

**Attenuation, Delay and Noise properties of light
slowed by Electromagnetically Induced
Transparency.**

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Declaration

This thesis is an account of research undertaken between February 2004 and October 2006 at The Department of Physics, Faculty of Science, The Australian National University, Canberra, Australia.

Except where acknowledged in the customary manner, the material presented in this thesis is, to the best of my knowledge, original and has not been submitted in whole or part for a degree in any university.

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October, 2006

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Abstract

Electromagnetically Induced Transparency is an effect that allows light to propagate through an otherwise opaque material. Due to the dispersion properties of the EIT system, it also slows the information carried on the sideband frequencies of a light beam. The effect can also be used to store information in an atomic medium. It was therefore proposed that EIT systems can be applied in optical circuits, specifically with regard to quantum information processing where it is necessary to delay, or store quantum states of light. Hence it is necessary to understand how EIT systems influence the quantum state of a light beam and whether EIT systems can ever have the potential to delay and store quantum information at the quantum noise limit.

This thesis examines the transfer of an optical signal at the sideband frequencies through an EIT medium. A comprehensive characterisation of an EIT system in terms of delay and signal transmission is performed. This measurement technique provides data on EIT performance over a range of signal frequencies and a range of experimental parameters. Experiments were conducted with both pure Rubidium vapour and Rubidium vapour mixed with a noble gas.

Studies have been made into a new system of measurement that will allow delay, attenuation and added noise spectra to be assessed via a single-shot measurement.

Contents

Declaration	iii
Acknowledgements	v
Abstract	vii
1 Introduction	3
1.1 History	3
1.2 Motivation	4
1.3 Thesis Structure	4
2 Theory	7
2.1 Electromagnetically Induced Transparency	7
2.1.1 The Dark State	8
2.1.2 Equations of Motion	9
2.2 Susceptibility and the Group Velocity	10
2.3 Homodyne detectors	13
2.4 Conditional Variance	14
2.4.1 Ideal Conditional Variance	15
2.4.2 Practical Conditional Variance Measurement	16
3 Experiment	19
3.1 Laser Source	19
3.1.1 The Laser	19
3.1.2 Multimode Behaviour	20
3.1.3 Feedback	20
3.1.4 Mode Quality	20
3.2 Rubidium Structure	20
3.3 Laser Frequency Control	22
3.3.1 Doppler Broadening	22
3.3.2 Saturated Absorption Spectroscopy	23
3.4 Locking Loops	24
3.5 Beam preparation	27
3.6 Rubidium Cell	27
3.7 Homodyne Detectors	28
3.8 Data Acquisition	29
3.8.1 Network Analyser	29
3.8.2 PXI	29
3.9 Data Processing	29
3.9.1 Delay measurement	29
3.9.2 Normalisation to Quantum Noise	30
3.9.3 Conditional Variance and Attenuation	30

4	Results	31
4.1	Laser Noise	31
4.2	Delay Measurements	32
4.2.1	Unbuffered Rubidium Cell	32
4.2.2	Buffered Rubidium Cell	34
4.2.3	Spatial Effects	36
4.3	PXI Measurements and Post Processing	36
4.3.1	Noise Characteristics of the Measurement System	36
4.3.2	Post Processed Delay Measurements	37
4.3.3	Post Processed Attenuation Measurements	37
4.3.4	Conditional Variance	38
5	Conclusion and Future Directions	41
5.1	Conclusions	41
5.2	Future Directions	41
	Bibliography	43

List of Figures

2.1	Three level λ system.	7
2.2	Susceptibility near an atomic Resonance.	11
2.3	Real and imaginary parts of $\chi(\omega, z)$ for an EIT system. This system has $\gamma_{ab}/2\pi = 3$ MHz, $\gamma_{bc} = 1$ kHz and $\Omega_c = 2\gamma_{ab}$. The application of a coupling beam causes a transparency on transition.	11
2.4	Real and imaginary parts of $\chi(\omega, z)$ for an EIT system. This system has $\gamma_{ab}/2\pi = 3$ MHz, $\gamma_{bc} = 1$ kHz and $\Omega_c = 0.5\gamma_{ab}$. The smaller coupling value causes the transparency window to decrease and the slope of the refractive index to increase.	12
2.5	A Homodyne Detector	13
2.6	An ideal conditional variance measurement. This measurement is impossible as it requires detecting the beam both before and after the EIT mechanism.	15
2.7	A conditional variance measurement	16
3.1	D1 line hyperfine level structure for Rubidium-87.	21
3.2	The full energy level diagram for the Rubidium 87 D1-line F=2 to F=1 transition.	22
3.3	Saturated Absorption Setup	24
3.4	Locking	25
3.5	Saturated Absorption spectra and associated error signals for natural rubidium. Transitions are labelled by isotope number and angular momenta of the ground state and excited state.	26
3.6	Setup for the experiment. Figure 3.3 shows the experimental setup before the optical fibre. AM is an amplitude modulator, PM is a phase modulator, GT is a Glan Thompson polariser. $\lambda/2$ is a half waveplate, $\lambda/4$ is a quarter waveplate.	27
3.7	Experimental Setup.	30
4.1	Laser noise, Quantum Noise and Detector Noise plots. 4mW of light incident on the detector out of 80mW total laser power. Two frequency scales are shown.	31
4.2	Delay and Transmission for an unbuffered Rubidium Cell. 50 μ W Probe beam.	32
4.3	Delay and Transmission for an unbuffered Rubidium Cell. 100 μ W Probe beam.	33
4.4	Delay and Transmission for an unbuffered Rubidium Cell. 150 μ W Probe beam.	33
4.5	Delay and Transmission for a buffered Rubidium Cell. 200 μ W Probe beam.	34
4.6	Delay and Transmission for a buffered Rubidium Cell. 300 μ W Probe beam.	35
4.7	Delay and Transmission for a buffered Rubidium Cell. 400 μ W Probe beam.	35
4.8	Noise measurements with 6mW local oscillator power.	37
4.9	Delay versus modulation frequency for the Buffer Gas cell.	38

4.10	Gain calculated using post-processed PXI data.	39
4.11	Conditional Variance, normalised to Quantum Noise and via Equation 2.16	40
4.12	Normalisation test of Conditional Variance algorithm.	40

Introduction

Electromagnetically Induced Transparency (EIT) is a physical effect where a laser beam is used to change an opaque material so that light can pass through it. This effect also causes a reduction in the speed of information carried by the light. The ability to delay light is of crucial importance for synchronising optical circuits, particularly in the emerging field of quantum information science. Aside from the practical applications, understanding the details of this effect will give greater understanding of atom-light interactions in general.

1.1 History

The groundwork for EIT is a process called Coherent Population Trapping (CPT), first demonstrated in 1976 by Alzetta et.al. [1]. To briefly explain this effect I will adapt an explanation given by Harris [2]. Classically, the interaction of light and matter can be understood as the light causing the electrons to vibrate at a certain frequency. If the electrons are vibrating at a very specific frequency then they will undergo a transition to another energy state. If two vibrations were out of phase, then their effects would destructively interfere. This would mean that the electrons could not vibrate at that specific frequency and so could not undergo a transition to an excited state. This coherence effect causes the population of electrons to be trapped in the lower energy states.

EIT was first demonstrated in 1991 [3] where it was shown that Strontium vapour could be made transparent. CPT is the mechanism that allows EIT to work. Classically, light is absorbed when it drives an electron into a higher energy state. If this energy state is forbidden by CPT, then light cannot be absorbed and the material becomes transparent.

Eight years later, slowing of light pulses was demonstrated using EIT in a Bose-Einstein Condensate of Sodium atoms [4]. In a Bose-Einstein Condensate, atoms are so cold that they are effectively stationary and occupy the same quantum state. In this experiment light pulses were slowed down to 17ms^{-1} , showing that EIT also causes a reduction in the group velocity of light.

Since these pulses of light were travelling so slowly, they were spatially compressed as well. They could be compressed enough so that the entire pulse fitted inside the atomic medium. Philips et. al. [5] showed that if the coupling laser beam was turned off while the pulse was inside the medium, then the pulse was stored and could be later retrieved by turning on the coupling laser beam. Philips et. al. [5] and later Bajcsy et. al. [6] showed that pulses of light could be stored in atomic vapour cells.

Liu et.al. [7] showed that storage of pulses was possible in a cold atomic cloud. Slow light and pulse storage via EIT have also been observed in cryogenic solids [8], with a record delay time of greater than one second being set by Longdell et. al. [9]. Other impressive results include storage of single photons by Chanelière et. al. [10].

In continuous wave experiments, group velocities of 90ms^{-1} were achieved for an optical beam in a hot Rubidium gas [11]. Heated Rubidium vapour is the material used in the experiments of this thesis, the main advantage of the material being that experiments can be conducted at room temperature.

Theoretical calculations [12, 13, 14, 15] have claimed that the effect of EIT can delay a squeezed or entangled state. As there is very little absorption of the light, only a small amount of vacuum noise should degrade the squeezed or entangled state, enabling EIT to be used as a quantum limited delay device. Akamatsu et. al. [16] have shown that a squeezed state can survive transmission through an EIT medium, although this was under the conditions of very small delay times.

However, Hsu et. al. [17] have shown that a continuous wave signal delayed using EIT in a hot Rubidium gas experiences a lot more added noise than theoretical calculations would suggest. The source of this extra noise is not understood and it can only be identified by varying parameters of the experiment and determining which parameters make a change to the amount of added noise. Atomic decoherences in the form of Radiation Trapping [18] or atomic collisions have been suggested as mechanisms by which this extra noise could be added to the system.

Experimental results of light storage via EIT have been called into question by Alexandrov and Zapasskii [19], [20] claiming that the delay phenomena observed have for the most part been over interpreted and that the pulse delays observed can be explained in terms of much simpler physics. The results have also been called into question after Akulshin et. al. [21] demonstrated qualitatively similar results to those of Phillips et. al. [5] under experimental conditions where EIT was not possible. The main criticism is that Saturable Absorption causes pulse-resaping. If the group velocity is defined as the speed of the peak of a pulse, and if the pulse is reshaped, then the speed of the peak of a pulse is no longer an accurate measure of the group velocity.

Selden [22] listed criteria that would confirm that delays were due to EIT and not due to Saturable Absorption effects. The most relevant of these is that a continuous wave signal must be delayed by at least a quarter of a cycle for the delay to unambiguously be ascribed to EIT.

1.2 Motivation

What is not well known is the source of the extra noise found by Hsu et. al. [17]. One of the effects that EIT relies upon is Coherent Population Trapping. If the atoms were to lose their coherence with each other then noise could be added to the light transmitted through an EIT system. Given that the material we are working with is a hot atomic vapour, atom-atom collisions could easily introduce decoherences.

It is also desirable to develop techniques to quickly measure the noise added to the system. I have worked on post-processing digitally acquired data to measure conditional variance, which is a measure of noise. These measurements could theoretically be made single-shot.

1.3 Thesis Structure

This thesis is broken into Theory, Experiment and Results Section. The Theory section covers the background theory of EIT and Conditional Variance measurements. Conditional

variance is a measure of the sameness of two signals and is used to calculate the noise added to a signal by EIT. The experimental section covers all aspects of the experiment, from the laser itself, to the structure of the atoms, to the detection systems used. Results shows the results I was able to obtain, and details some of the difficulties encountered.

Theory

The aim of this experiment is to measure properties of the EIT system, specifically delay, attenuation and added noise as a function of sideband frequency. It is necessary to understand what mechanism creates the transparency and delay. Also, to properly choose what measurements to take and how to analyse the data it is necessary to consider some quantum optics theory.

2.1 Electromagnetically Induced Transparency

In this treatment of EIT the coupling beam will be assumed to be classical, and much larger than the probe field. The probe field and atoms will be treated quantum mechanically. It is also assumed that there is no detuning from the atomic transitions.

The following equation for the system's Hamiltonian comes from [23] Eq. (2.15). the detuning terms in the original equation have been dropped. Many of the following equations have been adapted from [23] chapter 2.

$$\tilde{H}_{int} = -\frac{\hbar N}{l} \int g\hat{\mathcal{E}}(z,t)\hat{\sigma}_{ab}(z,t) + g\hat{\mathcal{E}}^\dagger(z,t)\hat{\sigma}_{ba}(z,t) + \Omega_c\hat{\sigma}_{ac}(z,t) + \Omega_c\hat{\sigma}_{ca}(z,t) dz \quad (2.1)$$

The Hamiltonian has been written in terms of the atomic operators ($\hat{\sigma}_{\mu\nu}$), the probe field operator ($\hat{\mathcal{E}}$) and the Rabi frequency (Ω_c), taken to be real, of the coupling beam.

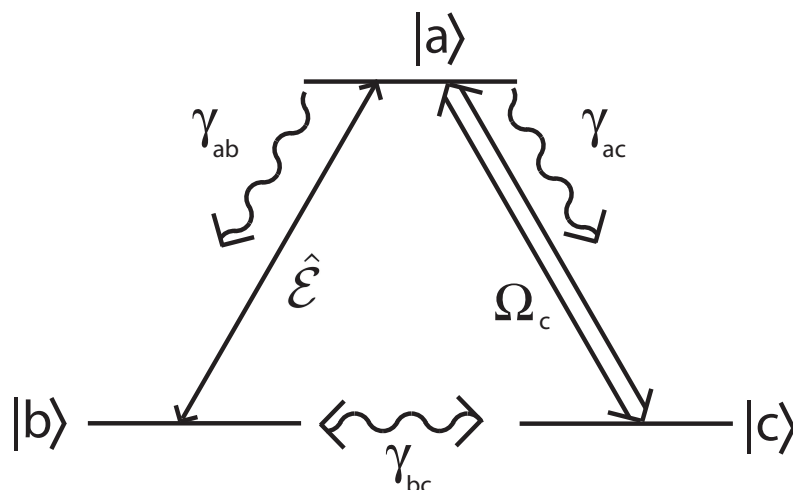


Figure 2.1: Three level λ system.

$g = d_{ba}\sqrt{\omega_{ab}/2\epsilon_0 V\hbar}$ is the atom-field coupling constant, where d_{ba} is the atomic dipole moment of the transition and ω_{ba} is the frequency of the transition. N is the number of atoms, l is length of the cell.

The atomic operators ($\hat{\sigma}_{\mu\nu}$) refer to a class of atoms with zero velocity

$$\hat{\sigma}_{\mu\nu}(z, t) = \frac{1}{n\mathcal{A}\delta z} \sum_{z_j \in [z, z+\delta z]} \hat{\sigma}_{\mu\nu}^j(z, t) e^{i\frac{\omega_{\mu\nu}}{c}(z_j - ct)}$$

Where n is the number of atoms, \mathcal{A} is the area of a small transverse slice of the cell and $\hat{\sigma}_{\mu\nu}^j$ is the j th atom in the small volume $\mathcal{A}\delta z$.

