP(J),AMP(J)*W(J)/9.81,
0223 + W(J)*T0+360.0*PHI(J)/(2.0*PI)
0224 810 FORMAT(1X,F11.4,3X,F11.2,3X,F11.3,3X,F11.0)
0225 620 CONTINUE
0226 C
0227 C OUTPUT HISTOGRAM VALUES
0228 IF(IBC.EQ.0)GO TO 899
0229 WRITE(7,117)
0230 117 FORMAT(1X, 'HISTOGRAM DATA',/)
0231 DO 116 I=1,41
0232 VBIN=I*0.005
0233 WRITE(7,118)VBIN,NABIN(I)
0234 118 FORMAT(5X, 'BIN-',F6.3,5X, ' # ACC POINTS-',I6)
0235 116 CONTINUE
0236 899 IF(A2.EQ.0)GO TO 99999
0237 C
0238 900 CONTINUE
0239 C
0240 99999 RETURN
0241 END

```



```

0001      FUNCTION SUMSIN(N,A,WT,WDELTA,PHI)
0002      C
0003      C Calculates sum of sinusoids. From SUMSIN.FUN in
0004      C CART program. Stored by Dale W. Hiltner/ SEPT-82
0005      C
0006      REAL A(50),WT(50),WDELTA(50),PHI(50)
0007      SUMSIN=0.0
0008      DO 10 11=1,N
0009      WT(11)=WT(11)+WDELTA(11)
0010      SUMSIN=SUMSIN+A(11)*SIN(WT(11)+PHI(11))
0011      10 CONTINUE
0012      END

```

```

0001      FUNCTION FPOWER(W)
0002      C
0003      C Calculates filter amplitudes for magnitude
0004      C scaling in dBs. From POWER.FUN in CART
0005      C program. Stored by Dale W. Hiltner/ OCT-82
0006      INTEGER FILTER
0007      COMMON/FILCOM/FILTER,POLE
0008      C
0009      TEMP=(ATAN(W/POLE))/POLE
0010      GO TO (1,2,3)FILTER
0011      C
0012      C FIRST ORDER FILTER
0013      1 FPOWER=TEMP
0014      GO TO 9999
0015      C
0016      C SECOND ORDER FILTER
0017      2 PP=POLE*POLE
0018      FPOWER=(W/(PP+W*W)+TEMP)/(2.0*PP)
0019      GO TO 9999
0020      C
0021      C THIRD ORDER FILTER
0022      3 FP=POLE*POLE
0023      PI=PP+W*W
0024      FPOWER=-((W*W*W)/(PI*PI)-(5.0*W/PI+3*TEMP)*0.5)/(4.0*PP*PP)
0025      GO TO 9999
0026      C
0027      9999 END

```

PRIMS.DAT

12
5
7
9
11
13
17
19
23
29
32
37
41

DSOS.DAT

81.92,12,0.01221,0,253.0,1,0.16,2.05,0.20,0.010

AMPAG.DAT

0.08,0.09,0.09,0.10,0.10,0.13,0.13,0.12,0.13,0.11,0.11,0.13

SET ACCEL AMPS PER AMPAG.DAT? (Y/N)
0
MULTIPLE DEL FRECUENCIAS? (Y/N)
0
CALCULATE HISTOGRAM DATA? (Y/N)
1

INPUT PARAMETERS

TRUN= 81.920
NSINES= 12
FFREQ= 0.012210
FLAT= 0
DEL= 253.000
FILTER= 1
FPOLE= 0.160
FPS= 2.05
FACC= 0.200
TLOOR= 0.010

PROFILE DESCRIPTION

MAX ACCEL= 0.113
VELMAX= 0.6290
% USAGE OF TRACK=100.00
STARTING POSITION=-0.236
SCALE= 0.2502

FREQ. (HZ)	AMP1	(M/S)	AMP2	(G)	PHASE (DEG)
0.0610		0.12		0.005	0.
0.0854		0.13		0.007	253.
0.1099		0.12		0.008	117.
0.1343		0.11		0.009	40.
0.1587		0.12		0.012	293.
0.2075		0.11		0.015	186.
0.2319		0.10		0.014	79.
0.2808		0.11		0.019	332.
0.3540		0.09		0.021	226.
0.3906		0.07		0.019	119.
0.4517		0.07		0.021	12.
0.5005		0.06		0.020	265.

HISTOGRAM DATA

ACC BIN= 0.005	# ACC POINTS= 862
ACC BIN= 0.010	# ACC POINTS= 1059
ACC BIN= 0.015	# ACC POINTS= 1048
ACC BIN= 0.020	# ACC POINTS= 808
ACC BIN= 0.025	# ACC POINTS= 684
ACC BIN= 0.030	# ACC POINTS= 551
ACC BIN= 0.035	# ACC POINTS= 559
ACC BIN= 0.040	# ACC POINTS= 426

ACC BIN= 0.045	ACC POINTS=	355
ACC BIN= 0.050	ACC POINTS=	342
ACC BIN= 0.055	ACC POINTS=	313
ACC BIN= 0.060	ACC POINTS=	210
ACC BIN= 0.065	ACC POINTS=	207
ACC BIN= 0.070	ACC POINTS=	175
ACC BIN= 0.075	ACC POINTS=	180
ACC BIN= 0.080	ACC POINTS=	126
ACC BIN= 0.085	ACC POINTS=	83
ACC BIN= 0.090	ACC POINTS=	41
ACC BIN= 0.095	ACC POINTS=	26
ACC BIN= 0.100	ACC POINTS=	30
ACC BIN= 0.105	ACC POINTS=	49
ACC BIN= 0.110	ACC POINTS=	37
ACC BIN= 0.115	ACC POINTS=	21
ACC BIN= 0.120	ACC POINTS=	0
ACC BIN= 0.125	ACC POINTS=	0
ACC BIN= 0.130	ACC POINTS=	0
ACC BIN= 0.135	ACC POINTS=	0
ACC BIN= 0.140	ACC POINTS=	0
ACC BIN= 0.145	ACC POINTS=	0
ACC BIN= 0.150	ACC POINTS=	0
ACC BIN= 0.155	ACC POINTS=	0
ACC BIN= 0.160	ACC POINTS=	0
ACC BIN= 0.165	ACC POINTS=	0
ACC BIN= 0.170	ACC POINTS=	0
ACC BIN= 0.175	ACC POINTS=	0
ACC BIN= 0.180	ACC POINTS=	0
ACC BIN= 0.185	ACC POINTS=	0
ACC BIN= 0.190	ACC POINTS=	0
ACC BIN= 0.195	ACC POINTS=	0
ACC BIN= 0.200	ACC POINTS=	0
ACC BIN= 0.205	ACC POINTS=	0

APPENDIX C

DATA REDUCTION PROGRAMS

The programs used in the data reduction are listed. Their hierarchy is as follows:

Program	Subroutines	Data File	Page
H1PIKY.DHF	H1CNDY.DHF		148-151
H1CY.DHF	H1FFTC.DHF, H1TRFY.DHF H1TY.DHF, H1NFC.DHF		152-159
H1PIKP.DHF	H1CNDP.DHF, H1SUM.DHF	H1RATK.DAT	160-166

Program H1PIKY.DHF is the first program used in the data reduction process. It accesses and concatenates the 8192 points of channels 1,2,3,4, and 6 stored in the run data file into 1024 points. It then prints out the statistics of each channel and outputs the 1024 points of channels 2,3,4, and 6 into separate files. (The statistics are output in count units so the calibrations of Chapter 3 must be used to convert them to engineering units. The position data is not used beyond finding its statistics.) Program H1CY.DHF then accesses these files, performs the FFT, and outputs the desired response results. A sample run is shown.

Program H1PIKP.DHF is used to generate statistics for groups of data files. Up to 30 data files with the same first three letters in their filenames, and therefore from the same test session, can be processed in one run. All points contained in full blocks are used in the calculations. (If the last block in the data file is not full, the data in this block is not used.) Individual statistics are output as with H1PIKY.DHF, but in engineering units. The file H1RATK.DAT stores the position, velocity,

acceleration, and command RMS values from the no subject runs, or any other baseline values desired. These values are then used to produce the RMS ratios which are output in summary form. A sample input and output is shown.

```

0001      PROGRAM HIFINY
C
C   Author: Jen-Nuans Nuans/Dale W. Hiltner
C   Creation Date: 22-APR-82/SEPT-82
C
C   Pick up sled data(5 channels, 256 word record, 385 blocks)
C   Currently assumes input data file is 10240 ticks long, with data
C   points per time tick
C   Program writes out a data point for every tick, for one specified
C   channel of data ==> can run HICY to reduce data.
C
0002      DIMENSION INBUF(256),IOBUF(256)
0003      REAL JSCALE,TRUN
0004      INTEGER OUT1(10240)
0005      INTEGER PRFILE(30)
0006      LOGICAL*1 ZFILE(10),FILEP(3),FILEN(120),FILEE(4)
0007      COMMON/CM1/IBR,II9
0008      COMMON/CM2/JSCALE,TRUN
0009      DATA FILEE/' ','C','V','4'//
C
0010      WRITE(7,100)
0011      100 FORMAT(' Pick up input/output sled data. Channels 1,2,3,4,6. ',/,
+ ' Currently accepts maximum of 10240 samples per channel ')
C
C
0012      WRITE(7,121)
0013      121 FORMAT(/,2X,'ENTER JSCALE FACTOR,RUN TIME TRUN')
0014      READ(5,123)JSCALE,TRUN
0015      123 FORMAT(F6.2,F9.4)
C
0016      101 WRITE(7,102)
0017      102 FORMAT(/,2X,'ENTER INPUT FILE & LETTER CODE (XXXXXX.CV4)')
0018      READ(5,120)((FILEP(I),I=1,3),(FILEN(I),I=1,3))
0019      120 FORMAT(3A1,3A1)
C
C   122      FORMAT(/,2X,'ENTER #FILES TO BE REDUCED')
C   READ(5,124)NFILE
C   124      FORMAT(I2)
C   WRITE(7,126)
C   126      FORMAT(/,2X,'ENTER PR FILE #, DATA FILE NUMBER.
+ ONE SET PER LINE')
C   DO 127 I7=1,NFILE
C   READ(5,128)PRFILE(I7),(FILEN(I),I=(I7-1)*3+1,(I7-1)*3+3)
C   128      FORMAT(I3,3A1)
C   127      CONTINUE
0020      NFILE=1
0021      INC=0
C
0022      DO 710 I7=1,NFILE
0023      INF=I7
0024      DO 720 I2=1,3
0025      720 ZFILE(I2)=FILEP(I2)
0026      DO 740 I4=1,3
0027      IN=3+I4
0028      INC=INC+1
0029      740 ZFILE(IN)=FILEN(INC)
0030      DO 760 I6=1,4
0031      IE=6+I6
0032      760 ZFILE(IE)=FILEE(I6)
C
C
0033      CALL ASSIGN(1,ZFILE,10)
0034      DEFINE FILE 1 (0,256,U,NREC)
C
0035      WRITE(7,770)(ZFILE(I),I=1,10)
0036      770 FORMAT('0',2X,'CURRENT FILE',3X,10A1)
0037      DO 600 I9=1,6
C
C   DEFINE FILE 1 (0,256,U,NREC)
C
0038      IF(I9.EQ.5)GO TO 600
0040      II9=I9
C

```

```

0041      DO 130 K=1,10240
0042 130 OUT1(K)=2048      ! Initialization
0043      NREC=2
0044      DO 150 J=1,384
0045      READ(1,NREC,END=200) INBUF
0046      ISAVE=32*(J-1)
0047      DO 140 I=1,32
0048      I1=8*(I-1)+19
0049      OUT1(ISAVE+I)=INBUF(I1)
0050 140 CONTINUE
0051 150 CONTINUE
      C
      C 200      IBR=INT((J-2)/8.)+1
0052 200 IBR=INT((J-1)/8.0)
0053      IF(I19.GT.1)GO TO 220
0055      WRITE(7,210) J-1,IBR
0056 210 FORMAT(2X,I4,' Records(256 words/record) => ',I2,' Records')
      C
      C CALL H1CNDY TO CONCATENATE DATA
      C
0057 220 CALL H1CNDY(OUT1)
      C
0058 600 CONTINUE
      C
0059      CALL CLOSE(1)
      C
0060 710 CONTINUE
      C
0061      END
H1FIKY

```

```

0001      SUBROUTINE H1CNDY(OUT1)
C
C Author: Jen-Kuang Huans/Dale W. Hiltner
C Creation Date: 8-JUN-82/ SEPT-82
C
C Concatenates sled data(1 channel, 256 word record, up to 48 blocks)
C Method: Window filter (Window: two standard deviation)
C Currently assumes input data file has max 12288 ticks long, with
C data points per time tick
C Program writes out 1024 data points, for two
C channels of data ==> can run H1CY to output final results.
C
0002      INTEGER OUT(1024)
0003      INTEGER OUT1(10240)
0004      REAL JSCALE,TRUN
0005      COMMON/CM1/IBR,II9
0006      COMMON/CM2/JSCALE,TRUN
C
0007      IF(II9.GT.1)GO TO 120
0009      WRITE(7,100)
0010      100 FORMAT(' Sled data concatenation. Channels 1,2,3,4,6.,',/,
+ ' Currently accepts maximum of 10240(256*40) samples ')
C
0011      120 INTVLE=INT(256*IBR/1024+0.)
0012      IF(II9.GT.1)GO TO 130
0014      WRITE(7,116) INTVLE
0015      116 FORMAT(2X,'Pick up one point from every ',I2,' points.')
0016      WRITE(7,125)
0017      125 FORMAT(' ')
C
0018      130      NREC=1
0019              IR=0
0020              SUM=0.
0021              RMSE=0.
C
C PERFORM WINDOW FILTERING
C
0022      DO 190 L=1,1024
0023          TSUM1=0.
0024          TSUM2=0.
0025          TRMSE=0.
0026          IA=0
0027          IC=INTVLE*(L-1)
0028          DO 170 M=1,INTVLE
0029              P1=OUT1(IC+M)-2048
0030              TSUM1=TSUM1+OUT1(IC+M)-2048
0031              TRMSE=TRMSE+P1*P1
0032      170 CONTINUE
0033          TAUG=TSUM1/INTVLE
0034          TRMS=SQRT(TRMSE/INTVLE)
0035          TSTD=SQRT(TRMS**2-TAUG**2)
0036          ULIM=TAUG+TSTD*2
0037          LLIM=TAUG-TSTD*2
0038          DO 180 N=1,INTVLE
0039              P1=OUT1(IC+N)-2048
0040              IF(P1.GT.ULIM) GOTO 180
0042              IF(P1.LT.LLIM) GOTO 180
0044              TSUM2=TSUM2+OUT1(IC+N)-2048
0045              IA=IA+1
0046      180 CONTINUE
0047              OUT(L)=INT(2048+TSUM2/IA)
C              OUT(L)=2048+TSUM2/IA
0048              IR=IR+INTVLE-IA
0049      190 CONTINUE
C
C FIND STATISTICS FOR 1024 POINTS
C
0050      DO 305 I=1,1024
0051          P1=OUT(I)-2048
0052          SUM=SUM+OUT(I)-2048
0053          RMSE=RMSE+P1*P1
0054      305 CONTINUE
0055          AVG=SUM/1024
0056          RMS=SQRT(RMSE/1024)
0057          STD=SQRT(RMS**2-AVG**2)
0058          RPC=IR*100./INTVLE/1024.
C
0059      GO TO(510,520,530,540,550,560)II9

```



```

C
C OUTPUT 1024 POINTS TO RESPECTIVE FILES
C
0060 510 WRITE(7,515)
0061 GO TO 307
0062 520 WRITE(7,525)
C
0063 CALL ASSIGN(2,'H1TMPV.DAT')
0064 DEFINE FILE 2 (1,2048,U,IREC2)
0065 WRITE(2'1)OUT
0066 CALL CLOSE(2)
C
0067 GO TO 307
0068 530 CALL ASSIGN(2,'H1TMPA.DAT')
0069 DEFINE FILE 2 (1,2048,U,IREC2)
0070 WRITE(2'1)OUT
0071 CALL CLOSE(2)
C
0072 WRITE(7,535)
0073 GO TO 307
0074 540 WRITE(7,545)
C
0075 CALL ASSIGN(2,'H1TMPG.DAT')
0076 DEFINE FILE 2 (1,2048,U,IREC2)
0077 WRITE(2'1)OUT
0078 CALL CLOSE(2)
C
0079 GO TO 307
0080 550 WRITE(7,555)
0081 GO TO 9999
0082 560 CALL ASSIGN(2,'H1TMPJ.DAT')
0083 DEFINE FILE 2 (1,2048,U,IREC2)
0084 WRITE(2'1)OUT
0085 CALL CLOSE(2)
0086 WRITE(7,565)
C
0087 307 WRITE(7,309) RPC,RMS,AVG,STD
0088 309 FORMAT(10X,F10.2,5X,3(F8.4,2X))
C
C OUTPUT STATISTICS
C
0089 WRITE(7,570)
0090 570 FORMAT(' ')
C
0091 515 FORMAT(' POSITION',6X,'REJECT(Z)',3X,'RMS',7X,'AVG',7X,'STD')
0092 525 FORMAT(' VELOCITY',6X,'REJECT(Z)',3X,'RMS',7X,'AVG',7X,'STD')
0093 535 FORMAT(' ACCEL ',6X,'REJECT(Z)',3X,'RMS',7X,'AVG',7X,'STD')
0094 545 FORMAT(' COMMAND ',6X,'REJECT(Z)',3X,'RMS',7X,'AVG',7X,'STD')
0095 555 FORMAT(' !! ERROR !! II9=5. EXIT COND'D')
0096 565 FORMAT(' JOYSTICK',6X,'REJECT(Z)',3X,'RMS',7X,'AVG',7X,'STD')
C
0097 9999 RETURN
0098 400 END
HICNDY

```

```

0001      PROGRAM H1CY
      C
      C
      C Written by Dale W. Hiltner. Sept-82
      C Computes FFT from files created by H1PKIY and
      C outputs response results.
      C
      C
0002      INTEGER I1R(1024)
0003      REAL A1R,JSCALE,TRUN
0004      REAL AMP1(60),PHASE1(60),AI1R(1024)
0005      COMPLEX A1C(1024)
0006      COMMON/CM2/JSCALE,TRUN
      C
      C
0007      WRITE(7,02)
0008      02 FORMAT(5X,'INPUT JSCALE,RUN TIME TRUN')
0009      READ(7,04)JSCALE,TRUN
0010      04 FORMAT(F9.4,F9.4)
      C
0011      CALL CLOSE(5)
0012      CALL CLOSE(6)
0013      CALL CLOSE(7)
0014      CALL CLOSE(8)
      C
0015      II=1
0016      900 GO TO (10,20,30,40,50)II
      C
      C RRAD IN 1024 DATA POINTS AND SCALE PER CALIBRATION BEFORE GOING
      C TO FFT. THEN STORE FIRST 60 AMPLITUDE AND PHASE VALUES
      C IN ANOTHER FILE.
      C
0017      10 DEFINE FILE 1 (1,2048,U,I1REC1)
0018      CALL ASSIGN(1,'H1TMPA.DAT')
0019      READ(1'1)I1R
0020      DO 12 I=1,1024
0021      12 AI1R(I)=I1R(I)*0.01
0022      GO TO 100
0023      210 CALL CLOSE(1)
0024      CALL ASSIGN(5,'H1OUTA.DAT')
0025      WRITE(5,*)(AMP1(I),I=1,60)
0026      WRITE(5,*)(PHASE1(I),I=1,60)
0027      CALL CLOSE(5)
0028      II=2
0029      GO TO 900
      C
0030      20 DEFINE FILE 2 (1,2048,U,I1REC2)
0031      CALL ASSIGN(2,'H1TMPJ.DAT')
0032      READ(2'1)I1R
0033      DO 22 I=1,1024
0034      22 AI1R(I)=I1R(I)*0.003998/JSCALE
0035      GO TO 100
0036      220 CALL CLOSE(2)
0037      CALL ASSIGN(6,'H1OUTJ.DAT')
0038      WRITE(6,*)(AMP1(I),I=1,60)
0039      WRITE(6,*)(PHASE1(I),I=1,60)
0040      CALL CLOSE(6)
0041      II=3
0042      GO TO 900
      C
0043      30 DEFINE FILE 3 (1,2048,U,I1REC3)
0044      CALL ASSIGN(3,'H1TMPV.DAT')
0045      READ(3'1)I1R
0046      DO 32 I=1,1024
0047      32 AI1R(I)=I1R(I)*0.002722
0048      GO TO 100
0049      230 CALL CLOSE(3)
0050      CALL ASSIGN(7,'H1OUTV.DAT')
0051      WRITE(7,*)(AMP1(I),I=1,60)
0052      WRITE(7,*)(PHASE1(I),I=1,60)
0053      CALL CLOSE(7)
0054      II=4
0055      GO TO 900
      C
0056      40 DEFINE FILE 4 (1,2048,U,I1REC4)
0057      CALL ASSIGN(4,'H1TMPD.DAT')
0058      READ(4'1)I1R
0059      DO 42 I=1,1024
0060      42 AI1R(I)=I1R(I)*0.003998
0061      GO TO 100

```

```

0062      240  CALL CLOSE(4)
0063      CALL ASSIGN(B,'H1OUTC.DAT')
0064      WRITE(B,*)(AMP1(I),I=1,60)
0065      WRITE(B,*)(PHASE1(I),I=1,60)
0066      CALL CLOSE(B)
0067      II=5
0068      GO TO 900

C
0069      100  DO 110 I1=1,1024
0070          AIR=AI1R(I1)
0071          A1C(I1)=CMPLX(A1R,0.0)
0072      110  CONTINUE

C
C  CALL FFT SUBROUTINE
C
0073      CALL H1FTCY(A1C,10,1024)

C
C  CALCULATE AMPLITUDE AND PHASE FROM FFT OUTPUTS
C
0074          DO 120 I2=1,60
0075          AMP1(I2)=SQRT(REAL(A1C(I2+1))**2+AIMAG(A1C(I2+1))**2)
0076          PHASE1(I2)=ATAN2(REAL(A1C(I2+1)),AIMAG(A1C(I2+1)))*57.29577949
0077      120  CONTINUE

C
0078      GO TO (210,220,230,240)II

C
C  CALL H1TRFY TO CALCULATE AND OUTPUT RESPONSE RESULTS
C
0079      50  CALL H1TRFY

C
C
C
C
0080      STOP
0081      9999  END
H1CY

```

```

0001      SUBROUTINE H1FTCY(A,M,N)
      C
      C
      C Fortran FFT subroutine. Stored by Dale W. Hiltner 16-sept-82
      C
      C
0002      COMPLEX A(N),U,W,T
0003      INTEGER N,M
      C
      C
0004      N=2**M
0005      NV2=N/2
0006      NM1=N-1
0007      J=1
0008      DO 7 I=1,NM1
0009      IF(I.GE.J)GO TO 5
0010      T=A(J)
0011      A(J)=A(I)
0012      A(I)=T
0013
0014      5 K=NV2
0015      6 IF(K.GE.J)GO TO 7
0016      J=J-K
0017      K=K/2
0018      GO TO 6
0019      7 J=J+K
0020
0021      PI=3.141593  !653589793
0022      DO 20 L=1,M
0023      LE=2**L
0024      LE1=LE/2
0025      U=(1.0,0.0)
0026      W=CMPLX(COS(PI/LE1),SIN(PI/LE1))
0027      DO 20 J=1,LE1
0028      DO 10 I=J,N,LE
0029      IP=I+LE1
0030      T=A(IP)*U
0031      A(IP)=A(I)-T
0032      10 A(I)=A(I)+T
0033      20 U=U*W
0034      RETURN
0035      END
H1FTCY

```

```

0001      SUBROUTINE H1TRFY
C
C      Author: Jen-Kuang Huang/Dale W. Hiltner
C      Date:   14-Mar-82/AUG-82
C
C      Accesses phase and amplitude output from H1CY and
C      outputs transfer function and frequency spectrum data.
C
0002      REAL PHASE(60),FREQ(60),FREQR(60),LFREQR(60)
0003      REAL GAINDB(60),LGAIN(60)
0004      REAL AMPA(60),AMPJ(60),AMPV(60),AMPC(60)
0005      REAL PHASEA(60),PHASEJ(60),PHASEV(60),PHASEC(60)
0006      REAL JSCALE,TRUN,NFC(25)
0007      COMMON/CM2/JSCALE,TRUN
C
C      ACCESS DATA FILES
C
0008      CALL ASSIGN(5,'H1OUTA.DAT')
0009      READ(5,*)(AMPA(I),I=1,60)
0010      READ(5,*)(PHASEA(I),I=1,60)
0011      CALL CLOSE(5)
C
0012      CALL ASSIGN(6,'H1OUTJ.DAT')
0013      READ(6,*)(AMPJ(I),I=1,60)
0014      READ(6,*)(PHASEJ(I),I=1,60)
0015      CALL CLOSE(6)
C
0016      CALL ASSIGN(7,'H1OUTV.DAT')
0017      READ(7,*)(AMPV(I),I=1,60)
0018      READ(7,*)(PHASEV(I),I=1,60)
0019      CALL CLOSE(7)
C
0020      CALL ASSIGN(8,'H1OUTC.DAT')
0021      READ(8,*)(AMPC(I),I=1,60)
0022      READ(8,*)(PHASEC(I),I=1,60)
0023      CALL CLOSE(8)
C
C
0024      DO 5 I=1,60
C
C      COMPUTE TRANSFER FUNCTION DATA
C
0025      IF(AMPJ(I).EQ.0.0)GO TO 11
0027      IF(AMPA(I).EQ.0.0)GO TO 11
C
0029      GAINDB(I)=20.*ALOG10(AMPJ(I)/AMPA(I))
0030      GO TO 12
0031      11 GAINDB(I)=0.0
0032      12 LGAIN(I)=GAINDB(I)/20.0
0033      PHASE(I)=PHASEJ(I)-PHASEA(I)+180.0
0034      IF(PHASE(I).LT.-360.0)PHASE(I)=PHASE(I)+360.0
0036      IF(PHASE(I).GT.0.0)PHASE(I)=PHASE(I)-360.0
0038      FREQ(I)=I/TRUN
0039      FREQR(I)=I/TRUN*6.2832
0040      LFREQR(I)=ALOG10(FREQR(I))
0041      5 CONTINUE
C
C      OUTPUT TRANSFER FUNCTION DATA
C
0042      WRITE(7,220)
0043      220 FORMAT(//,5X,'N',2X,'NF',4X,'FREQH',4X,'FREQR',4X,'LOG F',8X,
+      'GAINDB',7X,'LGAIN',8X,'PHASE',/)
0044      WRITE(7,230)(I,I,FREQ(I),FREQR(I),LFREQR(I),
+      GAINDB(I),LGAIN(I),PHASE(I),I=1,45)
0045      230 FORMAT(2X,2I4,2X,F7.3,2X,F7.3,2X,F7.3,3F12.3)
C

