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## Research Article

# Modeling and Identification of Podded Propulsion Unmanned Surface Vehicle and Its Course Control Research

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The response model of podded propulsion unmanned surface vehicle (USV) is established and identified; then considering the USV has characteristic of high speed, the course controller with fast convergence speed is proposed. The idea of MMG separate modeling is used to establish three-DOF planar motion model of the podded propulsion USV, and then the model is simplified as a response model. Then based on field experiments, the parameters of the response model are obtained by the method of system identification. Unlike ordinary ships, USV has the advantages of fast speed and small size, so the controller needs fast convergence speed and strong robustness. Based on the theory of multimode control, a fast nonsingular terminal sliding mode (FNTSM) course controller is proposed. In order to reduce the chattering of system, disturbance observer is used to compensate the disturbance to reduce the control gain and RBF neural network is applied to approximate the symbolic function. At the same time, fuzzy algorithm is employed to realize the mode soft switching, which avoids the unnecessary chattering when the mode is switched. Finally the rapidity and robustness of the proposed control approach are demonstrated by simulations and comparison studies.

## 1. Introduction

With the development of marine technology, the USV has the advantages of being fast, small volume, low cost, and so on. It can work efficiently in harsh environments and is widely used in the field of ocean surveillance, search, rescue, and military [1]. In order to better accomplish the tasks in the complex ocean, USV puts forward higher requirements for its propulsion system. Podded propulsor is a new type of propulsion device, which can save space and improve ship propulsion efficiency. In the absence of rudder and side thruster [2], ship can achieve rotation, reverse and horizontal movement, and so forth. It is one of the most promising new technologies in the field of ship propulsion [3], so for USV, the use of POD propeller can not only meet the basic needs of the operation, but also improve the performance of navigation.

For now, the modeling and identification of propellerrudder propulsion ship are very mature, but in contrast only a few articles study podded propulsion ship. For a small POD propeller USV, on the basis of studying the operating characteristics under different speeds, the model of three-DOF is obtained by the method of system identification, and the identification result is verified [4]. Japanese scholars Haraguchi and Nimura use the changing of course to study the model of podded propulsion ship and evaluate its operating performance [5]. Aiming at a large podded propulsion ship, [6] uses the idea of separate modeling to propose the concept of equivalent rudder. For a podded propulsion semisubmersible ship, normal velocity field Kijima model and low velocity field Yoshimura model are combined to establish a three-DOF model, but the accuracy of the model is not verified [7]. Paper [8] uses regression formula to estimate the thrust force generated by POD, and four-DOF model of semisubmersible ship is established. Paper [9] establishes a three-DOF model of podded propulsion ship by regression formula method, and based on this model, ADRC control theory is used to design the course controller. Generally, the research on modeling of podded propulsion ship is mostly limited to MMG model, but in fact in the design process of the actual ship controller, the most widely application is response model [10]. So the first step of this paper is to establish the USV response model of podded propulsion; then the method of system identification is employed to get the model param-

The design of course controller is an important research topic of ship motion control, because it is related to not only the economy and security of the ship's operation, but also the foundation of USV's autonomous navigation. Sliding mode control (SMC) is applied to design ship course controller in [11], but the SMC high frequency chattering phenomenon is difficult to solve. In order to solve the chattering problem, [12] uses fuzzy control to optimize control signal to design the fuzzy sliding mode controller, and simulation results show that the performance is better than conventional SMC control. Paper [13] combines SMC and ADRC to design ship course and track controller, and the sideslip angle caused by wave current is solved by means of coordinate transformation. In [14], backstepping and terminal sliding mode are combined to design the course controller, and the fuzzy approximation method is used to solve the model uncertainty. In [15], the adaptive backstepping is adopted to design course controller, but the backstepping's "computation explosion" problem has brought great difficulty to the controller design. In view of this problem, in [16], technique of dynamic surface is used to solve the problem of multiple derivation function effectively. On the basis of considering the input saturation, backstepping is used to design course controller and neural network approximation is employed to solve the problem of model uncertainty [17].

As previously mentioned, such as [5–9], there is little research on the identification of response model of podded propulsion ship. Meanwhile, unlike ordinary ships, USV has a faster speed and smaller volume, so the controller needs faster convergence speed and better anti-interference ability. That is, to a certain extent, the course controller of an ordinary ship is no longer applicable to USV (such as the methods proposed in [11–17]). So the contributions of this paper are that firstly, by analyzing the force acting on the hull and the thrust force of POD, the response model is derived; secondly, on the basis of field experiments, the response model of USV is identified by the method of system identification; finally, according to the characteristics of USV, based on the theory of multimode control and fuzzy weighted idea, a fast convergent course controller is proposed.

The rest of the paper is organized as follows. Section 2 introduces the experimental subject of Dalian Maritime University Lanxin USV and describes how the MMG model is simplified as a response model. In Section 3, the parameters of response model are identified and the result of identification is verified. Section 4 introduces the FNTSM course controller design. In Section 5, numerical simulations are carried out to show the rapidity and robustness of our design. Finally, some conclusions are made and future research directions are introduced in Section 6.

## 2. USV Modeling

At present, there are two main types of hydrodynamic model of the ship maneuvering: Abkowitz model and MMG model [18]. From the practical experience, an overly complex or simple model cannot better describe the characteristics of system. Weighing the pros and cons of the two models,



FIGURE 1: Lanxin USV.



FIGURE 2: The structure of POD propeller device.

MMG model is employed to study the mathematical model of Lanxin USV.

2.1. Lanxin USV and Its POD. Lanxin USV is a small intelligence equipment platform of Dalian Maritime University, which has the function of water sampling, marine monitoring, maritime rescue, and so forth. It is shown in Figure 1.

Its POD propeller device is shown in Figure 2.

It can be seen from Figure 2 that its propeller and rudder are integrated. When the propulsor changes its propulsion angle, the vector thrust produced by the propeller can be decomposed into two directions: keep the forward thrust of the ship and produce lateral thrust to achieve steering. In the application process, the actual propulsion angle range is  $|\delta| \le 35^\circ$  and its rate of change is limited to  $|\dot{\delta}| \le 10^\circ/\text{s}$ .

2.2. Planar Motion Model. Conventionally, the attached body coordinate system and the inertial coordinate system are adopted to study the model of USV. As shown in Figure 3,  $O-X_0Y_0Z_0$  is the inertial coordinate system and o-xyz is the attached body coordinate system. USV has six DOF including the surge velocity u, sway velocity v, heave velocity w, yaw rate r, rolling rate p, and pitching angle q.  $\psi$  is course angle and  $\delta$  is propulsion angle.

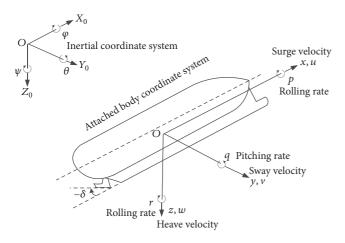


FIGURE 3: The inertial coordinate system and the coordinate system.

