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THE RATINGS OF A POWER TYPE TRANSISTOR

A Thesis

Presented to

the Faculty of the Department of Electrical Engineering
University of New Mexico

In Partial Fulfillment

of the Requirements for the Degree

Master of Science in Electrical Engineering

by
Richard Lee Mann
June 1957

This thesis, directed and approved by the candidate's committee, has been accepted by the Graduate Committee of the University of New Mexico in partial fulfillment of the requirements for the degree of

MASTER OF SCIENCE

EsCastetter

may 28, 1957

Thesis committee

Bob M. Jannin

Charles & Williams Jr.

THE RATINGS OF A POWER TYPE TRANSISTOR

by Richard Lee Mann Junior Engineer

Electrical Engineering Department University of New Mexico Engineering Experiment Station

TECHNICAL REPORT EE-2
Report No. 1 on Project 56-2EE
September, 1956

The work described in this report was performed under Sandia Corporation Purchase Order No. 53-4020.

^{*}This paper is also being used as a thesis in partial fulfillment of the requirements for the Degree of Master of Science in Electrical Engineering at the University of New Mexico.

THE RATINGS OF A POWER STOR TRABELSTOR

by Hichard Lee Menn Junior Engineer

Electrical Engineering Department University of New Mexico Engineering Experiment Station

Report No. 1 on Project 66-288 September, 1956

The work described in this report, was performed under Sandia Corporation Purchase Order Wo. 55-4020.

This paper is also being used as a thosis in partial faifillment of the requirements for the Degree of Marker of Science in Slectrical Engineering at the University of New Markers

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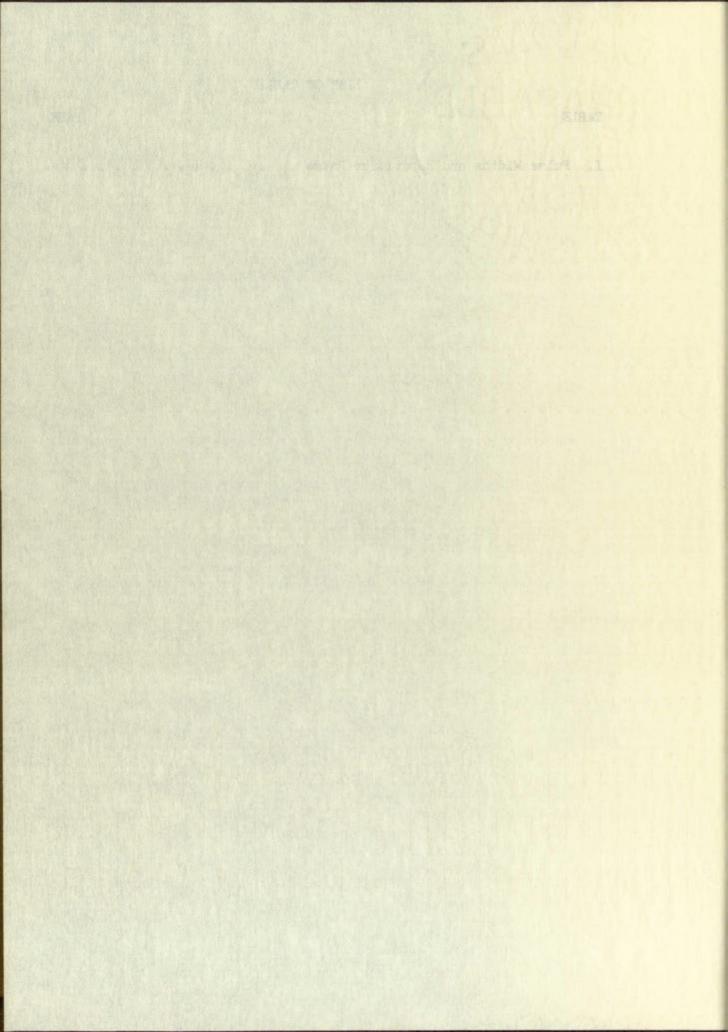
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CHAPTER I

INTRODUCTION

The transistor, first publicly announced in June 1948, is an offshoot of the germanium crystal rectifier which played an important role in World War II in radar receivers. The germanium crystal rectifier had high-inverse-voltage characteristics which were lacking in preceding type rectifiers, thus making it the center of attraction of electronic interest. By adding an electrode to the crystal rectifier in order to control the surface charges, a new semi-conducting device, the "transistor" (TRANSfer resISTOR) came into existance.

Transistors are replacing electronic vacuum tubes in many applications since they are much smaller in size and weight and the power requirements are much less. They are capable of being used in circuits to provide amplification, oscillation, pulse generation, gating, and many other functions. The miniaturization of electronic equipment has even furthered the importance of the transistor. Great efforts are being made to extend the useage of transistors by increasing their current carrying capacity and useful frequency range.

The large transistors, which are very small compared with an equivalent tube, are referred to as power type transistors. While all transistors, in a general sense, are power transistors, they have been referred to as being capable of carrying 0.5 amperes or greater.

lLeRoy A. Griffith, Germanium Power Transistors (Minneapolis-Honeywell Regulator Company, Minneapolis 8, Minnesota, Transistor Division, paper dated February 9, 1955, TR73pl.)

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Like most new devices, the standardization of terms and ratings is a slow process. Manufacturing skills and methods are continually changing in an attempt to perfect a device which can be mass-produced with reasonably similar characteristics so that they will be interchangeable. The problem of interchangeability of power type transistors has not been solved as yet which makes transistor circuit design more difficult than tube circuit design. As this paper will show, the characteristics of a type transistor may vary widely from one transistor to another.

Ratings for a type of transistor are furnished by the manufacturer. These ratings are direct current ratings only and the characteristic curves furnished are typical characteristics which may be quite useless as far as a specific circuit design is concerned.

Sandia Corporation personnel have witnessed failures of transistors in their circuitry design due to what was thought of as being large amplitude current pulses of short time duration. In an effort to establish definitely the causes of these failures and the permissible limits of these pulses, a contract was negotiated with the University of New Mexico to do the research described in this paper.

The purpose of this paper, then, is to describe the tests performed and present the results obtained in an attempt to establish the maximum permissible direct current and pulse ratings for the type CTP-1003 (Clevite Transistor Products) power transistors under typical operating conditions. The manufacturer's specifications for the type CTP-1003 power transistor are located in Appendix A.

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Tests were performed at 165°F. ambient temperatures to establish the maximum permissible ratings (i.e., current, voltage, and dissipation or combinations of these quantities applied to any terminals of the transistor) over the active operating region of the transistor characteristics. Maximum permissible ratings are defined as those values at which the transistor may operate with good stability, long life and no irreversible changes in characteristics.

The effects of duty cycle, heat sink characteristics, spread of transistor properties, ambient temperature, and transistor connection on the ratings were established for

- 1. Ambient temperature: +165°F.
- 2. Duty cycle range: Pulse widths between 10 microseconds and 10 milliseconds; pulse repetition frequencies between 60 CPS and 2000 CPS.

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CHAPTER II

THE METHOD OF TEST

I. PREPARATION OF TESTING

Each transistor was numbered with red paint such as, C-l.

"C" for Clevite to distinguish from other transistors being tested
in the laboratory.

Prior to all testing, the transistor was securely fastened to a 6" x 6" heat sink by 3 - #6 x 3/8" machine screws and placed in the temperature chamber and allowed to reach ambient temperature. A thermocouple (iron-constantan) was placed in a small groove cut in the heat sink just under the base of the transistor to measure the transistor temperature. Care was taken so that no air gap existed between the transistor and the heat sink. A picture was then taken of the transistor characteristics as the thermocouple passed through a temperature, ten degrees above ambient temperature. This is necessary since the curve tracer, operating intermittently at two amperes or better (collector current), causes a heating of the transistor, thus changing the transistor characteristics.

II. TESTING PROCEDURE

<u>Pulse testing at 165° F.</u> This test was designed for the purpose of establishing the maximum permissible pulse ratings and the effects of duty cycle on the power transistor at 165° F. ambient temperature.

Pulse widths and repetition rates shown in Table I were used in establishing the maximum pulse ratings.

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TABLE I
PULSE WIDTHS AND REPETITION RATES

Repetition Rates, CPS	60	200	600	2000
Pulse Widths	10	10	10	10
in	100	100	100	100
Microseconds	1000	1000	1000	
rijei osecolius	10000			

The pulse widths in Table I at a given repetition rate was applied to the base of the transistor connected common emitter. The pulse amplitude was then varied in steps from a minimum value to a value which destroyed the transistor. The size of the pulse steps was decreased as the temperature of the transistor increased. The amplitude of the collector pulse was held constant until a constant transistor temperature was maintained prior to advancing to the next step. Continuous observation of the input and output pulses, transistor and ambient temperatures was made throughout the pulsing test.

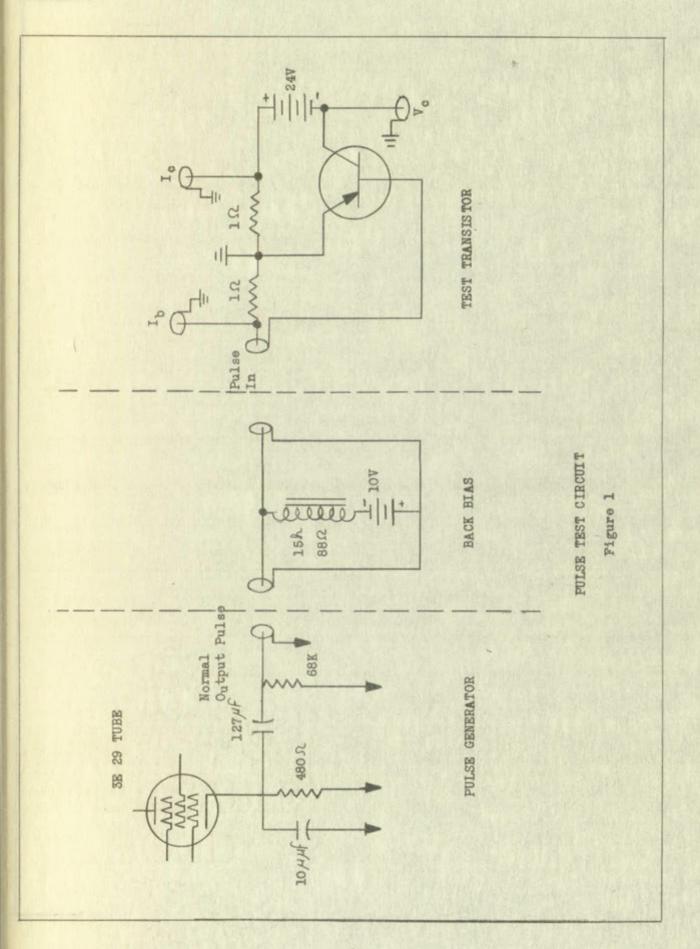
Where feasible, at least one transistor was made to fail at each of the above repetition rates and pulse widths by increasing the pulse amplitude. Where time allowed, not less than five transistors were pulsed at each repetition rate and pulse width and they were pulsed where possible beyond the point where failure existed on the preceding transistor. The transistor characteristics were checked against the original characteristics whenever there was any indication of an unusual change in output or sharp temperature change. Figure 1

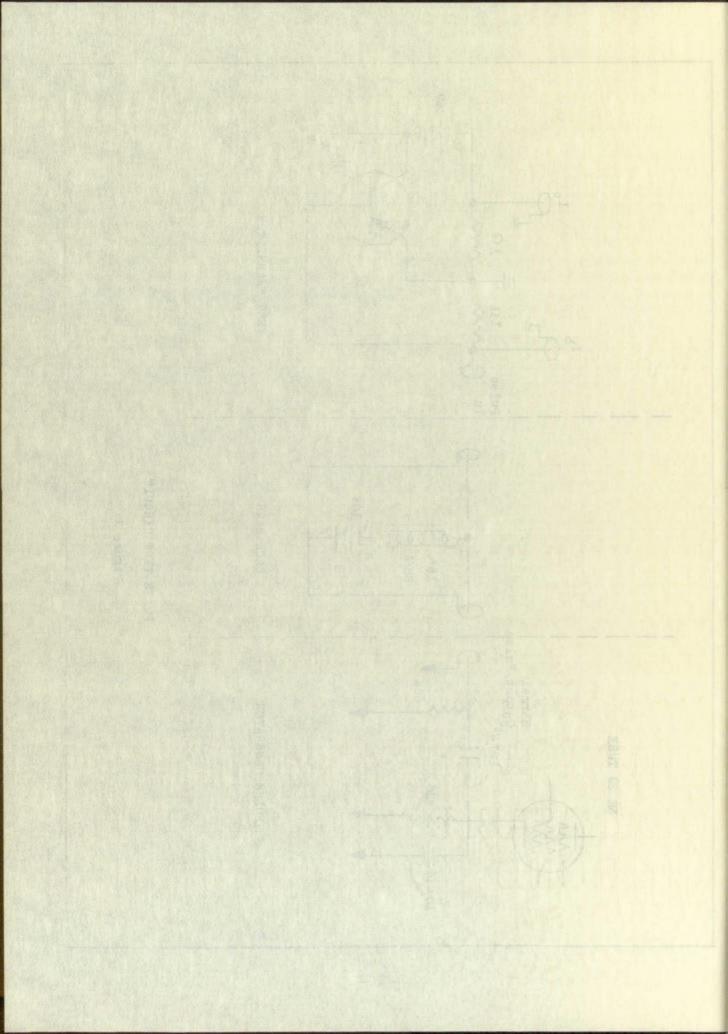
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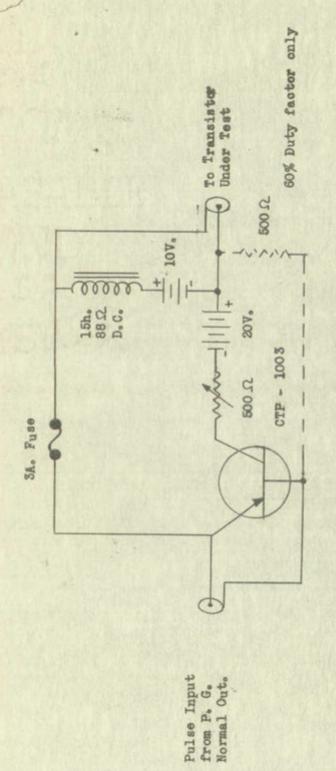
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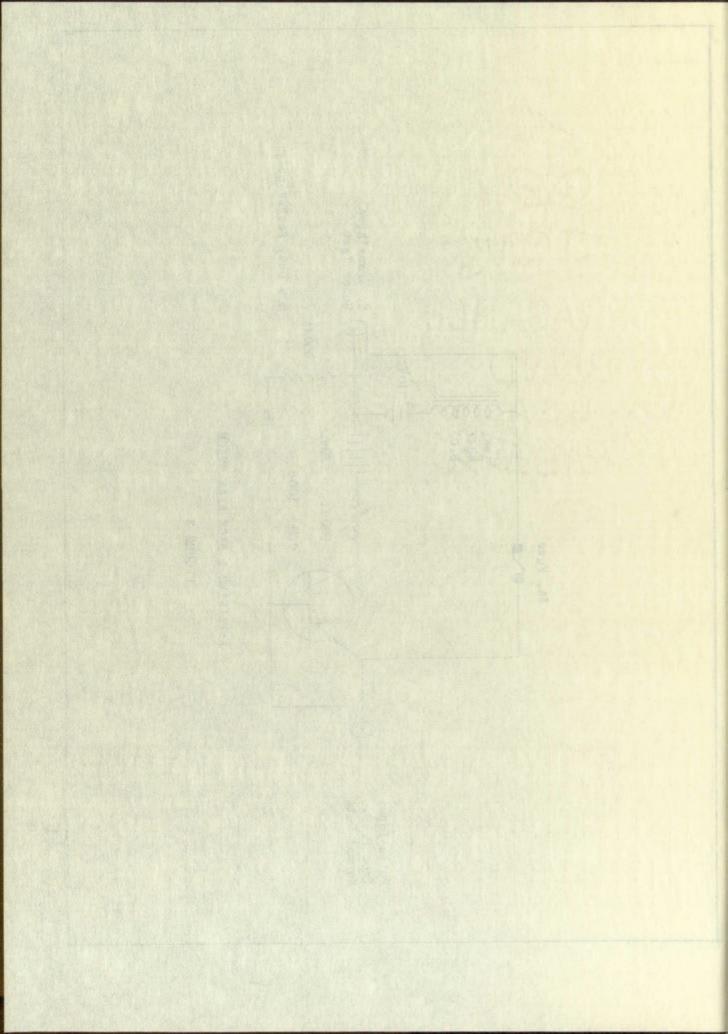






AMPLIFIER & BACK BIAS STAGE

FIGURE 2



shows the pulse testing circuit. In order to obtain greater pulse amplitudes for the 100 microsecond pulses and larger, a transistor stage of amplification was used. This circuit is shown in Figure 2.

The pulse generator's normal output pulse is limited to twentyfive percent duty factor. Where the duty factor exceeded this rating,
pulses were taken from the delay unit of the pulse generator. The
dash line and 500 ohm in Figure 2 shows the positive feedback circuit
required for the sixty percent duty factor pulses. This feedback
circuit eliminates the deleterious signals between the 250 volt power
supply ground and the oscilloscope ground which caused a distorted
signal in the test circuit. Figure 3 shows an interstage of amplification required for the sixty percent duty factor pulses.

Direct current test at 165° R. This test is designed for the purpose of establishing the maximum D. C. power rating of the power transistor. In the circuit shown in Figure 4, the collector voltage and current was varied until the failure point was reached. In this case, the base circuit is open. Transistor and ambient temperatures were recorded at each step. Using the circuit in Figure 5, Ib (base current) was held at 200 milliamperes while Vc and Ic were varied until failure. Five randomly picked transistors were used for each of these tests. Vc was measured by a calibrated oscilloscope and Ib and Ic by laboratory instruments.

Heat sink characteristics at 165° F. This test was designed to establish the effects of the heat sink on the power ratings of the transistor.