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```

C
C CALL HITY FOR DATA REORDRING
C
0046 CALL HITY(AMPA,AMPA)
0047 CALL HITY(AMPJ,AMPJ)
0048 CALL HITY(AMPV,AMPV)
0049 CALL HITY(AMPC,AMPC)
0050 CALL HITY(PHASEA,PHASEA)
0051 CALL HITY(PHASEJ,PHASEJ)
0052 CALL HITY(PHASEV,PHASEV)
0053 CALL HITY(PHASEC,PHASEC)
C
0054 CALL HINFC(NFC,NFC)
C
C
C OUTPUT FREQUENCY SPECTRUM DATA
C
0055 WRITE(7,260)
0056 260 FORMAT(//,5X,'I',3X,'NF',7X,'INAMPACC',8X,'INPHS',5X,
+          'OUTAMPJOY',7X,'OUTPHS',4X,'VELAMP',10X,'VELPHS',4X,
+          'COMAMP',10X,'COMPHS',/)
0057 WRITE(7,270)(I,NFC(I),AMPA(I),PHASEA(I),
+          AMPJ(I),PHASEJ(I),
+          AMPV(I),PHASEV(I),
+          AMPC(I),PHASEC(I),I=1,25)
0058 270 FORMAT(4X,I2,2X,F5.1,1X,3X,E12.5,2X,F9.4,3X,E12.5,2X,F9.4,
+          3X,E12.5,2X,F9.4,3X,E12.5,2X,F9.4)
C
C
C
0059 RETURN
0060 END
HITRFY

```

```

0001      SUBROUTINE HITY(AIN,ADUT)
      C
      C C Output reordering. Remnant average calculations
      C C
0002      REAL AIN(60),ADUT(60)
      C
      C
0003      ADUT(1)=(AIN(1)+AIN(2)+AIN(3)+AIN(4))/4.0
0004      DO 20 I=2,10
0005      20 ADUT(I)=AIN(I+3)
      C
0006      ADUT(11)=(AIN(14)+AIN(15)+AIN(16))/3.0
0007      DO 40 I=12,14
0008      40 ADUT(I)=AIN(I+5)
      C
0009      ADUT(15)=(AIN(20)+AIN(21)+AIN(22))/3.0
0010      ADUT(16)=AIN(23)
0011      ADUT(17)=(AIN(24)+AIN(25)+AIN(26)+AIN(27)+AIN(28))/5.0
0012      ADUT(18)=AIN(29)
0013      ADUT(19)=(AIN(30)+AIN(31))/2.0
0014      ADUT(20)=AIN(32)
0015      ADUT(21)=(AIN(33)+AIN(34)+AIN(35)+AIN(36))/4.0
0016      ADUT(22)=AIN(37)
0017      ADUT(23)=(AIN(38)+AIN(39)+AIN(40))/3.0
0018      ADUT(24)=AIN(41)
0019      ADUT(25)=(AIN(42)+AIN(43)+AIN(44))/3.0
0020      RETURN
0021      END
HITY

```

```

0001      SUBROUTINE H1NFC(A1,A)
      C
      C C Set up N values for frequency spectrum output.
      C C
0002      REAL A(25),A1(25)
0003      A(1)=2.5
0004      DO 10 I=2,10
0005      10 A(I)=I+3.0
0006      A(11)=15.0
0007      A(12)=17.0
0008      A(13)=18.0
0009      A(14)=19.0
0010      A(15)=21.0
0011      A(16)=23.0
0012      A(17)=26.0
0013      A(18)=29.0
0014      A(19)=30.5
0015      A(20)=32.0
0016      A(21)=34.5
0017      A(22)=37.0
0018      A(23)=39.0
0019      A(24)=41.0
0020      A(25)=43.0
0021      RETURN
0022      END
H1NFC

```

.RUN HIPIKY

Pick up input/output sled data. Channels 1,2,3,4,6.
Currently accepts maximum of 10240 samples per channel

ENTER JSCALE FACTOR,RUN TIME TRUN
2.1,81.92

ENTER INPUT FILE & LETTER CODE (XXXXXX.CV4)
HI1506

CURRENT FILE HI1506.CV4
256 Records(256 words/record) => 32 Records
Sled data concatenation. Channels 1,2,3,4,6.
Currently accepts maximum of 10240(256x40) samples
Pick up one point from every 8 points.

POSITION	REJECT(X)	RMS	AVG	STD
	5.54	366.2113	-73.0781	358.8458
VELOCITY	REJECT(X)	RMS	AVG	STD
	6.51	98.9566	-23.1777	96.2039
ACCEL	REJECT(X)	RMS	AVG	STD
	6.97	58.4783	-4.1484	58.3310
COMMAND	REJECT(X)	RMS	AVG	STD
	0.78	49.0192	4.6982	48.7935
JOYSTICK	REJECT(X)	RMS	AVG	STD
	3.76	85.1156	13.8730	83.9774

STOP --

.RUN HICY

INPUT JSCALE,RUN TIME TRUN
2.1,81.92

N	NF	FREQH	FREQR	LOG F	GAINDB	LGAIN	PHASE
1	1	0.012	0.077	-1.115	11.859	0.593	-2.409
2	2	0.024	0.153	-0.814	13.306	0.665	-185.733
3	3	0.037	0.230	-0.638	10.653	0.533	-309.691
4	4	0.049	0.307	-0.513	9.321	0.466	-112.378
5	5	0.061	0.383	-0.416	4.831	0.242	-20.619
6	6	0.073	0.460	-0.337	-9.560	-0.478	-123.847
7	7	0.085	0.537	-0.270	9.920	0.496	-11.427
8	8	0.098	0.614	-0.212	0.476	0.024	-275.422
9	9	0.110	0.690	-0.161	0.173	0.009	-104.327
10	10	0.122	0.767	-0.115	1.763	0.088	-207.855
11	11	0.134	0.844	-0.074	-1.527	-0.076	-91.992
12	12	0.146	0.920	-0.036	-11.614	-0.581	-196.790
13	13	0.159	0.997	-0.001	-3.277	-0.164	-111.106
14	14	0.171	1.074	0.031	-0.039	-0.002	39.861
15	15	0.183	1.150	0.061	5.334	0.267	-155.023
16	16	0.195	1.227	0.089	-6.209	-0.310	-255.527
17	17	0.208	1.304	0.115	-5.477	-0.274	-110.918
18	18	0.220	1.381	0.140	6.256	0.313	-194.005
19	19	0.232	1.457	0.164	-7.620	-0.381	-149.175
20	20	0.244	1.534	0.186	-0.286	-0.014	-256.559
21	21	0.256	1.611	0.207	-1.779	-0.089	-202.910
22	22	0.269	1.687	0.227	-13.788	-0.689	-261.260
23	23	0.281	1.764	0.247	-8.311	-0.416	-123.637
24	24	0.293	1.841	0.265	-11.249	-0.562	-270.350
25	25	0.305	1.917	0.283	-21.572	-1.079	-241.995
26	26	0.317	1.994	0.300	-6.288	-0.314	-254.745
27	27	0.330	2.071	0.316	-12.571	-0.629	-237.121
28	28	0.342	2.148	0.332	-10.857	-0.543	-235.816
29	29	0.354	2.224	0.347	-17.166	-0.858	-122.504
30	30	0.366	2.301	0.362	-3.425	-0.171	110.388
31	31	0.378	2.378	0.376	-7.832	-0.392	-271.848
32	32	0.391	2.454	0.390	-11.362	-0.568	-158.459
33	33	0.403	2.531	0.403	-8.445	-0.422	80.213
34	34	0.415	2.608	0.416	-5.589	-0.279	111.461
35	35	0.427	2.684	0.429	-7.037	-0.352	-309.420
36	36	0.439	2.761	0.441	-3.497	-0.175	-238.041
37	37	0.452	2.838	0.453	-17.499	-0.875	-169.986
38	38	0.464	2.915	0.465	-9.906	-0.495	-225.813
39	39	0.476	2.991	0.476	-10.603	-0.530	-259.397
40	40	0.488	3.068	0.487	-6.041	-0.302	-245.417

41	41	0.500	3.145	0.498	-15.469	-0.773	-167.939
42	42	0.513	3.221	0.508	-7.248	-0.362	96.345
43	43	0.525	3.298	0.518	-11.626	-0.581	-106.323
44	44	0.537	3.375	0.528	-12.078	-0.604	-243.301
45	45	0.549	3.451	0.538	-13.509	-0.675	-212.177

I	NF	INAMPACC	INPHS	OUTAMPJOY	OUTPHS	VELAMP	VELPHS
1	2.5	0.66695E+01	60.2222	0.25057E+02	-2.3307	0.16548E+02	-11.7246
2	5.0	0.24067E+02	79.6106	0.41973E+02	-121.0087	0.16966E+02	148.8168
3	6.0	0.91114E+01	-132.0222	0.30311E+01	-75.8689	0.76074E+01	128.5295
4	7.0	0.10824E+02	-49.6864	0.33915E+02	118.8866	0.21012E+02	74.9835
5	8.0	0.93956E+01	-81.6831	0.99246E+01	-177.1055	0.43357E+01	71.1522
6	9.0	0.22955E+02	-79.9227	0.23416E+02	-4.2493	0.24299E+02	-3.3412
7	10.0	0.73120E+01	92.6190	0.89580E+01	64.7642	0.11401E+02	-127.4970
8	11.0	0.25363E+02	159.1630	0.21273E+02	-112.8292	0.22157E+02	-94.0382
9	12.0	0.59945E+01	108.0695	0.15742E+01	91.2791	0.45189E+01	8.6655
10	13.0	0.40620E+02	73.7615	0.27855E+02	142.6557	0.13312E+02	166.0447
11	15.0	0.83275E+01	-78.1470	0.64369E+01	-21.7099	0.87842E+01	3.4708
12	17.0	0.48100E+02	-11.5155	0.25604E+02	57.5666	0.22678E+02	81.9786
13	18.0	0.28506E+01	-89.1315	0.58582E+01	-102.1369	0.71245E+01	-66.0477
14	19.0	0.59440E+02	-104.3898	0.24722E+02	-73.5648	0.12970E+02	-12.4083
15	21.0	0.81336E+01	87.9839	0.42859E+01	27.7408	0.44437E+01	-21.2059
16	23.0	0.73136E+02	154.6477	0.28091E+02	-148.9891	0.23776E+02	-114.4484
17	26.0	0.11571E+02	59.7801	0.33207E+01	-8.2254	0.64729E+01	-3.0979
18	29.0	0.91271E+02	61.6987	0.12648E+02	119.1949	0.32483E+02	151.7377
19	30.5	0.96311E+01	9.3120	0.46275E+01	108.5822	0.49632E+01	-14.3616
20	32.0	0.79010E+02	-15.8167	0.21360E+02	5.7247	0.22982E+02	88.7080
21	34.5	0.13236E+02	-39.2202	0.64904E+01	51.8331	0.60775E+01	-60.4718
22	37.0	0.12468E+03	-115.5391	0.16629E+02	-105.5256	0.24003E+02	-23.7358
23	39.0	0.17577E+02	20.0071	0.63231E+01	-43.5354	0.52098E+01	23.4820
24	41.0	0.10299E+03	153.1838	0.17352E+02	165.2443	0.15809E+02	-115.9404
25	43.0	0.12054E+02	3.8769	0.40758E+01	99.4505	0.58476E+01	-25.7926

STOP --

COMAMP	COMPMS
0.22681E+02	-4.2415
0.40994E+02	-19.7886
0.38364E+01	-85.9229
0.35627E+02	-98.2199
0.86879E+01	-179.0743
0.38705E+02	-178.6416
0.82803E+01	64.8018
0.36411E+02	84.6310
0.19444E+01	126.5418
0.39206E+02	-10.1903
0.58240E+01	-28.3612
0.34272E+02	-107.2774
0.42274E+01	-106.7055
0.36801E+02	170.1163
0.43229E+01	18.0284
0.36560E+02	71.6213
0.30703E+01	47.1517
0.37374E+02	-25.5031
0.44440E+01	97.1317
0.31863E+02	-100.3662
0.54566E+01	40.1723
0.37331E+02	161.9321
0.56506E+01	-59.5094
0.31416E+02	71.4905
0.33353E+01	57.9965

```

0001        PROGRAM H1PINP
C
C    Author: Jen-Kuang Huans/Dale W. Hiltner
C    Creation Date: 22-APR-82/ Sept-82
C
C    Pick up sled data(5 channels, 256 word record, 385 blocks)
C    Currently assumes input data file is 10240 ticks long.
C    Program finds RMS ratios, AVG, STD for all points in full
C    blocks. Can access groups of files with one program run.
C    Uses H1SUM.DHF to output data by channel number.
C    This program is a variant of H1PIKY.DHF
C
0002        DIMENSION INBUF(256),IOBUF(256)
0003        REAL DURT(30),JSCALE
0004        REAL POSR(30),FOSS(30),POSA(30),POSRAT(30)
0005        REAL VELR(30),VELS(30),VELA(30),VELRAT(30)
0006        REAL ACCR(30),ACCS(30),ACCA(30),ACCRAT(30)
0007        REAL COMR(30),COMS(30),COMA(30),COMRAT(30)
0008        REAL JOYR(30),JOYS(30),JOYA(30)
0009        INTEGER OUT1(10240),PRFILE(30)
0010        LOGICAL*1 ZFILE(10),FILEP(3),FILEN(120),FILEE(4)
0011        COMMON/CM1/OUT1,IBR,II9,JSCALE,FRFILE
0012        COMMON/SUMP/POSR,POSS,POSA,POSRAT
0013        COMMON/SUMV/VELR,VELS,VELA,VELRAT
0014        COMMON/SUMA/ACCR,ACCS,ACCA,ACCRAT
0015        COMMON/SUMC/COMR,COMS,COMA,COMRAT
0016        COMMON/SUMJ/JOYR,JOYS,JOYA
0017        COMMON/WFILE/FILEP,FILEN,FILEE
0018        COMMON/CFILE/NFILE,INF,DURT
0019        DATA FILEE/' ','C','U','4'/
C
0020        WRITE(7,100)
0021        100 FORMAT(' Pick up input/output sled data. Channels 1,2,3,4,6.',/
+        '        Currently accepts maximum of 10240 samples per channel ')
C
C
0022        WRITE(7,121)
0023        121 FORMAT(/,2X,'ENTER JSCALE FACTOR')
0024        READ(5,123)JSCALE
0025        123 FORMAT(F6.2)
C
C ENTER MULTIPLE FILES
C
0026        101 WRITE(7,102)
0027        102 FORMAT(/,2X,'ENTER INPUT FILE 3 LETTER CODE')
0028        READ(5,120)(FILEP(I),I=1,3)
0029        120 FORMAT(3A1)
0030        WRITE(7,122)
0031        122 FORMAT(/,2X,'ENTER #FILES TO BE REDUCED')
0032        READ(5,124)NFILE
0033        124 FORMAT(I2)
0034        WRITE(7,126)
0035        126 FORMAT(/,2X,'ENTER PR FILE #, DATA FILE NUMBER.
+        '        ONE SET PER LINE')
C
0036        DO 127 I7=1,NFILE
0037        READ(5,128)PRFILE(I7),(FILEN(I),I=(I7-1)*3+1,(I7-1)*3+3)
0038        128 FORMAT(I3,3A1)
0039        127 CONTINUE
0040        INC=0
C
0041        DO 710 I7=1,NFILE
0042        INF=I7
0043        DO 720 I2=1,3
0044        ZFILE(I2)=FILEP(I2)
0045        DO 740 I4=1,3
0046        IN=3+I4
0047        INC=INC+1
0048        740 ZFILE(IN)=FILEN(INC)
0049        DO 760 I6=1,4
0050        IE=6+I6
0051        760 ZFILE(IE)=FILEE(I6)
C
C
0052        CALL ASSIGN(1,ZFILE,10)
0053        DEFINE FILE 1 (0,256,U,NREC)
C
0054        WRITE(7,770)(ZFILE(I),I=1,10)
0055        770 FORMAT('0',2X,'CURRENT FILE',3X,10A1)
0056        DO 600 I9=1,6
C

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0057      IF(I9.EQ.5)GO TO 600
0059      II9=I9
C
0060      DO 130 K=1,10240
0061 130  OUT1(K)=2048          ! Initialization
0062      NREC=2
0063      DO 150 J=1,384
0064      READ(1,NREC,END=200) INBUF
0065      ISAVE=32*(J-1)
0066      DO 140 I=1,32
0067      I1=8*(I-1)+I9
0068      OUT1(ISAVE+I)=INBUF(I1)
0069 140  CONTINUE
0070 150  CONTINUE
C
C 200      IBR=INT((J-2)/8.)+1
0071 200  IBR=INT((J-1)/8.0)
0072      IF(II9.GT.1)GO TO 220
0074      WRITE(7,210) J-1,IBR
0075 210  FORMAT(2X,I4,' Records(256 words/record) => ',I2,' Records')
C
C CALL H1CNDP TO CALCULATE STATISTICS
C
0076 220  CALL H1CNDP
C
0077 600  CONTINUE
C
0078      CALL CLOSE(1)
0079      DURT(INF)=IBR*256*0.01
C
0080 710  CONTINUE
C
C CALL H1SUM TO OUTPUT STATISTICS BY CHANNEL NUMBER
C
0081      CALL H1SUM
C
0082      END
H1PIKP

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0001      SUBROUTINE H1CNDP
C
C   Author: Jen-Kuang Huang/ Dale W. Hiltner
C   Creation Date: 8-JUN-82/ SEPT-82
C
C   Program finds RMS ratios, AVG, STD of 256*ibr points.
C   The no-subject RMS values are stored in file H1RATK.DAT
C   This is the maximum number of points contained in all
C   full blocks of the file. This program is a variant of
C   H1CNDY.DHF
C
C
0002      DIMENSION INBUF(256)
0003      REAL POSR(30),POSS(30),POSA(30),POSRAT(30)
0004      REAL VELR(30),VELS(30),VELA(30),VELRAT(30)
0005      REAL ACCR(30),ACCS(30),ACCA(30),ACCRAT(30)
0006      REAL COMR(30),COMS(30),COMA(30),COMRAT(30)
0007      REAL JOYR(30),JOYS(30),JOYA(30),DURT(30),JSCALE,IRATK(17,4)
0008      LOGICAL*1 FILEP(3),FILEN(120),FILEE(4)
0009      INTEGER OUT1(10240),PRFILE(30)
0010      COMMON/SUMF/POSR,POSS,POSA,POSRAT
0011      COMMON/SUMV/VELR,VELS,VELA,VELRAT
0012      COMMON/SUMA/ACCR,ACCS,ACCA,ACCRAT
0013      COMMON/SUMC/COMR,COMS,COMA,COMRAT
0014      COMMON/SUMJ/JOYR,JOYS,JOYA
0015      COMMON/CFILE/NFILE,INF,DURT
0016      COMMON/CM1/ OUT1,IBR,II9,JSCALE,PRFILE
C
C   FILE H1RATK.DHF CONTAINS ALL THE NO-SUBJECT RMS VALUES
C
0017      CALL ASSIGN(10,'H1RATK.DAT','NC')
0018      DO 10 I=1,17
0019      DO 20 J=1,4
0020      20 RATK(I,J)=0.0
0021      10 CONTINUE
C
0022      READ(10,*)((RATK(I,J),J=1,4),I=1,7)
0023      READ(10,*)((RATK(I,J),J=1,4),I=10,17)
C
0024      IF(II9.GT.1)GO TO 120
0026      WRITE(7,100)
0027      100 FORMAT(' Sled data reduction. Channels 1,2,3,4,6.,/,
+ ' Currently accepts maximum of 10240(256*40) samples ')
C
0028      NPNTS=IBR*256
0029      DURT(INF)=NPNTS*0.01
0030      IPR=PRFILE(INF)
C
0031      WRITE(7,116)NPNTS
0032      116 FORMAT(2X,'Analysis of 'IS,' points.')
0033      WRITE(7,125)
0034      125 FORMAT(' ')
C
0035      120      SUM=0.
0036      RMSE=0.
C
C   CALCULATE STATISTICS USING 2*VARIANCE WINDOW FILTER
C
0037      DO 305 I=1,NPNTS
0038      P1=OUT1(I)-2048
0039      SUM=SUM+OUT1(I)-2048
0040      RMSE=RMSE+P1*P1
0041      305 CONTINUE
0042      AVG=SUM/NPNTS
0043      RMS=SQRT(RMSE/NPNTS)
0044      STD=SQRT(RMS**2-AVG**2)
C
C   SCALE AND STORE VALUES IN FILES TO BE OUTPUT PER CHANNEL
C   NUMBER BY H1SUM.DHF
C
0045      GO TO(510,520,530,540,550,560)II9
0046      510      RMS=RMS*0.002
0047      STD=STD*0.002
0048      AVG=AVG*0.002
0049      WRITE(7,515)
0050      POSR(INF)=RMS
0051      POSS(INF)=STD
0052      POSA(INF)=AVG
C

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0053     POSRAT(INF)=POSR(INF)/RATK(IPR,1)
0054     GO TO 307
0055 520 RMS=RMS*0.002722
0056     STD=STD*0.002722
0057     AVG=AVG*0.002722
0058     WRITE(7,525)
0059     VELR(INF)=RMS
0060     VELS(INF)=STD
0061     VELA(INF)=AVG
0062     VELRAT(INF)=VELR(INF)/RATK(IPR,2)
0063     GO TO 307
0064 530 RMS=RMS*0.01
0065     STD=STD*0.01
0066     AVG=AVG*0.01
0067     WRITE(7,535)
0068     ACCR(INF)=RMS
0069     ACCS(INF)=STD
0070     ACCA(INF)=AVG
0071     ACCRAT(INF)=ACCR(INF)/RATK(IPR,3)
0072     GO TO 307
0073 540 RMS=RMS*0.003998
0074     STD=STD*0.003998
0075     AVG=AVG*0.003998
0076     WRITE(7,545)
0077     COMR(INF)=RMS
0078     COMS(INF)=STD
0079     COMA(INF)=AVG
0080     COMRAT(INF)=COMR(INF)/RATK(IPR,4)
0081     GO TO 307
0082 550 WRITE(7,555)
0083     GO TO 9999
0084 560 RMS=RMS*0.003998/JSCALE
0085     STD=STD*0.003998/JSCALE
0086     AVG=AVG*0.003998/JSCALE
0087     WRITE(7,565)
0088     JOYR(INF)=RMS
0089     JOYS(INF)=STD
0090     JOYA(INF)=AVG
0091     GO TO 307

C
C OUTPUT STATISTICS FOR EACH INDIVIDUAL RUN
C
0092 307 WRITE(7,309)RMS,AVG,STD
0093 309 FORMAT(15X,F8.4,2X,F8.4,2X,F8.4)
C
0094     WRITE(7,570)
0095 570 FORMAT(' ')
C
0096 515 FORMAT(' POSITION',9X,'RMS',7X,'AVG',7X,'STD')
0097 525 FORMAT(' VELOCITY',9X,'RMS',7X,'AVG',7X,'STD')
0098 535 FORMAT(' ACCEL',9X,'RMS',7X,'AVG',7X,'STD')
0099 545 FORMAT(' COMMAND',9X,'RMS',7X,'AVG',7X,'STD')
0100 555 FORMAT(' !! ERROR !! II9=5. EXIT HICNDP')
0101 565 FORMAT(' JOYSTICK',9X,'RMS',7X,'AVG',7X,'STD')
C
0102     CALL CLOSE(10)
0103 9999 RETURN
0104 400 END
HICNDP

