



Cisneros-Magaña, Rafael and Medina, Aurelio and Anaya-Lara, Olimpo (2018) Time-domain voltage sag state estimation based on the unscented Kalman filter for power systems with nonlinear components. Energies. ISSN 1996-1073 (In Press) ,

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1 Article

2 Time-domain voltage sag state estimation based on 3 the unscented Kalman filter for power systems with 4 nonlinear components

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12 Received: date; Accepted: date; Published: date

13 **Abstract:** This paper proposes a time-domain methodology based on the unscented Kalman filter
14 to estimate voltage sags and their characteristics, such as magnitude and duration in power systems
15 represented by nonlinear models. Partial and noisy measurements from the electrical network with
16 nonlinear loads, used as data, are assumed. The characteristics of voltage sags can be calculated in
17 a discrete form with the unscented Kalman filter to estimate all the busbar voltages; being possible
18 to determine the rms voltage magnitude and the voltage sag starting and ending time, respectively.
19 Voltage sag state estimation results can be used to obtain the power quality indices for monitored
20 and unmonitored busbars in the power grid and to design adequate mitigating techniques. The
21 proposed methodology is successfully validated against the results obtained with the time-domain
22 system simulation for the power system with nonlinear components, being the normalized root
23 mean square error less than 3%.

24 **Keywords:** Nonlinear dynamic system; power quality; power system simulation; state estimation;
25 unscented Kalman filter; voltage fluctuation
26

27 1. Introduction

28 Power quality (PQ) is an important operation issue of any power system. Utilities must comply
29 with strict standards, relating primarily harmonics, transients and voltage sags [1-4]. PQ depends on
30 the power supply, the transmission and distribution systems and the electrical load condition.
31 Voltage sags are among the adverse PQ effects; they can cause malfunction of electronic loads, and
32 can reset voltage-sensitive loads [5-6]. The voltage sags characteristics in magnitude and duration are
33 necessary to determine their effect in the grid and its loads. They constitute the majority of PQ
34 problems, representing about 60% of them [7-8]. **Among the problems that the nonlinear electrical
35 components introduce to the power grid is the increase of harmonic distortion, which is an important
36 effect to mitigate.** Voltage sags have increased due to the use of nonlinear varying loads such as
37 power electronic devices, smelters, arc furnaces and electric welders, the starting of large electrical
38 loads, switching transients, connection of transformers and transmission lines, network faults,
39 lightning strikes, network switching operations, among others [9].

40 Kalman filter (KF) and the least squares method have been used to estimate the voltage
41 fluctuations in linear power systems [10-13]. PQ state estimation based on the KF uses a linear model,
42 partial and noisy measurements from the system. In [14] the number of sags is estimated using a
43 limited number of monitored busbars, recording the number of voltage sags during a determined
44 period.

45 This research work proposes as an innovation, an alternative methodology based on the
46 unscented Kalman filter (UKF) to perform the voltage sags state estimation (VSSE) in nonlinear load
47 power networks; this method can also be applied to nonlinear micro grids. The VSSE determines the
48 magnitude, duration and beginning-ending time of sags, with an observable system condition for the
49 busbars voltages using the available measurements.

50 The KF has been applied to estimate harmonics and voltage transients in a signal [15], KF gain
51 can be modified during the state estimation to reduce the estimation error [16], both references assess
52 linear cases; [17] has proposed the UKF to detect sags in a voltage waveform. In this work, the UKF
53 is extended to the nonlinear case to solve the time-domain VSSE, to estimate voltage sags in all
54 busbars of a power system including nonlinear components. The UKF makes use of a power grid
55 nonlinear model and noisy measurements from the same electrical network to estimate all the busbar
56 voltages.

57 The extended Kalman filter (EKF) can be also applied to solve the nonlinear state estimation.
58 The UKF error is slightly smaller when compared to the EKF error. This state estimation error
59 increases in the filters when sudden variations are present, both being of about the same accuracy.
60 The EKF can lead to divergence more easily than UKF, which shows good numerical stability
61 properties.

62 The state estimation receives measurements from the power network, through a wide area
63 measurement system (WAMS) and estimates the state vector, using algorithms such as the UKF.
64 Practical implementation of the time-domain state estimation can be achieved with measuring
65 instruments and data acquisition cards, capable of recording the voltage and current waveforms
66 synchronously during several cycles, e.g. using the global positioning system (GPS) to time stamp
67 the measurements [18-21]. The use of adequate communication channels like especially dedicated
68 optical fibre links, allows to the measurements be sent to the control centre with high data updating
69 rate, where they are received and numerically processed using computational systems with sufficient
70 memory and adequate capability [9].

71 Measurement technology for VSSE is currently limited, making the system underdetermined,
72 due to economic reasons. The VSSE presents different problems from those of the traditional power
73 system state estimation, where redundancy of measurements is possible [22].

74 The VSSE has been assessed in the frequency domain [14, 23]. In this work, the UKF is proposed
75 as an alternative method to obtain the time-domain VSSE. This approach makes possible the use of
76 nonlinear models to represent more accurately the power system components and to obtain the
77 results with a low state estimation error. The state estimation obtains the global or total system state
78 that can be used to take corrective actions to mitigate the adverse effects of voltage sags, such as the
79 network configuration change or control of flexible alternating current transmission system (FACTS)
80 devices, e.g. the static synchronous compensator (STATCOM).

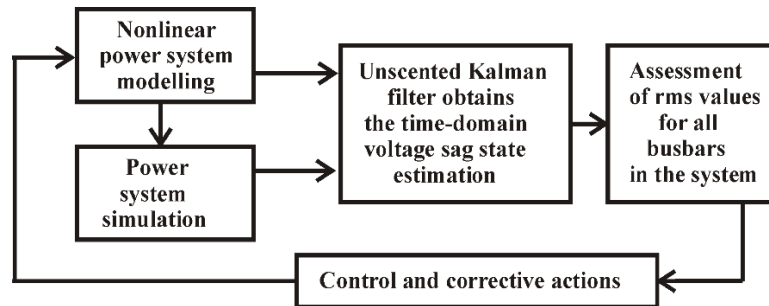
81 The time-domain UKF state estimation methodology can be used not only to estimate voltage
82 sags but also to estimate over voltages, over currents or electromagnetic transients. The main
83 objective of this work is to apply the UKF to obtain the VSSE, by addressing the dynamics of the
84 nonlinear electrical networks and by estimating and delimiting the voltage sags in the time-domain.
85 The case studies address short circuit faults and transient load conditions. The results are validated
86 against the actual time-domain response of the power grid.

87 2. Dynamic state estimation

88 The network model can be a set of first order differential equations to describe the dynamic state
89 performance. The dynamic estimation data are the grid model with its inputs and a measurement set
90 of selected outputs from the system during a determined number of cycles to define the measurement
91 equation.

92 The KF dynamically follows the variations in the states, i.e. currents and voltages, detecting
93 changes in the voltage waveform within less than half of a cycle and it is a good tool for instantaneous
94 tracking and detection of voltage sags [24-25].

95 The KF solves the dynamic estimation, due to its recursive process [26-27]; being applied in
 96 linear cases. The UKF solves the dynamic estimation in nonlinear cases. In this work, the UKF
 97 estimates the nonlinear power system state under transient conditions, e.g., voltage sags [28]. Figure
 98 1 describes the proposed VSSE methodology. The main steps are the nonlinear power system
 99 modelling and simulation, then UKF is applied to obtain the time-domain VSSE, and lastly the
 100 assessment of rms busbar voltages.



101
102

Figure 1. Time-domain UKF VSSE.

