

Energy balancing for robot aided clustered wireless sensor networks using mobility diversity algorithms

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Abstract—We consider the problem of energy balancing in a clustered wireless sensor network (WSN) deployed randomly in a large field and aided by a mobile robot (MR). The sensor nodes (SNs) are tasked to monitor a region of interest (ROI) and report their test statistics to the cluster heads (CHs), which subsequently report to the fusion center (FC) over a wireless fading channel. To maximize the lifetime of the WSN, the MR is deployed to act as an adaptive relay between a subset of the CHs and the FC. To achieve this we develop a *multiple-link* mobility diversity algorithm (MDA) executed by the MR that will allow to compensate simultaneously for the small-scale fading at the established wireless links (i.e., the MR-to-FC as well as various CH-to-MR communication links). Simulation results show that the proposed MR aided technique is able to significantly reduce the transmission power required and thus extend the operational lifetime of the WSN. We also show how the effect of small-scale fading at various wireless links is mitigated by using the proposed *multiple-link* MDA executed by a MR equipped with a single antenna.

Index Terms—Wireless sensor network, cluster, mobile robot, fading, mobility diversity

I. INTRODUCTION

Monitoring a region of interest (ROI) is one of the most important applications of wireless sensor networks (WSNs) [1], [2]. Multiple low-cost sensor nodes (SNs) are often spatially deployed over a large ROI to observe different events and estimate parameters of interest. In general, the SNs process the local observations and report back to a fusion center (FC) that optimally combines the individual reports to reach a global decision. Being geographically dispersed to cover large areas, the SNs are constrained in both bandwidth and power. To allow a *low-latency* WSN when the ROI is very large, the WSN is divided into multiple clusters to manage the large number of SNs needed to provide reliable coverage (e.g., see Fig. 2). Each cluster head (CH) receives data from each SN within the cluster, which subsequently reports to the FC where the ultimate decision is taken.

The framework of centralized decision for a single FC (i.e., single CH) network configuration has been extensively studied in [3]–[5], to name but a few references. There are some

recent publications [6], [7] (in the context of estimation) and [8] (in the context of detection) that considered the effect of *inter-sensor* collaboration on the WSN performance; after the collaboration stage, the SNs (which in general can be a subset of all SNs) report to a FC where the final decision is made. While the authors in [6] claim to reduce the FC control overhead, [7] derives the optimum power allocation scheme for a given maximum total network power budget in order to improve the estimation quality.

Now, clustered WSNs [9] has been extensively studied in various contexts such as energy management [10], [11] and fusion rules design [12]. In the context of clustering algorithms, the authors in [13] propose a *d-hop* cluster partitioning to deal with the load imbalance among CHs. In this paper, we adopt the network configuration in [5], i.e., we consider equal-sized clusters of SNs (e.g., [14]) that report their information on a regular basis to a FC. For the CHs that are too far from the FC and when communication link between the CH and the FC is poor, a mobile robot (MR) is deployed to act as an adaptive relay. We propose a *multiple-link* MDA to extend the operational lifetime of the WSN and to deal with the imbalanced load among the CHs. The transmission (CHs-to-MR as well as MR-to-FC) links are assumed to experience shadowing and multipath fading. We show that the proposed MDA effectively deals with the energy imbalance in a cluster WSN.

II. SYSTEM MODEL

Consider the problem of monitoring a large ROI by a WSN consisting of a FC, M spatially distributed SNs that are networked in N equal-sized clusters and a MR; all equipped with a single antenna. The wireless channels CHs-to-MR and the channel MR-to-FC are assumed to experience shadowing as well as small scale fading¹. Since most of the WSNs are bandwidth constrained, we assume narrow band communications so that the communication channels can be modeled as non-frequency selective. The case where the

¹The small scale fading is assumed to be time-variant over a coherence time τ .