2.1.1 The Dark State

The eigenstate of this Hamiltonian is special in that it has no population in the excited state $|a\rangle$. This means the state cannot spontaneously emit light, hence it is called a dark state.

To calculate the eigenstate of the Hamiltonian, the probe beam will be treated classically, and the spatial dependence will be removed. This simplified Hamiltonian is then;

$$\tilde{H}_{int} = -\hbar N \Omega_p(t) \hat{\sigma}_{ab}(t) + \Omega_p^*(t) \hat{\sigma}_{ba}(t) + \Omega_c \hat{\sigma}_{ac}(t) + \Omega_c \hat{\sigma}_{ca}(t)$$

Where $\Omega_p = g\langle \hat{\mathcal{E}} \rangle$. The eigenstate of this Hamiltonian is found by;

$$\begin{aligned} \hat{H}_{int} (a(t)|a\rangle + b(t)|b\rangle + c(t)|c\rangle) &= \lambda (a(t)|a\rangle + b(t)|b\rangle + c(t)|c\rangle) \\ -\hbar N (g\Omega_p(t)b(t)|a\rangle + \Omega_p^*(t)a(t)|b\rangle + \Omega_c c(t)|a\rangle + \Omega_c a(t)|c\rangle) &= \lambda (a(t)|a\rangle + b(t)|b\rangle + c(t)|c\rangle) \end{aligned}$$

Equating coefficients of the basis states gives;

$$\begin{aligned} \Omega_c a(t) &= \lambda c(t) \\ g\Omega_p^*(t)a(t) &= \lambda b(t) \\ g\hat{\mathcal{E}}(t)b(t) + \Omega_c c(t) &= \lambda a(t) \end{aligned}$$

The only nontrivial solution to this is;

$$\begin{aligned} \lambda &= 0 \\ a(t) &= 0 \\ b(t) &= \Omega_c \\ c(t) &= -\Omega_p(t) \end{aligned}$$

As explained before, this eigenstate has no population in the excited state $|a\rangle$. Therefore atoms cannot spontaneously emit light and there is very little absorption. The lack of spontaneous emission earns this eigenstate the name 'Dark State'. The dark state is a superposition of the two ground states.

$$|D\rangle = \frac{\Omega_c|b\rangle - \Omega_p(t)|c\rangle}{\sqrt{|\Omega_c|^2 + |\Omega_p(t)|^2}}$$

2.1.2 Equations of Motion

Assuming there is no decoherence, i.e. an ideal three level system. The equations of motion of the atomic operators comes from the Hamiltonian (Eq. 2.1) via the Schrodinger equation. From here on the space and time dependence of the atomic and probe field operators will not be explicitly shown.

$$i\hbar \frac{\partial \hat{\sigma}_{\mu\nu}}{\partial t} = [\hat{\sigma}_{\mu\nu}, \bar{H}_{int}]$$

$$\begin{aligned} \dot{\hat{\sigma}}_{aa} &= ig(\hat{\mathcal{E}}\hat{\sigma}_{ab} - \hat{\mathcal{E}}^\dagger\hat{\sigma}_{ba}) + i\Omega_c(\hat{\sigma}_{ac} - \hat{\sigma}_{ca}) \\ \dot{\hat{\sigma}}_{bb} &= ig(-\hat{\mathcal{E}}\hat{\sigma}_{ab} + \hat{\mathcal{E}}^\dagger\hat{\sigma}_{ba}) \\ \dot{\hat{\sigma}}_{cc} &= i\Omega_c(-\hat{\sigma}_{ac} + \hat{\sigma}_{ca}) \\ \dot{\hat{\sigma}}_{ba} &= ig(\hat{\mathcal{E}}\hat{\sigma}_{bb} - \hat{\mathcal{E}}^\dagger\hat{\sigma}_{aa}) + i\Omega_c\hat{\sigma}_{bc} \\ \dot{\hat{\sigma}}_{bc} &= -ig\hat{\mathcal{E}}\hat{\sigma}_{ac} + i\Omega_c\hat{\sigma}_{ba} \\ \dot{\hat{\sigma}}_{ac} &= -ig\hat{\mathcal{E}}^\dagger\hat{\sigma}_{bc} + i\Omega_c(\hat{\sigma}_{aa} - \hat{\sigma}_{cc}) \end{aligned}$$

Spontaneous emission, and decoherence are two loss effects which need to be accounted for. These effects are represented by the γ terms. To preserve the commutation relations it is necessary to add the Langevin noise operators, represented by the \hat{F} terms. These terms can be phenomenologically added to the equations of motion, although they can also be added rigorously.

$$\begin{aligned} \dot{\hat{\sigma}}_{aa} &= -\gamma_{aa}\hat{\sigma}_{aa} + ig(\hat{\mathcal{E}}\hat{\sigma}_{ab} - \hat{\mathcal{E}}^\dagger\hat{\sigma}_{ba}) + i\Omega_c(\hat{\sigma}_{ac} - \hat{\sigma}_{ca}) + \hat{F}_{aa} \\ \dot{\hat{\sigma}}_{bb} &= \gamma_{bb}\hat{\sigma}_{aa} + ig(-\hat{\mathcal{E}}\hat{\sigma}_{ab} + \hat{\mathcal{E}}^\dagger\hat{\sigma}_{ba}) + \hat{F}_{bb} \\ \dot{\hat{\sigma}}_{cc} &= \gamma_{cc}\hat{\sigma}_{aa} + i\Omega_c(-\hat{\sigma}_{ac} + \hat{\sigma}_{ca}) + \hat{F}_{cc} \\ \dot{\hat{\sigma}}_{ba} &= -\gamma_{ba}\hat{\sigma}_{ba} + ig(\hat{\mathcal{E}}\hat{\sigma}_{bb} - \hat{\mathcal{E}}^\dagger\hat{\sigma}_{aa}) + i\Omega_c\hat{\sigma}_{bc} + \hat{F}_{ba} \\ \dot{\hat{\sigma}}_{bc} &= -\gamma_{bc}\hat{\sigma}_{bc} - ig\hat{\mathcal{E}}\hat{\sigma}_{ac} + i\Omega_c\hat{\sigma}_{ba} + \hat{F}_{bc} \\ \dot{\hat{\sigma}}_{ac} &= -\gamma_{ac}\hat{\sigma}_{ac} - ig\hat{\mathcal{E}}^\dagger\hat{\sigma}_{bc} + i\Omega_c(\hat{\sigma}_{aa} - \hat{\sigma}_{cc}) + \hat{F}_{ac} \end{aligned}$$

The Maxwell equation to describe light in this medium is;

$$\left(\frac{\partial}{\partial t} + c\frac{\partial}{\partial z}\right)\hat{\mathcal{E}} = ig^*N\sigma_{ba}$$

It has been assumed that the coupling beam is much stronger than the probe beam. It is also assumed that in the initial state the entire population is in $|b\rangle$ so $\sigma_{bb} = 1$. Then $\hat{\sigma}_{aa}$ and $\hat{\sigma}_{ac}$ are zero to first order. Also the probe beam has a signal on it and so is a function of time, whereas the coupling beam is roughly constant. Hence the following equations can be formed.

$$\begin{aligned}\dot{\hat{\sigma}}_{ba} &= -\gamma_{ba}\hat{\sigma}_{ba} + ig\hat{\mathcal{E}}(t) + i\Omega_c\hat{\sigma}_{bc} + \hat{F}_{ba} \\ \dot{\hat{\sigma}}_{bc} &= -\gamma_{bc}\hat{\sigma}_{bc} + i\Omega_c\hat{\sigma}_{ba} + \hat{F}_{bc}\end{aligned}$$

These equations can be solved by transforming to Fourier Space via

$$\tilde{\mathcal{O}}(z, \omega) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} \hat{\mathcal{O}}(z, t) e^{i\omega t} dt$$

Giving;

$$\begin{aligned}-i\omega\tilde{\sigma}_{ba} &= -\gamma_{ba}\tilde{\sigma}_{ba} + ig\tilde{\mathcal{E}}(\omega) + i\Omega_c\tilde{\sigma}_{bc} + \tilde{F}_{ba} \\ -i\omega\tilde{\sigma}_{bc} &= -\gamma_{bc}\tilde{\sigma}_{bc} + i\Omega_c\tilde{\sigma}_{ba} + \tilde{F}_{bc} \\ \left(-i\omega + c\frac{\partial}{\partial z}\right)\tilde{\mathcal{E}} &= ig^*N\tilde{\sigma}_{ba}\end{aligned}\tag{2.2}$$

At this point we are only interested in deriving the susceptibility of the EIT medium. Only the classical terms are needed for this. $\tilde{\sigma}_{\mu\nu}$ will be replaced with $\langle\tilde{\sigma}_{\mu\nu}\rangle$, the probe field $g\tilde{\mathcal{E}}$ will be instead written as a Rabi frequency Ω_c , and the Langevin operators will be ignored. A full quantum treatment of these equations is given in ref. [23], the conclusion reached there is that the Langevin operators only add the amount of vacuum noise expected from the absorption of the medium. So, disregarding the Langevin operators, the equations 2.2 can be written as;

$$\left(\frac{\frac{|g|^2N}{c}(\gamma_{bc} - i\omega)}{(\gamma_{ba} - i\omega)(\gamma_{bc} - i\omega) + \Omega_c^2} - i\omega + c\frac{\partial}{\partial z}\right)\Omega_p = 0$$

let

$$\Lambda(\omega, z) = \frac{\frac{|g|^2N}{c}(\gamma_{bc} - i\omega)}{(\gamma_{ba} - i\omega)(\gamma_{bc} - i\omega) + \Omega_c^2} - i\omega\tag{2.3}$$

then;

$$\Omega_p\left(\frac{l}{2}, \omega\right) = e^{-\int_{-\frac{l}{2}}^{\frac{l}{2}} \Lambda(\omega, z) dz} \Omega_p\left(-\frac{l}{2}, \omega\right)\tag{2.4}$$

Equation 2.4 can be interpreted as the probe beam being attenuated and phase shifted by $\Lambda(\omega, z)$.

The probe susceptibility is defined as;

$$\frac{ik}{2}\chi(\omega, z) = -\left(\Lambda(\omega, z) + \frac{i\omega}{c}\right)\tag{2.5}$$

2.2 Susceptibility and the Group Velocity

The susceptibility of a medium determines both the refractive index, n , and the absorption coefficient, κ .

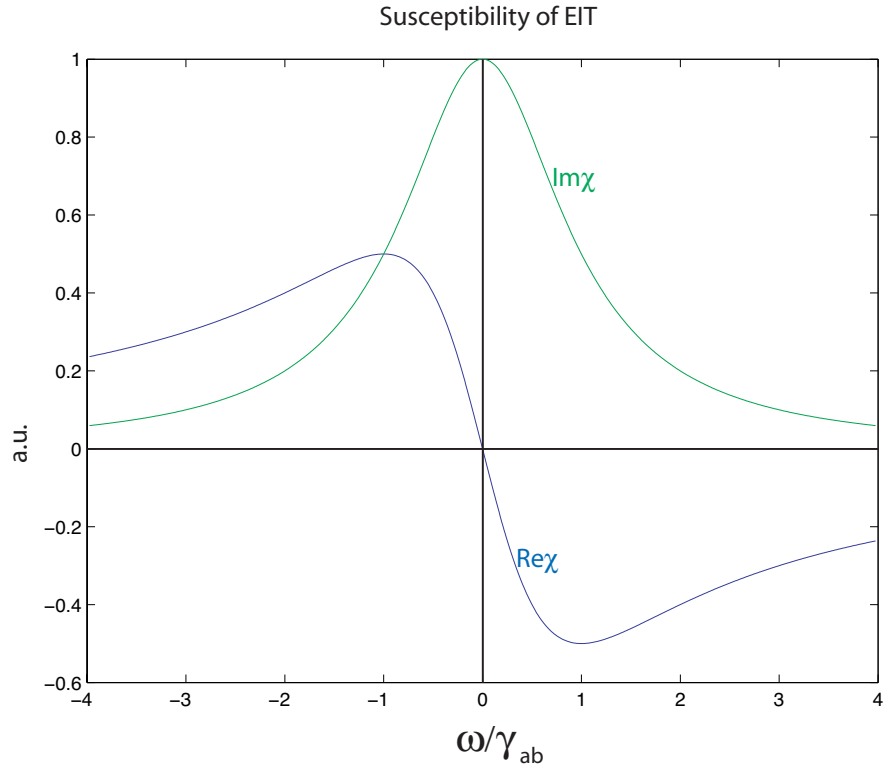


Figure 2.2: Real and imaginary parts of $\chi(\omega, z)$ for a typical two level system. This is what the susceptibility looks like when the coupling beam is turned off and there is no EIT.

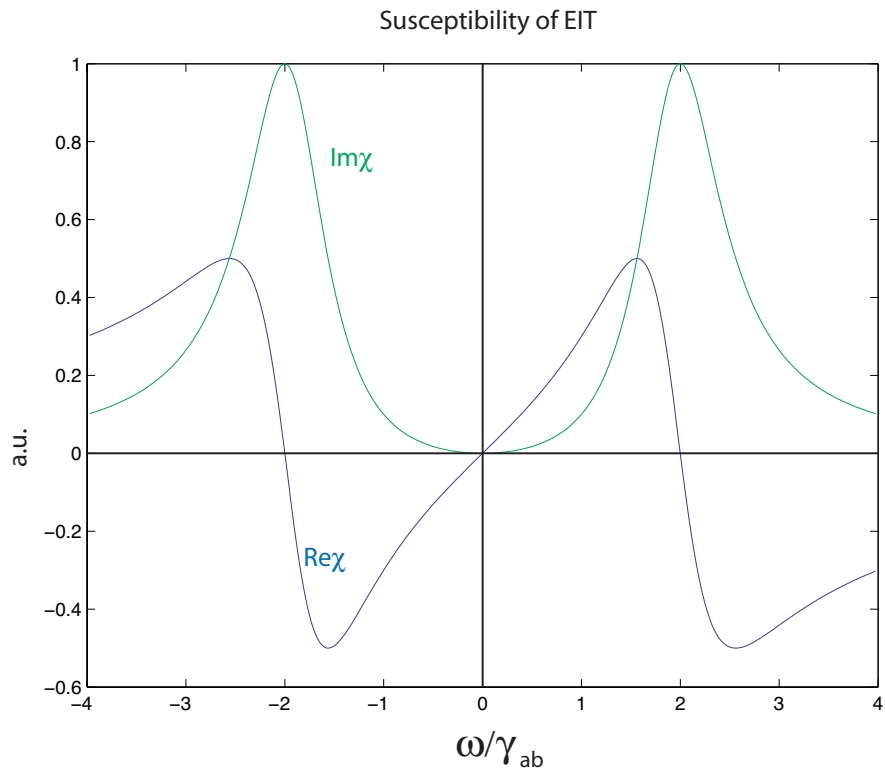


Figure 2.3: Real and imaginary parts of $\chi(\omega, z)$ for an EIT system. This system has $\gamma_{ab}/2\pi = 3$ MHz, $\gamma_{bc} = 1$ kHz and $\Omega_c = 2\gamma_{ab}$. The application of a coupling beam causes a transparency on transition.