103 The UKF applies a deterministic sampling technique; i.e. the unscented transform (UT), which
 104 takes a set of sigma points near of their mean value. These points are propagated through the
 105 nonlinear model by evaluating the estimated mean and covariance [25]. The mean and covariance
 106 are encoded in the set of sigma points; these points are treated as elements of a discrete probability
 107 distribution, which has mean and covariance equal to those originally given. The distribution is
 108 propagated by applying the non-linear function to each point. The mean and the covariance of the
 109 transformed points represent the transformed estimate.

110 The main advantage of the UKF is the derivative free nonlinear state estimation, thus avoiding
 111 analytical or numerical derivatives [29-30]. The UT avoids the need of linearization using the Jacobian
 112 matrix as in the EKF, and it can be applied to any function, independently if it is differentiable or not.
 113 The UKF includes a Cholesky decomposition with an inverse matrix to evaluate the sigma points at
 114 each time step.

115 Inaccuracies of the model and its parameters can be taken into account with a statistical term w ,
 116 called noise process. It accounts for the existence of phenomena such as the thermal noise of the
 117 electrical elements and the ambiguity in the accuracy of the parameters. Metering devices have errors
 118 and noise; they are represented by a statistical term v . In most cases, w and v have a Gaussian
 119 distribution. UKF is able to operate with partial, noisy, and inaccurate measurements [31-32].

120 3. Unscented Kalman Filter Methodology

121 The UT is based on the mean and covariance propagation by a nonlinear transform. The system
 122 and measurement nonlinear models can be represented as,

$$123 \quad dx/dt = f(x, u, w) \quad (1)$$

$$124 \quad y = h(x, u, v) \quad (2)$$

125 where $x \in \mathbb{R}^{n \times 1}$ is the state vector, u the known input vector of variable order, y the variable order
 126 output vector, f a nonlinear state function and h is a nonlinear output function, with n states and m
 127 measurements.

128 UKF uses a deterministic approach for mean and covariance calculation; $2n+1$ sigma points are
 129 defined by using a square root decomposition of prior covariance. Sigma points propagation through
 130 the model (1) obtains the weighted mean and covariance. W_i represents the scalar weights, defined
 131 as,

$$132 \quad W_0^{(m)} = \lambda / (n + \lambda) \quad (3)$$

$$133 \quad W_0^{(c)} = \lambda / (n + \lambda) + (1 + \alpha^2 + \beta) \quad (4)$$

$$134 \quad W_i^{(m)} = W_i^{(c)} = 1/(2(n + \lambda)), \quad i = 1, \dots, 2n \quad (5)$$

$$135 \quad \lambda = \alpha^2(n + \kappa) - n \quad (6)$$

$$136 \quad \gamma = \sqrt{n + \lambda} \quad (7)$$

137 where λ and γ are scaling parameters, α and κ determine the spread of sigma points; β is associated
138 with the distribution of x . If Gaussian $\beta=2$ is optimal, $\alpha=10^{-3}$ and $\kappa=0$ are normal values [30].

139 UT takes the sigma points with their mean and covariance values, and transform them by
140 applying the nonlinear function f , and then the mean and covariance can be calculated for the
141 transformed points. A weight W_i is assigned to each point.

142 UKF defines the n -state discrete-time nonlinear system from (1) and (2) as,

$$143 \quad \mathbf{x}_{k+1} = \mathbf{f}(\mathbf{x}_k, \mathbf{u}_k, \mathbf{w}_k, \mathbf{t}_k) \quad (8)$$

$$144 \quad \mathbf{y}_k = \mathbf{h}(\mathbf{x}_k, \mathbf{u}_k, \mathbf{v}_k, \mathbf{t}_k) \quad (9)$$

$$145 \quad \mathbf{w}_k \sim N(0, \mathbf{Q}_k) \quad (10)$$

$$146 \quad \mathbf{v}_k \sim N(0, \mathbf{R}_k) \quad (11)$$

147 Process noise \mathbf{w} and measurement noise \mathbf{v} are assumed stationary, zero-averaged and
148 uncorrelated, $\mathbf{Q} \in \mathbb{R}^{n \times n}$ and $\mathbf{R} \in \mathbb{R}^{m \times m}$ are the covariance matrices for noises \mathbf{w} and \mathbf{v} , respectively.

149 UKF applies the following steps:

150 a) Initialization, $k=0$.

$$151 \quad \hat{\mathbf{x}}_0^+ = \mathbf{E}(x_0) \quad (12)$$

$$152 \quad \mathbf{P}_0^+ = \mathbf{E}[(x_0 - \hat{\mathbf{x}}_0^+)(x_0 - \hat{\mathbf{x}}_0^+)^T] \quad (13)$$

153 \mathbf{E} is the expected value, \mathbf{P} is the error covariance matrix, + indicates update estimate or a
154 posteriori estimate and - project estimate or a priori estimate. Subscripts k and $k-1$ denote time
155 instants $t=k\Delta t$ and $t=(k-1)\Delta t$, respectively, Δt is the time step.

156 b) Sigma points assessment in matrix form by columns:

$$157 \quad \mathbf{X}_{k-1} = [\hat{\mathbf{x}}_{k-1} \quad \hat{\mathbf{x}}_{k-1} + \gamma\sqrt{\mathbf{P}_{k-1}} \quad \hat{\mathbf{x}}_{k-1} - \gamma\sqrt{\mathbf{P}_{k-1}}] \quad (14)$$

158 c) Update time step k from $k-1$.

$$159 \quad \mathbf{X}_{k|k-1}^* = \mathbf{f}[\mathbf{X}_{k-1}, \mathbf{u}_{k-1}] \quad (15)$$

$$160 \quad \hat{\mathbf{x}}_k^- = \sum_{i=0}^{2n} W_i^{(m)} \mathbf{X}_{i,k|k-1}^* \quad (16)$$

$$161 \quad \mathbf{P}_k^- = \sum_{i=0}^{2n} W_i^{(c)} [\mathbf{X}_{i,k|k-1}^* - \hat{\mathbf{x}}_k^-] [\mathbf{X}_{i,k|k-1}^* - \hat{\mathbf{x}}_k^-]^T + \mathbf{Q}_k \quad (17)$$

$$162 \quad \mathbf{X}_{k|k-1}^- = [\hat{\mathbf{x}}_k^- \quad \hat{\mathbf{x}}_k^- + \gamma\sqrt{\mathbf{P}_k^-} \quad \hat{\mathbf{x}}_k^- - \gamma\sqrt{\mathbf{P}_k^-}] \quad (18)$$

$$163 \quad \mathbf{Y}_{k|k-1}^* = \mathbf{h}[\mathbf{X}_{k|k-1}^-] \quad (19)$$

$$164 \quad \hat{\mathbf{y}}_k^- = \sum_{i=0}^{2n} W_i^{(m)} \mathbf{Y}_{i,k|k-1}^* \quad (20)$$

165 χ matrix represents the sigma points; χ^* matrix represents the updated sigma points and \mathbf{y}^*
 166 the updated output vector with sigma points.