spatially distributed CHs report to the FC via a dedicated parallel access channel (PAC) is investigated in e.g., [11], [12]. Here, we propose to deploy a single MR to act as an adaptive relay for forwarding the test statistics from the CHs to the FC. So, at the MR (positioned at point $\mathbf{p}(t)$), the test statistic received from the j th CH at time t is:

$$\hat{T}_j(t) = \left(\frac{s(\mathbf{p}(t), \mathbf{q}_j)h(\mathbf{p}(t), \mathbf{q}_j, t)}{\|\mathbf{p}(t) - \mathbf{q}_j\|_2^{\alpha/2}} \right) T_j(t) + n(t) \quad (1)$$

where \mathbf{q}_j is the position of the j th CH, $s(\mathbf{p}(t), \mathbf{q}_j)$ represents the shadowing which is modeled by a lognormal random variable whose normalized spatial correlation function is exponential, $T_j(t)$ is the test statistic transmitted from the j th CH, and $n(t) \sim \mathcal{N}(0, \sigma_i^2)$; $h(\mathbf{p}, \mathbf{q}_j, t)$ represents the small scale fading assumed to follow Jakes' model, i.e., its normalized spatial correlation is:

$$\rho(\mathbf{p}, \mathbf{q}) = \mathbb{E} [h(\mathbf{o}, \mathbf{p}, t)h^*(\mathbf{o}, \mathbf{q}, t)] = J_0 \left(\frac{2\pi\|\mathbf{p} - \mathbf{q}\|}{\lambda} \right) \quad (2)$$

where $\mathbf{o}, \mathbf{p}, \mathbf{q} \in \mathbb{R}^2$ are arbitrary points in the space; $J_0(\cdot)$ is the Bessel function of first order and zeroth degree while λ is the wavelength of the carrier used for the transmission. In addition, the small-scale fading is assumed to remain constant over the coherence time τ (i.e., $h(\mathbf{p}, \mathbf{q}_j, t) = h_k(\mathbf{p}, \mathbf{q}_j)$ and for $t \in [k\tau, (k+1)\tau)$), $h_{k_1}(\mathbf{p}, \mathbf{q}_j)$ and $h_{k_2}(\mathbf{p}, \mathbf{q}_j)$ are assumed statistically independent for $k_1 \neq k_2$.

Now, without loss of generality, we assume that the distance between the CHs is significantly larger compared to λ and so, for $j \neq i$, $h_k(\mathbf{p}, \mathbf{q}_j)$ and $h_k(\mathbf{p}, \mathbf{q}_i)$ are considered to be statistically independent.

Finally, to satisfy a certain average reference power P_{ref} at the receiver, the CHs and the MR use transmit power control mechanism. At the j th CH, the average transmit power is:

$$P_j = \frac{\|\mathbf{p}(t) - \mathbf{q}_j\|_2^\alpha P_{ref}}{s_j^2(\mathbf{p}, \mathbf{q}_j) |h_k(\mathbf{p}, \mathbf{q}_j)|^2} \quad (3)$$

where $t \in [k\tau, (k+1)\tau)$, and α is the path loss coefficient.

III. PROPOSED SOLUTION

To extend the operational lifetime of the WSN and to deal with the imbalanced load among the CHs, we propose a *double-link* mobility diversity algorithm and derive a MR path planner that we describe next in Section III-B. We assume that the FC, at time instant $t = k\tau$, has full knowledge of the channel gains ($h(\mathbf{p}, \mathbf{q}_j), \forall j = 1, 2, \dots, N$) from CHs to FC. Based on this information, the FC determines the L CHs with the lowest (CH-to-FC) channel gain and forward the corresponding CHs' identities to the MR. Then, the MR will act as a decode and forward relay that will establish communication links only with these L (FC selected) CHs. Nevertheless, due to small-scale fading, these communication links may significantly reduce the communication quality and eventually a larger amount of CH transmit power may be required to satisfy the average transmit power (3). Since the CH nodes are battery operated, optimizing their transmit power

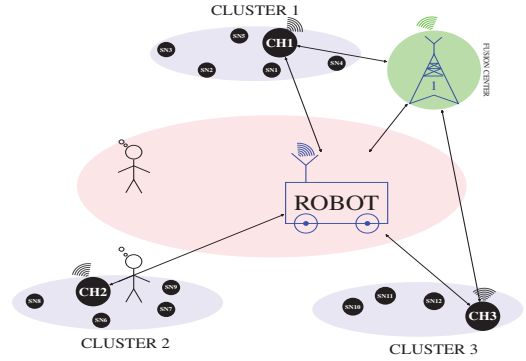


Fig. 1. Schematic communication architecture among peripheral CHs, MR, and FC. The i th CH generates a test statistic (T_i) by combining the observations received from the SNs within the cluster. The CH can communicate with the FC directly or via the MR.

is of a particular importance in extending the WSN operational lifetime.