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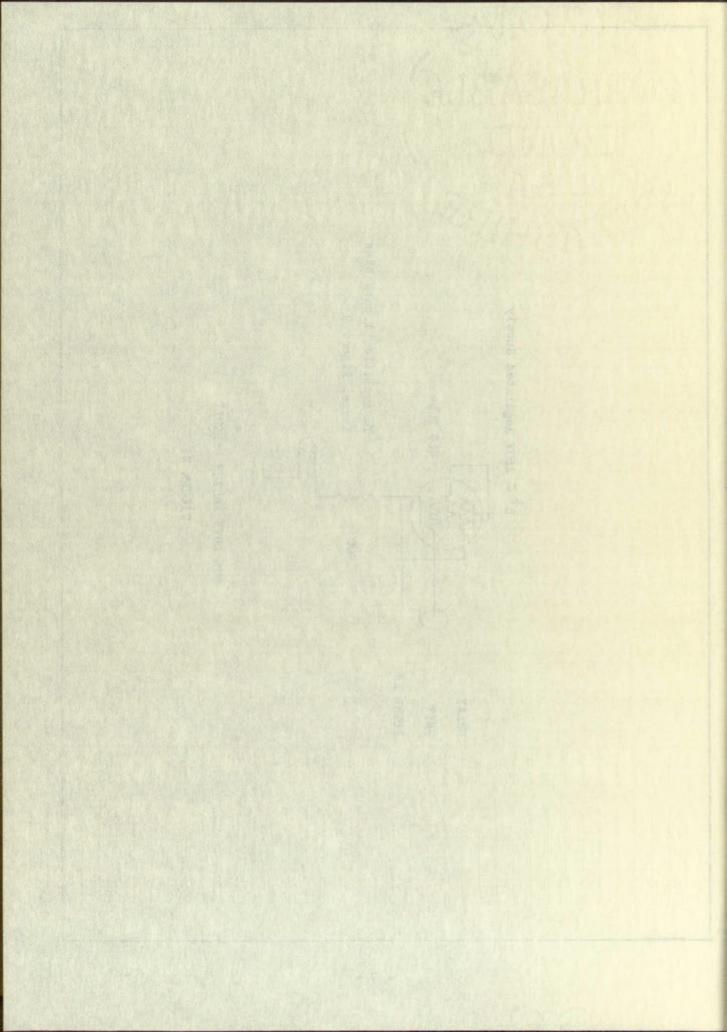
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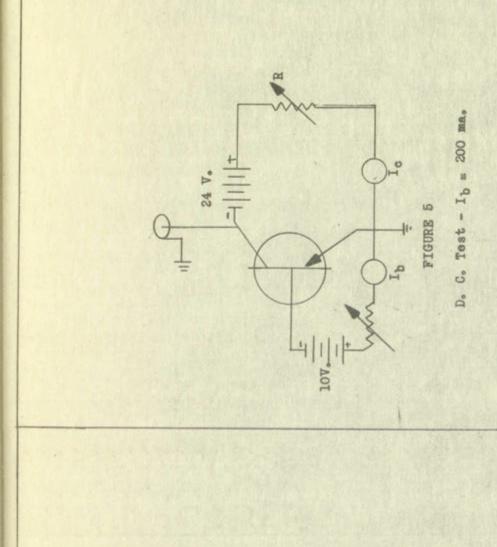
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400% Stage, Figure 9

60% DUTY PACTOR CIRCUIT

FIGURE 3

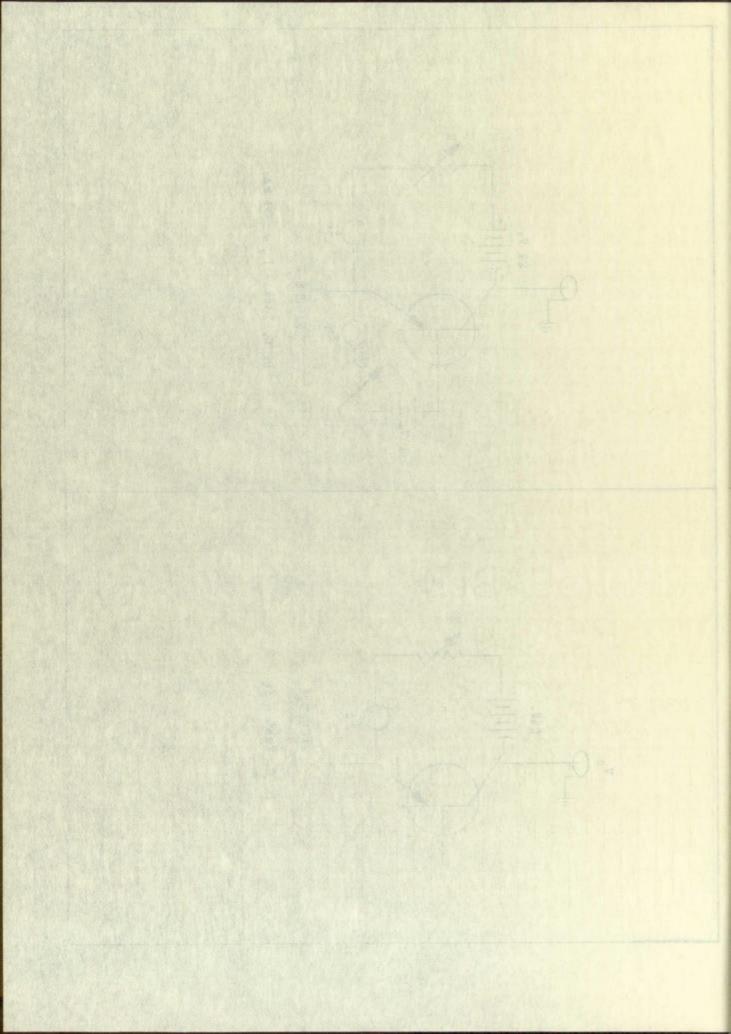




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D. C. Test - Ib = 0

FIGURE 4



The heat sinks for this test were made from AN-A-13 Alclad, S-T 24 Alcoa Aluminum, .064" thick. Three sizes of heat sinks were used: (1) 3" x 3", (2) 4 1/4" x 4 1/4", and (3) 6" x 6". The circuit in Figure 4 was also used for this test. Three sets of data were taken for each heat sink using a different transistor for each set of data. Heat sink characteristics were also taken at room temperature.

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CHAPTER III

THE DESCRIPTION OF THE TEST EQUIPMENT I. TEMPERATURE CHAMBER

An electric oven, 9 1/2" x 11 1/2" x 9 1/2", thermostatically controlled, manufactured by Schaar & Company was used for the temperature chamber. A few additions were made to the original oven: (1) an aluminum sheet, 0.40" thick, was placed above the heating elements to diffuse the heat, (2) a four-inch cage type fan was placed on the rear wall of the chamber to circulate the heat in effort to prevent hot spots, (3) two strands of asbestos covered wire spaced four inches apart were suspended 3 1/2" from the top of the chamber as supports for the heat sink containing the transistor under test, (4) coaxial receptacles and metering resistors were installed on the outside surface of the oven, and (5) the interior was wired with 50 ohm coax cable, all wiring being as short as possible and free from sharp bends.

Two mercury type, 30 - 180° F. thermometers, designated by X₁ and X₂, and three iron-constantan thermocouples, designated by T₂, T₃, and T₄ were used for measuring the ambient temperature. Figure 6 shows the location of these temperature measuring devices. A potentiometer calibrated for 0-400° F., was used in obtaining the thermocouple temperatures.

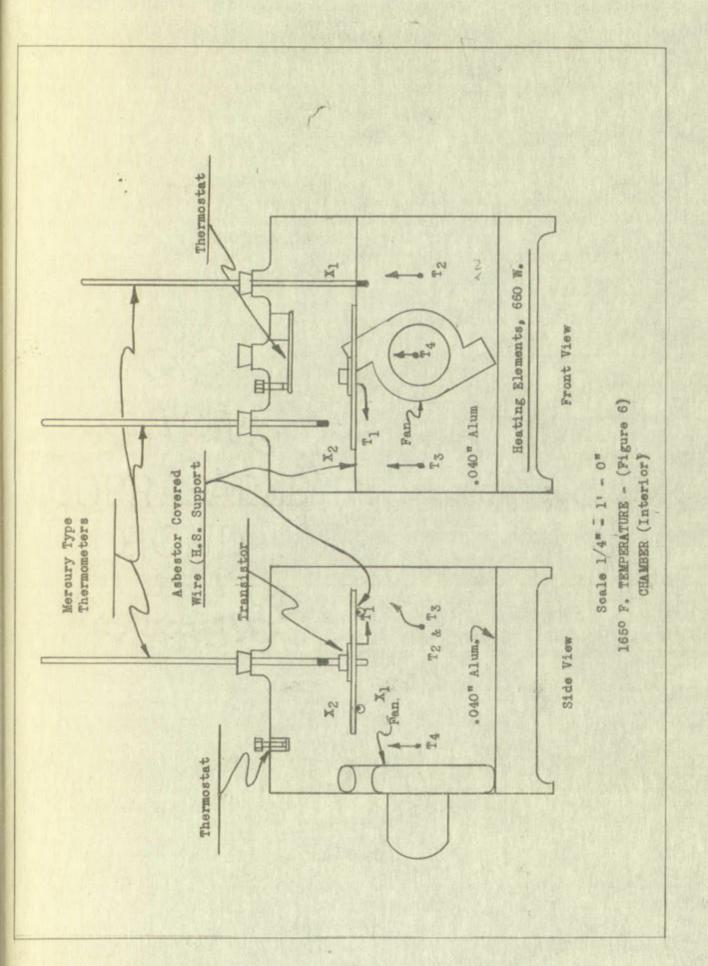
During a test, the thermostat was adjusted so that thermometer X₁ maintained a reading of 165° F. ± 1° F. The thermocouple T₁ located in a groove cut in the heat sink just under the base of the transistor,

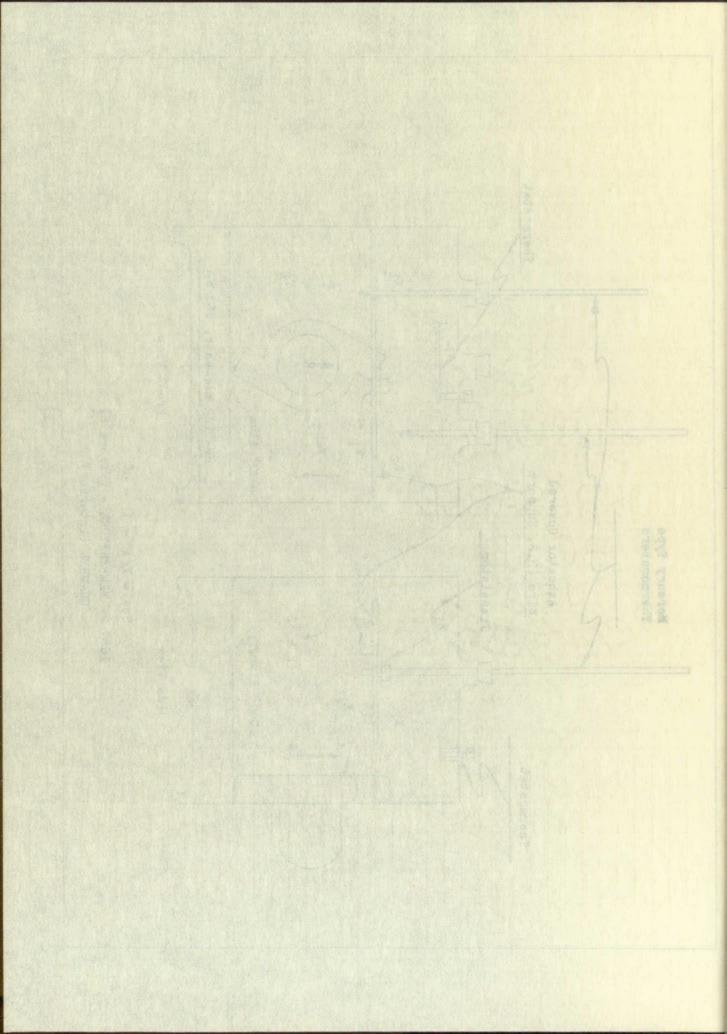
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was employed to read transistor temperature.

The following shows the temperatures at the various locations inside the temperature chamber in a typical case:

Prior to test

$$X_1 = 165^{\circ} \text{ F.}$$
 $X_2 = 161^{\circ} \text{ F.}$ $T_1 = 169^{\circ} \text{ F.}$

While the transistor is dissipating 135 watts peak power at a duty cycle of six percent

$$X_1 = 165^{\circ} F_{\bullet}$$
 $X_2 = 164^{\circ} F_{\bullet}$ $T_1 = 206^{\circ} F_{\bullet}$

Figure 7 shows the temperature chamber in operation, except the door is open to display the interior. The metering resistors on the exterior of the chamber are the non-inductive type (Koolohm) used in measuring the collector current and base current by using a calibrated oscilloscope.

II. EQUIPMENT FOR OBTAINING TRANSISTOR CHARACTERISTICS

The equipment used in obtaining the characteristics of the transistor is shown in Figure 8. A block diagram, Figure 9, is a general circuit description of this equipment. Figure 10 shows three sets of curves describing the transistor characteristics with the transistor connected common emitter. These curves were displayed on the face of the cathode ray tube and photographed with a polaroid land camera. The circuitry used in building the transistor characteristics curve tracer is shown in Figure 11.

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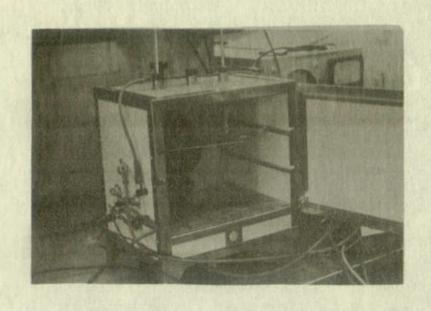
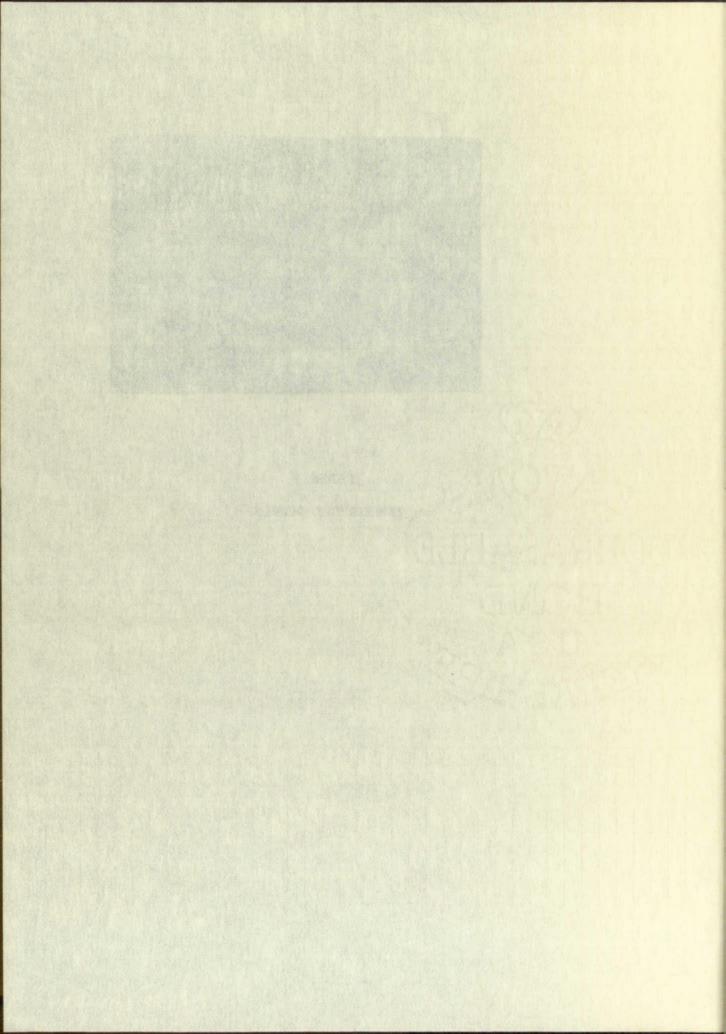


FIGURE 7
TEMPERATURE CHAMBER



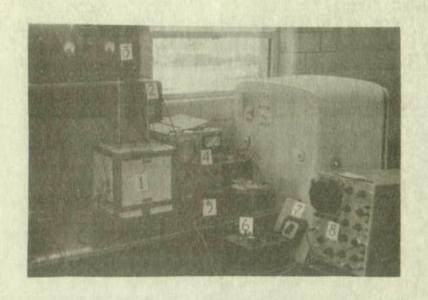


FIGURE 8

TEMPERATURE CHAMBER AND CURVE TRACER EQUIPMENT

1. Oven, electric.

2. Collector voltage supply (Battery Booster 6-12 volt).

3. Power Supply, D. C., 300 wolt regulated, 250 milliampere capacity.

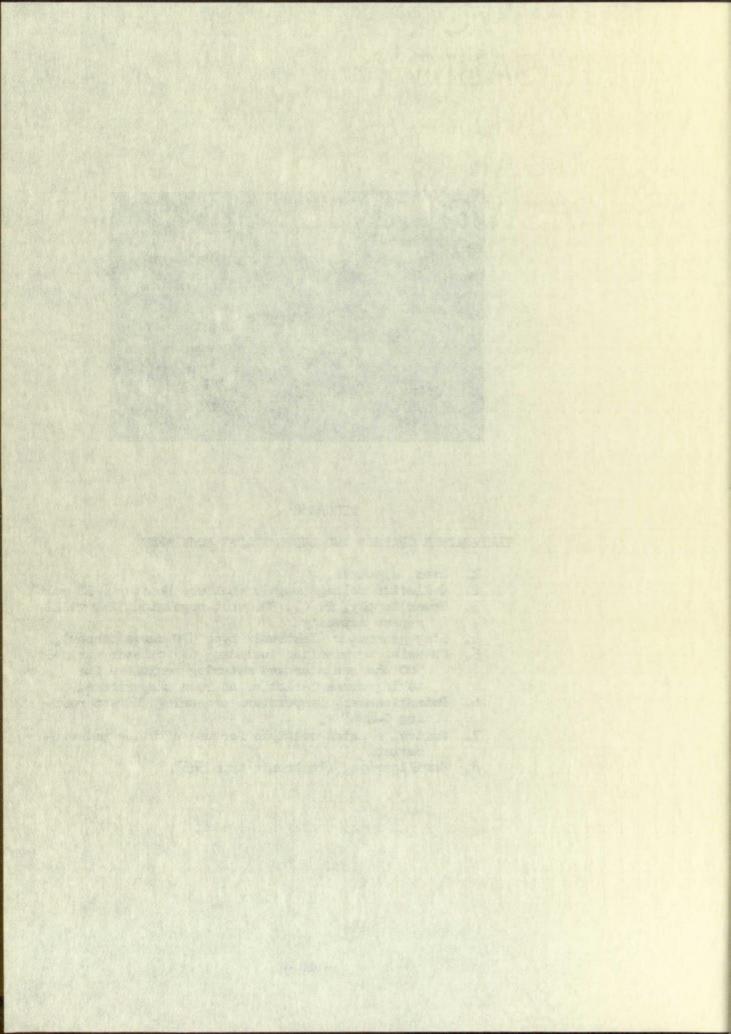
4. Step generator (Tektronix type 570 curve tracer).

