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Evanescent Wave Spectroscopy using Hollow Cylindrical Waveguide Probes

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Declaration

I hearby certify that this material, which I now submit for assessment on the programme of study leading to the award of Master of Science is entirely my own work and has not been taken from the work of others save to and to the extent that such work has been cited and acknowledged within the text of my work

Signed Deirche Gloma ID Number 94970840 Date <u>S-July-1996</u>

Dedication

For Mam and Dad

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I would like to thank Dr Vince Ruddy for his unending help and guidance throughout this project I would also like to show my appreciation to my fellow members in the Optical Sensors Group, both old and new, for their companionship and brain power Vincent Murphy, Tom Butler, Ger O'Keeffe, James Walsh, Fergus Connolly

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Abstract

Optical waveguides carry bound modes which consist of a core E and H field, which is oscillatory across the waveguide and evanescent in the waveguide cladding Both the core and cladding component of each mode has the same frequency and propagation constant. When the frequency of the light carried by the waveguide matches an absorption transition of the material of the cladding, the mode loses optical power as it propagates due to the attenuation of the evanescent cladding portion of the mode. This process is called attenuated total reflection spectroscopy (ATR) or evanescent wave spectrophotometry. As in simple transmission spectrophotometry the absorbance of the mode is related to the interaction length of the waveguide with the absorbing cladding, the evanescent waves of the various modes.

This work firstly represents a theoretical analysis of the bound modes that can exist in a step index hollow cylindrical waveguide, their evanescent power fraction and the effective length of such a waveguide when located in an absorbing cladding material. The waveguide is found to have a normalized frequency or effective V number whose magnitude determines the total number of bound modes and influences the mean evanescent power fraction between modes. This effective V number reduces to that of the solid step index fiber waveguide in the limit of a zero radius inner cavity Likewise the expression for the mean evanescent mode power fraction becomes - in the limit of zero inner radius - identical to that of the fiber waveguide. The evanescent absorbance of such a hollow waveguide located in an absorbing fluid is modeled in terms of the bulk absorption coefficient of the fluid and the waveguide dimensions.

In the second part of the thesis a set of experimental absorbance values for ATR spectrophotometry using a hollow silica waveguide probe are reported Good correspondence is found between the theoretical model and the experimental data

Schematic of Hollow Waveguide Evanescent Probe

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1. Introduction to evanescent wave spectrophotometry.

1.1 Modes in Waveguides.

When light travels within a waveguide, the path that the light follows is defined by the shape and dimensions of the waveguide According to wave theory, the electromagnetic E and H fields must also be solutions of the wave equation

$$\nabla^2 E - \frac{1}{v_p^2} \frac{\partial^2 E}{\partial t^2} = 0 \qquad v_p = \frac{c}{n} \qquad Eqn \ 111$$

These allowed light paths are considered bound within the waveguide, and are called modes An optical waveguide can only support a finite number of bound modes Each mode consists of two distinct parts, that which exists within the core of the waveguide (which is oscillatory in behaviour) and that part which travels along the interface of the core and cladding. The second part is called the evanescent wave, with an electric field amplitude that falls off exponentially with distance from the interface A mode is identified by its' core and cladding mode parameters, U and W U and W are related by the propagation constant β , which represents the component of the wavevector along the waveguide axis. If the wave has a wavenumber k, in free space, then in the core (medium with refractive index n₁, where n₁ > n₂) its' value is n₁k, so

$$\beta = n_1 k Cos \theta_z \qquad Eqn \ 1 \ 1 \ 2$$

where θ_z is the angle the wavevector makes with the waveguide axis (z axis)



Figure 1-1 The propagation constant β of a particular mode

The transverse component is therefore

$$\sqrt{n_1^2k^2-\beta^2}$$

For a core dimension of 2a, the core and cladding mode parameters are given by

$$U = a\sqrt{n_1^2 k^2 - \beta^2}$$

$$W = a\sqrt{\beta^2 - n_2^2 k^2}$$

Eqn 1 1 3

U and W combine to give

$$U^2 + W^2 = a^2 k^2 \left(n_1^2 - n_2^2 \right)$$

This is usually denoted

$$U^{2} + W^{2} = V^{2}$$

$$V = ak\sqrt{n_{1}^{2} - n_{2}^{2}}$$

Eqn 1 1 4

V is called the normalized frequency of a waveguide, and relates the core diameter (2a) to the numerical aperture, N A = $(n_1^2 - n_2^2)^{1/2}$, and the wavenumber k of the light V is a property only of the waveguide and of wavelength, λ Because of the above equation,

$$0 \le U \le V$$

$$Eqn \ 1 \ 15$$

and when one is large, the other is small

1.2 The Planar Waveguide.

Figure 1-2 shows a planar waveguide



Figure 1 2 A Planar Waveguide

The solution to this waveguide is derived from the wave equation (equation 1 1 1) and is

$$\frac{\partial^2 E}{\partial x^2} + \frac{\partial^2 E}{\partial y^2} + \frac{\partial^2 E}{\partial z^2} - \frac{n^2}{c^2} \frac{\partial^2 E}{\partial t^2} = 0 \qquad \qquad Eqn \ 1 \ 2 \ 1$$

in cartesian coordinates Assuming that the E and H field components of the electromagnetic wave have a time (t) and distance (z) dependence of the form

$$\exp i(\omega t - \beta z) \qquad \qquad Eqn \ 1 \ 2 \ 2$$

then a trial solution of

$$E = E_x \exp i(\omega t - \beta z) \qquad Eqn \ 1 \ 2 \ 3$$

will yield the wave equation m the core and cladding as

$$a^{2} \frac{\partial^{2} E_{x}}{\partial x^{2}} + U^{2} E_{x} = 0 \quad @ \quad x \le a$$

$$a^{2} \frac{\partial^{2} E_{x}}{\partial x^{2}} - W^{2} E_{x} = 0 \quad @ \quad x \ge a$$
Eqn 124

U and W, the mode core and cladding parameters are given in section 1.1 The wavenumber k is given by $2\pi/\lambda$, λ being the free space wavelength. The solutions to the above equations are given by

$$E = A_{1}Sin\left(\frac{Ux}{a}\right) \quad or$$

$$E = A_{2}Cos\left(\frac{Ux}{a}\right) \quad for \quad x \le a$$

$$E = A_{3}\exp\left(-\frac{Wx}{a}\right) \quad for \quad x \ge a$$

where A_1 , A_2 and A_3 are amplitude coefficients Within the planar waveguide core the waves are sinusoidal (or cosinusoidal), and are evanescent in the cladding

1.3 The Cylindrical Waveguide.



Figure 1 3 Cylindrical Waveguide

The cylindrical waveguide is shown in figure 1-3 The solution of the wave equation in a cylindrical waveguide is derived as follows

$$\nabla^2 E - \frac{1}{v_p^2} \frac{\partial^2 E}{\partial t^2} = 0 \qquad \qquad Eqn \ 13 \ 1$$

or

$$\nabla^2 E - \frac{n^2 k^2}{c^2} \frac{\partial^2 E}{\partial t^2} = 0 \qquad \qquad Eqn \ 1 \ 3 \ 2$$

where $k = 2\pi/\lambda$, and λ is the free space wavelength

Expressing equation 1.3.1 in cylindrical polar coordinates (r, ϕ, z) and inserting a trial solution of

$$E = E(r,\phi) \exp i(\omega t - \beta z)$$

= $E_i \exp i(\omega t - \beta z)$
Eqn 1 3 3

gives

$$\frac{\partial^2 E_i}{\partial r^2} + \frac{1}{r} \frac{\partial E_i}{\partial r} + \frac{1}{r^2} \frac{\partial^2 E_i}{\partial \phi^2} + (n^2 k^2 - \beta^2) E_i = 0 \qquad \text{Eqn 1 3 4}$$

Changing to a normalised radius $\mathbf{R} = \mathbf{r}/\mathbf{a}$ gives

$$\frac{\partial^2 E_i}{\partial R^2} + \frac{1}{R} \frac{\partial E_i}{\partial R} + \frac{1}{R^2} \frac{\partial^2 E_i}{\partial \phi^2} + a^2 (n^2 k^2 - \beta^2) E_i = 0 \qquad \text{Eqn 135}$$

Since the medium has cylindrical symmetry we can write

$$E_t = F(R)\Phi(\phi) \qquad \qquad E_{qn} 1 3 6$$

1 e separating E_t in radial (R) and azimuthal (ϕ) components Equation 1 3 5 then gives

$$\frac{R^{2}}{F}\left\{\frac{d^{2}F}{dR^{2}} + \frac{1}{R}\frac{dF}{dR}\right\} + R^{2}\left(n^{2}k^{2} - \beta^{2}\right) = -\frac{1}{\Phi}\frac{d^{2}\Phi}{d\phi^{2}} \qquad Eqn \ 137$$

If we write equation 1 3 7 as some positive quantity $+ l^2$ then

$$-\frac{1}{\Phi}\frac{d^2\Phi}{d\Phi^2} = l^2 \qquad \qquad Eqn \ 1 \ 3 \ 8$$

which has (S H M) solutions of the form Cos lo or Sin lo For the function to be single valued, i e

$$\Phi(\phi) = \Phi(\phi + 2\pi) \qquad \qquad Eqn \ 1 \ 3 \ 9$$

we must have l = 0, 1, 2, 3 Since for each value of l there may be two independent states of polarisation, modes with $l \ge 1$ are four fold degenerate while l = 0 modes, being ϕ independent are two fold degenerate. In equation 1.3.7 above the second term \mathbf{R}^2 ($\mathbf{n}^2 \mathbf{k}^2 \cdot \beta^2$) may be positive or negative depending on the relative magnitude of nk and β

a) When $n_1^2 k^2 > \beta^2 > n_2^2 k^2$

For β in this range the radial fields F(R) are oscillatory in the core and decay in the cladding (evanescent) These are known as guided modes or bound modes Recalling that β represents the z component of the wavevector n₁k in the core, Snell's Law for rays says that guided or internally reflected rays occur if

$$\theta > S_{1}n^{-1}\frac{n_{2}}{n_{1}}$$

$$n_{1} S_{1}n\theta > n_{2}$$

$$n_{1}k S_{1}n\theta > n_{2}k$$
Eqn 1 3 10

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Where θ is the angle the ray makes with the normal at the interface Now $\theta + \theta_z = \pi/2$ therefore Snell's Law gives

$$n_1 k \cos \theta_z > n_2 k$$

$$i e \quad \beta > n_2 k$$
Eqn 1311

for total internal reflection Since $\beta = n_1 k \cos \theta_z$, β_{max} is $n_1 k$ so we have

$$n_1 k > \beta > n_2 k \qquad \qquad Eqn \ 1 \ 3 \ 12$$

for total internal reflection (θ_z is shown in Figure 1-1)

b) $\beta^2 < n_2^2 k^2$

For such values of β the radial fields F(R) are oscillatory in the cladding. These are known as radiation modes and correspond to refracted light in the cladding.