167 d) Evaluate the error covariance matrices as,

$$168 \quad \mathbf{P}_{\hat{\mathbf{y}}_k \hat{\mathbf{y}}_k} = \sum_{i=0}^{2n} W_i^{(c)} \left[\mathbf{y}_{i,k|k-1}^* - \hat{\mathbf{y}}_k^- \right] \left[\mathbf{y}_{i,k|k-1}^* - \hat{\mathbf{y}}_k^- \right]^T + \mathbf{R}_k \quad (21)$$

$$169 \quad \mathbf{P}_{\mathbf{x}_k \mathbf{y}_k} = \sum_{i=0}^{2n} W_i^{(c)} \left[\chi_{i,k|k-1}^* - \hat{\mathbf{x}}_k^- \right] \left[\mathbf{y}_{i,k|k-1}^* - \hat{\mathbf{y}}_k^- \right]^T \quad (22)$$

170 e) UKF algorithm evaluates the filter gain \mathbf{K}_k and updates the estimated state and the error
 171 covariance matrix.

$$172 \quad \mathbf{K}_k = \mathbf{P}_{\mathbf{x}_k \mathbf{y}_k} \mathbf{P}_{\hat{\mathbf{y}}_k \hat{\mathbf{y}}_k}^{-1} \quad (23)$$

$$173 \quad \hat{\mathbf{x}}_k^+ = \hat{\mathbf{x}}_k^- + \mathbf{K}_k (\mathbf{y}_k - \hat{\mathbf{y}}_k^-) \quad (24)$$

$$174 \quad \mathbf{P}_k^+ = \mathbf{P}_k^- + \mathbf{K}_k \mathbf{P}_{\hat{\mathbf{y}}_k \hat{\mathbf{y}}_k} \mathbf{K}_k^T \quad (25)$$

175 The steps (b-d), equations (14)-(22), define the prediction stage, and the last step (e), equations
 176 (23)-(25), defines the update stage, as in the KF algorithm [33-34]. The main objective of this work is
 177 to use the UKF formulation to estimate the busbar voltage waveforms, mainly at unmonitored
 178 busbars in the presence of voltage sags generated by faults and load transients.

179 Waveforms can be contaminated with noise, and the assumption of constant values for \mathbf{Q} and \mathbf{R}
 180 is valid when the noise characteristics are constant, like its standard deviation and variance. If the
 181 noise is varying, \mathbf{Q} and \mathbf{R} should be computed at each time step and an adaptive KF is a requirement
 182 [16]. UKF algorithm tracks the time-varying model and noise through the on-line calculation of \mathbf{Q}
 183 and \mathbf{R} . In this work, \mathbf{Q} and \mathbf{R} matrices are assumed constant, in order to mainly analyse the UKF
 184 application to time-domain VSSE.

185 UKF identifies the interval where the sags are present, as well as their magnitude, with an
 186 acceptable precision. By increasing the number of cycles, the UKF can identify the voltage
 187 characteristics during fault transient periods.

188 The number of points per cycle is of important concern to evaluate the time-domain state
 189 estimation with periodic signals. This number defines the sampling rate for the monitored signals.
 190 The sampled waveform is a sequence of values taken at defined time intervals and represents the
 191 measured variable. Interpolation can be used to adjust the number of points per cycle, linearly or
 192 nonlinearly [35]. In addition, the interpolation should be used carefully with discrete signals to satisfy
 193 the sampling theorem. The sampling rate defines the speed at which the input channels are sampled;
 194 this rate is defined in samples per cycle. To detect transients, high sampling rates compared with the
 195 fundamental frequency may be necessary [36].

196 3.1 Rms value of discrete waveforms and normalized root mean square error.

197 The rms voltage magnitude can be determined by processing the discrete values for the voltage
 198 waveform according to the used data window size and the sampling frequency. The rms voltage
 199 magnitude V_{rms} for a discrete voltage signal can be calculated as,

$$200 \quad V_{rms}(iN) = \sqrt{\left(\frac{1}{N} \sum_{j=(i-1)N+1}^{iN} V_j^2 \right)} \quad i \geq 1 \quad (26)$$

201 where V_j is the sample voltage j and N is the number of samples per cycle taken in the sampling
 202 window; i is the sampled cycle. This expression can be applied to discrete voltage and current
 203 waveforms [22].

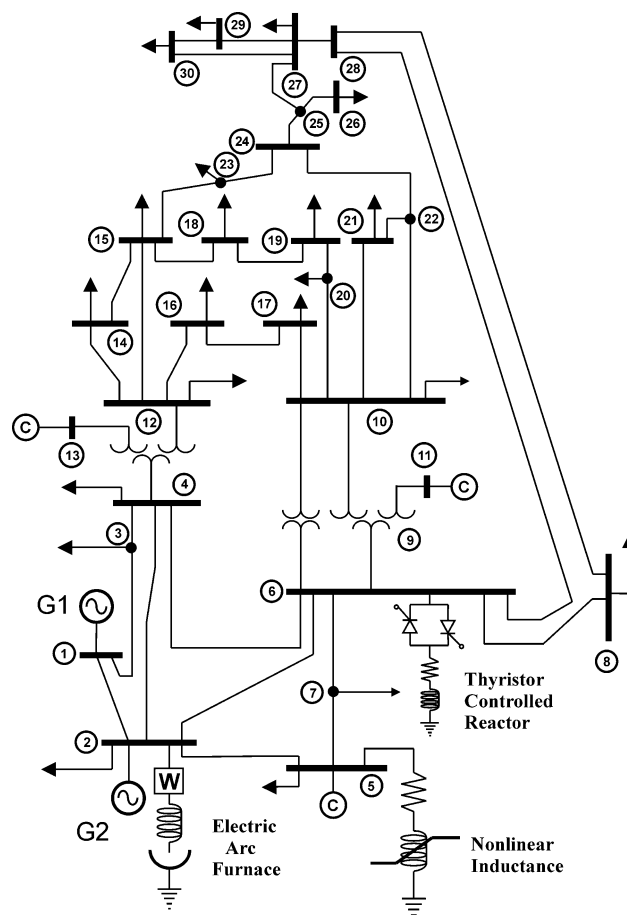
204 Normalized root mean square error (NRMSE) is used to validate the UKF-VSSE methodology;
 205 this error evaluates the state estimation residual between actually observed values and the estimated
 206 values; lower residual indicates less state estimation error. NRMSE is defined as,

$$207 \quad NRMSE = \sqrt{\frac{\sum_{t=1}^{np} (\hat{y}_t - y_t)^2}{np}} / (y_{max} - y_{min}) \quad (27)$$

208 \hat{y} is the estimated vector, y is the real or actually observed vector and np the number of elements
 209 of these vectors.

210 4. Case Studies

211 Figure 2 shows the modified IEEE 30 bus test system used in the case studies described next,
 212 assuming a three-phase base power of 100 MVA and a phase-to-phase base voltage of 230 kV. Lines
 213 1-2, 1-4, 2-4, 2-5, 2-6, 4-6 and 5-6 are represented by an equivalent pi model and by series impedance
 214 the rest of lines; transformers 6-10, 4-12-13, 6-10-11, are represented by an inductive reactance,
 215 according to the IEEE 30-bus test power system [37].



216

217 **Figure 2.** Modified IEEE 30-bus test power system with nonlinear loads at busbars 2, 5, and 6.

218 The system is modified adding three nonlinear electrical loads, i.e. an electric arc furnace (EAF)
 219 to busbar 2, a nonlinear inductance to busbar 5 and a thyristor-controlled reactor (TCR) to busbar 6.
 220 The addition of these nonlinear elements gives the nonlinearity of (1) and (2). Appendix A gives
 221 additional parameters of nonlinear loads. Appendix B presents the nonlinear load models and their
 222 differential equations.

223 Generators are modelled as voltage sources connected to busbars through a series inductance.
 224 Linear electric loads are represented as constant impedances. Busbar voltages, line and load currents
 225 are defined as state variables to obtain the state space model for the power network; the
 226 measurements are function of these state variables.

227 The measurement locations are selected so that the busbar voltages are observable. Tables 1 and
 228 2 show x and z vectors, respectively, to form the measurement equation by obtaining 103
 229 measurements to estimate 110 state variables ($n=110$, $m=103$). The observation equation with this set
 230 of measurements has an underdetermined condition, but all the busbar voltages are observable to
 231 estimate the voltage sags. When busbar voltages are assessed and estimated other variables can be
 232 calculated, i.e. line currents or the TCR current.