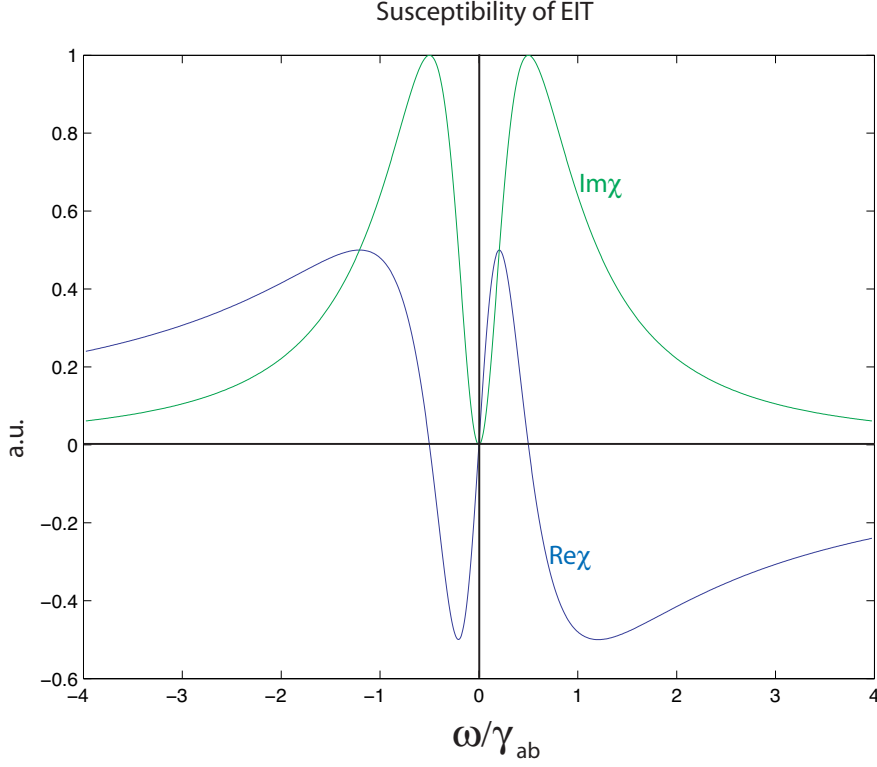


Figure 2.4: Real and imaginary parts of $\chi(\omega, z)$ for an EIT system. This system has $\gamma_{ab}/2\pi = 3$ MHz, $\gamma_{bc} = 1$ kHz and $\Omega_c = 0.5\gamma_{ab}$. The smaller coupling value causes the transparency window to decrease and the slope of the refractive index to increase.

$$(n + i\kappa)^2 = 1 + \chi$$

$$n = \mathcal{R}(\sqrt{1 + \chi}) \quad (2.6)$$

$$k = \mathcal{I}(\sqrt{1 + \chi}) \quad (2.7)$$

The refractive index describes the phase velocity of the the electromagnetic field. This is the speed at which individual wave peaks move.

Information, however, travels at the group velocity of a medium. The group velocity describes the velocity of the wave envelope and is given by;

$$v_g = \frac{c}{n + \omega \frac{\partial n}{\partial \omega}} \quad (2.8)$$

So, the steepness of the real part of the susceptibility in corresponds to a slow group velocity. As the slope gets steeper, the group velocity decreases. Figure 2.2 shows the susceptibility if there where no coupling beam, and hence no EIT effect. Figure 2.3 shows the susceptibility for a strong coupling beam. There is a large bandwidth of transmission. Figure 2.4 shows the susceptibility for a lower coupling beam strength. The group velocity is much slower but the bandwidth of transparency is less. This is one of the trade-offs with EIT. Large bandwidth corresponds to small delay time, and small bandwidth corresponds to a large delay time.

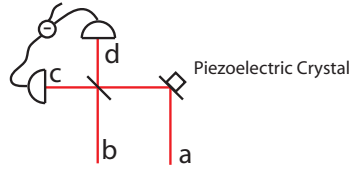


Figure 2.5: A Homodyne Detector

2.3 Homodyne detectors

The amplitude and phase of light are conjugate variables. It is only possible to measure one or the other without incurring a noise penalty. Amplitude quadrature is defined by;

$$X^0 = a + a^\dagger$$

Phase quadrature is defined by;

$$X^{\pi/2} = i(a - a^\dagger)$$

In this experiment I detect δX^0 at the sideband frequencies.

A homodyne detector is used to make measurements of either the amplitude or phase quadrature of light. Consider two input optical fields, a and b , incident on a 50-50 beam-splitter. The two optical fields will interfere and two beams, c and d will result. The detectors measure the fields c and d given by;

$$c = \sqrt{\frac{1}{2}}(a + b)$$

$$d = \sqrt{\frac{1}{2}}(a - b)$$

Each detector has a photocurrent proportional to $c^\dagger c$ and $d^\dagger d$ respectively.

$$c^\dagger c = \frac{1}{2}(a^\dagger a + b^\dagger b + a^\dagger b + b^\dagger a)$$

$$d^\dagger d = \frac{1}{2}(a^\dagger a + b^\dagger b - a^\dagger b - b^\dagger a)$$

The difference of the detector photocurrents is then;

$$c^\dagger c - d^\dagger d = a^\dagger b + b^\dagger a$$

The field can be written in terms of a mean and fluctuations about the mean. $a = A + \delta a$ where $A = \langle a \rangle$. Also, as the two beams come from the same source, they must have the same frequency. We can write a phase difference θ between the fields.

$$a = (A + \delta a)e^{i\theta}$$

$$b = B + \delta b$$

$$c^\dagger c - d^\dagger d = \left((A + \delta a^\dagger)e^{i\theta}(B + \delta b) + (B + \delta b^\dagger)(A + \delta a)e^{i\theta} \right)$$

Making the approximation that the fluctuations are small compared to the mean field, the δb and δa cross terms are discarded.

$$c^\dagger c - d^\dagger d = 2AB \cos \theta + A(\delta b e^{-i\theta} + \delta b^\dagger e^{i\theta}) + B(\delta a e^{i\theta} + \delta a^\dagger e^{-i\theta})$$

Now in the case of the homodyne detector, one input is the signal. The other input is the local oscillator and is much larger than the signal. So we assume $A \gg B$.

$$c^\dagger c - d^\dagger d = 2AB \cos \theta + A\delta X_b^\theta$$

By varying θ either the amplitude or phase quadrature can be detected. Experimentally θ can be adjusted by applying a voltage to a piezoelectric crystal mounted on a mirror where the mirror reflects one of the input beams. See figure 2.5. Changing the piezoelectric crystal changes the path length difference between the two beams.

2.4 Conditional Variance

A signal on a laser field will be written as \hat{X} in the time domain and as \tilde{X} in the frequency domain. $\delta\tilde{X}$ refers to the AC component of the field.

The variance of a signal describes how much power is in the fluctuations about the mean field and is given by;

$$V = \langle |\delta\tilde{X}|^2 \rangle$$

The correlation between two signals \tilde{X}_1 and \tilde{X}_2 is given by;

$$C_{1,2} = \frac{\langle \delta\tilde{X}_1 \delta\tilde{X}_2 \rangle}{\sqrt{V_1 V_2}}$$

A correlation of 1 means that the two signals are the same.

The correlation, however, will show two signals as less than perfectly correlated if one signal is attenuated with respect to another.

The conditional variance is a measure of signal similarity independent of attenuation. It is defined by;

$$V_{1|2} = V_1(1 - C_{1,2}^2) \quad (2.9)$$

If the two signals are the same it can be seen that the conditional variance is zero and if the two signals are uncorrelated then the conditional variance is just the V_1 .

This is equivalent to the following definition. We want find the difference between to signals, and we want it to be independent of the scale of the two signals. So minimise the difference between two signals after scaling one of them by a gain factor.

$$V_{1|2} = \min_g \langle |\delta\tilde{X}_1 - g\delta\tilde{X}_2| \rangle \quad (2.10)$$

$$V_{1|2} = \min_g \langle \delta\tilde{X}_1^2 \rangle - 2g\langle \delta\tilde{X}_1 \delta\tilde{X}_2 \rangle + g^2 \langle \delta\tilde{X}_2^2 \rangle \quad (2.11)$$

$$V_{1|2} = \min_g V_1 - 2gC_{1,2}\sqrt{V_1 V_2} + g^2 V_2 \quad (2.12)$$

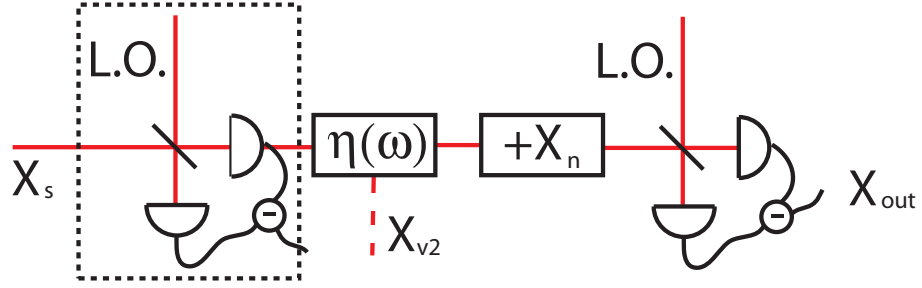


Figure 2.6: An ideal conditional variance measurement. This measurement is impossible as it requires detecting the beam both before and after the EIT mechanism.

The optimum g is given by setting the derivative to zero.

$$0 = -2C_{1,2}\sqrt{V_1V_2} + 2gV_2$$

$$g = C_{1,2}\sqrt{\frac{V_1}{V_2}}$$

Substituting this back into Equation 2.12.

$$V_{1|2} = V_1 - 2C_{1,2}^2V_1 + C_{1,2}V_1 \quad (2.13)$$

$$V_{1|2} = V_1(1 - C_{1,2}^2) \quad (2.14)$$

Which is the same as Equation 2.9. By substituting in the definition of the correlation the conditional variance is;

$$V_{1|2} = V_1 - \frac{|\langle \delta\tilde{X}_1\delta\tilde{X}_2 \rangle|^2}{V_2}$$

2.4.1 Ideal Conditional Variance

Figure 2.6 shows an experimental setup for an ideal conditional variance measurement. It requires the impossible step of detecting both the incoming beam and the outgoing beam. X_s is the signal encoded on the beam, X_{v2} is a vacuum noise term due to attenuation by EIT and X_n is any additional noise added by the EIT system not accounted for by the Langevin noise operators. X_{out} is the signal after modification by the EIT medium.

The Conditional Variance between X_{out} and X_s is.

$$V_{out|s} = V_{out} - \frac{|\langle \delta\tilde{X}_s\delta\tilde{X}_{out} \rangle|^2}{V_s}$$

$$V_{out|s} = \eta(\omega)V_s + (1 - \eta(\omega)) + V_n - \frac{|\langle \delta\tilde{X}_s (\sqrt{\eta(\omega)}\delta\tilde{X}_s + \sqrt{1 - \eta(\omega)}\delta\tilde{X}_{v2} + \delta\tilde{X}_n) \rangle|^2}{V_s}$$

$$V_{out|s} = \eta(\omega)V_s + 1 - \eta(\omega) + V_n - \frac{|\sqrt{\eta(\omega)}V_s|^2}{V_s}$$

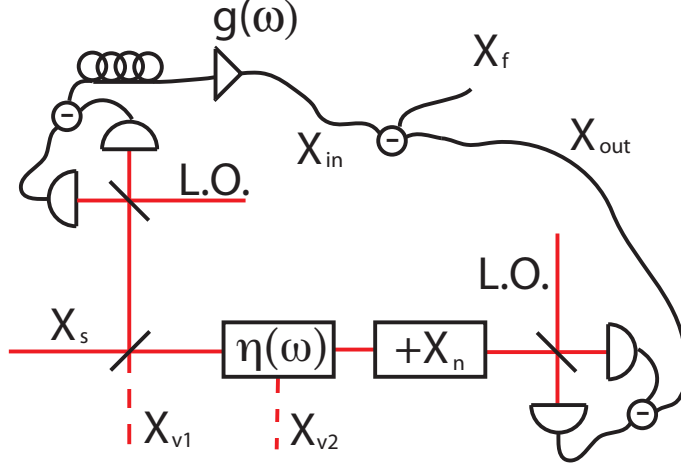


Figure 2.7: A conditional variance measurement

$$V_{out|s} = 1 - \eta(\omega) + V_n \quad (2.15)$$

2.4.2 Practical Conditional Variance Measurement

In this experiment measurements are made at two separate homodynes. An input homodyne and an output homodyne.

Figure 2.7 shows the experimental setup used to infer conditional variance. X_s is the signal encoded on the beam. X_{v1} and X_{v2} are vacuum noise terms. X_n is any additional noise added by the EIT system not accounted for by the quantum noise introduced with the attenuation. X_{in} is the signal from the first homodyne scaled by a gain factor $g(\omega)$. X_{out} is the signal from the second homodyne.

The input homodyne detects

$$X_{in} = \frac{g(\omega)}{\sqrt{2}}(X_s - X_{v1})$$

The output homodyne detects

$$X_{out} = \sqrt{\frac{\eta(\omega)}{2}}(X_s + X_{v1}) + \sqrt{1 - \eta(\omega)}X_{v2} + X_n$$

The subtraction of these two terms yields the final signal;

$$X_f = \frac{1}{\sqrt{2}}(\sqrt{\eta(\omega)} - g(\omega))X_s + \frac{1}{\sqrt{2}}(\sqrt{\eta(\omega)} + g(\omega))X_{v1} + \sqrt{1 - \eta(\omega)}X_{v2} + X_n$$

The variance of this signal is;

$$V_f = \frac{1}{2}(\sqrt{\eta(\omega)} - g(\omega))^2 V_s + \frac{1}{2}(\sqrt{\eta(\omega)} + g(\omega))^2 V_{v1} + (1 - \eta(\omega))V_{v2} + V_n$$

Now with $X_s \gg X_{v1}, X_{v2}, X_n$ the gain factor, $g(\omega)$ is set so as to minimise V_f . Therefore $g(\omega) = \sqrt{\eta(\omega)}$. Also, the variance of the vacuum noise is set to unity. So $V_{v1} = V_{v2} = 1$. This gives the final Conditional Variance measurement.

$$V_f = 1 + \eta(\omega) + V_n$$

$\eta(\omega)$ can be easily determined experimentally by applying a large signal and comparing the input and output signal strengths. The factor between these two signals will be $g(\omega) = \sqrt{\eta(\omega)}$

To infer the true conditional variance, given in 2.15, $2\eta(\omega)$ needs to be subtracted from V_s .

$$V_{out|s} = V_f - 2\eta(\omega) \tag{2.16}$$

Experiment

Multiple systems are necessary in order to make the required measurements for this experiment. Chief among these is the source of laser light used, various locking loops used, the atomic vapour cell, the data acquisition devices and the data processing techniques.