Based on [18], heave velocity, rolling rate, and heave velocity can be ignored. Meanwhile, we can premise that USV is symmetrical and the origin of the attached coordinate system and the barycenter of the USV are coincident. So six-DOF MMG model can be simplified to a three-DOF planar motion model.

$$(m + m_x) \dot{u} - (m + m_y) vr = X,$$

$$(m + m_y) \dot{v} + (m + m_x) ur = Y,$$

$$(I_{zz} + J_{zz}) \dot{r} = N,$$
(1)

where m is the mass of the USV,  $m_x$  is the additional mass in the x-axis direction,  $m_y$  is the additional mass in the y-axis direction,  $I_{zz}$  is the moment of inertia of the  $o_x$  axis,  $I_{zz}$  is the additional moment of inertia in the direction of the z-axis, X, Y, and N are the hydrodynamic forces and moments acting on the hull. According to [18], X, Y, and N can be divided as

$$X = X_H + X_p,$$
 
$$Y = Y_H + Y_p,$$
 
$$(2)$$
 
$$N = N_H + N_p,$$

where H is the hydrodynamic force acting on the bare hull and p is the propulsor force. The plane motion variables of USV are shown in Figure 4.

V is the speed of USV. According to paper [18], the hydrodynamic forces are

$$\begin{split} X_{H} &= X\left(u\right) + X_{Hvv}v^{2} + X_{Hvr}vr + H_{Hrr}r^{2}, \\ Y_{H} &= Y_{Hv}v + Y_{Hr}r + Y_{NL}, \\ N_{H} &= N_{Hv}v + N_{Hr}r + N_{NL}, \end{split} \tag{3}$$

where  $Y_{\rm NL}$  and  $N_{\rm NL}$  are nonlinear fluid dynamics and they can be ignored if considered as high order small quantity. According to [4], the propulsor thrust can be expressed as

$$T_{p} = cV\delta_{n} + d\left|\delta_{n}\right|\delta_{n},\tag{4}$$

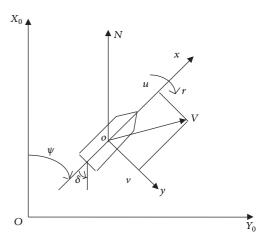


FIGURE 4: Planar motion variable description.

where  $T_p$  is propulsor thrust,  $\delta_n$  is the rotation speed of propeller, and c and d are the coefficients that are greater than zero. When the propulsion angle is  $\delta$ , the vector thrust in different directions is

$$X_{P} = (cV\delta_{n} + d |\delta_{n}| \delta_{n}) \cos \delta,$$

$$Y_{P} = (cV\delta_{n} + d |\delta_{n}| \delta_{n}) \sin \delta,$$

$$N_{P} = x_{P} (cV\delta_{n} + d |\delta_{n}| \delta_{n}) \sin \delta,$$
(5)

where  $x_p$  is the length from the center of rotation to the fulcrum of the propulsor.

2.3. Derivation and Simplification. Assume that the outside interference is small; at this point the linear hydrodynamic force acting on USV occupies a dominant position, and the higher order terms can be ignored. In the field of ship model research, the linear motion of the ship is regarded as the balance state. Assume  $u = u_0$ , v = 0, r = 0, and  $\delta = 0$ , where  $u_0$  is the longitudinal initial velocity of USV. When the USV is disturbed by interference,  $u = u_0 + \Delta u$ ,  $\Delta v = v$ ,  $\Delta r = r$ , and  $\Delta \delta = \delta$ . On the basis of this, formula (1) can be simplified as

$$(m + m_x) \Delta \dot{u} = X,$$

$$(m + m_y) \dot{v} + (m + m_x) u_0 r = Y,$$

$$(I_{zz} + J_{zz}) \dot{r} = N.$$
(6)

Keep first-order small quantities  $\Delta u$ , v, r,  $\delta$  and ignore high order small quantities. So

$$\begin{split} X_{H} &= X \left( u_{0} + \Delta u \right), \\ Y_{H} &= Y_{Hv}v + Y_{Hr}r, \\ N_{H} &= N_{Hv} + N_{Hr}r, \end{split} \tag{7}$$

where  $X_H$  is the direct resistance of USV.

$$X\left(u_0 + \Delta u\right) = -\frac{1}{2}\rho SC_t \left(u_0 + \Delta u\right)^2, \tag{8}$$

where *S* is the area of wet,  $\rho$  is water density,  $C_t$  is the total drag coefficient, and its essence is a function of speed.

$$X\left(u_{0} + \Delta u\right)$$

$$= -\frac{1}{2}\rho S \left[C_{t0} + \left(\frac{\partial C_{t}}{\partial \Delta u}\right)_{t0} \Delta u\right] \left(u_{0} + \Delta u\right)^{2}.$$
(9)

When the speed is  $u_0$ , its total resistance coefficient is  $C_{t0}$ .  $\Delta u$  of formula (8) is linearized as

$$X_{H} = -\frac{1}{2}\rho S C_{t0} u_0^2$$

$$-\frac{1}{2}\rho S \left[ 2C_{t0} u_0 + \left( \frac{\partial C_t}{\partial \Delta u} \right)_{u0} u_0^2 \right] \Delta u.$$
(10)

Define

$$\begin{split} X_0 &= -\frac{1}{2} \rho S C_{t0} u_0^2 \\ X_{Hu} &= -\frac{1}{2} \rho S \left[ 2 C_{t0} u_0 + \left( \frac{\partial C_t}{\partial \Delta u} \right)_{u0} u_0^2 \right], \end{split} \tag{11}$$

where  $X_0$  is the straight line resistance of the unmanned vehicle in the initial state.

Then  $X_H = X_0 + X_{Hu} \Delta u$ .

So formula (7) can be simplified as

$$X_{H} = X_{0} + X_{Hu}\Delta u,$$

$$Y_{H} = Y_{Hv}v + Y_{Hr}r,$$

$$N_{H} = N_{Hv} + N_{Hr}r.$$
(12)

Considering that  $\delta$  is small, so  $\sin \delta \approx \delta$  and  $\cos \delta \approx 1$ . Therefore formula (5) can be changed as

$$X_{P} = cV\delta_{n} + d \left| \delta_{n} \right| \delta_{n},$$

$$Y_{P} = \left( cV\delta_{n} + d \left| \delta_{n} \right| \delta_{n} \right) \delta,$$

$$N_{P} = x_{\delta s} \left( cV\delta_{n} + d \left| \delta_{n} \right| \delta_{n} \right) \delta.$$
(13)

Substituting (12) and (13) into (2) results in

$$X = X_0 + X_{pu}\Delta u + X_p,$$

$$Y = Y_H + Y_p = Y_{Hv}v + Y_{Hr}r + X_p\delta,$$

$$N = N_H + M_p = N_{Hv}v + N_{Hr}r + x_{\delta s}X_p\delta.$$
(14)