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```

0001      SUBROUTINE HISUM
C
C
C Written by Dale W. Hiltner / Sept-82
C Uses files generated by H1CNDP.DHF and
C outputs RMS ratio, AVG, STD values by
C channel content. This provides a quick
C summary of a test session.
C
C
0002      REAL POSR(30),POSS(30),POSA(30),POSRA(30)
0003      REAL VELR(30),VELS(30),VELA(30),VELRA(30)
0004      REAL ACCR(30),ACCS(30),ACCA(30),ACCRAT(30)
0005      REAL COMR(30),COMS(30),COMA(30),COMRAT(30)
0006      REAL JOYR(30),JOYS(30),JOYA(30),DURT(30)
0007      LOGICAL*1 FILEP(3),FILEN(120),FILEE(4)
0008      COMMON/SUMP/POSR,POSS,POSA,POSRA
0009      COMMON/SUMV/VELR,VELS,VELA,VELRA
0010      COMMON/SUMA/ACCR,ACCS,ACCA,ACCRAT
0011      COMMON/SUMC/COMR,COMS,COMA,COMRAT
0012      COMMON/SUMJ/JOYR,JOYS,JOYA
0013      COMMON/WFILE/FILEP,FILEN,FILEE
0014      COMMON/CFILE/NFILE,INF,DURT
C
C SUMMARY OF DATA FROM H1PIKP
C
0015      WRITE(7,600)
0016      600 FORMAT('0','H1PIKP DATA ANALYSIS SUMMARY')
C
0017      WRITE(7,610)(FILEP(I),I=1,3),(FILEE(I),I=1,4)
0018      610 FORMAT('0',5X,'DATA FILE',2X,3A1,'XXX',4A1)
C
C OUTPUT ALL DATA BY CHANNEL CONTENT
C
0019      DO 710 I1=1,5
C
0020      IF(I1.EQ.5)GO TO 500
0022      WRITE(7,720)
0023      720 FORMAT('0',4X,'CHANNEL',3X,'FILE #',7X,'RMS',7X,'RMSRAT',4X,
+           'STD',7X,'AVG',6X'DURATION')
C
0024      GO TO (100,200,300,400,500)I1
C
0025      100 DO 120 I2=1,NFILE
0026      120 WRITE(7,130)(FILEN(I),I=(I2-1)*3+1,(I2-1)*3+3),
+           POSR(I2),POSRA(I2),POSS(I2),POSA(I2),DURT(I2)
0027      130 FORMAT(5X,'POSITION',2X,3A1,7X,5(F8.4,2X))
0028      GO TO 710
C
0029      200 DO 220 I2=1,NFILE
0030      220 WRITE(7,230)(FILEN(I),I=(I2-1)*3+1,(I2-1)*3+3),
+           VELR(I2),VELRA(I2),VELS(I2),VELA(I2),DURT(I2)
0031      230 FORMAT(5X,'VELOCITY',2X,3A1,7X,5(F8.4,2X))
0032      GO TO 710
C
0033      300 DO 320 I2=1,NFILE
0034      320 WRITE(7,330)(FILEN(I),I=(I2-1)*3+1,(I2-1)*3+3),
+           ACCR(I2),ACCRAT(I2),ACCS(I2),ACCA(I2),DURT(I2)
0035      330 FORMAT(5X,'ACCEL',2X,3A1,7X,5(F8.4,2X))
0036      GO TO 710
C
0037      400 DO 420 I2=1,NFILE
0038      420 WRITE(7,430)(FILEN(I),I=(I2-1)*3+1,(I2-1)*3+3),
+           COMR(I2),COMRAT(I2),COMS(I2),COMA(I2),DURT(I2)
0039      430 FORMAT(5X,'COMMAND',2X,3A1,7X,5(F8.4,2X))
0040      GO TO 710
C
0041      500 WRITE(7,510)
0042      510 FORMAT('0',4X,'CHANNEL',3X,'FILE #',7X,'RMS',7X,'STD',7X,
+           'AVG',6X'DURATION')
0043      DO 520 I2=1,NFILE
0044      520 WRITE(7,530)(FILEN(I),I=(I2-1)*3+1,(I2-1)*3+3),
+           JOYR(I2),JOYS(I2),JOYA(I2),DURT(I2)
0045      530 FORMAT(5X,'JOYSTICK',2X,3A1,7X,4(F8.4,2X))
C
0046      710 CONTINUE
C
0047      RETURN
C
0048      END
HISUM

```

HIRATK.DAT

0.5373,0.5436,1.2651,0.5944
 0.8623,0.4777,0.9994,0.3959
 1.0367,0.5045,1.1612,0.4583
 0.5432,0.4749,1.0417,0.3381
 0.7032,0.4888,1.0931,0.2770
 0.4359,0.4646,1.0182,0.2993
 0.7495,0.5704,1.1825,0.3077
 1.0,1.0,1.0,1.0
 0.7937,0.5897,1.2313,0.1538
 0.7426,0.5662,1.2086,0.1870
 0.6731,0.5557,1.1438,0.1869
 0.5999,0.5690,1.1520,0.1700
 1.0452,0.6367,0.6537,0.1790
 0.3347,0.6107,0.7184,0.1949
 0.6214,0.6289,0.6776,0.1632/

RUN HIPIKP

Pick up input/output sled data. Channels 1,2,3,4,6.

Currently accepts maximum of 10240 samples per channel

ENTER JSCALE FACTOR

3.3

ENTER INPUT FILE 3 LETTER CODE

H1H

ENTER #FILES TO BE REDUCED

3

ENTER PR FILE #, DATA FILE NUMBER.

ONE SET PER LINE

11,404

12,407

13,409

CURRENT FILE H1H404.CV4

64 Records(256 words/record) => 8 Records

Sled data reduction. Channels 1,2,3,4,6.

Currently accepts maximum of 10240(256*40) samples

Analysis of 2048 points.

POSITION	RMS	AVG	STD
	1.0248	0.7592	0.6883
VELOCITY	RMS	AVG	STD
	0.4600	-0.1152	0.4453
ACCEL	RMS	AVG	STD
	0.9594	-0.0310	0.9589
COMMAND	RMS	AVG	STD
	0.2142	0.1147	0.1809
JOYSTICK	RMS	AVG	STD
	0.1397	0.0845	0.1112

CURRENT FILE H1H407.CV4

256 Records(256 words/record) => 32 Records

Sled data reduction. Channels 1,2,3,4,6.

Currently accepts maximum of 10240(256*40) samples

Analysis of 8192 points.

POSITION	RMS	AVG	STD
	1.0107	0.1444	1.0003
VELOCITY	RMS	AVG	STD