5. Chassis, terminating including 616 cathode variable 500 ohm resistor and metering resistors for taking characteristics at room temperature.

Potentiometer, temperature measuring, direct reading 0-400° F.

 Switch, 8 point multiple for use with the potentiometer.

8. Oscilloscope, (Tektronix type 545).



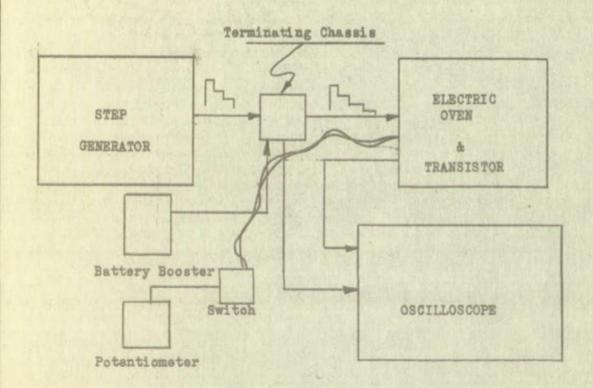
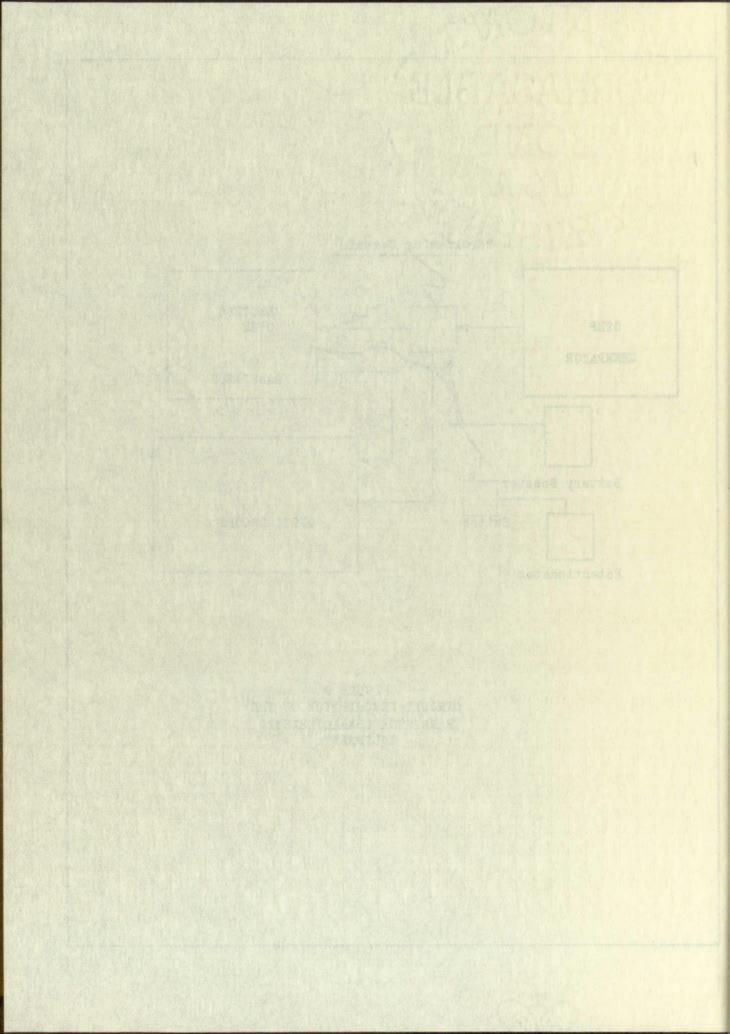
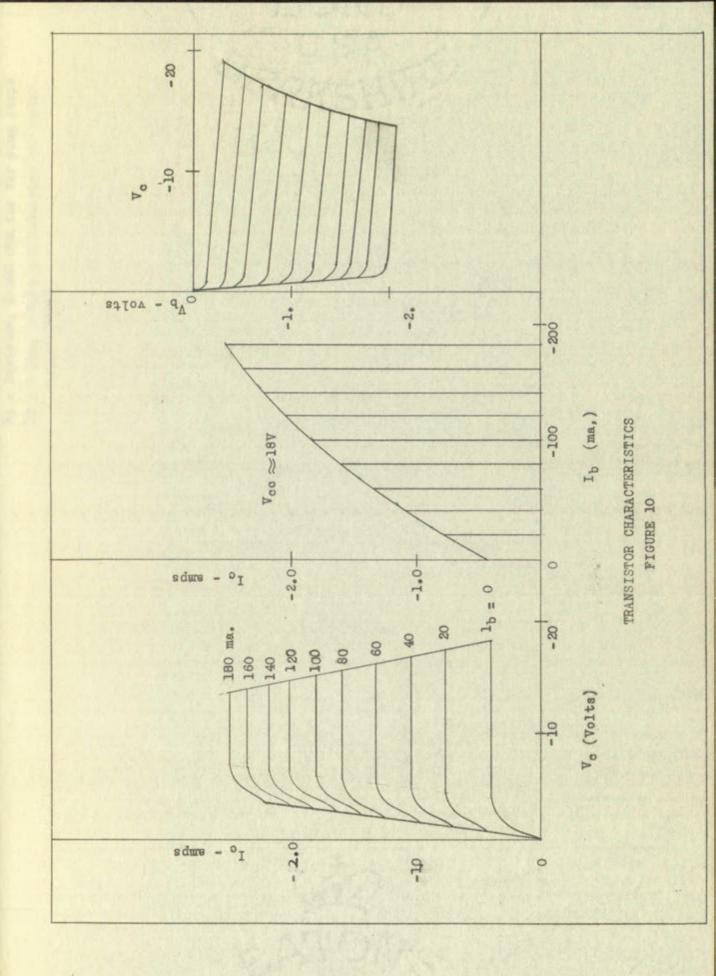
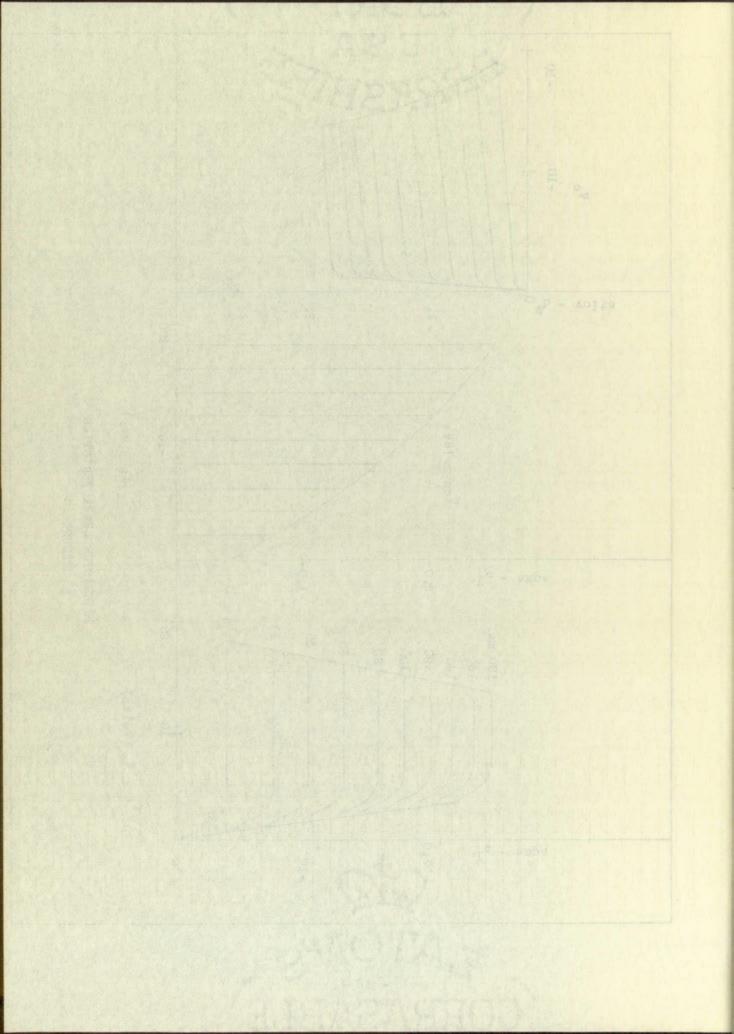
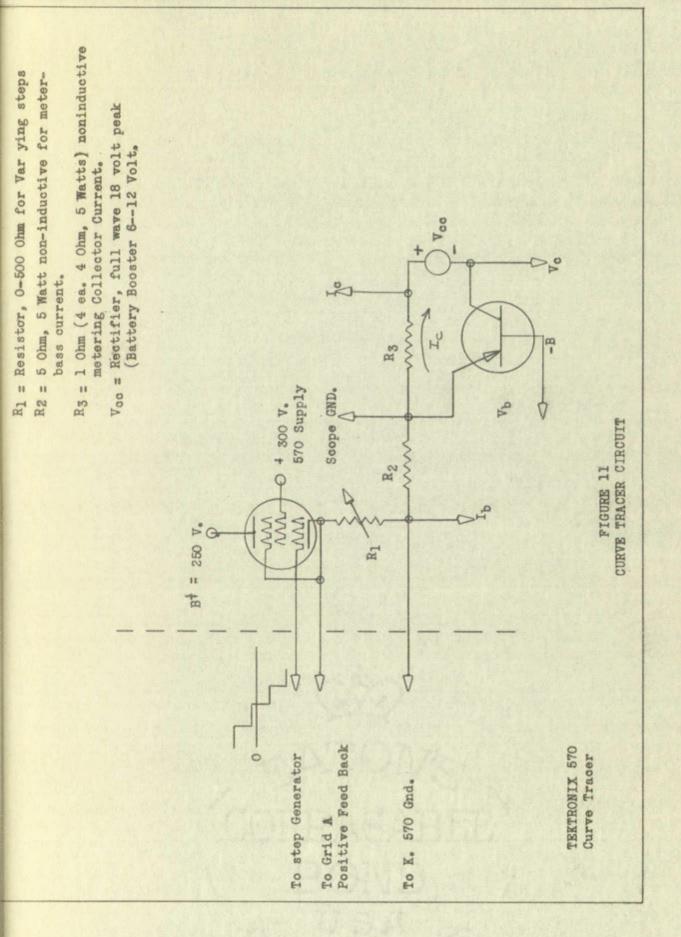


FIGURE 9
CIRCUIT DESCRIPTION OF THE
TRANSISTOR CHARACTERISTIC
EQUIPMENT









Theory of the Curve Tracer Circuit (General). Voltages of equal steps are impressed on the grid of the 616 tube, thus causing constant current steps in the base circuit of the transistor. A sweeping voltage of 18 volts, is impressed on the collector circuit. The base current steps are 1/120 second in duration but timed so that the step comes during the time that the collector voltage is a maximum or the change of collector voltage dvc is a minimum and the fall time of the current steps are not so critical. Figure 12 shows this time relationship of the base current, collector voltage and the collector current.

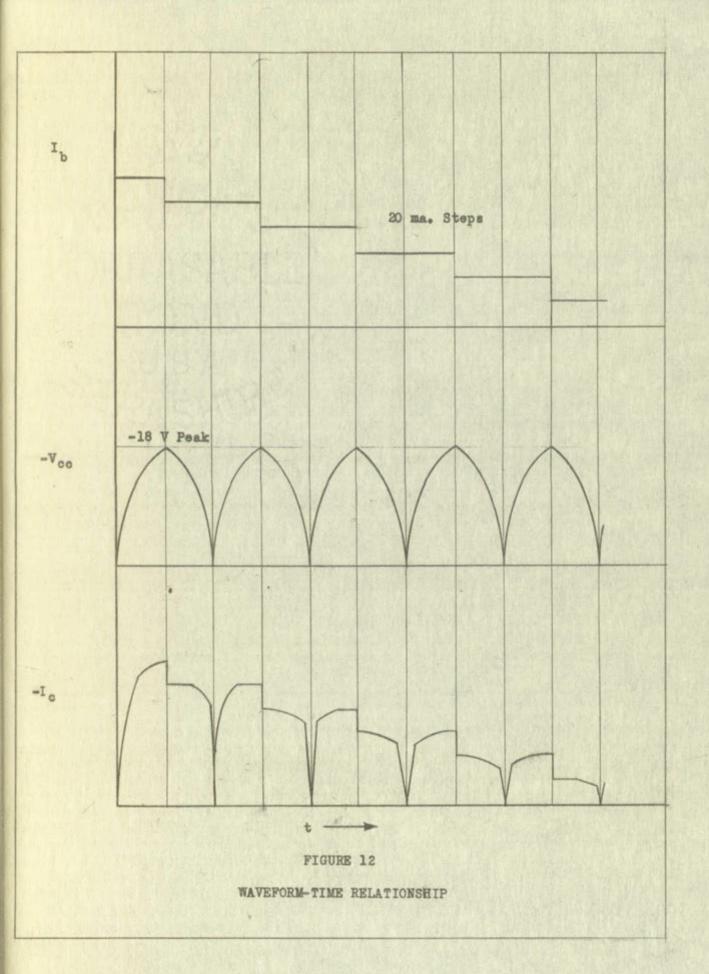
III. PULSE TESTING EQUIPMENT

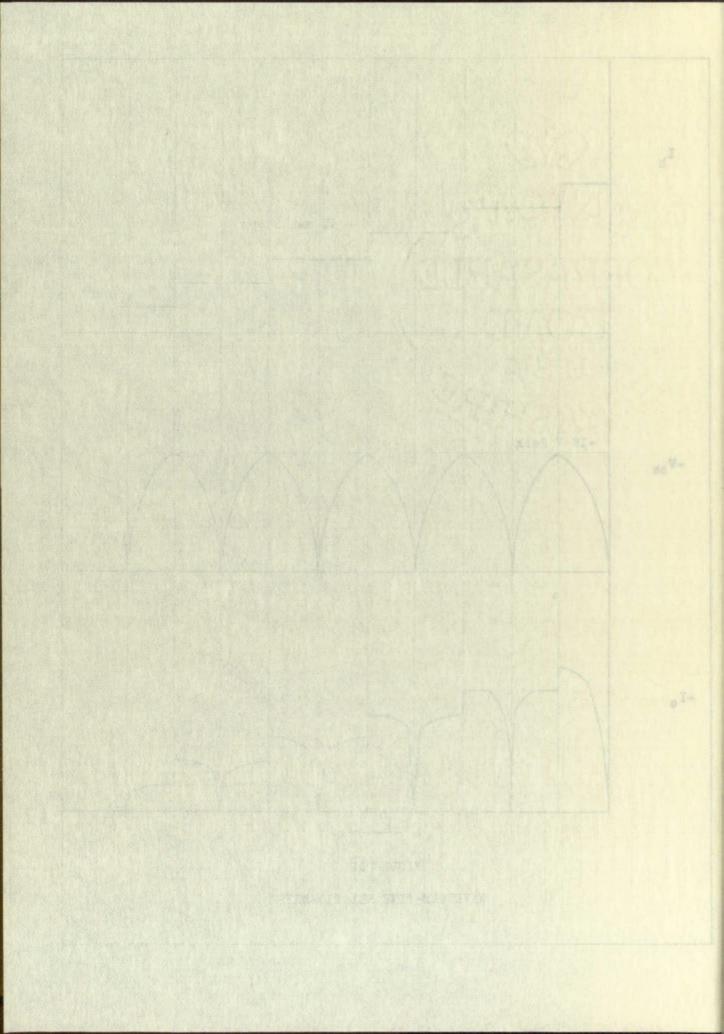
The overall pulse testing set up is shown in Figure 13 and a general description of the pulse testing circuit is shown in Figure 14.