Under the initial condition, the resistance of the USV is equal to the thrust of propulsor. That means  $X_0+X_p=0$ , so we can draw that

$$X = X_u \Delta u,$$

$$Y = Y_{Hv} v + Y_{Hr} r + X_p \delta,$$

$$N = N_{Hv} v + N_{Hr} r + x_{\delta s} X_p \delta.$$
(15)

Substituting (15) into (6) results in

$$(m + m_x) \Delta \dot{u} = X_u \Delta u,$$

$$(m + m_y) \dot{v} + (m + m_x) u_0 r = Y_{Hv} v + Y_{Hr} r + X_p \delta, \quad (16)$$

$$(I_{zz} + J_{zz}) \dot{r} = N_{Hv} v + N_{Hr} r + x_{\delta s} X_p \delta.$$

We can assume that USV is disturbed by small disturbances and the longitudinal velocity is constant. Formula (16) can be divided as (17) and (18).

$$(m+m_r)\Delta\dot{u} = X_u\Delta u,\tag{17}$$

$$(m + m_y)\dot{v} + (m + m_x)u_0r = Y_{Hv}v + Y_{Hr}r + X_p\delta, (I_{zz} + J_{zz})\dot{r} = N_{Hv}v + N_{Hr}r + x_{\delta\delta}X_p\delta.$$
(18)

Define  $Y_{H\nu}=Y_{\nu}, Y_{Hr}=Y_{r}, X_{p}=Y_{\delta}, N_{H\nu}=N_{\nu}, N_{Hr}=N_{r}$ , and  $x_{\delta s}X_{p}=N_{\delta}$ . Then formula (18) can be simplified as

$$(m + m_y)\dot{v} = Y_v v + (Y_r - (m + m_x)u_0 r) + Y_\delta \delta,$$

$$(I_{zz} + J_{zz})\dot{r} = N_v v + N_r r + N_\delta \delta.$$
(19)

In order to simplify the problem, assuming that the initial state is uniform motion, and all the motion variables have zero initial value, then  $\Delta u(0) = 0$ , v(0) = 0, r(0) = 0,  $\dot{v}(0) = 0$ ,  $\delta(0) = 0$ , and  $\dot{\delta}(0) = 0$ . After Laplace transformation, formula (19) is converted to (20) and (21).

$$(m + m_y) sv(s) = Y_v v(s) + (Y_r - (m + m_x) u_0) r(s) + Y_\delta \delta(s),$$
(20)

$$(I_{zz} + I_{zz}) sr(s) = N_{\nu} v(s) + N_{r} r(s) + N_{\delta} \delta(s).$$
 (21)

For ship control, we pay more attention to the relationship between the course and the propulsion angle, so we only analyze formula (21). By Laplace inverse transform, formula (21) can be transformed into a linear response equation in time domain.

$$T_1 T_2 \ddot{r} + (T_1 + T_2) \dot{r} + r = K (\delta + T_3 \dot{\delta}).$$
 (22)

The relationships between  $T_1$ ,  $T_2$ ,  $T_3$ , and K are shown in formulas (22)~(27). Their specific significance can refer to [18].

$$T_1 T_2 = \left(\frac{L}{V}\right) \frac{\left(m' + m'_{y}\right) \left(I'_{zz} + I'_{zz}\right)}{C'},\tag{23}$$

$$T_1 + T_2 = \left(\frac{L}{V}\right) \frac{-\left(m' + m'_y\right) N'_r - \left(I'_{zz} + J'_{zz}\right) Y'_v}{C'},\tag{24}$$

$$K = \left(\frac{V}{L}\right) \frac{N_{\nu}' Y_{\delta}' - N_{\delta}' Y_{\nu}'}{C'},\tag{25}$$

$$KT_3 = \frac{\left(m' + m_y'\right)N_\delta'}{C'},\tag{26}$$

$$C' = Y_{\nu}' N_{r}' - \left[ Y_{r}' - \left( m' + m_{x}' \right) \right] N_{\nu}', \tag{27}$$

where m' is the normalization of m. Similarly  $m'_x$ ,  $m'_y$ , and  $I'_{zz}$  and so on with superscript express their normalization. In this state, the fluid force derivative is no longer the slope when  $\nu$  and r are zero. So in the nonlinear case, various linear hydrodynamic derivatives will depend on the changing of r. When r is changed, formulas  $(23)\sim(27)$  are bound to change. C' is known as the stability index, when C'>0 the ship has the stability of the course; when C'<0 the ship does not have the stability of the course; when  $C'\cong0$  the ship shows a very strong nonlinear characteristic. So Saku [19] proposes formula (27).

$$C' = C_0' + nr^2. (28)$$

When r = 0,  $C' = C'_0$ . Taking into account that C' has the symmetry of r, so n as a new constant is introduced. Substituting (28) into (22) results in

$$\left(\frac{L}{V}\right)^{2} \left(m' + m'_{y}\right) \left(I'_{zz} + J'_{zz}\right) \ddot{r} 
+ \left(\frac{L}{V}\right) \left[\left(m' + m'_{y}\right) N'_{r} - \left(I'_{zz} + J'_{zz}\right) Y'_{v}\right] \dot{r} 
+ \left(C'_{0} + nr^{2}\right) r 
= \left(\frac{L}{V}\right) \left(N'_{v} Y'_{\delta} - N'_{\delta} Y'_{v}\right) \delta + \left(m' + m'_{y}\right) N'_{\delta} \dot{\delta}.$$
(29)

The left and right sides of formula (29) are divided by  $C'_0$  at the same time.

$$\left(\frac{L}{V}\right)^{2} \frac{\left(m' + m'_{y}\right)\left(I'_{zz} + J'_{zz}\right)}{C'_{0}} \ddot{r} 
+ \left(\frac{L}{V}\right) \frac{\left(m' + m'_{y}\right)N'_{r} - \left(I'_{zz} + J'_{zz}\right)Y'_{v}}{C'_{0}} \dot{r} 
+ \frac{C'_{0} + nr^{2}}{C'_{0}} r 
= \left(\frac{L}{V}\right) \frac{N'_{v}Y'_{\delta} - N'_{\delta}Y'_{v}}{C'_{0}} \delta + \frac{\left(m' + m'_{y}\right)N'_{\delta}}{C'_{0}} \dot{\delta}.$$
(30)

Define  $\alpha = n/C'_0$ ; then formula (30) is simplified as

$$T_1 T_2 \ddot{r} + (T_1 + T_2) \dot{r} + r + \alpha r^3 = K \delta + K T_3 \dot{\delta},$$
 (31)

where  $\alpha$  is a newly introduced constant. Formula (31) is simplified as a first-order system.

$$T\ddot{r} + r + \alpha r^3 = K\delta. \tag{32}$$

It can be seen that the response model of podded propulsion USV still conforms to the classic Norrbin model structure and it can be extended to ordinary vector propulsion ship.

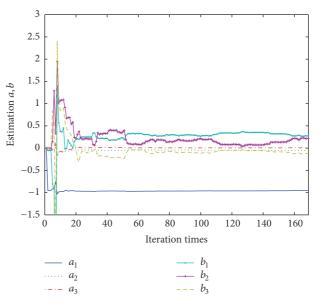


FIGURE 5: Identification curves.

