

Gain Antenna Measurement using Single Cut Near Field Measurements

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Abstract— Some antennas require rapid validation at a reduced measurement distance while maintaining sufficient accuracy in the determination of pertinent antenna parameters such as gain. In particular, for cellular base station antennas in production phase the measurement time can be a limitation. In these cases, a rapid check of the radiation performance in the two main planes is sufficient. Other examples are phase arrays with high degree of steering that would require considerable measurement time for characterizing all steering positions.

This paper presents a near-field antenna test procedure providing single or double main plane patterns including the gain. The procedure is applicable to antennas, with separable excitation in the two main planes. The test set-up is based on an azimuth positioner and near to far-field transformation based on expansion in cylindrical modes. The paper shows results for gain measurements. Near to far-field transformation is performed using the cylindrical modes expansion assuming a zero-height cylinder. This allows the use of a FFT in the calculation of the far field pattern including probe correction. In the case of gain, the near to far-field transformation factor is calculated for bore sight direction, taking advantage of the separable excitation properties of the antenna. This factor is used in the gain calculation by comparison technique.

I. INTRODUCTION

In the literature, different methods for reducing antenna measurement time have been explored. One of these techniques is single cut near field to far field transformation. For single cut, near to far-field transformation, one of the first methods is the discrete beam sampling method (DBSM), and it is explained in [1-3]. That algorithm is based on the Fourier series decomposition of the $1/R^2$ term of the field propagation in spherical coordinates, and it is valid for Fresnel Zone Measurements, where the $1/R^3$ term can be neglected. Technical University of Madrid (UPM) implemented this method for the measurements of the ASAR panels of the ENVISAT satellite [4] in the two main planes of the forward hemisphere ($-90^\circ < \theta < 90^\circ$). Later, it was adapted for the measurement of the two main planes and gain of BTS antennas in the whole angular range by processing the measurements in two halves and then combining the results [5]. One of the limitations of this algorithm is the absence of probe correction.

F. Las Heras proposed in [6] another algorithm based on a least-squares algorithm to include probe correction for the same

application. Also, in [7], F. Las Heras solved the problem of the calculation of the main plane far-field radiation with a method based on the reconstruction of equivalent magnetic currents (EMCs) using decoupled integral equations and one-dimensional source components.

The authors presented the algorithm used in this paper for the calculation of the radiation pattern in [8]. This algorithm is based in the cylindrical mode decomposition of the measurements on a ring, considered as a cylinder of zero height, and the fundamentals are explained in the next section.

R. Cornelius et al. presented in [9] the comparison of results using three different algorithms: this last algorithm, the one based on the Fourier series decomposition and one implemented by R. Cornelius based on the spherical wave expansion of the measurement in one cut. That paper analysed the advantages and drawbacks of each of the three methods, concluding that the method based on cylindrical mode expansion was more appropriate since it could work directly with the full ring and included probe correction. That paper only presented the antenna radiation patterns, without gain determination. This paper shows the method for the antenna gain calculation for one or two cuts and shows measured results for some specific antennas.

The paper is structured in this way: section 2 presents the algorithms for single or double cuts gain determination using single cut near to far field transformation. Section 3 presents the validation of the method through results of the measurement of different antennas (horns, base station arrays, reflectors) to analyze the limitations of the algorithm. Section 4 includes the conclusions and the possible implementation in a production test system.

II. ALGORITHM FOR ANTENNA GAIN CALCULATION

This section explains briefly the fundamentals of the single cut to NF/FF transformation and gain calculation algorithm. The algorithm is based on the cylindrical NF/FF transformation [10], but applied to a zero height cylinder. In that sense, any of the rings of the sphere can be considered as a zero height cylinder. Since only one spatial dimension is considered, the complexity of the algorithm is reduced. However, probe correction works in the same way of the classical cylindrical NF/FF transformation. The work presented in [8] explained the fundamentals of this single cut NF/FF transformation. The presented results were satisfactory for the measurement of the main planes of antennas with separable excitation.

Figure 1 shows, as an example, the measurement of the FF for one single cut of an X-band slotted array antenna having a diameter of approximately 31 cm. The measurement of only the E-plane of the antenna was performed at 9.3 GHz and at measurement distance of 2 meters. The FF distance for this antenna is approximately 6 meters at 9.3 GHz. In red, it is shown the results using this method, in blue using conventional spherical NF/FF transformation technique [11-12]. As it can be observed, the results agree pretty well for both the co-polar and cross-polar radiation pattern.

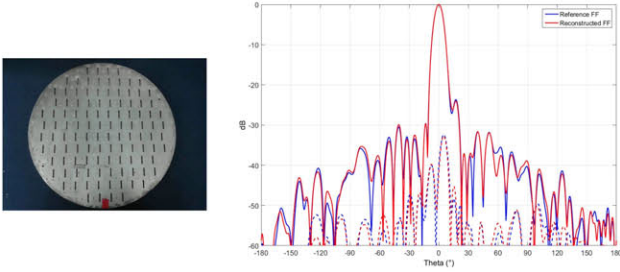


Figure 1. Application of single cut NF/FF transformation to the E-plane of a slotted array antenna.