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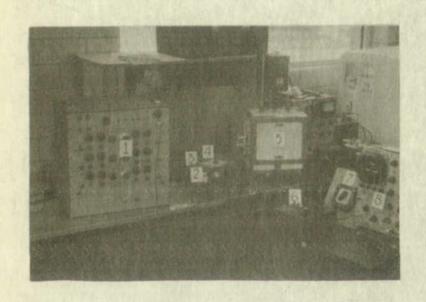


FIGURE 13

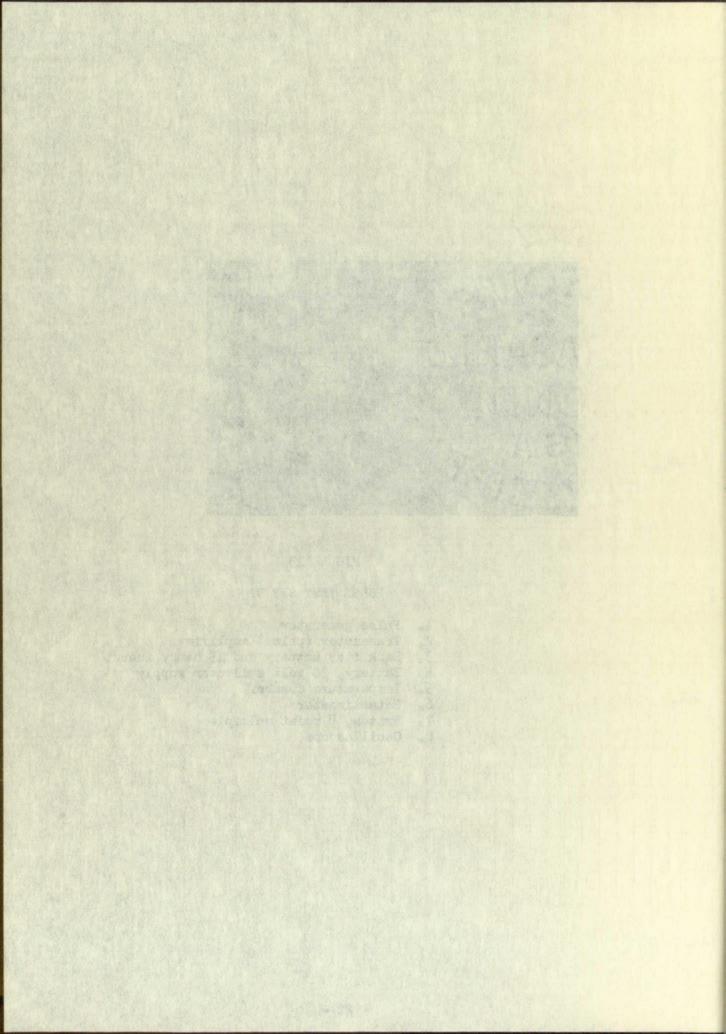
PULSE TEST SET UP

1. Pulse generator

2. Transistor (pulse) amplifier

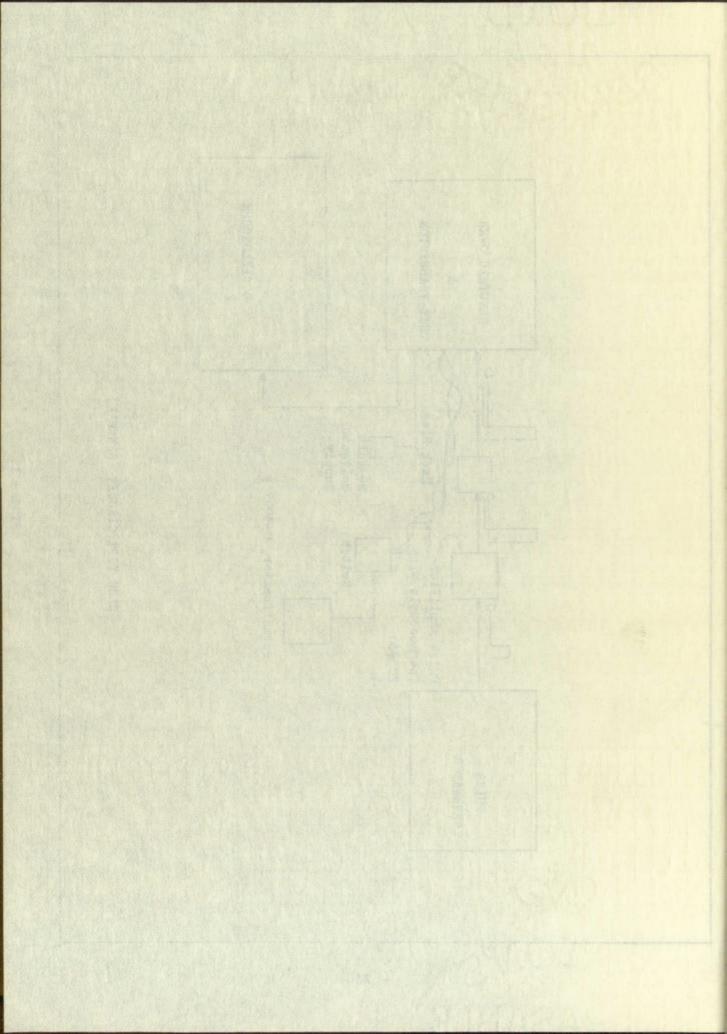
- 3. Back bias battery and 15 henry inductor
 4. Battery, 20 volt collector supply
 5. Temperature chamber
 6. Potentiometer

7. Switch, 8 point multiple 8. Oscilloscope



- 23 -

FIGURE 14



CHAPTER IV

THE RESULTS

I. PULSE TESTING

The transistors were placed in the temperature chamber and tested as described in Chapter II, The Method of Test. Where the available input pulse was too small in amplitude to cause an appreciable change in transistor temperature, the number of transistors pulsed at this duty factor was decreased expecting to devise a pulse generator at a later date which would extend these amplitudes.

The permissible peak power in the pulses at various repetition frequencies and pulse widths are presented in Figures 15, 16, 17, and 18. The solid flat-topped lines are the last peak power readings at which the transistor reproduced stable pulses and reached a constant temperature. The dash lines extending the flat-topped lines are the points where either the transistor failed completely or became unstable. The pulsing circuit, in many cases, would not provide pulses of large enough amplitude to destroy the transistor. These points are designated by a pointed line.

The maximum transistor temperature, average power of the permissible peak power reached, and the transistor used is also placed on the graphs.

The effects of duty factor or duty cycle are shown in Figure 19. Since failures occur in the temperature region of 200° F. and 216° F. transistor temperature, this information is plotted from curves in

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Appendix C at the 200° F. temperature point. Scarcely any failures occurred from pulsing below 200° F., the majority failed just after 200° F. For this reason then, in order to overcome the difference in transistor characteristics, the effects of duty cycle are taken at a safe operating temperature, 200° F., which is only slightly different from the maximum pulse ratings.

Since it was impossible to destroy the type CTP-1003 transistor in the small duty factor range with the existing pulsing equipment. peak power versus transistor temperature was plotted at these duty factors'so that the curves could be extrapolated into the failure region. This set of curves is present in Appendix C. The six percent and smaller duty factor curves deviate from the expected powertemperature curves in that they have a tendency to droop or flatten out in the upper temperature region. Above six percent, the curves are straight lines as would be expected since dissipated energy is directly related to temperature and also to power. That is energy = power times time (Joule's law). Figure 20 is a typical curve showing this drooping effect in a manner to emphasize the reason for this deviation from the straight line relationship between dissipated power and temperature. This is to say that base current cannot be ignored in calculating the total power dissipated in a transistor at small duty factors in the upper temperature region.

II. DIRECT CURRENT TESTING

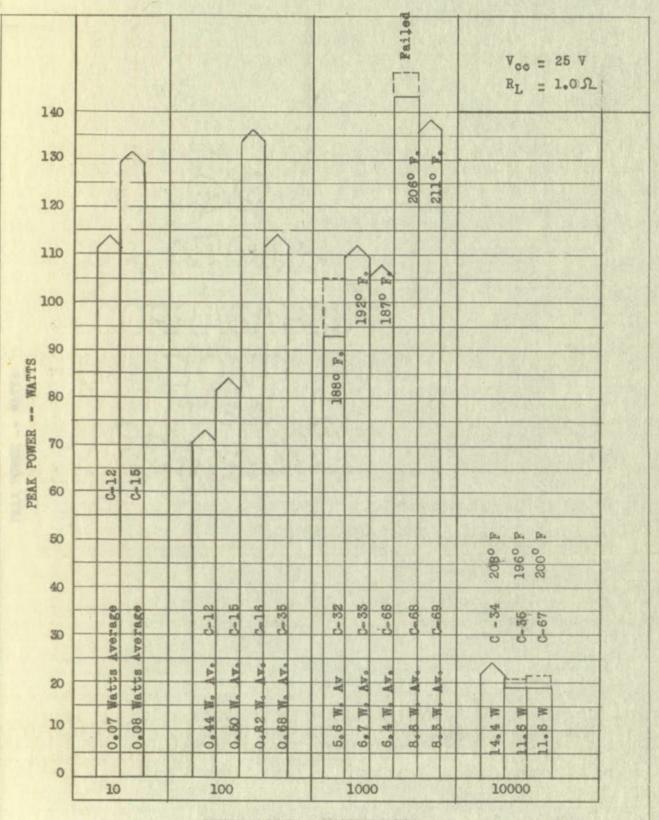
The two D. C. tests were performed as previously described in Chapter II, The Method of Test. With these tests, the transistors can

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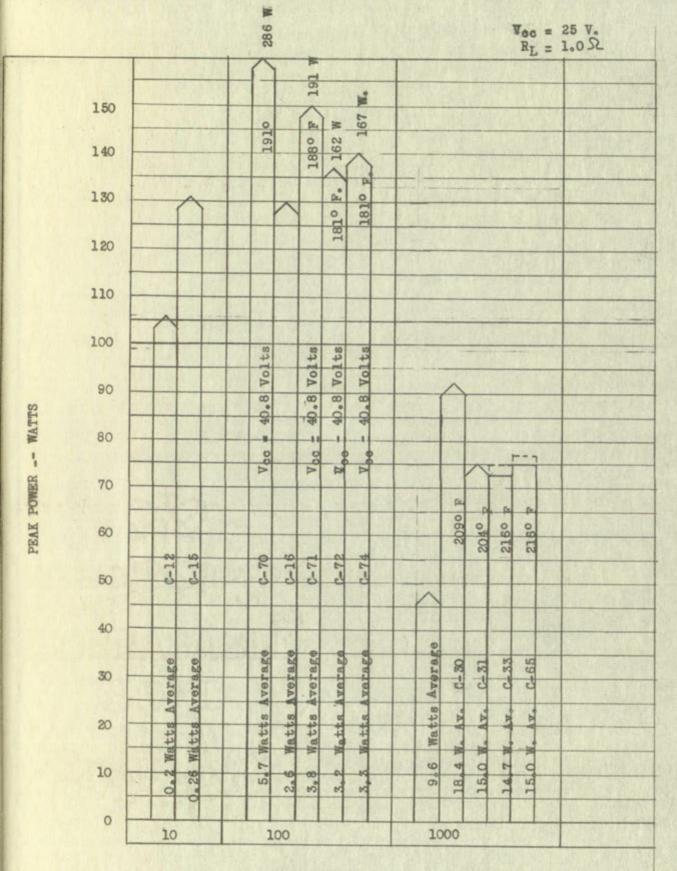
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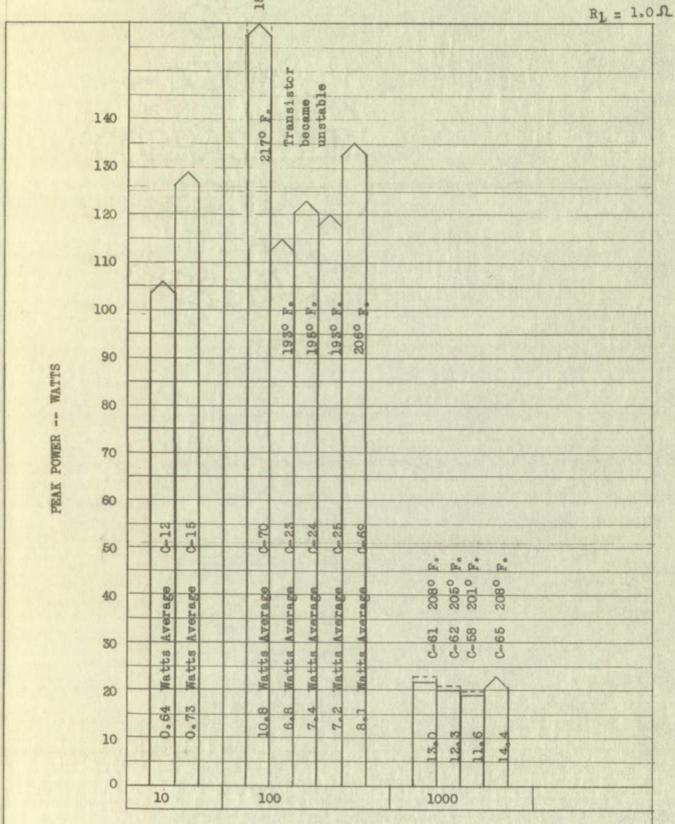
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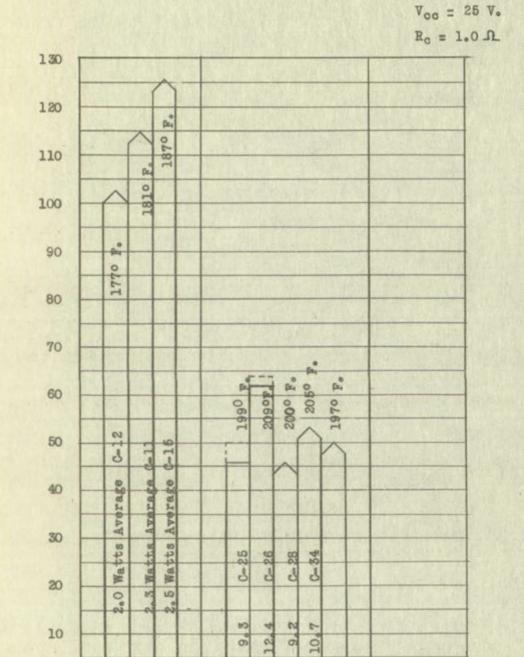
PULSE WIDEHS - MICROSECONDS FIGURE 15 PULSE POWER RATINGS -- 60 CPS



Pulse Widths -- Microseconds Figure 16 Pulse Power Ratings - 200 CPS 165° F. Ambient