## 3. Model Identification

3.1. Parameter Identification. In order to identify the response model, field experiments need to be done first. Field experiments include 5°, 8°, 12°, 15°, 17°, 18°, and 20° turning test and 15°/15° zig-zag test. It is noteworthy that field experiments need to be carried out in a relatively stable sea surface. The essence of Norrbin model is a linear Nomoto model with a nonlinear term  $\alpha$ , so K and T can be obtained by identifying the Nomoto model. Formula (33) is Nomoto model.

$$T\ddot{r} + r = K\delta. \tag{33}$$

Based on 15°/15° zig-zag test data, Nomoto model is identified by recursive least squares. The curves of identification are displayed in Figure 5.

 $a_1 \sim a_3$  and  $b_1 \sim b_3$  are the coefficients of discrete transfer function. The transfer function is

$$\frac{0.2743z^2 + 0.2122z - 0.118}{z^3 - 0.9537z^2 - 0.5476z + 0.007182}.$$
 (34)

Then formula (34) is converted to a continuous transfer function (35).

$$\frac{1.65s^3 - 10.63s^2 + 88.19s + 225.9}{s^4 + 14.26s^3 + 106.8s^2 + 322s - 0.7904}.$$
 (35)

Because the coefficients of higher order and low order differ greatly, formula (35) can be simplified as

$$\frac{225.9}{106.8s^2 + 332s}. (36)$$

Formula (33) is changed to a transfer function and then compared with formula (36). We can get that K = 0.707 and T = 0.332.  $\alpha$  is fitted by turning test data, and during the

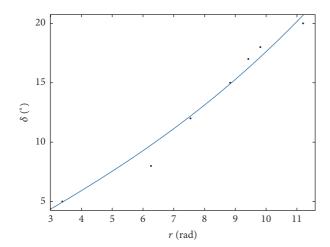


FIGURE 6: Fitting curve.

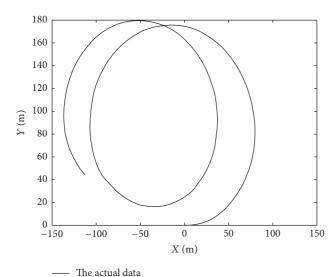


FIGURE 7: Actual turning test curve.

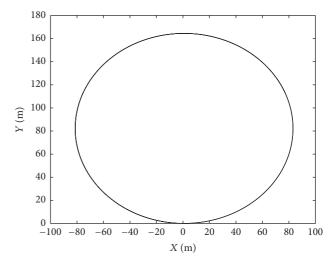
turning test,  $\dot{r}$ ,  $\ddot{r}$ , and  $\ddot{\delta}$  are equal to zero, so formula (32) can be simplified as

$$\alpha r^3 + r = K\delta. \tag{37}$$

It can be seen from formula (37) that if K and  $\alpha$  are constants,  $\delta$  and r are one by one. So according to the multigroup turning tests,  $\delta$  and the corresponding average r can be obtained; then  $\alpha$  can be fitted out. The fitting curve is shown in Figure 6 and the final result is  $\alpha = 0.001$ .

3.2. Model Validation. In order to verify the correctness of the identification results, the simulation 5° turning test is carried out and compared with the actual data. The actual turning test curve is shown in Figure 7 and the simulation turning test curve is shown in Figure 8.

As can be seen from Figures 7 and 8, the actual radius of gyration is 85.25 m and the simulation is 86.11 m. The error of numerical simulation and field experiment is small. This is to



— Simulation data

FIGURE 8: Simulation turning test curve.

say, the results of modeling and identification are correct and credible.

#### 4. Course Controller

4.1. Problem Formulation. Select state variables  $x_1 = \psi$ ,  $x_2 = \dot{\psi} = r$ , and  $u = \delta$ , so formula (32) can be simplified as

$$\dot{x}_1 = x_2,$$
 $\dot{x}_2 = f(x) + b(x)u + g(x),$ 
 $y = x_1,$ 
(38)

where  $f(x) = -(1/T)x_2 - (\alpha/T)x_2^3$ , b(x) = K/T,  $u \in R$ , and  $y \in R$  are the input and output of the system, respectively; g(x) is the uncertainty of model and environmental interference, and it can be seen as a total disturbance;  $\|g(x)\| < l_g$ ,  $l_g > 0$ .

Control Objective. In the presence of model uncertainties and external disturbances, design control input u to make the course  $\psi$  follow the target course  $x_d$ .

4.2. Course Controller Design. Sliding mode control has good adaptability and high robustness for system disturbances and parameter perturbation. It had been successfully applied to industrial control [20, 21].

Define tracking error

$$e_1 = x_1 - x_d,$$
  
 $e_2 = \dot{x}_1 - \dot{x}_d.$  (39)

For system (38), a linear sliding mode (LSM) control method and its proof of stability are proposed in [22]. The switching function of LSM is defined as

$$s = c_1 e_1 + e_2, (40)$$

where  $c_1$  is a constant and  $c_1 > 0$ .

The control law is designed as

$$u = -b(x)^{-1} \left[ f(x) + c_1 e_2 + \left( l_q + \eta_1 \right) \operatorname{sgn}(s) - \ddot{x}_d \right], \quad (41)$$

where  $\eta_1$  is constant and  $\eta_1 > 0$ .

The shortcoming of LSM is that the system state cannot converge to zero in a limited time. In order to improve the performance of the system, the control method of terminal sliding mode is proposed and the better convergence performance than LAM control is obtained. According to [23], under certain circumstances, the terminal sliding mode has a singular problem which can cause the system to be out of control. So [24] proposes nonsingular terminal sliding mode (NTSM), which can overcome the singularity problem of terminal sliding mode and it has a wide range of applications in the field of mechanical arm, motor control [25].

For system (38), the NTSM control method and its proof of stability are proposed in the paper [26]. The switching function of NTSM is defined as

$$s = e_1 + \frac{1}{\lambda_1} e_2^{p_1/q_1},\tag{42}$$

where  $\lambda_1 > 0$  is the sliding mode surface parameter to be designed,  $p_1$  and  $q_1$  are odd numbers,  $p_1 > q_1$ , and  $1 < p_1/q_1 < 2$ .

The control law is designed as

$$u = -b(x)^{-1} \left[ f(x) + \lambda_1 \frac{q_1}{p_1} e_2^{2 - (p_1/q_1)} + (l_g + \eta_2) \operatorname{sgn}(s) - \ddot{x}_d \right].$$
(43)

*Remark 1.* In order to solve the uncertainty of the model and the influence of external disturbance, (41) and (43) must have

a larger  $\eta_1$  and  $\eta_2$ . However, there is a fatal weakness that it will produce strong chattering, which is not allowed by the mechanical mechanism. So we must adopt the corresponding solutions to ensure the convergence speed and robustness based on the weakening of the chattering.