The gain can be calculated through the NF substitution technique using a Standard Gain Horn (SGH) which gain has been already calibrated (1)

$$G_{AUT}(dBi) = G_{SGH}(dBi) + P_{AUT,FF,\varphi=\varphi_0}(dB) - P_{SGH,FF,\varphi=\varphi_0}(dB) \quad (1)$$

where G_{AUT} is the gain of the Antenna Under Test (AUT), G_{SGH} is the gain of the SGH, and $P_{AUT,FF,SGH,FF}$ are the received power at a specific angular direction using either AUT either SGH. Since, the measurement is carried out in NF, $P_{AUT,FF,\varphi=\varphi_0}$ and $P_{SGH,FF,\varphi=\varphi_0}$ are not known. The measured values are the ones in NF and can be expressed as the sum of the FF values and a NF/FF transformation factor as reported in (2)

$$P_{AUT/SGH,NF,\varphi=\varphi_0}(dB) = P_{AUT/SGH,FF,\varphi=\varphi_0}(dB) + Ww_{AUT/SGH} \quad (2)$$

The NF/FF factor accounts for the modification of the AUT pattern when transforming from NF to FF and it can be computed as the difference between the FF and NF directivity in a specified direction (e.g. $\theta = 0$).

If the FF pattern is normalized maintaining constant the power radiated through the ring in both NF and FF, the NF/FF transformation factor becomes close to zero when the measurement is performed in FF. Therefore, in the case of the SGH, and for most of the cases, this factor can be neglected, making the measurement process faster.

In the case of the AUT, it is assumed that the excitation in the antenna aperture is separable. Therefore, the field can be expressed as a multiplication of two factors, each one depending on the aperture excitation on each main plane. In the same way, the NF/FF factor can therefore be expressed as the product of the NF/FF factor on each main plane. Also in the case of linear arrays, or thin apertures, it can be considered that the NF/FF factor in the thin side is also close to zero. In this

case, only one cut would be necessary to characterize the gain of the AUT. The key point of this procedure is thus the calculation of this NF/FF factor.

Figure 2 shows the calculation of the NF/FF factor at difference measurement distances for the slotted array shown in Figure 1. Red trace shows the NF/FF factor calculated from full sphere; blue trace shows the one computed as the product of the NF/FF factor for the two main cuts. The green curve shows instead the difference between the other two traces. Good agreement is obtained for this antenna if the measurement is not carried out at an extreme NF distance.

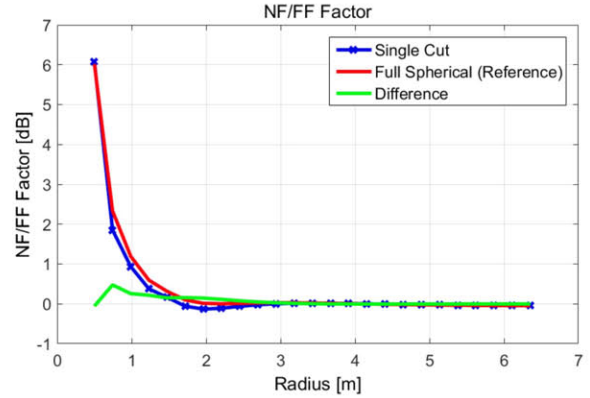


Figure 2. Comparison of the NF/FF factor for a planar slotted array calculated through the information of the full sphere or the product of the two main planes.

III. APPLICATION TO DIFFERENT ANTENNAS

This section shows the validation of the gain calculation algorithm for different cases. The antennas have been measured in the facilities of LEHA-UPM [13]. The results are compared with the gain calculated using Spherical NF/FF transformation software [12] and full spherical acquisition and also with measurements performed in other ranges (where possible). Three different antennas have been tested; some of them have been measured in NF, others in Fresnel zone and some of them close to far field distance.

A. Gain measurement results for BTS1940 antenna.

The first antenna is the MVI BTS1940 antenna [14] which dimensions are 1400x200 mm. This antenna is a linear array used as reference antenna to evaluate the quality of base station antenna measurement systems. Figure 3 shows the antenna itself and the comparison of the results of the gain when the antenna has been measured in Fresnel Zone (5.35 meters), while the FF distance is between 21 and 31 m, depending on the frequency. Single or dual cut slightly overestimate the gain, but the difference is bounded in 0.5 dB using one cut (violet trace) and 0.4 dB using two cuts (green trace) in the worst case.