PULSE WIDTHS -- MICROSECONDS FIGURE 17 PULSE POWER RATINGS -- 600 CPS 1650 F. Ambient



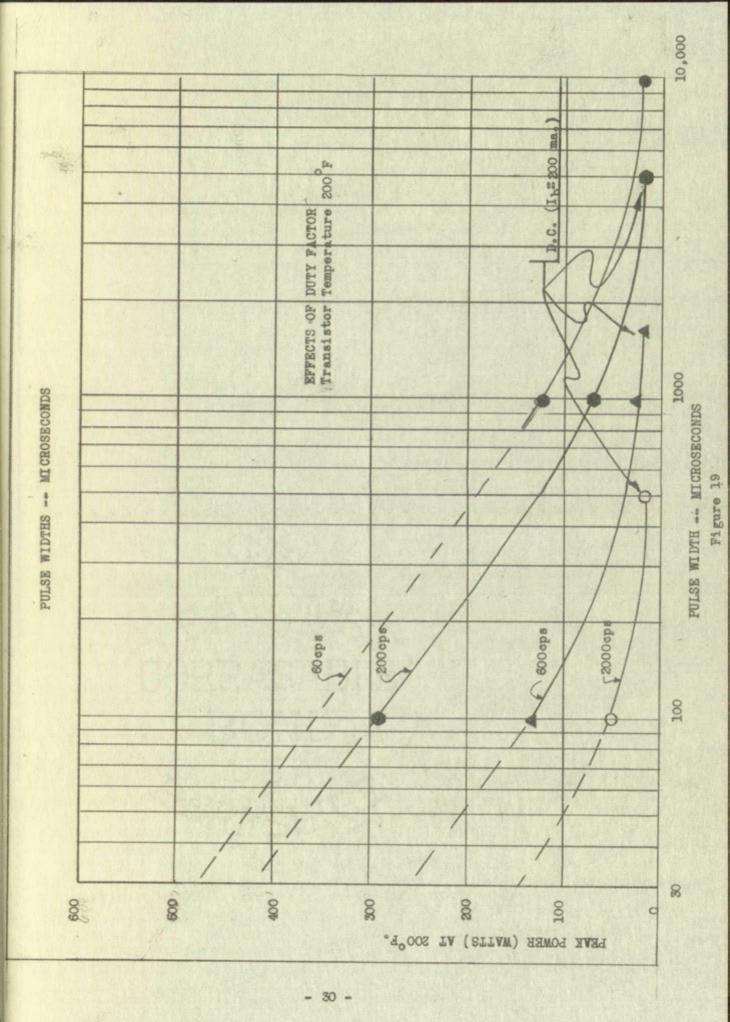
Pulse Power - Watts

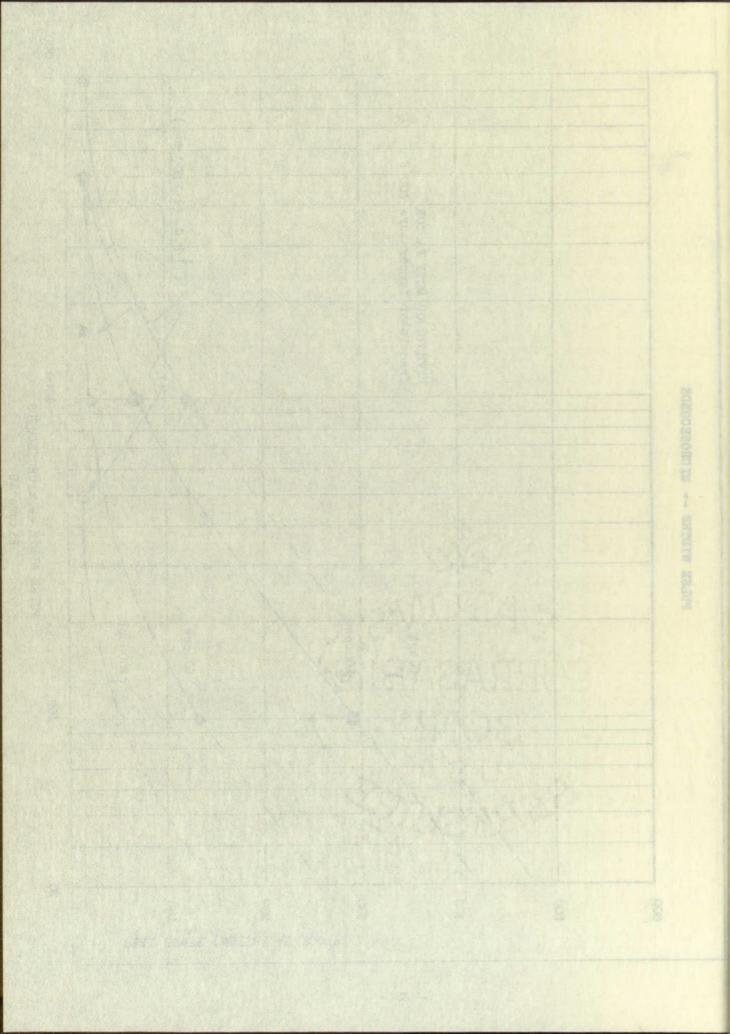
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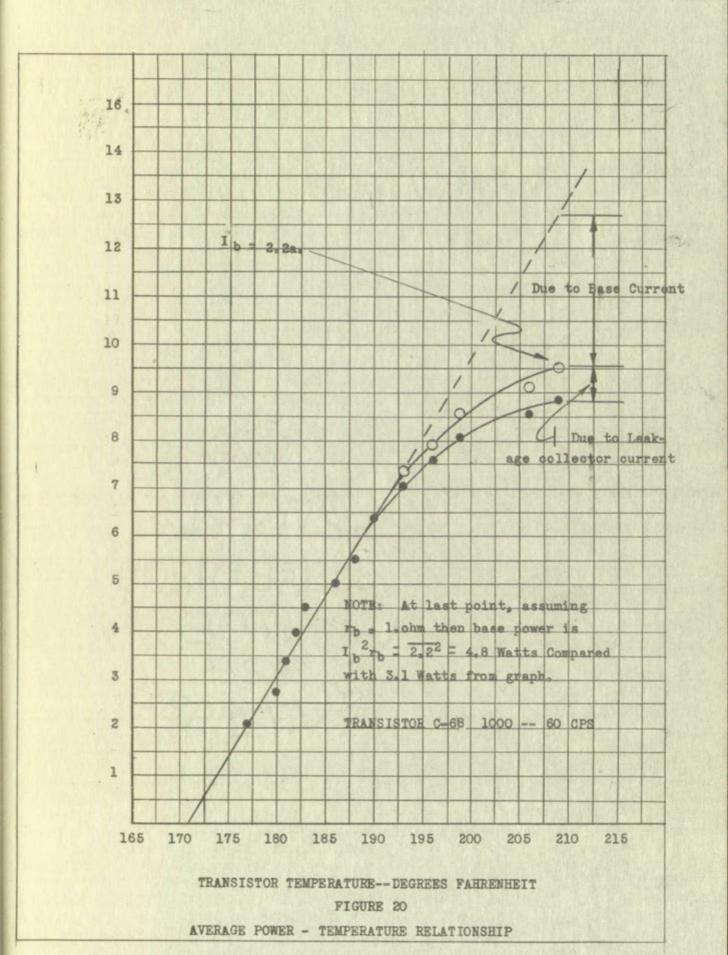
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Pulse Widths - Microseconds
Figure 18
Pulse Rower Ratings -- 2000 CPS
1650 F. Ambient

100







be controlled to prevent failure and yet the failure point can be determined. This makes it possible to use the same transistors for both tests.

The results of these tests are presented in Figures 21 and 22.

III. HEAT SINK CHARACTERISTICS

This test shows that approximately fifty percent more power can be dissipated in the transistor by doubling the area of the heat sink (in the heat sink range tested) and approximately three hundred percent more power can be dissipated in going from no heat sink to a three inch square sink. This information is contained in Figure 23. A comparison of heat sink characteristics of 85° F. and 165° F. ambient temperatures is made in Figure 24.

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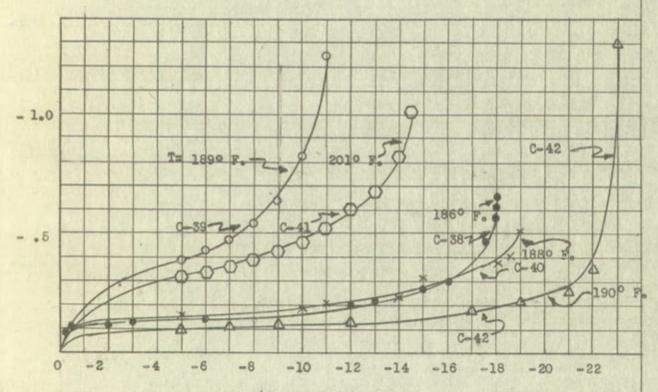
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Average Collector Volts = 16.4

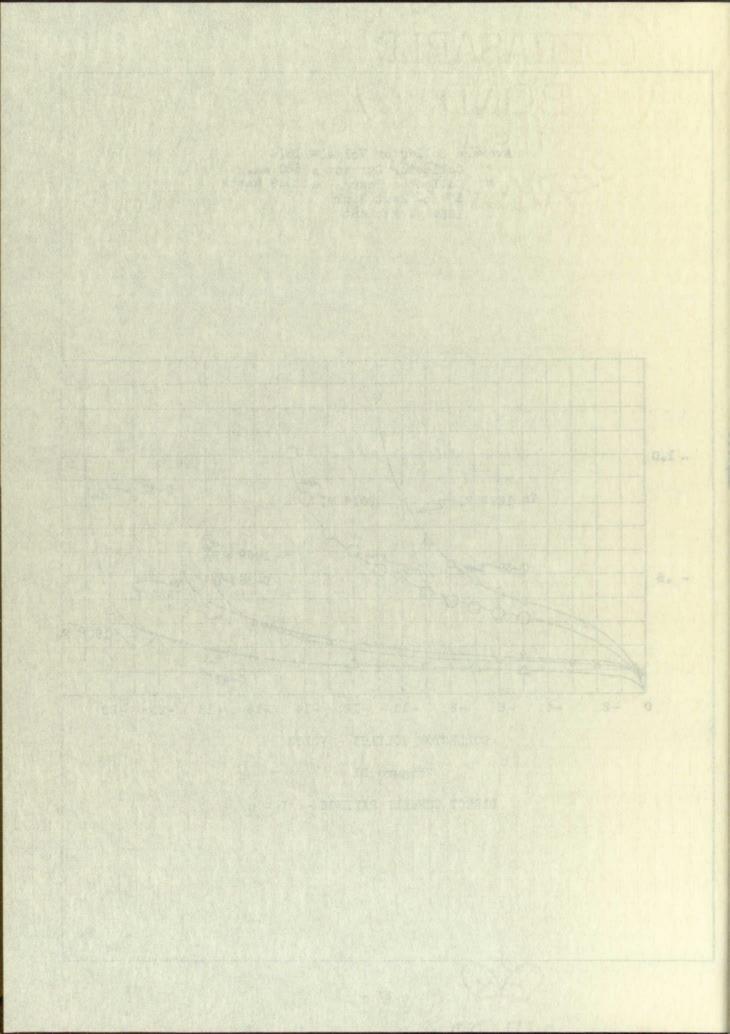
" Collector Current = 660 ma.

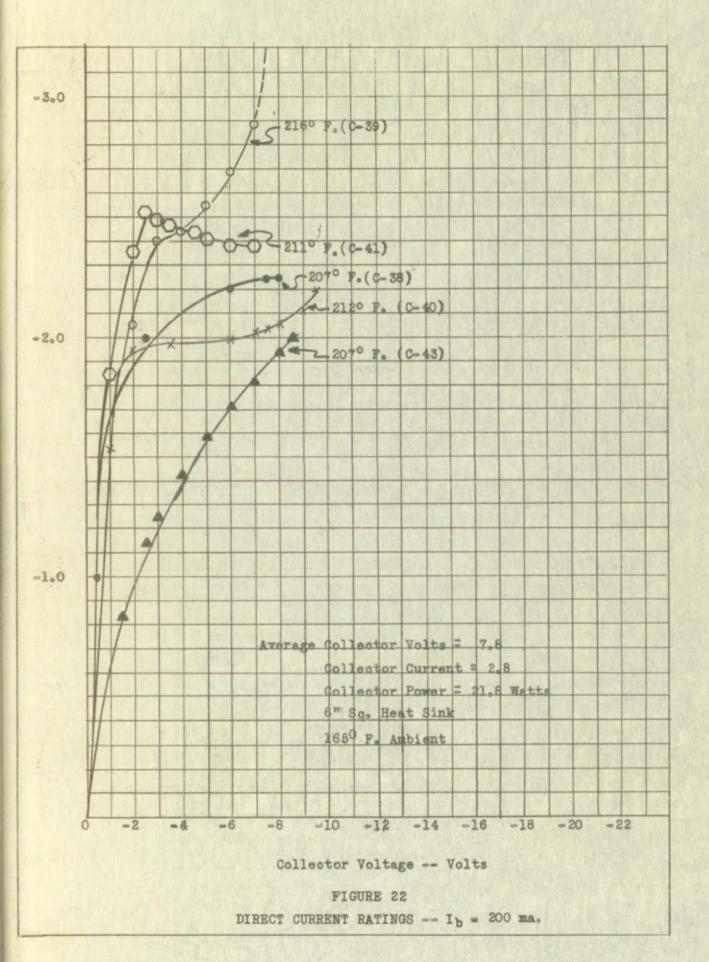
" Collector Power = 10.8 Watts
6" sq. Heat Sink
1650 F. Ambient

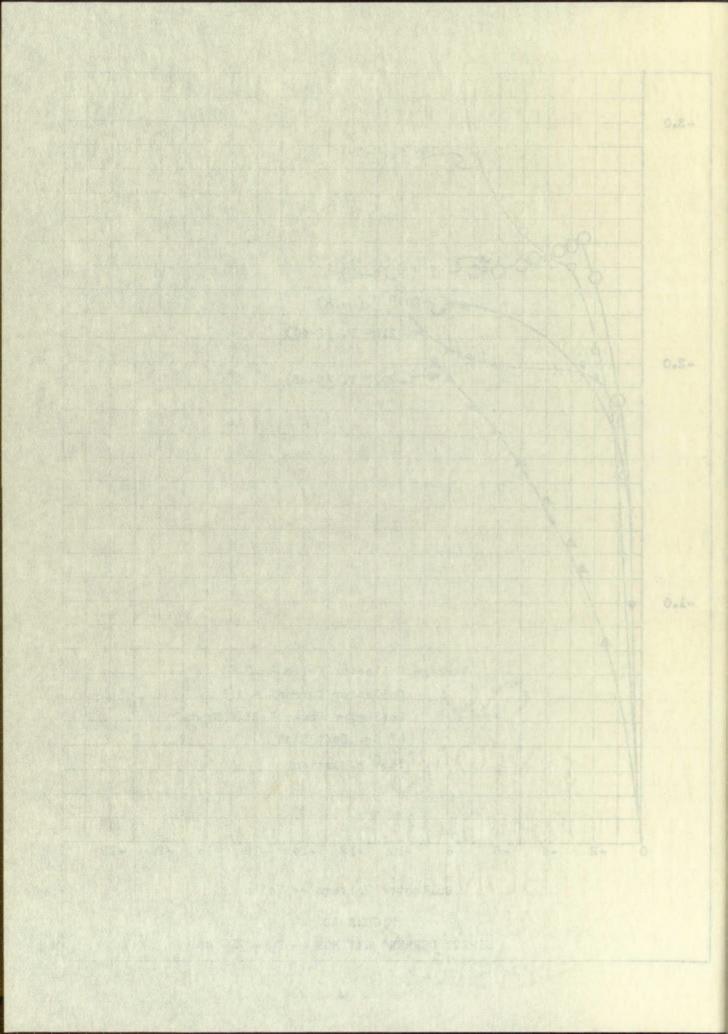


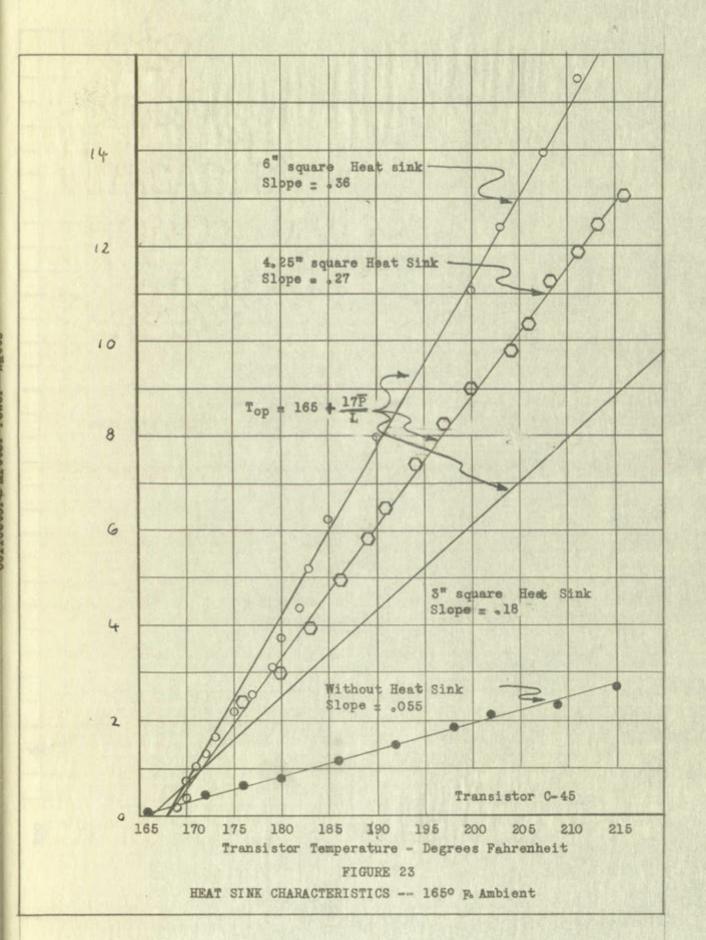
COLLECTOR VOLTAGE - VOLTS

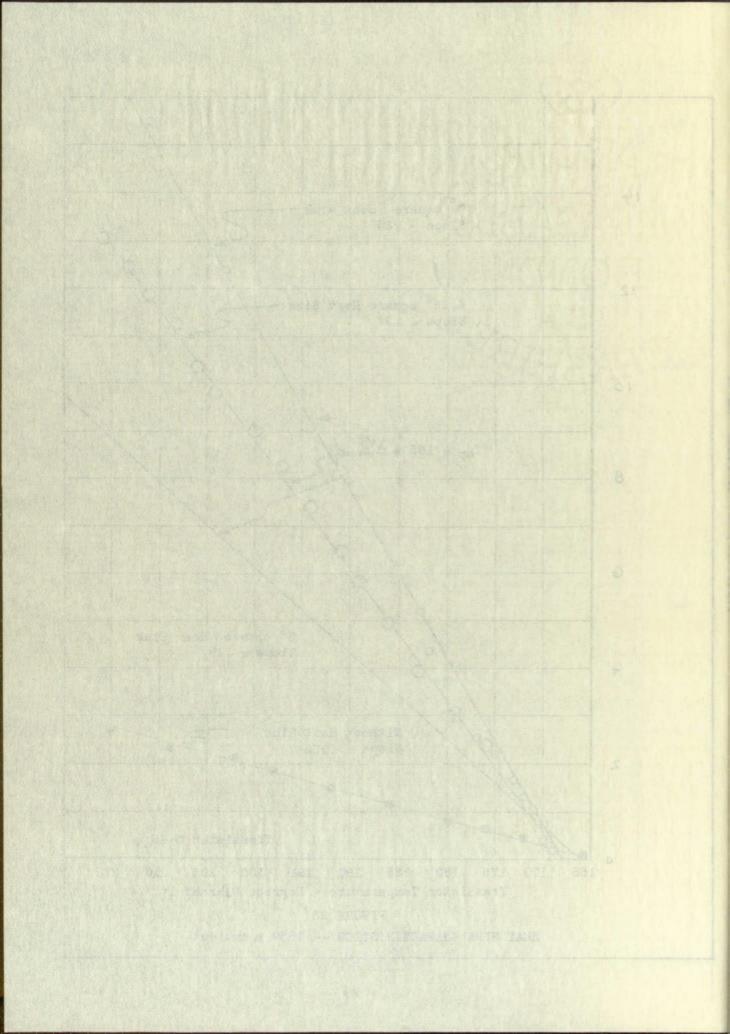
Figure 21
DIRECT CURRENT RATINGS -- Ib= 0