	0.3862	-0.0353	0.3846
ACCEL	RMS	AVG	STD
	1.0753	-0.0524	1.0740
COMMAND	RMS	AVG	STD
	0.2026	-0.0156	0.2020
JOYSTICK	RMS	AVG	STD
	0.1356	-0.0090	0.1353

CURRENT FILE H1H409.CV4
 224 Records(256 words/record) = 28 Records
 Sled data reduction. Channels 1,2,3,4,6.
 Currently accepts maximum of 10240(256*40) samples
 Analysis of 7168 Points.

POSITION	RMS	AVG	STD
	0.7708	-0.6053	0.4773
VELOCITY	RMS	AVG	STD
	0.3891	-0.0532	0.3855
ACCEL	RMS	AVG	STD
	0.9971	-0.0661	0.9949
COMMAND	RMS	AVG	STD
	0.1915	0.0196	0.1905
JOYSTICK	RMS	AVG	STD
	0.1303	0.0168	0.1292

H1FIKP DATA ANALYSIS SUMMARY

DATA FILE H1HXXX.CV4

CHANNEL	FILE #	RMS	RMSRAT	STD	AVG	DURATION
POSITION	404	1.0248	1.2911	0.6883	0.7592	20.4800
POSITION	407	1.0107	1.3610	1.0003	0.1444	81.9200
POSITION	409	0.7708	1.1452	0.4773	-0.6053	71.6800

CHANNEL	FILE #	RMS	RMSRAT	STD	AVG	DURATION
VELOCITY	404	0.4600	0.7800	0.4453	-0.1152	20.4800
VELOCITY	407	0.3862	0.6822	0.3846	-0.0353	81.9200
VELOCITY	409	0.3891	0.7002	0.3855	-0.0532	71.6800

CHANNEL	FILE #	RMS	RMSRAT	STD	AVG	DURATION
ACCEL	404	0.9594	0.7792	0.9589	-0.0310	20.4800
ACCEL	407	1.0753	0.8897	1.0740	-0.0524	81.9200
ACCEL	409	0.9971	0.8717	0.9949	-0.0661	71.6800

CHANNEL	FILE #	RMS	RMSRAT	STD	AVG	DURATION
COMMAND	404	0.2142	1.3926	0.1809	0.1147	20.4800
COMMAND	407	0.2026	1.0833	0.2020	-0.0156	81.9200
COMMAND	409	0.1915	1.0245	0.1905	0.0196	71.6800

CHANNEL	FILE #	RMS	STD	AVG	DURATION
JOYSTICK	404	0.1397	0.1112	0.0845	20.4800
JOYSTICK	407	0.1356	0.1353	-0.0090	81.9200
JOYSTICK	409	0.1303	0.1292	0.0168	71.6800

STOP --

Appendix D

Test Procedure Checklist

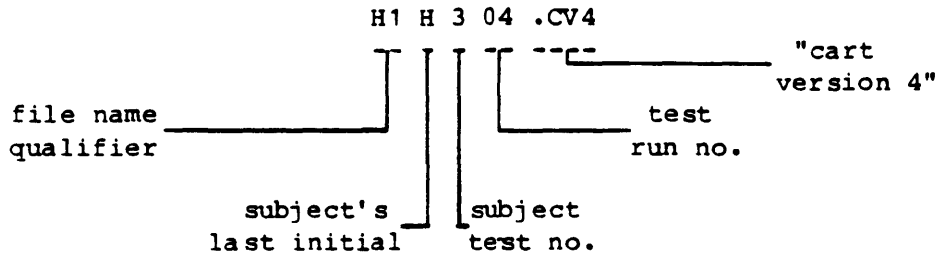
The formal Test Procedure Checklist is listed on pages 169-170. The Sled General Checklist, which covers the Sled start-up and shutdown procedures, is listed on pages 171-173. The Test Procedure Checklist is a step-by-step instruction list that should be used to conduct the tests properly. It lists all the steps used to obtain the final results of this work. The subject testing procedure, and the data reduction procedure, are both described in the checklist. It is recommended that this checklist be adhered to in all further testing.

The log sheet form used during the testing of this work is shown on page 174. This form was found to be invaluable during the analysis of the data. In addition to providing a log of all the tests, it helped reveal the various response trends of the subjects. It is recommended that this log form be used to take vigorous notes during all testing. All runs, practice and data, should be noted on the log sheets. COMMENTS should include any difficulties, irregularities, or pertinent observations of the test conductor or subject regarding the subjects performance and the test proceedings.

D.1 Data Filename Convention

The data filenames are defined by the test conductor according to the following convention: A filename consists of a series of three characters plus three numbers. The three characters begin with H1, and the third character is the first letter of the subject's last name (key subject

initial). The three numbers begin with the test session number (a 1 for the first session, 2 for the second, etc...), followed by two which indicate the run numbers. The run numbers are automatically set in sequential order during the test session. An illustration follows.



Test Procedure Checklist

1. Start-Up Procedure: Follow Sled General Checklist through section seven for all standard procedures. Add the following changes.
 - 1.1 Check AA1 cable (joystick command).
 - 1.2 RUN H1CAR7 and set program parameters DA, PR, and JO. (Data filename, protocol file, and JSSCALE)
 - 1.3 Explain the task to the subject.
 - 1.4 Seat subject in chair and adjust head restraint height, shoulder pads seatbelt and goggles.
 - 1.5 Attach joystick with voltmeter wired in and set to +2.0v. Check that joystick calibration is within +/- .17v.
 - 1.6 Put masking noise level at maximum (as long as subject is comfortable.)

2. Standard Run Procedure: Use to conduct each run.
 - 2.1 Enable sled, push START.
 - 2.2 SEND protocol file number command.