233

Table 1. State variable vector x

Description	State variable
Line currents	1-41
Busbar voltages	42-71
Generator currents	72-77
Busbar load currents	78-106
Nonlinear inductor magnetic flux	107
EAF current and arc radius	108-109
TCR current	110

234

Table 2. Measurements vector z

Description	Output variable
Line currents	1-38
Busbar voltages	42-68
Generator currents	72-77
Busbar load currents	78-106
Nonlinear inductor current	107
EAF real power	108-109

235 The EAF real power and the nonlinear inductance current are included as nonlinear functions
 236 in the measurement equation ($z=Hx$) represented in the formulation by (2).

237 In the measurement matrix $H \in \mathbb{R}^{m \times n}$, each measurement is associated with its corresponding
 238 state variable (Table 2). The sampling frequency is at least 30.72 kHz, to obtain 512 samples per cycle,
 239 for a fundamental frequency of 60 Hz [24].

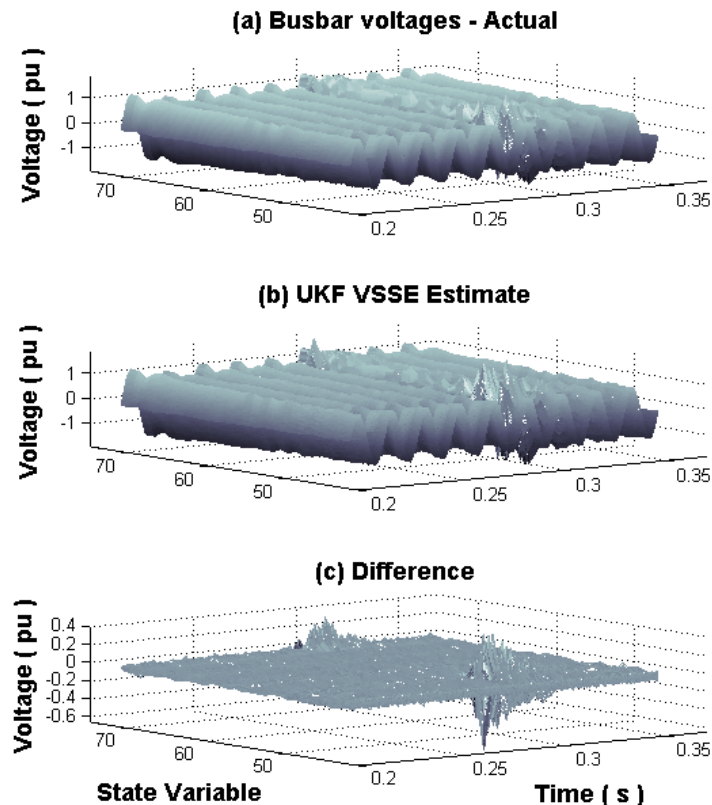
240 The conventional trapezoidal rule is used to solve the 110 first order ordinary differential
 241 equations set. To represent the power system, busbar voltages, line and load currents are defined as
 242 state variables; a step size of 512 points per period is used, i.e., 32.5 microseconds. The simulation
 243 time is set to 0.4 seconds or 24 cycles. The measurements are taken from this simulation and then are
 244 contaminated using randomly generated noise.

245 4.1. Case study: UKF VSSE short-circuit fault at busbar 4

246 A transient condition is simulated by applying a single-phase to ground fault at busbar 4. The
 247 fault impedance is of 0.1 pu, to simulate a short-circuit fault, starting in cycle 13 (0.216 s) and ending
 248 in cycle 17 (0.283 s). This fault generates busbar voltage sags and swells, which can be estimated with
 249 the power network model, partial and noisy measurements from the system, and the UKF algorithm.
 250 The criterion to select this case study is to represent a transient fault in the transmission system and
 251 verify the proposed VSSE method.

252 Measurement noise is assumed with a signal to noise ratio (SNR) of 0.025 pu or 2.5%; while a
 253 SNR of 0.001 pu or 0.1% is assumed for the noise process. Figure 3 shows the busbar voltages 1-30,
 254 where the actual, the proposed UKF estimate and the difference between instantaneous values during
 255 the fault at busbar 4 are shown, corresponding to state variables 42-71.

256 The largest estimation error is present when the fault condition is removed at 0.283 s; this error
 257 is due to sudden changes in the busbar voltages. It is approximately 7%, but quickly decreases in the
 258 next three cycles to 1%. These voltage fluctuations are due to the short-circuit transient condition at
 259 busbar 4.



260

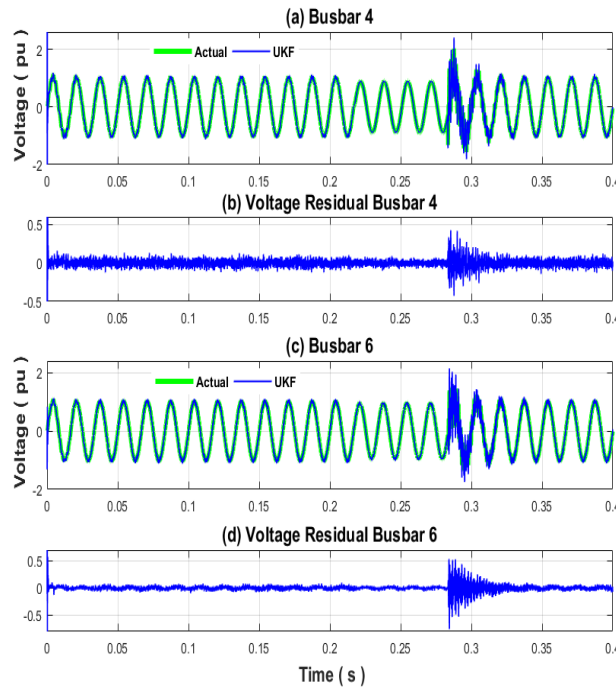
261 **Figure 3.** Busbar voltages (a) Actual, (b) UKF VSSE, (c) Difference, short-circuit at busbar 4 from 0.216
 262 to 0.283 s.

263 Voltage waveforms for the faulted busbar 4 and for busbar 6, near to fault, are shown in Figure
 264 4. Actual, UKF estimation and residual waveforms are illustrated. The presence of a voltage sag/swell
 265 condition at these busbars can be observed. Voltage sag lasts 4 cycles, while the fault condition is
 266 present, originating a reduction in the voltage magnitude of 12% for busbar 4 and 8% for busbar 6.
 267 Post-fault period begins at cycle 18, when the short circuit fault is removed. A voltage swell
 268 condition is present with a duration of two cycles and then the voltage eventually reaches the steady state.
 269 Residuals take considerable values during the voltage swell condition, the first two cycles of the post-
 270 fault period, and are due to the fast fluctuations of the state variables.

271 In Figure 4, NRMSE has been calculated using (27) to evaluate the state estimation error between
 272 actual and UKF estimated waveforms for the voltage busbars 4, and 6, during the 24 cycles under
 273 analysis, resulting on 2.5% and 1.2%, respectively.

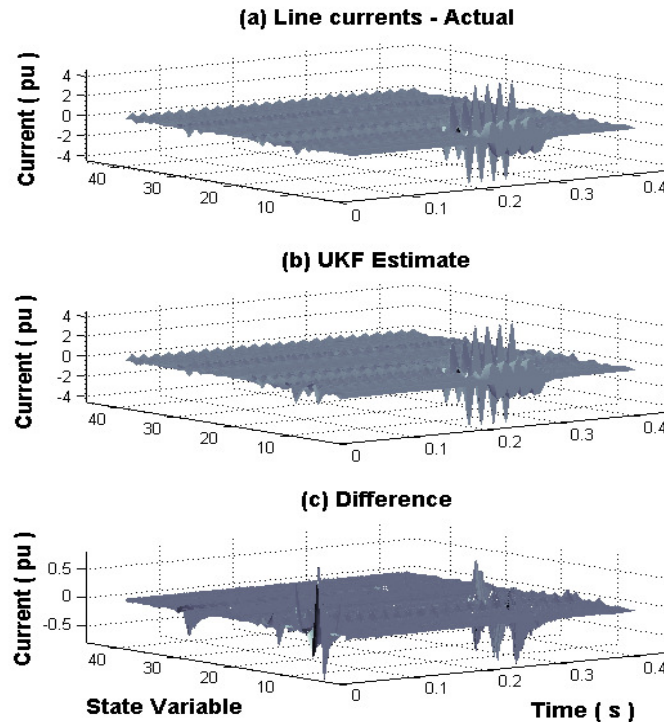
274 Busbars 3-30 show a similar behavior as for busbars 4 and 6 during and after the short-circuit
 275 fault. The busbar voltage magnitude reduction mainly depends on the network topology, the load
 276 condition and the line impedance between the busbars. Fluctuations in the voltage waveforms at

277 busbars are due to noisy measurements, network modelling, and the short-circuit fault used for the
 278 voltage sag/swell transient state estimation.