When s = 0, formula (40) can be simplified as

$$e_2 = -c_1 e_1. (44)$$

Formula (42) can be simplified as

$$e_2 = (-\lambda_1 e_1)^{q_1/p_1}. (45)$$

Contrasting formulas (44) and (45), due to 0.5  $< q_1/p_1 < 1$ , when the system is near the equilibrium point, the convergence rate of NTSM is higher than LSM; when the system is far away from the equilibrium state, the convergence rate of NTSM is lower than LSM [27]. In the light of these characteristics, FNTSM control method is proposed in [28], and it is proved that FNTSM can converge in finite time.

The switching function of FNTSM is defined as

$$s = \begin{cases} c_2 e_1 + e_2 & |e_1| > \varepsilon \\ e_1 + \frac{1}{\lambda_2} e_2^{p_2/q_2} & |e_1| \le \varepsilon, \end{cases}$$
 (46)

where  $c_2$  and  $\lambda_2$  are constants and  $c_2 > 0$  and  $\lambda_2 > 0$ ;  $p_2$  and  $q_2$  are odd numbers;  $p_2 > q_2$  and  $1 < p_2/q_2 < 2$ . It can be seen from formula (46) that the switching function of FNTSM is connected by the switching function of LSM and NTSM. According to the preceding derivation, the control law of FNTSM is shown in formula (47).

$$u = \begin{cases} -b(x)^{-1} \left[ f(x) + c_2 e_2 + \left( l_g + \eta_3 \right) \operatorname{sgn}(s) - \ddot{x}_d \right] & |e_1| > \varepsilon \\ -b(x)^{-1} \left[ f(x) + \lambda_2 \frac{q_2}{p_2} e_2^{2 - (p_2/q_2)} + \left( l_g + \eta_4 \right) \operatorname{sgn}(s) - \ddot{x}_d \right] & |e_1| \le \varepsilon. \end{cases}$$

$$(47)$$

**Theorem 2.** For system (38), under the control law of FNTSM, the system will reach the sliding surface in a limited time, and the tracking error on the sliding surface will converge to zero in a limited time.

According to Theorem 2, an important condition for the stability of Lyapunov is that g(x) meets  $||g(x)|| < l_g$ .  $l_g$  is the gain of sliding mode control switching, and it directly determines the severity of the chattering phenomenon. In order to eliminate the influence of g(x) on the control system, disturbance observer is designed to solve this problem.

Remark 3. Environmental interference is caused by the wind, flow, and two-order wave, and the frequency of these disturbances with respect to the dynamics of the ship is very low [29], so we can assume that the derivative of the uncertainty of model and environmental interference is equal to zero [30].

4.3. Disturbance Observer. The disturbance observer is designed as

$$\begin{split} \dot{\widetilde{x}}_2 &= f\left(x\right) + \widehat{g}\left(x\right) + b\left(x\right)u + k_{1i}\widetilde{x}_2, \\ \dot{\widehat{g}}_i\left(x\right) &= k_{2i}\widetilde{x}_2, \end{split} \tag{48}$$

where  $k_{1i}, k_{2i} > 0$ , i = 1, 2, are the parameters to be designed. When  $|e_1| > \varepsilon$ , i = 1; when  $|e_1| < \varepsilon$ , i = 2.  $\widehat{g}(x)$  is the estimated value of g(x) and  $\widehat{x}_2$  is the estimated value of  $x_2$ .  $\widetilde{g}(x) = g(x) - \widehat{g}(x)$  and  $\widetilde{x}_2 = x_2 - \widehat{x}_2$ ;  $\widetilde{g}(x)$  and  $\widetilde{x}_2$  are, respectively, estimated error.

Remark 4. In practical engineering, the same disturbance observer can be used to compensate the disturbance. That means  $k_{11} = k_{12}$ ,  $k_{21} = k_{22}$ , so  $\widehat{g}_1(x) = \widehat{g}_2(x) = \widehat{g}(x)$ .

So control law (47) becomes

$$u = \begin{cases} -b(x)^{-1} \left[ f(x) + c_2 e_2 + \widehat{g}(x) + \eta_3 \operatorname{sgn}(s) - \ddot{x}_d \right] & |e_1| > \varepsilon \\ -b(x)^{-1} \left[ f(x) + \lambda_2 \frac{q_2}{p_2} e_2^{2 - p_2/q_2} + \widehat{g}(x) + \eta_4 \operatorname{sgn}(s) - \ddot{x}_d \right] & |e_1| \le \varepsilon. \end{cases}$$

$$(49)$$

When the control law is (49), the stability of the system is proved.

When  $|e_1| > \varepsilon$ , formula (46) is derivative.

$$\dot{s} = c_2 e_2 + f(x) + b(x)u + g(x) - \ddot{x}_d. \tag{50}$$

Define the first Lyapunov function.

$$V_1 = \frac{1}{2}s^2 + \frac{1}{2k_{21}}\tilde{g}(x)^2 + \frac{1}{2}\tilde{x}_2^2.$$
 (51)

Then the time derivation of  $V_1$  along the solution of (48), (49), and (50) is

$$\dot{V}_{1} = s\dot{s} + \frac{1}{k_{21}}\tilde{g}(x)\dot{\tilde{g}}(x) + \tilde{x}_{2}\dot{\tilde{x}}_{2}$$

$$= s\left[g(x) - \hat{g}(x) - \eta_{3}\operatorname{sgn}(s)\right]$$

$$+ \frac{1}{k_{21}}\tilde{g}(x)\left(\dot{g}(x) - \dot{\tilde{g}}(x)\right) + \tilde{x}_{2}\left(\dot{x}_{2} - \dot{\tilde{x}}_{2}\right)$$

$$= s\left[\tilde{g}(x) - \eta_{3}\operatorname{sgn}(s)\right] - \tilde{g}(x)\tilde{x}_{2}$$

$$+ \tilde{x}_{2}\left(\tilde{g}(x) - k_{11}\tilde{x}_{2}\right)$$

$$= s\left[\tilde{g}(x) - \eta_{3}\operatorname{sgn}(s)\right] - k_{11}\tilde{x}_{2}^{2}$$

$$\leq -\left(\eta_{3} - \tilde{g}(x)\right)|s| - k_{11}\tilde{x}_{2}^{2}.$$
(52)

Define  $\eta_3 = |\widetilde{g}(x)| + \xi_1 \ (\xi_1 > 0)$ . Then  $\dot{V}_1 \le -\xi_1 |s| - k_{11} \widetilde{x}_2^2 \le 0$ . So it is asymptotically stable in the sense of Lyapunov. When  $|e_1| \le \varepsilon$ , formula (46) is derivative.

$$\dot{s} = e_2 + \frac{1}{\lambda_2} \frac{p_2}{q_2} e_2^{p_2/q_2 - 1} \left( f(x) + b(x) u + g(x) - \ddot{x}_d \right). \tag{53}$$

Define  $\phi(e_2) = (1/\lambda_2)(p_2/q_2)e_2^{p_2/q_2-1}$ .