Returning to equation 1 3 7 the wave equation for guided or bound modes becomes

$$R^{2} \frac{d^{2} F}{dR^{2}} + R \frac{dF}{dR} + (U^{2} R^{2} - l^{2})F = 0 \qquad R < 1$$

$$R^{2} \frac{d^{2} F}{dR^{2}} + R \frac{dF}{dR} - (W^{2} R^{2} + l^{2})F = 0 \qquad R > 1$$
Eqn 1.3.13

where

$$U^{2} = a^{2} (n_{1}^{2} k^{2} - \beta^{2})$$

$$W^{2} = a^{2} (\beta^{2} - n_{1}^{2} k^{2})$$

Eqn 1 3 14

Equations 1 3 13 are of the standard form of Bessel Equations with solution $J_t(UR)$, $Y_t(UR)$ in the R < 1 core region and modified Bessel functions $K_t(WR)$ and $I_t(UR)$ in the cladding

The equations above give the allowed solutions

$$E_{t} = A_{1} J_{i}(UR) Cos(l\phi)$$

or $E_{t} = A_{1} J_{i}(UR) Sin(l\phi)$
$$for R \leq 1$$

$$E_{t} = A_{1} K_{i}(WR) Cos(l\phi)$$

or $E_{t} = A_{1} K_{i}(WR) Sin(l\phi)$
$$for R \geq 1$$

where A_1 is the wave amplitude The $Y_t(UR)$ and $I_t(UR)$ Bessel functions are not allowed solutions in this case as $Y_t(UR)$ is infinite at R = 0, and $I_t(UR)$ is infinite at R = 1 (Abramowitz and Stegun, Figures 9 1 and 9 8)

1.4 The Evanescent Power Fraction.

For both planar and cylindrical waveguides the evanescent field

- (1) is approximately exponentially decaying away from the interface
- (11) Has a penetration depth \cong a/W, or

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$$d_p \approx \frac{1}{\sqrt{\beta^2 - n_2^2 k^2}}$$

By taking the Poynting vector ($\vec{E} \times \vec{H}$) and integrating from R = 1 to $R = \infty$, the power of the evanescent wave may be calculated Thus the power fraction of a mode which exists as an evanescent wave can be determined This fraction f, was shown by Gloge (1971) to be

$$f \approx \frac{U^2}{V^2 \sqrt{W^2 + l^2 + 1}}$$
 Eqn 1.4.1

For modes close to cut-off, i.e. modes for which $W \to 0$, this fraction will be large, $(W \to 0 \text{ as } U \to V)$

$$f \cong \frac{1}{l}$$

while for modes far from cutoff (W \rightarrow V, as U \rightarrow 0) the power fraction will be negligible

1.5 The Waveguide as a Sensor.

In order for a waveguide to be used as a sensor, an analytical wavelength of the cladding material must match that of the light being carried by the core. The cladding will then absorb photons from the modes at a rate determined by the bulk absorption coefficient of the cladding material (which may be solid or fluid). The amount of absorption that occurs also depends on the distance over which the core is in contact with the absorbing medium. The sensing mechanism is the detection of the size of this power loss into the absorber, and relating the power loss to the concentration of absorber present. Various evanescent wave sensor geometries are possible. Harrick (1987) Chapter 4 discusses rectangular waveguide designs. Kapany et al. (1963) I and II, Hansen (1963) and Harrick (1964) describe solid rod waveguides used in attenuated wave spectrophotometry.

1.6 Absorbance of a Sensor Probe.



Figure 1-4 Power distribution in a waveguide

Because only light being transmitted as an evanescent wave comes in contact with the absorber, only a small fraction of the total light launched into the waveguide is used to provide sensor information. Therefore a modified version of the Lambert-Beer law applies, being

$$I = I_0 \exp(-f\alpha z) \qquad \qquad Eqn \ 1 \ 6 \ 1$$

This can be seen as follows The evanescent power at point A in Figure 1 4 is

$$I(z) = fI Eqn \ 1 \ 6 \ 2$$

The loss in power of the mode in travelling Δz is ΔI , where

$$\frac{\Delta I}{I(z)} = -\alpha(\Delta z) \qquad \qquad Eqn \ 1 \ 6 \ 3$$

Using equation 1 6 2 gives

$$\frac{\Delta I}{I} = -f \alpha \Delta z \qquad \qquad Eqn \ 1 \ 6 \ 4$$

On integrating from z = 0 to z, for $I = I_0$ to I we get

$$\log \frac{I}{I_0} = -f\alpha z \qquad \qquad Eqn \ 165$$

$$I = I_0 \exp(-\alpha fz) \qquad \qquad Eqn \ 1 \ 6 \ 6$$

$$A' = (0.434) f \alpha z$$
 Eqn 167

For a given mode, specified by the core mode parameter U, and cladding mode parameter W, then the above equation gives the evanescent absorbance A' using equation 1 4 1 for f - as

$$A' = (0.434) \frac{U^2 \alpha z}{V^2 \sqrt{W^2 + l^2 + 1}}$$
 Eqn 168

In this case z is the length of the waveguide in contact with the absorber Where many modes are excited, each with the same incident power (I_0 / N) , the transmitted power will be

$$I = \sum_{N} \left(\frac{I_0}{N} \right) \exp\left\{ \frac{-U^2 \alpha z}{V^2 \sqrt{W^2 + l^2 + 1}} \right\}$$
Eqn 169
$$I_{I_0} = \frac{1}{N} \sum_{N} \exp\left\{ \frac{-U^2 \alpha z}{\sqrt{W^2 + l^2 + 1}} \right\}$$
Eqn 1610

$$A' = -\log_{10}\left[\frac{1}{N}\sum_{N}\exp\left\{\frac{\alpha z U^{2}}{\sqrt{W^{2} + l^{2} + 1}}\right\}\right] \qquad Eqn \ 1 \ 6 \ 11$$

where the summation is carried out over N modes A' is the evanescent absorbance

1.7 Conclusions.

This chapter provides the basis from which theoretical analysis on the hollow cylindrical waveguide will be done. The methods shown above will be expanded to describe the hollow waveguide in similar mathematical terms, so that the hollow waveguide will describe both the planar and fibre waveguides when the dimensions of the hollow waveguide are sufficiently large to be considered planar or small enough to be considered a solid fibre.

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2. The hollow cylindrical waveguide - a theoretical mode analysis.





Figure 2 1 Hollow cylindrical waveguide

Figure 2-1 above shows a hollow cylindrical waveguide The inner hole (0 < r < a) and outer region $(b < r < \infty)$ are the lower index media, and the glass region a < r < b is the higher refractive index medium. In this configuration the glass annulus acts as a light guide, but with cladding at two interfaces giving two surfaces where evanescent absorption can take place

2.2 The E and H fields of modes in a hollow waveguide.

The hollow cylinder is described as having an inner diameter of 2a and outer diameter 2b, this region being made of a glass with refractive index n_1 , and surrounded internally and externally by a medium of refractive index n_2 with $n_1 > n_2$ The following analysis is based on the method used in Unger(1980) for E and H fields in doubly clad cylindrical fibre waveguides modified for the hollow cylinder

Barlow (1981, 1983) developed the first published work on the "three concentric layer cylindrical waveguide" In his analysis - pertaining to fiber waveguides - the waveguide dimensions are small (typically 100 μ m) and the refractive indices of the 3 layers (n, n₂, n₃) are very close together, obeying the so-called "weakly guiding approximation" The latter condition is valid only in very limited cases but the thrust of the mode analysis can form a basis for a more general theory Tsao et al (1989) carried out further 3 layer fiber mode characterisations, again invoking a weakly guiding condition Brunner et al (1995) published some work on Attenuated Total Reflection Spectrophotometry using "capillary optical fiber" probes using the mode analysis technique of Barlow (1983) for their work Mode analysis for a 3 layered cylindrical waveguide of a hollow cylinder shape, where the guiding glass annulus is surrounded by 2 media of the same refractive index (as shown in figure 2 1) is carried out by the author without recourse to the weakly guiding condition (n₁ \equiv n₂) This is the most general case and the analysis described here is the first representation of such a treatment

Solutions of the wave equations in a medium in which the phase velocity of light is v = c/n

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$$\vec{\nabla}^2 \vec{E} - \frac{1}{v^2} \frac{\partial \vec{E}}{\partial t^2} = 0$$

$$\vec{\nabla}^2 \vec{H} - \frac{1}{v^2} \frac{\partial \vec{H}}{\partial t^2} = 0$$
Eqn 2.2.1

(where $\overline{\nabla}^2$ is a vector operator) may be expressed in terms of the two waveguide parameters U and W, defined in equation 1.1.3 The two vector differentials 2.2.1 for E and H can be broken into six differential scalar equations. Of these, four involve more than one E of H field component. The other two involve the z components E_z or H_z alone. These can be written as

$$\left\{\frac{\partial^2}{\partial R^2} + \frac{1}{R}\frac{\partial}{\partial R} + \frac{1}{R^2}\frac{\partial^2}{\partial \phi^2} + U^2\right\}E_z = 0 \quad @1 < R < C$$

$$\left\{\frac{\partial^2}{\partial R^2} + \frac{1}{R}\frac{\partial}{\partial R} + \frac{1}{R^2}\frac{\partial^2}{\partial \phi^2} - W^2\right\}E_z = 0 \quad @\begin{cases}0 < R < 1\\C < R < \infty\end{cases} \qquad Eqn \ 2 \ 2 \ 2$$

where R is the normalised radius r/a and C is given by b/a. The ϕ dependence of the fields may be represented by Cos $l\phi$ or Sin $l\phi$ terms giving a degeneracy of four for l > 0 in general and a degeneracy of two - corresponding to only two orthogonal polarisations - for the l = 0 modes Similar equations can be written for the H_z fields

The solutions of these equations in the three zones of interest, (within the centre of the tube, the glass itself and outside the tube) ignoring those whose values become infinite at any of the boundaries ($\mathbf{R} = 0$, $\mathbf{R} = 1$, $\mathbf{R} = \infty$) are

$$I_{l}(WR)Cos(l\phi)$$

$$I_{l}(WR)Sin(l\phi)$$

$$I_{l}(WR)Sin(l\phi)$$

$$I_{l}(UR)Cos(l\phi) \text{ and } Y_{l}(UR)Cos(l\phi)$$

$$I_{l}(UR)Sin(l\phi) \text{ and } Y_{l}(UR)Sin(l\phi)$$

$$I_{l}(UR)Sin(l\phi) \text{ and } Y_{l}(UR)Sin(l\phi)$$

$$K_{l}(WR)Cos(l\phi)$$

$$K_{l}(WR)Sin(l\phi)$$

$$In C \leq R \leq \infty \qquad Eqn \ 2 \ 2 \ 5$$

where J_t and K_t are Bessel functions of the first and second kind (of order *t*) and represent oscillatory functions, I_t and Y_t are modified Bessel functions of the first and second kind and represent exponentially varying functions of R, respectively (Abramowitz and Stegun, 1964) If each field is normalised so that it is unity in the R = 1 interface, the fields of the even modes in the glass can be written

$$E_{z} = i(A+B)\frac{I_{l}(WR)}{I_{l}(W)}Cos \ l\phi \qquad R < 1$$

$$E_{z} = i\left\{A\frac{J_{l}(UR)}{J_{l}(U)} + B\frac{Y_{l}(UR)}{Y_{l}(U)}\right\}Cos \ l\phi \qquad 1 < R < C \qquad Eqn \ 2 \ 2 \ 6$$

$$E_{z} = iE\frac{K_{l}(WR)}{K_{l}(W)}Cos \ l\phi \qquad C < R$$

with the Hz fields as

$$\begin{aligned} H_z &= \iota (G+D) \frac{I_l(WR)}{I_l(W)} S_{ln} l \phi & R < 1 \\ H_z &= \iota \Biggl\{ G \frac{J_l(UR)}{J_l(U)} + D \frac{Y_l(UR)}{Y_l(U)} \Biggr\} S_{ln} l \phi & 1 < R < C & Eqn \, 2 \, 2 \, 7 \\ H_z &= \iota F \frac{K_l(WR)}{K_l(W)} S_{ln} l \phi & C < R \end{aligned}$$

In equations 225 and 226 above the constants A B, G, D, E F represent six field amplitudes The Bessel functions are normalised to have unit values at the inner interface $R = 1 E_z$ and H_z are chosen to be imaginary in order that the transverse components E_r , E_{ϕ} and H_r , H_{ϕ} are real The radial and azimuthal (transverse) fields can be derived from the axial (i e z) field components using the well known relationships (Snyder and Love, 1983)

$$E_{r} = \frac{i}{a(n^{2}k^{2} - \beta^{2})} \left\{ \beta \frac{\partial E_{z}}{\partial R} + \frac{kZ}{R} \frac{\partial H_{z}}{\partial \phi} \right\}$$

$$H_{r} = \frac{i}{a(n^{2}k^{2} - \beta^{2})} \left\{ \beta \frac{\partial H_{z}}{\partial R} - \frac{n^{2}k}{ZR} \frac{\partial E_{z}}{\partial \phi} \right\}$$

$$E_{\phi} = \frac{i}{a(n^{2}k^{2} - \beta^{2})} \left\{ \frac{\beta}{R} \frac{\partial E_{z}}{\partial \phi} - kZ \frac{\partial H_{z}}{\partial R} \right\}$$

$$H_{\phi} = \frac{i}{a(n^{2}k^{2} - \beta^{2})} \left\{ \frac{\beta}{R} \frac{\partial H_{z}}{\partial \phi} + \frac{n^{2}k}{Z} \frac{\partial E_{z}}{\partial R} \right\}$$

where $Z^2 = (\mu_0/\epsilon_0)$ is the characteristic impedance of free space squared Differentiation of equations 2.2.5 and 2.2.6 allow the E_r, H_r, E_{\$,} and H_{\$,} field components to be evaluated in all three zones of the waveguide in terms of the various Bessel functions, their first derivatives (Snyder and Love, 1983) with respect to R, the mode parameters U and W and the 6 amplitude coefficients A - F