 - 2.3 Notify subject when program has STARTED.
 - 2.4 Notify subject when sled is MOVING to home position.
 - 2.5 When digital display on the sled control panel reads 333, start stopwatch and turn on masking noise.
 - 2.6 Fill in Data Sheet with RANGE, STOP, DURATION and any other observations.
 - 2.7 Check that the subject is comfortable, record any comments.
 - 2.8 If there will be more than a few moments between runs, press STOP.
 - 2.9 If sled has triggered limit switch, press STOP, disable sled, and push sled off of limit switch while holding down START, and press STOP.
 - 2.10 Check joystick calibration.

3. Test Procedure
 - 3.1 Explain RIDE ONLY to the subject.
 - 3.2 Disable joystick and data storage.
 - 3.3 Run practice profile.
 - 3.4 Explain PRACTICE RUN to the subject.
 - 3.5 Check joystick calibration, enable joystick control.
 - 3.6 Run practice profile as per standard run procedures. Continue sequence (calibration check, profile run) until subject does not improve performance or starts completing runs.
 - 3.7 Explain DATA RUN to subject and inform subject of RIDE ONLY.
 - 3.8 Disable joystick.
 - 3.9 Run data file, RIDE ONLY.
 - 3.10 Check joystick calibration, enable joystick.
 - 3.11 Notify subject of DATA RUN.
 - 3.12 Run data profile. Continue sequence (calibration check, data run) until 4-5 completed runs have been stored in the computer.

4. Power Shutdown Procedure: Follow Sled General Checklist.

5. Data Reduction

5.1 Delete generated files as file space allows if NOT COMPLETE.

5.2 LIST saved files.

5.3 RUN H1PIKY for each complete data file.

5.3.1 Inputs: JSCALE, RUN TIME, FILE NAME.

5.4 RUN H1CY immediately after running H1PIKY.

5.4.1 Inputs: JSCALE, RUN TIME.

5.5 Plot results.

5.5.1 Plot velocity Frequency Spectrum on graph with no-subject data already plotted.

5.5.2 Plot joystick frequency spectrum.

5.5.3 Pick three "best" runs to make BODE plot.

5.5.4 Record velocity RMS RATIO, joystick RMS.

5.5.5 Find average, variance of Log(GAIN) and phase.

5.5.6 Plot Bode plot.

Sled General Checklist

PRE-EXPERIMENT

- 1.00 Power On
 - 1.1 SLED ELECTRONICS CABINET: sled disabled.
 - 1.2 240 VOLT MAIN POWER BOX: main power on.
 - 1.3 CART: Remove covers from rails.
 - 1.4 SLED POWER SWITCH: Press START.

- 2.00 Mechanical Safety Checks
 - 2.1 CART: Before moving the sled make sure that everything on the sled is secured.
 - 2.2 CART: Check to see that the subject's panic button is in working order.
 - 2.3 SLED POWER SWITCH: Press START.
 - 2.4 CART: Check to see that the limit switches at both ends of the track are working: manually slide the sled over left limit switch, press START (on sled power switch), slide sled over right limit switch, and press START (on sled power switch) again.
 - 2.5 CART: Inspect the umbilical cable attached to the back of the sled to make sure that everything is in working order.

- 3.00 Wiring
 - 3.1 SLED CONTROL PANEL: Check cables to make sure they are in the proper configurations.
 - 3.1.1 CABLE-D1 (cart position)
 - 3.1.2 CABLE-D2 (cart velocity)
 - 3.1.3 CABLE-D3 (cart acceleration)
 - 3.1.4 CABLE-D11 (velocity command signal)
 - 3.1.5 CABLE-C11 ("SEND" signal)
 - 3.1.6 CABLE-C12 ("ABORT" signal)

- 4.00 Power Check
 - 4.1 SLED POWER SWITCH: Make sure that the Sled Power Switch "STOP" has been pressed before going to computer room.

- 5.00 Computer Program
 - 5.1 Make sure that the computer is free by checking that the last user has logged out in the log book.
 - 5.2 Log in the log book.
 - 5.3 PDP 11/34 CONTROL PANEL: Boot the computer: press and hold down "CONTROL" and then press "BOOT".
 - 5.4 KEYBOARD: type DPO (return).
 - 5.5 KEYBOARD: enter date and time when prompted by computer.
 - 5.6 KEYBOARD: run cart-control program by typing RUN ALCART (return).
 - 5.7 PATCH PANEL CABINET: make sure that the sled patch panel is in the patch panel holder.
 - 5.8 PATCH PANEL CABINET: "Sled General" cables 1,2,7 and 8 are in their respective 1,2,7,8 receptors on the back of the patch panel cabinet.