279
 280 **Figure 4.** Actual, UKF estimation, and residuals of busbars 4 and 6, voltage sag 14-17 cycles from 0.216
 281 to 0.283 s, voltage swell 18-19 cycles, short-circuit at busbar 4.

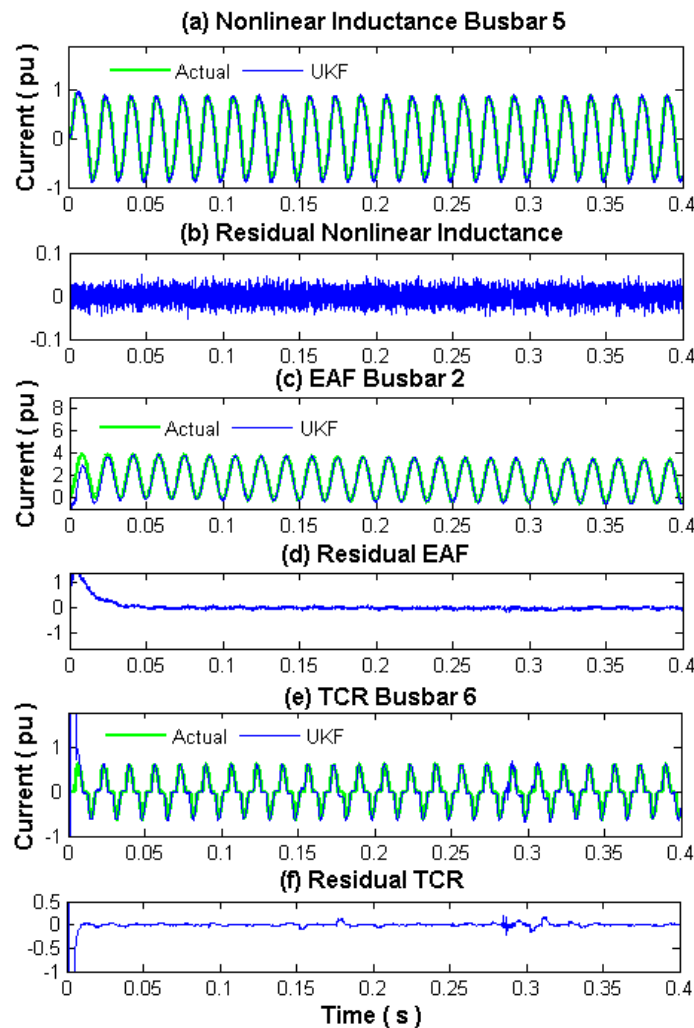
282 Line currents are shown in Figure 5 for the actual, UKF estimate and difference, respectively;
 283 with the fault condition at busbar 4 from 0.216 to 0.283 s.



284
 285 **Figure 5.** Line currents (a) Actual, (b) UKF estimation, (c) Difference, short-circuit at busbar 4 from
 286 0.216 to 0.283 s.

287 The distribution of line currents in the power system is shown for the interval of study. This
 288 distribution represents the fault currents from generators to the faulted busbar 4, which can be
 289 observed in Figure 5 by the current fluctuations in the first state variables during and after the fault
 290 period. During the first cycle after fault clearance, the error increases to 12%, but once, this cycle ends
 291 the error decreases to around 1% in the post-fault period. The difference graph (c) presents this error
 292 at 0.283 seconds for the state variables representing the currents from generators to the faulted busbar
 293 4.

294 Actual, UKF estimated currents and residuals of nonlinear components are illustrated in Figure
 295 6, for the nonlinear inductance (a, b), the EAF (c, d) and the TCR (e, f).



296

297 **Figure 6.** Nonlinear load currents, actual, UKF, and residuals, short-circuit at busbar 4 from 0.216 to
 298 0.283 s.

299 These state variables show small variations for the considered fault condition. Only in the post-
 300 fault period, TCR current differs by approximately 2.5%, but this difference decreases quickly after
 301 one cycle to negligible proportions, i.e., approximately to 1%. This error is due to the fast changes in
 302 the state variables which make the numerical process of state estimation difficult.

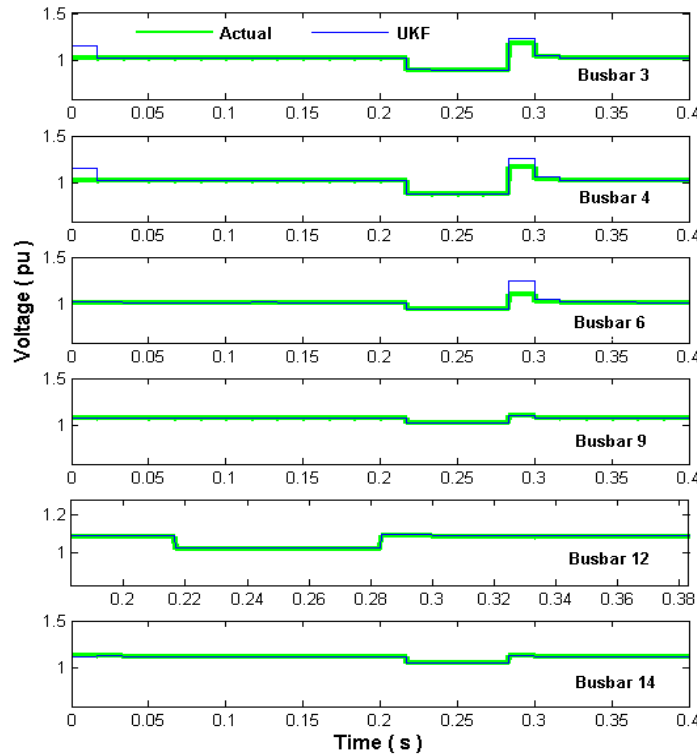
303 The NRMSE between actual and UKF estimated waveforms for nonlinear load currents in Figure
 304 6 gives 0.8% for the nonlinear inductance, 1.35% for the EAF, and 2.16% for the TCR.

305 4.2. RMS busbar voltages under the short-circuit fault at busbar 4

306 The voltage sags can be detected directly from the instantaneous or rms values of the nodal
 307 voltage waveforms, which are defined as state variables, by comparing the voltage values in the time
 308 interval under analysis. If these values vary, a voltage fluctuation (sag or swell) occurs.

309 Figure 7 shows the rms voltage magnitude for the faulted busbar 4 and for busbars 3, 6, 9, 12
 310 and 14; these busbars are near to busbar 4 and present the largest voltage sags.

311 The rms magnitude of these voltages is computed using (26), the initial step when the voltage
 312 sag begins is due to the short-circuit fault; this time is at cycle 13 or 0.216 seconds. During the first
 313 cycle of post-fault period (cycle 18 or 0.283 seconds), a noticeable difference is present in the rms
 314 voltage of the nearby busbars. The largest difference is 20% for busbar 6, but this error is reduced
 315 drastically in the next cycle to 4.5%, being of negligible proportions during the following cycles
 316 (approximately 1%). This effect is due to sudden variations in the state variables during and after the
 317 fault is removed, which are difficult to follow exactly with the UKF algorithm.