2.3 The mode eigenvalue equation.

The six field components e_z , e_{ϕ} , e_r , h_z , h_{ϕ} , h_r) in each of the three zones (0 < R < 1 1 < R < C, R > C) were determined These are listed in Appendix A The continuity of the field components at the two interfaces R = 1 and R = C generate a set of equations, the solutions of which give the allowed values of the parameters U and W (or the allowed values of the propagation constant β)

The solutions of these six equations were obtained by placing them in a matrix, and calculating the determinant of the matrix. The equations used to form the determinant are as follows, with the prime indicating differentiation with respect to the argument

$$A\frac{J_{l}(UC)}{J_{l}(U)} + B\frac{Y_{l}(UC)}{Y_{l}(U)} - E\frac{K_{l}(WC)}{K_{l}(W)} = 0 \qquad e_{z} \text{ at } R = C \qquad Eqn \ 2 \ 3 \ 1$$

$$G\frac{J_{l}(UC)}{J_{l}(U)} + D\frac{Y_{l}(UC)}{Y_{l}(U)} - F\frac{K_{l}(WC)}{K_{l}(W)} = 0 \qquad h_{z} \text{ at } R = C \qquad Eqn \ 2 \ 3 \ 2$$

$$-A\beta l \left(\frac{1}{U^{2}} + \frac{1}{W^{2}}\right) - B\beta l \left(\frac{1}{U^{2}} + \frac{1}{W^{2}}\right) + G\left\{\frac{p}{W}\frac{I_{l}'(W)}{I_{l}(W)} + \frac{p}{U}\frac{J_{l}'(U)}{J_{l}(U)}\right\} + D\left\{\frac{p}{W}\frac{I_{l}'(W)}{I_{l}(W)} + \frac{p}{U}\frac{Y_{l}'(U)}{Y_{l}(U)}\right\} = 0$$

 e_{ϕ} at R = 1 Eqn 2 3 3

$$A\left\{\frac{qn_2^2}{W}\frac{I_l'(W)}{I_l(W)} + \frac{qn_1^2}{U}\frac{J_l'(U)}{J_l(U)}\right\} + B\left\{\frac{qn_2^2}{W}\frac{I_l'(W)}{I_l(W)} + \frac{qn_1^2}{U}\frac{Y_l'(U)}{Y_l(U)}\right\} + G\beta l\left(\frac{1}{U^2} + \frac{1}{W^2}\right) + D\beta l\left(\frac{1}{U^2} + \frac{1}{W^2}\right) = 0$$

 h_{ϕ} at R = 1 Eqn 234

$$A\frac{\beta l}{U^{2}C}\frac{J_{l}(UC)}{J_{l}(U)} + B\frac{\beta l}{U^{2}C}\frac{Y_{l}(UC)}{Y_{l}(U)} - G\frac{p}{U}\frac{J_{l}'(UC)}{J_{l}(U)} -D\frac{p}{U}\frac{Y_{l}'(UC)}{Y_{l}(U)} + E\frac{\beta l}{W^{2}C}\frac{K_{l}(WC)}{K_{l}(W)} - F\frac{p}{W}\frac{K_{l}'(WC)}{K_{l}(W)} = 0$$

 e_{ϕ} at R = C Eqn 2.35

$$-A\frac{qn_{1}^{2}}{U}\frac{J_{l}'(UC)}{J_{l}(U)} - B\frac{qn_{1}^{2}}{U}\frac{Y_{l}'(UC)}{Y_{l}(U)} - G\frac{\beta l}{U^{2}C}\frac{J_{l}(UC)}{J_{l}(U)}$$
$$-D\frac{\beta l}{U^{2}C}\frac{Y_{l}(UC)}{Y_{l}(U)} - E\frac{qn_{2}^{2}}{W}\frac{K_{l}'(WC)}{K_{l}(W)} - F\frac{\beta l}{W^{2}C}\frac{K_{l}(WC)}{K_{l}(W)} = 0$$

 h_{ϕ} at R = C Eqn 2 3 6

Each pair of U and W values (same β value) that allow the value of the determinant to be zero (for a range of *l* values) is considered to be a valid solution, and therefore an allowed mode, provided it also meets the cut-off condition (which is discussed in section 2.4)



Figure 2-2 Radial E field of (3, 3) mode

A computationally generated radial field for the (3, 3) mode is shown in figure 2-2 The continuity of E_r at both interfaces R = 1 (i e r = a) and R = C (i e r = b) may be seen Figure 2-3 shows the radial E field of the (1, 3) mode in the annulus only (from R = 1 to 1 5) It can be seen from the graph that there are two sets of intensity maxima, each 180° degrees apart, and lessening in intensity as the distance from the centre of the annulus increases The number of intensity maxima in each row was found to be equal to (l + 1), and the number of rows was found to be equal to (2m)





Figure 2-3: Radial E field in three dimensions.

2.4 Mode cut-off condition.

A mode is considered to be cut-off when W = 0 ($\beta = n_2 k$). Applying this condition to the six by six determinant gives line 1 as

$$A\frac{J_{l}(UC)}{J_{l}(U)} + B\frac{Y_{l}(UC)}{Y_{l}(U)} = 0 \qquad Eqn \ 2.4.1$$

 $(K_t(0) \rightarrow \infty)$

likewise line 2 gives

$$G\frac{J_{l}(UC)}{J_{l}(U)} + D\frac{Y_{l}(UC)}{Y_{l}(U)} = 0$$
 Eqn 2.4.2

Mulitplying line 3 by W² gives

$$-\beta lA - \beta lB = 0$$

$$\Rightarrow A + B = 0$$

Eqn 2.4.3

Similarly line 4 by W² gives

$$\beta lG + \beta lD = 0$$

$$\Rightarrow G + D = 0$$
Eqn 2 4 4

Mulitplying line 5 by W^2 gives

$$E=0$$
 Eqn 245

And line 6 by W^2 gives

 $F = 0 \qquad \qquad Eqn \ 2 \ 4 \ 6$

The solution of equations 2.4 1 and 2.4.3 give the cut-off condition

$$\begin{vmatrix} J_{i}(UC) & Y_{i}(UC) \\ J_{i}(U) & Y_{i}(U) \\ 1 & 1 \end{vmatrix} = 0$$

eqn 2 4 7
or $J_{i}(UC)Y_{i}(U) - Y_{i}(UC)J_{i}(U) = 0$

For each integer value of l(0, 1, 2) the above equation has many roots, which are specified by m = 1, 2, 3 Thus an array of U values indexed by l and m (U_{4,m}) mode can be created

Each member of the family of $(U_{l,m})_{cut-off}$ values is the lowest possible value for a solution to the six by six determinant for its' given (l, m) values

$$U_{lm} > (U_{lm})_{cul-off}$$
 Eqn 248

Since the maximum value of U is V (the V number of the waveguide) at which W = 0, the cut-off condition for the (l, m) mode becomes

$$J_{l}(VC)Y_{l}(V) - Y_{l}(VC)J_{l}(V) = 0 \qquad Eqn \ 249$$

Solutions to the above equation are given in Abramowitz and Stegun (1964), equation 9-5-28, page 374 as

$$V = (U_{l,m})_{c} \cong \beta + \frac{p}{\beta} + \frac{q - p^{2}}{\beta^{3}} + \frac{e + 4pq - 2p^{3}}{\beta^{5}} + E_{qn\,2\,4\,10}$$

where
$$\beta = m\pi/C - 1$$

 $\mu = 4t^2$
 $p = (\mu - 1) / 8C$
 $q = (\mu - 1) (\mu - 25) (C^3 - 1) / 384C^3 (C - 1)$
 $r = (\mu - 1) (\mu^2 - 114\mu + 1073) (C^5 - 1) / 5120C^5 (C - 1)$

(β above is not the propagation constant defined earlier in equation 1.1.2)

Equation 2.4.10 is valid only for $\beta >> p / \beta$ or

$$m >> l^{\frac{2}{3}} \left(\frac{C-1}{2\pi C^{\frac{1}{3}}} \right)$$
 Eqn 2 4 11

2.5 Mode indices (/m).

For the fundamental mode (l = 0, m = 1) single mode operation exists for V = 12.6 with C = 1.25 and for V = 6.27 with C = 1.5 (V < π / (C - 1) approximately) Equation 2.4.10 can be used to extract the maximum value of m for l = 0 modes and yields the value

$$m_{\max} = \frac{(C-1)}{\pi} V \qquad Eqn \ 25 \ 1$$

Expanding equation 2 4 10 to the second term yields a functional relationship between l and m of

$$El^2 + Fm + Gm^2 = 0 \qquad \qquad Eqn \ 252$$

where E, F and G are functions of V and C (E = 4, F = 8π CV and G = $8\pi^2$ C / (C - 1)) For a large V number waveguide (V >>> 1) the third term in equation 2.5.2 can be disregarded so that

$$El^2 + Fm = 0 \qquad \qquad Eqn \ 2 \ 5 \ 3$$

1 e l and m are related in a parabolic fashion for

$$m >> l^{\frac{2}{3}} \left(\frac{C-1}{2\pi C^{\frac{1}{3}}} \right)$$
 Eqn 2.5.4

For the other extreme, $i \in small m$ and large l the approximation of equation 2.4.10 no longer holds Numerical modelling indicates that an equation of the form

$$m = Sl^2 + Tl + U \qquad Eqn \ 255$$

applies, i.e. the m versus l graph is parabolic in shape (S, T and U are constant for a particular waveguide) Furthermore the data generated for waveguides of different V and C values indicate that to a good approxiation l_{max} is given by

$$l_{\max} = \frac{2(C+1)}{\pi} V \qquad Eqn \, 256$$



Figure 2 4 Example of l versus m graph

2.6 Limiting values of l and m.

The total number of (non-degenerate) modes is given by four times the area of the bounded region shown in figure 2-4 As the area of a parabola is

$$\frac{2}{3} x_{\max} \quad y_{\max} \qquad Eqn \ 2 \ 6 \ 1$$

the total number of bound modes in the waveguide (N) is

$$N = 4 \times \frac{2}{3} \times \frac{2(C+1)}{\pi} V \times \frac{(C-1)}{\pi} V$$

or
$$N = 0 \quad 54(C^2 - 1)V^2$$

Eqn 2 6 2

2.7 An effective V number for the waveguide (V).

Substituting C = b/a and equation 1 1 4 into equation 2 6 2 gives

$$N = 0 \quad 54 \left(\frac{b^2}{a^2} - 1\right) \left(ka\sqrt{n_1^2 - n_2^2}\right)^2$$

$$N \equiv \frac{V'^2}{2}$$

Eqn 2.7 1

where

$$V' = \frac{2\pi}{\lambda} \sqrt{n_1^2 - n_2^2} \sqrt{(b^2 - a^2)}$$
 Eqn 2.72

In the limit $a \rightarrow 0$ V' reduces to the V number of the step index fibre waveguide of equation 1 1 4, and therefore the total number of modes N becomes V²/2, which is predicted by Gloge, equation 36 (1971) for such a waveguide V' is called the effective V number of the hollow cylindrical waveguide This effective V number for the hollow cylindrical waveguide has not been reported to date Tsao et al (1989) in their treatment of a 3 layered fiber structure derive an expression

$$\frac{2\pi}{\lambda}(b-a)\sqrt{n_1^2-n_2^2} \qquad \qquad Eqn \ 2\ 7\ 3$$

for its V number This is significantly different to equation 2.7.2 especially when b >> a