We note that after a duration τ , the wireless communication links may change due to the temporal dependence of the small-scale fading term. As a result, the FC is required to continuously estimate the set of L CHs with the lowest CHs-to-FC channel gains and forward the corresponding CHs identities to the MR.

Before introducing the proposed algorithm, we next briefly describe few existing fading compensation techniques.

A. Related Compensating Techniques

Few existing techniques deal with the fading channel compensation in practice. A widely used technique is the multi-antenna diversity. In the context of WSNs, clearly this technique requires multiple antennae transceivers mounted on each of the CH node. This not only increases the cost of the node but also in many practical scenarios, may not be even feasible due to both the node size limitations and transmit power constraint. Hence, single antenna transceivers are desired in practice.

Now, since the small-scale fading term is time varying, temporal diversity technique could be used. Because this work consider the scenarios where the coherence time is significantly greater than the symbol duration, the temporal diversity is not suitable as would introduce a large delay. In this case, the MDAs [15] are suitable techniques to compensate the small-scale fading in WSNs.

B. Multiple Link Mobility Diversity Algorithm

MDAs are a new type of diversity technique that exploit the spatial variations of the small-scale fading and the mobility of the MRs. Their operation is divided in two phases [15], [16]: (i) exploration phase; and (ii) selection phase. During the exploration phase, the MR explores a series of K stopping points located in its vicinity (from where it estimates the

channel gain) that are optimized based on a path planner. After the exploration phase, the MR uses a selection rule to decide on the optimum position for establishing a communication link.

The existing MDAs (e.g., [15]) are only applicable to the compensation of a single small-scale fading channel. Here, we require a simultaneously small-scale fading compensation technique of $L + 1$ communication links (i.e., the L MR-to-CHs as well as the MR-to-FC links). Therefore, in this paper we extend our previous work in [15], [16] and develop a *multiple-link* MDA. To the best of our knowledge, this is the first time that a *multiple-link* MDA is proposed. Next, we develop the path planner and the MDA selection rule.

1) *MDA development*: In the context of MDAs (as previously stated), the set of distances among all stopping points are small (see [15]). As a result, the shadowing term of the j th CH-to-MR link at time t is assumed to be constant, i.e., $s(\mathbf{p}(t), \mathbf{q}_j) \approx s_j, \forall j = 1, 2, \dots, N$. In addition, the distance traveled by the MR is significantly small compared to the MR-to-FC and MR-to-CH distances and so $\|\mathbf{p}(t) - \mathbf{q}_j\|_2 \approx d_j$ for all t .

In this paper, we develop a trajectory planner with memory that uses channel gain measurements both at the current and previous states in order to estimate the MR's next position. This trajectory planner requires small-scale fading predictors such as the one used in [15]. Here, for simplicity, we choose the first order predictor². However, using the results presented here, the development of MDA with higher memory order predictors can be easily established.

The small-scale fading predictor at time instant t_{n+1} given the estimate $\hat{h}(\mathbf{p}(t_n), \mathbf{q}_j)$ is [15]:

$$\begin{aligned} \tilde{h}(\mathbf{p}(t_{n+1}), \mathbf{q}_j) &= \rho(\mathbf{p}(t_{n+1}), \mathbf{p}(t_n)) \hat{h}(\mathbf{p}(t_n), \mathbf{q}_j) \\ &+ \left(\sqrt{1 - \rho^2(\mathbf{p}(t_{n+1}), \mathbf{p}(t_n))} \right) u_{j,n} \end{aligned} \quad (4)$$

where $t_{n+1} - t_n \ll \tau$; $\hat{h}(\mathbf{p}(t_n), \mathbf{q}_j)$ is the estimate of $h(\mathbf{p}(t_n), \mathbf{q}_j)$, $\rho(\cdot)$ as in (2), and $u_{j,n}$ is a set of Normal independent and identically distributed random variables for $0 \leq j \leq L, 1 \leq n \leq K$.