$$\dot{s} = e_2 + \phi_1(e_2) \left( f(x) + b(x) u + g(x) - \ddot{x}_d \right). \tag{54}$$

Define the second Lyapunov functions.

$$V_2 = \frac{1}{2}s^2 + \frac{1}{2k_{22}}\tilde{g}(x)^2 + \frac{1}{2}\tilde{x}_2^2.$$
 (55)

Then the time derivation of  $V_2$  along the solution of (48), (49), and (54) is

$$\dot{V}_{2} = s\dot{s} + \frac{1}{k_{22}}\tilde{g}(x)\dot{\tilde{g}}(x) + \tilde{x}_{2}\dot{\tilde{x}}_{2}$$

$$= s\phi(e_{2})\left[g(x) - \hat{g}(x) - \eta_{4}\operatorname{sgn}(s)\right]$$

$$+ \frac{1}{k_{22}}\tilde{g}(x)\left(\dot{g}(x) - \dot{\tilde{g}}(x)\right) + \tilde{x}_{2}\left(\dot{x}_{2} - \dot{\tilde{x}}_{2}\right)$$

$$= s\phi(e_{2})\left[\tilde{g}(x) - \eta_{4}\operatorname{sgn}(s)\right] - \tilde{g}(x)\tilde{x}_{2}$$

$$+ \tilde{x}_{2}\left(\tilde{g}(x) - k_{11}\tilde{x}_{2}\right)$$

$$= s\phi(e_{2})\left[\tilde{g}(x) - \eta_{4}\operatorname{sgn}(s)\right] - k_{11}\tilde{x}_{2}^{2}$$

$$\leq -\phi(e_{2})\left(\eta_{4} - \tilde{g}(x)\right)|s| - k_{11}\tilde{x}_{2}^{2}.$$
(56)

Define  $\eta_4 = |\widetilde{g}(x)| + \xi_2$   $(\xi_2 > 0)$  and  $\phi(e_2) > 0$ . Then  $\dot{V}_2 \le -\phi(e_2)\xi_2|s| - k_{12}\widetilde{x}_2^2 \le 0$ , so it is asymptotically stable in the sense of Lyapunov.

4.4. Weakened Chattering. Although the disturbance observer can reduce the control gain, because of the discontinuity of the switching function, the chattering is generated in the vicinity of the sliding mode.

Because the rapidity is mainly reflected in the LSM phase and the robustness is mainly embodied in the sliding mode of NTSM, when  $|e_1| > \varepsilon$ , symbolic function  $\eta_3 \text{sgn}(s)$  is still being used; when  $|e_1| \le \varepsilon$ , RBF neural network [31] is used to approximate sign function  $\eta_4 \text{sgn}(s)$ . Define  $p_u = \eta_4 \text{sgn}(s)$ .

The algorithm principle of RBF network algorithm is shown as follows:

$$h_{j} = \exp\left(-\frac{\|z - w_{j}\|^{2}}{2o_{j}^{2}}\right), \quad j = 1, 2, ..., m,$$

$$p_{u} = W^{T}h(x) + \omega,$$
(57)

where z is the input signal of the network; j is the number of nodes in the hidden layer of the network;  $h = [h_1, h_2, \ldots, h_m]^T$  is the output of the Gauss basis function;  $o_j$  is the base width parameter for the node j,  $o_j > 0$ ; W is the ideal weights of neural network;  $\omega$  is the approximation error of neural network,  $|\omega| \le \omega_{\max}$ , and  $\omega_{\max} > 0$ . In this paper, define z = s, j = 20, and  $o_j = 5$ .

The output of RBF neural network is

$$\widehat{p}_{u} = \widehat{W}^{T} h(x), \qquad (58)$$

where  $\widehat{p}_u$  is the estimated value of  $p_u$  and  $\widehat{W}$  is the estimated value of W. Then  $\widetilde{p}_u = p_u - \widehat{p}_u$  and  $\widetilde{W} = W - \widehat{W}$ . Define

 $\dot{\widehat{W}} = \gamma s \phi(e_2) h(x)$ , where  $\gamma$  is a design parameter and it is greater than zero. So the control law is

$$u = \begin{cases} -b(x)^{-1} \left[ f(x) + c_2 e_2 + \hat{g}(x) + \eta_3 \operatorname{sgn}(s) - \ddot{x}_d \right] & |e_1| > \varepsilon \\ -b(x)^{-1} \left[ f(x) + \lambda_2 \frac{q_2}{p_2} e_2^{2 - p_2/q_2} + \hat{g}(x) + \hat{p}_u - \ddot{x}_d \right] & |e_1| \le \varepsilon. \end{cases}$$
(59)

When the control law is (59), the stability of the system is proved.

When  $|e_1| \le \varepsilon$ 

$$\dot{s} = e_{2} + \phi(e_{2}) \left( f(x) + b(x) u + g(x) - \ddot{x}_{d} \right) = e_{2} 
+ \phi(e_{2}) \left( -\lambda_{2} \frac{q_{2}}{p_{2}} e_{2}^{2-p_{2}/q_{2}} - \widehat{p}_{u} + g(x) - \widehat{g}(x) \right) 
= e_{2} - e_{2} + \phi(e_{2}) \left( -\widehat{p}_{u} + g(x) - \widehat{g}(x) \right) = \phi(e_{2}) 
\cdot \left( -\widehat{p}_{u} + g(x) - \widehat{g}(x) \right) = \phi(e_{2}) 
\cdot \left( -\widehat{p}_{u} + p_{u} - p_{u} + g(x) - \widehat{g}(x) \right) = \phi(e_{2}) 
\cdot \left( -\widehat{W}^{T} h(x) + W^{T} h(x) - p_{u} + g(x) - \widehat{g}(x) + \omega \right) 
= \phi(e_{2}) \left( \widehat{W}^{T} h(x) - p_{u} + g(x) - \widehat{g}(x) + \omega \right).$$
(60)

Define the third Lyapunov function.

$$V_{3} = \frac{1}{2}s^{2} + \frac{1}{2\gamma}\widetilde{W}^{T}\widetilde{W} + \frac{1}{2k_{22}}\widetilde{g}(x)^{2} + \frac{1}{2}\widetilde{x}_{2}^{2}.$$
 (61)