2.8 Single mode operation.

The condition for m_{max} given in equation 2.5.1 may be used to derive the condition for single mode operation of this waveguide Putting $m_{max} = 1$ yields

$$(C-1)V < \pi$$

or $\left(\frac{b}{a}-1\right)ka(NA) < \pi$

$$\Rightarrow (b-a) < \frac{\pi}{(k)NA}$$

$$\Rightarrow (b-a) < \frac{\lambda}{2NA}$$

Eqn 2.8.1

i e if the waveguide thickness (b - a) is less than the light wavelength divided by twice the waveguide numerical aperture NA $[NA^2 = n_1^2 - n_2^2]$, only the fundamental (0, 1) mode can propagate in the waveguide This condition is quoted by Tsao et al (1989) for single mode operation of what they refer to as a "Ring fibre waveguide" [Equation 9 123 of Tsao 1989] This is referenced in Tsao (1992) which treats the three layered cylindrical waveguide using Debye potentials The field functions (E_r, E_φ, E_z) and (H_r, H_φ, H_z) obtained by Tsao (1992) are identical to those quoted in this analysis when his third layer refractive index n₃ is equated to n₂ in this analysis. The author was not aware of this paper when the enclosed analysis was carried out

2.9 The evanescent power fraction of a mode.

The mode power in the z direction in all three zones in the waveguide may be obtained from the Poynting vector

$$P_{z} = \pi a^{2} \int \left(E_{r} H_{\phi} - E_{\phi} H_{r} \right) R dR \qquad Eqn \ 291$$

using the limits appropriate to the zone in question, i.e. (0, 1), (1, C) and (C, ∞) for the inner cladding the core and the outer cladding, with the integration being made over the cross-sectional

area (The radial and azimuthal fields are given previously in equation 2.2.8) As the power ratios in the evanescent fields only are of interest, the factor of 2π and

$$\int_{0}^{2\pi} \cos^{2}(l\phi) d\phi = \pi \qquad \qquad Eqn \ 29 \ 2$$

are ignored

2.10 The evanescent power fraction of a mode.

The evanescent power fraction of each mode within a waveguide may be calculated from

$$f_{lm} = \frac{\left[P_{z}\right]_{R=0}^{1} + \left[P_{z}\right]_{R=C}^{\infty}}{\left[P_{z}\right]_{R=0}^{1} + \left[P_{z}\right]_{R=1}^{C} + \left[P_{z}\right]_{R=C}^{\infty}} \qquad Eqn \ 2 \ 10 \ 1$$

using equation 2.9.1 to evaluate P_z Summing over *l* and m for all allowed modes within a waveguide, and dividing by the number of modes gives the average evanescent power fraction for a mode within a particular waveguide

2.11 Conclusions.

The above analysis shows that a hollow cylindrical waveguide can act as a light guide when that light travels in the allowed modes dictated by the eigenvalue equations and the boundary conditions The core and cladding parameters U and W can be predicted for a waveguide of any given dimension for the equations discussed The number of modes that can be sustained by the waveguide can also be predicted using the above theoretical derivations. The power distribution of core guided light to evanescently bound light can also be described for each mode in the hollow cylindrical waveguide

2.12 References.

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Gloge D, "Weakly Guiding Fibres", Applied Optics 10 pp2252 - 2258 (1971) Eqn 36

Snyder A W and Love J D "Optical Waveguide Theory", (Chapman & Hall (1983)) Eqn 30-9 p593

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Unger H G, "Planar Optical Waveguides and Fibres", (Clarendon Press, Oxford (1980))

3. Model of hollow cylindrical waveguides - a computational analysis.

3.1 Introduction.

This chapter describes the computational methods and computer programs used to create an accurate simulation of the bound modes in a hollow cylindrical waveguide probe Each of the programs used was written in the Matlab language (which is based on matrices), and runs only in the Matlab environment A program called Modes m was written to control the other programmes

3.2 Program to determine mode cut-off values.

The following flow-chart (figure 3 1) describes the construction of the program used to calculate the cut-off values for a given waveguide The main program is called Cuts m, and the program which evaluates the cut-off condition at a particular U, l and C is called Cutout m. The cut-off condition in matrix form is given previously in equation 2.4.7 Any value of U the core mode parameter, which allows the value of the cut-off condition to be zero is considered to be a cut-off value for a particular l and m (the mode indices). It can be seen from equation 2.4.7 that the only other parameter in the equation is the C (= b/a) value. This means that there is only one set of cut-off values for any C, regardless of the actual dimensions of the waveguide. This set of cut-off values is the output of Cuts m, and is saved in matrix form. The code for both of these programs is given in Appendices A and B.


Figure 3-1 Flow chart of program Cuts m



Figure 3-2 Flow chart of Holl m.

3.3 Computer program to solve the eigenvalue equation.

This program was called **Holl m**, the code is listed in Appendix C. The purpose of this program is to calculate the allowed U values in the hollow cylindrical waveguide based on the variable parameters entered by the user. To to this, the six eigenvalue equations of equations 2.3.1 to 2.3.6 were placed in a matrix form. The determinant of the matrix was then calculated. Each value of U (with a particular *l* and m) for which the determinant of the matrix equalled zero was considered to be a bound mode within the waveguide, provided the U value was greater then the appropriate cut-off value for its' given mode indices (*l* and m). An incremental substitution method was used to find the correct U values, with the starting U value being the correct cut-off value, so that this condition was fulfilled automatically. The flow-chart in figure 3-2 shows the logic steps used in **Holl m**, the code is given in Appendix C.

3.4 Program to derive E and H field component amplitudes in core and cladding.

The program designed to derive the E and H field component amplitudes was called Holl_se m The code for this program is given in Appendix D The purpose of the program was to solve the the six equations 2 3 1 to 2 3 6 to find A, B, G D, E and for each mode found by Holl m This was done by setting A = 1, and solving the resultant 5 equations simultaneously to find B, G, D, E and F In matrix form this is described as

$$\begin{bmatrix} a_{2} & a_{3} & a_{4} & a_{5} & a_{6} \\ b_{2} & b_{3} & b_{4} & b_{5} & b_{6} \\ c_{2} & c_{3} & c_{4} & c_{5} & c_{6} \\ d_{2} & d_{3} & d_{4} & d_{5} & d_{6} \\ e_{2} & e_{3} & e_{4} & e_{5} & e_{6} \end{bmatrix} \bullet \begin{bmatrix} B \\ G \\ D \\ E \\ F \end{bmatrix} = \begin{bmatrix} -a_{1} \\ -b_{1} \\ -c_{1} \\ -d_{1} \\ -e_{1} \end{bmatrix}$$
Eqn 3 4 1

or in vector notation

(

Thus the five amplitudes B, G, D, E and F are obtained from the vector \overline{x} in equation 3.4.2 For each (l, m) mode for which a U and W value are known, the above operation is used to determine the wave amplitudes in the core and in the two cladding regions. The flow-chart to describe the steps in the program Holl_se m is shown in figure 3-3 [Any of the six amplitudes could be set at a fixed value, the choice of A = 1 is purely arbitrary]

3.5 Program to evaluate mean evanescent power fraction \overline{f} among modes.

The Matlab program Hollpow m was written to calculate the mean evanescent power fraction \overline{f} among modes This was done by calculating the power contained in the core and evanescent fields of each individual mode being guided by the hollow cylindrical waveguide, as identified by the method described above, then finding the fractional representation of evanescent power and calculating the mean over all modes in the waveguide. The formulas used to find the evanescent power fraction are given previously in sections 29 and 210 (equations 291 and 2101) The flow-chart to describe Hollpow m is shown in figures 3-4 and 3-5 and the code is given in Appendix

Ε



Figure 3-3 Flow-chart for Holl_se m



Figure 3-4 Flow-chart for Hollpow m (continued in figure 3 5)



Figure 3-5 Continuation of flow-chart for Hollpow m.

3.6 Mode power fraction distribution.

Individual evanescent mode power fractions derived by numerical integration of the Poynting vectors (using Hollpow m, and equation 2 10 1) were used to generate a histogram of such fractions and their mean value averaged over all the modes of a particular waveguide For example, for a waveguide for which a = 80µm, C = 1 3, $n_1 = 1.46$, $n_2 = 1.45$ and $\lambda = lµm$ (1 e V = 71.226) the modes are predominantly strongly guided with very low evanescent power fractions.

close to cut-off scales as 1/V' (or 1 4% in this case) It can be seen from table 3-1 that only 0 9% of modes have an evanescent power fraction with a value greater than 0 75 The distribution of modes is shown in figure 3-6

f Value	Percentage of Modes	
f < 0 01	40%	
f > 0 1	5 %	
f > 0 5	2%	
f > 0 75	0 9%	

Table 3-1 f values with corresponding mode percentages



Figure 3-6 Histogram of power fraction f distribution

3.7 Dependence of \overline{f} on V' and (b/a).

Following the same argument as Gloge (1971) the dependence of the average evanescent power fraction f on the V number was investigated. To do this the programs listed above were executed in sequence several times, each time using different inner and outer radii for the dimensions of the hollow waveguide (but the same C value), while leaving all the other parameters the same (refractive indices and wavelength of transmitted light). This yielded a series of average evanescent power tractions \overline{f} with corresponding V numbers. In figures 3-7 and 3-8 the power fractions for the C = 1.5 and C = 1.2 waveguides for $n_1 = 1.46$, $n_2 = 1.45$, $\lambda = l \mu m$ light is plotted against

to investigate a possible 1/V' dependence



Power contained in cladding regions, C = 1.5

Figure 3-7 f versus 1/V' C = 15

It can be seen that the numerical model data is scattered about a straight line through the origin, in both cases The scatter is not unexpected since at a particular V' number there will be one

mode extremely close to cut-off (W = 0) which will give an unusually high evanescent power fraction for that mode



Figure 3 8 f versus 1/V C = 12

Gloge (1971) has shown that for a step index fibre waveguide the mean evanescent power fraction f averaged over all bound modes is proportional to 1/V Payne and Hale (1993) find the same dependency with a different multiplicative constant. Both use an approximation in their analysis that W >>> l for all modes. Numerical modelling of the evanescent fraction in this fibre case without the approximation (W >>> l) indicates that

$$\bar{f} \equiv \frac{0.8}{V} \qquad \qquad Eqn \ 3.72$$

By plotting \overline{f} against C (figure 3-9) it was found from the data obtained for the hollow cylindrical waveguide model that like the fibre waveguide \overline{f} scaled linearly with 1/V' with a small dependence on C given by

$$\bar{f} = \frac{P}{V} \left\{ \frac{C+2P}{C-1} \right\}^{\frac{2}{\pi}}$$
 Eqn 3 7 3

where P = 0.867 In the limit of a = 0 or $C = \infty$ the effective V number V' reduces to the value V and equation 3.7.2 becomes identical to equation 3.7.3



Figure 3-9 Graph of \overline{f} versus C

3.8 Dependence of N on V'.

It has been shown in section 2.6 that the total number of bound waveguide modes is predicted to scale approximately as V'^2 or as $(b^2 - a^2)'$ (V' is defined in equation 2.7.2) This was verified numerically by counting modes for various waveguide dimensions, as shown in figures 3-10 and 3-11 with C = 1.2 and 1.5 respectively. The mode number values were provided by the Hollpow m program A very good fit for the total number of modes versus $(b^2 - a^2)$ was obtained, verifying the N = $V'^2/2$ type of relationship for the hollow waveguide Figure 3-11 C = 12





3.9 Conclusions.

The computer programs listed above represent a model of the effect a hollow cylindrical waveguide has on light being transmitted through it. The model parameters can be varied in terms of refractive index of the core and cladding, the waveguide dimensions and the wavelength of light being transmitted through it. The model yielded theoretical relationships between the number of modes, N, the mean evanescent power fraction, \vec{f} , and the effective V number, V', and showed the inter dependence of the various parameters

3.10 References.

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4. Evanescent wave spectrophotometry using a hollow waveguide probe.