3.1 Laser Source

Various techniques are required to produce laser light of sufficient quality to conduct the experiment.

3.1.1 The Laser

The laser source we have used is a diode laser (Model number DL-100) produced by Toptica Photonics. This model of diode laser is simple in design. The laser cavity consists of a mirror and a grating. Light incident on the grating has an angle of reflection dependent on its wavelength. This effect ensures that only one frequency is resonant in the cavity. The angle of this grating is controlled by a piezoelectric crystal. The voltage to the piezoelectric crystal controls the angle of the diffraction grating and hence the frequency of the laser, however it also changes the direction of the light coming out of the laser.

The optical gain medium is the diode. This diode is kept at a constant temperature by a Peltier device. Current is supplied to the diode, and by varying this current the power of the laser can be varied. Changes to the diode current primarily affects the power output of the laser, however the current also affects the refractive index of the diode, and hence the resonant wavelength of the cavity. This means changes to laser power also change the frequency of the laser and this has to be compensated for by changing the grating angle.

When scanning the laser frequency, the motion of the piezoelectric crystal causes the diffraction grating to move back and forth as well as changing its angle. This changes the resonant wavelength of the cavity, and if nothing is done, the laser will mode hop. To compensate for this the diode current is also modulated. As explained previously this also causes a change in the resonant wavelength of the cavity. By tuning this effect to cancel out the effect of the grating movement, the laser can be frequency scanned without mode hop. The penalty is that a frequency modulation also comes with an amplitude modulation. This does not matter for our application as the laser is only scanned to find the correct transition. Once the transition is found the laser is not scanned, and is kept at that frequency.

3.1.2 Multimode Behaviour

If the diffraction grating is poorly aligned, it is possible for the laser to run with multiple frequencies resonant in the cavity. To avoid this, the spectrum of the laser was checked with a Fabry Perot Cavity. A Fabry-Perot cavity is a diagnostic device used to accurately measure frequency spectra. It is a high finesse cavity that has a tunable length due to piezoelectric control of the cavity mirrors. Scanning the length of the cavity generates a spectrum of the the incoming laser light.

It was found that the laser's threshold power was several mW above the specified values, suggesting that the cavity was not well aligned. This would have allowed multiple frequency modes to propagate inside the cavity. By changing the diffraction grating position to optimise laser threshold power and checking the spectrum of the laser with the Fabry-Perot Cavity, multimode behaviour was eliminated.

3.1.3 Feedback

Back reflections from optical components in the experimental setup can re-enter the laser cavity. This causes interference inside the laser cavity and leads to large amounts of amplitude noise. To remove this effect it is necessary to pass the laser beam through an optical isolator. This device, based on the Faraday effect, attenuates light passing through in one direction, but not the other.

3.1.4 Mode Quality

A disadvantage of diode lasers is their poor mode shape. In this experiment homodyne detectors are used. These detectors rely on interference and so it is desirable to have a clean mode shape. Also spatial effects in EIT are poorly understood and so it is desirable to have a simple mode shape propagating through the cell.

Mode quality is improved by passing the beam through a single-mode fibre. Single mode fibres only accept light in a certain mode, so it is necessary to shape the beam with lenses to improve the amount of light which can couple into the fiber. Some of the light not coupled into the fibre is reflected. Even though an optical isolator separates the laser and the optical fibre, the tiniest amount of back reflection into the laser cavity can cause multi-mode behaviour. It was therefore necessary to use a fibre with an angle-cut face to insure that these reflections did not make it back to the laser cavity.

3.2 Rubidium Structure

Natural Rubidium comes in two isotopes. The stable Rubidium-85, making up 72.2%. Rubidium-87, makes up the remaining 27.8% of natural rubidium. Rubidium-87 is radioactive, although it's half-life is 5×10^{10} years so it is effectively stable. Rubidium is a group I element, so it has a single electron in it's valence shell. This electron normally resides in the S orbital, but can be excited to the P orbital by light of approximately 795nm.

The spin and orbital angular momentum interact with the spin of the atomic nuclei, causing the energy levels of the P and S orbitals to split. The total angular momentum of an electron includes it's intrinsic spin, it's orbital angular momentum and the spin-orbit coupling with the nuclear angular momentum. The total angular momentum is denoted by F. The transition of interest in this experiment is the S orbital, F=2 to the P orbital, F=1

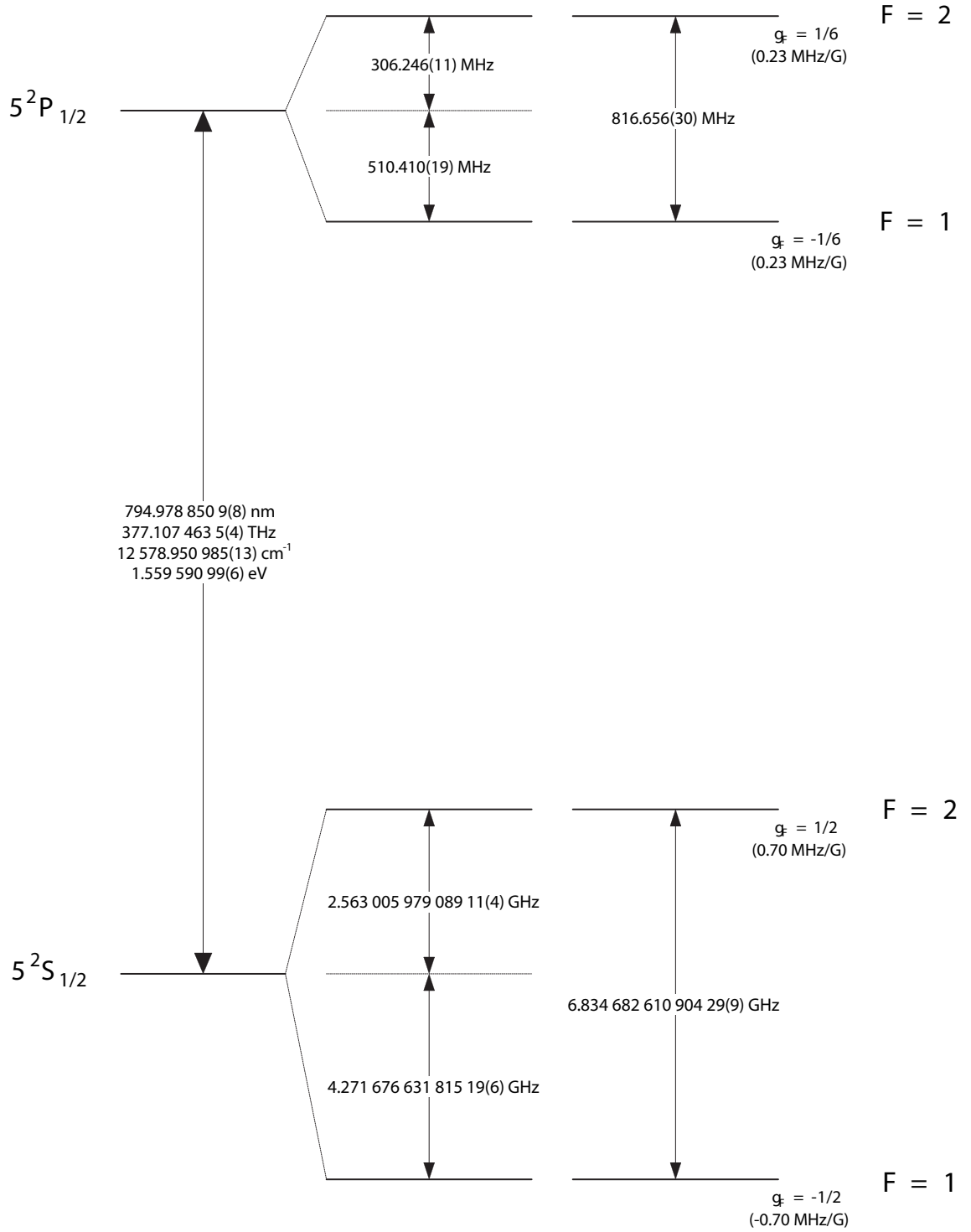


Figure 3.1: D line hyperfine level structure for Rubidium-87. This figure is reproduced from [24].

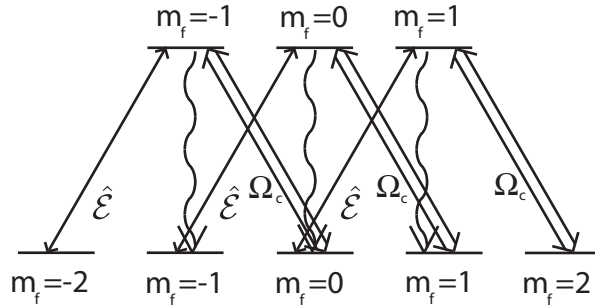


Figure 3.2: The full energy level diagram for the Rubidium 87 D1-line F=2 to F=1 transition.

in Rubidium-87. Other transitions of similar energy are present in both Rubidium-87 and Rubidium-85. Figure 3.4 in shows Saturated absorption spectra for both the transition of interest and other nearby transitions.

Figure 3.1 shows the atomic level structure for rubidium.

It should be noted that the transition has a net change of angular momentum for the electron. To satisfy conservation of angular momentum, the atom will absorb circularly polarised light.

It can also noticed that the lower energy state has an angular momentum of $F = 2$ which has five projections $m_f = -2, -1, 0, 1, 2$, and that the upper level has an angular momentum of $F = 1$ giving three projections $m_f = -1, 0, 1$. This should give an energy level diagram like Figure 3.2.

However the fact that the coupling beam, Ω_c , is much stronger than the probe beam, \mathcal{E} , means that the population pumped into the three far left states, $m_f = -2, -1$. When pumped into these three states, that population is effectively in a three level λ structure as shown in Figure 2.1.

3.3 Laser Frequency Control

It is necessary for the wavelength of the laser light to be at the correct wavelength for the atomic transitions we are interested in. Atomic transitions have very well defined frequencies, so we utilise the Rubidium transition of interest as a reference to set the laser frequency.

A cell containing natural Rubidium is used as a reference to set the laser frequency. The cell itself contains rubidium in a dual-phase state. By heating the cell. the solid rubidium sublimates into vapour and the density of the atomic vapour increases, thereby increasing the visibility of the transition. This heating is applied via heating wire wrapped around each end of the cell. It is necessary to heat the cell from the ends so that the faces of the cell are hotter than the sides. If this where not the case then the vapour would condense onto the end faces of the cell, plating them with Rubidium and turning the cell opaque.

Because the Rubidium is a hot gas, it is difficult to resolve the transition of interest due to Doppler Broadening.

3.3.1 Doppler Broadening

Atoms in the gas have an average speed, which can be found via statistical mechanics. The average velocity for an ideal gas molecule is taken from Equation (8.7) in [25]

$$\bar{v} = 2230ms^{-1} \sqrt{\frac{T \times 1 amu}{300K \times M}}$$

A rough calculation, taking $T = 330K$, $M = 85.5 amu$ gives;

$$\bar{v} \approx 250ms^{-1}$$

So, some atoms are moving with respect to the laser and they will see the laser light as doppler shifted. This effect is the same as when a siren sound higher pitched coming towards you than when it is moving away from you. The effective frequency of laser light observed by these atoms is;

$$f' = (1 + v/c)f$$

Where f' is the frequency the atom observes, v is the atoms velocity and f is the laser frequency. Using Equation (8.6) in [25], the FWHM of the Doppler broadened transmission peak is;

$$\begin{aligned} \delta\omega_D &= 2\sqrt{\ln 2} \frac{\bar{v}}{c} \omega_o \\ \delta\omega_D &= 2\sqrt{\ln 2} \frac{\bar{v}}{\lambda_o} \\ \delta\omega_D &\approx 520MHz \end{aligned}$$

As we require a greater accuracy for the frequency than 500MHz, we must use a technique called Saturable Absorption Spectroscopy to bypass the Doppler broadening effect.

3.3.2 Saturated Absorption Spectroscopy

It is necessary to keep the laser at the frequency of the Rubidium-87 D1-line 2-1 transition. To do this, it is first necessary to detect this transition's frequency, with greater accuracy than the 500MHz Doppler broadening would normally allow. The technique for doing this is called Saturated Absorption Spectroscopy. Figure 3.3 shows the experimental setup to conduct a saturated absorption measurement.

Light from the laser first passes through a polarising beamsplitter. It then passes through lenses. These lenses are not necessary, but serve to expand the beam so as to increase the number of atoms the beam interacts with. The expanded beam passes through the gas of rubidium atoms, exciting those atoms it is resonant with. The beam then passes through a quarter waveplate, reflects off a mirror and then returns through the quarter waveplate. The net effect of this is to convert the beam from horizontally polarized to vertically polarised. The beam then passes back through the cell.

In this setup, there are two counter-propagating beams in the cell. If the laser beam is slightly detuned (δ) from the atomic transition (ν), then atoms with velocity in the direction of beam propagation equal to $\pm c\delta/\nu$ will be resonant with one of the counter propagating beams. This means that only those atoms that are stationary will be resonant with the beams going in both directions. When the laser is on resonance, it will saturate

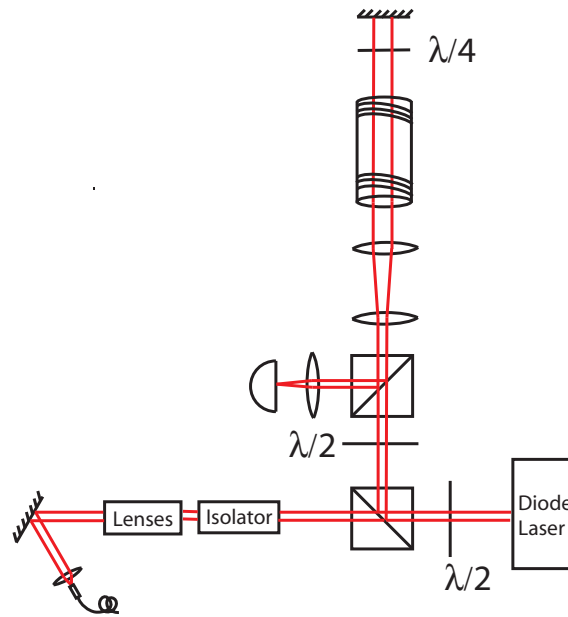


Figure 3.3: Saturated Absorption Setup

some of the material going in the forward direction and will see a transparency when returning in the other direction. If the Laser is off resonance, it will be attenuated in both directions. The corresponding peak in transmission is no longer affected by the Doppler broadening and is consequently much narrower.