Then the time derivation of  $V_3$  along the solution of (48), (59), and (60) is

$$\begin{split} \dot{V}_{3} &= s\dot{s} - \frac{1}{\gamma}\widetilde{W}^{T}\dot{\widehat{W}} + \frac{1}{k_{22}}\widetilde{g}\left(x\right)\dot{\widetilde{g}}\left(x\right) + \widetilde{x}_{2}\dot{\widetilde{x}}_{2} \\ &= s\phi\left(e_{2}\right)\left(\widetilde{W}^{T}h\left(x\right) - p_{u} + g\left(x\right) - \widehat{g}\left(x\right) + \omega\right) \\ &- \frac{1}{\gamma}\widetilde{W}^{T}\dot{\widehat{W}} + \frac{1}{k_{22}}\widetilde{g}\left(x\right)\dot{\widetilde{g}}\left(x\right) + \widetilde{x}_{2}\dot{\widetilde{x}}_{2} \\ &= s\phi\left(e_{2}\right)\widetilde{W}^{T}h\left(x\right) - \frac{1}{\gamma}\widetilde{W}^{T}\dot{\widehat{W}} \\ &+ s\phi\left(e_{2}\right)\left(-p_{u} + g\left(x\right) - \widehat{g}\left(x\right) + \omega\right) \\ &+ \frac{1}{k_{22}}\widetilde{g}\left(x\right)\dot{\widetilde{g}}\left(x\right) + \widetilde{x}_{2}\dot{\widetilde{x}}_{2} \\ &\leq s\phi\left(e_{2}\right)\left(\omega - \eta_{4}\mathrm{sgn}\left(s\right) + \widetilde{g}\left(x\right)\right) \\ &+ \frac{1}{k_{22}}\widetilde{g}\left(x\right)\left(\dot{g}\left(x\right) - \dot{\overline{g}}\left(x\right)\right) + \widetilde{x}_{2}\left(\dot{x}_{2} - \dot{\overline{x}}_{2}\right) \\ &\leq \phi\left(e_{2}\right)\left(s\omega + s\widetilde{g}\left(x\right) - \eta_{4}\left|s\right|\right) - \widetilde{g}\left(x\right)\widetilde{x}_{2} \end{split}$$

$$+ \widetilde{x}_{2} \left( g\left( x \right) - \widehat{g}\left( x \right) - k_{12} \widetilde{x}_{2} \right)$$

$$\leq -\phi \left( e_{2} \right) \left| s \right| \left( \eta_{4} - \widetilde{g}\left( x \right) - \omega \right) - k_{12} \widetilde{x}_{2}^{2}.$$
(62)

Considering  $\phi(e_2) > 0$  and  $k_{12} > 0$  and defining  $\eta_4 = |\tilde{g}(x)| + |\omega_{\max}| + \sigma \ (\sigma > 0)$ , then

$$\dot{V}_3 \le -\phi(e_2)\,\sigma|s| - k_{12}\tilde{x}_2^2 \le 0.$$
 (63)

So it is asymptotically stable in the sense of Lyapunov. Define  $\Omega(t)=\phi(e_2)\sigma|s|+k_{12}\tilde{x}_2^2=-\dot{V}_4$  Seeking the time integral of  $\Omega(t)$ 

$$\int_{0}^{t} \Omega\left(\tau\right) d\tau = V_{4}\left(s\left(0\right), \widetilde{x}_{2}\left(0\right)\right) - V_{4}\left(s\left(t\right), \widetilde{x}_{2}\left(t\right)\right). \tag{64}$$

Because  $\dot{V}_4 \leq 0$ , it can be concluded that  $V_4(t)$  is a nonincreasing function; then  $V_4(s(0),\widetilde{x}_2(0))$  is bounded. So  $V_4(s(t),\widetilde{x}_2(t)) \geq 0$  and  $0 \leq V_4(s(t),\widetilde{x}_2(t)) \leq \infty$ . We can draw that

$$\int_{0}^{t} \Omega\left(\tau\right) d\tau < \infty. \tag{65}$$

Obviously,  $\dot{\Omega}(\tau)$  is bounded. By Barbalat lemma

$$\lim_{t \to 0} \Omega(t) = 0. \tag{66}$$

This shows that when  $t \to \infty$ ,  $s(t) \to 0$  and  $\tilde{x}_2(t) \to 0$ , and then  $e_1(t) \to 0$ ,  $e_2(t) \to 0$ . In total, the system can be globally asymptotically stable.

- 4.5. Fuzzy Weight. When the system error is  $\varepsilon$ , the control mode will change. If the option of  $\varepsilon$  is not appropriate, the controller will produce chattering. In this paper, fuzzy weighting method is used to design the controller gain vector in the mode switching process, which can improve the transient response of the system. The basic idea is that, according to the system state error, the control gain v of LSM controller is obtained by using the fuzzy inference method; then the control gain of NTSM is 1 v.
- (1) The fuzzy controller is a single input and single output system. The input is the system error  $e_1$ , and the output is the control gain  $\nu$  of the LSM controller. Firstly, the systematic error  $e_1$  is normalized as  $\begin{bmatrix} -6 & 6 \end{bmatrix}$ , and the domain of  $\nu$  is  $\begin{bmatrix} 0 & 1 \end{bmatrix}$ . Their membership functions are shown in Figures 9 and 10, respectively.

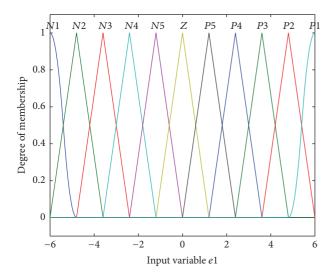


Figure 9: The membership function of  $e_1$ .

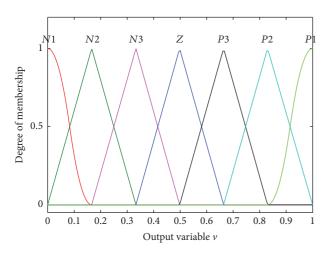
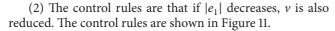


FIGURE 10: The membership function of v.



(3) Zadeh and Max-min method are used in fuzzy reasoning, and the centroid area of gravity method is used in the inverse model.

At this point, the final control law is

$$u = \left[ -b(x)^{-1} \left[ f(x) + c_{2}e_{2} + \widehat{g}(x) + \eta_{3}\operatorname{sgn}(s) - \ddot{x}_{d} \right] \right]$$

$$\cdot v + \left[ -b(x)^{-1} \right]$$

$$\cdot \left[ f(x) + \lambda_{2} \frac{q_{2}}{p_{2}} e_{2}^{2 - p_{2}/q_{2}} + \widehat{g}(x) + \widehat{p}(z) - \ddot{x}_{d} \right] \right] (1$$

$$- v).$$
(67)

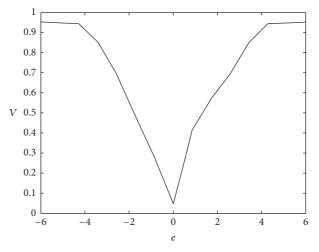


FIGURE 11: Control rules.

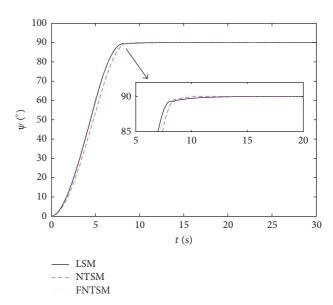


FIGURE 12: Course keeping curves under weak disturbance.