4.1 Introduction.

The design of an Attenuated Total Reflectance (ATR) Hollow Cylindrical Sensing probe and an absorbance detection system is described. Uniform mode excitation is achieved using a series of launch step-index fibers butt coupled to the ATR probe

4.2 The ATR probe.

A length of fused silica hollow tubing of inner diameter (2a) 9 32 mm and outer diameter (2b) 11 863 mm was chosen as the sensor probe The rod was cut to a length of 280 mm, using a diamond saw and polished at both ends on a Logitech PM2A Lapping Machine, with a set of water based grits of decreasing diameter from 10 μ m to 1 μ m For the polishing the silica was supported in a stainless steel disk, and kept vertical by strapping to the central shank of the "polishing tree" Regular inspection of the end (i e cut) surfaces was carried out to ensure that all surface blemishes were removed by the coarse grit lapping and polished to a high transparency by the final fine grit

4.3 Excitation of modes in the probe.

As discussed briefly in section 4.1 excitation of the modes in the hollow waveguide was achieved by butt coupling an array of step index fibers (CeramOptec GmbH OPTRAN H-UV 1000/1035) to one end of the probe This was done using a machined aluminum plug to which the fibers were epoxied as shown in figure 4-1. The plug was lodged into the waveguide using an o-ring seal



Figure 4-1 Aluminum plug with fibers attached

Light from a 20 W Tungsten halogen light source, powered by a 6 V dc supply was focussed by a 50 mm diameter 65 mm focal length convex-concave lens combination (convex effect) into the fiber bundle as shown in figure 4-2 below. To provide stable emittance conditions the lamp was run below its' 3 3 A rating, a figure of 2 5 A was found to be sufficient. With no liquid surrounding the probe a bright ring of light was observed at the other end of the probe indicating a uniform excitation of the modes in the waveguide probe



Figure 4-2 Light being focussed into fiber bundle

4.4 Theoretical absorbance of hollow waveguide probe.

The probe dimensions listed in 4 1 lead to a value of the dimensionless constant C (= b/a)of 1 2728 from the average values of a series of measurements of a and b The refractive index n_1 of fused silica was taken as 1 46 In order to determine the effective V number of the waveguide its' numerical aperture NA must be determined, i e NA = $(n_1^2 - n_2^2)^{1/2}$ The solution used for evaluation of the probe was Eosin Yellow (C₂₀ H₆ Br₄ Na₂ O₅) in Methanol Eosin yellow has an absorption band centred at 524 nm At the concentrations used the refractive index of the solutions were that of Methanol namely 1 3276 Taking $n_1 = 1$ 46, $n_2 = 1$ 3276 gives a probe numerical aperture of 0 607 Because this numerical aperture is quite large and because the probe was to be excited by light from a set of step index fibers (of small numerical aperture) butt coupled to one end, the limiting numerical aperture of the system was the smaller of the two, which in this case was the NA of the fibers. This excitation using fibers will be discussed in a subsequent section. Here the value of $(n_1^2 - n_2^2)^{1/2}$ will be taken as the numerical aperture of the fibers, namely 0 37 Using equation 2.7.2 with $\lambda = 524$ nm the effective V number of the hollow tube waveguide is then 16281 214 Using equation 3.7.3 the theoretical mean evanescent power fraction \overline{f} is then 245 389 x 10⁶ (where P is set at 0.867 and C at 1.2728)

It has been shown in section 1 6 that for an individual mode the transmitted intensity is related to the launched intensity (I_0) by

$$I = I_0 \exp(-\alpha fz) \qquad \qquad Eqn \ 44 \ 1$$

If all N modes are excited with equal incident power (I_0/N) then the transmitted power (after a length z of absorbing region) is

$$I = \sum_{n=1}^{N} \frac{I_0}{N} \exp(-\alpha f_n z) \qquad \qquad Eqn \, 442$$

so that the absorbance $A' = \log_{10} (I_0 / I)$ is given by

$$A' = -\log_{10}\left\{\frac{\sum_{n=1}^{N} \exp(-\alpha f_n z)}{N}\right\} \qquad Eqn \ 4 \ 4 \ 3$$

When $\alpha fz \ll 1$ for all modes equation 4.4.3 can be evaluated using the expansion exp (-x) \cong 1 - x in which case

$$A' = -0 \ 434 \log_{e} \left\{ \frac{\sum_{n=1}^{N} 1 - \alpha f_{n} z}{N} \right\}$$

= -0 \ 434 \log_{e} \left\{ 1 - \alpha z \frac{\sum_{f_{n}}}{\sum_{N}} \right\}
= -0 \ 434 \log_{e} \left(1 - \alpha z \overline{f_{n}} \right)
Eqn 444

where \overline{f} is the mean evanescent power fraction. Using the expansion log $(1-x) \equiv -x$ for small x therefore gives

$$A' = 0 434 \alpha z \bar{f}$$
 provided $\bar{f} \alpha z \ll 1$ Eqn 4.4.5

ie absorbance scales linearly with $\overline{f}\alpha z$

When $\alpha z \ \overline{f} >> 1$ the absorbance of equation 4 4 3 is more complicated and leads to a saturation of absorbance with increasing α , or increasing z, or increasing αz



Figure 4-3 Example of saturation of absorbance with increasing depth

4.5 Absorbance measurement technique.

As shown in figure 4-2 light was launched into the probe using a beam splitter A small fraction of this light is back reflected at the distal end where the probes annular tip interfaces with the liquid Because the numerical aperture of the probe is quite small, modes strike the end face at angles very close to the normal and are reflected with a power reflection coefficient of approximately

$$R = \left[\frac{n_1 - n_2}{n_1 + n_2}\right]^2 \qquad \qquad Eqn \ 45 \ 1$$

(the Fresnel reflection coefficient for normal incidence) [It is shown in Appendix G that for a practical range of incident angles θ_1 the reflection coefficient R is approximately independent of θ_1]

If an intensity I_0 is launched into the probe (at a wavelength λ) then $I_0 \exp(-\overline{f\alpha z})$ reaches the endface, $R_0 I_0 \exp(-\overline{f\alpha z})$ is reflected and $R_0 I_0 \exp(-2\overline{f\alpha z})$ is returned to the launch fibers. This is based on absorption occuring at an analytical wavelength. At a wavelength well removed from the absorption band - the so-called reference wavelength - a back reflected intensity of R I_0 occurs (i.e. no attenuation). Thus an evanescent power absorbance (A') of

$$A' = \log_{10} \left\{ \frac{RI_0}{RI_0 \exp(-2\bar{f}\alpha z)} \right\}$$

= (0 434)(2 $\bar{f}\alpha z$), $\bar{f}\alpha z \ll 1$

is obtained This is the previously derived expression of equation 4.4.5 but doubled for 2 way travel of the evanescent wave along the probe length z. As before equation 4.5.2 applies provided $f\alpha z \ll 1$ As previously defined α is the bulk attenuation coefficient of the absorber and z is the immersion depth of the probe in the absorber By comparing the back-reflected light intensity at the analytical and reference wavelength then the absorbance of the probe can be measured

Two interference filters were used to isolate a wavelength band centered at $\lambda = 525$ nm (where Eosin Yellow has an absorption band) and $\lambda = 430$ nm in the blue to one side of the absorption band. The filters were supported on a mechanical slide and could be placed in turn in front of the entrance window of a Hamamatsu 931A photomultiplier tube. The photomultiplier was operated in the "grounded anode" mode, the detector signal being extracted as a voltage across a 10M Ω resistor. The light entering the launch fibers was modulated using a mechanical chopper which operated at a chopping frequency of 330 Hz. The chopper drive unit (Scitec Instruments optical chopper) supplied the square wave pulse train to synchronise phase sensitive detection with a lock-in amplifier. The PM output was fed by coaxial cable to the signal input of an EG&G model lock-in amplifier (model 950VG) whose post filter time constant was set at 3 seconds.

An absorbance measurement then involved recording two output voltages from the lock-in amplifier corresponding to each optical filter being in the light beam returning to the photomultiplier detector A series of back reflected intensity measurements, with the probe enclosed in a light tight box, was made to determine if any drift occured in either (i) the light source intensity or (ii) the photomultiplier output. It was found that the lamp intensity stabilised in 45 minutes after switching on, and there was no detectable photomultiplier drift over a 2 hour period. The PM was powered by a 500 V EHT unit (EMI electron tube division Power Supply PM28B)

4.6 Conclusions.

A relatively simple back-reflection ATR probe, excited by light from a tungsten halogen lamp via an array of step index fibers butt coupled to one end and operated in phase sensitive detection, was constructed. The evanescent light in both the inner hole and surrounding medium may be used to analyse fluids with absorption bands in the visible using a dip-stick style of approach. All optics are concentrated at one end of the probe

5. Experimental absorbances using hollow silica waveguide.

5.1 Introduction.

Absorbance measurements made with the ATR analyser probe discussed in chapter 4 are reported here Results are compared to the mode model predictions of chapters 2 and 3

5.2 Bulk properties of the absorbing cladding.

As stated earlier a solution of Eosin Yellow in Methanol was chosen as the absorbing cladding with which the hollow cylindrical ATR probe was to be evaluated The bulk absorption properties of this chemical at $\lambda = 524$ nm were examined using a SHIMADZU (UV - 1201) UV-VIS spectrophotometer Solutions of various molar concentrations (1M = 0 69186 gram/cc of solute) were prepared and their bulk absorbances (at 524 nm) were measured. In each case a cuvette containing the solution was inserted in the spectrophotometer beam and the beam attenuation compared to that of the solute (methanol) on its own. A graph of absorbance versus solution concentration (figure 5-1) was prepared and a least-square fit line generated throught the data points (using an "Origm" subroutine). From the best-fit slope the bulk absorption coefficient α for Eosin Yellow was found. For a 1M solution α was found to be $\alpha = 23250.0 \text{ mm}^{-1}$ (or 0.02325 mm⁻¹ μ M⁻¹).

For a weaker solution - say of concentration 10³ M, the corresponding α value is 1000 times smaller For evanescent wave absorption the parameter of interest is $\overline{f\alpha}$, where \overline{f} is the mean evanescent power fraction among the modes



Figure 5-1 Bulk absorbance versus solution concentration

5.3 Evanescent absorbance as a function of probe immersion depth.

We have seen in section 4.4 that the evanescent absorbance A' is predicted to scale linearly with the product $\overline{f}\alpha z$. For a probe of constant dimensions (a, b) the evanescent absorbance is predicted, therefore, to scale linearly with the immersion depth z. This was investigated experimentally

A solution of 28 756 μ M Eosin Yellow in Methanol was prepared and placed in a graduated cylinder which could be raised or lowered around the hollow ATR probe A series of absorbance measurements (at $\lambda = 524$ nm) were made as a function of the immersion depth z These results are shown in figure 5-2 The functional relationship of equation 4.4.2 is vindicated in this straight line graph. The saturation effect alluded to in section 4.5 for high concentrations of solution or large immersion depths (i.e. when $\overline{f}\alpha z \ge 1$) was observed with this probe



Figure 5.2 Evanescent absorbance versus depth (28 756 µM solution) A 306 902 µM solution was used with the same immersion depth Results are shown in figure 5-3



Figure 5-3 Evanescent absorbance versus depth, 306 902 µM concentration

It can be seen that the complete absorption of some modes or the differential attenuation of the modes - varying from weak attenuation for modes far from cut-off (U << V) to strong attenuation for modes near cut-off (U \approx V) - gives rise to a non-linear dependence of absorbance on immersion depth (z), for fixed concentrations This effect was predicted in section 4.3 This effect also occurs in solid cylindrical fibre evanescent wave probes as observed by Ruddy (1994) In the intervening region, i.e. of concentration from 28 μ M to 306 μ M a family of absorbance versus immersion depth curves were obtained at various concentrations These showed a gradual transition from linearity (for low concentrations of ~ 30 μ M) to saturation for higher concentrations

5.4 The experimental \overline{f} value.

For the weak solution ($\overline{f\alpha z} < 1$) the linear absorbance graph of figure 5-2 can be used to extract an experimental \overline{f} value. The graph slope (s) of 0 0003 combined with a bulk absorption coefficient α of 0 668577 mm⁻¹ (for 28 756 μ M Eosin Yellow solution) yields an experimental \overline{f} value of 224 3571 x10⁻⁶ This may be compared to the theoretical value of equation 3 7 3 taking P = 0 867, C = 1 2728, V' = 16281 214 of \overline{f} = 245 38957 x10⁻⁶ This is a difference of 8 57% It can be seen that the mode modelling of chapters 2 and 3 and the experimental measurement of chapter 5 are in very good agreement.