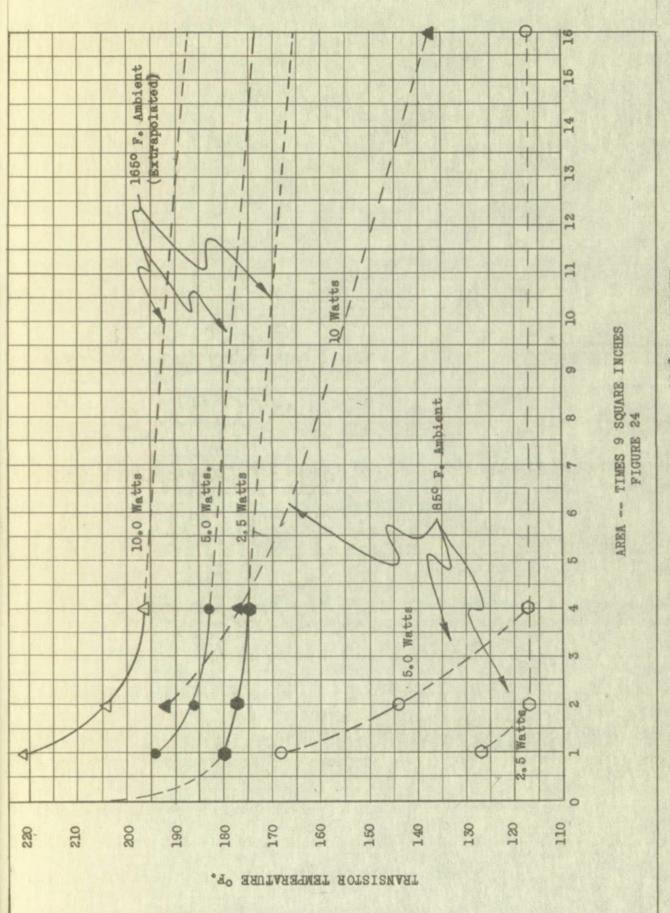




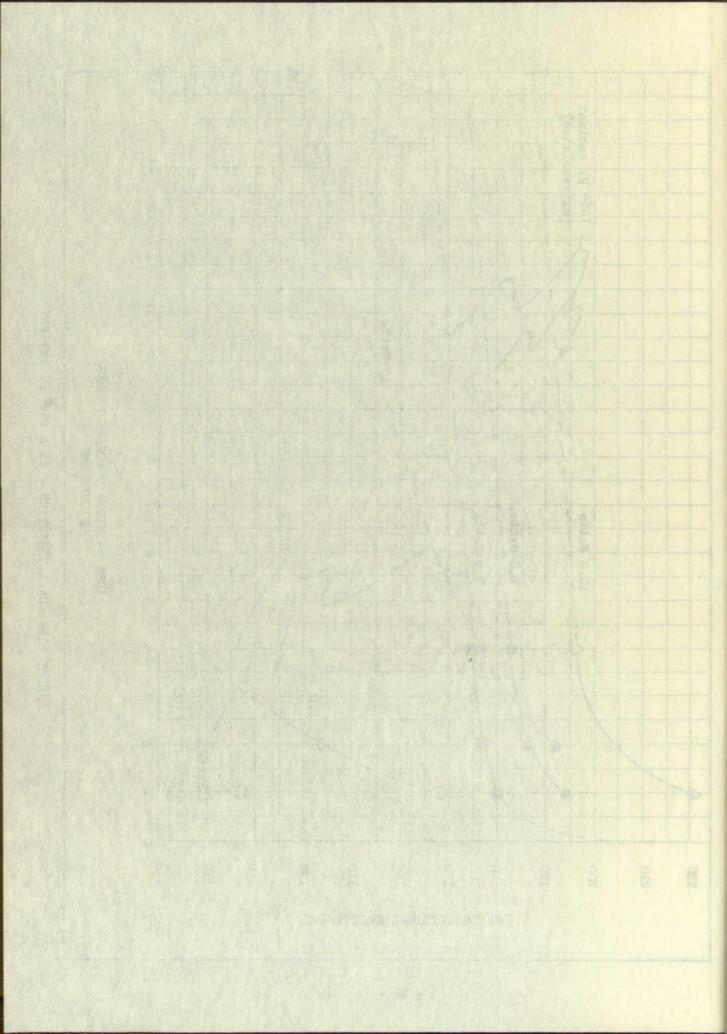








HEAT SINK CHARACTERISTICS - 85 & 1650 F. AMBIRNT



CHAPTER V

DISCUSSION OF RESULTS AND CONCLUSIONS

The results of the pulse and D. C. tests point strongly in the direction that temperature destroys transistors. It could be stated further that average collector power destroys transistors except for the low duty cycle cases. If the collector power is assumed to be the only power increasing the temperature of the transistor, then less average power is required to raise the transistor temperature at the low duty factors than at the higher duty factors. This certainly can not be possible. That is to say that although the base resistance is quite small the base current becomes very large for the low duty factor pulses, thus increasing the total power input to the transistor and affecting the transistor temperature. The base power then, cannot be overlooked in the low gain region.

Efforts were made to pulse the transistor with 0.1 microsecond pulses but it was found that reproduction was very poor below 100 microsecond pulses. Ten microsecond pulses could be amplified but the rise time was equal to the pulse width while the fall time was usually about four times the pulse width.

The spread of transistor characteristics (taken from the characteristics of approximately seventy transistors) located in Appendix B, is so wide that a sample of five transistors does not give a very reliable picture of the maximum permissible ratings. Possibly a sample of twenty-five at each duty factor would be more reliable.

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The D. C. tests show clearly that in order for the transistor circuit to be stable at 165° F. ambient with the collector voltage of any size that the base circuit must be back biased. Back bias of ten volts was used for all tests except the D.C. tests which did not require back bias since a large resistor was used in the collector circuits to limit the collector current.

From the heat sink characteristics at 165° F. ambient an equation was developed which makes it possible to predetermine the operating temperature of a transistor at an average D. C. power dissipated in the transistor for an arbitrary heat sink. The equation

$$T_{op} = 165 + \frac{17 \, \overline{P}}{L}$$

is developed in Appendix D. T_{op} is the operating temperature of the transistor when dissipating an average power, \overline{P} using a square heat sink with L the length of a side.

Thus, if an average power of five watts were to be dissipated in a transistor using a square heat sink of a length of three inches.

then
$$T_{\text{op}} = 165 + \frac{17}{3} = 165 + \frac{80}{3} = 193^{\circ} \text{ F.}$$
 as compared with 194° F. from Figure 23.

The failures witnessed by the Sandia Corporation as mentioned in Chapter I, Introduction, were most likely due to a large D. C. component of power which did not exceed the limits of the transistor until the large current spikes were induced.

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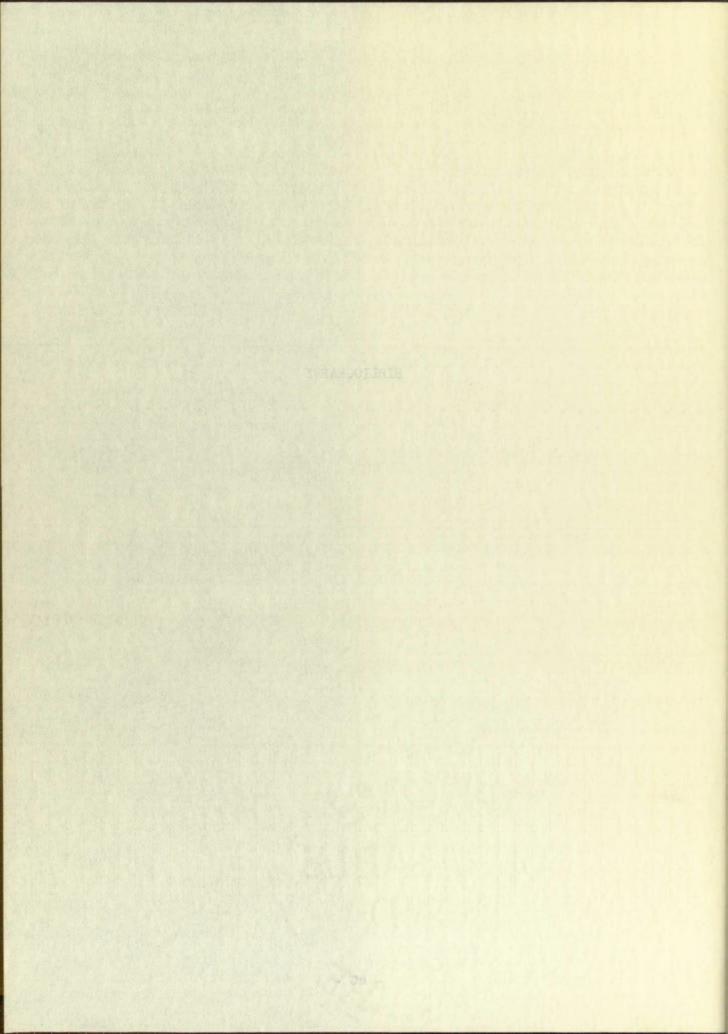
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Further testing could be done to prove this point by pulsing the collector circuit with large power pulses which were known not to contain a D. C. component.

BIBLIOGRAPHY



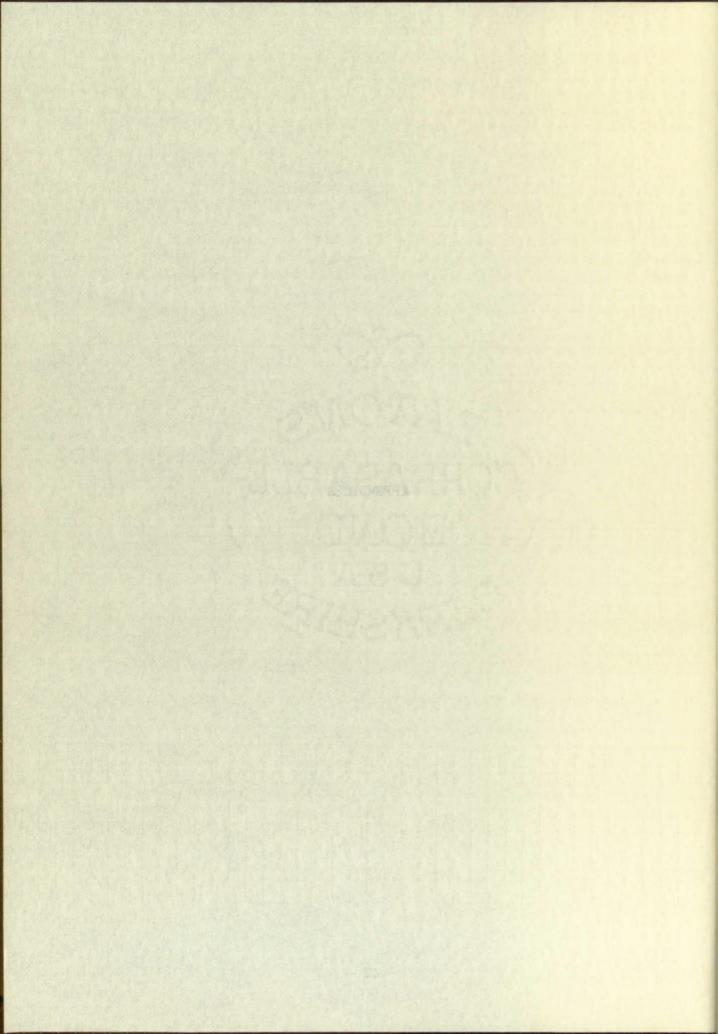
BIBLIOGRAPHY

Griffith, LeRoy A., Germanium Power Transistors, Minneapolis-Honeywell Regulator Company, Minneapolis 8, Minnesota, Transistor Division, paper dated February 9, 1955, TR73pl.

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APPENDICES



APPENDIX A

MANUFACTURER'S SPECIFICATIONS OF THE TYPE CTP-1003 TRANSISTORS

Description. The CTP-1003 transistor is a hermetically sealed PNP germanium power transistor designed for general use at nominal collector voltage up to 28 volts D.C. It is characterized by low collector saturation voltage, moderate current gain, and high current handling capacity. The collector is electrically connected to the metal housing. Where the collector must be insulated from the chassis, the large area of the mounting flange permits adequate heat flow through the insulating medium (e.g., mica).

Electrical Specifications.

Absolute maximum ratings

Instantaneous collector to base voltage	60 V.
Collector supply voltage	30 ₹.
Junction temperature	85° C.
Instantaneous total peak power	25 W.
Average total power with infinite	
heat sink at 25° C.	25 W.
Average total power (no heat sink)	
in free air at 25° C.	2.25 W.

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Description. The CTP-1003 translator to a bernetically scaled law personal are at scaled collector personal are at scaled collector release up to 16 volts D.C. It is dispensived at low collector saturation voltage, noderate correct grin, and also converte the scaling and also converted to the scaled converted to the scaled converted to the scaled converted the converted to the scale from the charge, where the collector was be insulated from the charges, the large area of the scanting flangs primite adoptate heat flow through the insulating scaling (1.5. mice).

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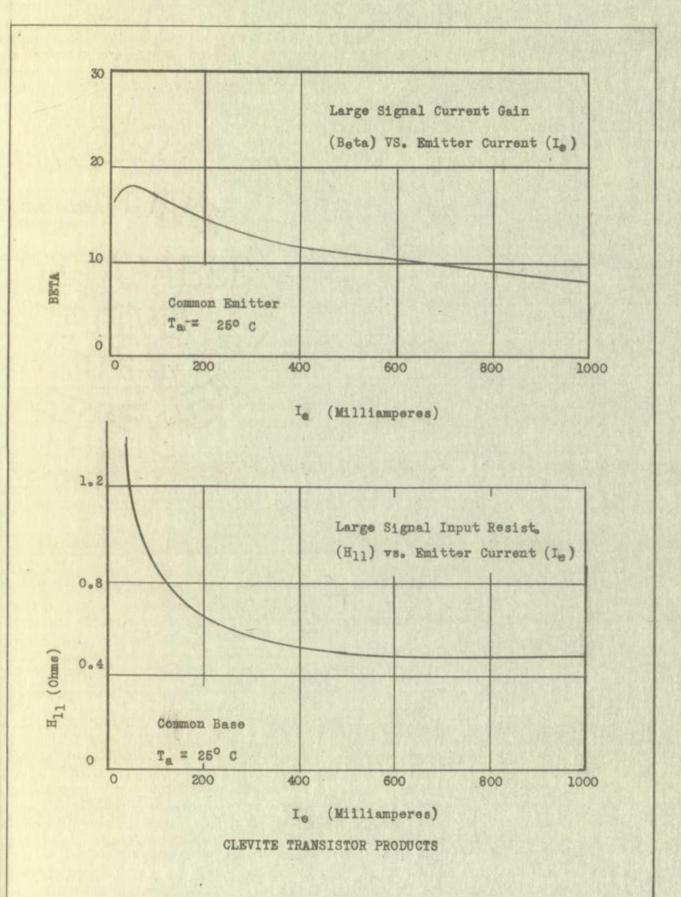
Characteristics at 25° C.*

Test	Conditions Symbol Min Max U	nit
Collector cut-off current	V _{cb} 60V I _{co} 2.0	ma
	l _e - 0	
Emitter cut-off current	Veb - 6 V Ieo .20	ma
	1 _c - 0	
Power gain, common	Vcc - 14.2 V Pg 23.0	db
emitter, transformer	I _c 37 A	
coupled	R _L = 30 ohms	
	Rg - 10 ohms	
Power gain cut-off	Same as for fpg 4 (250 M.S.)	Kc
frequency	Power gain (250 M.S.)	

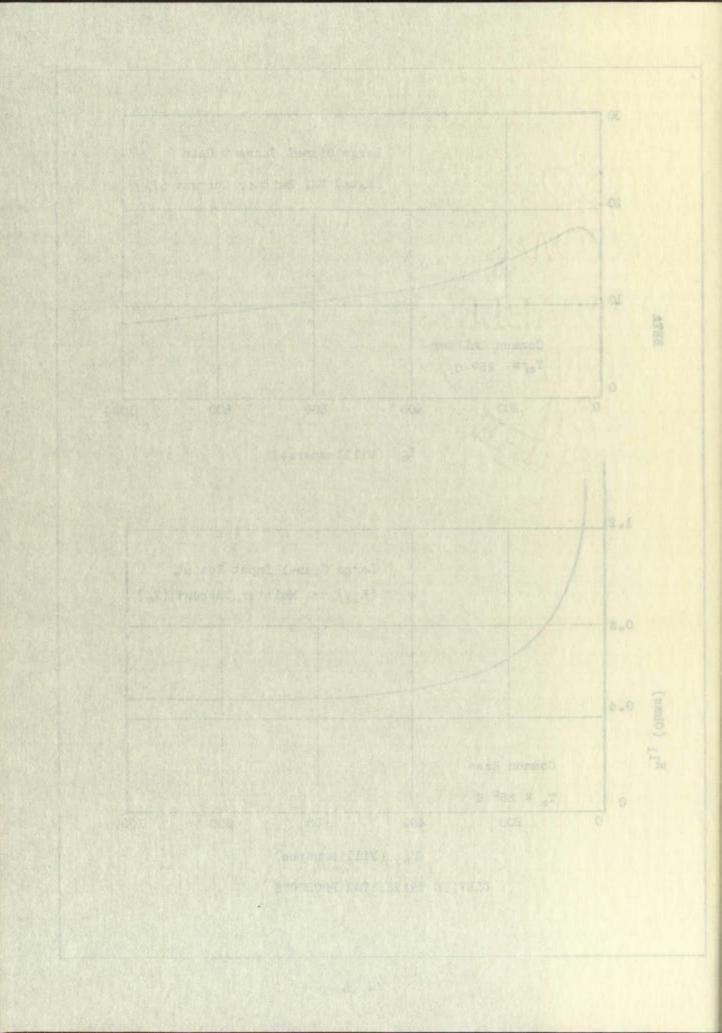
^{*}Cooling must be provided equivalent to mounting the transistor in the center of a 3" x 3" piece of aluminum, 1/16" thick, suspended in free air.

Sonractoriations 250 A.*

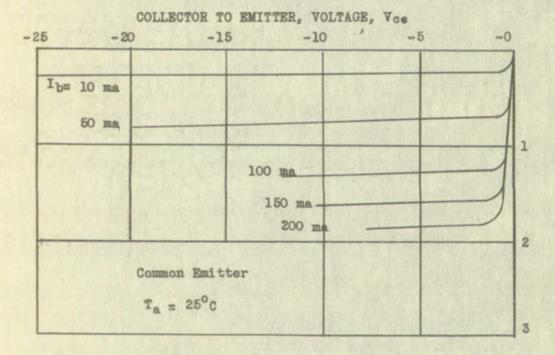
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CTP-1003
TYPICAL CHARACTERISTICS



CLEVITE TRANSISTOR PRODUCTS

APPENDIX B

THE SPREAD OF TRANSISTOR CHARACTERISTICS

Preceding all tests, a photograph of the input (base) and output (collector) characteristics of each transistor as taken with the transistor connected common emitter.

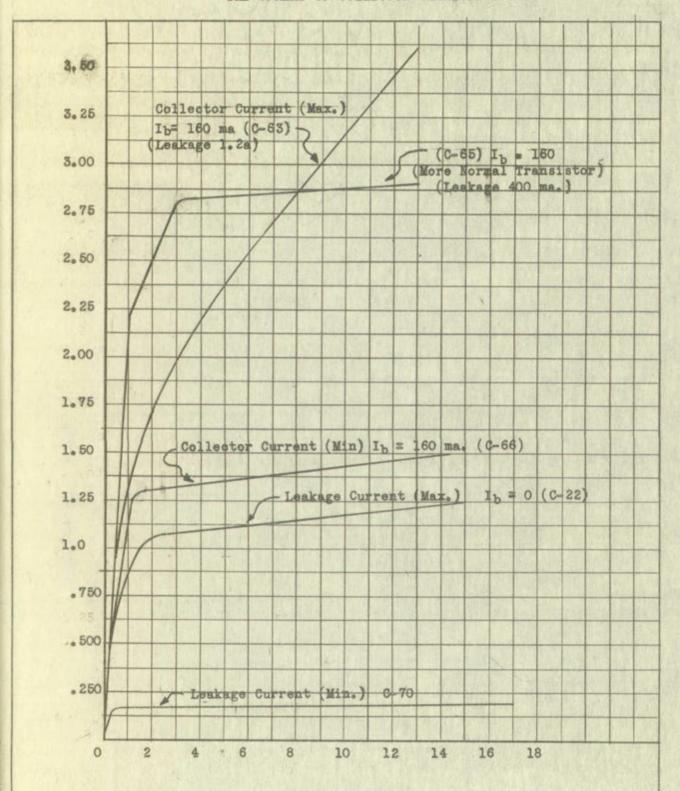
It is from these characteristics that the following spread of collector and base characteristics was taken.