318 **Figure 7.** Actual, UKF VSSE, rms voltage magnitude for faulted busbar 4 and busbars 3, 6, 9, 12,
 319 and 14. Sags of different magnitude are present from 0.216 to 0.283 s. Swells are present at first post-fault
 320 cycle after 0.283 s.
 321

322 Table 3 shows the actual and estimated voltage sags at the network busbars, referred to the pre-
 323 fault magnitudes, due to the single-phase to ground fault at busbar 4. These values are computed
 324 again using (26); not listed busbars have a voltage variation of less than 0.01 pu during the fault. The
 325 magnitude of the estimated voltage sags closely matches the actual values, thus validating the
 326 proposed UKF VSSE methodology.

Table 3. Actual and UKF VSSE voltage sags (pu)

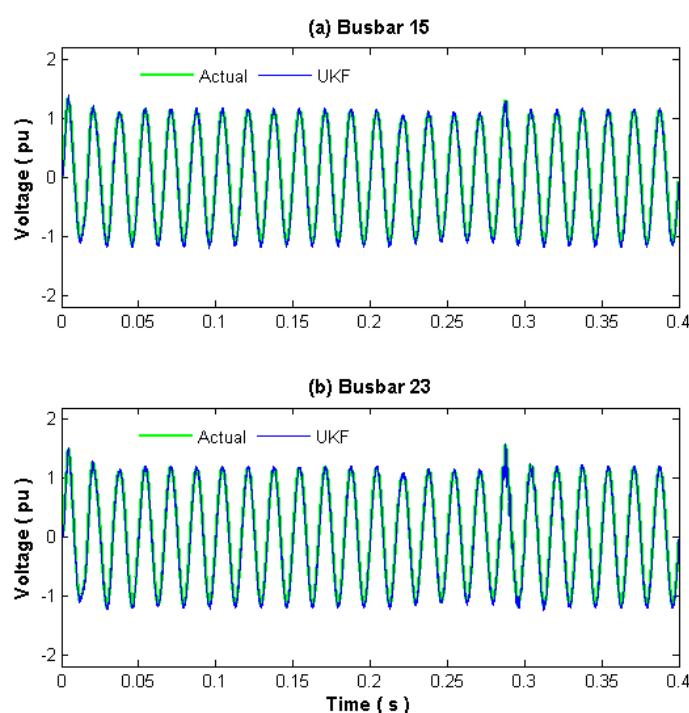
Busbar	Actual	UKF	Busbar	Actual	UKF
3	0.752	0.753	19	0.889	0.890
4	0.713	0.717	20	0.889	0.890
6	0.858	0.860	21	0.892	0.892
7	0.908	0.910	22	0.892	0.893
9	0.870	0.876	23	0.893	0.893

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10	0.890	0.900	24	0.887	0.888
12	0.870	0.880	25	0.888	0.889
14	0.880	0.885	26	0.892	0.892
15	0.872	0.873	27	0.891	0.892
16	0.880	0.890	28	0.880	0.881
17	0.892	0.895	29	0.884	0.885
18	0.875	0.880	30	0.907	0.909

328 4.3. Case study: UKF VSSE single-phase to ground fault at busbar 15

329 This case study reviews the UKF VSSE when a single-phase to ground fault is applied at busbar
 330 15; the fault impedance is 0.35 pu. This impedance is used to decrease the fault effect in the transient
 331 system condition. Busbar 15 has no voltage measurement, however, the state estimation is able to
 332 assess its voltage and the voltage of the nearby busbars with the same measurement points of the
 333 previous case. Measurements are contaminated with a 2.5% SNR noise. This case study is addressed
 334 to represent a short-circuit in the distribution network to assess the VSSE. **The state estimation**
 335 **assessment of high power load switching can be also addressed.** Figure 8 shows results under the
 336 short-circuit fault condition for busbar voltages 15 and 23; these are the busbars that present the
 337 largest voltage sag during the examined transient condition.

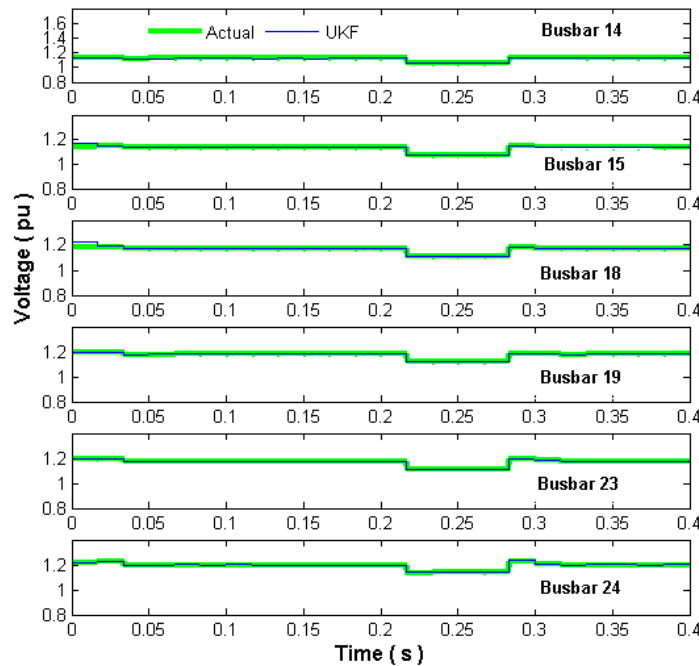


338

339 **Figure 8.** Actual and UKF estimated voltage waveforms of busbars 15 and 23, voltage sag from 0.216
 340 to 0.283 s, cycles 14-17, voltage swell during cycle 18, short-circuit at busbar 15.

341 A close agreement between the actual and UKF estimated signals including the post-fault period
 342 is achieved. Note the swell condition after the fault period. The UKF NRMSE for voltage at busbars
 343 15 and 23 are 1.5%, and 0.65%, respectively.

344 Figure 9 shows the rms busbar voltages near of the busbar 15. The proposed UKF algorithm
 345 gives acceptable estimates for the voltage sag magnitude and duration, mainly for the transient
 346 starting and ending time, respectively. This data can be used to classify the type of voltage sags. After
 347 the fault period, a voltage swell condition of different magnitude is present during the next two
 348 cycles, disappearing when the system transits to its steady state.



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Figure 9. Actual and UKF rms voltage, short-circuit at busbar 15 during 14-17 cycles, from 0.216 to 0.283 s.

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4.4. Case study: UKF VSSE transient load condition at busbar 24

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The proposed UKF-VSSE methodology is applied to estimate a transient load condition; this condition originates a fluctuating voltage sag/swell. The load at busbar 24 varies from cycles 6.25 to 18.75, generating a 12.5 cycle voltage transient in the busbar voltage waveforms. The current demanded by the load at busbar 24 increases 3 times during the first 4.25 cycles of the transient period and 6 times during the next 4 cycles. It then goes back to three times of the initial load current over the following 4.25 cycles, giving a transient condition during 12.5 cycles. Table 4 gives these load changes; the variations may represent mechanical load transients of an electrical motor, the commutation of linear and nonlinear electric loads at the power system busbars, faults, heavy motors starting, or electric heaters turning on, among others. This case study addresses a transient load condition in the distribution system.