5.5 Conclusions.

Experimental measurements of absorbance using a hollow cylindrical ATR probe were used to extract a mean evanescent power fraction (\vec{f}) among all the bound modes of the waveguide Good correlation between the experimentally derived \vec{f} values and that predicted by rigorous mode analysis for such a waveguide indicates the latter. It should be stated that the mode analysis carried out does not assume the "weakly guiding" approximation ($n_1 \cong n_2$) as in general with glass based probes and liquid absorber solutions that approximation (commonly used in step-index fibre mode analysis) is not valid

5.6 References.

Ruddy V, "Non linearity of absorbance with sample concentration and path length in evanescent wave spectroscopy using optical fibre sensors", Optical Engineering 33, no 12, pp (3891-2893) (1994)

Appendix A

The following is the Matlab code for the Cuts m program Any lines starting with a '%' sign are comments on the code, not part of the code itself, and are ignored by the Matlab complier

% Title Cuts m

% Aim To calculate cut off values for a hollow cylindrical % waveguide by stepping through the cut % off equation until the zeros are located

$delta_x = 1$,	% step	size in	U
_				

x1 = 1, % starting value for U

for J =0 J_max, %	6 loop through all orders, where 1 i	s the order l
-------------------	--------------------------------------	---------------

count = 1, % resets the count of U values in each l order

 $y_1 = cutout(j,x_1,C),$ % calculate value of cut-off equation

while $(x_1 < x_max)$, % step through all U values

$x2 = x1 + delta_x,$	% increment U by delta U
$\lim t = x^2$,	% stores largest U value tested so far
$y^2 = cutout(j,x^2,C),$	% calculate value of cut-off equation
test = $x1$,	% stores next largest U value tested so

uf((y2/y1) < 0),	% if sign change occurs.	, root is isolated
------------------	--------------------------	--------------------

while (test < limit),

$x^2 = x^1 + (delta_x/10),$	% increase U value by 10% of delta U
$y^2 = cutout(y x^2 C)$	% calculate value of cut-off equation
$f(y^2/y^1) > 0,$	% if sign change has not occured

far

else

slope =
$$(y^2 - y^1)/(x^2 - x^1)$$
, % sign change has occured

U = x1 - (y1/slope),	% calculate correct root value
ucut(j+1,count) = U,	% save U value in array
count = count + 1,	% increment counter

test = limit+1,	%	make	'test'	>	'lımıt'
-----------------	---	------	--------	---	---------

end,

else

y1 = y2,	% prepare for next run through loop
x1 = x2,	% prepare for next run through loop
end,	% end of 'while (test < limit),' statement

end, % end of 'if (y2/y1) < 0)' statement

if count == 1,	% if no cutoffs are found

```
break, \% exit 'while (x1 < x_max),' loop
```

end,

ıf j > 0,

If $ucut(j,1) < x_{max}$, % If all roots of present order ! have not yet been tound,

x1 = ucut(j+1,1), % start searching at last known U value

end,

else

x1 = 6, % reset to lowest U value, to search next l order

end,

ucut(j,) % output to screen all found U values

end, % end of 'j =0 j_max' loop

Appendix B

This program evaluates the cut-off equation at values passed to it from the program that called it in this case the calling program is Cuts m

% Title Cutout m

% Aim To calculate value of cut off equation at a given U value

function[y_val] = cutout(j,x,C) % defines the function name and the number and value of
variables % the function will use

y1 = (bessely(j,x) * bessely(j,C*x)), % defines the equation parts used in the function

 $y^2 = (bessely(j,C^*x) * bessely(j,x)),$

 $y_val = y1 - y2$, % calculates the value of the function at the specified parameter values,

% and returns this value to the calling program

Appendix C

This program finds the correct U values for each order by using the cut-off values calculated by Cuts m

% Program title holl8 m

% Purpose To establish the correct modes for each order of t function in a hollow cylindrical

% waveguide

% Method To cycle through a range of U values and calculate the determinant of the matrix for each

% valu	e until al	l the value	ies foi	which	the	determinant is	zero	are foun	d
--------	------------	-------------	---------	-------	-----	----------------	------	----------	---

global 1, % declares '1' as a variable used throughout the program

 $k = (2 * p_i)/lambda,$ % calculates the wavenumber k

Z = 377, % defines the characteristic impedance of free space

j = 1,

$\mathbf{p} = -\mathbf{k} * \mathbf{Z},$	% substitutional variable
q = k / Z,	% substitutional variable

beta(1) = n1*k,	% sets the value of the porpagation constant
$U_{max} = in_{rad} * sqrt((n1^2*k^2) - (n2^2*k^2)),$	% calculates the maximum U value
delta = 0.01,	% sets the step size in U
$limit = (U_max*100)$ -delta,	% calculates the limit in steps of delta
u(1) = 0 0,	% sets starting $U = 0$

t = 1,

fid = fop	en('holl8a.tmp','w');	% opens a file for output.
for i = 1	elength(ucut),	% length(ucut) is the size of the array holding cut-off values.
	if ucut(i,1) > U_max,	% if first value of any row (i) is > U_{max} .
	$j_{lim} = i-2;$	% sets limit of 'j' loop.
	break;	% break out of loop.
	end;	% end of 'if' statement.
end;		% end of 'for' loop.
x = 0;		% initialise x variable.
for $j = 0$:j_lim;	% for every order (j) from 0 to j_lim in steps of 1.
	x = x + 1;	% increment x.
	i = 3;	
	t = 1;	
	while i < limit,	% limit is defined at U_max - delta.

% u(i) is incremented from cut off value

% there is no zeroth row in array, so order 0 values stored in1st row.

u(i) = ucut(j+1,x) + (t*delta);	% u set at cut-off	value + (t*delta),
t = t+1;	% counter of num	aber of increments on u.
$beta(i) = sqrt((n1^2*k^2) - (u(i)^2/k^2))$	/in_rad^2));	% calculate porpagation const.
$\mathbf{U}=\mathbf{u}(\mathbf{i});$	% set va	lue of core mode parameter.
$W = in_rad * sqrt(beta(i)^2 - (n2^2))$	2*k^2)); % calcul	late cladding mode parameter.
$UC = out_rad * sqrt((n1^2*k^2) - 1)$	beta(i)^2);	% calculate U*C.

 $WC = out_rad * sqrt(beta(1)^2 - (n2^2*k^2)), \qquad \% calculate W*C$

BJU = bessely(J,U),	% calculate value of Bessel J at U
$DBJU = -bessel_J(j+1,U) + (j/U)*BJU,$	% calculate value of Bessel J derivative

BJUC = besselj(j,UC), % calculate value of Bessel J at UC DBJUC = -besselj(j+1,UC) + (j/UC)*BJUC, % calculate value of Bessel J derivative

BYU = bessely(J,U),	% calculate value of Bessel Y at U
DBYU = -bessely(j+1,U) + (j/U)*BYU,	%calculate value of Bessel Y derivative

BYUC = bessely(j, UC),	% calculate value of Bessel Y at UC
DBYUC = -bessely(j+1,UC) + (j/UC)*B	UC, % calculate value of Bessel Y
	% derivative

```
BIW = besseli(j,W), \qquad \% calculate value of Bessel I at WDBIW = besseli(j+1,W) + (j/W)*BIW, \qquad \% calculate value of Bessel I derivative
```

$BKWC = besselk(J,WC), \qquad \%$		6 calculate value of Bessel I at WC	
DBKWC = -besselk(j+1,WC) + (j/WC)*B	KWC,	% calculate value of Bessel I	
		% derivative	

BKW = besselk(J,W), % calculate value of Bessel K at W

DBKW = -besselk(j+1,W) + (j/W) * BKW, %calculate value of Bessel K derivative

% x1-4 and y1-6 are substitutional variables used in the matrix

x1 = BJUC / BJU, x2 = BYUC / BYU, x3 = BKWC / BKW, $x4 = (1/U^2) + (1/W^2),$

y1 = DBIW / BIW, y2 = DBJU / BJU, y3 = DBYU / BYU, y4 = DBYUC / BYU, y5 = DBKWC / BKW, y6 = DBJUC / BJU,

% the following are the elements of the 6*6 matrix,

$$a1 = x1$$
,
 $a2 = x2$,
 $a3 = 0$,
 $a4 = 0$,
 $a5 = -x3$,
 $a6 = 0$,
$$b1 = 0$$
,
 $b2 = 0$,
 $b3 = x1$,
 $b4 = x2$,
 $b5 = 0$,
 $b6 = -x3$,

$$c1 = (-beta(1) * J) * x4,$$

$$c2 = (-beta(1) * J) * x4,$$

$$c3 = ((p/W) * y1) + ((p/U) * y2),$$

$$c4 = ((p/W) * y1) + ((p/U) * y3),$$

$$c5 = 0,$$

$$c6 = 0,$$

$$d1 = ((q * n2^{2} * y1) / W) + ((q * n1^{2} * y2) / U),$$

$$d2 = ((q * n2^{2} * y1) / W) + ((q * n1^{2} * y3) / U),$$

$$d3 = beta(1) * j * x4,$$

$$d4 = beta(1) * j * x4,$$

$$d5 = 0,$$

$$d6 = 0,$$

$$e1 = (beta(1) * j * x1) / (U^{2} * C),$$

$$e^2 = (beta(1) * J * x^2) / (U^2 * C),$$

$$e3 = (-p * y6) / U,$$

$$e4 = (-p * y4) / U,$$

$$e5 = (beta(i) * j * x3) / (W^2 * C),$$

$$e6 = (-p * y5) / W,$$

 $f1 = (-q * n1^{2} * y6) / U,$ $f2 = (-q * n1^{2} * y4) / U,$ $f3 = (-beta(1) * j * x1) / (U^{2} * C),$ $f4 = (-beta(1) * j * x2) / (U^{2} * C),$ $f5 = (-q * n2^{2} * y5) / W,$ $f6 = (-beta(1) * j * x3) / (W^{2} * C),$

matr1 = [a1 a2 a3 a4 a5 a6 b1 b2 b3 b4 b5 b6 c1 c2 c3 c4 c5 c6 d1 d2 d3 d4 d5 d6 e1 e2 e3 e4 e5 e6 f1 f2 f3 f4 f5 f6], %

% places each element in the matrix

deter(i) = det(matr1),	% calculates the determinant of the matrix
m(i) = deter(i)/deter(i-1),	% divides determinany value of matrix by
	% previous determinant value
if(m(i) < 0 0),	% if sign change has occured

ıf (deter	(1-2)) = 0 0, %	eliminates fals	se roots	
	$U_t(j+1,x) = u(i)$	- ((deter(1) * d	lelta) / (deter(1) - deter(1-1))),	
	% calculates exac	t value of core	e mode parameter	
	<pre>beta_t(x) = sqrt((</pre>	n1^2*k^2) - (1	U_t(j+1,x)^2/in_rad^2)),	
	% calculates valu	e of correspon	nding propagation constant	
	$W_t(j+1,x) = in_{-1}$	rad* sqrt(beta	$t(x)^2 - (n2^2 k^2)),$	
	% calculates valu	e of correspor	nding cladding mode parameter	
	fprintf(fid,'% Of	% 0f %f	%f\n',j,x,U_t(j+1,x),W_t(j+1,x))),
	% prints l, m, U a	and W to the o	utput file	
	flag = 1,	% 11	needs to be incremented by 10	
	save djc mat,	% sa	ives all variables in the workspace	;
end,	% end o	f true roots 'il	f' statement	
end,	% end o	f sign change	'ıf' statement	
if flag == 1,	% if a ro	oot has been fo	ound	
x = x +	1, % increa	ment mode co	unt	
t = 1,	% reset	delta step sıze	to 1	
1 = 1 + 1	0 % to sep	oarate stored d	eterminant values in 'deter' array	
$\mathbf{flag} = 0$), % reset	flag		
end,	% end o	f positive flag	'if' statement	
1 = 1 +1,	% increa	ment '1' count	er	
ux = ucut(y+1,x)	+ (t*delta) + 0 01	,% ux is a var	able used for checking	
ıf ux >= (U_max	a - 2*delta)	% if the next	cut-off value is greater than	

U_max

i = limit,	% to break out of loop
$\mathbf{x}=0,$	% reset mode counter
$\mathbf{u}=\mathbf{u}^{*}0,$	% reset u array
deter = deter*0,	% reset deter array values
	% end of if loop

end,	% end of 'while i < limit' loop
end.	% end of 'for $j = 0$ j_lim' loop
fclose('all'),	% closes all open files

end,

Appendix D

% Program title holl_se m

% Purpose To compute the coefficients apporpriate to each allowed mode in a hollow cylindrical

% waveguide

% Method A 5 * 5 determinant will be inverted and multiplied by a column matrix to extract the

% coefficients

% The 5*5 matrix consists of 5 lines taken from the 6*6 matrix in holl m, with all of the first column % moved to a column matrix to form the constants

j = 1,	% sets j back to 1 after exiting the previous program
p = - k * Z,	% substitutional variable
q = k / Z,	% substitutional variable
$\mathbf{x} = 1,$	% sets x back to 1 after exiting the previous program
beta(1) = n1*k,	% sets first propagation constant value
fid2 = fopen('hollse dat','w'),	% opens a file for the output data
$[limit,maxmode] = size(U_t),$	% finds the size of the Matrix containing the true U

% (core mode parameter) values

i = 1,

for $j = 0$ limit-1,	% loops through all mode orders
$\mathbf{x} = 1$,	% sets mode index (within each order) to 1

while x <= maxmode,

$\cos = \cos * 0;$	% matrix to hold	coefficient values set to zero.
$U = U_t(j+1,x);$	% U value is retri	eved from the holding matrix.
$beta(i) = sqrt((n1^2*k^2) - (U^2/in$	_rad^2));	% claculates the propagation constant.
$W = W_t(j+1,x);$	% W value is retr	ieved from the holding matrix.
$UC = out_rad * sqrt((n1^2*k^2) - k^2)$	beta(i)^2);	% calculates U*C.
WC = out_rad * $sqrt(beta(i)^2 - (n))$	2^2*k^2));	% calculates W*C.