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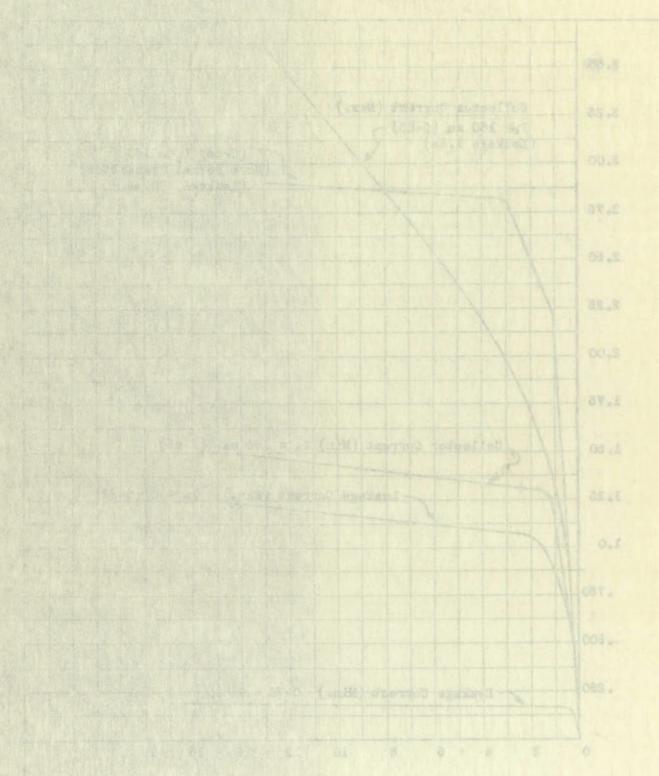
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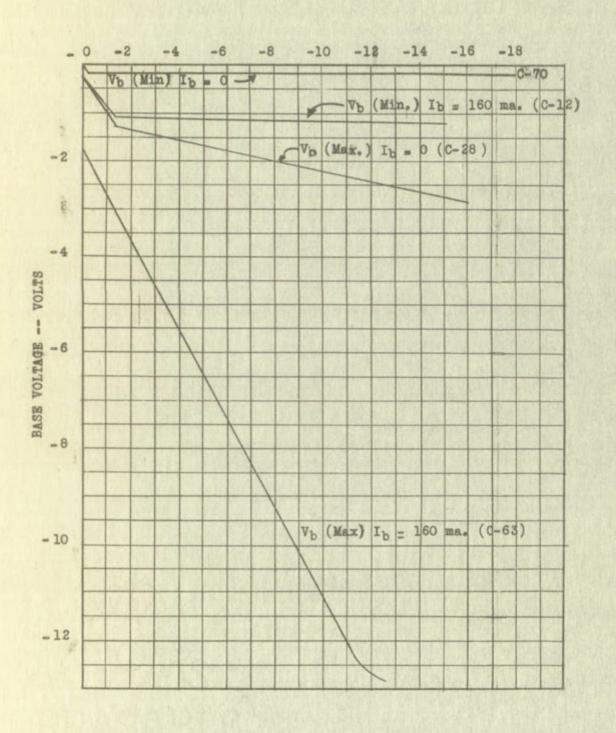


Collector Voltage - Volts



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THE SPREAD OF BASE CHARACTERISTICS Collector Voltage -- Volts



APPENDIX C

PERCENT DUTY FACTOR CURVES

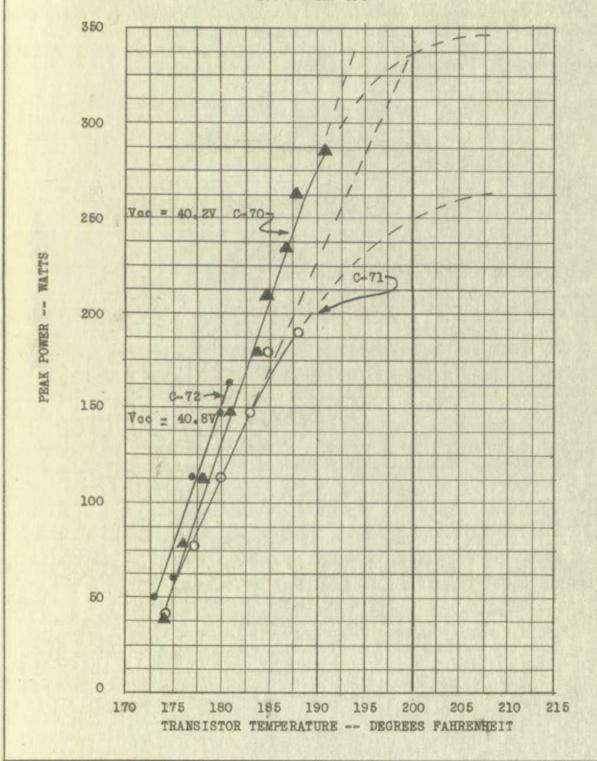
The curves contained in this appendix are the plot of the various collector peak power points against transistor temperature taken directly from the experimental data. This set of curves was plotted for the purpose of extending the pulse power into the failure region which had not been done experimentally.

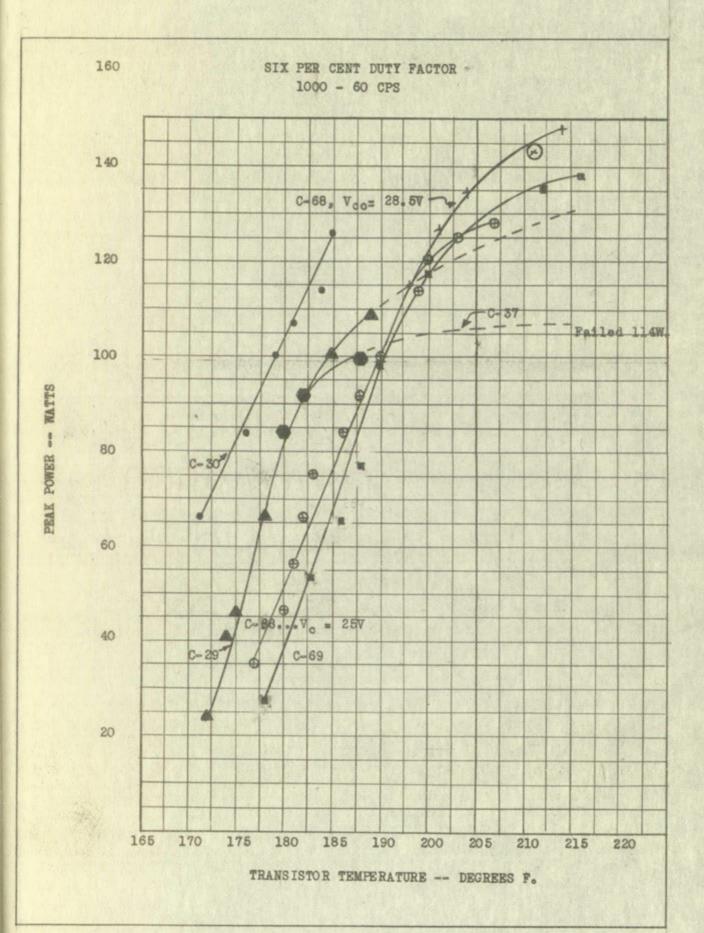
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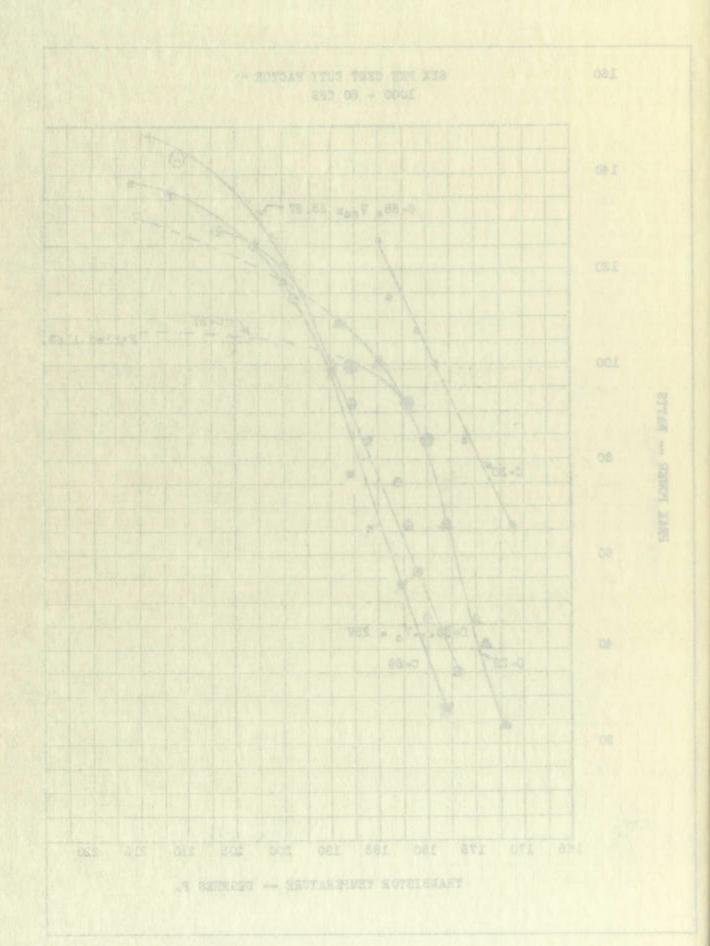
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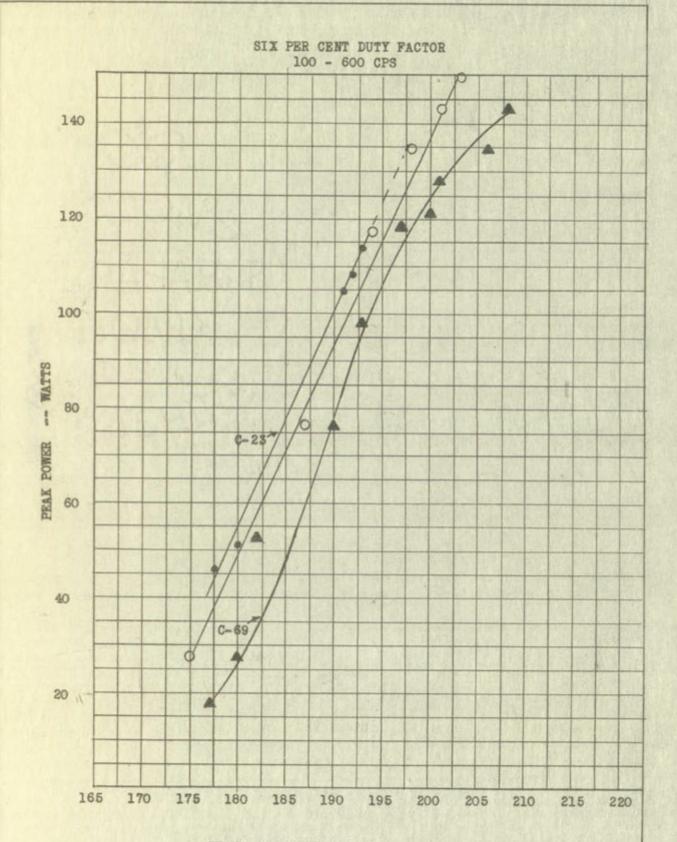
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100 -- 200 CPS

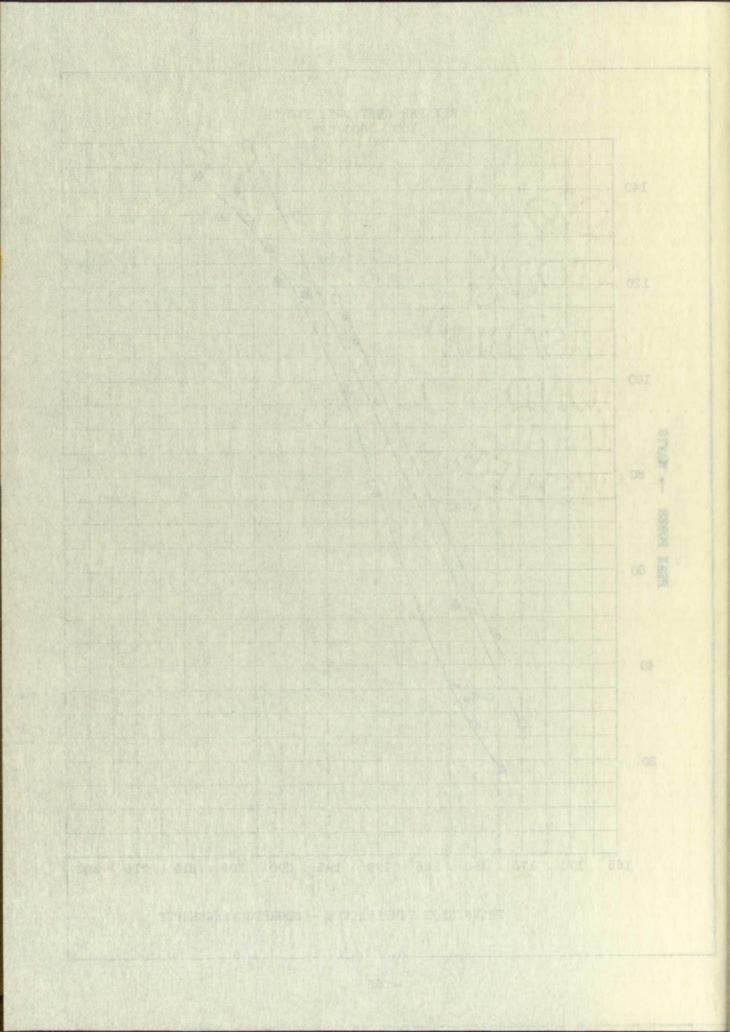




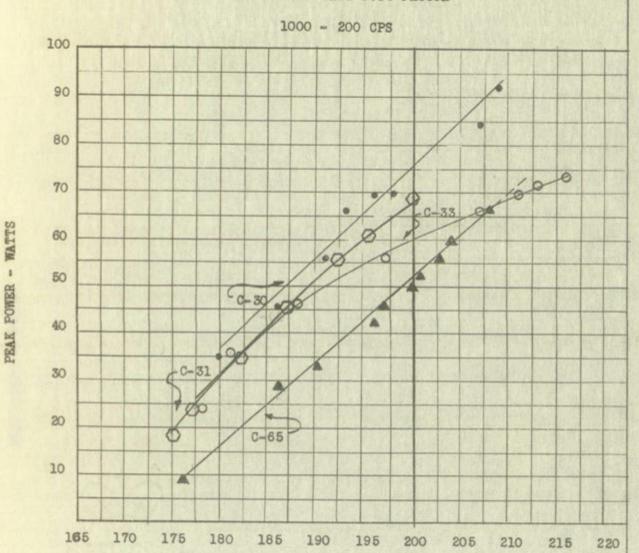




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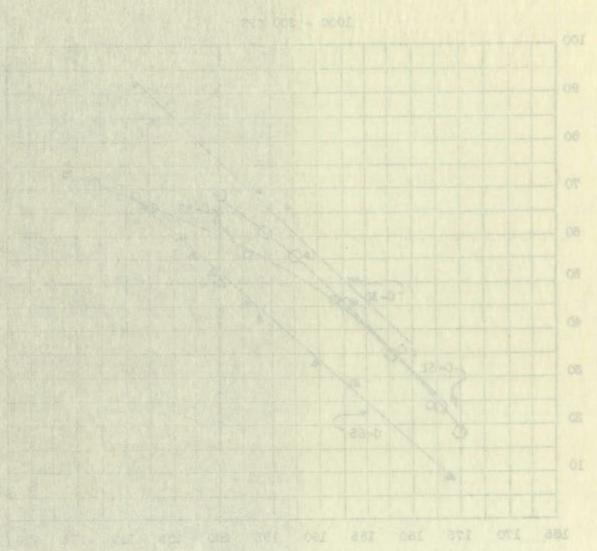