363

Table 4. Transient load condition

Period	Cycles	Time (s)	Load current (pu)
Initial	00.00-06.25	0.000-0.104	1.00
Load transient 1	06.25-10.50	0.104-0.175	3.00
Load transient 2	10.50-14.50	0.175-0.241	6.00
Load transient 3	14.50-18.75	0.241-0.312	3.00
Final	18.75-24.00	0.312-0.400	1.00

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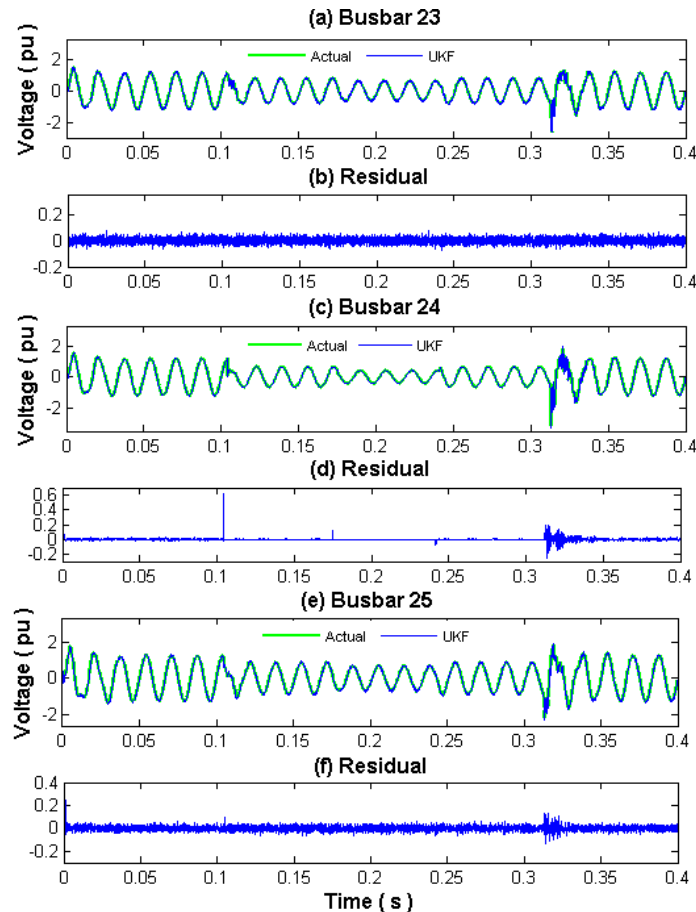
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Figure 10 shows the voltage waveforms at busbars 23, 24, and 25 during the transient load condition. The busbar voltages show the largest fluctuations as a result of the varying load at busbar 24. When the load current increases 3 times, the busbar voltages tend to drop generating a voltage sag. The voltage drops during the first 4.25 cycles of the transient period (6.25 to 10.5 cycles) then again decreases over the next 4 cycles to show the effect of the load current, which increases 6 times

370 during those 4 cycles (10.5 to 14.5 cycles). Finally, the current goes back to three times of the value at
 371 the initial period (14.5 to 18.75 cycles).

372 Load transient initiates at 6.25 cycles instead of 6 cycles to evaluate a more critical transient;
 373 similarly, the load transient finishes at 18.75 cycles instead of 18 cycles.

374 The transient state lasts 12.5 cycles (0.208 s), ending at 18.75 cycles (0.312 s), a voltage swell
 375 condition is present during the three cycles of the final transient period; voltage waveforms
 376 eventually reach the steady state close to the pre-fault operating condition.

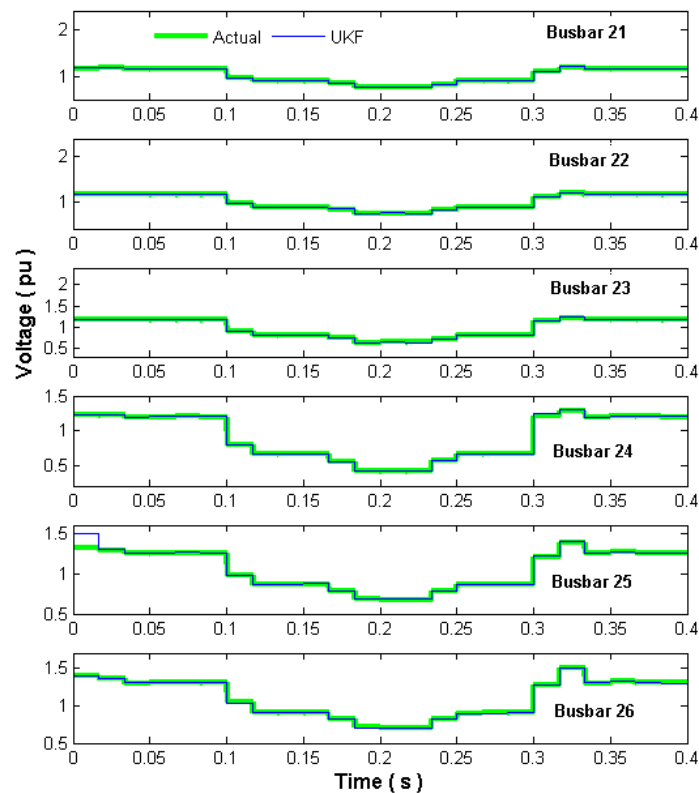


377

378 **Figure 10.** Voltage waveforms, actual, UKF, and residuals of busbars 23, 24, and 25; transient load
 379 condition at busbar 24 (0.104 to 0.312 s).

380 NRMSE between actual and UKF estimated waveforms for voltage at busbars 23, 24 and 25 in
 381 Figure 10, are 0.45%, 0.40% and 2.43%, respectively.

382 The rms voltage magnitudes have been calculated using (26) for actual and UKF estimated
 383 waveforms during the transient load condition. Figure 11 shows the rms voltage magnitude for each
 384 cycle at busbars 21-26, which are close to the load transient of busbar 24.



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Figure 11. Actual and UKF rms voltage magnitudes, transient load condition at busbar 24 from 0.104 to 0.312 s, during 6.25–18.75 cycles.

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The obtained rms voltage magnitudes represent the initial, transient and final operation periods, as well as the intermediate transient generating a fluctuating voltage sag. Actual and UKF estimate rms magnitudes closely agree. Please notice the voltage swell of different magnitude during the final period. The proposed UKF VSSE methodology closely estimates these voltage variations.

The use of detailed models to represent the power system components can reduce the state estimation error. Parameters should be close to their real values, filtering the noise from measurements before the assessment of the estimation, and increasing the available measurements.

It should be noted from the above case studies, that the UKF implemented in Matlab script language is still slow to be used in real-time applications. However, with adequate computational techniques such as parallel processing, better computational capability and programs compilation, the execution time can be significantly reduced.

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5. Conclusions

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A time-domain state estimation methodology for voltage sags in power networks using the UKF has been proposed. Nonlinear models for system and measurement equation have been used. It has been demonstrated that the UKF can be applied to precisely assess the voltage sag state estimation in power systems with nonlinear components. The proposed method has been verified using a modified version of the IEEE 30-bus test power system and noisy measurements.

It has been shown that the proposed UKF method dynamically follows the generation of voltage sags, by executing the estimator continuously, to record the voltage sags originated during the power network operation, especially for unmonitored busbars. This requires of an accurate model, a set of synchronized measurements preferably with low noise, sufficient to obtain an observable condition of busbar voltages. The measurement sampling frequency should satisfy the sampling theorem. The

410 rms value can be computed from discrete waveforms; this value gives the information to define the
411 sag magnitude, delimiting the sag time interval.

412 From the conducted case studies, it has been observed that when the power system goes under
413 fast transients, the UKF estimator error is more noticeable; however, as the network evolves to steady
414 state, the error quickly decreases to negligible proportions, i.e. on average 1%. In most cases, this
415 period is short compared with the voltage sag estimation interval. This condition is present during
416 the final period of the reviewed case studies, when the fault or transient condition is removed. It
417 should be noted that usually at this time, a voltage swell is generated.