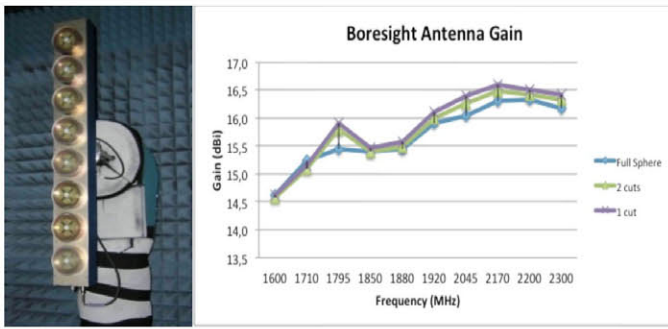


Figure 3. Comparison of gain measurements for BTS1940 antenna.

B. Gain measurement results for SH2000 antenna.

The second antenna is the MVI dual ridge horn antenna SH2000 [14] working in the frequency band from 2 to 32 GHz. The overall dimensions are 62.4x108x103.7 mm. This antenna is a standard gain horn based on a ridged design combining stable gain performance and low VSWR with ultra-wide band frequency operation. The horn is single linear polarized with excellent cross-polar discrimination, ideal for gain calibration of antenna measurement systems or as wideband probes in classical FF test ranges. The antenna has been measured at the LEHA-UPM spherical near field system, at a measurement distance of 3.86 meters, in the frequency band from 18.2 to 31 GHz for this exercise. For these frequencies, the FF goes from 1.9 to 3.2 m, therefore the antenna is in FF. In any case, it is observed in Figure 4 that the double cut NF/FF transformation (green trace) corrects the small error occurred when finite FF distance (violet trace) is used (between 0.1 and 0.2 dB in this case depending on the frequency).

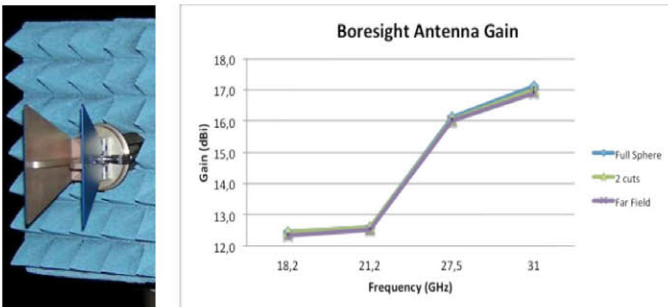


Figure 4. Comparison of gain measurements for SH2000 antenna.

C. Gain measurement results for SR40 antenna.

The third antenna is the MVI SR40 offset reflector antenna [14]. The dimensions of the antenna are 561x400x315mm. It is fabricated with Aluminium, and excited with the SH2000 feeder (2-32 GHz). The antenna has been measured in the frequency band from 10.7 to 38 GHz, at a distance of 5.45m. The far field distance is from 33 to 120 m at those frequencies, therefore, the measurement is clearly performed at a NF distance.

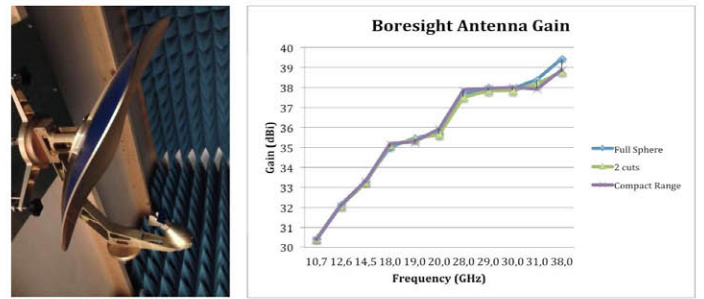


Figure 5. Comparison of gain measurements for SR40 antenna.

Figure. 5 shows the gain measurements using full sphere NF method (blue trace), the two cuts method (green trace) and the measurement in a compact range (violet trace). In the worst case, the difference between full sphere and 2 cuts is 0.6 dB (at 38 GHz). However, even in this case, the 2 cuts agrees pretty well with the measurement in compact range. For the other cases, the difference between full sphere and 2 cuts in less than 0.2 dB.

IV. CONCLUSIONS

Single cut NF/FF transformation has been applied in the gain measurement of different separable excitation antennas using substitution technique. In the case of a BTS antenna (linear array), the method can be applied with only information of the main cut (vertical pattern). In the case of the other two antennas (the horn and offset reflector) the NF/FF transformation factor, and therefore, the gain, can be calculated through the measurement and individual processing of the NF/FF transformation factor in the two main cuts.

The measurements of these three antennas, performed in different measurement scenarios (Fresnel Zone, FF at a finite distance and NF) conclude that this algorithm improves the results with respect FF measurement and gets enough accurate results (comparable to other measurement systems as compact range) and the uncertainty is bounded into a reasonable value. Future works will emphasize in the limitations of this algorithm and the performance depending on the measurement distance.

The main application for this work is the possibility of making very fast measurement in the production phase of commodity antennas. Antenna gain and pattern in the main pattern cuts can be characterized (almost) in real time using this method in combination with a multiprobe system.

ACKNOWLEDGEMENTS

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