To develop the path planner for the *multiple-link* MDA, the MR position at time instant t_{n+1} (i.e., $\mathbf{p}(t_{n+1})$), is chosen such that the minimum channel gain is then maximized over $L + 1$ links. So, our optimisation problem is:

$$\begin{aligned} &\text{maximize}_{\ell_n \in [\ell_d, \ell_u]} G_1(\mathbf{p}(t_{n+1})) \\ &\text{s.t.} \\ &\mathbf{p}(t_{n+1}) = \mathbf{p}(t_n) + \ell_n [\cos(\phi_n) \quad \sin(\phi_n)]^T \end{aligned} \quad (5)$$

where

$$G_1(\mathbf{p}(t_{n+1})) = \mathbb{E} \left[\min_{j=0,1,\dots,L} \left\{ \frac{s_j |\tilde{h}(\mathbf{p}(t_{n+1}), \mathbf{q}_j)|}{d_j^{\alpha/2}} \right\} \right] \quad (6)$$

²This predictor considers only the measurements of the channel at the current MR's position to predict the small-scale fading term at the next position.

and ℓ_n is the distance traveled by the MR between the current and the next position, ϕ_n represents the MR movement direction, and finally $\mathbb{E}[\cdot]$ denotes the expected value with respect to the random variables set $\{u_{j,n}\}_{\forall j}$. Note that in (6), \mathbf{q}_j , for $j = 0$ denotes the FC location. Since the predictor (4) is Gaussian distributed, it can be easily shown that:

$$G_1(\mathbf{p}(t_{n+1})) = \int_0^\infty \prod_{j=0}^L Q_1 \left(\frac{\nu_j}{\sigma_j}, \frac{x}{\sigma_j} \right) dx \quad (7)$$

where $Q_1(\cdot, \cdot)$ is the Marcum Q function with

$$\sigma_j = \frac{s_j \sqrt{1 - \rho^2(\mathbf{p}(t_{n+1}), \mathbf{p}(t_n))}}{d_j^{\alpha/2}} \quad (8)$$

$$\nu_j = \frac{s_j \rho(\mathbf{p}(t_{n+1}), \mathbf{p}(t_n)) |\hat{h}(\mathbf{p}(t_n), \mathbf{q}_j)|}{d_j^{\alpha/2}}. \quad (9)$$

Solving the optimisation problem (7) is computationally expensive in general³. However, we note that each multiplicative term in (7) is a monotonically decreasing function that tends to zero. Then, there exists a value X_0 such that:

$$\int_0^\infty \prod_{j=0}^L Q_1 \left(\frac{\nu_j}{\sigma_j}, \frac{x}{\sigma_j} \right) dx \approx \int_0^{X_0} \prod_{j=0}^L Q_1 \left(\frac{\nu_j}{\sigma_j}, \frac{x}{\sigma_j} \right) dx \quad (10)$$

Using Chebyshev's inequality:

$$\begin{aligned} \frac{\int_0^{X_0} \prod_{j=0}^L Q_1 \left(\frac{\nu_j}{\sigma_j}, \frac{x}{\sigma_j} \right) dx}{X_0} &\geq \frac{\prod_{j=0}^L \int_0^{X_0} Q_1 \left(\frac{\nu_j}{\sigma_j}, \frac{x}{\sigma_j} \right) dx}{X_0^L} \\ &= \frac{1}{X_0^L} \prod_{j=0}^L \left\{ \sigma_j \sqrt{\frac{\pi}{2}} L_{1/2} \left(\frac{-\nu_j^2}{2\sigma_j^2} \right) \right\} = \frac{1}{X_0^L} G_2(\mathbf{p}(t_{n+1})) \end{aligned} \quad (11)$$

where $L_{1/2}(\cdot)$ is Laguerre's polynomial of degree 1/2. Now, our optimisation problem becomes:

$$\begin{aligned} &\text{maximize}_{\ell_n \in [\ell_d, \ell_u]} G_2(\mathbf{p}(t_{n+1})) \\ &\text{s.t.} \\ &\mathbf{p}(t_{n+1}) = \mathbf{p}(t_n) + \ell_n [\cos(\phi_n) \quad \sin(\phi_n)]^T \end{aligned} \quad (12)$$

where $G_2(\mathbf{p}(t_{n+1}))$ is defined in (11), ℓ_n is defined over the interval $[\ell_d, \ell_u]$ and determines the correlation between the small-scale fading terms (see (2)). We would like to make it clear that ℓ_d is the design parameter and $0 < \ell_d < \ell_u$. Here, ℓ_u is the smallest distance ℓ such that $J_0(2\pi\ell/\lambda) = 0$ (i.e., the smallest distance such that the small-scale fading terms in (4) are independent). Defining ℓ_n as above yields a correlation factor $\rho(\mathbf{p}(t_{n+1}), \mathbf{p}(t_n))$ defined over the interval $[0, 1)$.

From the extensive numerical results, we have observed that the optimisation problem (12) yields an optimum value ℓ_n^o equal to ℓ_d or ℓ_u with a probability approaching 1 (i.e., with a very high probability). Hence, to further simplify the optimization process and to reduce the MR processing burden, (12) is solved only for $\ell_n \in \{\ell_d, \ell_u\}$. It is worth noting that the optimisation is performed at time instant t_n by making use

³The integral in the cost function (7) needs to be calculated numerically since it can not be evaluated analytically. This not only incur delays into the MDA algorithm but also require more MR processing power.

of the observed communication channel measurements at MR position ($\mathbf{p}(t_n)$).

Clearly, solving the optimisation problem (12) will yield a set of optimum MR stopping points. Now, the final step is to decide on the single optimum MR position such that the overall WSN performance is improved. In this paper, we select this optimum stopping point as the one that maximizes the minimum channel gain (i.e., incorporating both the MR-to-CH and MR-to-FC links).

IV. SIMULATIONS

We evaluate numerically the performance of our proposed *multiple-link* MDA. We simulate a WSN deployed in a 120×120 ROI and M SNs divided into $N = 3$ clusters with arbitrary SN geometry, where the distances between the MR and CHs are assumed to be known. The spatial configuration is shown in Fig. 2. We let the pathloss coefficient

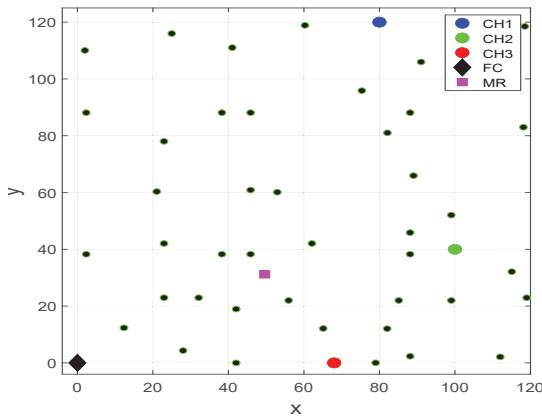


Fig. 2. Spatial configuration of the WSN where the SNs are represented with green.

$\alpha = 2$, the variance of $s_j(\mathbf{p}, \mathbf{q}_j)$ in (3) is taken such that $\text{Var}\{10 \log_{10}(s_j(\cdot))\} = 1\text{dB}$, $\forall j$ and the reference power $P_{ref} = 1\mu\text{W}$. Finally, we note that both $G_1(\mathbf{p}(t_{n+1}))$ and $G_2(\mathbf{p}(t_{n+1}))$ are independent of the MR movement direction and we take $\phi_n = \phi$, $\forall n$ in (12).

In Fig. 3, we show the mean MR and CHs transmit power (after executing the MDA) for different values of L compared to the *non-fading* case⁴. Clearly, as expected, as the number of stopping points increases, the MR probability of finding a stopping point with a larger channel gain is also increasing. As a result, the CHs transmit power decreases. However, when considering the MR (i.e., sub-figure d)), the MR transmit power increases when the number of stopping points L increases. This is expected as an increase in L will cause the MR to consider a larger number of communication links. Hence, it decreases the degree of freedom in obtaining simultaneously large channel gains for both itself and the L CHs.