After passing back through the cell, the beam is shrunk by the lens, and then reflects off the polarising beam-splitter to hit the detector. This reflection is the reason that the quarter waveplate was used to rotate the polarisation to horizontal.

The setup so far gives a peak in light on the detector when the laser is on resonance with the Rubidium transition. This effect needs to be converted into a signal which can be applied to the piezoelectric crystal to adjust the wavelength of the laser.

3.4 Locking Loops

Saturated Absorption gives information of where the laser frequency is relative to the Rb transition of interest. This information must be converted into a control signal to hold the laser frequency on the resonance. This process is called locking, the system is said to be 'locked' when the feedback loop is turned on and working correctly. One example of locking in this experiment is locking the laser's frequency to the Rubidium-87 D-line 2-1 transition.

The system will generally drift away from a required condition due to thermal or acoustic effects. Some quantity will be a maximum when the system is at the required point. In the case of the laser frequency, the quantity is the amount of light transmitted through the rubidium cell. It is locally maximum when the laser is on transition.

Figure 3.4 shows a modulation signal $S_{in}(t)$ being applied to some arbitrary function which has a local maximum. The modulation frequency is faster than thermal or acoustic drift. The function itself, $M(\nu)$ can be written as a Taylor expansion about the mean value of the modulation signal, S_{in} . If the modulation strength of the input signal is

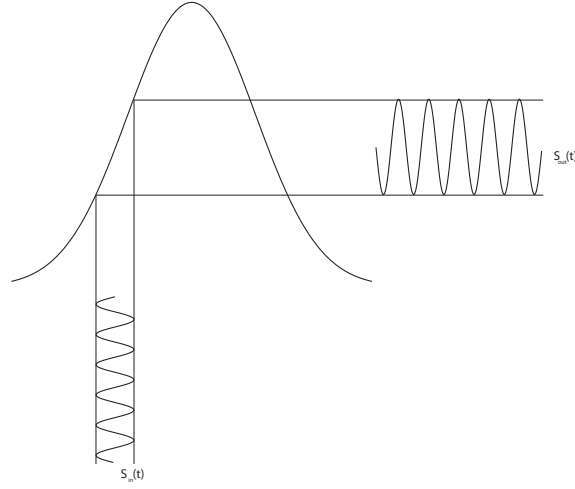


Figure 3.4: Locking

small compared to the maximum's higher order structure. Then the Taylor Series can be truncated at first order.

$$S_{in}(t) = \nu + \alpha \sin(\omega t)$$

S_{out} is the detected when a modulation is applied to $M(\nu)$.

$$S_{out}(t) = M(S_{in}(t))$$

$$S_{out}(t) = M(\nu) + M'(\nu)(S_{in}(t) - \nu)$$

$$S_{out}(t) = M(\nu) + \alpha M'(\nu) \sin(\omega t)$$

The modulation of the input signal, $\delta S_{in}(t)$ and the output signal $S_{out}(t)$ can be mixed together, then filtered with a low pass filter ...

$$LP(S_{out}(t)S_{in}(t)) = \frac{\alpha^2}{2} M'(\nu)$$

This resulting signal, called the error signal, is proportional to the derivative of $M(\nu)$. It is therefore zero crossing at the maximum and can be used as a feedback signal to keep the system on the maximum.

For a more graphical explanation, it can be seen in Figure 3.4 that if the modulation signal were to be applied at the maximum point, then it would translate to a zero output modulation. Also, it can be seen that if the modulation phase was applied on the other side of the maximum then the output would have opposite phase.

In the example of the laser locking, the error signal is applied to the piezoelectric crystal governing the frequency of the laser. At the correct frequency the error signal is zero, and so the piezoelectric crystal is at a neutral position. Should the frequency of the laser change, the error signal will cause the piezoelectric crystal to move back to the equilibrium position, thereby cancelling the drift. Figure 3.4 shows the saturated absorption spectra and locking signals for the transitions near the Rb-87 D-line 2-1 transition.

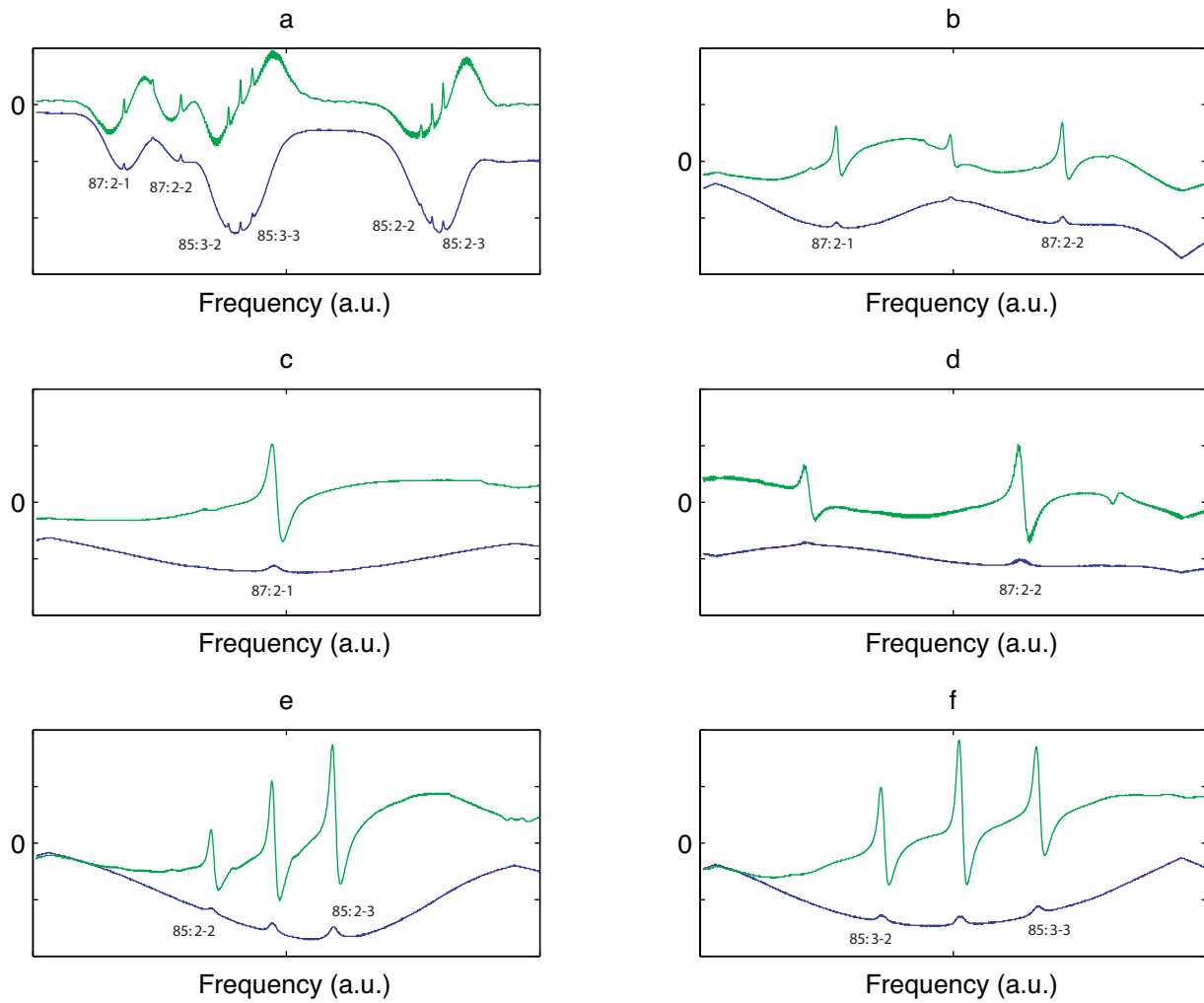


Figure 3.5: Saturated Absorption spectra and associated error signals for natural rubidium. Transitions are labelled by isotope number and angular momenta of the ground state and excited state.

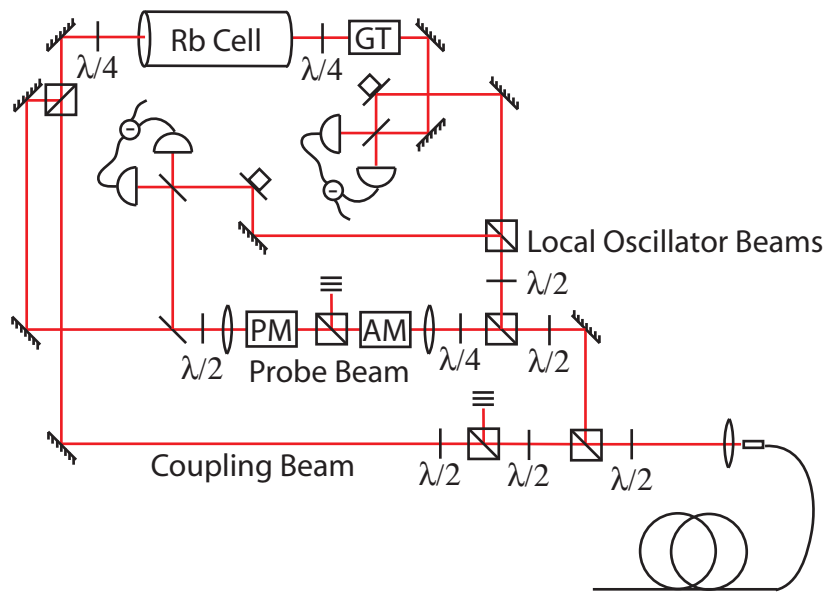


Figure 3.6: Setup for the experiment. Figure 3.3 shows the experimental setup before the optical fibre. AM is an amplitude modulator, PM is a phase modulator, GT is a Glan Thompson polariser. $\lambda/2$ is a half waveplate, $\lambda/4$ is a quarter waveplate.

3.5 Beam preparation

The mode cleaned and frequency locked laser light coming out of the fibre is separated into the local oscillator beams, the probe beam and the coupling beam. The coupling beam passes through a half waveplate and a Polarising Beamsplitter, which together act as a variable attenuator.

In this experiment a signal is applied to the probe beam. This is done with electro-optic modulators. A phase modulator and an amplitude modulator are both used. The phase modulator was only used to provide the modulation necessary for locking the homodyne detectors. It can however be used to encode phase information on the beam and examine the amount of noise added to the phase quadrature of the delayed signal.

The amplitude modulator applies the signal to the beam. It works by changing the polarisation of the light and then passing the light through a polarising beamsplitter.

Due to the small aperture size of the modulators, it was necessary to focus the beam down into the modulators using a pair of lenses.

3.6 Rubidium Cell

Before entering the Rubidium cell, the probe and coupling beams are recombined on a polarising beamsplitter so that they copropagate through the cell. The cell itself is contained in a plastic shell through which hot water is pumped. This heating increases the concentration of rubidium in the vapour phase and serves to keep the cell at a constant temperature. The water is pumped and heated by a Julabo unit.

The entire plastic shell is itself contained in a μ metal sleeve to eliminate any magnetic fields. Magnetic fields would be undesirable as they would cause the atomic energy levels to shift, changing the frequency of the transition. The ends of this sleeve are covered with tape, except for a hole small enough to admit the beams. Covering the ends with tape

serves to reduce thermal air currents caused by the hot cell.

Thermal convection causes major problems with the output homodyne due to the varying phase shift caused by changes in refractive index in the hot and cold air.

Before the Rubidium cell, quarter waveplates rotate the probe and coupling beams into orthogonal circular polarisations. After the Rubidium cell, a quarter waveplate restores the coupling and probe beam to their original polarisations. The coupling beam is then extinguished by a Glan Thompson polariser.

One problem associated with this setup is that the Coupling beam can undergo polarisation self-rotation [26] while passing through the atomic medium. This was checked for in each experiment, always the amount of self-rotated Coupling Beam was less than the amount of transmitted probe beam.

One proposed source of noise in the system is decoherence caused by atom-atom collisions within the gas. To test this hypothesis, two cells were used. The first cell consisted of pure Rubidium-87. The second cell had a buffer gas added to the cell. The idea of a buffer gas is that instead of colliding with other Rubidium atoms and decohering, the Rubidium atoms mostly collide with buffer gas atoms. The buffer gas is a noble gas that does not cause a decoherence upon collision with Rubidium.

3.7 Homodyne Detectors

As outlined in the theory section, homodyne detectors can be adjusted to detect either phase or quadrature. Experimentally, thermal and acoustic effects will change the phase difference between the local oscillator beam and the signal beam. To keep this phase difference constant, it is necessary to lock the homodyne detectors.

As given in the theory section, the subtraction of the photocurrent from the two detectors in the homodyne detector is;

$$c^\dagger c - d^\dagger d = AB \cos \theta + 2A\delta X_b^\theta$$

When θ is zero, amplitude quadrature is being detected. When θ is $\pi/2$ phase quadrature is being detected. The DC component of the signal is a maximum, so to lock to amplitude quadrature involves locking to a maximum, similar to the laser frequency locking.

A modulation at 9.8MHz is applied to the phase modulator. This modulation is picked up by the homodyne detector. The error signal is generated as described before and is fed back into the piezoelectric crystal governing the phase of the local oscillator beam. This mechanism ensures that the phase between the local oscillator and the input signal remains constant and that the amplitude quadrature is always being detected.

As homodyne detectors rely on interference, it is necessary for the probe beam to have the same mode and to have the same polarisation as the local oscillator. The polarisation and mode-matching does not have to be perfect, the imperfections correspond to a reduced visibility of the interference.

3.8 Data Acquisition

3.8.1 Network Analyser

A network analyser is a device which outputs a swept sine wave. The model used was a Anritsu MS4630B. This output signal is applied to some part of the experiment and then a signal is fed back to the network analyser. The network analyser can calculate the phase shift between these two signals and hence the delay. It is convenient because it can take data on both the delay and transmission of the EIT as a function of frequency very quickly. This speed makes it practical to take this data for a range of probe power and coupling beam power values. This large amount of data makes it possible to build up the contour plots of transmission and delay vs frequency and power shown in the results section.

3.8.2 PXI

The PXI is a fast digital oscilloscope that can capture data on four channels. The model used was a National Instruments PXI-1042. In this experiment I've only used two of those channels, one for the input homodyne and the other for the output homodyne detector. Data acquired through this system can be post processed to obtain information on delay, attenuation and conditional variance. The advantage of this system over the Network Analyser is that it can deliver a conditional variance measurement. As explained in the theory section, this is the quantity used to infer added noise in the system.

A disadvantage of the PXI system is that it has a limited dynamic range. With only 8-bit analogue to digital converters, the number of values the digitisers can take is 2^8 which corresponds to a dynamic range of 24dB.

3.9 Data Processing

Data taken on the PXI needs to be processed before it can be usefully interpreted. Matlab is used to analyse the data. It is theoretically possible to modulate the probe beam with broadband noise and get all the required information from post-processing of this signal. The noise could just be filtered at each frequency of interest and delay, attenuation and conditional variance could then be calculated. This type of process would allow a broad range of information to be inferred from one single shot measurement. However, it was necessary to take data at discrete frequencies by modulating the probe beam at stepped frequencies and taking data at each frequency. Single-shot measurement proved to be too noisy due to limited amounts of data.