- 5.9 CABINET 2: check that the green digital input/output cables in back of the cabinet are hooked to the digital input/output receptors.
- 5.10 KEYBOARD: set cart program parameters.
- 5.11 KEYBOARD: load protocol file in cart control program.
- 5.12 KEYBOARD: issue "REMOTE" command to computer.
- 5.13 Post signs indicating a "REMOTE OPERATED EXPERIMENT IN PROGRESS".

6.00 Cart Preparation

- 6.1 SLED CONTROL PANEL: verify blinking minus sign on digital LED display. If not there, return to computer room, check program status and wiring configurations.
- 6.2 Make sure all personnel are clear of the sled area.
- 6.3 SLED ELECTRONICS CABINET: enable sled controller.
- 6.4 SLED CONTROL PANEL: press START.
- 6.5 SLED CONTROL PANEL: if you are using a NEW protocol file, run each entry with the sled EMPTY. If you are using an OLD file, run ONE example of each type of profile (e.g. sine, step, etc.) with the sled EMPTY.
- 6.6 SLED POWER SWITCH: push STOP (i.e. put the brake on.)

7.00 Subject Preparation

- 7.1 Log experiment in SLED log book.
- 7.2 Explain to the subject the experiment and any risks involved.
- 7.3 Have subject read and sign "Informed Consent Form".
- 7.4 SLED ELECTRONICS CABINET: sled disabled.
- 7.5 SLED POWER SWITCH: press sled power STOP.
- 7.6 Have subject enter the sled, making sure that he/she does not step on the rails or the chair frame.
- 7.7 CART: demonstrate panic button and give to the subject.
- 7.8 SUBJECT/CART: adjust head restraint height, foot rests and shoulder pads.
- 7.9 CART: complete specialized instrumentation (e.g. biteboards, electrodes, camera focus, etc.).
- 7.10 SUBJECT/CART: tighten seat belt.
- 7.11 SUBJECT/CART: tighten chest straps (optional).
- 7.12 SUBJECT/CART: tighten and adjust forehead and chin straps.
- 7.13 SUBJECT/CART/SLED CONTROL PANEL: determine masking noise level.
- 7.14 SUBJECT: check to make sure that the subject is comfortable.
- 7.15 CART: lower hood.
- 7.16 CART: attach cowl.
- 7.17 SLED ELECTRONICS CABINET: turn ventilation fan on.

EXPERIMENT

- 8.00 Consult respective protocol for individual experiment.

POST-EXPERIMENT

- 9.00 Subject egress
- 9.1 SLED CONTROL PANEL: sled power off.
- 9.2 SLED ELECTRONICS CABINET: sled disabled.
- 9.3 CART: cowl off.
- 9.4 CART: raise hood.
- 9.5 CART/SUBJECT: disconnect specialized instrumentation.

9.6 SUBJECT: remove restraints.

9.7 SUBJECT: subject egress, again with no stepping on rails or chair frame.

10.00 Shutting Down

10.1 CART: lower and secure hood and all items on the cart.

10.2 SLED ELECTRONICS CABINET: disable sled (with sled power START).

10.3 CART: move sled manually to home position.

10.4 CART: place covers on rail.

10.5 SLED CONTROL PANEL: send control to computer terminal ("2000" command).

10.6 240 V MAIN POWER BOX: main power OFF.

APPENDIX E

THE EXPERIMENTAL RESULTS

The plotted results are presented in the next pages. The Bode plots are shown first followed by the frequency spectrum plots. Bessel's corrected one sigma deviations have been calculated and plotted for all points of the Bode plots. If no deviation is shown, the deviation is within the point symbol. An explanation of the meaning of these plots follows.

E.1 Plot Format Discussion

The Bode plot represents the transfer function of the Human Operator (HO). It relates the acceleration input, in m/s^2 , to the joystick commanded velocity, in m/s . The GAIN is calculated from the amplitude of the joystick velocity divided by the amplitude of the acceleration. These amplitudes are obtained directly from the FFT output at the disturbance frequencies. The GAIN is plotted in units of $\log(\text{GAIN})$, which may also be stated as units of $DB/20$. DB was not used in the plots as it is felt that the factor of 20 is not meaningful for the type of work involved in this thesis. The frequencies have been plotted on a linear scale of $\log(\text{freq.}, \text{rad/sec})$ but are labeled in Hz . The $\log(\text{GAIN})$ section of the Bode plot shows the factor, or GAIN, by which the input acceleration has been increased by the HO to obtain the output joystick velocity command.