⁴In the *non-fading* case, the pathloss only communication channel is considered and the MR is position at its initial point as in Fig. 2.

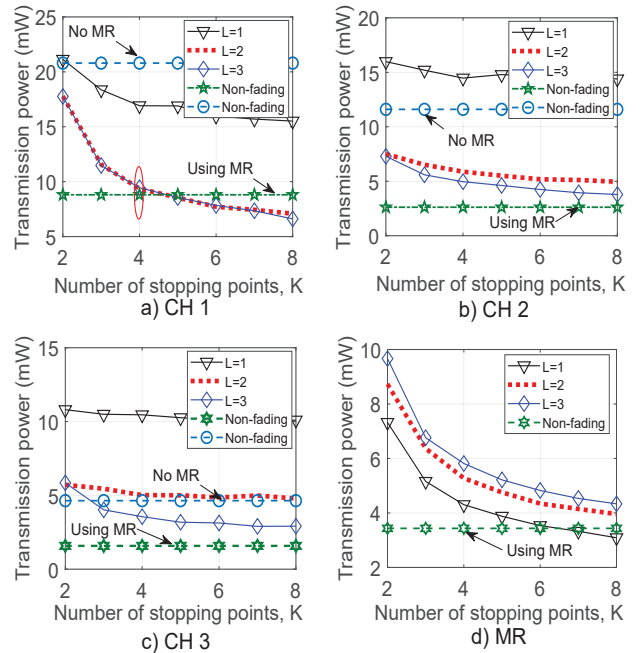


Fig. 3. Average CHs transmission power in (3) versus the number of stopping points (K), parametrized on the number of CHs that use the MR as a relay (L) with $P_{ref} = 1\mu\text{W}$, and $\alpha = 2$.

Interesting, in Fig. 3 (i.e., sub-figure a)), we can observe that when the MDA is used, the CH1 transmit power is lower than the *non-fading* case when the number of stopping points K is larger than 4 and for $L = 2$ and $L = 3$. This is due to the fact that the MDA takes advantage of the fading to improve the channel gain.

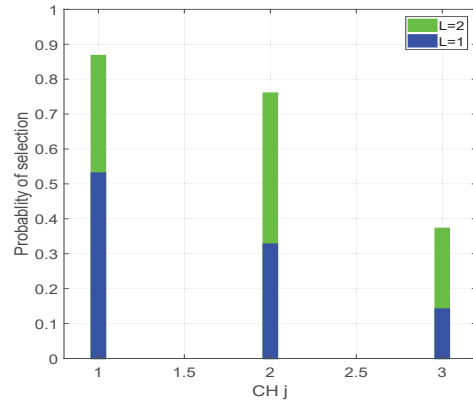


Fig. 4. CH's selection probability to use the MR as a relay versus the CH (j), parametrized on the number of CHs that use the MR as a relay (L) with $P_{ref} = 1\mu\text{W}$, and $\alpha = 2$.

Now, to further validate our results, in Fig. 4 we observe the CH selection probability (by the MR) for $L = 1$ and $L = 2$ (when $L = 3$ all CHs are selected). Clearly, there is a correlation between the selected CHs (i.e., the L CHs with the lowest channel gain) and the corresponding selection probability. For example, CH1 (experiencing the worst com-

munication channel) has a higher selection probability than CH2 and CH3. Furthermore, we can observe an increase in selection probability for all CH as L increases. This is as expected since an increase in L also increases the opportunity of a particular CH to use the MR as a relay in improving its communication link.

V. CONCLUSIONS

In this paper, we propose an efficient *multi-link* MDA to balance the CHs energy and extend their operational lifetime in random clustered WSNs. We have shown how by using an MR as a relay with the proposed MDA, the CH's mean transmit power can be significantly reduced. Finally, we have also shown that the proposed MDA results in a lower CH's transmit power compared to the non-fading communication channel case. Future work will investigate the analysis of the problem for fully distributed solution (i.e., where there is no FC).

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