BJU = besselj(j,U);	% calculates value of bessel J at U.	
DBJU = -besseli(i+1,U) + (i/U)*BJU;	% calculates derivative of bessel J at U	

BJUC = besselj(j,UC); % calculates value of bessel J at UC.

DBJUC = -besselj(j+1,UC) + (j/UC)*BJUC; % calculates derivative of bessel J at UC.

BYU = bessely(j, U);	% calculates value of bessel Y at U.
DBYU = -bessely(j+1,U) + (j/U)*BYU;	% calculates derivative of bessel Y at U.

BYUC = bessely(j,UC); % calculates value of bessel Y at UC.
DBYUC = -bessely(j+1,UC) + (j/UC)*BYUC; % calculates derivative of bessel Y at UC.

```
BIW = besseli(j,W); \\ \label{eq:BIW} & \mbox{calculates value of bessel I at W}. \\ DBIW = besseli(j+1,W) + (j/W)*BIW; \\ \mbox{\% calculates derivative of bessel I at W}. \\ \label{eq:BIW}
```

BKWC = besselk(J,WC), % calculates value of bessel K at WC DBKWC = -besselk(J+1,WC) + (J/WC)*BKWC, % calculates derivative of bessel K at WC

BKW = besselk(j,W), % calculates value of bessel K at W DBKW = -besselk(j+1,W) + (j/W) * BKW, % calculates derivative of bessel K at W

% x1-4 and y1-6 are substitutional variables used in the matrix

x1 = BJUC / BJU, x2 = BYUC / BYU, x3 = BKWC / BKW, x4 = (1/U^2) + (1/W^2),

temp2 = $((n2^2 * y1)/W) + ((n1^2 * y3)/U)$, coe(1) = (-temp1)/(temp2), % first coefficient is found coe(4) = (x1 + (x2 * coe(1))) / x3, % fourth coefficient is found coe(2) = 0, % all others are set to zero coe(3) = 0, coe(5) = 0, else % if mode order is not zero

% the following are the elements of the 5*5 matrix,

$$a2 = x2,$$

$$a3 = 0,$$

$$a4 = 0,$$

$$a5 = -x3,$$

$$a6 = 0,$$

$$c2 = (-beta(1) * j) * x4,$$

$$c3 = ((p/W) * y1) + ((p, -1)) + ((p,$$

c6 = 0,

$$d2 = ((q * n2^2 * y1) / W) + ((q * n1^2 * y3) / U),$$

$$d3 = beta(1) * j * x4,$$

$$d4 = beta(1) * j * x4,$$

((p/U) * y2),

((p/U) * y3),

d5 = 0,d6 = 0,

e2 = (beta(1) * j * x2) / (U^2 * C), e3 = (-p * y6) / U, e4 = (-p * y4) / U,

 $e5 = (beta(1) * 1 * x3) / (W^2 * C),$

e6 = (-p * y5) / W,

$$f2 = (-q * n1^2 * y4) / U,$$

 $f3 = (-beta(1) * j * x1) / (U^2 * C),$

 $f4 = (-beta(1) * 1 * x2) / (U^2 * C),$

 $f5 = (-q * n2^2 * y5) / W,$

$$f6 = (-beta(1) * 1 * x3) / (W^2 * C),$$

matr l	= [a2 a3 a4 a5 a6	% puts each element of the 5*5 matrix in the correct
	c2 c3 c4 c5 c6	% position
	d2 d3 d4 d5 d6	
	e2 e3 e4 e5 e6	
	f2 f3 f4 f5 f6],	

% the following are the elements of the column matrix

a1 = -x1,

c1 = (beta(i) * j) * x4,

$$d1 = ((-q * n2^2 * y1) / W) - ((q * n1^2 * y2) / U),$$

$$e1 = (-beta(i) * j * x1) / (U^2 * C),$$

$$f1 = (q * n1^2 * y6) / U,$$

```
matr2 = [a1, c1, d1, e1, f1], % sets up the column matrix

coe = matr1 \setminus matr2, % multiplies inverse of matr1 by column matrix

end, 
% the '\' is a special matlab character for this operation

fprintf(fid2,'\n% 0f % 0f % 4f % 4f',j,x,U_t(j+1,x),W_t(j+1,x)),

fprintf(fid2,' % 5f % 5f % 5f % 5f % 5f % 5f\n',coe(1),coe(2),coe(3),coe(4),coe(5)),
```

% the order, mode index m, U value, W value and the corresponding coefficients are output to a file

if x < maxmode, % if the last mode in a given order has not been reached

if $U_t(j+1,x+1) == 0$, % if the next mode of that order is zero

x = maxmode + 1, % make x > maxmode

end,

	end,	% end of 'x < maxmode' loop
	$\mathbf{x} = \mathbf{x} + 1,$	% increment x
end,		% end of 'while x <= maxmode' loop
end,		% end of 'for $j = 0$ limit-1' loop
fclose('a	ຟ'),	% close all open files
save c \	oollse mat	% save variable values in a matlab workspace file

Appendix E

% Title hollpow m

- % Object To calculate the power ratio in a hollow cylindrical waveguide
- % Method To numerically integrate over all allowed modes in annulus

J = 1,	% set mode order no back to 1
p = -k * Z,	% substitutional variable
q = k / Z,	% substitutional variable
x = 1,	% set mode order index to 1
beta(1) = n1*k,	% set first value of propagation constant

AA = 1,	% AA is first coefficient, and was set to 1 in previous program to enable

the

% calculation of the other coefficients

fid = fopen('hollse dat', 'r'),	% opens the file to which Hollse m saved the results of its run
ın_data = fscanf(fid,'%f\n'),	% reads in the data as one column
fclose(fid),	% closes the file
lumit = length(in_data),	% calculates the number of variables in the column of data

1 = 1,	% set index equal to 1, the first element of the data column	
while 1 <= limit,	% read in nine values for each single mode	
$j = in_data(i),$	% each input value is assigned a label and stored	

 $x = in_data(i+1),$

 $U_t(j+1,x) = in_data(i+2),$

 $W_t(j+1,x) = in_data(i+3),$

 $BB(j+1,x) = in_{data(i+4)},$

 $CC(j+1,x) = in_{data(i+5)},$

 $DD(j+1,x) = in_data(i+6),$

 $EE(j+1,x) = in_data(i+7),$

 $FF(j+1,x) = in_data(i+8),$

1 = I+9, % index incremented by nine to the next set of nine values

% end of data storage loop

end,

$j_limit = in_data(1-9),$	% sets the maximum value of j
$[qwe,x_limit] = size(U_t),$	% finds the size of the array holding the U values

 $U_t(j_{1}, x_{1}, x_{$

fid1 = fopen('hollpow5 dat', 'w'), % opens a file for output data

F1 = 0, % the variables that will store the power fractions are set to 0

F2 = 0,

F3 = 0,

$mode_count = 0,$	% the mode counter is set to 0
for $j = 0$ j_limit,	% loop through all the mode orders

x = 0, % sets mode index to 0

while $x < x_{lim}$, % while mode index is less than its maximum value

x = x + 1, % increment mode index

 $U = U_t(j+1,x)$, % retrieve corresponding U value

 $W = W_t(y+1,x)$, % retrieve corresponding W value

beta = sqrt($(n1^2*k^2) - (U^2/n_rad^2)$), % calculate corresponding prop const

BJU = bessely(J,U),	% calculates bessel J at U	
$DBJU = -bessel_J(j+1,U) + (j/U)*BJU,$	% calculates derivative of bessel J at U	

BYU = bessely(j,U),	% calculates bessel Y at U
DBYU = -bessely(j+1,U) + (j/U)*BYU,	% calculates derivative of bessel Y at U

BIW = besseli(j,W),	% calculates bessel I at W

 $DBIW = besseli(j+1,W) + (j/W)*BIW , \qquad \% \text{ calculates derivative of bessel I at W}$

BKW = besselk(J,W),	% calculates bessel K at W
DBKW = -bessel(1+1, W) + (1/W) * BKW,	% calculates derivative of bessel K at W

% the following are substitutional variables

$$a1 = (beta * (AA + BB(j+1,x))) / (W * BIW),$$

$$a2 = (p * j * (CC(j+1,x)+DD(j+1,x))) / (W^2 * BIW),$$

$$a3 = (beta * j * (CC(j+1,x)+DD(j+1,x))) / (W^2 * BIW),$$

$$a4 = (q * n^2^2 * (AA+BB(j+1,x))) / (W * BIW),$$

b1 = (-beta * AA) / (U * BJU),

$$b5 = (-beta * j * CC(j+1,x)) / (U^2 * BJU),$$

$$b6 = (-beta * j * DD(j+1,x)) / (U^2 * BYU),$$

$$b7 = (-q * n1^2 * AA) / (U * BJU),$$

$$b8 = (-q * n1^2 * BB(j+1,x)) / (U * BYU),$$

$$c1 = (beta * EE(j+1,x)) / (W * BKW),$$

 $c2 = (p * j * FF(j+1,x)) / (W^2 * BKW),$

$$d1 = (beta * j * FF(j+1,x)) / (W^2 * BKW),$$

$$d2 = (q * n2^2 * EE(j+1,x)) / (W * BKW),$$

$$e1 = (-beta * j * (AA+BB(j+1,x))) / (W^2 * BIW),$$

$$e^{2} = (p * (CC(j+1,x)+DD(j+1,x))) / (W * BIW),$$

$$f1 = (beta * (CC(j+1,x)+DD(j+1,x))) / (W * BIW),$$

$$f2 = (-n2^2 * q * j * (AA+BB(j+1,x))) / (W^2 * BIW),$$

$$g1 = (heta * j * AA) / (U^2 * BJU),$$

 $g2 = (heta * j * BB(j+1,x)) / (U^2 * BYU),$

$$g3 = (-p * CC(j+1,x)) / (U * BJU),$$

$$g4 = (-p * DD(j+1,x)) / (U * BYU),$$

$$h1 = (-beta * CC(j+1,x)) / (U * BJU),$$

$$h2 = (-beta * DD(j+1,x)) / (U * BYU),$$

$$h3 = (nI^2 * q * AA) / (U^2 * BJU),$$

$$h4 = (n1^2 * q * BB(j+1,x)) / (U^2 * BYU),$$

$$11 = (-beta * J * EE(J+1,x)) / (W^2 * BKW),$$

 $12 = (p * FF(J+1,x)) / (W * BKW),$

$$J1 = (beta * FF(J+1,x)) / (W * BKW),$$

$$j2 = (-n2^2 * q * j * EE(j+1,x)) / (W^2 * BKW),$$

	$\mathbf{R}_{\mathbf{m}\mathbf{n}1}=0\ 0,$	% sets the smallest radius of the waveguide
	$R_{min}^2 = 1,$	% radius at inner interface of cladding and core
	$R_{min3} = C,$	% radius at inner interface of cladding and core
	$R_{min4} = 2*C,$	% radius at outer surface of cladding
	deltaR = 0.01,	% step size used to increment radius
$maxN1 = R_mm2 / deltaR$, % number of steps from $R = 0$ to $R = 1$		
$maxN2 = ((R_min3 - R_min2)*2) / deltaR, % number of steps from R = 1 to R = 1$		
	$maxN3 = (R_min4 - R_n)$	nun3) / deltaR, % number of steps from $R = C$ to $2*C$