The phase data of the Bode plot is calculated from the phase angle of the joystick velocity command minus the phase angle of the acceleration. 180 deg. is then added to this value to correct for the negative sign of the joystick signal, since it opposes the acceleration. The resulting phase is

then that for the HO, and not that for the HO transfer function block in the closed-loop block diagram. The phase angles are taken directly from the FFT output. The phase section of the Bode plot shows the lead or lag the HO has applied to the acceleration disturbance in order to obtain the joystick velocity command. The $\log(\text{GAIN})$ and phase data together show the capabilities of the HO to respond to the acceleration disturbance in attempting to perform the velocity nulling task.

The frequency spectrum plots show the velocity and joystick amplitudes obtained directly from the FFT. In the velocity plot, the squares show the amplitudes obtained from running the disturbance profile without a subject in the cart, the no subject case. This shows the disturbance input to the subject. The dashed line connects the amplitudes at the disturbance frequencies. The solid line connects the amplitudes of the remnant. The remnant data is found by averaging all the frequency and amplitude data obtained from the FFT between the disturbance frequencies. The end remnant points are found by averaging the three frequencies and amplitudes before and after the first and last disturbance frequencies. (It is noted that the no subject data represents the disturbance profile. The true disturbance profile would have no remnant, as it is generated by a sum of sines signal. Since the no subject data is obtained from the Sled system itself, however, errors are introduced which cause the non-zero remnant. These errors probably occur mainly from noise in the acceleration signal and the averaging needed to reduce the number of data points to 1024. Thusly, the remnant shows the limitations of the test system, and should be considered as a rough reference value for the zero amplitude level.)

The circles show the velocity amplitudes obtained from the FFT from runs

with subject control. As in the no subject case the dashed line connects the disturbance frequency amplitudes and the solid line connects the remnant amplitudes. Comparing the two dashed lines shows how effectively the subject performed the velocity nulling task. Ideally the amplitudes obtained with subject control should be much lower than the amplitudes of the no subject case, meaning very effective velocity nulling by the HO. Comparing the two solid lines shows how exclusively the subject responded to the disturbance velocity. Ideally the remnant obtained with subject control should be about the same as the remnant of the no subject case, meaning that the HO was responding only to the disturbance.

The joystick frequency spectrum shows the amplitudes of the joystick velocity command obtained directly from the FFT. Since there is no joystick output for the no subject case only the subject control case can be shown. As for the velocity plot the dashed lines connect the amplitudes at the disturbance frequencies and the solid line connects the remnant amplitudes. The disturbance frequency amplitudes show the level of the joystick velocity output, or the level of control the HO used to perform the velocity nulling task. The remnant amplitudes show how exclusively the HO responded to the disturbance, as similarly indicated by the velocity remnant. Ideally, the joystick amplitudes at the disturbance frequencies should be high, and the remnant amplitudes should be low. This gives a wide joystick disturbance frequency amplitude/remnant separation, which is needed for accurate data. (Chapter 6)

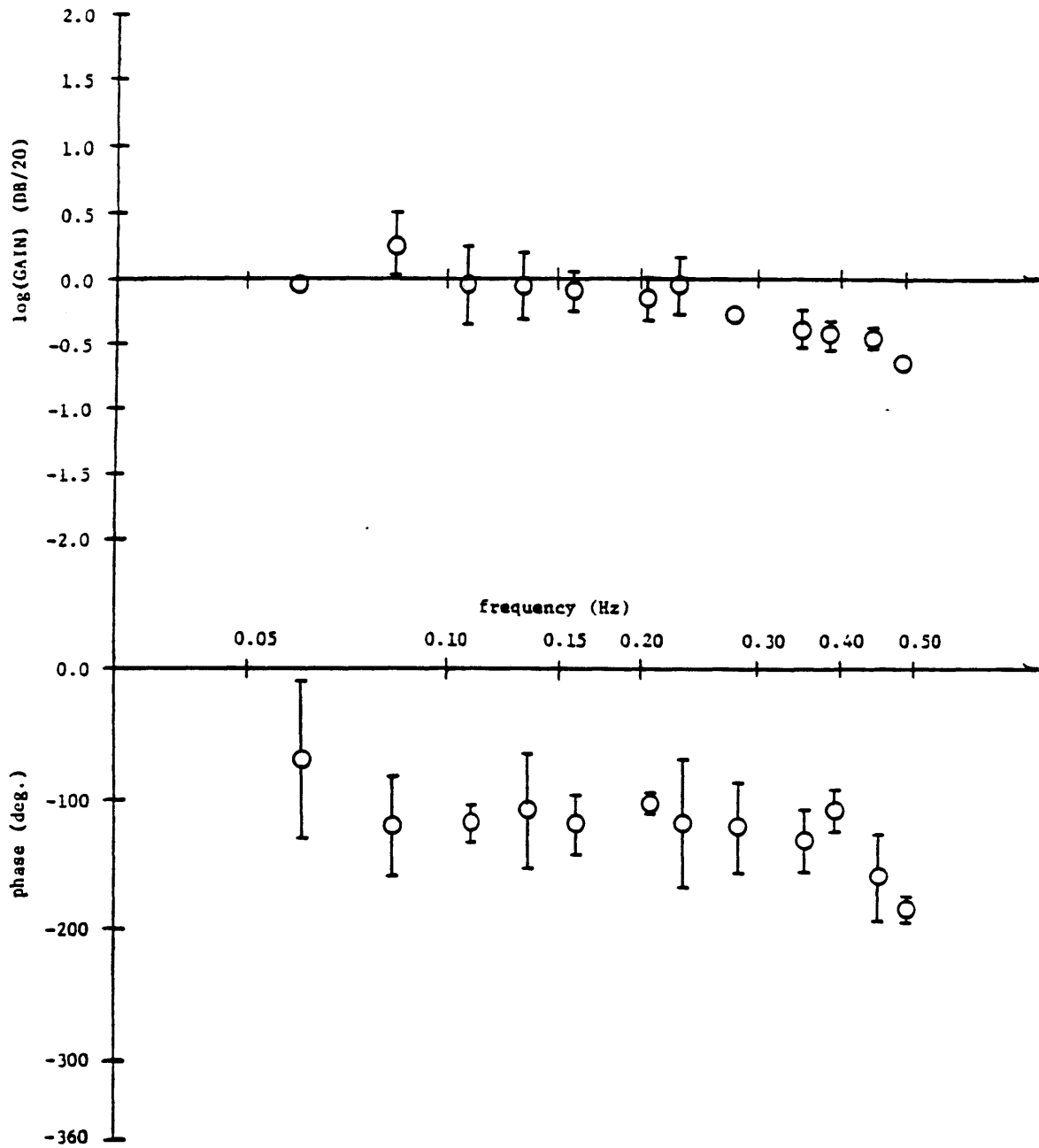


Figure E6.1.01 Bode Plot. Subject MS, Final Profile
Average of Runs 07,08,09

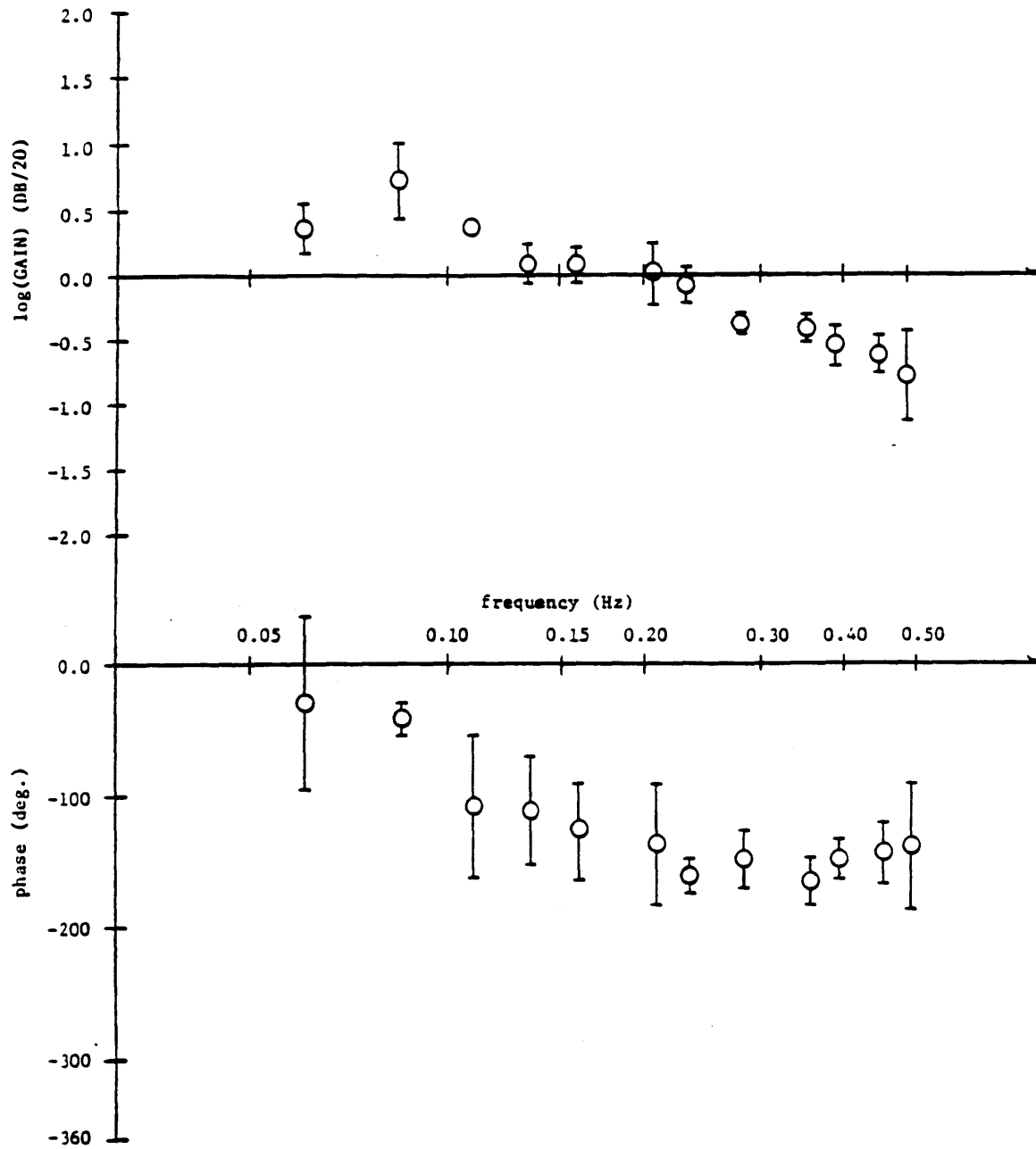


Figure E6.1.02 Bode Plot. Subject MM, Final Profile
Average of Runs 03,09,12

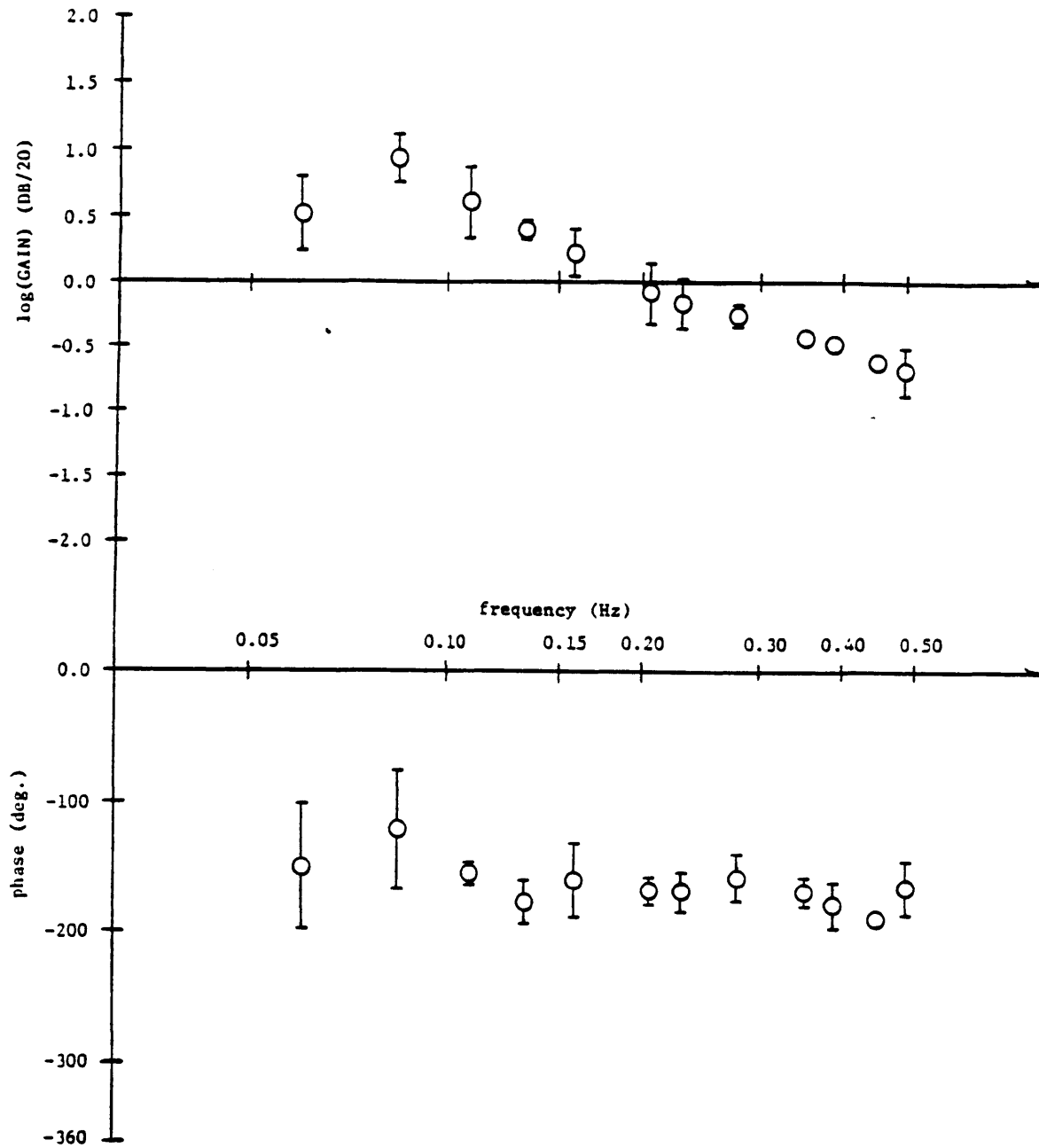


Figure E6.1.03 Bode Plot. Subject JH, Final Profile