3.9.1 Delay measurement

To calculate the delay between the input and the output signals, data was filtered with a gaussian filter at the frequency of interest. The filter serves to remove noise at any frequencies other than the frequency of interest. Signals were then multiplied together in the Fourier domain and then transformed back into the time domain. This is effectively a convolution between two sine waves. The maximum of this convolution corresponds to the time delay.

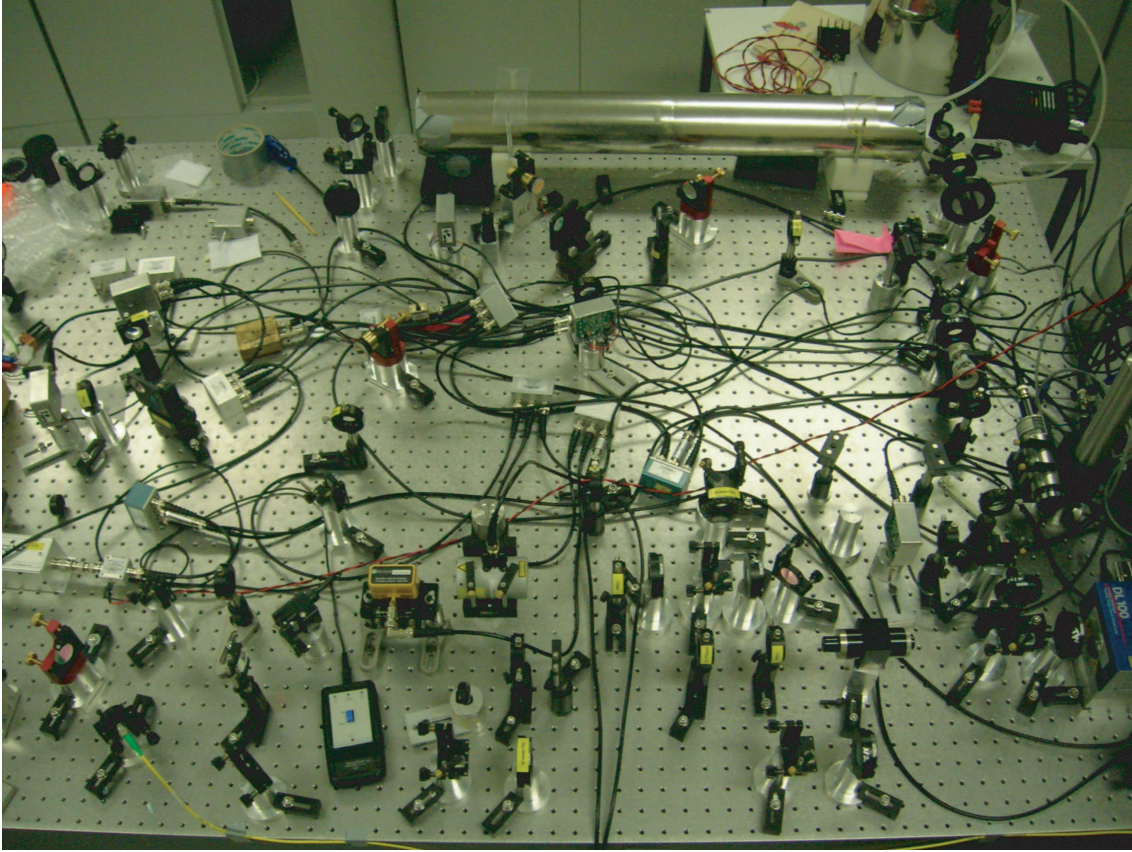


Figure 3.7: Experimental Setup.

3.9.2 Normalisation to Quantum Noise

By blocking both the probe and coupling beams, only the local oscillator beams are incident on the homodyne detectors. The two photodetectors in each homodyne setup are balanced, so the subtraction of the two photocurrents gives the quantum noise. This noise was recorded using the PXI to create a noise benchmark. The quantum noise data was filtered with a gaussian filter at each frequency, then the variance was calculated at each frequency. This was calculated for each homodyne. The resulting power vs frequency data is used to normalise the measurements of the conditional variance.

3.9.3 Conditional Variance and Attenuation

The algorithm used to find the conditional variance and the attenuation comes from Equation 2.10. Data is first shifted by the calculated delay, then filtered with a Gaussian filter. Data from each homodyne detector is then divided by the relevant quantum noise data to get the correct normalisation. A minimisation procedure based on 2.10 then finds the conditional variance, $V_{1|2}$, and the corresponding attenuation, g .

Due to the difference between an ideal and a practical conditional variance measurement, a factor of $2g^2$ must be subtracted. The reasons for this are explained in section 2.4.

Results

4.1 Laser Noise

Before taking any EIT data, the laser was characterised. Laser light was split with a 50-50 beamsplitter and imaged onto two balanced detectors. A spectrum analyser was used to find the noise spectrum of the laser. These results are shown in Figure 4.1. Also shown is the quantum noise and the dark noise of the detectors. Dark noise is simply the signal from the detectors when no light is incident upon them. Quantum noise is found by subtracting the signals from the two detectors, laser noise is cancelled out and quantum noise, being uncorrelated, is detected. Laser noise is then the sum of the two detectors. The laser noise is seen to be quite flat, and about 8dB above quantum noise.

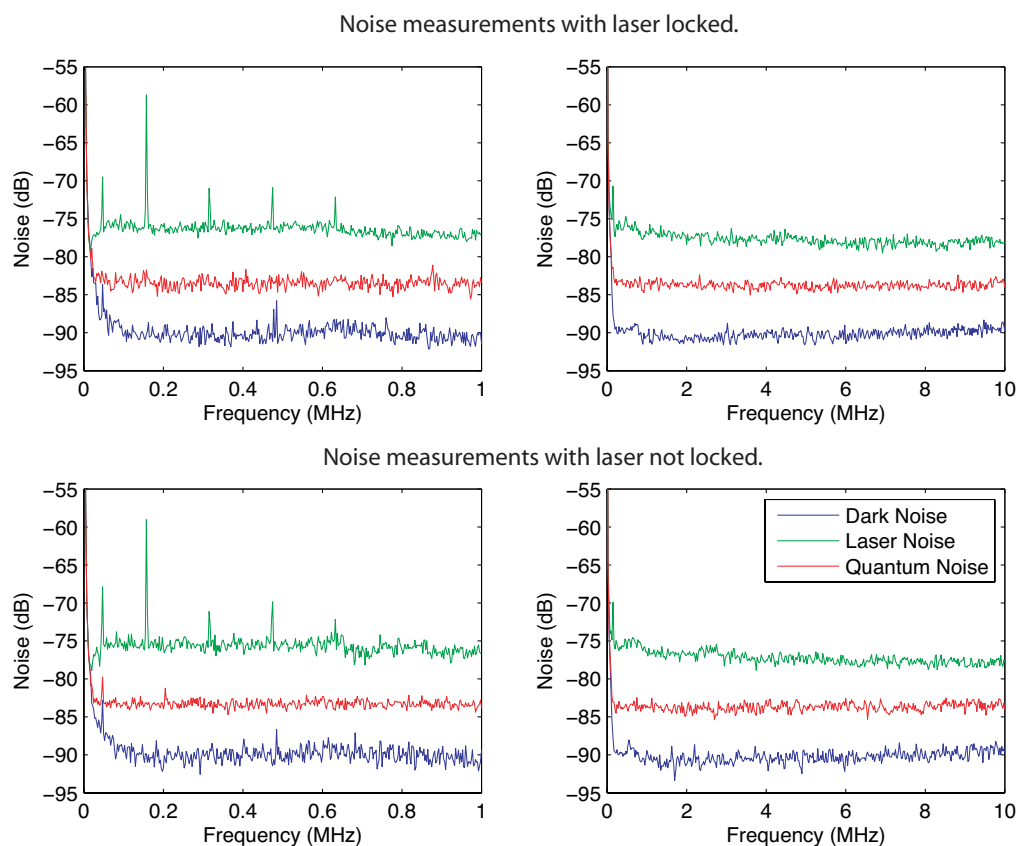


Figure 4.1: Laser noise, Quantum Noise and Detector Noise plots. 4mW of light incident on the detector out of 80mW total laser power. Two frequency scales are shown.

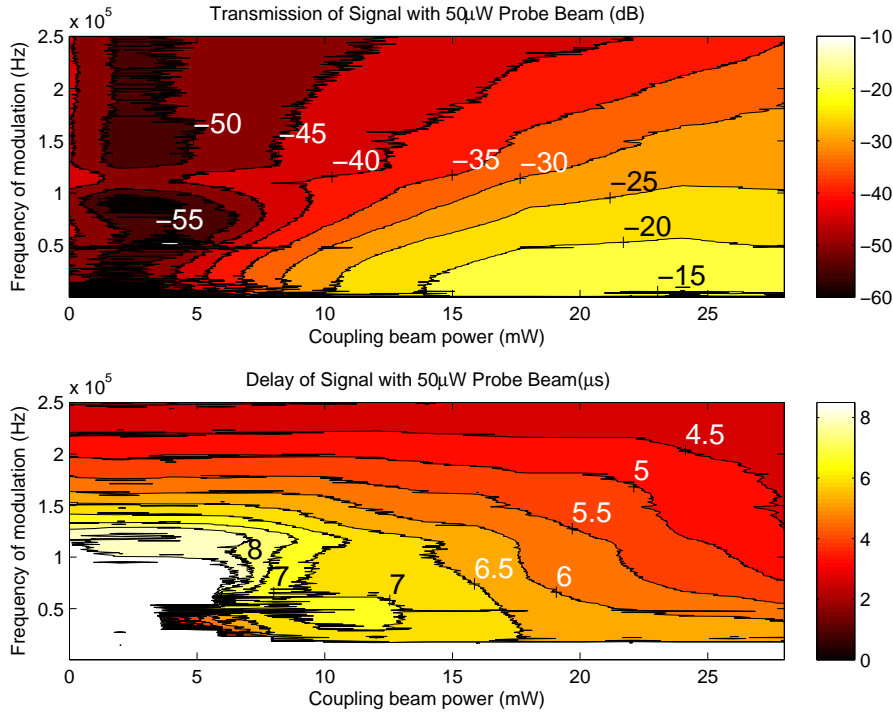


Figure 4.2: Delay and Transmission for an unbuffered Rubidium Cell. 50μW Probe beam.

4.2 Delay Measurements

By using the network analyser plots of delay vs. frequency and plots of attenuation vs. frequency can be generated very quickly. By taking this data for multiple values of the coupling beam strength, it was possible to build up a contour plot showing the parameters that correspond to the best delay. All data was taken with the Rubidium cell heated to 58°C.

4.2.1 Unbuffered Rubidium Cell

Using an isotopically purified Rubidium-87 vapour cell, delay and transmission were measured as a function of frequency and coupling beam power for three different Probe beam powers. These results are displayed in Figures 4.2, 4.3, 4.4.

Universally these plots show that with increasing frequency of modulation, the delay time decreases and the transmission decreases. This is consistent with the fact the EIT has a limited bandwidth. On all three plots the maximum delay occurs when the frequency of modulation and coupling power are both very small. This is, however, not useful because this region also corresponds to a minimum in transmission. A local maximum in delay also occurs for moderate coupling beam strengths with small modulation frequency. These local maxima in delay time corresponds to reasonable signal transmissions and so constitute a useful result.

These local maxima make sense. An increase in coupling beam strength would lead to a broadening of the transparency bandwidth. This would mean the susceptibility gradient would be shallower, corresponding to a smaller delay time. On the other hand, as the coupling beam gets smaller, the bandwidth of transparency gets smaller and starts

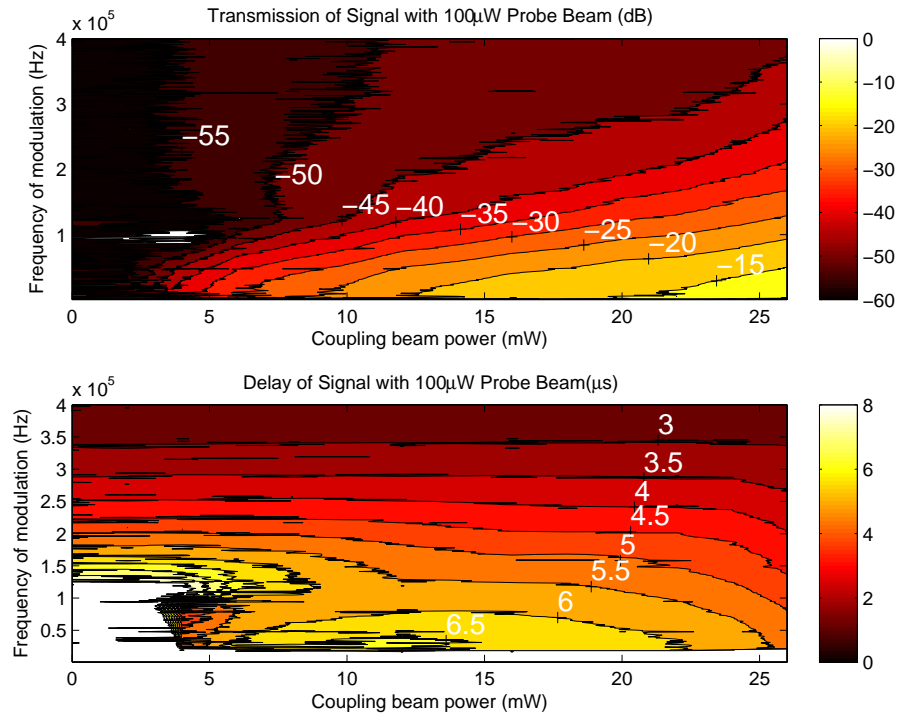


Figure 4.3: Delay and Transmission for an unbuffered Rubidium Cell. 100µW Probe beam.