## 5. Numerical Simulations

In order to verify the performance of the FNTSM controller, numerical simulations of course keeping and course tracking for Lanxin USV are carried out. The control parameters of (67) are selected as  $c_2=0.55$ ,  $\eta_3=10.9$ ,  $\lambda_2=0.45$ ,  $p_2=11$ ,  $q_2=9$ ,  $k_{11}=2$ ,  $k_{21}=15$ ,  $k_{12}=2$ ,  $k_{22}=15$ , and  $\gamma=0.1$ .

To show the rapidity and robustness of the proposed algorithm, control laws (41) and (43) are compared with FNTSM. The parameters of (41) are  $c_1 = 0.55$ ,  $\eta_1 = 8.44$  and the parameters of (43) are  $\lambda_1 = 1.2$ ,  $p_1 = 11$ ,  $q_1 = 9$ ,  $\eta_2 = 7.58$ .

*Case 1* (numerical simulation of course keeping under weak disturbance). The purpose of this numerical simulation is to verify the rapidity of FNTSM controller.

When  $g(x) = 2.5 \sin(0.6t)$ , course keeping simulation results are shown in Figures 12–14.

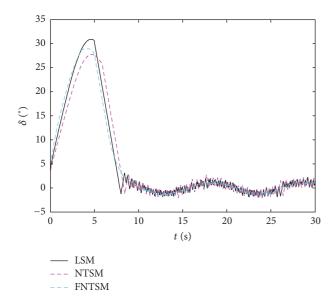


FIGURE 13: Propulsion angle curves under weak disturbance.

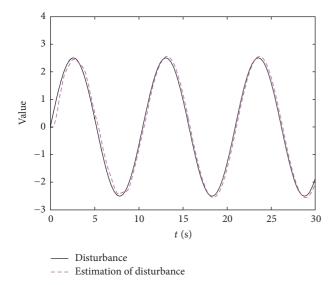


FIGURE 14: Estimation of disturbance under weak disturbance.

Figure 12 shows comparison results of course keeping performance among the control laws of LSM, NTSM, and FNTSM. It is observed that the proposed FNTSM control law has the fastest convergence rate than LSM and NTSM.

Figure 13 shows the control inputs of FNTSM, LSM, and NTSM. It is observed that in order to offset disturbance, the propulsion angle of all the control laws has certain chattering, but it can be clearly seen that the chattering of FNTSM is the smallest.

Figure 14 shows the estimated value of disturbance. It is observed that the error of the two curves is very small, and this can be proved that disturbance observer works well.

*Case 2* (numerical simulation of course keeping under strong disturbance). The purpose of this numerical simulation is to verify the robustness of FNTSM controller.

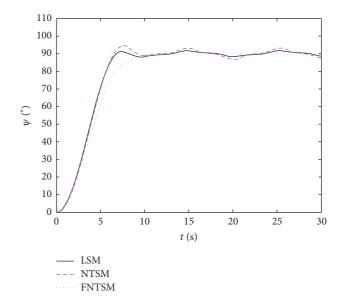


FIGURE 15: Course keeping curves under strong disturbance.

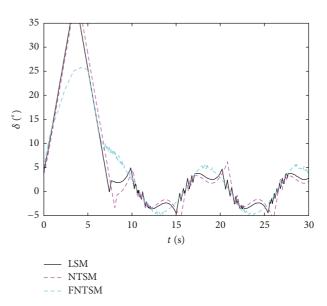
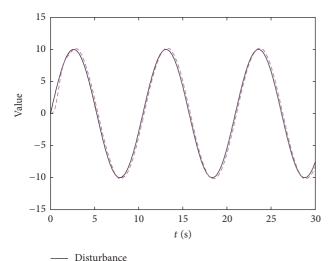


FIGURE 16: Propulsion angle curves under strong disturbance.

In the case of all control parameters being not changed and  $g(x) = 10 \sin(0.6t)$ , course keeping simulation results are shown in Figures 15–17.

It can be seen from Figures 15–17 that, under the strong disturbance, FNTSM still has a good control effect, and it can keep course stability at target value, but LSM and NTSM have shown overshoot and oscillation. The propulsion angle is still in a reasonable range and the disturbance observer still has a good performance.

Case 3 (numerical simulation of course tracking under strong disturbance). In the case of all control parameters being not changed and  $g(x) = 10 \sin(0.6t)$ , course tracking simulation results are shown in Figures 18–20.



--- Estimation of disturbance

FIGURE 17: Estimation of disturbance.

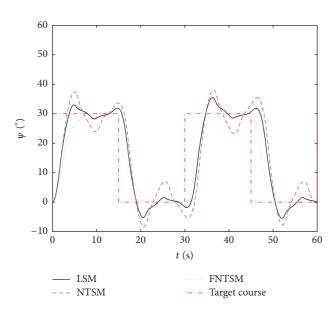


FIGURE 18: Course tracking curves under strong disturbance.

It can be seen from Figures 18–20 that, under the strong disturbance, FNTSM can make the course closely track the target value and not overshoot, while NTSM and LSM are obviously out of control. Meanwhile, the propulsion angle fluctuation range of FNTSM is the smallest and the disturbance observer can estimate the disturbance value very well.

#### 6. Conclusions

According to MMG separation model theory, the response model of podded propulsion USV is proved to be still in compliance with the Norrbin model, and then the parameters of response model are obtained by the method of system identification. Through the model validation we found that the error between simulation result and field experiment is

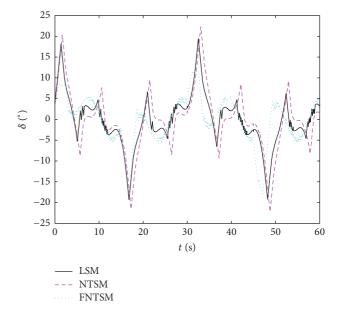
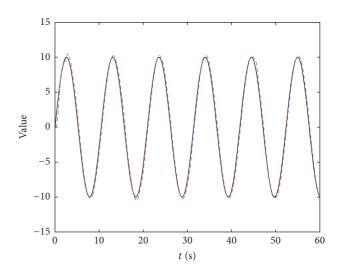


FIGURE 19: Propulsion angle curves under strong disturbance.



— Disturbance

--- Estimation of disturbance

FIGURE 20: Estimation of disturbance.

very small, which proves the correctness of modeling and identification. In view of the characteristic of USV with high speed, based on the multimodel control theory, FNTSM controller is proposed. Meanwhile, disturbance observer, RBF neural network, and fuzzy weighting are used to weaken the chattering of the system. Finally, the rapidity and robustness of FNTSM controller are demonstrated through numerical simulations. Obviously, the results of this paper can be extended for more general ship. In the next step, the FNTSM control law will be applied to the real ship experiment for verification.

## **Conflicts of Interest**

The authors declare that they have no conflicts of interest.

## Acknowledgments

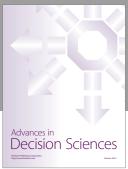
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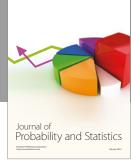
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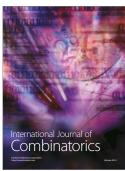








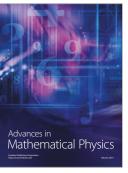






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