418 The state estimation error increases when sudden transient variations are present. The results
419 obtained with the proposed UKF VSSE methodology have been successfully compared against actual
420 values taken from a simulation of the test power system under the same transient condition. A close
421 agreement has been achieved in all cases between the compared responses.

422 **Acknowledgements:** The authors gratefully acknowledge the Universidad Michoacana de San
423 Nicolás de Hidalgo through the Facultad de Ingeniería Eléctrica, División de Estudios de Posgrado
424 (FIE-DEP) Morelia, México, for the facilities granted to carry out this investigation. First two authors
425 acknowledge financial assistance from CONACYT to conduct this investigation.

426 **Author Contributions:** Rafael Cisneros-Magaña performed the simulation and modelling, analyzed
427 the data, and wrote the paper. Aurelio Medina analyzed the results, reviewed the modeling and text,
428 and supervised the related research work. Olimpo Anaya-Lara provided critical comments and
429 revised the paper.

430 **Conflicts of Interest:** The authors declare no conflict of interest.

431 Nomenclature

432 *List of Abbreviations*

433	EAF	Electric arc furnace
434	FACTS	Flexible alternating current transmission system
435	KF	Kalman filter
436	NRMSE	Normalized root mean square error
437	PQ	Power quality
438	SNR	Signal to noise ratio
439	STATCOM	Static synchronous compensator
440	TCR	Thyristor-controlled rectifier
441	UKF	Unscented Kalman filter
442	UT	Unscented transform
443	VSSE	Voltage sags state estimation
444	WAMS	Wide area measurement system

445 *List of Symbols*

446	e	State estimation error vector
447	f	Nonlinear state function
448	h	Nonlinear output function
449	k	Time instant $t=k\Delta t$
450	$k+1$	Time instant $t=(k+1)\Delta t$
451	m	Number of measurements
452	n	Number of state variables
453	t	Time vector
454	u	Input vector
455	v	Process noise vector
456	w	Measurement noise vector
457	x	State vector
458	\hat{x}	Estimated state vector
459	y	Output vector

460	z	Measurement vector
461	E	Expected value
462	H	Measurements matrix
463	K	Kalman filter gain matrix
464	N	Normal distribution
465	P	Error covariance matrix
466	Q	Process noise covariance matrix
467	R	Measurement noise covariance matrix
468	V_{rms}	Rms voltage magnitude
469	W	Scalar weights
470	+	A posteriori or after measurement estimate
471	-	A priori or before measurement estimate
472	Δt	Step time
473	α	Parameter to determine the spread of sigma points
474	β	Parameter to determine the distribution of x
475	λ	Scaling parameter
476	γ	Scaling parameter
477	κ	Parameter to determine the spread of sigma points
478	χ	Sigma points matrix

479 Appendix A Per unit additional nonlinear load parameters

480 EAF busbar 2: $Leaf=0.5$, $k1=0.004$, $k2=0.0005$, $k3=0.005$, $m=0$, $n=2.0$, initial condition EAF arc
481 radius=0.1

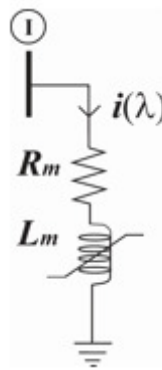
482 Nonlinear inductance busbar 5: $Rm=4.0$, $Lm=1.0$, $n=5.0$, $a=0$, $b=0.3$

483 TCR busbar 6: $Rtcr=1.0$, $Ltcr=0.5$, firing angle $\alpha=100$ deg.

484 Appendix B Nonlinear models

485 *Nonlinear inductor*

486 Figure B.1 shows a nonlinear inductor.



487

488

Figure B.1. Nonlinear inductance.

489 According to KVL, the first-order differential equation to represent the nonlinear inductance is:

$$490 \quad \frac{d\lambda}{dt} = v_l - R_m i(\lambda) \quad (B.1)$$

491 The discrete form of (B.1) to define (8-9) is given by,

$$492 \quad \lambda_{(k+1)} = \lambda_{(k)} + \Delta t [d\lambda/dt]|_k = \lambda_{(k)} + \Delta t [v_{l(k)} - R_m i(\lambda_{(k)})] \quad (B.2)$$

493 where Δt is the time step and k indicates the evaluation at time $t(k)$.

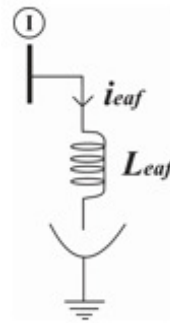
494 The nonlinear solution of (B.1), is represented by $i(\lambda)$, λ is the nonlinear inductor magnetic flux,
 495 the polynomial approximation for $i(\lambda)$ is,

$$496 \quad i(\lambda) = a\lambda + b\lambda^n \quad (\text{B.3})$$

497 n is an odd number due to the odd symmetry of (B.3). Coefficients a , b and n adjust the nonlinear
 498 saturation curve. The rational fractions and hyperbolic approximations are alternative methods to
 499 represent this nonlinearity [38-39].

500 *Electric arc furnace*

501 Figure B.2 shows the EAF model which can be expressed mathematically by two first-order
 502 nonlinear differential equations based on the energy conservation law, where the state variables are
 503 the arc radius r_{eaf} and the EAF current i_{eaf} [39].



504

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Figure B.2. Electric arc furnace.

506 The first-order nonlinear differential equations to represent the EAF are:

$$507 \quad dr_{eaf}/dt = (k_3/k_2)r_{eaf}^{(-m-3)}i_{eaf}^2 - (k_1/k_2)r_{eaf}^{(n-1)}, \quad (\text{B.4})$$

$$508 \quad di_{eaf}/dt = (1/L_{eaf})(v_l - k_3r_{eaf}^{(-m-2)}i_{eaf}), \quad (\text{B.5})$$

509 where n represents the arc cooling effect and m the arc column resistivity [38-39].

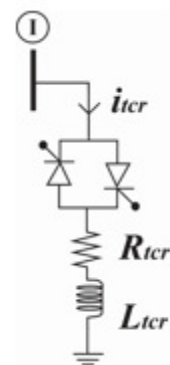
510 The following expressions give the discrete forms of (B.4) and (B.5) to define (8-9),

$$511 \quad r_{eaf(k+1)} = r_{eaf(k)} + \Delta t[(k_3/k_2)r_{eaf(k)}^{(-m-3)}i_{eaf(k)}^2 - (k_1/k_2)r_{eaf(k)}^{(n-1)}], \quad (\text{B.6})$$

$$512 \quad i_{eaf(k+1)} = i_{eaf(k)} + \Delta t[(1/L_{eaf})(v_{l(k)} - k_3r_{eaf(k)}^{(-m-2)}i_{eaf(k)})] \quad (\text{B.7})$$

513 *Thyristor controlled reactor*

514 A thyristor pair back-to-back connection represents the TCR jointly with an RL circuit. The TCR
 515 current is the state variable, the TCR model is shown in Figure B.3.



516

517

Figure B.3. Thyristor controlled reactor.

518 According to KVL, the first-order nonlinear differential equation modelling the TCR is:

$$519 \quad di_{tcr}/dt = s(v_l - i_{tcr}R_{tcr})/L_{tcr} \quad (B.8)$$

520 The discrete form of (B.8) to define (8-9) is given by,

$$521 \quad i_{tcr(k+1)} = i_{tcr(k)} + \Delta t[s_{(k)}(v_{l(k)} - i_{tcr(k)}R_{tcr})/L_{tcr}] \quad (B.9)$$

522 The TCR current is controlled by the thyristor-firing angle α , the variable s represents this
 523 dependency being the switching function to turn on the thyristors, which varies according to the
 524 desired firing angle α . This generates harmonic distortion in the voltage and current waveforms.
 525 Because of this distortion, the TCR can be considered as a nonlinear component.

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