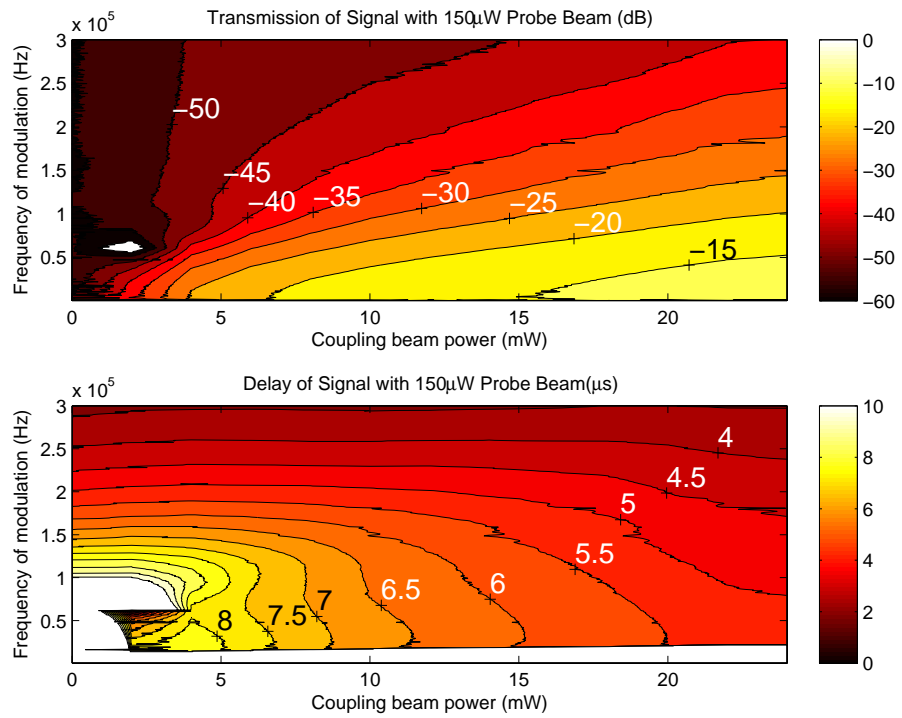


Figure 4.4: Delay and Transmission for an unbuffered Rubidium Cell. 150µW Probe beam.

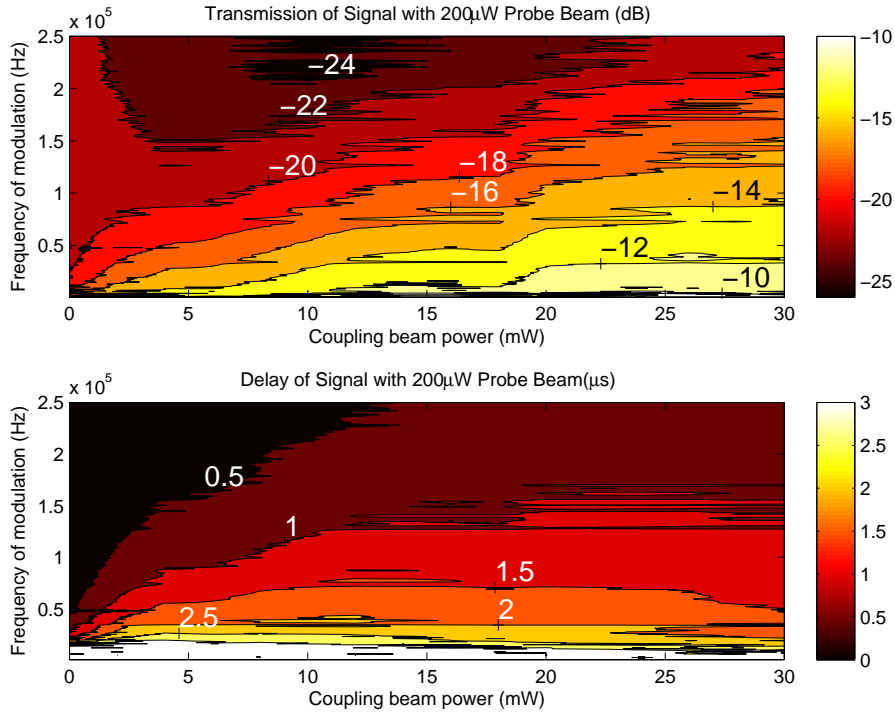


Figure 4.5: Delay and Transmission for a buffered Rubidium Cell. 200μW Probe beam.

to attenuate light at the sideband frequencies.

Importantly, $8\mu\text{s}$ delay is observed in Figure 4.4 at a modulation frequency of 50kHz. As 50kHz corresponds to a period of $20\mu\text{s}$, the $8\mu\text{s}$ satisfies the criteria of the delay time being greater than one quarter of a period of the modulation given by [22]. This means that the delay cannot be explained by saturable absorption effects.

4.2.2 Buffered Rubidium Cell

Isotopically pure Rubidium-87 and a buffer gas are mixed together in a vapour cell. Delay and Transmission were measured as a function of frequency and coupling beam power. These results are displayed in Figures 4.5, 4.6, 4.7. Data was also taken for 100μW probe beam, however signal was too low and the data was too noisy to be useful. It should be noted that the beam sizes of the probe and coupling beams were changed between the measurements for the buffered and unbuffered cells. As such, it is not useful to compare the power of the probe beam between the buffered and unbuffered cells.

The buffered cell shows no maximum in delay for small coupling power and small modulation frequency. Also, there seems to be no local maximum of delay time, with delay increasing for lower modulation frequencies. Other than these differences, the data is qualitatively similar to that of the unbuffered cell.

The maximum delay achieved for this cell is lower than that of the unbuffered cell. Also the overall amount of attenuation is less. Both of these effects are due to the fact that the two types of cell are operating at the same temperature, however the buffer gas cell has the buffer gas present, which takes a partial pressure. This allows less Rubidium into the vapour phase and so the number of atoms that the beam addresses is smaller. Equation 2.3, 2.5, 2.6 and 2.8 show that an decrease in N leads to an increase in the group

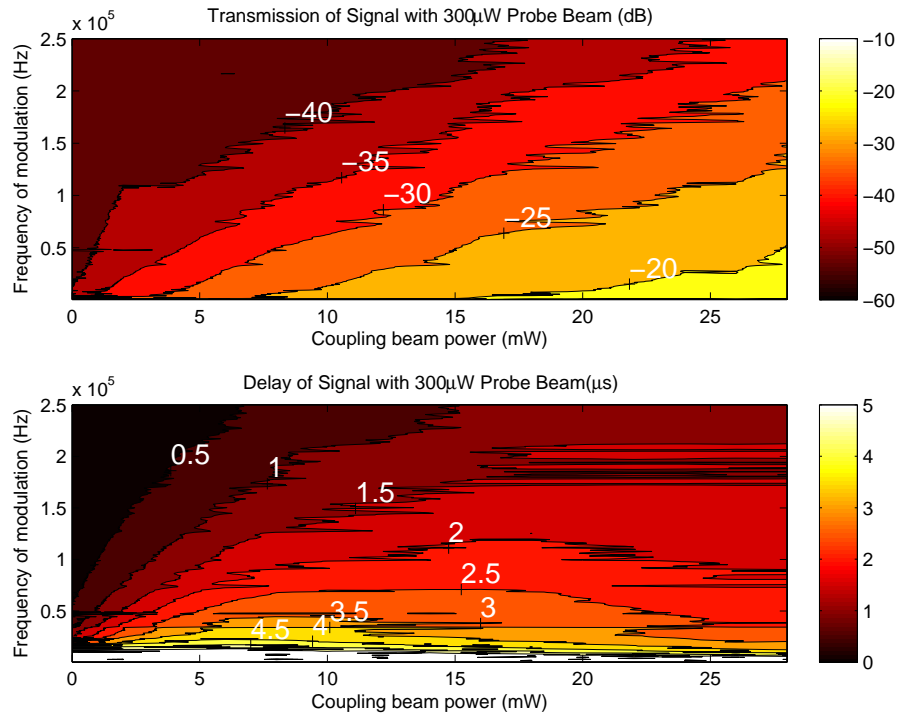


Figure 4.6: Delay and Transmission for a buffered Rubidium Cell. 300µW Probe beam.

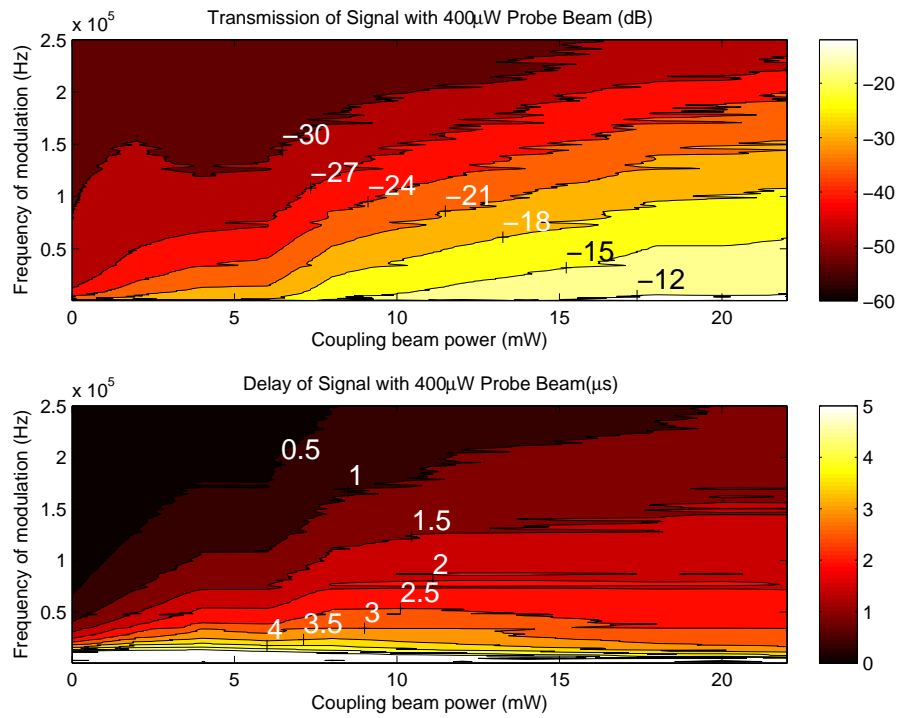


Figure 4.7: Delay and Transmission for a buffered Rubidium Cell. 400µW Probe beam.

velocity.

4.2.3 Spatial Effects

One of the non-quantitative results to come from these experiments is that EIT has a large dependence on spatial effects. Mode quality, beam alignment and the size ratio between pump and probe all had an effect on the quality of the delay achieved.

In fact, for some configurations, transmission was actually worsened by the application of the coupling beam. We believe this particular effect was caused by having the probe and coupling beam misaligned, or having the coupling beam much larger than the probe beam. This would mean that the coupling and probe beams would be operating in different spatial regions of the cell, and so they could not produce EIT. The separate regions, however, would still be coupled by the thermal drift of the atoms and so optical pumping could still occur.

Given the need to constantly adjust for thermal drift in the optical components, the beam configuration would not have been the same for both cells.

The buffered Rubidium cell did not have very flat optical surfaces, resulting in mode shape distortion after passage through the cell. This resulted in lower fringe visibility in the output homodyne detector.

Further it was necessary to tilt the cells slightly to avoid back reflections from the uncoated glass faces of the cells.

4.3 PXI Measurements and Post Processing

To characterise the extra noise added to the beam by EIT it is necessary to use the PXI. The PXI saves raw data, from which it is possible to calculate the conditional variance, which is a measure of noise. Before calculating the variance however it is necessary to calculate the delay. These calculations are done via the methods described in section 3.9.

All of the data shown in this section was taken with a coupling beam strength of 2.5mW, and a probe strength of 200 μ W. The area of the coupling beam was approximately 3.5 times greater than that of the probe.

4.3.1 Noise Characteristics of the Measurement System

Figure 4.8 shows the noise measurements of the system taken with a homodyne detector. The PXI noise is calculated by unplugging the detectors from the PXI.

The noise of the detectors is calculated by blocking all laser light from hitting the detectors. The peaks in the detector noise are not expected. It can clearly be seen that the detector noise peaks affect the measurement of quantum noise and even laser noise. The noise peaks shown in the detector noise are a big problem as they do not show up in the earlier measurement done with the spectrum analyser, results of which are shown in Figure 4.1. A more recent check of detector noise was made with the network analyser, which also showed a flat noise spectrum. This suggests that the peaks are most likely due to something in either the PXI or in the data processing methods. Experiments using the PXI should be repeated to see if these noise peaks are repeatable.

Quantum noise is calculated by putting the local oscillator beam into the homodyne detector. Any noise on the laser beam is detected equally by both detectors and is subtracted away, the only source of noise left in the detector signal is the quantum noise

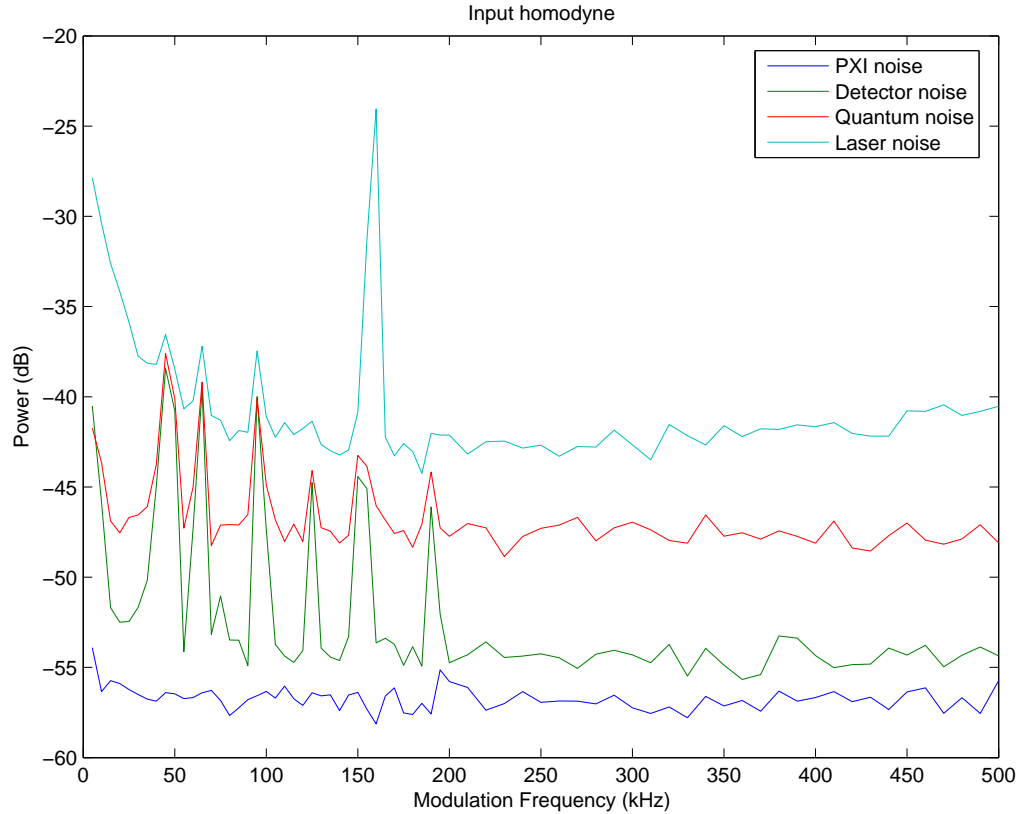


Figure 4.8: Noise measurements with 6mW local oscillator power.

introduced by the empty port of the beamsplitter. For a fuller description of this effect see Section 2.3.