Average of Runs 06,08,09

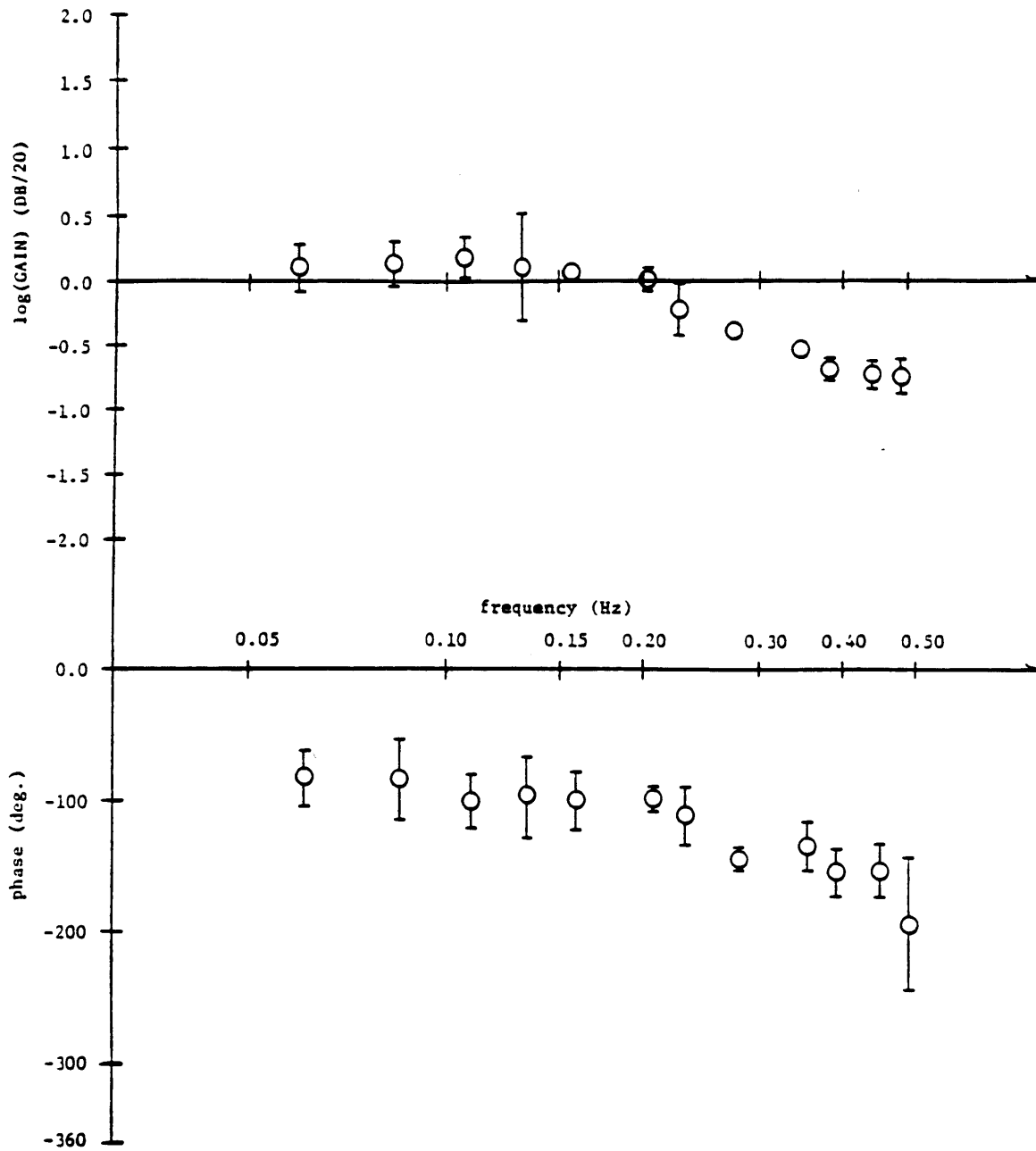


Figure E6.1.04 Bode Plot. Subject DH, Final Profile
Average of Runs 15,21,22

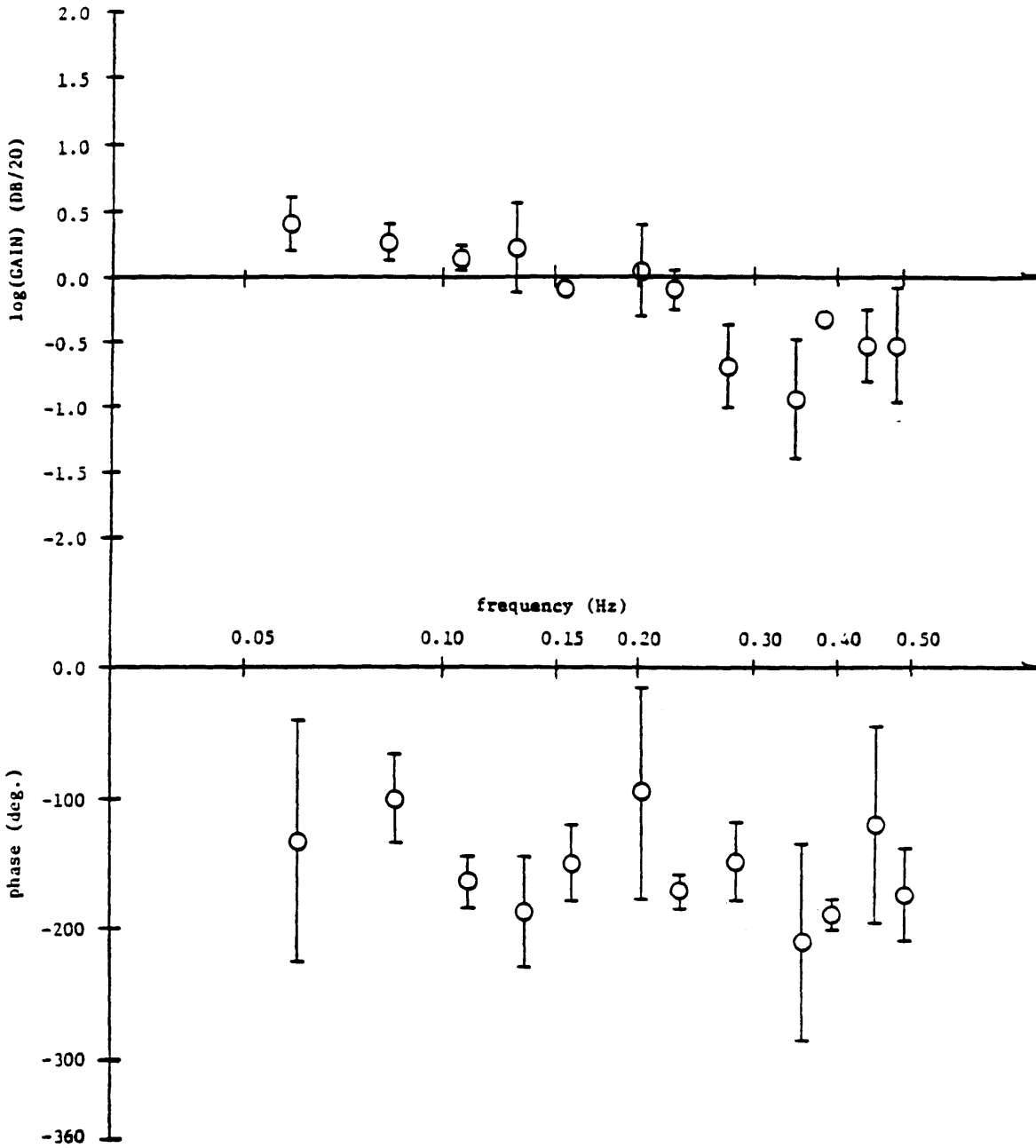


Figure E6.1.05 Bode Plot. Subject LR, Final Profile
Average of Runs 02,03,06

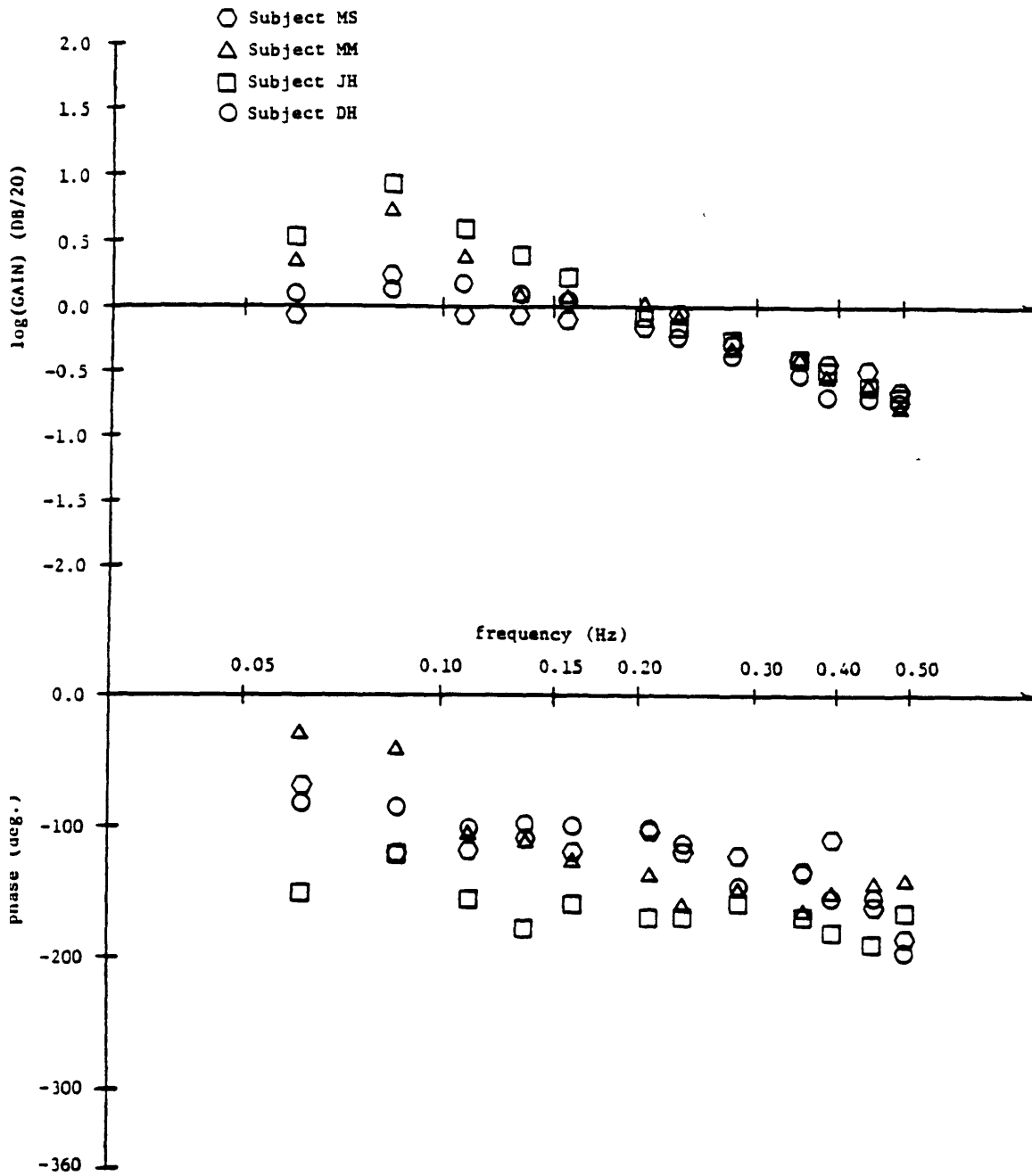


Figure E6.1.06 Summary Bode Plot, All Subjects, Final Profile

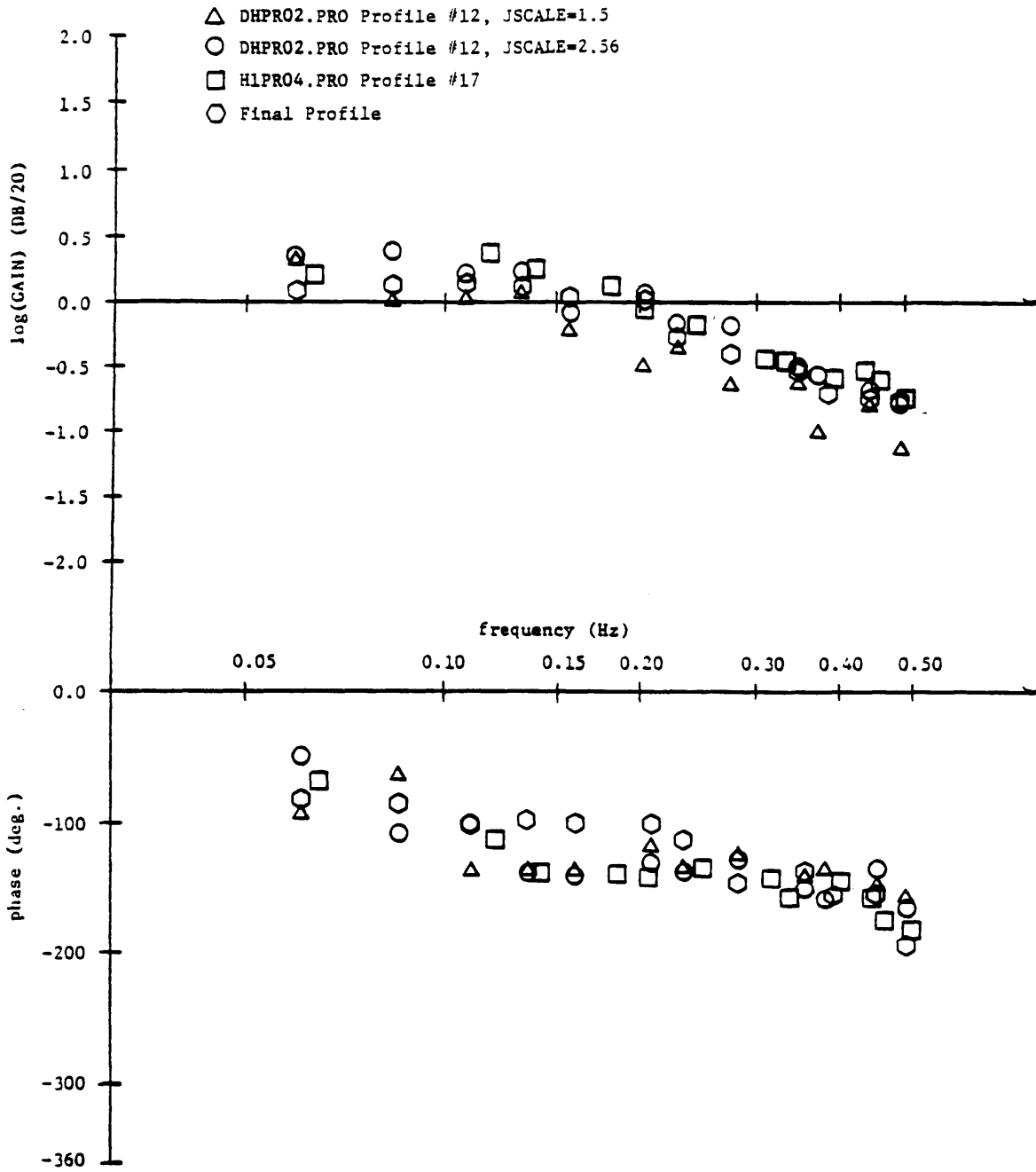


Figure E6.1.07 Summary Bode Plot, Subject DH

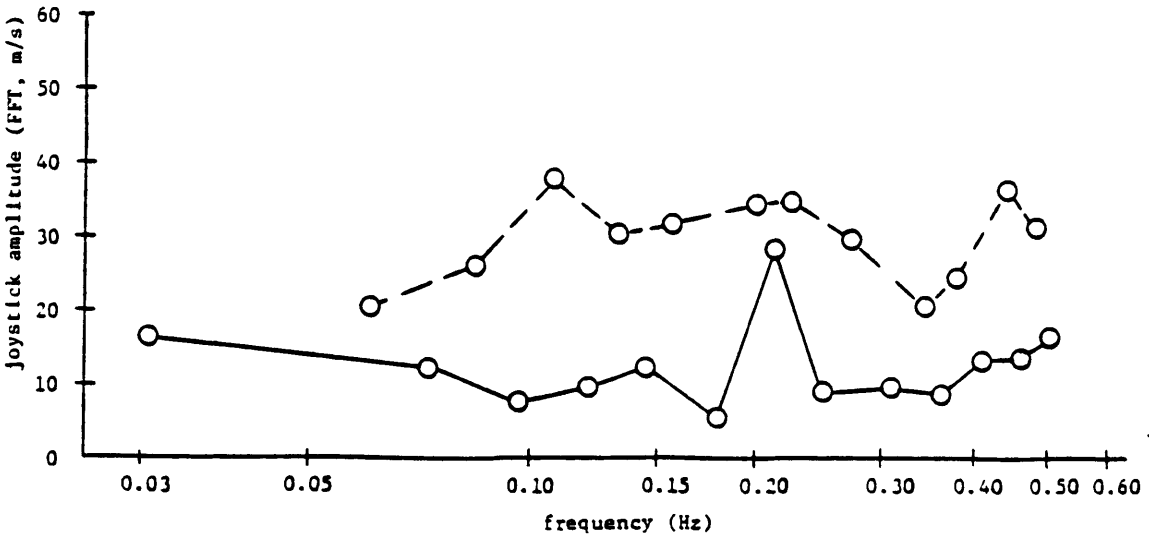
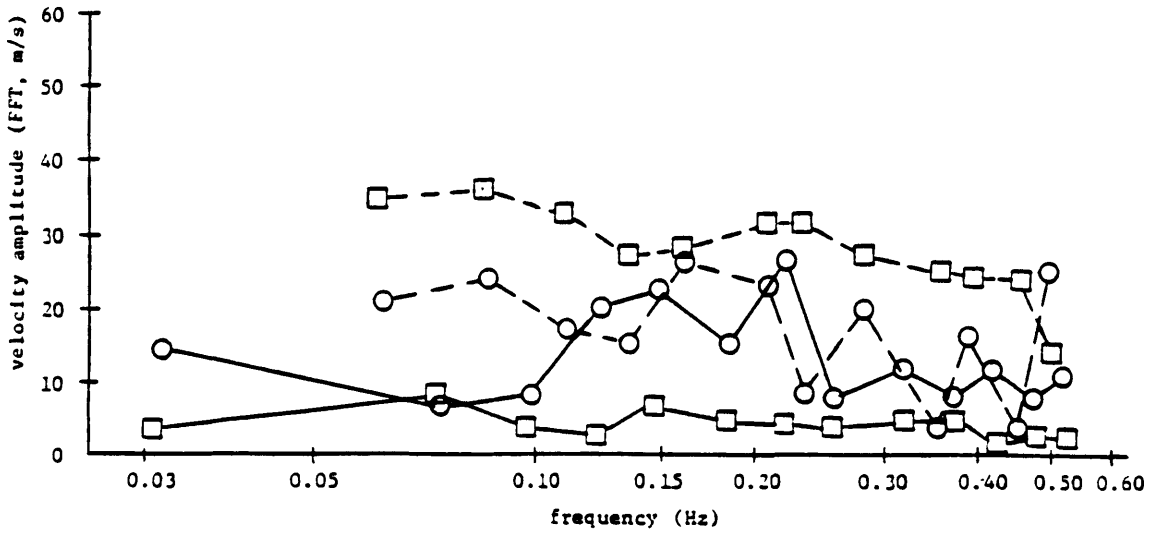


Figure E6.1.08 Frequency Spectrum, Subject MS
Run 02, Final Profile

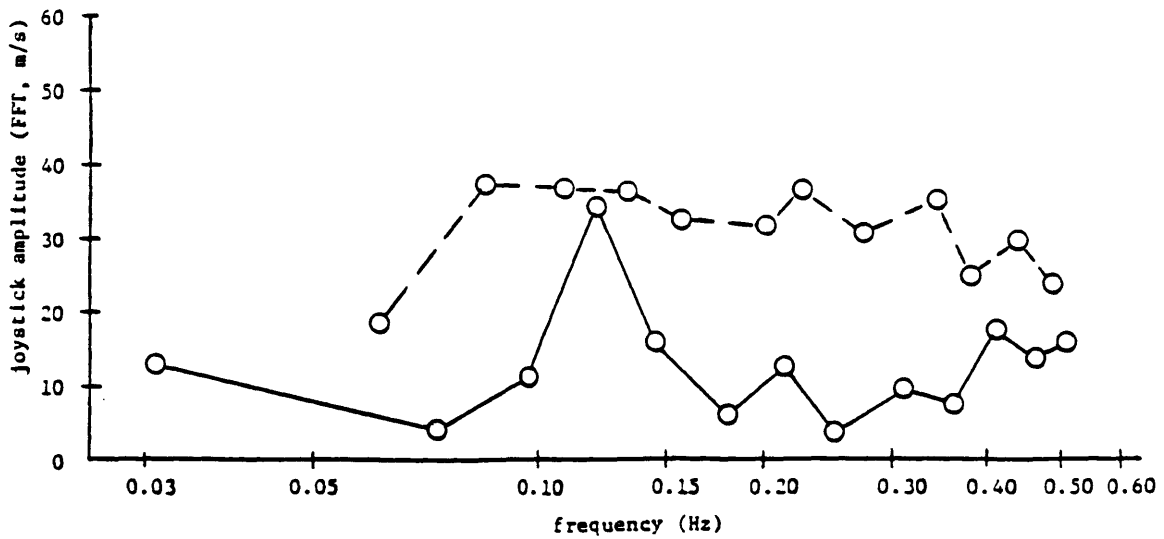
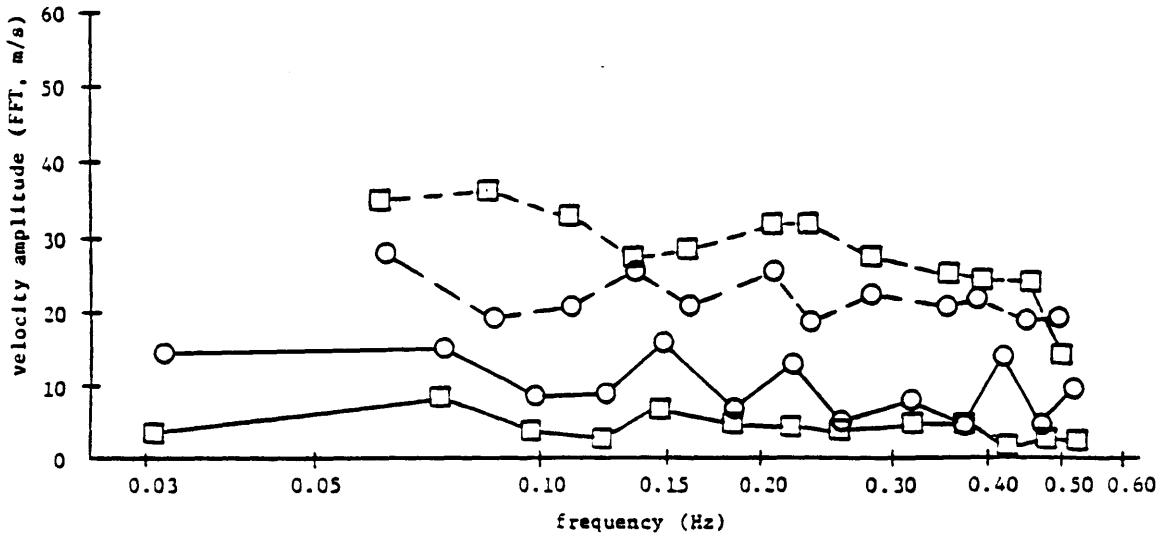


Figure E6.1.09 Frequency Spectrum, Subject MS
Run 07, Final Profile

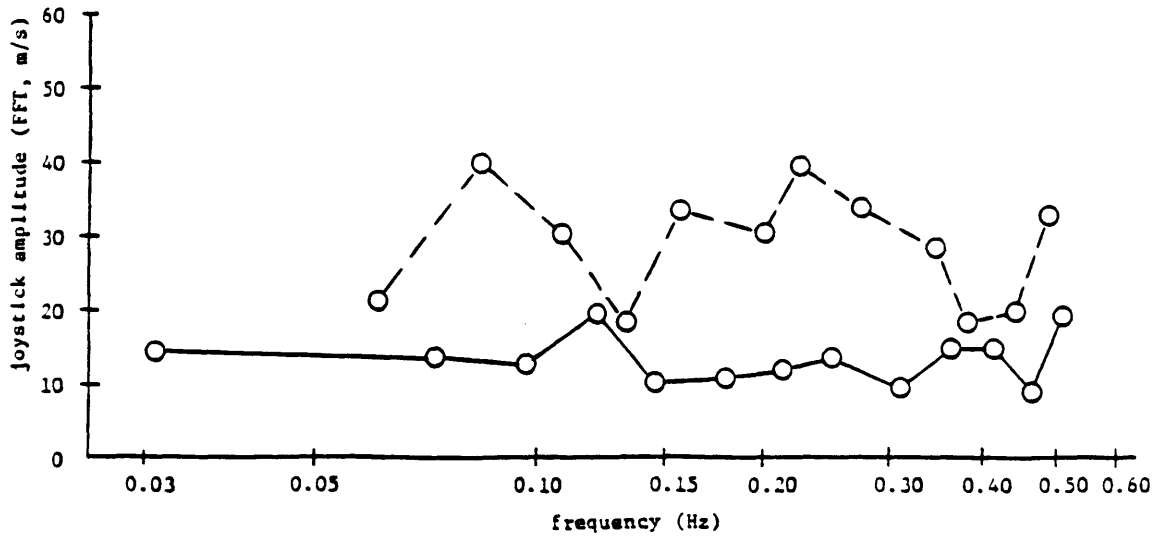
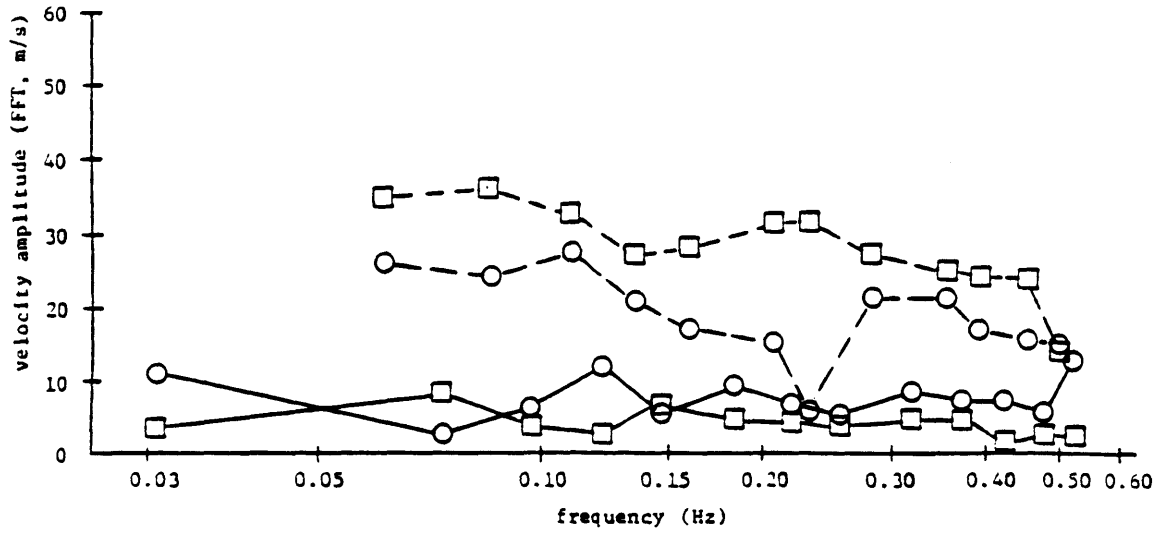


Figure E6.1.10 Frequency Spectrum, Subject MS
Run 08, Final Profile

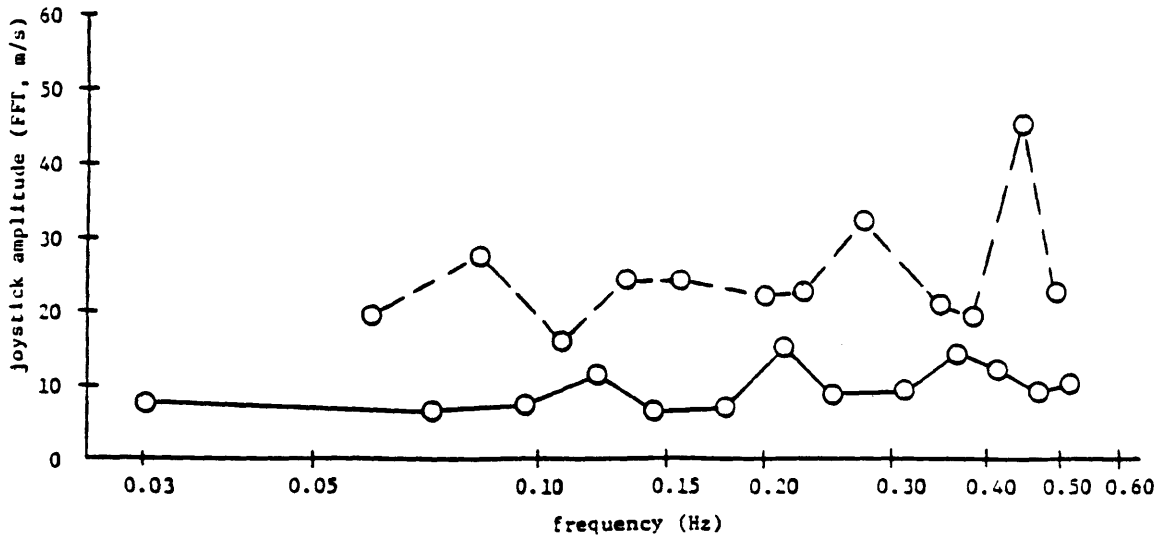
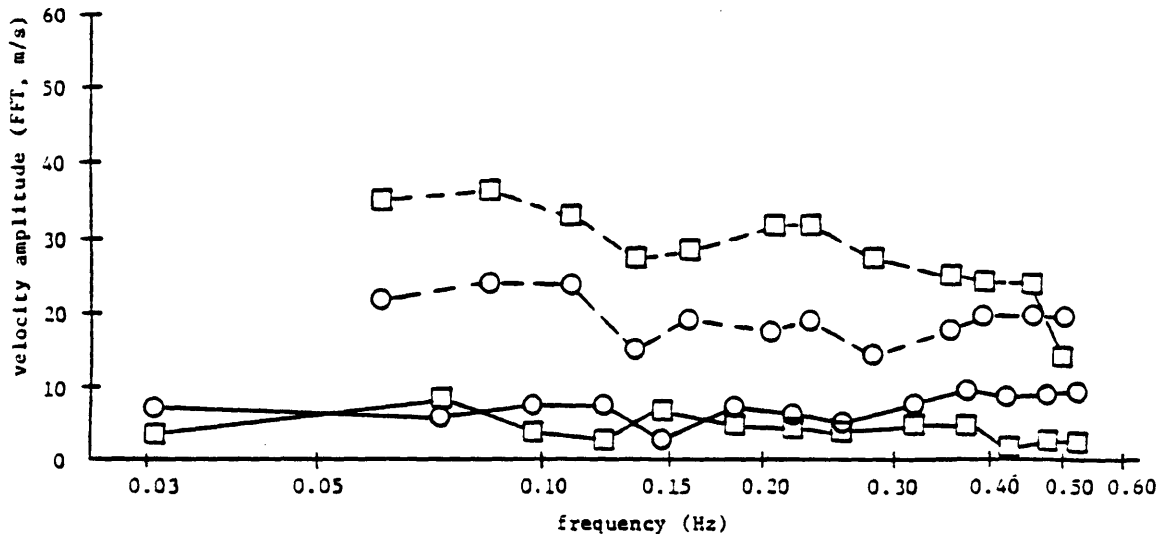


Figure E6.1.11 Frequency Spectrum, Subject MS
Run 09, Final Profile

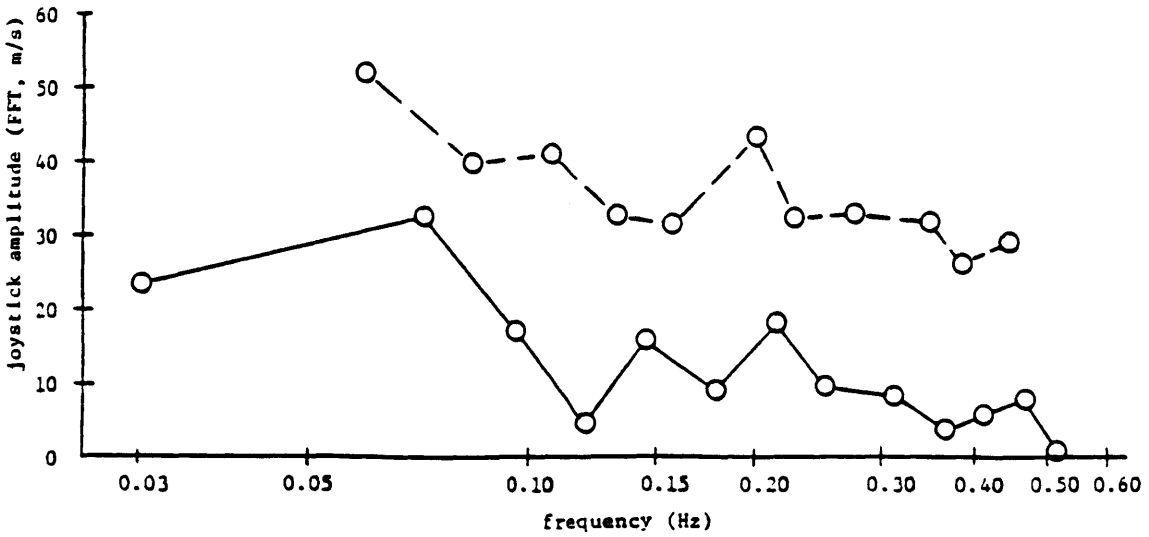
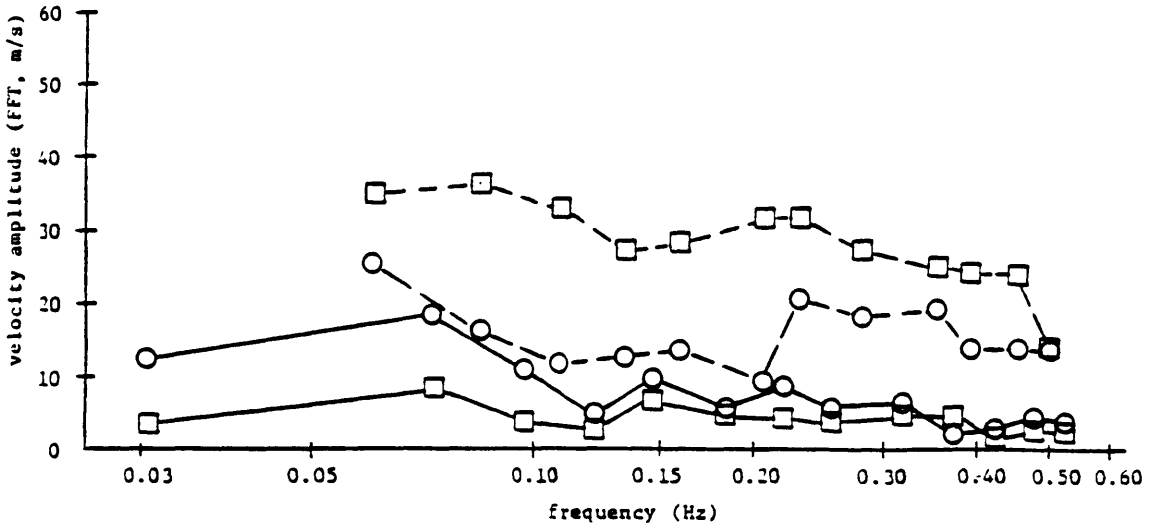


Figure E6.1.12 Frequency Spectrum, Subject MM
Run 03, Final Profile

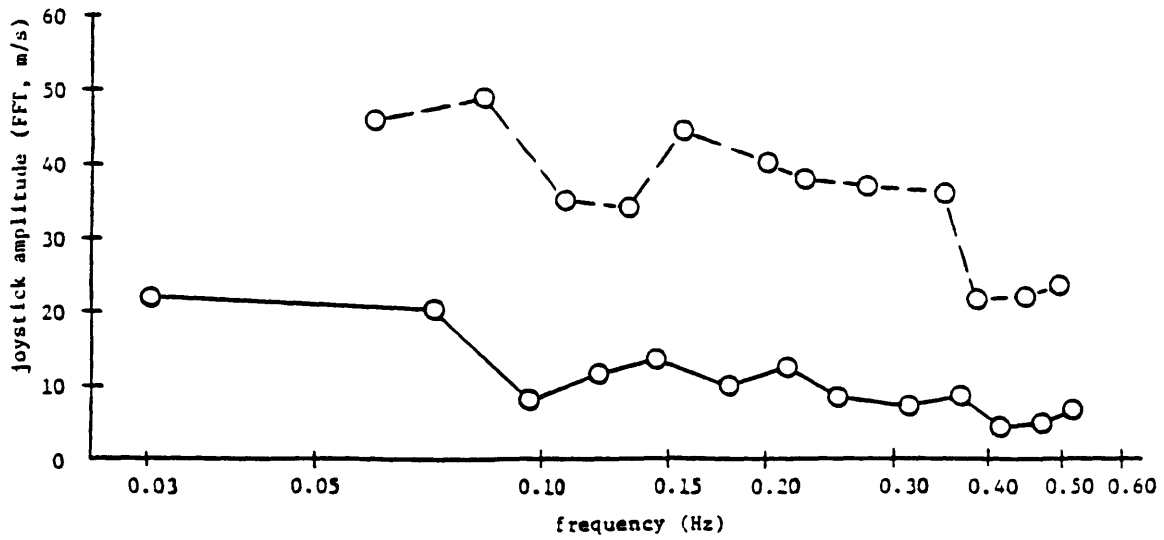
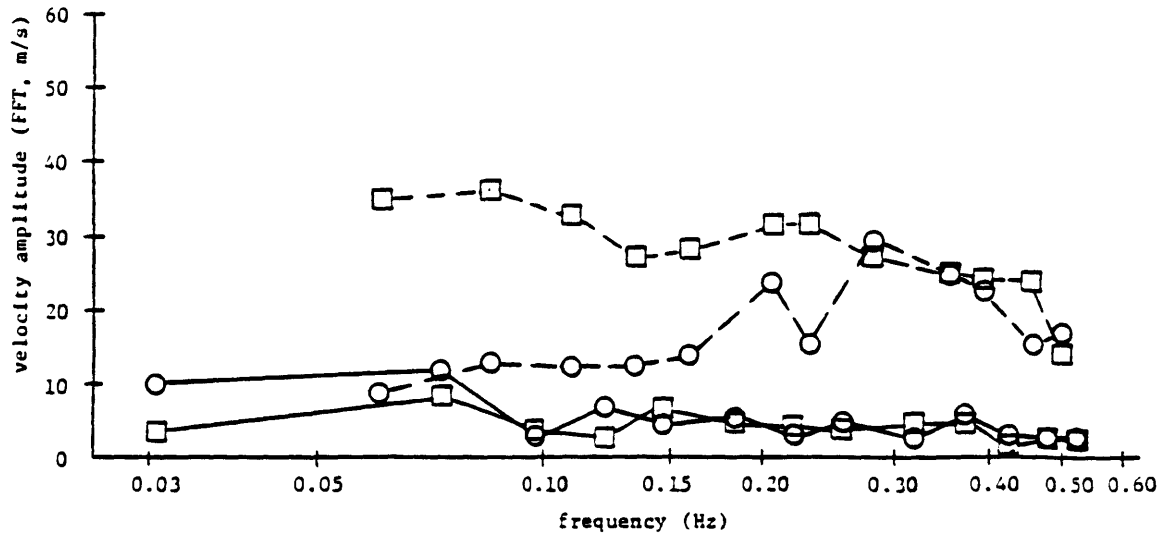


Figure E6.1.13 Frequency Spectrum, Subject MM
Run 09, Final Profile

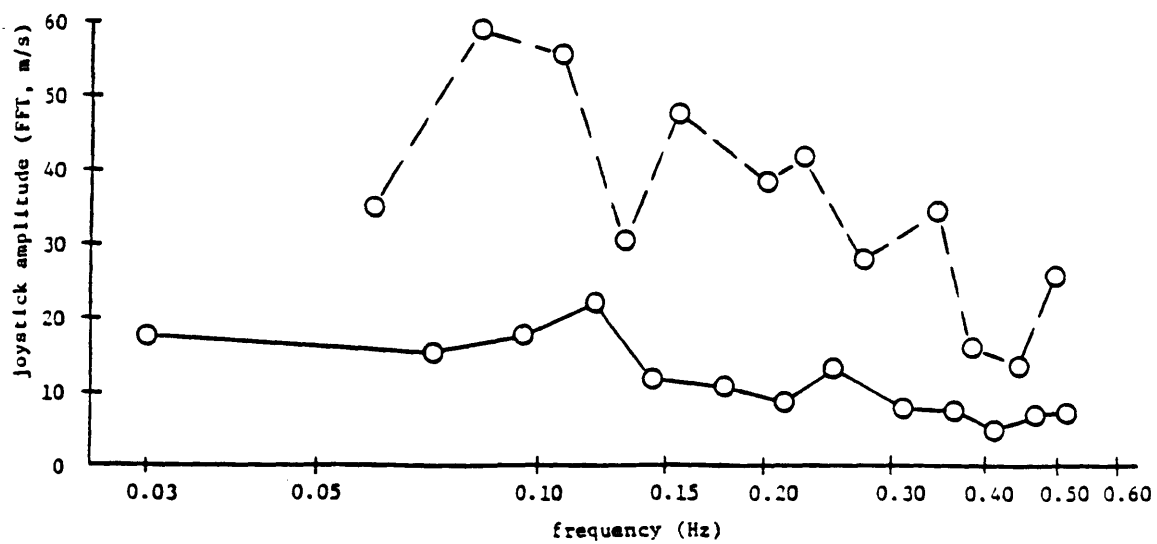
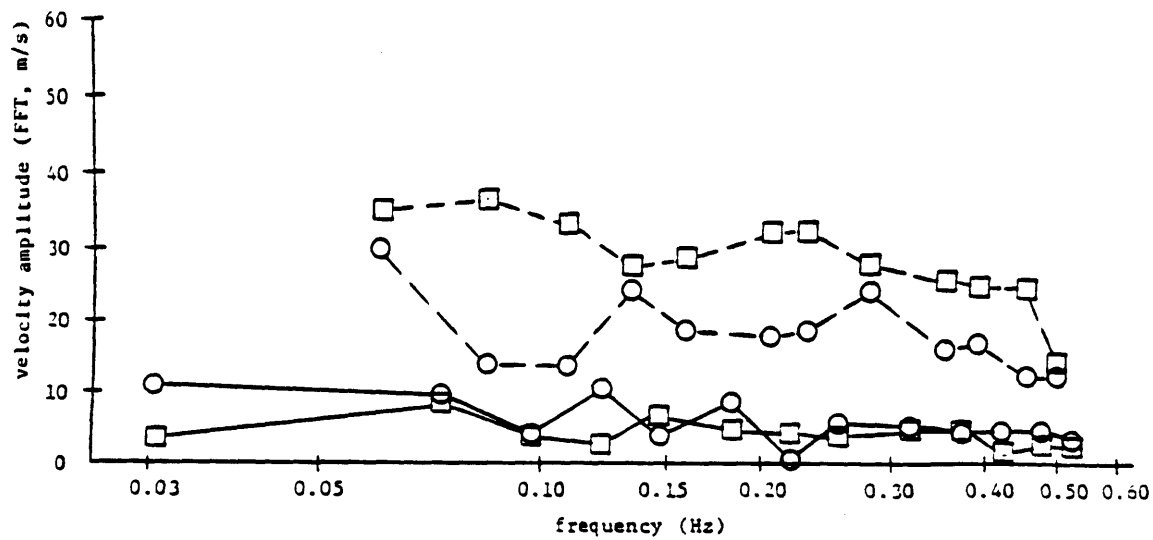


Figure E6.1.14 Frequency Spectrum, Subject MM
Run 12, Final Profile

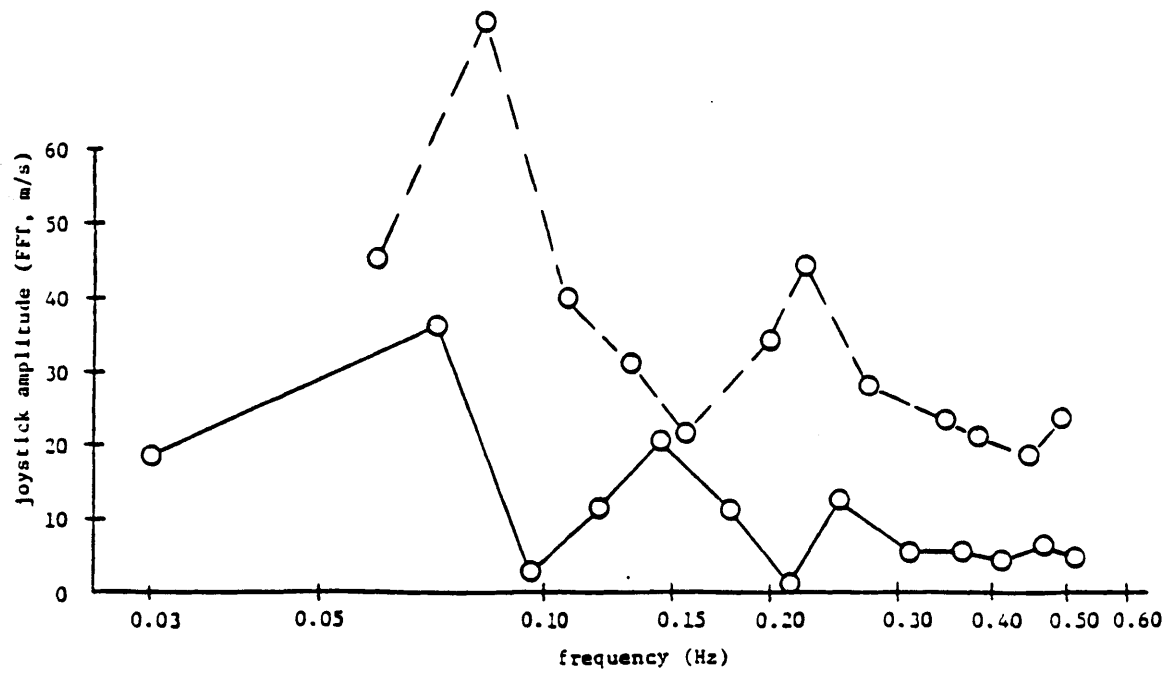
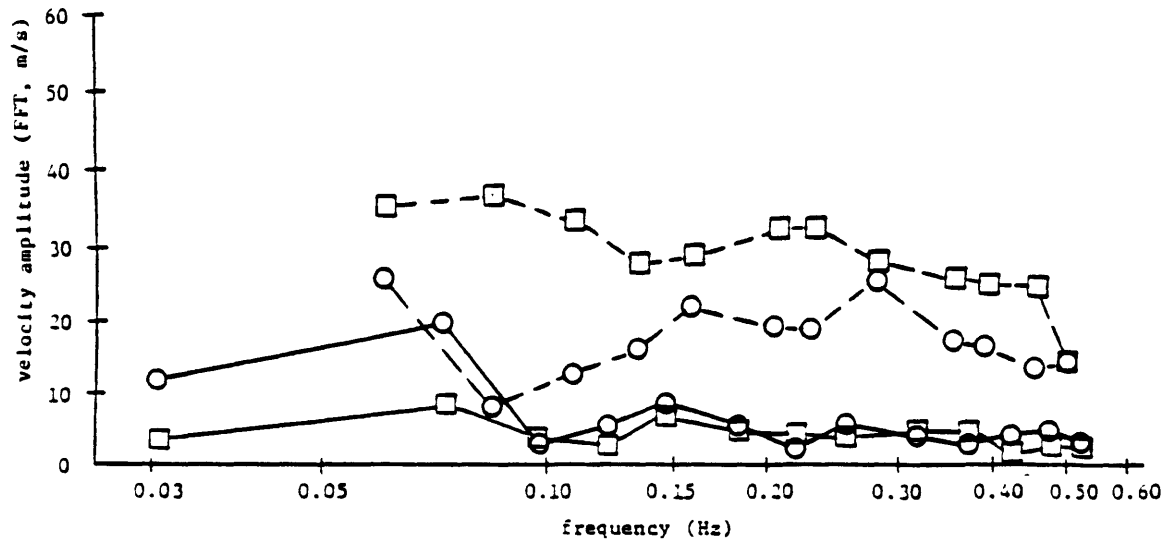


Figure E6.1.15 Frequency Spectrum, Subject MM
Run 15, Final Profile

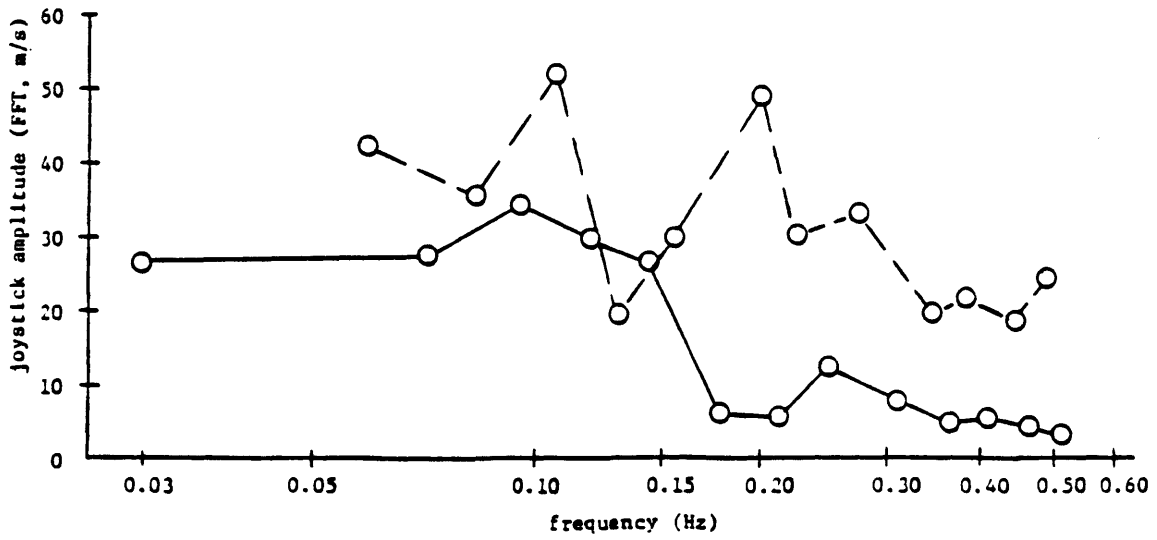
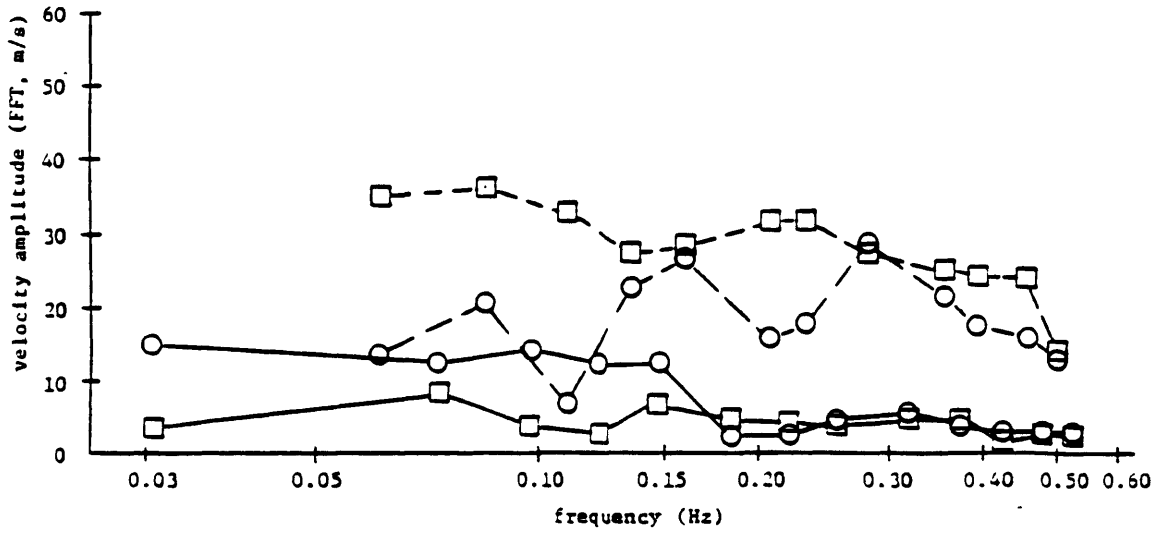


Figure E6.1.16 Frequency Spectrum, Subject MM
Run 18, Final Profile

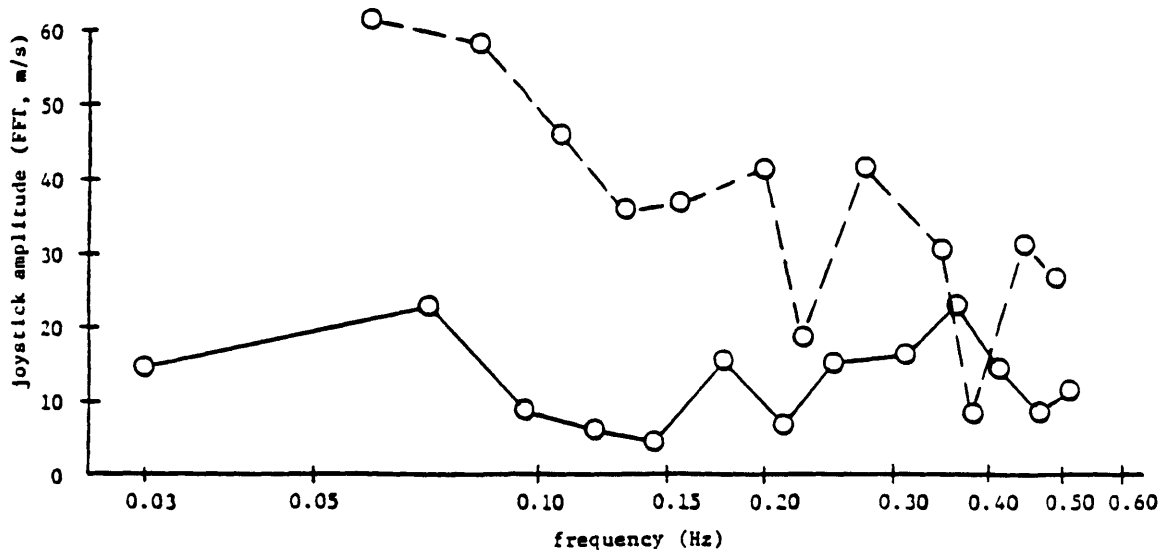
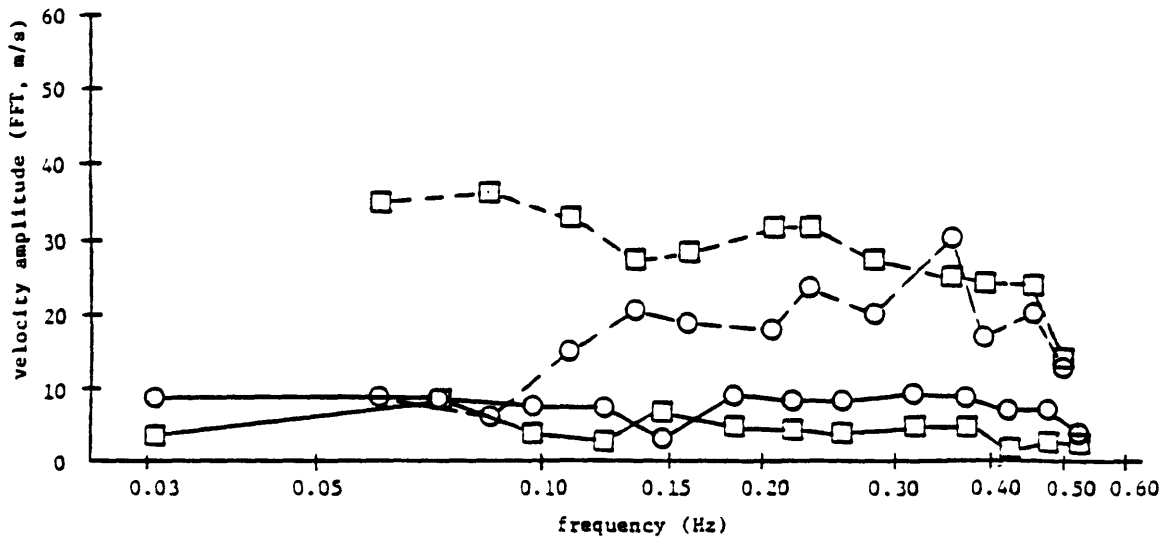


Figure E6.1.17 Frequency Spectrum, Subject JH
Run 03, Final Profile

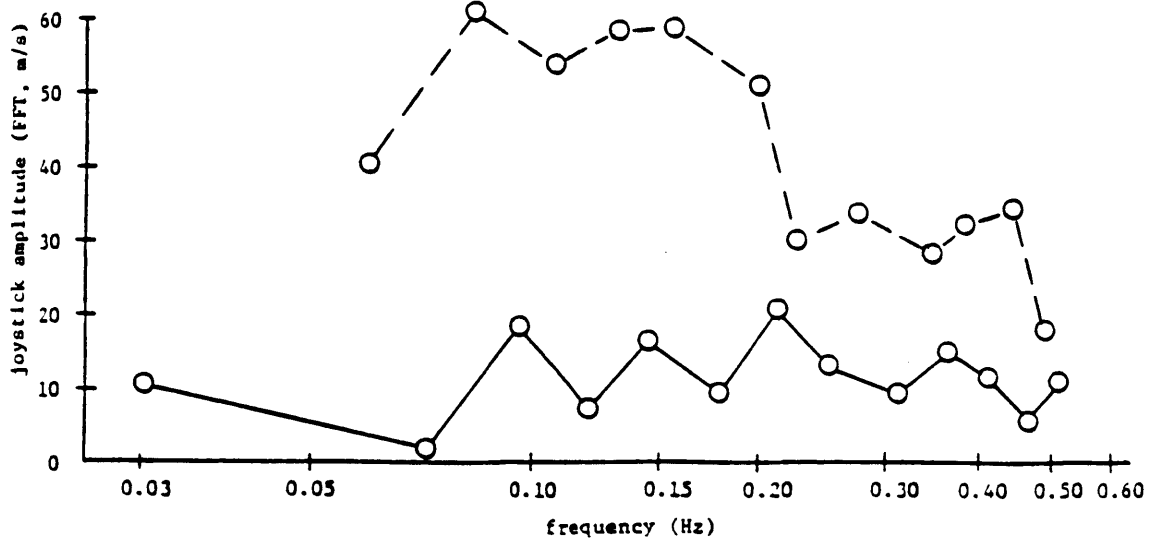
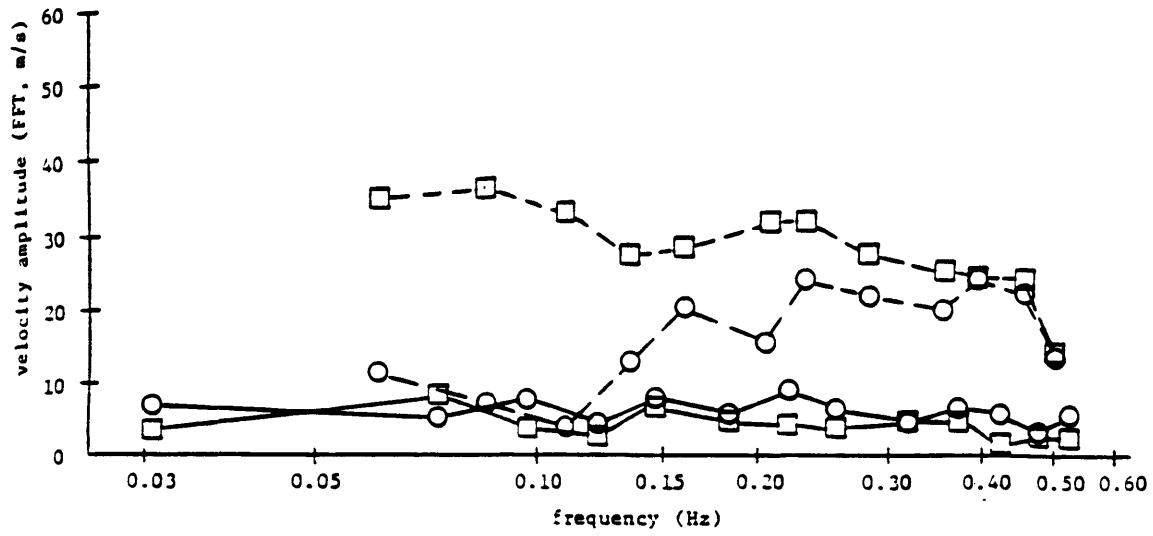


Figure E6.1.18 Frequency Spectrum, Subject JH
Run 06, Final Profile

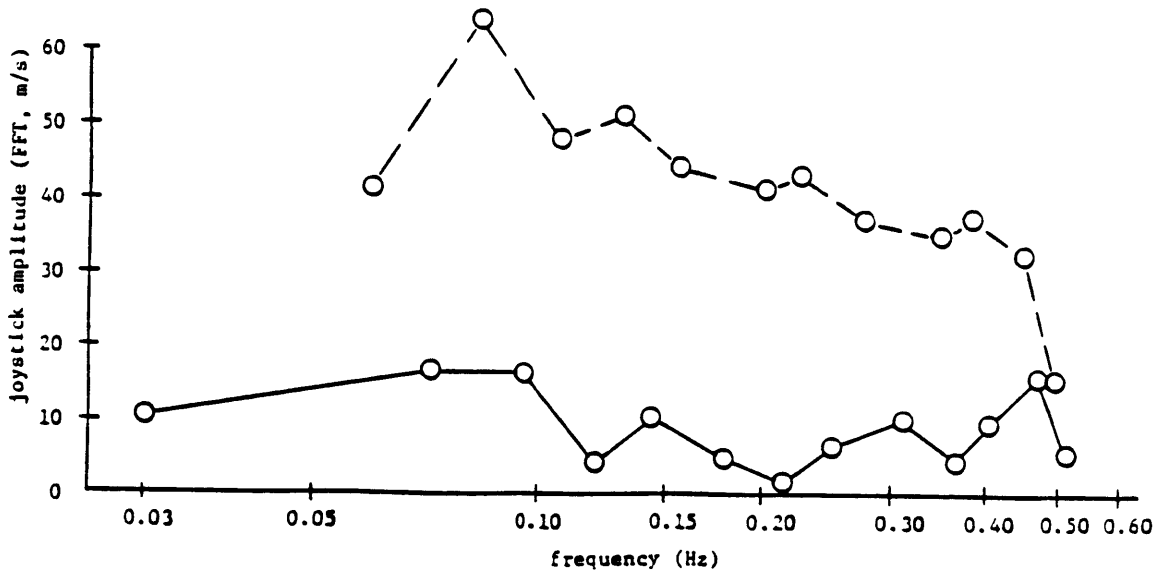
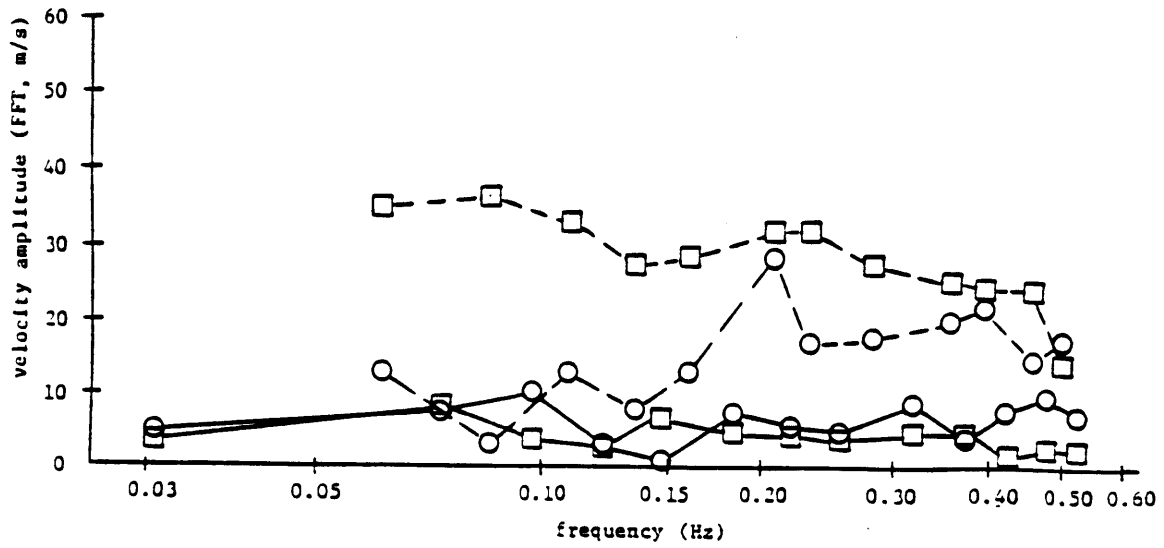


Figure E6.1.19 Frequency Spectrum, Subject JH
Run 08, Final Profile

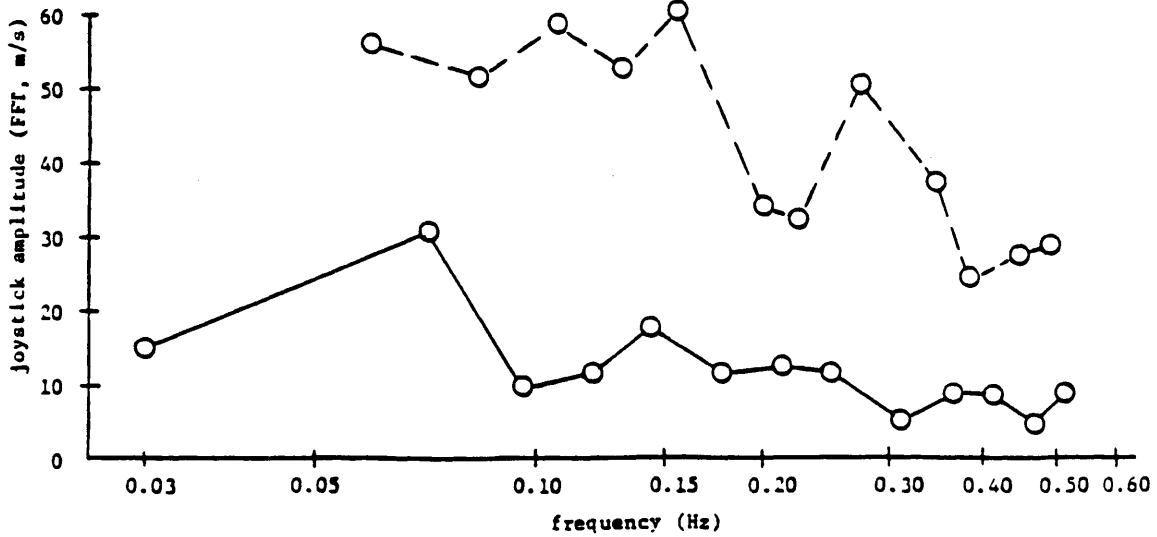
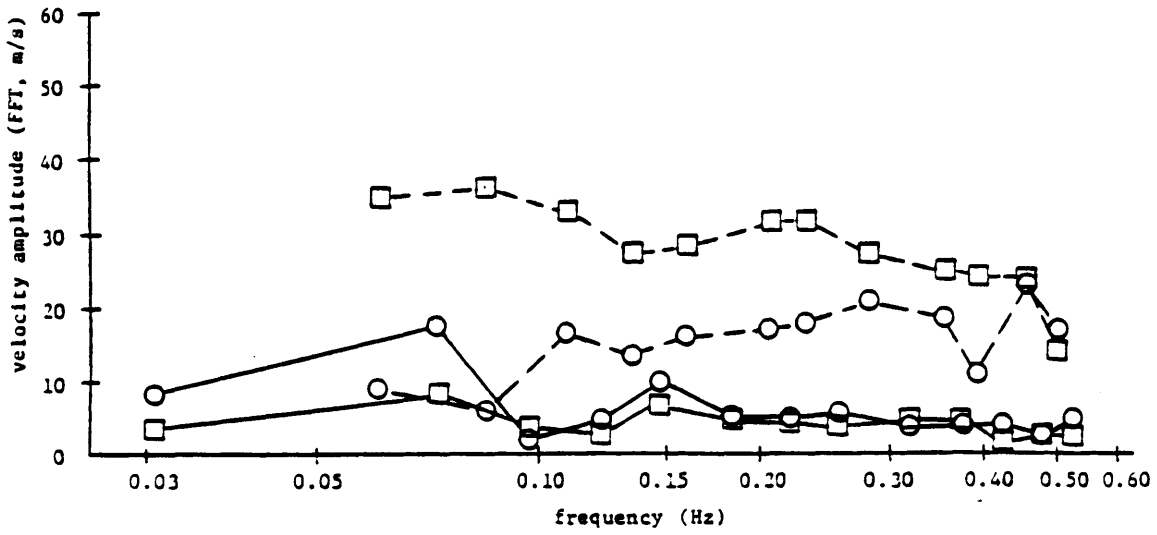


Figure E6.1.20 Frequency Spectrum, Subject JH
Run 09, Final Profile

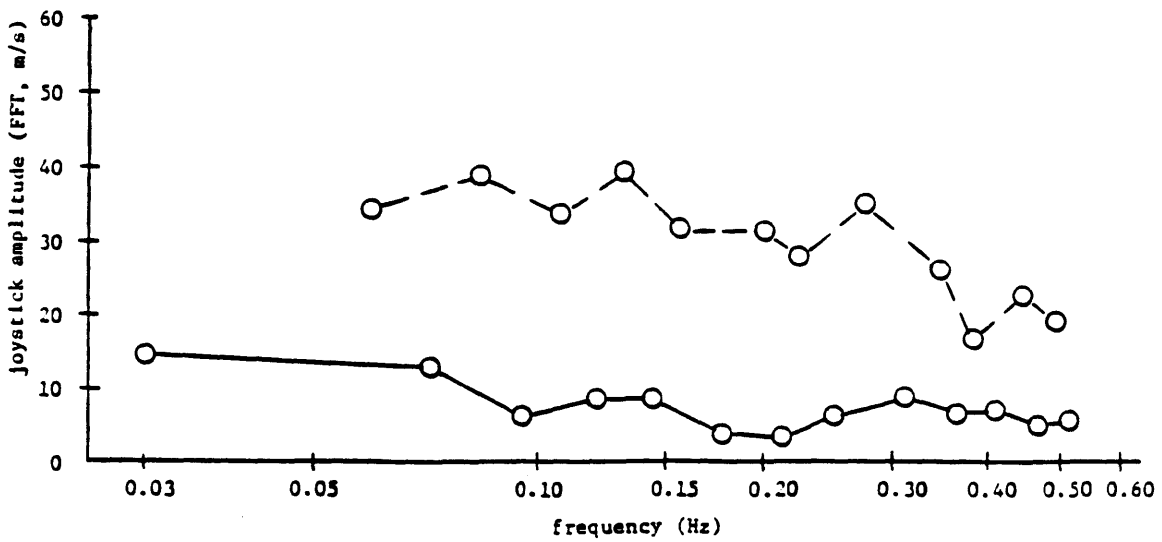
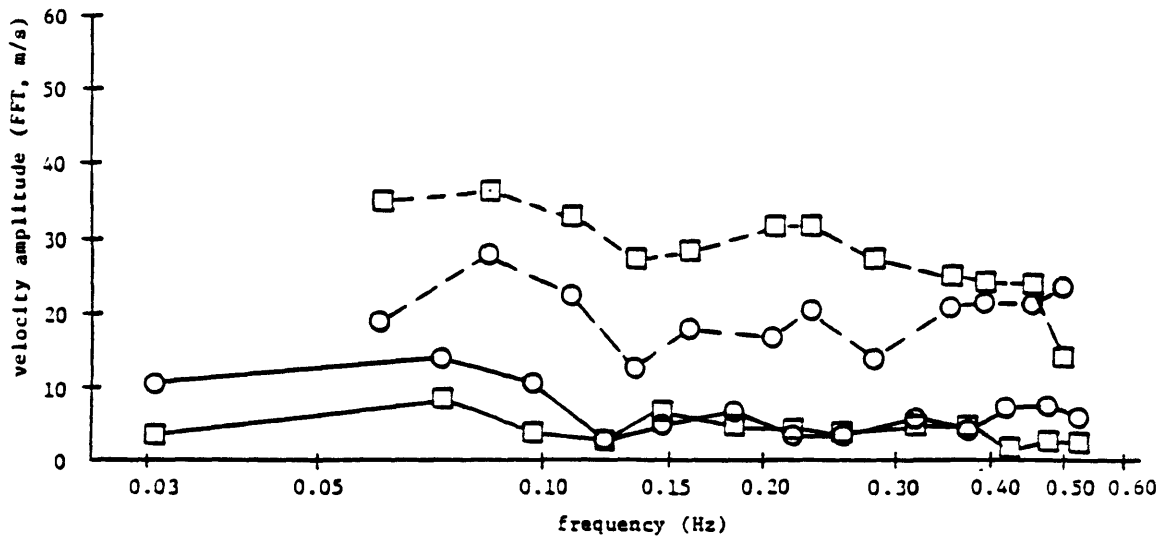


Figure E6.1.21 Frequency Spectrum, Subject DH
Run 12, Final Profile

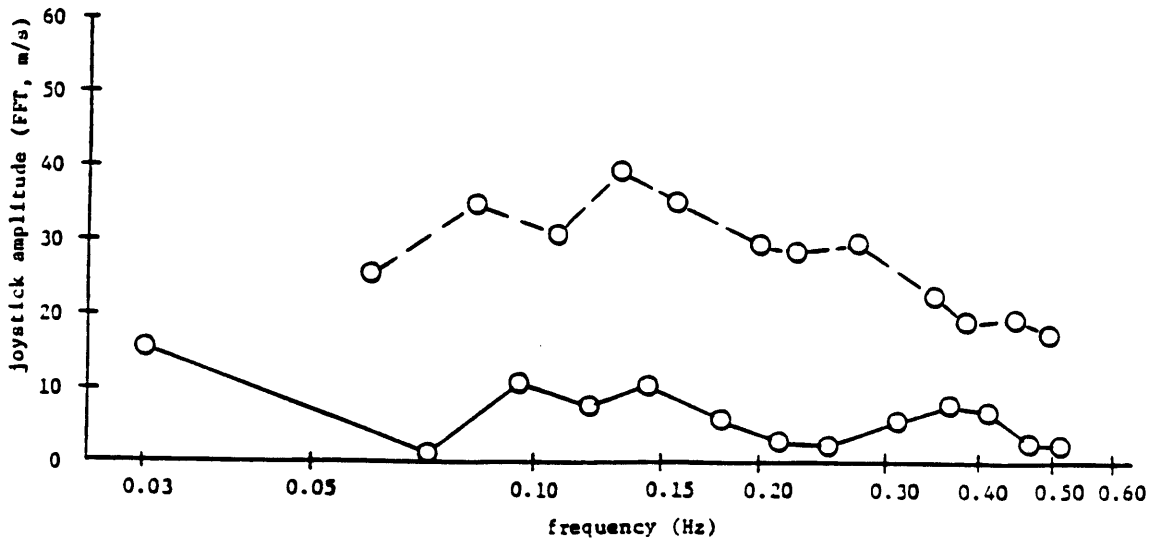
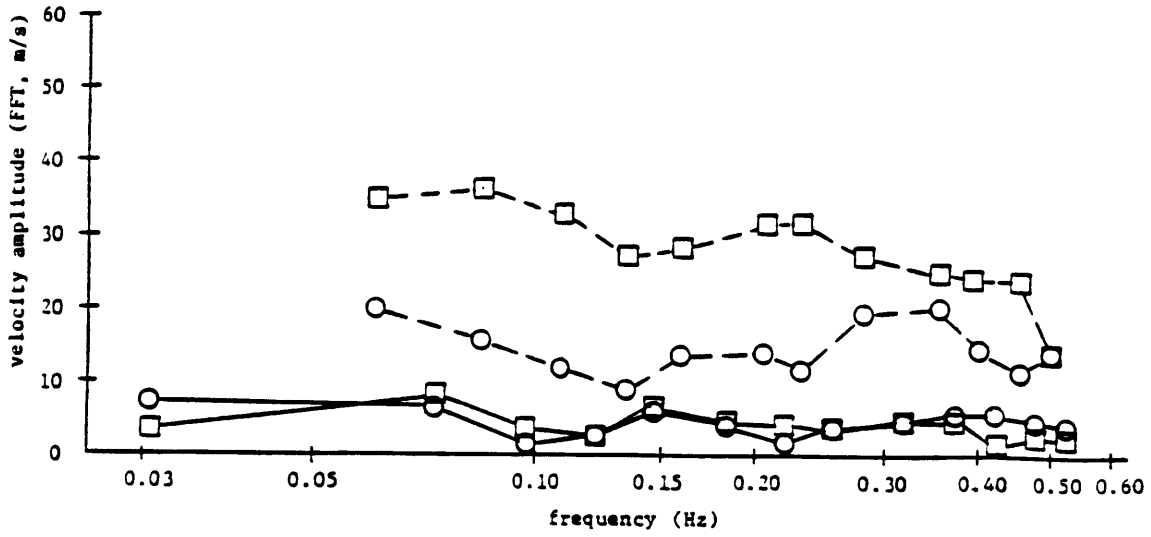


Figure E6.1.22 Frequency Spectrum, Subject DH
Run 15, Final Profile

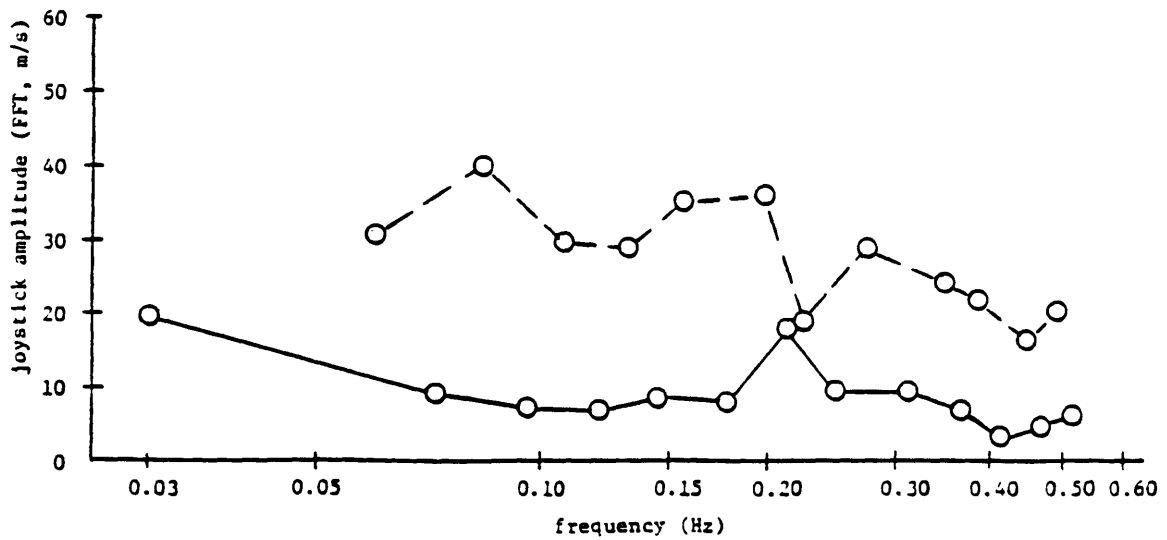
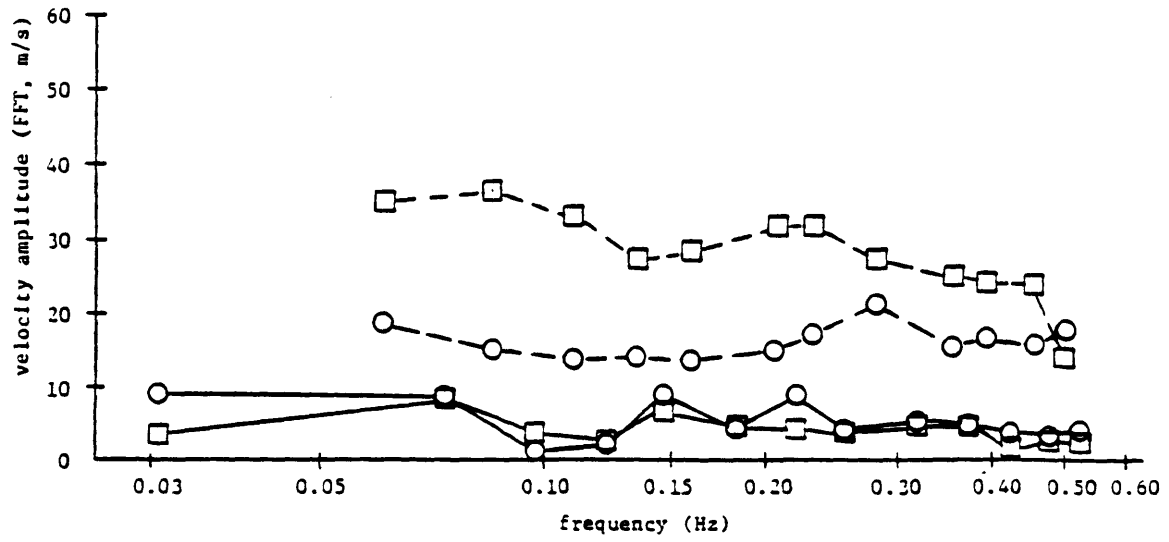


Figure E6.1.23 Frequency Spectrum, Subject DH
Run 21, Final Profile

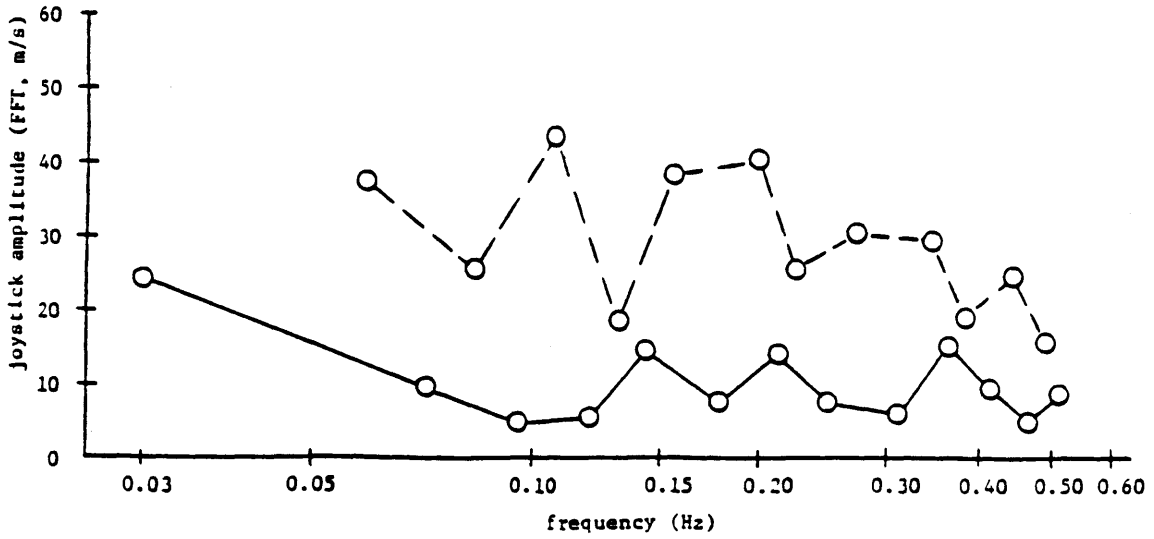
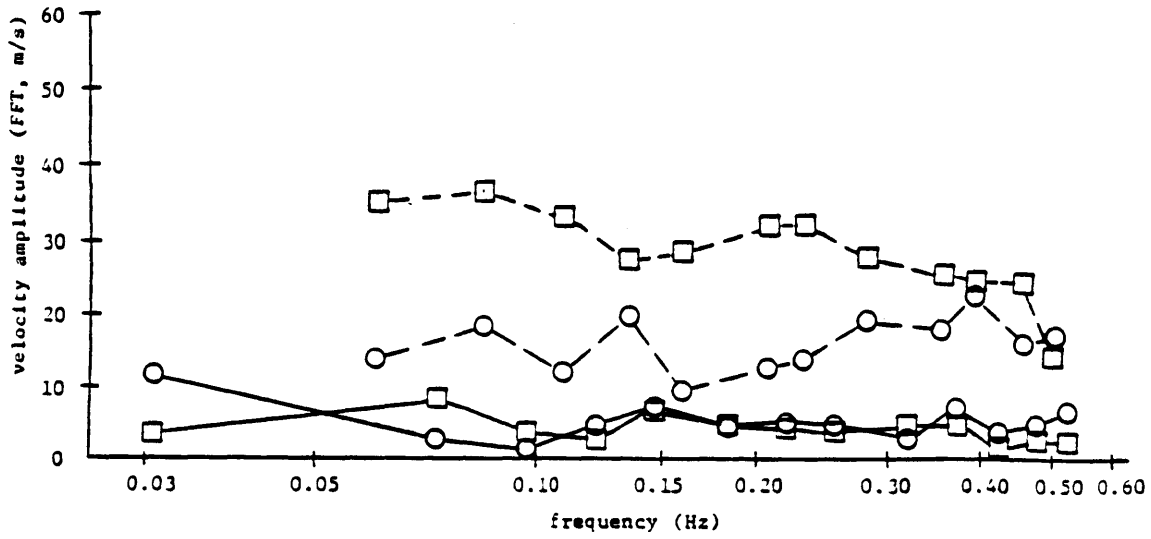


Figure E6.1.24 Frequency Spectrum, Subject DH
Run 22, Final Profile

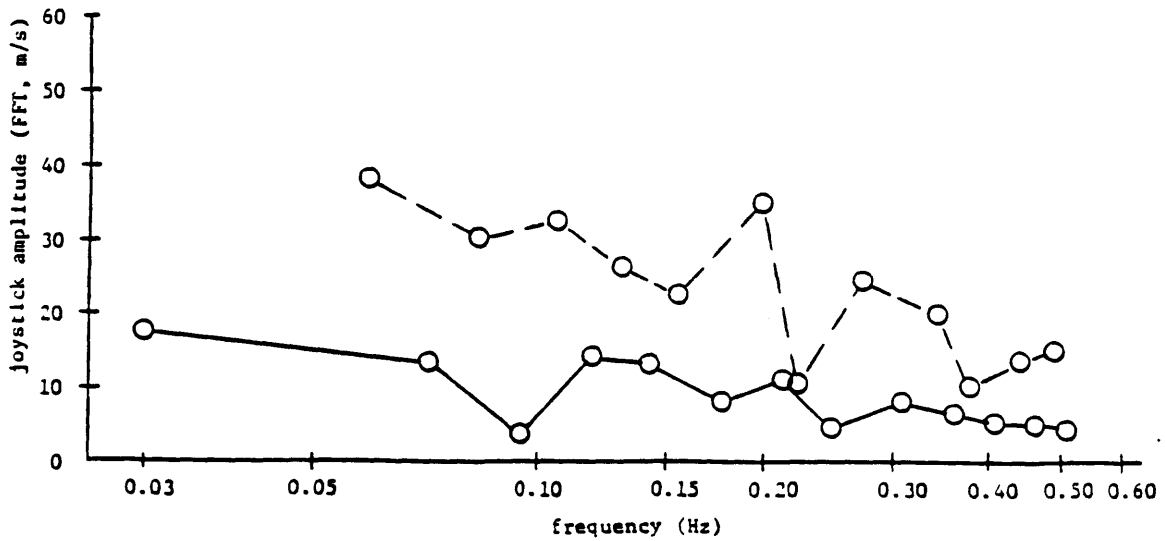
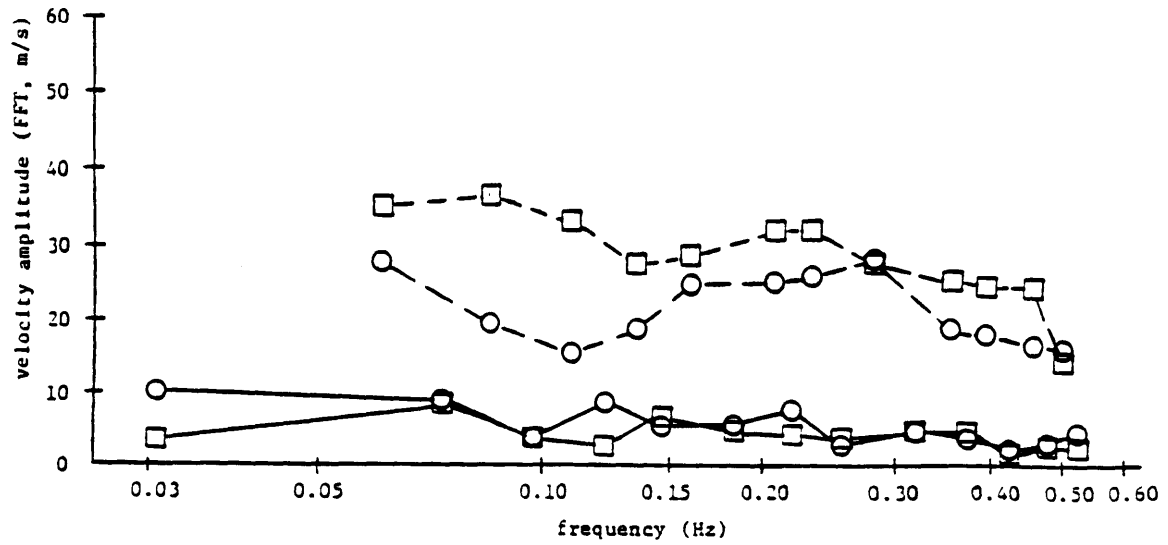


Figure E6.1.25 Frequency Spectrum, Subject DH
Run 23, Final Profile

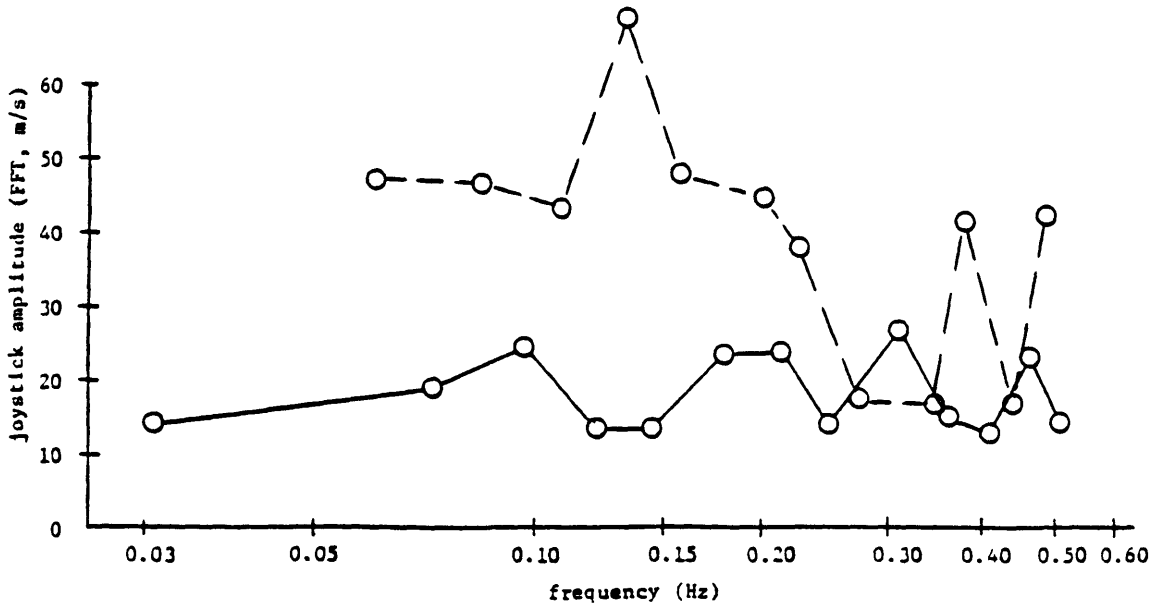
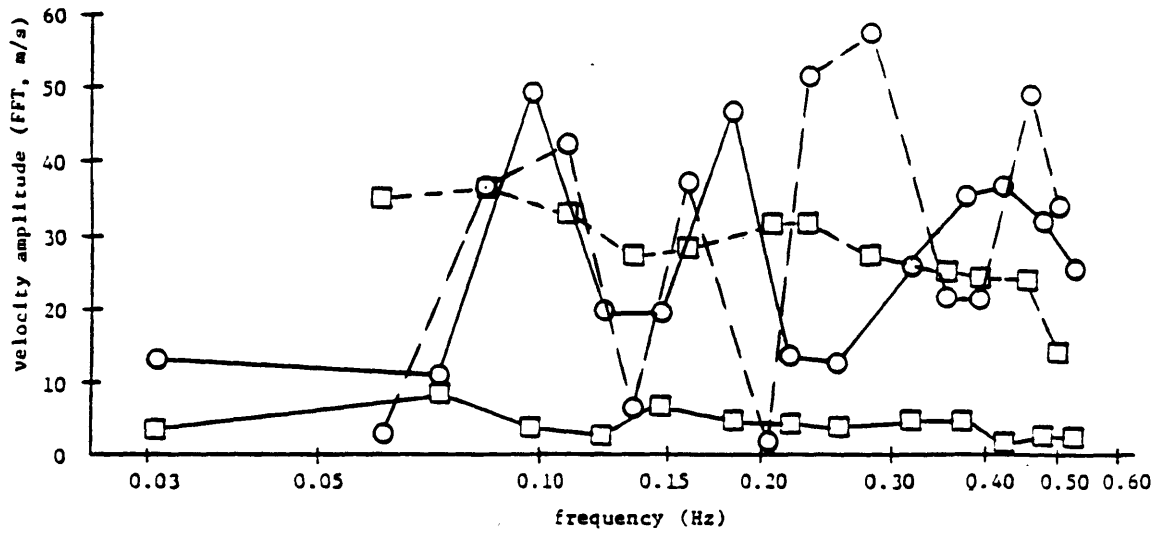


Figure E6.1.26 Frequency Spectrum, Subject LR
Run 02, Final Profile

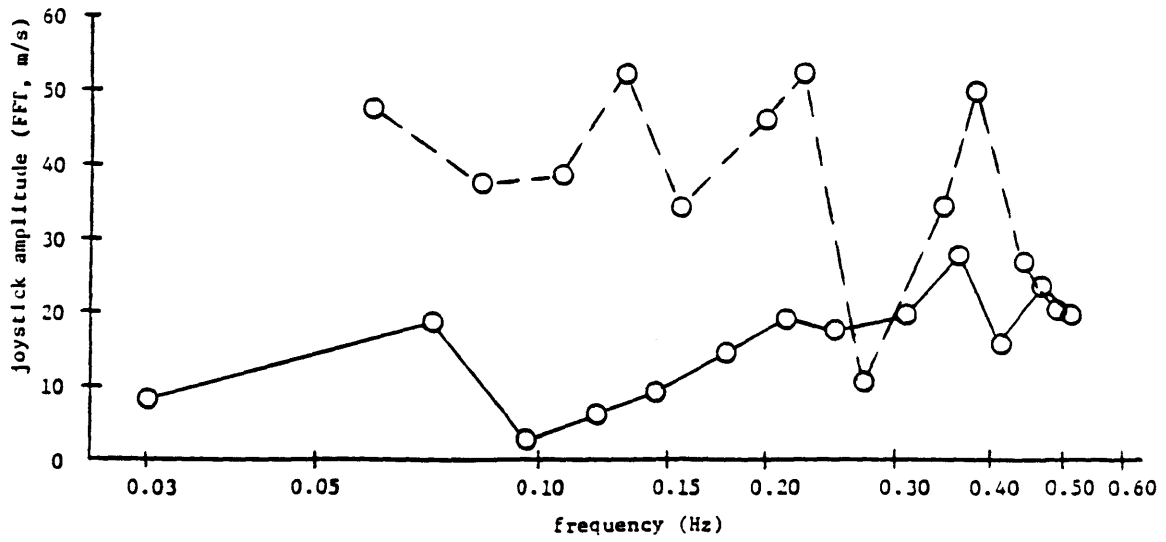
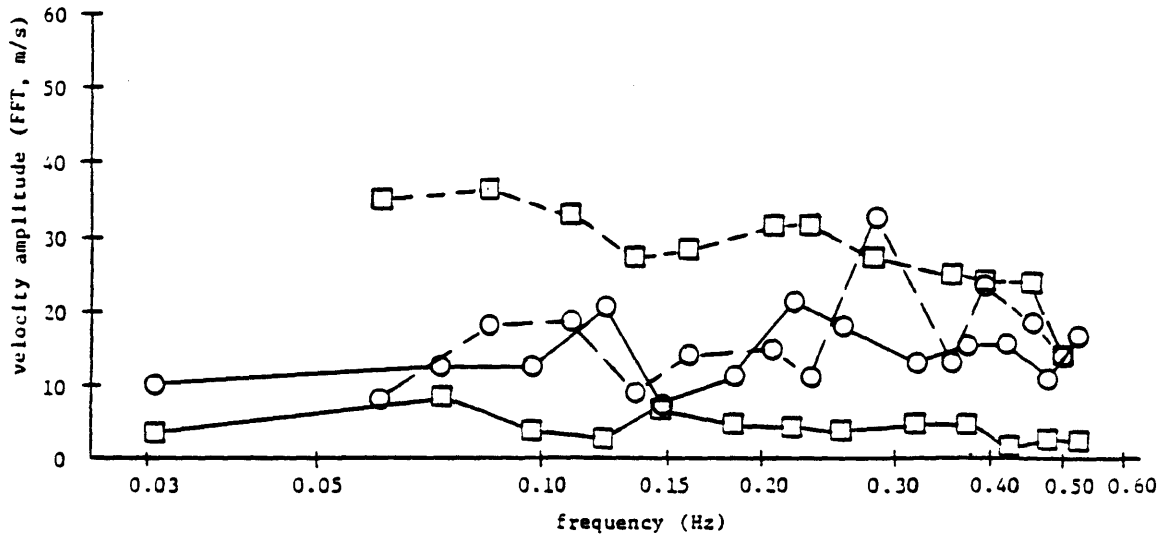


Figure E6.1.27 Frequency Spectrum, Subject LR
Run 03, Final Profile

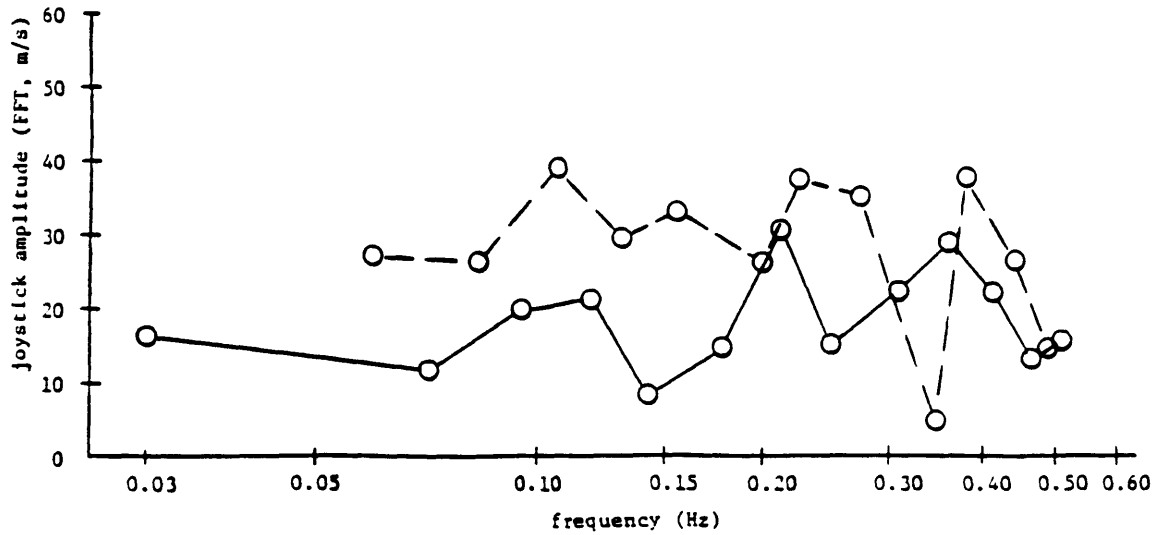
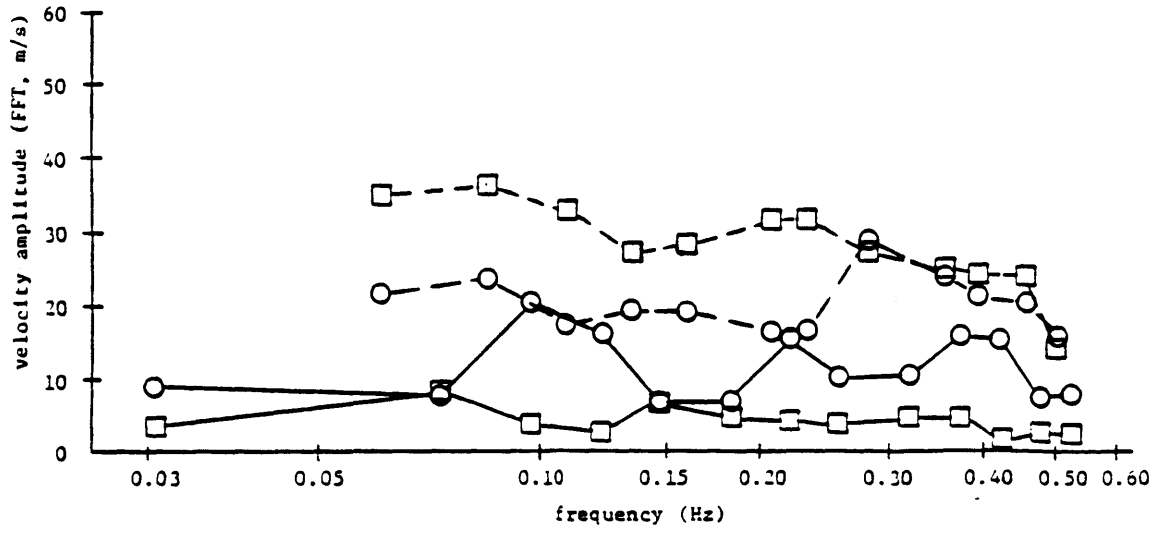


Figure E6.1.28 Frequency Spectrum. Subject LR
Run 06, Final Profile

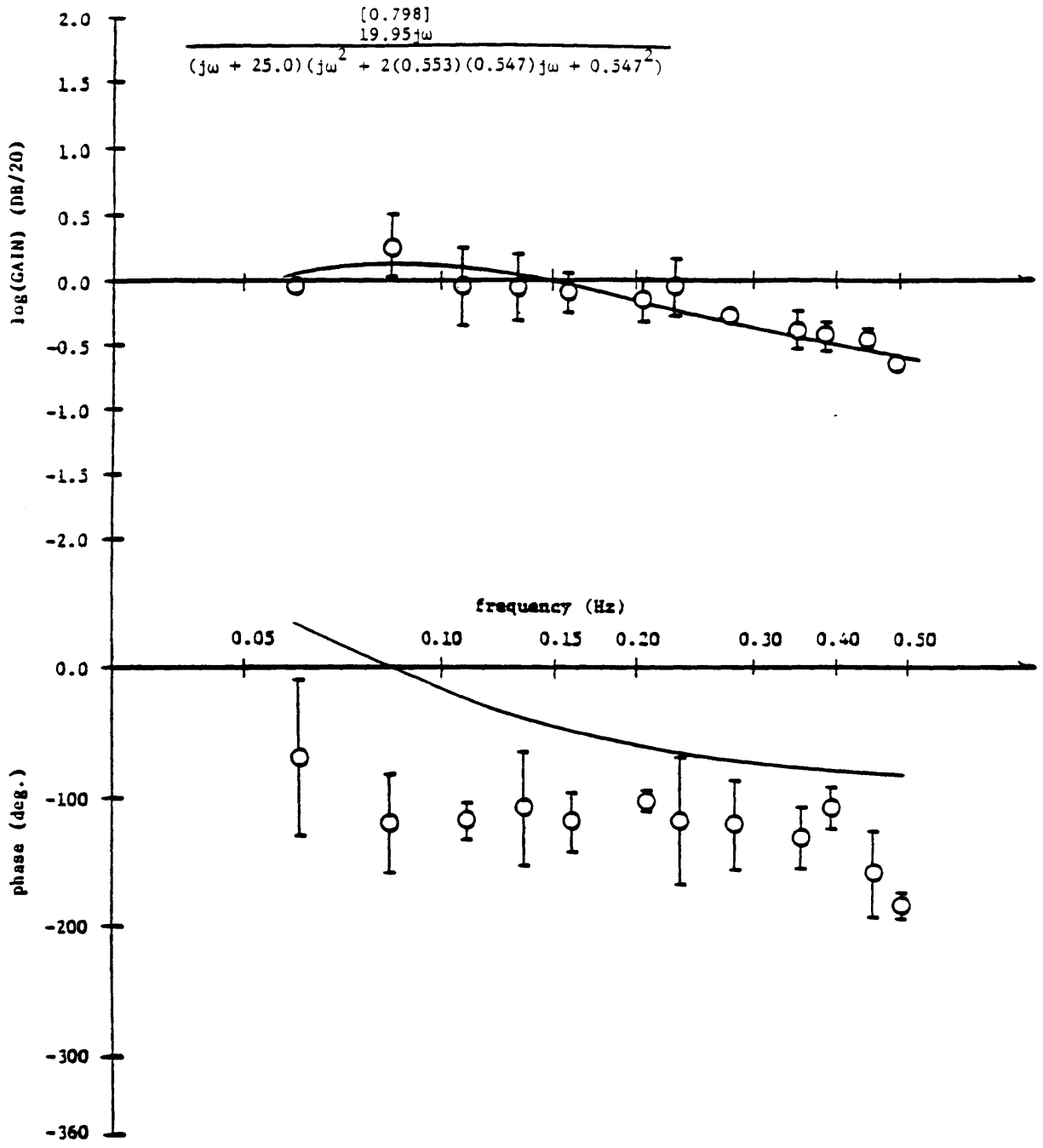


Figure E6.3.01 Model and Curve Fit. Subject MS

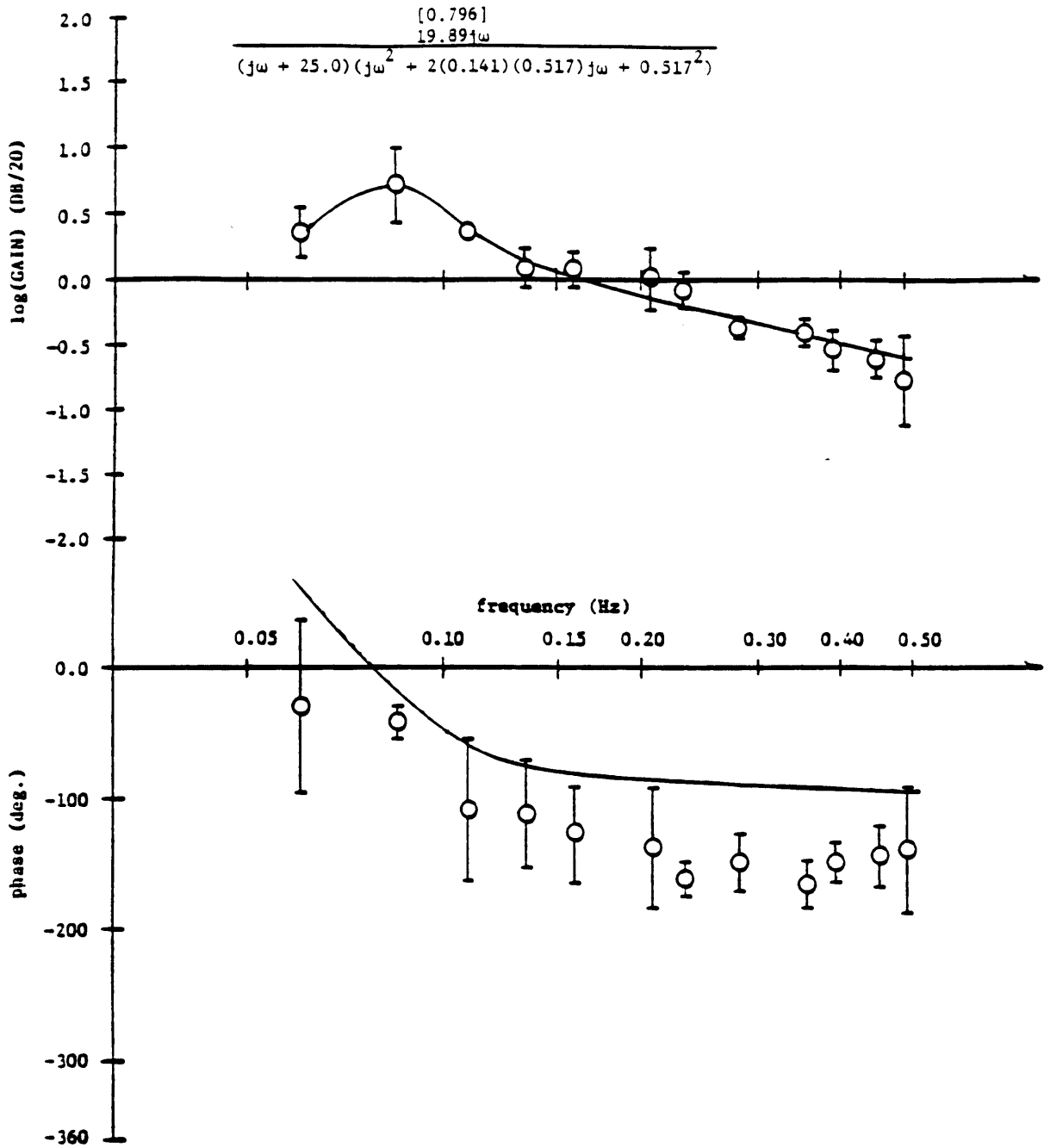


Figure E6.3.02 Model and Curve Fit. Subject MM

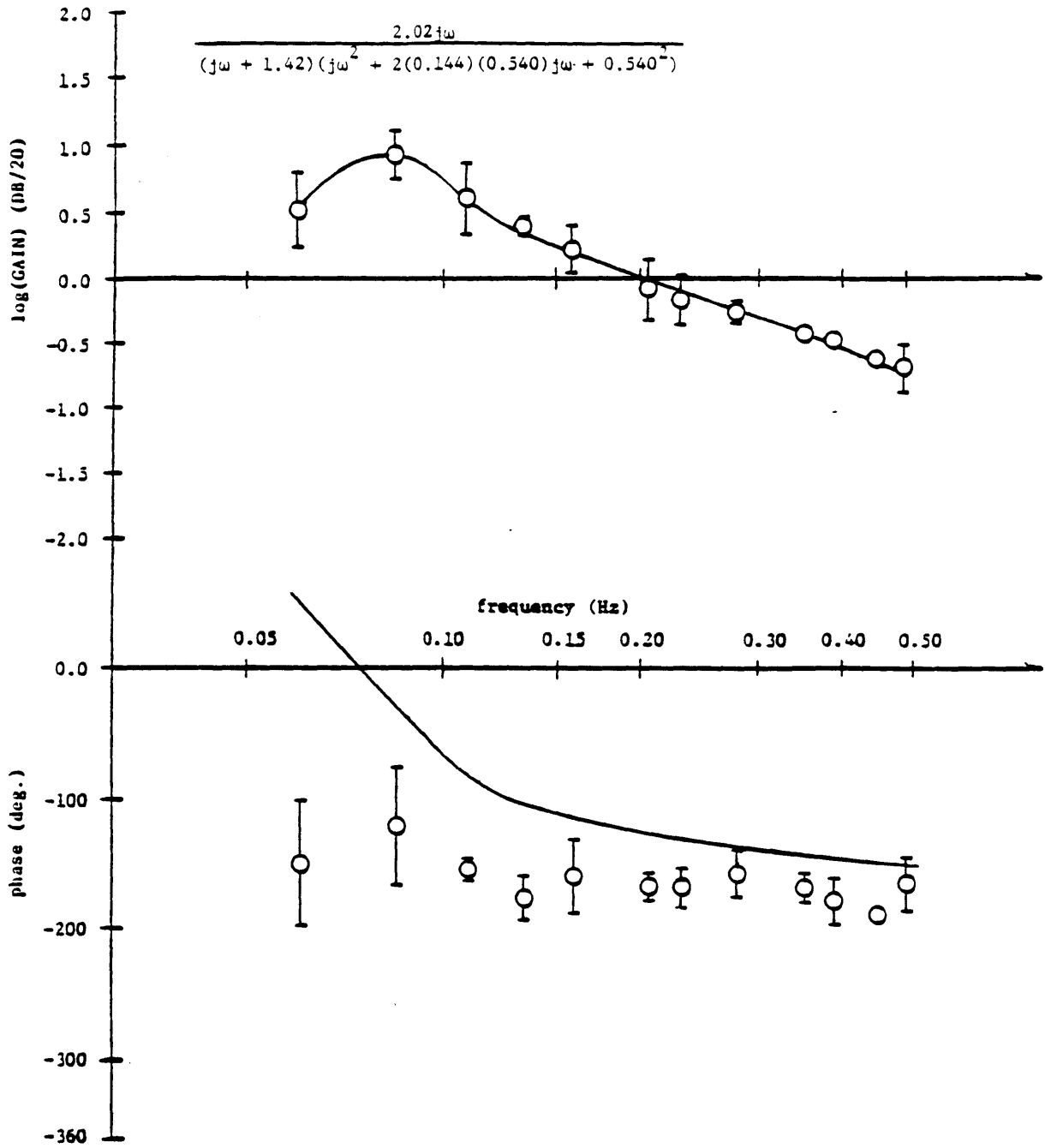


Figure E6.3.03 Model and Curve Fit. Subject JH

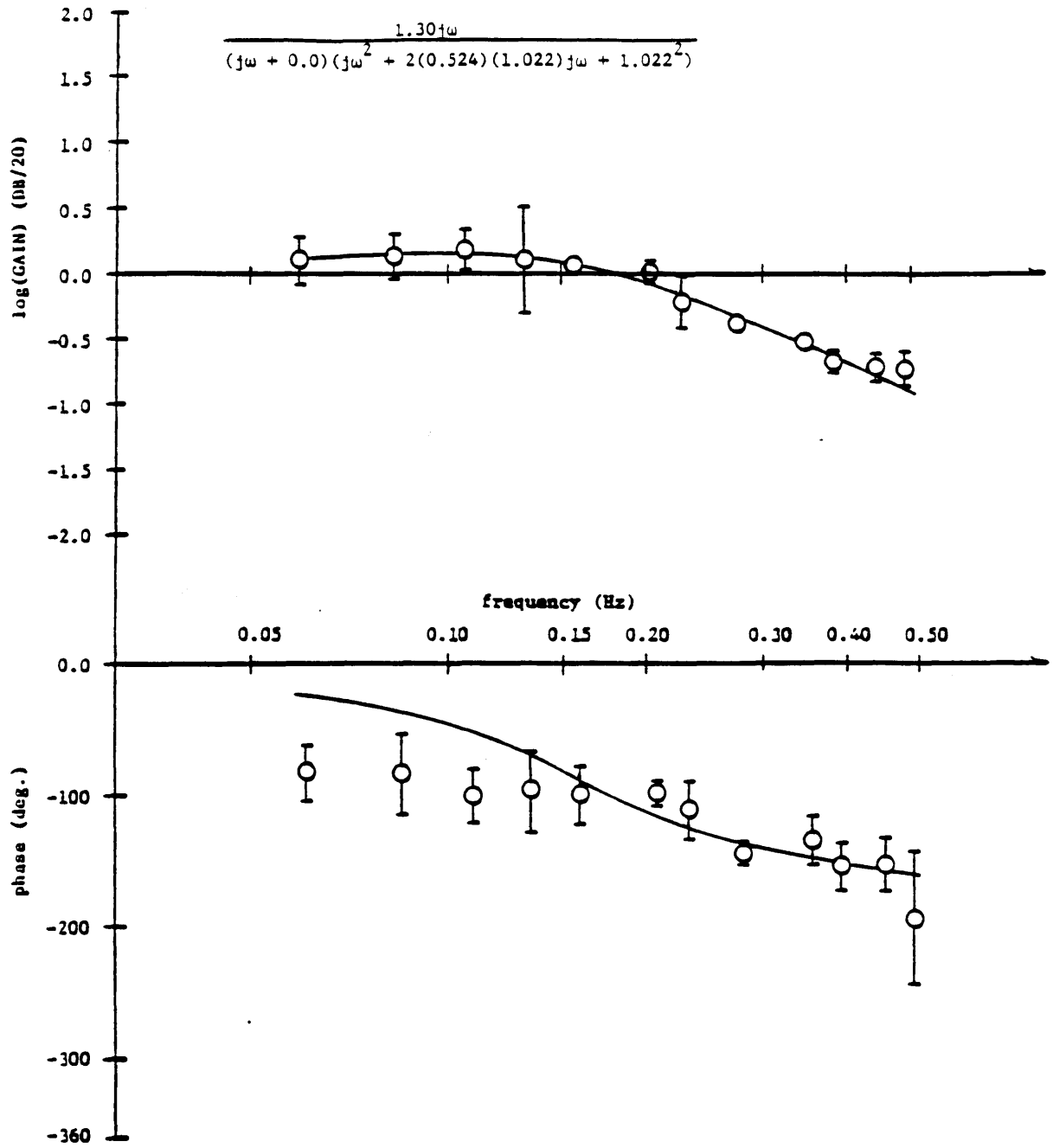


Figure E6.3.04 Model and Curve Fit. Subject DH

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