sum1 = 0, for $N = 1 \max N1$, % from R = 0 to R = 1 $R = R_{min1} + (N * deltaR),$ % calculate radius $UR = R * in_rad * sqrt((n1^2*k^2) - beta^2),$ % U*R $WR = R * m_rad * sqrt(beta^2 - (n2^2 * k^2)),$ %W*R BIWR = besseli(J,WR),% bessel I at W*R DBIWR = besseli(j+1,WR) + (j/WR)*BIWR,% derivative of bessel % I at W*R term1 = (a1 * DBIWR) + ((a2 * BIWR)/R), % substitutional variables term2 = ((a3 * BIWR)/R) + (a4 * DBIWR),term3 = ((e1 * BIWR)/R) + (e2 * DBIWR), $\operatorname{term4} = (f1 * DBIWR) + ((f2 * BIWR)/R),$ total(1) = (term1 * term2) - (term3 * term4),total(1) = abs(total(1)), % calculates power at each radius sum1 = sum1 + (total(1) * R * deltaR), % running total of power values end, % end of inner section loop

sum 2 = 0,

for $N = 1 \max N2$	% from $R = 1$ to $R = C$

 $R = R_m n^2 + (N * deltaR),$ % calculates radius

$$UR = R * in_rad * sqrt((n1^2*k^2) - beta^2), \qquad \% U*R$$
$$WR = R * in_rad * sqrt(beta^2 - (n2^2*k^2)), \qquad \% W*R$$

BJUR = bessely(j, UR),

DBJUR = -bessel j(j+1,UR) + (j/UR)*BJUR, % derivative of bessel J at UR

% bessel J at UR

BYUR = bessely(j,UR), % bessel Y at UR DBYUR = -bessely(j+1,UR) + (j/UR)*BYUR, % derivative of bessel % Y at UR

% substitutional variables

term1 = (b1 * DBJUR) + (b2 * DBYUR) + ((b3 * BJUR)/R) + ((b4 * B

BYUR)/R),

term2 = ((b5 * BJUR)/R) + ((b6 * BYUR)/R) + (b7 * DBJUR) + (b8 * BYUR)/R) + (b8 * BYUR)/R

DBYUR),

term3 = ((g1 * BJUR)/R) + ((g2 * BYUR)/R) + (g3 * DBJUR) + (g4 * BYUR)/R) + (g4 * BYUR)/R

DBYUR),

term4 = (h1 * DBJUR) + (h2 * DBYUR) + ((h3 * BJUR)/R) + ((h4 * B

BYUR)/R),

total(2) = (term1 * term2) - (term3 * term4),

total(2) = abs(total(2)),% calculates power at each radius

sum2 = sum2 + (total(2) * R * deltaR),% running total of power values

end, % end of glass annulus loop

sum3 = 0, for N = 1 maxN3, % from R = C to 2*C $R = R_min3 + (N * deltaR)$, % calculates radius $UR = R * in_rad * sqrt((n1^2*k^2) - beta^2)$, % U*R $WR = R * in_rad * sqrt(beta^2 - (n2^2*k^2))$, % W*R

BKWR = besselk(J,WR),	% bessel K of WR	
DBKWR = -besselk(j+1,WR) + (j/WR)*BKWR,	% derivative of bessel

% K of WR

% substitutional variables

term1 = (c1 * DBKWR) + ((c2 * BKWR)/R), term2 = ((d1 * BKWR)/R) + (d2 * DBKWR), term3 = (((1 * BKWR)/R) + ((12 * DBKWR)), term4 = ((1 * DBKWR) + (((12 * BKWR)/R)), total(3) = (term1 * term2) - (term3 * term4), total(3) = abs(total(3)), % calculates power at each radius sum3 = sum3 + (total(3) * R * deltaR), % running total of power values % end of outer cladding loop

% power fractions for each section of the waveguide are calculated

end,

R1 = sum1/(sum1 + sum2 + sum3),

R2 = sum3/(sum1 + sum2 + sum3),

R3 = sum2/(sum1 + sum2 + sum3),

	fprintf(fid1,'% 0f % 0f % 4f	% 4f % 4e % 4e % 4e % 4e\n',
j,x,U,W,R1,R2,F	R3,R1+R2),	% outputs mode parameters and power fractions
	F1 = F1 + R1,	% running totals of power fractions
	F2 = F2 + R2,	
	F3 = F3 + R3,	
	<pre>mode_count = mode_count + 1,</pre>	% mode count is incremented
	$II \cup_{i} (j+1, x+1) == 0,$	% if last mode for this order has been reached
	$x = x_{limit},$	% break out of mode index loop, move to next
		% mode
	end,	
end,		%end of x loop
end,		%end of j loop
fclose(fid1),		% close output file
$FI = F1/mode_{-}$	count	% output average power fractions to screen
$F2 = F2/mode_{-}$	count	
$F3 = F3/mode_{0}$	counts	

Appendix F

This program controls the running of all the other programs needed to calculate power fractions for the hollow cylindrical waveguide probe Variables named in this program become common to all the programs called by this one

% Title Modes m

% Aim To combine all steps to calculate power fractions in a hollow cylindrical waveguide

% The following are commands to obtain information from the program user

n1 = input('Enter the refractive index of the glass '),

n2 = mput ('Enter the refractive index of the cladding '),

lambda = input('What is the excitation wavelength ?'),

 $n_rad = nput(What is the inner radius dimension ?'),$

out_rad = input('What is the outer radius dimension ? '),

% the following are examples of values that could be used by the program

%n1 = 1 46

%n2 = 1 45

%lambda = 1e-6

%in_rad = 80e-6

%out_rad = 100e-6

 $C = out_rad/m_rad$, % ratio of inner radius to outer radius

$\mathbf{k} = (2 * \mathbf{p}_1) / \mathbf{lambda},$	% defines the wavenumber, k	
Z = 377,	% the characteristis impedance of free space	
x_max = (in_rad * sqrt((n1^2*k^2) - $n2^2 * k^2$)) + 30, % max x value used in finding cut-off	
	% values	
$j_max = x_max + 35,$	% max j value used in finding cut-off values	
cuts,	% each program is called in order of use	
holl,		
holl_se,		
hollpow,		

Appendix G

In the text of Chapter 4, section 2 6, it was assumed that all modes in the hollow waveguide were reflected with the same reflection coefficient R at the end face (silica-silver interface) This is proved in this appendix for the range of incident angles within which the quasi plane waves strike the glass-silver interface

For reflection at an interface (for light whose E field is in the plane of incidence) the reflected E field amplitude E_r is related to the incidnet value by the expression (due to Fresnel)

$$\frac{E_r}{E_1} = \frac{Z_2 Cos\theta_2 - Z_1 Cos\theta_1}{Z_2 Cos\theta_2 + Z_1 Cos\theta}$$
 Eqn G-1

(see Pain, (1968))



Figure G-1 Reflection and transmission at an interface between two media. Now $Z_2 = Z_0 / n_2$ and $Z_1 = Z_0 / n_1 (Z_0 = 377 \Omega = characteristic impedance of free-space (or vacuum)$

For silica $(n_1 = 1.46)$ surrounded by a methanol solution $(n_2 = 1.329)$ the critical angle is 65.5° For a launch numerical aperture of 0.37 we can write

$$0 \quad 37 = 1Sin\theta_{incident} = 1 \quad 46Sin\theta_{max}$$

$$u e \quad \theta_{max} = 14 \quad 68^{\circ}$$
Eqn G-2

 θ_{max} is the maximum value of the incident angle within the waveguide



Figure G-2 Light ray undergoing total internal reflection

Because of total internal reflection, the maximum value of θ_1 at the distal end is by symmetry also 14 68° By Snell's Law the transmitted angle θ_2 (maximum) is given by

$$146 \sin 1468^\circ = 1329 \sin (\theta_2)$$

 $1 e \theta_{2(\text{maximum})} = 16 \ 165^{\circ}$

Using equation G-1, R is then

$$\frac{(1 \ 46)[Cos(16 \ 16)] - 1 \ 329[Cos(14 \ 68)]}{(1 \ 46)[Cos(16 \ 16)] + 1 \ 329[Cos(14 \ 68)]} = 0 \ 0434 \qquad Eqn \ G-3$$

At normal incidence $\theta_1 = \theta_2 = 0$, the amplitude reflection coefficient is

$$\frac{1}{1} \frac{46 - 1}{46 + 1} \frac{329}{329} = 0 \quad 0487 \qquad \qquad Eqn \ G-4$$

or about 12% larger than the minimum possible value (which occurs at $\theta_{1 \text{ (max)}}$ of 16 16°) shown in equation G-3 above. Over the full range of possible mode incident angles, then it is not unreasonable to assume a constant reflection coefficient R for all modes as adopted in equation 4.5.2

Appendix F References.

Pain HJ "The Physics of Vibrations and Waves" (Wiley & Sons NY, 1968) Chapter 7, p208

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$\frac{J_l(UC)}{J_l(U)}$	$\frac{Y_l(UC)}{Y_l(U)}$	0
0	0	$\frac{J_l(UC)}{J_l(U)}$
$-\beta l \left(\frac{1}{U^2} + \frac{1}{W^2}\right)$	$-\beta l \left(\frac{1}{U^2} + \frac{1}{W^2}\right)$	$\frac{p}{W}\frac{I_l'(W)}{I_l(W)} + \frac{p}{U}\frac{J_l'(U)}{J_l(U)}$
$\frac{qn_2^2}{W}\frac{I_l'(W)}{I_l(W)} + \frac{qn_1^2}{U}\frac{J_l'(U)}{J_l(U)}$	$\frac{qn_2^2}{W} \frac{I_i'(W)}{I_i(W)} + \frac{qn_1^2}{U} \frac{Y_i'(U)}{Y_i(U)}$	$\beta l \left(\frac{1}{U^2} + \frac{1}{W^2} \right)$
$\frac{\beta l}{U^2 C} \frac{J_l(UC)}{J_l(U)}$	$\frac{\beta l}{U^2 C} \frac{Y_l(UC)}{Y_l(U)}$	$-\frac{p}{U}\frac{J_{l}'(UC)}{J_{l}(U)}$
$-\frac{qn_1^2}{U}\frac{J_l'(UC)}{J_l(U)}$	$-\frac{qn_1^2}{U}\frac{Y_l'(UC)}{Y_l(U)}$	$-\frac{\beta l}{U^2 C} \frac{J_l(UC)}{J_l(U)}$

0	$-\frac{K_{l}(WC)}{K_{l}(W)}$	
V V(UC)		K(WC)
$\frac{I_l(UC)}{Y_l(U)}$	0	$-\frac{K_{l}(WC)}{K_{l}(W)}$
$\frac{p}{W}\frac{I_l'(W)}{I_l(W)} + \frac{p}{U}\frac{Y_l'(U)}{Y_l(U)}$	0	0
$\beta l \left(\frac{1}{U^2} + \frac{1}{W^2} \right)$	0	0
$-\frac{p}{U}\frac{Y_l(UC)}{Y_l(U)}$	$\frac{\beta l}{W^2 C} \frac{K_l(WC)}{K_l(W)}$	$-\frac{p}{W}\frac{K_l'(WC)}{K_l(W)}$
$-\frac{\beta l}{U^2 C} \frac{Y_l(UC)}{Y_l(U)}$	$-\frac{qn_2^2}{W}\frac{K_l'(WC)}{K_l(W)}$	$-\frac{\beta l}{W^2 C} \frac{K_l(WC)}{K_l(W)}$

Appendix H

1