Laser noise is calculated by replacing the subtracter between the two detectors with an adder. The peak at 150 kHz is highly problematic as it is more than 30dB above the PXI noise. Given that the dynamic range of the PXI is only 24dB this means that the PXI must be saturating at this point.

4.3.2 Post Processed Delay Measurements

Figure 4.9 shows the calculated delay vs frequency for probe beam. This shows the narrow bandwidth of EIT and is a good proof of principle as it shows great similarity to the data taken with the Network Analyser as presented in Figures 4.5, 4.6 and 4.7. Except for one obviously erroneous point the trend is well defined.

This delay was checked by performing the same measurements with the laser off-transition. These measurements showed a delay of zero, so there is no detectable delay in the measurement system which would need to be accounted for by an offset.

4.3.3 Post Processed Attenuation Measurements

After normalising each signal to the quantum noise, the attenuation can be calculated by finding the attenuation value that minimises the variance between two signals. This is explained for fully in section 2.4. Here, the data is not so smooth as for the delay. There are at least 8 data points which depart from any discernable trend, and there is

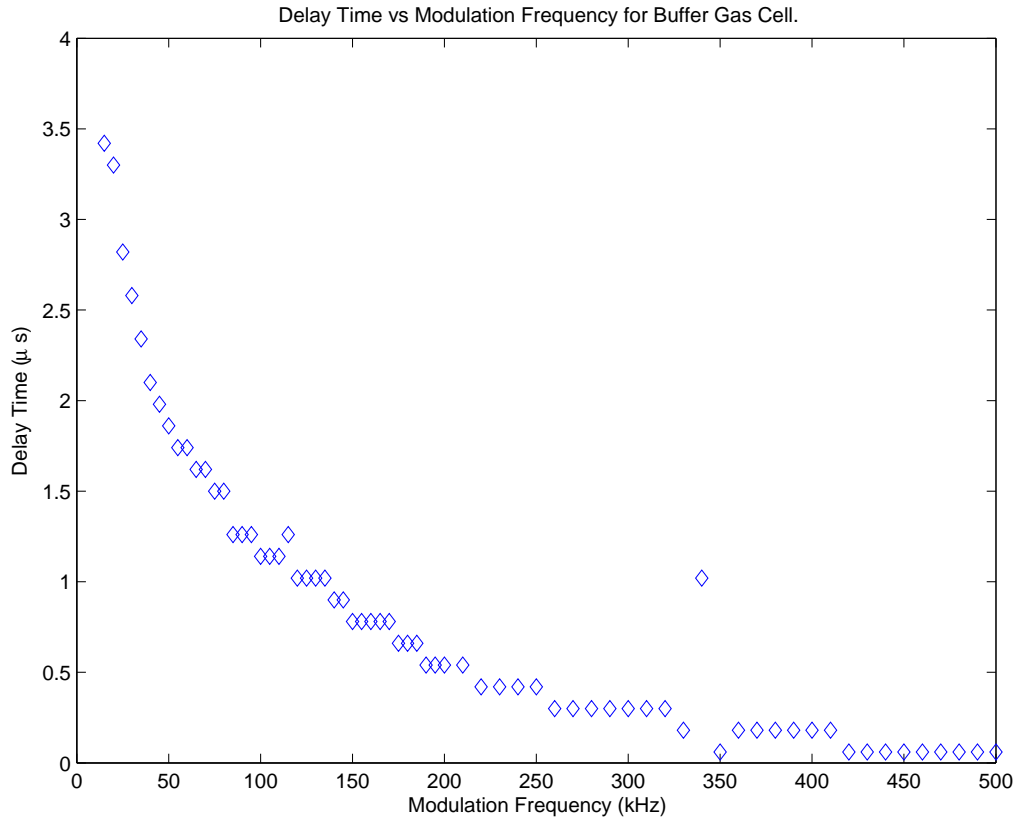


Figure 4.9: Delay versus modulation frequency for the Buffer Gas cell.

an inexplicable rise in transmission at between 300kHz and 350kHz. Finally, the signal to noise ratio for the plot is disappointingly low. Given that over 100,000 points of data were taken with the PXI, enough detail should be present to determine the gain with greater accuracy than this. When the inaccuracy on these measurements are compared to the smooth contour plots generated by the network analyser, the PXI is disappointing. Other than that, the data looks as expected, showing a drop in transmission with increasing modulation frequency.

4.3.4 Conditional Variance

The Conditional Variance is the most troubling result of all. Values in the thousands do not agree with Hsu's earlier work [17]. In fact they are hundreds of times larger than Hsu's results. One potential source of this problem is an error in the normalisation part of the algorithm described in Section 3.9.

To check for this error, I resubstituted the Quantum Noise data back into the Conditional Variance algorithm to check that the normalisation code was working correctly. Figure 4.12 shows the results of this test, and it shows that in the main part, the normalisation is correct, or at least not incorrect enough to account for the huge size of the Conditional Variance.

This leaves the process itself or the PXI as explanations for the exceptional values of the conditional variance. To properly narrow down the error, more data needs to be taken to see if the errors are repeatable.

Hsu's method of calculating the conditional variance was entirely electronic. Delay

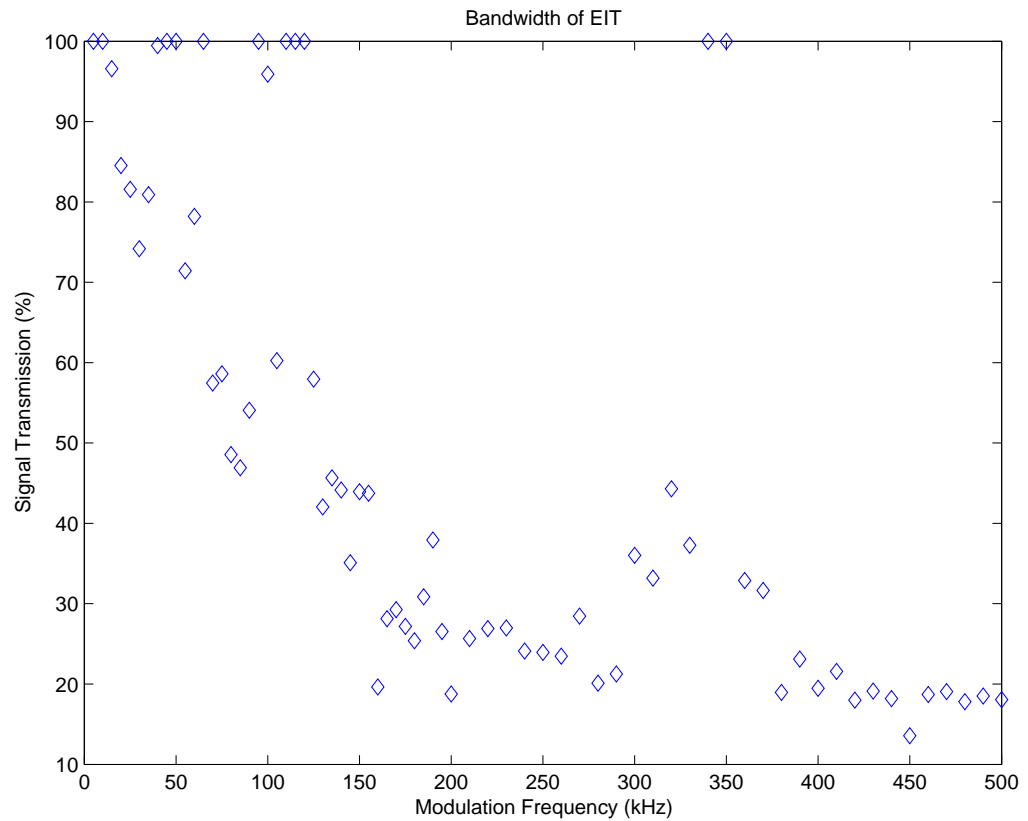


Figure 4.10: Gain calculated using post-processed PXI data.

was manually adjusted by varying the length of a BNC cable and the gain was manually adjusted to minimise the correlation between the input and output homodynes. The main advantage the PXI method has is that it is much faster. Even when modulating at discrete frequencies the data taking process is much faster. Unfortunately difficulty analysing the data prevented further data being taken.

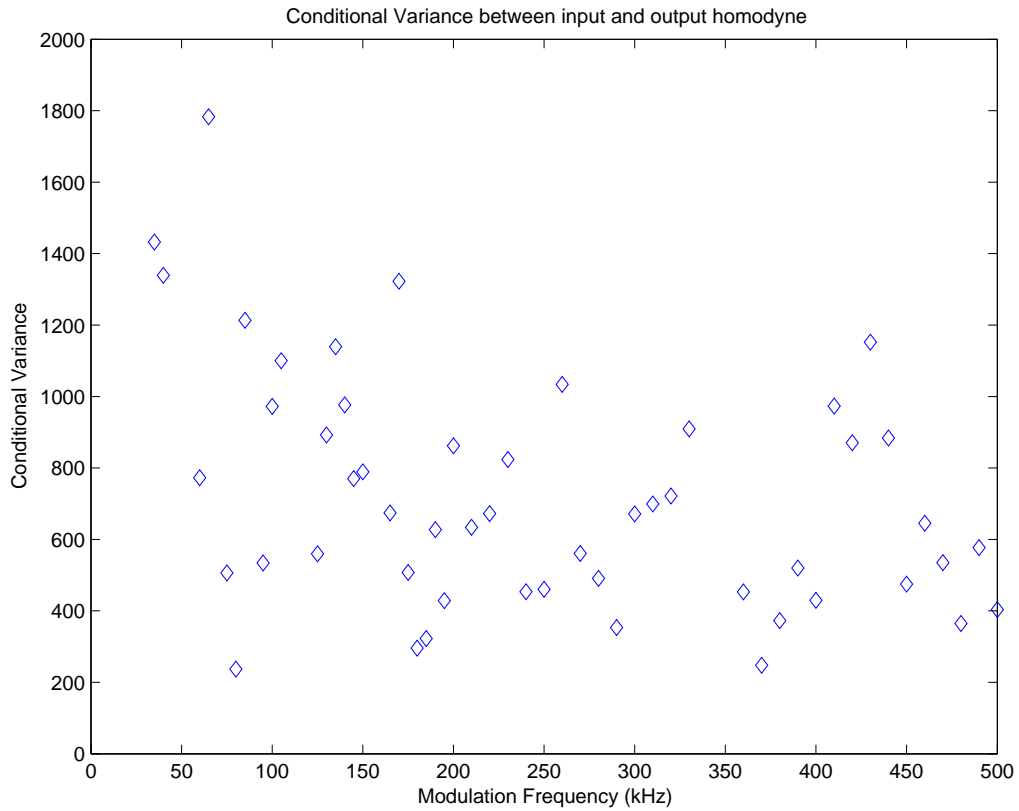


Figure 4.11: Conditional Variance, normalised to Quantum Noise and via Equation 2.16

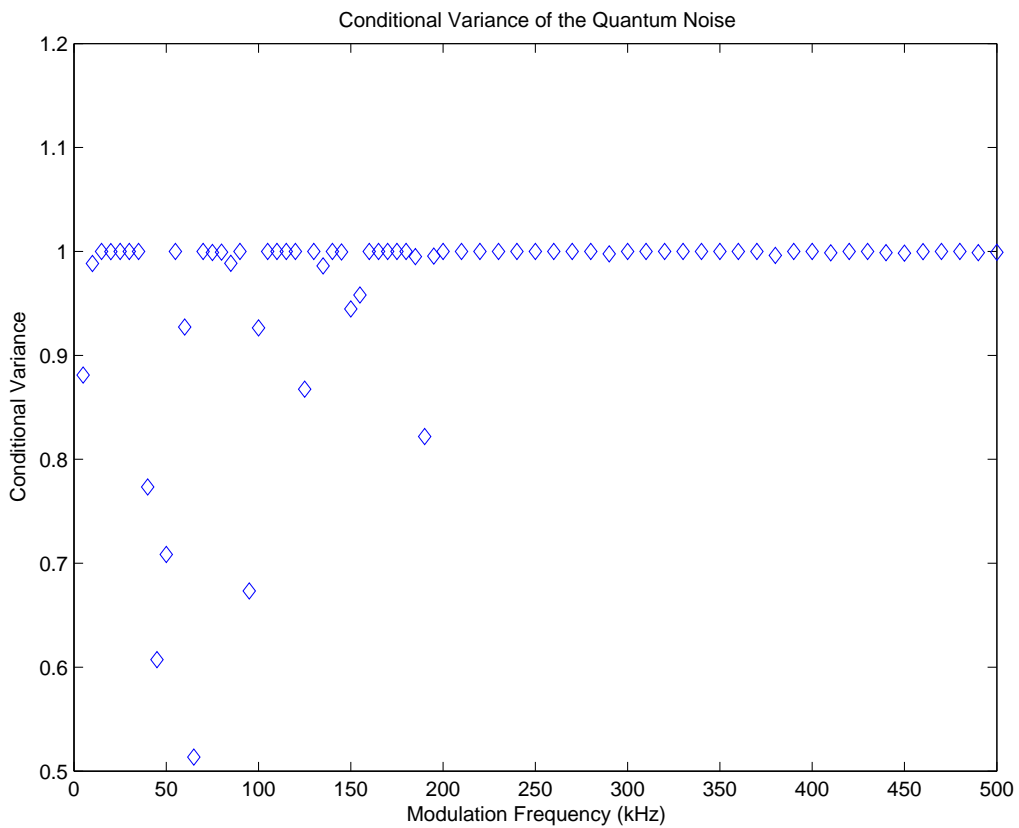


Figure 4.12: Normalisation test of Conditional Variance algorithm.

Conclusion and Future Directions

5.1 Conclusions

In this thesis, experiments have been done to build up a comprehensive characterisation of EIT signal delay and transmission as a function of modulation frequency and coupling beam power. This same method can be used with future EIT systems to find points of optimal delay, transmission and bandwidth.

Furthermore these results show that for cells at equal temperature, better delay can be achieved with an unbuffered cell. However, for a better comparison, the cells should be compared with equal atomic densities since the buffered cell will have a lower atomic density due to the presence of the buffer gas.

Attempts at making single-shot measurements to determine delay, transmission and added noise have shown that more work needs to be done to properly understand the methodology. In order to verify the noise results obtained, more measurements need to be made.¹

5.2 Future Directions

Future experiments should focus on understanding the PXI system and how to do single-shot measurements of the amount of excess noise in an EIT system. Problems to be solved include fitting all modulations inside the dynamic range of the PXI. Also the algorithms for post processing the data can be investigated. Gaussian filters and top hat filters were chosen for their simplicity although more optimal filter shapes may well exist.

Also, the high dependance on spatial effects could be investigated in more detail. In particular the hypothesis that coupling and probe beams can effect each other via the thermal drift of the atoms should be tested. Some work has already been done in this area [27, 28, 29].

¹I will be working on verification of these results over the next few months as we will possibly be preparing a manuscript for publication.

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