TWENTY PER CENT DUTY FACTOR



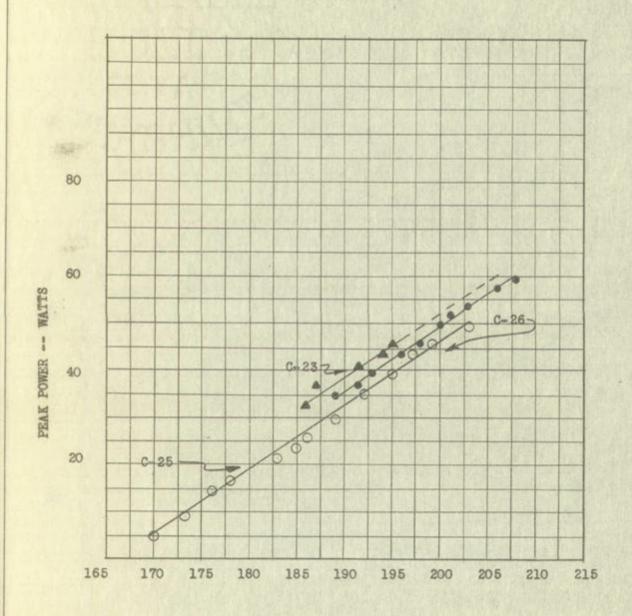
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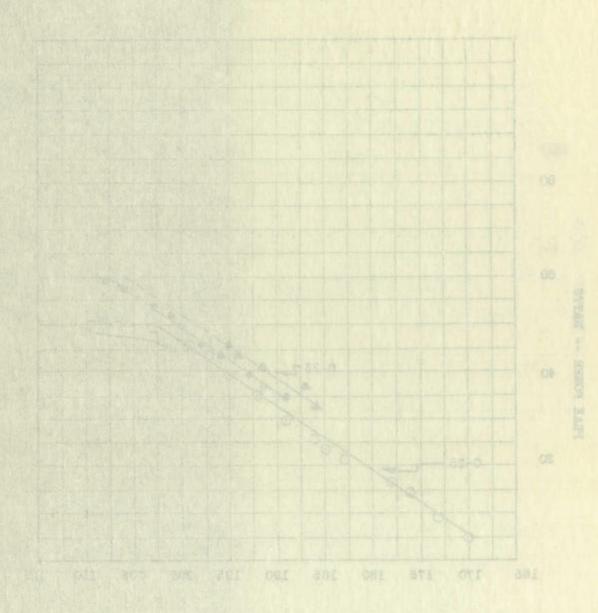
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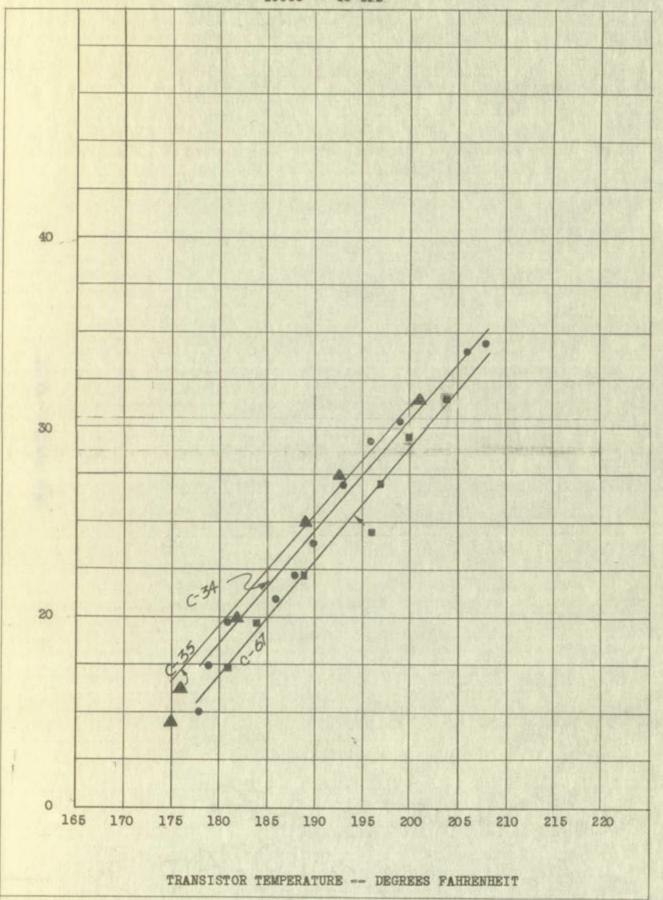
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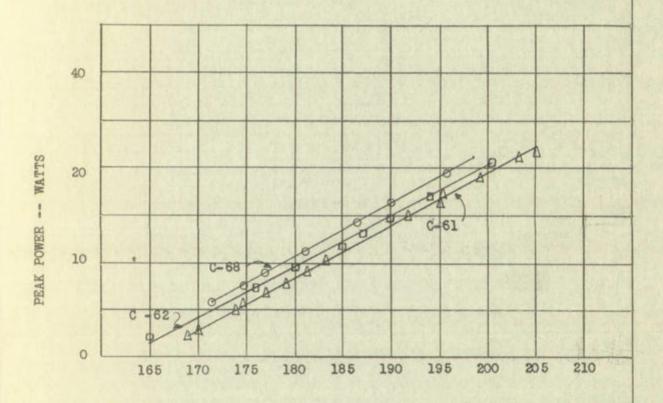


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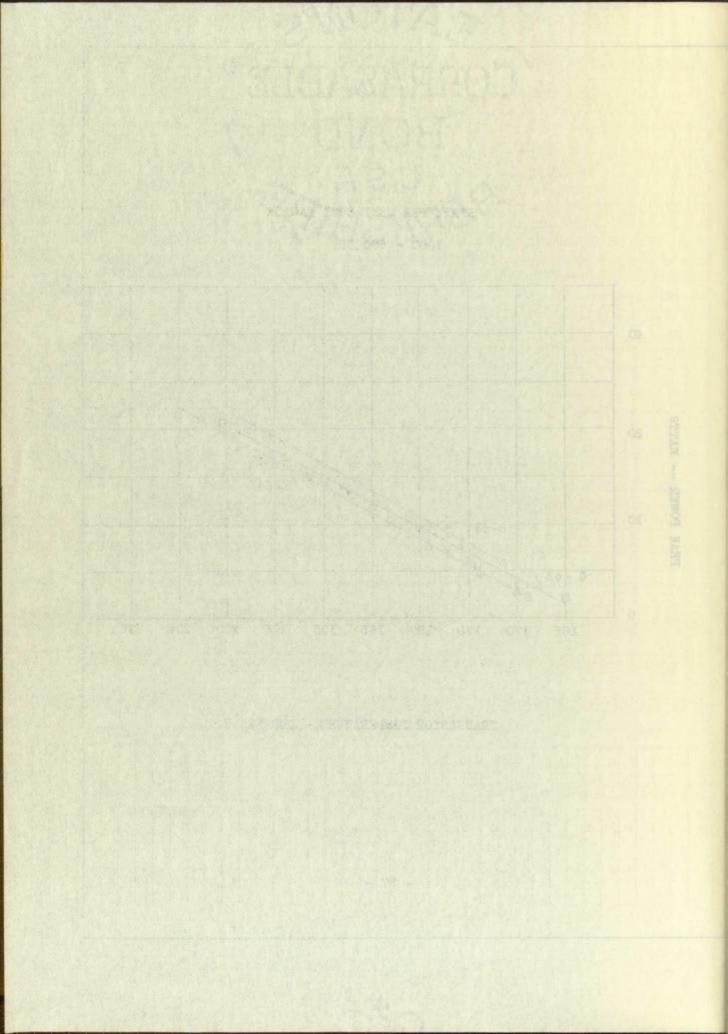
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SIXTY PER CENT DUTY FACTOR

1000 - 600 CPS



TRANSISTOR TEMPERATURE - DEGREES F.



APPENDIX D

DEVELOPMENT OF THE OPERATING TEMPERATURE EQUATION

The equation for the power-temperature curves in terms of temperature is

$$T_{op} = 165 + \frac{\overline{P}}{\tan \theta} \tag{2}$$

where 0 is the angle between the curve and the temperature axis. The transistor temperature rise per watt of average power is dependent on the thermal resistance of the transistor and the heat tan Q is a measurement of this thermal resistance which is dependent on the area and material used in making the transistor and heat sink. Since only one type of transistor and heat sink is involved here, then $\frac{1}{\tan \theta}$ is directly proportional to K, a constant dependent on thermal resistance of the materials, and inversely proportional to the area of the heat sink.

Then an equation can be developed using the length of the side of the heat sink which should be helpful in circuit design,

First using a 6th square sink, the slope of the curve is _______ Then using only the last part of the equation

 $\frac{\overline{P}}{\tan \theta} = \frac{\overline{P}}{.36}$ $\frac{K\overline{P}}{L^2} \quad \text{where } K = kL, \text{ both } k's \text{ are constant};$ so for the 6th sink

$$\frac{k_1 L_1 \overline{P}}{(\text{area})_1} = \frac{k_1 L_1 \overline{P}}{L_1^2} = \frac{\overline{P}}{.36}$$

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then

$$k_1 = \frac{I_1}{\tan \theta} = \frac{6}{.36} = .167 \approx 17$$

for the 4.25" sink

$$\frac{k_2}{L_2} = \frac{1}{\tan \theta_2} = \frac{1}{.27} , k_2 = \frac{L_2}{.27} = \frac{1}{.27}$$

$$\frac{1}{.27} = 15.7 \approx 17$$

for the 3" square sink

$$\frac{k_3}{L_3} = \frac{1}{.18}, \quad k_3 = \frac{3}{.18} = 16.6 \approx 17$$

Since k1 & k2 & k3, equation 2 can be expressed as follows:

$$T_{\rm op} = 165 \Rightarrow \frac{17\overline{P}}{L}$$

where

T = temperature indegrees fahrenheit of a transistor with a square heat sink of side L inches long dissipating as average power P. The heat sink material is described in Chapter II, THE METHOD OF TEST.

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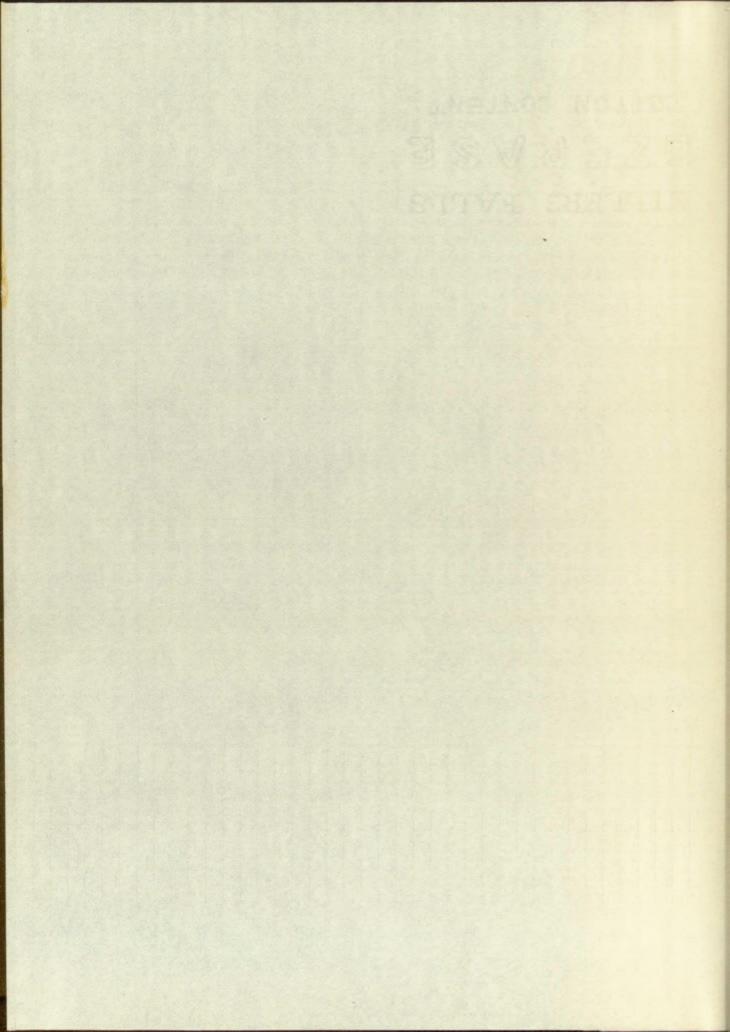
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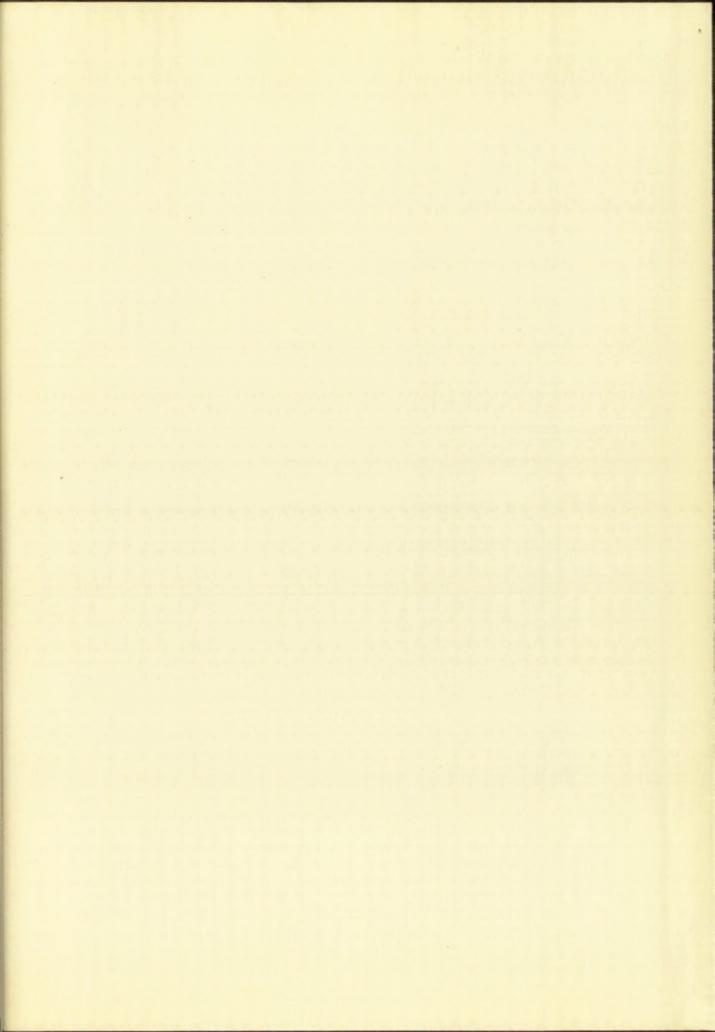
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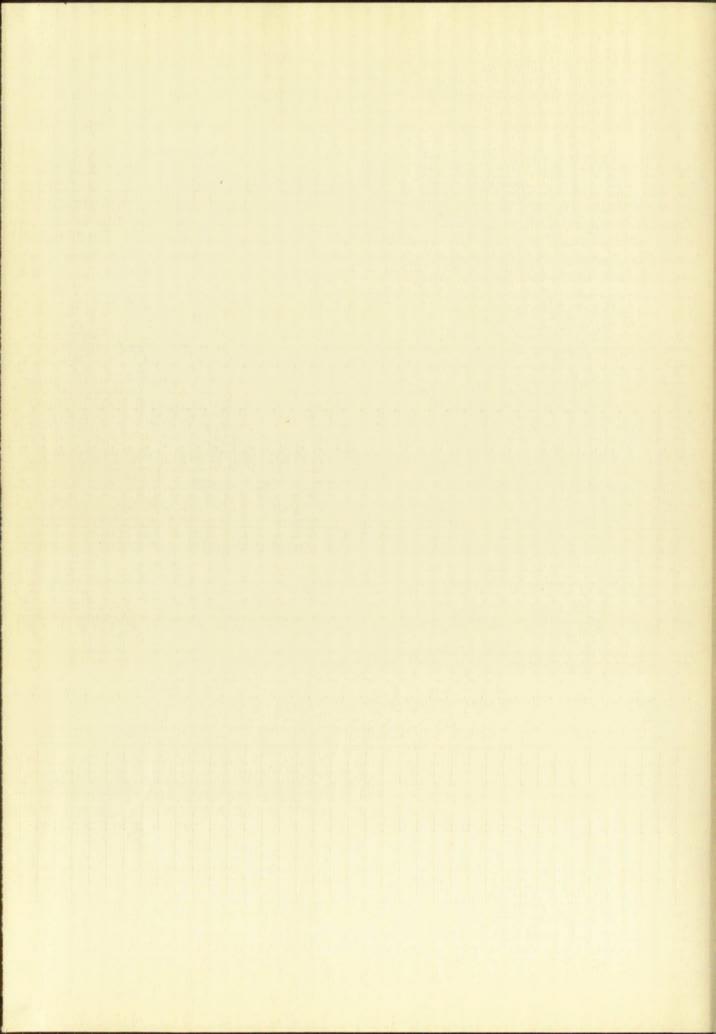
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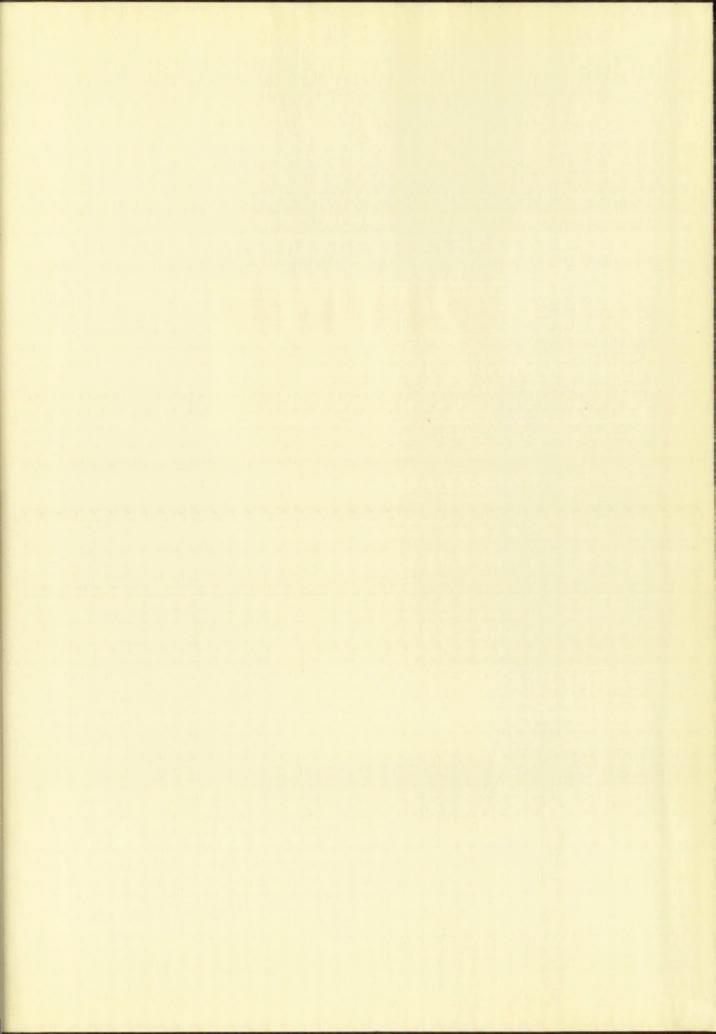
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